

MAX4173

Low-Cost, SOT23, Voltage-Output, High-Side Current-Sense Amplifier

General Description

The MAX4173 low-cost, precision, high-side current-sense amplifier is available in a tiny SOT23-6 package. It features a voltage output that eliminates the need for gain-setting resistors and it is ideal for today's notebook computers, cell phones, and other systems where current monitoring is critical. High-side current monitoring is especially useful in battery-powered systems, since it does not interfere with the ground path of the battery charger. The input common-mode range of 0 to +28V is independent of the supply voltage and ensures that the current-sense feedback remains viable even when connected to a battery in deep discharge. The MAX4173's wide 1.7MHz bandwidth makes it suitable for use inside battery charger control loops.

The combination of three gain versions and a user-selectable external sense resistor sets the full-scale current reading. This feature offers a high level of integration, resulting in a simple and compact current-sense solution.

The MAX4173 operates from a single +3V to +28V supply, typically draws only 420μA of supply current over the extended operating temperature range (-40°C to +85°C), and is offered in the space-saving SOT23 package.

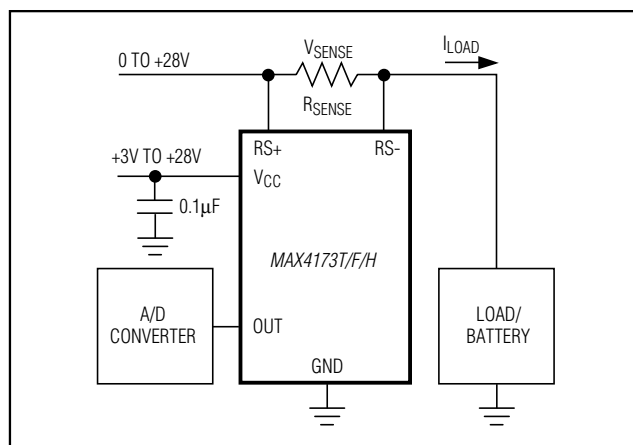
Applications

- Notebook Computers
- Portable/Battery-Powered Systems
- Smart Battery Packs/Chargers
- Cell Phones
- Power-Management Systems
- General System/Board-Level Current Monitoring
- PA Bias Control
- Precision Current Sources

Features

- ◆ Low-Cost, Compact Current-Sense Solution
- ◆ High Accuracy +2V to +28V Common-Mode Range, Functional Down to 0V, Independent of Supply Voltage
- ◆ Three Gain Versions Available
 - +20V/V (MAX4173T)
 - +50V/V (MAX4173F)
 - +100V/V (MAX4173H)
- ◆ ±0.5% Full-Scale Accuracy
- ◆ ±3mV Input Offset Voltage (MAX4173T)
- ◆ Wide 1.7MHz Bandwidth (MAX4173T)
- ◆ 420μA Supply Current
- ◆ Available in Space-Saving SOT23 Package

Typical Operating Circuit



Ordering Information

PART	GAIN (V/V)	TEMP RANGE	PIN-PACKAGE	SOT TOP MARK
MAX4173TEUT+T	20	-40°C to +85°C	6 SOT23	AABN
MAX4173TESA+	20	-40°C to +85°C	8 SO	—
MAX4173FEUT+T	50	-40°C to +85°C	6 SOT23	AABO
MAX4173FESA+	50	-40°C to +85°C	8 SO	—
MAX4173HEUT+T	100	-40°C to +85°C	6 SOT23	AABP
MAX4173HESA+	100	-40°C to +85°C	8 SO	—

+ Denotes a lead(Pb)-free/RoHS-compliant package.
T = Tape and reel.

Pin Configurations appear at end of data sheet.

