

TJA1048

Dual high-speed CAN transceiver with Standby mode

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Product data sheet

1. General description

The TJA1048 is a dual high-speed CAN transceiver that provides an interface between a Controller Area Network (CAN) protocol controller and the physical two-wire CAN bus. The transceiver is designed for high-speed CAN applications in the automotive industry, providing the differential transmit and receive capability to (a microcontroller with) a CAN protocol controller.

The TJA1048 belongs to the third generation of high-speed CAN transceivers from NXP Semiconductors, offering significant improvements over first- and second-generation devices such as the TJA1040. It offers improved Electro Magnetic Compatibility (EMC) and ElectroStatic Discharge (ESD) performance, and also features:

- Ideal passive behavior to the CAN bus when the supply voltage is off
- A very low-current Standby mode with bus wake-up capability on both channels
- Can be interfaced directly to microcontrollers with supply voltages from 3 V to 5 V

The TJA1048 implements the CAN physical layer as defined in the current ISO11898 standard (ISO11898-2:2003, ISO11898-5:2007) and the pending updated version of ISO 11898-2:2016. Pending the release of the updated version of ISO11898-2:2016 including CAN FD and SAE J2284-4/5, additional timing parameters defining loop delay symmetry are specified. This implementation enables reliable communication in the CAN FD fast phase at data rates up to 5 Mbit/s.

These features make the TJA1048 an excellent choice for all types of HS-CAN networks containing more than one HS-CAN interface that require a low-power mode with wake-up capability via the CAN bus, especially for Body Control and Gateway units.

2. Features and benefits

2.1 General

- Two TJA1042/3 HS-CAN transceivers combined monolithically in a single package
- Fully ISO 11898-2:2003 and ISO 11898-5:2007 compliant
- Timing guaranteed for data rates up to 5 Mbit/s in the CAN FD fast phase
- Suitable for 12 V and 24 V systems
- Low ElectroMagnetic Emission (EME) and high ElectroMagnetic Immunity (EMI)
- V_{IO} input allows for direct interfacing with 3 V to 5 V microcontrollers
- Available in SO14 and HVSON14 packages
- Leadless HVSON14 package (3.0 mm × 4.5 mm) with improved Automated Optical Inspection (AOI) capability



- Dark green product (halogen free and Restriction of Hazardous Substances (RoHS) compliant)
- AEC-Q100 qualified

2.2 Low-power management

- Very low-current Standby mode with host and bus wake-up capability
- Functional behavior predictable under all supply conditions
- Transceiver disengages from the bus when not powered up (zero load)
- Wake-up receiver powered by V_{IO} ; allows shut down of V_{CC}

2.3 Protection

- High ESD handling capability on the bus pins
- Bus pins protected against transients in automotive environments
- Transmit Data (TXD) dominant time-out function
- Undervoltage detection on pins V_{CC} and V_{IO}
- Thermally protected

3. Quick reference data

Table 1. Quick reference data

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{CC}	supply voltage		4.5	-	5.5	V
V_{IO}	supply voltage on pin V_{IO}		2.8	-	5.5	V
$V_{uvd}(V_{CC})$	undervoltage detection voltage on pin V_{CC}		3.5	-	4.5	V
$V_{uvd}(V_{IO})$	undervoltage detection voltage on pin V_{IO}		1.3	2.0	2.7	V
I_{CC}	supply current	Standby mode	-	0.5	2	μ A
		Normal mode				
		both channels recessive	-	-	20	mA
		one channel dominant	-	-	80	mA
		both channels dominant	-	90	140	mA
I_{IO}	supply current on pin V_{IO}	Standby mode; $V_{TXD} = V_{IO}$	-	16.5	26	μ A
		Normal mode				
		both channels recessive	-	-	35	μ A
		one channel dominant	-	-	300	μ A
		both channels dominant	-	-	550	μ A
V_{ESD}	electrostatic discharge voltage	IEC 61000-4-2 at pins CANHx and CANLx	-6	-	+6	kV
V_{CANH}	voltage on pin CANH	pins CANH1 and CANH2	-58	-	+58	V
V_{CANL}	voltage on pin CANL	pins CANL1 and CANL2	-58	-	+58	V
T_{vj}	virtual junction temperature		-40	-	+150	$^{\circ}$ C

4. Ordering information

Table 2. Ordering information

Type number	Package		
	Name	Description	Version
TJA1048T	SO14	plastic small outline package; 14 leads; body width 3.9 mm	SOT108-1
TJA1048TK	HVSON14	plastic, thermal enhanced very thin small outline package; no leads; 14 terminals; body 3 × 4.5 × 0.85 mm	SOT1086-2

