

## MCP2021A/2A

## LIN Transceiver with Voltage Regulator

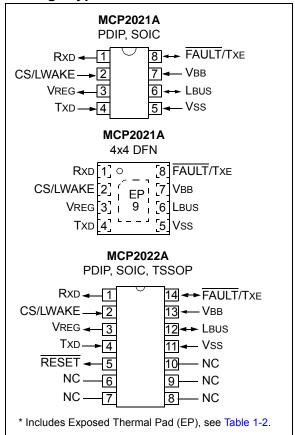
#### Features:

- The MCP2021A/2A are compliant with LIN Bus Specifications Version 1.3, 2.1 and with SAE J2602-2
- · Support Baud Rates up to 20 kBaud
- · 43V Load Dump Protected
- · Maximum Continuous Input Voltage: 30V
- Wide LIN-Compliant Supply Voltage: 6.0 18.0V
- Extended Temperature Range: -40 to +125°C
- Interface to PIC<sup>®</sup> MCU EUSART and Standard USARTs
- · Wake-Up on LIN Bus Activity or Local Wake Input
- · Local Interconnect Network (LIN) Bus Pin:
  - Internal Pull-Up Termination Resistor and Diode for Slave Node
  - Protected Against VBAT Shorts
  - Protected Against Loss of Ground
  - High-Current Drive
- · TXD and LIN Bus Dominant Time-Out Function
- Two Low-Power Modes:
  - Transmitter Off: 90 μA (typical)
  - Power Down: 4.5 µA (typical)
- Output Indicating Internal Reset State (POR or Sleep Wake)
- MCP2021A/2A On-Chip Voltage Regulator:
  - Output Voltage of 5.0V or 3.3V at 70 mA Capability with Tolerances of ±3% Over the Temperature Range
  - Internal Short Circuit Current Limit
  - External Components Limited to Filter Capacitor and Load Capacitor
- · Automatic Thermal Shutdown
- High Electromagnetic Immunity (EMI), Low Electromagnetic Emission (EME)
- Robust ESD Performance: ±15 kV for LBUs and VBB pin (IEC61000-4-2)
- Transient Protection for LBUS and VBB Pins in Automotive Environment (ISO7637)
- Meets Stringent Automotive Design Requirements, including "OEM Hardware Requirements for LIN, CAN and FlexRay Interfaces in Automotive Applications", Version 1.2, March 2011
- Multiple Package Options, including Small 4x4 mm DFN Package

#### **Description:**

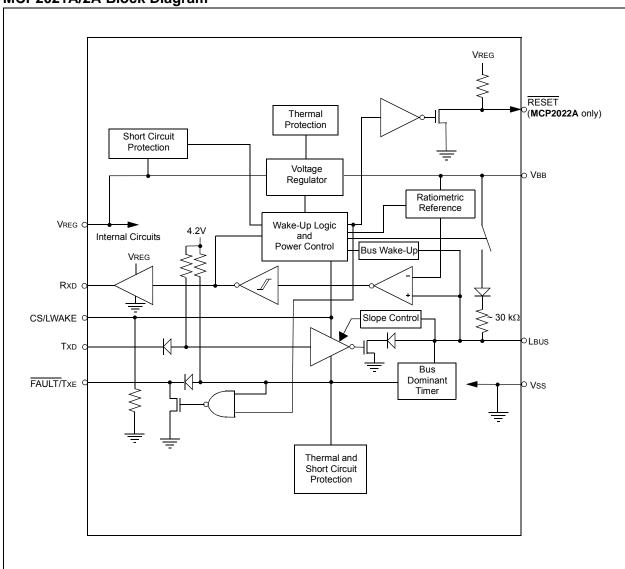
The MCP2021A/2A provide a bidirectional, half-duplex communication physical interface to meet the LIN bus specification Revision 2.1 and SAE J2602-2. The devices incorporate a voltage regulator with 5V or 3.3V at 70 mA regulated power supply output. The devices have been designed to meet the stringent quiescent current requirements of the automotive industry and will survive +43V load dump transients and double battery jumps.

#### **Package Types**





#### MCP2021A/2A Block Diagram



#### 1.0 DEVICE OVERVIEW

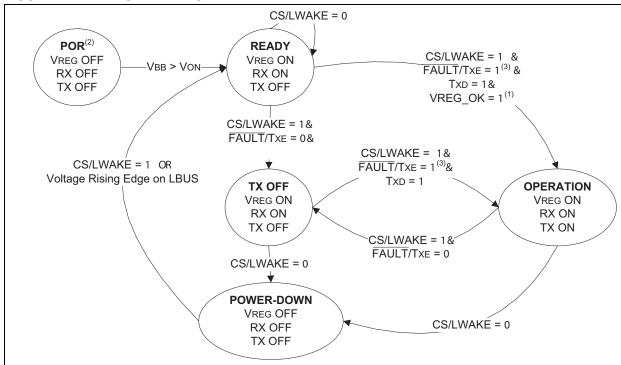
The MCP2021A/2A devices provide a physical interface between a microcontroller and a LIN half-duplex bus. They are intended for automotive and industrial applications with serial bus baud rates up to 20 kBaud. These devices will translate the CMOS/TTL logic levels to LIN logic levels and vice versa.

The MCP2021A/2A offer optimum EMI and ESD performance and can withstand high voltage on the LIN bus. The devices support two low-power modes to meet automotive industry power consumption requirements. The MCP2021A/2A also provide a +5V or 3.3V regulated power output at 70 mA.

#### 1.1 Modes of Operation

The MCP2021A/2A work in five modes: Power-On Reset, Power-Down, Ready, Operation and Transmitter Off. For an overview of all operational modes, please refer to Table 1-1. For the operational mode transition, please refer to Figure 1-1.

FIGURE 1-1: STATE DIAGRAM



- Note 1: VREG\_OK: Regulator Output Voltage > 0.8VREG\_NOM.
  - 2: If the voltage on pin VBB falls below VOFF, the device will enter Power-On Reset mode from all other modes, which is not shown in the figure.
  - 3: FAULT/TXE = 1 represents input high and no fault conditions. FAULT/TXE = 0 represents input low or a fault condition. Refer to Table 1-5.

#### 1.1.1 POWER-ON RESET MODE

Upon application of VBB or whenever the voltage on VBB is below the threshold of regulator turn-off voltage VOFF (typically 4.50V), the device enters Power-On Reset (POR) mode. During this mode, the device maintains the digital section in a Reset mode and waits until the voltage on the VBB pin rises above the threshold of regulator turn-on voltage VoN (typically 5.75V) to enter Ready mode. In Power-On Reset mode, the LIN physical layer and voltage regulator are disabled and the RESET output (MCP2022A only) is forced to low.

#### 1.1.2 READY MODE

The device enters Ready mode from POR mode after the voltage on VBB rises above the threshold of regulator turn-on voltage Von or from Power-Down mode when a remote or local wake-up event happens.

Upon entering Ready mode, the voltage regulator and the receiver section of the transceiver are powered up. The transmitter remains in an off state. The device is ready to receive data but not to transmit. In order to minimize the power consumption, the regulator operates in a reduced-power mode. It has a lower GBW product and it is thus slower. However, the 70 mA drive capability is unchanged.

The device stays in Ready mode until the output of the voltage regulator has stabilized and the CS/LWAKE pin is high ('1').

#### 1.1.3 OPERATION MODE

If VREG is OK (VREG > 0.8 VREG\_NORM) and the CS/LWAKE, FAULT/TXE and TXD pins are high, the part enters Operation mode from either Ready or Transmitter Off mode.

In this mode, all internal modules are operational. The internal pull-up resistor between Lbus and Vbb is connected only in this mode.

The device goes into Power-Down mode at the falling edge on CS/LWAKE or into Transmitter Off mode at the falling edge on FAULT/TXE while CS/LWAKE stays high.

#### 1.1.4 TRANSMITTER OFF MODE

In Transmitter Off mode, the receiver is enabled but the LBUS transmitter is off. It is a lower power mode.

In order to minimize power consumption, the regulator operates in a reduced-power mode. It has a lower GBW product and it is thus slower. However, the 70 mA drive capability is unchanged.

The transmitter may be re-enabled whenever the FAULT/TxE signal returns high, by removing the internal fault condition and by driving FAULT/TxE high. The transmitter will not be enabled even if the FAULT/TxE pin is brought high externally, when the internal fault is still present. However, externally forcing the FAULT/TxE high while the internal fault is still present should be avoided, since this will induce high current and power dissipation in the FAULT/TxE pin.

The transmitter is also turned off whenever the voltage regulator is unstable or recovering from a fault. This prevents unwanted disruption of the bus during times of uncertain operation.

#### 1.1.5 POWER-DOWN MODE

In Power-Down mode, the transceiver and the voltage regulator are both off. Only the bus wake-up section and the CS/LWAKE pin wake-up circuits are in operation. This is the lowest power mode.

If any bus activity (e.g., a Break character) occurs during Power-Down mode, the device will immediately enter Ready mode and enable the voltage regulator. Then, once the regulator output has stabilized (approximately 0.3 ms to 1.2 ms), it goes into Operation mode. Refer to Section 1.1.6 "Remote Wake-Up".

The part will also enter Ready mode from Power-Down mode, followed by the Operation mode, if the CS/LWAKE pin becomes active high ('1').

#### 1.1.6 REMOTE WAKE-UP

The Remote Wake-Up sub-module observes the LBUS in order to detect bus activity. In Power-Down mode, normal LIN recessive/dominant threshold is disabled and the LIN bus wake-up voltage threshold VWK(LBUS) is used to detect bus activities. Bus activity is detected when the voltage on the LBUS falls below the LIN bus wake-up voltage threshold VWK(LBUS) (approximately 3.5V) for at least tBDB (a typical duration of 80 µs) followed by a rising edge. Such a condition causes the device to leave Power-Down mode.

TABLE 1-1: OVERVIEW OF OPERATIONAL MODES

State	Transmitter	Receiver	Internal Wake Module	Voltage Regulator	Operation	Comments
POR	OFF	OFF	OFF	OFF	Proceed to Ready mode after VBB > VON	
Ready	OFF	ON	OFF	ON	If CS/LWAKE is high, then proceed to Operation or Transmitter Off mode.	Bus Off state
Operation	ON	ON	OFF	ON	If CS/LWAKE is low, then proceed to Power-Down mode. If FAULT/TxE is low, then proceed to Transmitter Off mode.	Normal Operation mode
Power-Down	OFF	OFF	ON Activity Detect	OFF	On LIN bus rising edge or CS/LWAKE high level, go to Ready mode.	Lowest power mode
Transmitter Off	OFF	ON	OFF	ON	If CS/LWAKE is low, then proceed to Power-Down mode. If FAULT/TXE is high, then proceed to Operation mode.	Bus Off state, lower power mode

#### 1.2 Pin Descriptions

The descriptions of the pins are listed in Table 1-2.

