



# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

MAX5109

## General Description

The MAX5109 dual 8-bit digital-to-analog converters (DACs) feature nonvolatile registers. These nonvolatile registers store the DAC operating modes and output states, allowing the DACs to initialize to specified configurations at power-up.

Precision on-chip output buffers swing rail-to-rail, and provide 8 $\mu$ s settling time. The I<sup>2</sup>C\*-compatible, 2-wire serial interface allows for a maximum clock frequency of 400kHz.

The MAX5109 has independent high and low reference inputs allowing maximum output voltage range flexibility. The reference rails accept voltage inputs that range from ground to the positive supply rail.

This device operates from a single +2.7V to +5.25V supply and consumes 200 $\mu$ A per DAC. A software-controlled power-down mode decreases supply current to less than 25 $\mu$ A. A software-controlled mute mode sets each DAC, or both DACs simultaneously, to their respective REFL\_ voltages. The MAX5109 also includes an asynchronous MUTE input, that drives both DAC outputs simultaneously to their respective REFL\_ voltages.

The MAX5109 is available in a 16-pin QSOP and is specified for operation over the extended (-40°C to +85°C) temperature range.

## Applications

Digital Gain and Offset Adjustments  
 Programmable Attenuators  
 Portable Instruments  
 Power-Amp Bias Control  
 ATE Calibration  
 Laser Biasing

**Pin Configuration and Typical Operating Circuit appear at end of data sheet.**

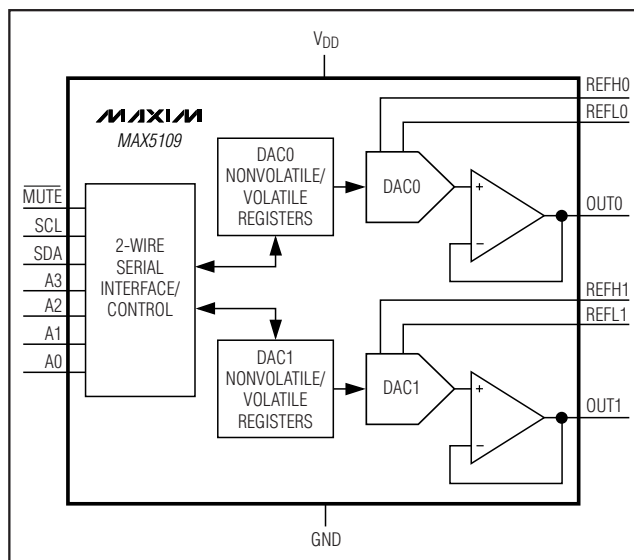
## Features

- ◆ Nonvolatile Registers Initialize DACs to Stored States
- ◆ +2.7V to +5.25V Single-Supply Operation
- ◆ Dual 8-Bit DACs with Independent High and Low Reference Inputs
- ◆ Rail-to-Rail Output Buffers
- ◆ Low 200 $\mu$ A per DAC Supply Current
- ◆ Power-Down Mode Reduces Supply Current to 25 $\mu$ A (max)
- ◆ 400kHz, I<sup>2</sup>C-Compatible, 2-Wire Serial Interface
- ◆ Asynchronous  $\overline{\text{MUTE}}$  Input
- ◆ Small 16-Pin QSOP Package

## Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX5109EEE	-40°C to +85°C	16 QSOP

## Simplified Diagram



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## ABSOLUTE MAXIMUM RATINGS

(All voltages referenced to GND, unless otherwise noted.)

V<sub>DD</sub>, A0, A1, A2, A3, SCL, SDA, MUTE.....-0.3V to +6.0V

OUT0, OUT1, REFH0, REFH1

REFL0, REFL1 .....-0.3V to (V<sub>DD</sub> + 0.3V)

Maximum Current into Any Pin .....±50mA

Power Dissipation (T<sub>A</sub> = +70°C)

16-Pin QSOP (derate 8.3mW/°C above +70°C).....667mW

Operating Temperature Range .....-40°C to +85°C

Junction Temperature .....+150°C

Storage Temperature Range .....-60°C to +150°C

Lead Temperature (soldering, 10s) .....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

(V<sub>DD</sub> = +2.7V to +5.25V, GND = 0, REFH<sub>-</sub> = V<sub>DD</sub>, REFL<sub>-</sub> = GND, R<sub>LOAD</sub> = 5kΩ, C<sub>L</sub> = 100pF, T<sub>A</sub> = -40°C to +85°C, unless otherwise noted. Typical values are at V<sub>DD</sub> = +3.0V and T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>STATIC ACCURACY</b>						
Resolution			8			Bits
Integral Nonlinearity	INL	Code range 0A hex to F0 hex			±1	LSB
		Full code range			±2	
Differential Nonlinearity (Note 2)	DNL	Code range 0A hex to F0 hex			±0.5	LSB
		Full code range			±1	
Offset Error	ZCE	Code = 0A hex			±20	mV
Offset Temperature Coefficient		Code = 0A hex		±20		μV/°C
Gain Error		Code = F0 hex (Note 3)			±1	LSB
Gain-Error Temperature Coefficient		Code = F0 hex		±0.002		LSB/°C
Power-Supply Rejection Ratio	PSRR	Code = FF hex or 0A hex, V <sub>REFH-</sub> = 2.5V, V <sub>REFL-</sub> = 0, f = DC			1	LSB/V
<b>REFERENCE INPUT (REFH<sub>-</sub>, REFL<sub>-</sub>)</b>						
Input Voltage Range	V <sub>REFH-</sub> , V <sub>REFL-</sub>	V <sub>REFH-</sub> ≥ V <sub>REFL-</sub>	0		V <sub>DD</sub>	V
Input Resistance			320	460	600	kΩ
Input-Resistance Temperature Coefficient				±35		ppm/°C
Input Capacitance				10		pF
<b>DAC OUTPUTS (OUT<sub>-</sub>)</b>						
Load Regulation		Code = F0 hex, R <sub>LOAD</sub> ≥ 5kΩ		±0.5	±1	LSB
Output Leakage		DAC powered down, not muted			±10	μA
Amplifier Output Resistance		0.5V ≤ V <sub>OUT-</sub> ≤ (V <sub>DD</sub> - 0.5V)		0.5		Ω
<b>DIGITAL INPUTS (A<sub>-</sub>, MUTE)</b>						
Input High Voltage (Note 4)	V <sub>IH</sub>	2.7V ≤ V <sub>DD</sub> < 3.6V		0.7 × V <sub>DD</sub>		V
		3.6V ≤ V <sub>DD</sub> ≤ 5.25V		2.52		

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## ELECTRICAL CHARACTERISTICS (continued)

( $V_{DD} = +2.7V$  to  $+5.25V$ ,  $GND = 0$ ,  $REFH_{-} = V_{DD}$ ,  $REFL_{-} = GND$ ,  $R_{LOAD} = 5k\Omega$ ,  $C_L = 100pF$ ,  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ , unless otherwise noted. Typical values are at  $V_{DD} = +3.0V$  and  $T_A = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Low Voltage (Note 4)	$V_{IL}$	$2.7V \leq V_{DD} < 3.6V$			$0.3 \times V_{DD}$	V
		$3.6V \leq V_{DD} \leq 5.25V$			1.1	
Input Hysteresis	$V_{HYS}$			$0.05 \times V_{DD}$		V
Input Leakage Current	$I_{IN}$	$V_{IN} = 0$ or $V_{DD}$			$\pm 1$	$\mu A$
Input Capacitance	$C_{IN}$			10		pF
<b>DIGITAL OUTPUT (SDA)</b>						
Output Low Voltage	$V_{OL}$	$I_{SINK} = 3mA$			0.4	V
		$I_{SINK} = 6mA$			0.6	
Tri-State Leakage	$I_L$				$\pm 1$	$\mu A$
Tri-State Output Capacitance	$C_{OUT}$			15		pF
<b>DYNAMIC PERFORMANCE</b>						
SCL to OUT_ Settling	$t_{COS}$	(Note 5)		8		$\mu s$
Crosstalk		$V_{REFH_{-}} = 2.5V_{P-P}$ at 10kHz (Note 6)		55		dB
Multiplying Signal-to-Noise Plus Distortion	SINAD	$V_{REFH_{-}} = 2.5V_{P-P}$ at 1kHz		65		dB
		$V_{REFH_{-}} = 2.5V_{P-P}$ at 10kHz		52		
Multiplying Bandwidth		$V_{REFH_{-}} = 0.5V_{P-P}$ , 3dB bandwidth		325		kHz
Reference Feedthrough		$V_{REFH_{-}} = 2.5V_{P-P}$ at $f = 10kHz$ (Note 7)		88		dB
Clock Feedthrough				2.5		nVs
Output Noise	$e_N$			800		$nV/\sqrt{Hz}$
Power-Up Time	$t_{SDR}$	From power-down state		4		$\mu s$
Power-Down Time	$t_{SDN}$			1.5		$\mu s$
<b>INTERFACE PORTS (SCL, SDA)</b>						
Input Voltage	$V_{IL}$				$0.3 \times V_{DD}$	V
	$V_{IH}$		$0.7 \times V_{DD}$			
Input Hysteresis	$V_{HYS}$			$0.05 \times V_{DD}$		V
Input Current	$I_{IN}$				$\pm 1$	$\mu A$
Input Capacitance	$C_{IN}$			5		pF
<b>POWER SUPPLIES</b>						
Power-Supply Voltage	$V_{DD}$		2.70		5.25	V
Supply Current	$I_{DD}$	$I_{LOAD} = 0$ , digital inputs at GND or $V_{DD}$	Normal operation	0.4	0.7	mA
			During nonvolatile write		2	
Power-Down Current					25	$\mu A$

