



÷1, ÷2 DIFFERENTIAL-TO-LVPECL CLOCK GENERATOR

ICS87321I

GENERAL DESCRIPTION

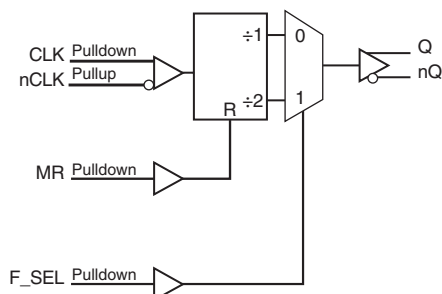


The ICS87321I is a high performance ÷1, ÷2 Differential-to-LVPECL Clock Generator and a member of the HiPerClockS™ family of High Performance Clock Solutions from IDT. The CLK, nCLK pair can accept most standard differential input levels. The ICS87321I is characterized to operate from a 3.3V or 2.5V power supply. Guaranteed part-to-part skew characteristics make the ICS87321I ideal for those clock distribution applications demanding well defined performance and repeatability.

FEATURES

- One differential LVPECL output
- One CLK, nCLK input pair
- CLK, nCLK pair can accept the following differential input levels: LVPECL, LVDS, LVHSTL, SSTL, HCSL
- Maximum clock input frequency: 700MHz
- Translates any single ended input signal (LVCMOS, LVTTTL, GTL) to LVPECL levels with resistor bias on nCLK input
- Part-to-part skew: 600ps (maximum)
- Propagation delay: 1.8ns (maximum)
- Additive phase Jitter, RMS: 0.18ps
- Full 3.3V or 2.5V operating supply
- -40°C to 85°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages

BLOCK DIAGRAM



PIN ASSIGNMENT

CLK	1	8	Vcc
nCLK	2	7	Q
MR	3	6	nQ
F_SEL	4	5	VEE

ICS87321I

8-Lead SOIC

3.90mm x 4.90mm x 1.37mm package body

M Package

Top View

TABLE 1. PIN DESCRIPTIONS

Number	Name	Type		Description
1	CLK	Input	Pulldown	Non-inverting differential clock input.
2	nCLK	Input	Pullup	Inverting differential clock input.
3	MR	Input	Pulldown	Active High Master Reset. When logic HIGH, the internal dividers are reset causing the true outputs (Q) to go low and the inverted outputs (nQ) to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS / LVTTTL interface levels. See Table 3.
4	F_SEL	Input	Pulldown	Selects divider value for Q, nQ outputs as described in Table 3. LVCMOS / LVTTTL interface levels.
5	V _{EE}	Power		Negative supply pin.
6, 7	nQ, Q	Output		Differential output pair. LVPECL interface levels.
8	V _{CC}	Power		Positive supply pin.

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance			4		pF
R _{PULLUP}	Input Pullup Resistor			51		kΩ
R _{PULLDOWN}	Input Pulldown Resistor			51		kΩ

TABLE 3. FUNCTION TABLE

MR	F_SEL	Divide Value
1	X	Reset: Q output low, nQ output high
0	0	÷1
0	1	÷2

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V_{CC}	4.6V
Inputs, V_I	-0.5V to $V_{CC} + 0.5$ V
Outputs, I_O	
Continuous Current	50mA
Surge Current	100mA
Package Thermal Impedance, θ_{JA}	95°C/W (0 lfpm)
Storage Temperature, T_{STG}	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 4A. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ\text{C}$ TO 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Positive Supply Voltage		3.135	3.3	3.465	V
I_{EE}	Power Supply Current				18	mA

TABLE 4B. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ\text{C}$ TO 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Positive Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current				16	mA

TABLE 4C. LVCMOS/LVTTL DC CHARACTERISTICS, $V_{CC} = 3.3V \pm 5\%$ OR $2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ\text{C}$ TO 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage	$V_{CC} = 3.3V$	2		$V_{CC} + 0.3$	V
		$V_{CC} = 2.5V$	1.7		$V_{CC} + 0.3$	V
V_{IL}	Input Low Voltage	$V_{CC} = 3.3V$	-0.3		0.8	V
		$V_{CC} = 2.5V$	-0.3		0.7	V
V_{HYS}	Input Hysteresis	MR, F_SEL	100			mV
I_{IH}	Input High Current	MR, F_SEL $V_{CC} = V_{IN} = 3.465V$ or $2.625V$			150	μA
I_{IL}	Input Low Current	MR, F_SEL $V_{CC} = 3.465V$, or $2.625V$ $V_{IN} = 0V$	-5			μA

