

DDR3L SDRAM

MT41K512M4 – 64 Meg x 4 x 8 banks

MT41K256M8 – 32 Meg x 8 x 8 banks

MT41K128M16 – 16 Meg x 16 x 8 banks

Description

The 1.35V DDR3L SDRAM device is a low-voltage version of the 1.5V DDR3 SDRAM device. Refer to the DDR3 (1.5V) SDRAM data sheet specifications when running in 1.5V compatible mode.

Features

- $V_{DD} = V_{DDQ} = 1.35V$ (1.283–1.45V)
- Backward-compatible to $V_{DD} = V_{DDQ} = 1.5V \pm 0.075V$
- Differential bidirectional data strobe
- 8n-bit prefetch architecture
- Differential clock inputs (CK, CK#)
- 8 internal banks
- Nominal and dynamic on-die termination (ODT) for data, strobe, and mask signals
- Programmable CAS (READ) latency (CL)
- Programmable posted CAS additive latency (AL)
- Programmable CAS (WRITE) latency (CWL)
- Fixed burst length (BL) of 8 and burst chop (BC) of 4 (via the mode register set [MRS])
- Selectable BC4 or BL8 on-the-fly (OTF)
- Self refresh mode
- T_C of 0°C to +95°C
 - 64ms, 8192-cycle refresh at 0°C to +85°C
 - 32ms at +85°C to +95°C
- Self refresh temperature (SRT)

- Automatic self refresh (ASR)
- Write leveling
- Multipurpose register
- Output driver calibration

Options	Marking
• Configuration	
– 512 Meg x 4	512M4
– 256 Meg x 8	256M8
– 128 Meg x 16	128M16
• FBGA package (Pb-free) – x4, x8	
– 78-ball (8mm x 10.5mm) Rev. M, K	DA
– 78-ball FBGA (9mm x 11.5mm) Rev. D	HX
• FBGA package (Pb-free) – x16	
– 96-ball FBGA (9mm x 14mm) Rev. D	HA
– 96-ball FBGA (8mm x 14mm) Rev. K	JT
• Timing – cycle time	
– 1.071ns @ CL = 13 (DDR3-1866)	-107
– 1.25ns @ CL = 11 (DDR3-1600)	-125
– 1.5ns @ CL = 9 (DDR3-1333)	-15E
– 1.875ns @ CL = 7 (DDR3-1066)	-187E
• Operating temperature	
– Commercial (0°C $\leq T_C \leq$ +95°C)	None
– Industrial (-40°C $\leq T_C \leq$ +95°C)	IT
• Revision	:D / :M / :K

Table 1: Key Timing Parameters

Speed Grade	Data Rate (MT/s)	Target t_{RCD} - t_{RP} -CL	t_{RCD} (ns)	t_{RP} (ns)	CL (ns)
-107 ^{1, 2, 3}	1866	13-13-13	13.91	13.91	13.91
-125 ^{1, 2}	1600	11-11-11	13.75	13.75	13.75
-15E ¹	1333	9-9-9	13.5	13.5	13.5
-187E	1066	7-7-7	13.1	13.1	13.1

Notes:

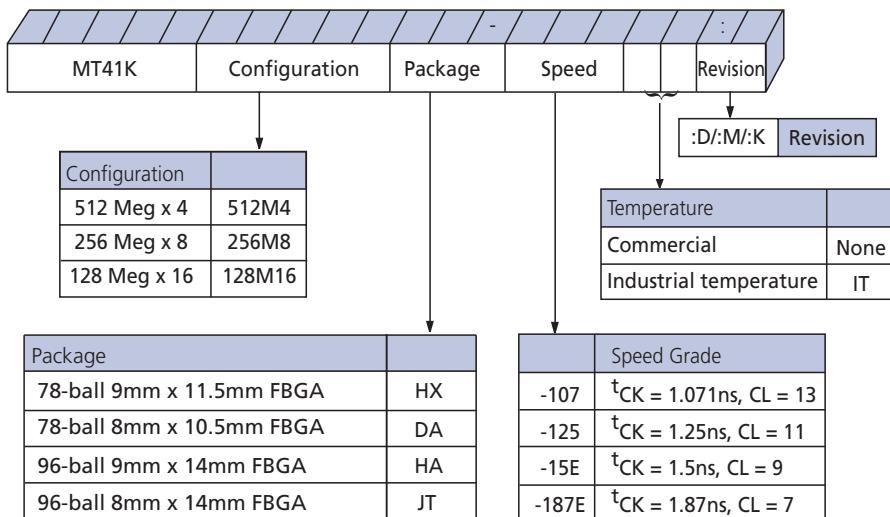
1. Backward compatible to 1066, CL = 7 (-187E).
2. Backward compatible to 1333, CL = 9 (-15E).
3. Backward compatible to 1600, CL = 11 (-107).

Table 2: Addressing

Parameter	512 Meg x 4	256 Meg x 8	128 Meg x 16
Configuration	64 Meg x 4 x 8 banks	32 Meg x 8 x 8 banks	16 Meg x 16 x 8 banks
Refresh count	8K	8K	8K
Row address	32K A[14:0]	32K A[14:0]	16K A[13:0]
Bank address	8 BA[2:0]	8 BA[2:0]	8 BA[2:0]
Column address	2K A[11, 9:0]	1K A[9:0]	1K A[9:0]

Figure 1: DDR3L Part Numbers

Example Part Number: MT41K256M8DA-125:K



Note: 1. Not all options listed can be combined to define an offered product. Use the part catalog search on <http://www.micron.com> for available offerings.

FBGA Part Marking Decoder

Due to space limitations, FBGA-packaged components have an abbreviated part marking that is different from the part number. For a quick conversion of an FBGA code, see the FBGA Part Marking Decoder on Micron's Web site: <http://www.micron.com>.

Contents

State Diagram	11
Functional Description	12
Industrial Temperature	12
General Notes	12
Functional Block Diagrams	14
Ball Assignments and Descriptions	16
Package Dimensions	22
Electrical Specifications	26
Absolute Ratings	26
Input/Output Capacitance	27
Thermal Characteristics	28
Electrical Specifications – I_{DD} Specifications and Conditions	30
Electrical Characteristics – I_{DD} Specifications	41
Electrical Specifications – DC and AC	44
DC Operating Conditions	44
Input Operating Conditions	45
DDR3L 1.35V AC Overshoot/Undershoot Specification	49
DDR3L 1.35V Slew Rate Definitions for Single-Ended Input Signals	52
DDR3L 1.35V Slew Rate Definitions for Differential Input Signals	54
ODT Characteristics	55
1.35V ODT Resistors	56
ODT Sensitivity	57
ODT Timing Definitions	57
Output Driver Impedance	61
34 Ohm Output Driver Impedance	62
DDR3L 34 Ohm Driver	63
DDR3L 34 Ohm Output Driver Sensitivity	64
DDR3L Alternative 40 Ohm Driver	65
DDR3L 40 Ohm Output Driver Sensitivity	65
Output Characteristics and Operating Conditions	67
Reference Output Load	70
Slew Rate Definitions for Single-Ended Output Signals	70
Slew Rate Definitions for Differential Output Signals	72
Speed Bin Tables	73
Electrical Characteristics and AC Operating Conditions	77
Command and Address Setup, Hold, and Derating	97
Data Setup, Hold, and Derating	104
Commands – Truth Tables	112
Commands	115
DESELECT	115
NO OPERATION	115
ZQ CALIBRATION LONG	115
ZQ CALIBRATION SHORT	115
ACTIVATE	115
READ	115
WRITE	116
PRECHARGE	117
REFRESH	117
SELF REFRESH	118
DLL Disable Mode	119

Input Clock Frequency Change	123
Write Leveling	125
Write Leveling Procedure	127
Write Leveling Mode Exit Procedure	129
Initialization	130
Voltage Initialization / Change	132
V _{DD} Voltage Switching	133
Mode Registers	134
Mode Register 0 (MR0)	135
Burst Length	135
Burst Type	136
DLL RESET	137
Write Recovery	137
Precharge Power-Down (Precharge PD)	138
CAS Latency (CL)	138
Mode Register 1 (MR1)	139
DLL Enable/DLL Disable	139
Output Drive Strength	140
OUTPUT ENABLE/DISABLE	140
TDQS Enable	140
On-Die Termination	141
WRITE LEVELING	141
POSTED CAS ADDITIVE Latency	141
Mode Register 2 (MR2)	142
CAS Write Latency (CWL)	143
AUTO SELF REFRESH (ASR)	143
SELF REFRESH TEMPERATURE (SRT)	144
SRT vs. ASR	144
DYNAMIC ODT	144
Mode Register 3 (MR3)	145
MULTIPURPOSE REGISTER (MPR)	145
MPR Functional Description	146
MPR Register Address Definitions and Bursting Order	147
MPR Read Predefined Pattern	153
MODE REGISTER SET (MRS) Command	153
ZQ CALIBRATION Operation	154
ACTIVATE Operation	155
READ Operation	157
WRITE Operation	168
DQ Input Timing	176
PRECHARGE Operation	178
SELF REFRESH Operation	178
Extended Temperature Usage	180
Power-Down Mode	181
RESET Operation	189
On-Die Termination (ODT)	191
Functional Representation of ODT	191
Nominal ODT	191
Dynamic ODT	193
Dynamic ODT Special Use Case	193
Functional Description	193
Synchronous ODT Mode	199

ODT Latency and Posted ODT	199
Timing Parameters	199
ODT Off During READs	202
Asynchronous ODT Mode	204
Synchronous to Asynchronous ODT Mode Transition (Power-Down Entry)	206
Asynchronous to Synchronous ODT Mode Transition (Power-Down Exit)	208
Asynchronous to Synchronous ODT Mode Transition (Short CKE Pulse)	210

List of Figures

Figure 1: DDR3L Part Numbers	2
Figure 2: Simplified State Diagram	11
Figure 3: 512 Meg x 4 Functional Block Diagram	14
Figure 4: 256 Meg x 8 Functional Block Diagram	15
Figure 5: 128 Meg x 16 Functional Block Diagram	15
Figure 6: 78-Ball FBGA – x4, x8 Ball Assignments (Top View)	16
Figure 7: 96-Ball FBGA – x16 Ball Assignments (Top View)	17
Figure 8: 78-Ball FBGA – x4, x8 (DA)	22
Figure 9: 78-Ball FBGA – x4, x8 (HX)	23
Figure 10: 96-Ball FBGA – x16 (HA)	24
Figure 11: 96-Ball FBGA – x16 (JT)	25
Figure 12: Thermal Measurement Point	29
Figure 13: DDR3L 1.35V Input Signal	48
Figure 14: Overshoot	49
Figure 15: Undershoot	49
Figure 16: V_{IX} for Differential Signals	50
Figure 17: Single-Ended Requirements for Differential Signals	50
Figure 18: Definition of Differential AC-Swing and t_{DVAC}	51
Figure 19: Nominal Slew Rate Definition for Single-Ended Input Signals	53
Figure 20: DDR3L 1.35V Nominal Differential Input Slew Rate Definition for DQS, DQS# and CK, CK#	54
Figure 21: ODT Levels and I-V Characteristics	55
Figure 22: ODT Timing Reference Load	58
Figure 23: t_{AON} and t_{AOF} Definitions	59
Figure 24: t_{AONPD} and t_{AOFPD} Definitions	59
Figure 25: t_{ADC} Definition	60
Figure 26: Output Driver	61
Figure 27: DQ Output Signal	68
Figure 28: Differential Output Signal	69
Figure 29: Reference Output Load for AC Timing and Output Slew Rate	70
Figure 30: Nominal Slew Rate Definition for Single-Ended Output Signals	71
Figure 31: Nominal Differential Output Slew Rate Definition for DQS, DQS#	72
Figure 32: Nominal Slew Rate and t_{VAC} for t_{IS} (Command and Address – Clock)	100
Figure 33: Nominal Slew Rate for t_{IH} (Command and Address – Clock)	101
Figure 34: Tangent Line for t_{IS} (Command and Address – Clock)	102
Figure 35: Tangent Line for t_{IH} (Command and Address – Clock)	103
Figure 36: Nominal Slew Rate and t_{VAC} for t_{DS} (DQ – Strobe)	108
Figure 37: Nominal Slew Rate for t_{DH} (DQ – Strobe)	109
Figure 38: Tangent Line for t_{DS} (DQ – Strobe)	110
Figure 39: Tangent Line for t_{DH} (DQ – Strobe)	111
Figure 40: Refresh Mode	118
Figure 41: DLL Enable Mode to DLL Disable Mode	120
Figure 42: DLL Disable Mode to DLL Enable Mode	121
Figure 43: DLL Disable t_{DQSCK}	122
Figure 44: Change Frequency During Precharge Power-Down	124
Figure 45: Write Leveling Concept	125
Figure 46: Write Leveling Sequence	128
Figure 47: Write Leveling Exit Procedure	129
Figure 48: Initialization Sequence	131
Figure 49: V_{DD} Voltage Switching	133
Figure 50: MRS to MRS Command Timing (t_{MRD})	134

Figure 51: MRS to nonMRS Command Timing (t_{MOD})	135
Figure 52: Mode Register 0 (MR0) Definitions	136
Figure 53: READ Latency	138
Figure 54: Mode Register 1 (MR1) Definition	139
Figure 55: READ Latency (AL = 5, CL = 6)	142
Figure 56: Mode Register 2 (MR2) Definition	143
Figure 57: CAS Write Latency	143
Figure 58: Mode Register 3 (MR3) Definition	145
Figure 59: Multipurpose Register (MPR) Block Diagram	146
Figure 60: MPR System Read Calibration with BL8: Fixed Burst Order Single Readout	149
Figure 61: MPR System Read Calibration with BL8: Fixed Burst Order, Back-to-Back Readout	150
Figure 62: MPR System Read Calibration with BC4: Lower Nibble, Then Upper Nibble	151
Figure 63: MPR System Read Calibration with BC4: Upper Nibble, Then Lower Nibble	152
Figure 64: ZQ CALIBRATION Timing (ZQCL and ZQCS)	154
Figure 65: Example: Meeting t_{RRD} (MIN) and t_{RCD} (MIN)	155
Figure 66: Example: t_{FAW}	156
Figure 67: READ Latency	157
Figure 68: Consecutive READ Bursts (BL8)	159
Figure 69: Consecutive READ Bursts (BC4)	159
Figure 70: Nonconsecutive READ Bursts	160
Figure 71: READ (BL8) to WRITE (BL8)	160
Figure 72: READ (BC4) to WRITE (BC4) OTF	161
Figure 73: READ to PRECHARGE (BL8)	161
Figure 74: READ to PRECHARGE (BC4)	162
Figure 75: READ to PRECHARGE (AL = 5, CL = 6)	162
Figure 76: READ with Auto Precharge (AL = 4, CL = 6)	162
Figure 77: Data Output Timing – t_{DQSQ} and Data Valid Window	164
Figure 78: Data Strobe Timing – READs	165
Figure 79: Method for Calculating t_{LZ} and t_{HZ}	166
Figure 80: t_{RPRE} Timing	166
Figure 81: t_{RPST} Timing	167
Figure 82: t_{WPRE} Timing	169
Figure 83: t_{WPST} Timing	169
Figure 84: WRITE Burst	170
Figure 85: Consecutive WRITE (BL8) to WRITE (BL8)	171
Figure 86: Consecutive WRITE (BC4) to WRITE (BC4) via OTF	171
Figure 87: Nonconsecutive WRITE to WRITE	172
Figure 88: WRITE (BL8) to READ (BL8)	172
Figure 89: WRITE to READ (BC4 Mode Register Setting)	173
Figure 90: WRITE (BC4 OTF) to READ (BC4 OTF)	174
Figure 91: WRITE (BL8) to PRECHARGE	175
Figure 92: WRITE (BC4 Mode Register Setting) to PRECHARGE	175
Figure 93: WRITE (BC4 OTF) to PRECHARGE	176
Figure 94: Data Input Timing	177
Figure 95: Self Refresh Entry/Exit Timing	179
Figure 96: Active Power-Down Entry and Exit	183
Figure 97: Precharge Power-Down (Fast-Exit Mode) Entry and Exit	183
Figure 98: Precharge Power-Down (Slow-Exit Mode) Entry and Exit	184
Figure 99: Power-Down Entry After READ or READ with Auto Precharge (RDAP)	184
Figure 100: Power-Down Entry After WRITE	185
Figure 101: Power-Down Entry After WRITE with Auto Precharge (WRAP)	185
Figure 102: REFRESH to Power-Down Entry	186

Figure 103: ACTIVATE to Power-Down Entry	186
Figure 104: PRECHARGE to Power-Down Entry	187
Figure 105: MRS Command to Power-Down Entry	187
Figure 106: Power-Down Exit to Refresh to Power-Down Entry	188
Figure 107: RESET Sequence	190
Figure 108: On-Die Termination	191
Figure 109: Dynamic ODT: ODT Asserted Before and After the WRITE, BC4	196
Figure 110: Dynamic ODT: Without WRITE Command	196
Figure 111: Dynamic ODT: ODT Pin Asserted Together with WRITE Command for 6 Clock Cycles, BL8	197
Figure 112: Dynamic ODT: ODT Pin Asserted with WRITE Command for 6 Clock Cycles, BC4	198
Figure 113: Dynamic ODT: ODT Pin Asserted with WRITE Command for 4 Clock Cycles, BC4	198
Figure 114: Synchronous ODT	200
Figure 115: Synchronous ODT (BC4)	201
Figure 116: ODT During READs	203
Figure 117: Asynchronous ODT Timing with Fast ODT Transition	205
Figure 118: Synchronous to Asynchronous Transition During Precharge Power-Down (DLL Off) Entry	207
Figure 119: Asynchronous to Synchronous Transition During Precharge Power-Down (DLL Off) Exit	209
Figure 120: Transition Period for Short CKE LOW Cycles with Entry and Exit Period Overlapping	211
Figure 121: Transition Period for Short CKE HIGH Cycles with Entry and Exit Period Overlapping	211

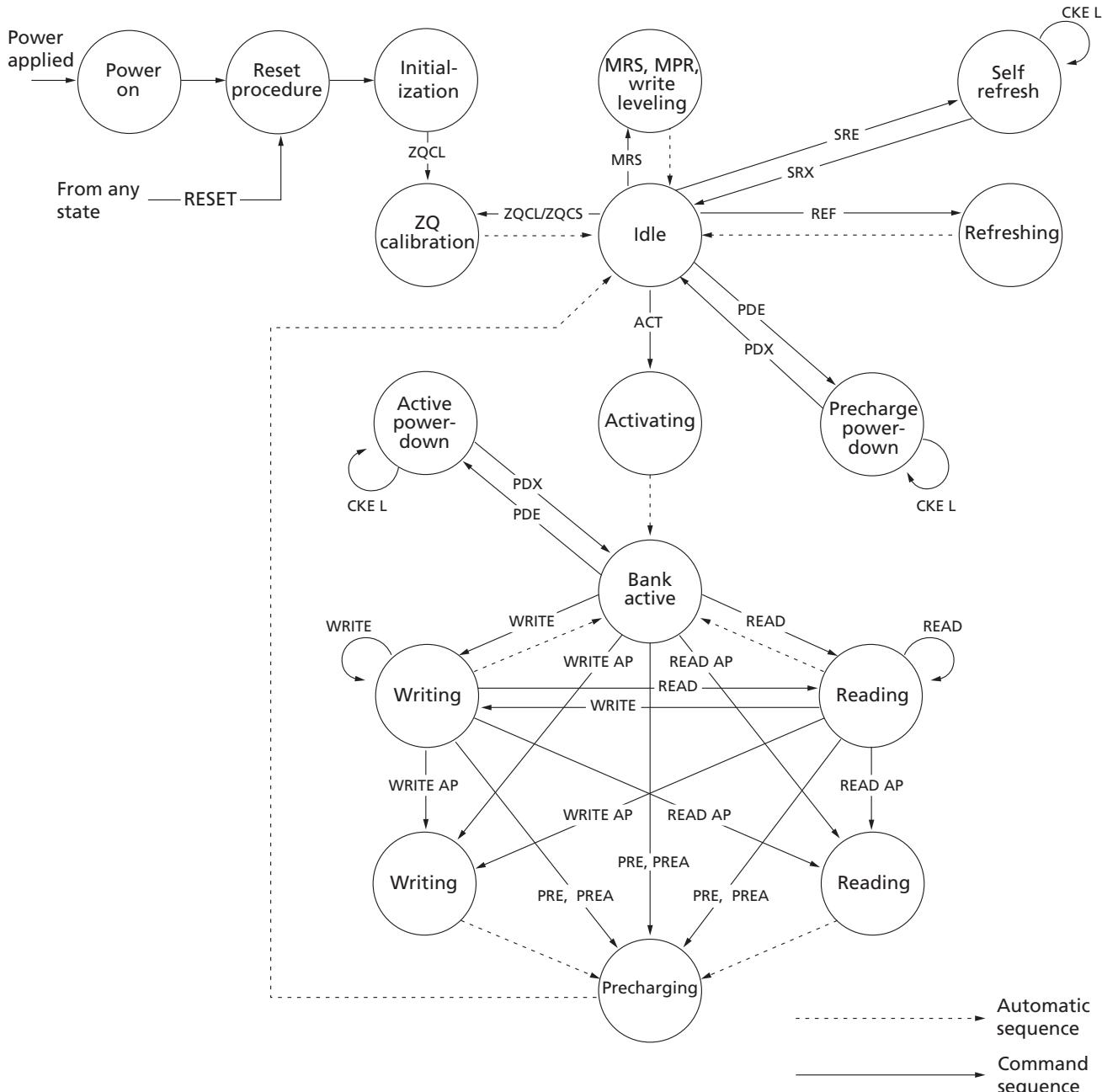
List of Tables

Table 1: Key Timing Parameters	1
Table 2: Addressing	2
Table 3: 78-Ball FBGA – x4, x8 Ball Descriptions	18
Table 4: 96-Ball FBGA – x16 Ball Descriptions	20
Table 5: Absolute Maximum Ratings	26
Table 6: DDR3L Input/Output Capacitance	27
Table 7: Thermal Characteristics	28
Table 8: DDR3L Timing Parameters Used for I_{DD} Measurements – Clock Units	30
Table 9: DDR3L I_{DD0} Measurement Loop	31
Table 10: DDR3L I_{DD1} Measurement Loop	32
Table 11: DDR3L I_{DD} Measurement Conditions for Power-Down Currents	33
Table 12: DDR3L I_{DD2N} and I_{DD3N} Measurement Loop	34
Table 13: DDR3L I_{DD2NT} Measurement Loop	34
Table 14: DDR3L I_{DD4R} Measurement Loop	35
Table 15: DDR3L I_{DD4W} Measurement Loop	36
Table 16: DDR3L I_{DD5B} Measurement Loop	37
Table 17: DDR3L I_{DD} Measurement Conditions for I_{DD6} , I_{DD6ET} , and I_{DD8}	38
Table 18: DDR3L I_{DD7} Measurement Loop	39
Table 19: I_{DD} Maximum Limits – Die Rev. D	41
Table 20: I_{DD} Maximum Limits – Die Rev. M	42
Table 21: I_{DD} Maximum Limits – Die Rev. K	43
Table 22: DDR3L 1.35V DC Electrical Characteristics and Operating Conditions	44
Table 23: DDR3L 1.35V DC Electrical Characteristics and Input Conditions	45
Table 24: DDR3L 1.35V Input Switching Conditions - Command and Address	46
Table 25: DDR3L 1.35V Differential Input Operating Conditions (CK, CK# and DQS, DQS#)	47
Table 26: DDR3L Control and Address Pins	49
Table 27: DDR3L 1.35V Clock, Data, Strobe, and Mask Pins	49
Table 28: DDR3L 1.35V - Minimum Required Time 'DVAC for CK/CK#, DQS/DQS# Differential for AC Ringback	51
Table 29: Single-Ended Input Slew Rate Definition	52
Table 30: DDR3L 1.35V Differential Input Slew Rate Definition	54
Table 31: On-Die Termination DC Electrical Characteristics	55
Table 32: 1.35V R_{TT} Effective Impedance	56
Table 33: ODT Sensitivity Definition	57
Table 34: ODT Temperature and Voltage Sensitivity	57
Table 35: ODT Timing Definitions	58
Table 36: DDR3L(1.35V) Reference Settings for ODT Timing Measurements	58
Table 37: DDR3L 34 Ohm Driver Impedance Characteristics	62
Table 38: DDR3L 34 Ohm Driver Pull-Up and Pull-Down Impedance Calculations	63
Table 39: DDR3L 34 Ohm Driver I_{OH}/I_{OL} Characteristics: $V_{DD} = V_{DDQ} = \text{DDR3L@1.35V}$	63
Table 40: DDR3L 34 Ohm Driver I_{OH}/I_{OL} Characteristics: $V_{DD} = V_{DDQ} = \text{DDR3L@1.45V}$	63
Table 41: DDR3L 34 Ohm Driver I_{OH}/I_{OL} Characteristics: $V_{DD} = V_{DDQ} = \text{DDR3L@1.283}$	64
Table 42: DDR3L 34 Ohm Output Driver Sensitivity Definition	64
Table 43: DDR3L 34 Ohm Output Driver Voltage and Temperature Sensitivity	64
Table 44: DDR3L 40 Ohm Driver Impedance Characteristics	65
Table 45: DDR3L 40 Ohm Output Driver Sensitivity Definition	65
Table 46: 40 Ohm Output Driver Voltage and Temperature Sensitivity	66
Table 47: DDR3L Single-Ended Output Driver Characteristics	67
Table 48: DDR3L Differential Output Driver Characteristics	68
Table 49: DDR3L Differential Output Driver Characteristics $V_{OX(AC)}$	69
Table 50: Single-Ended Output Slew Rate Definition	70

Table 51: Differential Output Slew Rate Definition	72
Table 52: DDR3L-1066 Speed Bins	73
Table 53: DDR3L-1333 Speed Bins	74
Table 54: DDR3L-1600 Speed Bins	75
Table 55: DDR3L-1866 Speed Bins	76
Table 56: Electrical Characteristics and AC Operating Conditions	77
Table 57: Electrical Characteristics and AC Operating Conditions	79
Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions	89
Table 59: DDR3L Command and Address Setup and Hold Values 1 V/ns Referenced – AC/DC-Based	98
Table 60: DDR3L-800/1066/1333/1600 Derating Values t_{IS}/t_{IH} – AC160/DC90-Based	98
Table 61: DDR3L-800/1066/1333/1600 Derating Values for t_{IS}/t_{IH} – AC135/DC90-Based	98
Table 62: DDR3L-1866 Derating Values for t_{IS}/t_{IH} – AC125/DC90-Based	99
Table 63: DDR3L Minimum Required Time t_{VAC} Above $V_{IH(AC)}$ (Below $V_{IL(AC)}$) for Valid ADD/CMD Transition ..	99
Table 64: DDR3L Data Setup and Hold Values at 1 V/ns (DQS, DQS# at 2 V/ns) – AC/DC-Based	105
Table 65: DDR3L Derating Values for t_{DS}/t_{DH} – AC160/DC90-Based	105
Table 66: DDR3L Derating Values for t_{DS}/t_{DH} – AC135/DC100-Based	105
Table 67: DDR3L Derating Values for t_{DS}/t_{DH} – AC130/DC100-Based at 2V/ns	106
Table 68: DDR3L Minimum Required Time t_{VAC} Above $V_{IH(AC)}$ (Below $V_{IL(AC)}$) for Valid DQ Transition	107
Table 69: Truth Table – Command	112
Table 70: Truth Table – CKE	114
Table 71: READ Command Summary	116
Table 72: WRITE Command Summary	116
Table 73: READ Electrical Characteristics, DLL Disable Mode	122
Table 74: Write Leveling Matrix	126
Table 75: Burst Order	137
Table 76: MPR Functional Description of MR3 Bits	146
Table 77: MPR Readouts and Burst Order Bit Mapping	147
Table 78: Self Refresh Temperature and Auto Self Refresh Description	180
Table 79: Self Refresh Mode Summary	180
Table 80: Command to Power-Down Entry Parameters	181
Table 81: Power-Down Modes	182
Table 82: Truth Table – ODT (Nominal)	192
Table 83: ODT Parameters	192
Table 84: Write Leveling with Dynamic ODT Special Case	193
Table 85: Dynamic ODT Specific Parameters	194
Table 86: Mode Registers for $R_{TT,nom}$	194
Table 87: Mode Registers for $R_{TT(WR)}$	195
Table 88: Timing Diagrams for Dynamic ODT	195
Table 89: Synchronous ODT Parameters	200
Table 90: Asynchronous ODT Timing Parameters for All Speed Bins	205
Table 91: ODT Parameters for Power-Down (DLL Off) Entry and Exit Transition Period	207

State Diagram

Figure 2: Simplified State Diagram



ACT = ACTIVATE
MPR = Multipurpose register
MRS = Mode register set
PDE = Power-down entry
PDX = Power-down exit
PRE = PRECHARGE

PREA = PRECHARGE ALL
READ = RD, RDS4, RDS8
READ AP = RDAP, RDAPS4, RDAPS8
REF = REFRESH
RESET = START RESET PROCEDURE
SRE = Self refresh entry

SRX = Self refresh exit
WRITE = WR, WRS4, WR8
WRITE AP = WRAP, WRAPS4, WRAPS8
ZQCL = ZQ LONG CALIBRATION
ZQCS = ZQ SHORT CALIBRATION

Functional Description

DDR3 SDRAM uses a double data rate architecture to achieve high-speed operation. The double data rate architecture is an $8n$ -prefetch architecture with an interface designed to transfer two data words per clock cycle at the I/O pins. A single read or write operation for the DDR3 SDRAM effectively consists of a single $8n$ -bit-wide, four-clock-cycle data transfer at the internal DRAM core and eight corresponding n -bit-wide, one-half-clock-cycle data transfers at the I/O pins.

The differential data strobe (DQS, DQS#) is transmitted externally, along with data, for use in data capture at the DDR3 SDRAM input receiver. DQS is center-aligned with data for WRITEs. The read data is transmitted by the DDR3 SDRAM and edge-aligned to the data strobes.

The DDR3 SDRAM operates from a differential clock (CK and CK#). The crossing of CK going HIGH and CK# going LOW is referred to as the positive edge of CK. Control, command, and address signals are registered at every positive edge of CK. Input data is registered on the first rising edge of DQS after the WRITE preamble, and output data is referenced on the first rising edge of DQS after the READ preamble.

Read and write accesses to the DDR3 SDRAM are burst-oriented. Accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVATE command, which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVATE command are used to select the bank and row to be accessed. The address bits registered coincident with the READ or WRITE commands are used to select the bank and the starting column location for the burst access.

The device uses a READ and WRITE BL8 and BC4. An auto precharge function may be enabled to provide a self-timed row precharge that is initiated at the end of the burst access.

As with standard DDR SDRAM, the pipelined, multibank architecture of DDR3 SDRAM allows for concurrent operation, thereby providing high bandwidth by hiding row precharge and activation time.

A self refresh mode is provided, along with a power-saving, power-down mode.

Industrial Temperature

The industrial temperature (IT) device requires that the case temperature not exceed -40°C or 95°C . JEDEC specifications require the refresh rate to double when T_C exceeds 85°C ; this also requires use of the high-temperature self refresh option. Additionally, ODT resistance and the input/output impedance must be derated when T_C is $< 0^{\circ}\text{C}$ or $> 95^{\circ}\text{C}$.

General Notes

- The functionality and the timing specifications discussed in this data sheet are for the DLL enable mode of operation (normal operation).
- Throughout this data sheet, various figures and text refer to DQs as “DQ.” DQ is to be interpreted as any and all DQ collectively, unless specifically stated otherwise.
- The terms “DQS” and “CK” found throughout this data sheet are to be interpreted as DQS, DQS# and CK, CK# respectively, unless specifically stated otherwise.

- Complete functionality may be described throughout the document; any page or diagram may have been simplified to convey a topic and may not be inclusive of all requirements.
- Any specific requirement takes precedence over a general statement.
- Any functionality not specifically stated is considered undefined, illegal, and not supported, and can result in unknown operation.
- Row addressing is denoted as $A[n:0]$. *For example*, 1Gb: $n = 12$ (x16); 1Gb: $n = 13$ (x4, x8); 2Gb: $n = 13$ (x16) and 2Gb: $n = 14$ (x4, x8); 4Gb: $n = 14$ (x16); and 4Gb: $n = 15$ (x4, x8).
- Dynamic ODT has a special use case: when DDR3 devices are architected for use in a single rank memory array, the ODT ball can be wired HIGH rather than routed. Refer to the Dynamic ODT Special Use Case section.
- A x16 device's DQ bus is comprised of two bytes. If only one of the bytes needs to be used, use the lower byte for data transfers and terminate the upper byte as noted:
 - Connect UDQS to ground via $1k\Omega^*$ resistor.
 - Connect UDQS# to V_{DD} via $1k\Omega^*$ resistor.
 - Connect UDM to V_{DD} via $1k\Omega^*$ resistor.
 - Connect DQ[15:8] individually to either V_{SS} , V_{DD} , or V_{REF} via $1k\Omega$ resistors,* or float DQ[15:8].

*If ODT is used, $1k\Omega$ resistor should be changed to 4x that of the selected ODT.

Functional Block Diagrams

DDR3 SDRAM is a high-speed, CMOS dynamic random access memory. It is internally configured as an 8-bank DRAM.

Figure 3: 512 Meg x 4 Functional Block Diagram

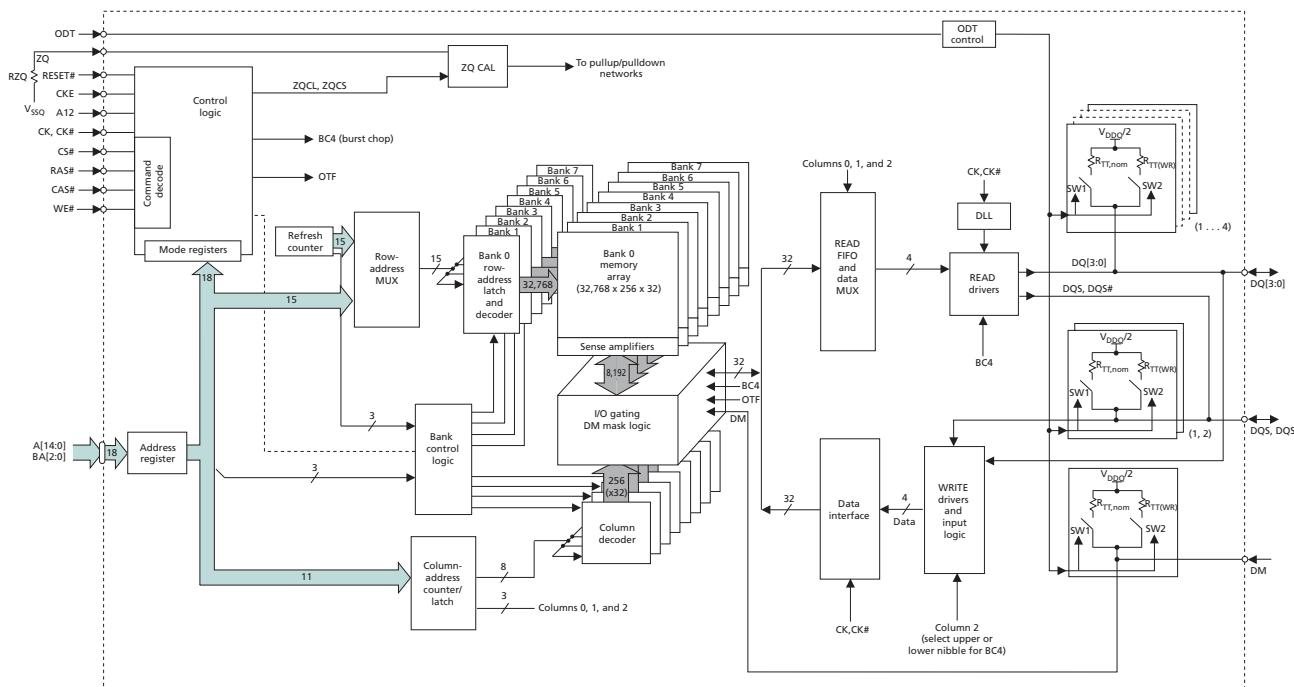
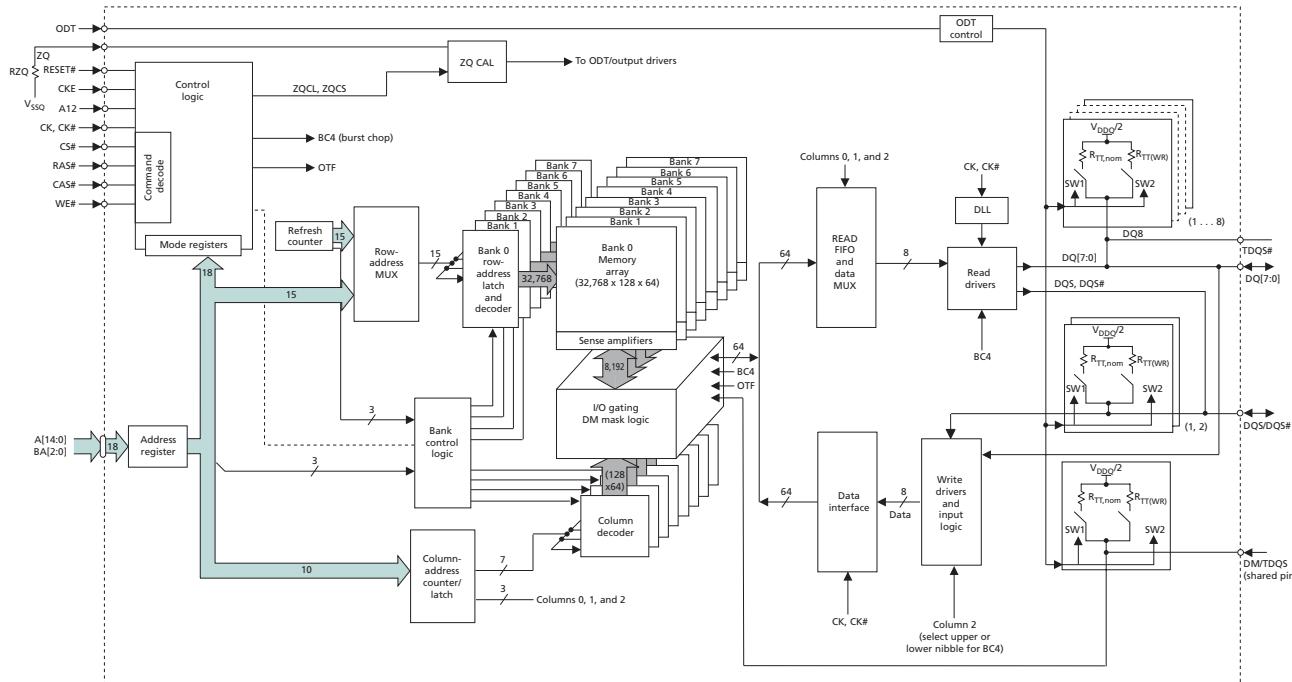
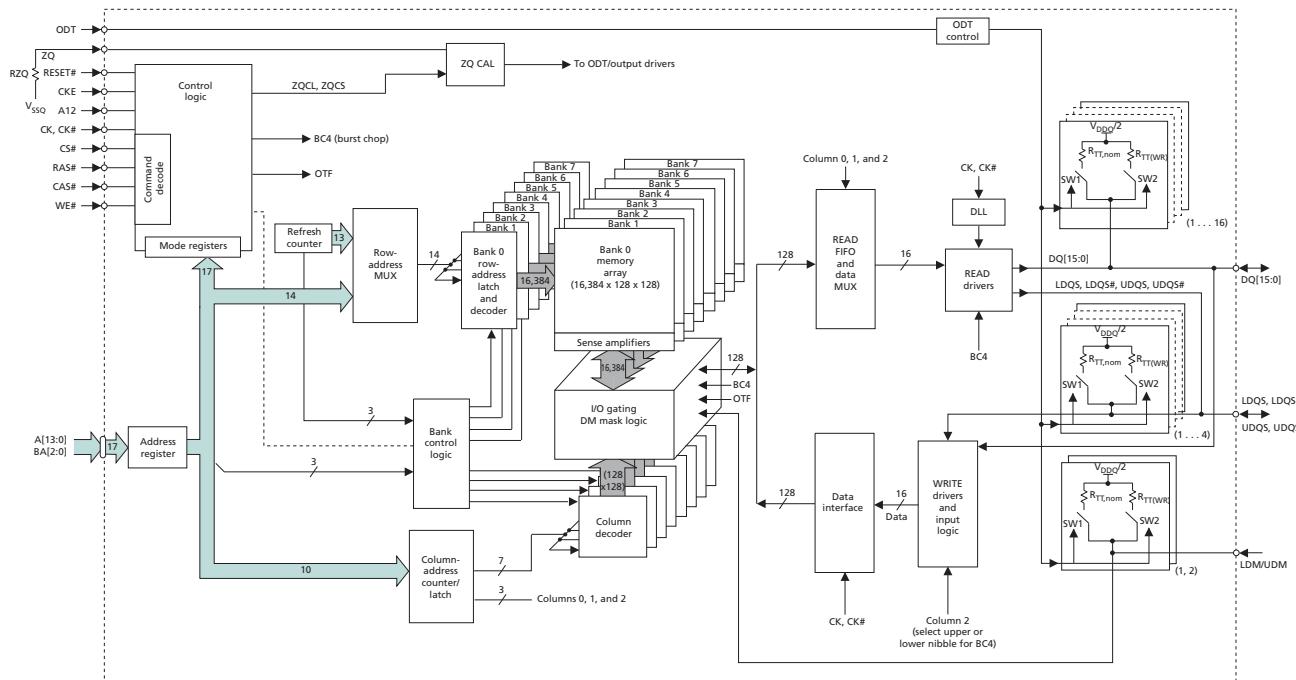


Figure 4: 256 Meg x 8 Functional Block Diagram

Figure 5: 128 Meg x 16 Functional Block Diagram


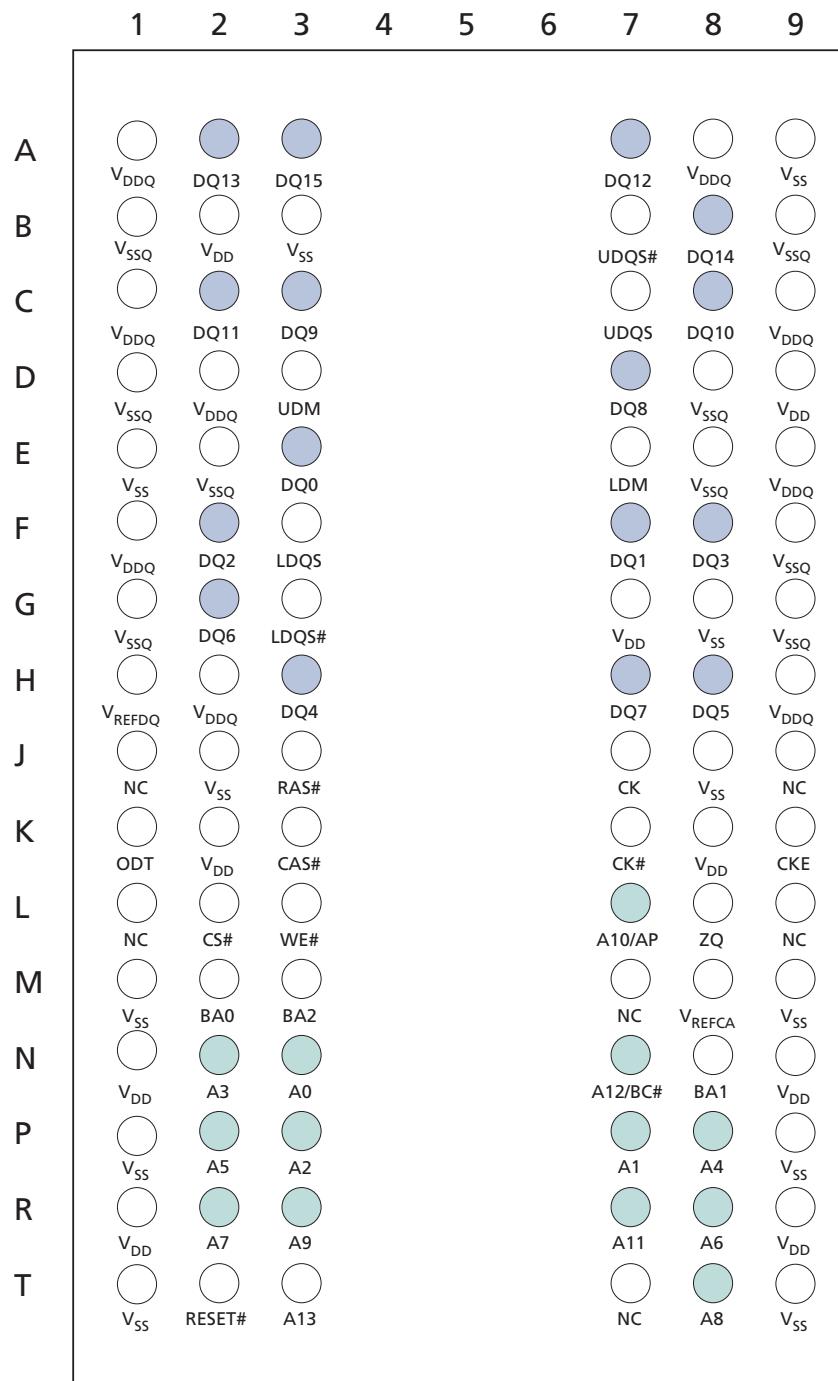
Ball Assignments and Descriptions

Figure 6: 78-Ball FBGA – x4, x8 Ball Assignments (Top View)

	1	2	3	4	5	6	7	8	9
A									
V _{SS}	V _{DD}	NC				NF, NF/TDQS#	V _{SS}	V _{DD}	
B									
V _{SS}	V _{SSQ}	DQ0				DM, DM/TDQS	V _{SSQ}	V _{DDQ}	
C									
V _{DDQ}	DQ2	DQS				DQ1	DQ3	V _{SSQ}	
D									
V _{SSQ}	NF, DQ6	DQS#				V _{DD}	V _{SS}	V _{SSQ}	
E									
V _{REFDQ}	V _{DDQ}	NF, DQ4				NF, DQ7	NF, DQ5	V _{DDQ}	
F									
NC	V _{SS}	RAS#				CK	V _{SS}	NC	
G									
ODT	V _{DD}	CAS#				CK#	V _{DD}	CKE	
H									
NC	CS#	WE#				A10/AP	ZQ	NC	
J									
V _{SS}	BA0	BA2				NC	V _{REFCA}	V _{SS}	
K									
V _{DD}	A3	A0				A12/BC#	BA1	V _{DD}	
L									
V _{SS}	A5	A2				A1	A4	V _{SS}	
M									
V _{DD}	A7	A9				A11	A6	V _{DD}	
N									
V _{SS}	RESET#	A13				A14	A8	V _{SS}	

Notes:

1. Ball descriptions listed in Table 3 (page 18) are listed as "x4, x8" if unique; otherwise, x4 and x8 are the same.
2. A comma separates the configuration; a slash defines a selectable function.
Example: D7 = NF, NF/TDQS#. NF applies to the x4 configuration only. NF/TDQS# applies to the x8 configuration only—selectable between NF or TDQS# via MRS (symbols are defined in Table 3).

Figure 7: 96-Ball FBGA – x16 Ball Assignments (Top View)


Notes: 1. Ball descriptions listed in Table 4 (page 20) are listed as "x16."
 2. A comma separates the configuration; a slash defines a selectable function.

Table 3: 78-Ball FBGA – x4, x8 Ball Descriptions

Symbol	Type	Description
A[14:13], A12/BC#, A11, A10/AP, A[9:0]	Input	Address inputs: Provide the row address for ACTIVATE commands, and the column address and auto precharge bit (A10) for READ/WRITE commands, to select one location out of the memory array in the respective bank. A10 sampled during a PRECHARGE command determines whether the PRECHARGE applies to one bank (A10 LOW, bank selected by BA[2:0]) or all banks (A10 HIGH). The address inputs also provide the op-code during a LOAD MODE command. Address inputs are referenced to V_{REFCA} . A12/BC#: When enabled in the mode register (MR), A12 is sampled during READ and WRITE commands to determine whether burst chop (on-the-fly) will be performed (HIGH = BL8 or no burst chop, LOW = BC4 burst chop). See Truth Table - Command.
BA[2:0]	Input	Bank address inputs: BA[2:0] define the bank to which an ACTIVATE, READ, WRITE, or PRECHARGE command is being applied. BA[2:0] define which mode register (MR0, MR1, MR2, or MR3) is loaded during the LOAD MODE command. BA[2:0] are referenced to V_{REFCA} .
CK, CK#	Input	Clock: CK and CK# are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and the negative edge of CK#. Output data strobe (DQS, DQS#) is referenced to the crossings of CK and CK#.
CKE	Input	Clock enable: CKE enables (registered HIGH) and disables (registered LOW) internal circuitry and clocks on the DRAM. The specific circuitry that is enabled/disabled is dependent upon the DDR3 SDRAM configuration and operating mode. Taking CKE LOW provides PRECHARGE power-down and SELF REFRESH operations (all banks idle) or active power-down (row active in any bank). CKE is synchronous for power-down entry and exit and for self refresh entry. CKE is asynchronous for self refresh exit. Input buffers (excluding CK, CK#, CKE, RESET#, and ODT) are disabled during power-down. Input buffers (excluding CKE and RESET#) are disabled during SELF REFRESH. CKE is referenced to V_{REFCA} .
CS#	Input	Chip select: CS# enables (registered LOW) and disables (registered HIGH) the command decoder. All commands are masked when CS# is registered HIGH. CS# provides for external rank selection on systems with multiple ranks. CS# is considered part of the command code. CS# is referenced to V_{REFCA} .
DM	Input	Input data mask: DM is an input mask signal for write data. Input data is masked when DM is sampled HIGH along with the input data during a write access. Although the DM ball is input-only, the DM loading is designed to match that of the DQ and DQS balls. DM is referenced to V_{REFDQ} . DM has an optional use as TDQS on the x8 device.
ODT	Input	On-die termination: ODT enables (registered HIGH) and disables (registered LOW) termination resistance internal to the DDR3 SDRAM. When enabled in normal operation, ODT is only applied to each of the following balls: DQ[7:0], DQS, DQS#, and DM for the x8; DQ[3:0], DQS, DQS#, and DM for the x4. The ODT input is ignored if disabled via the LOAD MODE command. ODT is referenced to V_{REFCA} .
RAS#, CAS#, WE#	Input	Command inputs: RAS#, CAS#, and WE# (along with CS#) define the command being entered and are referenced to V_{REFCA} .
RESET#	Input	Reset: RESET# is an active LOW CMOS input referenced to V_{SS} . The RESET# input receiver is a CMOS input defined as a rail-to-rail signal with DC HIGH $\geq 0.8 \times V_{DDQ}$ and DC LOW $\leq 0.2 \times V_{DDQ}$. RESET# assertion and deassertion are asynchronous.
DQ[3:0]	I/O	Data input/output: Bidirectional data bus for the x4 configuration. DQ[3:0] are referenced to V_{REFDQ} .

Table 3: 78-Ball FBGA – x4, x8 Ball Descriptions (Continued)

Symbol	Type	Description
DQ[7:0]	I/O	Data input/output: Bidirectional data bus for the x8 configuration. DQ[7:0] are referenced to V_{REFDQ} .
DQS, DQS#	I/O	Data strobe: Output with read data. Edge-aligned with read data. Input with write data. Center-aligned to write data.
TDQS, TDQS#	I/O	Termination data strobe: Applies to the x8 configuration only. When TDQS is enabled, DM is disabled, and the TDQS and TDQS# balls provide termination resistance.
V_{DD}	Supply	Power supply: 1.35V, 1.283–1.45V operational; compatible to 1.5V operation.
V_{DDQ}	Supply	DQ power supply: 1.35V, 1.283–1.45V operational; compatible with 1.5V operation.
V_{REFCA}	Supply	Reference voltage for control, command, and address: V_{REFCA} must be maintained at all times (including self refresh) for proper device operation.
V_{REFDQ}	Supply	Reference voltage for data: V_{REFDQ} must be maintained at all times (including self refresh) for proper device operation.
V_{SS}	Supply	Ground.
V_{SSQ}	Supply	DQ ground: Isolated on the device for improved noise immunity.
ZQ	Reference	External reference ball for output drive calibration: This ball is tied to an external 240Ω resistor (R_{ZQ}), which is tied to V_{SSQ} .
NC	–	No connect: These balls should be left unconnected (the ball has no connection to the DRAM or to other balls).
NF	–	No function: When configured as a x4 device, these balls are NF. When configured as a x8 device, these balls are defined as TDQS#, DQ[7:4].

Table 4: 96-Ball FBGA – x16 Ball Descriptions

Symbol	Type	Description
A13, A12/BC#, A11, A10/AP, A[9:0]	Input	Address inputs: Provide the row address for ACTIVATE commands, and the column address and auto precharge bit (A10) for READ/WRITE commands, to select one location out of the memory array in the respective bank. A10 sampled during a PRECHARGE command determines whether the PRECHARGE applies to one bank (A10 LOW, bank selected by BA[2:0]) or all banks (A10 HIGH). The address inputs also provide the op-code during a LOAD MODE command. Address inputs are referenced to V_{REFCA} . A12/BC#: When enabled in the mode register (MR), A12 is sampled during READ and WRITE commands to determine whether burst chop (on-the-fly) will be performed (HIGH = BL8 or no burst chop, LOW = BC4 burst chop). See Truth Table - Command.
BA[2:0]	Input	Bank address inputs: BA[2:0] define the bank to which an ACTIVATE, READ, WRITE, or PRECHARGE command is being applied. BA[2:0] define which mode register (MR0, MR1, MR2, or MR3) is loaded during the LOAD MODE command. BA[2:0] are referenced to V_{REFCA} .
CK, CK#	Input	Clock: CK and CK# are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and the negative edge of CK#. Output data strobe (LDQS, LDQS#, UDQS, UDQS#) is referenced to the crossings of CK and CK#.
CKE	Input	Clock enable: CKE enables (registered HIGH) and disables (registered LOW) internal circuitry and clocks on the DRAM. The specific circuitry that is enabled/disabled is dependent upon the DDR3 SDRAM configuration and operating mode. Taking CKE LOW provides PRECHARGE power-down and SELF REFRESH operations (all banks idle) or active power-down (row active in any bank). CKE is synchronous for power-down entry and exit and for self refresh entry. CKE is asynchronous for self refresh exit. Input buffers (excluding CK, CK#, CKE, RESET#, and ODT) are disabled during power-down. Input buffers (excluding CKE and RESET#) are disabled during SELF REFRESH. CKE is referenced to V_{REFCA} .
CS#	Input	Chip select: CS# enables (registered LOW) and disables (registered HIGH) the command decoder. All commands are masked when CS# is registered HIGH. CS# provides for external rank selection on systems with multiple ranks. CS# is considered part of the command code. CS# is referenced to V_{REFCA} .
LDM	Input	Input data mask: LDM is a lower-byte, input mask signal for write data. Lower-byte input data is masked when LDM is sampled HIGH along with the input data during a write access. Although the LDM ball is input-only, the LDM loading is designed to match that of the DQ and LDQS balls. LDM is referenced to V_{REFDQ} .
ODT	Input	On-die termination: ODT enables (registered HIGH) and disables (registered LOW) termination resistance internal to the DDR3 SDRAM. When enabled in normal operation, ODT is only applied to each of the following balls: DQ[15:0], LDQS, LDQS#, UDQS, UDQS#, LDM, and UDM for the x16. The ODT input is ignored if disabled via the LOAD MODE command. ODT is referenced to V_{REFCA} .
RAS#, CAS#, WE#	Input	Command inputs: RAS#, CAS#, and WE# (along with CS#) define the command being entered and are referenced to V_{REFCA} .
RESET#	Input	Reset: RESET# is an active LOW CMOS input referenced to V_{SS} . The RESET# input receiver is a CMOS input defined as a rail-to-rail signal with DC HIGH $\geq 0.8 \times V_{DDQ}$ and DC LOW $\leq 0.2 \times V_{DDQ}$. RESET# assertion and deassertion are asynchronous.

Table 4: 96-Ball FBGA – x16 Ball Descriptions (Continued)

Symbol	Type	Description
UDM	Input	Input data mask: UDM is an upper-byte, input mask signal for write data. Upper-byte input data is masked when UDM is sampled HIGH along with the input data during a write access. Although the UDM ball is input-only, the UDM loading is designed to match that of the DQ and UDQS balls. UDM is referenced to V_{REFDQ} .
DQ[7:0]	I/O	Data input/output: Lower byte of bidirectional data bus for the x16 configuration. DQ[7:0] are referenced to V_{REFDQ} .
DQ[15:8]	I/O	Data input/output: Upper byte of bidirectional data bus for the x16 configuration. DQ[15:8] are referenced to V_{REFDQ} .
LDQS, LDQS#	I/O	Lower byte data strobe: Output with read data. Edge-aligned with read data. Input with write data. LDQS is center-aligned to write data.
UDQS, UDQS#	I/O	Upper byte data strobe: Output with read data. Edge-aligned with read data. Input with write data. UDQS is center-aligned to write data.
V_{DD}	Supply	Power supply: 1.35V, 1.283–1.45V operational; compatible to 1.5V operation.
V_{DDQ}	Supply	DQ power supply: 1.35V, 1.283–1.45V operational; compatible with 1.5V operation.
V_{REFCA}	Supply	Reference voltage for control, command, and address: V_{REFCA} must be maintained at all times (including self refresh) for proper device operation.
V_{REFDQ}	Supply	Reference voltage for data: V_{REFDQ} must be maintained at all times (including self refresh) for proper device operation.
V_{SS}	Supply	Ground.
V_{SSQ}	Supply	DQ ground: Isolated on the device for improved noise immunity.
ZQ	Reference	External reference ball for output drive calibration: This ball is tied to an external 240 Ω resistor (R_{ZQ}), which is tied to V_{SSQ} .
NC	–	No connect: These balls should be left unconnected (the ball has no connection to the DRAM or to other balls).

Package Dimensions

Figure 8: 78-Ball FBGA – x4, x8 (DA)

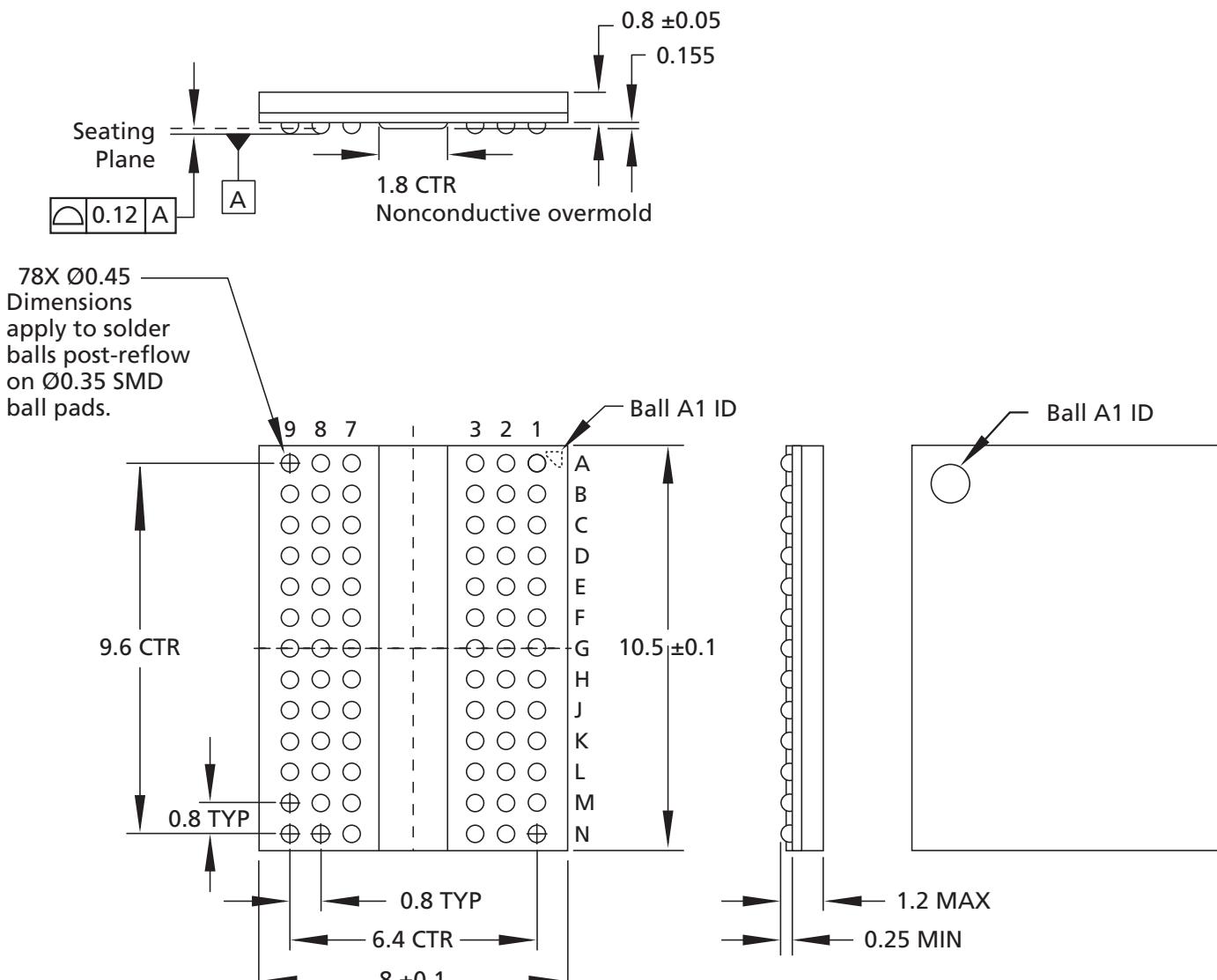


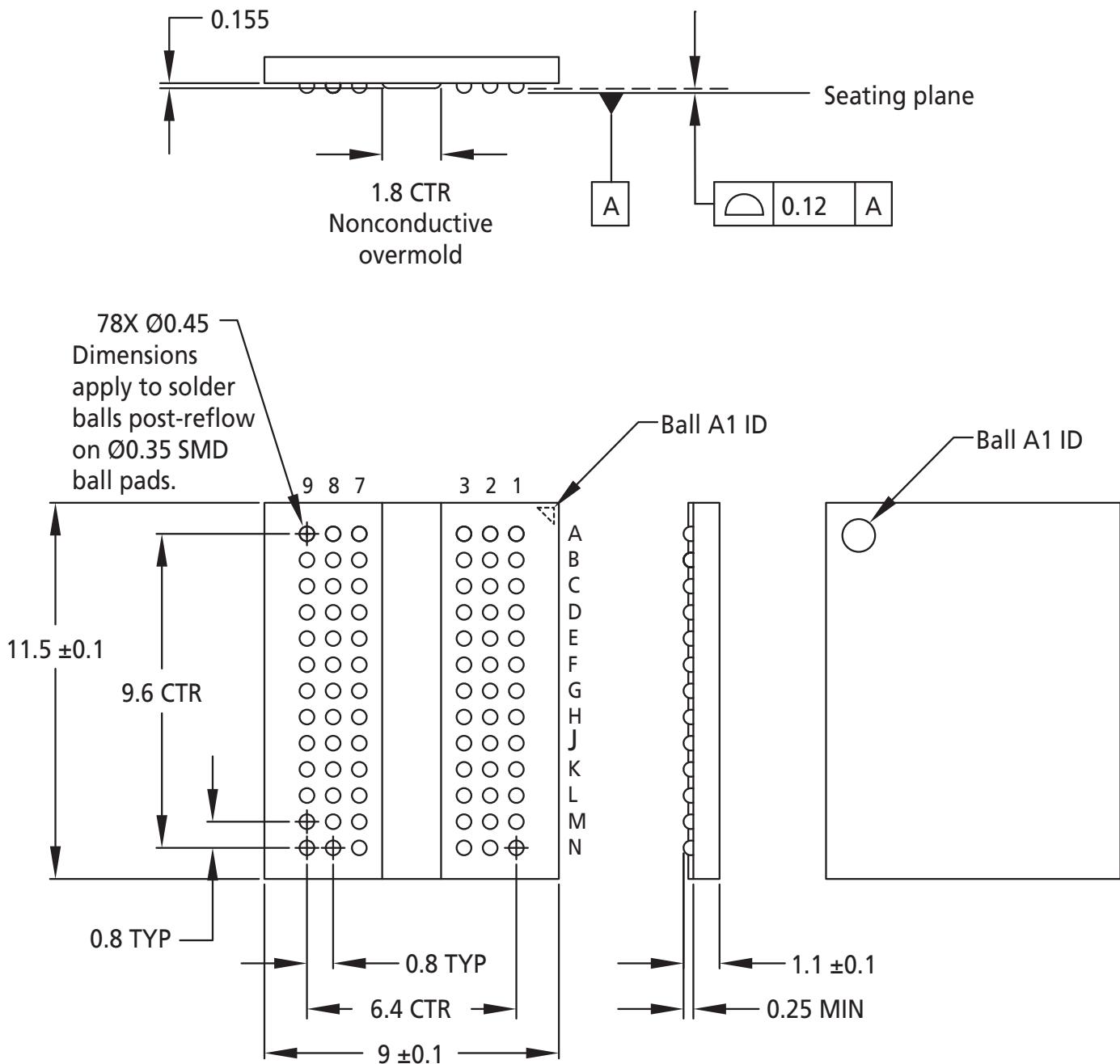
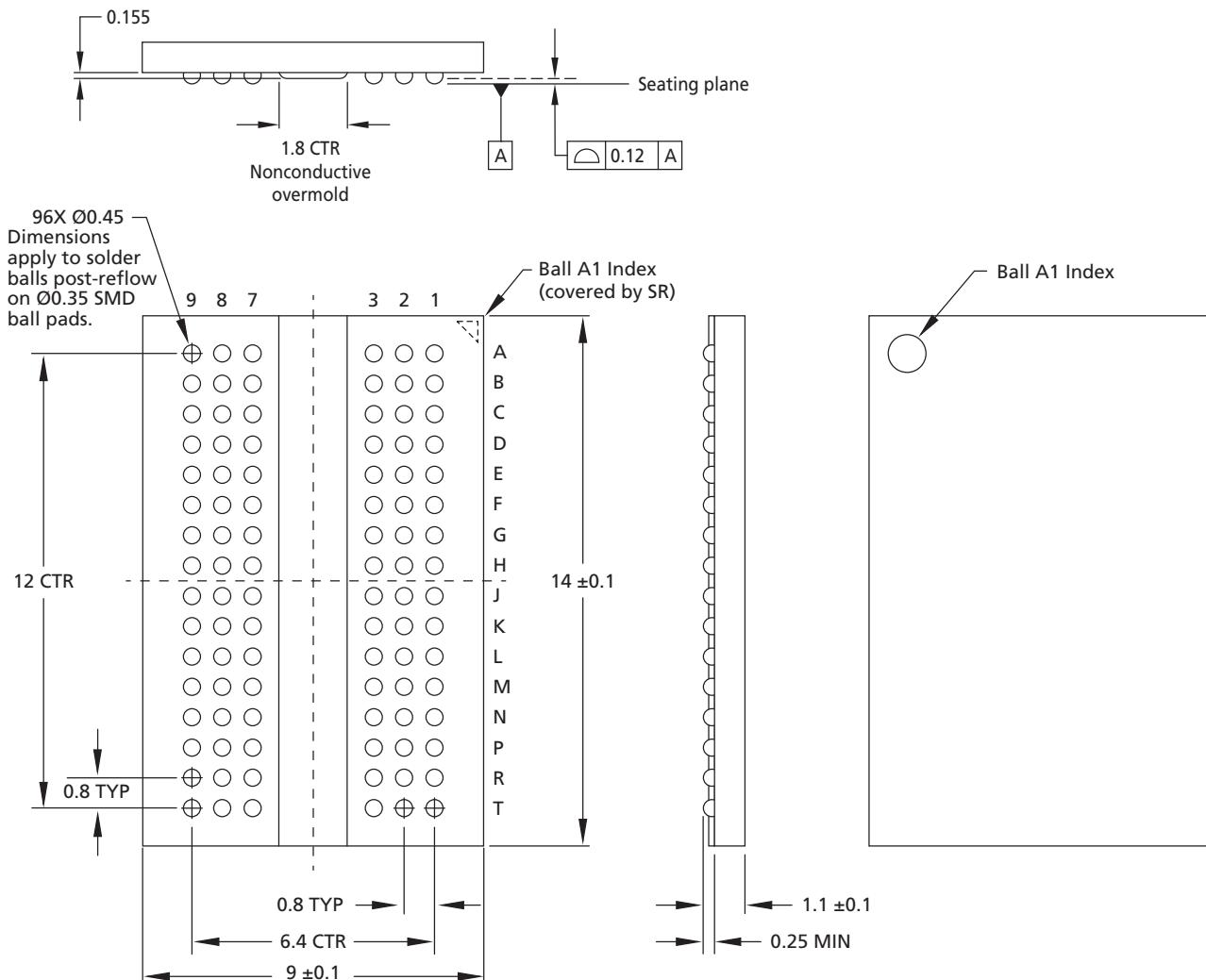
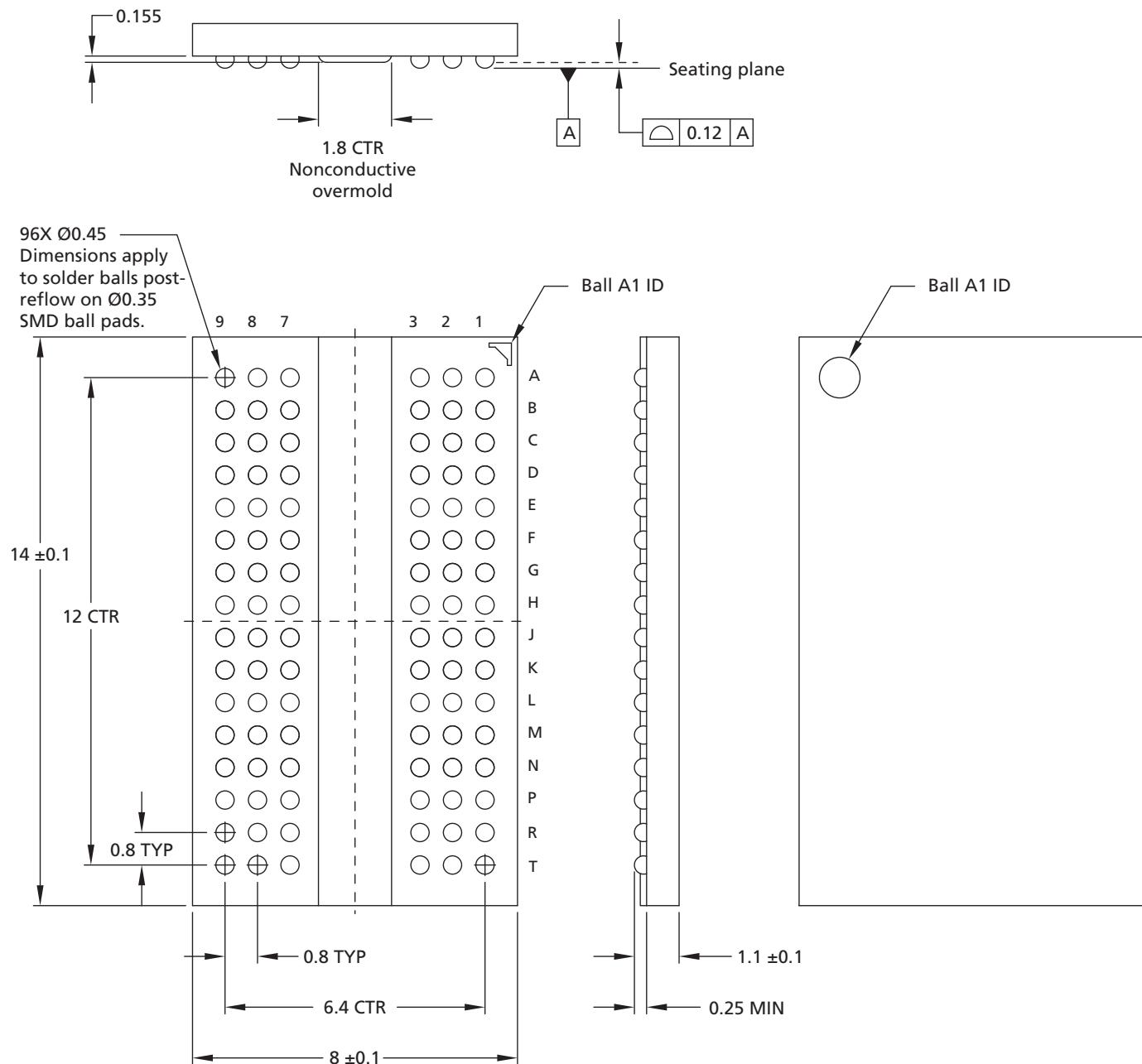
Figure 9: 78-Ball FBGA – x4, x8 (HX)


Figure 10: 96-Ball FBGA – x16 (HA)


Notes: 1. All dimensions are in millimeters.
 2. Solder ball material: SAC305 (96.5% Sn, 3% Ag, 0.5% Cu).

Figure 11: 96-Ball FBGA – x16 (JT)


Notes: 1. All dimensions are in millimeters.
 2. Solder ball material: SAC305 (96.5% Sn, 3% Ag, 0.5% Cu).

Electrical Specifications

Absolute Ratings

Stresses greater than those listed may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may adversely affect reliability.

Table 5: Absolute Maximum Ratings

Symbol	Parameter	Min	Max	Unit	Notes
V_{DD}	V_{DD} supply voltage relative to V_{SS}	-0.4	1.975	V	1
V_{DDQ}	V_{DD} supply voltage relative to V_{SSQ}	-0.4	1.975	V	
V_{IN}, V_{OUT}	Voltage on any pin relative to V_{SS}	-0.4	1.975	V	
T_C	Operating case temperature – Commercial	0	95	°C	2, 3
	Operating case temperature – Industrial	-40	95	°C	2, 3
	Operating case temperature – Automotive	-40	105	°C	2, 3
T_{STG}	Storage temperature	-55	150	°C	

Notes:

1. V_{DD} and V_{DDQ} must be within 300mV of each other at all times, and V_{REF} must not be greater than $0.6 \times V_{DDQ}$. When V_{DD} and V_{DDQ} are <500 mV, V_{REF} can be ≤ 300 mV.
2. MAX operating case temperature. T_C is measured in the center of the package.
3. Device functionality is not guaranteed if the DRAM device exceeds the maximum T_C during operation.

Input/Output Capacitance

Table 6: DDR3L Input/Output Capacitance

Note 1 applies to the entire table; gray-shaded cells are DDR3L unique values; all other values are the same for both DDR3L and DDR3

Capacitance Parameters	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		DDR3L-1866		Unit	Notes
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max		
CK and CK#	C_{CK}	0.8	1.6	0.8	1.6	0.8	1.4	0.8	1.4	0.8	1.3	pF	
ΔC : CK to CK#	C_{DCK}	0.0	0.15	0.0	0.15	0.0	0.15	0.0	0.15	0.0	0.15	pF	
Single-end I/O: DQ, DM	C_{IO}	1.4	2.5	1.4	2.5	1.4	2.3	1.4	2.2	1.4	2.1	pF	2
Differential I/O: DQS, DQS#, TDQS, TDQS#	C_{IO}	1.4	2.5	1.4	2.5	1.4	2.3	1.4	2.2	1.4	2.1	pF	3
ΔC : DQS to DQS#, TDQS, TDQS#	C_{DDQS}	0.0	0.2	0.0	0.2	0.0	0.15	0.0	0.15	0.0	0.15	pF	3
ΔC : DQ to DQS	C_{DIO}	-0.5	0.3	-0.5	0.3	-0.5	0.3	-0.5	0.3	-0.5	0.3	pF	4
Inputs (CTRL, CMD, ADDR)	C_I	0.75	1.3	0.75	1.3	0.75	1.3	0.75	1.2	0.75	1.2	pF	5
ΔC : CTRL to CK	C_{DI_CTRL}	-0.5	0.3	-0.5	0.3	-0.4	0.2	-0.4	0.2	-0.4	0.2	pF	6
ΔC : CMD_ADDR to CK	$C_{DI_CMD_ADDR}$	-0.5	0.5	-0.5	0.5	-0.4	0.4	-0.4	0.4	-0.4	0.4	pF	7
ZQ pin capacitance	C_{ZQ}	-	3.0	-	3.0	-	3.0	-	3.0	-	3.0	pF	
Reset pin capacitance	C_{RE}	-	3.0	-	3.0	-	3.0	-	3.0	-	3.0	pF	

Notes: 1. $V_{DD} = 1.35V$ (1.283–1.45V), $V_{DDQ} = V_{DD}$, $V_{REF} = V_{SS}$, $f = 100$ MHz, $T_C = 25^\circ C$. $V_{OUT(DC)} = 0.5 \times V_{DDQ}$, $V_{OUT} = 0.1V$ (peak-to-peak).

