

ML65244/ML65L244

High Speed Dual Quad Buffer/Line Drivers

GENERAL DESCRIPTION

The ML65244 and ML65L244 are non-inverting dual quad buffer/line drivers. The high operating frequency (50MHz driving a 50pF load) and low propagation delay (ML65244 – 1.7ns, ML65L244 – 2ns) make them ideal for very high speed applications such as processor bus buffering and cache and main memory control.

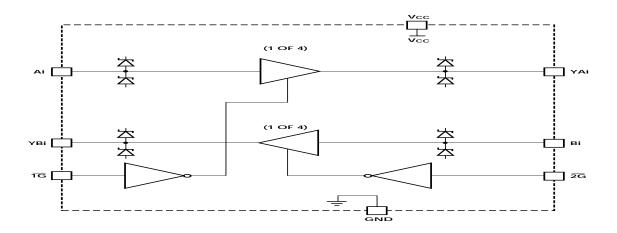
These buffers use a unique analog implementation to eliminate the delays inherent in traditional digital designs. Schottky clamps reduce under and overshoot, and special output driver circuits limit ground bounce. The ML65244 and ML65L244 conform to the pinout and functionality of the industry standard FCT244 and are intended for applications where propagation delay is critical to the system design.

Note: This part was previously numbered ML6582.

FEATURES

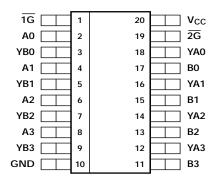
- Low propagation delay 1.7ns ML65244 2.0ns ML65L244
- Fast Dual 4-bit TTL level buffer/line driver with tristate capability on the output (two 4-bit sections)
- TTL compatible input and output levels
- Schottky diode clamps on all inputs to handle undershoot and overshoot
- Onboard schottky diodes minimize noise
- \blacksquare Reduced output swing of 0 4.1 volts
- Ground bounce controlled outputs, typically less than 400mV
- Industry standard FCT244 type pinout
- Applications include high speed cache memory, main memory, processor bus buffering, and graphics cards

BLOCK DIAGRAM



PIN CONFIGURATION

20-Pin SOIC, QSOP



TOP VIEW

PIN DESCRIPTION

NAME	I/O	DESCRIPTION
Ai	I	Data Bus A
YAi	Ο	Data Bus A
Bi	1	Data Bus B
YBi	Ο	Data Bus B
1 G	I	Output Enable for data bus A
2 G	1	Output Enable for data bus B
GND	1	Signal Ground
V_{CC}	1	+ 5V supply

FUNCTION TABLE

1G/2G	Ai/Bi	YAi/YBi	
Н	X	Z	
L	L	L	
L	Н	Н	

L = Logic Low

ABSOLUTE MAXIMUM RATINGS

V _{CC}	–0.3V to 7V
DC Input voltage	0.3 to $V_{CC} + 0.3V$
AC Input voltage (< 20ns)	
DC Output voltage	
Output sink current (per pin)	120mA
Storage temperature	
Junction temperature	
Thermal Impedance (θ_{IA})	
SOIC	96°C/W
QSOP	100°C/W

H = Logic High

X = Don't Care

Z = High Impedance

ELECTRICAL CHARACTERISTICS

Unless otherwise stated, these specifications apply for: $V_{CC} = 5.0 \pm 5\% V$, $T_A = 0^{\circ}C$ to $70^{\circ}C$ (Note 1).