TABLE 4D. DIFFERENTIAL DC CHARACTERISTICS, $V_{CC} = 3.3V \pm 5\%$ OR $2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ\text{C}$ TO 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
I_{IH}	Input High Current	CLK $V_{CC} = V_{IN} = 3.465V$ or $2.625V$			150	μA
		nCLK $V_{CC} = V_{IN} = 3.465V$ or $2.625V$			5	μA
I_{IL}	Input Low Current	CLK $V_{CC} = 3.465V$, or $2.625V$ $V_{IN} = 0V$	-5			μA
		nCLK $V_{CC} = 3.465V$ or $2.625V$, $V_{IN} = 0V$	-150			μA
V_{PP}	Peak-to-Peak Input Voltage; NOTE 1		0.15		1.3	V
V_{CMR}	Common Mode Input Voltage; NOTE 1, 2		$V_{EE} + 0.5$		$V_{CC} - 0.85$	V

NOTE 1: V_{IL} should not be less than -0.3V.

NOTE 2: Common mode voltage is defined as V_{IH} .

TABLE 4E. LVPECL DC CHARACTERISTICS, $V_{CC} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ TO $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CC} - 1.4$		$V_{CC} - 0.9$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CC} - 2.0$		$V_{CC} - 1.7$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.65		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CC} - 2V$.**TABLE 4E. LVPECL DC CHARACTERISTICS, $V_{CC} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ TO $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage; NOTE 1		$V_{CC} - 1.4$		$V_{CC} - 0.9$	V
V_{OL}	Output Low Voltage; NOTE 1		$V_{CC} - 2.0$		$V_{CC} - 1.5$	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.4		1.0	V

NOTE 1: Outputs terminated with 50Ω to $V_{CC} - 2V$.**TABLE 5A. AC CHARACTERISTICS, $V_{CC} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ TO $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f_{CLK}	Clock Input Frequency				700	MHz
t_{PD}	Propagation Delay; CLK to Q (Dif) NOTE 1		1.2		1.8	ns
t_{jit}	Buffer Additive Phase Jitter, RMS; refer to Additive Phase Jitter Section	155.52MHz, Integration Range: (12kHz - 20MHz)		0.18		ps
$t_{sk(pp)}$	Part-to-Part Skew; NOTE 2, 3				600	ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	100		600	ps
odc	Output Duty Cycle; NOTE 4		48		53	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs on different devices operating at the same supply voltages and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 4: Input duty cycle must be 50%.

TABLE 5B. AC CHARACTERISTICS, $V_{CC} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ TO $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f_{CLK}	Clock Input Frequency				700	MHz
t_{PD}	Propagation Delay; CLK to Q (Dif) NOTE 1		1.2		1.8	ns
f_{jit}	Buffer Additive Phase Jitter, RMS; refer to Additive Phase Jitter Section	155.52MHz, Integration Range: (12kHz - 20MHz)		0.18		ps
$t_{sk(pp)}$	Part-to-Part Skew; NOTE 2, 3				600	ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	100		600	ps
odc	Output Duty Cycle; NOTE 4		45		55	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs on different devices operating at the same supply voltages and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