5. Block diagram

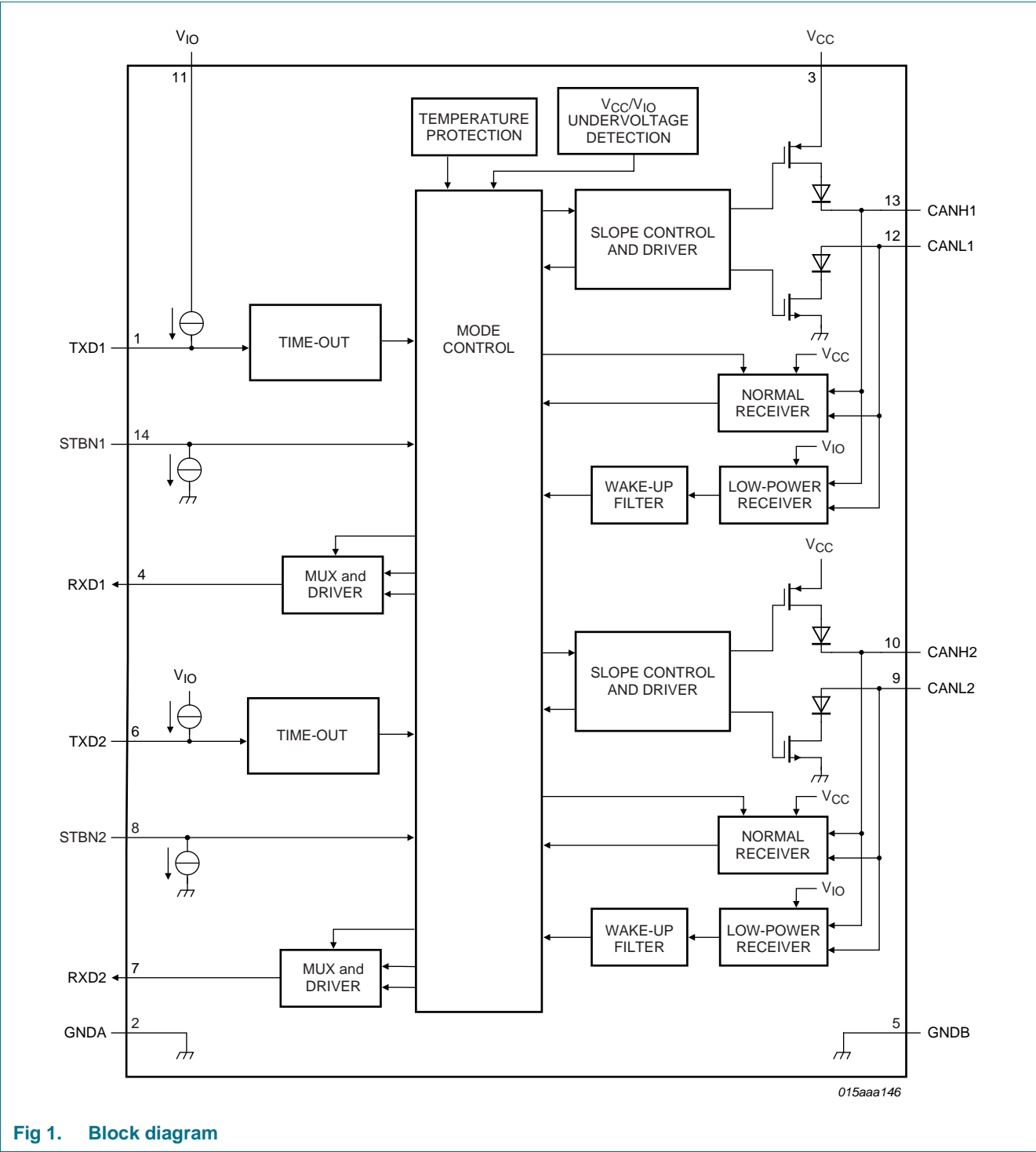


Fig 1. Block diagram

6. Pinning information

6.1 Pinning

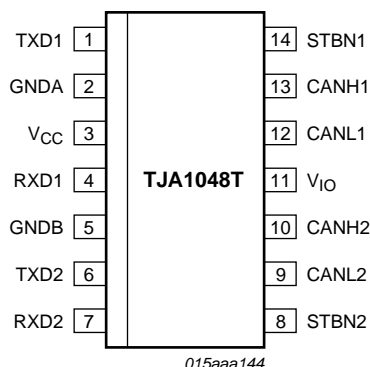


Fig 2. Pin configuration diagram: SO14

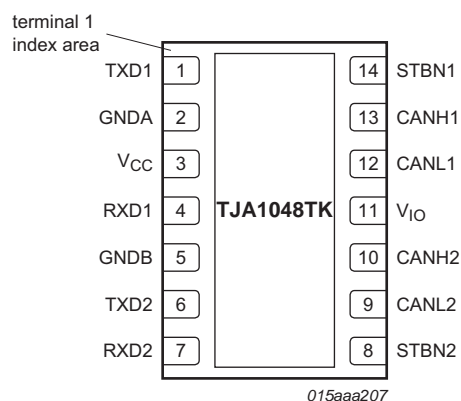


Fig 3. Pin configuration diagram: HVSON14

6.2 Pin description

Table 3. Pin description

Symbol	Pin	Description
TXD1	1	transmit data input 1
GNDA	2 ^[1]	transceiver ground
V _{CC}	3	transceiver supply voltage
RXD1	4	receive data output 1; reads out data from bus line1
GNDB	5 ^[1]	transceiver ground
TXD2	6	transmit data input 2
RXD2	7	receive data output 2; reads out data from bus line 2
STBN2	8	standby control input 2 (HIGH = Normal mode, LOW = Standby mode)
CANL2	9	LOW-level CAN bus line 2
CANH2	10	HIGH-level CAN bus line 2
V _{IO}	11	supply voltage for I/O level adapter
CANL1	12	LOW-level CAN bus line 1
CANH1	13	HIGH-level CAN bus line 1
STBN1	14	standby control input 1 (HIGH = Normal mode, LOW = Standby mode)

[1] Pins 2 and 5 must be connected together externally in the application. HVSON14 package die supply ground is connected to both the GND pin and the exposed center pad. The GND pin must be soldered to board ground. For enhanced thermal and electrical performance, it is recommended that the exposed center pad also be soldered to board ground.

7. Functional description

The TJA1048 is a dual HS-CAN stand-alone transceiver with Standby mode and robust ESD handling capability. It combines the functionality of two TJA1042/3 transceivers with improved EMC and quiescent current performance. Improved slope control and high DC handling capability on the bus pins provide additional application flexibility.

7.1 Operating modes

The TJA1048 supports two operating modes per transceiver, Normal and Standby. The operating mode can be selected independently for each transceiver via pins STBN1 and STBN2 (see [Table 4](#)).

Table 4. Operating modes

Mode	Pin STBN1/STBN2	Pin RXD1/RXD2	
		LOW	HIGH
Normal	HIGH	bus dominant	bus recessive
Standby	LOW	wake-up request detected	no wake-up request detected

7.1.1 Normal mode

A HIGH level on pin STBN1/STBN2 selects Normal mode. In this mode, the transceiver can transmit and receive data via the bus lines CANH1/CANL1 and CANH2/CANL2 (see [Figure 1](#) for the block diagram). The differential receiver converts the analog data on the bus lines into digital data which is output on pin RXD1/RXD2. The slopes of the output signals on the bus lines are controlled internally and are optimized in a way that guarantees the lowest possible EME.

7.1.2 Standby mode

A LOW level on pin STBN1/STBN2 selects Standby mode. In Standby mode, the transceiver is not able to transmit or correctly receive data via the bus lines. The transmitter and Normal-mode receiver blocks are switched off to reduce supply current, and only a low-power differential receiver monitors the bus lines for activity.

In Standby mode, the bus lines are biased to ground to minimize the system supply current. The low-power receiver is supplied by V_{IO} , and is capable of detecting CAN bus activity even if V_{IO} is the only supply voltage available. When pin RXD1/RXD2 goes LOW to signal a wake-up request, a transition to Normal mode will not be triggered until STBN1/STBN2 is forced HIGH.

7.1.3 Remote wake-up (via the CAN bus)

A dedicated wake-up sequence (specified in ISO11898-5:2007) must be received to wake-up the TJA1048 from a low-power mode. This filtering is necessary to avoid spurious wake-up events due to a dominant clamped CAN bus or dominant phases caused by noise or spikes on the bus.

A valid wake-up pattern consists of:

- A dominant phase of at least $t_{wake(busdom)}$ followed by
- A recessive phase of at least $t_{wake(busrec)}$ followed by
- A dominant phase of at least $t_{wake(busdom)}$

The complete dominant-recessive-dominant pattern must be received within $t_{to(wake)bus}$ to be recognized as a valid wake-up pattern (see [Figure 4](#)). Pin RXD1/RXD2 will remain recessive until the wake-up event has been triggered.

After a wake-up sequence has been detected, the TJA1048 will remain in Standby mode with the bus signals reflected on RXD1/RXD2. Note that dominant or recessive phases lasting less than $t_{fltr(wake)bus}$ will not be detected by the low-power differential receiver and will not be reflected on RXD1/RXD2 in Standby mode.

A wake-up event will not be registered if any of the following events occurs while a wake-up sequence is being transmitted:

- The TJA1048 switches to Normal mode
- The complete wake-up pattern was not received within $t_{to(wake)bus}$
- A V_{IO} undervoltage is detected ($V_{IO} < V_{uvd(VIO)}$; see [Section 7.2.3](#))

If any of these events occurs while a wake-up sequence is being received, the internal wake-up logic will be reset and the complete wake-up sequence will have to be re-transmitted to trigger a wake-up event.

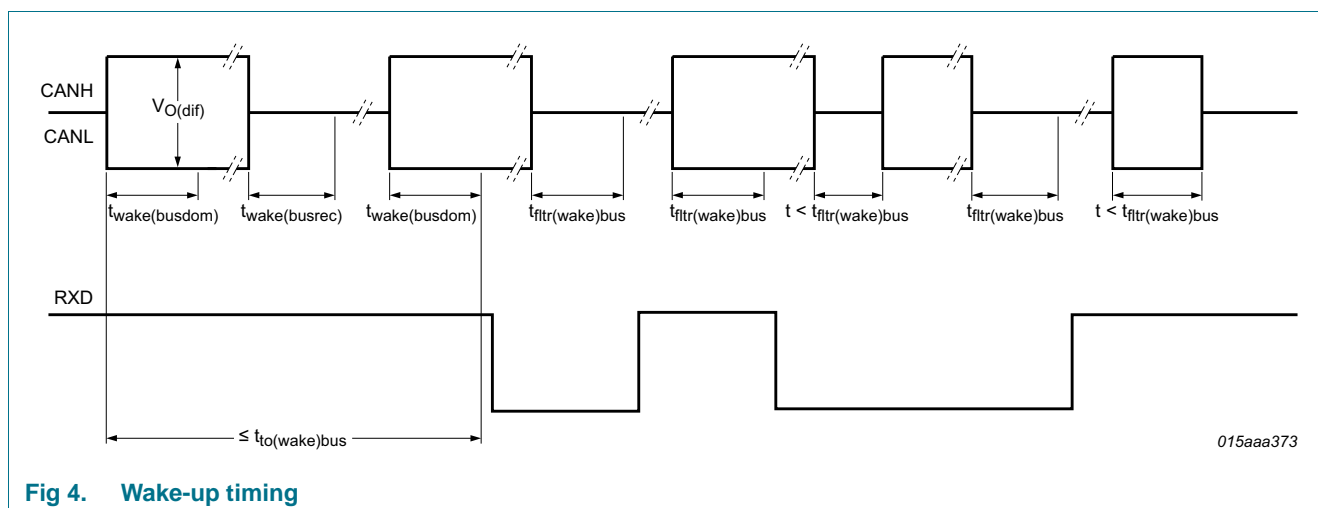


Fig 4. Wake-up timing

7.2 Fail-safe features

7.2.1 TXD dominant time-out function

A 'TXD dominant time-out' timer is started when pin TXD1/TXD2 is set LOW. If the LOW state on this pin persists for longer than $t_{to(dom)TXD}$, the transmitter is disabled, releasing the bus lines to recessive state. This function prevents a hardware and/or software application failure from driving the bus lines to a permanent dominant state (blocking all network communications). The TXD dominant time-out timer is reset when pin TXD1/TXD2 is set HIGH. The TXD dominant time-out time also defines the minimum possible bit rate of 40 kbit/s. The TJA1048 has two TXD dominant time-out timers that operate independently of each other.