SYMBOL	PARAMETER	ARAMETER CONDITIONS		MIN	TYP	MAX	UNITS
AC ELECTR	ICAL CHARACTERISTICS ($C_{LOAD} = 50$ pF, $R_{LOAD} = 500$ ý)					
t _{PLH} , t _{PHL}	Propagation delay	Ai to YAi, Bi to YBi (Note 2)	ML65244		1.4	1.7	ns
			ML65L244		1.6	2.0	ns
t _{OE}	Output enable time 1\overline{G}, 2\overline{G} to YAi/YBi				10	15	ns
t _{OD}	Output disable time 1\overline{G}, 2\overline{G} to YAi/YBi					10	ns
C _{IN}	Input capacitance				8		pF
DC ELECTR	RICAL CHARACTERISTICS ($(C_{LOAD} = 50pF, R_{LOAD} =)$					
V_{IH}	Input high voltage	Logic HIGH		2.0			V
V _{IL}	Input low voltage	Logic LOW				0.8	V
I _{IH}	Input high current	Per pin, V _{IN} = 3V	ML65244		0.5	1.5	mA
			ML65L244		0.3	0.5	mA
I _{IL}	Input low current	Per pin, V _{IN} = 0	ML65244		2.4	3.5	mA
			ML65L244		0.8	1.0	mA
I _{HI-Z}	Three-state output current	$V_{CC} = 5.25V, 0 < V_{IN} < V_{CC}$				5	μΑ
I _{OS}	Short circuit current	$V_{CC} = 5.25V, V_{O} = GND$ (Note 3)		-60		-225	mA
V _{IC}	Input clamp voltage	V _{CC} = 4.75V, I _{IN} = 18mA			-0.7	-1.2	V
V _{OH}	Output high voltage	$V_{CC} = 4.75V$, $I_{OH} = 100\mu A$ (Notes 4 & 5)		2.4			V
V_{OL}	Output low voltage	V _{CC} = 4.75V, I _{OL} = 25mA (Notes 4 & 5)				0.6	V
V _{OFF}	V _{IN} – V _{OUT} per buffer	V _{CC} = 4.75V (Note 4)	ML65244	0	100	200	mV
			ML65L244	0	200	300	mV
I _{CC}	Quiescent Power Supply Current	V _{CC} = 5.25V, Freq = 0Hz, Inputs/outputs open			55	80	mA

Note 1: Limits are guaranteed by 100% testing, sampling or correlation with worst case test conditions.

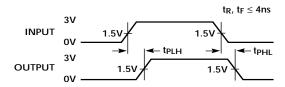
Note 2: One line switching, see Figure 3, t_{PLH}, t_{PHL} versus C_L.

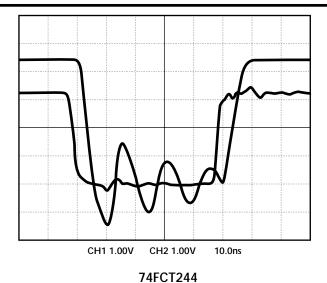
Note 3: Not more than one output should be shorted for more than a second.

Note 4: This is a true analog buffer. In the linear region, the output tracks the input with an offset (V_{OFF}). For V_{OH} , $V_{IN} = 2.7V$.

 $V_{OH\ MIN}$ includes V_{OFF} . For V_{OL} , V_{IN} = 0V, $V_{OL\ MAX}$ includes V_{OFF}

Note 5: See Figure 2 for I_{OH} versus V_{OH} and I_{OL} versus V_{OL} data.





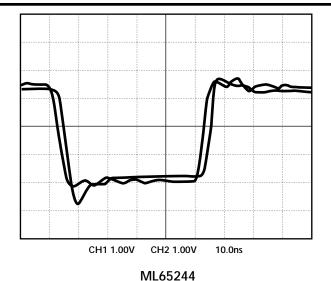
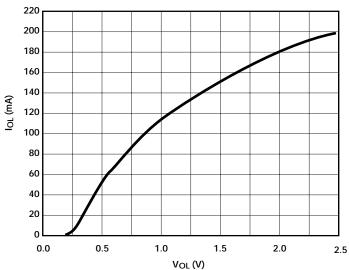


Figure 1. Ground Bounce Comparison, Four Outputs Switching into 50pF Loads.



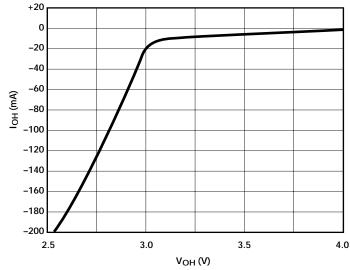
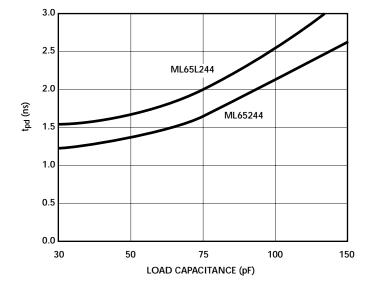


Figure 2a. Typical V_{OL} Versus I_{OL} for One Buffer Output.

Figure 2b. Typical V_{OH} Versus I_{OH} for One Buffer Output.