2. DM input is grouped with I/O pins, reflecting the fact that they are matched in loading.
3. Includes TDQS, TDQS#. C_{DDQS} is for DQS vs. DQS# and TDQS vs. TDQS# separately.
4. $C_{DIO} = C_{IO(DQ)} - 0.5 \times (C_{IO(DQS)} + C_{IO(DQS#)})$.
5. Excludes CK, CK#; CTRL = ODT, CS#, and CKE; CMD = RAS#, CAS#, and WE#; ADDR = A[n:0], BA[2:0].
6. $C_{DI_CTRL} = C_{I(CTRL)} - 0.5 \times (C_{CK(CK)} + C_{CK(CK#)})$.
7. $C_{DI_CMD_ADDR} = C_{I(CMD_ADDR)} - 0.5 \times (C_{CK(CK)} + C_{CK(CK#)})$.

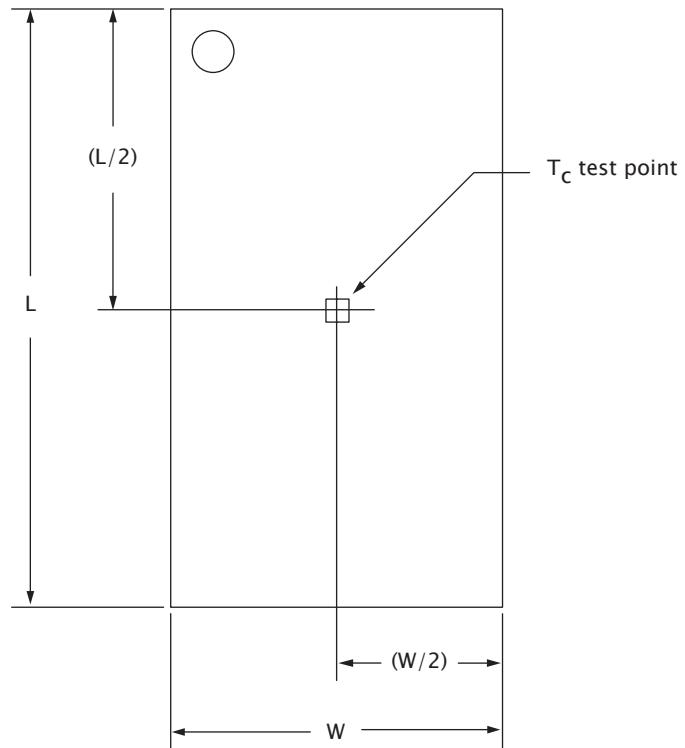
Thermal Characteristics

Table 7: Thermal Characteristics

Parameter/Condition		Value	Unit	Symbol	Notes
Operating case temperature – Commercial	0 to +85	°C	T_C	T_C	1, 2, 3
	0 to +95	°C	T_C		1, 2, 3, 4
Operating case temperature – Industrial	–40 to +85	°C	T_C	T_C	1, 2, 3
	–40 to +95	°C	T_C		1, 2, 3, 4
Operating case temperature – Automotive	–40 to +85	°C	T_C	T_C	1, 2, 3
	–40 to +105	°C	T_C		1, 2, 3, 4
Junction-to-case (TOP)	96-ball (JT)	6.5	°C/W	Θ_{JC}	5
	96-ball (HA)	3.9			
	78-ball (DA) M, K	6.5			
	78-ball (HX)	3.9			

Notes:

1. Maximum operating case temperature. T_C is measured in the center of the package.
2. A thermal solution must be designed to ensure the DRAM device does not exceed T_C MAX during operation.
3. Device functionality is not guaranteed if the DRAM device exceeds T_C MAX during operation.
4. If T_C exceeds 85°C, the DRAM must be refreshed externally at 2x refresh, which is a 3.9µs interval refresh rate. The use of SRT or ASR (if available) must be enabled.
5. Thermal resistance data is based on a number of samples from multiple lots and should be viewed as a typical number.

Figure 12: Thermal Measurement Point

Electrical Specifications – I_{DD} Specifications and Conditions

Within the following I_{DD} measurement tables, the following definitions and conditions are used, unless stated otherwise:

- LOW: V_{IN} ≤ V_{IL(AC)max}; HIGH: V_{IN} ≥ V_{IH(AC)min}.
- Midlevel: Inputs are V_{REF} = V_{DD}/2.
- R_{ON} set to RZQ/7 (34Ω).
- R_{TT,nom} set to RZQ/6 (40Ω).
- R_{TT(WR)} set to RZQ/2 (120Ω).
- Q_{OFF} is enabled in MR1.
- ODT is enabled in MR1 (R_{TT,nom}) and MR2 (R_{TT(WR)}).
- TDQS is disabled in MR1.
- External DQ/DQS/DM load resistor is 25Ω to V_{DDQ}/2.
- Burst lengths are BL8 fixed.
- AL equals 0 (except in I_{DD7}).
- I_{DD} specifications are tested after the device is properly initialized.
- Input slew rate is specified by AC parametric test conditions.
- Optional ASR is disabled.
- Read burst type uses nibble sequential (MR0[3] = 0).
- Loop patterns must be executed at least once before current measurements begin.

Table 8: DDR3L Timing Parameters Used for I_{DD} Measurements – Clock Units

I _{DD} Parameter	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		DDR3L-1866		Unit
	-25E	-25	-187E	-187	-15E	-15	-125E	-125	-107		
	5-5-5	6-6-6	7-7-7	8-8-8	9-9-9	10-10-10	10-10-10	11-11-11	13-13-13		
t _{CK} (MIN) I _{DD}	2.5		1.875		1.5		1.25		1.071	ns	
CL I _{DD}	5	6	7	8	9	10	10	11	13	CK	
t _{RCD} (MIN) I _{DD}	5	6	7	8	9	10	10	11	13	CK	
t _{RC} (MIN) I _{DD}	20	21	27	28	33	34	38	39	45	CK	
t _{RAS} (MIN) I _{DD}	15	15	20	20	24	24	28	28	32	CK	
t _{RP} (MIN)	5	6	7	8	9	10	10	11	13	CK	
t _{FAW}	x4, x8	16	16	20	20	20	24	24	26	CK	
	x16	20	20	27	27	30	32	32	33	CK	
t _{RRD} I _{DD}	x4, x8	4	4	4	4	4	5	5	5	CK	
	x16	4	4	6	6	5	6	6	6	CK	
t _{RFC}	1Gb	44	44	59	59	74	88	88	103	CK	
	2Gb	64	64	86	86	107	128	128	150	CK	
	4Gb	104	104	139	139	174	208	208	243	CK	
	8Gb	140	140	187	187	234	280	280	328	CK	

Table 9: DDR3L I_{DD0} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data														
Toggling Static HIGH	0		0	ACT	0	0	1	1	0	0	0	0	0	0	0	–														
			1	D	1	0	0	0	0	0	0	0	0	0	0	–														
			2	D	1	0	0	0	0	0	0	0	0	0	0	–														
			3	D#	1	1	1	1	0	0	0	0	0	0	0	–														
			4	D#	1	1	1	1	0	0	0	0	0	0	0	–														
		Repeat cycles 1 through 4 until nRAS - 1; truncate if needed																												
			nRAS	PRE	0	0	1	0	0	0	0	0	0	0	0	–														
			Repeat cycles 1 through 4 until nRC - 1; truncate if needed																											
			nRC	ACT	0	0	1	1	0	0	0	0	0	F	0	–														
			nRC + 1	D	1	0	0	0	0	0	0	0	0	F	0	–														
			nRC + 2	D	1	0	0	0	0	0	0	0	0	F	0	–														
			nRC + 3	D#	1	1	1	1	0	0	0	0	0	F	0	–														
			nRC + 4	D#	1	1	1	1	0	0	0	0	0	F	0	–														
			Repeat cycles nRC + 1 through nRC + 4 until nRC - 1 + nRAS - 1; truncate if needed																											
			nRC + nRAS	PRE	0	0	1	0	0	0	0	0	0	F	0	–														
			Repeat cycles nRC + 1 through nRC + 4 until 2 × RC - 1; truncate if needed																											
1	2 × nRC	Repeat sub-loop 0, use BA[2:0] = 1																												
		Repeat sub-loop 0, use BA[2:0] = 2																												
		Repeat sub-loop 0, use BA[2:0] = 3																												
		Repeat sub-loop 0, use BA[2:0] = 4																												
		Repeat sub-loop 0, use BA[2:0] = 5																												
		Repeat sub-loop 0, use BA[2:0] = 6																												
		Repeat sub-loop 0, use BA[2:0] = 7																												

Notes:

1. DQ, DQS, DQS# are midlevel.
2. DM is LOW.
3. Only selected bank (single) active.

Table 10: DDR3L I_{DD1} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data ²	
Toggling	Static HIGH	0	0	ACT	0	0	1	1	0	0	0	0	0	0	0	–	
			1	D	1	0	0	0	0	0	0	0	0	0	0	–	
			2	D	1	0	0	0	0	0	0	0	0	0	0	–	
			3	D#	1	1	1	1	0	0	0	0	0	0	0	–	
			4	D#	1	1	1	1	0	0	0	0	0	0	0	–	
			Repeat cycles 1 through 4 until nRCD - 1; truncate if needed														
			nRCD	RD	0	1	0	1	0	0	0	0	0	0	0	00000000	
			Repeat cycles 1 through 4 until nRAS - 1; truncate if needed														
			nRAS	PRE	0	0	1	0	0	0	0	0	0	0	0	–	
			Repeat cycles 1 through 4 until nRC - 1; truncate if needed														
			nRC	ACT	0	0	1	1	0	0	0	0	0	F	0	–	
			nRC + 1	D	1	0	0	0	0	0	0	0	0	F	0	–	
			nRC + 2	D	1	0	0	0	0	0	0	0	0	F	0	–	
			nRC + 3	D#	1	1	1	1	0	0	0	0	0	F	0	–	
			nRC + 4	D#	1	1	1	1	0	0	0	0	0	F	0	–	
			Repeat cycles nRC + 1 through nRC + 4 until nRC + nRCD - 1; truncate if needed														
			nRC + nRCD	RD	0	1	0	1	0	0	0	0	0	F	0	00110011	
			Repeat cycles nRC + 1 through nRC + 4 until nRC + nRAS - 1; truncate if needed														
			nRC + nRAS	PRE	0	0	1	0	0	0	0	0	0	F	0	–	
			Repeat cycle nRC + 1 through nRC + 4 until 2 × nRC - 1; truncate if needed														
			1	2 × nRC	Repeat sub-loop 0, use BA[2:0] = 1												
			2	4 × nRC	Repeat sub-loop 0, use BA[2:0] = 2												
			3	6 × nRC	Repeat sub-loop 0, use BA[2:0] = 3												
			4	8 × nRC	Repeat sub-loop 0, use BA[2:0] = 4												
			5	10 × nRC	Repeat sub-loop 0, use BA[2:0] = 5												
			6	12 × nRC	Repeat sub-loop 0, use BA[2:0] = 6												
			7	14 × nRC	Repeat sub-loop 0, use BA[2:0] = 7												

Notes:

1. DQ, DQS, DQS# are midlevel unless driven as required by the RD command.
2. DM is LOW.
3. Burst sequence is driven on each DQ signal by the RD command.
4. Only selected bank (single) active.

Table 11: DDR3L I_{DD} Measurement Conditions for Power-Down Currents

Name	I_{DD2P0} Precharge Power-Down Current (Slow Exit) ¹	I_{DD2P1} Precharge Power-Down Current (Fast Exit) ¹	I_{DD2Q} Precharge Quiet Standby Current	I_{DD3P} Active Power-Down Current
Timing pattern	N/A	N/A	N/A	N/A
CKE	LOW	LOW	HIGH	LOW
External clock	Toggling	Toggling	Toggling	Toggling
t_{CK}	t_{CK} (MIN) I_{DD}	t_{CK} (MIN) I_{DD}	t_{CK} (MIN) I_{DD}	t_{CK} (MIN) I_{DD}
t_{RC}	N/A	N/A	N/A	N/A
t_{RAS}	N/A	N/A	N/A	N/A
t_{RCD}	N/A	N/A	N/A	N/A
t_{RRD}	N/A	N/A	N/A	N/A
t_{RC}	N/A	N/A	N/A	N/A
CL	N/A	N/A	N/A	N/A
AL	N/A	N/A	N/A	N/A
CS#	HIGH	HIGH	HIGH	HIGH
Command inputs	LOW	LOW	LOW	LOW
Row/column addr	LOW	LOW	LOW	LOW
Bank addresses	LOW	LOW	LOW	LOW
DM	LOW	LOW	LOW	LOW
Data I/O	Midlevel	Midlevel	Midlevel	Midlevel
Output buffer DQ, DQS	Enabled	Enabled	Enabled	Enabled
ODT ²	Enabled, off	Enabled, off	Enabled, off	Enabled, off
Burst length	8	8	8	8
Active banks	None	None	None	All
Idle banks	All	All	All	None
Special notes	N/A	N/A	N/A	N/A

Notes:

1. MR0[12] defines DLL on/off behavior during precharge power-down only; DLL on (fast exit, MR0[12] = 1) and DLL off (slow exit, MR0[12] = 0).
2. "Enabled, off" means the MR bits are enabled, but the signal is LOW.

Table 12: DDR3L I_{DD2N} and I_{DD3N} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data
Toggling	Static HIGH	0	0	D	1	0	0	0	0	0	0	0	0	0	0	–
			1	D	1	0	0	0	0	0	0	0	0	0	0	–
			2	D#	1	1	1	1	0	0	0	0	0	F	0	–
			3	D#	1	1	1	1	0	0	0	0	0	F	0	–
		1	4–7	Repeat sub-loop 0, use BA[2:0] = 1												
			8–11	Repeat sub-loop 0, use BA[2:0] = 2												
			12–15	Repeat sub-loop 0, use BA[2:0] = 3												
			16–19	Repeat sub-loop 0, use BA[2:0] = 4												
			20–23	Repeat sub-loop 0, use BA[2:0] = 5												
			24–27	Repeat sub-loop 0, use BA[2:0] = 6												
			28–31	Repeat sub-loop 0, use BA[2:0] = 7												

Notes: 1. DQ, DQS, DQS# are midlevel.
 2. DM is LOW.
 3. All banks closed during I_{DD2N}; all banks open during I_{DD3N}.

Table 13: DDR3L I_{DD2NT} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data
Toggling	Static HIGH	0	0	D	1	0	0	0	0	0	0	0	0	0	0	–
			1	D	1	0	0	0	0	0	0	0	0	0	0	–
			2	D#	1	1	1	1	0	0	0	0	0	F	0	–
			3	D#	1	1	1	1	0	0	0	0	0	F	0	–
		1	4–7	Repeat sub-loop 0, use BA[2:0] = 1; ODT = 0												
			8–11	Repeat sub-loop 0, use BA[2:0] = 2; ODT = 1												
			12–15	Repeat sub-loop 0, use BA[2:0] = 3; ODT = 1												
			16–19	Repeat sub-loop 0, use BA[2:0] = 4; ODT = 0												
			20–23	Repeat sub-loop 0, use BA[2:0] = 5; ODT = 0												
			24–27	Repeat sub-loop 0, use BA[2:0] = 6; ODT = 1												
			28–31	Repeat sub-loop 0, use BA[2:0] = 7; ODT = 1												

Notes: 1. DQ, DQS, DQS# are midlevel.
 2. DM is LOW.
 3. All banks closed.

Table 14: DDR3L I_{DD4R} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data ³
Toggling Static HIGH		0	0	RD	0	1	0	1	0	0	0	0	0	0	0	00000000
			1	D	1	0	0	0	0	0	0	0	0	0	0	–
			2	D#	1	1	1	1	0	0	0	0	0	0	0	–
			3	D#	1	1	1	1	0	0	0	0	0	0	0	–
			4	RD	0	1	0	1	0	0	0	0	0	F	0	00110011
			5	D	1	0	0	0	0	0	0	0	0	F	0	–
			6	D#	1	1	1	1	0	0	0	0	0	F	0	–
			7	D#	1	1	1	1	0	0	0	0	0	F	0	–
		1	8–15	Repeat sub-loop 0, use BA[2:0] = 1												
			16–23	Repeat sub-loop 0, use BA[2:0] = 2												
			24–31	Repeat sub-loop 0, use BA[2:0] = 3												
			32–39	Repeat sub-loop 0, use BA[2:0] = 4												
			40–47	Repeat sub-loop 0, use BA[2:0] = 5												
			48–55	Repeat sub-loop 0, use BA[2:0] = 6												
			56–63	Repeat sub-loop 0, use BA[2:0] = 7												

Notes:

1. DQ, DQS, DQS# are midlevel when not driving in burst sequence.
2. DM is LOW.
3. Burst sequence is driven on each DQ signal by the RD command.
4. All banks open.

Table 15: DDR3L I_{DD4W} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data ³
Toggling Static HIGH		0	0	WR	0	1	0	0	1	0	0	0	0	0	0	00000000
			1	D	1	0	0	0	1	0	0	0	0	0	0	–
			2	D#	1	1	1	1	1	0	0	0	0	0	0	–
			3	D#	1	1	1	1	1	0	0	0	0	0	0	–
			4	WR	0	1	0	0	1	0	0	0	0	F	0	00110011
			5	D	1	0	0	0	1	0	0	0	0	F	0	–
			6	D#	1	1	1	1	1	0	0	0	0	F	0	–
			7	D#	1	1	1	1	1	0	0	0	0	F	0	–
		1	8–15	Repeat sub-loop 0, use BA[2:0] = 1												
			16–23	Repeat sub-loop 0, use BA[2:0] = 2												
			24–31	Repeat sub-loop 0, use BA[2:0] = 3												
			32–39	Repeat sub-loop 0, use BA[2:0] = 4												
			40–47	Repeat sub-loop 0, use BA[2:0] = 5												
			48–55	Repeat sub-loop 0, use BA[2:0] = 6												
			56–63	Repeat sub-loop 0, use BA[2:0] = 7												

Notes:

1. DQ, DQS, DQS# are midlevel when not driving in burst sequence.
2. DM is LOW.
3. Burst sequence is driven on each DQ signal by the WR command.
4. All banks open.

Table 16: DDR3L I_{DD5B} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data
Toggling Static HIGH		0	0	REF	0	0	0	1	0	0	0	0	0	0	0	0
		1a	1	D	1	0	0	0	0	0	0	0	0	0	0	0
			2	D	1	0	0	0	0	0	0	0	0	0	0	0
			3	D#	1	1	1	1	0	0	0	0	0	F	0	–
			4	D#	1	1	1	1	0	0	0	0	0	F	0	–
			1b	5–8	Repeat sub-loop 1a, use BA[2:0] = 1											
			1c	9–12	Repeat sub-loop 1a, use BA[2:0] = 2											
			1d	13–16	Repeat sub-loop 1a, use BA[2:0] = 3											
			1e	17–20	Repeat sub-loop 1a, use BA[2:0] = 4											
			1f	21–24	Repeat sub-loop 1a, use BA[2:0] = 5											
			1g	25–28	Repeat sub-loop 1a, use BA[2:0] = 6											
			1h	29–32	Repeat sub-loop 1a, use BA[2:0] = 7											
		2	33–nRFC - 1	Repeat sub-loop 1a through 1h until nRFC - 1; truncate if needed												

Notes: 1. DQ, DQS, DQS# are midlevel.
2. DM is LOW.

Table 17: DDR3L I_{DD} Measurement Conditions for I_{DD6} , I_{DD6ET} , and I_{DD8}

I_{DD} Test	I_{DD6} : Self Refresh Current Normal Temperature Range $T_C = 0^\circ\text{C}$ to $+85^\circ\text{C}$	I_{DD6ET} : Self Refresh Current Extended Temperature Range $T_C = 0^\circ\text{C}$ to $+95^\circ\text{C}$	I_{DD8} : Reset ²
CKE	LOW	LOW	Midlevel
External clock	Off, CK and CK# = LOW	Off, CK and CK# = LOW	Midlevel
t_{CK}	N/A	N/A	N/A
t_{RC}	N/A	N/A	N/A
t_{RAS}	N/A	N/A	N/A
t_{RCD}	N/A	N/A	N/A
t_{RRD}	N/A	N/A	N/A
t_{RC}	N/A	N/A	N/A
CL	N/A	N/A	N/A
AL	N/A	N/A	N/A
CS#	Midlevel	Midlevel	Midlevel
Command inputs	Midlevel	Midlevel	Midlevel
Row/column addresses	Midlevel	Midlevel	Midlevel
Bank addresses	Midlevel	Midlevel	Midlevel
Data I/O	Midlevel	Midlevel	Midlevel
Output buffer DQ, DQS	Enabled	Enabled	Midlevel
ODT ¹	Enabled, midlevel	Enabled, midlevel	Midlevel
Burst length	N/A	N/A	N/A
Active banks	N/A	N/A	None
Idle banks	N/A	N/A	All
SRT	Disabled (normal)	Enabled (extended)	N/A
ASR	Disabled	Disabled	N/A

Notes:

1. “Enabled, midlevel” means the MR command is enabled, but the signal is midlevel.
2. During a cold boot RESET (initialization), current reading is valid after power is stable and RESET has been LOW for 1ms; During a warm boot RESET (while operating), current reading is valid after RESET has been LOW for $200\text{ns} + t_{RFC}$.

Table 18: DDR3L I_{DD7} Measurement Loop

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data ³
Toggle Static HIGH		0	0	ACT	0	0	1	1	0	0	0	0	0	0	0	–
			1	RDA	0	1	0	1	0	0	0	1	0	0	0	00000000
			2	D	1	0	0	0	0	0	0	0	0	0	0	–
			3	Repeat cycle 2 until nRRD - 1												
		1	nRRD	ACT	0	0	1	1	0	1	0	0	0	F	0	–
			nRRD + 1	RDA	0	1	0	1	0	1	0	1	0	F	0	00110011
			nRRD + 2	D	1	0	0	0	0	1	0	0	0	F	0	–
			nRRD + 3	Repeat cycle nRRD + 2 until 2 × nRRD - 1												
		2	2 × nRRD	Repeat sub-loop 0, use BA[2:0] = 2												
		3	3 × nRRD	Repeat sub-loop 1, use BA[2:0] = 3												
		4	4 × nRRD	D	1	0	0	0	0	3	0	0	0	F	0	–
			4 × nRRD + 1	Repeat cycle 4 × nRRD until nFAW - 1, if needed												
		5	nFAW	Repeat sub-loop 0, use BA[2:0] = 4												
		6	nFAW + nRRD	Repeat sub-loop 1, use BA[2:0] = 5												
		7	nFAW + 2 × nRRD	Repeat sub-loop 0, use BA[2:0] = 6												
		8	nFAW + 3 × nRRD	Repeat sub-loop 1, use BA[2:0] = 7												
		9	nFAW + 4 × nRRD	D	1	0	0	0	0	7	0	0	0	F	0	–
			nFAW + 4 × nRRD + 1	Repeat cycle nFAW + 4 × nRRD until 2 × nFAW - 1, if needed												
		10	2 × nFAW	ACT	0	0	1	1	0	0	0	0	0	F	0	–
			2 × nFAW + 1	RDA	0	1	0	1	0	0	0	1	0	F	0	00110011
			2 × nFAW + 2	D	1	0	0	0	0	0	0	0	0	F	0	–
			2 × nFAW + 3	Repeat cycle 2 × nFAW + 2 until 2 × nFAW + nRRD - 1												
		11	2 × nFAW + nRRD	ACT	0	0	1	1	0	1	0	0	0	0	0	–
			2 × nFAW + nRRD + 1	RDA	0	1	0	1	0	1	0	1	0	0	0	00000000
			2 × nFAW + nRRD + 2	D	1	0	0	0	0	1	0	0	0	0	0	–
			2 × nFAW + nRRD + 3	Repeat cycle 2 × nFAW + nRRD + 2 until 2 × nFAW + 2 × nRRD - 1												
		12	2 × nFAW + 2 × nRRD	Repeat sub-loop 10, use BA[2:0] = 2												
		13	2 × nFAW + 3 × nRRD	Repeat sub-loop 11, use BA[2:0] = 3												
		14	2 × nFAW + 4 × nRRD	D	1	0	0	0	0	3	0	0	0	0	0	–
			2 × nFAW + 4 × nRRD + 1	Repeat cycle 2 × nFAW + 4 × nRRD until 3 × nFAW - 1, if needed												
		15	3 × nFAW	Repeat sub-loop 10, use BA[2:0] = 4												

Table 18: DDR3L I_{DD7} Measurement Loop (Continued)

CK, CK#	CKE	Sub-Loop	Cycle Number	Command	CS#	RAS#	CAS#	WE#	ODT	BA[2:0]	A[15:11]	A[10]	A[9:7]	A[6:3]	A[2:0]	Data ³	
Toggling Static HIGH		16	3 × nFAW + nRRD														Repeat sub-loop 11, use BA[2:0] = 5
		17	3 × nFAW + 2 × nRRD														Repeat sub-loop 10, use BA[2:0] = 6
		18	3 × nFAW + 3 × nRRD														Repeat sub-loop 11, use BA[2:0] = 7
		19	3 × nFAW + 4 × nRRD	D	1	0	0	0	0	7	0	0	0	0	0	–	
			3 × nFAW + 4 × nRRD + 1														Repeat cycle 3 × nFAW + 4 × nRRD until 4 × nFAW - 1, if needed

Notes:

1. DQ, DQS, DQS# are midlevel unless driven as required by the RD command.
2. DM is LOW.
3. Burst sequence is driven on each DQ signal by the RD command.
4. AL = CL-1.

Electrical Characteristics – I_{DD} Specifications

Table 19: I_{DD} Maximum Limits – Die Rev. D

Speed Bin		DDR3L-800	DDR3L-1066	DDR3L-1333	Units
I_{DD}	Width				
I_{DD0}	x4, 8	70	75	85	mA
	x16	85	90	100	mA
I_{DD1}	x4, 8	92	95	100	mA
	x16	122	125	130	mA
I_{DD2P0} (Slow)	All	12	12	12	mA
I_{DD2P1} (Fast)	x4, 8	22	25	30	mA
	x16	27	30	35	mA
I_{DD2Q}	All	27	30	35	mA
I_{DD2N}	All	28	32	37	mA
I_{DD2NT}	x4, 8	37	40	45	mA
	x16	52	55	60	mA
I_{DD3P}	x4, 8	27	30	35	mA
	x16	32	35	40	mA
I_{DD3N}	All	32	35	40	mA
I_{DD4R}	x4	110	125	145	mA
	x8	125	140	160	mA
	x16	160	200	245	mA
I_{DD4W}	x4	120	135	155	mA
	x8	130	145	165	mA
	x16	170	210	255	mA
I_{DD5B}	All	185	190	200	mA
I_{DD6}	All	12	12	12	mA
I_{DD6ET}	All	15	15	15	mA
I_{DD7}	x4, 8	290	335	385	mA
	x16	330	375	425	mA
I_{DD8}	All	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	mA

Notes:

- $T_C = 85^\circ\text{C}$; SRT and ASR are disabled.
- The I_{DD} values must be derated (increased) on IT-option devices when operated outside the range $0^\circ\text{C} \leq T_C \leq +85^\circ\text{C}$:
 - When $T_C < 0^\circ\text{C}$: I_{DD2P0} , I_{DD2P1} and I_{DD3P} must be derated by 4%; I_{DD4R} and I_{DD4W} must be derated by 2%; and I_{DD6} , I_{DD6ET} and I_{DD7} must be derated by 7%.
 - When $T_C > 85^\circ\text{C}$: I_{DD0} , I_{DD1} , I_{DD2N} , I_{DD2NT} , I_{DD2Q} , I_{DD3N} , I_{DD3P} , I_{DD4R} , I_{DD4W} , and I_{DD5B} must be derated by 2%; and I_{DD2Px} must be derated by 30%.

Table 20: I_{DD} Maximum Limits – Die Rev. M

Speed Bin		DDR3L-1066	DDR3L-1333	DDR3L-1600	Unit
I_{DD}	Width				
I_{DD0}	x4, 8	50	55	60	mA
I_{DD1}	x4, 8	65	70	75	mA
I_{DD2P0} (Slow)	x4, 8	12	12	12	mA
I_{DD2P1} (Fast)	x4, 8	23	28	33	mA
I_{DD2Q}	x4, 8	23	28	33	mA
I_{DD2N}	x4, 8	25	30	35	mA
I_{DD2NT}	x4, 8	30	35	40	mA
I_{DD3P}	x4, 8	37	42	47	mA
I_{DD3N}	x4, 8	42	47	52	mA
I_{DD4R}	x4	95	110	125	mA
	x8	110	125	140	mA
I_{DD4W}	x4	85	100	115	mA
	x8	95	110	125	mA
I_{DD5B}	x4, 8	180	185	190	mA
I_{DD6}	x4, 8	12	12	12	mA
I_{DD6ET}	x4, 8	15	15	15	mA
I_{DD7}	x4, 8	190	205	220	mA
I_{DD8}	x4, 8	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	mA

Note: 1. The I_{DD} values must be derated (increased) on IT-option devices when operated outside the range $0^{\circ}\text{C} \leq T_{\text{C}} \leq +85^{\circ}\text{C}$:

a. When $T_{\text{C}} < 0^{\circ}\text{C}$: I_{DD2P0} , I_{DD2P1} and I_{DD3P} must be derated by 4%; I_{DD4R} and I_{DD4W} must be derated by 2%; and I_{DD6} , I_{DD6ET} and I_{DD7} must be derated by 7%.

b. When $T_{\text{C}} > 85^{\circ}\text{C}$: I_{DD0} , I_{DD1} , I_{DD2N} , I_{DD2NT} , I_{DD2Q} , I_{DD3N} , I_{DD3P} , I_{DD4R} , I_{DD4W} , and I_{DD5B} must be derated by 2%; and I_{DD2Px} must be derated by 30%.

Table 21: I_{DD} Maximum Limits – Die Rev. K

Speed Bin		DDR3L-1066	DDR3L-1333	DDR3L-1600	DDR3L-1866	Units	Notes
I_{DD}	Width						
I_{DD0}	x4, x8	36	38	39	40	mA	1, 2
	x16	43	45	46	48	mA	1, 2
I_{DD1}	x4	43	47	49	52	mA	1, 2
	x8	46	50	52	54	mA	1, 2
	x16	58	63	65	68	mA	1, 2
I_{DD2P0} (Slow)	All	12	12	12	12	mA	1, 2
I_{DD2P1} (Fast)	All	14	14	14	14	mA	1, 2
I_{DD2Q}	All	20	20	20	20	mA	1, 2
I_{DD2N}	All	21	21	21	21	mA	1, 2
I_{DD2NT}	x4, x8	26	29	31	33	mA	1, 2
	x16	30	33	34	36	mA	1, 2
I_{DD3P}	All	21	21	21	21	mA	1, 2
I_{DD3N}	x4,x8	28	30	32	34	mA	1, 2
	x16	30	33	34	36	mA	1, 2
I_{DD4R}	x4	64	78	90	100	mA	1, 2
	x8	68	82	94	104	mA	1, 2
	x16	88	108	128	148	mA	1, 2
I_{DD4W}	x4	69	81	93	105	mA	1, 2
	x8	73	85	97	108	mA	1, 2
	x16	99	119	138	156	mA	1, 2
I_{DD5B}	All	177	179	180	182	mA	1, 2
I_{DD6}	All	12	12	12	12	mA	1, 2, 3
I_{DD6ET}	All	15	15	15	15	mA	2, 4
I_{DD7}	x4, 8	121	150	156	164	mA	1, 2
	x16	152	172	195	219	mA	1, 2
I_{DD8}	All	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	$I_{DD2P0} + 2\text{mA}$	mA	1, 2

Notes:

1. $T_C = 85^\circ\text{C}$; SRT and ASR are disabled.
2. Enabling ASR could increase I_{DDx} by up to an additional 2mA.
3. Restricted to T_C (MAX) = 85°C .
4. $T_C = 85^\circ\text{C}$; ASR and ODT are disabled; SRT is enabled.
5. The I_{DD} values must be derated (increased) on IT-option devices when operated outside of the range $0^\circ\text{C} \leq T_C \leq +85^\circ\text{C}$:
 - 5a. When $T_C < 0^\circ\text{C}$: I_{DD2P0} , I_{DD2P1} and I_{DD3P} must be derated by 4%; I_{DD4R} and I_{DD4W} must be derated by 2%; and I_{DD6} , I_{DD6ET} and I_{DD7} must be derated by 7%.
 - 5b. When $T_C > 85^\circ\text{C}$: I_{DD0} , I_{DD1} , I_{DD2N} , I_{DD2NT} , I_{DD2Q} , I_{DD3N} , I_{DD3P} , I_{DD4R} , I_{DD4W} , and I_{DD5B} must be derated by 2%; I_{DD2Px} must be derated by 30%.

Electrical Specifications – DC and AC

DC Operating Conditions

Table 22: DDR3L 1.35V DC Electrical Characteristics and Operating Conditions

All voltages are referenced to V_{SS}

Parameter/Condition	Symbol	Min	Nom	Max	Unit	Notes
Supply voltage	V_{DD}	1.283	1.35	1.45	V	1-7
I/O supply voltage	V_{DDQ}	1.283	1.35	1.45	V	1-7
Input leakage current Any input $0V \leq V_{IN} \leq V_{DD}$, V_{REF} pin $0V \leq V_{IN} \leq 1.1V$ (All other pins not under test = 0V)	I_I	-2	-	2	μA	
V_{REF} supply leakage current $V_{REFDQ} = V_{DD}/2$ or $V_{REFCA} = V_{DD}/2$ (All other pins not under test = 0V)	I_{VREF}	-1	-	1	μA	8, 9

Notes:

1. V_{DD} and V_{DDQ} must track one another. V_{DDQ} must be $\leq V_{DD}$. $V_{SS} = V_{SSQ}$.
2. V_{DD} and V_{DDQ} may include AC noise of $\pm 50mV$ (250 kHz to 20 MHz) in addition to the DC (0 Hz to 250 kHz) specifications. V_{DD} and V_{DDQ} must be at same level for valid AC timing parameters.
3. Maximum DC value may not be greater than 1.425V. The DC value is the linear average of $V_{DD}/V_{DDQ}(t)$ over a very long period of time (for example, 1 second).
4. Under these supply voltages, the device operates to this DDR3L specification.
5. If the maximum limit is exceeded, input levels shall be governed by DDR3 specifications.
6. Under 1.5V operation, this DDR3L device operates in accordance with the DDR3 specifications under the same speed timings as defined for this device.
7. Once initialized for DDR3L operation, DDR3 operation may only be used if the device is in reset while V_{DD} and V_{DDQ} are changed for DDR3 operation (see V_{DD} Voltage Switching (page 133)).
8. The minimum limit requirement is for testing purposes. The leakage current on the V_{REF} pin should be minimal.
9. V_{REF} (see Table 23).

Input Operating Conditions

Table 23: DDR3L 1.35V DC Electrical Characteristics and Input Conditions

All voltages are referenced to V_{SS}

Parameter/Condition	Symbol	Min	Nom	Max	Unit	Notes
V_{IN} low; DC/commands/address busses	V_{IL}	V_{SS}	N/A	See Table 24	V	
V_{IN} high; DC/commands/address busses	V_{IH}	See Table 24	N/A	V_{DD}	V	
Input reference voltage command/address bus	$V_{REFCA(DC)}$	$0.49 \times V_{DD}$	$0.5 \times V_{DD}$	$0.51 \times V_{DD}$	V	1, 2
I/O reference voltage DQ bus	$V_{REFDQ(DC)}$	$0.49 \times V_{DD}$	$0.5 \times V_{DD}$	$0.51 \times V_{DD}$	V	2, 3
I/O reference voltage DQ bus in SELF REFRESH	$V_{REFDQ(SR)}$	V_{SS}	$0.5 \times V_{DD}$	V_{DD}	V	4
Command/address termination voltage (system level, not direct DRAM input)	V_{TT}	–	$0.5 \times V_{DDQ}$	–	V	5

Notes:

1. $V_{REFCA(DC)}$ is expected to be approximately $0.5 \times V_{DD}$ and to track variations in the DC level. Externally generated peak noise (non-common mode) on V_{REFCA} may not exceed $\pm 1\% \times V_{DD}$ around the $V_{REFCA(DC)}$ value. Peak-to-peak AC noise on V_{REFCA} should not exceed $\pm 2\%$ of $V_{REFCA(DC)}$.
2. DC values are determined to be less than 20 MHz in frequency. DRAM must meet specifications if the DRAM induces additional AC noise greater than 20 MHz in frequency.
3. $V_{REFDQ(DC)}$ is expected to be approximately $0.5 \times V_{DD}$ and to track variations in the DC level. Externally generated peak noise (non-common mode) on V_{REFDQ} may not exceed $\pm 1\% \times V_{DD}$ around the $V_{REFDQ(DC)}$ value. Peak-to-peak AC noise on V_{REFDQ} should not exceed $\pm 2\%$ of $V_{REFDQ(DC)}$.
4. $V_{REFDQ(DC)}$ may transition to $V_{REFDQ(SR)}$ and back to $V_{REFDQ(DC)}$ when in SELF REFRESH, within restrictions outlined in the SELF REFRESH section.
5. V_{TT} is not applied directly to the device. V_{TT} is a system supply for signal termination resistors. Minimum and maximum values are system-dependent.

Table 24: DDR3L 1.35V Input Switching Conditions - Command and Address

Parameter/Condition	Symbol	DDR3L-800/1066	DDR3L-1333/1600	DDR3L-1866	Units
Command and Address					
Input high AC voltage: Logic 1	$V_{IH(AC160),min}^5$	160	160	–	mV
Input high AC voltage: Logic 1	$V_{IH(AC135),min}^5$	135	135	135	mV
Input high AC voltage: Logic 1	$V_{IH(AC125),min}^5$	–	–	125	mV
Input high DC voltage: Logic 1	$V_{IH(DC90),min}$	90	90	90	mV
Input low DC voltage: Logic 0	$V_{IL(DC90),min}$	–90	–90	–90	mV
Input low AC voltage: Logic 0	$V_{IL(AC125),min}^5$	–	–	–125	mV
Input low AC voltage: Logic 0	$V_{IL(AC135),min}^5$	–135	–135	–135	mV
Input low AC voltage: Logic 0	$V_{IL(AC160),min}^5$	–160	–160	–	mV
DQ and DM					
Input high AC voltage: Logic 1	$V_{IH(AC160),min}^5$	160	160	–	mV
Input high AC voltage: Logic 1	$V_{IH(AC135),min}^5$	135	135	135	mV
Input high AC voltage: Logic 1	$V_{IH(AC125),min}^5$	–	–	130	mV
Input high DC voltage: Logic 1	$V_{IH(DC90),min}$	90	90	90	mV
Input low DC voltage: Logic 0	$V_{IL(DC90),min}$	–90	–90	–90	mV
Input low AC voltage: Logic 0	$V_{IL(AC125),min}^5$	–	–	–130	mV
Input low AC voltage: Logic 0	$V_{IL(AC135),min}^5$	–135	–135	–135	mV
Input low AC voltage: Logic 0	$V_{IL(AC160),min}^5$	–160	–160	–	mV

Notes:

1. All voltages are referenced to V_{REF} . V_{REF} is V_{REFCA} for control, command, and address. All slew rates and setup/hold times are specified at the DRAM ball. V_{REF} is V_{REFDQ} for DQ and DM inputs.
2. Input setup timing parameters (t_{IS} and t_{DS}) are referenced at $V_{IL(AC)}/V_{IH(AC)}$, not $V_{REF(DC)}$.
3. Input hold timing parameters (t_{IH} and t_{DH}) are referenced at $V_{IL(DC)}/V_{IH(DC)}$, not $V_{REF(DC)}$.
4. Single-ended input slew rate = 1 V/ns; maximum input voltage swing under test is 900mV (peak-to-peak).
5. When two $V_{IH(AC)}$ values (and two corresponding $V_{IL(AC)}$ values) are listed for a specific speed bin, the user may choose either value for the input AC level. Whichever value is used, the associated setup time for that AC level must also be used. Additionally, one $V_{IH(AC)}$ value may be used for address/command inputs and the other $V_{IH(AC)}$ value may be used for data inputs.

For example, for DDR3-800, two input AC levels are defined: $V_{IH(AC160),min}$ and $V_{IH(AC135),min}$ (corresponding $V_{IL(AC160),min}$ and $V_{IL(AC135),min}$). For DDR3-800, the address/command inputs must use either $V_{IH(AC160),min}$ with $t_{IS}(AC160)$ of 210ps or $V_{IH(AC150),min}$ with $t_{IS}(AC135)$ of 365ps; independently, the data inputs must use either $V_{IH(AC160),min}$ with $t_{DS}(AC160)$ of 75ps or $V_{IH(AC150),min}$ with $t_{DS}(AC150)$ of 125ps.

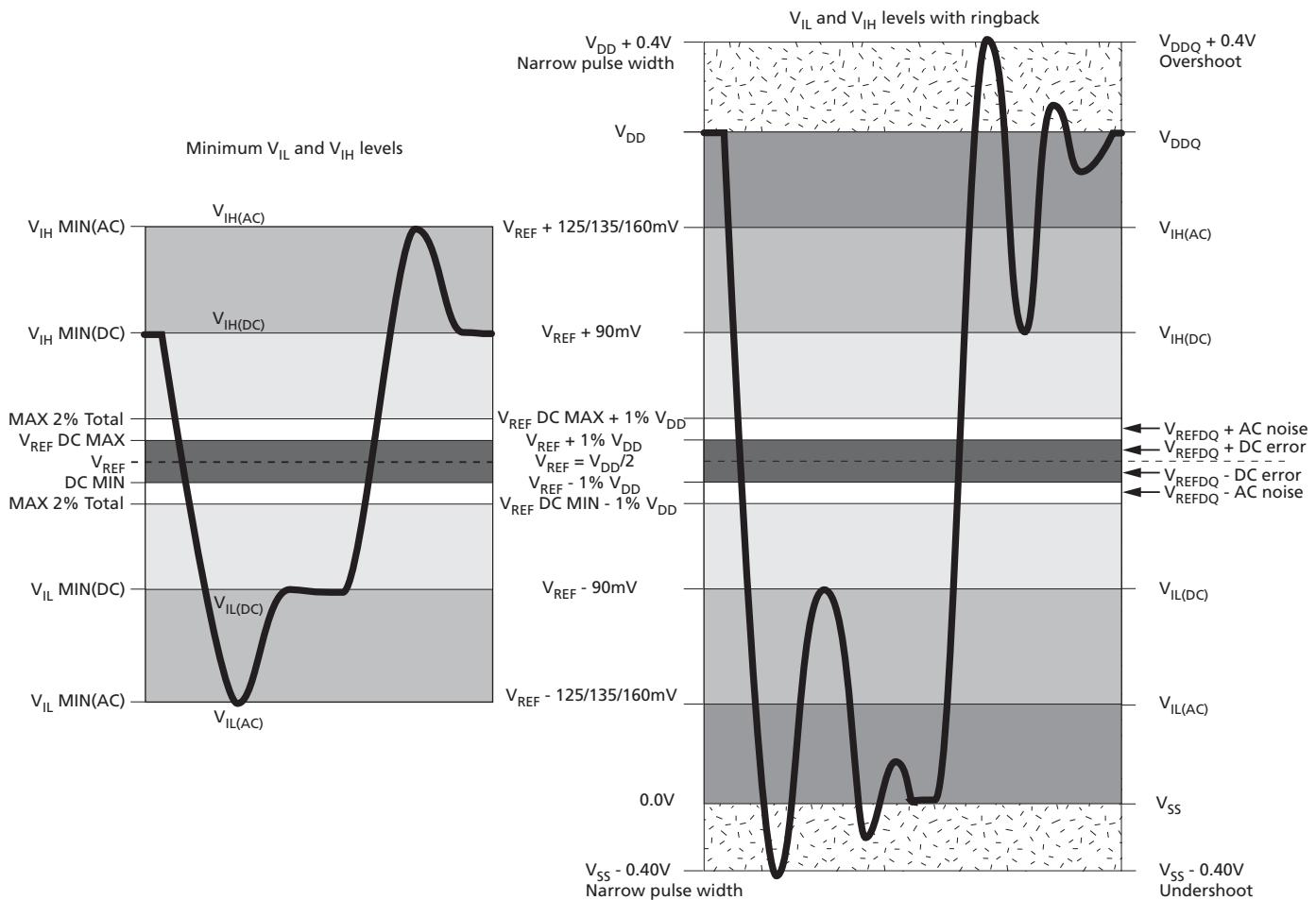
Table 25: DDR3L 1.35V Differential Input Operating Conditions (CK, CK# and DQS, DQS#)

Parameter/Condition	Symbol	Min	Max	Units	Notes
Differential input logic high – slew	$V_{IH,diff(AC)slew}$	180	N/A	mV	4
Differential input logic low – slew	$V_{IL,diff(AC)slew}$	N/A	-180	mV	4
Differential input logic high	$V_{IH,diff(AC)}$	$2 \times (V_{IH(AC)} - V_{REF})$	V_{DD}/V_{DDQ}	mV	5
Differential input logic low	$V_{IL,diff(AC)}$	V_{SS}/V_{SSQ}	$2 \times (V_{IL(AC)} - V_{REF})$	mV	6
Differential input crossing voltage relative to $V_{DD}/2$ for DQS, DQS#; CK, CK#	V_{IX}	$V_{REF(DC)} - 150$	$V_{REF(DC)} + 150$	mV	5, 7, 9
Differential input crossing voltage relative to $V_{DD}/2$ for CK, CK#	$V_{IX} (175)$	$V_{REF(DC)} - 175$	$V_{REF(DC)} + 175$	mV	5, 7–9
Single-ended high level for strobes	V_{SEH}	$V_{DDQ}/2 + 160$	V_{DDQ}	mV	5
Single-ended high level for CK, CK#		$V_{DD}/2 + 160$	V_{DD}	mV	5
Single-ended low level for strobes	V_{SEL}	V_{SSQ}	$V_{DDQ}/2 - 160$	mV	6
Single-ended low level for CK, CK#		V_{SS}	$V_{DD}/2 - 160$	mV	6

Notes:

1. Clock is referenced to V_{DD} and V_{SS} . Data strobe is referenced to V_{DDQ} and V_{SSQ} .
2. Reference is $V_{REFCA(DC)}$ for clock and $V_{REFDQ(DC)}$ for strobe.
3. Differential input slew rate = 2 V/ns.
4. Defines slew rate reference points, relative to input crossing voltages.
5. Minimum DC limit is relative to single-ended signals; overshoot specifications are applicable.
6. Maximum DC limit is relative to single-ended signals; undershoot specifications are applicable.
7. The typical value of $V_{IX(AC)}$ is expected to be about $0.5 \times V_{DD}$ of the transmitting device, and $V_{IX(AC)}$ is expected to track variations in V_{DD} . $V_{IX(AC)}$ indicates the voltage at which differential input signals must cross.
8. The V_{IX} extended range (± 175 mV) is allowed only for the clock; this V_{IX} extended range is only allowed when the following conditions are met: The single-ended input signals are monotonic, have the single-ended swing V_{SEL} , V_{SEH} of at least $V_{DD}/2 \pm 250$ mV, and the differential slew rate of CK, CK# is greater than 3 V/ns.
9. V_{IX} must provide 25mV (single-ended) of the voltages separation.

Figure 13: DDR3L 1.35V Input Signal



Note: 1. Numbers in diagrams reflect nominal values.

DDR3L 1.35V AC Overshoot/Undershoot Specification

Table 26: DDR3L Control and Address Pins

Parameter	DDR3L-800	DDR3L-1066	DDR3L-1333	DDR3L-1600	DDR3L-1866
Maximum peak amplitude allowed for overshoot area (see Figure 14)	0.4V	0.4V	0.4V	0.4V	0.4V
Maximum peak amplitude allowed for undershoot area (see Figure 15)	0.4V	0.4V	0.4V	0.4V	0.4V
Maximum overshoot area above V_{DD} (see Figure 14)	0.67 Vns	0.5 Vns	0.4 Vns	0.33 Vns	0.28 Vns
Maximum undershoot area below V_{SS} (see Figure 15)	0.67 Vns	0.5 Vns	0.4 Vns	0.33 Vns	0.28 Vns

Table 27: DDR3L 1.35V Clock, Data, Strobe, and Mask Pins

Parameter	DDR3L-800	DDR3L-1066	DDR3L-1333	DDR3L-1600	DDR3L-1866
Maximum peak amplitude allowed for overshoot area (see Figure 14)	0.4V	0.4V	0.4V	0.4V	0.4V
Maximum peak amplitude allowed for undershoot area (see Figure 15)	0.4V	0.4V	0.4V	0.4V	0.4V
Maximum overshoot area above V_{DD}/V_{DDQ} (see Figure 14)	0.25 Vns	0.19 Vns	0.15 Vns	0.13 Vns	0.11 Vns
Maximum undershoot area below V_{SS}/V_{SSQ} (see Figure 15)	0.25 Vns	0.19 Vns	0.15 Vns	0.13 Vns	0.11 Vns

Figure 14: Overshoot

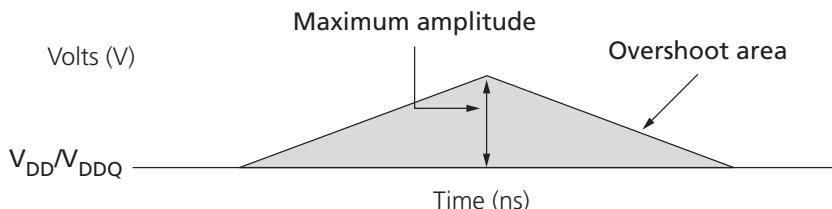


Figure 15: Undershoot

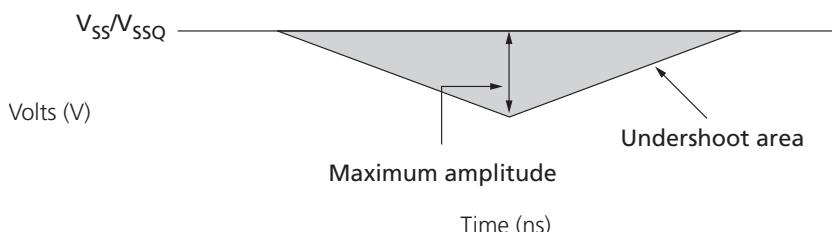


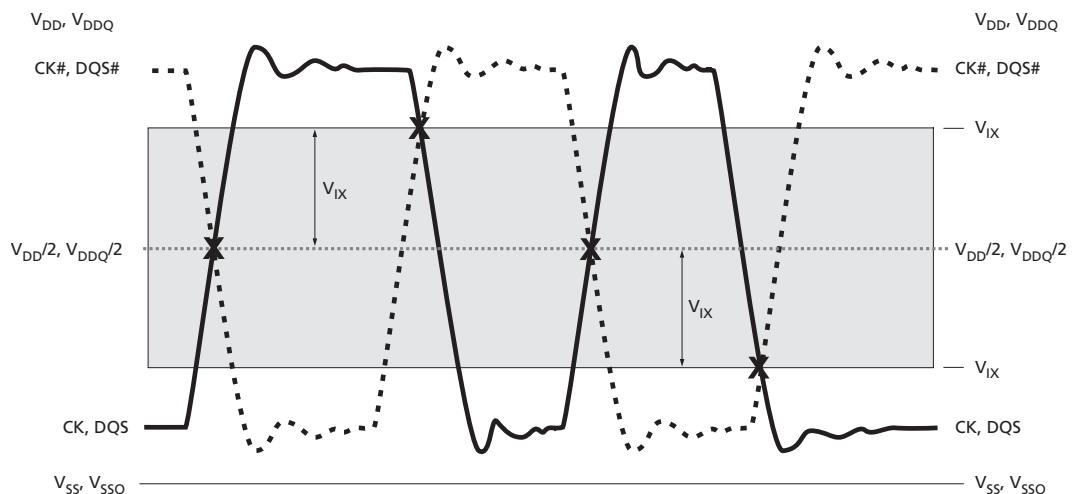
Figure 16: V_{IX} for Differential Signals


Figure 17: Single-Ended Requirements for Differential Signals

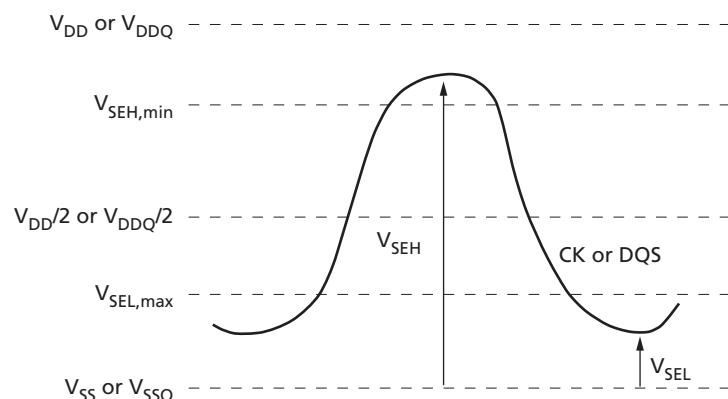
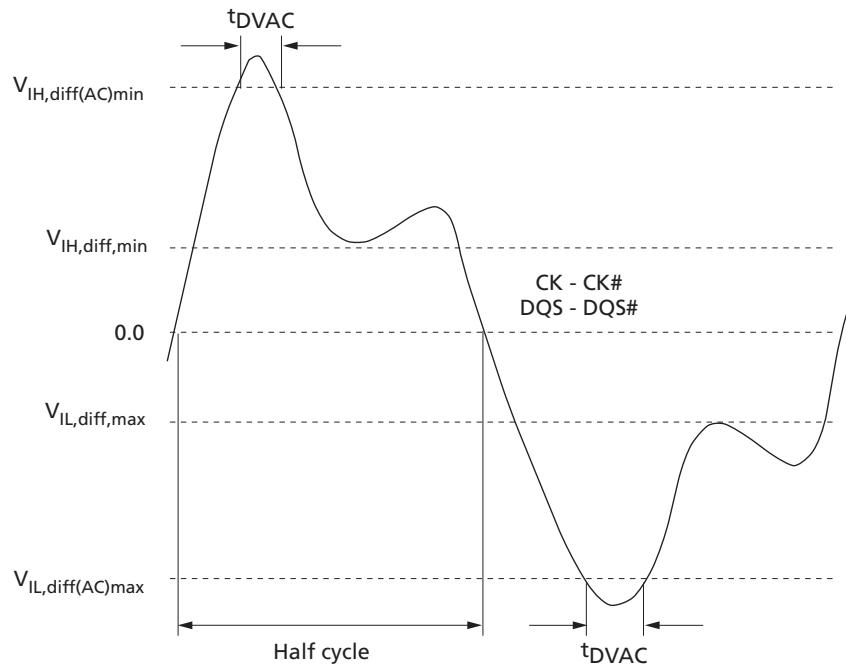


Figure 18: Definition of Differential AC-Swing and t_{DVAC}

Table 28: DDR3L 1.35V - Minimum Required Time t_{DVAC} for CK/CK#, DQS/DQS# Differential for AC Ringback

Slew Rate (V/ns)	DDR3L-800/1066/1333/1600		DDR3L-1866		
	t_{DVAC} at 320mV (ps)	t_{DVAC} at 270mV (ps)	t_{DVAC} at 270mV (ps)	t_{DVAC} at 250mV (ps)	t_{DVAC} at 260mV (ps)
>4.0	189	201	163	168	176
4.0	189	201	163	168	176
3.0	162	179	140	147	154
2.0	109	134	95	105	111
1.8	91	119	80	91	97
1.6	69	100	62	74	78
1.4	40	76	37	52	55
1.2	Note1	44	5	22	24
1.0	Note1	Note1	Note1	Note1	Note1
<1.0	Note1	Note1	Note1	Note1	Note1

Note: 1. Rising input signal shall become equal to or greater than $V_{IH(AC)}$ level and Falling input signal shall become equal to or less than $V_{IL(AC)}$ level.

DDR3L 1.35V Slew Rate Definitions for Single-Ended Input Signals

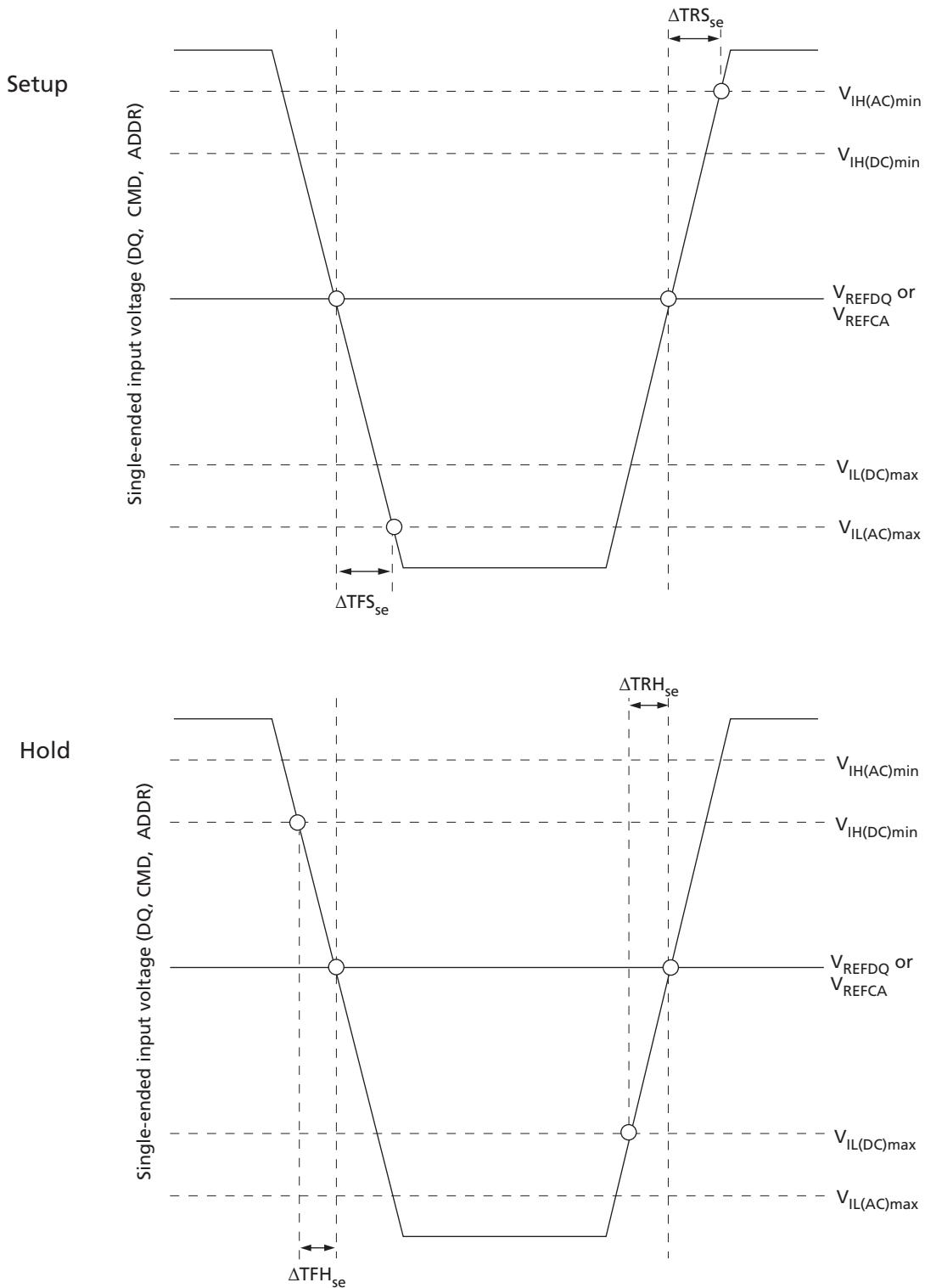
Setup (t_{IS} and t_{DS}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of V_{REF} and the first crossing of $V_{IH(AC)min}$. Setup (t_{IS} and t_{DS}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of V_{REF} and the first crossing of $V_{IL(AC)max}$.

Hold (t_{IH} and t_{DH}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{IL(DC)max}$ and the first crossing of V_{REF} . Hold (t_{IH} and t_{DH}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{IH(DC)min}$ and the first crossing of V_{REF} (see Figure 19 (page 53)).

Table 29: Single-Ended Input Slew Rate Definition

Input Slew Rates (Linear Signals)		Measured		Calculation
Input	Edge	From	To	
Setup	Rising	V_{REF}	$V_{IH(AC),min}$	$\frac{V_{IH(AC),min} - V_{REF}}{\Delta TRS_{se}}$
	Falling	V_{REF}	$V_{IL(AC),max}$	$\frac{V_{REF} - V_{IL(AC),max}}{\Delta TFS_{se}}$
Hold	Rising	$V_{IL(DC),max}$	V_{REF}	$\frac{V_{REF} - V_{IL(DC),max}}{\Delta TFH_{se}}$
	Falling	$V_{IH(DC),min}$	V_{REF}	$\frac{V_{IH(DC),min} - V_{REF}}{\Delta TRSH_{se}}$

Figure 19: Nominal Slew Rate Definition for Single-Ended Input Signals



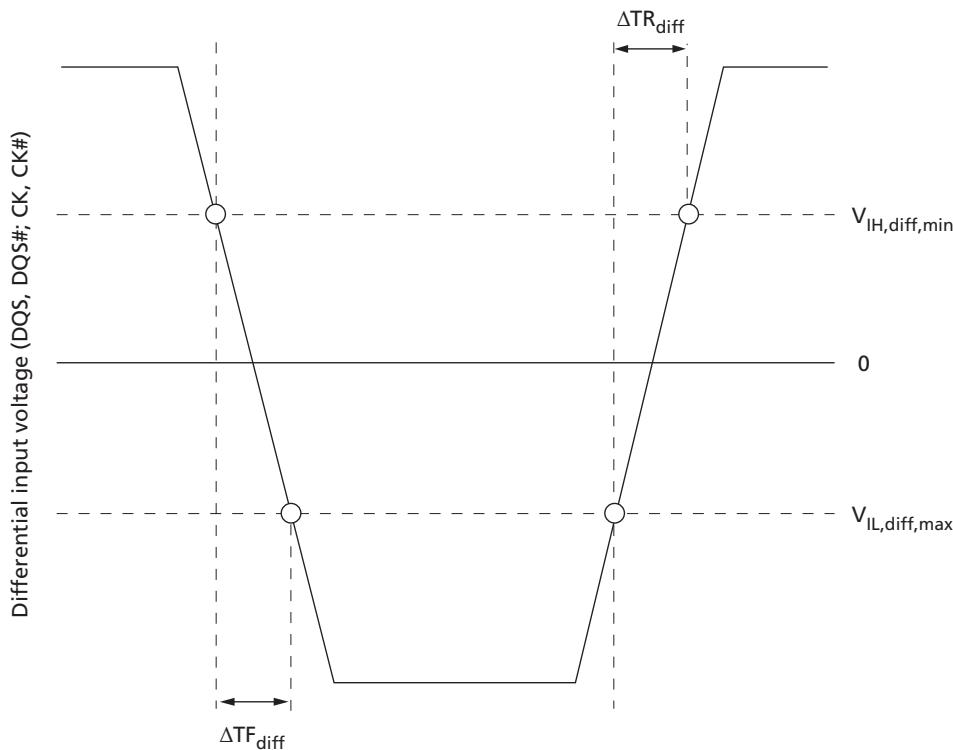
DDR3L 1.35V Slew Rate Definitions for Differential Input Signals

Input slew rate for differential signals (CK, CK# and DQS, DQS#) are defined and measured, as shown in Table 30 and Figure 20. The nominal slew rate for a rising signal is defined as the slew rate between $V_{IL,diff,max}$ and $V_{IH,diff,min}$. The nominal slew rate for a falling signal is defined as the slew rate between $V_{IH,diff,min}$ and $V_{IL,diff,max}$.

Table 30: DDR3L 1.35V Differential Input Slew Rate Definition

Differential Input Slew Rates (Linear Signals)		Measured		Calculation
Input	Edge	From	To	
CK and DQS reference	Rising	$V_{IL,diff,max}$	$V_{IH,diff,min}$	$\frac{V_{IH,diff,min} - V_{IL,diff,max}}{\Delta T_{R,diff}}$
	Falling	$V_{IH,diff,min}$	$V_{IL,diff,max}$	$\frac{V_{IH,diff,min} - V_{IL,diff,max}}{\Delta T_{F,diff}}$

Figure 20: DDR3L 1.35V Nominal Differential Input Slew Rate Definition for DQS, DQS# and CK, CK#



ODT Characteristics

The ODT effective resistance R_{TT} is defined by MR1[9, 6, and 2]. ODT is applied to the DQ, DM, DQS, DQS#, and TDQS, TDQS# balls (x8 devices only). The ODT target values and a functional representation are listed in Table 31 and Table 32 (page 56). The individual pull-up and pull-down resistors ($R_{TT(PU)}$ and $R_{TT(PD)}$) are defined as follows:

- $R_{TT(PU)} = (V_{DDQ} - V_{OUT})/|I_{OUT}|$, under the condition that $R_{TT(PD)}$ is turned off
- $R_{TT(PD)} = (V_{OUT})/|I_{OUT}|$, under the condition that $R_{TT(PU)}$ is turned off

Figure 21: ODT Levels and I-V Characteristics

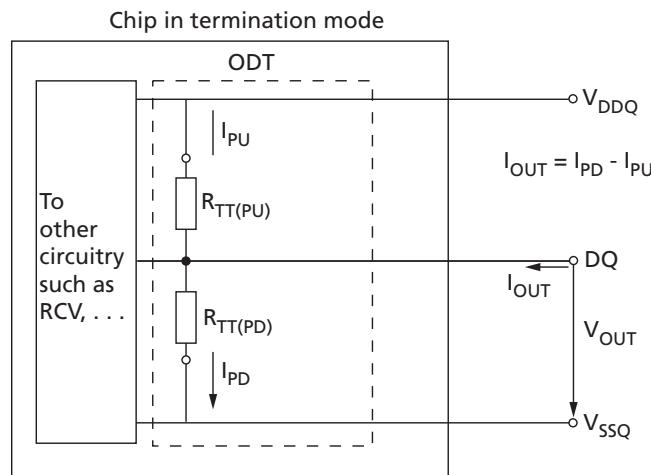


Table 31: On-Die Termination DC Electrical Characteristics

Parameter/Condition	Symbol	Min	Nom	Max	Unit	Notes
R_{TT} effective impedance	$R_{TT(EFF)}$		See Table 32 (page 56)			1, 2
Deviation of VM with respect to $V_{DDQ}/2$	ΔVM	-5		5	%	1, 2, 3

Notes:

1. Tolerance limits are applicable after proper ZQ calibration has been performed at a stable temperature and voltage ($V_{DDQ} = V_{DD}$, $V_{SSQ} = V_{SS}$). Refer to ODT Sensitivity (page 57) if either the temperature or voltage changes after calibration.
2. Measurement definition for R_{TT} : Apply $V_{IH(AC)}$ to pin under test and measure current $I[V_{IH(AC)}]$, then apply $V_{IL(AC)}$ to pin under test and measure current $I[V_{IL(AC)}]$:
$$R_{TT} = \frac{V_{IH(AC)} - V_{IL(AC)}}{I(V_{IH(AC)}) - I(V_{IL(AC)})}$$
3. Measure voltage (VM) at the tested pin with no load:
$$\Delta VM = \left(\frac{2 \times VM}{V_{DDQ}} - 1 \right) \times 100$$
4. For IT and AT devices, the minimum values are derated by 6% when the device operates between -40°C and 0°C (T_C).

1.35V ODT Resistors

Table 32 provides an overview of the ODT DC electrical characteristics. The values provided are not specification requirements; however, they can be used as design guidelines to indicate what R_{TT} is targeted to provide:

- R_{TT} 120Ω is made up of $R_{TT120(PD240)}$ and $R_{TT120(PU240)}$
- R_{TT} 60Ω is made up of $R_{TT60(PD120)}$ and $R_{TT60(PU120)}$
- R_{TT} 40Ω is made up of $R_{TT40(PD80)}$ and $R_{TT40(PU80)}$
- R_{TT} 30Ω is made up of $R_{TT30(PD60)}$ and $R_{TT30(PU60)}$
- R_{TT} 20Ω is made up of $R_{TT20(PD40)}$ and $R_{TT20(PU40)}$

Table 32: 1.35V R_{TT} Effective Impedance

Gray-shaded cells are DDR3L unique values; All other values are the same for both DDR3L and DDR3

MR1 [9, 6, 2]	R_{TT}	Resistor	V_{OUT}	Min	Nom	Max	Units
0, 1, 0	120Ω	$R_{TT,120PD240}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/1
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/1
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/1
	120Ω	$R_{TT,120PU240}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/1
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/1
			0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/1
		$V_{IL(AC)}$ to $V_{IH(AC)}$		0.9	1.0	1.65	RZQ/2
	60Ω	$R_{TT,60PD120}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/2
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/2
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/2
		$R_{TT,60PU120}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/2
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/2
			0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/2
0, 0, 1	60Ω	$V_{IL(AC)}$ to $V_{IH(AC)}$		0.9	1.0	1.65	RZQ/4
		$R_{TT,60PD120}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/2
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/2
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/2
	$R_{TT,60PU120}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/2	
		0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/2	
		0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/2	
	60Ω		$V_{IL(AC)}$ to $V_{IH(AC)}$	0.9	1.0	1.65	RZQ/4
0, 1, 1	40Ω	$R_{TT,40PD80}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/3
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/3
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/3
	$R_{TT,40PU80}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/3	
		0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/3	
		0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/3	
	40Ω		$V_{IL(AC)}$ to $V_{IH(AC)}$	0.9	1.0	1.65	RZQ/6
	30Ω	$R_{TT,30PD60}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/4
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/4
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/4
		$R_{TT,30PU60}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/4
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/4
			0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/4
	30Ω		$V_{IL(AC)}$ to $V_{IH(AC)}$	0.9	1.0	1.65	RZQ/8

Table 32: 1.35V R_{TT} Effective Impedance (Continued)

Gray-shaded cells are DDR3L unique values; All other values are the same for both DDR3L and DDR3

MR1 [9, 6, 2]	R_{TT}	Resistor	V_{OUT}	Min	Nom	Max	Units
1, 0, 0	20Ω	$R_{TT,20PD40}$	0.2 × V_{DDQ}	0.6	1.0	1.15	RZQ/6
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/6
			0.8 × V_{DDQ}	0.9	1.0	1.45	RZQ/6
	20Ω	$R_{TT,20PU40}$	0.2 × V_{DDQ}	0.9	1.0	1.45	RZQ/6
			0.5 × V_{DDQ}	0.9	1.0	1.15	RZQ/6
			0.8 × V_{DDQ}	0.6	1.0	1.15	RZQ/6
	20Ω		$V_{IL(AC)}$ to $V_{IH(AC)}$	0.9	1.0	1.65	RZQ/12

ODT Sensitivity

If either the temperature or voltage changes after I/O calibration, then the tolerance limits listed in Table 31 and Table 32 can be expected to widen according to Table 33 and Table 34.

Table 33: ODT Sensitivity Definition

Symbol	Min	Max	Unit
R_{TT}	0.9 - $dR_{TT}dT \times DT - dR_{TT}dV \times DV $	1.6 + $dR_{TT}dT \times DT + dR_{TT}dV \times DV $	RZQ/(2, 4, 6, 8, 12)

Note: 1. $\Delta T = T - T(@ \text{ calibration})$, $\Delta V = V_{DDQ} - V_{DDQ}(@ \text{ calibration})$ and $V_{DD} = V_{DDQ}$.

Table 34: ODT Temperature and Voltage Sensitivity

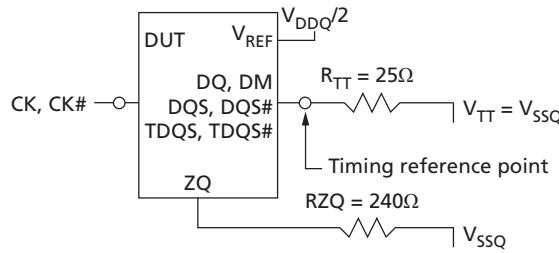
Change	Min	Max	Unit
$dR_{TT}dT$	0	1.5	%/°C
$dR_{TT}dV$	0	0.15	%/mV

Note: 1. $\Delta T = T - T(@ \text{ calibration})$, $\Delta V = V_{DDQ} - V_{DDQ}(@ \text{ calibration})$ and $V_{DD} = V_{DDQ}$.

ODT Timing Definitions

ODT loading differs from that used in AC timing measurements. The reference load for ODT timings is shown in Figure 22. Two parameters define when ODT turns on or off synchronously, two define when ODT turns on or off asynchronously, and another defines when ODT turns on or off dynamically. Table 35 and Table 36 (page 58) outline and provide definition and measurement references settings for each parameter.

ODT turn-on time begins when the output leaves High-Z and ODT resistance begins to turn on. ODT turn-off time begins when the output leaves Low-Z and ODT resistance begins to turn off.