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## ELECTRICAL CHARACTERISTICS (continued)

(V<sub>DD</sub> = +2.7V to +5.25V, GND = 0, REFH<sub>-</sub> = V<sub>DD</sub>, REFL<sub>-</sub> = GND, R<sub>LOAD</sub> = 5kΩ, C<sub>L</sub> = 100pF, T<sub>A</sub> = -40°C to +85°C, unless otherwise noted. Typical values are at V<sub>DD</sub> = +3.0V and T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>DIGITAL TIMING (Figure 3, Note 8)</b>						
SCL Clock Frequency	f <sub>SCL</sub>				400	kHz
Setup Time for START Condition	t <sub>SU:STA</sub>		0.6			μs
Hold Time for START Condition	t <sub>HD:STA</sub>		0.6			μs
SCL High Time	t <sub>HIGH</sub>		0.6			μs
SCL Low Time	t <sub>LOW</sub>		1.3			μs
Data Setup Time	t <sub>SU:DAT</sub>		100			ns
Data Hold Time	t <sub>HD:DAT</sub>		0		0.9	μs
SDA, SCL Rise Time	t <sub>R</sub>				300	ns
SDA, SCL Fall Time	t <sub>F</sub>				300	ns
Setup Time for STOP Condition	t <sub>SU:STO</sub>		0.6			μs
Bus Free Time Between a STOP and START Condition	t <sub>BUF</sub>		1.3			μs
Pulse Width of Spike Suppressed	t <sub>SP</sub>				50	ns
Maximum Capacitive Load for Each Bus Line	C <sub>B</sub>	(Note 9)		400		pF
Write NV Register Busy Time		(Note 10)			15	ms
<b>NONVOLATILE MEMORY RELIABILITY</b>						
Data Retention		T <sub>A</sub> = +85°C		50		Years
Endurance		T <sub>A</sub> = +25°C		200,000		Stores
		T <sub>A</sub> = +85°C		50,000		

**Note 1:** All devices are 100% production tested at T<sub>A</sub> = +25°C. All temperature limits are guaranteed by design.

**Note 2:** Guaranteed monotonic.

**Note 3:** Gain error is defined as:

$$\frac{256 \times (V_{F0,Meas} - ZCE - V_{F0,Ideal})}{V_{REFH_-}}$$

where V<sub>F0,Meas</sub> is the DAC voltage with input code F0 hex and V<sub>F0,Ideal</sub> is the ideal DAC voltage with input code F0 hex or (V<sub>REFH</sub> - V<sub>REFL</sub>) × (240 / 256) + V<sub>REFL</sub>.

**Note 4:** The device draws higher supply current when the digital inputs are driven with voltages between (V<sub>DD</sub> - 0.5V) and (GND + 0.5V). See Supply Current vs. Digital Input Voltage in the *Typical Operating Characteristics*.

**Note 5:** Output settling time is measured from the 50% point of the rising edge of the last SCL of the data byte to 0.5 LSB of OUT<sub>-</sub>'s final value for a code transition from 10 hex to F0 hex.

**Note 6:** Crosstalk is defined as the coupling from a DAC switching from code 00 hex to code FF hex to any other DAC that is in a steady state at code 00 hex.

**Note 7:** Reference feedthrough is defined as the coupling from one driven reference with input code = FF hex to any other DAC output with the reference of the DAC at a constant value and input code = 00 hex.

**Note 8:** SCL clock period includes rise and fall times t<sub>R</sub> and t<sub>F</sub>. All digital input signals are specified with t<sub>R</sub> = t<sub>F</sub> = 2ns and timed from a voltage level of (V<sub>IL</sub> + V<sub>IH</sub>) / 2.

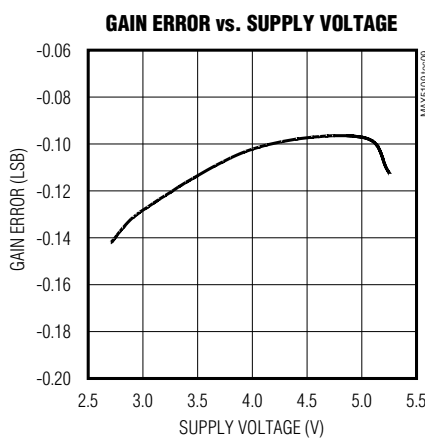
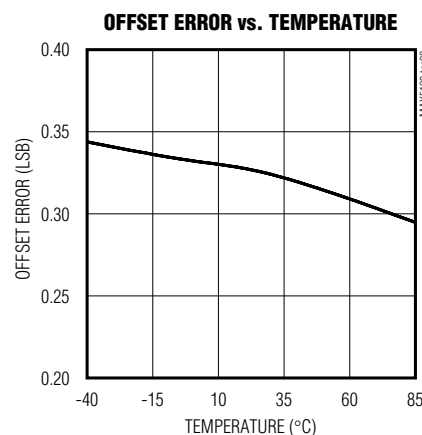
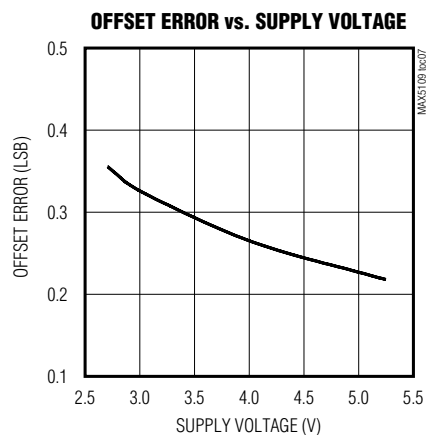
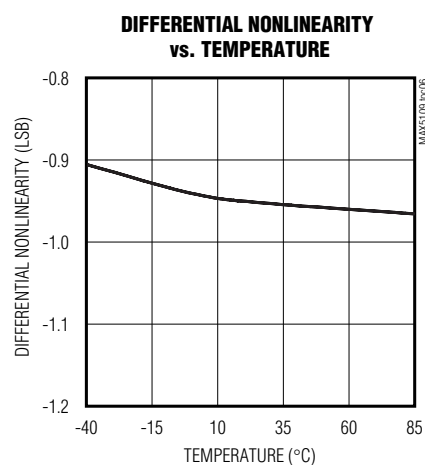
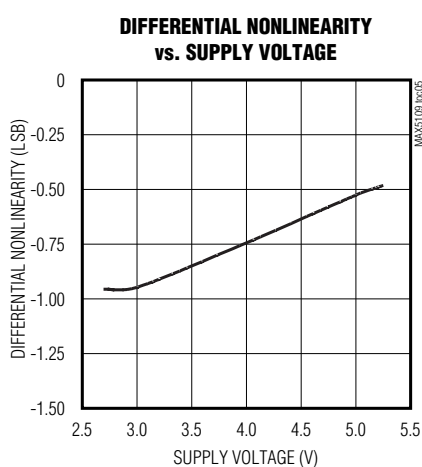
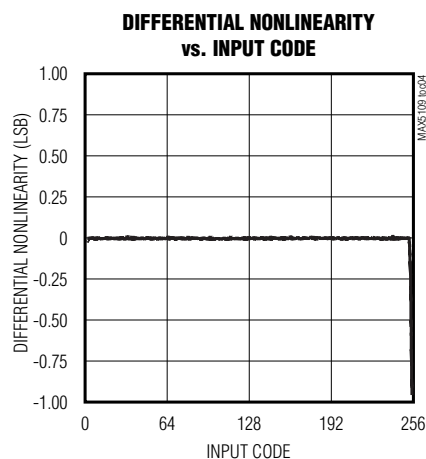
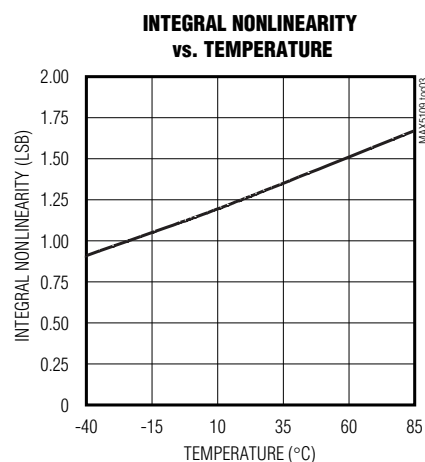
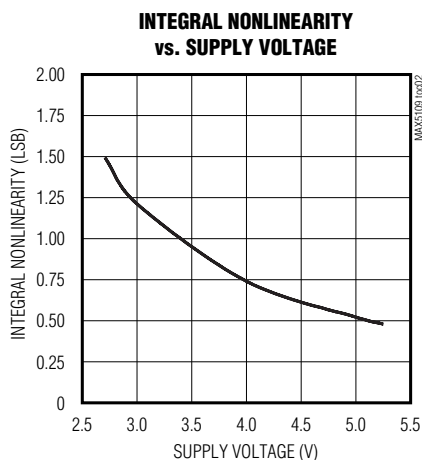
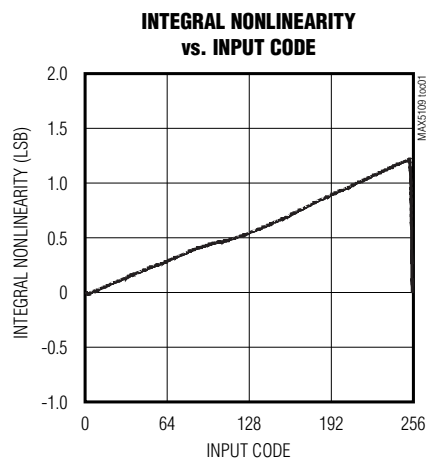
**Note 9:** An appropriate bus pullup resistance must be selected depending on board capacitance. Refer to the document linked to this web address: [www.semiconductors.philips.com/acrobat/literature/9398/39340011.pdf](http://www.semiconductors.philips.com/acrobat/literature/9398/39340011.pdf).