TABLE 1-2: PIN FUNCTION TABLE

		Pin Nun	nber			
Pin Name	8-lead PDIP, SOIC 4x4 DFN PDIP, SOIC, TSSOP		Pin Type	Description		
Rxd	1	1	1	Output	Receive Data Output	
CS/LWAKE	2	2	2	TTL Input, HV-tolerant	Chip Select and Local Wake-up Input	
VREG	3	3	3	Output	Voltage Regulator Output	
TXD	4	4	4	Input, HV-tolerant	Transmit Data Input	
RESET	_	_	5	Output	Reset Output	
NC	_	_	6–10	_	No Connection	
Vss	5	5	11	Power	Ground	
LBUS	6	6	12	I/O, HV	LIN Bus	
VBB	7	7	13	Power	Battery	
FAULT/TXE	8	8	14	I/O, HV-tolerant	Fault Detect Output/Transmitter Enable Input	
EP	_	9	_	_	Exposed Thermal Pad	

#### 1.2.1 RECEIVE DATA OUTPUT (RXD)

Receive Data Output pin. The RXD pin is a standard CMOS output pin and it follows the state of the LBUS pin.

## 1.2.2 CHIP SELECT AND LOCAL WAKE-UP INPUT (CS/LWAKE)

Chip Select and Local Wake-up Input pin (TTL level, high-voltage tolerant). This pin controls the device state transition. Refer to Figure 1-1.

If CS/LWAKE = 1, the device can work in Operation mode ( $\overline{FAULT}/TxE = 1$ ) or in Transmitter Off mode ( $\overline{FAULT}/TxE = 0$ ). If CS/LWAKE = 0, the device can work in Power-Down mode or in Ready mode.

An internal pull-down resistor will keep the CS/LWAKE pin low to ensure that no disruptive data will be present on the bus while the microcontroller is executing a Power-On Reset and I/O initialization sequence. When CS/LWAKE is '1', a weak pull-down (~600 k $\Omega$ ) is used to reduce current. When CS/LWAKE is '0', a stronger pull-down (~300 k $\Omega$ ) is used to maintain the logic level.

This pin may also be used as a local wake-up input (see Figure 1-1). The microcontroller will set the I/O pin to control the CS/LWAKE. An external switch or another source can then wake up both the transceiver and the microcontroller.

**Note:** CS/LWAKE should NOT be tied directly to

the VREG pin as this could force the MCP2021A/2A into Operation mode before the microcontroller is initialized.

#### 1.2.3 POWER OUTPUT (VREG)

Positive Supply Voltage Regulator Output pin. An on-chip LDO gives +5.0 or +3.3V at 70 mA regulated voltage on this pin.

#### 1.2.4 TRANSMIT DATA INPUT (TXD)

Transmit Data Input pin (TTL level, HV-compliant, adaptive pull-up). The transmitter reads the data stream on the TXD pin and sends it to the LIN bus. The LBUS pin is low (dominant) when TXD is low and high (recessive) when TXD is high.

TXD is internally pulled up to approximately 4.2V. When TXD is '0', a weak pull-up (~900 k $\Omega$ ) is used to reduce current. When TXD is '1', a stronger pull-up (~300 k $\Omega$ ) is used to maintain the logic level. A series reverse-blocking diode allows applying TXD input voltages greater than the internally generated 4.2V and renders the TXD pin HV-compliant up to 30V (see MCP2021A/2A Block Diagram).

#### 1.2.5 RESET (MCP2022A ONLY)

RESET output pin. This pin is open-drain with ~90 k $\Omega$  pull-up to VREG. It indicates the internal voltage has reached a valid, stable level. As long as the internal voltage is valid (above 0.8 VREG), this pin will remain high ('1'); otherwise, the RESET pin switches to low ('0').

#### 1.2.6 NO CONNECTION (NC)

No internal connection.

#### 1.2.7 GROUND (Vss)

Ground pin.

#### 1.2.8 LIN BUS (LBUS)

LIN Bus pin. LBUs is a bidirectional LIN bus interface pin and is controlled by the signal TxD. It has an open collector output with a current limitation. To reduce electromagnetic emission, the slopes during signal changes are controlled and the LBUs pin has corner-rounding control for both falling and rising edges.

The internal LIN receiver observes the activities on the LIN bus and generates the output signal RXD that follows the state of the LBUS. A 1<sup>st</sup> degree 160 kHz low-pass input filter optimizes electromagnetic immunity.

## 1.2.9 BATTERY POSITIVE SUPPLY VOLTAGE (VBB)

Battery Positive Supply Voltage pin. An external diode is connected in series to prevent the device from being reversely powered (refer to Figure 1-7).

# 1.2.10 FAULT DETECT OUTPUT/TRANSMITTER ENABLE INPUT (FAULT/TXE)

Fault Detect Output/Transmitter Enable Input pin. The output section is HV-tolerant open-drain (up to 30V). The input section is identical to the TxD section (TTL level, HV-compliant, adaptive pull-up). The internal pull-up resistor may be too weak for some applications. We recommend adding a 10 k $\Omega$  external pull-up resistor to ensure a logic high level. Its state is defined as shown in Table 1-5. The device is placed in Transmitter Off mode whenever this pin is low ('0'), either from an internal fault condition or by external drive

If CS/LWAKE is high ('1'), the FAULT/TXE signals a mismatch between the TXD input and the LBUS level. This can be used to detect a bus contention. Since the bus exhibits a propagation delay, the sampling of the internal compare is debounced to eliminate false faults.

After the device wakes up, the FAULT/TxE indicates what wakes the device if CS/LWAKE remains low ('0') (refer to Table 1-5).

The  $\overline{FAULT}/TxE$  pin sampled at a rate faster than every 10  $\mu s$ .

TABLE 1-3: FAULT/TXE TRUTH TABLE

Tvp	Dva	Libinus	Thermon	FAUL	T/TXE			
TXD In	RxD Out	LINBUS I/O	Thermal Override	External Input	Driven Output	Definition		
					CS = 1			
L	Н	VBB	OFF	H	L	<b>FAULT</b> , TXD driven low, LBUS shorted to VBB (Note 1) or LBUS/TXD permanent dominant detected and Transmit time-out shutdown.		
Н	Н	VBB	OFF	Н	Н	ок		
L	L	GND	OFF	Н	Н	ок		
Н	L	GND	OFF	Н	Н	<b>OK</b> , data is being received from LBUS		
х	Х	VBB	ON	Н	L	FAULT, Transceiver in thermal shutdown		
Х	Х	VBB	Х	L	Х	<b>NO FAULT</b> , the CPU is commanding the transceiver to turn off the transmitter driver		
	CS = 0							
Х	Х	Х	Х	х	L	Wake-up from LIN bus activity		
х	Х	Х	Х	Х	Н	Wake-up from POR		

**Legend:** x = Don't care

**Note 1:** The FAULT/TXE is valid after approximately 25 μs after the TXD falling edge. This is to eliminate false fault reporting during bus propagation delays.

#### 1.3 Fail-Safe Features

#### 1.3.1 GENERAL FAIL-SAFE FEATURES

- An internal pull-down resistor on CS/LWAKE pin disables the transmitter if the pin is floating.
- An internal pull-up resistor on the TXD pin places TXD in high and the LBUS in recessive if the TXD pin is floating.
- High-impedance and low-leakage current on LBUS during loss of power or ground.
- The current limit on LBUS protects the transceiver from being damaged if the pin is shorted to VBB.

#### 1.3.2 THERMAL PROTECTION

The thermal protection circuit monitors the die temperature and is able to shut down the LIN transmitter and voltage regulator.

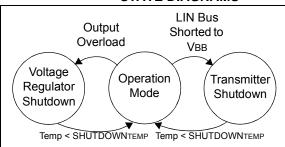
There are three causes for a thermal overload. A thermal shutdown can be triggered by any one, or a combination of, the following thermal overload conditions:

- · Voltage regulator overload
- · LIN bus output overload
- Increase in die temperature due to increase in environment temperature

The recovery time from the thermal shutdown is equal to adequate cooling time.

Driving the TXD and checking the RXD pin make it possible to determine whether there is a bus contention (TXD = high, RXD = low) or a thermal overload condition (TXD = low, RXD = high).

FIGURE 1-2: THERMAL SHUTDOWN STATE DIAGRAMS



#### 1.3.3 TXD/LBUS TIME-OUT TIMER

The LIN bus can be driven to a dominant level either from the TxD pin or externally. An internal timer deactivates the LBUs transmitter if a dominant status (low) on the LIN bus lasts longer than Bus Dominant Time-out Time, tTO(LIN) (approximately 20 milliseconds). At the same time, the RxD output is put in recessive (high), FAULT/TxE is also driven to low and the internal LIN pull-up resistor is disconnected. The timer is reset on any recessive LBUs status or POR mode. The recessive status on LBUs can be caused either by the bus being externally pulled up or by the TxD pin being returned high.

#### 1.4 Internal Voltage Regulator

The MCP2021A/2A have a positive regulator capable of supplying +5.00 or +3.30 VDC  $\pm3\%$  at up to 70 mA of load current over the entire operating temperature range of  $-40^{\circ}$ C to  $+125^{\circ}$ C. The regulator uses a LDO design, is short-circuit-protected and will turn the regulator output off if its output falls below the shutdown voltage threshold, VSD.

With a load current of 70 mA, the minimum input to output voltage differential required for the output to remain in regulation is typically +0.5V (+1V maximum over the full operating temperature range). Quiescent current is less than 100  $\mu A$  with a full 70 mA load current when the input to output voltage differential is greater than +3.00V.

Regarding the correlation between VBB, VREG and IDD, please refer to Figures 1-4 and 1-5. When the input voltage (VBB) drops below the differential needed to provide stable regulation, the voltage regulator output, VREG, will track the input down to approximately VOFF, at which point the regulator will turn off the output. This will allow PIC<sup>®</sup> microcontrollers with internal POR circuits to generate a clean arming of the POR trip point. The MCP2021A/2A will then monitor VBB and turn on the regulator when VBB is above the threshold of regulator turn-on voltage, VON.

In Power-Down mode, the VBB monitor is turned off.

