# 2.7-V TO 5.5-V LOW-POWER DUAL 8-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

SLAS236A –JUNE 1999 – REVISED JUNE 2000

features

- **Dual 8-Bit Voltage Output DAC**
- **Programmable Internal Reference**
- **Programmable Settling Time:**
  - **0.8 μs in Fast Mode ,**
  - **2.8 μs in Slow Mode**
- **Compatible With TMS320 and SPI™ Serial Ports**
- **Differential Nonlinearity <0.1 LSB Typ**
- **Monotonic Over Temperature**

D PACKAGE
(TOP VIEW)

```
        _____
DIN  [ 1        8 ] V_DD
SCLK [ 2        7 ] OUTB
 CS  [ 3        6 ] REF
OUTA [ 4        5 ] AGND
        _____
```
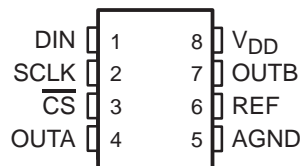
applications

- **Digital Servo Control Loops**
- **Digital Offset and Gain Adjustment**
- **Industrial Process Control**
- **Machine and Motion Control Devices**
- **Mass Storage Devices**

description

The TLV5626 is a dual 8-bit voltage output DAC with a flexible 3-wire serial interface.The serial interface allows glueless interface to TMS320 and SPI™, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing 2 control and 8 data bits.

The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed versus power dissipation. With its on-chip programmable precision voltage reference, the TLV5626 simplifies overall system design.

Because of its ability to source up to 1 mA, the reference can also be used as a system reference. Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package to reduce board space in standard commercial and industrial temperature ranges.

AVAILABLE OPTIONS

	PACKAGE
T$_A$	**SOIC (D)**
0°C to 70°C	TLV5626CD
−40°C to 85°C	TLV5626ID

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Microwire is a trademark of National Semiconductor Corporation.

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TEXAS INSTRUMENTS

functional block diagram

Terminal Functions
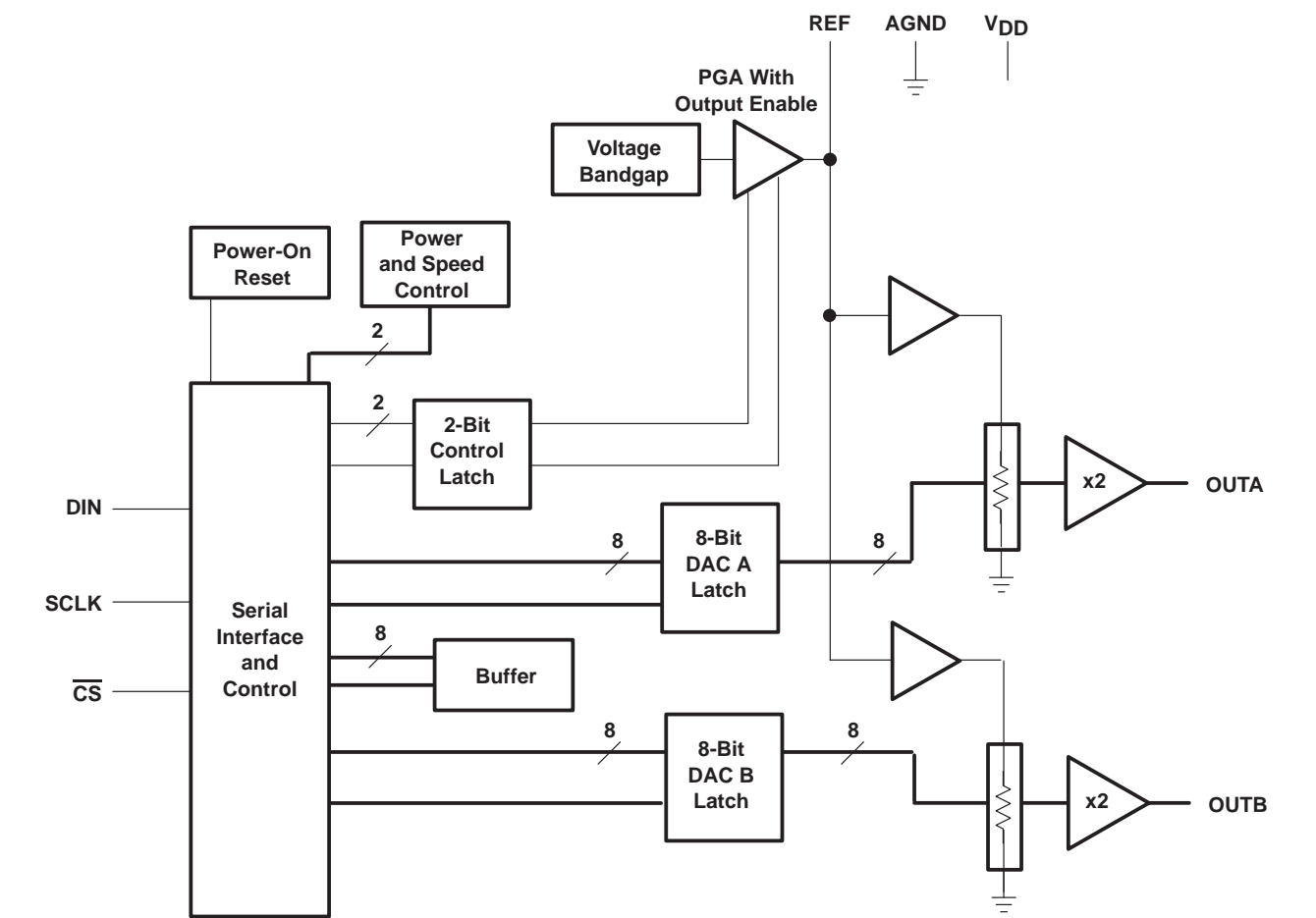
TERMINAL		I/O/P	DESCRIPTION
NAME	NO.		
AGND	5	P	Ground
$\overline{\text{CS}}$	3	I	Chip select. Digital input active low, used to enable/disable inputs
DIN	1	I	Digital serial data input
OUTA	4	I	DAC A analog voltage output
OUTB	7	O	DAC B analog voltage output
REF	6	I/O	Analog reference voltage input/output
SCLK	2	I	Digital serial clock input
V_{DD}	8	P	Positive power supply