**Note 10:** The busy time begins from the initiation of the stop pulse.

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Typical Operating Characteristics

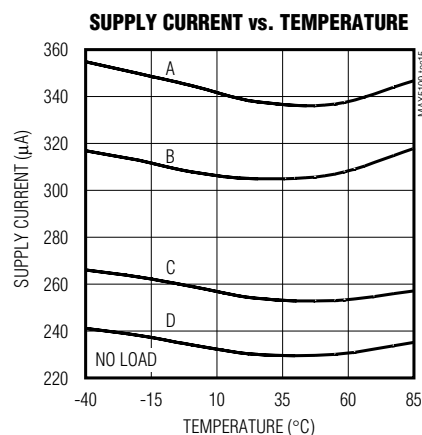
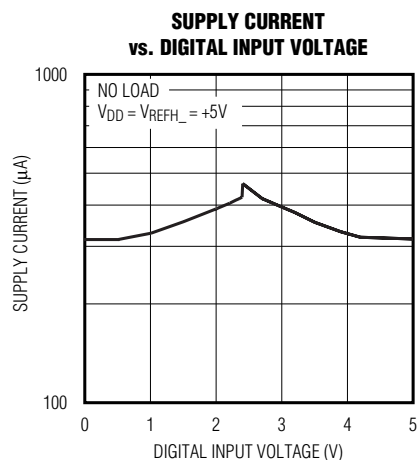
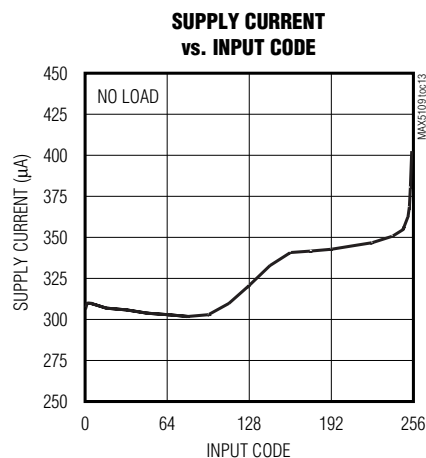
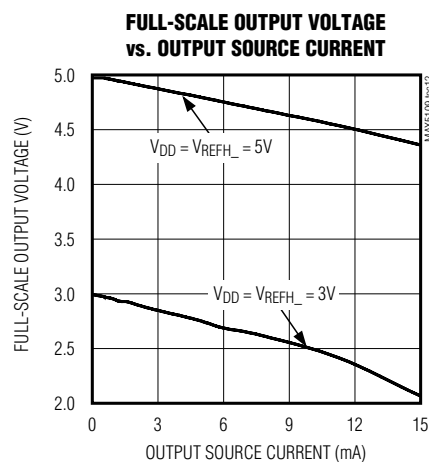
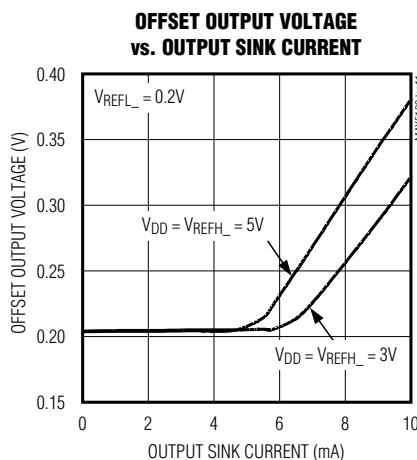
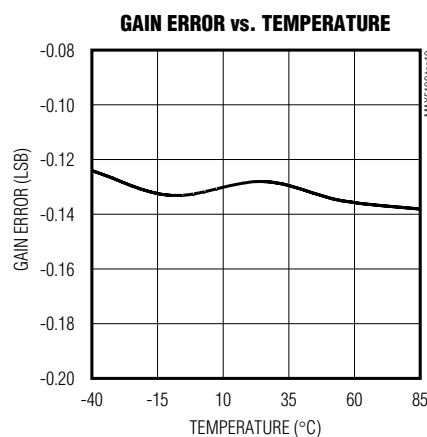
( $V_{DD} = +3V$ ,  $V_{REFH\_} = +3V$ ,  $V_{REFL\_} = GND$ ,  $R_L = 5k\Omega$ ,  $C_L = 100pF$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Typical Operating Characteristics (continued)

( $V_{DD} = +3V$ ,  $V_{REFH\_} = +3V$ ,  $V_{REFL\_} = GND$ ,  $R_L = 5k\Omega$ ,  $C_L = 100pF$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)