Figure 22: ODT Timing Reference Load

Table 35: ODT Timing Definitions

Symbol	Begin Point Definition	End Point Definition	Figure
^t AON	Rising edge of CK – CK# defined by the end point of ODTLon	Extrapolated point at V _{SSQ}	Figure 23 (page 59)
^t AOF	Rising edge of CK – CK# defined by the end point of ODTLooff	Extrapolated point at V _{RTT,nom}	Figure 23 (page 59)
^t AONPD	Rising edge of CK – CK# with ODT first being registered HIGH	Extrapolated point at V _{SSQ}	Figure 24 (page 59)
^t AOFPD	Rising edge of CK – CK# with ODT first being registered LOW	Extrapolated point at V _{RTT,nom}	Figure 24 (page 59)
^t ADC	Rising edge of CK – CK# defined by the end point of ODTLcnw, ODTLcwn4, or ODTLcwn8	Extrapolated points at V _{RTT(WR)} and V _{RTT,nom}	Figure 25 (page 60)

Table 36: DDR3L(1.35V) Reference Settings for ODT Timing Measurements

Gray-shaded cells are DDR3L unique values; All other values are the same for both DDR3L and DDR3

Measured Parameter	R _{TT,nom} Setting	R _{TT(WR)} Setting	V _{SW1}	V _{SW2}
^t AON	RZQ/4 (60Ω)	N/A	50mV	100mV
	RZQ/12 (20Ω)	N/A	100mV	200mV
^t AOF	RZQ/4 (60Ω)	N/A	50mV	100mV
	RZQ/12 (20Ω)	N/A	100mV	200mV
^t AONPD	RZQ/4 (60Ω)	N/A	50mV	100mV
	RZQ/12 (20Ω)	N/A	100mV	200mV
^t AOFPD	RZQ/4 (60Ω)	N/A	50mV	100mV
	RZQ/12 (20Ω)	N/A	100mV	200mV
^t ADC	RZQ/12 (20Ω)	RZQ/2 (20Ω)	200mV	250mV

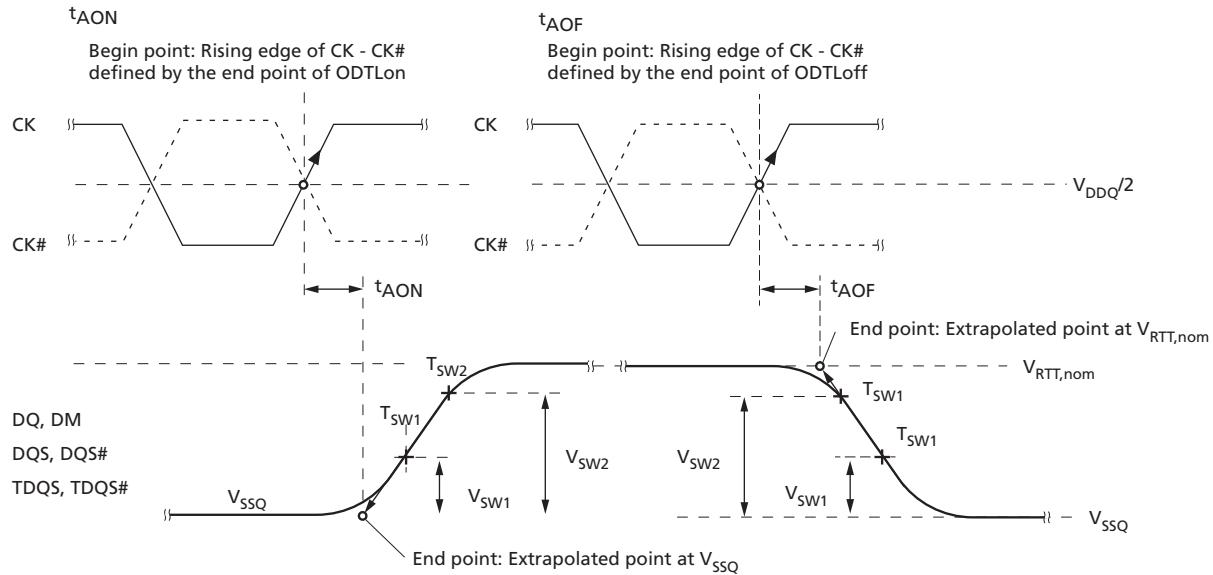
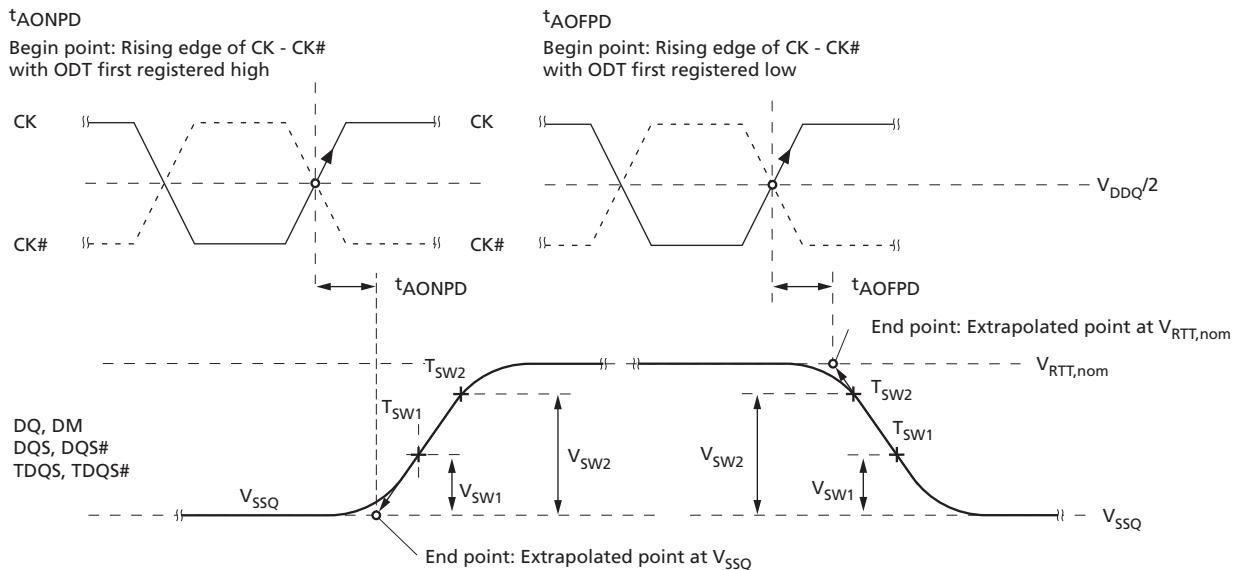
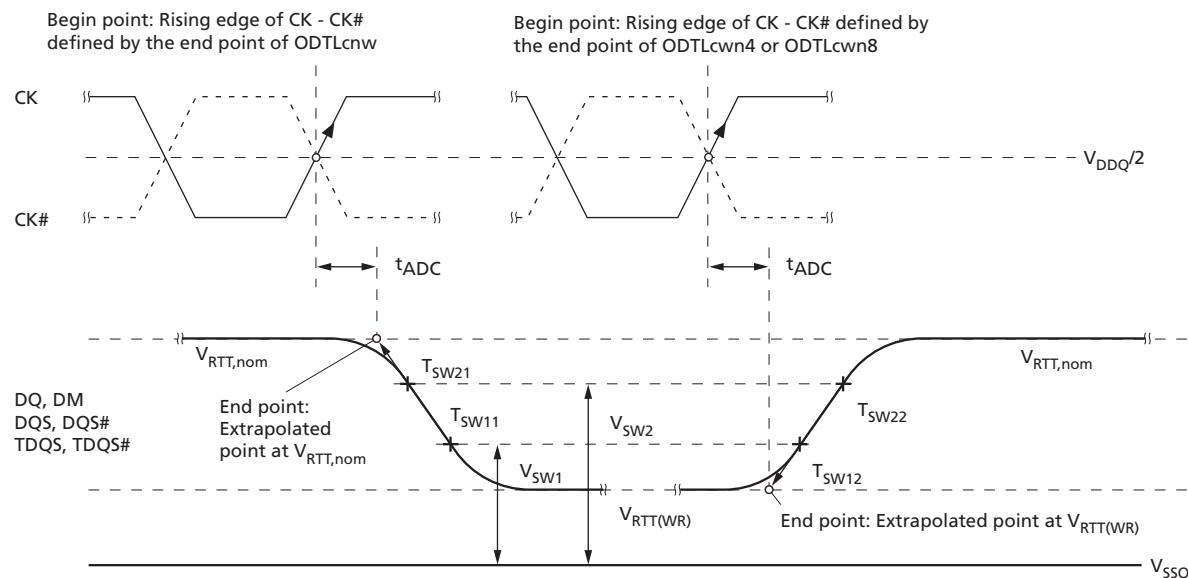
Figure 23: tAON and tAOF Definitions

Figure 24: tAONPD and tAOFPD Definitions


Figure 25: t_{ADC} Definition


Output Driver Impedance

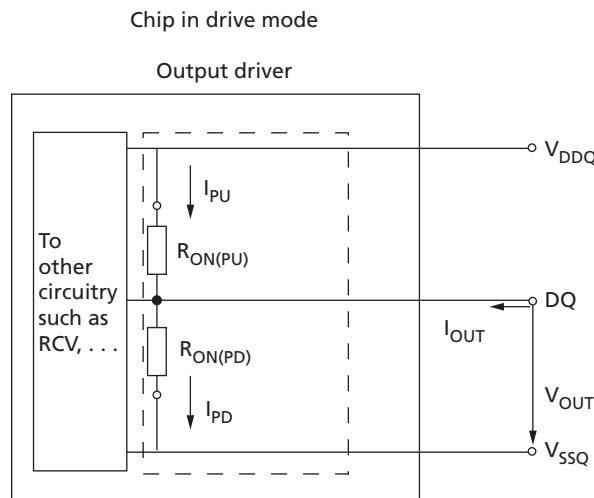
The output driver impedance is selected by MR1[5,1] during initialization. The selected value is able to maintain the tight tolerances specified if proper ZQ calibration is performed. Output specifications refer to the default output driver unless specifically stated otherwise. A functional representation of the output buffer is shown below. The output driver impedance R_{ON} is defined by the value of the external reference resistor R_{ZQ} as follows:

- $R_{ON,x} = R_{ZQ}/y$ (with $R_{ZQ} = 240\Omega \pm 1\%$; $x = 34\Omega$ or 40Ω with $y = 7$ or 6 , respectively)

The individual pull-up and pull-down resistors $R_{ON(PU)}$ and $R_{ON(PD)}$ are defined as follows:

- $R_{ON(PU)} = (V_{DDQ} - V_{OUT})/|I_{OUT}|$, when $R_{ON(PD)}$ is turned off
- $R_{ON(PD)} = (V_{OUT})/|I_{OUT}|$, when $R_{ON(PU)}$ is turned off

Figure 26: Output Driver



34 Ohm Output Driver Impedance

The 34Ω driver ($MR1[5, 1] = 01$) is the default driver. Unless otherwise stated, all timings and specifications listed herein apply to the 34Ω driver only. Its impedance R_{ON} is defined by the value of the external reference resistor RZQ as follows: $R_{ON34} = RZQ/7$ (with nominal $RZQ = 240\Omega \pm 1\%$) and is actually $34.3\Omega \pm 1\%$.

Table 37: DDR3L 34 Ohm Driver Impedance Characteristics

Gray-shaded cells are DDR3L unique values; All other values are the same for both DDR3L and DDR3

MR1 [5, 1]	R _{ON}	Resistor	V _{OUT}	Min	Nom	Max	Units
0, 1	34.3Ω	R _{ON,34PD}	0.2 × V _{DDQ}	0.6	1.0	1.15	RZQ/7
			0.5 × V _{DDQ}	0.9	1.0	1.15	RZQ/7
			0.8 × V _{DDQ}	0.9	1.0	1.45	RZQ/7
		R _{ON,34PU}	0.2 × V _{DDQ}	0.9	1.0	1.45	RZQ/7
			0.5 × V _{DDQ}	0.9	1.0	1.15	RZQ/7
			0.8 × V _{DDQ}	0.6	1.0	1.15	RZQ/7
Pull-up/pull-down mismatch (MM _{PUPD})		V _{IL(AC)} to V _{IH(AC)}		-10	N/A	10	%

Notes:

1. Tolerance limits assume RZQ of $240\Omega \pm 1\%$ and are applicable after proper ZQ calibration has been performed at a stable temperature and voltage: $V_{DDQ} = V_{DD}$; $V_{SSQ} = V_{SS}$. Refer to DDR3L 34 Ohm Output Driver Sensitivity (page 64) if either the temperature or the voltage changes after calibration.
2. Measurement definition for mismatch between pull-up and pull-down (MM_{PUPD}). Measure both $R_{ON(PU)}$ and $R_{ON(PD)}$ at $0.5 \times V_{DDQ}$:

$$MM_{PUPD} = \frac{R_{ON(PU)} - R_{ON(PD)}}{R_{ON,nom}} \times 100$$

3. For IT and AT (1Gb only) devices, the minimum values are derated by 6% when the device operates between -40°C and 0°C (T_C). A larger maximum limit will result in slightly lower minimum currents.

DDR3L 34 Ohm Driver

Using Table 38, the 34Ω driver's current range has been calculated and summarized in Table 39 (page 63) $V_{DD} = 1.35V$, Table 40 for $V_{DD} = 1.45V$, and Table 41 (page 64) for $V_{DD} = 1.283V$. The individual pull-up and pull-down resistors $R_{ON34(PD)}$ and $R_{ON34(PU)}$ are defined as follows:

- $R_{ON34(PD)} = (V_{OUT})/|I_{OUT}|$; $R_{ON34(PU)}$ is turned off
- $R_{ON34(PU)} = (V_{DDQ} - V_{OUT})/|I_{OUT}|$; $R_{ON34(PD)}$ is turned off

Table 38: DDR3L 34 Ohm Driver Pull-Up and Pull-Down Impedance Calculations

R _{ON}				Min	Nom	Max	Unit
$R_{ZQ} = 240\Omega \pm 1\%$				237.6	240	242.4	Ω
$R_{ZQ}/7 = (240\Omega \pm 1\%)/7$				33.9	34.3	34.6	Ω
MR1[5,1]	R _{ON}	Resistor	V _{OUT}	Min	Nom	Max	Unit
0, 1	34.3 Ω	R _{ON34(PD)}	0.2 \times V _{DDQ}	20.4	34.3	38.1	Ω
			0.5 \times V _{DDQ}	30.5	34.3	38.1	Ω
			0.8 \times V _{DDQ}	30.5	34.3	48.5	Ω
		R _{ON34(PU)}	0.2 \times V _{DDQ}	30.5	34.3	48.5	Ω
			0.5 \times V _{DDQ}	30.5	34.3	38.1	Ω
			0.8 \times V _{DDQ}	20.4	34.3	38.1	Ω

Table 39: DDR3L 34 Ohm Driver I_{OL}/I_{OL} Characteristics: V_{DD} = V_{DDQ} = DDR3L@1.35V

MR1[5,1]	R _{ON}	Resistor	V _{OUT}	Max	Nom	Min	Unit
0, 1	34.3 Ω	R _{ON34(PD)}	I _{OL} @ 0.2 \times V _{DDQ}	13.3	7.9	7.1	mA
			I _{OL} @ 0.5 \times V _{DDQ}	22.1	19.7	17.7	mA
			I _{OL} @ 0.8 \times V _{DDQ}	35.4	31.5	22.3	mA
		R _{ON34(PU)}	I _{OH} @ 0.2 \times V _{DDQ}	35.4	31.5	22.3	mA
			I _{OH} @ 0.5 \times V _{DDQ}	22.1	19.7	17.7	mA
			I _{OH} @ 0.8 \times V _{DDQ}	13.3	7.9	7.1	mA

Table 40: DDR3L 34 Ohm Driver I_{OL}/I_{OL} Characteristics: V_{DD} = V_{DDQ} = DDR3L@1.45V

MR1[5,1]	R _{ON}	Resistor	V _{OUT}	Max	Nom	Min	Unit
0, 1	34.3 Ω	R _{ON34(PD)}	I _{OL} @ 0.2 \times V _{DDQ}	14.2	8.5	7.6	mA
			I _{OL} @ 0.5 \times V _{DDQ}	23.7	21.1	19.0	mA
			I _{OL} @ 0.8 \times V _{DDQ}	38.0	33.8	23.9	mA
		R _{ON34(PU)}	I _{OH} @ 0.2 \times V _{DDQ}	38.0	33.8	23.9	mA
			I _{OH} @ 0.5 \times V _{DDQ}	23.7	21.1	19.0	mA
			I _{OH} @ 0.8 \times V _{DDQ}	14.2	8.5	7.6	mA

Table 41: DDR3L 34 Ohm Driver I_{OH}/I_{OL} Characteristics: $V_{DD} = V_{DDQ} = DDR3L@1.283$

MR1[5,1]	R _{ON}	Resistor	V _{OUT}	Max	Nom	Min	Unit
0, 1	34.3Ω	R _{ON34(PD)}	I _{OL} @ 0.2 × V _{DDQ}	12.6	7.5	6.7	mA
			I _{OL} @ 0.5 × V _{DDQ}	21.0	18.7	16.8	mA
			I _{OL} @ 0.8 × V _{DDQ}	33.6	29.9	21.2	mA
	34.3Ω	R _{ON34(PU)}	I _{OH} @ 0.2 × V _{DDQ}	33.6	29.9	21.2	mA
			I _{OH} @ 0.5 × V _{DDQ}	21.0	18.7	16.8	mA
			I _{OH} @ 0.8 × V _{DDQ}	12.6	7.5	6.7	mA

DDR3L 34 Ohm Output Driver Sensitivity

If either the temperature or the voltage changes after ZQ calibration, then the tolerance limits listed in Table 37 (page 62) can be expected to widen according to Table 42 and Table 43.

Table 42: DDR3L 34 Ohm Output Driver Sensitivity Definition

Symbol	Min	Max	Unit
R _{ON(PD)} @ 0.2 × V _{DDQ}	0.6 - dR _{OND} TL × ΔT - dR _{OND} VL × ΔV	1.1 + dR _{OND} TL × ΔT + dR _{OND} VL × ΔV	RZQ/7
R _{ON(PD)} @ 0.5 × V _{DDQ}	0.9 - dR _{OND} TM × ΔT - dR _{OND} VM × ΔV	1.1 + dR _{OND} TM × ΔT + dR _{OND} VM × ΔV	RZQ/7
R _{ON(PD)} @ 0.8 × V _{DDQ}	0.9 - dR _{OND} TH × ΔT - dR _{OND} VH × ΔV	1.4 + dR _{OND} TH × ΔT + dR _{OND} VH × ΔV	RZQ/7
R _{ON(PU)} @ 0.2 × V _{DDQ}	0.9 - dR _{OND} TL × ΔT - dR _{OND} VL × ΔV	1.4 + dR _{OND} TL × ΔT + dR _{OND} VL × ΔV	RZQ/7
R _{ON(PU)} @ 0.5 × V _{DDQ}	0.9 - dR _{OND} TM × ΔT - dR _{OND} VM × ΔV	1.1 + dR _{OND} TM × ΔT + dR _{OND} VM × ΔV	RZQ/7
R _{ON(PU)} @ 0.8 × V _{DDQ}	0.6 - dR _{OND} TH × ΔT - dR _{OND} VH × ΔV	1.1 + dR _{OND} TH × ΔT + dR _{OND} VH × ΔV	RZQ/7

Note: 1. $\Delta T = T - T_{(@CALIBRATION)}$; $\Delta V = V_{DDQ} - V_{DDQ(@CALIBRATION)}$; and $V_{DD} = V_{DDQ}$.

Table 43: DDR3L 34 Ohm Output Driver Voltage and Temperature Sensitivity

Change	Min	Max	Unit
dR _{OND} TM	0	1.5	%/°C
dR _{OND} VM	0	0.13	%/mV
dR _{OND} TL	0	1.5	%/°C
dR _{OND} VL	0	0.13	%/mV
dR _{OND} TH	0	1.5	%/°C
dR _{OND} VH	0	0.13	%/mV

DDR3L Alternative 40 Ohm Driver

Table 44: DDR3L 40 Ohm Driver Impedance Characteristics

Gray-shaded cells are DDR3L unique values; All other values are the same for both DDR3L and DDR3

MR1 [5, 1]	R _{ON}	Resistor	V _{OUT}	Min	Nom	Max	Units	
0, 0	40Ω	R _{ON,40PD}	0.2 × V _{DDQ}	0.6	1.0	1.15	RZQ/6	
			0.5 × V _{DDQ}	0.9	1.0	1.15	RZQ/6	
			0.8 × V _{DDQ}	0.9	1.0	1.45	RZQ/6	
	R _{ON,40PU}		0.2 × V _{DDQ}	0.9	1.0	1.45	RZQ/6	
			0.5 × V _{DDQ}	0.9	1.0	1.15	RZQ/6	
			0.8 × V _{DDQ}	0.6	1.0	1.15	RZQ/6	
Pull-up/pull-down mismatch (MM _{PUPD})		V _{IL(AC)} to V _{IH(AC)}	-10	N/A	10		%	

Notes: 1. Tolerance limits assume RZQ of 240Ω ±1% and are applicable after proper ZQ calibration has been performed at a stable temperature and voltage (V_{DDQ} = V_{DD}; V_{SSQ} = V_{SS}). Refer to DDR3L 40 Ohm Output Driver Sensitivity (page 65) if either the temperature or the voltage changes after calibration.

2. Measurement definition for mismatch between pull-up and pull-down (MM_{PUPD}). Measure both R_{ON(PU)} and R_{ON(PD)} at 0.5 × V_{DDQ}:

$$MM_{PUPD} = \frac{R_{ON(PU)} - R_{ON(PD)}}{R_{ON,nom}} \times 100$$

3. For IT and AT (1Gb only) devices, the minimum values are derated by 6% when the device operates between -40°C and 0°C (T_C). A larger maximum limit will result in slightly lower minimum currents.

DDR3L 40 Ohm Output Driver Sensitivity

If either the temperature or the voltage changes after I/O calibration, then the tolerance limits listed in Table 44 can be expected to widen according to Table 45 and Table 46 (page 66).

Table 45: DDR3L 40 Ohm Output Driver Sensitivity Definition

Symbol	Min	Max	Unit
R _{ON(PD)} @ 0.2 × V _{DDQ}	0.6 - dR _{OND} TL × ΔT - dR _{OND} VL × ΔV	1.1 + dR _{OND} TL × ΔT + dR _{OND} VL × ΔV	RZQ/6
R _{ON(PD)} @ 0.5 × V _{DDQ}	0.9 - dR _{OND} TM × ΔT - dR _{OND} VM × ΔV	1.1 + dR _{OND} TM × ΔT + dR _{OND} VM × ΔV	RZQ/6
R _{ON(PD)} @ 0.8 × V _{DDQ}	0.9 - dR _{OND} TH × ΔT - dR _{OND} VH × ΔV	1.4 + dR _{OND} TH × ΔT + dR _{OND} VH × ΔV	RZQ/6
R _{ON(PU)} @ 0.2 × V _{DDQ}	0.9 - dR _{OND} TL × ΔT - dR _{OND} VL × ΔV	1.4 + dR _{OND} TL × ΔT + dR _{OND} VL × ΔV	RZQ/6
R _{ON(PU)} @ 0.5 × V _{DDQ}	0.9 - dR _{OND} TM × ΔT - dR _{OND} VM × ΔV	1.1 + dR _{OND} TM × ΔT + dR _{OND} VM × ΔV	RZQ/6
R _{ON(PU)} @ 0.8 × V _{DDQ}	0.6 - dR _{OND} TH × ΔT - dR _{OND} VH × ΔV	1.1 + dR _{OND} TH × ΔT + dR _{OND} VH × ΔV	RZQ/6

Note: 1. $ΔT = T - T_{(@CALIBRATION)}$; $ΔV = V_{DDQ} - V_{DDQ(@CALIBRATION)}$; and $V_{DD} = V_{DDQ}$.

Table 46: 40 Ohm Output Driver Voltage and Temperature Sensitivity

Change	Min	Max	Unit
$dR_{ON}dTM$	0	1.5	%/°C
$dR_{ON}dVM$	0	0.15	%/mV
$dR_{ON}dTL$	0	1.5	%/°C
$dR_{ON}dVL$	0	0.15	%/mV
$dR_{ON}dTH$	0	1.5	%/°C
$dR_{ON}dVH$	0	0.15	%/mV

Output Characteristics and Operating Conditions

The DRAM uses both single-ended and differential output drivers. The single-ended output driver is summarized below, while the differential output driver is summarized in Table 48 (page 68).

Table 47: DDR3L Single-Ended Output Driver Characteristics

All voltages are referenced to V_{SS}

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Output leakage current: DQ are disabled; $0V \leq V_{OUT} \leq V_{DDQ}$; ODT is disabled; ODT is HIGH	I_{OZ}	-5	5	μA	1
Output slew rate: Single-ended; For rising and falling edges, measure between $V_{OL(AC)} = V_{REF} - 0.09 \times V_{DDQ}$ and $V_{OH(AC)} = V_{REF} + 0.09 \times V_{DDQ}$	SRQ_{se}	1.75	6	V/ns	1, 2, 3, 4
Single-ended DC high-level output voltage	$V_{OH(DC)}$	$0.8 \times V_{DDQ}$		V	1, 2, 5
Single-ended DC mid-point level output voltage	$V_{OM(DC)}$	$0.5 \times V_{DDQ}$		V	1, 2, 5
Single-ended DC low-level output voltage	$V_{OL(DC)}$	$0.2 \times V_{DDQ}$		V	1, 2, 5
Single-ended AC high-level output voltage	$V_{OH(AC)}$	$V_{TT} + 0.1 \times V_{DDQ}$		V	1, 2, 3, 6
Single-ended AC low-level output voltage	$V_{OL(AC)}$	$V_{TT} - 0.1 \times V_{DDQ}$		V	1, 2, 3, 6
Delta R_{ON} between pull-up and pull-down for DQ/DQS	MM_{PUPD}	-10	10	%	1, 7
Test load for AC timing and output slew rates	Output to V_{TT} ($V_{DDQ}/2$) via 25Ω resistor				3

Notes:

1. RZQ of $240\Omega \pm 1\%$ with RZQ/7 enabled (default 34Ω driver) and is applicable after proper ZQ calibration has been performed at a stable temperature and voltage ($V_{DDQ} = V_{DD}$; $V_{SSQ} = V_{SS}$).
2. $V_{TT} = V_{DDQ}/2$.
3. See Figure 29 (page 70) for the test load configuration.
4. The 6 V/ns maximum is applicable for a single DQ signal when it is switching either from HIGH to LOW or LOW to HIGH while the remaining DQ signals in the same byte lane are either all static or all switching in the opposite direction. For all other DQ signal switching combinations, the maximum limit of 6 V/ns is reduced to 5 V/ns.
5. See Figure 26 (page 61) for IV curve linearity. Do not use AC test load.
6. See Table 50 (page 70) for output slew rate.
7. See Figure 26 (page 61) for additional information.
8. See Figure 27 (page 68) for an example of a single-ended output signal.

Figure 27: DQ Output Signal

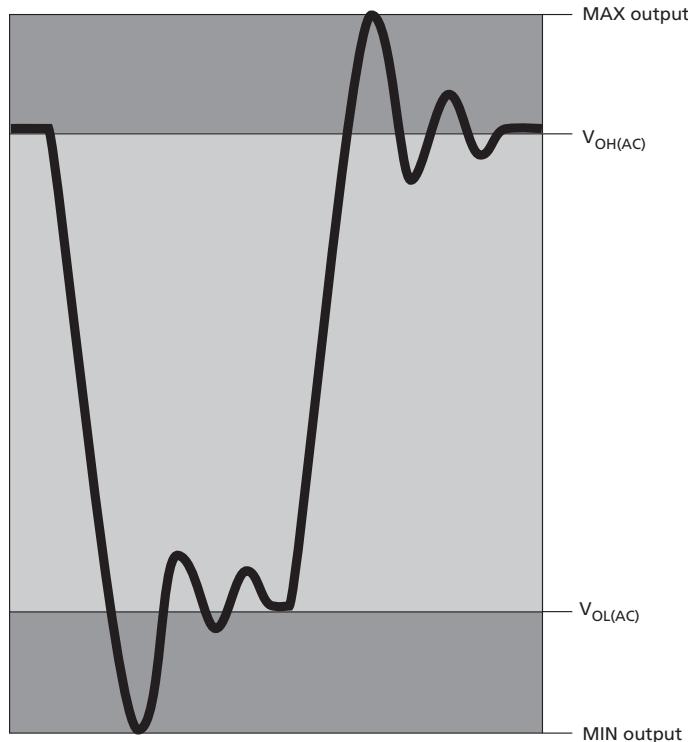


Table 48: DDR3L Differential Output Driver Characteristics

 All voltages are referenced to V_{SS}

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Output leakage current: DQ are disabled; $0V \leq V_{OUT} \leq V_{DDQ}$; ODT is disabled; ODT is HIGH	I_{OZ}	-5	5	μA	1
DDR3L Output slew rate: Differential; For rising and falling edges, measure between $V_{OL,diff(AC)} = -0.18 \times V_{DDQ}$ and $V_{OH,diff(AC)} = 0.18 \times V_{DDQ}$	SRQ_{diff}	3.5	12	V/ns	1
Differential high-level output voltage	$V_{OH,diff(AC)}$	$+0.2 \times V_{DDQ}$		V	1, 4
Differential low-level output voltage	$V_{OL,diff(AC)}$	$-0.2 \times V_{DDQ}$		V	1, 4
Delta Ron between pull-up and pull-down for DQ/DQS	MM_{PUPD}	-10	10	%	1, 5
Test load for AC timing and output slew rates	Output to V_{TT} ($V_{DDQ}/2$) via 25Ω resistor				3

Notes:

1. RZQ of $240\Omega \pm 1\%$ with RZQ/7 enabled (default 34Ω driver) and is applicable after proper ZQ calibration has been performed at a stable temperature and voltage ($V_{DDQ} = V_{DD}$; $V_{SSQ} = V_{SS}$).
2. $V_{REF} = V_{DDQ}/2$; slew rate @ 5 V/ns, interpolate for faster slew rate.
3. See Figure 29 (page 70) for the test load configuration.
4. See Table 51 (page 72) for the output slew rate.
5. See Table 37 (page 62) for additional information.
6. See Figure 28 (page 69) for an example of a differential output signal.

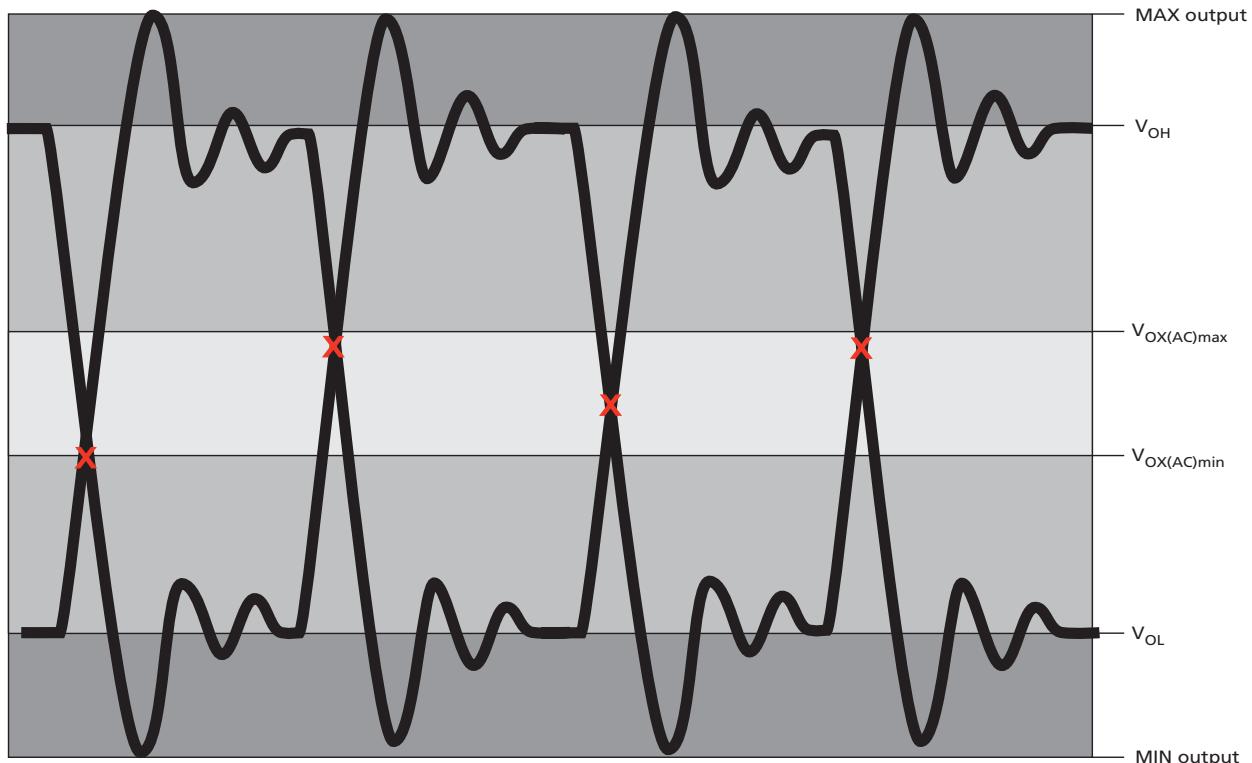
Table 49: DDR3L Differential Output Driver Characteristics $V_{OX(AC)}$

 All voltages are referenced to V_{SS}

Parameter/Condition	Symbol	DDR3L- 800/1066/1333 DQS/DQS# Differential Slew Rate										Unit
		3.5V/ns	4V/ns	5V/ns	6V/ns	7V/ns	8V/ns	9V/ns	10V/ns	12V/ns		
Output differential crosspoint voltage	$V_{OX(AC)}$	Max	+115	+130	+135	+195	+205	+205	+205	+205	+205	mV
		Min	-115	-130	-135	-195	-205	-205	-205	-205	-205	mV
Parameter/Condition	Symbol	DDR3L-1600/1866 DQS/DQS# Differential Slew Rate										Unit
		3.5V/ns	4V/ns	5V/ns	6V/ns	7V/ns	8V/ns	9V/ns	10V/ns	12V/ns		
Output differential crosspoint voltage	$V_{OX(AC)}$	Max	+90	+105	+135	+155	+180	+205	+205	+205	+205	mV
		Min	-90	-105	-135	-155	-180	-205	-205	-205	-205	mV

Notes:

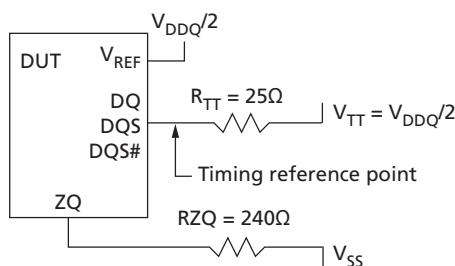
1. RZQ of $240\Omega \pm 1\%$ with RZQ/7 enabled (default 34Ω driver) and is applicable after proper ZQ calibration has been performed at a stable temperature and voltage ($V_{DDQ} = V_{DD}$; $V_{SSQ} = V_{SS}$).
2. See Figure 29 (page 70) for the test load configuration.
3. See Figure 28 (page 69) for an example of a differential output signal.
4. For a differential slew rate between the list values, the $V_{OX(AC)}$ value may be obtained by linear interpolation.

Figure 28: Differential Output Signal


Reference Output Load

Figure 29 represents the effective reference load of 25Ω used in defining the relevant device AC timing parameters (except ODT reference timing) as well as the output slew rate measurements. It is not intended to be a precise representation of a particular system environment or a depiction of the actual load presented by a production tester. System designers should use IBIS or other simulation tools to correlate the timing reference load to a system environment.

Figure 29: Reference Output Load for AC Timing and Output Slew Rate



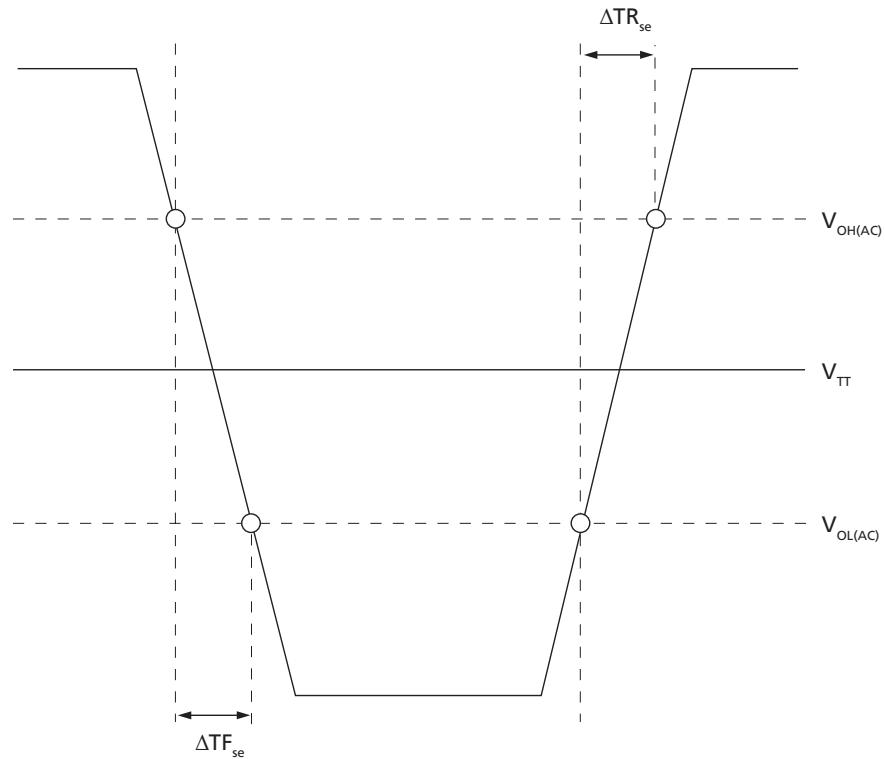
Slew Rate Definitions for Single-Ended Output Signals

The single-ended output driver is summarized in Table 47 (page 67). With the reference load for timing measurements, the output slew rate for falling and rising edges is defined and measured between $V_{OL(AC)}$ and $V_{OH(AC)}$ for single-ended signals.

Table 50: Single-Ended Output Slew Rate Definition

Single-Ended Output Slew Rates (Linear Signals)		Measured		Calculation
Output	Edge	From	To	
DQ	Rising	$V_{OL(AC)}$	$V_{OH(AC)}$	$\frac{V_{OH(AC)} - V_{OL(AC)}}{\Delta T R_{se}}$
	Falling	$V_{OH(AC)}$	$V_{OL(AC)}$	$\frac{V_{OH(AC)} - V_{OL(AC)}}{\Delta T F_{se}}$

Figure 30: Nominal Slew Rate Definition for Single-Ended Output Signals



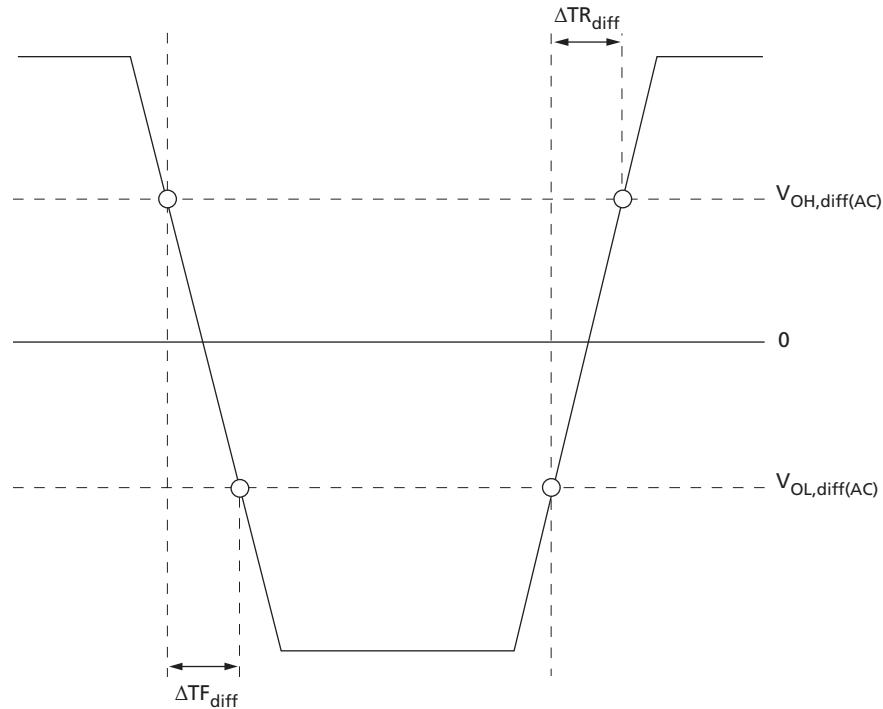
Slew Rate Definitions for Differential Output Signals

The differential output driver is summarized in Table 48 (page 68). With the reference load for timing measurements, the output slew rate for falling and rising edges is defined and measured between $V_{OL(AC)}$ and $V_{OH(AC)}$ for differential signals.

Table 51: Differential Output Slew Rate Definition

Differential Output Slew Rates (Linear Signals)		Measured		Calculation
Output	Edge	From	To	
DQS, DQS#	Rising	$V_{OL,diff(AC)}$	$V_{OH,diff(AC)}$	$\frac{V_{OH,diff(AC)} - V_{OL,diff(AC)}}{\Delta TR_{diff}}$
	Falling	$V_{OH,diff(AC)}$	$V_{OL,diff(AC)}$	$\frac{V_{OH,diff(AC)} - V_{OL,diff(AC)}}{\Delta TF_{diff}}$

Figure 31: Nominal Differential Output Slew Rate Definition for DQS, DQS#



Speed Bin Tables

Table 52: DDR3L-1066 Speed Bins

DDR3L-1066 Speed Bin			-187E		-187		Unit	Notes
CL-tRCD-tRP			7-7-7		8-8-8			
Parameter	Symbol		Min	Max	Min	Max		
Internal READ command to first data	^t AA		13.125	–	15	–	ns	
ACTIVATE to internal READ or WRITE delay time	^t RCD		13.125	–	15	–	ns	
PRECHARGE command period	^t RP		13.125	–	15	–	ns	
ACTIVATE-to-ACTIVATE or REFRESH command period	^t RC		50.625	–	52.5	–	ns	
ACTIVATE-to-PRECHARGE command period	^t RAS		37.5	9 x ^t REFI	37.5	9 x ^t REFI	ns	1
CL = 5	CWL = 5	^t CK (AVG)	3.0	3.3	3.0	3.3	ns	2
	CWL = 6	^t CK (AVG)	Reserved		Reserved		ns	3
CL = 6	CWL = 5	^t CK (AVG)	2.5	3.3	2.5	3.3	ns	2
	CWL = 6	^t CK (AVG)	Reserved		Reserved		ns	3
CL = 7	CWL = 5	^t CK (AVG)	Reserved		Reserved		ns	3
	CWL = 6	^t CK (AVG)	1.875	<2.5	Reserved		ns	2, 3
CL = 8	CWL = 5	^t CK (AVG)	Reserved		Reserved		ns	3
	CWL = 6	^t CK (AVG)	1.875	<2.5	1.875	<2.5	ns	2
Supported CL settings			5, 6, 7, 8		5, 6, 8		CK	
Supported CWL settings			5, 6		5, 6		CK	

Notes:

1. ^tREFI depends on T_{OPER} .
2. The CL and CWL settings result in ^tCK requirements. When making a selection of ^tCK, both CL and CWL requirement settings need to be fulfilled.
3. Reserved settings are not allowed.

Table 53: DDR3L-1333 Speed Bins

DDR3L-1333 Speed Bin			-15E ¹		-15 ²		Unit	Notes		
CL-tRCD-tRP			9-9-9		10-10-10					
Parameter	Symbol		Min	Max	Min	Max				
Internal READ command to first data	^t AA		13.5	–	15	–	ns			
ACTIVATE to internal READ or WRITE delay time	^t RCD		13.5	–	15	–	ns			
PRECHARGE command period	^t RP		13.5	–	15	–	ns			
ACTIVATE-to-ACTIVATE or REFRESH command period	^t RC		49.5	–	51	–	ns			
ACTIVATE-to-PRECHARGE command period	^t RAS		36	9 x ^t REFI	36	9 x ^t REFI	ns	3		
CL = 5	CWL = 5	^t CK (AVG)	3.0	3.3	3.0	3.3	ns	4		
	CWL = 6, 7	^t CK (AVG)	Reserved		Reserved		ns	5		
CL = 6	CWL = 5	^t CK (AVG)	2.5	3.3	2.5	3.3	ns	4		
	CWL = 6	^t CK (AVG)	Reserved		Reserved		ns	5		
	CWL = 7	^t CK (AVG)	Reserved		Reserved		ns	5		
CL = 7	CWL = 5	^t CK (AVG)	Reserved		Reserved		ns	5		
	CWL = 6	^t CK (AVG)	1.875	<2.5	Reserved		ns	4, 5		
	CWL = 7	^t CK (AVG)	Reserved		Reserved		ns	5		
CL = 8	CWL = 5	^t CK (AVG)	Reserved		Reserved		ns	5		
	CWL = 6	^t CK (AVG)	1.875	<2.5	1.875	<2.5	ns	4		
	CWL = 7	^t CK (AVG)	Reserved		Reserved		ns	5		
CL = 9	CWL = 5, 6	^t CK (AVG)	Reserved		Reserved		ns	5		
	CWL = 7	^t CK (AVG)	1.5	<1.875	Reserved		ns	4, 5		
CL = 10	CWL = 5, 6	^t CK (AVG)	Reserved		Reserved		ns	5		
	CWL = 7	^t CK (AVG)	1.5	<1.875	1.5	<1.875	ns	4		
Supported CL settings			5, 6, 7, 8, 9, 10		5, 6, 8, 10		CK			
Supported CWL settings			5, 6, 7		5, 6, 7		CK			

Notes:

1. The -15E speed grade is backward compatible with 1066, CL = 7 (-187E).
2. The -15 speed grade is backward compatible with 1066, CL = 8 (-187).
3. ^tREFI depends on T_{OPER} .
4. The CL and CWL settings result in ^tCK requirements. When making a selection of ^tCK, both CL and CWL requirement settings need to be fulfilled.
5. Reserved settings are not allowed.

Table 54: DDR3L-1600 Speed Bins

DDR3L-1600 Speed Bin		-125 ¹		Unit	Notes
CL-tRCD-tRP		11-11-11			
Parameter	Symbol	Min	Max		
Internal READ command to first data	^t AA	13.75	–	ns	
ACTIVATE to internal READ or WRITE delay time	^t RCD	13.75	–	ns	
PRECHARGE command period	^t RP	13.75	–	ns	
ACTIVATE-to-ACTIVATE or REFRESH command period	^t RC	48.75	–	ns	
ACTIVATE-to-PRECHARGE command period	^t RAS	35	9 x ^t REFI	ns	2
CL = 5	CWL = 5	^t CK (AVG)	3.0	3.3	ns 3
	CWL = 6, 7, 8	^t CK (AVG)	Reserved		ns 4
CL = 6	CWL = 5	^t CK (AVG)	2.5	3.3	ns 3
	CWL = 6	^t CK (AVG)	Reserved		ns 4
	CWL = 7, 8	^t CK (AVG)	Reserved		ns 4
CL = 7	CWL = 5	^t CK (AVG)	Reserved		ns 4
	CWL = 6	^t CK (AVG)	1.875	<2.5	ns 3
	CWL = 7	^t CK (AVG)	Reserved		ns 4
	CWL = 8	^t CK (AVG)	Reserved		ns 4
CL = 8	CWL = 5	^t CK (AVG)	Reserved		ns 4
	CWL = 6	^t CK (AVG)	1.875	<2.5	ns 3
	CWL = 7	^t CK (AVG)	Reserved		ns 4
	CWL = 8	^t CK (AVG)	Reserved		ns 4
CL = 9	CWL = 5, 6	^t CK (AVG)	Reserved		ns 4
	CWL = 7	^t CK (AVG)	1.5	<1.875	ns 3
	CWL = 8	^t CK (AVG)	Reserved		ns 4
CL = 10	CWL = 5, 6	^t CK (AVG)	Reserved		ns 4
	CWL = 7	^t CK (AVG)	1.5	<1.875	ns 3
	CWL = 8	^t CK (AVG)	Reserved		ns 4
CL = 11	CWL = 5, 6, 7	^t CK (AVG)	Reserved		ns 4
	CWL = 8	^t CK (AVG)	1.25	<1.5	ns 3
Supported CL settings		5, 6, 7, 8, 9, 10, 11		CK	
Supported CWL settings		5, 6, 7, 8		CK	

Notes:

1. The -125 speed grade is backward compatible with 1333, CL = 9 (-15E) and 1066, CL = 7 (-187E).
2. ^tREFI depends on T_{OPER} .
3. The CL and CWL settings result in ^tCK requirements. When making a selection of ^tCK, both CL and CWL requirement settings need to be fulfilled.
4. Reserved settings are not allowed.

Table 55: DDR3L-1866 Speed Bins

DDR3L-1866 Speed Bin			-107 ¹		Unit	Notes		
CL-tRCD-tRP			13-13-13					
Parameter	Symbol		Min	Max				
Internal READ command to first data	^t AA		13.91	20				
ACTIVATE to internal READ or WRITE delay time	^t RCD		13.91	—	ns			
PRECHARGE command period	^t RP		13.91	—	ns			
ACTIVATE-to-ACTIVATE or REFRESH command period	^t RC		47.91	—	ns			
ACTIVATE-to-PRECHARGE command period	^t RAS		34	9 x ^t REFI	ns	2		
CL = 5	CWL = 5	^t CK (AVG)	3.0	3.3	ns	3		
	CWL = 6, 7, 8, 9	^t CK (AVG)	Reserved		ns	4		
CL = 6	CWL = 5	^t CK (AVG)	2.5	3.3	ns	3		
	CWL = 6, 7, 8, 9	^t CK (AVG)	Reserved		ns	4		
CL = 7	CWL = 5, 7, 8, 9	^t CK (AVG)	Reserved		ns	4		
	CWL = 6	^t CK (AVG)	1.875	<2.5	ns	3		
CL = 8	CWL = 5, 8, 9	^t CK (AVG)	Reserved		ns	4		
	CWL = 6	^t CK (AVG)	1.875	<2.5	ns	3		
	CWL = 7	^t CK (AVG)	Reserved		ns	4		
CL = 9	CWL = 5, 6, 8, 9	^t CK (AVG)	Reserved		ns	4		
	CWL = 7	^t CK (AVG)	1.5	<1.875	ns	3		
CL = 10	CWL = 5, 6, 9	^t CK (AVG)	Reserved		ns	4		
	CWL = 7	^t CK (AVG)	1.5	<1.875	ns	3		
	CWL = 8	^t CK (AVG)	Reserved		ns	4		
CL = 11	CWL = 5, 6, 7	^t CK (AVG)	Reserved		ns	4		
	CWL = 8	^t CK (AVG)	1.25	<1.5	ns	3		
	CWL = 9	^t CK (AVG)	Reserved		ns	4		
CL = 12	CWL = 5, 6, 7, 8	^t CK (AVG)	Reserved		ns	4		
	CWL = 9	^t CK (AVG)	Reserved		ns	4		
CL = 13	CWL = 5, 6, 7, 8	^t CK (AVG)	Reserved		ns	4		
	CWL = 9	^t CK (AVG)	1.071	<1.25	ns	3		
Supported CL settings			5, 6, 7, 8, 9, 10, 11, 13		CK			
Supported CWL settings			5, 6, 7, 8, 9		CK			

Notes:

1. The -107 speed grade is backward compatible with 1600, CL = 11 (-125), 1333, CL = 9 (-15E) and 1066, CL = 7 (-187E).
2. ^tREFI depends on T_{OPER} .
3. The CL and CWL settings result in ^tCK requirements. When making a selection of ^tCK, both CL and CWL requirement settings need to be fulfilled.
4. Reserved settings are not allowed.

Electrical Characteristics and AC Operating Conditions

Table 56: Electrical Characteristics and AC Operating Conditions

Note 1 below applies to base timing specifications; for additional specifications see the DDR3 Electrical Characteristics and AC Operating Condition Tables and Notes

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		DDR3L-1866		Units	
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max		
DQ Input Timing													
Data setup time to DQS, DQS#	Base (specification)	t_{DS} (AC160)	90	—	40	—	N/A	—	N/A	—	N/A	—	ps
	V_{REF} @ 1 V/ns		250	—	200	—	N/A	—	N/A	—	N/A	—	ps
Data setup time to DQS, DQS#	Base (specification)	t_{DS} (AC135)	140	—	90	—	45	—	25	—	N/A	—	ps
	V_{REF} @ 1 V/ns		275	—	225	—	180	—	160	—	N/A	—	ps
Data hold time from DQS, DQS#	Base (specification)	t_{DH} (DC90)	160	—	110	—	75	—	55	—	N/A	—	ps
	V_{REF} @ 1 V/ns		250	—	200	—	165	—	145	—	N/A	—	ps
Data setup time to DQS, DQS#	Base (specification)	t_{DS} (AC130)	N/A	—	N/A	—	N/A	—	N/A	—	70	—	ps
	V_{REF} @ 2 V/ns		N/A	—	N/A	—	N/A	—	N/A	—	135	—	ps
Data hold time from DQS, DQS#	Base (specification)	t_{DH} (DC90)	N/A	—	N/A	—	N/A	—	N/A	—	75	—	ps
	V_{REF} @ 2 V/ns		N/A	—	N/A	—	N/A	—	N/A	—	110	—	ps
Command and Address Timing													
CTRL, CMD, ADDR setup to CK, CK#	Base (specification)	t_{IS} (AC160)	215	—	140	—	80	—	60	—	N/A	—	ps
	V_{REF} @ 1 V/ns		375	—	300	—	240	—	220	—	N/A	—	ps
CTRL, CMD, ADDR setup to CK, CK#	Base (specification)	t_{IS} (AC135)	365	—	290	—	205	—	185	—	65	—	ps
	V_{REF} @ 1 V/ns		500	—	425	—	340	—	320	—	200	—	ps
CTRL, CMD, ADDR setup to CK, CK#	Base (specification)	t_{IS} (AC125)	N/A	—	N/A	—	N/A	—	N/A	—	150	—	ps
	V_{REF} @ 1 V/ns		N/A	—	N/A	—	N/A	—	N/A	—	275	—	ps
CTRL, CMD, ADDR hold from CK, CK#	Base (specification)	t_{IH} (DC90)	285	—	210	—	150	—	130	—	110	—	ps
	V_{REF} @ 1 V/ns		375	—	300	—	240	—	220	—	200	—	ps

Notes: 1. When two $V_{IH(AC)}$ values (and two corresponding $V_{IL(AC)}$ values) are listed for a specific speed bin, the user may choose either value for the input AC level. Whichever value is used, the associated setup time for that AC level must also be used. Additionally, one $V_{IH(AC)}$ value may be used for address/command inputs and the other $V_{IH(AC)}$ value may be used for data inputs.

For example, for DDR3-800, two input AC levels are defined: $V_{IH(AC160),min}$ and $V_{IH(AC135),min}$ (corresponding $V_{IL(AC160),min}$ and $V_{IL(AC135),min}$). For DDR3-800, the address/command inputs must use either $V_{IH(AC160),min}$ with $t_{IS}(AC160)$ of 215ps or $V_{IH(AC135),min}$

with $t_{IS(AC135)}$ of 365ps; independently, the data inputs must use either $V_{IH(AC160),min}$ with $t_{DS(AC160)}$ of 90ps or $V_{IH(AC135),min}$ with $t_{DS(AC135)}$ of 140ps.

2. When DQ single-ended slew rate is 1V/ns, the DQS differential slew rate is 2V/ns; when DQ single-ended slew rate is 2V/ns, the DQS differential slew rate is 4V/ns.

Table 57: Electrical Characteristics and AC Operating Conditions

Notes 1–8 apply to the entire table

Table 57: Electrical Characteristics and AC Operating Conditions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		Unit	Notes
		Min	Max	Min	Max	Min	Max	Min	Max		
Data setup time to DQS, DQS#	tDS (AC175)	75	–	25	–	–	–	–	–	ps	18, 19, 44
		250	–	200	–	–	–	–	–	ps	19, 20
Data setup time to DQS, DQS#	tDS (AC150)	125	–	75	–	30	–	10	–	ps	18, 19, 44
		275	–	250	–	180	–	160	–	ps	19, 20
Data setup time to DQS, DQS#	tDS (AC135)	–	–	–	–	–	–	–	–	ps	18, 19
		–	–	–	–	–	–	–	–	ps	19, 20
Data hold time from DQS, DQS#	tDH (DC100)	150	–	100	–	65	–	45	–	ps	18, 19
		250	–	200	–	165	–	145	–	ps	19, 20
Minimum data pulse width	tDIPW	600	–	490	–	400	–	360	–	ps	41
DQ Output Timing											
DQS, DQS# to DQ skew, per access	tDQSQ	–	200	–	150	–	125	–	100	ps	
DQ output hold time from DQS, DQS#	tQH	0.38	–	0.38	–	0.38	–	0.38	–	tCK (AVG)	21
DQ Low-Z time from CK, CK#	tLZDQ	–800	400	–600	300	–500	250	–450	225	ps	22, 23
DQ High-Z time from CK, CK#	tHZDQ	–	400	–	300	–	250	–	225	ps	22, 23
DQ Strobe Input Timing											
DQS, DQS# rising to CK, CK# rising	tDQSS	–0.25	0.25	–0.25	0.25	–0.25	0.25	–0.27	0.27	CK	25
DQS, DQS# differential input low pulse width	tDQSL	0.45	0.55	0.45	0.55	0.45	0.55	0.45	0.55	CK	
DQS, DQS# differential input high pulse width	tDQSH	0.45	0.55	0.45	0.55	0.45	0.55	0.45	0.55	CK	
DQS, DQS# falling setup to CK, CK# rising	tDSS	0.2	–	0.2	–	0.2	–	0.18	–	CK	25
DQS, DQS# falling hold from CK, CK# rising	tDSH	0.2	–	0.2	–	0.2	–	0.18	–	CK	25
DQS, DQS# differential WRITE preamble	tWPRE	0.9	–	0.9	–	0.9	–	0.9	–	CK	
DQS, DQS# differential WRITE postamble	tWPST	0.3	–	0.3	–	0.3	–	0.3	–	CK	
DQ Strobe Output Timing											
DQS, DQS# rising to/from rising CK, CK#	tDQSCK	–400	400	–300	300	–255	255	–225	225	ps	23
DQS, DQS# rising to/from rising CK, CK# when DLL is disabled	tDQSCK (DLL_DIS)	1	10	1	10	1	10	1	10	ns	26
DQS, DQS# differential output high time	tQSH	0.38	–	0.38	–	0.40	–	0.40	–	CK	21

Table 57: Electrical Characteristics and AC Operating Conditions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		Unit	Notes	
		Min	Max	Min	Max	Min	Max	Min	Max			
DQS, DQS# differential output low time	t_{QSL}	0.38	–	0.38	–	0.40	–	0.40	–	CK	21	
DQS, DQS# Low-Z time (RL - 1)	t_{LZDQS}	–800	400	–600	300	–500	250	–450	225	ps	22, 23	
DQS, DQS# High-Z time (RL + BL/2)	t_{HZDQS}	–	400	–	300	–	250	–	225	ps	22, 23	
DQS, DQS# differential READ preamble	t_{RPRE}	0.9	Note 24	0.9	Note 24	0.9	Note 24	0.9	Note 24	CK	23, 24	
DQS, DQS# differential READ postamble	t_{RPST}	0.3	Note 27	0.3	Note 27	0.3	Note 27	0.3	Note 27	CK	23, 27	
Command and Address Timing												
DLL locking time	t_{DLLK}	512	–	512	–	512	–	512	–	CK	28	
CTRL, CMD, ADDR setup to CK,CK#	Base (specification)	t_{IS} (AC175)	200	–	125	–	65	–	45	–	ps	29, 30, 44
			375	–	300	–	240	–	220	–	ps	20, 30
CTRL, CMD, ADDR setup to CK,CK#	Base (specification)	t_{IS} (AC150)	350	–	275	–	190	–	170	–	ps	29, 30, 44
			500	–	425	–	340	–	320	–	ps	20, 30
CTRL, CMD, ADDR hold from CK,CK#	Base (specification)	t_{IH} (DC100)	275	–	200	–	140	–	120	–	ps	29, 30
			375	–	300	–	240	–	220	–	ps	20, 30
Minimum CTRL, CMD, ADDR pulse width	t_{IPW}	900	–	780	–	620	–	560	–	ps	41	
ACTIVATE to internal READ or WRITE delay	t_{RCD}	See Speed Bin Tables (page 73) for t_{RCD}								ns	31	
PRECHARGE command period	t_{RP}	See Speed Bin Tables (page 73) for t_{RP}								ns	31	
ACTIVATE-to-PRECHARGE command period	t_{RAS}	See Speed Bin Tables (page 73) for t_{RAS}								ns	31, 32	
ACTIVATE-to-ACTIVATE command period	t_{RC}	See Speed Bin Tables (page 73) for t_{RC}								ns	31, 43	
ACTIVATE-to-ACTIVATE minimum command period	x4/x8 (1KB page size)	t_{RRD}	MIN = greater of 4CK or 10ns		MIN = greater of 4CK or 7.5ns		MIN = greater of 4CK or 6ns		MIN = greater of 4CK or 6ns		CK	31
	x16 (2KB page size)		MIN = greater of 4CK or 10ns						MIN = greater of 4CK or 7.5ns		CK	31
Four ACTIVATE windows	x4/x8 (1KB page size)	t_{FAW}	40	–	37.5	–	30	–	30	–	ns	31
	x16 (2KB page size)		50	–	50	–	45	–	40	–	ns	31
Write recovery time	t_{WR}	MIN = 15ns; MAX = N/A								ns	31, 32, 33, 34	
Delay from start of internal WRITE transaction to internal READ command	t_{WTR}	MIN = greater of 4CK or 7.5ns; MAX = N/A								CK	31, 34	
READ-to-PRECHARGE time	t_{RTP}	MIN = greater of 4CK or 7.5ns; MAX = N/A								CK	31, 32	

Table 57: Electrical Characteristics and AC Operating Conditions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		Unit	Notes				
		Min	Max	Min	Max	Min	Max	Min	Max						
Exit self refresh to commands requiring a locked DLL	t_{XSDLL}	MIN = t_{DLLK} (MIN); MAX = N/A						CK		28					
Minimum CKE low pulse width for self refresh entry to self refresh exit timing	t_{CKESR}	MIN = t_{CKE} (MIN) + CK; MAX = N/A						CK							
Valid clocks after self refresh entry or power-down entry	t_{CKSRE}	MIN = greater of 5CK or 10ns; MAX = N/A						CK							
Valid clocks before self refresh exit, power-down exit, or reset exit	t_{CKSRX}	MIN = greater of 5CK or 10ns; MAX = N/A						CK							
Power-Down Timing															
CKE MIN pulse width	t_{CKE} (MIN)	Greater of 3CK or 7.5ns		Greater of 3CK or 5.625ns		Greater of 3CK or 5.625ns		Greater of 3CK or 5ns		CK					
Command pass disable delay	t_{CPDED}	MIN = 1; MAX = N/A						CK							
Power-down entry to power-down exit timing	t_{PD}	MIN = t_{CKE} (MIN); MAX = 9 × t_{REFI}						CK							
Begin power-down period prior to CKE registered HIGH	t_{ANPD}	WL - 1CK						CK							
Power-down entry period: ODT either synchronous or asynchronous	PDE	Greater of t_{ANPD} or t_{RFC} - REFRESH command to CKE LOW time						CK							
Power-down exit period: ODT either synchronous or asynchronous	PDX	$t_{ANPD} + t_{XPDLL}$						CK							
Power-Down Entry Minimum Timing															
ACTIVATE command to power-down entry	$t_{ACTPDEN}$	MIN = 1						CK							
PRECHARGE/PRECHARGE ALL command to power-down entry	t_{PRPDEN}	MIN = 1						CK							
REFRESH command to power-down entry	$t_{REFPDEN}$	MIN = 1						CK		37					
MRS command to power-down entry	$t_{MRSPDEN}$	MIN = t_{MOD} (MIN)						CK							
READ/READ with auto precharge command to power-down entry	t_{RDPDEN}	MIN = RL + 4 + 1						CK							
WRITE command to power-down entry	BL8 (OTF, MRS) BC4OTF	t_{WRPDEN}	MIN = WL + 4 + t_{WR}/t_{CK} (AVG)						CK						
	BC4MRS	t_{WRPDEN}	MIN = WL + 2 + t_{WR}/t_{CK} (AVG)						CK						

Table 57: Electrical Characteristics and AC Operating Conditions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		Unit	Notes
		Min	Max	Min	Max	Min	Max	Min	Max		
WRITE with auto precharge command to power-down entry	BL8 (OTF, MRS) BC4OTF	^t WRAP-DEN		MIN = WL + 4 + WR + 1						CK	
	BC4MRS	^t WRAP-DEN		MIN = WL + 2 + WR + 1						CK	
Power-Down Exit Timing											
DLL on, any valid command, or DLL off to commands not requiring locked DLL	^t XP	MIN = greater of 3CK or 7.5ns; MAX = N/A		MIN = greater of 3CK or 6ns; MAX = N/A						CK	
Precharge power-down with DLL off to commands requiring a locked DLL	^t XPDLL	MIN = greater of 10CK or 24ns; MAX = N/A								CK	28
ODT Timing											
R_{TT} synchronous turn-on delay	ODTLon			CWL + AL - 2CK						CK	38
R_{TT} synchronous turn-off delay	ODTLooff			CWL + AL - 2CK						CK	40
R_{TT} turn-on from ODTL on reference	^t AON	-400	400	-300	300	-250	250	-225	225	ps	23, 38
R_{TT} turn-off from ODTL off reference	^t AOF	0.3	0.7	0.3	0.7	0.3	0.7	0.3	0.7	CK	39, 40
Asynchronous R_{TT} turn-on delay (power-down with DLL off)	^t AONPD	MIN = 2; MAX = 8.5								ns	38
Asynchronous R_{TT} turn-off delay (power-down with DLL off)	^t AOPD	MIN = 2; MAX = 8.5								ns	40
ODT HIGH time with WRITE command and BL8	ODTH8	MIN = 6; MAX = N/A								CK	
ODT HIGH time without WRITE command or with WRITE command and BC4	ODTH4	MIN = 4; MAX = N/A								CK	
Dynamic ODT Timing											
$R_{TT,nom}$ -to- $R_{TT(WR)}$ change skew	ODTLcnw			WL - 2CK						CK	
$R_{TT(WR)}$ -to- $R_{TT,nom}$ change skew - BC4	ODTLcwn4			4CK + ODTLooff						CK	
$R_{TT(WR)}$ -to- $R_{TT,nom}$ change skew - BL8	ODTLcwn8			6CK + ODTLooff						CK	
R_{TT} dynamic change skew	^t ADC	0.3	0.7	0.3	0.7	0.3	0.7	0.3	0.7	CK	39
Write Leveling Timing											
First DQS, DQS# rising edge	^t WLMRD	40	–	40	–	40	–	40	–	CK	
DQS, DQS# delay	^t WLQSEN	25	–	25	–	25	–	25	–	CK	
Write leveling setup from rising CK, CK# crossing to rising DQS, DQS# crossing	^t WLS	325	–	245	–	195	–	165	–	ps	

Table 57: Electrical Characteristics and AC Operating Conditions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-800		DDR3L-1066		DDR3L-1333		DDR3L-1600		Unit	Notes
		Min	Max	Min	Max	Min	Max	Min	Max		
Write leveling hold from rising DQS, DQS# crossing to rising CK, CK# crossing	^t WLH	325	–	245	–	195	–	165	–	ps	
Write leveling output delay	^t WLO	0	9	0	9	0	9	0	7.5	ns	
Write leveling output error	^t WLOE	0	2	0	2	0	2	0	2	ns	

Notes: 1. AC timing parameters are valid from specified T_C MIN to T_C MAX values.
2. All voltages are referenced to V_{SS} .
3. Output timings are only valid for RO_{N34} output buffer selection.
4. The unit t_{CK} (AVG) represents the actual t_{CK} (AVG) of the input clock under operation. The unit CK represents one clock cycle of the input clock, counting the actual clock edges.
5. AC timing and I_{DD} tests may use a V_{IL} -to- V_{IH} swing of up to 900mV in the test environment, but input timing is still referenced to V_{REF} (except t_{IS} , t_{IH} , t_{DS} , and t_{DH} use the AC/DC trip points and CK, CK# and DQS, DQS# use their crossing points). The minimum slew rate for the input signals used to test the device is 1 V/ns for single-ended inputs and 2 V/ns for differential inputs in the range between $V_{IL(AC)}$ and $V_{IH(AC)}$.
6. All timings that use time-based values (ns, μ s, ms) should use t_{CK} (AVG) to determine the correct number of clocks (Table 57 (page 79) uses CK or t_{CK} [AVG] interchangeably). In the case of noninteger results, all minimum limits are to be rounded up to the nearest whole integer, and all maximum limits are to be rounded down to the nearest whole integer.
7. Strobe or DQS_{diff} refers to the DQS and DQS# differential crossing point when DQS is the rising edge. Clock or CK refers to the CK and CK# differential crossing point when CK is the rising edge.
8. This output load is used for all AC timing (except ODT reference timing) and slew rates. The actual test load may be different. The output signal voltage reference point is $V_{DDQ}/2$ for single-ended signals and the crossing point for differential signals (see Figure 28 (page 69)).
9. When operating in DLL disable mode, Micron does not warrant compliance with normal mode timings or functionality.
10. The clock's t_{CK} (AVG) is the average clock over any 200 consecutive clocks and t_{CK} (AVG) MIN is the smallest clock rate allowed, with the exception of a deviation due to clock jitter. Input clock jitter is allowed provided it does not exceed values specified and must be of a random Gaussian distribution in nature.
11. Spread spectrum is not included in the jitter specification values. However, the input clock can accommodate spread-spectrum at a sweep rate in the range of 20–60 kHz with an additional 1% of t_{CK} (AVG) as a long-term jitter component; however, the spread spectrum may not use a clock rate below t_{CK} (AVG) MIN.
12. The clock's t_{CH} (AVG) and t_{CL} (AVG) are the average half clock period over any 200 consecutive clocks and is the smallest clock half period allowed, with the exception of a deviation due to clock jitter. Input clock jitter is allowed provided it does not exceed values specified and must be of a random Gaussian distribution in nature.
13. The period jitter (t_{JITper}) is the maximum deviation in the clock period from the average or nominal clock. It is allowed in either the positive or negative direction.
14. t_{CH} (ABS) is the absolute instantaneous clock high pulse width as measured from one rising edge to the following falling edge.
15. t_{CL} (ABS) is the absolute instantaneous clock low pulse width as measured from one falling edge to the following rising edge.
16. The cycle-to-cycle jitter t_{JITcc} is the amount the clock period can deviate from one cycle to the next. It is important to keep cycle-to-cycle jitter at a minimum during the DLL locking time.
17. The cumulative jitter error $t_{ERRnper}$, where n is the number of clocks between 2 and 50, is the amount of clock time allowed to accumulate consecutively away from the average clock over n number of clock cycles.
18. t_{DS} (base) and t_{DH} (base) values are for a single-ended 1 V/ns slew rate DQs and 2 V/ns slew rate differential DQS, DQS#.
19. These parameters are measured from a data signal (DM, DQ0, DQ1, and so forth) transition edge to its respective data strobe signal (DQS, DQS#) crossing.

20. The setup and hold times are listed converting the base specification values (to which derating tables apply) to V_{REF} when the slew rate is 1 V/ns. These values, with a slew rate of 1 V/ns, are for reference only.
21. When the device is operated with input clock jitter, this parameter needs to be derated by the actual tJITper (larger of tJITper (MIN) or tJITper (MAX)) of the input clock (output deratings are relative to the SDRAM input clock).
22. Single-ended signal parameter.
23. The DRAM output timing is aligned to the nominal or average clock. Most output parameters must be derated by the actual jitter error when input clock jitter is present, even when within specification. This results in each parameter becoming larger. The following parameters are required to be derated by subtracting tERR10per (MAX): tDQSCK (MIN), tLZDQS (MIN), tLZDQ (MIN), and tAON (MIN). The following parameters are required to be derated by subtracting tERR10per (MIN): tDQSCK (MAX), tHZ (MAX), tLZDQS (MAX), tLZDQ MAX, and tAON (MAX). The parameter tRPRE (MIN) is derated by subtracting tJITper (MAX), while tRPRE (MAX) is derated by subtracting tJITper (MIN).
24. The maximum preamble is bound by tLZDQS (MAX).
25. These parameters are measured from a data strobe signal (DQS, DQS#) crossing to its respective clock signal (CK, CK#) crossing. The specification values are not affected by the amount of clock jitter applied, as these are relative to the clock signal crossing. These parameters should be met whether clock jitter is present.
26. The tDQSCK (DLL_DIS) parameter begins CL + AL - 1 cycles after the READ command.
27. The maximum postamble is bound by tHZDQS (MAX).
28. Commands requiring a locked DLL are: READ (and RDAP) and synchronous ODT commands. In addition, after any change of latency tXPDLL , timing must be met.
29. tIS (base) and tIH (base) values are for a single-ended 1 V/ns control/command/address slew rate and 2 V/ns CK, CK# differential slew rate.
30. These parameters are measured from a command/address signal transition edge to its respective clock (CK, CK#) signal crossing. The specification values are not affected by the amount of clock jitter applied as the setup and hold times are relative to the clock signal crossing that latches the command/address. These parameters should be met whether clock jitter is present.
31. For these parameters, the DDR3 SDRAM device supports tnPARAM (nCK) = $RU(^tPARAM$ [ns]/ tCK [AVG] [ns]), assuming all input clock jitter specifications are satisfied. For example, the device will support tnRP (nCK) = $RU(^tRP$ / tCK [AVG]) if all input clock jitter specifications are met. This means that for DDR3-800 6-6-6, of which tRP = 5ns, the device will support tnRP = $RU(^tRP$ / tCK [AVG]) = 6 as long as the input clock jitter specifications are met. That is, the PRECHARGE command at T0 and the ACTIVATE command at T0 + 6 are valid even if six clocks are less than 15ns due to input clock jitter.
32. During READs and WRITEs with auto precharge, the DDR3 SDRAM will hold off the internal PRECHARGE command until tRAS (MIN) has been satisfied.
33. When operating in DLL disable mode, the greater of 4CK or 15ns is satisfied for tWR .
34. The start of the write recovery time is defined as follows:
 - For BL8 (fixed by MRS or OTF): Rising clock edge four clock cycles after WL
 - For BC4 (OTF): Rising clock edge four clock cycles after WL
 - For BC4 (fixed by MRS): Rising clock edge two clock cycles after WL
35. RESET# should be LOW as soon as power starts to ramp to ensure the outputs are in High-Z. Until RESET# is LOW, the outputs are at risk of driving and could result in excessive current, depending on bus activity.
36. The refresh period is 64ms when T_C is less than or equal to 85°C. This equates to an average refresh rate of 7.8125μs. However, nine REFRESH commands should be asserted at least once every 70.3μs. When T_C is greater than 85°C, the refresh period is 32ms.
37. Although CKE is allowed to be registered LOW after a REFRESH command when

$t_{REFPDEN}$ (MIN) is satisfied, there are cases where additional time such as t_{XPDLL} (MIN) is required.

38. ODT turn-on time MIN is when the device leaves High-Z and ODT resistance begins to turn on. ODT turn-on time maximum is when the ODT resistance is fully on. The ODT reference load is shown in Figure 22 (page 58). Designs that were created prior to JEDEC tightening the maximum limit from 9ns to 8.5ns will be allowed to have a 9ns maximum.
39. Half-clock output parameters must be derated by the actual $t_{ERR10per}$ and t_{JITdty} when input clock jitter is present. This results in each parameter becoming larger. The parameters t_{ADC} (MIN) and t_{AOF} (MIN) are each required to be derated by subtracting both $t_{ERR10per}$ (MAX) and t_{JITdty} (MAX). The parameters t_{ADC} (MAX) and t_{AOF} (MAX) are required to be derated by subtracting both $t_{ERR10per}$ (MAX) and t_{JITdty} (MAX).
40. ODT turn-off time minimum is when the device starts to turn off ODT resistance. ODT turn-off time maximum is when the DRAM buffer is in High-Z. The ODT reference load is shown in Figure 22 (page 58). This output load is used for ODT timings (see Figure 29 (page 70)).
41. Pulse width of a input signal is defined as the width between the first crossing of $V_{REF(DC)}$ and the consecutive crossing of $V_{REF(DC)}$.
42. Should the clock rate be larger than t_{RFC} (MIN), an AUTO REFRESH command should have at least one NOP command between it and another AUTO REFRESH command. Additionally, if the clock rate is slower than 40ns (25 MHz), all REFRESH commands should be followed by a PRECHARGE ALL command.
43. DRAM devices should be evenly addressed when being accessed. Disproportionate accesses to a particular row address may result in a reduction of REFRESH characteristics or product lifetime.
44. When two $V_{IH(AC)}$ values (and two corresponding $V_{IL(AC)}$ values) are listed for a specific speed bin, the user may choose either value for the input AC level. Whichever value is used, the associated setup time for that AC level must also be used. Additionally, one $V_{IH(AC)}$ value may be used for address/command inputs and the other $V_{IH(AC)}$ value may be used for data inputs.

For example, for DDR3-800, two input AC levels are defined: $V_{IH(AC175),min}$ and $V_{IH(AC150),min}$ (corresponding $V_{IL(AC175),min}$ and $V_{IL(AC150),min}$). For DDR3-800, the address/command inputs must use either $V_{IH(AC175),min}$ with $t_{IS(AC175)}$ of 200ps or $V_{IH(AC150),min}$ with $t_{IS(AC150)}$ of 350ps; independently, the data inputs must use either $V_{IH(AC175),min}$ with $t_{DS(AC175)}$ of 75ps or $V_{IH(AC150),min}$ with $t_{DS(AC150)}$ of 125ps.