Under specific ambient temperature and battery voltage range, the voltage regulator can output as high as 150 mA current. For current load capability of the voltage regulator, refer to Figures 2-8 and 2-9.

Note: The regulator has an overload current limit of approximately 250 mA. The regulator output voltage, VREG, is monitored. If output voltage VREG is lower than VSD, the voltage regulator will turn off. After a recovery time of about 3 ms, the VREG will be checked again. If there is no short circuit (VREG > VSD), then the voltage regulator remains on.

The regulator requires an external output bypass capacitor for stability. See Figure 2-1 for correct capacity and ESR for stable operation.

Note: A ceramic capacitor of at least 10 μF or a tantalum capacitor of at least 2.2 μF is recommended for stability.

In worst-case scenarios, the ceramic capacitor may derate by 50%, based on tolerance, voltage and temperature. Therefore, in order to ensure stability, ceramic capacitors smaller than 10  $\mu F$  may require a small series resistance to meet the ESR requirements, as shown in Table 1-4.

TABLE 1-4: RECOMMENDED SERIES
RESISTANCE FOR CERAMIC
CAPACITORS

Resistance	Capacitor
1Ω	1 μF
$0.47\Omega$	2.2 µF
$0.22\Omega$	4.7 µF
0.1Ω	6.8 µF

FIGURE 1-3: VOLTAGE REGULATOR BLOCK DIAGRAM

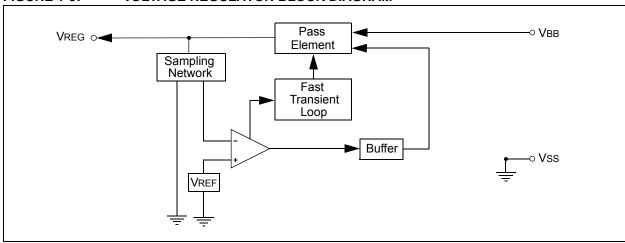


FIGURE 1-4: VOLTAGE REGULATOR OUTPUT ON POWER-ON RESET

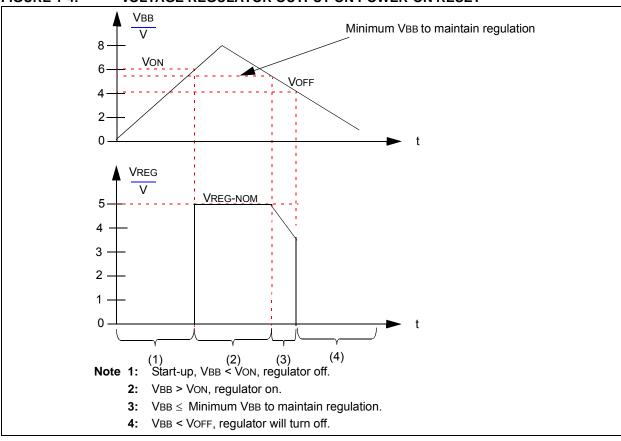
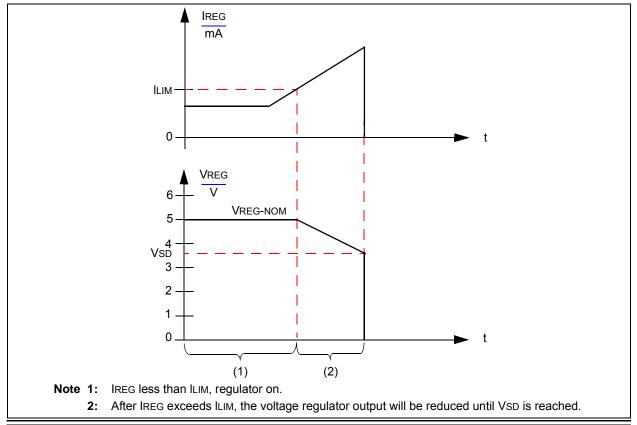


FIGURE 1-5: VOLTAGE REGULATOR OUTPUT ON OVERCURRENT PROTECTION



#### 1.5 Optional External Protection

#### 1.5.1 REVERSE BATTERY PROTECTION

An external reverse-battery-blocking diode should be used to provide polarity protection (see Figure 1-7).

## 1.5.2 TRANSIENT VOLTAGE PROTECTION (LOAD DUMP)

An external 43V transient suppressor (TVS) diode, between VBB and ground, with a transient protection resistor (RTP) in series with the battery supply and the VBB pin, protects the device from power transients and ESD events greater than 43V (see Figure 1-7). The maximum value for the RTP protection resistor depends upon two parameters: the minimum voltage the part will start at and the impacts of this RTP resistor on the VBB value, thus on the bus recessive level and slopes.

This leads to a set of three equations to fulfill.

Equation 1-1 provides a maximum RTP value according to the minimum battery voltage the user wants.

Equation 1-2 provides a maximum RTP value according to the maximum error on the recessive level, thus VBB, since the part uses VBB as the reference value for the recessive level.

Equation 1-3 provides a maximum RTP value according to the maximum relative variation the user can accept on the slope when IREG varies.

Since both Equations 1-1 and 1-2 must be fulfilled, the maximum allowed value for RTP is thus the smaller of the two values found when solving Equations 1-1 and 1-2.

Usually Equation 1-1 gives the higher constraint (smaller value) for RTP, as shown in the following example where  $V_{BATmin}$  is 8V.

However, the user needs to check that the value found with Equation 1-1 fulfills Equations 1-2 and 1-3.

While this protection is optional, it should be considered as good engineering practice.

#### **EQUATION 1-1:**

$$R_{TP} \le \frac{V_{BATmin} - 5.5V}{250 \text{ mA}}$$

$$5.5V = V_{OFF} + 1.0V$$

Where:

250 mA = Peak current at power-on when VBB = 5.5V

Assume  $V_{BATmin}$  = 8V. Equation 1-1 shows 10 $\Omega$ .

#### **EQUATION 1-2:**

$$R_{TP} \leq \frac{\Delta V_{RECESSIVE}}{I_{REGMAX}}$$

Where:

ΔVRECESSIVE = Maximum variation tolerated on the recessive level

Assume  $\Delta V$ RECESSIVE = 1V and IREGMAX = 50 mA. Equation 1-2 shows  $20\Omega$ .

#### **EQUATION 1-3:**

$$R_{TP} \leq \frac{\Delta Slope \times (V_{BATmin} - IV)}{I_{REGMAX}}$$

Where:

 $\Delta$ Slope = Maximum variation tolerated on the

slope level

IREGMAX = Maximum current the current will

provide to the load

V<sub>BATmin</sub> > Voff + 1.0V

Assume  $\Delta \text{Slope} = 15\%$ , VBATMIN = 8V and IREGMAX = 50 mA. Equation 1-2 shows  $20\Omega$ .

#### 1.5.3 CBAT CAPACITOR

Selecting CBAT = 10 x CREG is recommended. However, this leads to a high value capacitor. Lower values for CBAT capacitor can be used with respect to some rules. In any case, the voltage at the VBB pin should remain above VOFF when the device is turned on.

The current peak at start-up (due to the fast charge of the CREG and CBAT capacitors) may induce a significant drop on the VBB pin. This drop is proportional to the impedance of the VBAT connection (see Figure 1-7).

The VBAT connection is mainly inductive and resistive. Therefore, it can be modeled as a resistor (RTOT) in series with an inductor (L). RTOT and L can be measured.

The following formula gives an indication of the minimum value of CBAT using RTOT and L:

#### **EQUATION 1-4:**

$$\frac{C_{BAT}}{C_{REG}} = \sqrt{\frac{100L^2 + R_{TOT}^2}{1 + L^2 + \frac{R_{TOT}^2}{100}}}$$

Where:

L = Inductor (measured in mH)

RTOT = RLINE + RTP (measured in  $\Omega$ )

## MCP2021A/2A

Equation 1-4 allows lower CBAT/CREG values than the 10x ratio we recommend.

Assume that we have a good quality VBAT connection with RTOT =  $0.1\Omega$  and L = 0.1 mH.

Solving the equation gives CBAT/CREG = 1.

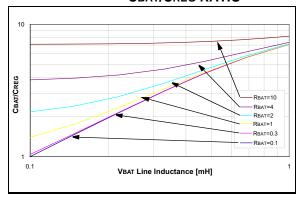
If we increase RTOT up to  $1\Omega$ , the result becomes CBAT/CREG = 1.4. However, if the connection is highly resistive or highly inductive (poor connection), the CBAT/CREG ratio greatly increases.

TABLE 1-5: CBAT/CREG RATIO BY VBAT CONNECTION TYPE

Connection Type	Rтот L		Сват/Спес Ratio		
Good	0.1Ω	0.1 mH	1		
Typical	1Ω	0.1 mH	1.4		
Highly inductive	0.1Ω	1 mH	7		
Highly resistive	10Ω	0.1 mH	7		

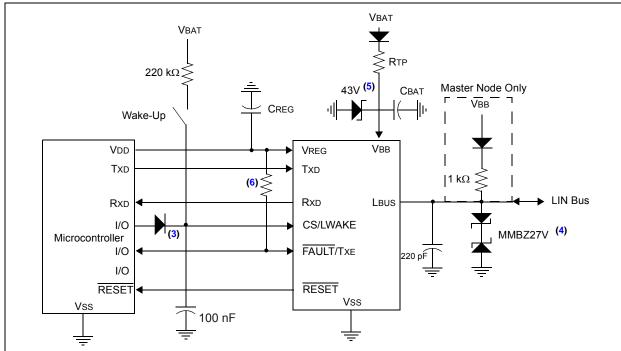
Figure 1-6 shows the minimum recommended CBAT/CREG ratio as a function of the impedance of the VBAT connection.

FIGURE 1-6: MINIMUM
RECOMMENDED
CBAT/CREG RATIO



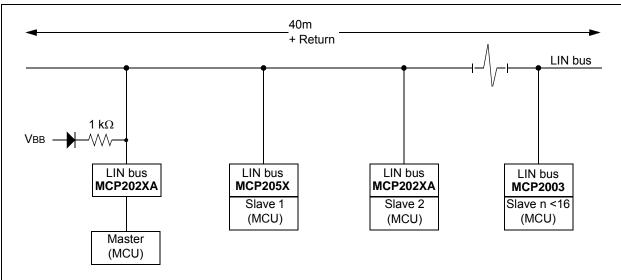
#### 1.6 Typical Applications

#### FIGURE 1-7: TYPICAL APPLICATION CIRCUIT



- Note 1: CREG, the load capacitor, should be ceramic or tantalum rated for extended temperatures,  $1.0 22 \mu F$ . See Figure 2-1 to select the correct ESR.
  - 2: CBAT is the filter capacitor for the external voltage supply. Typically 10 x CREG, with no ESR restriction. See Figure 1-6 to select the minimum recommended value for CBAT. The RTP value is added to the line resistance.
  - 3: This diode is only needed if CS/LWAKE is connected to the VBAT supply.
  - 4: ESD protection diode.
  - 5: This component is for additional load dump protection.
  - **6:** An external 10 k $\Omega$  resistor is recommended for some applications.