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage (V_{DD} to AGND) . . . . 7 V
Reference input voltage range . . . . -0.3 V to V_{DD} + 0.3 V
Digital input voltage range . . . . -0.3 V to V_{DD} + 0.3 V
Operating free-air temperature range, T_A: TLV5626C . . . . 0°C to 70°C
TLV5626I . . . . -40°C to 85°C
Storage temperature range, T_{stg} . . . . -65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds . . . . 260°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and
functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not
implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	V_{DD} = 5 V	4.5	5	5.5	V
	V_{DD} = 3 V	2.7	3	3.3	V
Power on threshold voltage, POR		0.55		2	V
High-level digital input voltage, V_{IH}	V_{DD} = 2.7 V to 5.5 V	2			V
Low-level digital input voltage, V_{IL}	V_{DD} = 2.7 V to 5.5 V			0.8	V
Reference voltage, V_{ref} to REF terminal	V_{DD} = 5 V (see Note 1)	AGND	2.048	$V_{DD}-1.5$	V
Reference voltage, V_{ref} to REF terminal	V_{DD} = 3 V (see Note 1)	AGND	1.024	$V_{DD}-1.5$	V
Load resistance, R_L		2			kΩ
Load capacitance, C_L				100	pF
Clock frequency, f_{CLK}				20	MHz
Operating free-air temperature, T_A	TLV5626C	0		70	°C
	TLV5626I	-40		85	

NOTE 1: Due to the x2 output buffer, a reference input voltage $\geq (V_{DD} - 0.4$ V)/2 causes clipping of the transfer function. The output buffer of the
internal reference must be disabled, if an external reference is used.

electrical characteristics over recommended operating conditions (unless otherwise noted)

power supply

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
I_{DD}	Power supply current	No load, All inputs = AGND or V_{DD}, DAC latch = 0x800	V_{DD} = 5 V, Int. ref.	Fast		4.2	7	mA
				Slow		2	3.6	mA
			V_{DD} = 3 V, Int. ref.	Fast		3.7	6.3	mA
				Slow		1.7	3.0	mA
			V_{DD} = 5 V, Ext. ref.	Fast		3.8	6.3	mA
				Slow		1.7	3.0	mA
			V_{DD} = 3 V, Ext. ref.	Fast		3.4	5.7	mA
				Slow		1.4	2.6	mA
	Power-down supply current					1		µA
PSRR	Power supply rejection ratio	Zero scale, See Note 2				−65		dB
		Full scale, See Note 3				−65		

NOTES: 2. Power supply rejection ratio at zero scale is measured by varying V_{DD} and is given by:
$$PSRR = 20 \log [(E_{ZS}(V_{DD}max) - E_{ZS}(V_{DD}min))/V_{DD}max]$$
3. Power supply rejection ratio at full scale is measured by varying V_{DD} and is given by:
$$PSRR = 20 \log [(E_{G}(V_{DD}max) - E_{G}(V_{DD}min))/V_{DD}max]$$

static DAC specifications

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Resolution			8		bits
INL	Integral nonlinearity, end point adjusted	See Note 4		±0.4	±1	LSB
DNL	Differential nonlinearity	See Note 5		±0.1	±0.5	LSB
E_{ZS}	Zero-scale error (offset error at zero scale)	See Note 6			±24	mV
E_{ZS} TC	Zero-scale-error temperature coefficient	See Note 7		10		ppm/°C
E_{G}	Gain error	See Note 8			±0.6	% full scale V
E_{G} T$_C$	Gain error temperature coefficient	See Note 9		10		ppm/°C

NOTES: 4. The relative accuracy or integral nonlinearity (INL) sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.
5. The differential nonlinearity (DNL) sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.
6. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.
7. Zero-scale-error temperature coefficient is given by: E_{ZS} TC = $[E_{ZS} (T_{max}) - E_{ZS} (T_{min})]/V_{ref} \times 10^6/(T_{max} - T_{min})$.
8. Gain error is the deviation from the ideal output ($2V_{ref} - 1$ LSB) with an output load of 10 k excluding the effects of the zero-error.
9. Gain temperature coefficient is given by: E_{G} TC = $[E_{G}(T_{max}) - E_{G} (T_{min})]/V_{ref} \times 10^6/(T_{max} - T_{min})$.

output specifications

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_O	Output voltage	R_L = 10 kΩ	0		V_{DD}−0.4	V
	Output load regulation accuracy	V_O = 4.096 V, 2.048 V, R_L = 2 kΩ vs 10 kⓒ			±0.25	% full scale V

electrical characteristics over recommended operating conditions (unless otherwise noted) (Continued)

reference pin configured as output (REF)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{ref(OUTL)}$	Low reference voltage		1.003	1.024	1.045	V
$V_{ref(OUTH)}$	High reference voltage	$V_{DD} > 4.75$ V	2.027	2.048	2.069	V
$I_{ref(source)}$	Output source current				1	mA
$I_{ref(sink)}$	Output sink current		−1			mA
	Load capacitance				100	pF
PSRR	Power supply rejection ratio			−65		dB

reference pin configured as input (REF)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_I	Input voltage			0		V_{DD}–1.5	V
R_I	Input resistance				10		MΩ
C_I	Input capacitance				5		pF
	Reference input bandwidth	REF = 0.2 V_{pp} + 1.024 V dc	Fast		1.3		MHz
			Slow		525		kHz
	Reference feedthrough	REF = 1 V_{pp} at 1 kHz + 1.024 V dc (see Note 10)			−80		dB