7.2.2 Internal biasing of TXD1, TXD2, STBN1 and STBN2 input pins

Pins TXD1 and TXD2 have internal pull-ups to V_{IO} and pins STBN1 and STBN2 have internal pull-downs to GNDA and GNDB. This ensures a safe, defined state if any of these pins is left floating. Pins GNDA and GNDB must be connected together in the application.

Pull-up/pull-down currents flow in these pins in all states. Pins TXD1 and TXD2 should be held HIGH in Standby mode to minimize the supply current; pins STBN1 and STBN2 should be held LOW.

7.2.3 Undervoltage detection on pins V_{CC} and V_{IO}

Should V_{CC} drop below the V_{CC} undervoltage detection level, $V_{uvd(VCC)}$, both transceivers will switch to Standby mode. The logic state of pins STBN1 and STBN2 will be ignored until V_{CC} has recovered.

Should V_{IO} drop below the V_{IO} undervoltage detection level, $V_{uvd(VIO)}$, the transceivers will switch off and disengage from the bus (zero load) until V_{IO} has recovered.

7.2.4 Overtemperature protection

The output drivers are protected against overtemperature conditions. If the virtual junction temperature exceeds the shutdown junction temperature, $T_{j(sd)}$, both output drivers will be disabled. When the virtual junction temperature drops below $T_{j(sd)}$ again, the output drivers will recover independently once TXD1/TXD2 has been reset to HIGH. Including the TXD1/TXD2 condition prevents output driver oscillation due to small variations in temperature.

7.3 V_{IO} supply pin

Pin V_{IO} should be connected to the microcontroller supply voltage (see [Figure 7](#)). This will adjust the signal levels of pins TXD1, TXD2, RXD1, RXD2, STBN1 and STBN2 to the I/O levels of the microcontroller. Pin V_{IO} also provides the internal supply voltage for the transceiver's low-power differential receiver. For applications running in low-power mode, this allows the bus lines to be monitored for activity even if there is no supply voltage on pin V_{CC} .

8. Limiting values

Table 5. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are referenced to GND.

Symbol	Parameter	Conditions	Min	Max	Unit
V_x	voltage on pin x ^[1]	on pins CANH1, CANL1, CANH2 and CANL2	-58	+58	V
		on any other pin	-0.3	+7	V
$V_{(CANH-CANL)}$	voltage between pin CANH and pin CANL		-27	+27	V
V_{trt}	transient voltage	on pins CANH1, CANL1, CANH2 and CANL2 ^[2]			
		pulse 1	-100	-	V
		pulse 2a	-	75	V
		pulse 3a	-150	-	V
		pulse 3b	-	100	V
V_{ESD}	electrostatic discharge voltage	IEC 61000-4-2 (150 pF, 330 Ω) ^[3]			
		on pins CANH1, CANL1, CANH2 and CANL2	-6	+6	kV
		Human Body Model (HBM); 100 pF, 1.5 k Ω ^[4]			
		on pins CANH1, CANL1, CANH2 and CANL2	-6	+6	kV
		at any other pin	-4	+4	kV
		Machine Model (MM); 200 pF, 0.75 μ H, 10 Ω ^[5]			
		at any pin	-300	+300	V
		Charged Device Model (CDM); field Induced charge; 4 pF ^[6]			
		at corner pins	-750	+750	V
		at any pin	-500	+500	V
T_{vj}	virtual junction temperature	^[7]	-40	+150	°C
T_{stg}	storage temperature		-55	+150	°C

- [1] The device can sustain voltages up to the specified values over the product lifetime, provided applied voltages (including transients) never exceed these values.
- [2] According to IEC TS 62228 (2007), Section 4.2.4; parameters for standard pulses defined in ISO7637 part 2: 2004-06.
- [3] According to IEC TS 62228 (2007), Section 4.3; DIN EN 61000-4-2.
- [4] According to AEC-Q100-002.
- [5] According to AEC-Q100-003.
- [6] According to AEC-Q100-011 Rev-C1. The classification level is C4B.
- [7] In accordance with IEC 60747-1. An alternative definition of virtual junction temperature is: $T_{vj} = T_{amb} + P \times R_{th(vj-a)}$, where $R_{th(vj-a)}$ is a fixed value to be used for the calculation of T_{vj} . The rating for T_{vj} limits the allowable combinations of power dissipation (P) and ambient temperature (T_{amb}).

9. Thermal characteristics

Table 6. Thermal characteristics

Values determined for free convection conditions on a JESD51-7 board.

Symbol	Parameter	Conditions	Value	Unit
$R_{th(vj-a)}$	thermal resistance from virtual junction to ambient	SO14	65	K/W
		HVSON14	42	K/W

10. Static characteristics

Table 7. Static characteristics

$T_{vj} = -40\text{ }^{\circ}\text{C}$ to $+150\text{ }^{\circ}\text{C}$; $V_{CC} = 4.5\text{ V}$ to 5.5 V ; $V_{IO} = 2.8\text{ V}$ to 5.5 V ; $R_L = 60\text{ }\Omega$ unless specified otherwise; all voltages are defined with respect to ground; positive currents flow into the device^[1].

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Supply; pin V_{CC}						
V _{CC}	supply voltage		4.5	-	5.5	V
I _{CC}	supply current	Standby mode; V _{TXD} = V _{IO} ^[2]	-	0.5	2	μA
		Normal mode				
		both channels recessive	-	-	20	mA
		one channel dominant	-	-	80	mA
		both channels dominant	-	90	140	mA
V _{uvd(VCC)}	undervoltage detection voltage on pin V _{CC}		3.5	-	4.5	V
I/O level adapter supply; pin V_{IO}						
V _{IO}	supply voltage on pin V _{IO}		2.8	-	5.5	V
I _{IO}	supply current on pin V _{IO}	Standby mode; V _{TXD} = V _{IO} ^[2]	-	16.5	26	μA
		Normal mode				
		both channels recessive	-	-	35	μA
		one channel dominant	-	-	300	μA
		both channels dominant	-	-	550	μA
V _{uvd(VIO)}	undervoltage detection voltage on pin V _{IO}		1.3	2.0	2.7	V
Standby mode control input; pins STBN1 and STBN2						
V _{IH}	HIGH-level input voltage		0.7V _{IO}	-	V _{IO} + 0.3	V
V _{IL}	LOW-level input voltage		-0.3	-	0.3V _{IO}	V
I _{IH}	HIGH-level input current	V _{STBN} ^[3] = V _{IO}	1	-	10	μA
I _{IL}	LOW-level input current	V _{STBN} = 0 V	-1	-	+1	μA
CAN transmit data input; pins TXD1 and TXD2						
V _{IH}	HIGH-level input voltage		0.7V _{IO}	-	V _{IO} + 0.3	V
V _{IL}	LOW-level input voltage		-0.3	-	0.3V _{IO}	V
I _{IH}	HIGH-level input current	V _{TXD} ^[4] = V _{IO}	-5	-	+5	μA
I _{IL}	LOW-level input current	V _{TXD} = 0 V	-260	-150	-30	μA
C _i	input capacitance	^[5]	-	5	10	pF
CAN receive data output; pins RXD1 and RXD2						
I _{OH}	HIGH-level output current	V _{RXD} ^[6] = V _{IO} - 0.4 V; V _{IO} = V _{CC}	-8	-3	-1	mA
I _{OL}	LOW-level output current	V _{RXD} = 0.4 V; bus dominant	2	5	12	mA
Bus lines; pins CANH1, CANL1, CANH2 and CANL2						
V _{O(dom)}	dominant output voltage	V _{TXD} = 0 V; t < t _{to(dom)TXD}				
		pin CANH1/CANH2; R _L = 50 Ω to 60 Ω	2.75	3.5	4.5	V
		pin CANL1/CANL2; R _L = 50 Ω to 60 Ω	0.5	1.5	2.25	V
V _{dom(TX)sym}	transmitter dominant voltage symmetry	V _{dom(TX)sym} = V _{CC} - V _{CANH} ^[7] - V _{CANL} ^[8]	-300	-	+300	mV

Table 7. Static characteristics ...continued

$T_{vj} = -40\text{ }^{\circ}\text{C}$ to $+150\text{ }^{\circ}\text{C}$; $V_{CC} = 4.5\text{ V}$ to 5.5 V ; $V_{IO} = 2.8\text{ V}$ to 5.5 V ; $R_L = 60\text{ }\Omega$ unless specified otherwise; all voltages are defined with respect to ground; positive currents flow into the device^[1].