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
Clock Timing					
Clock period average: DLL disable mode	$T_C = 0^\circ\text{C to } 85^\circ\text{C}$	$t_{\text{CK}}(\text{DLL_DIS})$	8	7800	ns
	$T_C = >85^\circ\text{C to } 95^\circ\text{C}$		8	3900	ns
Clock period average: DLL enable mode		$t_{\text{CK}}(\text{AVG})$	See Speed Bin Tables (page 73) for t_{CK} range allowed		ns
High pulse width average		$t_{\text{CH}}(\text{AVG})$	0.47	0.53	CK
Low pulse width average		$t_{\text{CL}}(\text{AVG})$	0.47	0.53	CK
Clock period jitter	DLL locked	t_{JITper}	-60	60	ps
	DLL locking	$t_{\text{JITper,lck}}$	-50	50	ps
Clock absolute period		$t_{\text{CK}}(\text{ABS})$	$\text{MIN} = t_{\text{CK}}(\text{AVG}) \text{ MIN} + t_{\text{JITper}} \text{ MIN};$ $\text{MAX} = t_{\text{CK}}(\text{AVG}) \text{ MAX} + t_{\text{JITper}} \text{ MAX}$		ps
Clock absolute high pulse width		$t_{\text{CH}}(\text{ABS})$	0.43	-	$t_{\text{CK}}(\text{AVG})$
Clock absolute low pulse width		$t_{\text{CL}}(\text{ABS})$	0.43	-	$t_{\text{CK}}(\text{AVG})$
Cycle-to-cycle jitter	DLL locked	t_{JITcc}	120		ps
	DLL locking	$t_{\text{JITcc,lck}}$	100		ps
Cumulative error across	2 cycles	t_{ERR2per}	-88	88	ps
	3 cycles	t_{ERR3per}	-105	105	ps
	4 cycles	t_{ERR4per}	-117	117	ps
	5 cycles	t_{ERR5per}	-126	126	ps
	6 cycles	t_{ERR6per}	-133	133	ps
	7 cycles	t_{ERR7per}	-139	139	ps
	8 cycles	t_{ERR8per}	-145	145	ps
	9 cycles	t_{ERR9per}	-150	150	ps
	10 cycles	t_{ERR10per}	-154	154	ps
	11 cycles	t_{ERR11per}	-158	158	ps
	12 cycles	t_{ERR12per}	-161	161	ps
	$n = 13, 14 \dots 49, 50$ cycles	t_{ERRnper}	$t_{\text{ERRnper}} \text{ MIN} = (1 + 0.68 \ln[n]) \times t_{\text{JITper}} \text{ MIN}$ $t_{\text{ERRnper}} \text{ MAX} = (1 + 0.68 \ln[n]) \times t_{\text{JITper}} \text{ MAX}$		ps
DQ Input Timing					
Data setup time to DQS, DQS#	Base (specification) @ 2 V/ns	$t_{\text{DS}}(\text{AC135})$	68	-	ps
	V_{REF} @ 2 V/ns		135	-	ps
					18, 19
					19, 20

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
Data hold time from DQS, DQS#	Base (specification) @ 2 V/ns	t_{DH} (DC100)	70	–	ps
	V_{REF} @ 2 V/ns		120	–	ps
Minimum data pulse width	t_{DIPW}	320	–	ps	41
DQ Output Timing					
DQS, DQS# to DQ skew, per access	t_{DQSQ}	–	85	ps	
DQ output hold time from DQS, DQS#	t_{QH}	0.38	–	t_{CK} (AVG)	21
DQ Low-Z time from CK, CK#	t_{LZDQ}	–390	195	ps	22, 23
DQ High-Z time from CK, CK#	t_{HZDQ}	–	195	ps	22, 23
DQ Strobe Input Timing					
DQS, DQS# rising to CK, CK# rising	t_{DQSS}	–0.27	0.27	CK	25
DQS, DQS# differential input low pulse width	t_{DQL}	0.45	0.55	CK	
DQS, DQS# differential input high pulse width	t_{DQH}	0.45	0.55	CK	
DQS, DQS# falling setup to CK, CK# rising	t_{DSS}	0.18	–	CK	25
DQS, DQS# falling hold from CK, CK# rising	t_{DSH}	0.18	–	CK	25
DQS, DQS# differential WRITE preamble	t_{WPRE}	0.9	–	CK	
DQS, DQS# differential WRITE postamble	t_{WPST}	0.3	–	CK	
DQ Strobe Output Timing					
DQS, DQS# rising to/from rising CK, CK#	t_{DQCK}	–195	195	ps	23
DQS, DQS# rising to/from rising CK, CK# when DLL is disabled	t_{DQCK} (DLL_DIS)	1	10	ns	26
DQS, DQS# differential output high time	t_{QSH}	0.40	–	CK	21
DQS, DQS# differential output low time	t_{QSL}	0.40	–	CK	21
DQS, DQS# Low-Z time (RL - 1)	t_{LZDQS}	–390	195	ps	22, 23
DQS, DQS# High-Z time (RL + BL/2)	t_{HZDQS}	–	195	ps	22, 23
DQS, DQS# differential READ preamble	t_{RPRE}	0.9	Note 24	CK	23, 24
DQS, DQS# differential READ postamble	t_{RPST}	0.3	Note 27	CK	23, 27
Command and Address Timing					
DLL locking time	t_{DLLK}	512	–	CK	28
CTRL, CMD, ADDR setup to CK, CK#	Base (specification)	t_{IS} (AC135)	65	–	ps
	V_{REF} @ 1 V/ns		200	–	ps
CTRL, CMD, ADDR setup to CK, CK#	Base (specification)	t_{IS} (AC125)	150	–	ps
	V_{REF} @ 1 V/ns		275	–	ps
CTRL, CMD, ADDR hold from CK, CK#	Base (specification)	t_{IH} (DC100)	100	–	ps
	V_{REF} @ 1 V/ns		200	–	ps
Minimum CTRL, CMD, ADDR pulse width	t_{IPW}	535	–	ps	41

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
ACTIVATE to internal READ or WRITE delay	t_{RCD}	See Speed Bin Tables (page 73) for t_{RCD}		ns	31
PRECHARGE command period	t_{RP}	See Speed Bin Tables (page 73) for t_{RP}		ns	31
ACTIVATE-to-PRECHARGE command period	t_{RAS}	See Speed Bin Tables (page 73) for t_{RAS}		ns	31, 32
ACTIVATE-to-ACTIVATE command period	t_{RC}	See Speed Bin Tables (page 73) for t_{RC}		ns	31, 43
ACTIVATE-to-ACTIVATE minimum command period	1KB page size	t_{RRD}	MIN = greater of 4CK or 5ns		CK
	2KB page size		MIN = greater of 4CK or 6ns		CK
Four ACTIVATE windows	1KB page size	t_{FAW}	27	–	ns
	2KB page size		35	–	ns
Write recovery time	t_{WR}	MIN = 15ns; MAX = N/A		ns	31, 32, 33
Delay from start of internal WRITE transaction to internal READ command	t_{WTR}	MIN = greater of 4CK or 7.5ns; MAX = N/A		CK	31, 34
READ-to-PRECHARGE time	t_{RTP}	MIN = greater of 4CK or 7.5ns; MAX = N/A		CK	31, 32
CAS#-to-CAS# command delay	t_{CCD}	MIN = 4CK; MAX = N/A		CK	
Auto precharge write recovery + precharge time	t_{DAL}	MIN = WR + t_{RP}/t_{CK} (AVG); MAX = N/A		CK	
MODE REGISTER SET command cycle time	t_{MRD}	MIN = 4CK; MAX = N/A		CK	
MODE REGISTER SET command update delay	t_{MOD}	MIN = greater of 12CK or 15ns; MAX = N/A		CK	
MULTIPURPOSE REGISTER READ burst end to mode register set for multipurpose register exit	t_{MPRR}	MIN = 1CK; MAX = N/A		CK	
Calibration Timing					
ZQCL command: Long calibration time	POWER-UP and RESET operation	t_{ZQinit}	MIN = N/A MAX = MAX(512nCK, 640ns)		CK
	Normal operation	t_{ZQoper}	MIN = N/A MAX = MAX(256nCK, 320ns)		CK
ZQCS command: Short calibration time		MIN = N/A MAX = MAX(64nCK, 80ns)		CK	
Initialization and Reset Timing					
Exit reset from CKE HIGH to a valid command	t_{XPR}	MIN = greater of 5CK or $t_{RFC} + 10$ ns; MAX = N/A		CK	
Begin power supply ramp to power supplies stable	t_{VDDPR}	MIN = N/A; MAX = 200		ms	
RESET# LOW to power supplies stable	t_{RPS}	MIN = 0; MAX = 200		ms	
RESET# LOW to I/O and R_{TT} High-Z	t_{IOZ}	MIN = N/A; MAX = 20		ns	35
Refresh Timing					

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
REFRESH-to-ACTIVATE or REFRESH command period	^t RFC – 1Gb	MIN = 110; MAX = 70,200		ns	
	^t RFC – 2Gb	MIN = 160; MAX = 70,200		ns	
	^t RFC – 4Gb	MIN = 260; MAX = 70,200		ns	
	^t RFC – 8Gb	MIN = 350; MAX = 70,200		ns	
Maximum refresh period	$T_C \leq 85^\circ\text{C}$	–	64 (1X)	ms	36
	$T_C > 85^\circ\text{C}$		32 (2X)	ms	36
Maximum average periodic refresh	$T_C \leq 85^\circ\text{C}$	^t REFI	7.8 (64ms/8192)	μs	36
	$T_C > 85^\circ\text{C}$		3.9 (32ms/8192)	μs	36
Self Refresh Timing					
Exit self refresh to commands not requiring a locked DLL		^t XS	MIN = greater of 5CK or ${}^t\text{RFC} + 10\text{ns}$; MAX = N/A	CK	
Exit self refresh to commands requiring a locked DLL		^t XSDL	MIN = ${}^t\text{DLLK}$ (MIN); MAX = N/A	CK	28
Minimum CKE low pulse width for self refresh entry to self refresh exit timing		^t CKESR	MIN = ${}^t\text{CKE}$ (MIN) + CK; MAX = N/A	CK	
Valid clocks after self refresh entry or power-down entry		^t CKSRE	MIN = greater of 5CK or 10ns; MAX = N/A	CK	
Valid clocks before self refresh exit, power-down exit, or reset exit		^t CKSRX	MIN = greater of 5CK or 10ns; MAX = N/A	CK	
Power-Down Timing					
CKE MIN pulse width		^t CKE (MIN)	Greater of 3CK or 5ns	CK	
Command pass disable delay		^t CPDED	MIN = 2; MAX = N/A	CK	
Power-down entry to power-down exit timing		^t PD	MIN = ${}^t\text{CKE}$ (MIN); MAX = $9 \times {}^t\text{REFI}$	CK	
Begin power-down period prior to CKE registered HIGH		^t ANPD	WL - 1CK	CK	
Power-down entry period: ODT either synchronous or asynchronous		PDE	Greater of ${}^t\text{ANPD}$ or ${}^t\text{RFC}$ - REFRESH command to CKE LOW time	CK	
Power-down exit period: ODT either synchronous or asynchronous		PDX	${}^t\text{ANPD} + {}^t\text{XPDL}$	CK	
Power-Down Entry Minimum Timing					
ACTIVATE command to power-down entry		^t ACTPDEN	MIN = 2	CK	
PRECHARGE/PRECHARGE ALL command to power-down entry		^t PRPDEN	MIN = 2	CK	
REFRESH command to power-down entry		^t REFPDEN	MIN = 2	CK	37
MRS command to power-down entry		^t MRSPDEN	MIN = ${}^t\text{MOD}$ (MIN)	CK	

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
READ/READ with auto precharge command to power-down entry	t_{RDPDEN}	MIN = RL + 4 + 1		CK	
WRITE command to power-down entry	BL8 (OTF, MRS) BC4OTF	t_{WRPDEN}	MIN = WL + 4 + t_{WR}/t_{CK} (AVG)		CK
	BC4MRS	t_{WRPDEN}	MIN = WL + 2 + t_{WR}/t_{CK} (AVG)		CK
WRITE with auto precharge command to power-down entry	BL8 (OTF, MRS) BC4OTF	$t_{WRAPDEN}$	MIN = WL + 4 + WR + 1		CK
	BC4MRS	$t_{WRAPDEN}$	MIN = WL + 2 + WR + 1		CK
Power-Down Exit Timing					
DLL on, any valid command, or DLL off to commands not requiring locked DLL	t_{XP}	MIN = greater of 3CK or 6ns; MAX = N/A		CK	
Precharge power-down with DLL off to commands requiring a locked DLL	$t_{XP DLL}$	MIN = greater of 10CK or 24ns; MAX = N/A		CK	28
ODT Timing					
R_{TT} synchronous turn-on delay	ODTL on	CWL + AL - 2CK		CK	38
R_{TT} synchronous turn-off delay	ODTL off	CWL + AL - 2CK		CK	40
R_{TT} turn-on from ODTL on reference	t_{AON}	-195	195	ps	23, 38
R_{TT} turn-off from ODTL off reference	t_{AOF}	0.3	0.7	CK	39, 40
Asynchronous R_{TT} turn-on delay (power-down with DLL off)	t_{AONPD}	MIN = 2; MAX = 8.5		ns	38
Asynchronous R_{TT} turn-off delay (power-down with DLL off)	t_{AOFPD}	MIN = 2; MAX = 8.5		ns	40
ODT HIGH time with WRITE command and BL8	ODTH8	MIN = 6; MAX = N/A		CK	
ODT HIGH time without WRITE command or with WRITE command and BC4	ODTH4	MIN = 4; MAX = N/A		CK	
Dynamic ODT Timing					
$R_{TT,nom}$ -to- $R_{TT(WR)}$ change skew	ODTLcnw	WL - 2CK		CK	
$R_{TT(WR)}$ -to- $R_{TT,nom}$ change skew - BC4	ODTLcwn4	4CK + ODTLoff		CK	
$R_{TT(WR)}$ -to- $R_{TT,nom}$ change skew - BL8	ODTLcwn8	6CK + ODTLoff		CK	
R_{TT} dynamic change skew	t_{ADC}	0.3	0.7	CK	39
Write Leveling Timing					
First DQS, DQS# rising edge	t_{WLMRD}	40	-	CK	
DQS, DQS# delay	$t_{WLDQSEN}$	25	-	CK	
Write leveling setup from rising CK, CK# crossing to rising DQS, DQS# crossing	t_{WLS}	140	-	ps	
Write leveling hold from rising DQS, DQS# crossing to rising CK, CK# crossing	t_{WLH}	140	-	ps	
Write leveling output delay	t_{WLO}	0	7.5	ns	

Table 58: Electrical Characteristics and AC Operating Conditions for Speed Extensions (Continued)

Notes 1–8 apply to the entire table

Parameter	Symbol	DDR3L-1866		Unit	Notes
		Min	Max		
Write leveling output error	t_{WLOE}	0	2	ns	

Notes:

1. AC timing parameters are valid from specified T_C MIN to T_C MAX values.
2. All voltages are referenced to V_{SS} .
3. Output timings are only valid for R_{ON34} output buffer selection.
4. The unit t_{CK} (AVG) represents the actual t_{CK} (AVG) of the input clock under operation. The unit CK represents one clock cycle of the input clock, counting the actual clock edges.
5. AC timing and I_{DD} tests may use a V_{IL} -to- V_{IH} swing of up to 900mV in the test environment, but input timing is still referenced to V_{REF} (except t_{IS} , t_{IH} , t_{DS} , and t_{DH} use the AC/DC trip points and CK, CK# and DQS, DQS# use their crossing points). The minimum slew rate for the input signals used to test the device is 1 V/ns for single-ended inputs (DQs are at 2V/ns for DDR3-1866 and DDR3-2133) and 2 V/ns for differential inputs in the range between $V_{IL(AC)}$ and $V_{IH(AC)}$.
6. All timings that use time-based values (ns, μ s, ms) should use t_{CK} (AVG) to determine the correct number of clocks (Table 58 (page 89) uses CK or t_{CK} [AVG] interchangeably). In the case of noninteger results, all minimum limits are to be rounded up to the nearest whole integer, and all maximum limits are to be rounded down to the nearest whole integer.
7. Strobe or DQSdiff refers to the DQS and DQS# differential crossing point when DQS is the rising edge. Clock or CK refers to the CK and CK# differential crossing point when CK is the rising edge.
8. This output load is used for all AC timing (except ODT reference timing) and slew rates. The actual test load may be different. The output signal voltage reference point is $V_{DDQ}/2$ for single-ended signals and the crossing point for differential signals (see Figure 28 (page 69)).
9. When operating in DLL disable mode, Micron does not warrant compliance with normal mode timings or functionality.
10. The clock's t_{CK} (AVG) is the average clock over any 200 consecutive clocks and t_{CK} (AVG) MIN is the smallest clock rate allowed, with the exception of a deviation due to clock jitter. Input clock jitter is allowed provided it does not exceed values specified and must be of a random Gaussian distribution in nature.
11. Spread spectrum is not included in the jitter specification values. However, the input clock can accommodate spread-spectrum at a sweep rate in the range of 20–60 kHz with an additional 1% of t_{CK} (AVG) as a long-term jitter component; however, the spread spectrum may not use a clock rate below t_{CK} (AVG) MIN.
12. The clock's t_{CH} (AVG) and t_{CL} (AVG) are the average half clock period over any 200 consecutive clocks and is the smallest clock half period allowed, with the exception of a deviation due to clock jitter. Input clock jitter is allowed provided it does not exceed values specified and must be of a random Gaussian distribution in nature.
13. The period jitter ('JITper) is the maximum deviation in the clock period from the average or nominal clock. It is allowed in either the positive or negative direction.
14. t_{CH} (ABS) is the absolute instantaneous clock high pulse width as measured from one rising edge to the following falling edge.
15. t_{CL} (ABS) is the absolute instantaneous clock low pulse width as measured from one falling edge to the following rising edge.

16. The cycle-to-cycle jitter t_{JITcc} is the amount the clock period can deviate from one cycle to the next. It is important to keep cycle-to-cycle jitter at a minimum during the DLL locking time.
17. The cumulative jitter error $t_{ERRnper}$, where n is the number of clocks between 2 and 50, is the amount of clock time allowed to accumulate consecutively away from the average clock over n number of clock cycles.
18. t_{DS} (base) and t_{DH} (base) values are for a single-ended 1 V/ns slew rate DQs (DQs are at 2V/ns for DDR3-1866 and DDR3-2133) and 2 V/ns slew rate differential DQS, DQS#.
19. These parameters are measured from a data signal (DM, DQ0, DQ1, and so forth) transition edge to its respective data strobe signal (DQS, DQS#) crossing.
20. The setup and hold times are listed converting the base specification values (to which derating tables apply) to V_{REF} when the slew rate is 1 V/ns (DQs are at 2V/ns for DDR3-1866 and DDR3-2133). These values, with a slew rate of 1 V/ns (DQs are at 2V/ns for DDR3-1866 and DDR3-2133), are for reference only.
21. When the device is operated with input clock jitter, this parameter needs to be derated by the actual t_{JITper} (larger of t_{JITper} (MIN) or t_{JITper} (MAX)) of the input clock (output deratings are relative to the SDRAM input clock).
22. Single-ended signal parameter.
23. The DRAM output timing is aligned to the nominal or average clock. Most output parameters must be derated by the actual jitter error when input clock jitter is present, even when within specification. This results in each parameter becoming larger. The following parameters are required to be derated by subtracting $t_{ERR10per}$ (MAX): t_{DQSCK} (MIN), t_{LZDQS} (MIN), t_{LZDQ} (MIN), and t_{AON} (MIN). The following parameters are required to be derated by subtracting $t_{ERR10per}$ (MIN): t_{DQSCK} (MAX), t_{HZ} (MAX), t_{LZDQS} (MAX), t_{LZDQ} (MAX), and t_{AON} (MAX). The parameter t_{RPRE} (MIN) is derated by subtracting t_{JITper} (MAX), while t_{RPRE} (MAX) is derated by subtracting t_{JITper} (MIN).
24. The maximum preamble is bound by t_{LZDQS} (MAX).
25. These parameters are measured from a data strobe signal (DQS, DQS#) crossing to its respective clock signal (CK, CK#) crossing. The specification values are not affected by the amount of clock jitter applied, as these are relative to the clock signal crossing. These parameters should be met whether clock jitter is present.
26. The t_{DQSCK} (DLL_DIS) parameter begins CL + AL - 1 cycles after the READ command.
27. The maximum postamble is bound by t_{HZDQS} (MAX).
28. Commands requiring a locked DLL are: READ (and RDAP) and synchronous ODT commands. In addition, after any change of latency $t_{XP DLL}$, timing must be met.
29. t_{IS} (base) and t_{IH} (base) values are for a single-ended 1 V/ns control/command/address slew rate and 2 V/ns CK, CK# differential slew rate.
30. These parameters are measured from a command/address signal transition edge to its respective clock (CK, CK#) signal crossing. The specification values are not affected by the amount of clock jitter applied as the setup and hold times are relative to the clock signal crossing that latches the command/address. These parameters should be met whether clock jitter is present.
31. For these parameters, the DDR3 SDRAM device supports t_{nPARAM} (n CK) = $RU(t_{PARAM} [ns]/t_{CK[AVG]} [ns])$, assuming all input clock jitter specifications are satisfied. For example, the device will support t_{nRP} (n CK) = $RU(t_{RP}/t_{CK[AVG]})$ if all input clock jitter specifications are met. This means that for DDR3-800 6-6-6, of which $t_{RP} = 5$ ns, the device will support $t_{nRP} = RU(t_{RP}/t_{CK[AVG]}) = 6$ as long as the input clock jitter specifications are met. That is, the PRECHARGE command at T0 and the ACTIVATE command at T0 + 6 are valid even if six clocks are less than 15ns due to input clock jitter.
32. During READs and WRITEs with auto precharge, the DDR3 SDRAM will hold off the internal PRECHARGE command until t_{RAS} (MIN) has been satisfied.
33. When operating in DLL disable mode, the greater of 4CK or 15ns is satisfied for t_{WR} .
34. The start of the write recovery time is defined as follows:

- For BL8 (fixed by MRS or OTF): Rising clock edge four clock cycles after WL
- For BC4 (OTF): Rising clock edge four clock cycles after WL
- For BC4 (fixed by MRS): Rising clock edge two clock cycles after WL

35. RESET# should be LOW as soon as power starts to ramp to ensure the outputs are in High-Z. Until RESET# is LOW, the outputs are at risk of driving and could result in excessive current, depending on bus activity.

36. The refresh period is 64ms when T_C is less than or equal to 85°C. This equates to an average refresh rate of 7.8125 μ s. However, nine REFRESH commands should be asserted at least once every 70.3 μ s. When T_C is greater than 85°C, the refresh period is 32ms.

37. Although CKE is allowed to be registered LOW after a REFRESH command when $t_{REFPDEN}$ (MIN) is satisfied, there are cases where additional time such as $t_{XP DLL}$ (MIN) is required.

38. ODT turn-on time MIN is when the device leaves High-Z and ODT resistance begins to turn on. ODT turn-on time maximum is when the ODT resistance is fully on. The ODT reference load is shown in Figure 22 (page 58). Designs that were created prior to JEDEC tightening the maximum limit from 9ns to 8.5ns will be allowed to have a 9ns maximum.

39. Half-clock output parameters must be derated by the actual $t_{ERR10per}$ and t_{JITdty} when input clock jitter is present. This results in each parameter becoming larger. The parameters t_{ADC} (MIN) and t_{AOF} (MIN) are each required to be derated by subtracting both $t_{ERR10per}$ (MAX) and t_{JITdty} (MAX). The parameters t_{ADC} (MAX) and t_{AOF} (MAX) are required to be derated by subtracting both $t_{ERR10per}$ (MAX) and t_{JITdty} (MAX).

40. ODT turn-off time minimum is when the device starts to turn off ODT resistance. ODT turn-off time maximum is when the DRAM buffer is in High-Z. The ODT reference load is shown in Figure 22 (page 58). This output load is used for ODT timings (see Figure 29 (page 70)).

41. Pulse width of a input signal is defined as the width between the first crossing of $V_{REF(DC)}$ and the consecutive crossing of $V_{REF(DC)}$.

42. Should the clock rate be larger than t_{RFC} (MIN), an AUTO REFRESH command should have at least one NOP command between it and another AUTO REFRESH command. Additionally, if the clock rate is slower than 40ns (25 MHz), all REFRESH commands should be followed by a PRECHARGE ALL command.

43. DRAM devices should be evenly addressed when being accessed. Disproportionate accesses to a particular row address may result in a reduction of REFRESH characteristics or product lifetime.

44. When two $V_{IH(AC)}$ values (and two corresponding $V_{IL(AC)}$ values) are listed for a specific speed bin, the user may choose either value for the input AC level. Whichever value is used, the associated setup time for that AC level must also be used. Additionally, one $V_{IH(AC)}$ value may be used for address/command inputs and the other $V_{IH(AC)}$ value may be used for data inputs.

For example, for DDR3-800, two input AC levels are defined: $V_{IH(AC175),min}$ and $V_{IH(AC150),min}$ (corresponding $V_{IL(AC175),min}$ and $V_{IL(AC150),min}$). For DDR3-800, the address/command inputs must use either $V_{IH(AC175),min}$ with $t_{IS(AC175)}$ of 200ps or $V_{IH(AC150),min}$ with $t_{IS(AC150)}$ of 350ps; independently, the data inputs must use either $V_{IH(AC175),min}$ with $t_{DS(AC175)}$ of 75ps or $V_{IH(AC150),min}$ with $t_{DS(AC150)}$ of 125ps.

Command and Address Setup, Hold, and Derating

The total t_{IS} (setup time) and t_{IH} (hold time) required is calculated by adding the data sheet t_{IS} (base) and t_{IH} (base) values (see Table 59; values come from Table 57 (page 79)) to the Δt_{IS} and Δt_{IH} derating values (see Table 60 (page 98), Table 61 (page 98) or Table 62 (page 99)) respectively. Example: t_{IS} (total setup time) = t_{IS} (base) + Δt_{IS} . For a valid transition, the input signal has to remain above/below $V_{IH(AC)}/V_{IL(AC)}$ for some time t_{VAC} (see Table 63 (page 99)).

Although the total setup time for slow slew rates might be negative (for example, a valid input signal will not have reached $V_{IH(AC)}/V_{IL(AC)}$ at the time of the rising clock transition), a valid input signal is still required to complete the transition and to reach $V_{IH(AC)}/V_{IL(AC)}$ (see Figure 13 (page 48) for input signal requirements). For slew rates that fall between the values listed in Table 60 (page 98) and Table 62 (page 99), the derating values may be obtained by linear interpolation.

Setup (t_{IS}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{REF(DC)}$ and the first crossing of $V_{IH(AC)min}$. Setup (t_{IS}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{REF(DC)}$ and the first crossing of $V_{IL(AC)max}$. If the actual signal is always earlier than the nominal slew rate line between the shaded $V_{REF(DC)}$ -to-AC region, use the nominal slew rate for derating value (see Figure 32 (page 100)). If the actual signal is later than the nominal slew rate line anywhere between the shaded $V_{REF(DC)}$ -to-AC region, the slew rate of a tangent line to the actual signal from the AC level to the DC level is used for derating value (see Figure 34 (page 102)).

Hold (t_{IH}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{IL(DC)max}$ and the first crossing of $V_{REF(DC)}$. Hold (t_{IH}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{IH(DC)min}$ and the first crossing of $V_{REF(DC)}$. If the actual signal is always later than the nominal slew rate line between the shaded DC-to- $V_{REF(DC)}$ region, use the nominal slew rate for derating value (see Figure 33 (page 101)). If the actual signal is earlier than the nominal slew rate line anywhere between the shaded DC-to- $V_{REF(DC)}$ region, the slew rate of a tangent line to the actual signal from the DC level to the $V_{REF(DC)}$ level is used for derating value (see Figure 35 (page 103)).

Table 59: DDR3L Command and Address Setup and Hold Values 1 V/ns Referenced – AC/DC-Based

Symbol	800	1066	1333	1600	1866	Unit	Reference
$t_{IS}(\text{base, AC160})$	215	140	80	60	–	ps	$V_{IH(\text{AC})}/V_{IL(\text{AC})}$
$t_{IS}(\text{base, AC135})$	365	290	205	185	65	ps	$V_{IH(\text{AC})}/V_{IL(\text{AC})}$
$t_{IS}(\text{base, AC125})$	–	–	–	–	150	ps	$V_{IH(\text{AC})}/V_{IL(\text{AC})}$
$t_{IH}(\text{base, DC90})$	285	210	150	130	110	ps	$V_{IH(\text{DC})}/V_{IL(\text{DC})}$

Table 60: DDR3L-800/1066/1333/1600 Derating Values t_{IS}/t_{IH} – AC160/DC90-Based

CMD/ADDR Slew Rate V/ns	$\Delta t_{IS}, \Delta t_{IH}$ Derating (ps) – AC/DC-Based															
	CK, CK# Differential Slew Rate															
	4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}
2.0	80	45	80	45	80	45	88	53	96	61	104	69	112	79	120	95
1.5	53	30	53	30	53	30	61	38	69	46	77	54	85	64	93	80
1.0	0	0	0	0	0	0	8	8	16	16	24	24	32	34	40	50
0.9	-1	-3	-1	-3	-1	-3	7	5	15	13	23	21	31	31	39	47
0.8	-3	-8	-3	-8	-3	-8	5	1	13	9	21	17	29	27	37	43
0.7	-5	-13	-5	-13	-5	-13	3	-5	11	3	19	11	27	21	35	37
0.6	-8	-20	-8	-20	-8	-20	0	-12	8	-4	16	4	24	14	32	30
0.5	-20	-30	-20	-30	-20	-30	-12	-22	-4	-14	4	-6	12	4	20	20
0.4	-40	-45	-40	-45	-40	-45	-32	-37	-24	-29	-16	-21	-8	-11	0	5

Table 61: DDR3L-800/1066/1333/1600 Derating Values for t_{IS}/t_{IH} – AC135/DC90-Based

CMD/ADDR Slew Rate V/ns	$\Delta t_{IS}, \Delta t_{IH}$ Derating (ps) – AC/DC-Based															
	CK, CK# Differential Slew Rate															
	4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}
2.0	68	45	68	45	68	45	76	53	84	61	92	69	100	79	108	95
1.5	45	30	45	30	45	30	53	38	61	46	69	54	77	64	85	80
1.0	0	0	0	0	0	0	8	8	16	16	24	24	32	34	40	50
0.9	2	-3	2	-3	2	-3	10	5	18	13	26	21	34	31	42	47
0.8	3	-8	3	-8	3	-8	11	1	19	9	27	17	35	27	43	43
0.7	6	-13	6	-13	6	-13	14	-5	22	3	30	11	38	21	46	37
0.6	9	-20	9	-20	9	-20	17	-12	25	-4	33	4	41	14	49	30
0.5	5	-30	5	-30	5	-30	13	-22	21	-14	29	-6	37	4	45	20
0.4	-3	-45	-3	-45	-3	-45	6	-37	14	-29	22	-21	30	-11	38	5

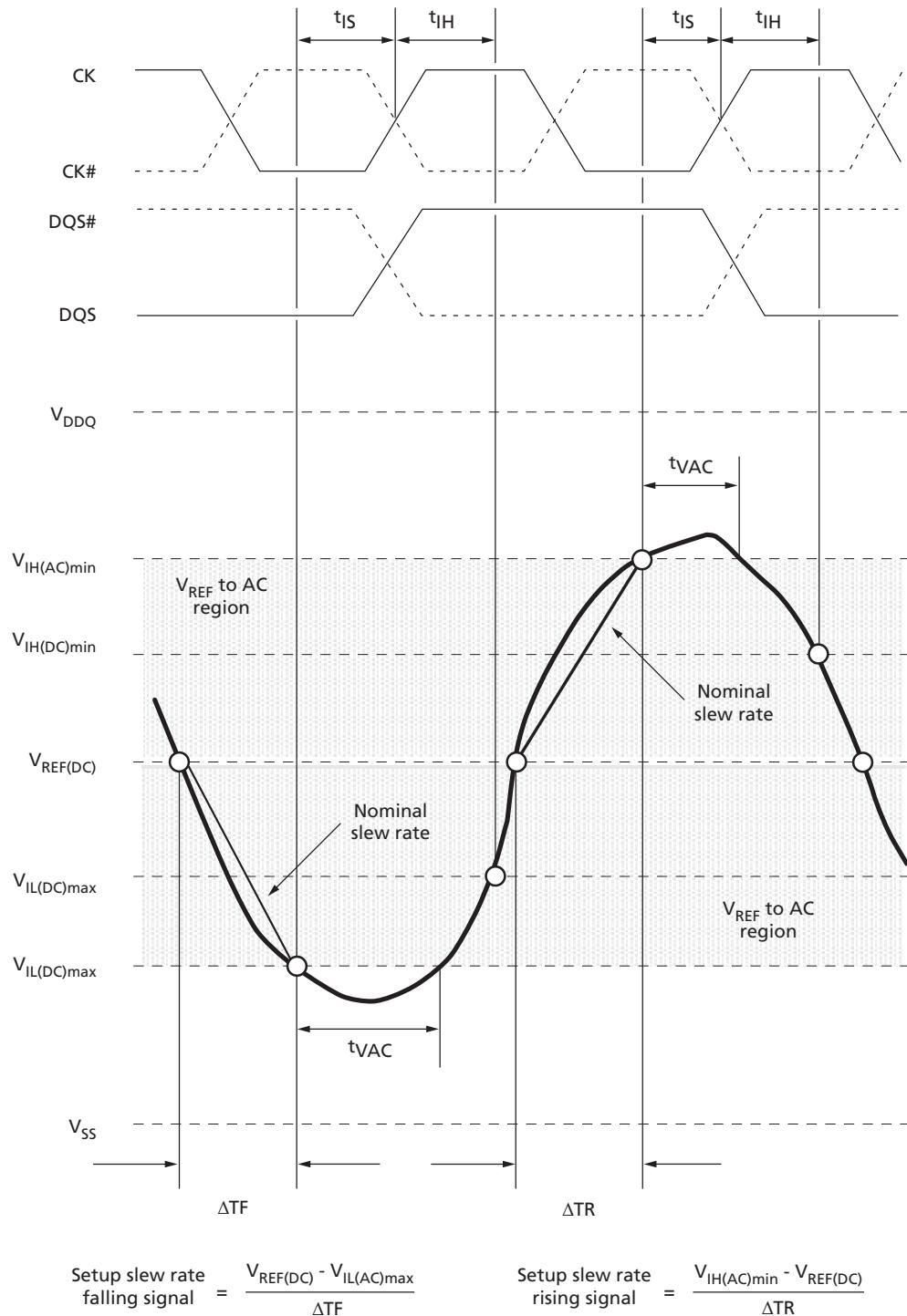
Table 62: DDR3L-1866 Derating Values for t_{IS}/t_{IH} – AC125/DC90-Based

CMD/ADDR Slew Rate V/ns	$\Delta t_{IS}, \Delta t_{IH}$ Derating (ps) – AC/DC-Based															
	CK, CK# Differential Slew Rate															
	4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}	Δt_{IS}	Δt_{IH}
2.0	63	45	63	45	63	45	71	53	79	61	87	69	95	79	103	95
1.5	42	30	42	30	42	30	50	38	58	46	66	54	74	64	82	80
1.0	0	0	0	0	0	0	8	8	16	16	24	24	32	34	40	50
0.9	3	-3	3	-3	3	-3	11	5	19	13	27	21	35	31	43	47
0.8	6	-8	6	-8	6	-8	14	1	22	9	30	17	38	27	46	43
0.7	10	-13	10	-13	10	-13	18	-5	26	3	34	11	42	21	50	37
0.6	16	-20	16	-20	16	-20	24	-12	32	-4	40	4	48	14	56	30
0.5	15	-30	15	-30	15	-30	23	-22	31	-14	39	-6	47	4	55	20
0.4	13	-45	13	-45	13	-45	21	-37	29	-29	37	-21	45	-11	53	5

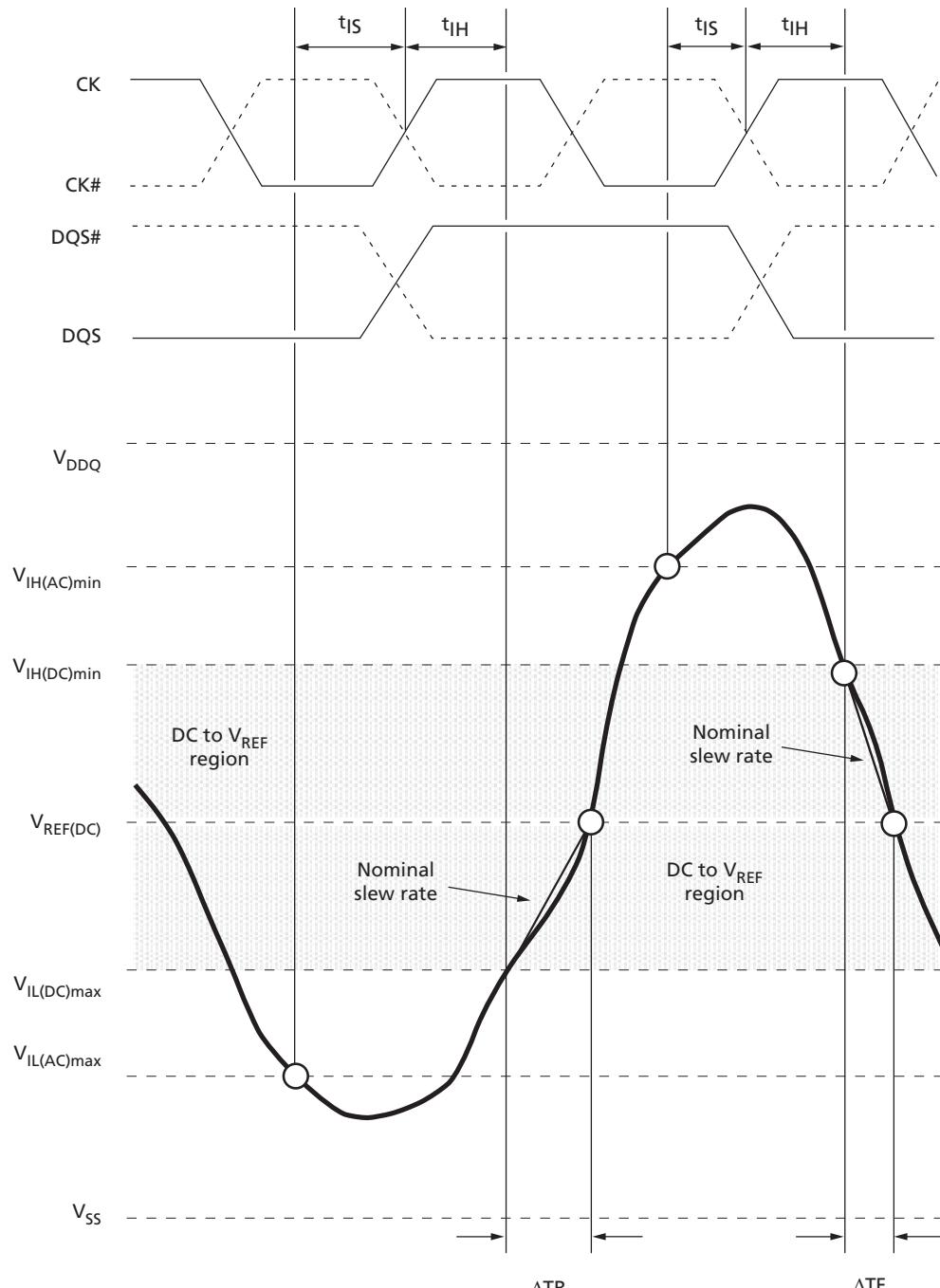
Table 63: DDR3L Minimum Required Time t_{VAC} Above $V_{IH(AC)}$ (Below $V_{IL(AC)}$) for Valid ADD/CMD Transition

Slew Rate (V/ns)	DDR3L-800/1066/1333/1600		DDR3L-1866	
	t_{VAC} at 160mV (ps)	t_{VAC} at 135mV (ps)	t_{VAC} at 135mV (ps)	t_{VAC} at 125mV (ps)
>2.0	70	209	200	205
2.0	53	198	200	205
1.5	47	194	178	184
1.0	35	186	133	143
0.9	31	184	118	129
0.8	26	181	99	111
0.7	20	177	75	89
0.6	12	171	43	59
0.5	Note 1	164	Note 1	18
<0.5	Note 1	164	Note 1	18

Note: 1. Rising input signal shall become equal to or greater than $V_{IH(AC)}$ level and Falling input signal shall become equal to or less than $V_{IL(AC)}$ level.

Figure 32: Nominal Slew Rate and t_{VAC} for t_{IS} (Command and Address – Clock)


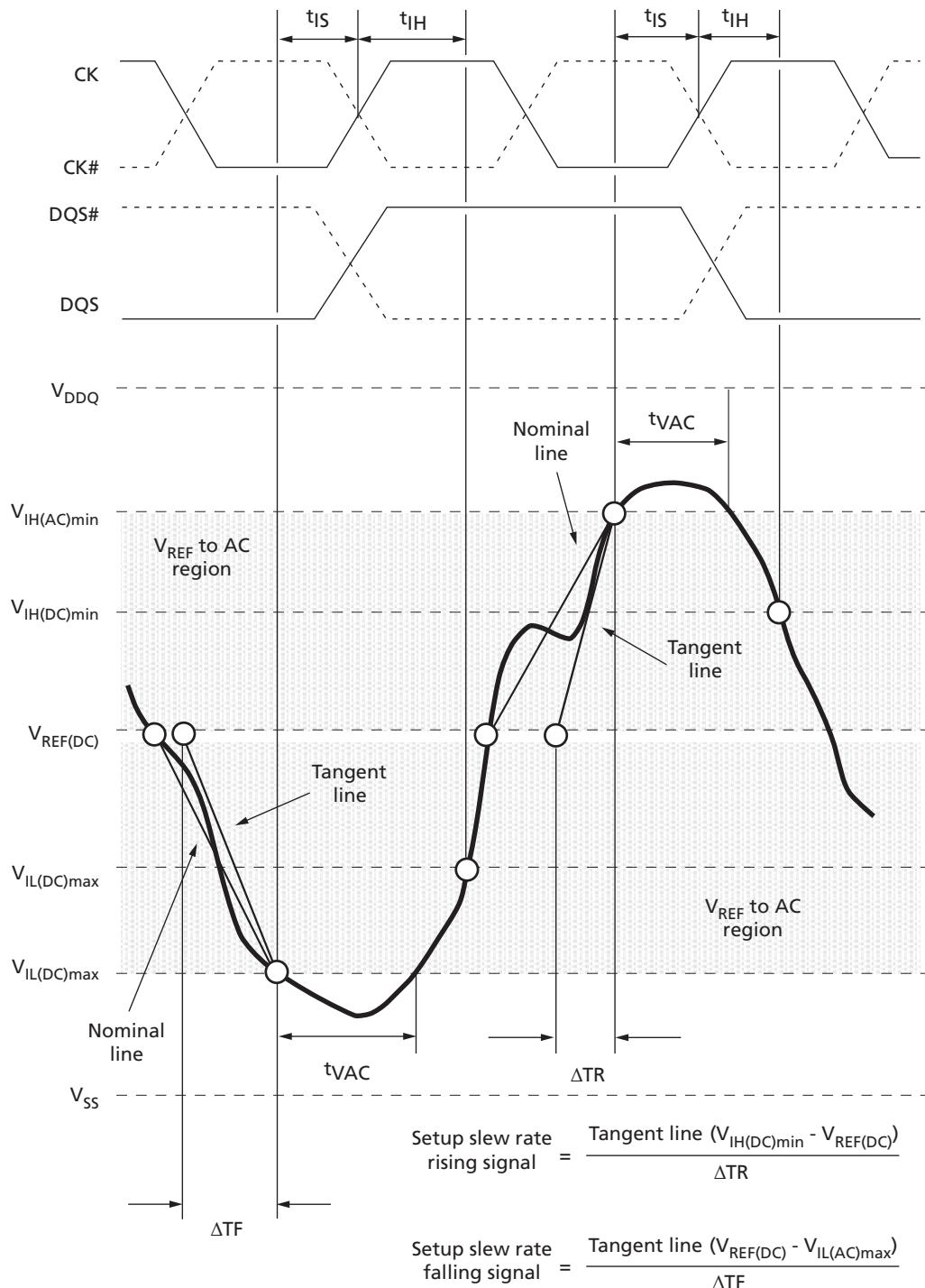
Note: 1. The clock and the strobe are drawn on different time scales.

Figure 33: Nominal Slew Rate for t_{IH} (Command and Address – Clock)


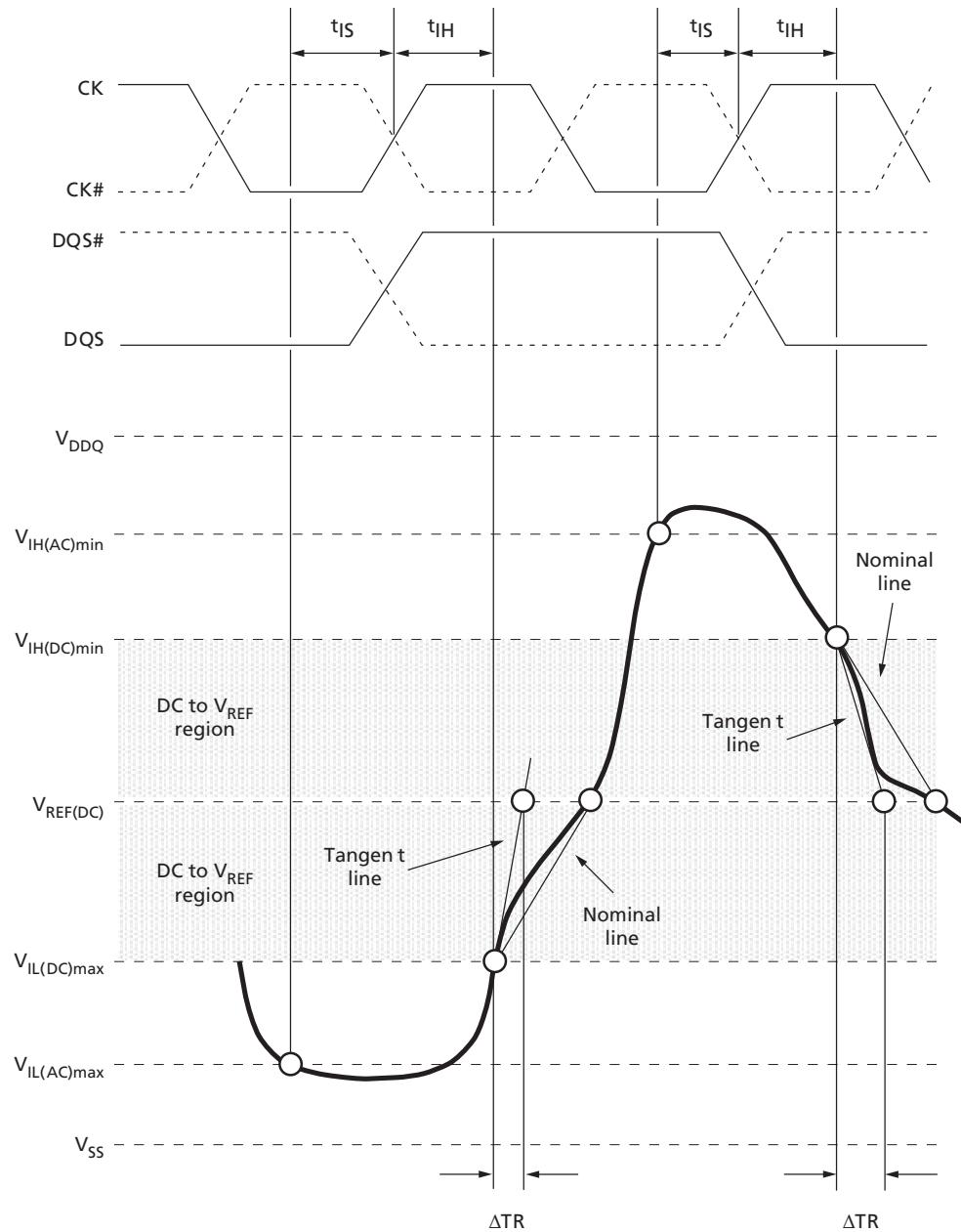
$$\text{Hold slew rate rising signal} = \frac{V_{REF(DC)} - V_{IL(DC)max}}{\Delta TR}$$

$$\text{Hold slew rate falling signal} = \frac{V_{IH(DC)min} - V_{REF(DC)}}{\Delta TF}$$

Note: 1. The clock and the strobe are drawn on different time scales.

Figure 34: Tangent Line for t_{IS} (Command and Address – Clock)


Note: 1. The clock and the strobe are drawn on different time scales.

Figure 35: Tangent Line for t_{IH} (Command and Address – Clock)


Note: 1. The clock and the strobe are drawn on different time scales.

Data Setup, Hold, and Derating

The total t_{DS} (setup time) and t_{DH} (hold time) required is calculated by adding the data sheet t_{DS} (base) and t_{DH} (base) values (see Table 64 (page 105); values come from Table 57 (page 79)) to the Δt_{DS} and Δt_{DH} derating values (see Table 65 (page 105)), Table 66 (page 105) or Table 66 (page 105) respectively. Example: t_{DS} (total setup time) = t_{DS} (base) + Δt_{DS} . For a valid transition, the input signal has to remain above/below $V_{IH(AC)}/V_{IL(AC)}$ for some time t_{VAC} (see Table 68 (page 107)).

Although the total setup time for slow slew rates might be negative (for example, a valid input signal will not have reached $V_{IH(AC)}/V_{IL(AC)}$ at the time of the rising clock transition), a valid input signal is still required to complete the transition and to reach $V_{IH}/V_{IL(AC)}$. For slew rates that fall between the values listed in Table 65 (page 105), Table 66 (page 105) or Table 66 (page 105), the derating values may be obtained by linear interpolation.

Setup (t_{DS}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{REF(DC)}$ and the first crossing of $V_{IH(AC)min}$. Setup (t_{DS}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{REF(DC)}$ and the first crossing of $V_{IL(AC)max}$. If the actual signal is always earlier than the nominal slew rate line between the shaded $V_{REF(DC)}$ -to-AC region, use the nominal slew rate for derating value (see Figure 36 (page 108)). If the actual signal is later than the nominal slew rate line anywhere between the shaded $V_{REF(DC)}$ -to-AC region, the slew rate of a tangent line to the actual signal from the AC level to the DC level is used for derating value (see Figure 38 (page 110)).

Hold (t_{DH}) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of $V_{IL(DC)max}$ and the first crossing of $V_{REF(DC)}$. Hold (t_{DH}) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of $V_{IH(DC)min}$ and the first crossing of $V_{REF(DC)}$. If the actual signal is always later than the nominal slew rate line between the shaded DC-to- $V_{REF(DC)}$ region, use the nominal slew rate for derating value (see Figure 37 (page 109)). If the actual signal is earlier than the nominal slew rate line anywhere between the shaded DC-to- $V_{REF(DC)}$ region, the slew rate of a tangent line to the actual signal from the DC-to- $V_{REF(DC)}$ region is used for derating value (see Figure 39 (page 111)).

Table 64: DDR3L Data Setup and Hold Values at 1 V/ns (DQS, DQS# at 2 V/ns) – AC/DC-Based

Symbol	800	1066	1333	1600	1866	Unit	Reference
t_{DS} (base) AC160	90	40	–	–	–	ps	$V_{IH(AC)}/V_{IL(AC)}$
t_{DS} (base) AC135	140	90	45	45	–	ps	$V_{IH(AC)}/V_{IL(AC)}$
t_{DS} (base) AC130	–	–	–	–	70	ps	$V_{IH(AC)}/V_{IL(AC)}$
t_{DH} (base) DC100	160	110	75	55	–	ps	$V_{IH(DC)}/V_{IL(DC)}$
t_{DH} (base) DC100	–	–	–	–	75	ps	$V_{IH(DC)}/V_{IL(DC)}$
Slew Rate Referenced	1	1	1	1	2	V/ns	

Table 65: DDR3L Derating Values for t_{DS}/t_{DH} – AC160/DC90-Based

$\Delta t_{DS}, \Delta t_{DH}$ Derating (ps) – AC/DC-Based																
DQ Slew Rate V/ns	DQS, DQS# Differential Slew Rate															
	4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}
2.0	80	45	80	45	80	45	–	–	–	–	–	–	–	–	–	–
1.5	53	30	53	30	53	30	61	38	–	–	–	–	–	–	–	–
1.0	0	0	0	0	0	0	8	8	16	16	–	–	–	–	–	–
0.9	–	–	–1	–3	–1	–3	7	5	15	13	23	21	–	–	–	–
0.8	–	–	–	–	–3	–8	5	1	13	9	21	17	29	27	–	–
0.7	–	–	–	–	–	–	–3	–5	11	3	19	11	27	21	35	37
0.6	–	–	–	–	–	–	–	–	8	–4	16	4	24	14	32	30
0.5	–	–	–	–	–	–	–	–	–	4	6	12	4	20	20	–
0.4	–	–	–	–	–	–	–	–	–	–	–	–8	–11	0	5	–

Table 66: DDR3L Derating Values for t_{DS}/t_{DH} – AC135/DC100-Based

$\Delta t_{DS}, \Delta t_{DH}$ Derating (ps) – AC/DC-Based																
DQ Slew Rate V/ns	DQS, DQS# Differential Slew Rate															
	4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}
2.0	68	45	68	45	68	45	–	–	–	–	–	–	–	–	–	–
1.5	45	30	45	30	45	30	53	38	–	–	–	–	–	–	–	–
1.0	0	0	0	0	0	0	8	8	16	16	–	–	–	–	–	–
0.9	–	–	2	–3	2	–3	10	5	18	13	26	21	–	–	–	–
0.8	–	–	–	–	3	–8	11	1	19	9	27	17	35	27	–	–
0.7	–	–	–	–	–	–	14	–5	22	3	30	11	38	21	46	37
0.6	–	–	–	–	–	–	–	25	–4	33	4	41	14	49	30	–
0.5	–	–	–	–	–	–	–	–	39	–6	37	4	45	20	–	–
0.4	–	–	–	–	–	–	–	–	–	30	–11	38	5	–	–	–



Table 67: DDR3L Derating Values for t_{DS}/t_{DH} – AC130/DC100-Based at 2V/ns

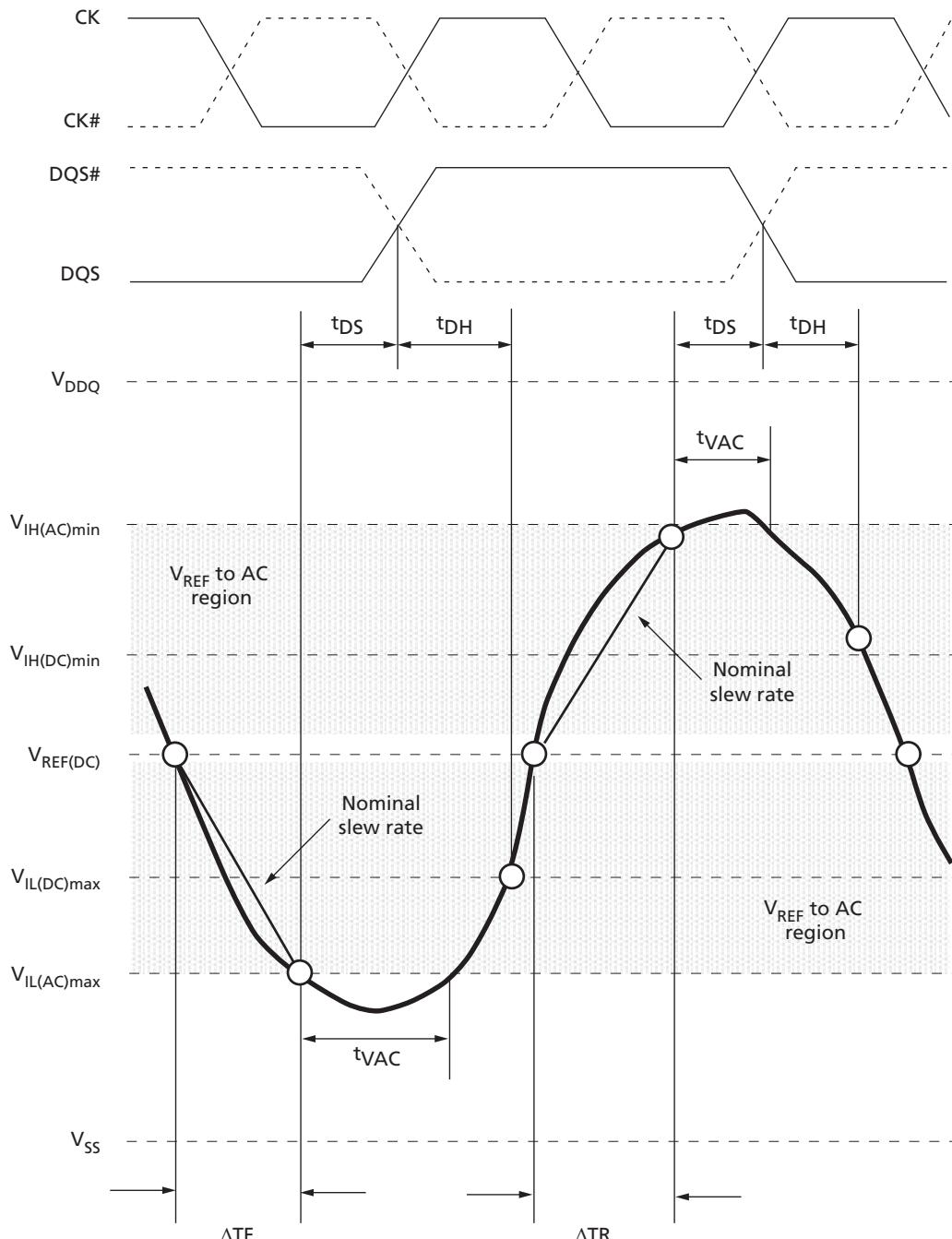
Shaded cells indicate slew rate combinations not supported

DQ Slew Rate V/ns		$\Delta t_{DS}, \Delta t_{DH}$ Derating (ps) – AC/DC-Based																							
		DQS, DQS# Differential Slew Rate																							
Δt_{DS}	Δt_{DH}	8.0 V/ns		7.0 V/ns		6.0 V/ns		5.0 V/ns		4.0 V/ns		3.0 V/ns		2.0 V/ns		1.8 V/ns		1.6 V/ns		1.4 V/ns		1.2 V/ns		1.0 V/ns	
		Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}	Δt_{DS}	Δt_{DH}		
4.0	33	23	33	23	33	23																			
3.5	28	19	28	19	28	19	28	19																	
3.0	22	15	22	15	22	15	22	15	22	15															
2.5			13	9	13	9	13	9	13	9	13	9	13	9											
2.0					0	0	0	0	0	0	0	0	0	0	0	0									
1.5							-22	-15	-22	-15	-22	-15	-22	-15	-22	-15	-14	-7							
1.0									-65	-45	-65	-45	-65	-45	-65	-45	-57	-37	-49	-29					
0.9											-62	-48	-62	-48	-62	-48	-54	-40	-46	-32	-38	-24			
0.8													-61	-53	-53	-45	-45	-37	-37	-29	-29	-29	-19		
0.7															-49	-50	-41	-42	-33	-34	-25	-24	-17	-8	
0.6																-37	-49	-29	-41	-21	-31	-13	-15		
0.5																		-31	-51	-23	-41	-15	-25		
0.4																			-28	-56	-20	-40			

Table 68: DDR3L Minimum Required Time t_{VAC} Above $V_{IH(AC)}$ (Below $V_{IL(AC)}$) for Valid DQ Transition

Slew Rate (V/ns)	DDR3L-800/1066 160mV (ps) min	DDR3L-800/1066/1333 135mV (ps) min	DDR3L-1866 130mV (ps) min
>2.0	165	113	95
2.0	165	113	95
1.5	138	90	73
1.0	85	45	30
0.9	67	30	16
0.8	45	11	Note1
0.7	16	Note1	—
0.6	Note1	Note1	—
0.5	Note1	Note1	—
<0.5	Note1	Note1	—

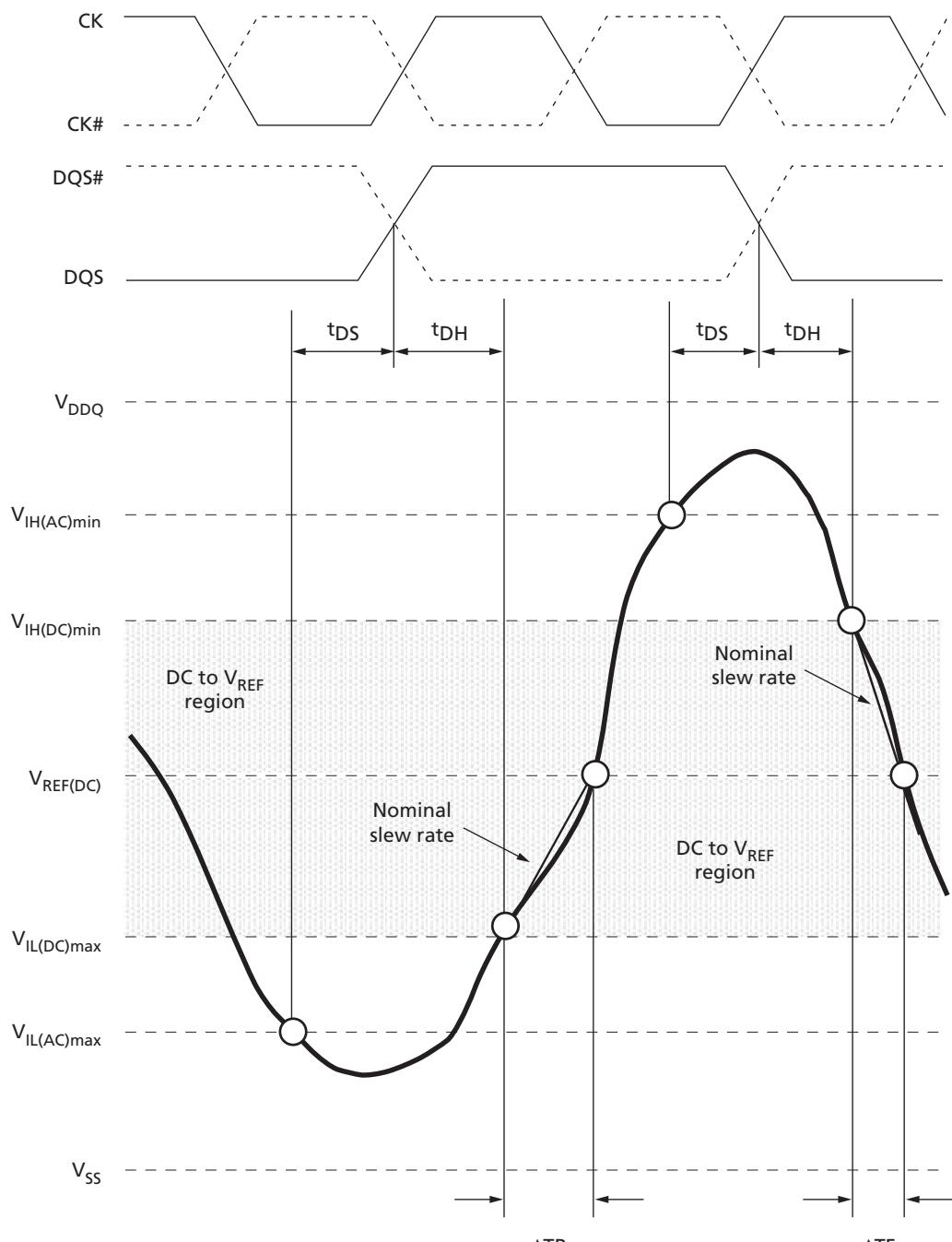
Note: 1. Rising input signal shall become equal to or greater than $V_{IH(AC)}$ level and Falling input signal shall become equal to or less than $V_{IL(AC)}$ level.

Figure 36: Nominal Slew Rate and t_{VAC} for t_{DS} (DQ – Strobe)


$$\text{Setup slew rate falling signal} = \frac{V_{REF(DC)} - V_{IL(AC)max}}{\Delta TF}$$

$$\text{Setup slew rate rising signal} = \frac{V_{IH(AC)min} - V_{REF(DC)}}{\Delta TR}$$

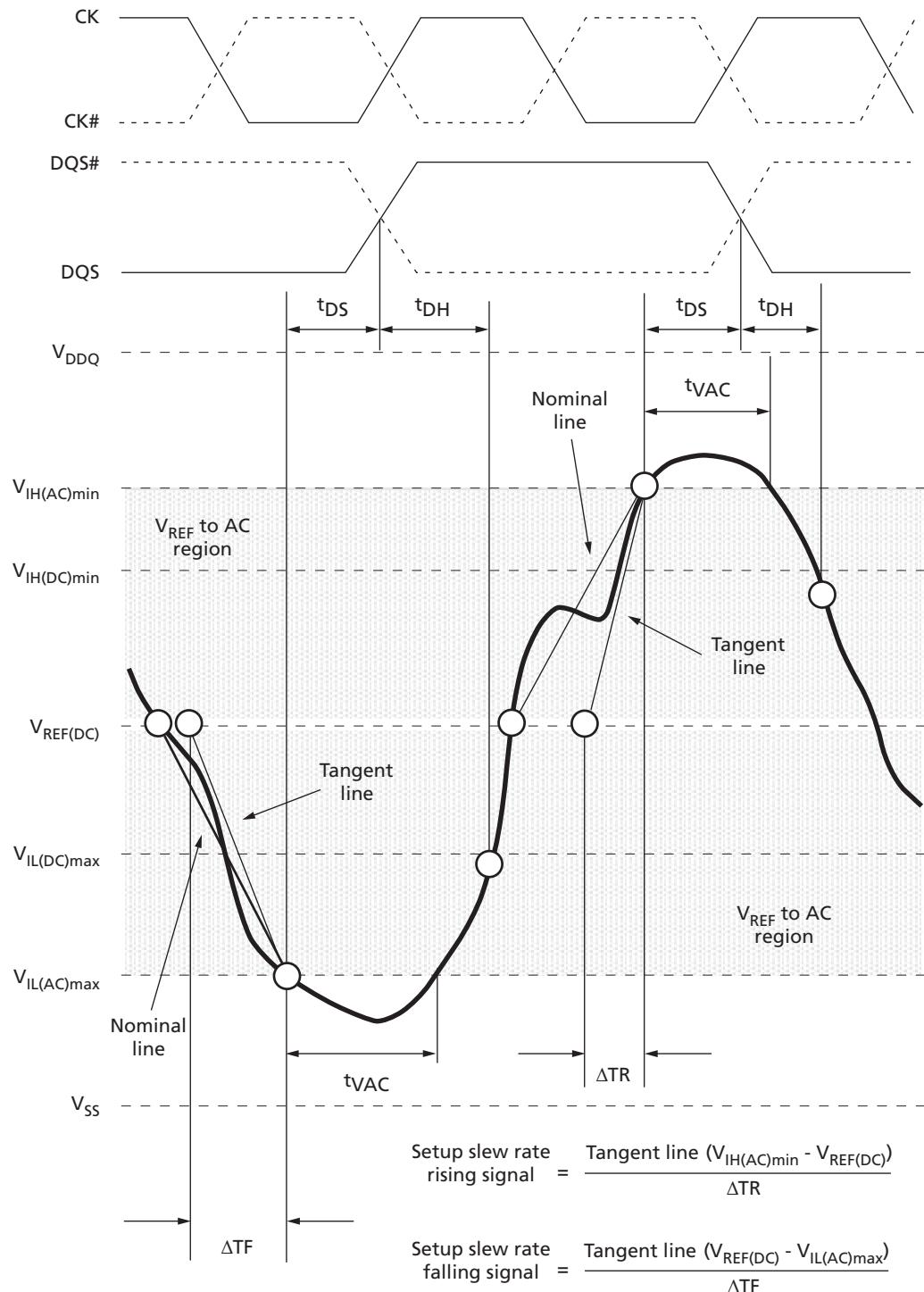
Note: 1. The clock and the strobe are drawn on different time scales.

Figure 37: Nominal Slew Rate for t_{DH} (DQ – Strobe)


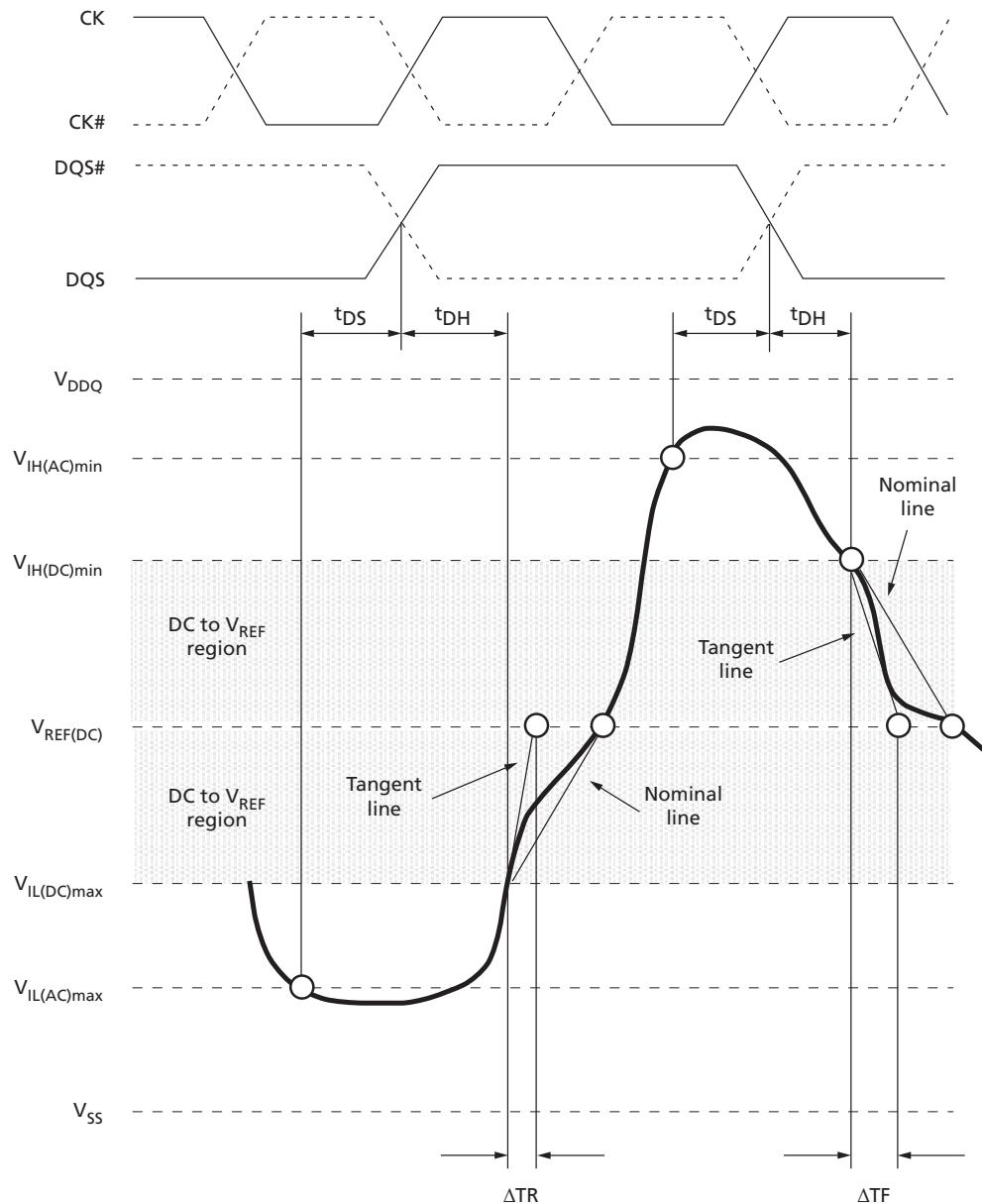
$$\text{Hold slew rate rising signal} = \frac{V_{REF(DC)} - V_{IL(DC)max}}{\Delta TR}$$

$$\text{Hold slew rate falling signal} = \frac{V_{IL(DC)min} - V_{REF(DC)}}{\Delta TF}$$

Note: 1. The clock and the strobe are drawn on different time scales.

Figure 38: Tangent Line for t_{DS} (DQ – Strobe)


Note: 1. The clock and the strobe are drawn on different time scales.

Figure 39: Tangent Line for t_{DH} (DQ – Strobe)


$$\text{Hold slew rate rising signal} = \frac{\text{Tangent line } (V_{REF(DC)} - V_{IL(DC)max})}{\Delta TR}$$

$$\text{Hold slew rate falling signal} = \frac{\text{Tangent line } (V_{IH(DC)min} - V_{REF(DC)})}{\Delta TF}$$

Note: 1. The clock and the strobe are drawn on different time scales.

Commands – Truth Tables

Table 69: Truth Table – Command

Notes 1–5 apply to the entire table

Function		Symbol	CKE		CS#	RAS#	CAS#	WE#	BA [2:0]	An	A12	A10	A[11, 9:0]	Notes
			Prev. Cycle	Next Cycle										
MODE REGISTER SET	MRS	H	H	L	L	L	L	L	BA	OP code				
REFRESH	REF	H	H	L	L	L	H	V	V	V	V	V	V	
Self refresh entry	SRE	H	L	L	L	L	H	V	V	V	V	V	V	6
Self refresh exit	SRX	L	H	H	V	V	V	V	V	V	V	V	V	6, 7
				L	H	H	H	V	V	V	V	V	V	
Single-bank PRECHARGE	PRE	H	H	L	L	H	L	BA	V	V	L	V		
PRECHARGE all banks	PREA	H	H	L	L	H	L	V		V	H	V		
Bank ACTIVATE	ACT	H	H	L	L	H	H	BA	Row address (RA)					
WRITE	BL8MRS, BC4MRS	WR	H	H	L	H	L	L	BA	RFU	V	L	CA	8
	BC4OTF	WRS4	H	H	L	H	L	L	BA	RFU	L	L	CA	8
	BL8OTF	WRS8	H	H	L	H	L	L	BA	RFU	H	L	CA	8
WRITE with auto precharge	BL8MRS, BC4MRS	WRAP	H	H	L	H	L	L	BA	RFU	V	H	CA	8
	BC4OTF	WRAPS4	H	H	L	H	L	L	BA	RFU	L	H	CA	8
	BL8OTF	WRAPS8	H	H	L	H	L	L	BA	RFU	H	H	CA	8
READ	BL8MRS, BC4MRS	RD	H	H	L	H	L	H	BA	RFU	V	L	CA	8
	BC4OTF	RDS4	H	H	L	H	L	H	BA	RFU	L	L	CA	8
	BL8OTF	RDS8	H	H	L	H	L	H	BA	RFU	H	L	CA	8
READ with auto precharge	BL8MRS, BC4MRS	RDAP	H	H	L	H	L	H	BA	RFU	V	H	CA	8
	BC4OTF	RDAPS4	H	H	L	H	L	H	BA	RFU	L	H	CA	8
	BL8OTF	RDAPS8	H	H	L	H	L	H	BA	RFU	H	H	CA	8
NO OPERATION		NOP	H	H		H	H	H	V	V	V	V	V	9
Device DESELECTED	DES	H	H	H	X	X	X	X	X	X	X	X	X	10
Power-down entry	PDE	H	L	L	H	H	H		V	V	V	V	V	6
				H	V	V	V							
Power-down exit	PDX	L	H	L	H	H	H	V	V	V	V	V	V	6, 11
				H	V	V	V							
ZQ CALIBRATION LONG	ZQCL	H	H	L	H	H	L	X	X	X	H	X		12
ZQ CALIBRATION SHORT	ZQCS	H	H	L	H	H	L	X	X	X	L	X		

Notes: 1. Commands are defined by the states of CS#, RAS#, CAS#, WE#, and CKE at the rising edge of the clock. The MSB of BA, RA, and CA are device-, density-, and configuration-dependent.

2. RESET# is enabled LOW and used only for asynchronous reset. Thus, RESET# must be held HIGH during any normal operation.
3. The state of ODT does not affect the states described in this table.
4. Operations apply to the bank defined by the bank address. For MRS, BA selects one of four mode registers.
5. "V" means "H" or "L" (a defined logic level), and "X" means "Don't Care."
6. See Table 70 (page 114) for additional information on CKE transition.
7. Self refresh exit is asynchronous.
8. Burst READs or WRITEs cannot be terminated or interrupted. MRS (fixed) and OTF BL/BC are defined in MRO.
9. The purpose of the NOP command is to prevent the DRAM from registering any unwanted commands. A NOP will not terminate an operation that is executing.
10. The DES and NOP commands perform similarly.
11. The power-down mode does not perform any REFRESH operations.
12. ZQ CALIBRATION LONG is used for either ZQinit (first ZQCL command during initialization) or ZQoper (ZQCL command after initialization).

Table 70: Truth Table – CKE

Notes 1–2 apply to the entire table; see Table 69 (page 112) for additional command details

Current State ³	CKE		Command ⁵ (RAS#, CAS#, WE#, CS#)	Action ⁵	Notes
	Previous Cycle ⁴ (n - 1)	Present Cycle ⁴ (n)			
Power-down	L	L	“Don’t Care”	Maintain power-down	
	L	H	DES or NOP	Power-down exit	
Self refresh	L	L	“Don’t Care”	Maintain self refresh	
	L	H	DES or NOP	Self refresh exit	
Bank(s) active	H	L	DES or NOP	Active power-down entry	
Reading	H	L	DES or NOP	Power-down entry	
Writing	H	L	DES or NOP	Power-down entry	
Precharging	H	L	DES or NOP	Power-down entry	
Refreshing	H	L	DES or NOP	Precharge power-down entry	
All banks idle	H	L	DES or NOP	Precharge power-down entry	6
	H	L	REFRESH	Self refresh	

Notes:

1. All states and sequences not shown are illegal or reserved unless explicitly described elsewhere in this document.
2. ^tCKE (MIN) means CKE must be registered at multiple consecutive positive clock edges. CKE must remain at the valid input level the entire time it takes to achieve the required number of registration clocks. Thus, after any CKE transition, CKE may not transition from its valid level during the time period of ^tIS + ^tCKE (MIN) + ^tIH.
3. Current state = The state of the DRAM immediately prior to clock edge n.
4. CKE (n) is the logic state of CKE at clock edge n; CKE (n - 1) was the state of CKE at the previous clock edge.
5. COMMAND is the command registered at the clock edge (must be a legal command as defined in Table 69 (page 112)). Action is a result of COMMAND. ODT does not affect the states described in this table and is not listed.
6. Idle state = All banks are closed, no data bursts are in progress, CKE is HIGH, and all timings from previous operations are satisfied. All self refresh exit and power-down exit parameters are also satisfied.

Commands

DESELECT

The DESELECT (DES) command (CS# HIGH) prevents new commands from being executed by the DRAM. Operations already in progress are not affected.

NO OPERATION

The NO OPERATION (NOP) command (CS# LOW) prevents unwanted commands from being registered during idle or wait states. Operations already in progress are not affected.

ZQ CALIBRATION LONG

The ZQ CALIBRATION LONG (ZQCL) command is used to perform the initial calibration during a power-up initialization and reset sequence (see Figure 48 (page 131)). This command may be issued at any time by the controller, depending on the system environment. The ZQCL command triggers the calibration engine inside the DRAM. After calibration is achieved, the calibrated values are transferred from the calibration engine to the DRAM I/O, which are reflected as updated R_{ON} and ODT values.

The DRAM is allowed a timing window defined by either t_{ZQinit} or t_{ZQoper} to perform a full calibration and transfer of values. When ZQCL is issued during the initialization sequence, the timing parameter t_{ZQinit} must be satisfied. When initialization is complete, subsequent ZQCL commands require the timing parameter t_{ZQoper} to be satisfied.

ZQ CALIBRATION SHORT

The ZQ CALIBRATION SHORT (ZQCS) command is used to perform periodic calibrations to account for small voltage and temperature variations. A shorter timing window is provided to perform the reduced calibration and transfer of values as defined by timing parameter t_{ZQCS} . A ZQCS command can effectively correct a minimum of 0.5% R_{ON} and R_{TT} impedance error within 64 clock cycles, assuming the maximum sensitivities specified in DDR3L 34 Ohm Output Driver Sensitivity (page 64).

ACTIVATE

The ACTIVATE command is used to open (or activate) a row in a particular bank for a subsequent access. The value on the BA[2:0] inputs selects the bank, and the address provided on inputs A[n:0] selects the row. This row remains open (or active) for accesses until a PRECHARGE command is issued to that bank.