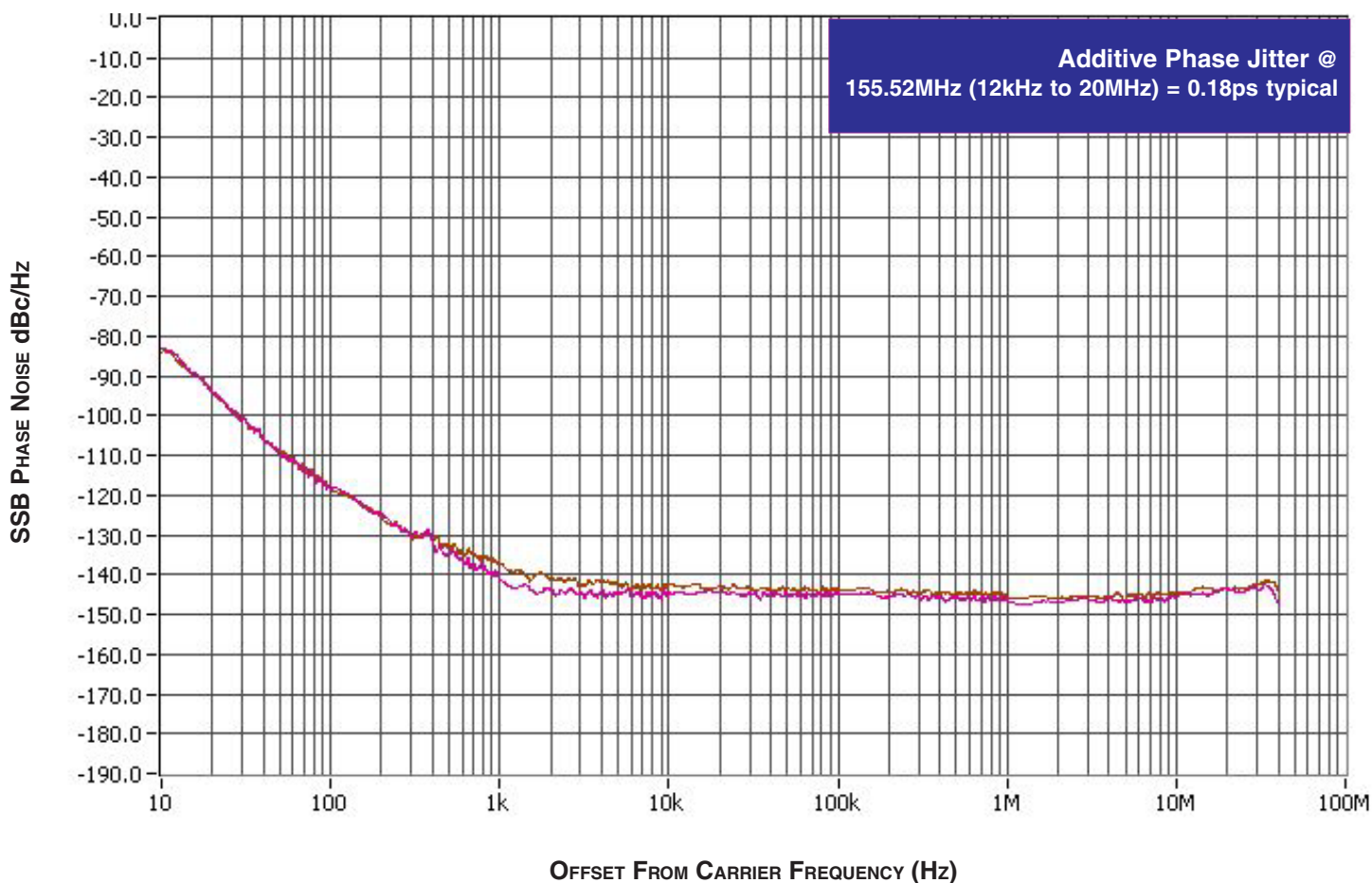
NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 4: Input duty cycle must be 50%.

ADDITIVE PHASE JITTER

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise**. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1Hz

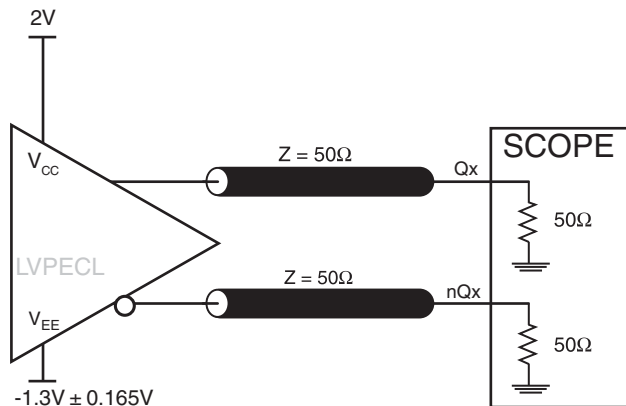
band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



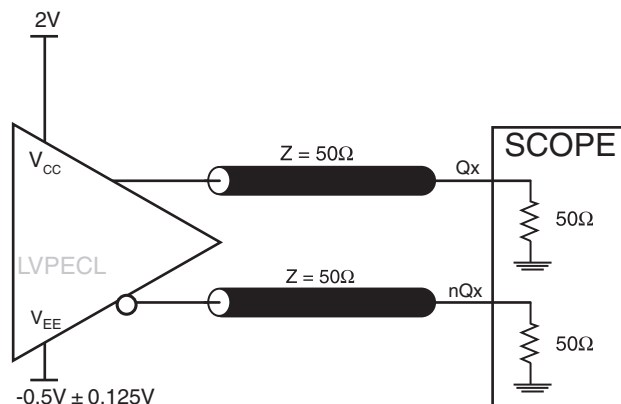
As with most timing specifications, phase noise measurements have issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the