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

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Low-Cost, SOT23, Voltage-Output, High-Side Current-Sense Amplifier

ABSOLUTE MAXIMUM RATINGS

V_{CC}, RS+, RS- to GND -0.3V to +30V
OUT to GND -0.3V to (V_{CC} + 0.3V)
Output Short-Circuit to V_{CC} or GND Continuous
Differential Input Voltage (V_{RS+} - V_{RS-}) ±0.3V
Current into Any Pin ±20mA

Continuous Power Dissipation (T_A = +70°C)
8-Pin SO (derate 5.88mW/°C above +70°C) 471mW
SOT23-6 (derate 8.7mW/°C above +70°C) 696mW
Operating Temperature Range -40°C to +85°C
Storage Temperature Range -65°C to +150°C
Lead Temperature (soldering, 10s) +300°C
Soldering Temperature (reflow) +260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{RS+} = 0 to +28V, V_{CC} = +3V to +28V, V_{SENSE} = 0V, T_A = T_{MIN} to T_{MAX}, R_{LOAD} = ∞ unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Operating Voltage Range	V _{CC}	Guaranteed by PSR test	3		28	V
Common-Mode Input Range	V _{CMR}	(Note 2)	0		28	V
Common-Mode Rejection	CMR	V _{RS+} > +2.0V		90		dB
Supply Current	I _{CC}	V _{RS+} > +2.0V, V _{CC} > 12V		0.42	1.0	mA
Leakage Current	I _{RS+} , I _{RS-}	V _{CC} = 0V, V _{RS+} = 28V		0.3	3	μA
Input Bias Current	I _{RS+}	V _{RS+} > +2.0V	0		50	μA
		V _{RS+} ≤ +2.0V	-350		50	
	I _{RS-}	V _{RS+} > +2.0V	0		100	
		V _{RS+} ≤ +2.0V	-700		100	
Full-Scale Sense Voltage	V _{SENSE}	V _{SENSE} = V _{RS+} - V _{RS-}		150		mV
Total OUT Voltage Error (Note 3)		V _{SENSE} = +100mV, V _{CC} = +12V, V _{RS+} = +12V		±0.5	5.75	%
		V _{SENSE} = +100mV, V _{CC} = +12V, V _{RS+} = +12V, T _A = +25°C		0.5	3.25	
		V _{SENSE} = +100mV, V _{CC} = +28V, V _{RS+} = +28V		0.5	5.75	
		V _{SENSE} = +100mV, V _{CC} = +12V, V _{RS+} = +0.1V		-9	±24	
		V _{CC} = +12V, V _{RS+} = +12V, V _{SENSE} = +6.25mV (Note 4)		±7.5		
Out High Voltage (Note 5)	(V _{CC} - V _{OH})	MAX4173T, V _{CC} = +3.0V, V _{RS+} = 28V, V _{SENSE} = 250mV		0.8	1.2	V
		MAX4173F, V _{CC} = +7.5V, V _{RS+} = 28V, V _{SENSE} = 250mV		0.8	1.2	
		MAX4173H, V _{CC} = +15V, V _{RS+} = 28V, V _{SENSE} = 250mV		0.8	1.2	
OUT Low Voltage	V _{OL}	MAX4173TEUT, V _{CC} = +5V, V _{RS+} = 0.89V, V _{SENSE} = 0mV	T _A = +25°C	1.2	5	mV
			T _A = -40°C to +85°C		40	

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ELECTRICAL CHARACTERISTICS (continued)

($V_{RS+} = 0$ to $+28V$, $V_{CC} = +3V$ to $+28V$, $V_{SENSE} = 0V$, $T_A = T_{MIN}$ to T_{MAX} , $R_{LOAD} = \infty$ unless otherwise noted. Typical values are at $T_A = +25^\circ C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Bandwidth	BW	$V_{RS+} = +12V$, $V_{CC} = +12V$, $C_{LOAD} = 5pF$	MAX4173T, $V_{SENSE} = +100mV$		1.7		MHz
			MAX4173F, $V_{SENSE} = +100mV$		1.4		
			MAX4173H, $V_{SENSE} = +100mV$		1.2		
			$V_{SENSE} = +6.25mV$ (Note 4)		0.6		
Gain	A_V	MAX4173T			20		V/V
		MAX4173F			50		
		MAX4173H			100		
Gain Accuracy	ΔA_V	MAX4173T/F $V_{SENSE} = +10mV$ to $+150mV$, $V_{CC} = V_{RS+} = 12V$	$T_A = +25^\circ C$		0.5	± 2.5	%
			$T_A = -40^\circ C$ to $+85^\circ C$			4.0	
	ΔA_V	MAX4173H $V_{SENSE} = +10mV$ to $+100mV$, $V_{CC} = V_{RS+} = 12V$	$T_A = +25^\circ C$		0.5	± 2.5	
			$T_A = -40^\circ C$ to $+85^\circ C$			4.0	
Input Offset Voltage (Note 6)	V_{OS}	MAX4173TEUT, $V_{RS+} = 12V$	$T_A = +25^\circ C$		0.3	± 3	mV
			$T_A = -40^\circ C$ to $+85^\circ C$			± 5	
OUT Settling Time to 1% of Final Value		$V_{CC} = +12V$, $V_{RS+} = 12V$, $C_{LOAD} = 5pF$	$V_{SENSE} = +6.25mV$ to $+100mV$		400		ns
			$V_{SENSE} = +100mV$ to $+6.25mV$		800		
OUT Output Resistance	R_{OUT}				12		$k\Omega$
Power-Supply Rejection	PSR	MAX4173T, $V_{SENSE} = 80mV$, $V_{RS+} \geq +2V$		60	84		dB
		MAX4173F, $V_{SENSE} = 32mV$, $V_{RS+} \geq +2V$		60	91		
		MAX4173H, $V_{SENSE} = 16mV$, $V_{RS+} \geq +2V$		60	95		
Power-Up Time to 1% of Final Value		$V_{SENSE} = +100mV$, $C_{LOAD} = 5pF$			10		μs
Saturation Recovery Time		$V_{CC} = +12V$, $V_{RS+} = 12V$ (Note 7)			10		μs