#### FIGURE 1-8: TYPICAL LIN NETWORK CONFIGURATION



## MCP2021A/2A

#### 1.7 ICSP™ Considerations

The following should be considered when the MCP2021A/2A are connected to pins supporting in-circuit programming:

- Power used for programming the microcontroller can be supplied from the programmer or from the MCP2021A/2A.
- The voltage on the VREG pin should not exceed the maximum value of VREG in DC Specifications.

#### 2.0 ELECTRICAL CHARACTERISTICS

#### 2.1 Absolute Maximum Ratings†

VIN DC Voltage on RXD and RESET	0.3V to VREG + 0.3
VIN DC Voltage on TxD, CS/LWAKE, FAULT/TxE	0.3 to +40V
VBB Battery Voltage, continuous, non-operating (Note 1)	0.3 to +40V
VBB Battery Voltage, non-operating (LIN bus recessive, no regulator load, t < 60s) (Note 2)	0.3 to +43V
VBB Battery Voltage, transient ISO 7637 Test 1	100V
VBB Battery Voltage, transient ISO 7637 Test 2a	+75V
VBB Battery Voltage, transient ISO 7637 Test 3a	
VBB Battery Voltage, transient ISO 7637 Test 3b	+100V
VLBUS Bus Voltage, continuous	18 to +30V
VLBUS Bus Voltage, transient (Note 3)	
ILBUS Bus Short Circuit Current Limit	200 mA
ESD protection on LIN, VBB (IEC 61000-4-2) (Note 4)	±15 kV
ESD protection on LIN, VBB (Human Body Model) (Note 5)	±8 kV
ESD protection on all other pins (Human Body Model) (Note 5)	±4 kV
ESD protection on all pins (Charge Device Model) (Note 6)	±1500V
ESD protection on all pins (Machine Model) (Note 7)	±200V
Maximum Junction Temperature	150°C
Storage Temperature	65 to +150°C

**† Notice:** Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

Note 1: LIN 2.x compliant specification.

- 2: SAE J2602-2 compliant specification.
- 3: ISO 7637/1 load dump compliant (t < 500 ms).
- **4:** According to IEC 61000-4-2,  $330\Omega$ , 150 pF and Transceiver EMC Test Specifications [2] to [4].
- 5: According to AEC-Q100-002/JESD22-A114.
- 6: According to AEC-Q100-011B.
- 7: According to AEC-Q100-003/JESD22-A115.

#### 2.2 Nomenclature Used in This Document

Some terms and names used in this data sheet deviate from those referred to in the LIN specifications. Equivalent values are shown below.

LIN 2.1 Name	Term used in the following tables	Definition
VBAT	not used	ECU operating voltage
Vsup	VBB	Supply voltage at device pin
VBUS_LIM	Isc	Current limit of driver
VBUSREC	VIH(LBUS)	Recessive state
VBUSDOM	VIL(LBUS)	Dominant state

## MCP2021A/2A

#### 2.3 DC Specifications

DC Specifications	Electrical Char VBB = 6.0V to 1					limits are specified for
Parameter	Sym	Min	Тур	Max	Units	Conditions
Power	•					
VBB Quiescent Operating Current	IBBQ	_	_	200	μA	IOUT = 0 mA LBUS recessive VREG = 5.0V
			_	200	μΑ	IOUT = 0 mA LBUS recessive VREG = 3.3V
VBB Ready Current	IBBRD	I		100	μΑ	IOUT = 0 mA LBUS recessive VREG = 5.0V
		I		100	μΑ	IOUT = 0 mA LBUS recessive VREG = 3.3V
VBB Transmitter-Off Current	Іввто	ľ	_	100	μА	With voltage regulator on, transmitter off, receiver on, FAULT/TXE = VIL, CS = VIH, VREG = 5.0V
		1	_	100	μA	With voltage regulator on, transmitter off, receiver on, FAULT/TXE = VIL, CS = VIH, VREG = 3.3V
VBB Power-Down Current	IBBPD	_	4.5	8	μА	With voltage regulator off, receiver on and transmitter off, FAULT/TXE = VIH, TXD = VIH, CS = VIL
VBB Current with Vss Floating	IBBNOGND	-1	_	1	mA	VBB = 12V, GND to VBB, VLIN = 0 – 18V
Microcontroller Interface						
High-Level Input Voltage (TxD, FAULT/TxE)	ViH	2.0	_	VREG +0.3	V	
Low-Level Input Voltage (TxD, FAULT/TxE)	VIL	-0.3		0.8	V	
High-Level Input Current (TXD, FAULT/TXE)	lін	-2.5		0.4	μА	Input voltage = 4.0V ~800 kΩ internal adaptive pull-up
Low-Level Input Current (TXD, FAULT/TXE)	lı∟	-10		_	μA	Input voltage = 0.5V ~800 kΩ internal adaptive pull-up
High-Level Input Voltage (CS/LWAKE)	VIH	2.0		VBB	V	Through a current-limiting resistor
Low-Level Input Voltage (CS/LWAKE)	VIL	-0.3		0.8	V	

Note 1: Internal current limited. 2.0 ms maximum recovery time (RLBUS = 0W, TX = 0, VLBUS = VBB).

<sup>2:</sup> Characterized, not 100% tested.

**<sup>3:</sup>** In Power-Down mode, normal LIN recessive/dominant threshold is disabled. VWK(LBUS) is used to detect bus activities.

#### 2.3 DC Specifications (Continued)

CS/LWAKE	DC Specifications	Electrical Char VBB = 6.0V to 18					limits are specified for
CS/LWAKE	Parameter	Sym	Min	Тур	Max	Units	Conditions
CS/LWAKE		ІІН	_	_	8.0	μA	
RXD    High-Level Output Voltage   VOHRXD    0.8VREG		li∟	-		5.0	μA	
(RXD)   Low-Level Output Voltage   VoLoD   —   —   1.0   V   IoL = 4 mA   (FAULT/TXE)   Low-Level Output Voltage   VoLRST   —   —   1.0   V   IoL = 4 mA   (RSET)   Bus Interface   High-Level Input Voltage   ViH(LBUS)   0.6 VBB   —   —   V   Recessive state   Low-Level Input Voltage   ViH(LBUS)   -8   —   0.4 VBB   V   Dominant state   Input Hysteresis   VHYS   —   —   0.175 VBB   V   VIH(LBUS)	,	Volrxd			0.2VREG	V	IOL = 2 mA
(FAULT/TXE)         Low-Level Output Voltage (RESET)         Volrst         —         1.0         V         IoL = 4 mA           Bus Interface         Bus Interface         High-Level Input Voltage         VIH(LBUS)         0.6 VBB         —         —         V         Recessive state           Low-Level Input Voltage         VIH(LBUS)         -8         —         0.4 VBB         V         Dominant state           Input Hysteresis         VHYS         —         —         0.175 VBB         V         VIH(LBUS) – VIL(LBUS)           Low-Level Output Current         IoL(LBUS)         40         —         200         mA         Output voltage = 0.1 VBB = 12V           Pull-Up Current on Input         IPU(LBUS)         -180         —         -72         µA         ~30 KΩ internal pull-L@ VH(LBUS)         -30 KΩ internal pull-L@ VH(LBUS)         -0.72         µA         Note 1         Note 1           High-Level Output Voltage         VOH(LBUS)         0.8 VBB         —         VBB         V           Driver Dominant Voltage         VoH(LBUS)         0.8 VBB         —         VBB         V           Up HisuP         —         —         1.1         V VBB = 7.3V RLOAD = 1000Ω         RLOAD = 1000Ω         Note 1         PRIOD PRIOD PRIOD PRIOD PRIOD PRIOD PRIOD		VOHRXD	0.8VREG	_	_	V	IOH = 2 mA
RESET   Bus Interface   High-Level Input Voltage   VIH(LBUS)   0.6 VBB   —   —   V   Recessive state		VOLOD		_	1.0	V	IOL = 4 mA
High-Level Input Voltage   Vih(LBus)   0.6 VBB   —   —   V   Recessive state		Volrst		_	1.0	V	IOL = 4 mA
Low-Level Input Voltage   Vii (LBus)   -8   - 0.4 VBB   V   Dominant state	Bus Interface						
Input Hysteresis	High-Level Input Voltage	VIH(LBUS)	0.6 VBB	_	_	V	Recessive state
Low-Level Output Current   IoL(LBus)   40	Low-Level Input Voltage	VIL(LBUS)	-8		0.4 VBB	V	Dominant state
Pull-Up Current on Input	Input Hysteresis	VHYS		_	0.175 Vвв	V	VIH(LBUS) – VIL(LBUS)
Bus_Pas_Rec   Current (at the receiver during recessive bus level)   Bus_No_GND   Classed Current (disconnected from ground)   Bus_No_PWR   Classed Current (disconnected from VBB)   Classed Current (disconnected VBB = CN V VBB = CN VBB VBB V VBB = CN VBB VBB V VBB = CN VBB V VBB = CN VBB VBB VBB VBB VBB VBB V VBB = CN VBB VBB VBB VBB VBB VBB VBB VBB VBB VB	Low-Level Output Current	IOL(LBUS)	40		200	mA	Output voltage = 0.1 VBB, VBB = 12V
High-Level Output Voltage	Pull-Up Current on Input	IPU(LBUS)	-180	_	-72	μA	~30 kΩ internal pull-up @ VIH(LBUS) = 0.7 VBB, VBB = 12V
	Short Circuit Current Limit	Isc	50	_	200	mA	Note 1
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	High-Level Output Voltage	Voh(LBUS)	0.8 VBB	_	VBB	V	
Input Leakage Current (at the receiver during dominant bus level)  Input Leakage Current (at the receiver during dominant bus level)  Input Leakage Current (at the receiver during recessive bus level)  Leakage Current (disconnected from ground)  IBUS_PAS_REC  -20  -20 $\mu$ A  Driver off 8V < VBB < 18V 8V < VBUS < 18V VBUS < 18V VBUS > VBUS < 18V VBUS	Driver Dominant Voltage	V_LOSUP	_	_	1.1	V	
(at the receiver during dominant bus level)  Input Leakage Current (at the receiver during recessive bus level)  Leakage Current (disconnected from ground)  Leakage Current (disconnected from VBB)  Receiver Center Voltage  VBUS_PAS_REC  -20  — 20  µA Driver off 8V < VBB < 18V 8V < VBUS < 18V VBUS ≥ VBB  AGNDDEVICE = VBB 0V < VBUS < 18V VBB = 12V  PAB = 12V  Leakage Current (disconnected from VBB)  Receiver Center Voltage  VBUS_CNT  0.475 VBB  VBUS_CNT  VBUS_CNT = (VIL(LBU VIH(LBUS))/2		V_HISUP	_	_	1.2	V	1
(at the receiver during recessive bus level)    A	(at the receiver during	IBUS_PAS_DOM	-1		_	mA	VBUS = 0V
	(at the receiver during	IBUS_PAS_REC	-20	_	20	μA	8V < VBB < 18V 8V < VBUS < 18V
(disconnected from VBB)         0 < VBUS < 18V		IBUS_NO_GND	-10	_	+10	μA	0V < VBUS < 18V
VBB VIH(LBUS))/2		IBUS_NO_PWR	-10		+10	μA	
Slave Termination RSLAVE 20 30 47 kO Note 2	Receiver Center Voltage	VBUS_CNT	0.475 VBB		0.525 VBB	V	VBUS_CNT = (VIL(LBUS) + VIH(LBUS))/2
	Slave Termination	RSLAVE	20	30	47	kΩ	Note 2