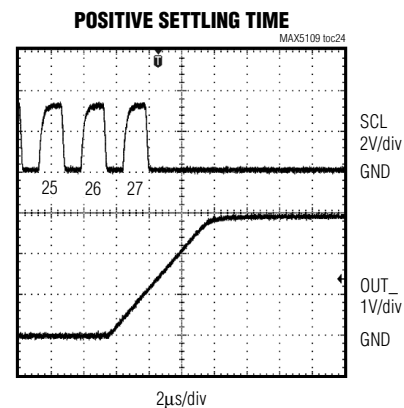
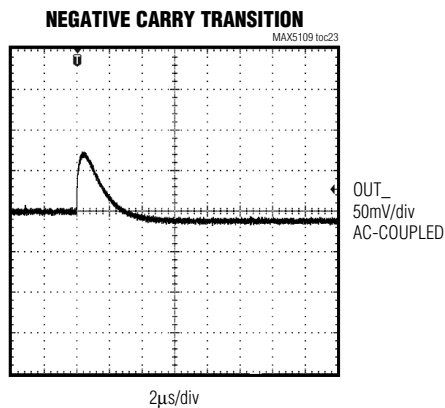
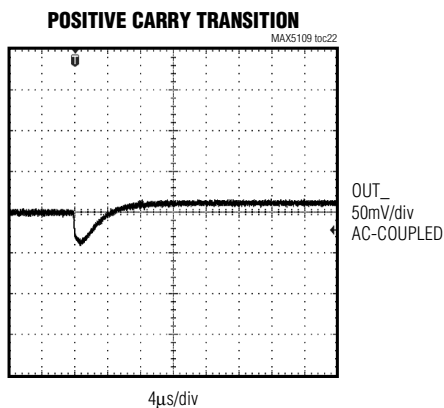
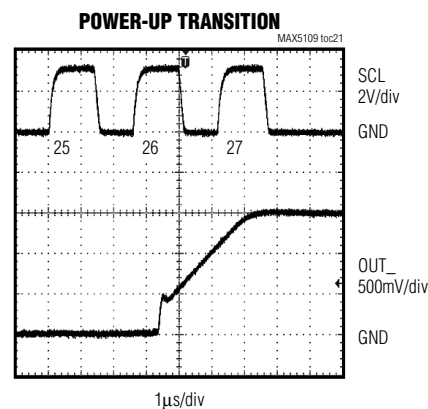
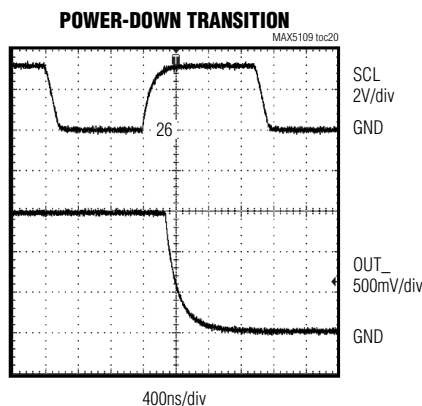
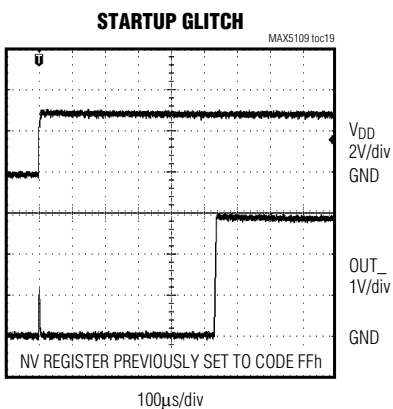
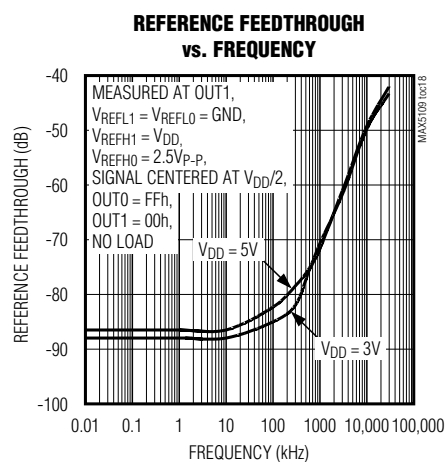
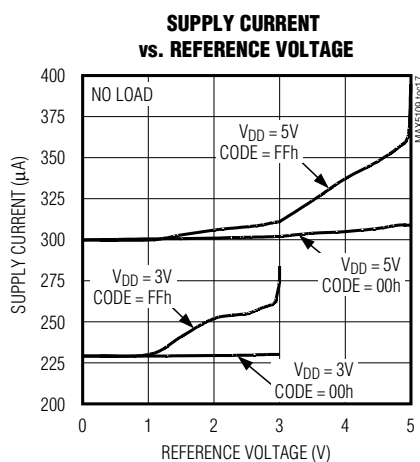
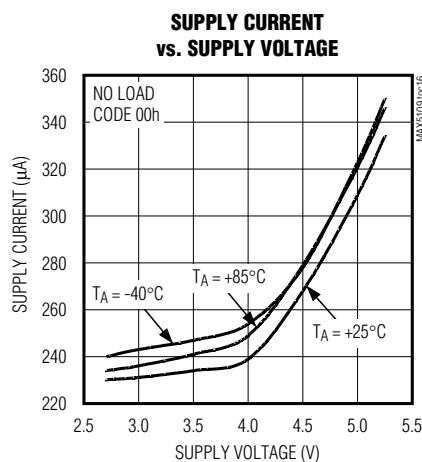


A:  $V_{DD} = 5V$ ,  $V_{REFH\_} = 4.096V$ , CODE = FFh  
 B:  $V_{DD} = 5V$ ,  $V_{REFH\_} = 4.096V$ , CODE = 00h  
 C:  $V_{DD} = 3V$ ,  $V_{REFH\_} = 2.5V$ , CODE = FFh  
 D:  $V_{DD} = 3V$ ,  $V_{REFH\_} = 2.5V$ , CODE = 00h

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Typical Operating Characteristics (continued)

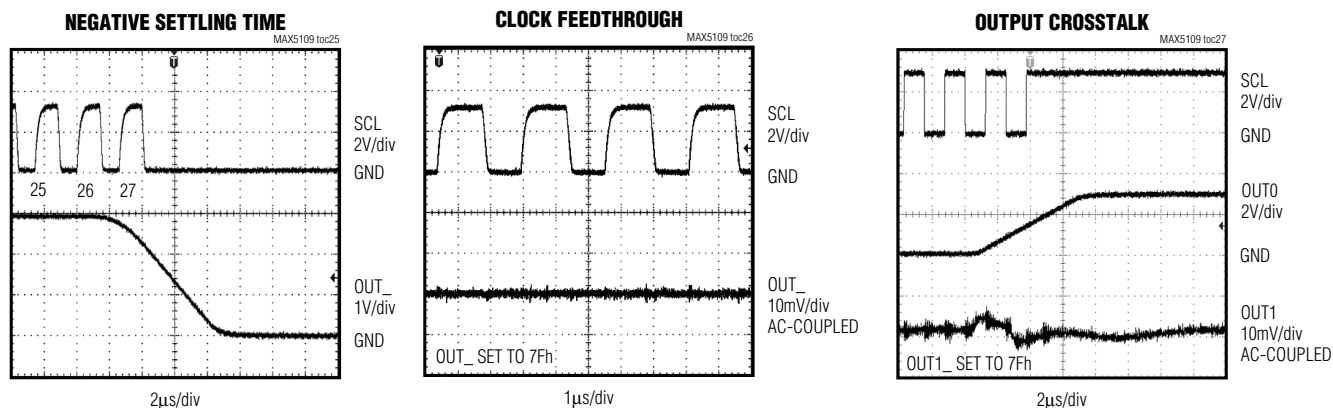
( $V_{DD} = +3V$ ,  $V_{REFH} = +3V$ ,  $V_{REFL} = GND$ ,  $R_L = 5k\Omega$ ,  $C_L = 100pF$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Typical Operating Characteristics (continued)

( $V_{DD} = +3V$ ,  $V_{REFH\_} = +3V$ ,  $V_{REFL\_} = GND$ ,  $R_L = 5k\Omega$ ,  $C_L = 100pF$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



## Pin Description

PIN	NAME	FUNCTION
1	A3	Address Select 3. Connect to $V_{DD}$ or GND to set the device address.
2	A2	Address Select 2. Connect to $V_{DD}$ or GND to set the device address.
3	A1	Address Select 1. Connect to $V_{DD}$ or GND to set the device address.
4	A0	Address Select 0. Connect to $V_{DD}$ or GND to set the device address.
5	REFH1	DAC1 High Reference Input. REFH1 must be equal to or greater than REFL1.
6	REFL1	DAC1 Low Reference Input. REFL1 must be equal to or less than REFH1.
7	OUT1	DAC1 Output. OUT1 is buffered with a unity-gain amplifier.
8	GND	Ground
9	$\overline{MUTE}$	Active-Low Mute Input. Connect $\overline{MUTE}$ low to drive all DAC outputs to their respective reference low voltages. Connect $\overline{MUTE}$ to $V_{DD}$ for normal operation.
10	N.C.	No Connection. Not internally connected.
11	OUT0	DAC0 Output. OUT0 is buffered with a unity-gain amplifier.
12	REFL0	DAC0 Low Reference Input. REFL0 must be equal to or less than REFH0.
13	REFH0	DAC0 High Reference Input. REFH0 must be equal to or greater than REFL0.
14	SCL	Serial Clock Input. Connect SCL to $V_{DD}$ through a $2.4k\Omega$ pullup resistor.
15	$V_{DD}$	Positive Power Input. Connect $V_{DD}$ to a +2.7 to +5.25V power supply. Bypass $V_{DD}$ to GND with a $0.1\mu F$ capacitor as close to the device as possible.
16	SDA	Serial Data Input/Output. Connect SDA to $V_{DD}$ through a $2.4k\Omega$ pullup resistor.



# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

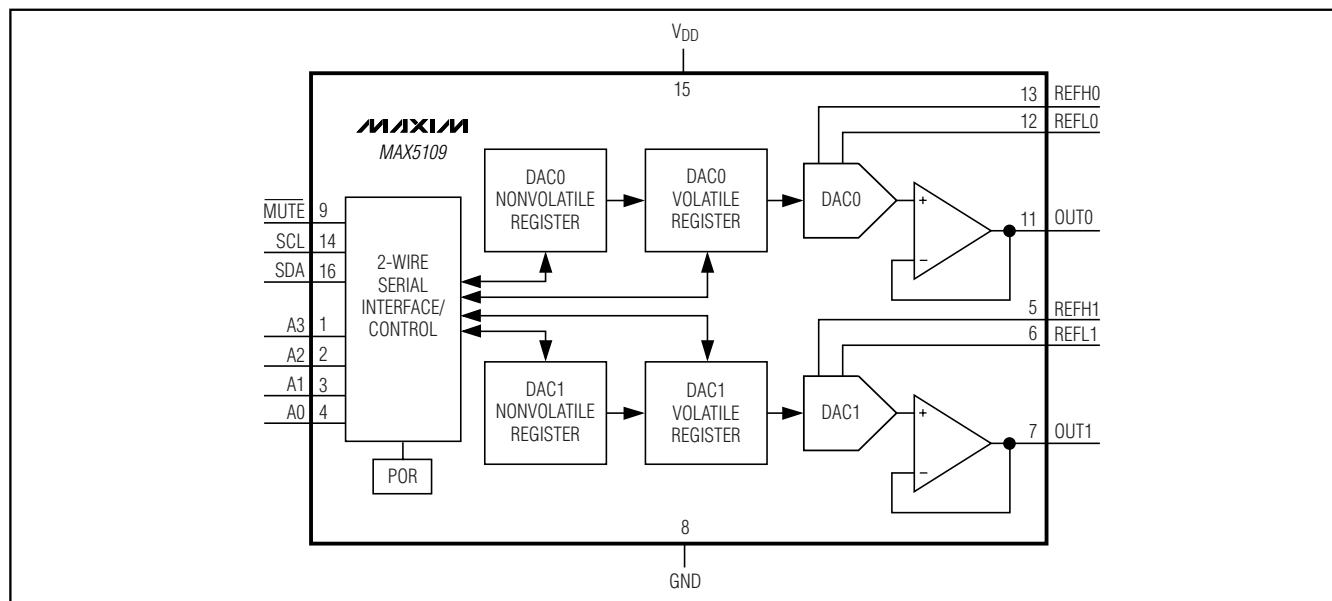


Figure 1. MAX5109 Functional Diagram

## Detailed Description

The MAX5109 8-bit DACs feature internal, nonvolatile registers that store the DAC states for initialization during power-up. This device consists of resistor-string DACs, rail-to-rail output buffers, a shift register, power-on reset (POR) circuitry, and volatile and nonvolatile memory registers (Figure 1). The shift register decodes the control and address bits, routing the data to the proper registers. Writing data to a selected volatile register immediately updates the DAC outputs.

The volatile registers retain data as long as the device is powered. Removing power clears the volatile registers. The nonvolatile registers retain data even after power is removed. On startup, when power is first applied, data from the nonvolatile registers is transferred to the volatile registers to automatically initialize the device. Read data from the nonvolatile or volatile registers using the 2-wire serial interface.

## DAC Operation

The MAX5109 uses a DAC matrix decoding architecture that saves power. A resistor string divides the difference between the external reference voltages,  $V_{REFH\_}$  and  $V_{REFL\_}$ . Row and column decoders select the appropriate tap from the resistor string, providing the equivalent analog voltage. The resistor string presents a code-independent input impedance to the reference and guarantees a monotonic output. Figure 2 shows a simplified diagram of one DAC.

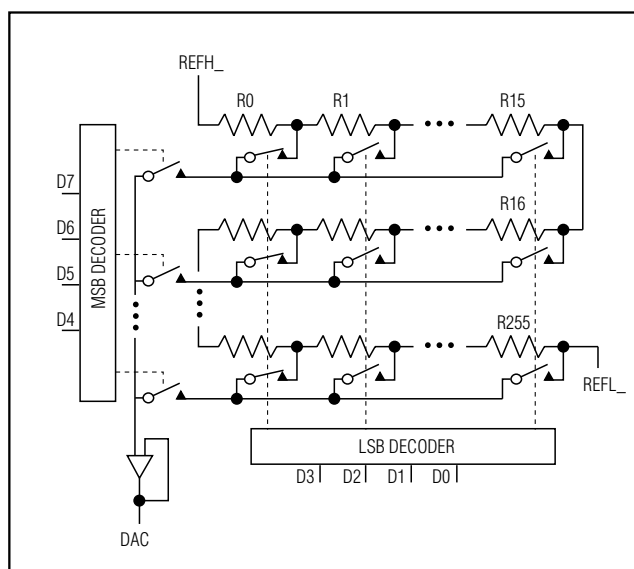


Figure 2. DAC Simplified Circuit Diagram

## Output Buffer Amplifiers

The MAX5109 analog outputs are internally buffered by precision unity-gain amplifiers. The outputs swing from GND to  $V_{DD}$  with a  $V_{REFL\_}$ -to- $V_{REFH\_}$  output transition. The amplifier outputs typically settle to  $\pm 0.5$  LSB in  $8\mu s$  when loaded with  $5k\Omega$  in parallel with  $100pF$ .

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

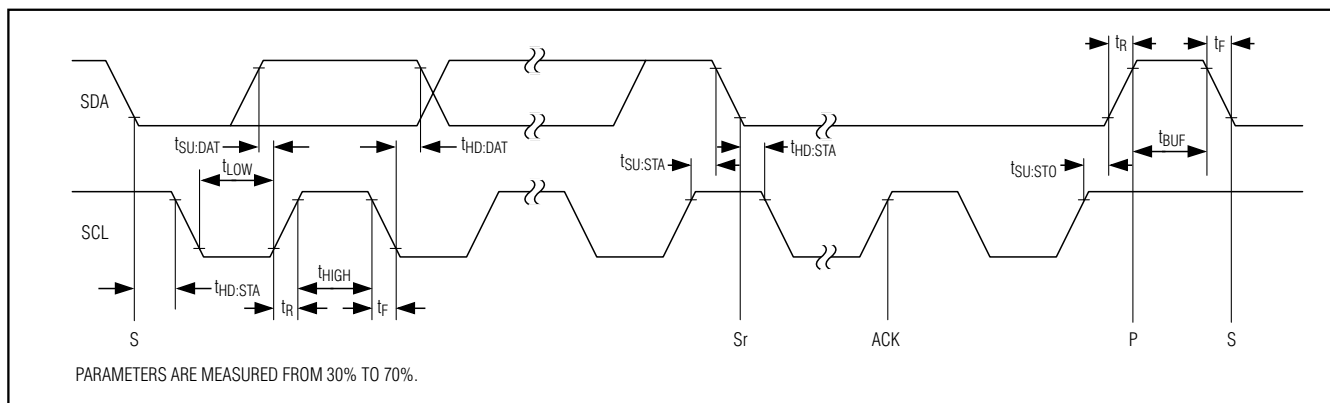


Figure 3. 2-Wire Serial-Interface Timing Diagram

## DAC Registers

The MAX5109 features two registers per DAC, a volatile and a nonvolatile register, that store the DAC data. The volatile DAC register holds the current value of each DAC. Write data to the volatile registers directly from the 2-wire serial interface or by loading the previously stored data from the respective nonvolatile register. Clear the volatile registers by removing power to the device. The volatile registers are read/write.

The nonvolatile register retains the DAC values even after power is removed. Read stored data using the 2-wire serial interface. On power-up, the device is automatically initialized with data stored in the nonvolatile registers. The nonvolatile registers are read/write and programmed to all zeros at the factory.

## Serial Interface

The MAX5109 features an I<sup>2</sup>C-compatible, 2-wire serial interface consisting of a bidirectional serial data line (SDA) and a serial clock line (SCL). SDA and SCL facilitate bidirectional communication between the MAX5109 and the master at rates up to 400kHz (Figure 3). The master (typically a microcontroller) initiates data transfer on the bus and generates SCL. SDA and SCL require pullup resistors (2.4k $\Omega$  or greater; see the *Typical Operating Circuit*). Optional resistors (24 $\Omega$ ) in series with SDA and SCL protect the device inputs from high-voltage spikes on the bus lines. Series resistors also minimize crosstalk and undershoot of the bus signals.