NOTE 10: Reference feedthrough is measured at the DAC output with an input code = 0x000.

digital inputs

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{IH}	High-level digital input current	$V_I = V_{DD}$			1	µA
I_{IL}	Low-level digital input current	$V_I = 0$ V	−1			µA
C_i	Input capacitance			8		pF

analog output dynamic performance

PARAMETER		TEST CONDITIONS			MIN	TYP	MAX	UNIT
$t_{s(FS)}$	Output settling time, full scale	$R_L = 10$ kΩ, See Note 11	$C_L = 100$ pF,	Fast		0.8	2.4	µs
				Slow		2.8	5.5	
$t_{s(CC)}$	Output settling time, code to code	$R_L = 10$ kΩ, See Note 12	$C_L = 100$ pF,	Fast		0.4	1.2	µs
				Slow		0.8	1.6	
SR	Slew rate	$R_L = 10$ kΩ, See Note 13	$C_L = 100$ pF,	Fast		12		V/µs
				Slow		1.8		
	Glitch energy	DIN = 0 to 1, $\overline{CS} = V_{DD}$	$f_{CLK} = 100$ kHz,			5		nV–S
SNR	Signal-to-noise ratio				53	57		dB
S/(N+D)	Signal-to-noise + distortion	$f_S = 480$ kSPS, $f_{out} = 1$ kHz,			48	47		
THD	Total harmonic distortion	$R_L = 10$ kΩ, $C_L = 100$ pF				−50	−48	
SFDR	Spurious free dynamic range				50	62		

NOTES: 11. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0xFD0 or 0xFD0 to 0x020 respectively. Not tested, assured by design.
    12. Settling time is the time for the output signal to remain within ± 0.5 LSB of the final measured value for a digital input code change of one count. Not tested, assured by design.
    13. Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.

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digital input timing requirements

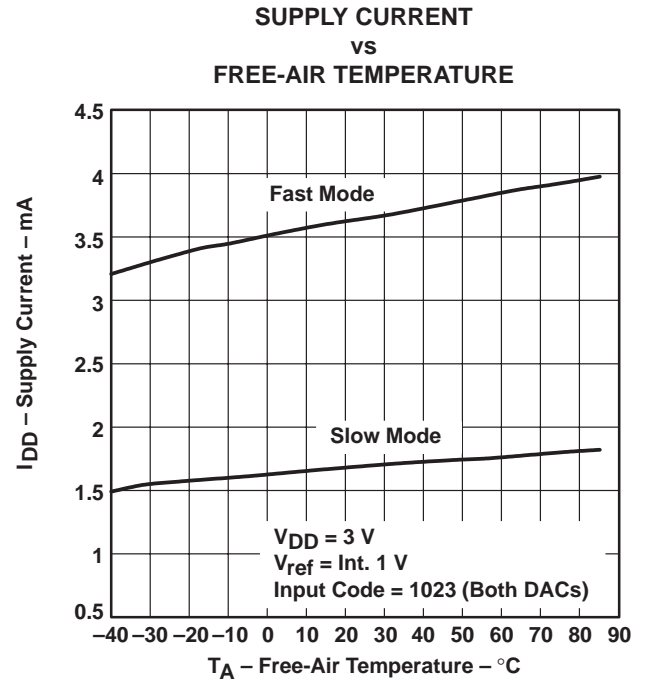|		MIN	NOM	MAX	UNIT
$t_{su(CS–CK)}$	Setup time, $\overline{CS}$ low before first negative SCLK edge	10			ns
$t_{su(C16-CS)}$	Setup time, 16th negative SCLK edge (when D0 is sampled) before $\overline{CS}$ rising edge	10			ns
t_{wH}	SCLK pulse width high	25			ns
t_{wL}	SCLK pulse width low	25			ns
$t_{su(D)}$	Setup time, data ready before SCLK falling edge	10			ns
$t_{h(D)}$	Hold time, data held valid after SCLK falling edge	5			ns

PARAMETER MEASUREMENT INFORMATION
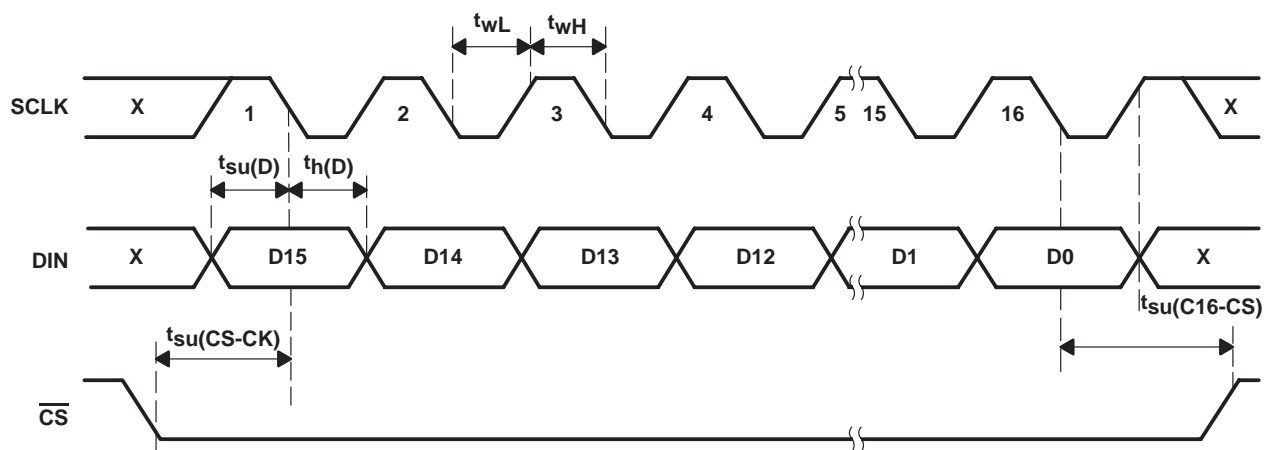
Figure 1. Timing Diagram