Note 1: Internal current limited. 2.0 ms maximum recovery time (RLBUS = 0W, TX = 0, VLBUS = VBB).

<sup>2:</sup> Characterized, not 100% tested.

**<sup>3:</sup>** In Power-Down mode, normal LIN recessive/dominant threshold is disabled. VWK(LBUS) is used to detect bus activities.

## MCP2021A/2A

#### 2.3 DC Specifications (Continued)

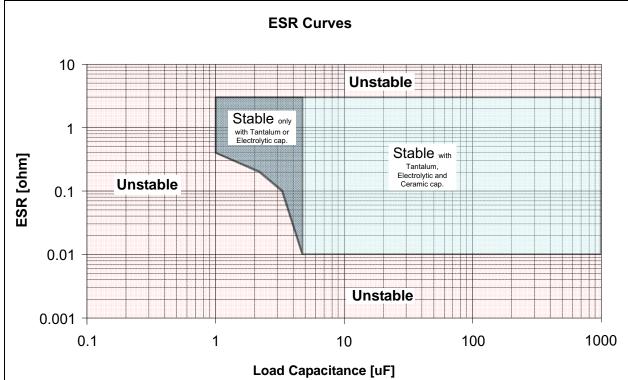
DC Specifications	Electrical Char VBB = 6.0V to 1					limits are specified for
Parameter	Sym	Min	Тур	Max	Units	Conditions
Capacitance of Slave Node	CSLAVE	_		50	pF	Note 2
Wake-Up Voltage Threshold on LIN Bus	VWK(LBUS)	_	_	3.4	V	Wake-up from Power-Down mode (Note 3)
Voltage Regulator – 5.0V						
Output Voltage Range	VREG	4.85	5.00	5.15	V	0 mA < IOUT < 70 mA
Line Regulation	ΔVουτ1	_	10	50	mV	IOUT = 1 mA 6.0V < VBB < 18V
Load Regulation	ΔVουτ2	_	10	50	mV	5 mA < IOUT < 70 mA 6.0V < VBB < 12V
Power Supply Ripple Reject	PSRR	_		50	dB	1 VPP @ 10 – 20 kHz ILOAD = 20 mA
Output Noise Voltage	eN	_		100	μVRM S	10 Hz – 40 MHz CFILTER = 10 µf CBP = 0.1 µf ILOAD = 20 mA
Shutdown Voltage Threshold	VsD	3.5	_	4.0	V	See Figure 1-5 (Note 2)
Input Voltage to Turn-Off Output	Voff	3.9	_	4.5	V	
Input Voltage to Turn-On Output	Von	5.25		6.0	V	
Voltage Regulator – 3.3V						
Output Voltage	VREG	3.20	3.30	3.40	V	0 mA < IOUT < 70 mA
Line Regulation	ΔVουτ1	_	10	50	mV	IOUT = 1 mA 6.0V < VBB < 18V
Load Regulation	ΔVουτ2	_	10	50	mV	5 mA < IOUT < 70 mA 6.0V < VBB < 12V
Power Supply Ripple Reject	PSRR	_		50	dB	1 VPP @ 10 – 20 kHz ILOAD = 20 mA
Output Noise Voltage	eN	_		100	μVRM s/√Hz	10 Hz – 40 MHz CFILTER = 10 μf CBP = 0.1 μf ILOAD = 20 mA
Shutdown Voltage Threshold	VsD	2.5		2.7	V	See Figure 1-5 (Note 2)
Input Voltage to Turn-Off Output	Voff	3.9		4.5	V	
Input Voltage to Turn-On Output	Von	5.25	_	6	V	

**Note 1:** Internal current limited. 2.0 ms maximum recovery time (RLBUS = 0W, TX = 0, VLBUS = VBB).

<sup>2:</sup> Characterized, not 100% tested.

**<sup>3:</sup>** In Power-Down mode, normal LIN recessive/dominant threshold is disabled. VWK(LBUS) is used to detect bus activities.

FIGURE 2-1: ESR CURVES FOR LOAD CAPACITOR SELECTION



**Note 1:** The graph shows the minimum required capacitance after derating due to tolerance, temperature and voltage.

## MCP2021A/2A

#### 2.4 AC Specifications

AC Characteristics	Characteristics Electrical Characteristics: Unless otherwise indicated, all limits are specified for VBB = 6.0V to 18.0V; TA = -40°C to +125°C.								
Parameter	Sym	Min	Тур	Max	Units	Conditions			
Bus Interface – Constant Slope Time Parameters									
Slope Rising and Falling Edges	tslope	3.5	_	22.5	μs	7.3V ≤ VBB ≤ 18V			
Propagation Delay of Transmitter	ttranspd	_		5.0	μs	ttranspd = max. (ttranspdr or ttranspdf)			
Propagation Delay of Receiver	trecpd	ı	1	6.0	μs	trecpd = max. (trecpdr or trecpdf)			
Symmetry of Propagation Delay of Receiver Rising Edge w.r.t. Falling Edge	trecsym	-2.0		2.0	μs	trecsym = max. (trecpdf – trecpdr) Rrxd = 2.4 k $\Omega$ to Vcc Crxd = 20 pF			
Symmetry of Propagation Delay of Transmitter Rising Edge w.r.t. Falling Edge	ttranssym	-2.0	_	2.0	μs	ttranssym = max. (ttranspdf - ttranspdr)			
Bus Dominant Time-Out Time	tto(lin)	1	25	_	ms				
Time to Sample FAULT/TXE for Bus Conflict Reporting	tfault	_	_	32.5	μs	tFAULT = max. (tTRANSPD + tSLOPE + tRECPD)			
Duty Cycle 1 @ 20.0 kbps		0.396	_	_	%tвіт	CBUS; RBUS conditions: 1 nF; 1 k $\Omega$   6.8 nF; 660 $\Omega$   10 nF; 500 $\Omega$ THREC(MAX) = 0.744 x VBB, THDOM(MAX) = 0.581 x VBB, VBB = 7.0V - 18V; tBIT = 50 $\mu$ S. D1 = tBUS_REC(MIN)/2 x tBIT			
Duty Cycle 2 @ 20.0 kbps				0.581	%tвіт	CBUS; RBUS conditions: 1 nF; 1 k $\Omega$   6.8 nF; 660 $\Omega$   10 nF; 500 $\Omega$ THREC(MAX) = 0.284 x VBB, THDOM(MAX) = 0.422 x VBB, VBB = 7.6V - 18V; tBIT = 50 $\mu$ S. D2 = tBUS_REC(MAX)/2 x tBIT			
Duty Cycle 3 @ 10.4 kbps		0.417	_	_	%tвіт	CBUS; RBUS conditions: 1 nF; 1 k $\Omega$   6.8 nF; 660 $\Omega$   10 nF; 500 $\Omega$ THREC(MAX) = 0.778 x VBB, THDOM(MAX) = 0.616 x VBB, VBB = 7.0V - 18V; tBIT = 96 $\mu$ S. D3 = tBUS_REC(MIN)/2 x tBIT			

**Note 1:** Time depends on external capacitance and load. Test condition: CREG = 4.7 μF, no resistor load.

2: Characterized, not 100% tested.

#### 2.4 AC Specifications (Continued)

AC Characteristics	Electrical Characteristics: Unless otherwise indicated, all limits are specified for VBB = 6.0V to 18.0V; TA = -40°C to +125°C.							
Parameter	Sym	Min	Тур	Max	Units	Conditions		
Duty Cycle 4 @ 10.4 kbps		_	_	0.590	%tвіт	CBUS; RBUS conditions: 1 nF; 1 k $\Omega$   6.8 nF; 660 $\Omega$   10 nF; 500 $\Omega$ THREC(MAX) = 0.251 x VBB, THDOM(MAX) = 0.389 x VBB, VBB = 7.6V - 18V; tBIT = 96 $\mu$ S. D4 = tBUS_REC(MAX)/2 x tBIT		
Voltage Regulator								
Bus Activity Debounce time	tBDB	30	80	250	μs			
Bus Activity to Voltage Regulator Enabled	<b>t</b> BACTIVE	35	_	200	μs			
Voltage Regulator Enabled to Ready	tvevr	300	_	1200	μs	Note 1		
Chip Select to Ready Mode	tcsr	_	_	230	μs			
Chip Select to Power-Down	tcspd	_	_	330	μs	Note 2		
Short Circuit to Shutdown	tshutdown	20	_	100	μs			
RESET Timing								
VREG OK Detect to RESET Inactive	tRPU	_	_	60	μs			
VREG Not OK Detect to RESET Active	tRPD	_	_	60	μs			

- Note 1: Time depends on external capacitance and load. Test condition: CREG = 4.7 μF, no resistor load.
  - 2: Characterized, not 100% tested.