Note 1: All devices are 100% production tested at $T_A = +25^\circ C$. All temperature limits are guaranteed by design.

Note 2: Guaranteed by Total Output Voltage Error Test.

Note 3: Total OUT Voltage Error is the sum of gain and offset voltage errors.

Note 4: $+6.25mV = 1/16$ of $+100mV$ full-scale voltage.

Note 5: V_{SENSE} such that output stage is in saturation.

Note 6: V_{OS} is extrapolated from the Gain Accuracy tests.

Note 7: The device does not experience phase reversal when overdriven.

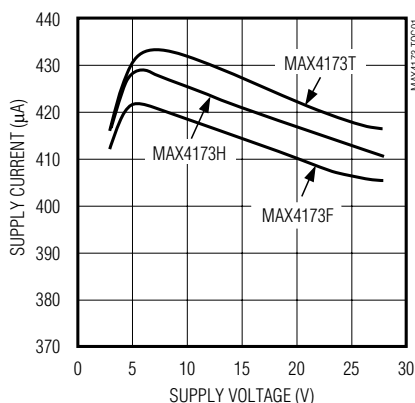
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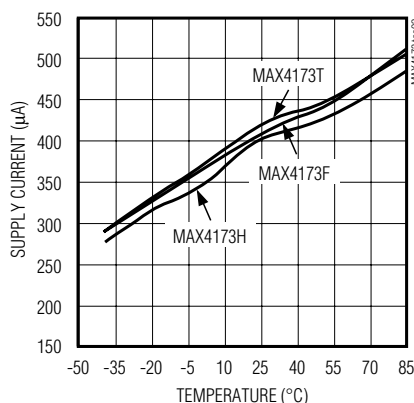
Typical Operating Characteristics

($V_{CC} = +12V$, $V_{RS+} = +12V$, $V_{SENSE} = +100mV$, $T_A = +25^{\circ}C$, unless otherwise noted.)