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TYPICAL CHARACTERISTICS

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

V_{DD} = 5 V
V_{ref} = Int. 2 V
Input Code = 1023 (Both DACs)

I_{DD} – Supply Current – mA

T_A – Free-Air Temperature – °C

Figure 2

SUPPLY CURRENT
vs
FREE-AIR TEMPERATURE

V_{DD} = 3 V
V_{ref} = Int. 1 V
Input Code = 1023 (Both DACs)

I_{DD} – Supply Current – mA

T_A – Free-Air Temperature – °C

Figure 3

POWER DOWN SUPPLY CURRENT
vs
TIME

I_{DD} – Power Down Supply Current – mA

t – Time – μs

Figure 4

OUTPUT VOLTAGE
vs
LOAD CURRENT

V_{DD} = 3 V
V_{ref} = Int. 1 V
Input Code = 4095

V_O – Output Voltage – V

Source Current – mA

Figure 5

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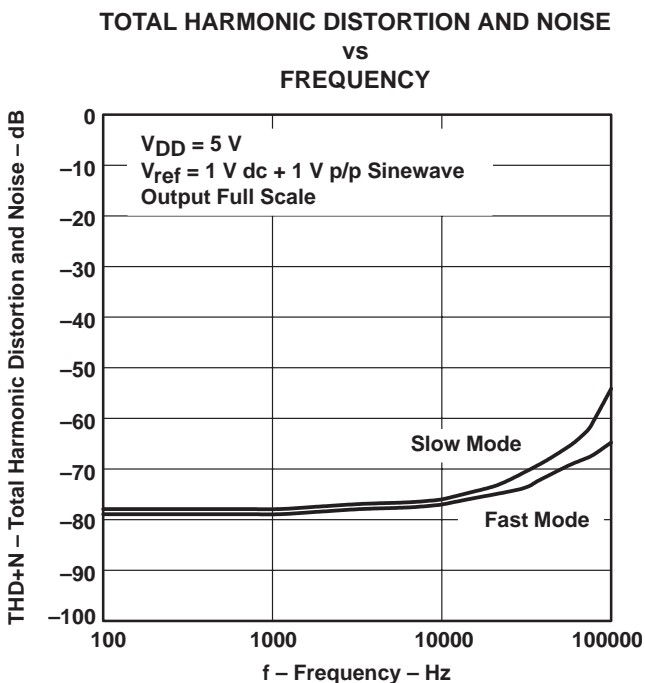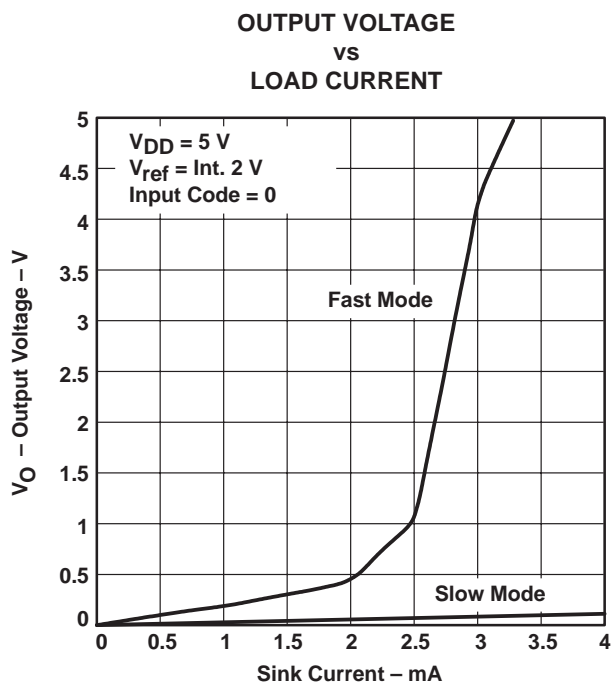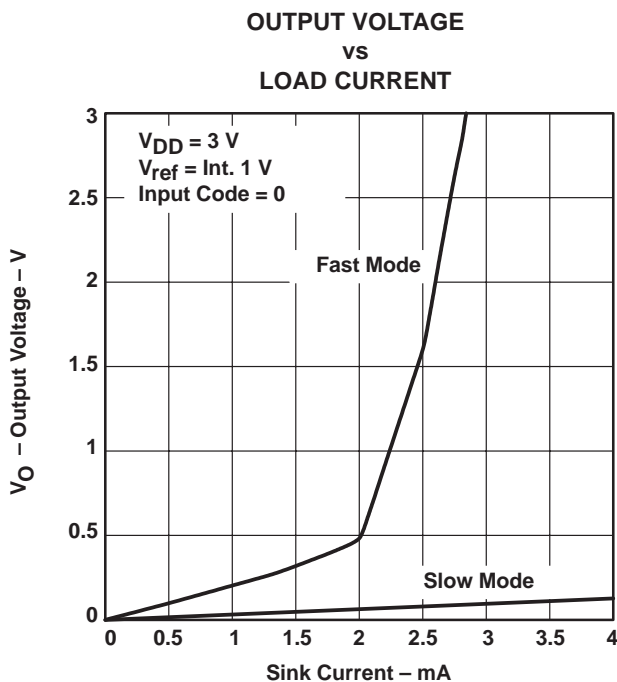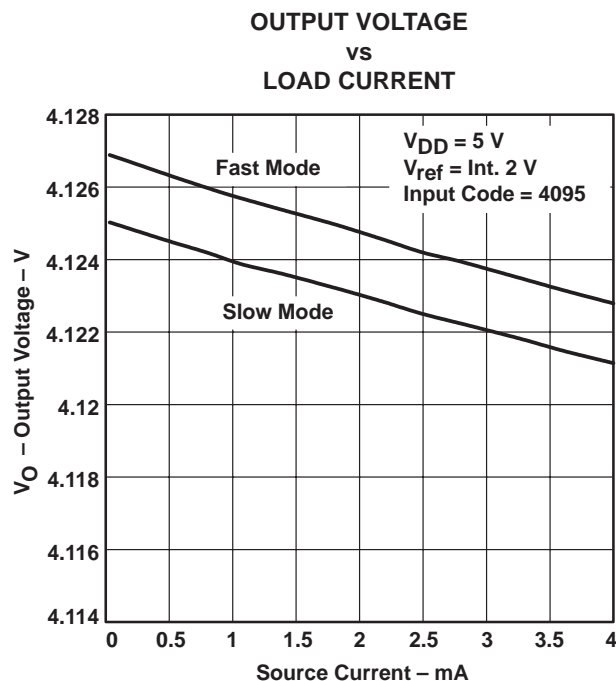POST OFFICE BOX 655303 ● DALLAS, TEXAS 75265