#### 2.5 Thermal Specifications

Parameter	Sym	Min	Тур	Max	Units	Test Conditions
Specified Temperature Range	TA	-40	_	+125	°C	
Maximum Junction Temperature	TJ	_	_	+150	°C	
Storage Temperature Range	TA	-65	_	+150	°C	
Recovery Temperature	$\theta$ RECOVERY	_	+140	_	°C	
Shutdown Temperature	θSHUTDOWN	_	+150	_	°C	
Short Circuit Recovery Time	ttherm	_	1.5	5.0	ms	
Thermal Package Resistances						
Thermal Resistance, 8L-PDIP	θJA	_	89.3	_	°C/W	
Thermal Resistance, 8L-SOIC	θJA	_	149.5	_	°C/W	
Thermal Resistance, 8L-DFN	θJA	_	48	_	°C/W	
Thermal Resistance, 14L-PDIP	θЈА	_	70	_	°C/W	
Thermal Resistance, 14L-SOIC	θJA	_	90.8	_	°C/W	
Thermal Resistance, 14L-TSSOP	θЈА	_	100	_	°C/W	

**Note 1:** The maximum power dissipation is a function of TJMAX,  $\theta$ JA, and ambient temperature, TA. The maximum allowable power dissipation at an ambient temperature is PD = (TJMAX – TA)  $\theta$ JA. If this dissipation is exceeded, the die temperature will rise above 150°C and the MCP2021A/2A will go into thermal shutdown.

#### 2.6 Typical Performance Curves

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

**Note:** Unless otherwise indicated, VBB = 6.0V to 18.0V; TA =  $-40^{\circ}C$  to  $+125^{\circ}C$ .

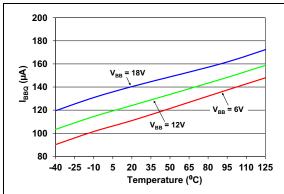
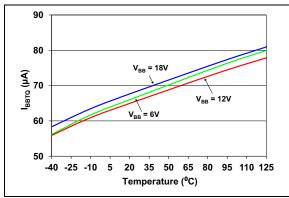


FIGURE 2-2: Typical IBBQ vs.

Temperature - 5.0V.



**FIGURE 2-3:** Typical IBBTO vs. Temperature – 5.0V.

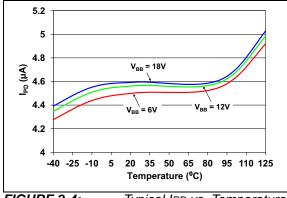


FIGURE 2-4: Typical IPD vs. Temperature – 5.0V.

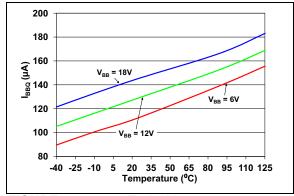


FIGURE 2-5: Typical IBBQ vs. Temperature – 3.3V.

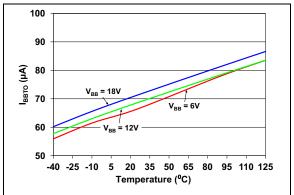


FIGURE 2-6: Typical IBBTO vs. Temperature – 3.3V.

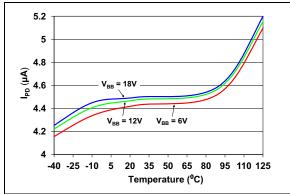
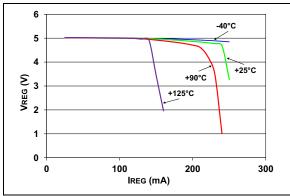
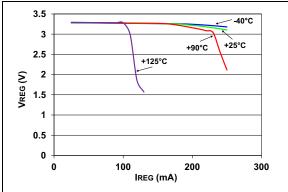


FIGURE 2-7: Typical IPD vs. Temperature – 3.3V.



**FIGURE 2-8:** 5.0V VREG vs. IREG at VBB = 12 V.



**FIGURE 2-9:** 3.3V VREG vs. IREG at VBB = 12V.

#### 2.7 Timing Diagrams and Specifications

FIGURE 2-10: BUS TIMING DIAGRAM

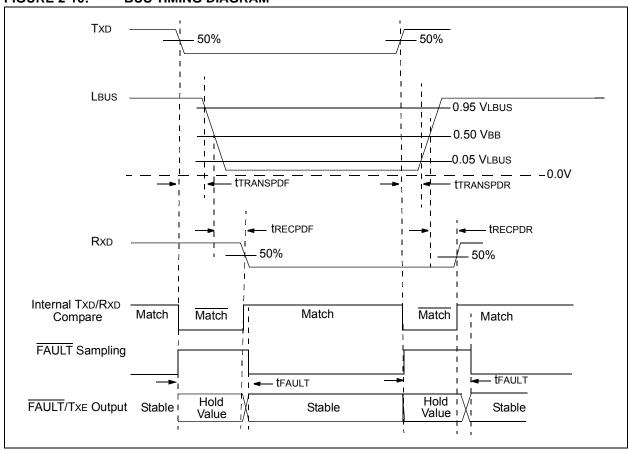
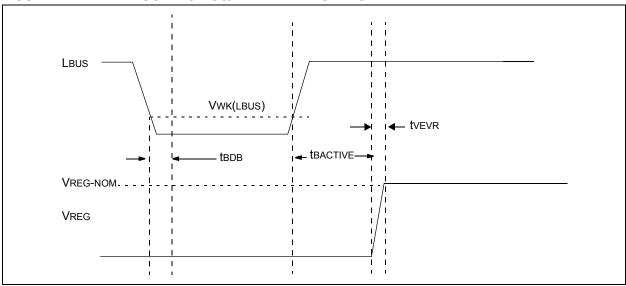
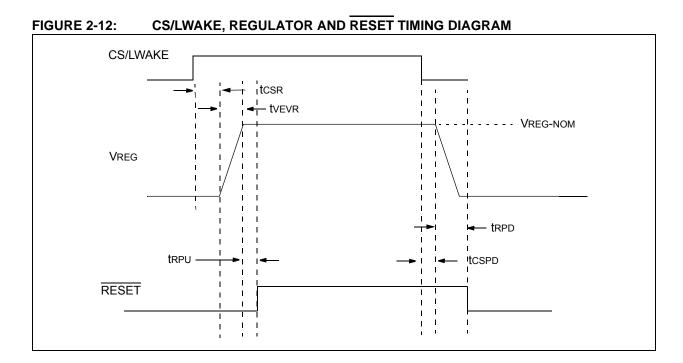


FIGURE 2-11: REGULATOR CS/LWAKE TIMING DIAGRAM

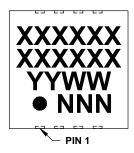




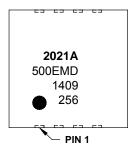
#### 3.0 PACKAGING INFORMATION

#### 3.1 Package Marking Information

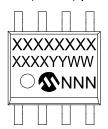
8-Lead DFN (4x4x0.9 mm) (MCP2021A)



Example



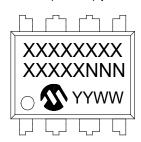
8-Lead SOIC (150 mil) (MCP2021A)



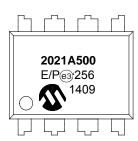
Example



8-Lead PDIP (300 mil) (MCP2021A)



Example



**Legend:** XX...X Customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

e3 Pb-free JEDEC® designator for Matte Tin (Sn)

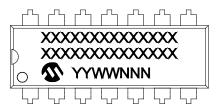
This package is Pb-free. The Pb-free JEDEC designator (e3)

can be found on the outer packaging for this package.

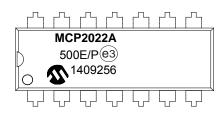
**Note**: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

#### **Package Marking Information (Continued)**

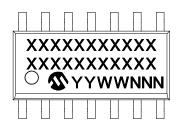
14-Lead PDIP (300 mil) (MCP2022A)



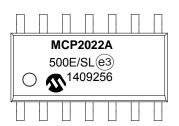
Example



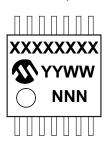
14-Lead SOIC (.150") (MC2022A)



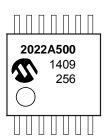
Example



14-Lead TSSOP (MCP2022A)

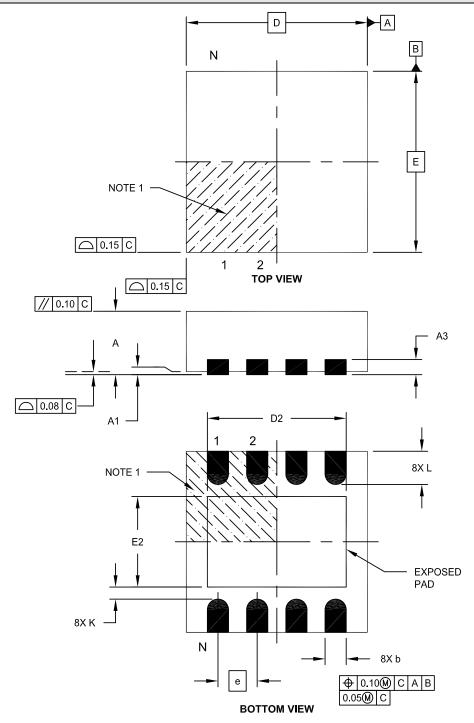


Example



#### 8-Lead Plastic Dual Flat, No Lead Package (MD) – 4x4x0.9 mm Body [DFN]

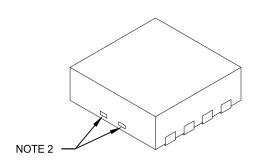
**te:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-131E Sheet 1 of 2

#### 8-Lead Plastic Dual Flat, No Lead Package (MD) - 4x4x0.9 mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX
Number of Pins	N		8	
Pitch	е		0.80 BSC	
Overall Height	Α	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Length	D		4.00 BSC	
Exposed Pad Width	E2	2.60	2.70	2.80
Overall Width	E	4.00 BSC		
Exposed Pad Length	D2	3.40	3.50	3.60
Contact Width	b	0.25	0.30	0.35
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Package may have one or more exposed tie bars at ends.
- 3. Package is saw singulated
- 4. Dimensioning and tolerancing per ASME Y14.5M

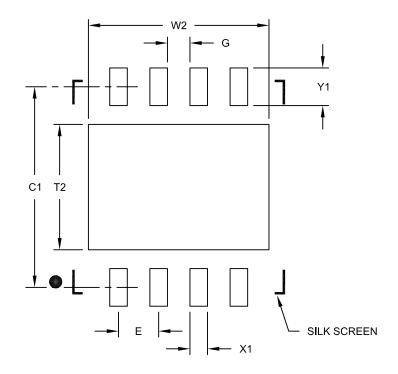
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-131E Sheet 2 of 2