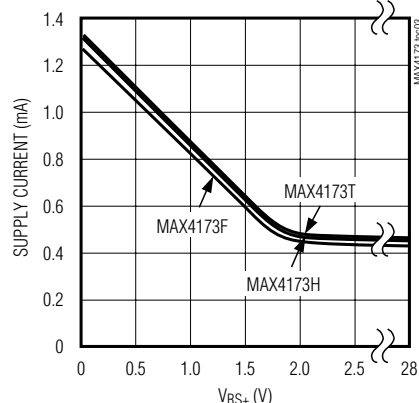
SUPPLY CURRENT vs. SUPPLY VOLTAGE



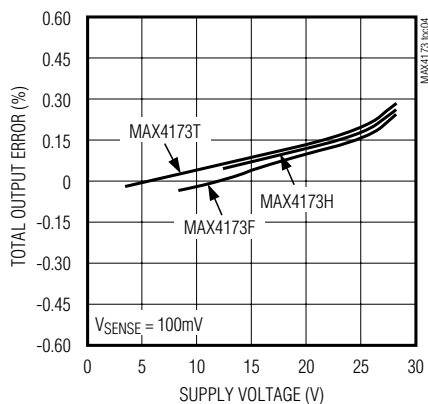
SUPPLY CURRENT vs. TEMPERATURE



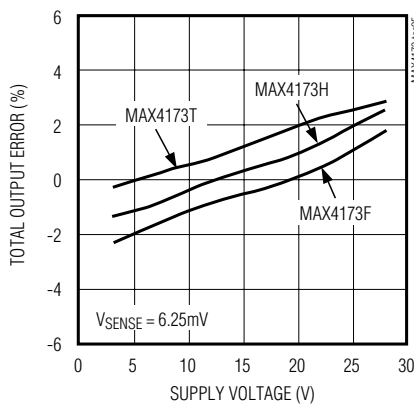
SUPPLY CURRENT vs. V_{RS+} VOLTAGE



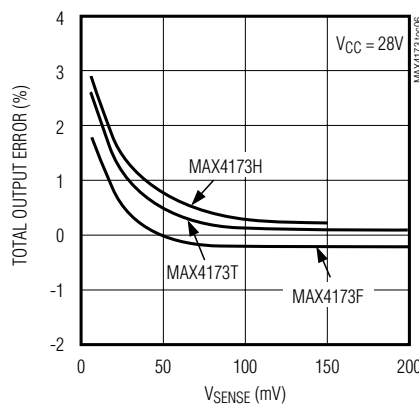
TOTAL OUTPUT ERROR vs. SUPPLY VOLTAGE



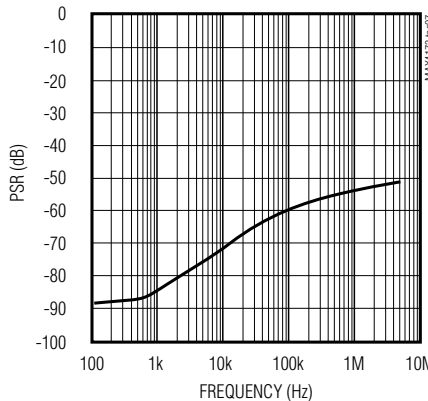
TOTAL OUTPUT ERROR vs. SUPPLY VOLTAGE



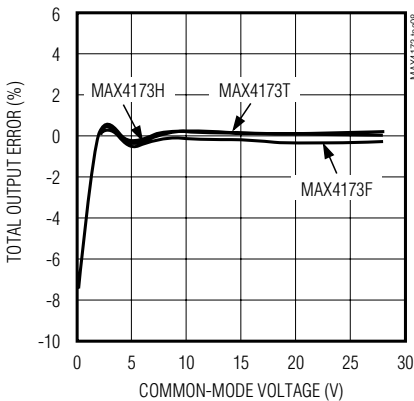
TOTAL OUTPUT ERROR vs. FULL-SCALE SENSE VOLTAGE