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TLV5626
2.7-V TO 5.5-V LOW-POWER DUAL 8-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

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TYPICAL CHARACTERISTICS

OUTPUT VOLTAGE vs LOAD CURRENT

V_{DD} = 5 V
V_{ref} = Int. 2 V
Input Code = 4095

Fast Mode
Slow Mode

V_O – Output Voltage – V
Source Current – mA

Figure 6

OUTPUT VOLTAGE vs LOAD CURRENT

V_{DD} = 3 V
V_{ref} = Int. 1 V
Input Code = 0

Fast Mode
Slow Mode

V_O – Output Voltage – V
Sink Current – mA

Figure 7

OUTPUT VOLTAGE vs LOAD CURRENT

V_{DD} = 5 V
V_{ref} = Int. 2 V
Input Code = 0

Fast Mode
Slow Mode

V_O – Output Voltage – V
Sink Current – mA

Figure 8

TOTAL HARMONIC DISTORTION AND NOISE vs FREQUENCY

V_{DD} = 5 V
V_{ref} = 1 V dc + 1 V p/p Sinewave
Output Full Scale

Slow Mode
Fast Mode

THD+N – Total Harmonic Distortion and Noise – dB
f – Frequency – Hz

Figure 9

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TYPICAL CHARACTERISTICS

TOTAL HARMONIC DISTORTION
vs
FREQUENCY
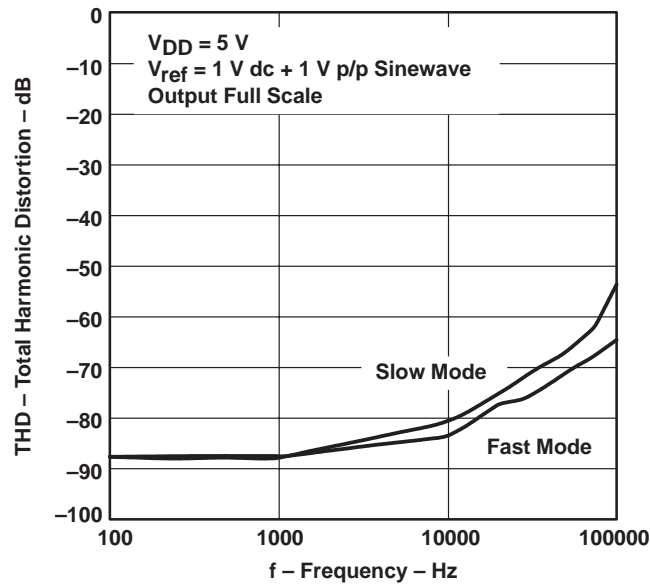
Figure 10