#### 8-Lead Plastic Dual Flat, No Lead Package (MD) - 4x4x0.9 mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

Units		N.	ILLIMETER:	S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	E	0.80 BSC		
Optional Center Pad Width	W2			3.60
Optional Center Pad Length	T2			2.50
Contact Pad Spacing	C1		4.00	
Contact Pad Width (X8)	X1			0.35
Contact Pad Length (X8)	Y1			0.75
Distance Between Pads	G	0.45		

#### Notes:

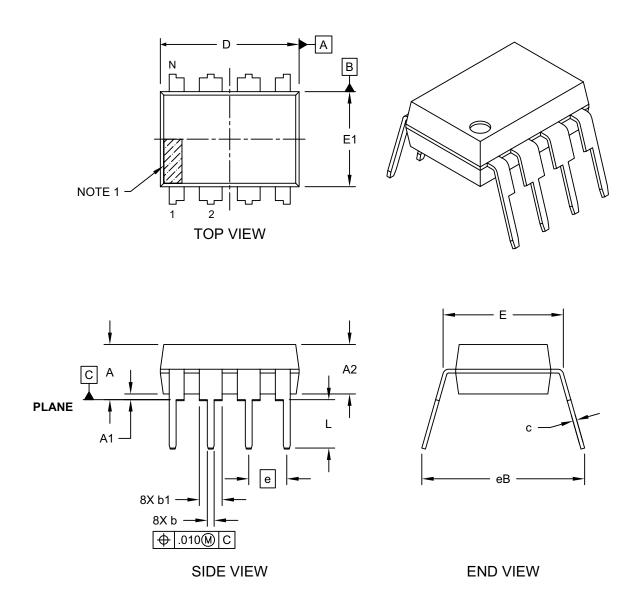
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2131C

#### 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

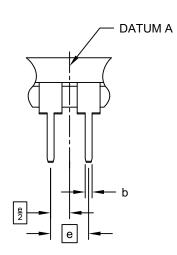
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



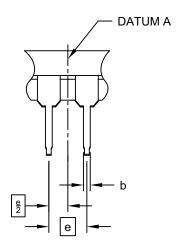
Microchip Technology Drawing No. C04-018D Sheet 1 of 2

#### 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



### ALTERNATE LEAD DESIGN (VENDOR DEPENDENT)



Units			INCHES	
Dimension	Limits	MIN	NOM	MAX
Number of Pins	N		8	
Pitch	е		.100 BSC	
Top to Seating Plane	Α	-	-	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	Е	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.348	.365	.400
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.040	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eВ	-	-	.430

#### Notes:

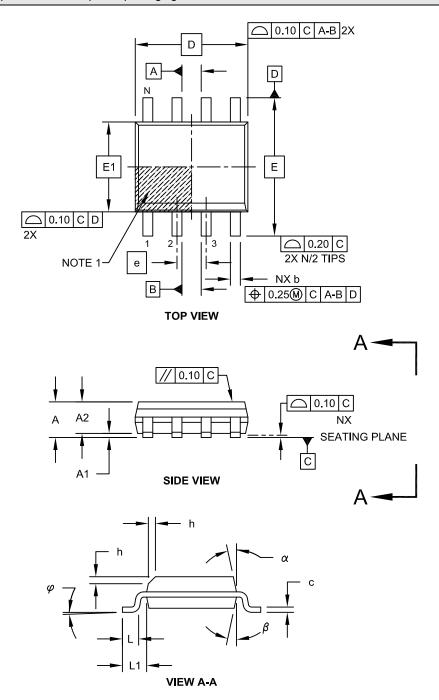
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-018D Sheet 2 of 2

#### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

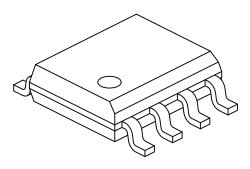
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing No. C04-057C Sheet 1 of 2

#### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

te: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		8	
Pitch	е		1.27 BSC	
Overall Height	Α	-	-	1.75
Molded Package Thickness	A2	1.25	-	-
Standoff §	A1	0.10	-	0.25
Overall Width	E	6.00 BSC		
Molded Package Width	E1	3.90 BSC		
Overall Length	D	4.90 BSC		
Chamfer (Optional)	h	0.25	-	0.50
Foot Length	L	0.40	-	1.27
Footprint	L1		1.04 REF	
Foot Angle	φ	0°	-	8°
Lead Thickness	С	0.17	-	0.25
Lead Width	b	0.31	-	0.51
Mold Draft Angle Top	α	5°	-	15°
Mold Draft Angle Bottom	β	5°	-	15°

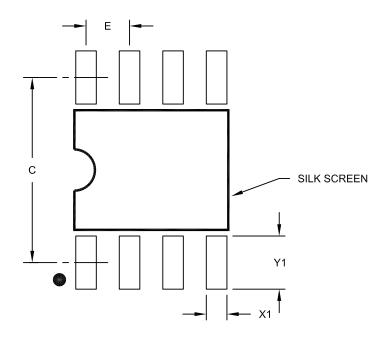
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-057C Sheet 2 of 2

#### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width (X8)	X1			0.60
Contact Pad Length (X8)	Y1			1.55

#### Notes:

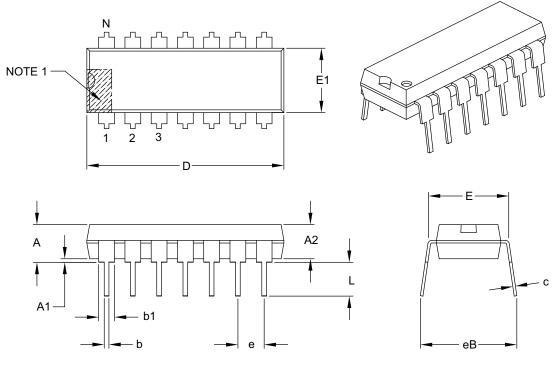
1. Dimensioning and tolerancing per ASME Y14.5M  $\,$ 

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2057A

#### 14-Lead Plastic Dual In-Line (P) – 300 mil Body [PDIP]

**>>te:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimension	n Limits	MIN	NOM	MAX
Number of Pins	N		14	
Pitch	е		.100 BSC	
Top to Seating Plane	Α	-	_	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	_	-
Shoulder to Shoulder Width	Е	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.735	.750	.775
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.045	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	_	_	.430

#### Notes:

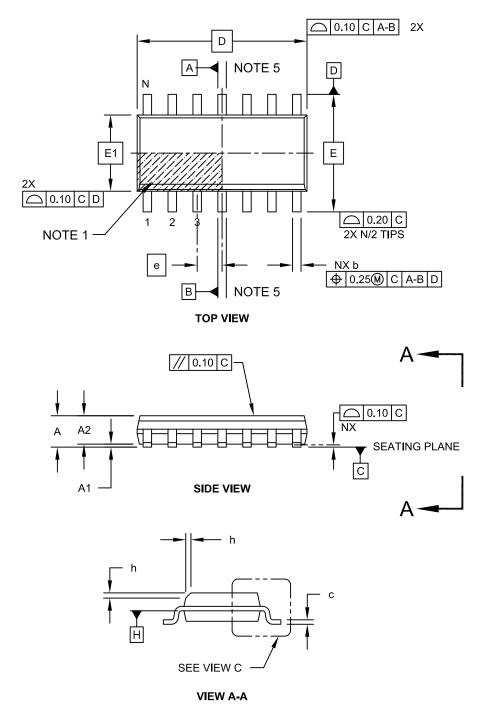
- 1. Pin 1 visual index feature may vary, but must be located with the hatched area.
- 2. § Significant Characteristic.
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-005B

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

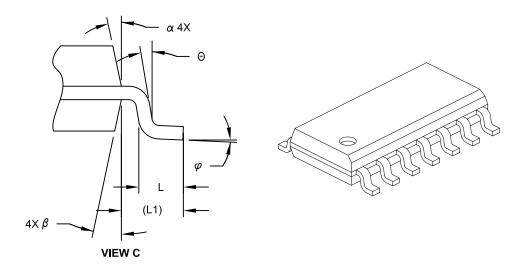
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing No. C04-065C Sheet 1 of 2

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		14	
Pitch	е		1.27 BSC	
Overall Height	Α	ı	-	1.75
Molded Package Thickness	A2	1.25	-	-
Standoff §	A1	0.10	-	0.25
Overall Width	Е		6.00 BSC	
Molded Package Width	E1		3.90 BSC	
Overall Length	D	8.65 BSC		
Chamfer (Optional)	h	0.25	-	0.50
Foot Length	L	0.40	-	1.27
Footprint	L1		1.04 REF	
Lead Angle	Θ	0°	-	-
Foot Angle	$\varphi$	0°	-	8°
Lead Thickness	С	0.10	-	0.25
Lead Width	b	0.31	-	0.51
Mold Draft Angle Top	α	5°	-	15°
Mold Draft Angle Bottom	β	5°	-	15°

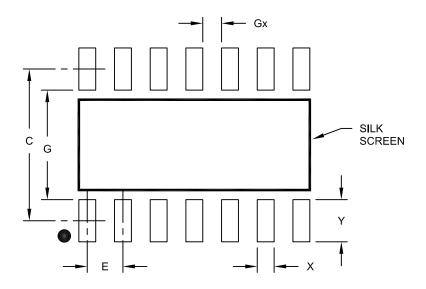
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-065C Sheet 2 of 2

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units		<i>I</i> ILLIMETER	S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width	Х			0.60
Contact Pad Length	Υ			1.50
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	3.90		

#### Notes

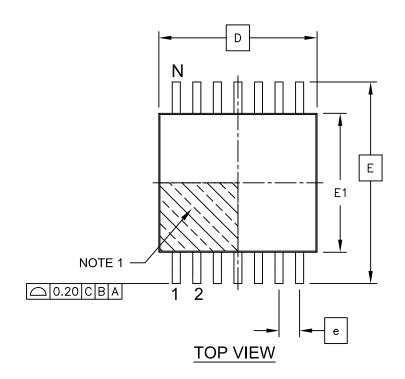
1. Dimensioning and tolerancing per ASME Y14.5M

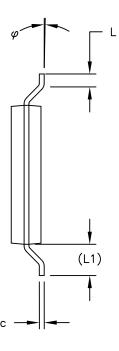
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

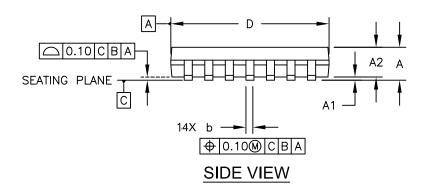
Microchip Technology Drawing No. C04-2065A