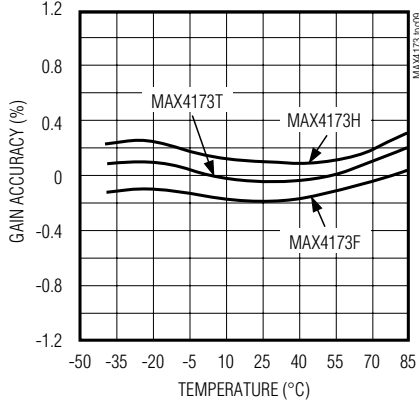
POWER-SUPPLY REJECTION vs. FREQUENCY



TOTAL OUTPUT ERROR vs. COMMON-MODE VOLTAGE



GAIN ACCURACY vs. TEMPERATURE



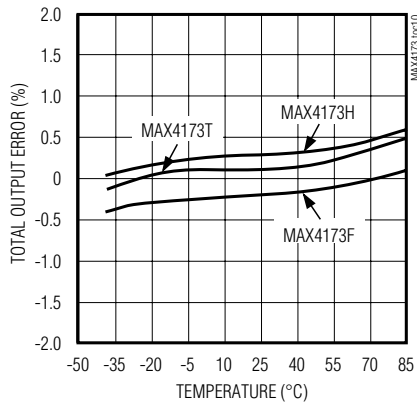
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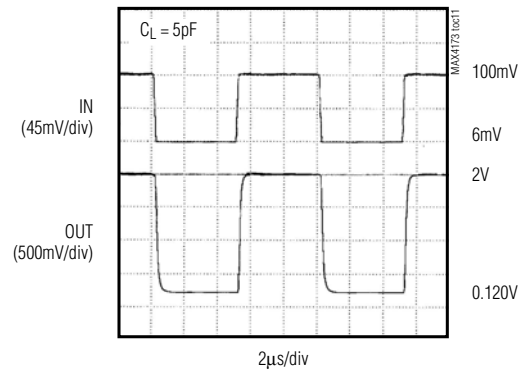
Typical Operating Characteristics (continued)

($V_{CC} = +12V$, $V_{RS+} = +12V$, $V_{SENSE} = +100mV$, $T_A = +25^{\circ}C$, unless otherwise noted.)

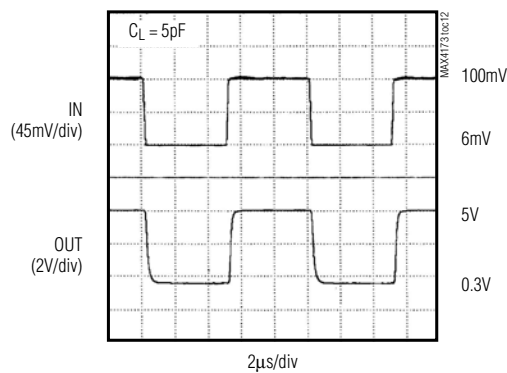
**TOTAL OUTPUT ERROR
vs. TEMPERATURE**



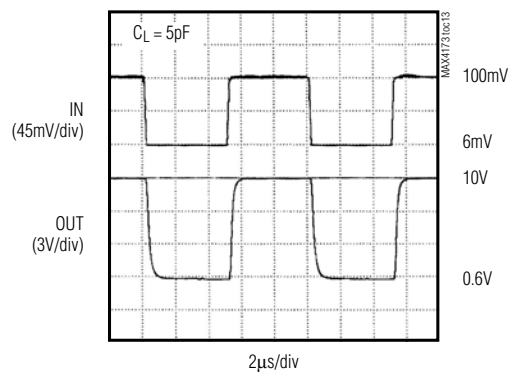
**MAX4173T
LARGE-SIGNAL TRANSIENT RESPONSE
($V_{SENSE} = 6mV$ to $100mV$)**



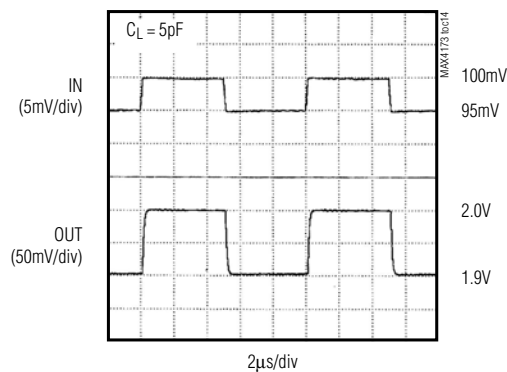
**MAX4173F
LARGE-SIGNAL TRANSIENT RESPONSE
($V_{SENSE} = 6mV$ to $100mV$)**



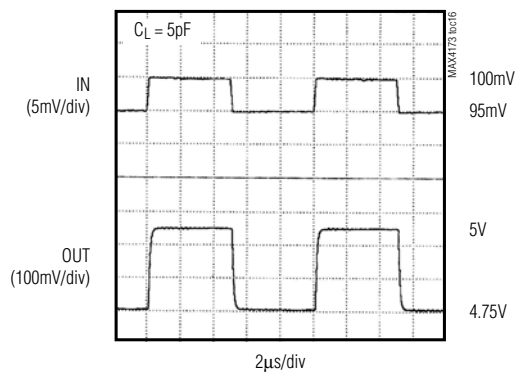
**MAX4173H
LARGE-SIGNAL TRANSIENT RESPONSE
($V_{SENSE} = 6mV$ to $100mV$)**