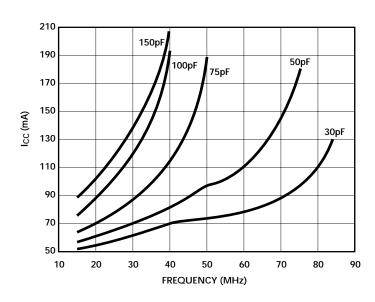


Figure 3. Propagation Delay (t_{PLH}, t_{PHL}) Versus Load Capacitance, One Output Switching.

Figure 4. I_{CC} Versus Frequency for Various Load Capacitances, Four Outputs Switching.

FUNCTIONAL DESCRIPTION

The ML65244 and ML65L244 are very high speed noninverting buffer/line drivers with three-state outputs which are ideally suited for bus-oriented applications. They provide a low propagation delay by using an analog design approach (a high speed unity gain buffer), as compared to conventional digital approaches. The ML65244 and ML65L244 follow the pinout and functionality of the industry standard FCT244 series of buffer/line drivers and are intended to replace them in designs where the propagation delay is a critical part of the system design considerations. The ML65244 and ML65L244 are capable of driving load capacitances several times larger than their input capacitance. They are configured so that the Ai inputs go to the YAi outputs, with the A side output enable controlled by $1\overline{G}$. Similarly, **2G** controls the Bi inputs which go to the YBi outputs.

These unity gain analog buffers achieve low propagation delays by having the output follow the input with a small offset. The output rise and fall times will closely match those of the input waveform. All inputs and outputs have Schottky clamp diodes to handle undershoot or overshoot noise suppression in unterminated applications. All outputs have ground bounce suppression (typically < 400mV), high drive output capability with almost immediate response to the input signal, and low output skew.

The I_{OL} current drive capability of a buffer/line driver is often interpreted as a measure of its ability to sink

current in a dynamic sense. This may be true for CMOS buffer/line drivers, but it is not true for the ML65244 and ML65L244. This is because their sink and source current capability depends on the voltage difference between the output and the input. The ML65244 can sink or source more than 100mA to a load when the load is switching due to the fact that during the transition, the difference between the input and output is large. I_{OL} is only significant as a DC specification, and is 25mA.

ARCHITECTURAL DESCRIPTION

Until now, buffer/line drivers have been implemented in CMOS logic and made to be TTL compatible by sizing the input devices appropriately. In order to buffer large capacitances with CMOS logic, it is necessary to cascade an even number of inverters, each successive inverter larger than the preceding, eventually leading to an inverter that will drive the required load capacitance at the required frequency. Each inverter stage represents an additional delay in the gating process because in order for a single gate to switch, the input must slew more than half of the supply voltage. The best of these CMOS buffers has managed to drive a 50pF load capacitance with a delay of 3.2ns. Micro Linear has produced a dual quad buffer/line driver with a delay less than 1.7ns by using a unique circuit architecture that does not require cascaded logic gates. The ML65244 uses a feedback technique to produce an output that follows the input. If the output voltage is not close to the input, then the feedback circuitry will source or sink enough current to the load capacitance to correct the discrepancy.

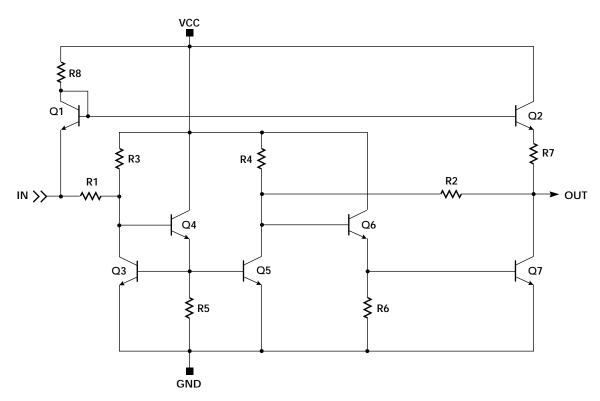


Figure 5. One buffer cell of the ML65244

REV. 1.0 10/25/2000 5

ML65244/ML65L244

The basic architecture of the ML65244 is shown in Figure 5. It is implemented on a $1.5\mu m$ BiCMOS process. However, in this particular circuit, all of the active devices are NPNs — the fastest devices available in the process.