## I<sup>2</sup>C Compatibility

The MAX5109 is compatible with existing I<sup>2</sup>C systems. SCL and SDA are high-impedance inputs; SDA has an open-drain output. The *Typical Operating Circuit* shows an I<sup>2</sup>C application. The communication protocol supports standard I<sup>2</sup>C 8-bit communications. The general call address is ignored, and CBUS formats are not sup-

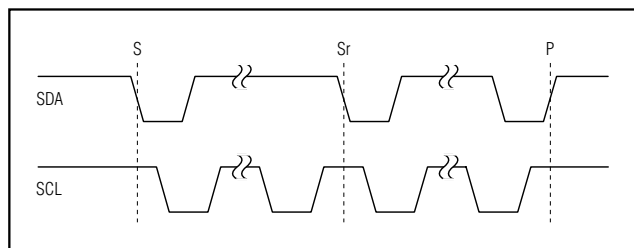


Figure 4. START and STOP Conditions

ported. The device's address is compatible with 7-bit I<sup>2</sup>C addressing protocol only. No 10-bit address formats are supported.

## Bit Transfer

One data bit transfers during each SCL rising edge. Nine clock cycles are required to transfer the data into or out of the MAX5109. The data on SDA must remain stable during the high period of the SCL clock pulse. Changes in SDA while SCL is high are read as control signals (see the *START and STOP Conditions* section). Both SDA and SCL idle high.

## START and STOP Conditions

The master initiates a transmission with a START condition (S), a high-to-low transition on SDA with SCL high. The master terminates a transmission with a STOP condition (P), a low-to-high transition on SDA while SCL is high (Figure 4). A START condition from the master signals the beginning of a transmission to the MAX5109. The master terminates transmission by issuing a STOP condition. The STOP condition frees the bus. If a REPEATED START condition (Sr) is generated instead of a STOP condition, the bus remains active.

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Early STOP Conditions

The MAX5109 recognizes a STOP condition at any point during transmission except if a STOP condition occurs in the same high pulse as a START condition (Figure 5). This condition is not a legal I<sup>2</sup>C format.

## REPEATED START Conditions

A REPEATED START (Sr) condition is used when the bus master is writing to several I<sup>2</sup>C devices and does not want to relinquish control of the bus. The MAX5109 serial interface supports continuous write operations with an Sr condition separating them. Continuous read operations require Sr conditions because of the change in direction of data flow.

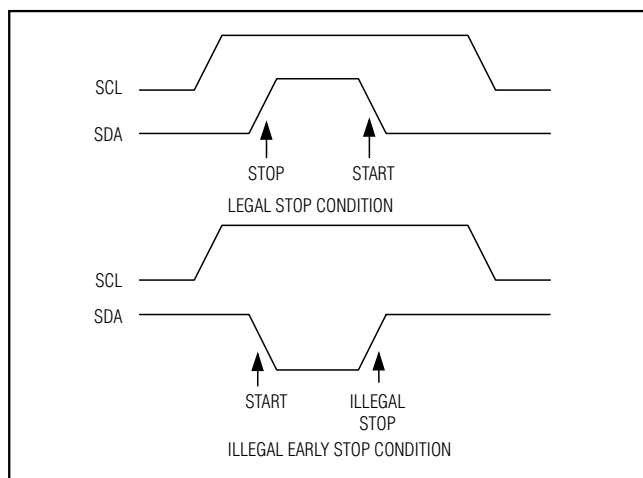


Figure 5. Early STOP Conditions

## Acknowledge Bit (ACK) and Not-Acknowledge Bit (NACK)

Successful data transfers are acknowledged with an acknowledge bit (ACK) or a not-acknowledge bit (NACK). Both the master and the MAX5109 (slave) generate acknowledge bits. To generate an acknowledge, the receiving device must pull SDA low before the rising edge of the acknowledge-related clock pulse (ninth pulse) and keep it low during the high period of the clock pulse (Figure 6). To generate a not acknowledge, the receiver allows SDA to be pulled high before the rising edge of the acknowledge-related clock pulse and leaves it high during the high period of the clock pulse.

Monitoring the acknowledge bits allows for detection of unsuccessful data transfers. An unsuccessful data transfer happens if a receiving device is busy or if a system fault has occurred. In the event of an unsuccessful data transfer, the master should reattempt communication at a later time.

## Slave Address

A master initiates communication with a slave device by issuing a START condition followed by a slave address (Figure 7). The slave address consists of 7 address bits and a read/write bit (R/W). When idle, the device continuously waits for a START condition followed by its slave address. When the device recognizes its slave address, it acquires the data byte and executes the command. The first 3 bits (MSBs) of the slave address have been factory programmed and are always 010. Connect A3–A0 to V<sub>DD</sub> or GND to program