A PRECHARGE command must be issued before opening a different row in the same bank.

READ

The READ command is used to initiate a burst read access to an active row. The address provided on inputs A[2:0] selects the starting column address, depending on the burst length and burst type selected (see Burst Order table for additional information). The value on input A10 determines whether auto precharge is used. If auto precharge is selected, the row being accessed will be precharged at the end of the READ burst. If auto

precharge is not selected, the row will remain open for subsequent accesses. The value on input A12 (if enabled in the mode register) when the READ command is issued determines whether BC4 (chop) or BL8 is used. After a READ command is issued, the READ burst may not be interrupted.

Table 71: READ Command Summary

Function		Symbol	CKE		CS#	RAS#	CAS#	WE#	BA [2:0]	An	A12	A10	A[11, 9:0]
			Prev. Cycle	Next Cycle									
READ	BL8MRS, BC4MRS	RD	H		L	H	L	H	BA	RFU	V	L	CA
	BC4OTF	RDS4	H		L	H	L	H	BA	RFU	L	L	CA
	BL8OTF	RDS8	H		L	H	L	H	BA	RFU	H	L	CA
READ with auto precharge	BL8MRS, BC4MRS	RDAP	H		L	H	L	H	BA	RFU	V	H	CA
	BC4OTF	RDAPS4	H		L	H	L	H	BA	RFU	L	H	CA
	BL8OTF	RDAPS8	H		L	H	L	H	BA	RFU	H	H	CA

WRITE

The WRITE command is used to initiate a burst write access to an active row. The value on the BA[2:0] inputs selects the bank. The value on input A10 determines whether auto precharge is used. The value on input A12 (if enabled in the MR) when the WRITE command is issued determines whether BC4 (chop) or BL8 is used.

Input data appearing on the DQ is written to the memory array subject to the DM input logic level appearing coincident with the data. If a given DM signal is registered LOW, the corresponding data will be written to memory. If the DM signal is registered HIGH, the corresponding data inputs will be ignored and a WRITE will not be executed to that byte/column location.

Table 72: WRITE Command Summary

Function		Symbol	CKE		CS#	RAS#	CAS#	WE#	BA [2:0]	An	A12	A10	A[11, 9:0]
			Prev. Cycle	Next Cycle									
WRITE	BL8MRS, BC4MRS	WR	H		L	H	L	L	BA	RFU	V	L	CA
	BC4OTF	WRS4	H		L	H	L	L	BA	RFU	L	L	CA
	BL8OTF	WRS8	H		L	H	L	L	BA	RFU	H	L	CA
WRITE with auto precharge	BL8MRS, BC4MRS	WRAP	H		L	H	L	L	BA	RFU	V	H	CA
	BC4OTF	WRAPS4	H		L	H	L	L	BA	RFU	L	H	CA
	BL8OTF	WRAPS8	H		L	H	L	L	BA	RFU	H	H	CA

PRECHARGE

The PRECHARGE command is used to de-activate the open row in a particular bank or in all banks. The bank(s) are available for a subsequent row access a specified time (t_{RP}) after the PRECHARGE command is issued, except in the case of concurrent auto pre-charge. A READ or WRITE command to a different bank is allowed during a concurrent auto precharge as long as it does not interrupt the data transfer in the current bank and does not violate any other timing parameters. Input A10 determines whether one or all banks are precharged. In the case where only one bank is precharged, inputs BA[2:0] select the bank; otherwise, BA[2:0] are treated as “Don’t Care.”

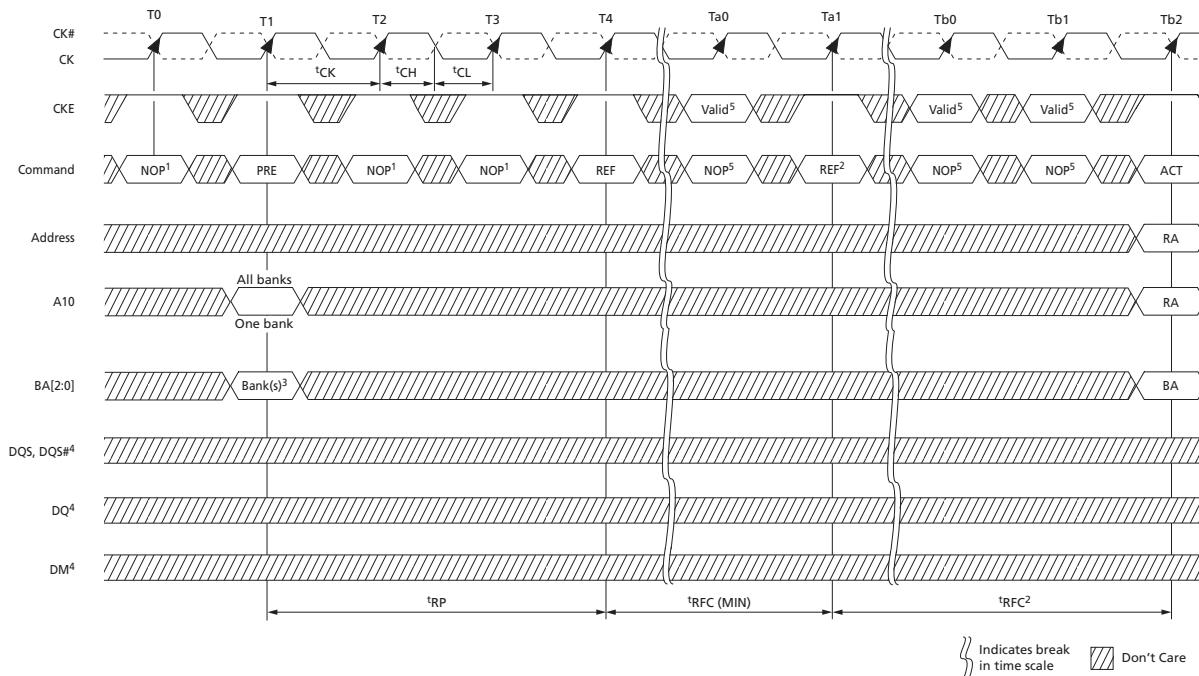
After a bank is precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank. A PRECHARGE command is treated as a NOP if there is no open row in that bank (idle state) or if the previously open row is already in the process of precharging. However, the precharge period is determined by the last PRECHARGE command issued to the bank.

REFRESH

The REFRESH command is used during normal operation of the DRAM and is analogous to CAS#-before-RAS# (CBR) refresh or auto refresh. This command is nonpersistent, so it must be issued each time a refresh is required. The addressing is generated by the internal refresh controller. This makes the address bits a “Don’t Care” during a REFRESH command. The DRAM requires REFRESH cycles at an average interval of $7.8\mu s$ (maximum when $T_C \leq 85^\circ C$ or $3.9\mu s$ maximum when $T_C \leq 95^\circ C$). The REFRESH period begins when the REFRESH command is registered and ends $t_{RFC}(\text{MIN})$ later.

To allow for improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. A maximum of eight REFRESH commands can be posted to any given DRAM, meaning that the maximum absolute interval between any REFRESH command and the next REFRESH command is nine times the maximum average interval refresh rate. Self refresh may be entered with up to eight REFRESH commands being posted. After exiting self refresh (when entered with posted REFRESH commands), additional posting of REFRESH commands is allowed to the extent that the maximum number of cumulative posted REFRESH commands (both pre- and post-self refresh) does not exceed eight REFRESH commands.

At any given time, a maximum of 16 REFRESH commands can be issued within $2 \times t_{REFI}$.

Figure 40: Refresh Mode


Notes:

1. NOP commands are shown for ease of illustration; other valid commands may be possible at these times. CKE must be active during the PRECHARGE, ACTIVATE, and REFRESH commands, but may be inactive at other times (see Power-Down Mode (page 181)).
2. The second REFRESH is not required, but two back-to-back REFRESH commands are shown.
3. "Don't Care" if A10 is HIGH at this point; however, A10 must be HIGH if more than one bank is active (must precharge all active banks).
4. For operations shown, DM, DQ, and DQS signals are all "Don't Care"/High-Z.
5. Only NOP and DES commands are allowed after a REFRESH command and until tRFC (MIN) is satisfied.

SELF REFRESH

The SELF REFRESH command is used to retain data in the DRAM, even if the rest of the system is powered down. When in self refresh mode, the DRAM retains data without external clocking. Self refresh mode is also a convenient method used to enable/disable the DLL as well as to change the clock frequency within the allowed synchronous operating range (see Input Clock Frequency Change (page 123)). All power supply inputs (including V_{REFCA} and V_{REFDQ}) must be maintained at valid levels upon entry/exit and during self refresh mode operation. V_{REFDQ} may float or not drive $V_{DDQ}/2$ while in self refresh mode under the following conditions:

- $V_{SS} < V_{REFDQ} < V_{DD}$ is maintained
- V_{REFDQ} is valid and stable prior to CKE going back HIGH
- The first WRITE operation may not occur earlier than 512 clocks after V_{REFDQ} is valid
- All other self refresh mode exit timing requirements are met

DLL Disable Mode

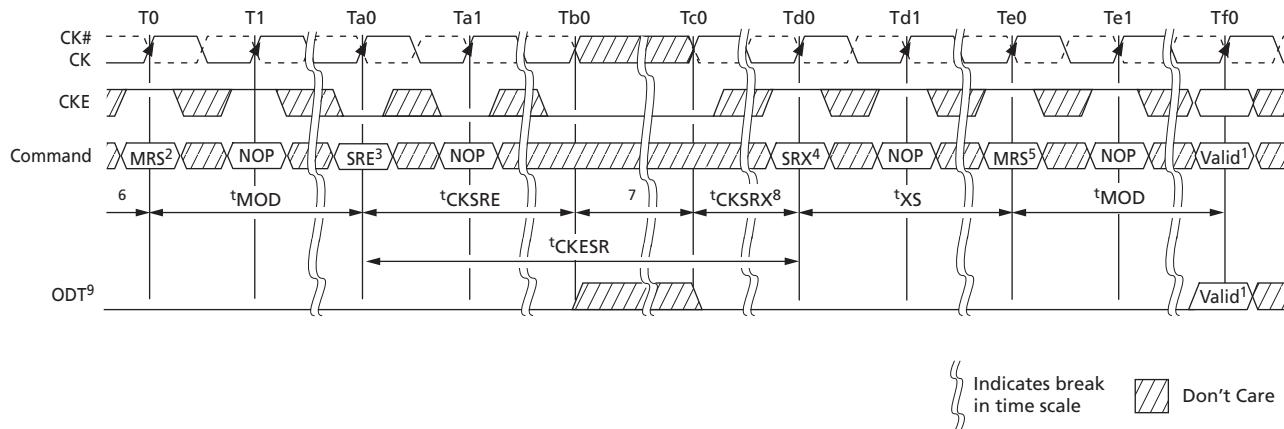
If the DLL is disabled by the mode register (MR1[0] can be switched during initialization or later), the DRAM is targeted, but not guaranteed, to operate similarly to the normal mode, with a few notable exceptions:

- The DRAM supports only one value of CAS latency (CL = 6) and one value of CAS WRITE latency (CWL = 6).
- DLL disable mode affects the read data clock-to-data strobe relationship (t_{DQSCK}), but not the read data-to-data strobe relationship (t_{DQSQ} , t_{QH}). Special attention is required to line up the read data with the controller time domain when the DLL is disabled.
- In normal operation (DLL on), t_{DQSCK} starts from the rising clock edge $AL + CL$ cycles after the READ command. In DLL disable mode, t_{DQSCK} starts $AL + CL - 1$ cycles after the READ command. Additionally, with the DLL disabled, the value of t_{DQSCK} could be larger than t_{CK} .

The ODT feature (including dynamic ODT) is not supported during DLL disable mode. The ODT resistors must be disabled by continuously registering the ODT ball LOW by programming $R_{TT,nom}$ MR1[9, 6, 2] and $R_{TT(WR)}$ MR2[10, 9] to 0 while in the DLL disable mode.

Specific steps must be followed to switch between the DLL enable and DLL disable modes due to a gap in the allowed clock rates between the two modes (t_{CK} [AVG] MAX and t_{CK} [DLL_DIS] MIN, respectively). The only time the clock is allowed to cross this clock rate gap is during self refresh mode. Thus, the required procedure for switching from the DLL enable mode to the DLL disable mode is to change frequency during self refresh:

1. Starting from the idle state (all banks are precharged, all timings are fulfilled, ODT is turned off, and $R_{TT,nom}$ and $R_{TT(WR)}$ are High-Z), set MR1[0] to 1 to disable the DLL.
2. Enter self refresh mode after t_{MOD} has been satisfied.
3. After t_{CKSRE} is satisfied, change the frequency to the desired clock rate.
4. Self refresh may be exited when the clock is stable with the new frequency for t_{CKSRX} . After t_{XS} is satisfied, update the mode registers with appropriate values.
5. The DRAM will be ready for its next command in the DLL disable mode after the greater of t_{MRD} or t_{MOD} has been satisfied. A ZQCL command should be issued with appropriate timings met.

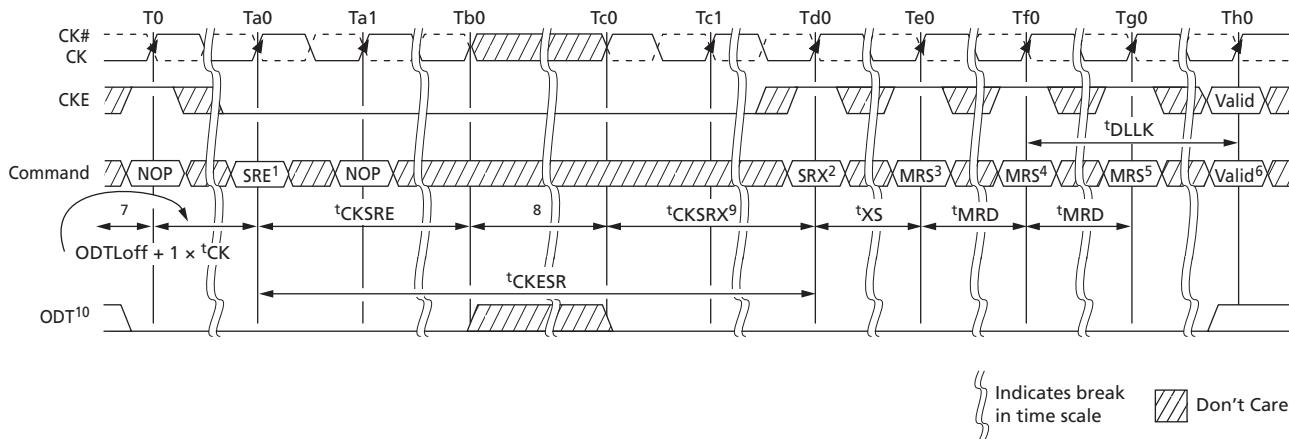
Figure 41: DLL Enable Mode to DLL Disable Mode


Notes:

1. Any valid command.
2. Disable DLL by setting MR1[0] to 1.
3. Enter SELF REFRESH.
4. Exit SELF REFRESH.
5. Update the mode registers with the DLL disable parameters setting.
6. Starting with the idle state, R_{TT} is in the High-Z state.
7. Change frequency.
8. Clock must be stable t_{CKSRX} .
9. Static LOW in the case that $R_{TT,nom}$ or $R_{TT(WR)}$ is enabled; otherwise, static LOW or HIGH.

A similar procedure is required for switching from the DLL disable mode back to the DLL enable mode. This also requires changing the frequency during self refresh mode (see Figure 42 (page 121)).

1. Starting from the idle state (all banks are precharged, all timings are fulfilled, ODT is turned off, and $R_{TT,nom}$ and $R_{TT(WR)}$ are High-Z), enter self refresh mode.
2. After t_{CKSRE} is satisfied, change the frequency to the new clock rate.
3. Self refresh may be exited when the clock is stable with the new frequency for t_{CKSRX} . After t_{XS} is satisfied, update the mode registers with the appropriate values. At a minimum, set MR1[0] to 0 to enable the DLL. Wait t_{MRD} , then set MR0[8] to 1 to enable DLL RESET.
4. After another t_{MRD} delay is satisfied, update the remaining mode registers with the appropriate values.
5. The DRAM will be ready for its next command in the DLL enable mode after the greater of t_{MRD} or t_{MOD} has been satisfied. However, before applying any command or function requiring a locked DLL, a delay of t_{DLLK} after DLL RESET must be satisfied. A ZQCL command should be issued with the appropriate timings met.

Figure 42: DLL Disable Mode to DLL Enable Mode


Notes:

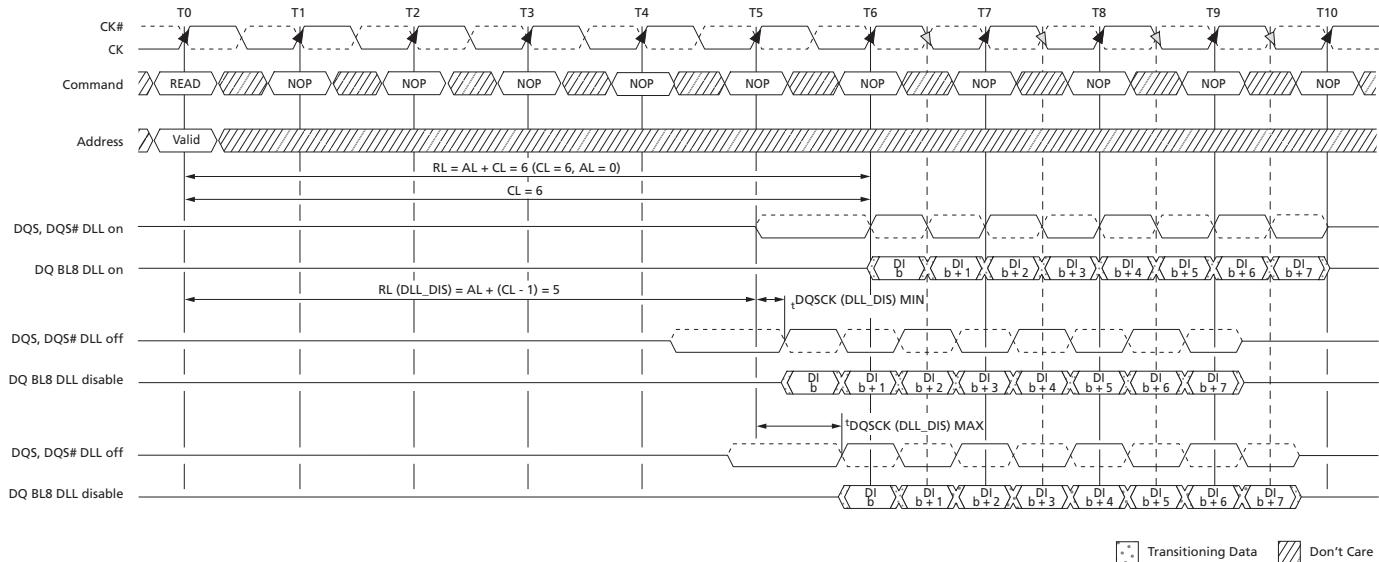
1. Enter SELF REFRESH.
2. Exit SELF REFRESH.
3. Wait t_{XS} , then set MR1[0] to 0 to enable DLL.
4. Wait t_{MRD} , then set MRO[8] to 1 to begin DLL RESET.
5. Wait t_{MRD} , update registers (CL, CWL, and write recovery may be necessary).
6. Wait t_{MOD} , any valid command.
7. Starting with the idle state.
8. Change frequency.
9. Clock must be stable at least t_{CKSRX} .
10. Static LOW in the case that $R_{TT,nom}$ or $R_{TT(WR)}$ is enabled; otherwise, static LOW or HIGH.

The clock frequency range for the DLL disable mode is specified by the parameter t_{CK} (DLL_DIS). Due to latency counter and timing restrictions, only CL = 6 and CWL = 6 are supported.

DLL disable mode will affect the read data clock to data strobe relationship (t_{DQSCK}) but not the data strobe to data relationship (t_{DQSQ} , t_{QH}). Special attention is needed to line up read data to the controller time domain.

Compared to the DLL on mode where t_{DQSCK} starts from the rising clock edge AL + CL cycles after the READ command, the DLL disable mode t_{DQSCK} starts AL + CL - 1 cycles after the READ command.

WRITE operations function similarly between the DLL enable and DLL disable modes; however, ODT functionality is not allowed with DLL disable mode.

Figure 43: DLL Disable t_{DQSCK}

Table 73: READ Electrical Characteristics, DLL Disable Mode

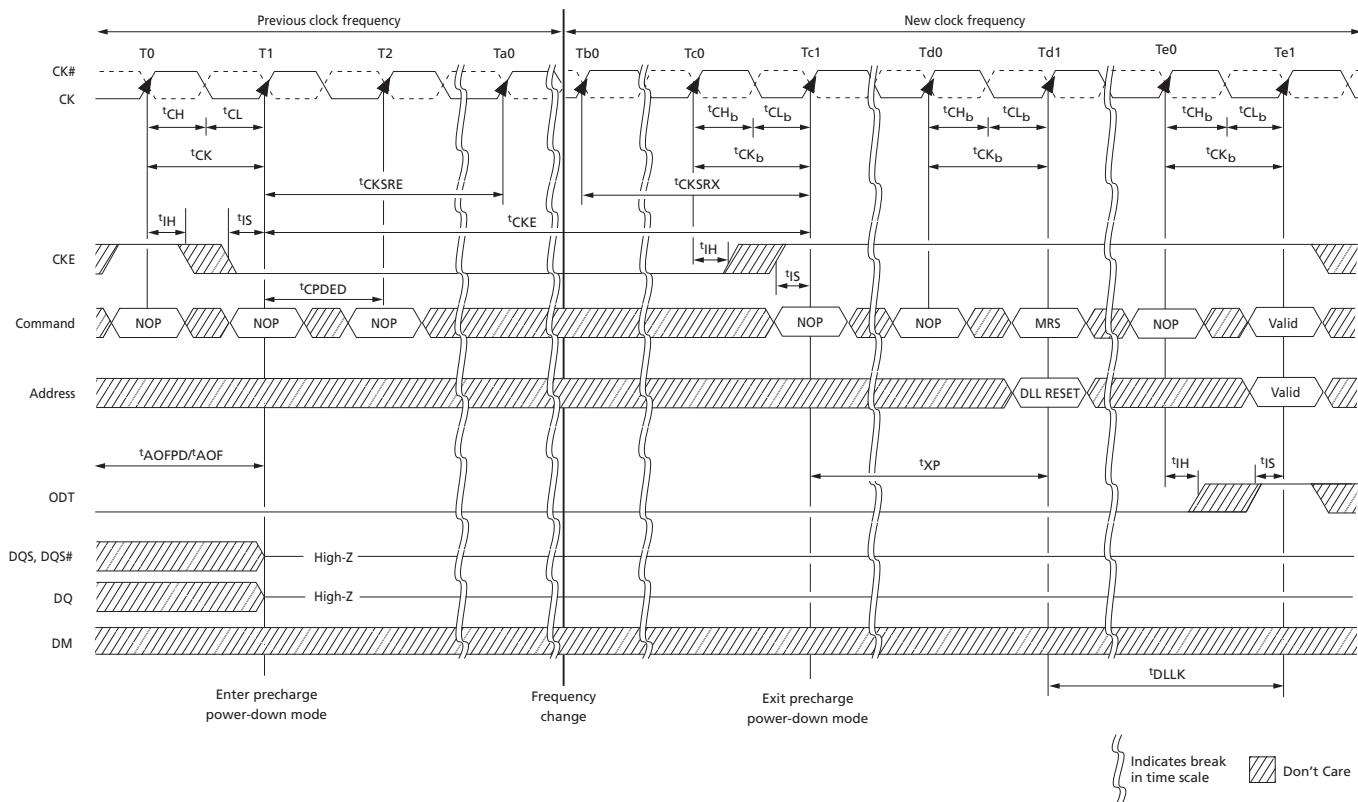
Parameter	Symbol	Min	Max	Unit
Access window of DQS from CK, CK#	$t_{DQSCK} (DLL_DIS)$	1	10	ns

Input Clock Frequency Change

When the DDR3 SDRAM is initialized, the clock must be stable during most normal states of operation. This means that after the clock frequency has been set to the stable state, the clock period is not allowed to deviate, except for what is allowed by the clock jitter and spread spectrum clocking (SSC) specifications.

The input clock frequency can be changed from one stable clock rate to another under two conditions: self refresh mode and precharge power-down mode. It is illegal to change the clock frequency outside of those two modes. For the self refresh mode condition, when the DDR3 SDRAM has been successfully placed into self refresh mode and t_{CKSRE} has been satisfied, the state of the clock becomes a “Don’t Care.” When the clock becomes a “Don’t Care,” changing the clock frequency is permissible if the new clock frequency is stable prior to t_{CKSRX} . When entering and exiting self refresh mode for the sole purpose of changing the clock frequency, the self refresh entry and exit specifications must still be met.

The precharge power-down mode condition is when the DDR3 SDRAM is in precharge power-down mode (either fast exit mode or slow exit mode). Either ODT must be at a logic LOW or $R_{TT,nom}$ and $R_{TT(WR)}$ must be disabled via MR1 and MR2. This ensures $R_{TT,nom}$ and $R_{TT(WR)}$ are in an off state prior to entering precharge power-down mode, and CKE must be at a logic LOW. A minimum of t_{CKSRE} must occur after CKE goes LOW before the clock frequency can change. The DDR3 SDRAM input clock frequency is allowed to change only within the minimum and maximum operating frequency specified for the particular speed grade ($t_{CK [AVG] MIN}$ to $t_{CK [AVG] MAX}$). During the input clock frequency change, CKE must be held at a stable LOW level. When the input clock frequency is changed, a stable clock must be provided to the DRAM t_{CKSRX} before precharge power-down may be exited. After precharge power-down is exited and t_{XP} has been satisfied, the DLL must be reset via the MRS. Depending on the new clock frequency, additional MRS commands may need to be issued. During the DLL lock time, $R_{TT,nom}$ and $R_{TT(WR)}$ must remain in an off state. After the DLL lock time, the DRAM is ready to operate with a new clock frequency.

Figure 44: Change Frequency During Precharge Power-Down


Notes:

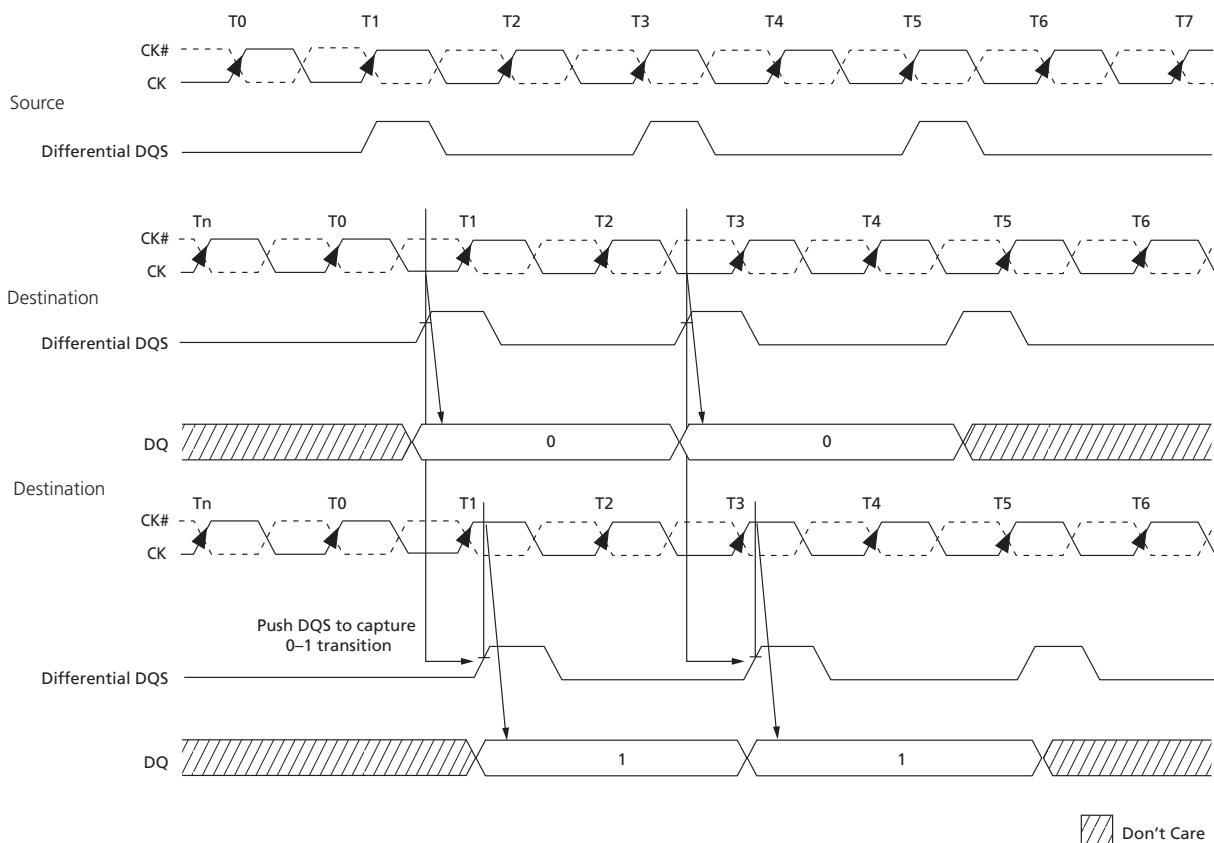
1. Applicable for both SLOW-EXIT and FAST-EXIT precharge power-down modes.
2. t_{AOPD} and t_{AOF} must be satisfied and outputs High-Z prior to T1 (see On-Die Termination (ODT) (page 191) for exact requirements).
3. If the $R_{TT,nom}$ feature was enabled in the mode register prior to entering precharge power-down mode, the ODT signal must be continuously registered LOW, ensuring R_{TT} is in an off state. If the $R_{TT,nom}$ feature was disabled in the mode register prior to entering precharge power-down mode, R_{TT} will remain in the off state. The ODT signal can be registered LOW or HIGH in this case.

Write Leveling

For better signal integrity, DDR3 SDRAM memory modules have adopted fly-by topology for the commands, addresses, control signals, and clocks. Write leveling is a scheme for the memory controller to adjust or de-skew the DQS strobe (DQS, DQS#) to CK relationship at the DRAM with a simple feedback feature provided by the DRAM. Write leveling is generally used as part of the initialization process, if required. For normal DRAM operation, this feature must be disabled. This is the only DRAM operation where the DQS functions as an input (to capture the incoming clock) and the DQ function as outputs (to report the state of the clock). Note that nonstandard ODT schemes are required.

The memory controller using the write leveling procedure must have adjustable delay settings on its DQS strobe to align the rising edge of DQS to the clock at the DRAM pins. This is accomplished when the DRAM asynchronously feeds back the CK status via the DQ bus and samples with the rising edge of DQS. The controller repeatedly delays the DQS strobe until a CK transition from 0 to 1 is detected. The DQS delay established by this procedure helps ensure t_{DQSS} , t_{DSS} , and t_{DSH} specifications in systems that use fly-by topology by de-skewing the trace length mismatch. A conceptual timing of this procedure is shown in Figure 45.

Figure 45: Write Leveling Concept



When write leveling is enabled, the rising edge of DQS samples CK, and the prime DQ outputs the sampled CK's status. The prime DQ for a x4 or x8 configuration is DQ0 with all other DQ (DQ[7:1]) driving LOW. The prime DQ for a x16 configuration is DQ0 for the lower byte and DQ8 for the upper byte. It outputs the status of CK sampled by LDQS and UDQS. All other DQ (DQ[7:1], DQ[15:9]) continue to drive LOW. Two prime DQ on a x16 enable each byte lane to be leveled independently.

The write leveling mode register interacts with other mode registers to correctly configure the write leveling functionality. Besides using MR1[7] to disable/enable write leveling, MR1[12] must be used to enable/disable the output buffers. The ODT value, burst length, and so forth need to be selected as well. This interaction is shown in Table 74. It should also be noted that when the outputs are enabled during write leveling mode, the DQS buffers are set as inputs, and the DQ are set as outputs. Additionally, during write leveling mode, only the DQS strobe terminations are activated and deactivated via the ODT ball. The DQ remain disabled and are not affected by the ODT ball.

Table 74: Write Leveling Matrix

Note 1 applies to the entire table

MR1[7]	MR1[12]	MR1[2, 6, 9]	DRAM ODT Ball	DRAM $R_{TT,nom}$		DRAM State	Case	Notes
Write Leveling	Output Buffers	$R_{TT,nom}$ Value		DQS	DQ			
Disabled	See normal operations					Write leveling not enabled	0	
Enabled (1)	Disabled (1)	n/a	Low	Off	Off	DQS not receiving: not terminated Prime DQ High-Z: not terminated Other DQ High-Z: not terminated	1	2
		20Ω, 30Ω, 40Ω, 60Ω, or 120Ω	High	On		DQS not receiving: terminated by R_{TT} Prime DQ High-Z: not terminated Other DQ High-Z: not terminated	2	
	Enabled (0)	n/a	Low	Off		DQS receiving: not terminated Prime DQ driving CK state: not terminated Other DQ driving LOW: not terminated	3	3
		40Ω, 60Ω, or 120Ω	High	On		DQS receiving: terminated by R_{TT} Prime DQ driving CK state: not terminated Other DQ driving LOW: not terminated	4	

Notes:

1. Expected usage if used during write leveling: Case 1 may be used when DRAM are on a dual-rank module and on the rank not being leveled or on any rank of a module not being leveled on a multislot system. Case 2 may be used when DRAM are on any rank of a module not being leveled on a multislot system. Case 3 is generally not used. Case 4 is generally used when DRAM are on the rank that is being leveled.
2. Since the DRAM DQS is not being driven (MR1[12] = 1), DQS ignores the input strobe, and all $R_{TT,nom}$ values are allowed. This simulates a normal standby state to DQS.
3. Since the DRAM DQS is being driven (MR1[12] = 0), DQS captures the input strobe, and only some $R_{TT,nom}$ values are allowed. This simulates a normal write state to DQS.

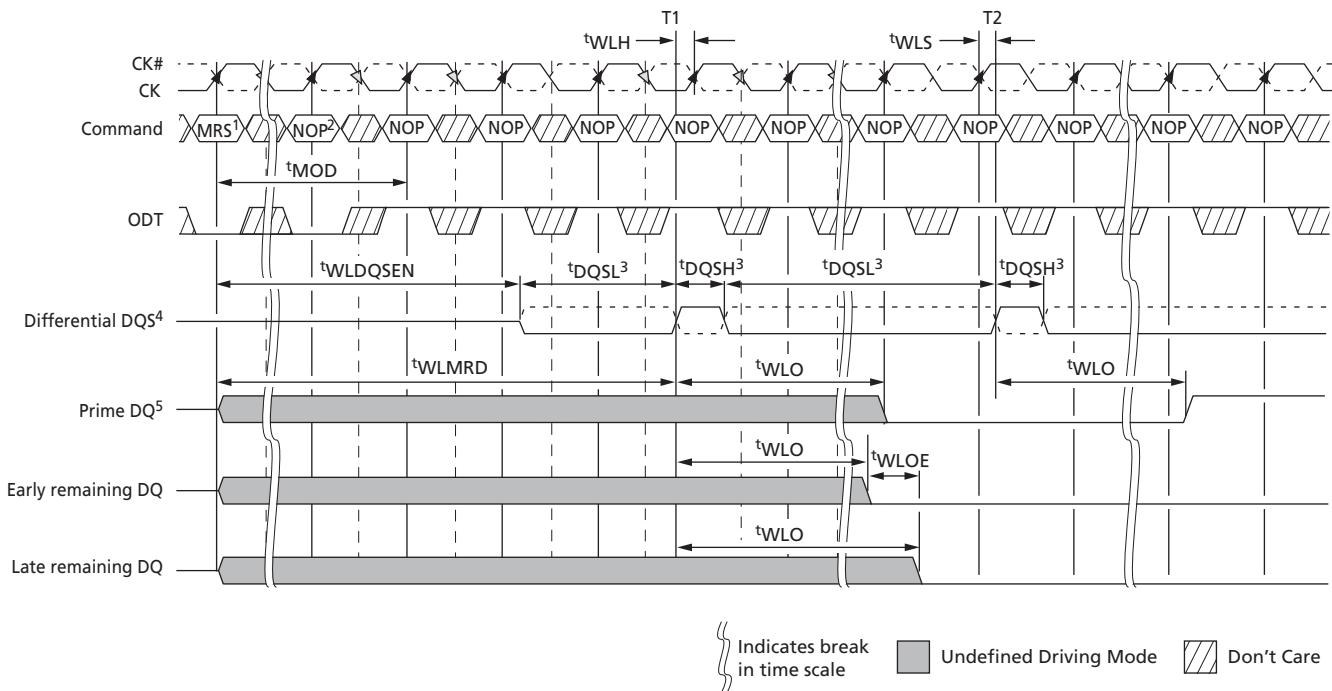
Write Leveling Procedure

A memory controller initiates the DRAM write leveling mode by setting MR1[7] to 1, assuming the other programmable features (MR0, MR1, MR2, and MR3) are first set and the DLL is fully reset and locked. The DQ balls enter the write leveling mode going from a High-Z state to an undefined driving state, so the DQ bus should not be driven. During write leveling mode, only the NOP or DES commands are allowed. The memory controller should attempt to level only one rank at a time; thus, the outputs of other ranks should be disabled by setting MR1[12] to 1 in the other ranks. The memory controller may assert ODT after a t_{MOD} delay, as the DRAM will be ready to process the ODT transition. ODT should be turned on prior to DQS being driven LOW by at least ODT_{Lon} delay (WL - 2 t_{CK}), provided it does not violate the aforementioned t_{MOD} delay requirement.

The memory controller may drive DQS LOW and DQS# HIGH after $t_{WLDQSEN}$ has been satisfied. The controller may begin to toggle DQS after t_{WLMRD} (one DQS toggle is DQS transitioning from a LOW state to a HIGH state with DQS# transitioning from a HIGH state to a LOW state, then both transition back to their original states). At a minimum, ODT_{Lon} and t_{AON} must be satisfied at least one clock prior to DQS toggling.

After t_{WLMRD} and a DQS LOW preamble (t_{WPRE}) have been satisfied, the memory controller may provide either a single DQS toggle or multiple DQS toggles to sample CK for a given DQS-to-CK skew. Each DQS toggle must not violate t_{DQSL} (MIN) and t_{DQSH} (MIN) specifications. t_{DQSL} (MAX) and t_{DQSH} (MAX) specifications are not applicable during write leveling mode. The DQS must be able to distinguish the CK's rising edge within t_{WLS} and t_{WLH} . The prime DQ will output the CK's status asynchronously from the associated DQS rising edge CK capture within t_{WLO} . The remaining DQ that always drive LOW when DQS is toggling must be LOW within t_{WLOE} after the first t_{WLO} is satisfied (the prime DQ going LOW). As previously noted, DQS is an input and not an output during this process. Figure 46 (page 128) depicts the basic timing parameters for the overall write leveling procedure.

The memory controller will most likely sample each applicable prime DQ state and determine whether to increment or decrement its DQS delay setting. After the memory controller performs enough DQS toggles to detect the CK's 0-to-1 transition, the memory controller should lock the DQS delay setting for that DRAM. After locking the DQS setting is locked, leveling for the rank will have been achieved, and the write leveling mode for the rank should be disabled or reprogrammed (if write leveling of another rank follows).

Figure 46: Write Leveling Sequence


Notes:

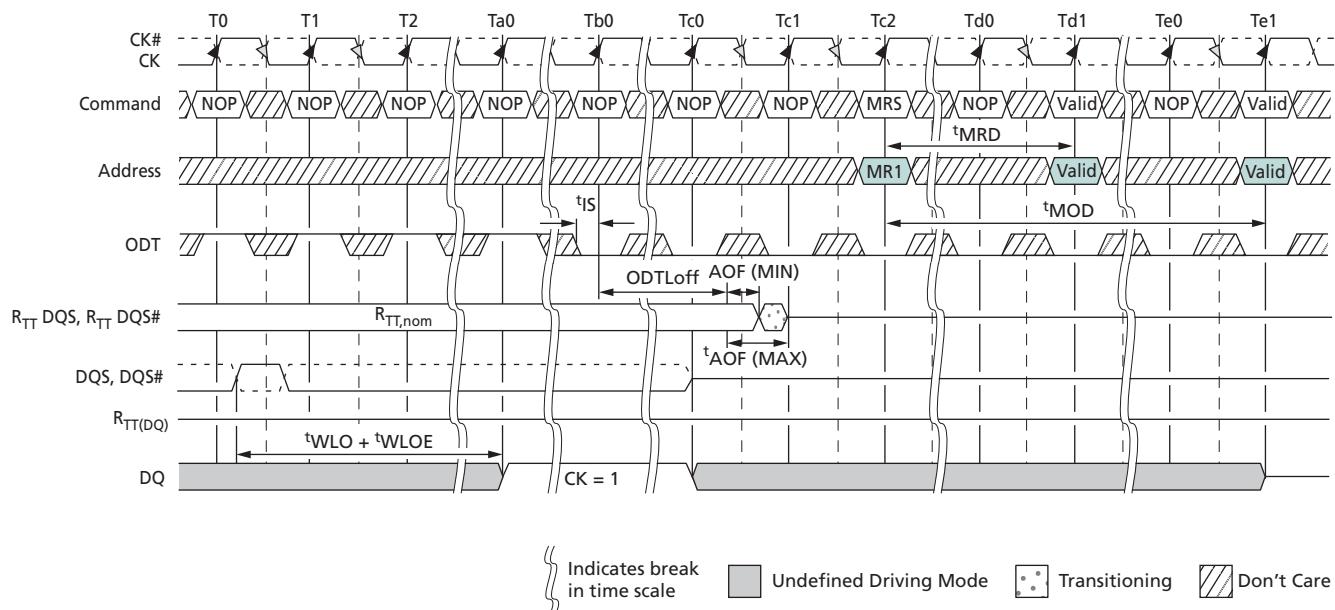
1. MRS: Load MR1 to enter write leveling mode.
2. NOP: NOP or DES.
3. DQS, DQS# needs to fulfill minimum pulse width requirements $t_{DQSH}^{(MIN)}$ and $t_{DQSL}^{(MIN)}$ as defined for regular writes. The maximum pulse width is system-dependent.
4. Differential DQS is the differential data strobe (DQS, DQS#). Timing reference points are the zero crossings. The solid line represents DQS; the dotted line represents DQS#.
5. DRAM drives leveling feedback on a prime DQ (DQ0 for x4 and x8). The remaining DQ are driven LOW and remain in this state throughout the leveling procedure.

Write Leveling Mode Exit Procedure

After the DRAM are leveled, they must exit from write leveling mode before the normal mode can be used. Figure 47 depicts a general procedure for exiting write leveling mode. After the last rising DQS (capturing a 1 at T0), the memory controller should stop driving the DQS signals after $t_{WLO}(\text{MAX})$ delay plus enough delay to enable the memory controller to capture the applicable prime DQ state (at $\sim T_{b0}$). The DQ balls become undefined when DQS no longer remains LOW, and they remain undefined until t_{MOD} after the MRS command (at T_{e1}).

The ODT input should be de-asserted LOW such that ODTLooff (MIN) expires after the DQS is no longer driving LOW. When ODT LOW satisfies t_{IS} , ODT must be kept LOW (at $\sim T_{b0}$) until the DRAM is ready for either another rank to be leveled or until the normal mode can be used. After DQS termination is switched off, write level mode should be disabled via the MRS command (at T_{c2}). After t_{MOD} is satisfied (at T_{e1}), any valid command may be registered by the DRAM. Some MRS commands may be issued after t_{MRD} (at T_{d1}).

Figure 47: Write Leveling Exit Procedure



Note: 1. The DQ result, = 1, between T_{a0} and T_{c0} , is a result of the DQS, DQS# signals capturing CK HIGH just after the T_0 state.

Initialization

The following sequence is required for power-up and initialization, as shown in Figure 48 (page 131):

1. Apply power. RESET# is recommended to be below $0.2 \times V_{DDQ}$ during power ramp to ensure the outputs remain disabled (High-Z) and ODT off (R_{TT} is also High-Z). All other inputs, including ODT, may be undefined.

During power-up, either of the following conditions may exist and must be met:

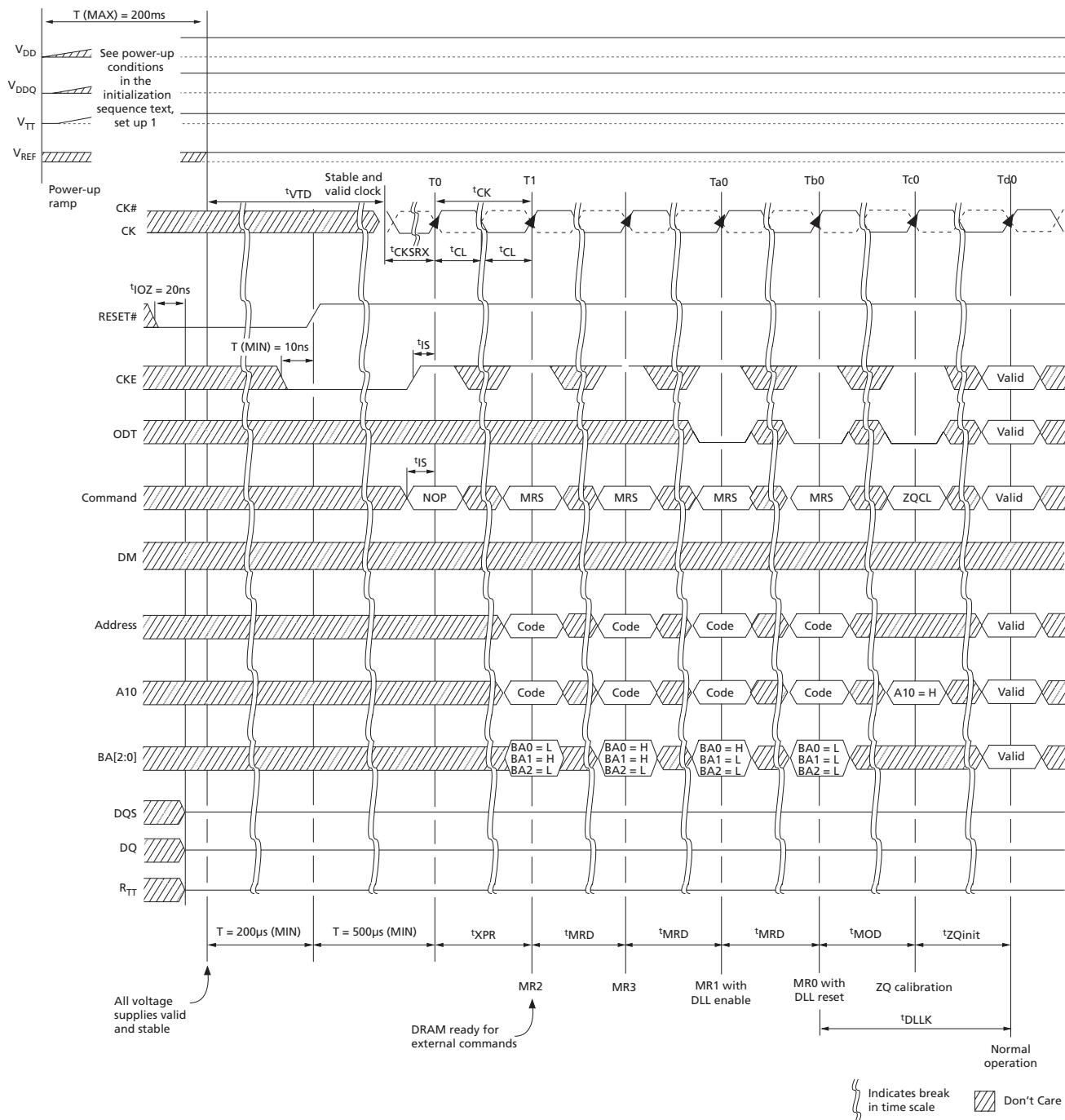
- Condition A:

- V_{DD} and V_{DDQ} are driven from a single-power converter output and are ramped with a maximum delta voltage between them of $\Delta V \leq 300\text{mV}$. Slope reversal of any power supply signal is allowed. The voltage levels on all balls other than V_{DD} , V_{DDQ} , V_{SS} , V_{SSQ} must be less than or equal to V_{DDQ} and V_{DD} on one side, and must be greater than or equal to V_{SSQ} and V_{SS} on the other side.
- Both V_{DD} and V_{DDQ} power supplies ramp to $V_{DD,\min}$ and $V_{DDQ,\min}$ within $t_{V_{DDPR}} = 200\text{ms}$.
- V_{REFDQ} tracks $V_{DD} \times 0.5$, V_{REFCA} tracks $V_{DD} \times 0.5$.
- V_{TT} is limited to 0.95V when the power ramp is complete and is not applied directly to the device; however, t_{VTD} should be greater than or equal to 0 to avoid device latchup.

- Condition B:

- V_{DD} may be applied before or at the same time as V_{DDQ} .
- V_{DDQ} may be applied before or at the same time as V_{TT} , V_{REFDQ} , and V_{REFCA} .
- No slope reversals are allowed in the power supply ramp for this condition.

2. Until stable power, maintain RESET# LOW to ensure the outputs remain disabled (High-Z). After the power is stable, RESET# must be LOW for at least 200 μs to begin the initialization process. ODT will remain in the High-Z state while RESET# is LOW and until CKE is registered HIGH.
3. CKE must be LOW 10ns prior to RESET# transitioning HIGH.
4. After RESET# transitions HIGH, wait 500 μs (minus one clock) with CKE LOW.
5. After the CKE LOW time, CKE may be brought HIGH (synchronously) and only NOP or DES commands may be issued. The clock must be present and valid for at least 10ns (and a minimum of five clocks) and ODT must be driven LOW at least t_{IS} prior to CKE being registered HIGH. When CKE is registered HIGH, it must be continuously registered HIGH until the full initialization process is complete.
6. After CKE is registered HIGH and after t_{XPR} has been satisfied, MRS commands may be issued. Issue an MRS (LOAD MODE) command to MR2 with the applicable settings (provide LOW to BA2 and BA0 and HIGH to BA1).
7. Issue an MRS command to MR3 with the applicable settings.
8. Issue an MRS command to MR1 with the applicable settings, including enabling the DLL and configuring ODT.
9. Issue an MRS command to MR0 with the applicable settings, including a DLL RESET command. t_{DLLK} (512) cycles of clock input are required to lock the DLL.
10. Issue a ZQCL command to calibrate R_{TT} and R_{ON} values for the process voltage temperature (PVT). Prior to normal operation, t_{ZQinit} must be satisfied.
11. When t_{DLLK} and t_{ZQinit} have been satisfied, the DDR3 SDRAM will be ready for normal operation.

Figure 48: Initialization Sequence


Voltage Initialization / Change

If the SDRAM is powered up and initialized for the 1.35V operating voltage range, voltage can be increased to the 1.5V operating range provided the following conditions are met (See Figure 49 (page 133)):

- Just prior to increasing the 1.35V operating voltages, no further commands are issued, other than NOPs or COMMAND INHIBITS, and all banks are in the precharge state.
- The 1.5V operating voltages are stable prior to issuing new commands, other than NOPs or COMMAND INHIBITS.
- The DLL is reset and relocked after the 1.5V operating voltages are stable and prior to any READ command.
- The ZQ calibration is performed. t_{ZQinit} must be satisfied after the 1.5V operating voltages are stable and prior to any READ command.

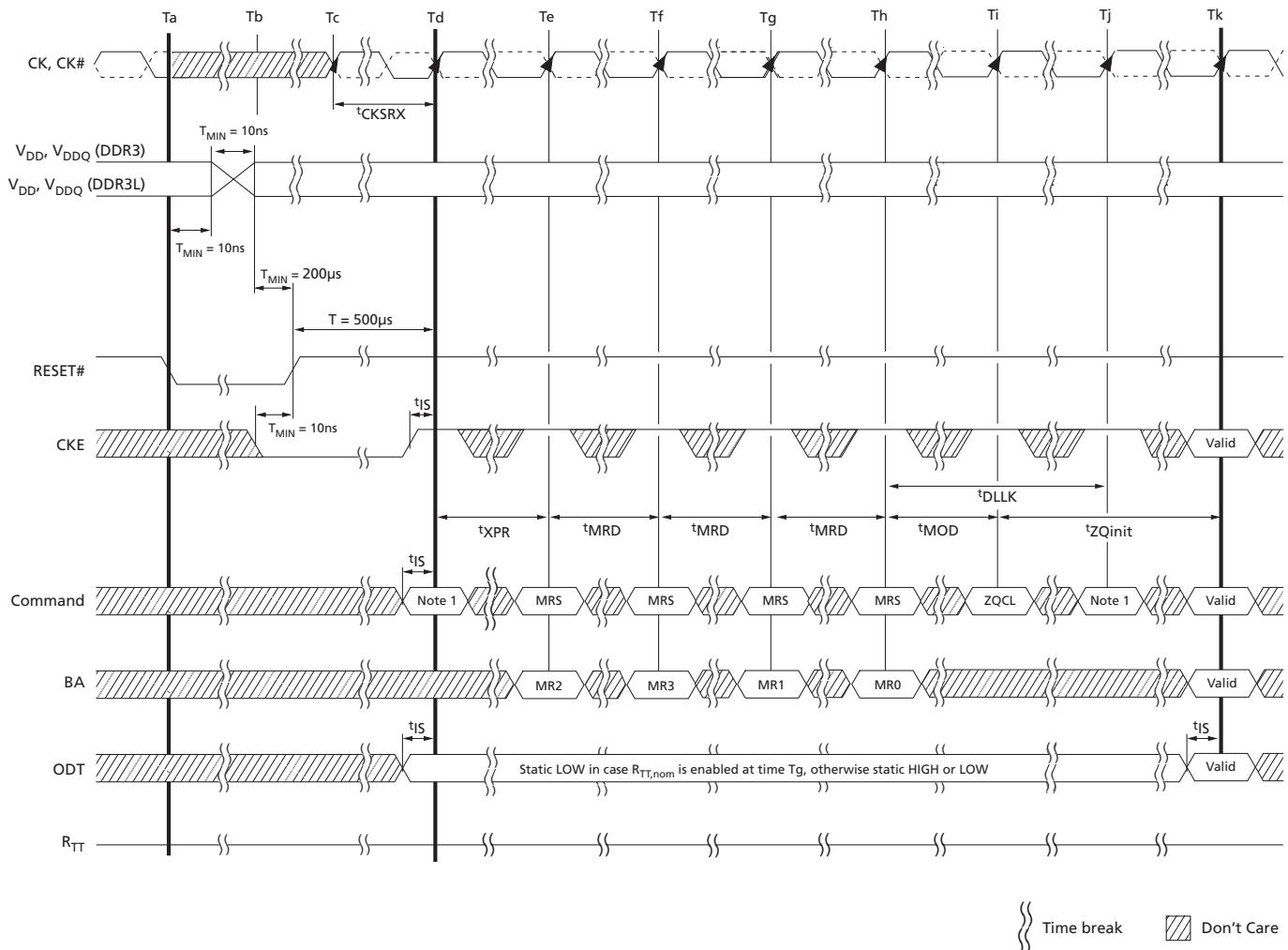
If the SDRAM is powered up and initialized for the 1.5V operating voltage range, voltage can be reduced to the 1.35V operation range provided the following conditions are met (See Figure 49 (page 133)) :

- Just prior to reducing the 1.5V operating voltages, no further commands are issued, other than NOPs or COMMAND INHIBITS, and all banks are in the precharge state.
- The 1.35V operating voltages are stable prior to issuing new commands, other than NOPs or COMMAND INHIBITS.
- The DLL is reset and relocked after the 1.35V operating voltages are stable and prior to any READ command.
- The ZQ calibration is performed. t_{ZQinit} must be satisfied after the 1.35V operating voltages are stable and prior to any READ command.

V_{DD} Voltage Switching

After the DDR3L DRAM is powered up and initialized, the power supply can be altered between the DDR3L and DDR3 levels, provided the sequence in Figure 49 is maintained.

Figure 49: V_{DD} Voltage Switching



Note: 1. From time point Td until Tk, NOP or DES commands must be applied between MRS and ZQCL commands.

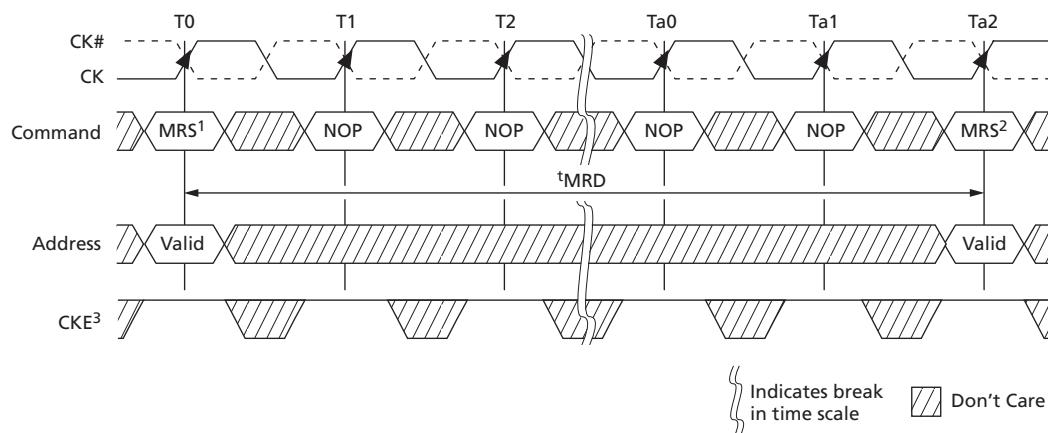
Mode Registers

Mode registers (MR0–MR3) are used to define various modes of programmable operations of the DDR3 SDRAM. A mode register is programmed via the mode register set (MRS) command during initialization, and it retains the stored information (except for MR0[8], which is self-clearing) until it is reprogrammed, RESET# goes LOW, the device loses power.

Contents of a mode register can be altered by re-executing the MRS command. Even if the user wants to modify only a subset of the mode register's variables, all variables must be programmed when the MRS command is issued. Reprogramming the mode register will not alter the contents of the memory array, provided it is performed correctly.

The MRS command can only be issued (or re-issued) when all banks are idle and in the precharged state (t_{RP} is satisfied and no data bursts are in progress). After an MRS command has been issued, two parameters must be satisfied: t_{MRD} and t_{MOD} . The controller must wait t_{MRD} before initiating any subsequent MRS commands.

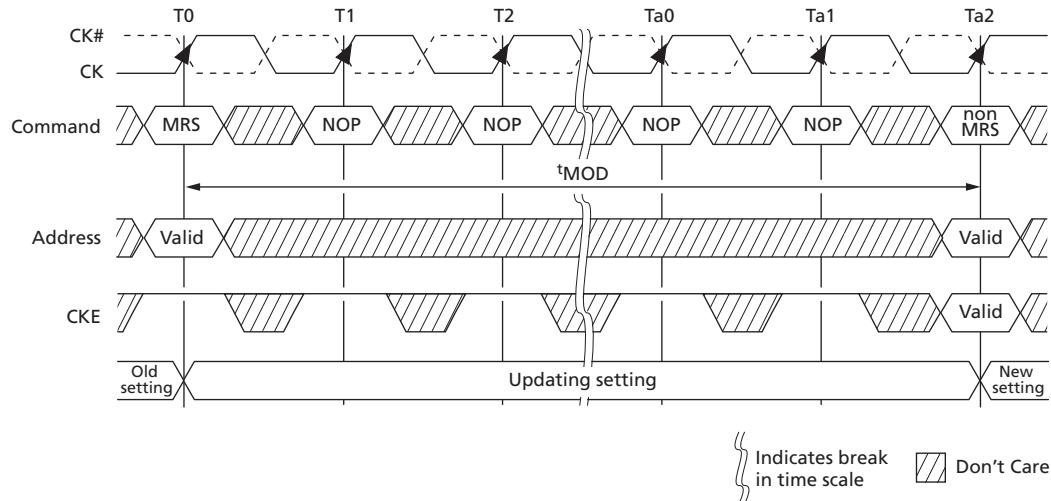
Figure 50: MRS to MRS Command Timing (t_{MRD})



Notes:

1. Prior to issuing the MRS command, all banks must be idle and precharged, t_{RP} (MIN) must be satisfied, and no data bursts can be in progress.
2. t_{MRD} specifies the MRS to MRS command minimum cycle time.
3. CKE must be registered HIGH from the MRS command until $t_{MRSPDEN}$ (MIN) (see Power-Down Mode (page 181)).
4. For a CAS latency change, $t_{XP DLL}$ timing must be met before any non-MRS command.

The controller must also wait t_{MOD} before initiating any non-MRS commands (excluding NOP and DES). The DRAM requires t_{MOD} in order to update the requested features, with the exception of DLL RESET, which requires additional time. Until t_{MOD} has been satisfied, the updated features are to be assumed unavailable.

Figure 51: MRS to nonMRS Command Timing (tMOD)


Notes:

1. Prior to issuing the MRS command, all banks must be idle (they must be precharged, tRP must be satisfied, and no data bursts can be in progress).
2. Prior to Ta2 when tMOD (MIN) is being satisfied, no commands (except NOP/DES) may be issued.
3. If R_{TT} was previously enabled, ODT must be registered LOW at T0 so that ODTL is satisfied prior to Ta1. ODT must also be registered LOW at each rising CK edge from T0 until tMODmin is satisfied at Ta2.
4. CKE must be registered HIGH from the MRS command until tMRSPDEN (MIN), at which time power-down may occur (see Power-Down Mode (page 181)).

Mode Register 0 (MR0)

The base register, mode register 0 (MR0), is used to define various DDR3 SDRAM modes of operation. These definitions include the selection of a burst length, burst type, CAS latency, operating mode, DLL RESET, write recovery, and precharge power-down mode (see Figure 52 (page 136)).

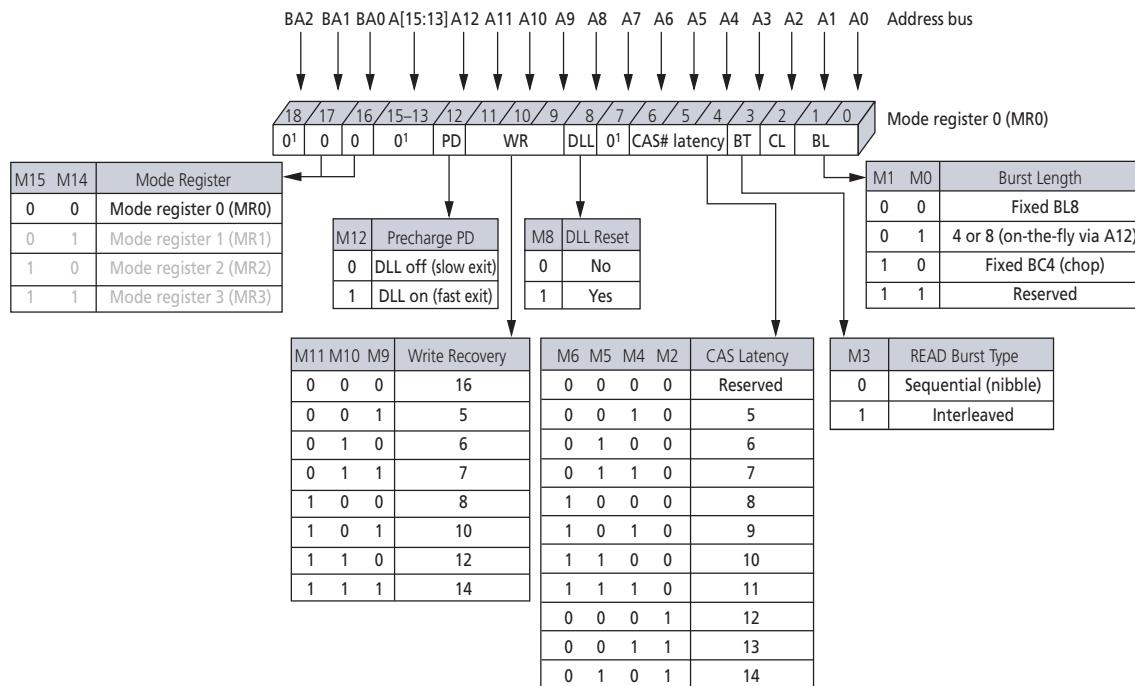
Burst Length

Burst length is defined by MR0[1:0]. Read and write accesses to the DDR3 SDRAM are burst-oriented, with the burst length being programmable to 4 (chop) mode, 8 (fixed) mode, or selectable using A12 during a READ/WRITE command (on-the-fly). The burst length determines the maximum number of column locations that can be accessed for a given READ or WRITE command. When MR0[1:0] is set to 01 during a READ/WRITE command, if A12 = 0, then BC4 mode is selected. If A12 = 1, then BL8 mode is selected. Specific timing diagrams, and turnaround between READ/WRITE, are shown in the READ/WRITE sections of this document.

When a READ or WRITE command is issued, a block of columns equal to the burst length is effectively selected. All accesses for that burst take place within this block, meaning that the burst will wrap within the block if a boundary is reached. The block is uniquely selected by A[i:2] when the burst length is set to 4 and by A[i:3] when the burst length is set to 8, where A_i is the most significant column address bit for a given configuration. The remaining (least significant) address bit(s) is (are) used to select the start-

ing location within the block. The programmed burst length applies to both READ and WRITE bursts.

Figure 52: Mode Register 0 (MR0) Definitions



Note: 1. MR0[18, 15:13, 7] are reserved for future use and must be programmed to 0.

Burst Type

Accesses within a given burst can be programmed to either a sequential or an interleaved order. The burst type is selected via MR0[3] (see Figure 52 (page 136)). The ordering of accesses within a burst is determined by the burst length, the burst type, and the starting column address. DDR3 only supports 4-bit burst chop and 8-bit burst access modes. Full interleave address ordering is supported for READs, while WRITEs are restricted to nibble (BC4) or word (BL8) boundaries.

Table 75: Burst Order

Burst Length	READ/ WRITE	Starting Column Address (A[2, 1, 0])	Burst Type = Sequential (Decimal)	Burst Type = Interleaved (Decimal)	Notes
4 (chop)	READ	0 0 0	0, 1, 2, 3, Z, Z, Z, Z	0, 1, 2, 3, Z, Z, Z, Z	1, 2
		0 0 1	1, 2, 3, 0, Z, Z, Z, Z	1, 0, 3, 2, Z, Z, Z, Z	1, 2
		0 1 0	2, 3, 0, 1, Z, Z, Z, Z	2, 3, 0, 1, Z, Z, Z, Z	1, 2
		0 1 1	3, 0, 1, 2, Z, Z, Z, Z	3, 2, 1, 0, Z, Z, Z, Z	1, 2
		1 0 0	4, 5, 6, 7, Z, Z, Z, Z	4, 5, 6, 7, Z, Z, Z, Z	1, 2
		1 0 1	5, 6, 7, 4, Z, Z, Z, Z	5, 4, 7, 6, Z, Z, Z, Z	1, 2
		1 1 0	6, 7, 4, 5, Z, Z, Z, Z	6, 7, 4, 5, Z, Z, Z, Z	1, 2
		1 1 1	7, 4, 5, 6, Z, Z, Z, Z	7, 6, 5, 4, Z, Z, Z, Z	1, 2
8 (fixed)	WRITE	0 V V	0, 1, 2, 3, X, X, X, X	0, 1, 2, 3, X, X, X, X	1, 3, 4
		1 V V	4, 5, 6, 7, X, X, X, X	4, 5, 6, 7, X, X, X, X	1, 3, 4
8 (fixed)	READ	0 0 0	0, 1, 2, 3, 4, 5, 6, 7	0, 1, 2, 3, 4, 5, 6, 7	1
		0 0 1	1, 2, 3, 0, 5, 6, 7, 4	1, 0, 3, 2, 5, 4, 7, 6	1
		0 1 0	2, 3, 0, 1, 6, 7, 4, 5	2, 3, 0, 1, 6, 7, 4, 5	1
		0 1 1	3, 0, 1, 2, 7, 4, 5, 6	3, 2, 1, 0, 7, 6, 5, 4	1
		1 0 0	4, 5, 6, 7, 0, 1, 2, 3	4, 5, 6, 7, 0, 1, 2, 3	1
		1 0 1	5, 6, 7, 4, 1, 2, 3, 0	5, 4, 7, 6, 1, 0, 3, 2	1
		1 1 0	6, 7, 4, 5, 2, 3, 0, 1	6, 7, 4, 5, 2, 3, 0, 1	1
		1 1 1	7, 4, 5, 6, 3, 0, 1, 2	7, 6, 5, 4, 3, 2, 1, 0	1
8 (fixed)	WRITE	V V V	0, 1, 2, 3, 4, 5, 6, 7	0, 1, 2, 3, 4, 5, 6, 7	1, 3

Notes:

1. Internal READ and WRITE operations start at the same point in time for BC4 as they do for BL8.
2. Z = Data and strobe output drivers are in tri-state.
3. V = A valid logic level (0 or 1), but the respective input buffer ignores level-on input pins.
4. X = "Don't Care."

DLL RESET

DLL RESET is defined by MR0[8] (see Figure 52 (page 136)). Programming MR0[8] to 1 activates the DLL RESET function. MR0[8] is self-clearing, meaning it returns to a value of 0 after the DLL RESET function has been initiated.

Anytime the DLL RESET function is initiated, CKE must be HIGH and the clock held stable for 512 (^tDLLK) clock cycles before a READ command can be issued. This is to allow time for the internal clock to be synchronized with the external clock. Failing to wait for synchronization can result in invalid output timing specifications, such as ^tDQSK timings.

Write Recovery

WRITE recovery time is defined by MR0[11:9] (see Figure 52 (page 136)). Write recovery values of 5, 6, 7, 8, 10, or 12 can be used by programming MR0[11:9]. The user is re-

quired to program the correct value of write recovery, which is calculated by dividing t_{WR} (ns) by t_{CK} (ns) and rounding up a noninteger value to the next integer:
 WR (cycles) = roundup (t_{WR} (ns) / t_{CK} (ns)).

Precharge Power-Down (Precharge PD)

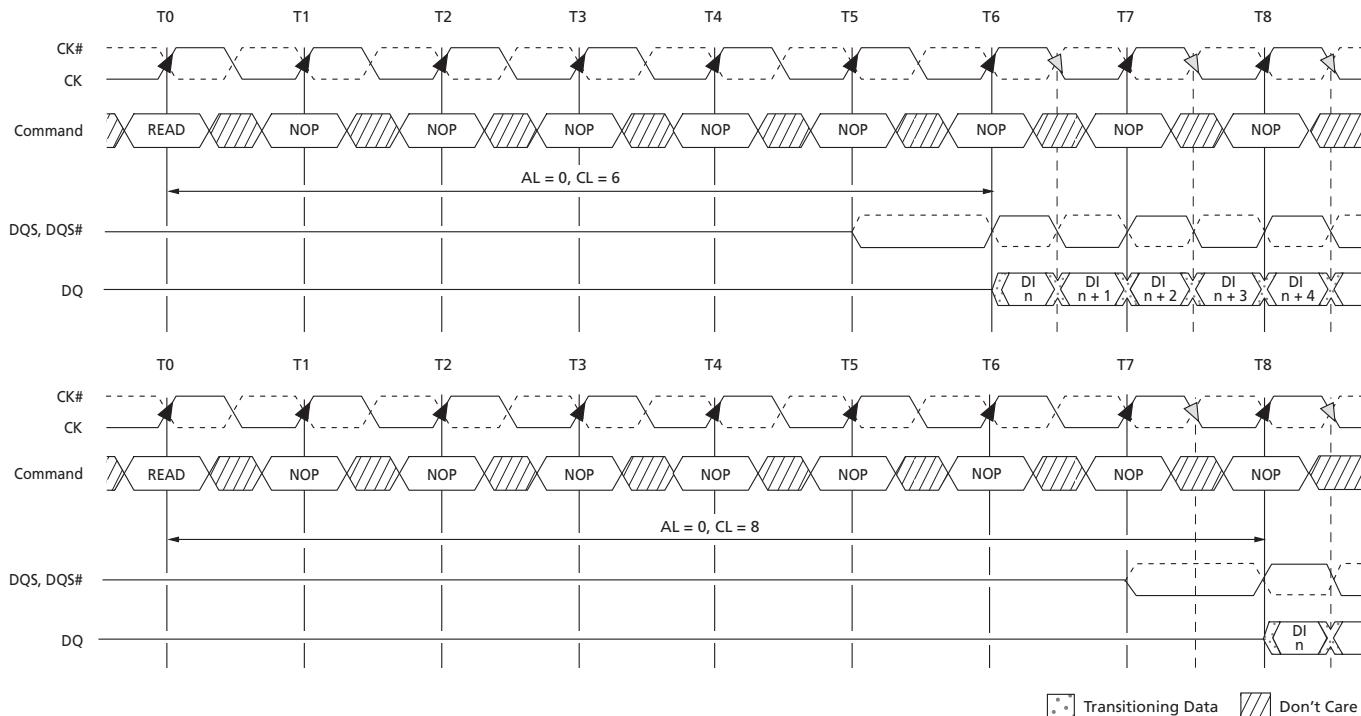
The precharge power-down (precharge PD) bit applies only when precharge power-down mode is being used. When MR0[12] is set to 0, the DLL is off during precharge power-down, providing a lower standby current mode; however, $t_{XP DLL}$ must be satisfied when exiting. When MR0[12] is set to 1, the DLL continues to run during precharge power-down mode to enable a faster exit of precharge power-down mode; however, t_{XP} must be satisfied when exiting (see Power-Down Mode (page 181)).

CAS Latency (CL)

CAS latency (CL) is defined by MR0[6:4], as shown in Figure 52 (page 136). CAS latency is the delay, in clock cycles, between the internal READ command and the availability of the first bit of output data. CL can be set to 5 through 14. DDR3 SDRAM do not support half-clock latencies.

Examples of CL = 6 and CL = 8 are shown below. If an internal READ command is registered at clock edge n , and the CAS latency is m clocks, the data will be available nominally coincident with clock edge $n + m$. See Speed Bin Tables for the CLs supported at various operating frequencies.

Figure 53: READ Latency



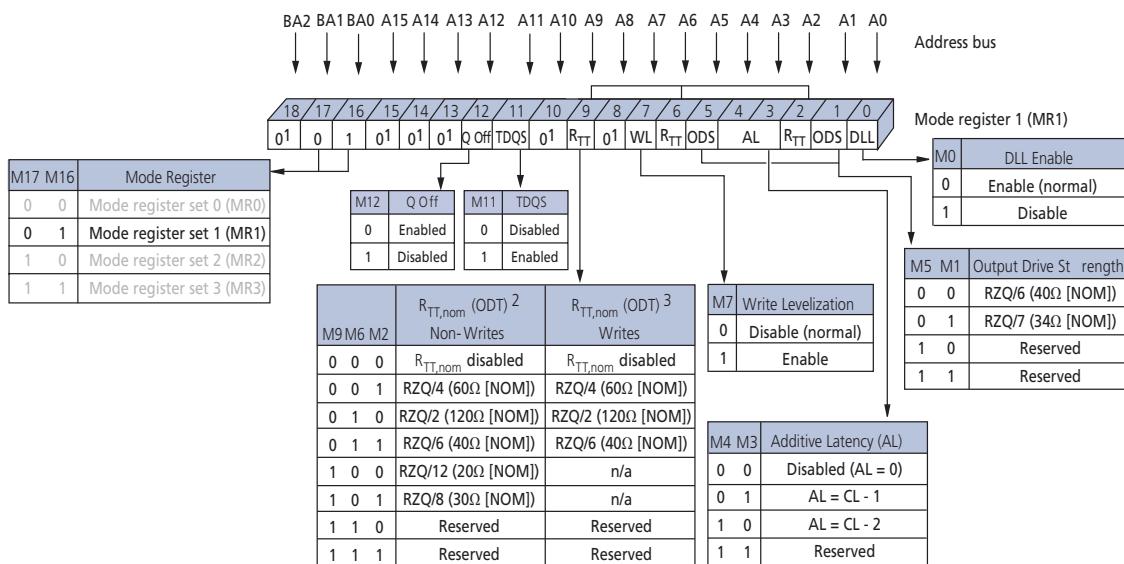
Notes: 1. For illustration purposes, only CL = 6 and CL = 8 are shown. Other CL values are possible.
2. Shown with nominal t_{DQSCK} and nominal t_{DSDQ} .

Mode Register 1 (MR1)

The mode register 1 (MR1) controls additional functions and features not available in the other mode registers: Q OFF (OUTPUT DISABLE), TDQS (for the x8 configuration only), DLL ENABLE/DLL DISABLE, $R_{TT,nom}$ value (ODT), WRITE LEVELING, POSTED CAS ADDITIVE latency, and OUTPUT DRIVE STRENGTH. These functions are controlled via the bits shown in Figure 54 (page 139). The MR1 register is programmed via the MRS command and retains the stored information until it is reprogrammed, until RESET# goes LOW, or until the device loses power. Reprogramming the MR1 register will not alter the contents of the memory array, provided it is performed correctly.

The MR1 register must be loaded when all banks are idle and no bursts are in progress. The controller must satisfy the specified timing parameters t_{MRD} and t_{MOD} before initiating a subsequent operation.