device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

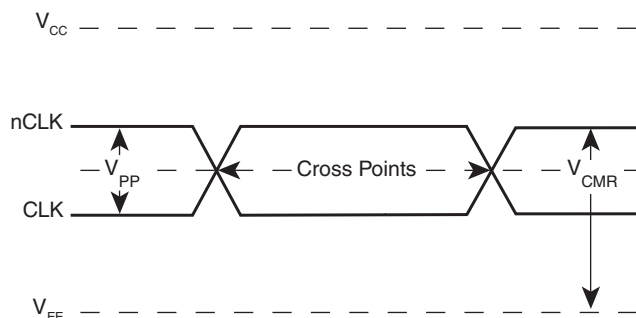
PARAMETER MEASUREMENT INFORMATION



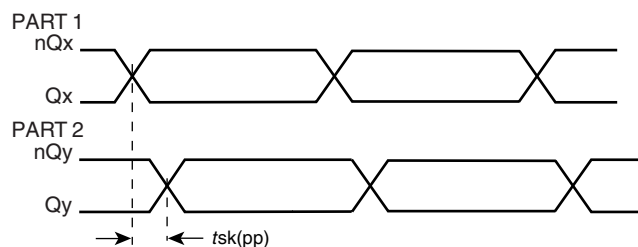
3.3V OUTPUT LOAD AC TEST CIRCUIT



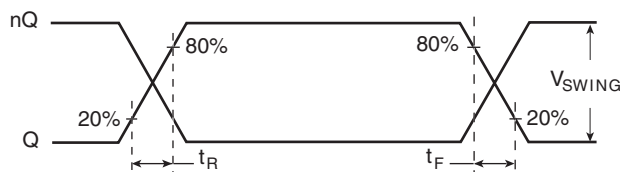
2.5V OUTPUT LOAD AC TEST CIRCUIT



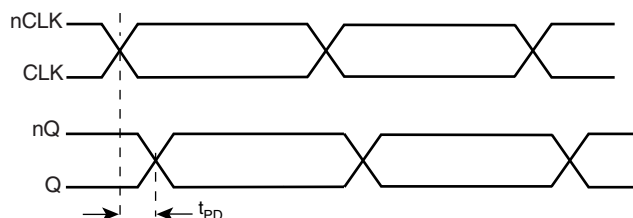
DIFFERENTIAL INPUT LEVEL



PART-TO-PART SKEW

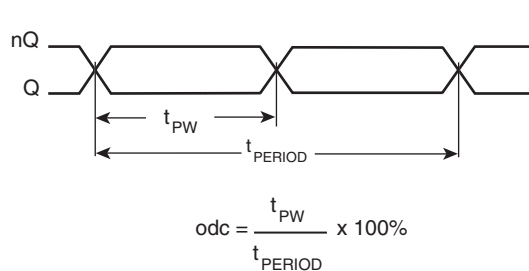


OUTPUT RISE/FALL TIME



PROPAGATION DELAY

PARAMETER MEASUREMENT INFORMATION, CONTINUED



OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD

APPLICATION INFORMATION

WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 1 shows how the differential input can be wired to accept single ended levels. The reference voltage $V_{REF} = V_{CC}/2$ is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio

of R1 and R2 might need to be adjusted to position the V_{REF} in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and $V_{CC} = 3.3V$, V_{REF} should be 1.25V and $R2/R1 = 0.609$.

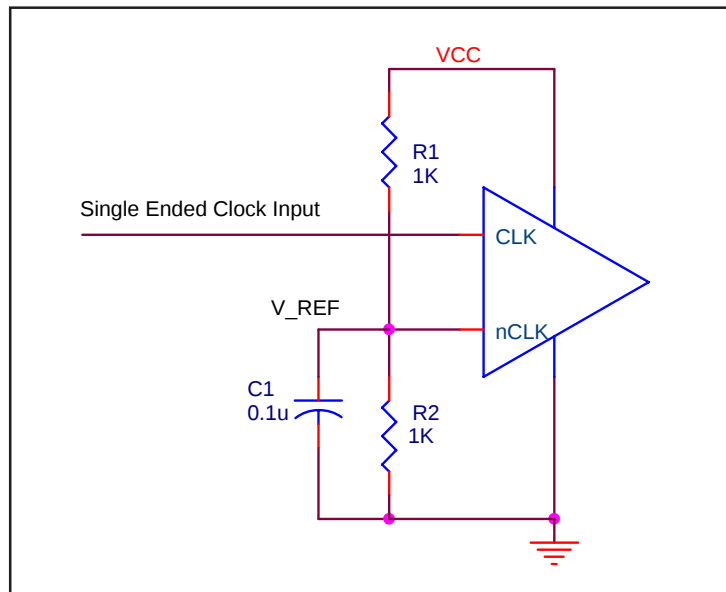


FIGURE 1. SINGLE ENDED SIGNAL DRIVING DIFFERENTIAL INPUT

DIFFERENTIAL CLOCK INPUT INTERFACE

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSSL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. Figures 2A to 2E show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are

examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in *Figure 2A*, the input termination applies for IDT HiPerClockS LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

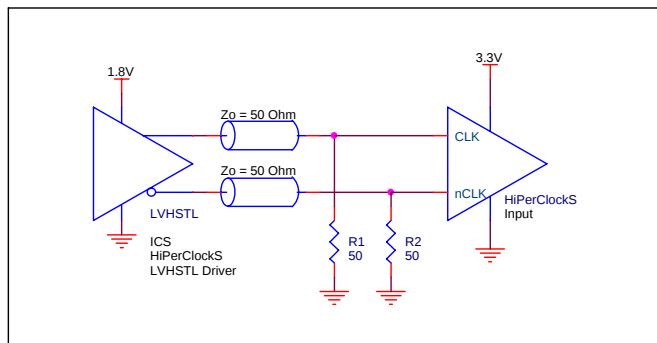


FIGURE 2A. HiPerClockS CLK/nCLK INPUT DRIVEN BY IDT HiPerClockS LVHSTL DRIVER

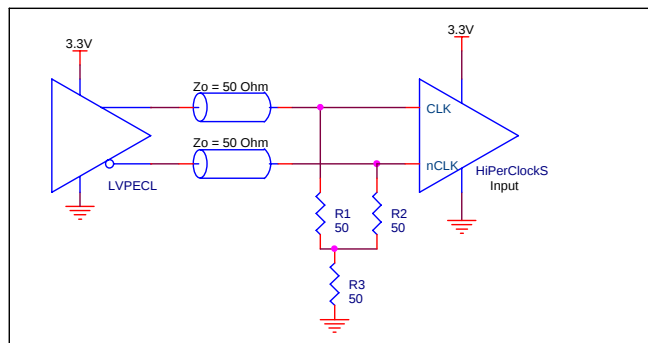


FIGURE 2B. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

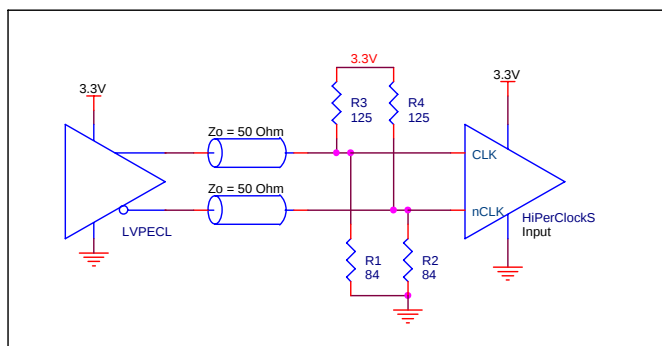


FIGURE 2C. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

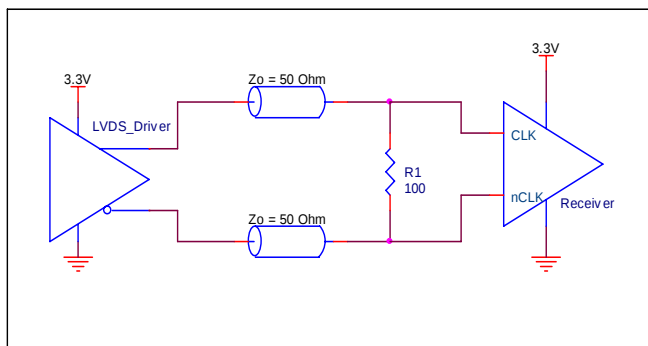


FIGURE 2D. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVDS DRIVER

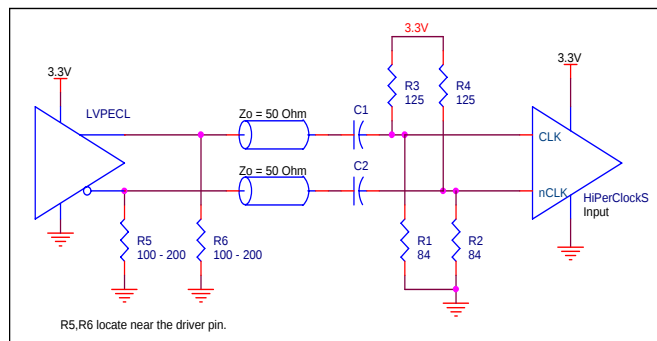


FIGURE 2E. HiPerClockS CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER WITH AC COUPLE