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{TXsym}	transmitter voltage symmetry	$V_{TXsym} = V_{CANH}$ ^[7] + V_{CANL} ^[8] ; $f_{TXD} = 250\text{ kHz}$; $C_{SPLIT} = 4.7\text{ nF}$ ^[9]	$0.9V_{CC}$	-	$1.1V_{CC}$	V
$V_{O(dif)}$	differential output voltage	dominant: Normal mode				
		$V_{TXD} = 0\text{ V}$; $t < t_{to(dom)TXD}$; $V_{CC} = 4.75\text{ V}$ to 5.25 V ; $R_L = 50\text{ }\Omega$ to $65\text{ }\Omega$	1.5	-	3	V
		$V_{TXD} = 0\text{ V}$; $t < t_{to(dom)TXD}$; $V_{CC} = 4.75\text{ V}$ to 5.25 V ; $R_L = 45\text{ }\Omega$ to $70\text{ }\Omega$	1.4	-	3.3	V
		$V_{TXD} = 0\text{ V}$; $t < t_{to(dom)TXD}$; $V_{CC} = 4.75\text{ V}$ to 5.25 V ; $R_L = 2240\text{ }\Omega$	1.5	-	5	V
		recessive				
		Normal mode: $V_{TXD} = V_{IO}$; no load	-50	-	+50	mV
		Standby mode	-0.2	-	+0.2	V
$V_{O(rec)}$	recessive output voltage	Normal mode; $V_{TXD} = V_{IO}$; no load	2	$0.5V_{CC}$	3	V
		Standby mode; no load	-0.1	-	+0.1	V
$V_{th(RX)dif}$	differential receiver threshold voltage	Normal mode; $-30\text{ V} \leq V_{CANL} \leq +30\text{ V}$; $-30\text{ V} \leq V_{CANH} \leq +30\text{ V}$	0.5	0.7	0.9	V
		Standby mode; $-12\text{ V} \leq V_{CANL} \leq +12\text{ V}$; $-12\text{ V} \leq V_{CANH} \leq +12\text{ V}$	0.4	0.7	1.15	V
$V_{rec(RX)}$	receiver recessive voltage	Normal mode; $-12\text{ V} \leq V_{CANL} \leq +12\text{ V}$; $-12\text{ V} \leq V_{CANH} \leq +12\text{ V}$	-3	-	0.5	V
$V_{dom(RX)}$	receiver dominant voltage	Normal mode; $-12\text{ V} \leq V_{CANL} \leq +12\text{ V}$; $-12\text{ V} \leq V_{CANH} \leq +12\text{ V}$	0.9	-	8.0	V
$V_{hys(RX)dif}$	differential receiver hysteresis voltage	Normal mode; $-30\text{ V} \leq V_{CANL} \leq +30\text{ V}$; $-30\text{ V} \leq V_{CANH} \leq +30\text{ V}$	50	120	200	mV
$I_{O(sc)dom}$	dominant short-circuit output current	$V_{TXD} = 0\text{ V}$; $t < t_{to(dom)TXD}$; $V_{CC} = 5\text{ V}$				
		pin CANH1/CANH2; $V_{CANH} = -3\text{ V}$ to $+40\text{ V}$	-100	-70	-40	mA
		pin CANL1/CANL2; $V_{CANL} = -3\text{ V}$ to $+40\text{ V}$	40	70	100	mA
$I_{O(sc)rec}$	recessive short-circuit output current	Normal mode; $V_{TXD} = V_{IO}$; $V_{CANH} = V_{CANL} = -40\text{ V}$ to $+40\text{ V}$	-5	-	+5	mA
I_L	leakage current	$V_{CC} = V_{IO} = 0\text{ V}$ or $V_{CC} = V_{IO} = \text{shorted to ground via } 47\text{ k}\Omega$; $V_{CANH} = V_{CANL} = 5\text{ V}$	-5	-	+5	μA
R_i	input resistance		9	15	28	k Ω
ΔR_i	input resistance deviation	between pin CANH1/CANH2 and pin; CANL1/CANL2	-1	-	+1	%

Table 7. Static characteristics ...continued

$T_{vj} = -40\text{ }^{\circ}\text{C}$ to $+150\text{ }^{\circ}\text{C}$; $V_{CC} = 4.5\text{ V}$ to 5.5 V ; $V_{IO} = 2.8\text{ V}$ to 5.5 V ; $R_L = 60\text{ }\Omega$ unless specified otherwise; all voltages are defined with respect to ground; positive currents flow into the device^[1].

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_{i(dif)}$	differential input resistance		19	30	52	k Ω
$C_{i(cm)}$	common-mode input capacitance	[5]	-	-	20	pF
$C_{i(dif)}$	differential input capacitance	[5]	-	-	10	pF
Temperature detection						
$T_{j(sd)}$	shutdown junction temperature	[5]	-	190	-	$^{\circ}\text{C}$

- [1] All parameters are guaranteed over the virtual junction temperature range by design. Factory testing uses correlated test conditions to cover the specified temperature and power supply voltage range.
- [2] Total supply current ($I_{CC} + I_{IO}$) in Standby mode is typically 17 μA , with a maximum value of 26 μA .
- [3] STBN refers to the input signal on pin STBN1 or pin STBN2.
- [4] TXD refers to the input signal on pin TXD1 or pin TXD2.
- [5] Not tested in production; guaranteed by design.
- [6] RXD refers to the output signal on pin RXD1 or pin RXD2.
- [7] CANH refers to the input/output signal on pin CANH1 or pin CANH2.
- [8] CANL refers to the input/output signal on pin CANL1 or pin CANL2.
- [9] The test circuit used to measure the bus output voltage symmetry (which includes CSPLIT) is shown in [Figure 9](#).

11. Dynamic characteristics

Table 8. Dynamic characteristics

$T_{vj} = -40\text{ }^{\circ}\text{C}$ to $+150\text{ }^{\circ}\text{C}$; $V_{CC} = 4.5\text{ V}$ to 5.5 V ; $V_{IO} = 2.8\text{ V}$ to 5.5 V ; $R_L = 60\text{ }\Omega$ unless specified otherwise; all voltages are defined with respect to ground; positive currents flow into the device^[1].

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Transceiver timing; pins CANH1, CANH2, CANL1, CANL2, TXD1, TXD2, RXD1 and RXD2; see Figure 8 and Figure 5						
$t_{d(\text{TXD-busdom})}$	delay time from TXD to bus dominant	Normal mode	-	65	-	ns
$t_{d(\text{TXD-busrec})}$	delay time from TXD to bus recessive	Normal mode	-	90	-	ns
$t_{d(\text{busdom-RXD})}$	delay time from bus dominant to RXD	Normal mode	-	60	-	ns
$t_{d(\text{busrec-RXD})}$	delay time from bus recessive to RXD	Normal mode	-	65	-	ns
$t_{d(\text{TXDL-RXDL})}$	delay time from TXD LOW to RXD LOW	Normal mode	60	-	250	ns
$t_{d(\text{TXDH-RXDH})}$	delay time from TXD HIGH to RXD HIGH	Normal mode	60	-	250	ns
$t_{\text{bit}(\text{bus})}$	transmitted recessive bit width	$t_{\text{bit}(\text{TXD})} = 500\text{ ns}$	[2] 435	-	530	ns
		$t_{\text{bit}(\text{TXD})} = 200\text{ ns}$	[2] 155	-	210	ns
$t_{\text{bit}(\text{RXD})}$	bit time on pin RXD	$t_{\text{bit}(\text{TXD})} = 500\text{ ns}$	[2] 400	-	550	ns
		$t_{\text{bit}(\text{TXD})} = 200\text{ ns}$	[2] 120	-	220	ns
Δt_{rec}	receiver timing symmetry	$t_{\text{bit}(\text{TXD})} = 500\text{ ns}$	-65	-	+40	ns
		$t_{\text{bit}(\text{TXD})} = 200\text{ ns}$	-45	-	+15	ns
$t_{\text{to}(\text{dom})\text{TXD}}$	TXD dominant time-out time	$V_{\text{TXD}} = 0\text{ V}$; Normal mode	0.5	2	5	ms
$t_{d(\text{stb-norm})}$	standby to normal mode delay time		7	25	47	μs
$t_{\text{wake}(\text{busdom})}$	bus dominant wake-up time	Standby mode	0.5	-	5	μs
$t_{\text{wake}(\text{busrec})}$	bus recessive wake-up time	Standby mode	0.5	-	5	μs
$t_{\text{to}(\text{wake})\text{bus}}$	bus wake-up time-out time		0.5	2	5	ms
$t_{\text{fltr}(\text{wake})\text{bus}}$	bus wake-up filter time	Standby mode	0.5	1.5	5	μs

[1] All parameters are guaranteed over the virtual junction temperature range by design. Factory testing uses correlated test conditions to cover the specified temperature and power supply voltage range.