Figure 54: Mode Register 1 (MR1) Definition



Notes:

1. MR1[18, 15:13, 10, 8] are reserved for future use and must be programmed to 0.
2. During write leveling, if MR1[7] and MR1[12] are 1, then all $R_{TT,nom}$ values are available for use.
3. During write leveling, if MR1[7] is a 1, but MR1[12] is a 0, then only $R_{TT,nom}$ write values are available for use.

DLL Enable/DLL Disable

The DLL may be enabled or disabled by programming MR1[0] during the LOAD MODE command, as shown in Figure 54 (page 139). The DLL must be enabled for normal operation. DLL enable is required during power-up initialization and upon returning to normal operation after having disabled the DLL for the purpose of debugging or evaluation. Enabling the DLL should always be followed by resetting the DLL using the appropriate LOAD MODE command.

If the DLL is enabled prior to entering self refresh mode, the DLL is automatically disabled when entering SELF REFRESH operation and is automatically reenabled and reset

upon exit of SELF REFRESH operation. If the DLL is disabled prior to entering self refresh mode, the DLL remains disabled even upon exit of SELF REFRESH operation until it is reenabled and reset.

The DRAM is not tested to check—nor does Micron warrant compliance with—normal mode timings or functionality when the DLL is disabled. An attempt has been made to have the DRAM operate in the normal mode where reasonably possible when the DLL has been disabled; however, by industry standard, a few known exceptions are defined:

- ODT is not allowed to be used
- The output data is no longer edge-aligned to the clock
- CL and CWL can only be six clocks

When the DLL is disabled, timing and functionality can vary from the normal operation specifications when the DLL is enabled (see DLL Disable Mode (page 119)). Disabling the DLL also implies the need to change the clock frequency (see Input Clock Frequency Change (page 123)).

Output Drive Strength

The DDR3 SDRAM uses a programmable impedance output buffer. The drive strength mode register setting is defined by MR1[5, 1]. RZQ/7 (34Ω [NOM]) is the primary output driver impedance setting for DDR3 SDRAM devices. To calibrate the output driver impedance, an external precision resistor (RZQ) is connected between the ZQ ball and VSSQ. The value of the resistor must be 240Ω ±1%.

The output impedance is set during initialization. Additional impedance calibration updates do not affect device operation, and all data sheet timings and current specifications are met during an update.

To meet the 34Ω specification, the output drive strength must be set to 34Ω during initialization. To obtain a calibrated output driver impedance after power-up, the DDR3 SDRAM needs a calibration command that is part of the initialization and reset procedure.

OUTPUT ENABLE/DISABLE

The OUTPUT ENABLE function is defined by MR1[12], as shown in Figure 54 (page 139). When enabled (MR1[12] = 0), all outputs (DQ, DQS, DQS#) function when in the normal mode of operation. When disabled (MR1[12] = 1), all DDR3 SDRAM outputs (DQ and DQS, DQS#) are tri-stated. The output disable feature is intended to be used during I_{DD} characterization of the READ current and during t_{DQSS} margining (write leveling) only.

TDQS Enable

Termination data strobe (TDQS) is a feature of the x8 DDR3 SDRAM configuration that provides termination resistance (R_{TT}) and may be useful in some system configurations. TDQS is not supported in x4 or x16 configurations. When enabled via the mode register (MR1[11]), the R_{TT} that is applied to DQS and DQS# is also applied to TDQS and TDQS#. In contrast to the RDQS function of DDR2 SDRAM, DDR3's TDQS provides the termination resistance R_{TT} only. The OUTPUT DATA STROBE function of RDQS is not provided by TDQS; thus, R_{ON} does not apply to TDQS and TDQS#. The TDQS and DM functions share the same ball. When the TDQS function is enabled via the mode register, the DM function is not supported. When the TDQS function is disabled, the DM function is pro-

vided, and the TDQS# ball is not used. The TDQS function is available in the x8 DDR3 SDRAM configuration only and must be disabled via the mode register for the x4 and x16 configurations.

On-Die Termination

ODT resistance $R_{TT,nom}$ is defined by MR1[9, 6, 2] (see Figure 54 (page 139)). The R_{TT} termination value applies to the DQ, DM, DQS, DQS#, and TDQS, TDQS# balls. DDR3 supports multiple R_{TT} termination values based on RZQ/n where n can be 2, 4, 6, 8, or 12 and RZQ is 240Ω .

Unlike DDR2, DDR3 ODT must be turned off prior to reading data out and must remain off during a READ burst. $R_{TT,nom}$ termination is allowed any time after the DRAM is initialized, calibrated, and not performing read access, or when it is not in self refresh mode. Additionally, write accesses with dynamic ODT enabled ($R_{TT(WR)}$) temporarily replaces $R_{TT,nom}$ with $R_{TT(WR)}$.

The actual effective termination, $R_{TT(EFF)}$, may be different from the R_{TT} targeted due to nonlinearity of the termination. For $R_{TT(EFF)}$ values and calculations (see On-Die Termination (ODT) (page 191)).

The ODT feature is designed to improve signal integrity of the memory channel by enabling the DDR3 SDRAM controller to independently turn on/off ODT for any or all devices. The ODT input control pin is used to determine when R_{TT} is turned on (ODTL on) and off (ODTL off), assuming ODT has been enabled via MR1[9, 6, 2].

Timings for ODT are detailed in On-Die Termination (ODT) (page 191).

WRITE LEVELING

The WRITE LEVELING function is enabled by MR1[7], as shown in Figure 54 (page 139). Write leveling is used (during initialization) to deskew the DQS strobe to clock offset as a result of fly-by topology designs. For better signal integrity, DDR3 SDRAM memory modules adopted fly-by topology for the commands, addresses, control signals, and clocks.

The fly-by topology benefits from a reduced number of stubs and their lengths. However, fly-by topology induces flight time skews between the clock and DQS strobe (and DQ) at each DRAM on the DIMM. Controllers will have a difficult time maintaining t_{DQSS} , t_{DSS} , and t_{DSH} specifications without supporting write leveling in systems which use fly-by topology-based modules. Write leveling timing and detailed operation information is provided in Write Leveling (page 125).

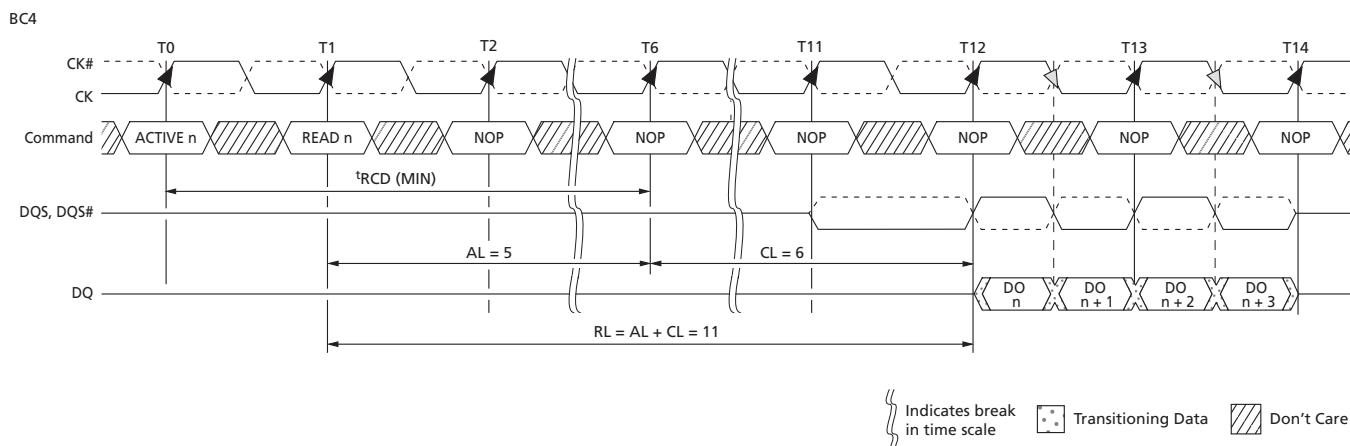
POSTED CAS ADDITIVE Latency

POSTED CAS ADDITIVE latency (AL) is supported to make the command and data bus efficient for sustainable bandwidths in DDR3 SDRAM. MR1[4, 3] define the value of AL, as shown in Figure 55 (page 142). MR1[4, 3] enable the user to program the DDR3 SDRAM with AL = 0, CL - 1, or CL - 2.

With this feature, the DDR3 SDRAM enables a READ or WRITE command to be issued after the ACTIVATE command for that bank prior to t_{RCD} (MIN). The only restriction is ACTIVATE to READ or WRITE + AL $\geq t_{RCD}$ (MIN) must be satisfied. Assuming t_{RCD} (MIN) = CL, a typical application using this feature sets AL = CL - 1 t_{CK} = t_{RCD} (MIN) - 1 t_{CK} . The READ or WRITE command is held for the time of the AL before it is released internally to the DDR3 SDRAM device. READ latency (RL) is controlled by the sum of

the AL and CAS latency (CL), $RL = AL + CL$. WRITE latency (WL) is the sum of CAS WRITE latency and AL, $WL = AL + CWL$ (see Mode Register 2 (MR2) (page 142)). Examples of READ and WRITE latencies are shown in Figure 55 (page 142) and Figure 57 (page 143).

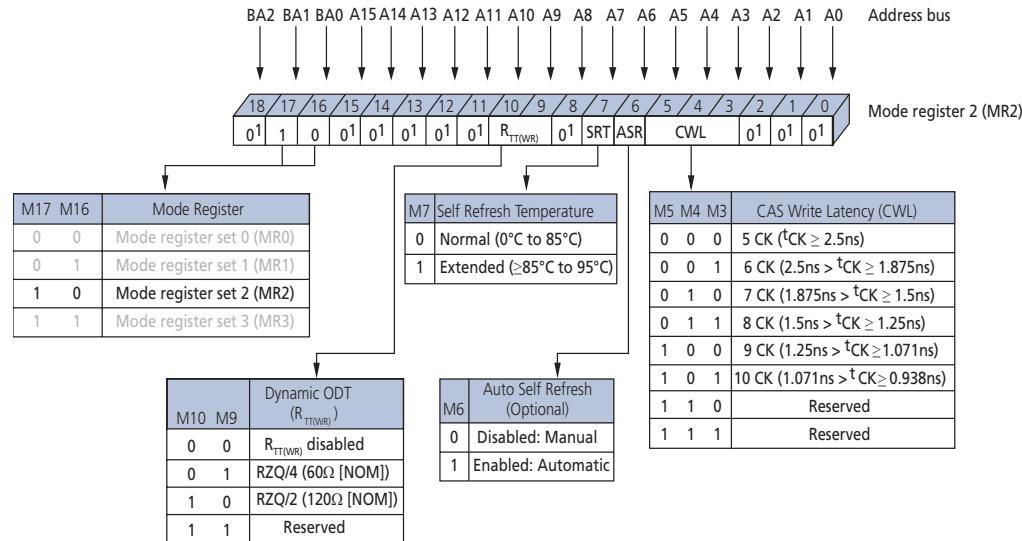
Figure 55: READ Latency (AL = 5, CL = 6)



Mode Register 2 (MR2)

The mode register 2 (MR2) controls additional functions and features not available in the other mode registers. These additional functions are CAS WRITE latency (CWL), AUTO SELF REFRESH (ASR), SELF REFRESH TEMPERATURE (SRT), and DYNAMIC ODT ($R_{TT(WR)}$). These functions are controlled via the bits shown in Figure 56. The MR2 is programmed via the MRS command and will retain the stored information until it is programmed again or until the device loses power. Reprogramming the MR2 register will not alter the contents of the memory array, provided it is performed correctly. The MR2 register must be loaded when all banks are idle and no data bursts are in progress, and the controller must wait the specified time t_{MRD} and t_{MOD} before initiating a subsequent operation.

Figure 56: Mode Register 2 (MR2) Definition

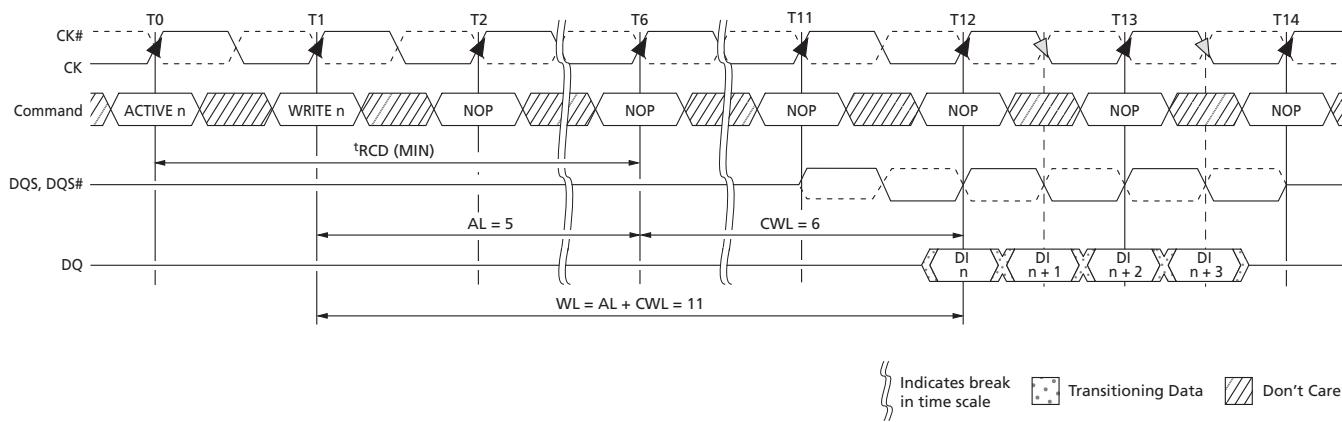


Note: 1. MR2[18, 15:11, 8, and 2:0] are reserved for future use and must all be programmed to 0.

CAS Write Latency (CWL)

CWL is defined by MR2[5:3] and is the delay, in clock cycles, from the releasing of the internal write to the latching of the first data in. CWL must be correctly set to the corresponding operating clock frequency (see Figure 56 (page 143)). The overall WRITE latency (WL) is equal to CWL + AL (Figure 54 (page 139)).

Figure 57: CAS Write Latency



AUTO SELF REFRESH (ASR)

Mode register MR2[6] is used to disable/enable the ASR function. When ASR is disabled, the self refresh mode's refresh rate is assumed to be at the normal 85°C limit (sometimes referred to as 1x refresh rate). In the disabled mode, ASR requires the user to en-

sure the DRAM never exceeds a T_C of 85°C while in self refresh unless the user enables the SRT feature listed below when the T_C is between 85°C and 95°C.

Enabling ASR assumes the DRAM self refresh rate is changed automatically from 1x to 2x when the case temperature exceeds 85°C. This enables the user to operate the DRAM beyond the standard 85°C limit up to the optional extended temperature range of 95°C while in self refresh mode.

The standard self refresh current test specifies test conditions to normal case temperature (85°C) only, meaning if ASR is enabled, the standard self refresh current specifications do not apply (see Extended Temperature Usage (page 180)).

SELF REFRESH TEMPERATURE (SRT)

Mode register MR2[7] is used to disable/enable the SRT function. When SRT is disabled, the self refresh mode's refresh rate is assumed to be at the normal 85°C limit (sometimes referred to as 1x refresh rate). In the disabled mode, SRT requires the user to ensure the DRAM never exceeds a T_C of 85°C while in self refresh mode unless the user enables ASR.

When SRT is enabled, the DRAM self refresh is changed internally from 1x to 2x, regardless of the case temperature. This enables the user to operate the DRAM beyond the standard 85°C limit up to the optional extended temperature range of 95°C while in self refresh mode. The standard self refresh current test specifies test conditions to normal case temperature (85°C) only, meaning if SRT is enabled, the standard self refresh current specifications do not apply (see Extended Temperature Usage (page 180)).

SRT vs. ASR

If the normal case temperature limit of 85°C is not exceeded, then neither SRT nor ASR is required, and both can be disabled throughout operation. However, if the extended temperature option of 95°C is needed, the user is required to provide a 2x refresh rate during (manual) refresh and to enable either the SRT or the ASR to ensure self refresh is performed at the 2x rate.

SRT forces the DRAM to switch the internal self refresh rate from 1x to 2x. Self refresh is performed at the 2x refresh rate regardless of the case temperature.

ASR automatically switches the DRAM's internal self refresh rate from 1x to 2x. However, while in self refresh mode, ASR enables the refresh rate to automatically adjust between 1x to 2x over the supported temperature range. One other disadvantage with ASR is the DRAM cannot always switch from a 1x to a 2x refresh rate at an exact case temperature of 85°C. Although the DRAM will support data integrity when it switches from a 1x to a 2x refresh rate, it may switch at a lower temperature than 85°C.

Since only one mode is necessary, SRT and ASR cannot be enabled at the same time.

DYNAMIC ODT

The dynamic ODT ($R_{TT(WR)}$) feature is defined by MR2[10, 9]. Dynamic ODT is enabled when a value is selected. This new DDR3 SDRAM feature enables the ODT termination value to change without issuing an MRS command, essentially changing the ODT termination on-the-fly.

With dynamic ODT ($R_{TT(WR)}$) enabled, the DRAM switches from normal ODT ($R_{TT,nom}$) to dynamic ODT ($R_{TT(WR)}$) when beginning a WRITE burst and subsequently switches

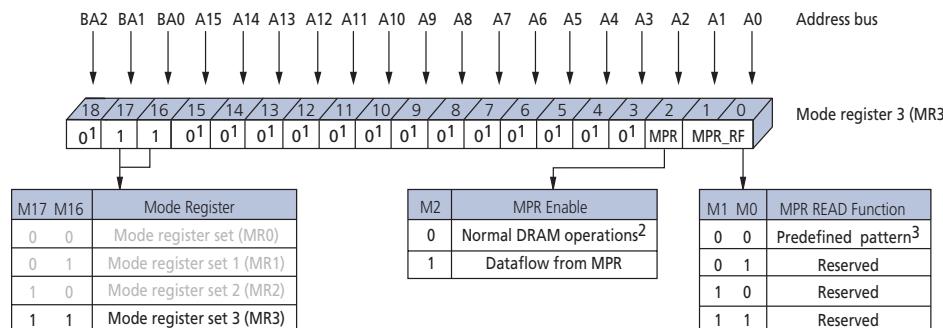
back to ODT ($R_{TT,nom}$) at the completion of the WRITE burst. If $R_{TT,nom}$ is disabled, the $R_{TT,nom}$ value will be High-Z. Special timing parameters must be adhered to when dynamic ODT ($R_{TT(WR)}$) is enabled: ODTLcnw, ODTLcnw4, ODTLcnw8, ODTH4, ODTH8, and t_{ADC} .

Dynamic ODT is only applicable during WRITE cycles. If ODT ($R_{TT,nom}$) is disabled, dynamic ODT ($R_{TT(WR)}$) is still permitted. $R_{TT,nom}$ and $R_{TT(WR)}$ can be used independent of one other. Dynamic ODT is not available during write leveling mode, regardless of the state of ODT ($R_{TT,nom}$). For details on dynamic ODT operation, refer to Dynamic ODT (page 193).

Mode Register 3 (MR3)

The mode register 3 (MR3) controls additional functions and features not available in the other mode registers. Currently defined is the MULTIPURPOSE REGISTER (MPR). This function is controlled via the bits shown in Figure 58 (page 145). The MR3 is programmed via the LOAD MODE command and retains the stored information until it is programmed again or until the device loses power. Reprogramming the MR3 register will not alter the contents of the memory array, provided it is performed correctly. The MR3 register must be loaded when all banks are idle and no data bursts are in progress, and the controller must wait the specified time t_{MRD} and t_{MOD} before initiating a subsequent operation.

Figure 58: Mode Register 3 (MR3) Definition



Notes:

1. MR3[18 and 15:3] are reserved for future use and must all be programmed to 0.
2. When MPR control is set for normal DRAM operation, MR3[1, 0] will be ignored.
3. Intended to be used for READ synchronization.

Multipurpose Register (MPR)

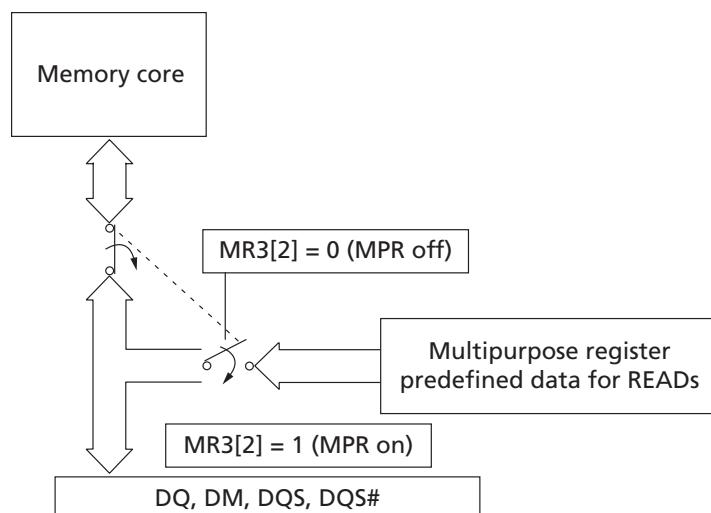
The MULTIPURPOSE REGISTER function is used to output a predefined system timing calibration bit sequence. Bit 2 is the master bit that enables or disables access to the MPR register, and bits 1 and 0 determine which mode the MPR is placed in. The basic concept of the multipurpose register is shown in Figure 59 (page 146).

If MR3[2] is a 0, then the MPR access is disabled, and the DRAM operates in normal mode. However, if MR3[2] is a 1, then the DRAM no longer outputs normal read data but outputs MPR data as defined by MR3[0, 1]. If MR3[0, 1] is equal to 00, then a predefined read pattern for system calibration is selected.

To enable the MPR, the MRS command is issued to MR3, and MR3[2] = 1. Prior to issuing the MRS command, all banks must be in the idle state (all banks are precharged,

and t_{RP} is met). When the MPR is enabled, any subsequent READ or RDAP commands are redirected to the multipurpose register. The resulting operation when either a READ or a RDAP command is issued, is defined by MR3[1:0] when the MPR is enabled (see Table 77 (page 147)). When the MPR is enabled, only READ or RDAP commands are allowed until a subsequent MRS command is issued with the MPR disabled (MR3[2] = 0). Power-down mode, self refresh, and any other nonREAD/RDAP commands are not allowed during MPR enable mode. The RESET function is supported during MPR enable mode.

Figure 59: Multipurpose Register (MPR) Block Diagram



Notes:

1. A predefined data pattern can be read out of the MPR with an external READ command.
2. MR3[2] defines whether the data flow comes from the memory core or the MPR. When the data flow is defined, the MPR contents can be read out continuously with a regular READ or RDAP command.

Table 76: MPR Functional Description of MR3 Bits

MR3[2]	MR3[1:0]	Function
MPR	MPR READ Function	Function
0	"Don't Care"	Normal operation, no MPR transaction All subsequent READs come from the DRAM memory array All subsequent WRITEs go to the DRAM memory array
1	A[1:0] (see Table 77 (page 147))	Enable MPR mode, subsequent READ/RDAP commands defined by bits 1 and 2

MPR Functional Description

The MPR JEDEC definition enables either a prime DQ (DQ0 on a x4 and a x8; on a x16, DQ0 = lower byte and DQ8 = upper byte) to output the MPR data with the remaining DQs driven LOW, or for all DQs to output the MPR data. The MPR readout supports fixed READ burst and READ burst chop (MRS and OTF via A12/BC#) with regular READ latencies and AC timings applicable, provided the DLL is locked as required.

MPR addressing for a valid MPR read is as follows:

- A[1:0] must be set to 00 as the burst order is fixed per nibble
- A2 selects the burst order:
 - BL8, A2 is set to 0, and the burst order is fixed to 0, 1, 2, 3, 4, 5, 6, 7
- For burst chop 4 cases, the burst order is switched on the nibble base along with the following:
 - A2 = 0; burst order = 0, 1, 2, 3
 - A2 = 1; burst order = 4, 5, 6, 7
- Burst order bit 0 (the first bit) is assigned to LSB, and burst order bit 7 (the last bit) is assigned to MSB
- A[9:3] are a “Don’t Care”
- A10 is a “Don’t Care”
- A11 is a “Don’t Care”
- A12: Selects burst chop mode on-the-fly, if enabled within MR0
- A13 is a “Don’t Care”
- BA[2:0] are a “Don’t Care”

MPR Register Address Definitions and Bursting Order

The MPR currently supports a single data format. This data format is a predefined read pattern for system calibration. The predefined pattern is always a repeating 0-1 bit pattern.

Examples of the different types of predefined READ pattern bursts are shown in the following figures.

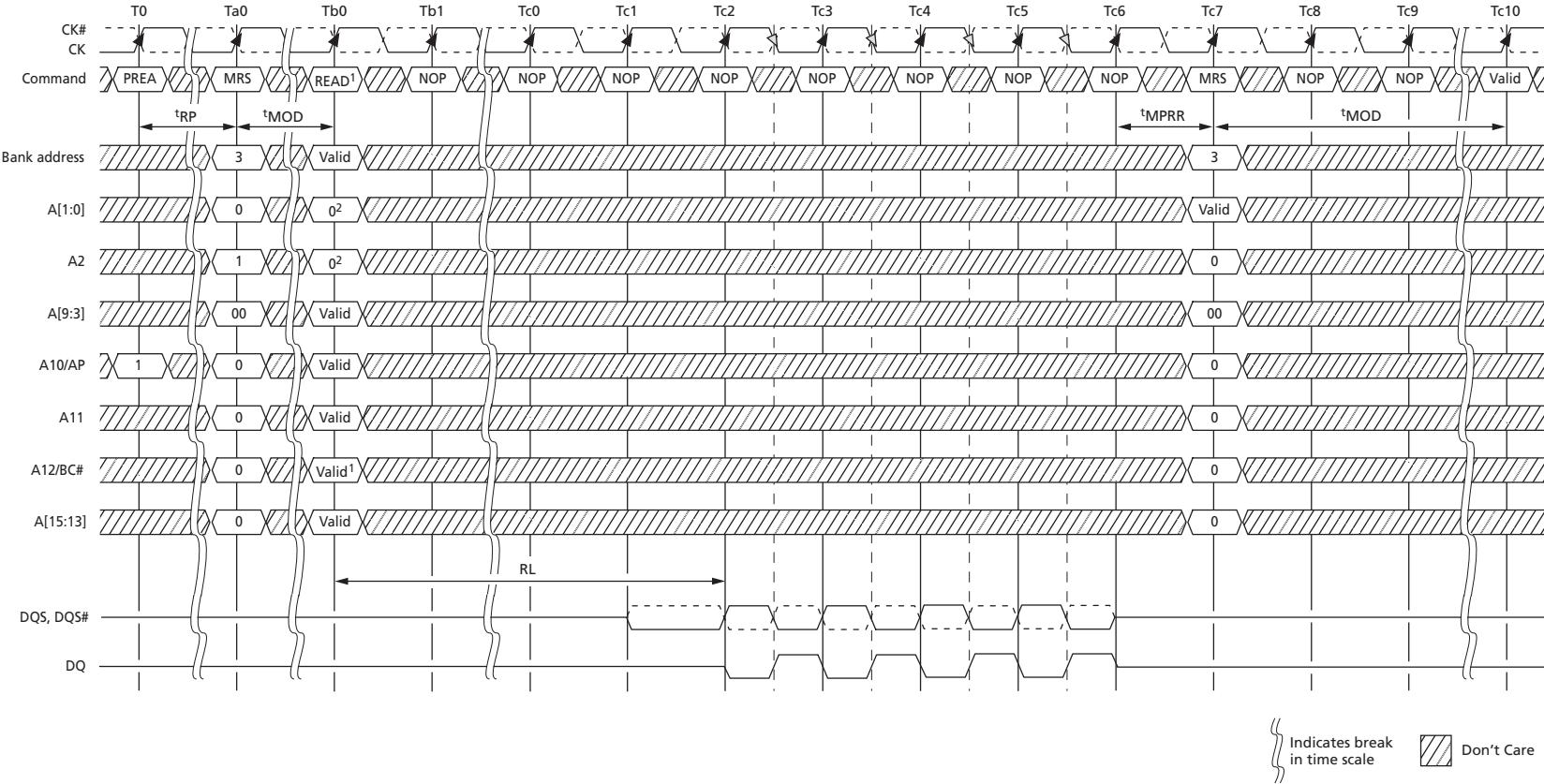
Table 77: MPR Readouts and Burst Order Bit Mapping

MR3[2]	MR3[1:0]	Function	Burst Length	Read A[2:0]	Burst Order and Data Pattern
1	00	READ predefined pattern for system calibration	BL8	000	Burst order: 0, 1, 2, 3, 4, 5, 6, 7 Predefined pattern: 0, 1, 0, 1, 0, 1, 0, 1
			BC4	000	Burst order: 0, 1, 2, 3 Predefined pattern: 0, 1, 0, 1
			BC4	100	Burst order: 4, 5, 6, 7 Predefined pattern: 0, 1, 0, 1
1	01	RFU	N/A	N/A	N/A
			N/A	N/A	N/A
			N/A	N/A	N/A
1	10	RFU	N/A	N/A	N/A
			N/A	N/A	N/A
			N/A	N/A	N/A

Table 77: MPR Readouts and Burst Order Bit Mapping (Continued)

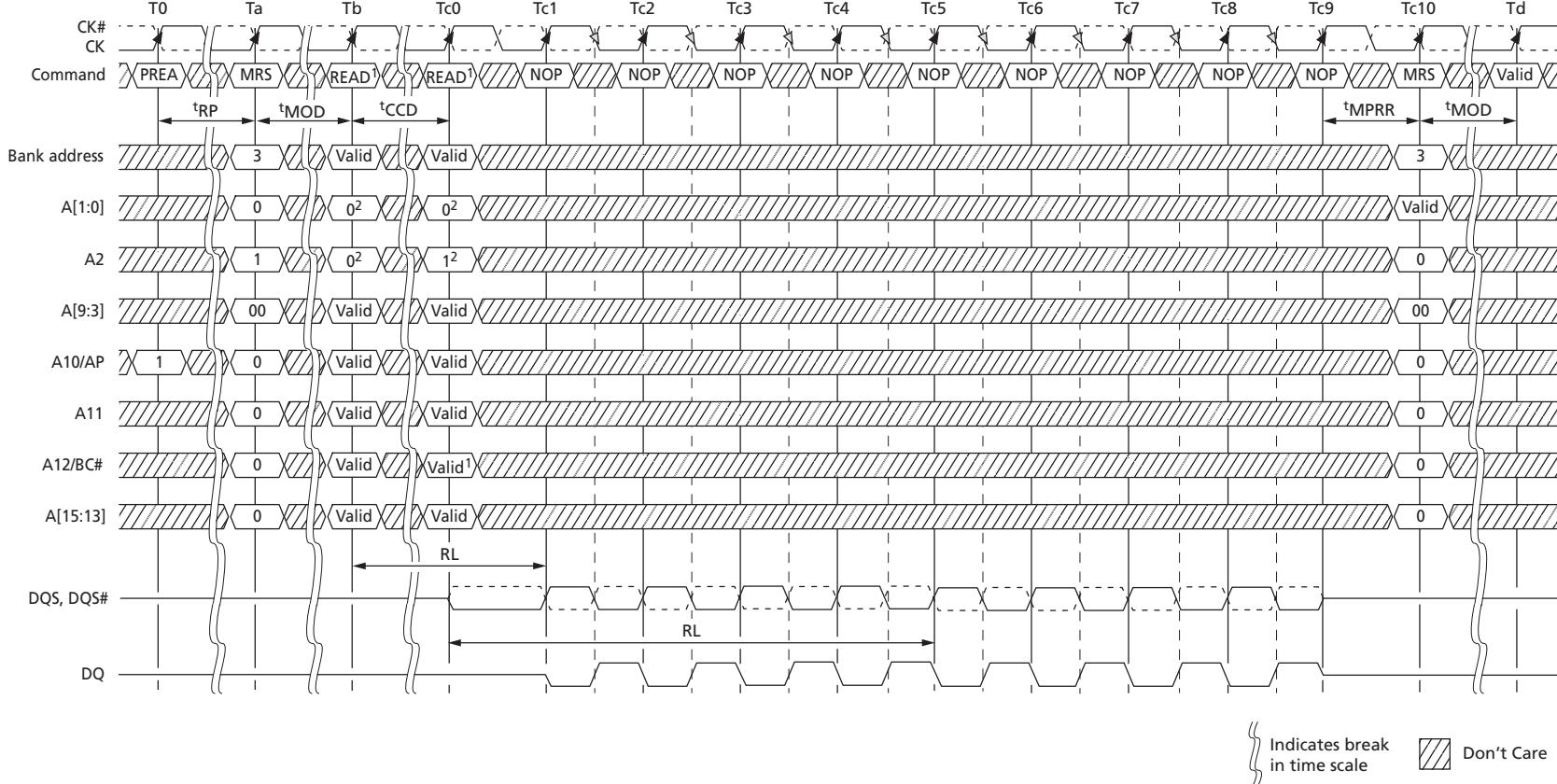
MR3[2]	MR3[1:0]	Function	Burst Length	Read A[2:0]	Burst Order and Data Pattern
1	11	RFU	N/A	N/A	N/A
			N/A	N/A	N/A
			N/A	N/A	N/A

Note: 1. Burst order bit 0 is assigned to LSB, and burst order bit 7 is assigned to MSB of the selected MPR agent.

Figure 60: MPR System Read Calibration with BL8: Fixed Burst Order Single Readout


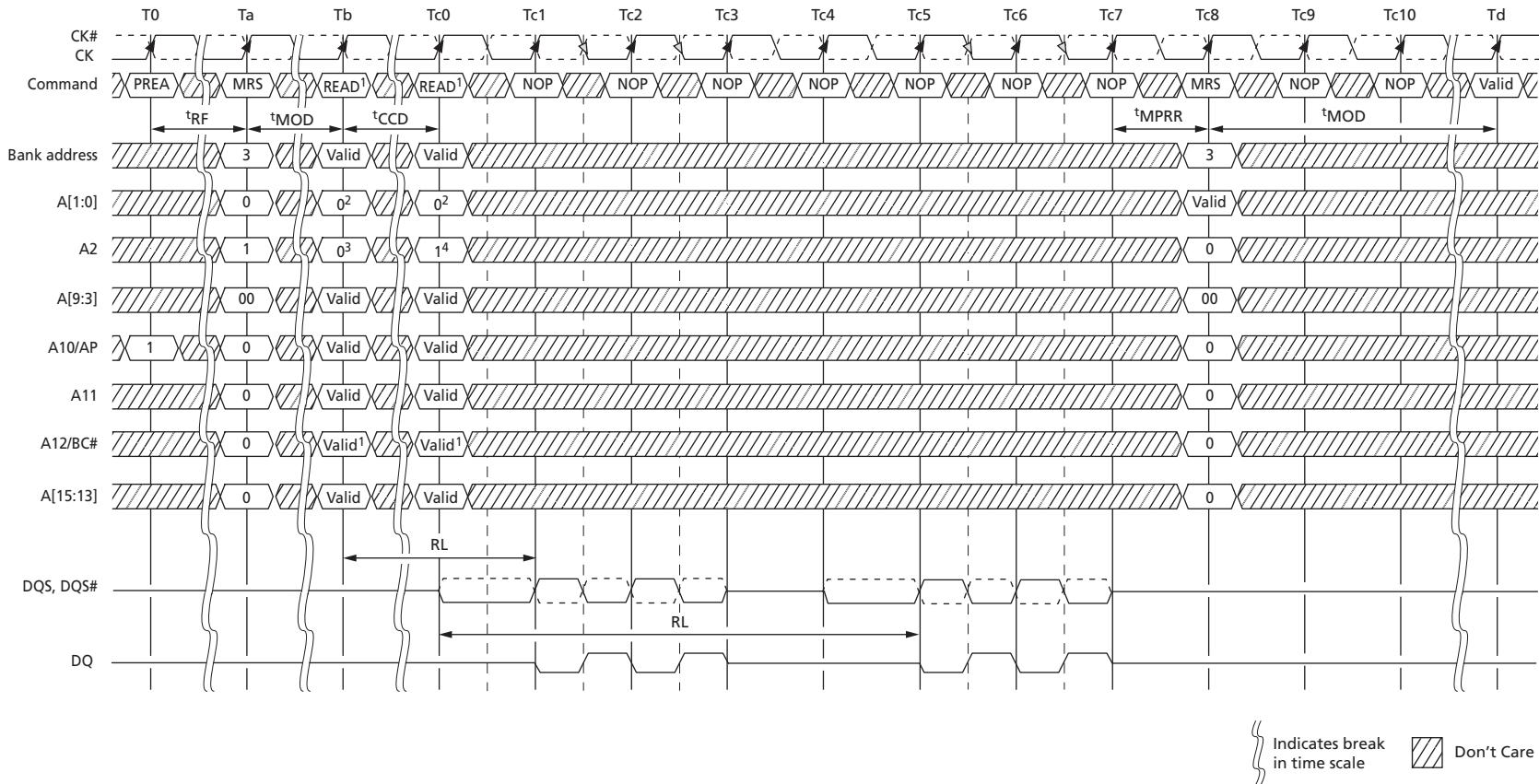
Notes:

1. READ with BL8 either by MRS or OTF.
2. Memory controller must drive 0 on A[2:0].

Figure 61: MPR System Read Calibration with BL8: Fixed Burst Order, Back-to-Back Readout


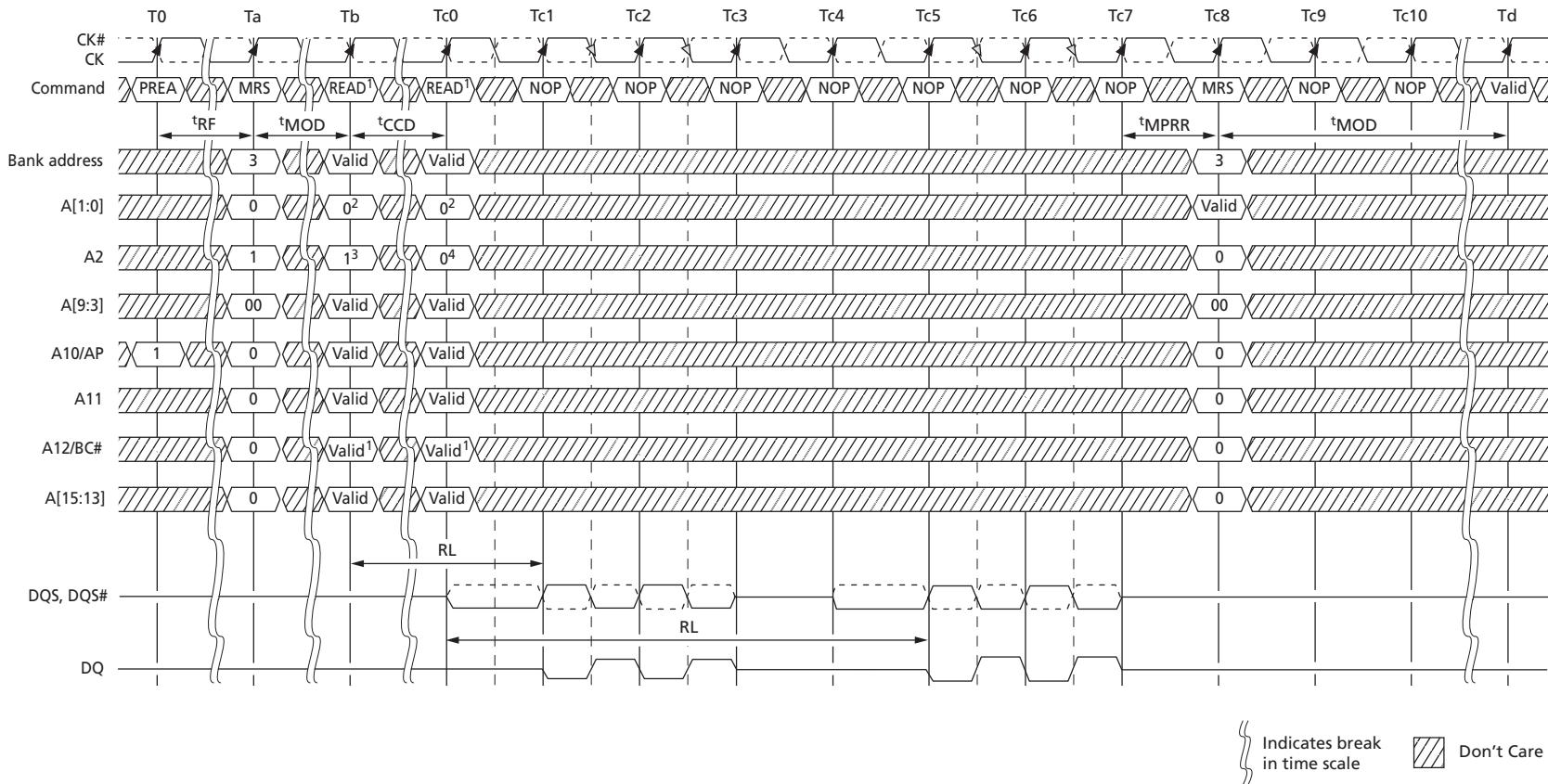
Notes:

1. READ with BL8 either by MRS or OTF.
2. Memory controller must drive 0 on A[2:0].

Figure 62: MPR System Read Calibration with BC4: Lower Nibble, Then Upper Nibble


Notes:

1. READ with BC4 either by MRS or OTF.
2. Memory controller must drive 0 on A[1:0].
3. A2 = 0 selects lower 4 nibble bits 0 . . . 3.
4. A2 = 1 selects upper 4 nibble bits 4 . . . 7.

Figure 63: MPR System Read Calibration with BC4: Upper Nibble, Then Lower Nibble


Notes:

1. READ with BC4 either by MRS or OTF.
2. Memory controller must drive 0 on A[1:0].
3. A2 = 1 selects upper 4 nibble bits 4 . . . 7.
4. A2 = 0 selects lower 4 nibble bits 0 . . . 3.

MPR Read Predefined Pattern

The predetermined read calibration pattern is a fixed pattern of 0, 1, 0, 1, 0, 1, 0, 1. The following is an example of using the read out predetermined read calibration pattern. The example is to perform multiple reads from the multipurpose register to do system level read timing calibration based on the predetermined and standardized pattern.

The following protocol outlines the steps used to perform the read calibration:

1. Precharge all banks
2. After t_{RP} is satisfied, set MRS, $MR3[2] = 1$ and $MR3[1:0] = 00$. This redirects all subsequent reads and loads the predefined pattern into the MPR. As soon as t_{MRD} and t_{MOD} are satisfied, the MPR is available
3. Data WRITE operations are not allowed until the MPR returns to the normal DRAM state
4. Issue a read with burst order information (all other address pins are “Don’t Care”):
 - $A[1:0] = 00$ (data burst order is fixed starting at nibble)
 - $A2 = 0$ (for BL8, burst order is fixed as 0, 1, 2, 3, 4, 5, 6, 7)
 - $A12 = 1$ (use BL8)
5. After $RL = AL + CL$, the DRAM bursts out the predefined read calibration pattern (0, 1, 0, 1, 0, 1, 0, 1)
6. The memory controller repeats the calibration reads until read data capture at memory controller is optimized
7. After the last MPR READ burst and after t_{MPRR} has been satisfied, issue MRS, $MR3[2] = 0$, and $MR3[1:0] = “Don’t Care”$ to the normal DRAM state. All subsequent read and write accesses will be regular reads and writes from/to the DRAM array
8. When t_{MRD} and t_{MOD} are satisfied from the last MRS, the regular DRAM commands (such as activate a memory bank for regular read or write access) are permitted

MODE REGISTER SET (MRS) Command

The mode registers are loaded via inputs $BA[2:0]$, $A[13:0]$. $BA[2:0]$ determine which mode register is programmed:

- $BA2 = 0$, $BA1 = 0$, $BA0 = 0$ for MR0
- $BA2 = 0$, $BA1 = 0$, $BA0 = 1$ for MR1
- $BA2 = 0$, $BA1 = 1$, $BA0 = 0$ for MR2
- $BA2 = 0$, $BA1 = 1$, $BA0 = 1$ for MR3

The MRS command can only be issued (or re-issued) when all banks are idle and in the precharged state (t_{RP} is satisfied and no data bursts are in progress). The controller must wait the specified time t_{MRD} before initiating a subsequent operation such as an ACTIVATE command (see Figure 50 (page 134)). There is also a restriction after issuing an MRS command with regard to when the updated functions become available. This parameter is specified by t_{MOD} . Both t_{MRD} and t_{MOD} parameters are shown in Figure 50 (page 134) and Figure 51 (page 135). Violating either of these requirements will result in unspecified operation.

ZQ CALIBRATION Operation

The ZQ CALIBRATION command is used to calibrate the DRAM output drivers (R_{ON}) and ODT values (R_{TT}) over process, voltage, and temperature, provided a dedicated 240Ω ($\pm 1\%$) external resistor is connected from the DRAM's ZQ ball to V_{SSQ} .

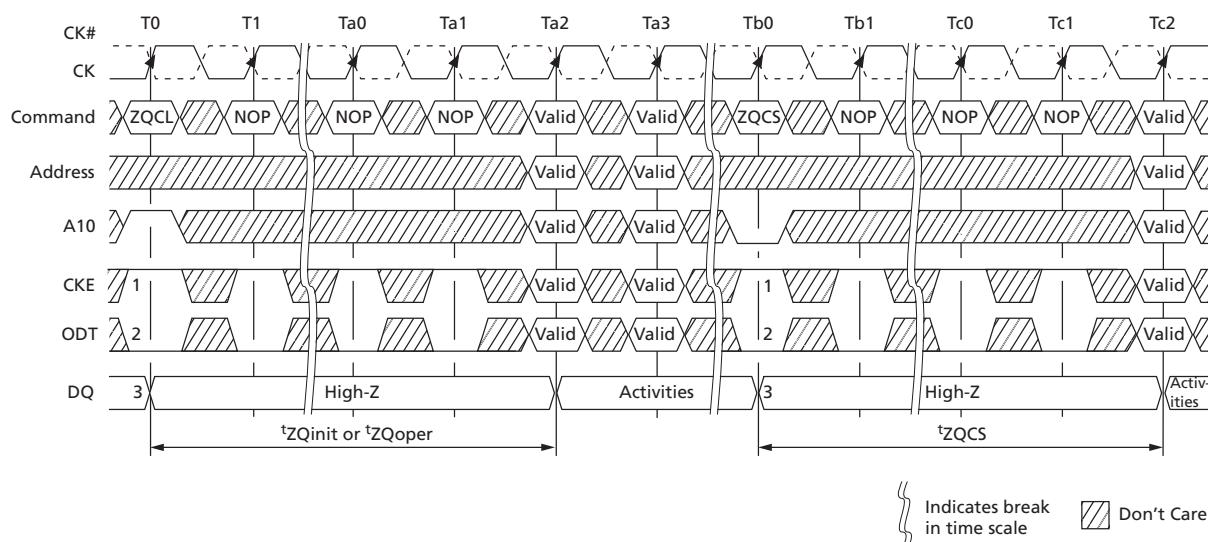
DDR3 SDRAM require a longer time to calibrate R_{ON} and ODT at power-up initialization and self refresh exit, and a relatively shorter time to perform periodic calibrations. DDR3 SDRAM defines two ZQ CALIBRATION commands: ZQCL and ZQCS. An example of ZQ calibration timing is shown below.

All banks must be precharged and t_{RP} must be met before ZQCL or ZQCS commands can be issued to the DRAM. No other activities (other than issuing another ZQCL or ZQCS command) can be performed on the DRAM channel by the controller for the duration of t_{ZQinit} or t_{ZQoper} . The quiet time on the DRAM channel helps accurately calibrate R_{ON} and ODT. After DRAM calibration is achieved, the DRAM should disable the ZQ ball's current consumption path to reduce power.

ZQ CALIBRATION commands can be issued in parallel to DLL RESET and locking time. Upon self refresh exit, an explicit ZQCL is required if ZQ calibration is desired.

In dual-rank systems that share the ZQ resistor between devices, the controller must not enable overlap of t_{ZQinit} , t_{ZQoper} , or t_{ZQCS} between ranks.

Figure 64: ZQ CALIBRATION Timing (ZQCL and ZQCS)



Notes:

1. CKE must be continuously registered HIGH during the calibration procedure.
2. ODT must be disabled via the ODT signal or the MRS during the calibration procedure.
3. All devices connected to the DQ bus should be High-Z during calibration.

ACTIVATE Operation

Before any READ or WRITE commands can be issued to a bank within the DRAM, a row in that bank must be opened (activated). This is accomplished via the ACTIVATE command, which selects both the bank and the row to be activated.

After a row is opened with an ACTIVATE command, a READ or WRITE command may be issued to that row, subject to the t_{RCD} specification. However, if the additive latency is programmed correctly, a READ or WRITE command may be issued prior to t_{RCD} (MIN). In this operation, the DRAM enables a READ or WRITE command to be issued after the ACTIVATE command for that bank, but prior to t_{RCD} (MIN) with the requirement that $(\text{ACTIVATE-to-READ/WRITE}) + \text{AL} \geq t_{RCD} \text{ (MIN)}$ (see Posted CAS Additive Latency). t_{RCD} (MIN) should be divided by the clock period and rounded up to the next whole number to determine the earliest clock edge after the ACTIVATE command on which a READ or WRITE command can be entered. The same procedure is used to convert other specification limits from time units to clock cycles.

When at least one bank is open, any READ-to-READ command delay or WRITE-to-WRITE command delay is restricted to t_{CCD} (MIN).

A subsequent ACTIVATE command to a different row in the same bank can only be issued after the previous active row has been closed (precharged). The minimum time interval between successive ACTIVATE commands to the same bank is defined by t_{RC} .

A subsequent ACTIVATE command to another bank can be issued while the first bank is being accessed, which results in a reduction of total row-access overhead. The minimum time interval between successive ACTIVATE commands to different banks is defined by t_{RRD} . No more than four bank ACTIVATE commands may be issued in a given t_{FAW} (MIN) period, and the t_{RRD} (MIN) restriction still applies. The t_{FAW} (MIN) parameter applies, regardless of the number of banks already opened or closed.

Figure 65: Example: Meeting t_{RRD} (MIN) and t_{RCD} (MIN)

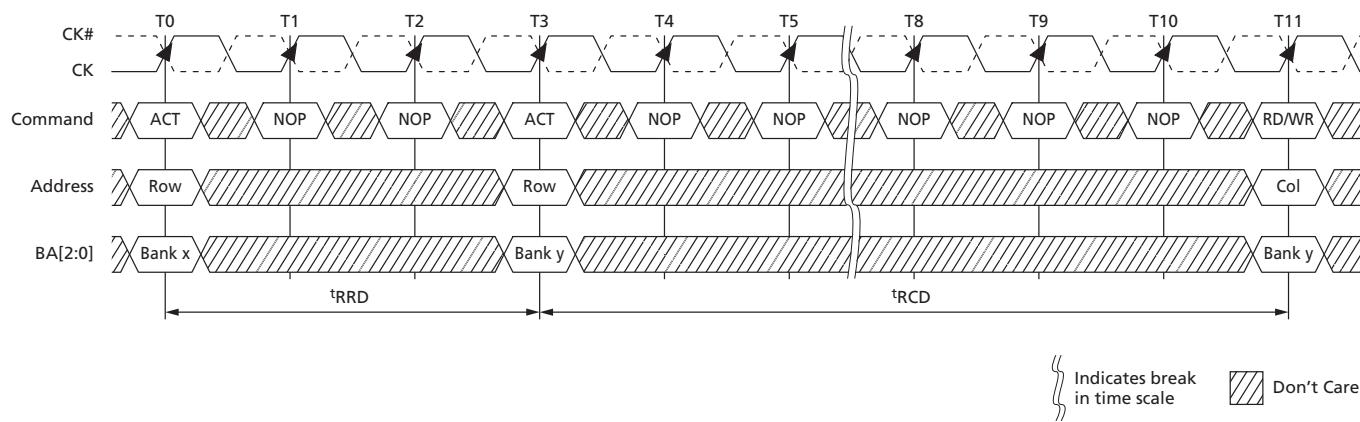
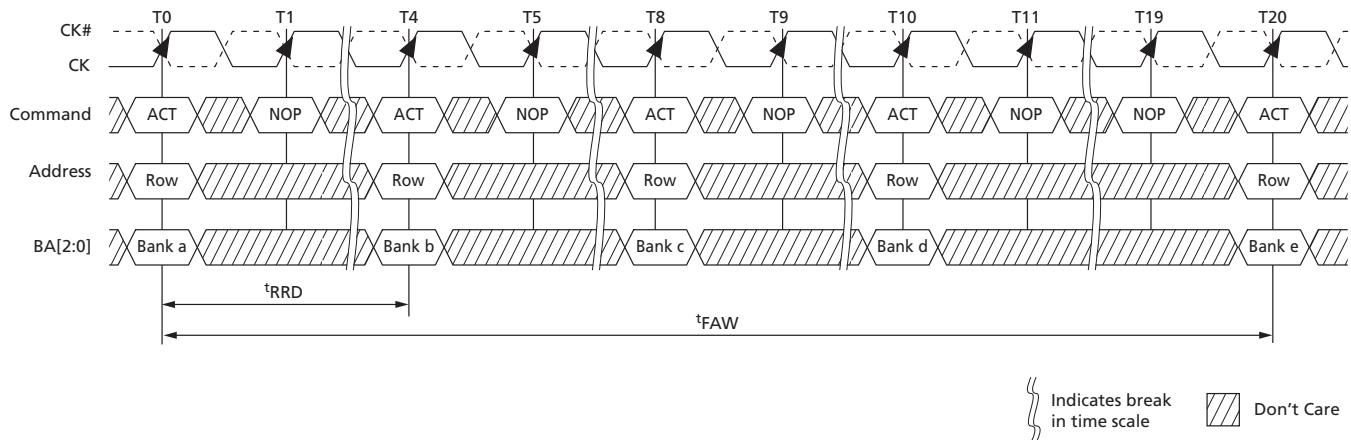


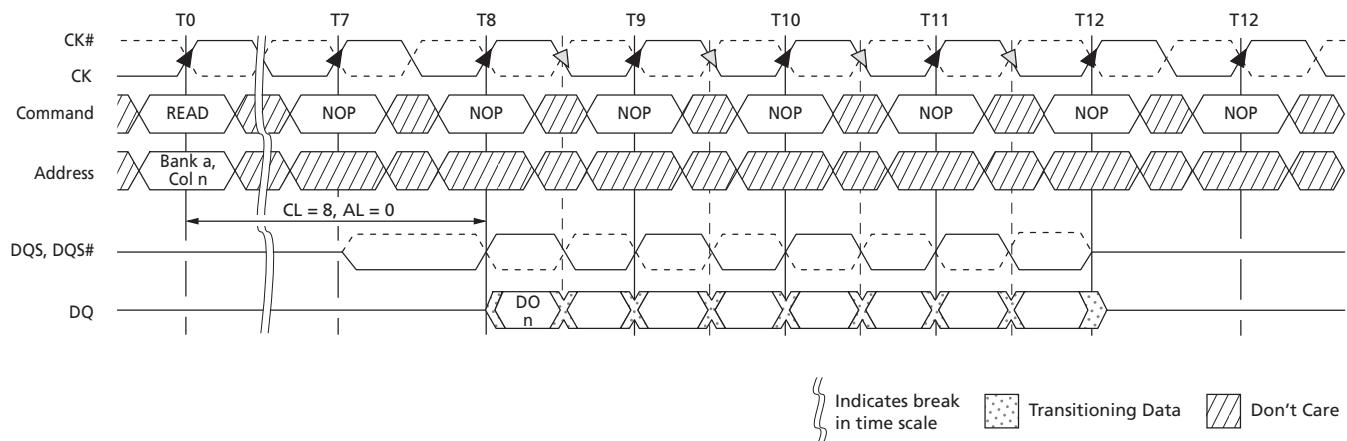
Figure 66: Example: t_{FAW}


READ Operation

READ bursts are initiated with a READ command. The starting column and bank addresses are provided with the READ command and auto precharge is either enabled or disabled for that burst access. If auto precharge is enabled, the row being accessed is automatically precharged at the completion of the burst. If auto precharge is disabled, the row will be left open after the completion of the burst.

During READ bursts, the valid data-out element from the starting column address is available READ latency (RL) clocks later. RL is defined as the sum of posted CAS additive latency (AL) and CAS latency (CL) ($RL = AL + CL$). The value of AL and CL is programmable in the mode register via the MRS command. Each subsequent data-out element is valid nominally at the next positive or negative clock edge (that is, at the next crossing of CK and CK#). Figure 67 shows an example of RL based on a CL setting of 8 and an AL setting of 0.

Figure 67: READ Latency



Notes:

1. DO n = data-out from column n .
2. Subsequent elements of data-out appear in the programmed order following DO n .

DQS, DQS# is driven by the DRAM along with the output data. The initial LOW state on DQS and HIGH state on DQS# is known as the READ preamble (t_{RPRE}). The LOW state on DQS and the HIGH state on DQS#, coincident with the last data-out element, is known as the READ postamble (t_{RPST}). Upon completion of a burst, assuming no other commands have been initiated, the DQ goes High-Z. A detailed explanation of t_{DQSQ} (valid data-out skew), t_{QH} (data-out window hold), and the valid data window are depicted in Figure 78 (page 165). A detailed explanation of t_{DQSK} (DQS transition skew to CK) is also depicted in Figure 78 (page 165).

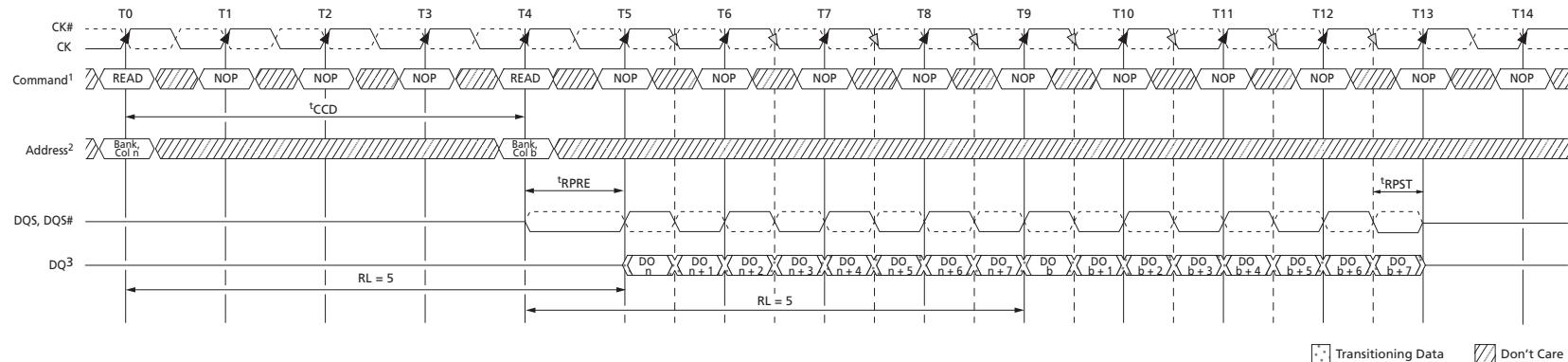
Data from any READ burst may be concatenated with data from a subsequent READ command to provide a continuous flow of data. The first data element from the new burst follows the last element of a completed burst. The new READ command should be issued t_{CCD} cycles after the first READ command. This is shown for BL8 in Figure 68 (page 159). If BC4 is enabled, t_{CCD} must still be met, which will cause a gap in the data output, as shown in Figure 69 (page 159). Nonconsecutive READ data is reflected in

Figure 70 (page 160). DDR3 SDRAM does not allow interrupting or truncating any READ burst.

Data from any READ burst must be completed before a subsequent WRITE burst is allowed. An example of a READ burst followed by a WRITE burst for BL8 is shown in Figure 71 (page 160) (BC4 is shown in Figure 72 (page 161)). To ensure the READ data is completed before the WRITE data is on the bus, the minimum READ-to-WRITE timing is $RL + t_{CCD} - WL + 2t_{CK}$.

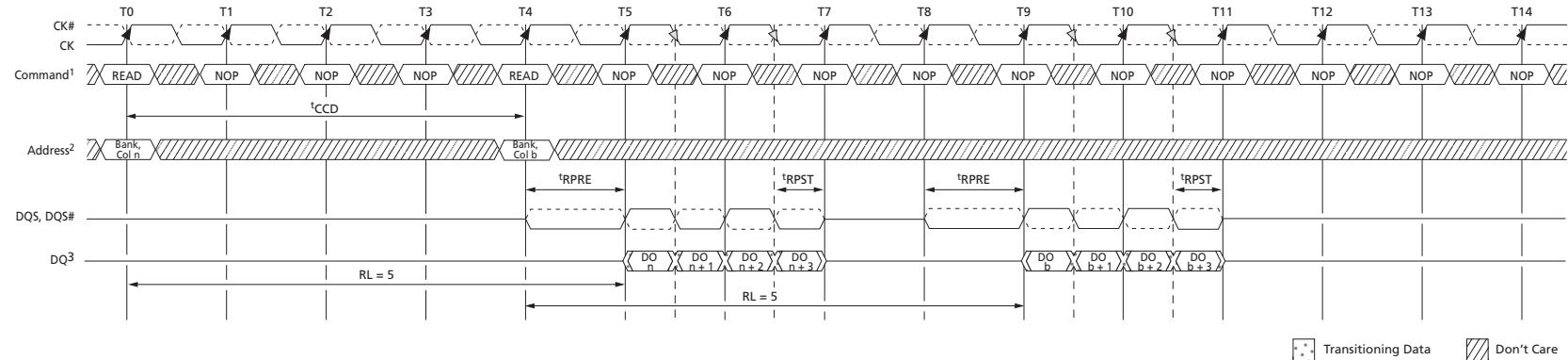
A READ burst may be followed by a PRECHARGE command to the same bank, provided auto precharge is not activated. The minimum READ-to-PRECHARGE command spacing to the same bank is four clocks and must also satisfy a minimum analog time from the READ command. This time is called t_{RTP} (READ-to-PRECHARGE). t_{RTP} starts AL cycles later than the READ command. Examples for BL8 are shown in Figure 73 (page 161) and BC4 in Figure 74 (page 162). Following the PRECHARGE command, a subsequent command to the same bank cannot be issued until t_{RP} is met. The PRECHARGE command followed by another PRECHARGE command to the same bank is allowed. However, the precharge period will be determined by the last PRECHARGE command issued to the bank.

If A10 is HIGH when a READ command is issued, the READ with auto precharge function is engaged. The DRAM starts an auto precharge operation on the rising edge, which is $AL + t_{RTP}$ cycles after the READ command. DRAM support a t_{RAS} lockout feature (see Figure 76 (page 162)). If t_{RAS} (MIN) is not satisfied at the edge, the starting point of the auto precharge operation will be delayed until t_{RAS} (MIN) is satisfied. If t_{RTP} (MIN) is not satisfied at the edge, the starting point of the auto precharge operation is delayed until t_{RTP} (MIN) is satisfied. In case the internal precharge is pushed out by t_{RTP} , t_{RP} starts at the point at which the internal precharge happens (not at the next rising clock edge after this event). The time from READ with auto precharge to the next ACTIVATE command to the same bank is $AL + (t_{RTP} + t_{RP})^*$, where * means rounded up to the next integer. In any event, internal precharge does not start earlier than four clocks after the last 8n-bit prefetch.

Figure 68: Consecutive READ Bursts (BL8)


Notes:

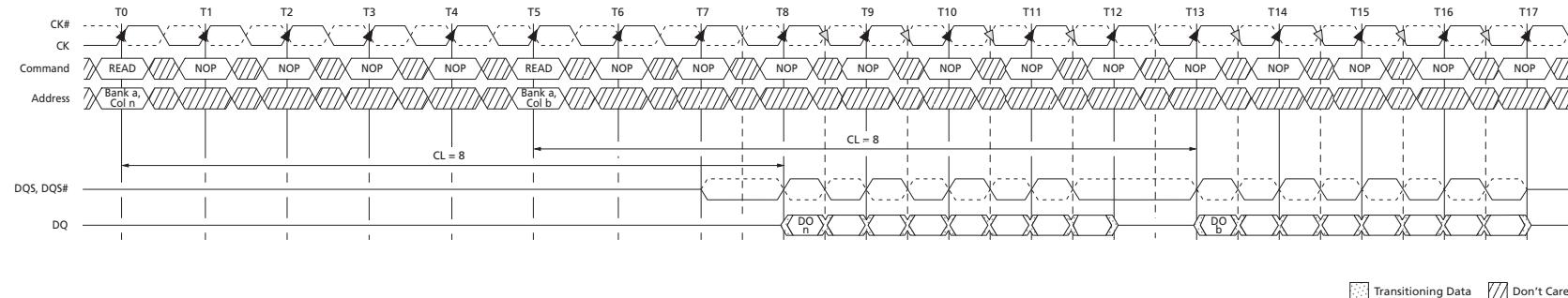
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BL8 setting is activated by either $MRO[1:0] = 00$ or $MRO[1:0] = 01$ and $A12 = 1$ during READ command at T0 and T4.
3. $DO\ n$ (or b) = data-out from column n (or column b).
4. BL8, $RL = 5$ ($CL = 5$, $AL = 0$).

Figure 69: Consecutive READ Bursts (BC4)


Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BC4 setting is activated by either $MRO[1:0] = 10$ or $MRO[1:0] = 01$ and $A12 = 0$ during READ command at T0 and T4.
3. $DO\ n$ (or b) = data-out from column n (or column b).
4. BC4, $RL = 5$ ($CL = 5$, $AL = 0$).

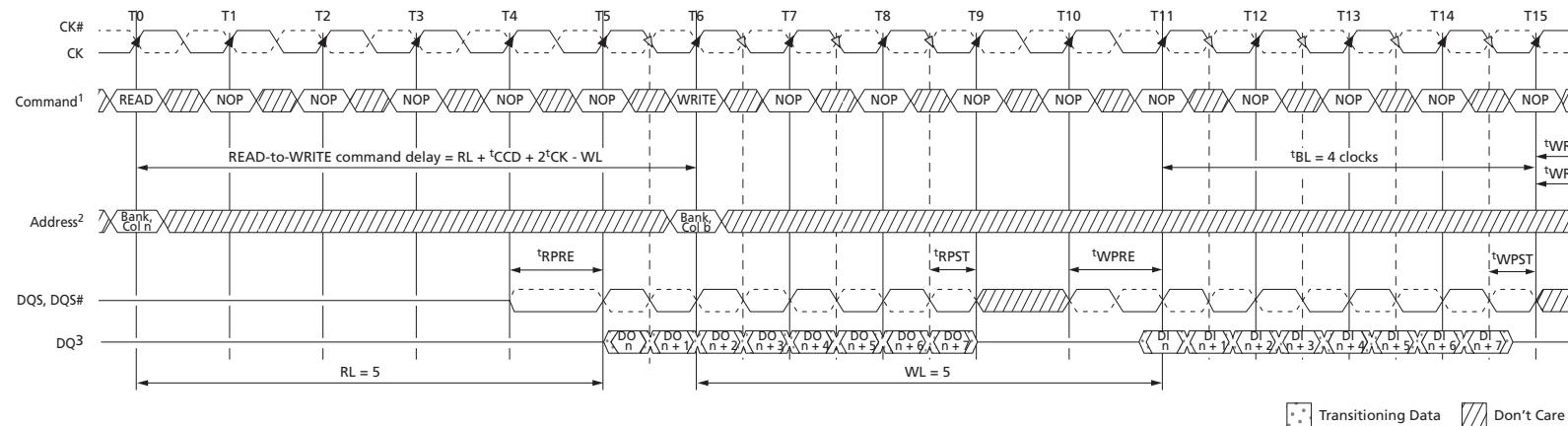
Figure 70: Nonconsecutive READ Bursts



Notes:

1. AL = 0, RL = 8.
2. DO *n* (or *b*) = data-out from column *n* (or column *b*).
3. Seven subsequent elements of data-out appear in the programmed order following DO *n*.
4. Seven subsequent elements of data-out appear in the programmed order following DO *b*.

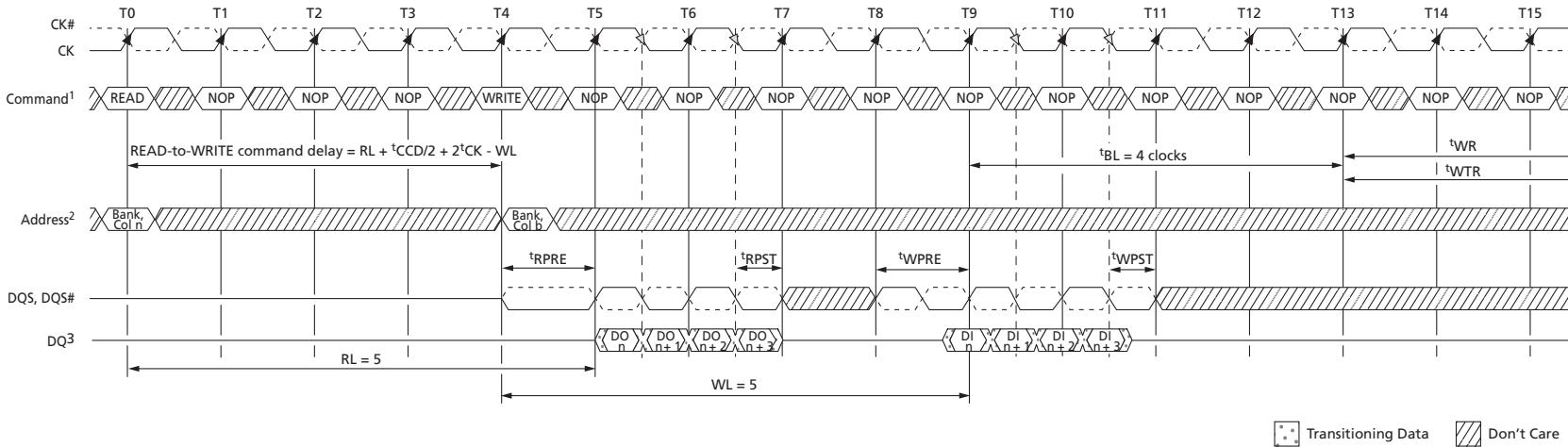
Figure 71: READ (BL8) to WRITE (BL8)



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BL8 setting is activated by either MR0[1:0] = 00 or MR0[1:0] = 01 and A12 = 1 during the READ command at T0, and the WRITE command at T6.
3. DO *n* = data-out from column, DI *b* = data-in for column *b*.
4. BL8, RL = 5 (AL = 0, CL = 5), WL = 5 (AL = 0, CWL = 5).

Figure 72: READ (BC4) to WRITE (BC4) OTF



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BC4 OTF setting is activated by MRO[1:0] and A12 = 0 during READ command at T0 and WRITE command at T4.
3. DO n = data-out from column n ; DI n = data-in from column b .
4. BC4, RL = 5 (AL = 0, CL = 5), WL = 5 (AL = 0, CWL = 5).

Figure 73: READ to PRECHARGE (BL8)

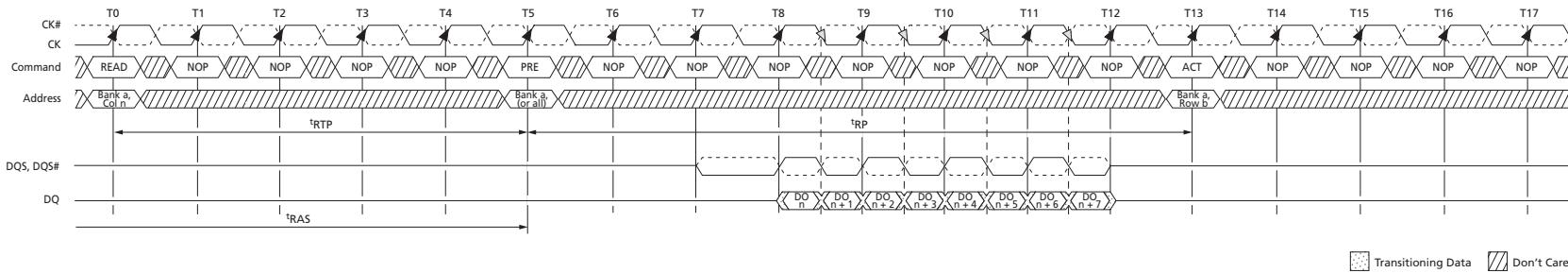


Figure 74: READ to PRECHARGE (BC4)

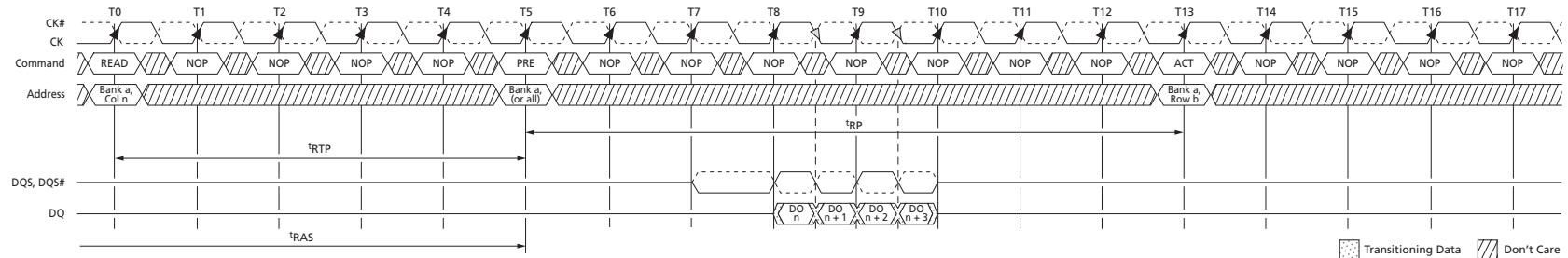


Figure 75: READ to PRECHARGE (AL = 5, CL = 6)

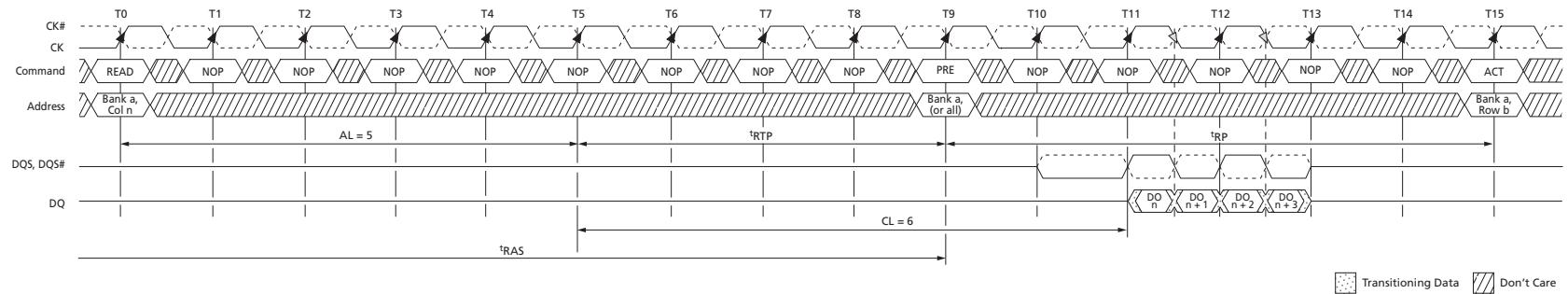
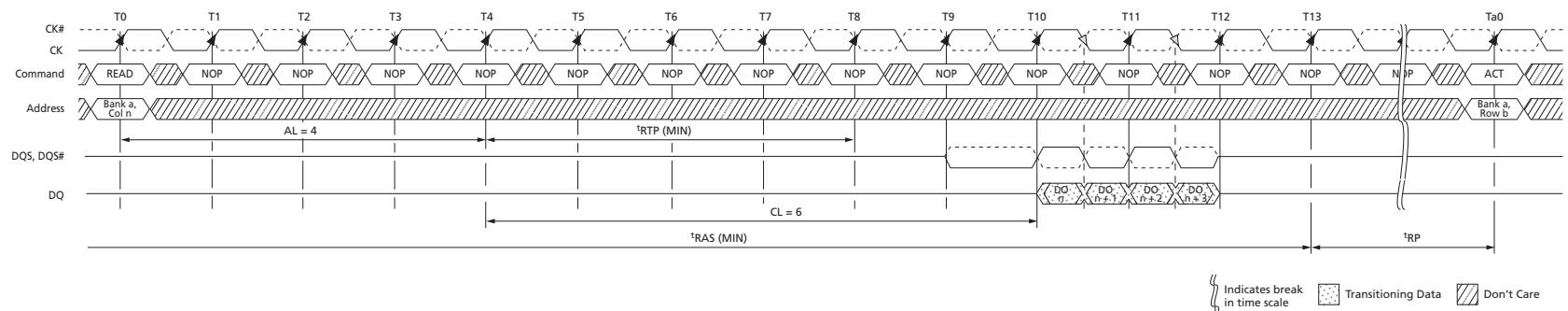


Figure 76: READ with Auto Precharge (AL = 4, CL = 6)



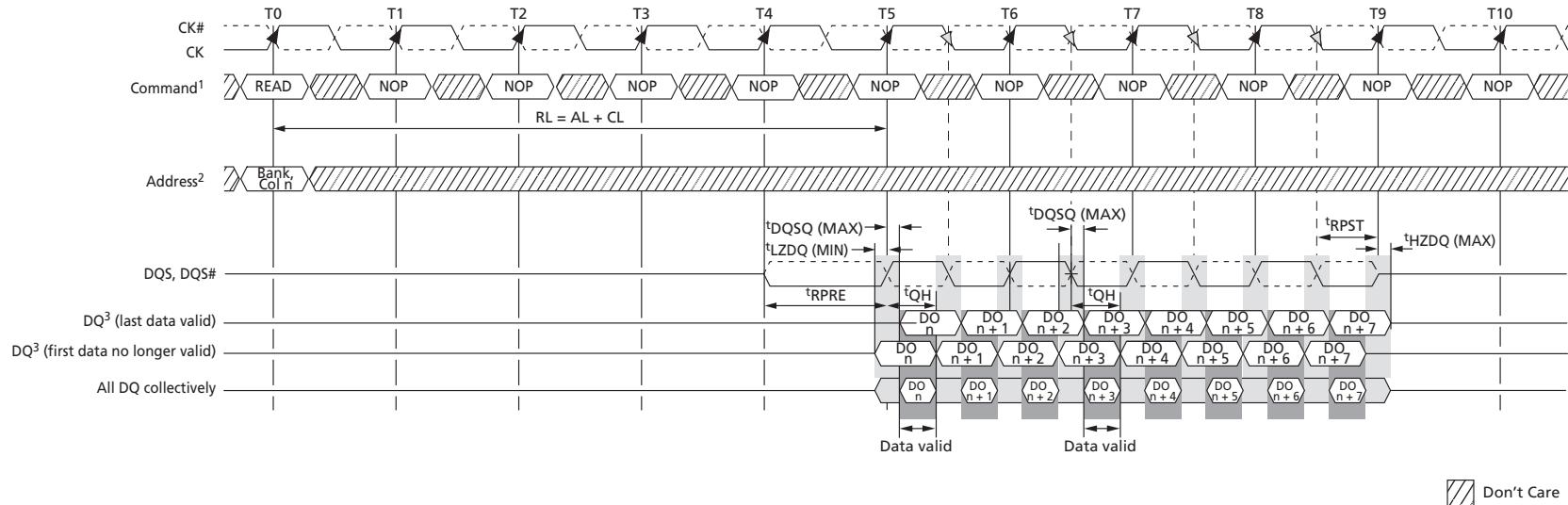
DQS to DQ output timing is shown in Figure 77 (page 164). The DQ transitions between valid data outputs must be within t_{DQSQ} of the crossing point of DQS, DQS#. DQS must also maintain a minimum HIGH and LOW time of t_{QSH} and t_{QSL} . Prior to the READ preamble, the DQ balls will either be floating or terminated, depending on the status of the ODT signal.