**MAX4173T
SMALL-SIGNAL TRANSIENT RESPONSE
($V_{SENSE} = 95mV$ TO $100mV$)**



**MAX4173F
SMALL-SIGNAL TRANSIENT RESPONSE
($V_{SENSE} = 95mV$ TO $100mV$)**

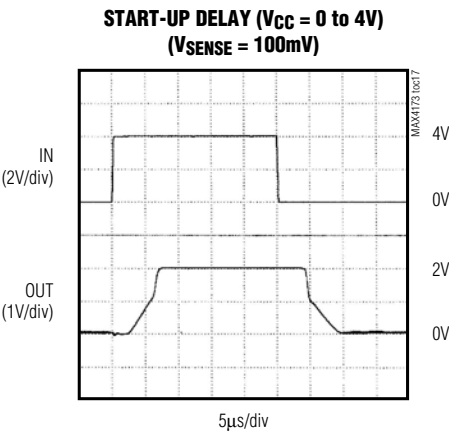
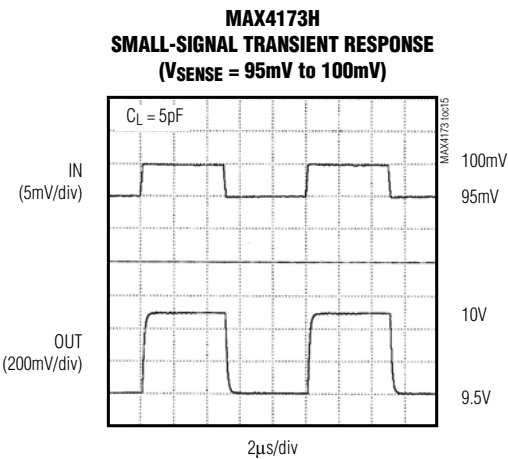


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Typical Operating Characteristics (continued)

(V_{CC} = +12V, V_{RS+} = +12V, V_{SENSE} = +100mV, T_A = +25°C, unless otherwise noted.)



Pin Description

PIN		NAME	FUNCTION
SOT23	SO		
1, 2	3	GND	Ground
3	1	V _{CC}	Supply Voltage Input. Bypass to GND with a 0.1µF capacitor.
4	8	RS+	Power-Side Connection to the External Sense Resistor
5	6	RS-	Load-Side Connection for the External Sense Resistor
6	4	OUT	Voltage Output. V _{OUT} is proportional to V _{SENSE} (V _{RS+} - V _{RS-}). Output impedance is approximately 12kΩ.
–	2, 5, 7	N.C.	No Connection. Not internally connected.

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Detailed Description

The MAX4173 high-side current-sense amplifier features a 0 to +28V input common-mode range that is independent of supply voltage. This feature allows the monitoring of current out of a battery in deep discharge and also enables high-side current sensing at voltages greater than the supply voltage (V_{CC}).

The MAX4173 operates as follows: Current from the source flows through R_{SENSE} to the load (Figure 1). Since the internal-sense amplifier's inverting input has high impedance, negligible current flows through R_{G2} (neglecting the input bias current). Therefore, the sense amplifier's inverting-input voltage equals $V_{SOURCE} - (I_{LOAD})(R_{SENSE})$. The amplifier's open-loop gain forces its noninverting input to the same voltage as the inverting input. Therefore, the drop across R_{G1} equals $(I_{LOAD})(R_{SENSE})$. Since I_{RG1} flows through R_{G1} , $I_{RG1} = (I_{LOAD})(R_{SENSE}) / R_{G1}$. The internal current mirror multiplies I_{RG1} by a current gain factor, β , to give $I_{RGD} = \beta \cdot I_{RG1}$. Solving $I_{RGD} = \beta \cdot (I_{LOAD})(R_{SENSE}) / R_{G1}$. Assuming infinite output impedance, $V_{OUT} = (I_{RGD})(R_{GD})$. Substituting in for I_{RGD} and rearranging, $V_{OUT} = \beta \cdot (R_{GD} / R_{G1})(R_{SENSE} \cdot I_{LOAD})$. The parts gain equals $\beta \cdot R_{GD} / R_{G1}$. Therefore, $V_{OUT} = (GAIN)(R_{SENSE})(I_{LOAD})$, where $GAIN = 20$ for MAX4173T, $GAIN = 50$ for MAX4173F, and $GAIN = 100$ for MAX4173H.