RECOMMENDATIONS FOR UNUSED INPUT PINS

INPUTS:

LVC MOS CONTROL PINS

All control pins have internal pulldowns; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

TERMINATION FOR 3.3V LVPECL OUTPUTS

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 3A and 3B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

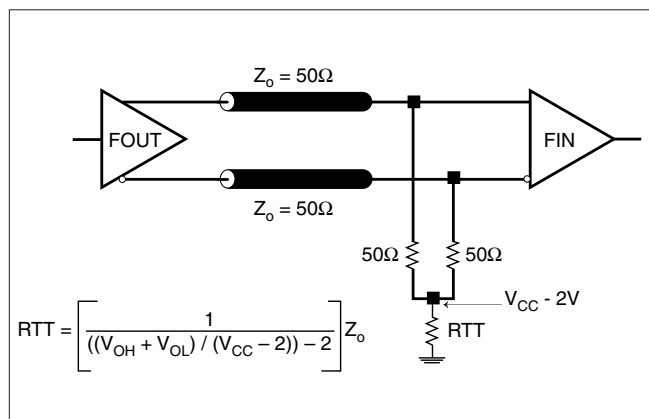


FIGURE 3A. LVPECL OUTPUT TERMINATION

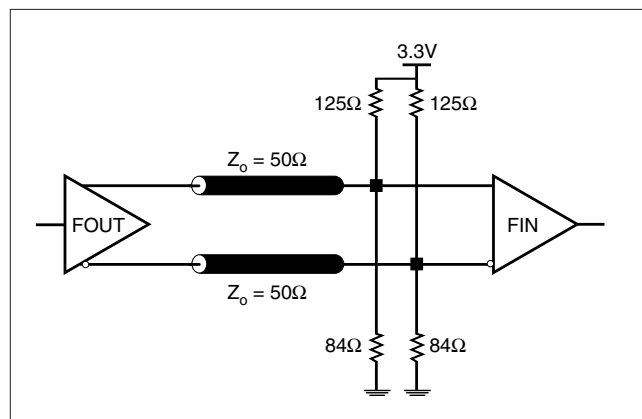


FIGURE 3B. LVPECL OUTPUT TERMINATION

TERMINATION FOR 2.5V LVPECL OUTPUTS

Figure 4A and Figure 4B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{CC} - 2V$. For $V_{CC} = 2.5V$, the $V_{CC} - 2V$ is very close to ground

level. The R3 in Figure 4B can be eliminated and the termination is shown in Figure 4C.