Figure 78 (page 165) shows the strobe-to-clock timing during a READ. The crossing point DQS, DQS# must transition within $\pm t_{DQSK}$ of the clock crossing point. The data out has no timing relationship to CK, only to DQS, as shown in Figure 78 (page 165).

Figure 78 (page 165) also shows the READ preamble and postamble. Typically, both DQS and DQS# are High-Z to save power (V_{DDQ}). Prior to data output from the DRAM, DQS is driven LOW and DQS# is HIGH for t_{RPRE} . This is known as the READ preamble.

The READ postamble, t_{RPST} , is one half clock from the last DQS, DQS# transition. During the READ postamble, DQS is driven LOW and DQS# is HIGH. When complete, the DQ is disabled or continues terminating, depending on the state of the ODT signal. Figure 81 (page 167) demonstrates how to measure t_{RPST} .

Figure 77: Data Output Timing – t_{DQSQ} and Data Valid Window



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BL8 setting is activated by either $MR0[1, 0] = 0, 0$ or $MR0[0, 1] = 0, 1$ and $A12 = 1$ during READ command at T0.
3. DO n = data-out from column n.
4. BL8, $RL = 5$ ($AL = 0$, $CL = 5$).
5. Output timings are referenced to $V_{DDQ}/2$ and DLL on and locked.
6. t_{DQSQ} defines the skew between DQS, DQS# to data and does not define DQS, DQS# to CK.
7. Early data transitions may not always happen at the same DQ. Data transitions of a DQ can be early or late within a burst.

t_{HZ} and t_{LZ} transitions occur in the same access time as valid data transitions. These parameters are referenced to a specific voltage level that specifies when the device output is no longer driving t_{HZDQS} and t_{HZDQ} , or begins driving t_{LZDQS} , t_{LZDQ} . Figure 79 (page 166) shows a method of calculating the point when the device is no longer driving t_{HZDQS} and t_{HZDQ} , or begins driving t_{LZDQS} , t_{LZDQ} , by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent. The parameters t_{LZDQS} , t_{LZDQ} , t_{HZDQS} , and t_{HZDQ} are defined as single-ended.

Figure 78: Data Strobe Timing – READs

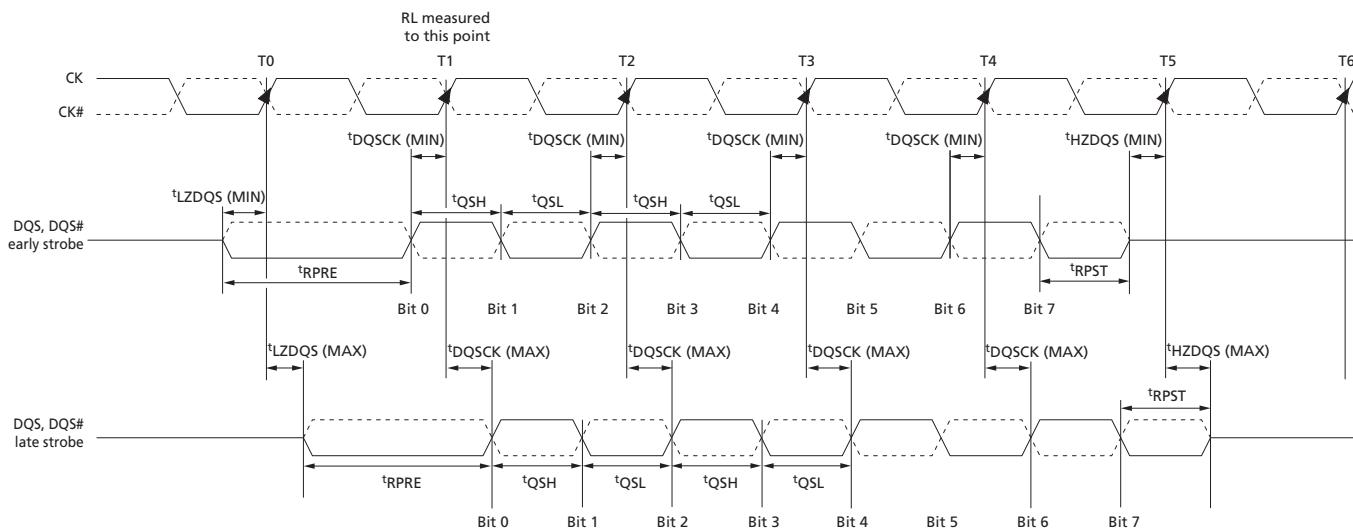
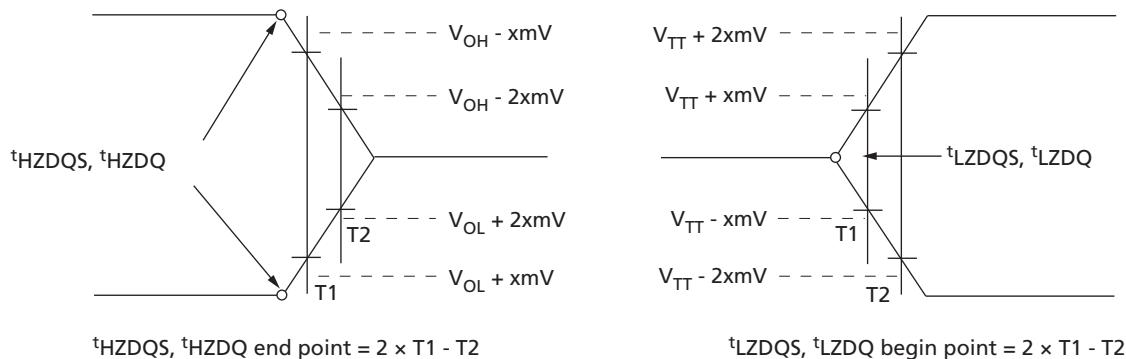


Figure 79: Method for Calculating t_{LZ} and t_{HZ}


Notes:

1. Within a burst, the rising strobe edge is not necessarily fixed at t_{DQSCK} (MIN) or t_{DQSCK} (MAX). Instead, the rising strobe edge can vary between t_{DQSCK} (MIN) and t_{DQSCK} (MAX).
2. The DQS HIGH pulse width is defined by t_{QSH} , and the DQS LOW pulse width is defined by t_{QSL} . Likewise, t_{LZDQS} (MIN) and t_{HZDQS} (MIN) are not tied to t_{DQSCK} (MIN) (early strobe case), and t_{LZDQS} (MAX) and t_{HZDQS} (MAX) are not tied to t_{DQSCK} (MAX) (late strobe case); however, they tend to track one another.
3. The minimum pulse width of the READ preamble is defined by t_{RPRE} (MIN). The minimum pulse width of the READ postamble is defined by t_{RPST} (MIN).

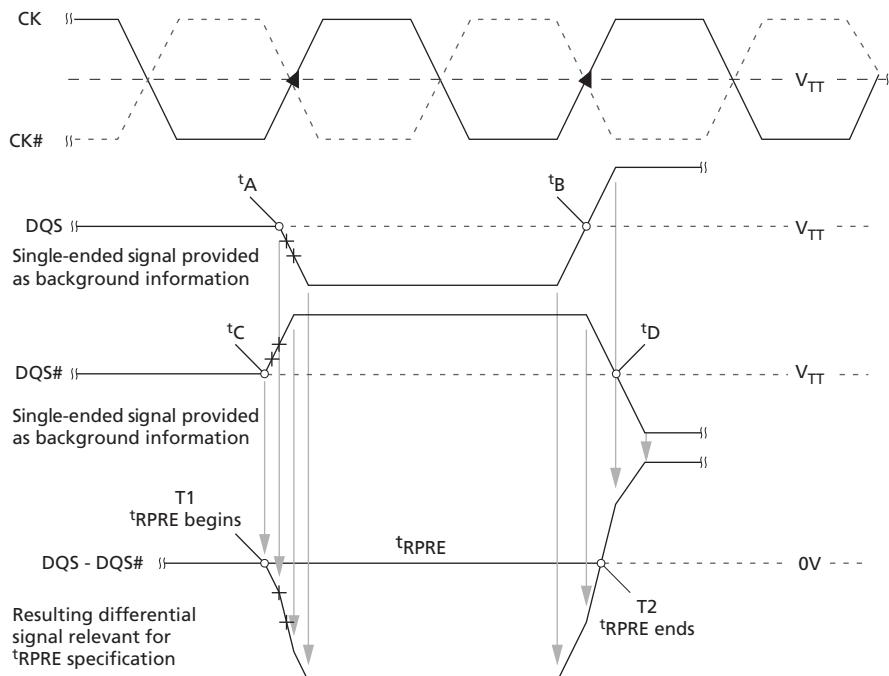
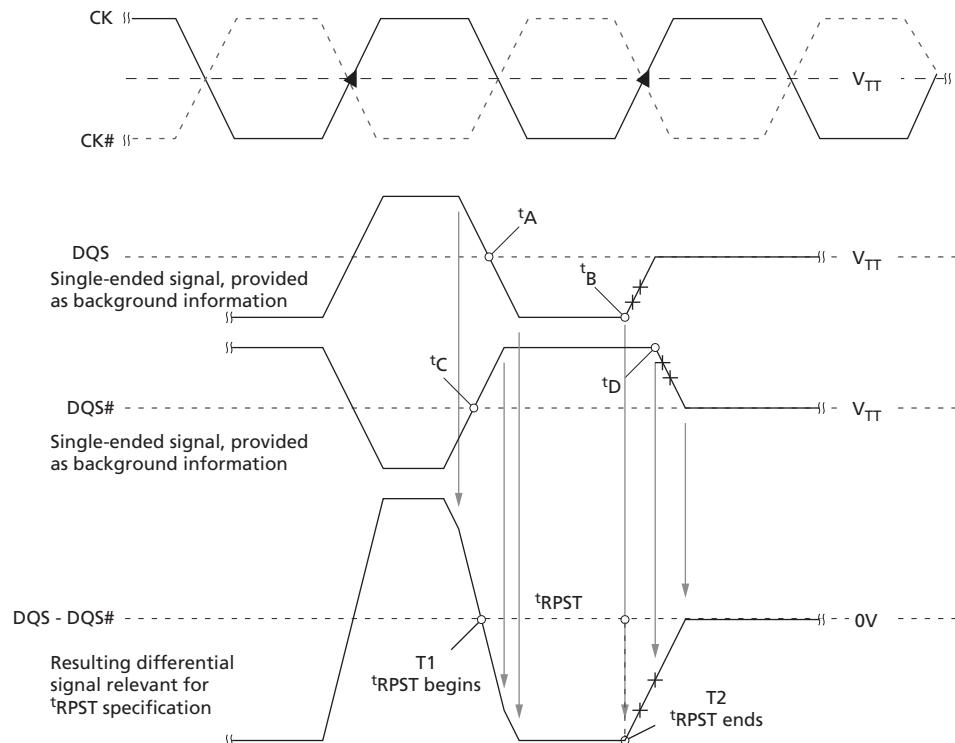
Figure 80: t_{RPRE} Timing


Figure 81: t_{RPST} Timing


WRITE Operation

WRITE bursts are initiated with a WRITE command. The starting column and bank addresses are provided with the WRITE command, and auto precharge is either enabled or disabled for that access. If auto precharge is selected, the row being accessed is precharged at the end of the WRITE burst. If auto precharge is not selected, the row will remain open for subsequent accesses. After a WRITE command has been issued, the WRITE burst may not be interrupted. For the generic WRITE commands used in Figure 84 (page 170) through Figure 92 (page 175), auto precharge is disabled.

During WRITE bursts, the first valid data-in element is registered on a rising edge of DQS following the WRITE latency (WL) clocks later and subsequent data elements will be registered on successive edges of DQS. WRITE latency (WL) is defined as the sum of posted CAS additive latency (AL) and CAS WRITE latency (CWL): $WL = AL + CWL$. The values of AL and CWL are programmed in the MR0 and MR2 registers, respectively. Prior to the first valid DQS edge, a full cycle is needed (including a dummy crossover of DQS, DQS#) and specified as the WRITE preamble shown in Figure 84 (page 170). The half cycle on DQS following the last data-in element is known as the WRITE postamble.

The time between the WRITE command and the first valid edge of DQS is WL clocks $\pm t_{DQSS}$. Figure 85 (page 171) through Figure 92 (page 175) show the nominal case where $t_{DQSS} = 0\text{ns}$; however, Figure 84 (page 170) includes $t_{DQSS}(\text{MIN})$ and $t_{DQSS}(\text{MAX})$ cases.

Data may be masked from completing a WRITE using data mask. The data mask occurs on the DM ball aligned to the WRITE data. If DM is LOW, the WRITE completes normally. If DM is HIGH, that bit of data is masked.

Upon completion of a burst, assuming no other commands have been initiated, the DQ will remain High-Z, and any additional input data will be ignored.

Data for any WRITE burst may be concatenated with a subsequent WRITE command to provide a continuous flow of input data. The new WRITE command can be t_{CCD} clocks following the previous WRITE command. The first data element from the new burst is applied after the last element of a completed burst. Figure 85 (page 171) and Figure 86 (page 171) show concatenated bursts. An example of nonconsecutive WRITEs is shown in Figure 87 (page 172).

Data for any WRITE burst may be followed by a subsequent READ command after t_{WTR} has been met (see Figure 88 (page 172), Figure 89 (page 173), and Figure 90 (page 174)).

Data for any WRITE burst may be followed by a subsequent PRECHARGE command, providing t_{WR} has been met, as shown in Figure 91 (page 175) and Figure 92 (page 175).

Both t_{WTR} and t_{WR} starting time may vary, depending on the mode register settings (fixed BC4, BL8 versus OTF).

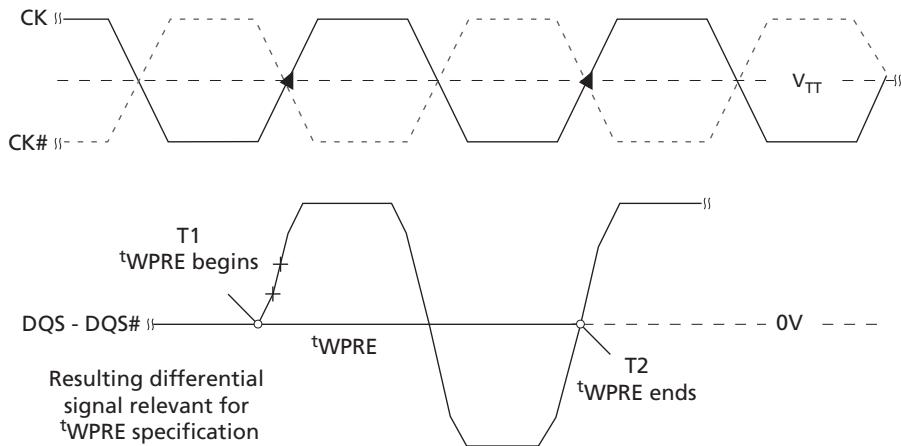
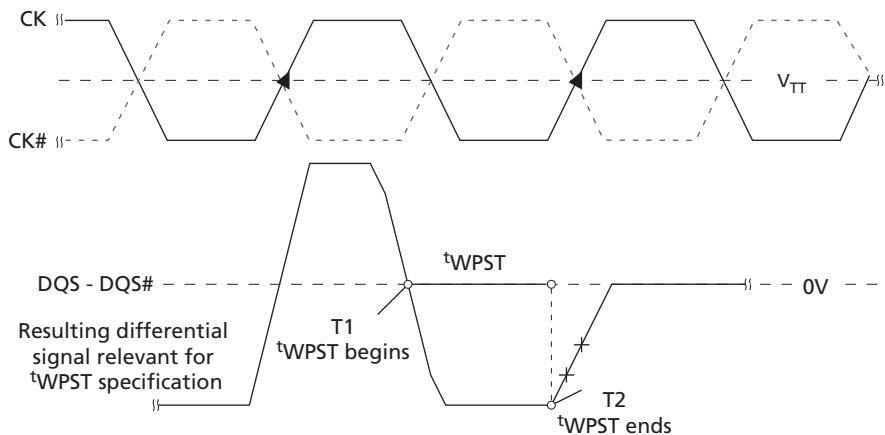
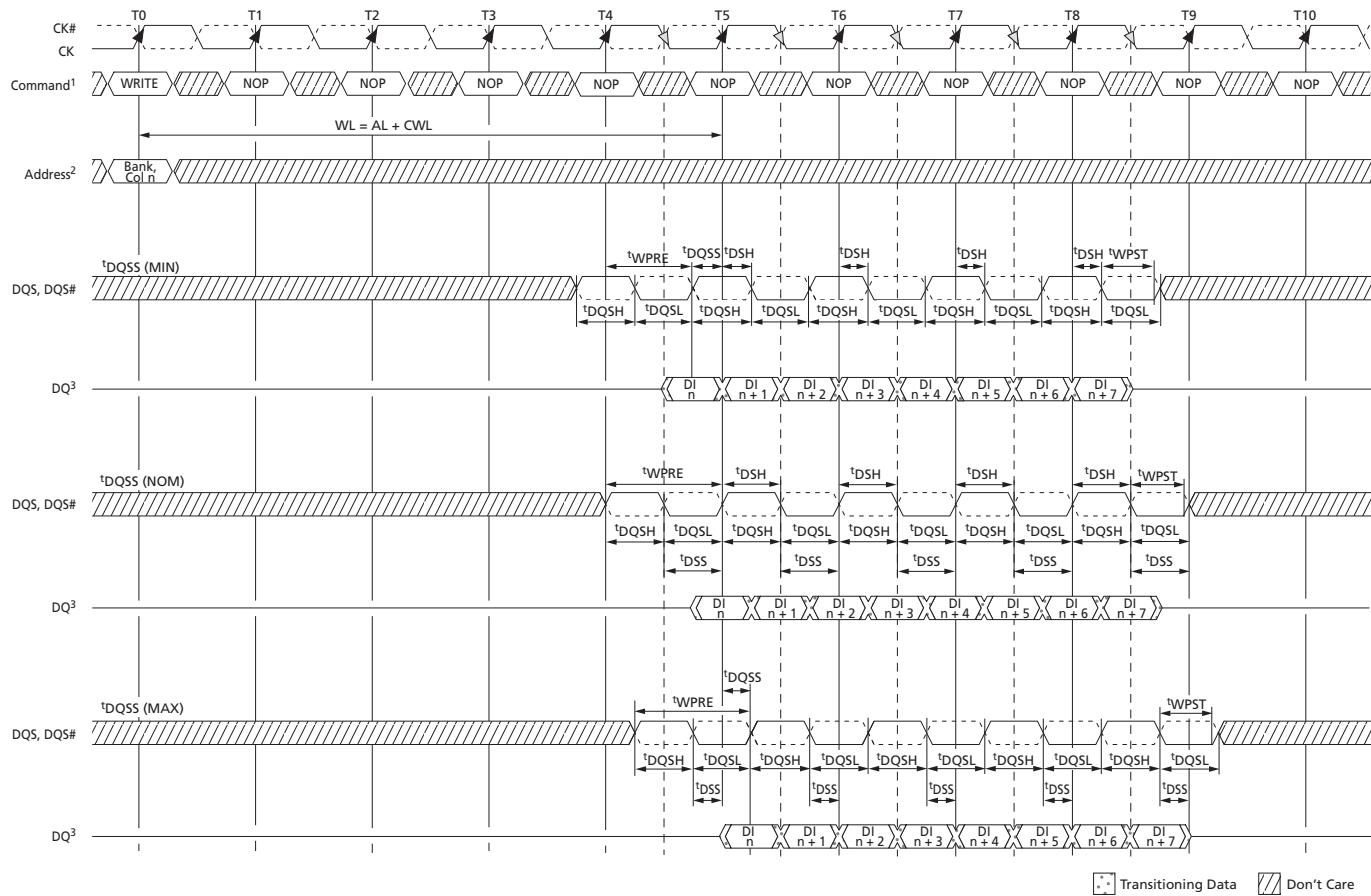
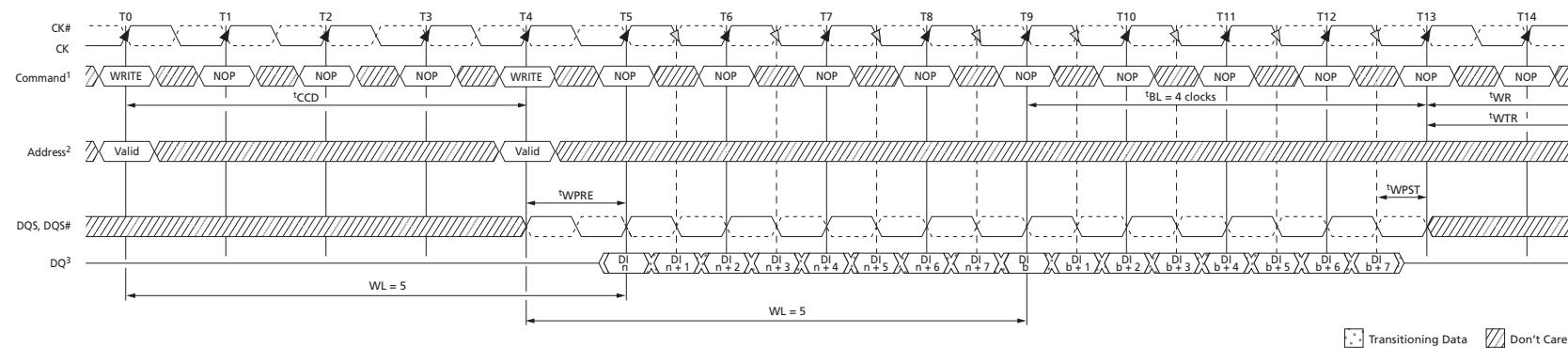
Figure 82: t_{WPRE} Timing

Figure 83: t_{WPST} Timing


Figure 84: WRITE Burst


Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BL8 setting is activated by either MR0[1:0] = 00 or MR0[1:0] = 01 and A12 = 1 during the WRITE command at T0.
3. DI n = data-in for column n .
4. BL8, WL = 5 (AL = 0, CWL = 5).
5. t^{DQSS} must be met at each rising clock edge.
6. t^{WPST} is usually depicted as ending at the crossing of DQS, DQS#; however, t^{WPST} actually ends when DQS no longer drives LOW and DQS# no longer drives HIGH.

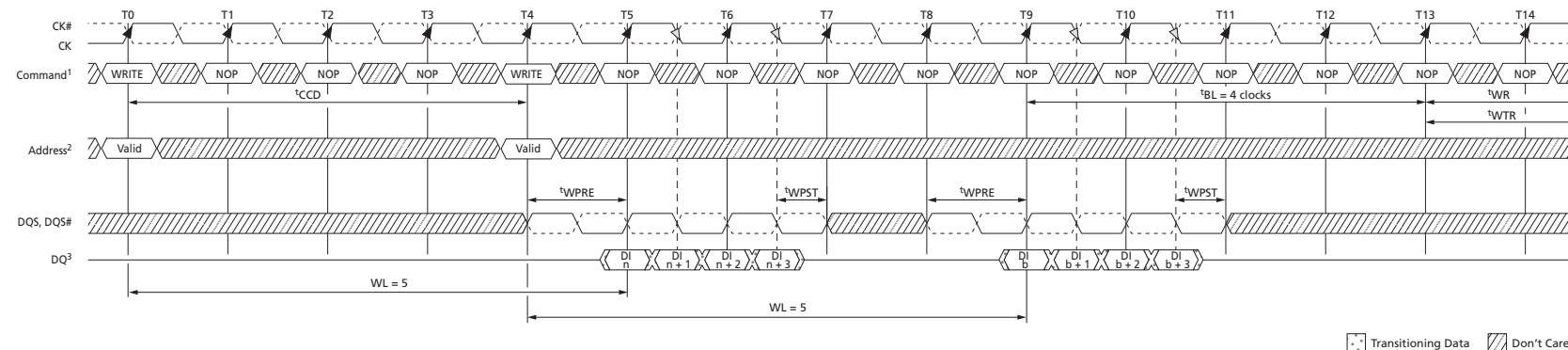
Figure 85: Consecutive WRITE (BL8) to WRITE (BL8)



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The BL8 setting is activated by either MR0[1:0] = 00 or MR0[1:0] = 01 and A12 = 1 during the WRITE commands at T0 and T4.
3. DI n (or b) = data-in for column n (or column b).
4. BL8, WL = 5 (AL = 0, CWL = 5).

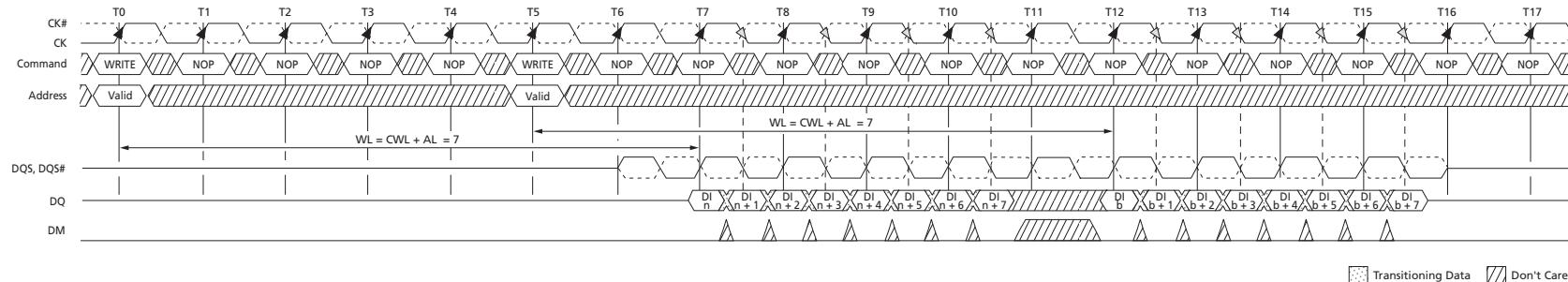
Figure 86: Consecutive WRITE (BC4) to WRITE (BC4) via OTF



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. BC4, WL = 5 (AL = 0, CWL = 5).
3. DI n (or b) = data-in for column n (or column b).
4. The BC4 setting is activated by MR0[1:0] = 01 and A12 = 0 during the WRITE command at T0 and T4.
5. If set via MRS (fixed) tWR and tWTR would start T11 (2 cycles earlier).

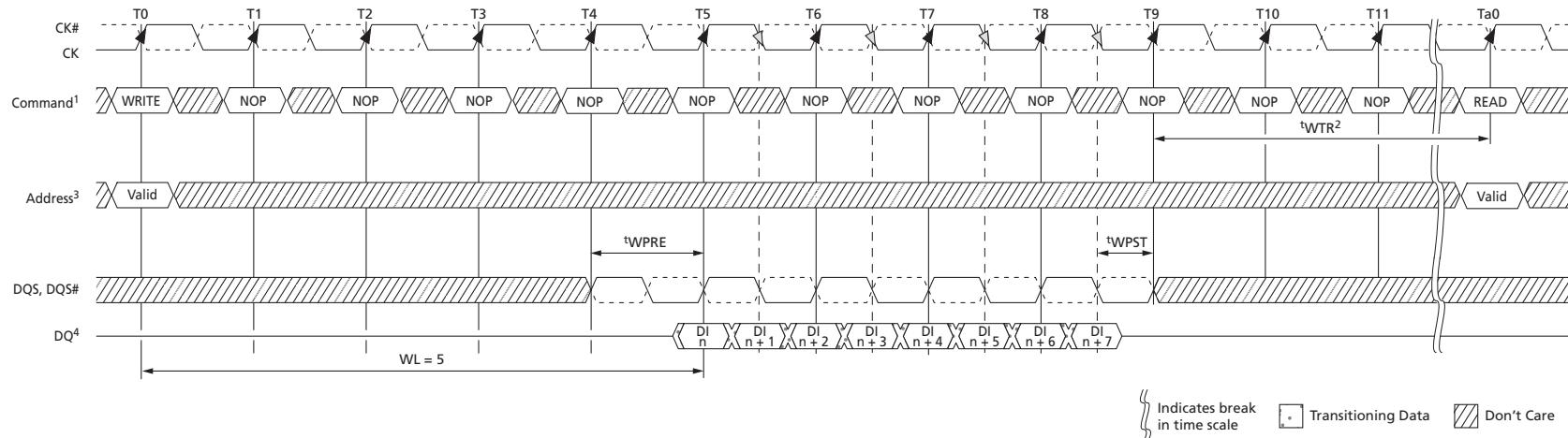
Figure 87: Nonconsecutive WRITE to WRITE



Notes:

1. DI n (or b) = data-in for column n (or column b).
2. Seven subsequent elements of data-in are applied in the programmed order following DO n .
3. Each WRITE command may be to any bank.
4. Shown for WL = 7 (CWL = 7, AL = 0).

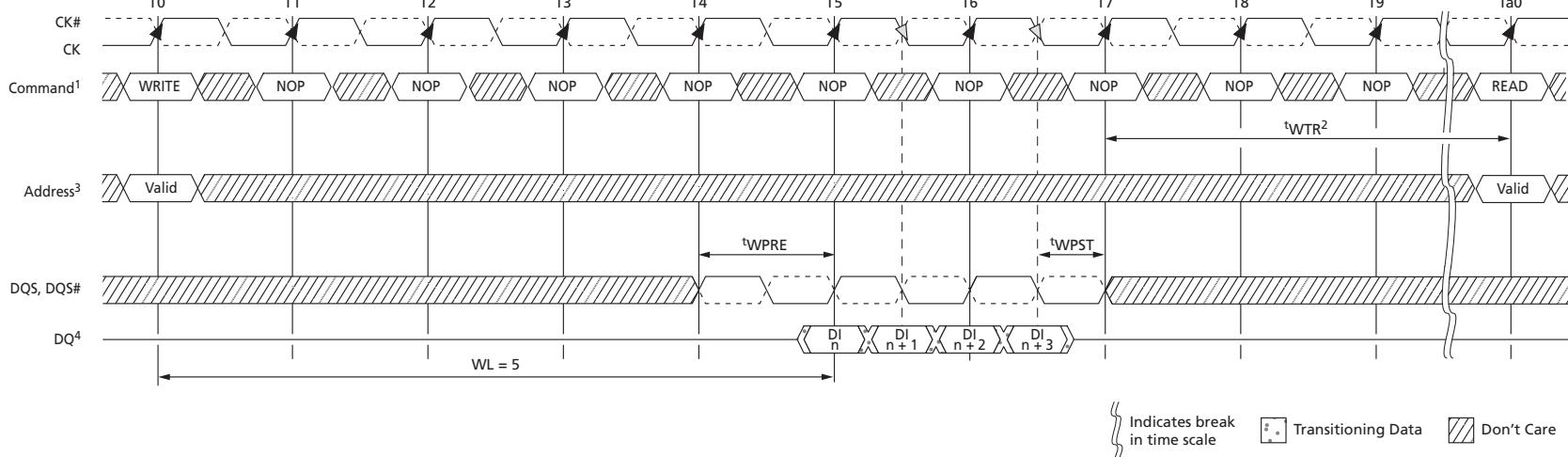
Figure 88: WRITE (BL8) to READ (BL8)



Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. t_{WTR} controls the WRITE-to-READ delay to the same device and starts with the first rising clock edge after the last write data shown at T9.
3. The BL8 setting is activated by either MR0[1:0] = 00 or MR0[1:0] = 01 and MR0[12] = 1 during the WRITE command at T0. The READ command at Ta0 can be either BC4 or BL8, depending on MR0[1:0] and the A12 status at Ta0.
4. DI n = data-in for column n .
5. RL = 5 (AL = 0, CL = 5), WL = 5 (AL = 0, CWL = 5).

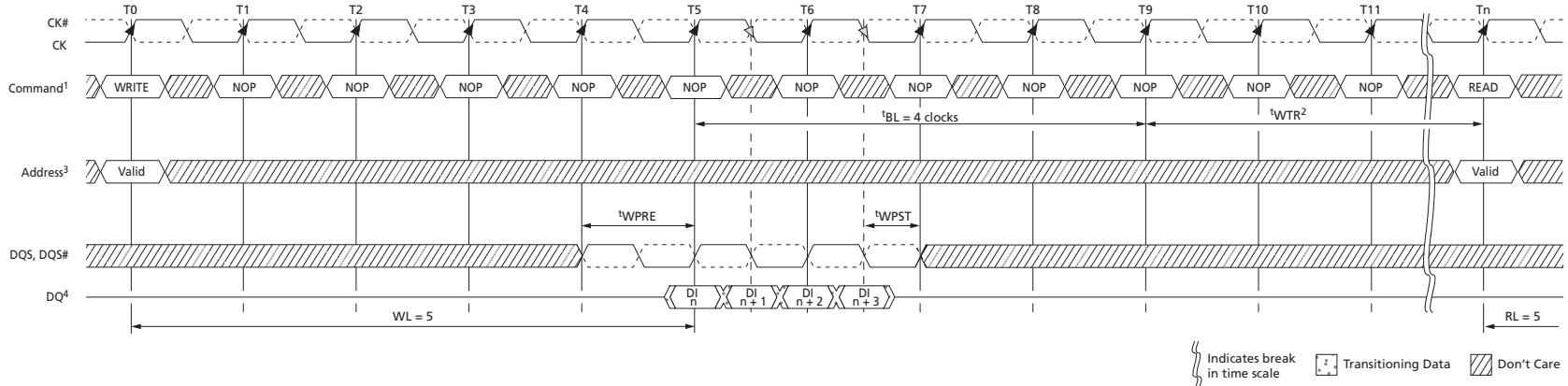
Figure 89: WRITE to READ (BC4 Mode Register Setting)



Notes:

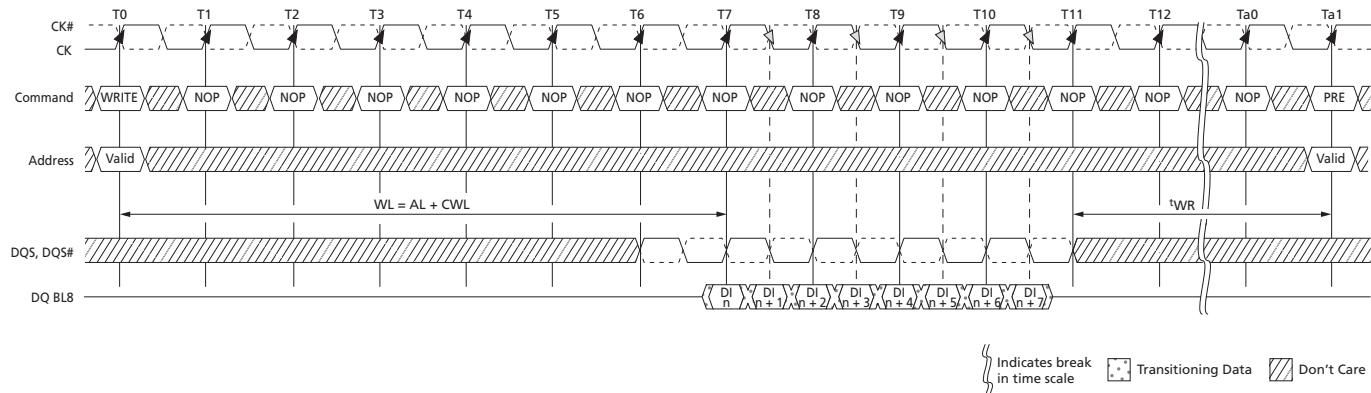
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. t_{WTR} controls the WRITE-to-READ delay to the same device and starts with the first rising clock edge after the last write data shown at T7.
3. The fixed BC4 setting is activated by $MR0[1:0] = 10$ during the WRITE command at T0 and the READ command at Ta0.
4. DI n = data-in for column n .
5. BC4 (fixed), $WL = 5$ ($AL = 0$, $CWL = 5$), $RL = 5$ ($AL = 0$, $CL = 5$).

Figure 90: WRITE (BC4 OTF) to READ (BC4 OTF)



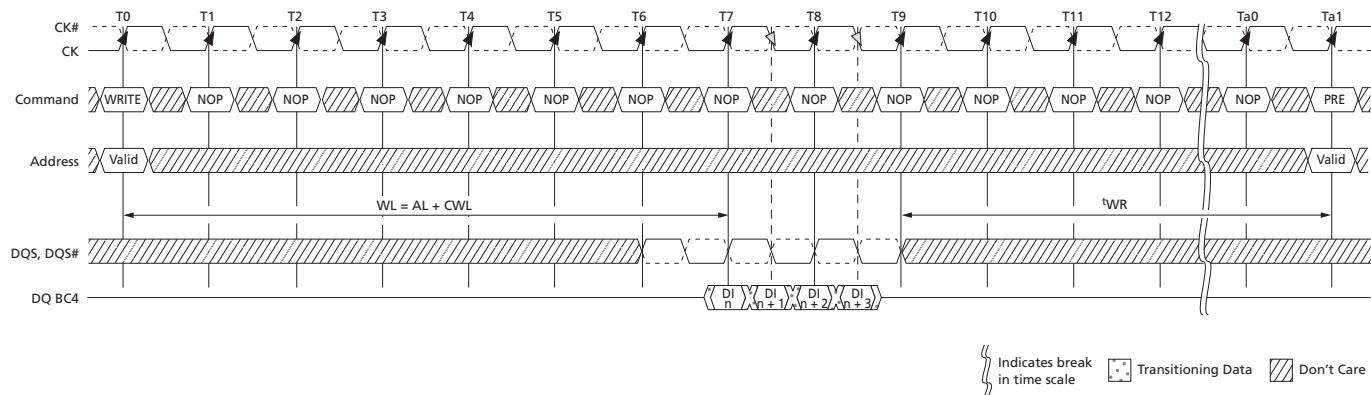
Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. t_{WTR} controls the WRITE-to-READ delay to the same device and starts after t_{BL} .
3. The BC4 OTF setting is activated by MRO[1:0] = 01 and A12 = 0 during the WRITE command at T0 and the READ command at T_n .
4. DI n = data-in for column n .
5. BC4, RL = 5 (AL = 0, CL = 5), WL = 5 (AL = 0, CWL = 5).

Figure 91: WRITE (BL8) to PRECHARGE


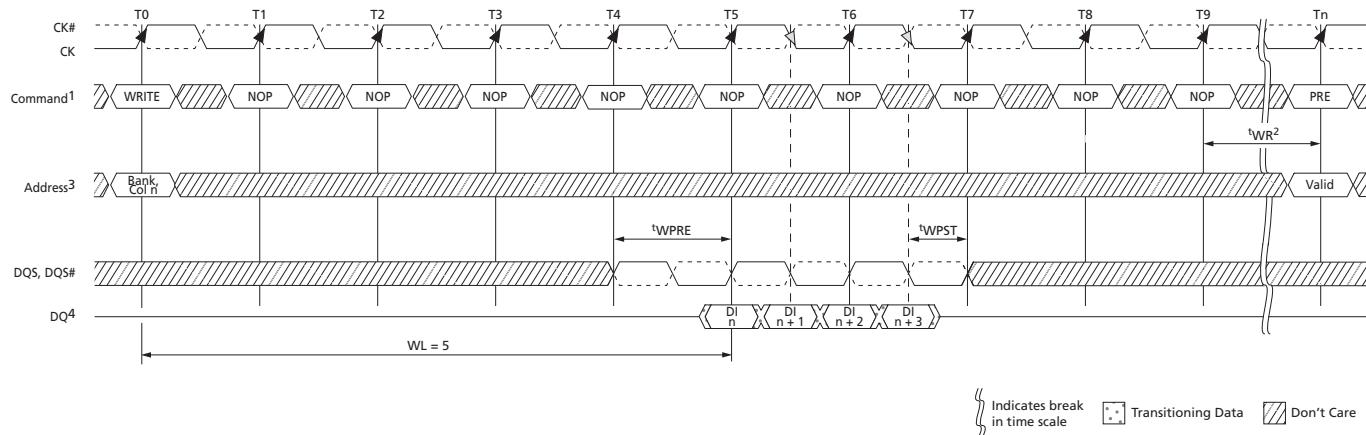
Notes:

1. DI n = data-in from column n .
2. Seven subsequent elements of data-in are applied in the programmed order following DO n .
3. Shown for WL = 7 (AL = 0, CWL = 7).

Figure 92: WRITE (BC4 Mode Register Setting) to PRECHARGE


Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The write recovery time (t_{WR}) is referenced from the first rising clock edge after the last write data is shown at T7. t_{WR} specifies the last burst WRITE cycle until the PRECHARGE command can be issued to the same bank.
3. The fixed BC4 setting is activated by MR0[1:0] = 10 during the WRITE command at T0.
4. DI n = data-in for column n .
5. BC4 (fixed), WL = 5, RL = 5.

Figure 93: WRITE (BC4 OTF) to PRECHARGE


Notes:

1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
2. The write recovery time (t^{WR}) is referenced from the rising clock edge at T9. t^{WR} specifies the last burst WRITE cycle until the PRECHARGE command can be issued to the same bank.
3. The BC4 setting is activated by MRO[1:0] = 01 and A12 = 0 during the WRITE command at T0.
4. DI n = data-in for column n .
5. BC4 (OTF), WL = 5, RL = 5.

DQ Input Timing

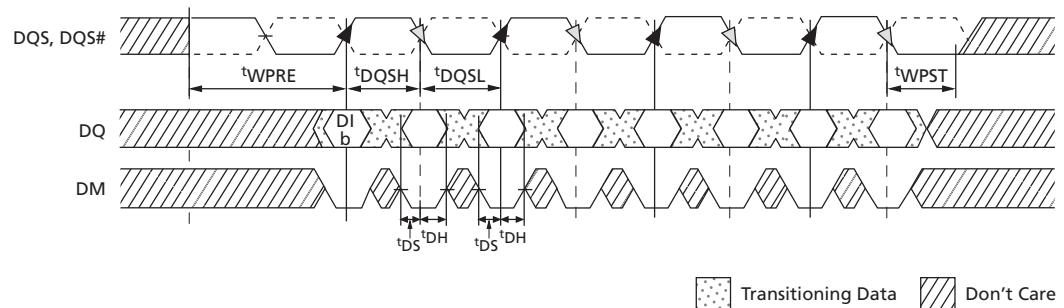
Figure 84 (page 170) shows the strobe-to-clock timing during a WRITE burst. DQS, DQS# must transition within $0.25t^{CK}$ of the clock transitions, as limited by t^{DQSS} . All data and data mask setup and hold timings are measured relative to the DQS, DQS# crossing, not the clock crossing.

The WRITE preamble and postamble are also shown in Figure 84 (page 170). One clock prior to data input to the DRAM, DQS must be HIGH and DQS# must be LOW. Then for a half clock, DQS is driven LOW (DQS# is driven HIGH) during the WRITE preamble, t^{WPRE} . Likewise, DQS must be kept LOW by the controller after the last data is written to the DRAM during the WRITE postamble, t^{WPST} .

Data setup and hold times are also shown in Figure 84 (page 170). All setup and hold times are measured from the crossing points of DQS and DQS#. These setup and hold values pertain to data input and data mask input.

Additionally, the half period of the data input strobe is specified by t^{DQSH} and t^{DQSL} .

Figure 94: Data Input Timing



PRECHARGE Operation

Input A10 determines whether one bank or all banks are to be precharged and, in the case where only one bank is to be precharged, inputs BA[2:0] select the bank.

When all banks are to be precharged, inputs BA[2:0] are treated as “Don’t Care.” After a bank is precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued.

SELF REFRESH Operation

The SELF REFRESH operation is initiated like a REFRESH command except CKE is LOW. The DLL is automatically disabled upon entering SELF REFRESH and is automatically enabled and reset upon exiting SELF REFRESH.

All power supply inputs (including V_{REFCA} and V_{REFDQ}) must be maintained at valid levels upon entry/exit and during self refresh mode operation. V_{REFDQ} may float or not drive $V_{DDQ}/2$ while in self refresh mode under certain conditions:

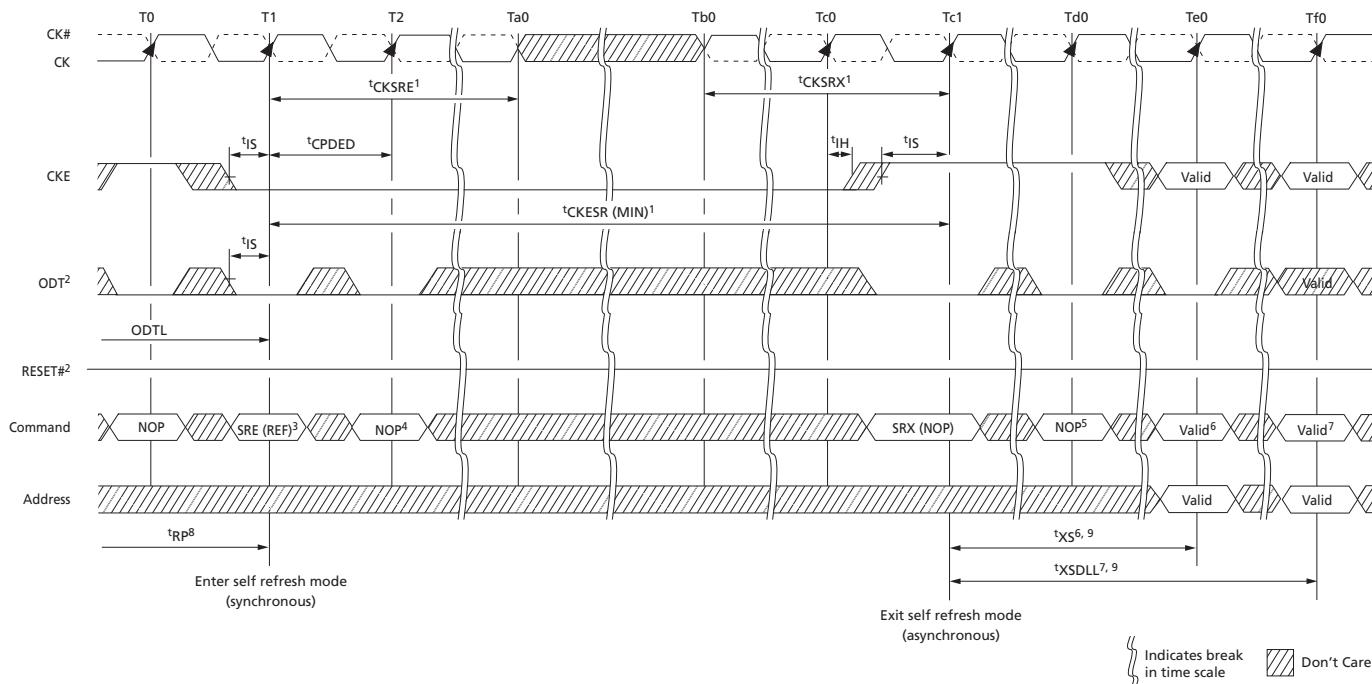
- $V_{SS} < V_{REFDQ} < V_{DD}$ is maintained.
- V_{REFDQ} is valid and stable prior to CKE going back HIGH.
- The first WRITE operation may not occur earlier than 512 clocks after V_{REFDQ} is valid.
- All other self refresh mode exit timing requirements are met.

The DRAM must be idle with all banks in the precharge state (t_{RP} is satisfied and no bursts are in progress) before a self refresh entry command can be issued. ODT must also be turned off before self refresh entry by registering the ODT ball LOW prior to the self refresh entry command (see On-Die Termination (ODT) (page 191) for timing requirements). If $R_{TT,nom}$ and $R_{TT(WR)}$ are disabled in the mode registers, ODT can be a “Don’t Care.” After the self refresh entry command is registered, CKE must be held LOW to keep the DRAM in self refresh mode.

After the DRAM has entered self refresh mode, all external control signals, except CKE and RESET#, are “Don’t Care.” The DRAM initiates a minimum of one REFRESH command internally within the t_{CKE} period when it enters self refresh mode.

The requirements for entering and exiting self refresh mode depend on the state of the clock during self refresh mode. First and foremost, the clock must be stable (meeting t_{CK} specifications) when self refresh mode is entered. If the clock remains stable and the frequency is not altered while in self refresh mode, then the DRAM is allowed to exit self refresh mode after t_{CKESR} is satisfied (CKE is allowed to transition HIGH t_{CKESR} later than when CKE was registered LOW). Since the clock remains stable in self refresh mode (no frequency change), t_{CKSRE} and t_{CKSRX} are not required. However, if the clock is altered during self refresh mode (if it is turned-off or its frequency changes), then t_{CKSRE} and t_{CKSRX} must be satisfied. When entering self refresh mode, t_{CKSRE} must be satisfied prior to altering the clock's frequency. Prior to exiting self refresh mode, t_{CKSRX} must be satisfied prior to registering CKE HIGH.

When CKE is HIGH during self refresh exit, NOP or DES must be issued for t_{XS} time. t_{XS} is required for the completion of any internal refresh already in progress and must be satisfied before a valid command not requiring a locked DLL can be issued to the device. t_{XS} is also the earliest time self refresh re-entry may occur. Before a command requiring a locked DLL can be applied, a ZQCL command must be issued, t_{ZQOPER} timing must be met, and t_{XSDL} must be satisfied. ODT must be off during t_{XSDL} .

Figure 95: Self Refresh Entry/Exit Timing


Notes:

1. The clock must be valid and stable, meeting t_{CK} specifications at least t_{CKSRE} after entering self refresh mode, and at least t_{CKSRX} prior to exiting self refresh mode, if the clock is stopped or altered between states $Ta0$ and $Tb0$. If the clock remains valid and unchanged from entry and during self refresh mode, then t_{CKSRE} and t_{CKSRX} do not apply; however, t_{CKESR} must be satisfied prior to exiting at SRX.
2. ODT must be disabled and R_{TT} off prior to entering self refresh at state $T1$. If both $R_{TT,nom}$ and $R_{TT(WR)}$ are disabled in the mode registers, ODT can be a "Don't Care."
3. Self refresh entry (SRE) is synchronous via a REFRESH command with CKE LOW.
4. A NOP or DES command is required at $T2$ after the SRE command is issued prior to the inputs becoming "Don't Care."
5. NOP or DES commands are required prior to exiting self refresh mode until state $Te0$.
6. t_{XS} is required before any commands not requiring a locked DLL.
7. t_{XSDLL} is required before any commands requiring a locked DLL.
8. The device must be in the all banks idle state prior to entering self refresh mode. For example, all banks must be precharged, t_{RP} must be met, and no data bursts can be in progress.
9. Self refresh exit is asynchronous; however, t_{XS} and t_{XSDLL} timings start at the first rising clock edge where CKE HIGH satisfies t_{ISXR} at $Tc1$. t_{CKSRX} timing is also measured so that t_{ISXR} is satisfied at $Tc1$.

Extended Temperature Usage

Micron's DDR3 SDRAM support the optional extended case temperature (T_C) range of 0°C to 95°C. Thus, the SRT and ASR options must be used at a minimum.

The extended temperature range DRAM must be refreshed externally at 2x (double refresh) anytime the case temperature is above 85°C (and does not exceed 95°C). The external refresh requirement is accomplished by reducing the refresh period from 64ms to 32ms. However, self refresh mode requires either ASR or SRT to support the extended temperature. Thus, either ASR or SRT must be enabled when T_C is above 85°C or self refresh cannot be used until T_C is at or below 85°C. Table 78 summarizes the two extended temperature options and Table 79 summarizes how the two extended temperature options relate to one another.

Table 78: Self Refresh Temperature and Auto Refresh Description

Field	MR2 Bits	Description
Self Refresh Temperature (SRT)		
SRT	7	If ASR is disabled (MR2[6] = 0), SRT must be programmed to indicate T_{OPER} during self refresh: *MR2[7] = 0: Normal operating temperature range (0°C to 85°C) *MR2[7] = 1: Extended operating temperature range (0°C to 95°C) If ASR is enabled (MR2[7] = 1), SRT must be set to 0, even if the extended temperature range is supported *MR2[7] = 0: SRT is disabled
Auto Self Refresh (ASR)		
ASR	6	When ASR is enabled, the DRAM automatically provides SELF REFRESH power management functions, (refresh rate for all supported operating temperature values) * MR2[6] = 1: ASR is enabled (M7 must = 0) When ASR is not enabled, the SRT bit must be programmed to indicate T_{OPER} during SELF REFRESH operation * MR2[6] = 0: ASR is disabled; must use manual self refresh temperature (SRT)

Table 79: Self Refresh Mode Summary

MR2[6] (ASR)	MR2[7] (SRT)	SELF REFRESH Operation	Permitted Operating Temperature Range for Self Refresh Mode
0	0	Self refresh mode is supported in the normal temperature range	Normal (0°C to 85°C)
0	1	Self refresh mode is supported in normal and extended temperature ranges; When SRT is enabled, it increases self refresh power consumption	Normal and extended (0°C to 95°C)
1	0	Self refresh mode is supported in normal and extended temperature ranges; Self refresh power consumption may be temperature-dependent	Normal and extended (0°C to 95°C)
1	1	Illegal	

Power-Down Mode

Power-down is synchronously entered when CKE is registered LOW coincident with a NOP or DES command. CKE is not allowed to go LOW while an MRS, MPR, ZQCAL, READ, or WRITE operation is in progress. CKE is allowed to go LOW while any of the other legal operations (such as ROW ACTIVATION, PRECHARGE, auto precharge, or REFRESH) are in progress. However, the power-down I_{DD} specifications are not applicable until such operations have completed. Depending on the previous DRAM state and the command issued prior to CKE going LOW, certain timing constraints must be satisfied (as noted in Table 80). Timing diagrams detailing the different power-down mode entry and exits are shown in Figure 96 (page 183) through Figure 105 (page 187).

Table 80: Command to Power-Down Entry Parameters

DRAM Status	Last Command Prior to CKE LOW ¹	Parameter (Min)	Parameter Value	Figure
Idle or active	ACTIVATE	$t_{ACTPDEN}$	1^{tCK}	Figure 103 (page 186)
Idle or active	PRECHARGE	t_{PRPDEN}	1^{tCK}	Figure 104 (page 187)
Active	READ or READAP	t_{RDPDEN}	$RL + 4^{tCK} + 1^{tCK}$	Figure 99 (page 184)
Active	WRITE: BL8OTF, BL8MRS, BC4OTF	t_{WRPDEN}	$WL + 4^{tCK} + t_{WR/tCK}$	Figure 100 (page 185)
			$WL + 2^{tCK} + t_{WR/tCK}$	Figure 100 (page 185)
Active	WRITERAP: BL8OTF, BL8MRS, BC4OTF	$t_{WRAPDEN}$	$WL + 4^{tCK} + WR + 1^{tCK}$	Figure 101 (page 185)
			$WL + 2^{tCK} + WR + 1^{tCK}$	Figure 101 (page 185)
Idle	REFRESH	$t_{REFPDEN}$	1^{tCK}	Figure 102 (page 186)
Power-down	REFRESH	t_{XPDLL}	Greater of 10^{tCK} or 24ns	Figure 106 (page 188)
Idle	MODE REGISTER SET	$t_{MRSPDEN}$	t_{MOD}	Figure 105 (page 187)

Note: 1. If slow-exit mode precharge power-down is enabled and entered, ODT becomes asynchronous t_{ANPD} prior to CKE going LOW and remains asynchronous until $t_{ANPD} + t_{XPDLL}$ after CKE goes HIGH.

Entering power-down disables the input and output buffers, excluding CK, CK#, ODT, CKE, and RESET#. NOP or DES commands are required until t_{CPDED} has been satisfied, at which time all specified input/output buffers are disabled. The DLL should be in a locked state when power-down is entered for the fastest power-down exit timing. If the DLL is not locked during power-down entry, the DLL must be reset after exiting power-down mode for proper READ operation as well as synchronous ODT operation.

During power-down entry, if any bank remains open after all in-progress commands are complete, the DRAM will be in active power-down mode. If all banks are closed after all in-progress commands are complete, the DRAM will be in precharge power-down mode. Precharge power-down mode must be programmed to exit with either a slow exit mode or a fast exit mode. When entering precharge power-down mode, the DLL is turned off in slow exit mode or kept on in fast exit mode.

The DLL also remains on when entering active power-down. ODT has special timing constraints when slow exit mode precharge power-down is enabled and entered. Refer to Asynchronous ODT Mode (page 204) for detailed ODT usage requirements in slow

exit mode precharge power-down. A summary of the two power-down modes is listed in Table 81 (page 182).

While in either power-down state, CKE is held LOW, RESET# is held HIGH, and a stable clock signal must be maintained. ODT must be in a valid state but all other input signals are “Don’t Care.” If RESET# goes LOW during power-down, the DRAM will switch out of power-down mode and go into the reset state. After CKE is registered LOW, CKE must remain LOW until $t_{PD}(\text{MIN})$ has been satisfied. The maximum time allowed for power-down duration is $t_{PD}(\text{MAX})$ ($9 \times t_{REFI}$).

The power-down states are synchronously exited when CKE is registered HIGH (with a required NOP or DES command). CKE must be maintained HIGH until t_{CKE} has been satisfied. A valid, executable command may be applied after power-down exit latency, t_{XP} , and t_{XPDLL} have been satisfied. A summary of the power-down modes is listed below.

For specific CKE-intensive operations, such as repeating a power-down-exit-to-refresh-to-power-down-entry sequence, the number of clock cycles between power-down exit and power-down entry may not be sufficient to keep the DLL properly updated. In addition to meeting t_{PD} when the REFRESH command is used between power-down exit and power-down entry, two other conditions must be met. First, t_{XP} must be satisfied before issuing the REFRESH command. Second, t_{XPDLL} must be satisfied before the next power-down may be entered. An example is shown in Figure 106 (page 188).

Table 81: Power-Down Modes

DRAM State	MR0[12]	DLL State	Power-Down Exit	Relevant Parameters
Active (any bank open)	“Don’t Care”	On	Fast	t_{XP} to any other valid command
Precharged (all banks precharged)	1	On	Fast	t_{XP} to any other valid command
	0	Off	Slow	t_{XPDLL} to commands that require the DLL to be locked (READ, RDAP, or ODT on); t_{XP} to any other valid command

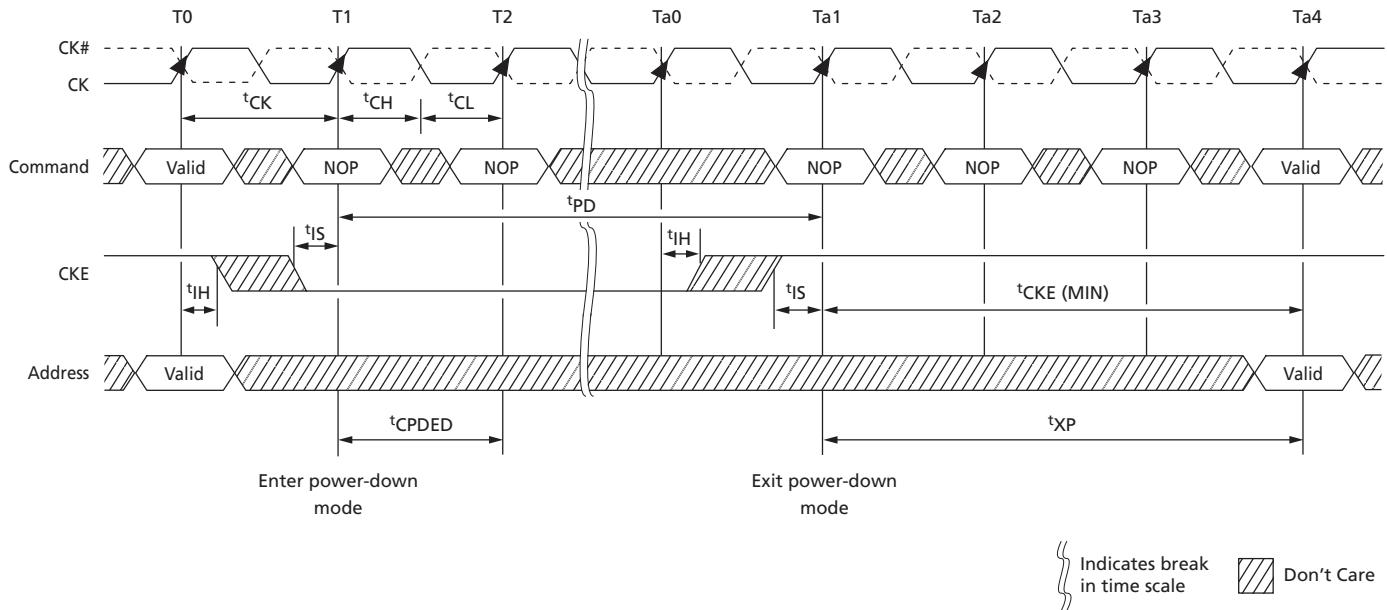
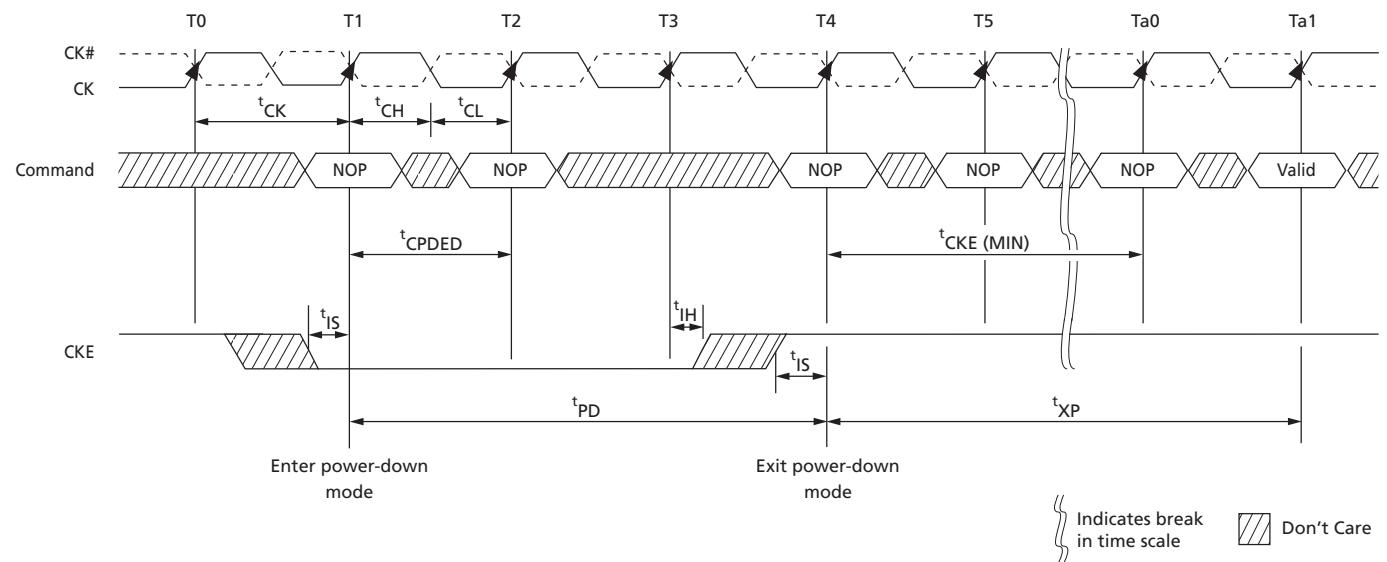
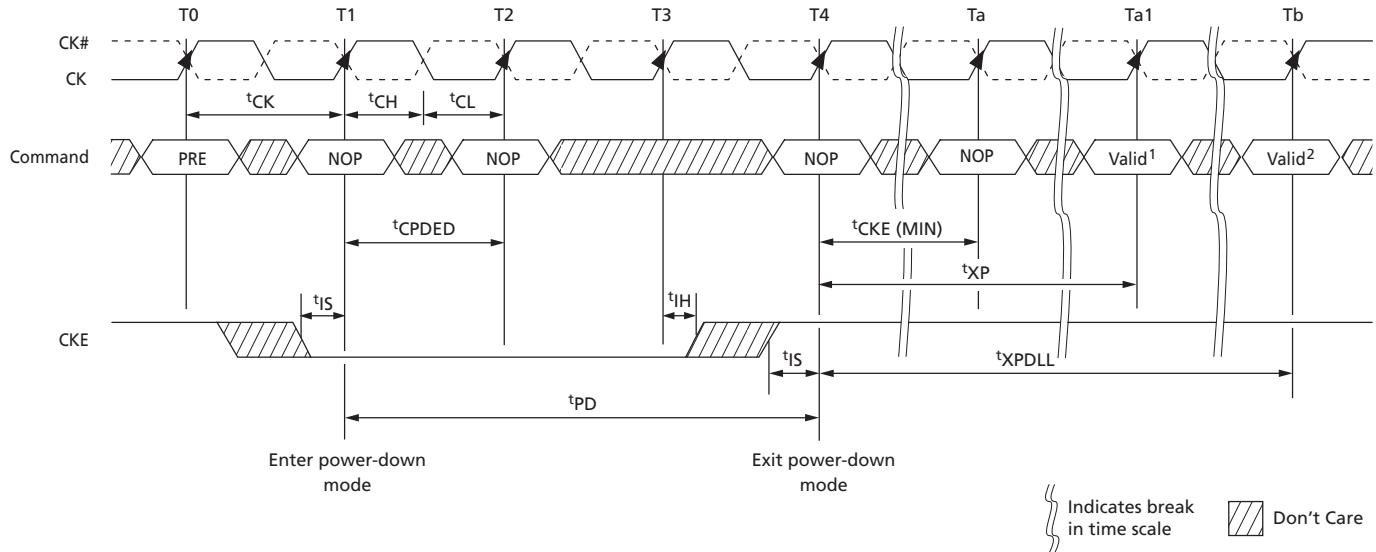
Figure 96: Active Power-Down Entry and Exit

Figure 97: Precharge Power-Down (Fast-Exit Mode) Entry and Exit


Figure 98: Precharge Power-Down (Slow-Exit Mode) Entry and Exit



Notes: 1. Any valid command not requiring a locked DLL.
2. Any valid command requiring a locked DLL.

Figure 99: Power-Down Entry After READ or READ with Auto Precharge (RDAP)

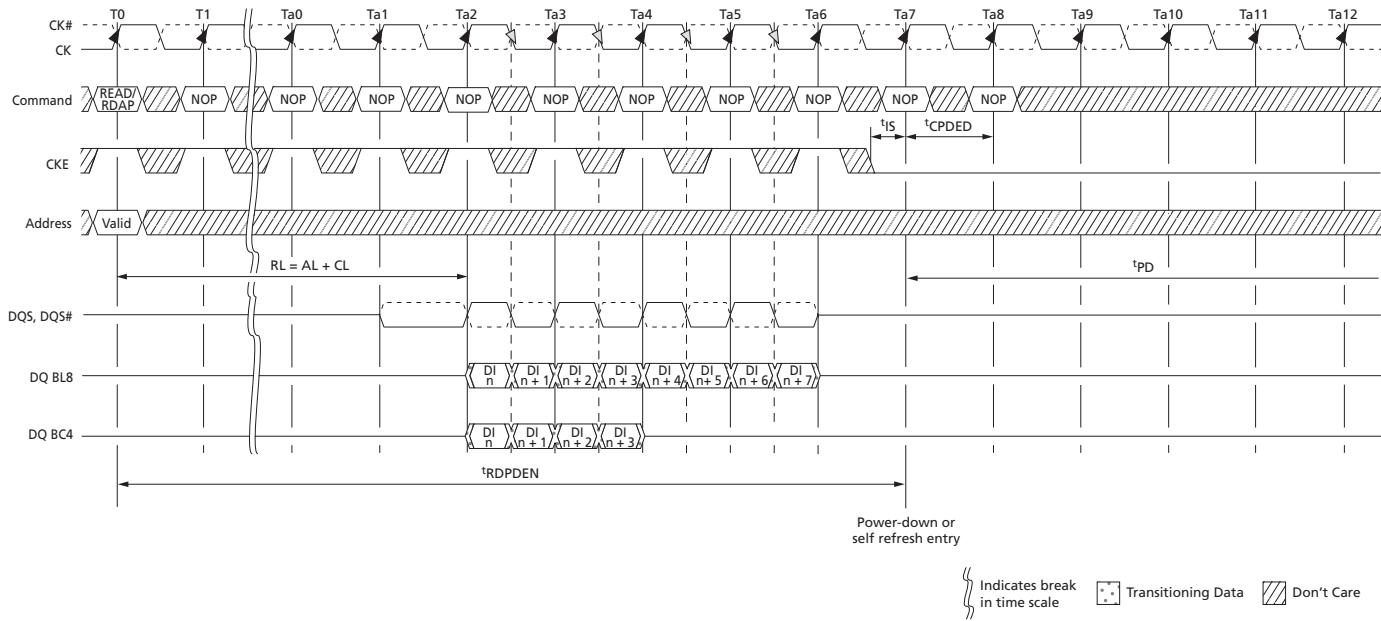
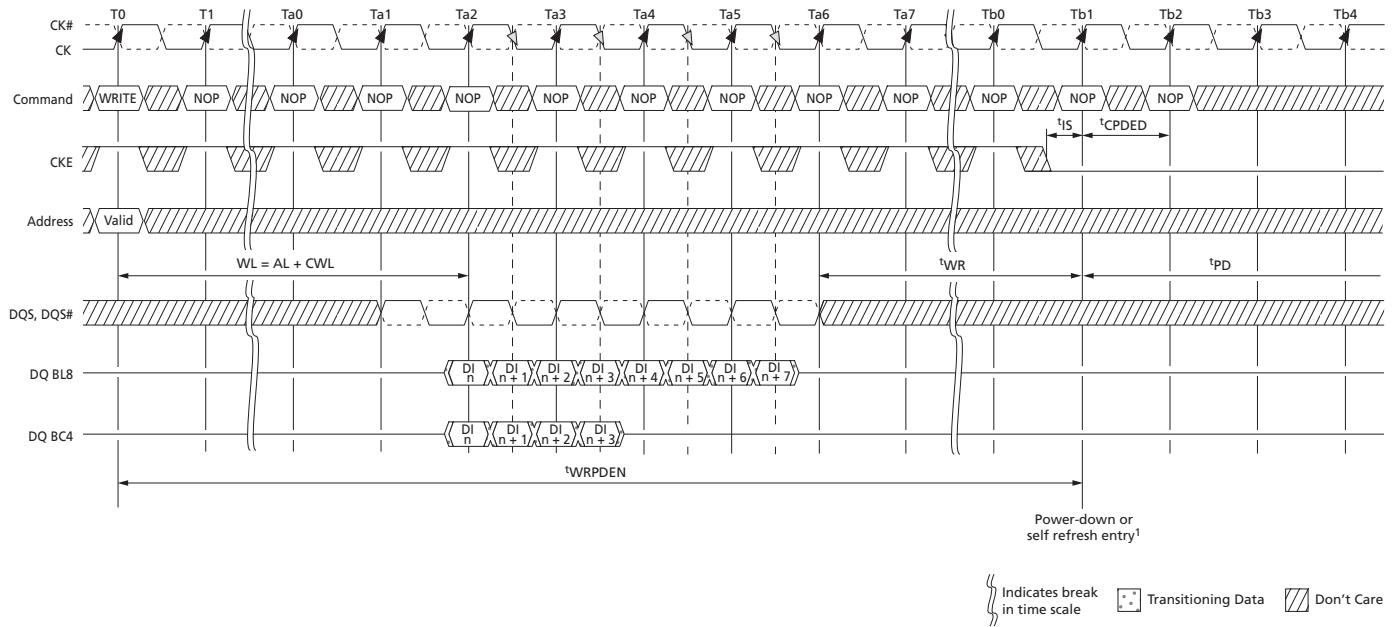
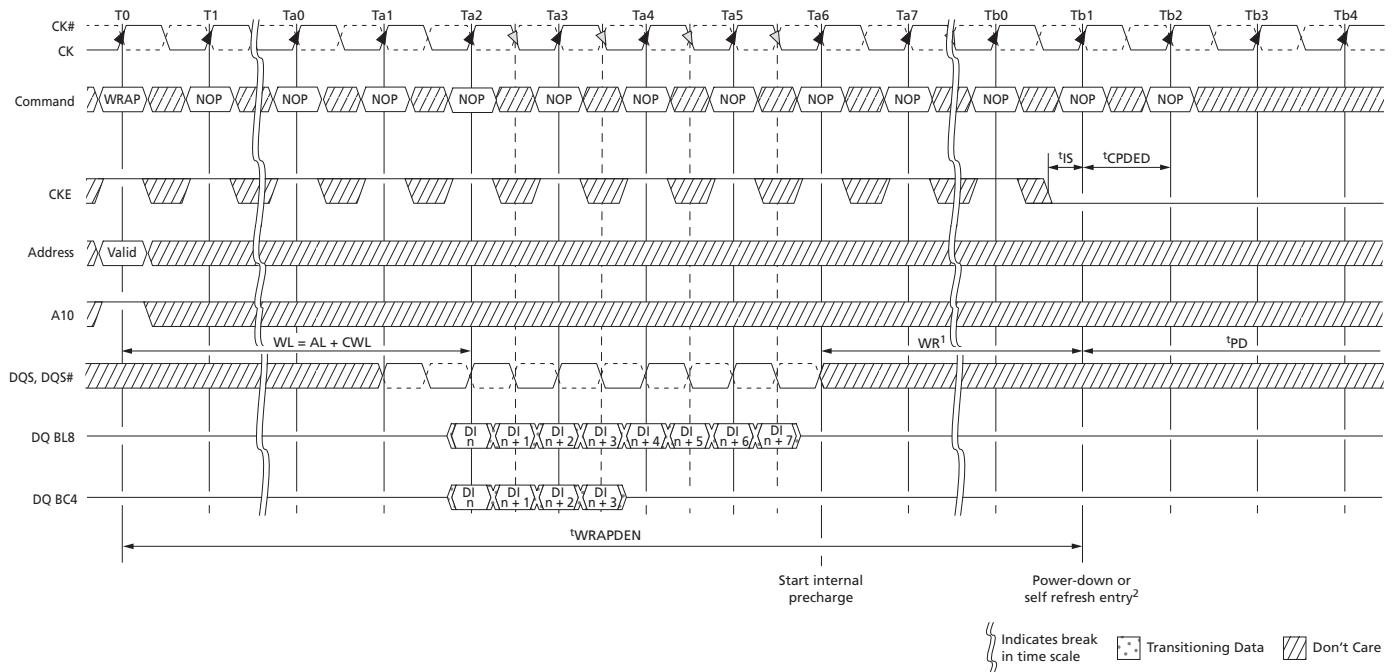
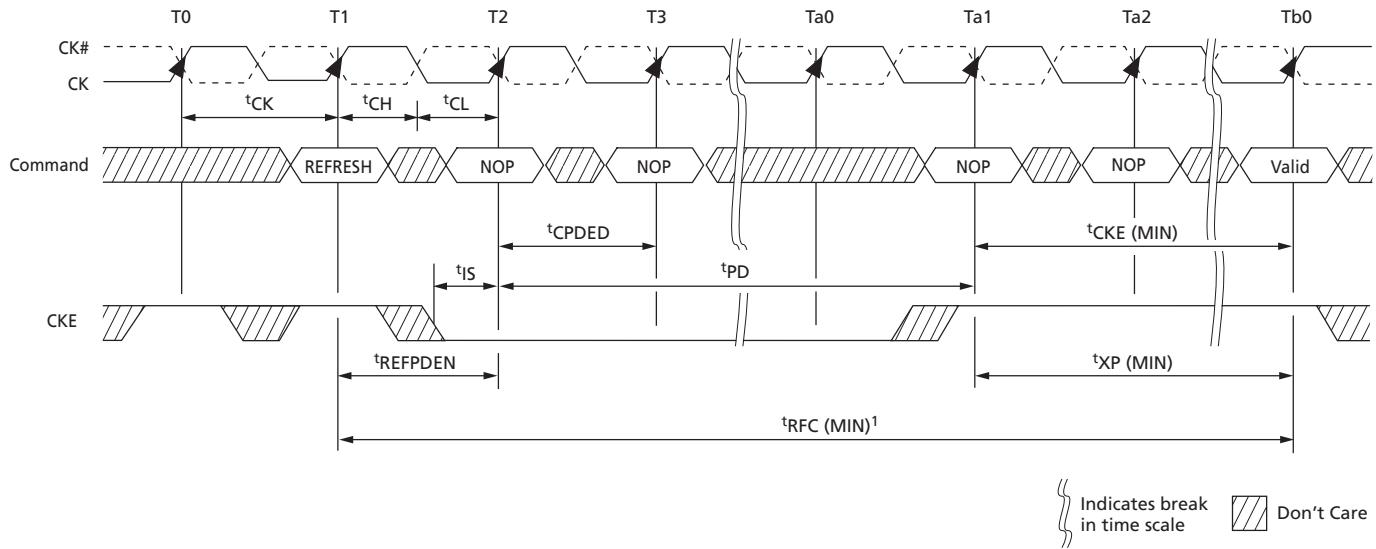


Figure 100: Power-Down Entry After WRITE


Note: 1. CKE can go LOW $2t_{CK}$ earlier if BC4MRS.