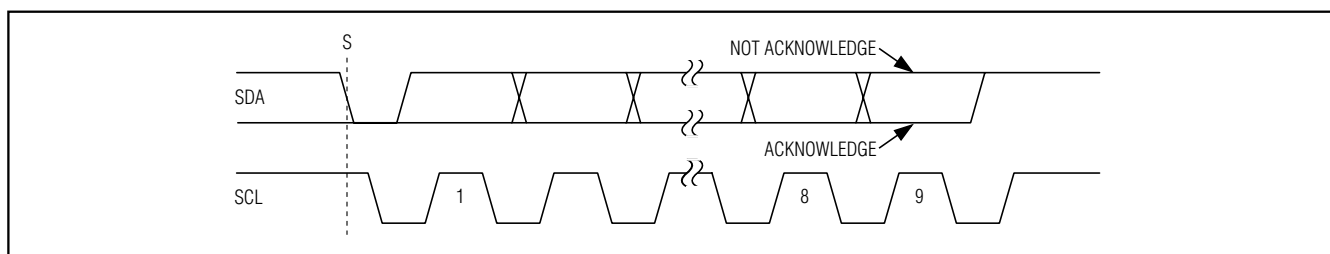


Figure 6. Acknowledge and Not-Acknowledge Bits

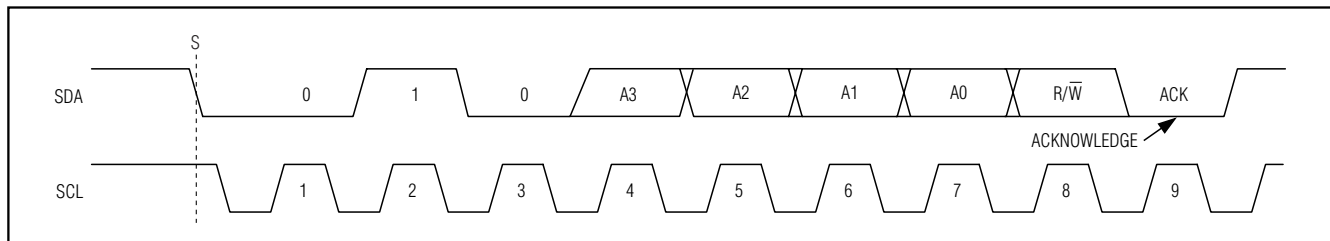


Figure 7. Slave Address Byte

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

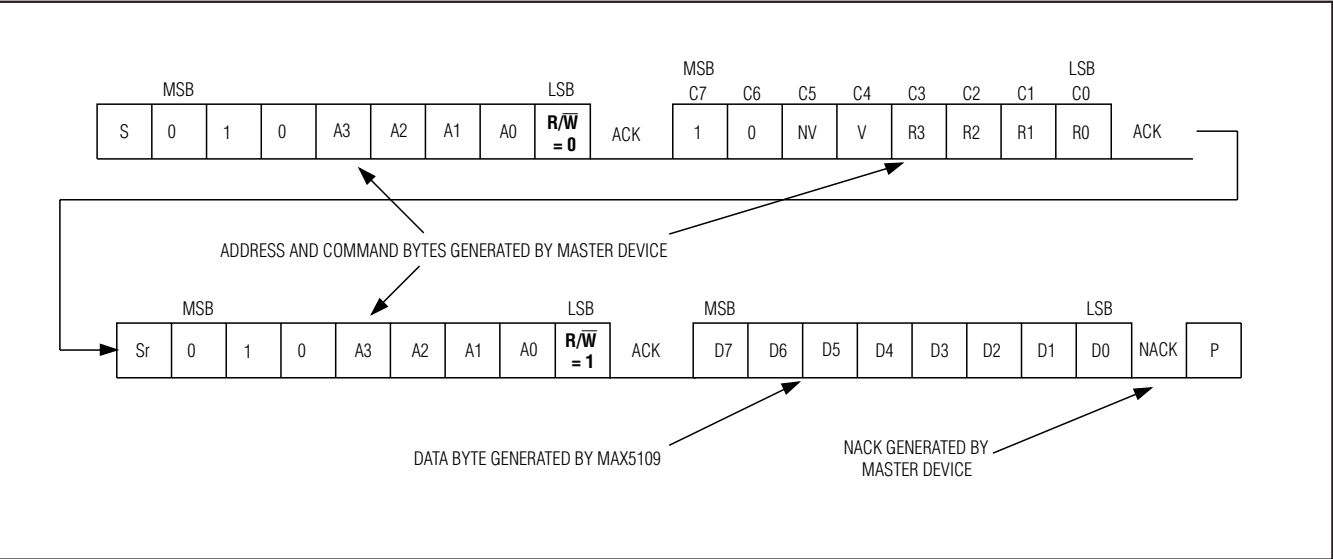


Figure 8. Example Read Word Data Sequence

the remaining 4 bits of the slave address. The least significant bit (LSB) of the address byte ( $R/\overline{W}$ ) determines whether the master is writing to or reading from the MAX5109. ( $R/\overline{W} = 0$  selects a write condition.  $R/\overline{W} = 1$  selects a read condition.) After receiving the address, the MAX5109 (slave) issues an acknowledge by pulling SDA low for one clock cycle.

### Write Cycle

The write command requires 27 clock cycles. In write mode ( $R/\overline{W} = 0$ ), the command byte that follows the address byte controls the MAX5109 (Table 1). For a write function, set bits C7 and C6 to zero. Set bits C5 and C4 to select the volatile or nonvolatile register (Table 2). Set bits C3–C0 to select the respective DAC register (Table 3). The registers update on the rising edge of the 26th SCL pulse. Prematurely aborting the

write cycle does not update the DAC. See Table 4 for a summary of the write commands.

### Read Cycle

A read command requires 36 clock cycles. In read mode, the MAX5109 sends the contents of the volatile and nonvolatile registers to the bus. Reading a register requires a REPEATED START ( $S_r$ ) condition. To read a register first, write a read command ( $R/\overline{W} = 0$ , Figure 8). Set the most significant 2 bits of the command byte to 10 ( $C_7 = 1$  and  $C_6 = 0$ ). Set bits C5 and C4 to read from either the volatile or nonvolatile register (Table 5). Set bits C3–C0 to select the desired DAC register (Table 6). After the command byte, send a  $S_r$  condition followed by the address of the device ( $R/\overline{W} = 1$ ). The MAX5109 then acknowledges and sends the data to the bus.

Table 1. Write Operation

	START	ADDRESS BYTE									COMMAND BYTE									DATA BYTE								STOP
									$R/\overline{W}$		C7	C6	C5	C4	C3	C2	C1	C0		D7	D6	D5	D4	D3	D2	D1	D0	
Master SDA	S	0	1	0	A3	A2	A1	A0	0		C7	C6	NV	V	R3	R2	R1	R0		D7–D0								P
Slave SDA									ACK										ACK									ACK

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

**Table 2. Volatile and Nonvolatile Write Selection**

NONVOLATILE (NV)	VOLATILE (V)	FUNCTION
0	0	Transfer data from NVREG_ to VREG_
0	1	Write to VREG_
1	0	Write to NVREG_
1	1	Write to NVREG and VREG_

**Table 3. DAC Write Selection**

R3	R2	R1	R0	FUNCTION
0	0	0	0	DAC0
0	0	0	1	DAC1
1	1	1	1	All DACs*

\*This option is only valid for a write to all volatile registers.

**Table 4. Write-Command Summary**

COMMAND	S T A R T	ADDRESS BYTE		A C K	COMMAND BYTE								A C K	DATA BYTE								A C K	STOP
			R/W		C 7	C 6	C 5	C 4	C 3	C 2	C 1	C 0		MSB	D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0	
Write VREG_	S		0		0	0	0	1	R 3	R 2	R 1	R 0		D7-D0									P
Write All VREG_*	S		0		0	0	0	1	1	1	1	1		D7-D0									P
Write NVREG_	S		0		0	0	1	0	R 3	R 2	R 1	R 0		D7-D0									P
Write VREG_ and NVREG_	S		0		0	0	1	1	R 3	R 2	R 1	R 0		D7-D0									P
Transfer NVREG_ to VREG_	S		0		0	0	0	0	R 3	R 2	R 1	R 0		—									P

\*This option is only valid for a write to all volatile registers.

## Mute/Power-Down Mode

The MAX5109 features software-controlled mute and power-down modes for each DAC. The power-down mode places the DAC output in a high-impedance state and reduces quiescent-current consumption (25µA (max) with all DACs powered-down).

Mute drives the selected DAC output to the corresponding REFL\_ voltage. The volatile DAC registers retain data and the output returns to its previous state when mute is

disabled. The MAX5109 also features an asynchronous MUTE input that mutes all DACs simultaneously.

The volatile and nonvolatile registers remain active while the MAX5109 is in mute and/or power-down modes. Writing to or reading from the volatile or nonvolatile registers does not remove the MAX5109 from mute or power-down mode. Writing or transferring data to the volatile registers while the device is muted or powered down updates the DAC outputs to the new state upon exiting mute or power-down mode.

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

**Table 5. Volatile and Nonvolatile Read Selection**

NONVOLATILE (NV)	VOLATILE (V)	FUNCTION
0	1	Read from VREG_
1	0	Read from NVREG_

**Table 6. DAC Read Selection**

R3	R2	R1	R0	FUNCTION
0	0	0	0	DAC0
0	0	0	1	DAC1

**Table 7. Mute/Power-Down Operation**

COMMAND	S T A R T	ADDRESS BYTE		A C K	COMMAND BYTE								A C K	DATA BYTE								A C K	STOP
														MSB				LSB					
			R/ W		C 7	C 6	C 5	C 4	C 3	C 2	C 1	C 0		D 7	D 6	D 5	D 4	D 3	D 2	D 1	D 0		
Write VCTL	S		0		0	0	0	1	0	1	0	0		Control register*									P
Write NVCTL	S		0		0	0	1	0	0	1	0	0		Control register*									P
Write VCTL and NVCTL	S		0		0	0	1	1	0	1	0	0		Control register*									P
Transfer NVCTL to VCTL	S		0		0	0	0	0	0	1	0	0		Control register*									P

\*See Mute/Power-Down Control Register (Table 8).

**Table 8. Mute/Power-Down Control Register**

	BIT IN REGISTER							
	D7 (MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)
CONTROLLING FUNCTION	X	X	Mute DAC1	Mute DAC0	X	X	Power-down DAC1	Power-down DAC0

X = Don't care.

## Mute/Power-Down Register and Operation

Separate nonvolatile and volatile control registers store and update the state of the mute/power-down mode for each DAC. Tables 7 and 8 show how to access and control each register. Register access is gained by setting control bits C3–C0 to 0100. Bits C5 and C4 indicate whether the nonvolatile or volatile control register is accessed. The volatile register maintains data while the device remains powered. The nonvolatile register maintains data even after power is removed. The MAX5109 starts up (power first applied) by transferring the mute/power-down modes from the nonvolatile con-

trol register to the volatile control register. The non-volatile control register is set to 00 hex at the factory.

## Power-On Reset

Power-on reset (POR) circuitry controls the initialization of the MAX5109. A POR loads the volatile registers with the data stored in the nonvolatile registers.

This initialization period takes 500μs (typ). During this time, the DAC outputs are held in mute mode. At the completion of the initialization period, the DAC outputs update in accordance with the data stored in the non-volatile registers.

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## DAC Data

The 8-bit DAC data is decoded as offset binary, MSB first, with 1 LSB =  $(V_{REFH\_} - V_{REFL\_}) / 256$ , and converted into the corresponding analog voltage as shown in Table 9.

## Applications Information

### DAC Linearity and Offset Voltage

The output buffer can have a negative input offset voltage that would normally drive the output negative, but with no negative supply, the output remains at GND (Figure 9). Determine linearity using the end-point method, measuring between code 10 (0A hex) and code 240 (F0 hex) after calibrating the offset and gain error (Figure 9).