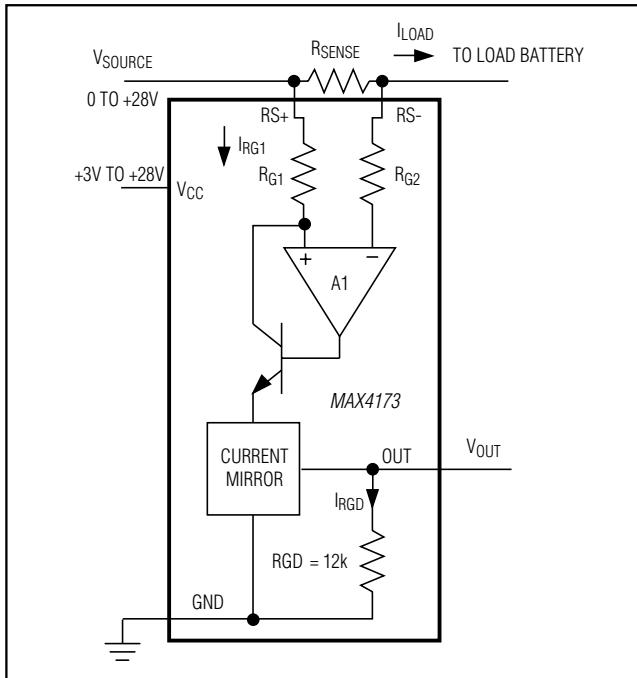


Figure 1. Functional Diagram

Set the full-scale output range by selecting R_{SENSE} and the appropriate gain version of the MAX4173.

Applications Information

Recommended Component Values

The MAX4173 senses a wide variety of currents with different sense resistor values. Table 1 lists common resistor values for typical operation of the MAX4173.

Choosing R_{SENSE}

To measure lower currents more accurately, use a high value for R_{SENSE} . The high value develops a higher sense voltage that reduces offset voltage errors of the internal op amp.

In applications monitoring very high currents, R_{SENSE} must be able to dissipate the I^2R losses. If the resistor's rated power dissipation is exceeded, its value may drift or it may fail altogether, causing a differential voltage across the terminals in excess of the absolute maximum ratings.

If I_{SENSE} has a large high-frequency component, minimize the inductance of R_{SENSE} . Wire-wound resistors have the highest inductance, metal-film resistors are somewhat better, and low-inductance metal-film resistors are best suited for these applications.

Using a PCB Trace as R_{SENSE}

If the cost of R_{SENSE} is an issue and accuracy is not critical, use the alternative solution shown in Figure 2. This solution uses copper PC board traces to create a sense resistor. The resistivity of a 0.1-inch-wide trace of 2-ounce copper is approximately 30mΩ/ft. The resistance-temperature coefficient of copper is fairly high (approximately 0.4%/°C), so systems that experience a wide temperature variance must compensate for this effect. In addition, do not exceed the maximum power dissipation of the copper trace.

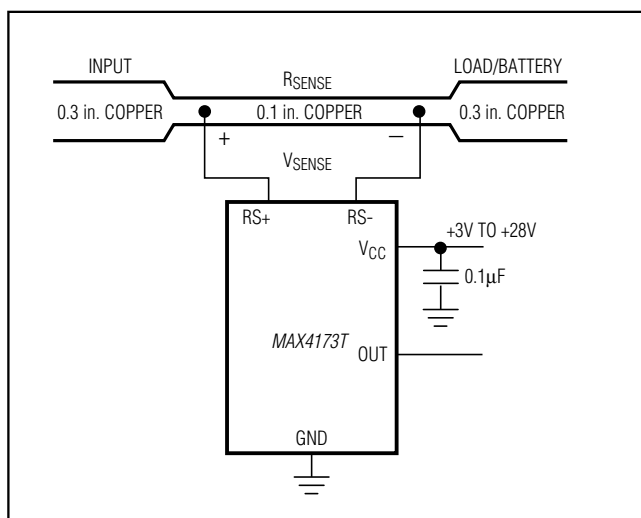
For example, the MAX4173T (with a maximum load current of 10A and an R_{SENSE} of 5mΩ) creates a full-scale V_{SENSE} of 50mV that yields a maximum V_{OUT} of 1V. R_{SENSE} in this case requires about 2 inches of 0.1 inch-wide copper trace.