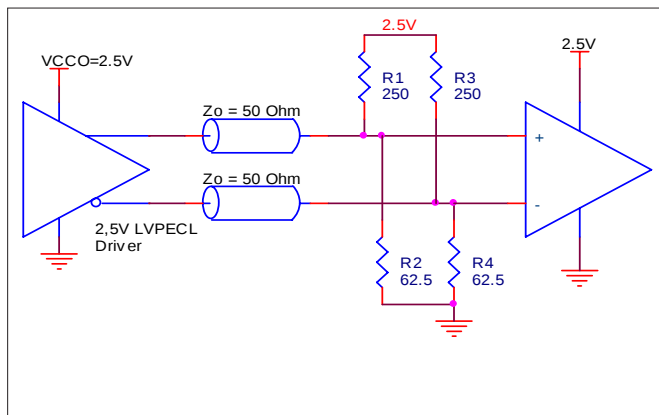


FIGURE 4A. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

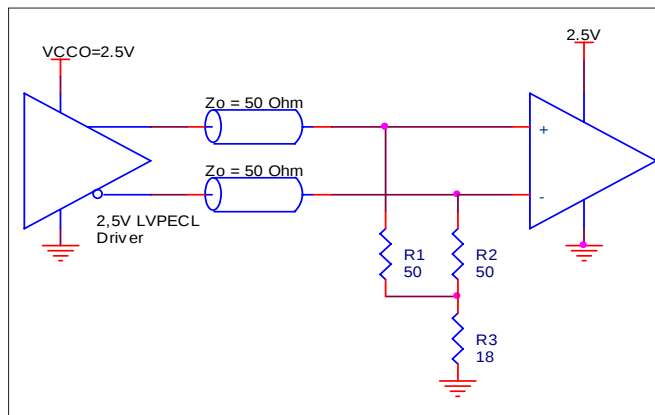


FIGURE 4B. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

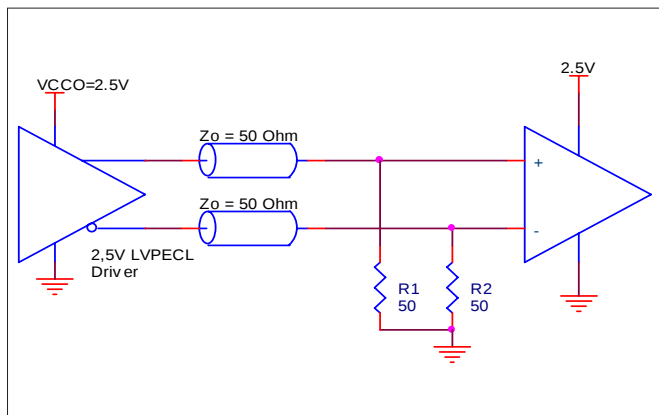


FIGURE 4C. 2.5V LVPECL TERMINATION EXAMPLE

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS87321I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS87321I is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipation into the load.

- Power (core)_{MAX} = $V_{CC_MAX} * I_{EE_MAX} = 3.465V * 18mA = 62.37mW$
- Power (outputs)_{MAX} = **30mW/Loaded Output pair**

$$\text{Total Power}_{MAX} (3.465V, \text{ with outputs switching}) = 62.37mW + 30mW = \mathbf{92.37mW}$$

2. Junction Temperature.

Junction temperature, T_j , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for T_j is as follows: $T_j = \theta_{JA} * Pd_total + T_A$

T_j = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 95.0°C/W per Table 6 below.

Therefore, T_j for an ambient temperature of 85°C with output switching is:

$$85^\circ C + 0.092W * 95^\circ C/W = 93.7^\circ C. \text{ This is well below the limit of } 125^\circ C.$$

This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (multi-layer).

TABLE 6. THERMAL RESISTANCE θ_{JA} FOR 8-PIN SOIC, FORCED CONVECTION

θ_{JA} by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	95.0°C/W	88.4°C/W	83.7°C/W

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in *Figure 5*.

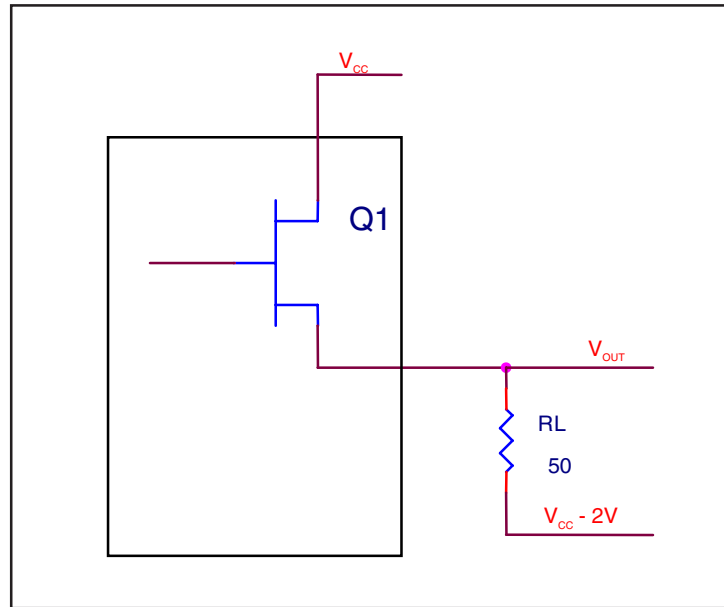


FIGURE 5. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CC} - 2V$.