In this circuit, there are two paths to the output. One path sources current to the load capacitance when the signal is asserted, and the other path sinks current from the output when the signal is negated.

The assertion path is the emitter follower path consisting of the level shift transistor Q1, the output transistor Q2, and the bias resistor R8. It sources current to the output through the 75ý resistor R7 which is bypassed by another NPN (not shown) during fast input transients. The negation path is a current differencing op amp connected in a follower configuration. The active components in this amplifier are transistors Q3–Q7. R3–R6 are bias resistors, and R1 and R2 are the feedback resistors. The key to understanding the operation of the current differencing op amp is to know that the currents in transistors Q3 and Q5 are the same at all times and that the voltages at the bases of Q4 and Q6 are roughly the same. If the output is higher than the input, then an error current will flow through R2. This error current will flow into the base of Q6 and be multiplied by β squared to the collector of Q7, closing the loop. The larger the discrepancy between the output and input, the larger the feedback current, and the harder Q7 sinks current from the load capacitor.

A number of MOSFETs are not shown in Figure 5. These MOSFETs are used to three-state dormant buffers. For instance, the feedback resistors R1 and R2 were implemented as resistive transmission gates to ensure that disabled buffers do not load the lines they are connected to. Similarly, there is a PMOS in series with R8 that is normally on but shuts off for disable. Other MOSFETs have been included to ensure that disabled

buffers consume no power.

TERMINATION

R7 in Figure 5 also acts as a termination resistor. This 75ý resistor is in series with the output and therefore helps suppress noise caused by transmission line effects such as reflections from mismatched impedances. System designers using CMOS transceivers commonly have to use external resistors in series with each transceiver output to suppress this noise. Systems using the ML65244 or ML65L244 may not have to use these external resistors.

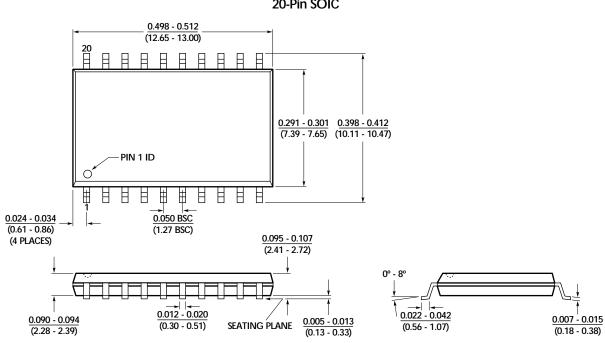
APPLICATIONS

There are a wide variety of needs for extremely fast buffers in high speed processor system designs like Pentium, PowerPC, Mips, Sparc, Alpha and other RISC processors. These applications are either in the cache memory area or the main memory (DRAM) area. In addition, fast buffers find applications in high speed graphics and multimedia applications. The high capacitive loading due to multiplexed address lines on the system bus demand external buffers to take up the excess drive current. The needed current to skew the transitions between rise and fall times must be done without adding excessive propagation delay. The ML65244 and ML65L244 are equipped with Schottky diodes to clean up ringing from overshoot and undershoot caused by reflections in unterminated board traces.

PHYSICAL DIMENSIONS inches (millimeters)

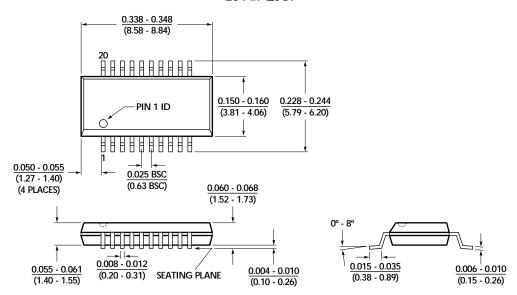
Package: \$20W 20-Pin SOIC





Package: K20 20-Pin QSOP

Package: K20 20-Pin QSOP



REV. 1.0 10/25/2000 7

ORDERING INFORMATION

PART NUMBER	SPEED	TEMPERATURE RANGE	PACKAGE
ML65244CK	1.7ns	0°C to 70°C	20-Pin QSOP (K20)
ML65244CS	1.7ns	0°C to 70°C	20-Pin SOIC (S20)
ML65L244CK	2.0ns	0°C to 70°C	20-Pin QSOP (K20)
ML65L244CS	2.0ns	0°C to 70°C	20-Pin SOIC (S20)

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