DIFFERENTIAL NONLINEARITY
vs
DIGITAL OUTPUT CODE

Figure 11

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TYPICAL CHARACTERISTICS

INTEGRAL NONLINEARITY
vs
DIGITAL OUTPUT CODE
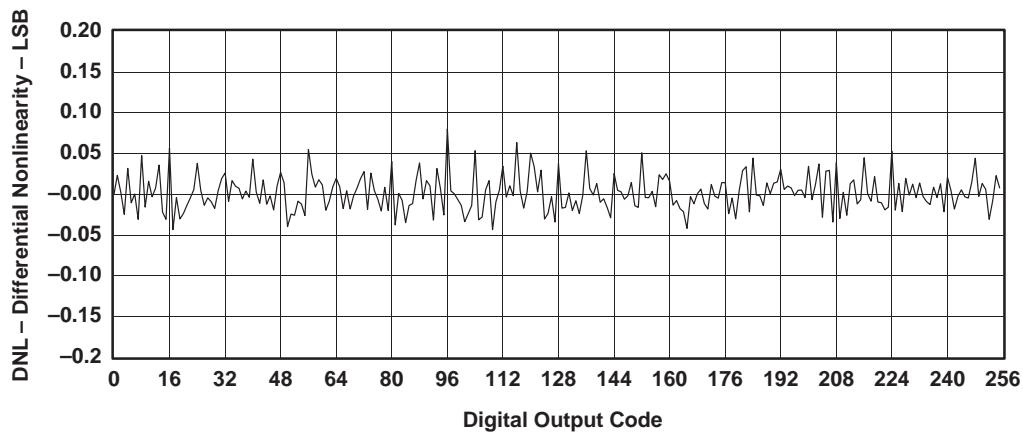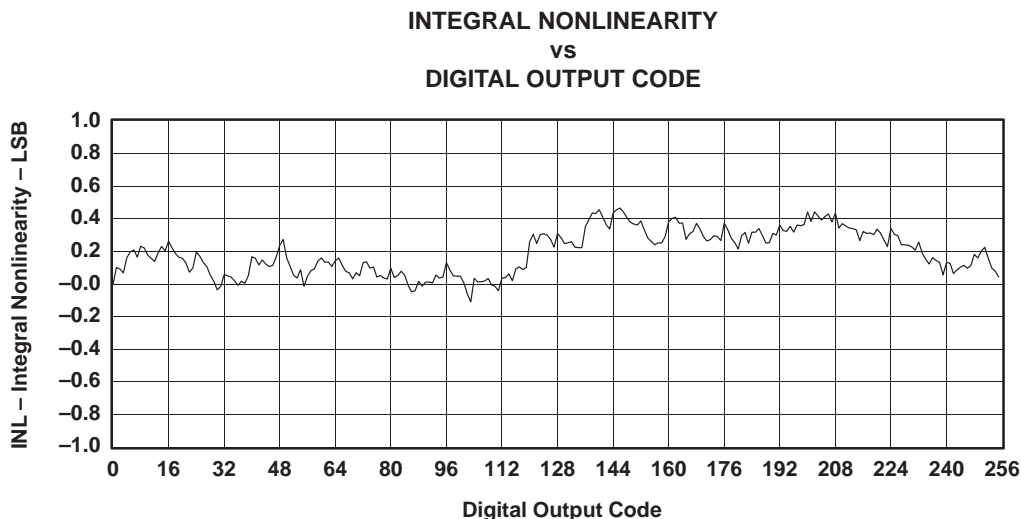

Figure 12

APPLICATION INFORMATION

general function

The TLV5626 is a dual 8-bit, single supply DAC, based on a resistor string architecture. It consists of a serial interface, a speed and power-down control logic, a programmable internal reference, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by reference) is given by:

$$2 \ REF \ \frac{CODE}{0x1000} \ [V]$$

Where REF is the reference voltage and CODE is the digital input value in the range 0x000 to 0xFF0. Bits 3 to 0 must be set to zero. A power-on reset initially puts the internal latches to a defined state (all bits zero).

serial interface

A falling edge of $\overline{CS}$ starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or $\overline{CS}$ rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 13 shows examples of how to connect the TLV5626 to TMS320, SPI™, and Microwire™.

Figure 13. Three-Wire Interface

APPLICATION INFORMATION

Notes on SPI™ and Microwire™: Before the controller starts the data transfer, the software has to generate a falling edge on the I/O pin connected to $\overline{CS}$. If the word width is 8 bits (SPI™ and Microwire™), two write operations must be performed to program the TLV5626. After the write operation(s), the holding registers or the control register are updated automatically on the 16th positive clock edge.

serial clock frequency and update rate

The maximum serial clock frequency is given by:

$$f_{sclkmax} = \frac{1}{t_{whmin} + t_{wlmin}} = 20 \text{ MHz}$$

The maximum update rate is:

$$f_{updatemax} = \frac{1}{16 \left(t_{whmin} + t_{wlmin}\right)} = 1.25 \text{ MHz}$$

The maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5626 has to be considered, too.

data format

The 16-bit data word for the TLV5626 consists of two parts:

- Program bits (D15..D12)
- New data (D11..D0)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R1	SPD	PWR	R0						12 Data bits						

SPD: Speed control bit   1 → fast mode   0 → slow mode
PWR: Power control bit   1 → power down   0 → normal operation

The following table lists the possible combination of the register select bits:

register select bits

R1	R0	REGISTER
0	0	Write data to DAC B and BUFFER
0	1	Write data to BUFFER
1	0	Write data to DAC A and update DAC B with BUFFER content
1	1	Write data to control register

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

data bits: DAC A, DAC B and BUFFER

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
				New DAC Value				0	0	0	0

If control is selected, then D1, D0 of the 12 data bits are used to program the reference voltage:

APPLICATION INFORMATION

data bits: CONTROL

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
X	X	X	X	X	X	X	X	X	X	REF1	REF0

X: don't care

REF1 and REF0 determine the reference source and, if internal reference is selected, the reference voltage.

reference bits

REF1	REF0	REFERENCE
0	0	External
0	1	1.024 V
1	0	2.048 V
1	1	External

CAUTION:
If external reference voltage is applied to the REF pin, external reference MUST be selected.

examples of operation:

- Set DAC A output, select fast mode, select internal reference at 2.048 V:

    1. Set reference voltage to 2.048 V (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	1	0

    2. Write new DAC A value and update DAC A output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	0	New DAC A output value								0	0	0	0

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set DAC B output, select fast mode, select external reference:

    3. Select external reference (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0

    4. Write new DAC B value to BUFFER and update DAC B output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	1	0	0	New BUFFER content and DAC B output value								0	0	0	0

X = Don't care

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

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APPLICATION INFORMATION

examples of operation: (continued)

- Set DAC A value, set DAC B value, update both simultaneously, select slow mode, select internal reference at 1.024 V:

    1. Set reference voltage to 1.024 V (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	1	0	0	0	0	0	0	0	0	0	0	0	1

    2. Write data for DAC B to BUFFER:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1				New DAC B value					0	0	0	0

X = Don't care

    3. Write new DAC A value and update DAC A and B simultaneously:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	0				New DAC A value					0	0	0	0

X = Don't care

Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

- Set power-down mode:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
X	X	1	X	X	X	X	X	X	X	X	X	X	X	X	X

X = Don't care

linearity, offset, and gain error using single ended supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 14.
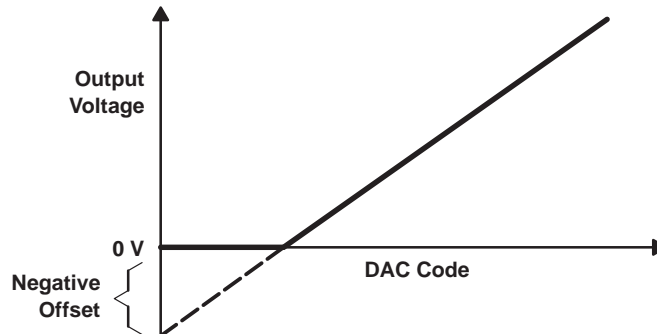

Figure 14. Effect of Negative Offset (single supply)

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APPLICATION INFORMATION

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.

definitions of specifications and terminology

integral nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

differential nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

zero-scale error (E_{ZS})

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

gain error (E_G)

Gain error is the error in slope of the DAC transfer function.

signal-to-noise ratio + distortion (S/N+D)

S/N+D is the ratio of the rms value of the measured input signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

spurious free dynamic range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.

TEXAS
INSTRUMENTS

MECHANICAL DATA

D (R-PDSO-G**)   **PLASTIC SMALL-OUTLINE PACKAGE**

14 PINS SHOWN

PINS **	8	14	16
DIM			
A MAX	0.197 (5,00)	0.344 (8,75)	0.394 (10,00)
A MIN	0.189 (4,80)	0.337 (8,55)	0.386 (9,80)

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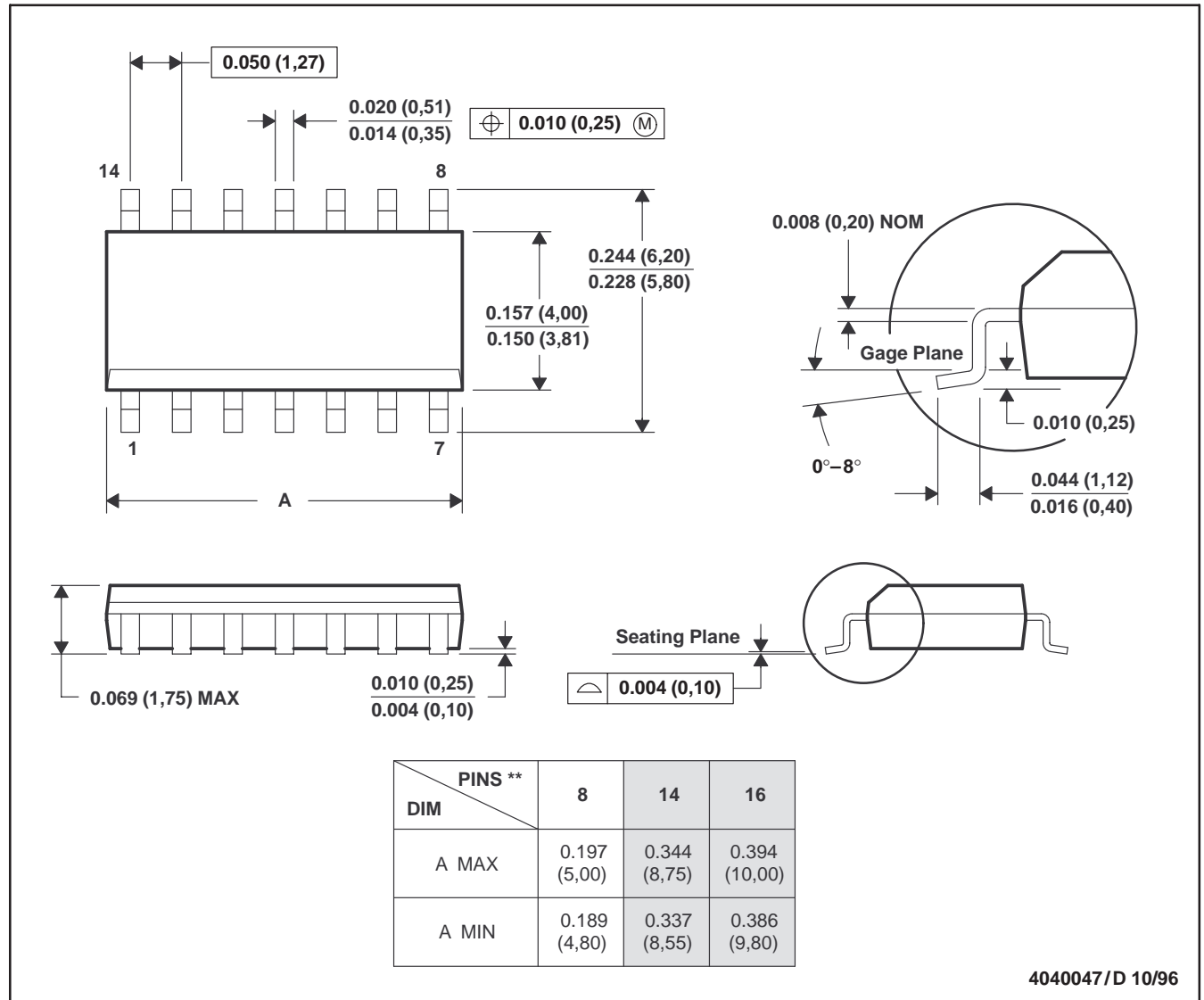
NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
D. Falls within JEDEC MS-012