[2] See [Figure 6](#).

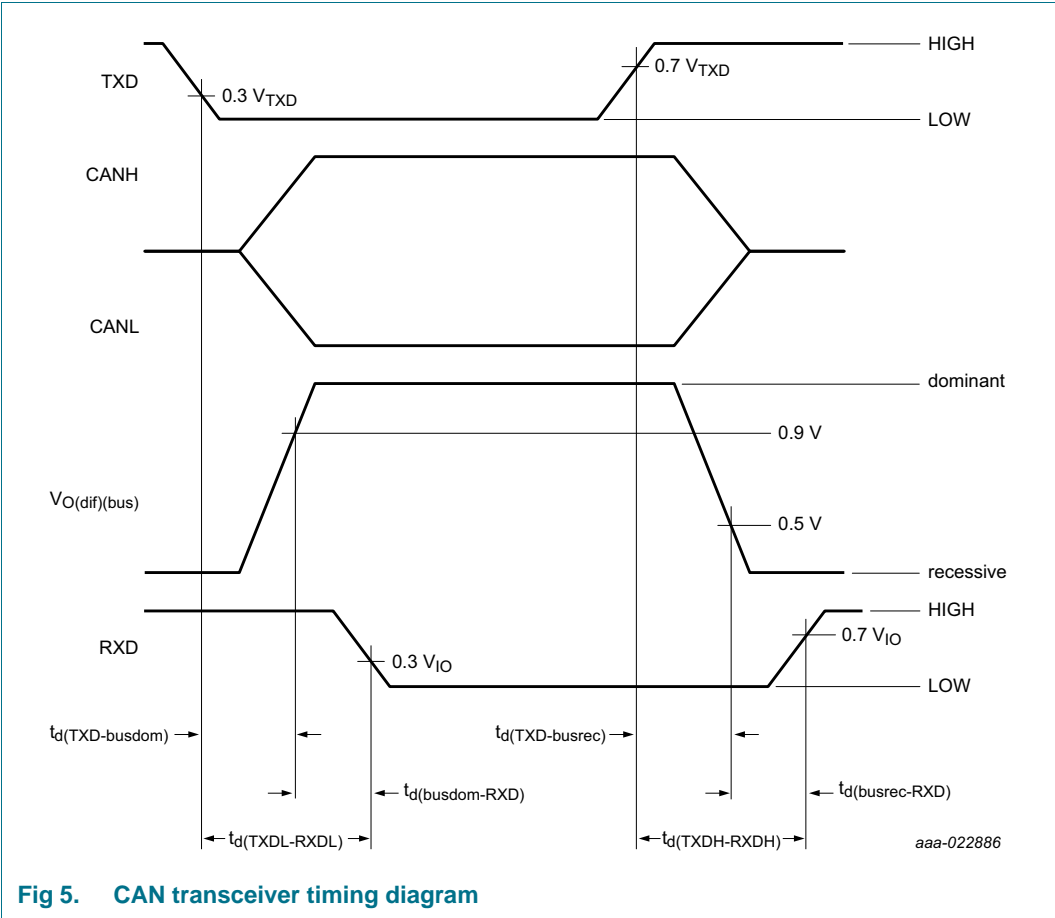


Fig 5. CAN transceiver timing diagram

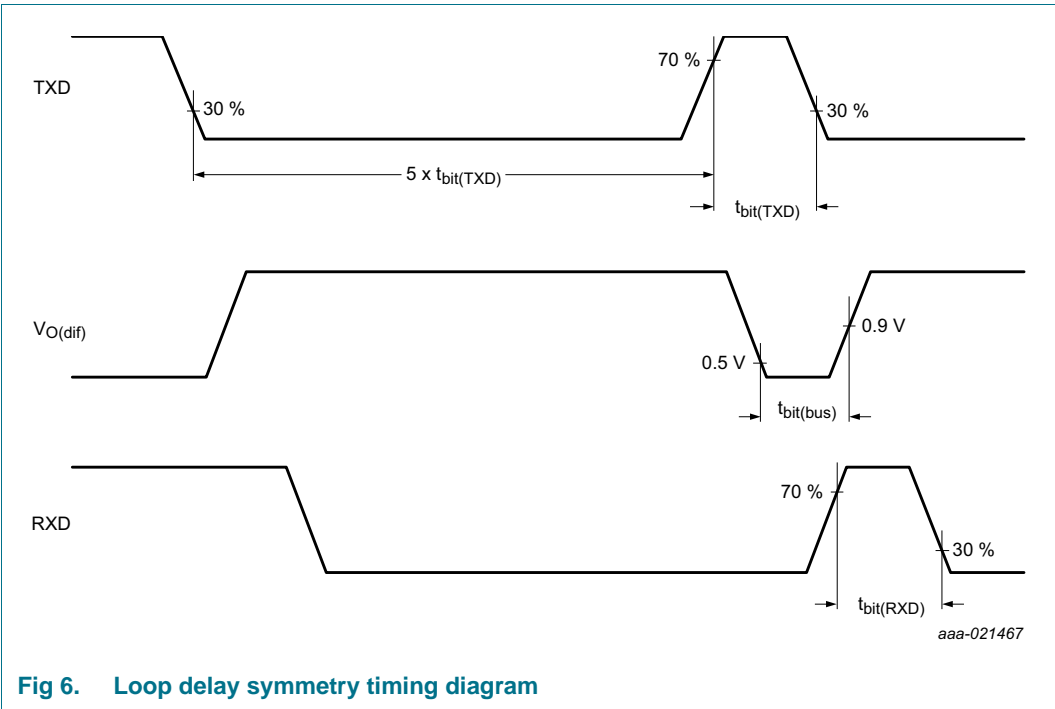
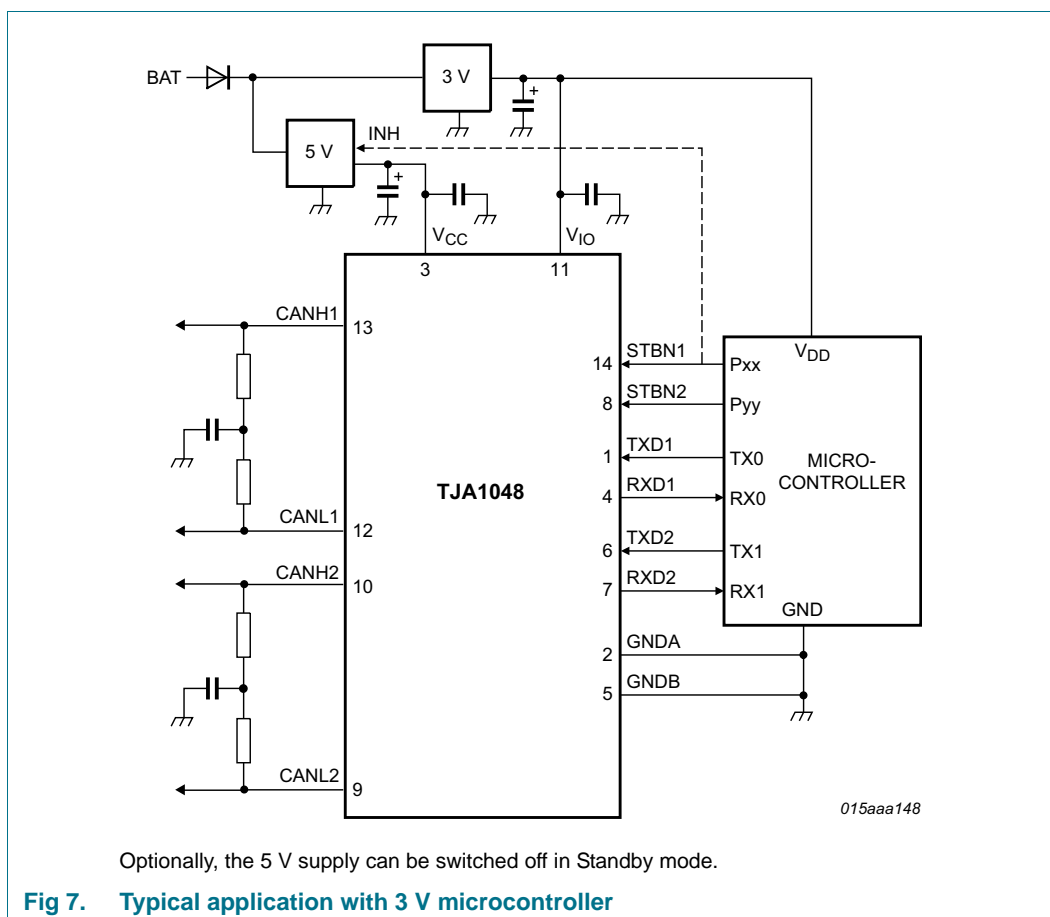