Output Impedance

The output of the MAX4173 is a current source driving a 12kΩ resistance. Resistive loading added to OUT reduces the output gain of the MAX4173. To minimize output errors for most applications, connect OUT to a high-impedance input stage. When output buffering is required, choose an op amp with a common-mode input range and an output voltage swing that includes ground when operating with a single supply. The op

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FULL-SCALE LOAD CURRENT I _{LOAD} (A)	CURRENT-SENSE RESISTOR R _{SENSE} (mΩ)	GAIN	FULL-SCALE OUTPUT VOLTAGE (FULL-SCALE V _{SENSE} = 100mV) V _{OUT} (V)
0.1	1000	20	2.0
		50	5.0
		100	10.0
1	100	20	2.0
		50	5.0
		100	10.0
5	20	20	2.0
		50	5.0
		100	10.0
10	10	20	2.0
		50	5.0
		100	10.0



amp's supply voltage range should be at least as high as any voltage the system may encounter.

$$\% \text{ERROR} = 100 \left(\frac{R_{\text{LOAD}}}{12\text{k}\Omega + R_{\text{LOAD}}} - 1 \right)$$

The diagram shows a precision current source circuit. A 'LOW-COST SWITCHING REGULATOR' provides a supply voltage of 0 TO +28V to the RS+ pin of the MAX4173 op-amp. The RS- pin is connected to the output of the regulator. A feedback network consisting of a resistor labeled R_{SENSE} and a sense resistor labeled V_{SENSE} is connected between the output and the RS- pin. The output current is labeled I_{LOAD} . The op-amp's non-inverting input (V_{CC}) is connected to a +3V TO +28V supply through a 0.1μF capacitor. The op-amp's output (OUT) is connected to the same +3V TO +28V supply. The op-amp is powered by a GND connection. The load is represented by a box labeled 'LOAD/ BATTERY' connected to the output and GND.

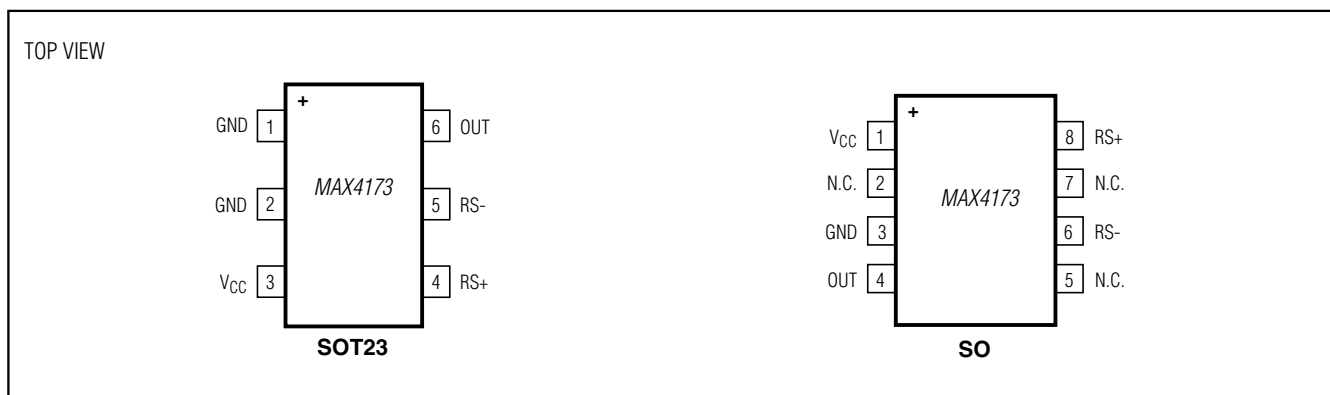
Current Source Circuit

Figure 3 shows a block diagram using the MAX4173 with a switching regulator to make a current source.

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Pin Configurations



Package Information

For the latest package outline information and land patterns (footprints), go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 SO	S8+4	21-0041	90-0096
6 SOT23	U6+2	21-0058	90-0175

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
4	6/10	Clarified 0V to 2V is not a high-accuracy range for the device, added lead-free options and soldering temperature	1, 2
5	4/11	Updated V_{RS+} conditions to synchronize with tested material	2, 3



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