TEXAS INSTRUMENTS

PACKAGING INFORMATION

Orderable Device	Status [1]	Package Type	Package Drawing	Pins	Package Qty	Eco Plan [2]	Lead/Ball Finish	MSL Peak Temp [3]
TLV5626CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5626IDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

[1] The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

[2] Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
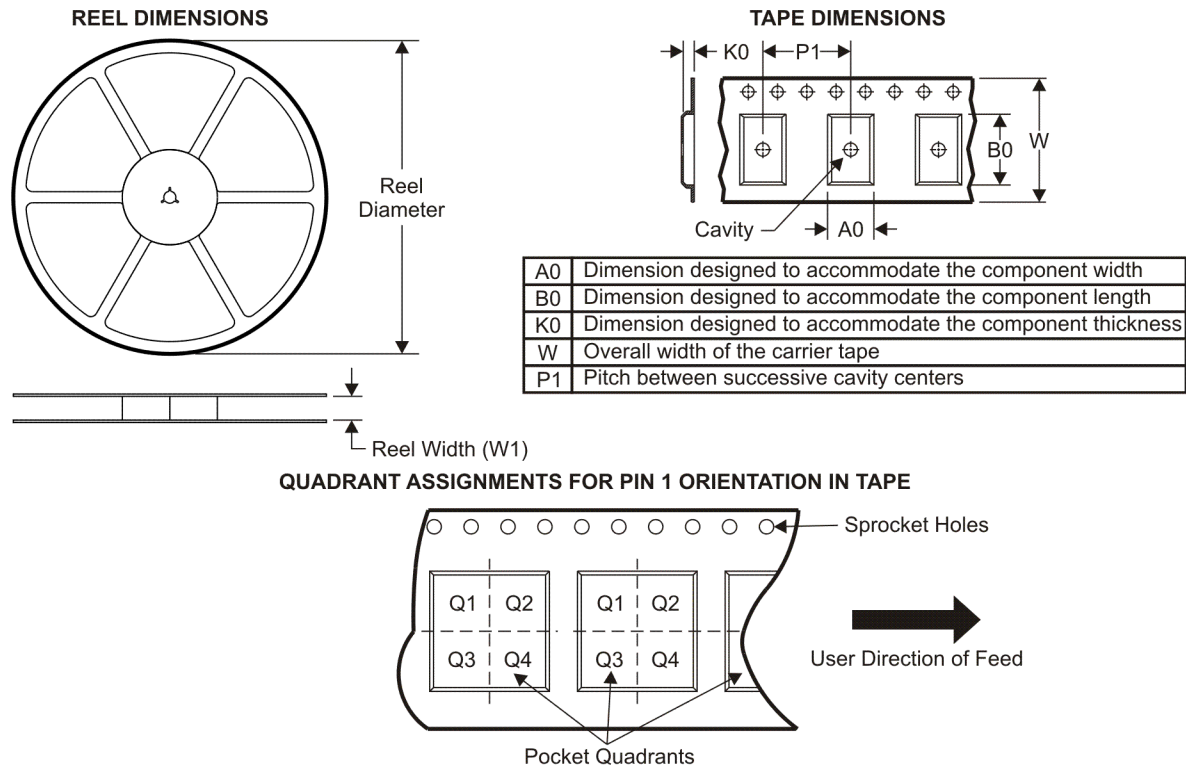**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

[3] MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5626CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5626IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS
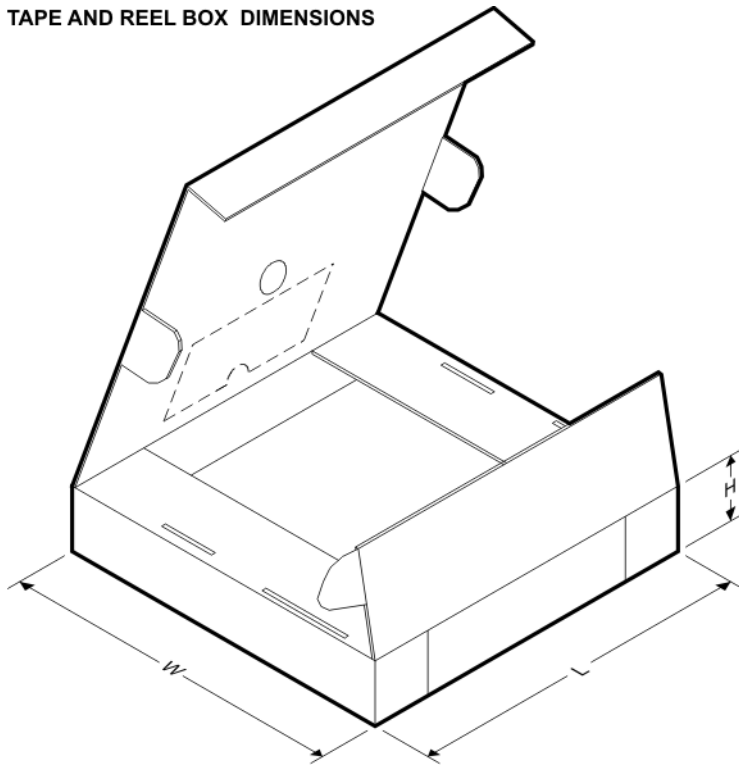
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV5626CDR	SOIC	D	8	2500	346.0	346.0	29.0
TLV5626IDR	SOIC	D	8	2500	346.0	346.0	29.0

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