Fig 6. Loop delay symmetry timing diagram

12. Application information

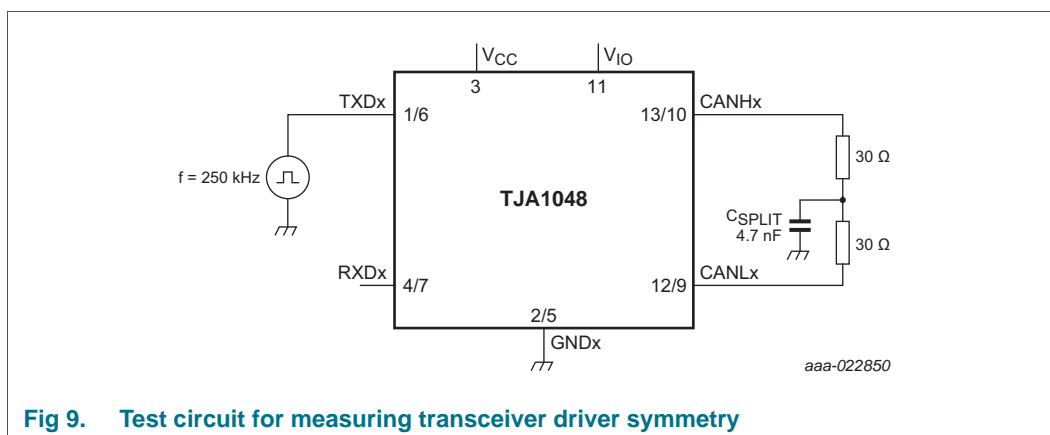
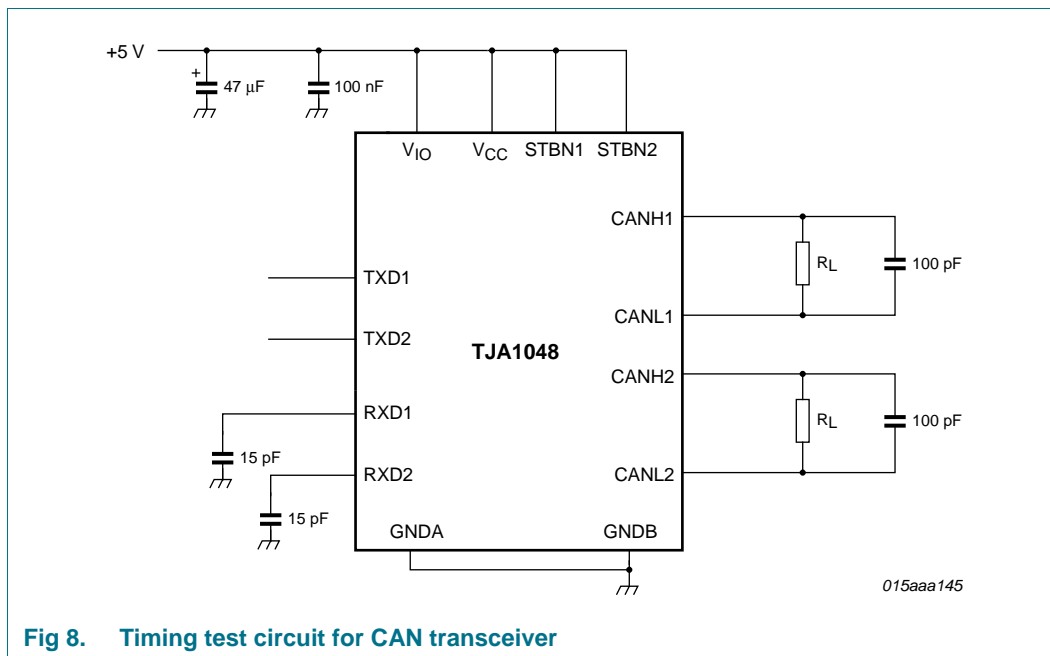
12.1 Application diagram



12.2 Application hints

Further information on the application of the TJA1048 can be found in NXP application hints *AH1014 Application Hints - Standalone high speed CAN transceiver TJA1042/TJA1043/TJA1048/TJA1051*.

13. Test information



13.1 Quality information

This product has been qualified in accordance with the Automotive Electronics Council (AEC) standard Q100 Rev-G - *Failure mechanism based stress test qualification for integrated circuits*, and is suitable for use in automotive applications.

14. Package outline

SO14: plastic small outline package; 14 leads; body width 3.9 mm

SOT108-1

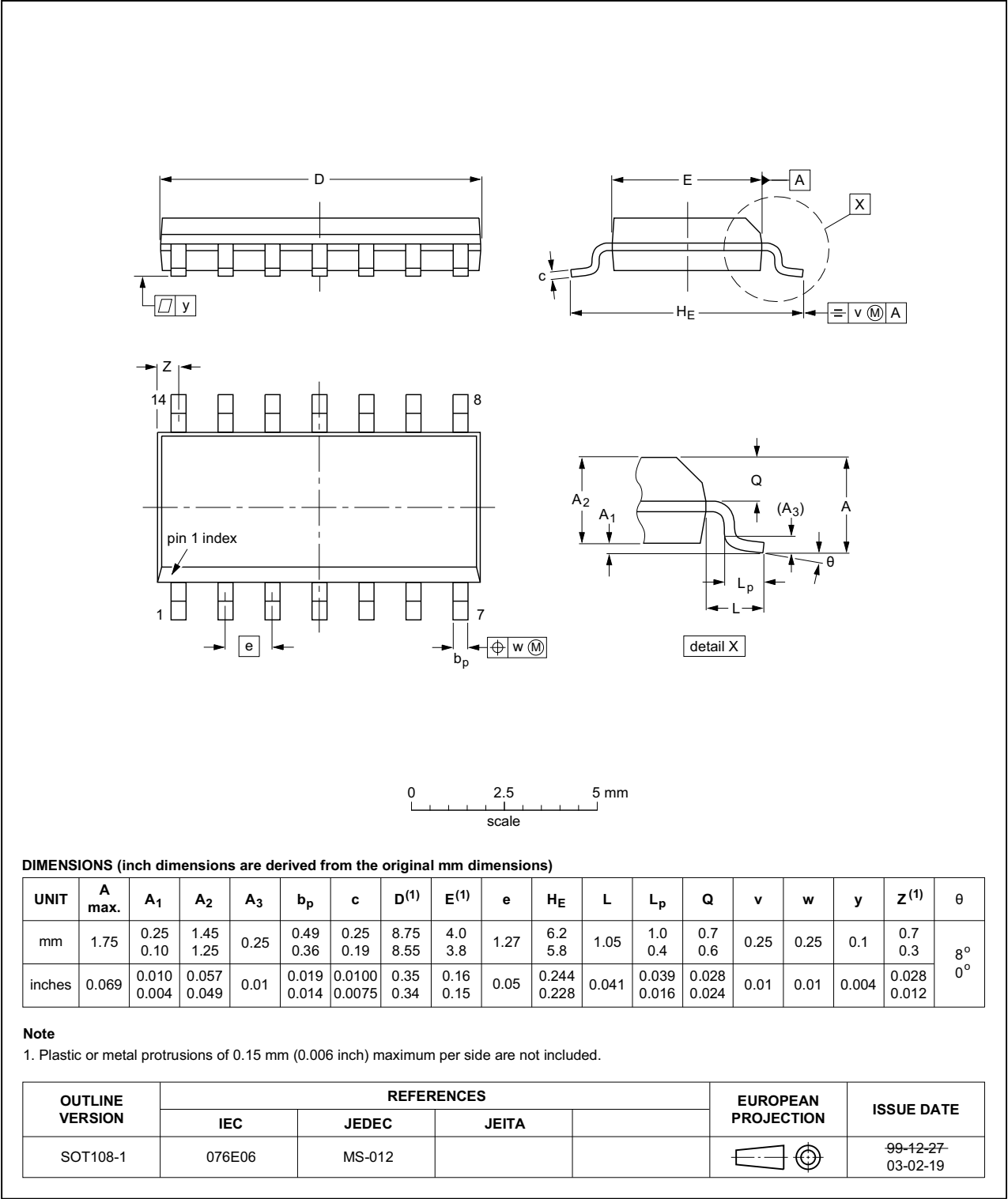


Fig 10. Package outline SOT108 (SO14)