### External Voltage Reference

The MAX5109 features two reference inputs for each DAC (REFH\_ and REFL\_). REFH\_ sets the full-scale voltage, while REFL\_ sets the zero code output. A 460k $\Omega$  typical input impedance is independent of the code.

### Power Sequencing

The voltage applied to REFH\_ and REFL\_ should not exceed  $V_{DD}$  at any time. If proper power sequencing is not possible, connect an external Schottky diode between REFH\_, REFL\_, and  $V_{DD}$  to ensure compliance with the absolute maximum ratings. Do not apply signals to the digital inputs before the device is fully powered.

### Power-Supply Bypassing and Ground Management

Digital or AC transient signals on GND can create noise at the analog output. Return GND to the highest-quality ground available. Bypass  $V_{DD}$ , REFH\_, and REFL\_ to GND with a 0.1 $\mu$ F capacitor, located as close to the device as possible. Careful PC board ground layout minimizes crosstalk between the DAC outputs and digital inputs.

Table 9. Unipolar Code Output Voltage

DAC CODE	OUTPUT VOLTAGE (V)
1111 1111	$\frac{255 \times (V_{REFH\_} - V_{REFL\_})}{256} + V_{REFL\_}$
1000 0000	$\frac{128 \times (V_{REFH\_} - V_{REFL\_})}{256} + V_{REFL\_}$
0000 0001	$\frac{(V_{REFH\_} - V_{REFL\_})}{256} + V_{REFL\_}$
0000 0000	$V_{REFL\_}$

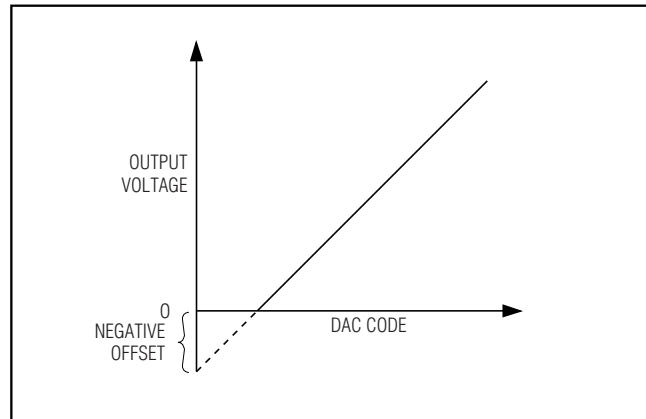
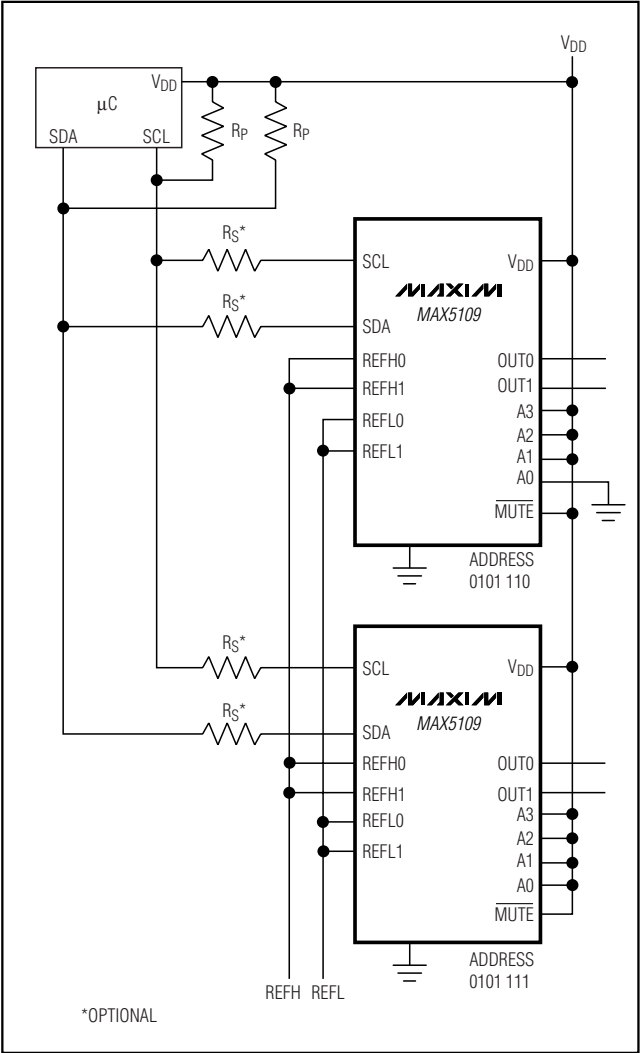


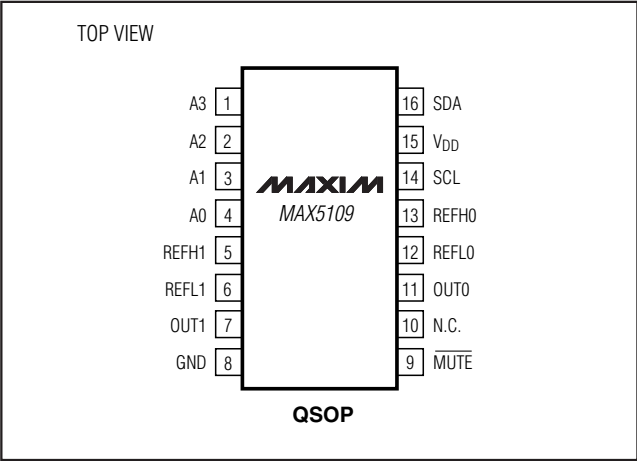
Figure 9. Effect of Negative Offset (Single Supply)

# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Typical Operating Circuit



## Pin Configuration



## Chip Information

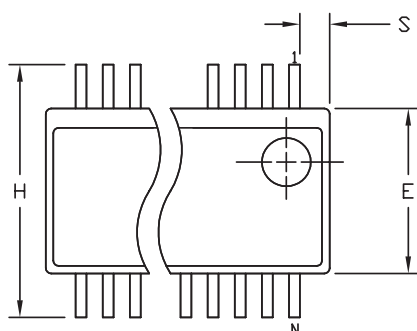
TRANSISTOR COUNT: 40,209  
PROCESS: BiCMOS



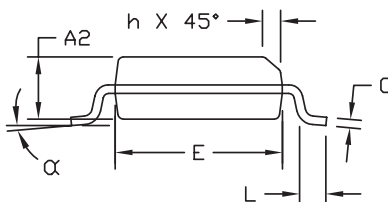
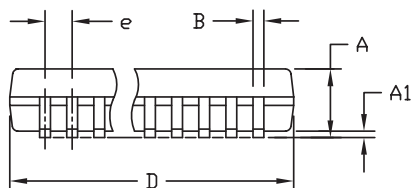
# Nonvolatile, Dual, 8-Bit DACs with 2-Wire Serial Interface

## Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to [www.maxim-ic.com/packages](http://www.maxim-ic.com/packages).)



DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.061	.068	1.55	1.73
A1	.004	.0098	0.102	0.249
A2	.055	.061	1.40	1.55
B	.008	.012	0.20	0.30
C	.0075	.0098	0.191	0.249
D	SEE VARIATIONS			
E	.150	.157	3.81	3.99
e	.025 BSC		0.635 BSC	
H	.230	.244	5.84	6.20
h	.010	.016	0.25	0.41
L	.016	.035	0.41	0.89
N	SEE VARIATIONS			
α	0°	8°	0°	8°



### VARIATIONS:

	INCHES		MILLIMETERS		N	
	MIN.	MAX.	MIN.	MAX.		
D	.189	.196	4.80	4.98	16	AB
S	.0020	.0070	0.05	0.18		
D	.337	.344	8.56	8.74	20	AD
S	.0500	.0550	1.270	1.397		
D	.337	.344	8.56	8.74	24	AE
S	.0250	.0300	0.635	0.762		
D	.386	.393	9.80	9.98	28	AF
S	.0250	.0300	0.635	0.762		

### NOTES:

- 1). D & E DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
- 2). MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .006" PER SIDE.
- 3). CONTROLLING DIMENSIONS: INCHES.
- 4). MEETS JEDEC MO137.

 <b>DALLAS</b> SEMICONDUCTOR			
PROPRIETARY INFORMATION			
TITLE: PACKAGE OUTLINE, QSOP .150", .025" LEAD PITCH			
APPROVAL 	DOCUMENT CONTROL NO. 21-0055	REV. E	1/1

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