Figure 101: Power-Down Entry After WRITE with Auto Precharge (WRAP)


Notes: 1. t_{WR} is programmed through MR0[11:9] and represents t_{WRmin} (ns) t_{CK} rounded up to the next integer t_{CK} .
2. CKE can go LOW $2t_{CK}$ earlier if BC4MRS.

Figure 102: REFRESH to Power-Down Entry


Note: 1. After CKE goes HIGH during t_{RFC} , CKE must remain HIGH until t_{RFC} is satisfied.

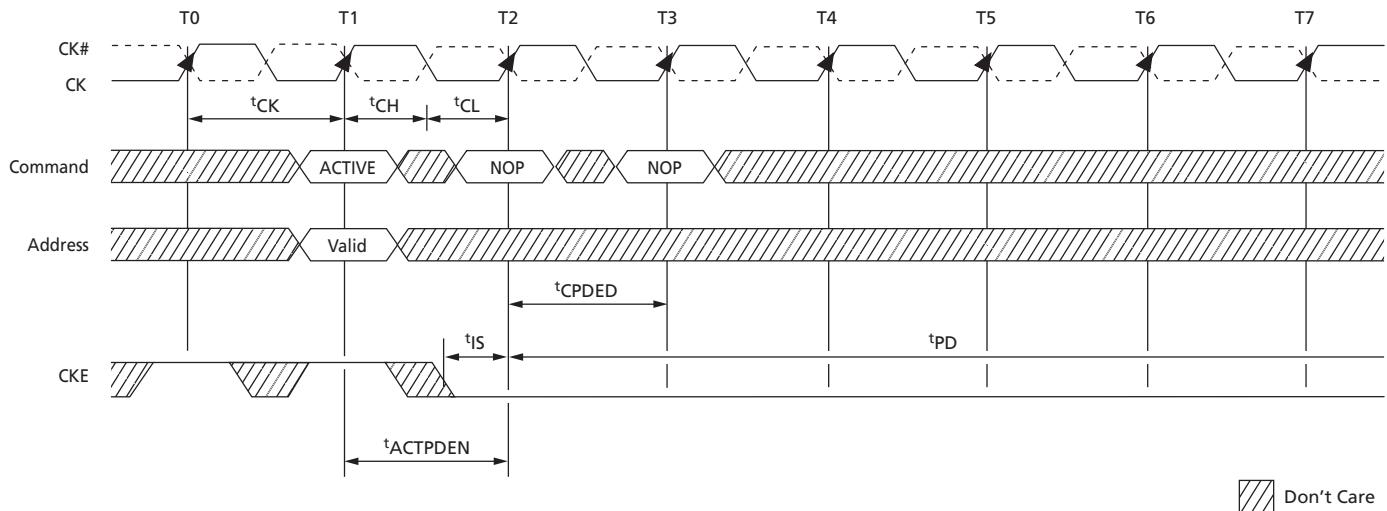
Figure 103: ACTIVATE to Power-Down Entry


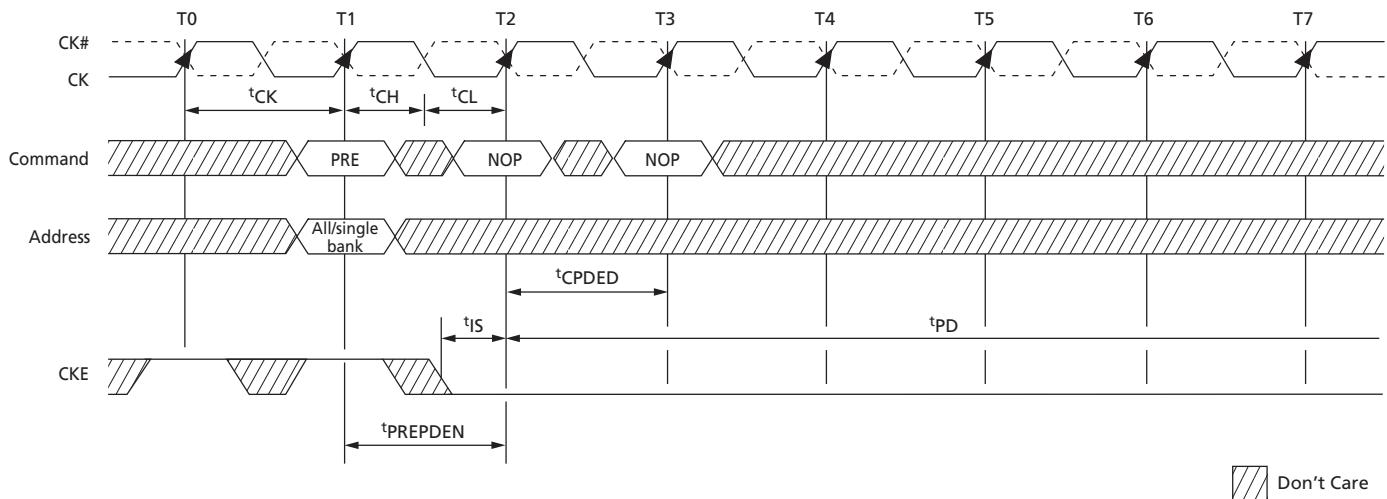
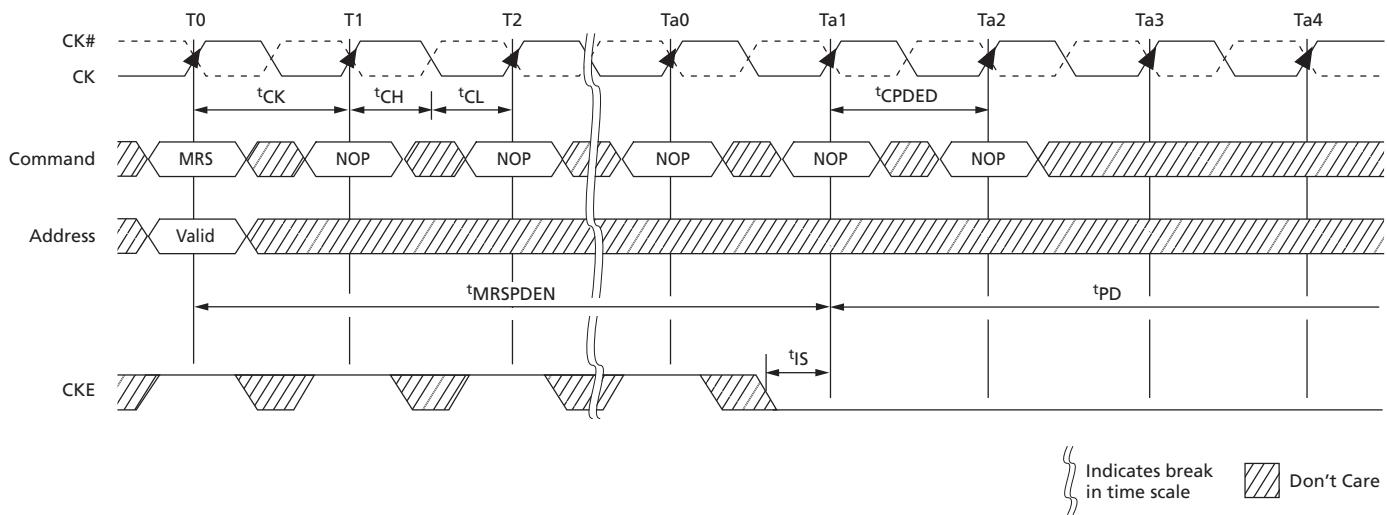
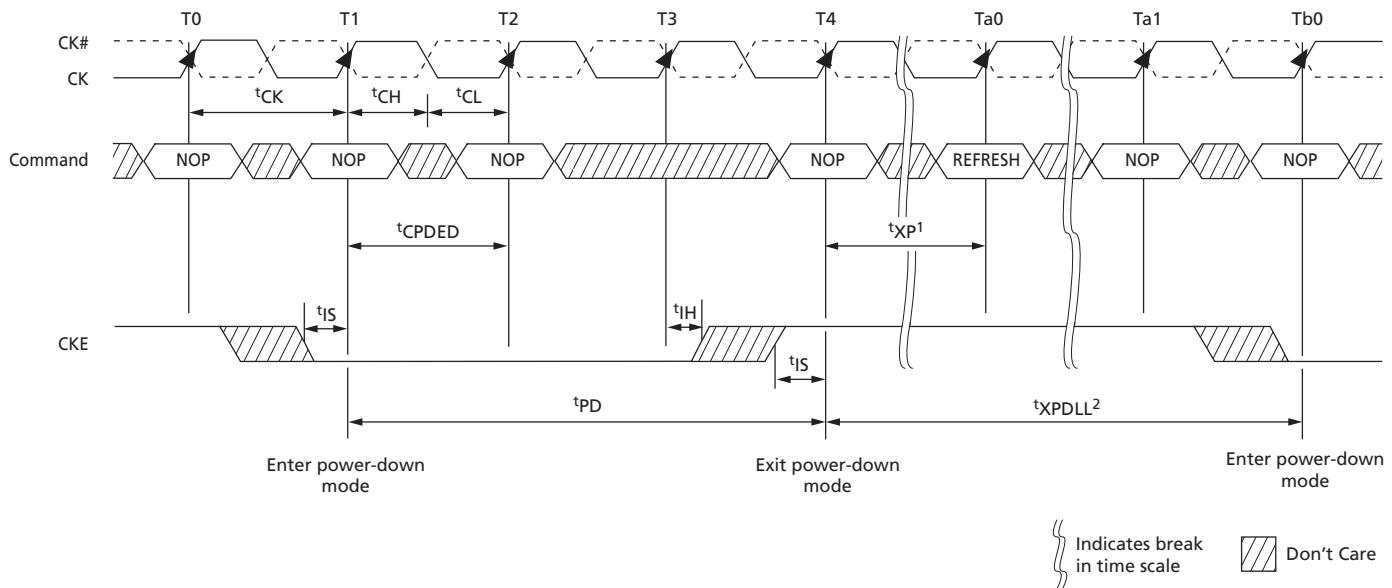
Figure 104: PRECHARGE to Power-Down Entry

Figure 105: MRS Command to Power-Down Entry


Figure 106: Power-Down Exit to Refresh to Power-Down Entry


Notes:

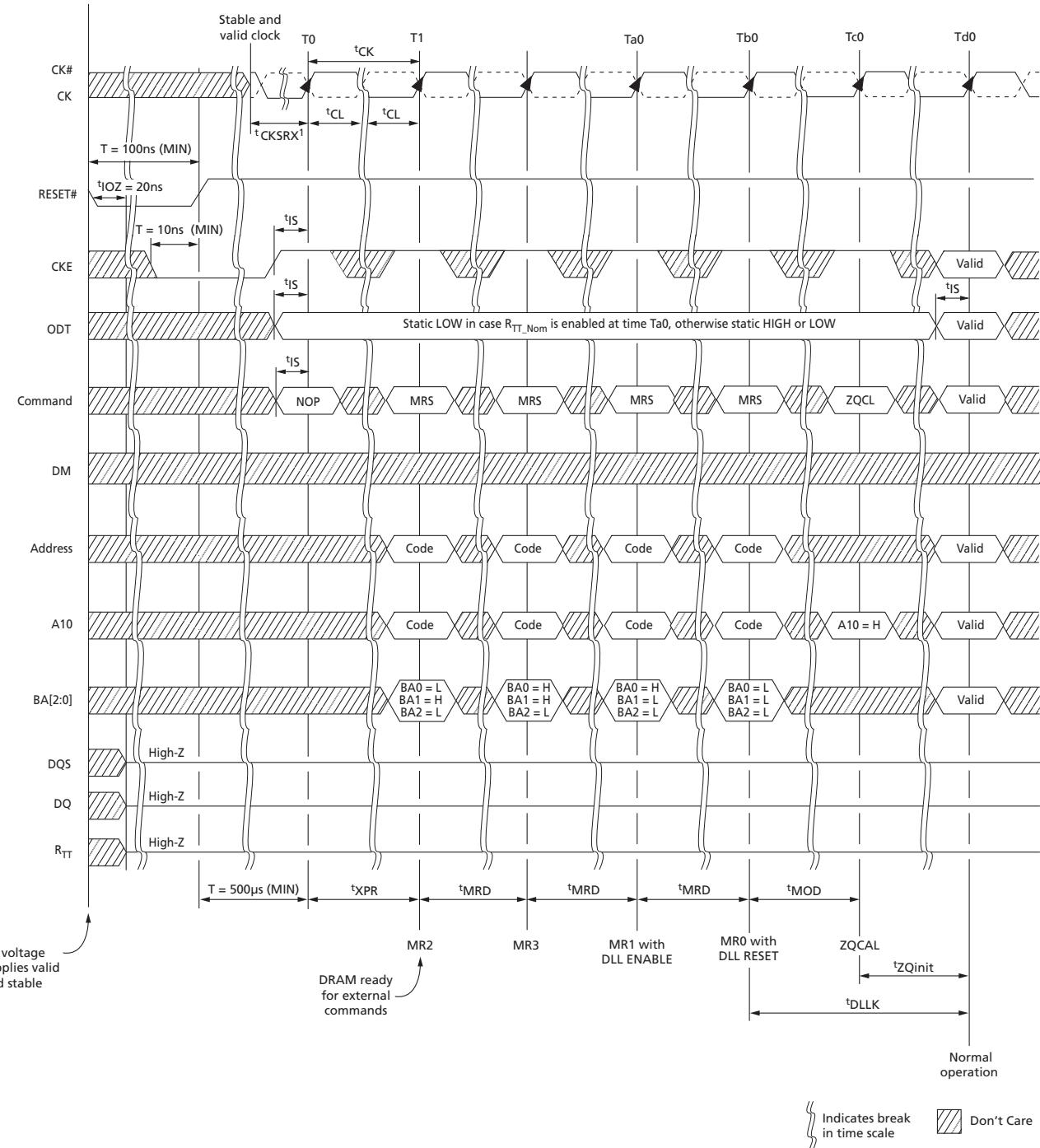
1. t_{XP} must be satisfied before issuing the command.
2. $t_{XP DLL}$ must be satisfied (referenced to the registration of power-down exit) before the next power-down can be entered.

RESET Operation

The RESET signal (RESET#) is an asynchronous reset signal that triggers any time it drops LOW, and there are no restrictions about when it can go LOW. After RESET# goes LOW, it must remain LOW for 100ns. During this time, the outputs are disabled, ODT (R_{TT}) turns off (High-Z), and the DRAM resets itself. CKE should be driven LOW prior to RESET# being driven HIGH. After RESET# goes HIGH, the DRAM must be re-initialized as though a normal power-up was executed. All refresh counters on the DRAM are reset, and data stored in the DRAM is assumed unknown after RESET# has gone LOW.

Figure 107: RESET Sequence

System RESET (warm boot)



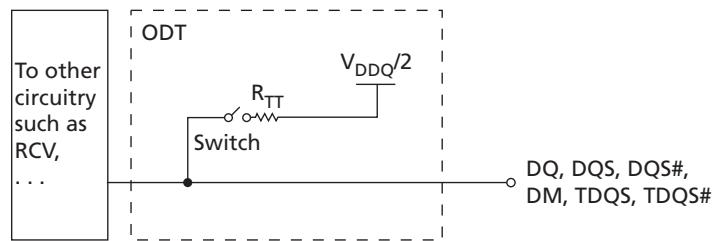
Note: 1. The minimum time required is the longer of 10ns or 5 clocks.

On-Die Termination (ODT)

On-die termination (ODT) is a feature that enables the DRAM to enable/disable and turn on/off termination resistance for each DQ, DQS, DQS#, and DM for the x4 and x8 configurations (and TDQS, TDQS# for the x8 configuration, when enabled). ODT is applied to each DQ, UDQS, UDQS#, LDQS, LDQS#, UDM, and LDM signal for the x16 configuration.

ODT is designed to improve signal integrity of the memory channel by enabling the DRAM controller to independently turn on/off the DRAM's internal termination resistance for any grouping of DRAM devices. ODT is not supported during DLL disable mode (simple functional representation shown below). The switch is enabled by the internal ODT control logic, which uses the external ODT ball and other control information.

Figure 108: On-Die Termination



Functional Representation of ODT

The value of R_{TT} (ODT termination resistance value) is determined by the settings of several mode register bits (see Table 87 (page 195)). The ODT ball is ignored while in self refresh mode (must be turned off prior to self refresh entry) or if mode registers MR1 and MR2 are programmed to disable ODT. ODT is comprised of nominal ODT and dynamic ODT modes and either of these can function in synchronous or asynchronous mode (when the DLL is off during precharge power-down or when the DLL is synchronizing). Nominal ODT is the base termination and is used in any allowable ODT state. Dynamic ODT is applied only during writes and provides OTF switching from no R_{TT} or $R_{TT,nom}$ to $R_{TT(WR)}$.

The actual effective termination, $R_{TT(EFF)}$, may be different from R_{TT} targeted due to nonlinearity of the termination. For $R_{TT(EFF)}$ values and calculations, see Table 32 (page 56).

Nominal ODT

ODT (NOM) is the base termination resistance for each applicable ball; it is enabled or disabled via MR1[9, 6, 2] (see Mode Register 1 (MR1) Definition), and it is turned on or off via the ODT ball.

Table 82: Truth Table – ODT (Nominal)

Note 1 applies to the entire table

MR1[9, 6, 2]	ODT Pin	DRAM Termination State	DRAM State	Notes
000	0	$R_{TT,nom}$ disabled, ODT off	Any valid	2
000	1	$R_{TT,nom}$ disabled, ODT on	Any valid except self refresh, read	3
000-101	0	$R_{TT,nom}$ enabled, ODT off	Any valid	2
000-101	1	$R_{TT,nom}$ enabled, ODT on	Any valid except self refresh, read	3
110 and 111	X	$R_{TT,nom}$ reserved, ODT on or off	Illegal	

Notes:

- Assumes dynamic ODT is disabled (see Dynamic ODT (page 193) when enabled).
- ODT is enabled and active during most writes for proper termination, but it is not illegal for it to be off during writes.
- ODT must be disabled during reads. The $R_{TT,nom}$ value is restricted during writes. Dynamic ODT is applicable if enabled.

Nominal ODT resistance $R_{TT,nom}$ is defined by MR1[9, 6, 2], as shown in Mode Register 1 (MR1) Definition. The $R_{TT,nom}$ termination value applies to the output pins previously mentioned. DDR3 SDRAM supports multiple $R_{TT,nom}$ values based on RZQ/n where n can be 2, 4, 6, 8, or 12 and RZQ is 240Ω . $R_{TT,nom}$ termination is allowed any time after the DRAM is initialized, calibrated, and not performing read access, or when it is not in self refresh mode.

Write accesses use $R_{TT,nom}$ if dynamic ODT ($R_{TT(WR)}$) is disabled. If $R_{TT,nom}$ is used during writes, only $RZQ/2$, $RZQ/4$, and $RZQ/6$ are allowed (see Table 86 (page 194)). ODT timings are summarized in Table 83 (page 192), as well as listed in the Electrical Characteristics and AC Operating Conditions table.

Examples of nominal ODT timing are shown in conjunction with the synchronous mode of operation in Synchronous ODT Mode (page 199).

Table 83: ODT Parameters

Symbol	Description	Begins at	Defined to	Definition for All DDR3L Speed Bins	Unit
ODTLon	ODT synchronous turn-on delay	ODT registered HIGH	$R_{TT(ON)} \pm t_{AON}$	CWL + AL - 2	t_{CK}
ODTloff	ODT synchronous turn-off delay	ODT registered HIGH	$R_{TT(OFF)} \pm t_{AOF}$	CWL + AL - 2	t_{CK}
t_{AONPD}	ODT asynchronous turn-on delay	ODT registered HIGH	$R_{TT(ON)}$	2-8.5	ns
t_{AOPD}	ODT asynchronous turn-off delay	ODT registered HIGH	$R_{TT(OFF)}$	2-8.5	ns
ODTH4	ODT minimum HIGH time after ODT assertion or write (BC4)	ODT registered HIGH or write registration with ODT HIGH	ODT registered LOW	$4t_{CK}$	t_{CK}
ODTH8	ODT minimum HIGH time after write (BL8)	Write registration with ODT HIGH	ODT registered LOW	$6t_{CK}$	t_{CK}
t_{AON}	ODT turn-on relative to ODTlon completion	Completion of ODTlon	$R_{TT(ON)}$	See Electrical Characteristics and AC Operating Conditions table	ps
t_{AOF}	ODT turn-off relative to ODTloff completion	Completion of ODTloff	$R_{TT(OFF)}$	$0.5t_{CK} \pm 0.2t_{CK}$	t_{CK}

Dynamic ODT

In certain application cases, and to further enhance signal integrity on the data bus, it is desirable that the termination strength of the DDR3 SDRAM can be changed without issuing an MRS command, essentially changing the ODT termination on the fly. With dynamic ODT $R_{TT(WR)}$ enabled, the DRAM switches from nominal ODT $R_{TT,nom}$ to dynamic ODT $R_{TT(WR)}$ when beginning a WRITE burst and subsequently switches back to nominal ODT $R_{TT,nom}$ at the completion of the WRITE burst. This requirement is supported by the dynamic ODT feature, as described below.

Dynamic ODT Special Use Case

When DDR3 devices are architect as a single rank memory array, dynamic ODT offers a special use case: the ODT ball can be wired high (via a current limiting resistor preferred) by having $R_{TT,nom}$ disabled via MR1 and $R_{TT(WR)}$ enabled via MR2. This will allow the ODT signal not to have to be routed yet the DRAM can provide ODT coverage during write accesses.

When enabling this special use case, some standard ODT spec conditions may be violated: ODT is sometimes suppose to be held low. Such ODT spec violation (ODT not LOW) is allowed under this special use case. Most notably, if Write Leveling is used, this would appear to be a problem since $R_{TT(WR)}$ can not be used (should be disabled) and $R_{TT(NOM)}$ should be used. For Write leveling during this special use case, with the DLL locked, then $R_{TT(NOM)}$ maybe enabled when entering Write Leveling mode and disabled when exiting Write Leveling mode. More so, $R_{TT(NOM)}$ must be enabled when enabling Write Leveling, via same MR1 load, and disabled when disabling Write Leveling, via same MR1 load if $R_{TT(NOM)}$ is to be used.

ODT will turn-on within a delay of ODT_{On} + t_{AON} + t_{MOD} + 1CK (enabling via MR1) or turn-off within a delay of ODT_{Off} + t_{AOF} + t_{MOD} + 1CK. As seen in the table below, between the Load Mode of MR1 and the previously specified delay, the value of ODT is uncertain. this means the DQ ODT termination could turn-on and then turn-off again during the period of stated uncertainty.

Table 84: Write Leveling with Dynamic ODT Special Case

Begin $R_{TT,nom}$ Uncertainty	End $R_{TT,nom}$ Uncertainty	I/Os	$R_{TT,nom}$ Final State
MR1 load mode command: Enable Write Leveling and $R_{TT(NOM)}$	ODT _{On} + t_{AON} + t_{MOD} + 1CK	DQS, DQS#	Drive $R_{TT,nom}$ value
		DQs	No $R_{TT,nom}$
MR1 load mode command: Disable Write Leveling and $R_{TT(NOM)}$	ODT _{Off} + t_{AOFF} + t_{MOD} + 1CK	DQS, DQS#	No $R_{TT,nom}$
		DQs	No $R_{TT,nom}$

Functional Description

The dynamic ODT mode is enabled if either MR2[9] or MR2[10] is set to 1. Dynamic ODT is not supported during DLL disable mode so $R_{TT(WR)}$ must be disabled. The dynamic ODT function is described below:

- Two R_{TT} values are available— $R_{TT,nom}$ and $R_{TT(WR)}$.
 - The value for $R_{TT,nom}$ is preselected via MR1[9, 6, 2].
 - The value for $R_{TT(WR)}$ is preselected via MR2[10, 9].

- During DRAM operation without READ or WRITE commands, the termination is controlled.
 - Nominal termination strength $R_{TT,nom}$ is used.
 - Termination on/off timing is controlled via the ODT ball and latencies ODTLon and ODTLoff.
- When a WRITE command (WR, WRAP, WRS4, WRS8, WRAPS4, WRAPS8) is registered, and if dynamic ODT is enabled, the ODT termination is controlled.
 - A latency of ODTLcnw after the WRITE command: termination strength $R_{TT,nom}$ switches to $R_{TT(WR)}$
 - A latency of ODTLcwn8 (for BL8, fixed or OTF) or ODTLcwn4 (for BC4, fixed or OTF) after the WRITE command: termination strength $R_{TT(WR)}$ switches back to $R_{TT,nom}$.
 - On/off termination timing is controlled via the ODT ball and determined by ODT-Lon, ODTLoff, ODTH4, and ODTH8.
 - During the t_{ADC} transition window, the value of R_{TT} is undefined.

ODT is constrained during writes and when dynamic ODT is enabled (see Table 85 (page 194)). ODT timings listed in Table 83 (page 192) also apply to dynamic ODT mode.

Table 85: Dynamic ODT Specific Parameters

Symbol	Description	Begins at	Defined to	Definition for All DDR3L Speed Bins	Unit
ODTLcnw	Change from $R_{TT,nom}$ to $R_{TT(WR)}$	Write registration	R_{TT} switched from $R_{TT,nom}$ to $R_{TT(WR)}$	WL - 2	t_{CK}
ODTLcwn4	Change from $R_{TT(WR)}$ to $R_{TT,nom}$ (BC4)	Write registration	R_{TT} switched from $R_{TT(WR)}$ to $R_{TT,nom}$	$4t_{CK} + ODTL$ off	t_{CK}
ODTLcwn8	Change from $R_{TT(WR)}$ to $R_{TT,nom}$ (BL8)	Write registration	R_{TT} switched from $R_{TT(WR)}$ to $R_{TT,nom}$	$6t_{CK} + ODTL$ off	t_{CK}
t_{ADC}	R_{TT} change skew	ODTLcnw completed	R_{TT} transition complete	$0.5t_{CK} \pm 0.2t_{CK}$	t_{CK}

Table 86: Mode Registers for $R_{TT,nom}$

MR1 ($R_{TT,nom}$)			$R_{TT,nom}$ (RZQ)	$R_{TT,nom}$ (Ohm)	$R_{TT,nom}$ Mode Restriction
M9	M6	M2			
0	0	0	Off	Off	n/a
0	0	1	RZQ/4	60	Self refresh
0	1	0	RZQ/2	120	
0	1	1	RZQ/6	40	
1	0	0	RZQ/12	20	Self refresh, write
1	0	1	RZQ/8	30	
1	1	0	Reserved	Reserved	n/a
1	1	1	Reserved	Reserved	n/a

Note: 1. $R_{ZQ} = 240\Omega$. If $R_{TT,nom}$ is used during WRITEs, only RZQ/2, RZQ/4, RZQ/6 are allowed.

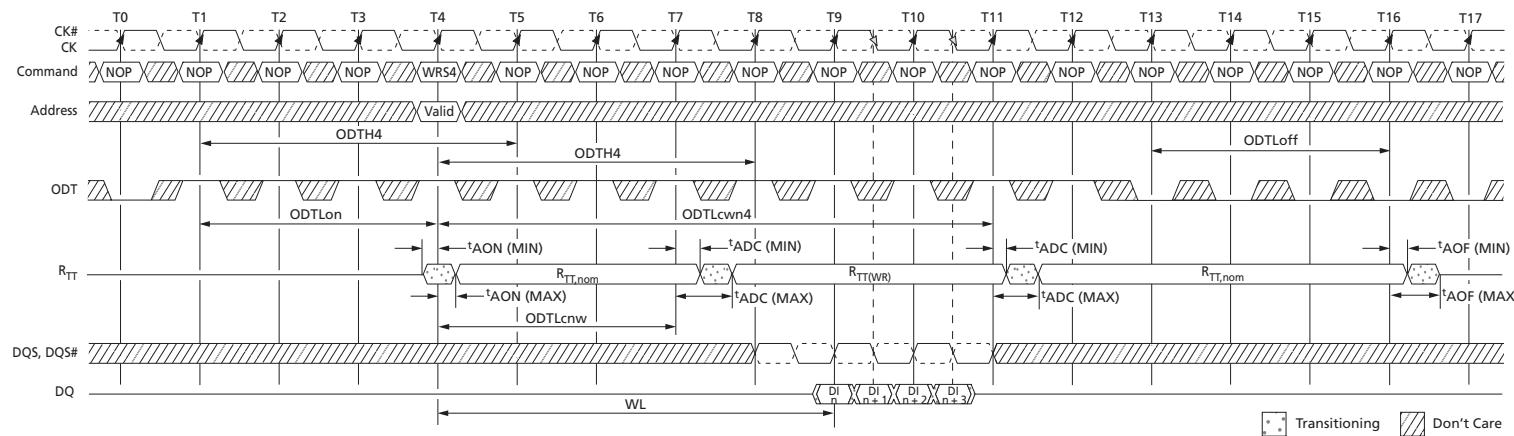
Table 87: Mode Registers for $R_{TT(WR)}$

MR2 ($R_{TT(WR)}$)		$R_{TT(WR)} (RZQ)$	$R_{TT(WR)} (\text{Ohm})$
M10	M9		
0	0	Dynamic ODT off: WRITE does not affect $R_{TT,nom}$	
0	1	RZQ/4	60
1	0	RZQ/2	120
1	1	Reserved	Reserved

Table 88: Timing Diagrams for Dynamic ODT

Figure and Page	Title
Figure 109 (page 196)	Dynamic ODT: ODT Asserted Before and After the WRITE, BC4
Figure 110 (page 196)	Dynamic ODT: Without WRITE Command
Figure 111 (page 197)	Dynamic ODT: ODT Pin Asserted Together with WRITE Command for 6 Clock Cycles, BL8
Figure 112 (page 198)	Dynamic ODT: ODT Pin Asserted with WRITE Command for 6 Clock Cycles, BC4
Figure 113 (page 198)	Dynamic ODT: ODT Pin Asserted with WRITE Command for 4 Clock Cycles, BC4

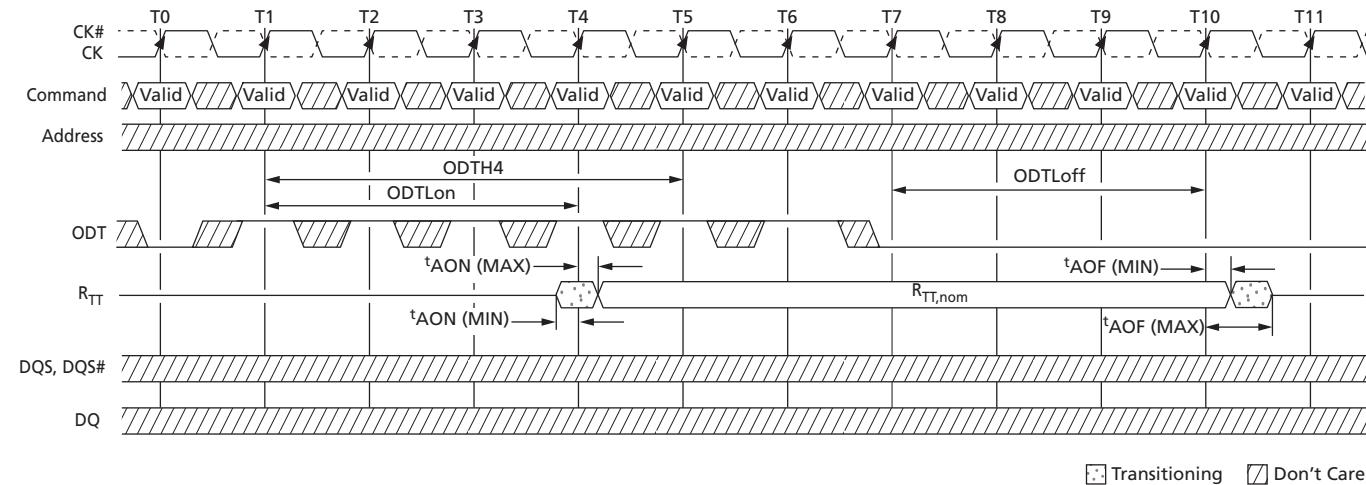
Figure 109: Dynamic ODT: ODT Asserted Before and After the WRITE, BC4



Notes:

1. Via MRS or OTF. AL = 0, CWL = 5. R_{TT,nom} and R_{TT(WR)} are enabled.
2. ODTH4 applies to first registering ODT HIGH and then to the registration of the WRITE command. In this example, ODTH4 is satisfied if ODT goes LOW at T8 (four clocks after the WRITE command).

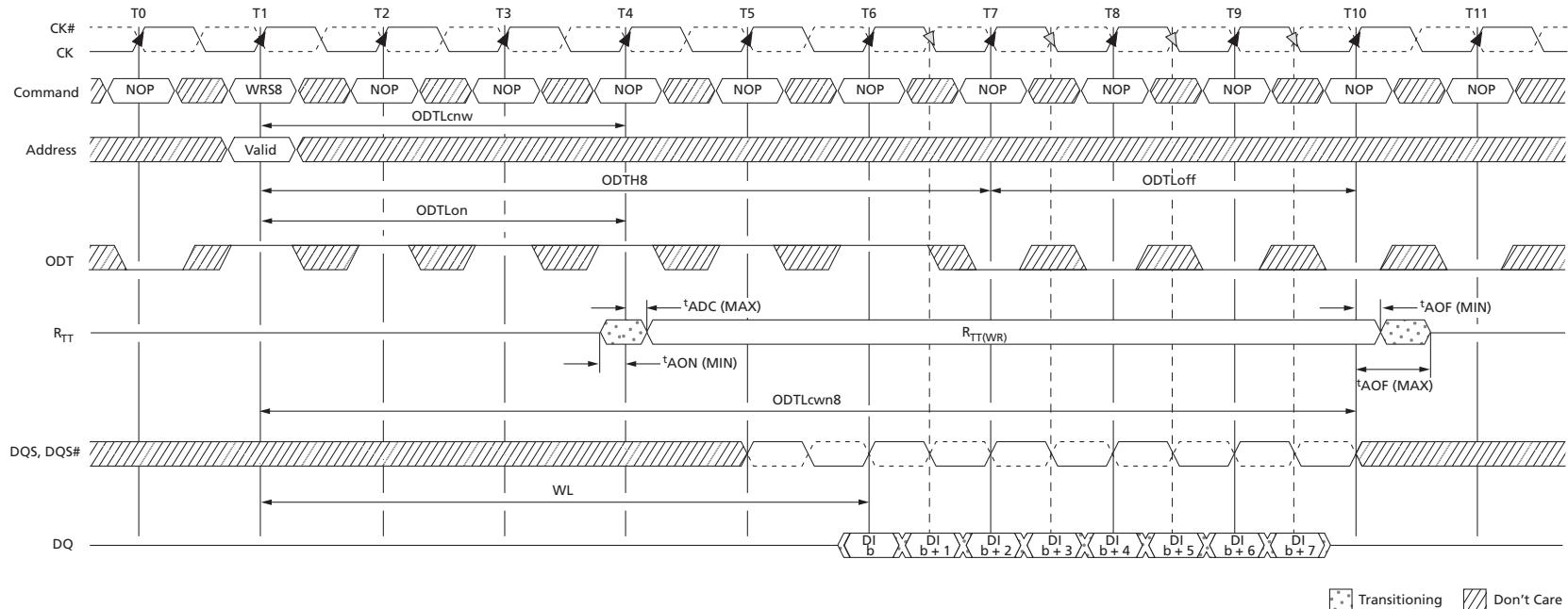
Figure 110: Dynamic ODT: Without WRITE Command



Notes:

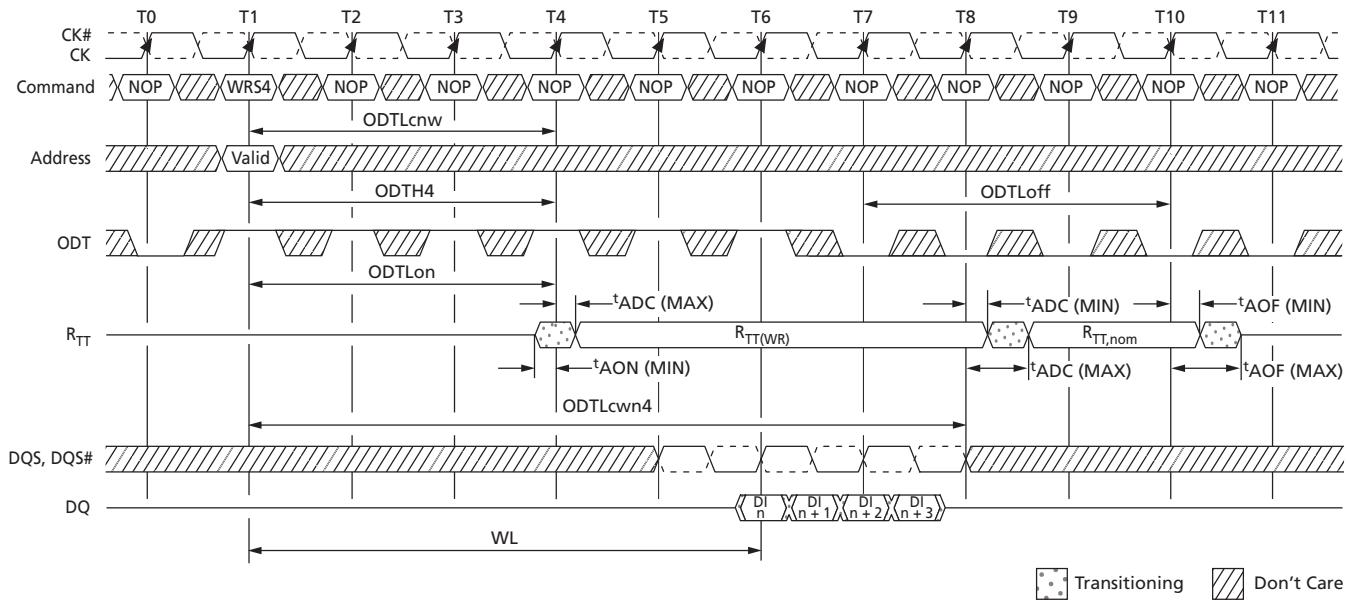
1. AL = 0, CWL = 5. R_{TT,nom} is enabled and R_{TT(WR)} is either enabled or disabled.
2. ODTH4 is defined from ODT registered HIGH to ODT registered LOW; in this example, ODTH4 is satisfied. ODT registered LOW at T5 is also legal.

Figure 111: Dynamic ODT: ODT Pin Asserted Together with WRITE Command for 6 Clock Cycles, BL8



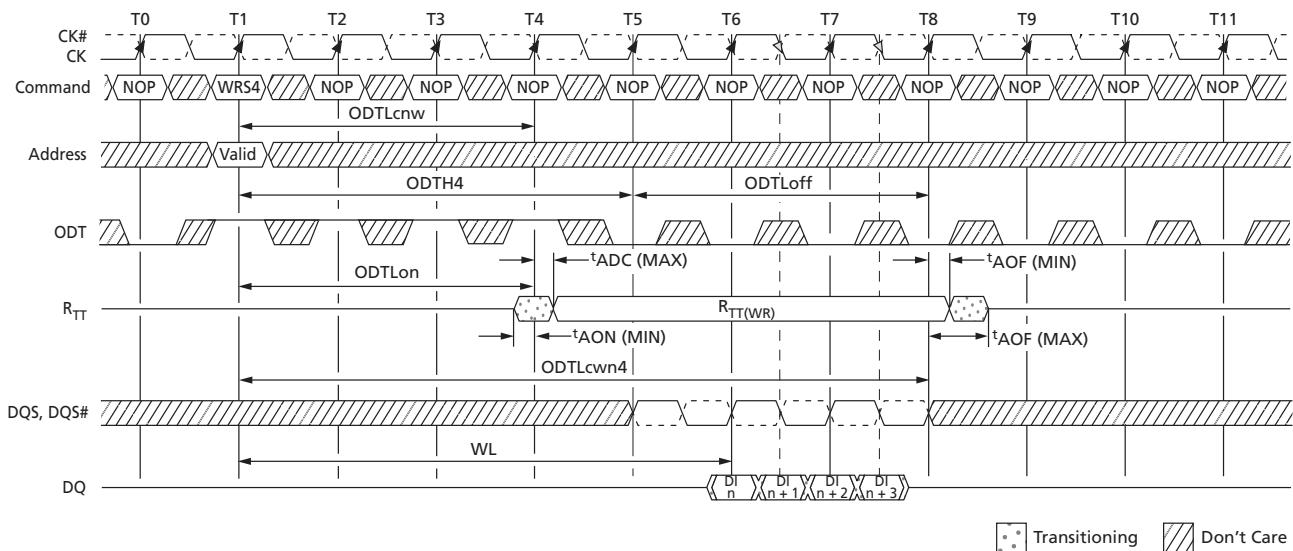
Notes:

1. Via MRS or OTF; AL = 0, CWL = 5. If R_{TT,nom} can be either enabled or disabled, ODT can be HIGH. R_{TT(WR)} is enabled.
2. In this example, ODTH8 = 6 is satisfied exactly.

Figure 112: Dynamic ODT: ODT Pin Asserted with WRITE Command for 6 Clock Cycles, BC4


Notes:

1. Via MRS or OTF. AL = 0, CWL = 5. R_{TT, nom} and R_{TT(WR)} are enabled.
2. ODTH4 is defined from ODT registered HIGH to ODT registered LOW, so in this example, ODTH4 is satisfied. ODT registered LOW at T5 is also legal.

Figure 113: Dynamic ODT: ODT Pin Asserted with WRITE Command for 4 Clock Cycles, BC4


Notes:

1. Via MRS or OTF. AL = 0, CWL = 5. R_{TT, nom} can be either enabled or disabled. If disabled, ODT can remain HIGH. R_{TT(WR)} is enabled.
2. In this example ODTH4 = 4 is satisfied exactly.

Synchronous ODT Mode

Synchronous ODT mode is selected whenever the DLL is turned on and locked and when either $R_{TT,nom}$ or $R_{TT(WR)}$ is enabled. Based on the power-down definition, these modes are:

- Any bank active with CKE HIGH
- Refresh mode with CKE HIGH
- Idle mode with CKE HIGH
- Active power-down mode (regardless of MR0[12])
- Precharge power-down mode if DLL is enabled by MR0[12] during precharge power-down

ODT Latency and Posted ODT

In synchronous ODT mode, R_{TT} turns on ODT_{Lon} clock cycles after ODT is sampled HIGH by a rising clock edge and turns off ODT_{loff} clock cycles after ODT is registered LOW by a rising clock edge. The actual on/off times varies by t_{AON} and t_{AOF} around each clock edge (see Table 89 (page 200)). The ODT latency is tied to the WRITE latency (WL) by ODT_{Lon} = WL - 2 and ODT_{loff} = WL - 2.

Since write latency is made up of CAS WRITE latency (CWL) and additive latency (AL), the AL programmed into the mode register (MR1[4, 3]) also applies to the ODT signal. The device's internal ODT signal is delayed a number of clock cycles defined by the AL relative to the external ODT signal. Thus, ODT_{Lon} = CWL + AL - 2 and ODT_{loff} = CWL + AL - 2.

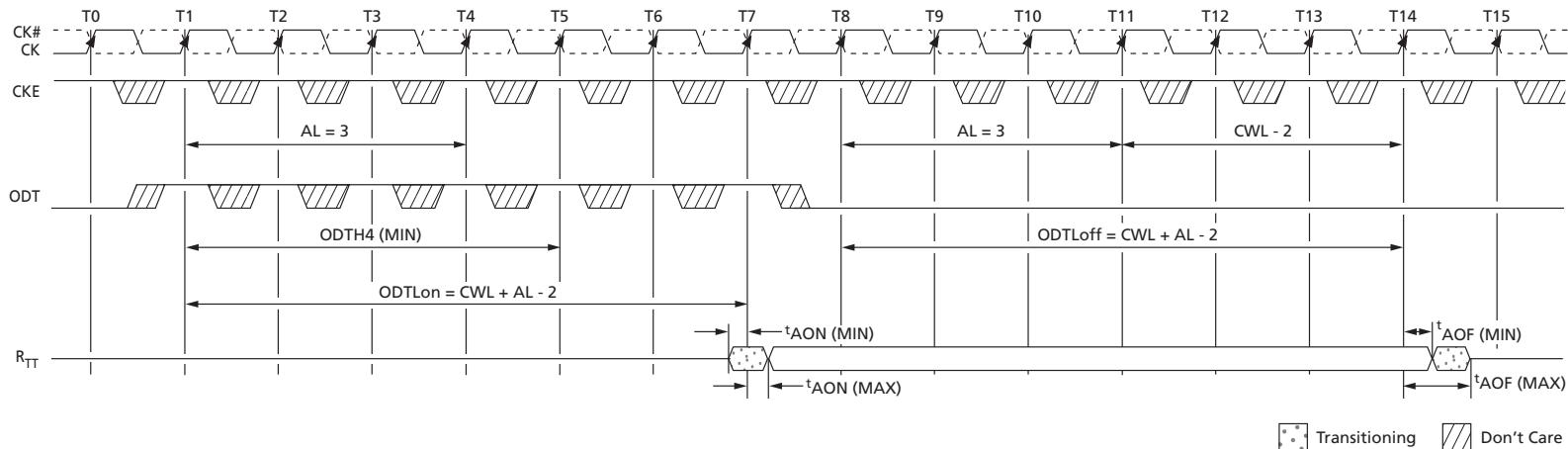
Timing Parameters

Synchronous ODT mode uses the following timing parameters: ODT_{Lon}, ODT_{loff}, ODTH4, ODTH8, t_{AON} , and t_{AOF} . The minimum R_{TT} turn-on time (t_{AON} [MIN]) is the point at which the device leaves High-Z and ODT resistance begins to turn on. Maximum R_{TT} turn-on time (t_{AON} [MAX]) is the point at which ODT resistance is fully on. Both are measured relative to ODT_{Lon}. The minimum R_{TT} turn-off time (t_{AOF} [MIN]) is the point at which the device starts to turn off ODT resistance. The maximum R_{TT} turn off time (t_{AOF} [MAX]) is the point at which ODT has reached High-Z. Both are measured from ODT_{loff}.

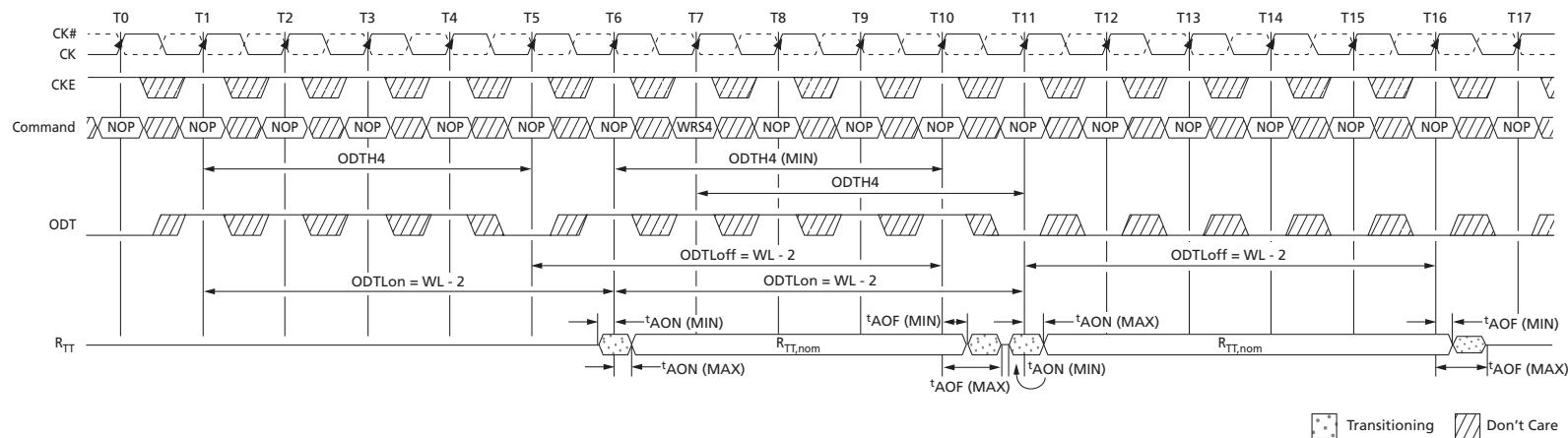
When ODT is asserted, it must remain HIGH until ODTH4 is satisfied. If a WRITE command is registered by the DRAM with ODT HIGH, then ODT must remain HIGH until ODTH4 (BC4) or ODTH8 (BL8) after the WRITE command (see Figure 115 (page 201)). ODTH4 and ODTH8 are measured from ODT registered HIGH to ODT registered LOW or from the registration of a WRITE command until ODT is registered LOW.

Table 89: Synchronous ODT Parameters

Symbol	Description	Begins at	Defined to	Definition for All DDR3L Speed Bins	Unit
ODTLon	ODT synchronous turn-on delay	ODT registered HIGH	$R_{TT(ON)} \pm t_{AON}$	CWL + AL - 2	t_{CK}
ODTloff	ODT synchronous turn-off delay	ODT registered HIGH	$R_{TT(OFF)} \pm t_{AOF}$	CWL + AL - 2	t_{CK}
ODTH4	ODT minimum HIGH time after ODT assertion or WRITE (BC4)	ODT registered HIGH or write registration with ODT HIGH	ODT registered LOW	$4t_{CK}$	t_{CK}
ODTH8	ODT minimum HIGH time after WRITE (BL8)	Write registration with ODT HIGH	ODT registered LOW	$6t_{CK}$	t_{CK}
t_{AON}	ODT turn-on relative to ODTlon completion	Completion of ODTlon	$R_{TT(ON)}$	See Electrical Characteristics and AC Operating Conditions table	ps
t_{AOF}	ODT turn-off relative to ODTloff completion	Completion of ODTloff	$R_{TT(OFF)}$	$0.5t_{CK} \pm 0.2t_{CK}$	t_{CK}

Figure 114: Synchronous ODT


Note: 1. AL = 3; CWL = 5; ODTlon = WL = 6.0; ODTloff = WL - 2 = 6. $R_{TT,nom}$ is enabled.

Figure 115: Synchronous ODT (BC4)


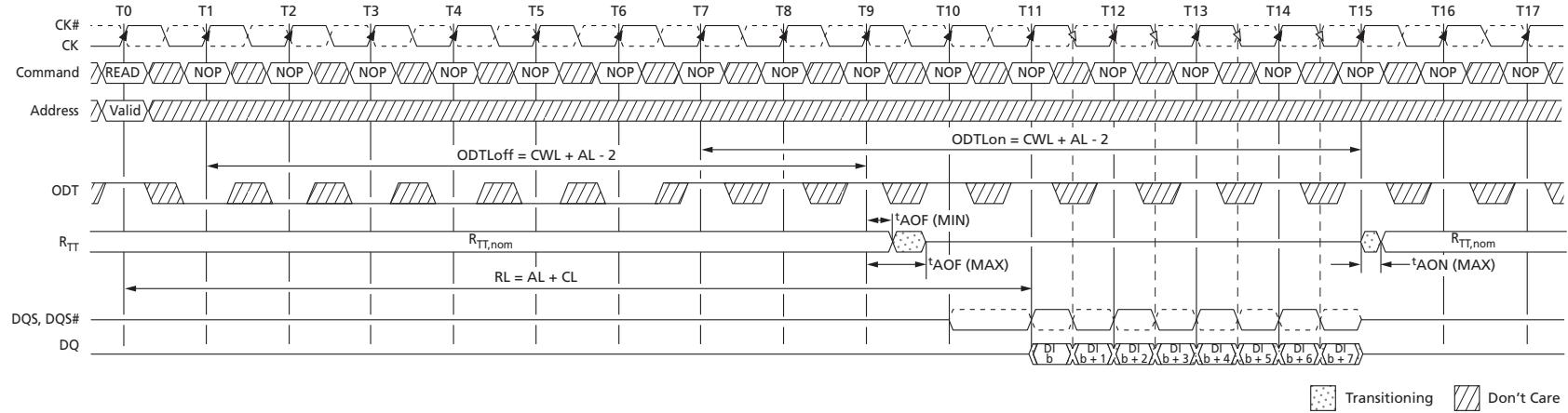
Notes:

1. $WL = 7$. $R_{TT,nom}$ is enabled. $R_{TT(WR)}$ is disabled.
2. ODT must be held HIGH for at least $ODTH4$ after assertion (T1).
3. ODT must be kept HIGH $ODTH4$ (BC4) or $ODTH8$ (BL8) after the WRITE command (T7).
4. ODTH is measured from ODT first registered HIGH to ODT first registered LOW or from the registration of the WRITE command with ODT HIGH to ODT registered LOW.
5. Although $ODTH4$ is satisfied from ODT registered HIGH at T6, ODT must not go LOW before T11 as $ODTH4$ must also be satisfied from the registration of the WRITE command at T7.

ODT Off During READs

Because the device cannot terminate and drive at the same time, R_{TT} must be disabled at least one-half clock cycle before the READ preamble by driving the ODT ball LOW (if either $R_{TT,nom}$ or $R_{TT(WR)}$ is enabled). R_{TT} may not be enabled until the end of the postamble, as shown in the following example.

Note: ODT may be disabled earlier and enabled later than shown in Figure 116 (page 203).

Figure 116: ODT During READs


Note: 1. ODT must be disabled externally during READs by driving ODT LOW. For example, CL = 6; AL = CL - 1 = 5; RL = AL + CL = 11; CWL = 5; ODT_{Lon} = CWL + AL - 2 = 8; ODT_{Loff} = CWL + AL - 2 = 8. R_{TT,nom} is enabled. R_{TT(WR)} is a "Don't Care."

Asynchronous ODT Mode

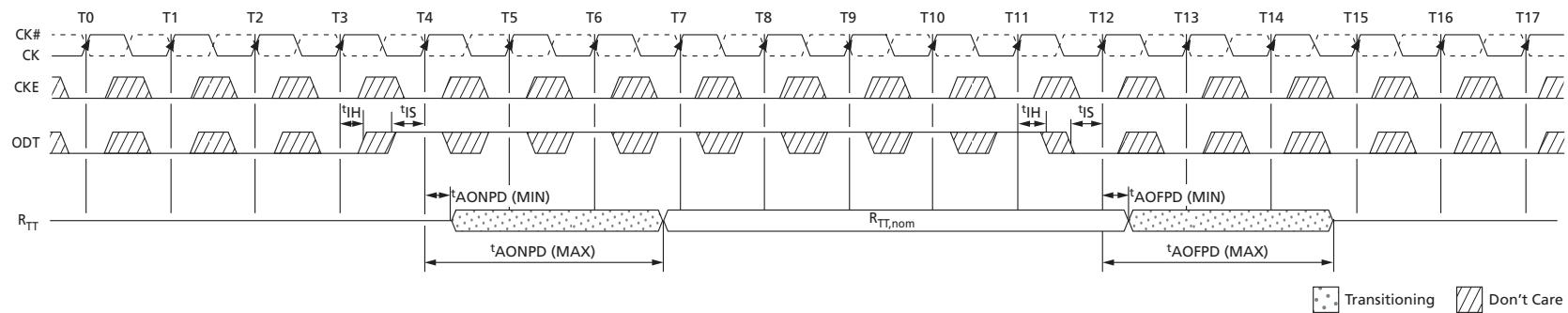
Asynchronous ODT mode is available when the DRAM runs in DLL on mode and when either $R_{TT,nom}$ or $R_{TT(WR)}$ is enabled; however, the DLL is temporarily turned off in pre-charged power-down standby (via MRO[12]). Additionally, ODT operates asynchronously when the DLL is synchronizing after being reset. See Power-Down Mode (page 181) for definition and guidance over power-down details.

In asynchronous ODT timing mode, the internal ODT command is not delayed by AL relative to the external ODT command. In asynchronous ODT mode, ODT controls R_{TT} by analog time. The timing parameters t_{AONPD} and t_{AOFPD} replace t_{ODTLon}/t_{AON} and $t_{ODTloff}/t_{AOF}$, respectively, when ODT operates asynchronously.

The minimum R_{TT} turn-on time (t_{AONPD} [MIN]) is the point at which the device termination circuit leaves High-Z and ODT resistance begins to turn on. Maximum R_{TT} turn-on time (t_{AONPD} [MAX]) is the point at which ODT resistance is fully on. t_{AONPD} (MIN) and t_{AONPD} (MAX) are measured from ODT being sampled HIGH.

The minimum R_{TT} turn-off time (t_{AOFPD} [MIN]) is the point at which the device termination circuit starts to turn off ODT resistance. Maximum R_{TT} turn-off time (t_{AOFPD} [MAX]) is the point at which ODT has reached High-Z. t_{AOFPD} (MIN) and t_{AOFPD} (MAX) are measured from ODT being sampled LOW.

Figure 117: Asynchronous ODT Timing with Fast ODT Transition



Note: 1. AL is ignored.

Table 90: Asynchronous ODT Timing Parameters for All Speed Bins

Symbol	Description	Min	Max	Unit
t_{AONPD}	Asynchronous R_{TT} turn-on delay (power-down with DLL off)	2	8.5	ns
t_{AOFPD}	Asynchronous R_{TT} turn-off delay (power-down with DLL off)	2	8.5	ns

Synchronous to Asynchronous ODT Mode Transition (Power-Down Entry)

There is a transition period around power-down entry (PDE) where the DRAM's ODT may exhibit either synchronous or asynchronous behavior. This transition period occurs if the DLL is selected to be off when in precharge power-down mode by the setting MR0[12] = 0. Power-down entry begins t_{ANPD} prior to CKE first being registered LOW, and ends when CKE is first registered LOW. t_{ANPD} is equal to the greater of ODTLoff + $1^{t_{CK}}$ or ODTLon + $1^{t_{CK}}$. If a REFRESH command has been issued, and it is in progress when CKE goes LOW, power-down entry ends t_{RFC} after the REFRESH command, rather than when CKE is first registered LOW. Power-down entry then becomes the greater of t_{ANPD} and t_{RFC} - REFRESH command to CKE registered LOW.

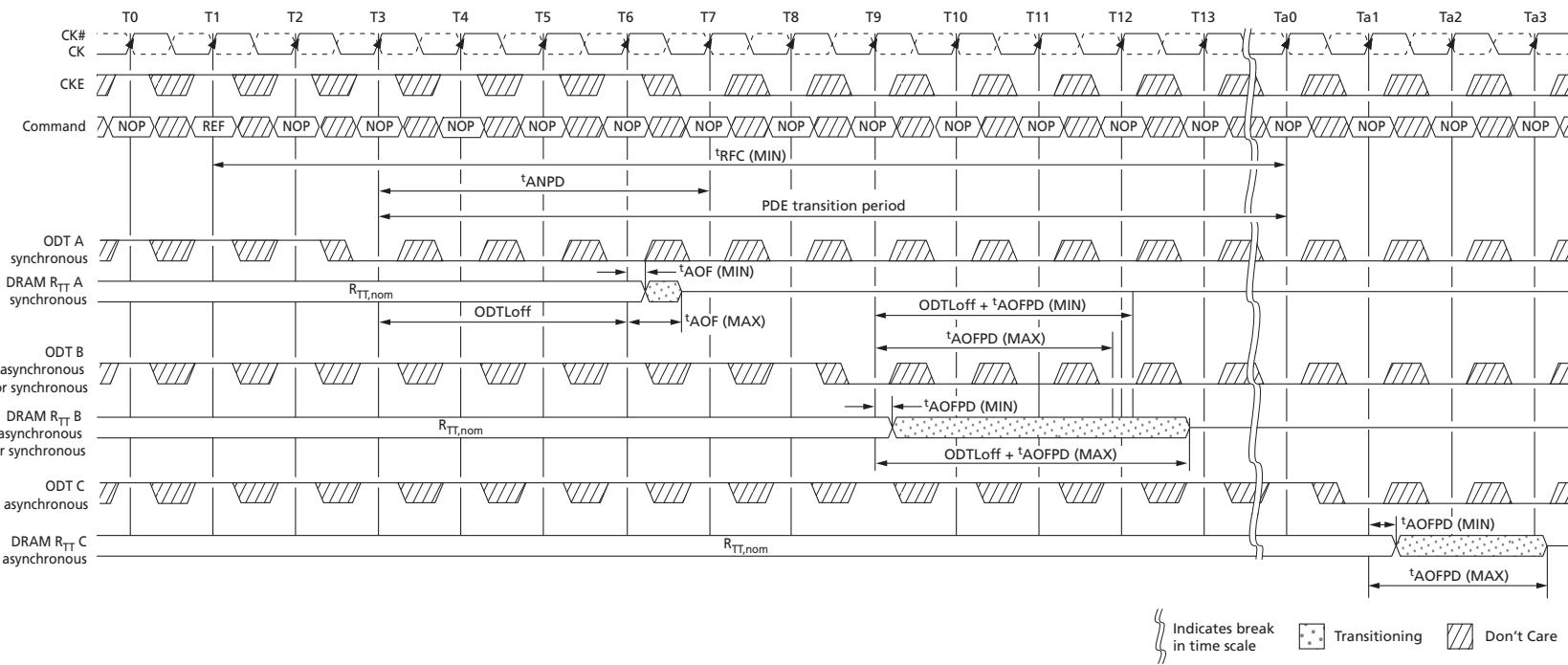
ODT assertion during power-down entry results in an R_{TT} change as early as the lesser of t_{AONPD} (MIN) and ODTLon \times t_{CK} + t_{AON} (MIN), or as late as the greater of t_{AONPD} (MAX) and ODTLon \times t_{CK} + t_{AON} (MAX). ODT de-assertion during power-down entry can result in an R_{TT} change as early as the lesser of t_{AOFPD} (MIN) and ODTLoff \times t_{CK} + t_{AOF} (MIN), or as late as the greater of t_{AOFPD} (MAX) and ODTLoff \times t_{CK} + t_{AOF} (MAX). Table 91 (page 207) summarizes these parameters.

If AL has a large value, the uncertainty of the state of R_{TT} becomes quite large. This is because ODTLon and ODTLoff are derived from the WL; and WL is equal to CWL + AL. Figure 118 (page 207) shows three different cases:

- ODT_A: Synchronous behavior before t_{ANPD} .
- ODT_B: ODT state changes during the transition period with t_{AONPD} (MIN) < ODTLon \times t_{CK} + t_{AON} (MIN) and t_{AONPD} (MAX) > ODTLon \times t_{CK} + t_{AON} (MAX).
- ODT_C: ODT state changes after the transition period with asynchronous behavior.

Table 91: ODT Parameters for Power-Down (DLL Off) Entry and Exit Transition Period

Description	Min	Max
Power-down entry transition period (power-down entry)	Greater of: t_{ANPD} or t_{RFC} - refresh to CKE LOW	
Power-down exit transition period (power-down exit)	$t_{ANPD} + t_{XP DLL}$	
ODT to R_{TT} turn-on delay (ODT _{Lon} = WL - 2)	Lesser of: t_{AONPD} (MIN) (2ns) or $ODT_{Lon} \times t_{CK} + t_{AON}$ (MIN)	Greater of: t_{AONPD} (MAX) (8.5ns) or $ODT_{Lon} \times t_{CK} + t_{AON}$ (MAX)
ODT to R_{TT} turn-off delay (ODT _{loff} = WL - 2)	Lesser of: t_{AOFPD} (MIN) (2ns) or $ODT_{loff} \times t_{CK} + t_{AOF}$ (MIN)	Greater of: t_{AOFPD} (MAX) (8.5ns) or $ODT_{loff} \times t_{CK} + t_{AOF}$ (MAX)
t_{ANPD}	WL - 1 (greater of ODT _{loff} + 1 or ODT _{lon} + 1)	

Figure 118: Synchronous to Asynchronous Transition During Precharge Power-Down (DLL Off) Entry


Note: 1. AL = 0; CWL = 5; ODT_{off} = WL - 2 = 3.

Asynchronous to Synchronous ODT Mode Transition (Power-Down Exit)

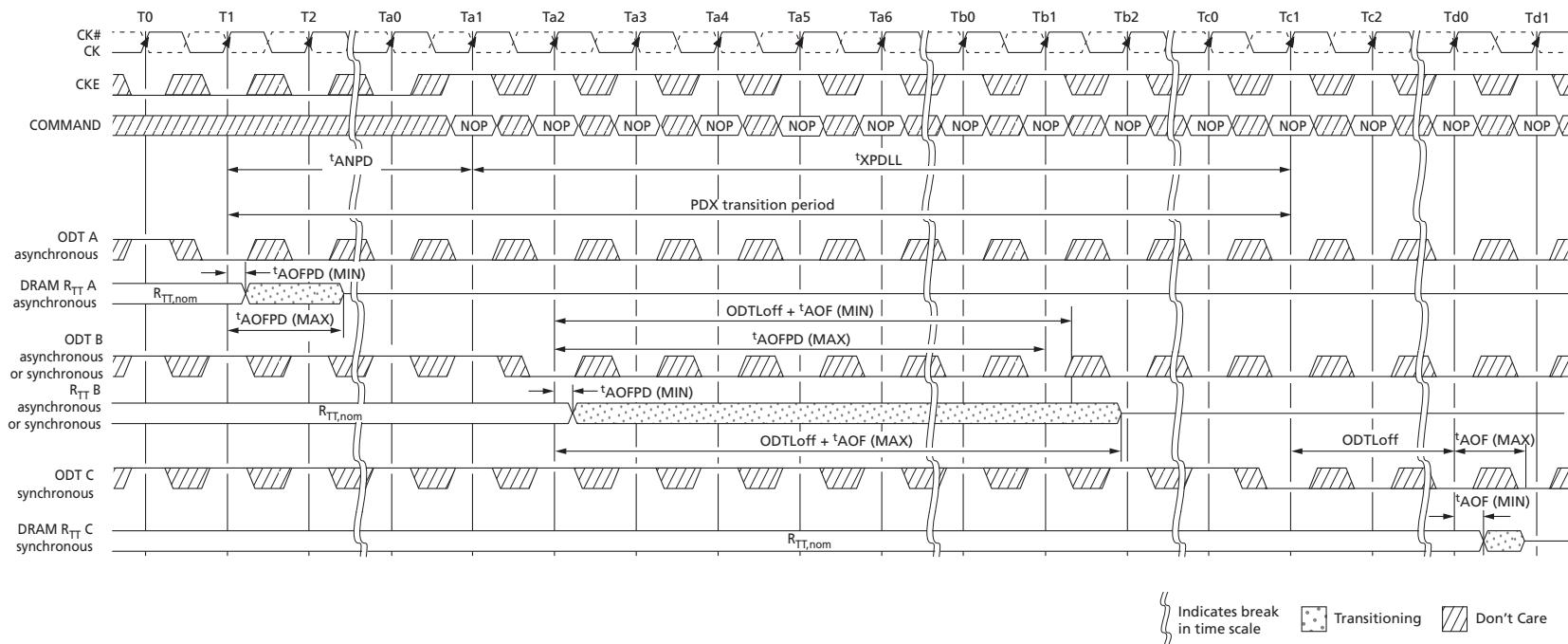
The DRAM's ODT can exhibit either asynchronous or synchronous behavior during power-down exit (PDX). This transition period occurs if the DLL is selected to be off when in precharge power-down mode by setting MR0[12] to 0. Power-down exit begins t_{ANPD} prior to CKE first being registered HIGH, and ends t_{XPDLL} after CKE is first registered HIGH. t_{ANPD} is equal to the greater of $ODT_{Loff} + t_{CK}$ or $ODT_{Lon} + t_{CK}$. The transition period is $t_{ANPD} + t_{XPDLL}$.

ODT assertion during power-down exit results in an R_{TT} change as early as the lesser of t_{AONPD} (MIN) and $ODT_{Lon} \times t_{CK} + t_{AON}$ (MIN), or as late as the greater of t_{AONPD} (MAX) and $ODT_{Lon} \times t_{CK} + t_{AON}$ (MAX). ODT de-assertion during power-down exit may result in an R_{TT} change as early as the lesser of t_{AOFPD} (MIN) and $ODT_{Loff} \times t_{CK} + t_{AOF}$ (MIN), or as late as the greater of t_{AOFPD} (MAX) and $ODT_{Loff} \times t_{CK} + t_{AOF}$ (MAX). Table 91 (page 207) summarizes these parameters.

If AL has a large value, the uncertainty of the R_{TT} state becomes quite large. This is because ODT_{Lon} and ODT_{Loff} are derived from WL, and WL is equal to CWL + AL. Figure 119 (page 209) shows three different cases:

- ODT C: Asynchronous behavior before t_{ANPD} .
- ODT B: ODT state changes during the transition period, with t_{AOFPD} (MIN) < $ODT_{Loff} \times t_{CK} + t_{AOF}$ (MIN), and $ODT_{Loff} \times t_{CK} + t_{AOF}$ (MAX) > t_{AOFPD} (MAX).
- ODT A: ODT state changes after the transition period with synchronous response.

Figure 119: Asynchronous to Synchronous Transition During Precharge Power-Down (DLL Off) Exit



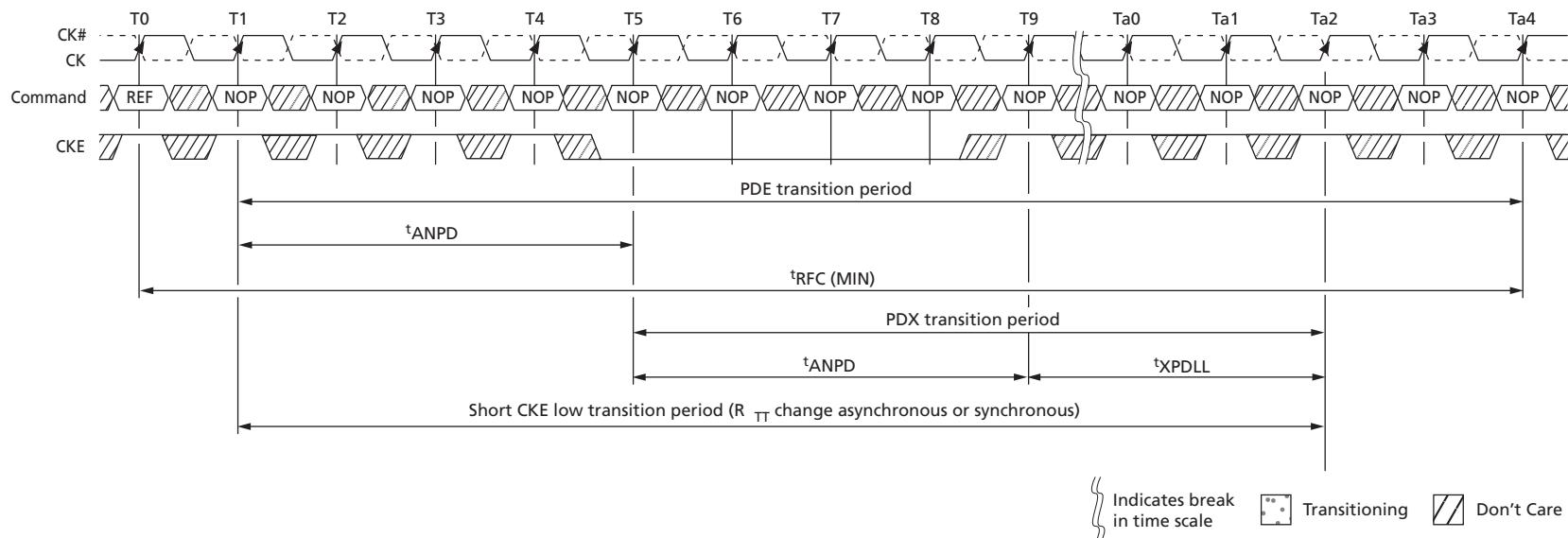
Note: 1. CL = 6; AL = CL - 1; CWL = 5; ODTloff = WL - 2 = 8.

Asynchronous to Synchronous ODT Mode Transition (Short CKE Pulse)

If the time in the precharge power-down or idle states is very short (short CKE LOW pulse), the power-down entry and power-down exit transition periods overlap. When overlap occurs, the response of the DRAM's R_{TT} to a change in the ODT state can be synchronous or asynchronous from the start of the power-down entry transition period to the end of the power-down exit transition period, even if the entry period ends later than the exit period.

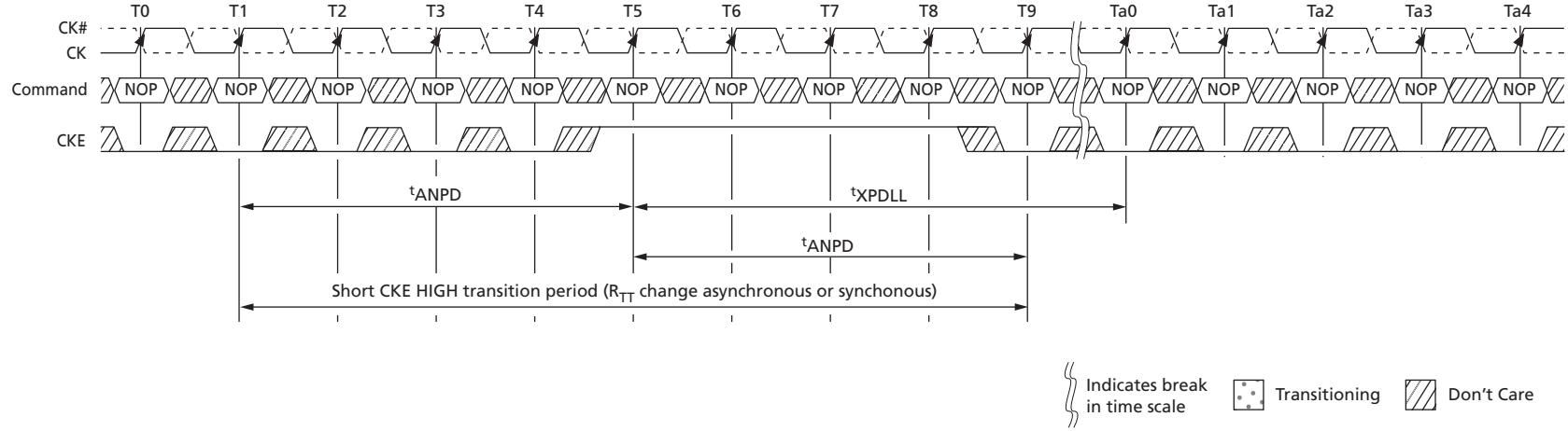
If the time in the idle state is very short (short CKE HIGH pulse), the power-down exit and power-down entry transition periods overlap. When this overlap occurs, the response of the DRAM's R_{TT} to a change in the ODT state may be synchronous or asynchronous from the start of power-down exit transition period to the end of the power-down entry transition period.

Figure 120: Transition Period for Short CKE LOW Cycles with Entry and Exit Period Overlapping



Note: 1. AL = 0, WL = 5, $t_{ANPD} = 4$.

Figure 121: Transition Period for Short CKE HIGH Cycles with Entry and Exit Period Overlapping



Note: 1. AL = 0, WL = 5, $t_{ANPD} = 4$.



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This data sheet contains minimum and maximum limits specified over the power supply and temperature range set forth herein. Although considered final, these specifications are subject to change, as further product development and data characterization sometimes occur.