HVSON14: plastic, thermal enhanced very thin small outline package; no leads;
 14 terminals; body 3 x 4.5 x 0.85 mm

SOT1086-2

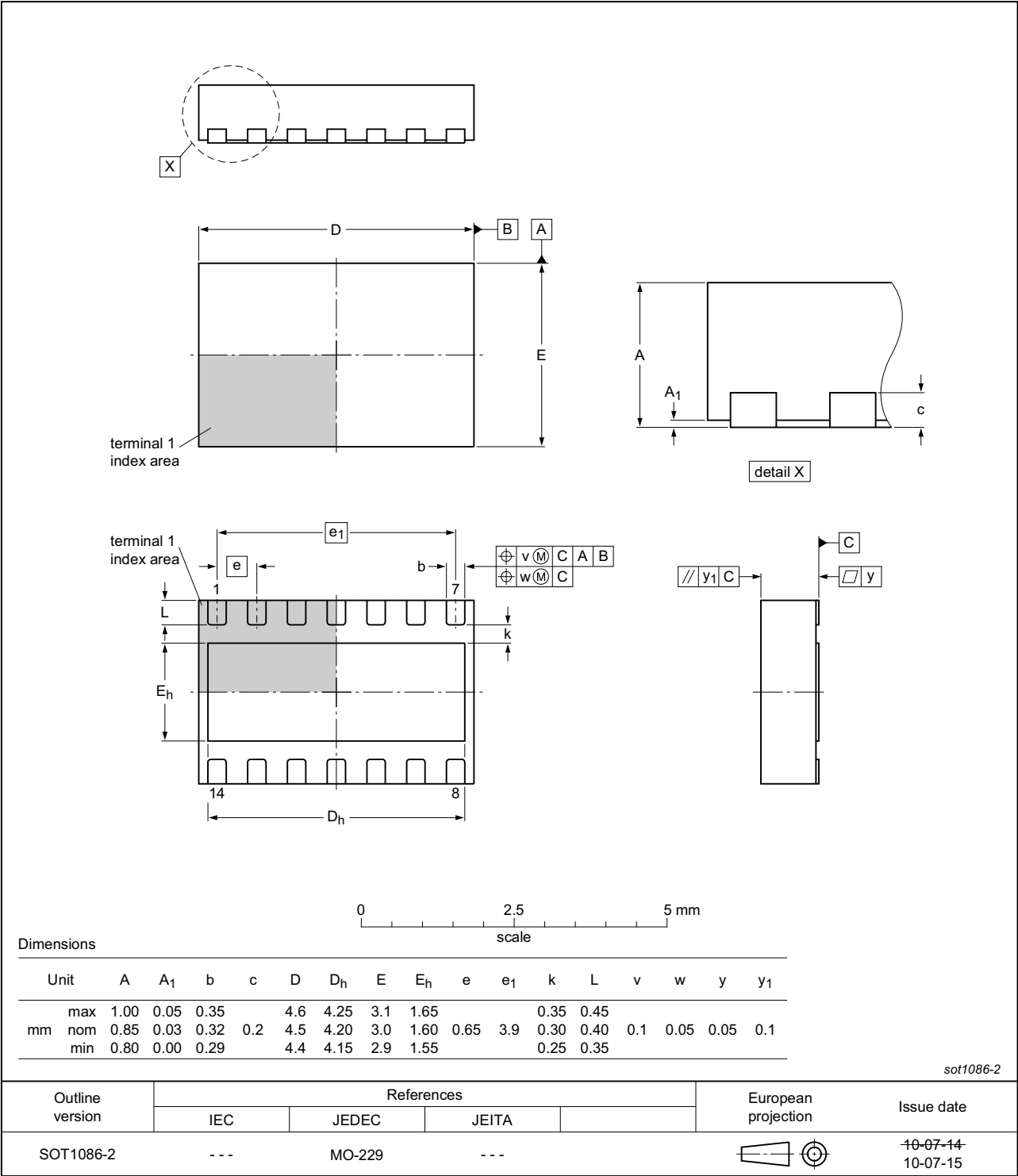


Fig 11. Package outline SOT1086 (HVSON14)

15. Handling information

All input and output pins are protected against ElectroStatic Discharge (ESD) under normal handling. When handling ensure that the appropriate precautions are taken as described in *JESD625-A* or equivalent standards.

16. Soldering of SMD packages

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note *AN10365 "Surface mount reflow soldering description"*.

16.1 Introduction to soldering

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

16.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- Through-hole components
- Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- Board specifications, including the board finish, solder masks and vias
- Package footprints, including solder thieves and orientation
- The moisture sensitivity level of the packages
- Package placement
- Inspection and repair
- Lead-free soldering versus SnPb soldering

16.3 Wave soldering

Key characteristics in wave soldering are:

- Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- Solder bath specifications, including temperature and impurities

16.4 Reflow soldering

Key characteristics in reflow soldering are:

- Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see [Figure 12](#)) than a SnPb process, thus reducing the process window
- Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with [Table 9](#) and [10](#)

Table 9. SnPb eutectic process (from J-STD-020D)

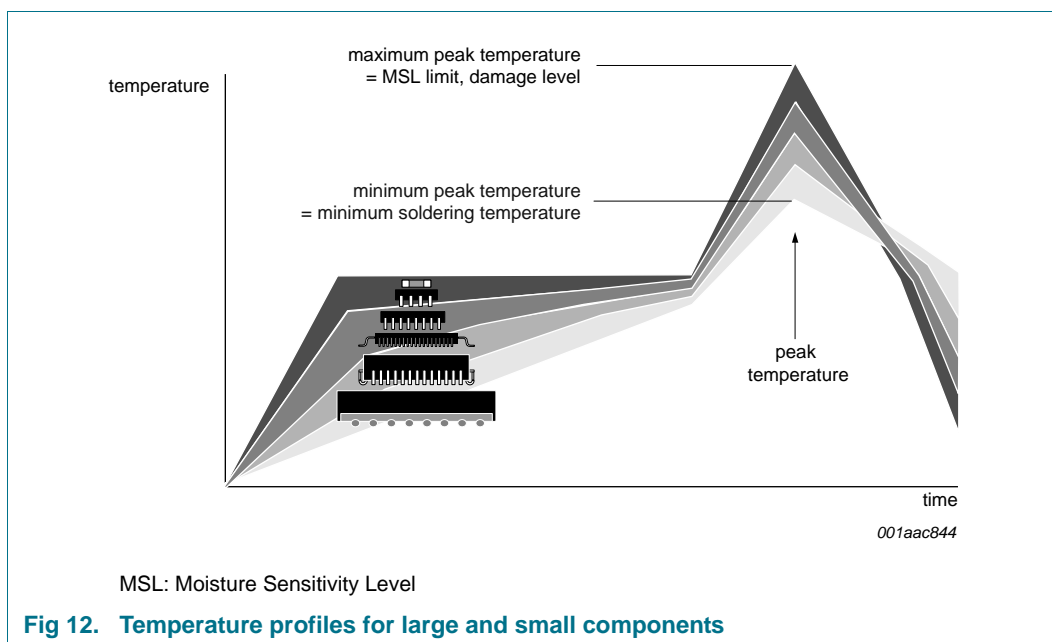
Package thickness (mm)	Package reflow temperature (°C)	
	Volume (mm ³)	
	< 350	≥ 350
< 2.5	235	220
≥ 2.5	220	220

Table 10. Lead-free process (from J-STD-020D)

Package thickness (mm)	Package reflow temperature (°C)		
	Volume (mm ³)		
	< 350	350 to 2000	> 2000
< 1.6	260	260	260
1.6 to 2.5	260	250	245
> 2.5	250	245	245

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see [Figure 12](#).



For further information on temperature profiles, refer to Application Note *AN10365* “Surface mount reflow soldering description”.