- For logic high, $V_{OUT} = V_{OH_MAX} = V_{CC_MAX} - 0.9V$
 $(V_{CC_MAX} - V_{OH_MAX}) = 0.9V$
- For logic low, $V_{OUT} = V_{OL_MAX} = V_{CC_MAX} - 1.7V$
 $(V_{CC_MAX} - V_{OL_MAX}) = 1.7V$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OH_MAX}) = [(2V - (V_{CC_MAX} - V_{OH_MAX}))/R_L] * (V_{CC_MAX} - V_{OH_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - (V_{CC_MAX} - V_{OL_MAX}))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

$$\text{Total Power Dissipation per output pair} = Pd_H + Pd_L = 30mW$$

RELIABILITY INFORMATION

TABLE 7. θ_{JA} vs. AIR FLOW TABLE FOR 8 LEAD SOIC

θ_{JA} by Velocity (Linear Feet per Minute)			
	0	200	500
Multi-Layer PCB, JEDEC Standard Test Boards	95.0°C/W	88.4°C/W	83.7°C/W

TRANSISTOR COUNT

The transistor count for ICS87321I is: 309

PACKAGE OUTLINE - M SUFFIX FOR 8 LEAD SOIC

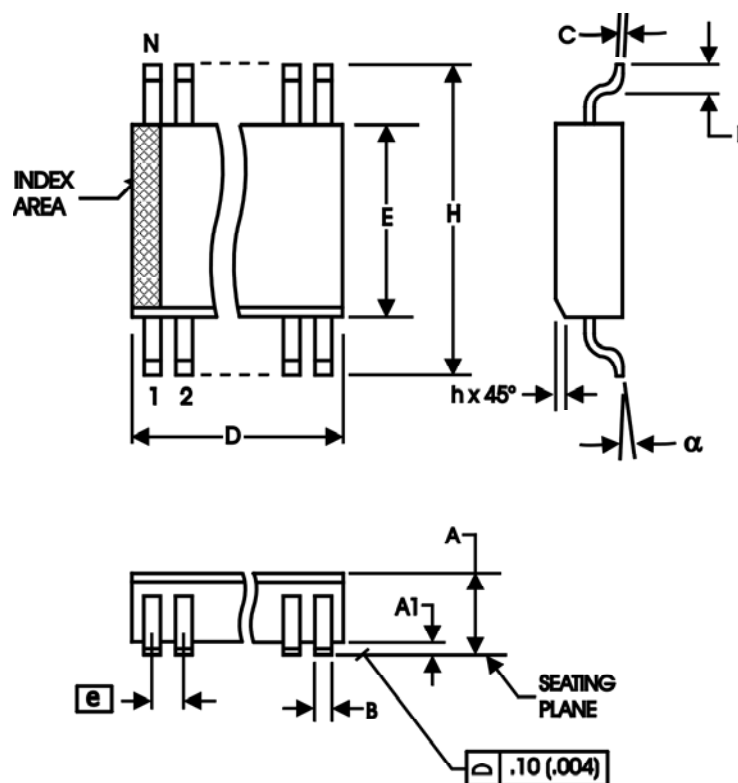


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millimeters	
	MINIMUM	MAXIMUM
N	8	
A	1.35	1.75
A1	0.10	0.25
B	0.33	0.51
C	0.19	0.25
D	4.80	5.00
E	3.80	4.00
e	1.27 BASIC	
H	5.80	6.20
h	0.25	0.50
L	0.40	1.27
α	0°	8°

Reference Document: JEDEC Publication 95, MS-012

TABLE 9. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
87321AMI	87321AMI	8 lead SOIC	tube	-40°C to 85°C
87321AMIT	87321AMI	8 lead SOIC	2500 tape & reel	-40°C to 85°C
87321AMILF	87321AIL	8 lead "Lead-Free" SOIC	tube	-40°C to 85°C
87321AMILFT	87321AIL	8 lead "Lead-Free" SOIC	2500 tape & reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

While the information presented herein has been checked for both accuracy and reliability, Integrated Device Technology, Incorporated (IDT) assumes no responsibility for either its use or for infringement of any patents or other rights of third parties, which would result from its use. No other circuits, patents, or licenses are implied. This product is intended for use in normal commercial and industrial applications. Any other applications such as those requiring high reliability or other extraordinary environmental requirements are not recommended without additional processing by IDT. IDT reserves the right to change any circuitry or specifications without notice. IDT does not authorize or warrant any IDT product for use in life support devices or critical medical instruments.

Innovate with IDT and accelerate your future networks. Contact:

www.IDT.com

For Sales

800-345-7015 (inside USA)
+408-284-8200 (outside USA)
Fax: 408-284-2775
www.IDT.com/go/contactIDT

For Tech Support

netcom@idt.com
+480-763-2056

Corporate Headquarters

Integrated Device Technology, Inc.
6024 Silver Creek Valley Road
San Jose, CA 95138
United States
800-345-7015 (inside USA)
+408-284-8200 (outside USA)