17. Soldering of HVSON packages

[Section 17](#) contains a brief introduction to the techniques most commonly used to solder Surface Mounted Devices (SMD). A more detailed discussion on soldering HVSON leadless package ICs can be found in the following application notes:

- *AN10365* “Surface mount reflow soldering description”
- *AN10366* “HVQFN application information” [Section 16](#)

18. Appendix: ISO 11898-2:2016 parameter cross-reference list

Table 11. ISO 11898-2:2016 to NXP data sheet parameter conversion

ISO 11898-2:2016		NXP data sheet	
Parameter	Notation	Symbol	Parameter
HS-PMA dominant output characteristics			
Single ended voltage on CAN_H	V _{CAN_H}	V _{O(dom)}	dominant output voltage
Single ended voltage on CAN_L	V _{CAN_L}		
Differential voltage on normal bus load	V _{Diff}	V _{O(dif)}	differential output voltage
Differential voltage on effective resistance during arbitration			
Optional: Differential voltage on extended bus load range			
HS-PMA driver symmetry			
Driver symmetry	V _{SYM}	V _{TXsym}	transmitter voltage symmetry
Maximum HS-PMA driver output current			
Absolute current on CAN_H	I _{CAN_H}	I _{O(sc)dom}	dominant short-circuit output current
Absolute current on CAN_L	I _{CAN_L}		
HS-PMA recessive output characteristics, bus biasing active/inactive			
Single ended output voltage on CAN_H	V _{CAN_H}	V _{O(rec)}	recessive output voltage
Single ended output voltage on CAN_L	V _{CAN_L}		
Differential output voltage	V _{Diff}	V _{O(dif)}	differential output voltage
Optional HS-PMA transmit dominant timeout			
Transmit dominant timeout, long	t _{dom}	t _{to(dom)TXD}	TXD dominant time-out time
Transmit dominant timeout, short			
HS-PMA static receiver input characteristics, bus biasing active/inactive			
Recessive state differential input voltage range	V _{Diff}	V _{th(RX)dif}	differential receiver threshold voltage
Dominant state differential input voltage range		V _{rec(RX)}	receiver recessive voltage
		V _{dom(RX)}	receiver dominant voltage
HS-PMA receiver input resistance (matching)			
Differential internal resistance	R _{Diff}	R _{i(dif)}	differential input resistance
Single ended internal resistance	R _{CAN_H} R _{CAN_L}	R _i	input resistance
Matching of internal resistance	MR	ΔR _i	input resistance deviation
HS-PMA implementation loop delay requirement			
Loop delay	t _{Loop}	t _{d(TXDH-RXDH)}	delay time from TXD HIGH to RXD HIGH
		t _{d(TXDL-RXDL)}	delay time from TXD LOW to RXD LOW
Optional HS-PMA implementation data signal timing requirements for use with bit rates above 1 Mbit/s up to 2 Mbit/s and above 2 Mbit/s up to 5 Mbit/s			
Transmitted recessive bit width @ 2 Mbit/s / @ 5 Mbit/s, intended	t _{Bit(Bus)}	t _{bit(bus)}	transmitted recessive bit width
Received recessive bit width @ 2 Mbit/s / @ 5 Mbit/s	t _{Bit(RXD)}	t _{bit(RXD)}	bit time on pin RXD
Receiver timing symmetry @ 2 Mbit/s / @ 5 Mbit/s	Δt _{Rec}	Δt _{rec}	receiver timing symmetry

Table 11. ISO 11898-2:2016 to NXP data sheet parameter conversion

ISO 11898-2:2016		NXP data sheet	
Parameter	Notation	Symbol	Parameter
HS-PMA maximum ratings of V _{CAN_H} , V _{CAN_L} and V _{Diff}			
Maximum rating V _{Diff}	V _{Diff}	V _(CANH-CANL)	voltage between pin CANH and pin CANL
General maximum rating V _{CAN_H} and V _{CAN_L}	V _{CAN_H}	V _x	voltage on pin x
Optional: Extended maximum rating V _{CAN_H} and V _{CAN_L}	V _{CAN_L}		
HS-PMA maximum leakage currents on CAN_H and CAN_L, unpowered			
Leakage current on CAN_H, CAN_L	I _{CAN_H} I _{CAN_L}	I _L	leakage current
HS-PMA bus biasing control timings			
CAN activity filter time, long	t _{Filter}	t _{wake(busdom)} ^[1]	bus dominant wake-up time
CAN activity filter time, short		t _{wake(busrec)} ^[1]	bus recessive wake-up time
Wake-up timeout, short	t _{Wake}	t _{to(wake)bus}	bus wake-up time-out time
Wake-up timeout, long			
Timeout for bus inactivity	t _{Silence}	t _{to(silence)}	bus silence time-out time
Bus Bias reaction time	t _{Bias}	t _{d(busact-bias)}	delay time from bus active to bias

[1] $t_{filtr(wake)bus}$ - bus wake-up filter time, in devices with basic wake-up functionality

19. Revision history

Table 12. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
TJA1048 v.5	20160523	Product data sheet	-	TJA1048 v.4
Modifications	<ul style="list-style-type: none"> • Table 3: Table note 1: text revised • Section 7.1.3 added • Figure 4 changed • Table 5: <ul style="list-style-type: none"> – Table note 1 added for parameter V_x – new parameter added: ($V_{CANH-CANL}$) – parameter V_{trt} reformatted • Section 12.1: Figure 7 modified • ISO 11898-2:2016 compliance <ul style="list-style-type: none"> – Section 1: text revised (3rd paragraph) – Section 2.1 third list item revised – Table 5: Table note 1 added for parameter V_x – Table 7: Parameter $V_{TX(sym)}$ added; Table note 9 added – Table 7: Parameters for $V_{O(dom)}$ modified – Table 7: Parameter $V_{O(dif)bus}$ changed to $V_{O(dif)}$; Conditions revised – Table 7: Parameter $V_{th(RX)dif}$ Conditions revised – Table 7: Parameter $V_{rec(RX)}$ added – Table 7: Parameter $V_{dom(RX)}$ added – Table 7: Parameter $V_{hys(RX)dif}$ Conditions revised – Table 7: Parameter $I_{O(sc)dom}$ Conditions revised – Table 7: Parameter I_L conditions revised – Table 7: Previous note 9 deleted – Section 11: Figure 5 and Figure 6 replaced – Table 8: parameters $t_{bit(bus)}$ and Δt_{rec} added – Table 8: parameter $t_{PD(RDX-RXD)}$ replaced by $t_d(TXDL-RXDL)$ and $t_d(TXDH-RXDH)$ – Table 8: additional condition and specification values added to parameter $t_{bit(RXD)}$ – Section 13: Figure 9 “Test circuit for measuring transceiver driver symmetry” added – Section 18 “Appendix: ISO 11898-2:2016 parameter cross-reference list” added 			
TJA1048 v.4	20150115	Product data sheet	-	TJA1048 v.3
TJA1048 v.3	20130424	Product data sheet	-	TJA1048 v.2
TJA1048 v.2	20110325	Product data sheet	-	TJA1048 v.1
TJA1048 v.1	20101103	Product data sheet	-	-

20. Legal information

20.1 Data sheet status

Document status ^{[1][2]}	Product status ^[3]	Definition
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Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

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22. Contents

1	General description	1	20.1	Data sheet status	25
2	Features and benefits	1	20.2	Definitions	25
2.1	General	1	20.3	Disclaimers	25
2.2	Low-power management	2	20.4	Trademarks	26
2.3	Protection	2	21	Contact information	26
3	Quick reference data	2	22	Contents	27
4	Ordering information	3			
5	Block diagram	4			
6	Pinning information	5			
6.1	Pinning	5			
6.2	Pin description	5			
7	Functional description	6			
7.1	Operating modes	6			
7.1.1	Normal mode	6			
7.1.2	Standby mode	6			
7.1.3	Remote wake-up (via the CAN bus)	6			
7.2	Fail-safe features	7			
7.2.1	TXD dominant time-out function	7			
7.2.2	Internal biasing of TXD1, TXD2, STBN1 and STBN2 input pins	8			
7.2.3	Undervoltage detection on pins V_{CC} and V_{IO}	8			
7.2.4	Overtemperature protection	8			
7.3	V_{IO} supply pin	8			
8	Limiting values	9			
9	Thermal characteristics	9			
10	Static characteristics	10			
11	Dynamic characteristics	13			
12	Application information	15			
12.1	Application diagram	15			
12.2	Application hints	15			
13	Test information	16			
13.1	Quality information	16			
14	Package outline	17			
15	Handling information	19			
16	Soldering of SMD packages	19			
16.1	Introduction to soldering	19			
16.2	Wave and reflow soldering	19			
16.3	Wave soldering	19			
16.4	Reflow soldering	20			
17	Soldering of HVSON packages	21			
18	Appendix: ISO 11898-2:2016 parameter cross-reference list	22			
19	Revision history	24			
20	Legal information	25			

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