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



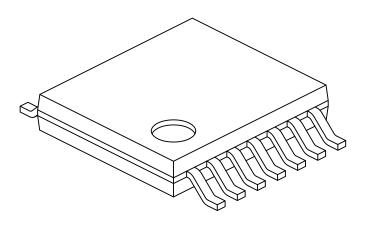




Microchip Technology Drawing C04-087C Sheet 1 of 2

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension I	Limits	MIN	NOM	MAX
Number of Pins	Z		14	
Pitch	е		0.65 BSC	
Overall Height	Α	ı	-	1.20
Molded Package Thickness	A2	0.80	1.00	1.05
Standoff	A1	0.05	ı	0.15
Overall Width	Е		6.40 BSC	
Molded Package Width	E1	4.30	4.40	4.50
Molded Package Length	О	4.90	5.00	5.10
Foot Length	Г	0.45	0.60	0.75
Footprint	(L1)	1.00 REF		
Foot Angle	$\varphi$	0°	-	8°
Lead Thickness	С	0.09	-	0.20
Lead Width	b	0.19	-	0.30

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M

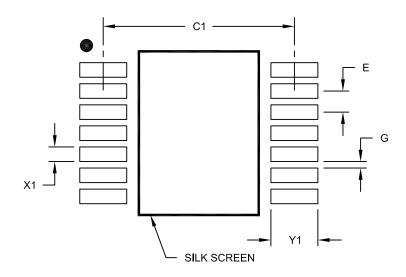
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-087C Sheet 2 of 2

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

Units		V	<b>IILLIMETER</b>	S
Dimension Limits		MIN	NOM	MAX
Contact Pitch	Е		0.65 BSC	
Contact Pad Spacing	C1		5.90	
Contact Pad Width (X14)	X1			0.45
Contact Pad Length (X14)	Y1			1.45
Distance Between Pads	G	0.20		

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2087A

#### APPENDIX A: REVISION HISTORY

#### Revision C (July 2014)

The following is the list of modifications:

- Updated Section 1.6 "Typical Applications" with values used during ESD tests.
- 2. Minor typographical corrections.
- 3. Updated 8-Lead PDIP Package.

#### **Revision B (December 2013)**

The following is the list of modifications:

- 1. Removed two notes in AC Specifications.
- 2. Updated Figure 1-3.
- 3. Added pull-up to FAULT/TXE pin in the pin description and typical applications.
- 4. Revised product identification examples for SOIC package.

#### Revision A (March 2012)

· Original Release of this Document.

#### PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO	-X -X /XX	Examples:
Device Volta	age Temperature Package Range	a) MCP2021A-330E/MD: 3.3V, 8-lead DFN package b) MCP2021A-500E/MD: 5.0V, 8-lead DFN package c) MCP2021AT-330E/MD: 3.3V, 8-lead DFN package, Tape and Reel d) MCP2021AT-500E/MD: 5.0V, 8-lead DFN package.
Device:	MCP2021A: LIN Transceiver with Voltage Regulator MCP2021AT: LIN Transceiver with Voltage Regulator (Tape and Reel)  MCP2022AT: LIN Transceiver with Voltage Regulator MCP2022AT: LIN Transceiver with Voltage Regulator (Tape and Reel)	Tape and Reel e) MCP2021A-330E/P: 3.3V, 8-lead PDIP package f) MCP2021A-500E/P: 5.0V, 8-lead PDIP package g) MCP2021A-330E/SN: 3.3V, 8-lead SOIC package h) MCP2021AT-330E/SN: 3.3V, 8-lead SOIC package, Tape and Reel i) MCP2021A-500E/SN: 5.0V, 8-lead SOIC package j) MCP2021AT-500E/SN: 5.0V, 8-lead SOIC package,
Voltage:	330 = 3.3V 500 = 5.0V	Tape and Reel  a) MCP2022A-330E/P: 3.3V, 14-lead PDIP package
Temperature Range:	E = -40°C to +125°C	b) MCP2022A-500E/P: 5.0V, 14-lead PDIP package c) MCP2022A-330E/SL: 3.3V, 14-lead SOIC package d) MCP2022AT-330E/SL: 3.3V, 14-lead SOIC package, Tape and Reel
Package:	MD = 8LD Plastic Dual Flat, No Lead – 4x4x0.8 mm Body P = 8LD/14LD Plastic Dual In-Line – 300 mil Body SN = 8LD Plastic Small Outline – Narrow, 3.90 mm Body SL = 14LD Plastic Small Outline – Narrow, 3.90 mm Body ST = 14LD Plastic Thin Shrink Small Outline – 4.4 mm Body	e) MCP2022A-500E/SL: 5.0V, 14-lead SOIC package f) MCP2022AT-500E/SL: 5.0V, 14-lead SOIC package, Tape and Reel g) MCP2022A-330E/ST: 3.3V, 14-lead TSSOP package, Tape and Reel i) MCP2022AT-330E/ST: 3.3V, 14-lead TSSOP package, Tape and Reel i) MCP2022AT-500E/ST: 5.0V, 14-lead TSSOP package j) MCP2022AT-500E/ST: 5.0V, 14-lead TSSOP package, Tape and Reel

#### Note the following details of the code protection feature on Microchip devices:

- · Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our
  knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data
  Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- · Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

#### Trademarks

The Microchip name and logo, the Microchip logo, dsPIC, FlashFlex, flexPWR, JukeBlox, KEELoQ, KEELoQ logo, Kleer, LANCheck, MediaLB, MOST, MOST logo, MPLAB, OptoLyzer, PIC, PICSTART, PIC<sup>32</sup> logo, RightTouch, SpyNIC, SST, SST Logo, SuperFlash and UNI/O are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

The Embedded Control Solutions Company and mTouch are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, BodyCom, chipKIT, chipKIT logo, CodeGuard, dsPICDEM, dsPICDEM.net, ECAN, In-Circuit Serial Programming, ICSP, Inter-Chip Connectivity, KleerNet, KleerNet logo, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, RightTouch logo, REAL ICE, SQI, Serial Quad I/O, Total Endurance, TSHARC, USBCheck, VariSense, ViewSpan, WiperLock, Wireless DNA, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

 $\ensuremath{\mathsf{SQTP}}$  is a service mark of Microchip Technology Incorporated in the U.S.A.

Silicon Storage Technology is a registered trademark of Microchip Technology Inc. in other countries.

GestIC is a registered trademarks of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2012-2014, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

ISBN: 978-1-63276-403-4

# QUALITY MANAGEMENT SYSTEM CERTIFIED BY DNV = ISO/TS 16949=

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



#### **Worldwide Sales and Service**

#### **AMERICAS**

Corporate Office 2355 West Chandler Blvd. Chandler, AZ 85224-6199

Chandler, AZ 85224-6199 Tel: 480-792-7200 Fax: 480-792-7277 Technical Support:

http://www.microchip.com/

support
Web Address:

www.microchip.com

Atlanta

Duluth, GA Tel: 678-957-9614 Fax: 678-957-1455

**Austin, TX** Tel: 512-257-3370

**Boston** 

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL

Tel: 630-285-0071 Fax: 630-285-0075

Cleveland

Independence, OH Tel: 216-447-0464 Fax: 216-447-0643

**Dallas** 

Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit Novi. MI

Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Indianapolis Noblesville, IN Tel: 317-773-8323

Fax: 317-773-5453

Los Angeles Mission Viejo, CA

Tel: 949-462-9523 Fax: 949-462-9608

New York, NY Tel: 631-435-6000

**San Jose, CA** Tel: 408-735-9110

**Canada - Toronto** Tel: 905-673-0699 Fax: 905-673-6509

#### ASIA/PACIFIC

**Asia Pacific Office** 

Suites 3707-14, 37th Floor Tower 6, The Gateway Harbour City, Kowloon Hong Kong

Tel: 852-2943-5100 Fax: 852-2401-3431

**Australia - Sydney** Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing

Tel: 86-10-8569-7000 Fax: 86-10-8528-2104

China - Chengdu Tel: 86-28-8665-5511

Fax: 86-28-8665-7889 China - Chongging

Tel: 86-23-8980-9588 Fax: 86-23-8980-9500

**China - Hangzhou** Tel: 86-571-8792-8115 Fax: 86-571-8792-8116

China - Hong Kong SAR

Tel: 852-2943-5100 Fax: 852-2401-3431

**China - Nanjing** Tel: 86-25-8473-2460 Fax: 86-25-8473-2470

**China - Qingdao** Tel: 86-532-8502-7355

Fax: 86-532-8502-7205 China - Shanghai

Tel: 86-21-5407-5533 Fax: 86-21-5407-5066

**China - Shenyang** Tel: 86-24-2334-2829 Fax: 86-24-2334-2393

**China - Shenzhen** Tel: 86-755-8864-2200 Fax: 86-755-8203-1760

**China - Wuhan** Tel: 86-27-5980-5300 Fax: 86-27-5980-5118

**China - Xian** Tel: 86-29-8833-7252 Fax: 86-29-8833-7256

**China - Xiamen** Tel: 86-592-2388138 Fax: 86-592-2388130

**China - Zhuhai** Tel: 86-756-3210040 Fax: 86-756-3210049

#### ASIA/PACIFIC

**India - Bangalore** Tel: 91-80-3090-4444

Fax: 91-80-3090-4123

India - New Delhi Tel: 91-11-4160-8631 Fax: 91-11-4160-8632

India - Pune

Tel: 91-20-3019-1500

**Japan - Osaka** Tel: 81-6-6152-7160 Fax: 81-6-6152-9310

**Japan - Tokyo** Tel: 81-3-6880- 3770 Fax: 81-3-6880-3771

**Korea - Daegu** Tel: 82-53-744-4301 Fax: 82-53-744-4302

**Korea - Seoul** Tel: 82-2-554-7200 Fax: 82-2-558-5932 or 82-2-558-5934

Malaysia - Kuala Lumpur Tel: 60-3-6201-9857

Fax: 60-3-6201-9859

Malaysia - Penang
Tel: 60-4-227-8870

Fax: 60-4-227-4068

Philippines - Manila

Tel: 63-2-634-9065 Fax: 63-2-634-9069

**Singapore** Tel: 65-6334-8870 Fax: 65-6334-8850

**Taiwan - Hsin Chu** Tel: 886-3-5778-366 Fax: 886-3-5770-955

**Taiwan - Kaohsiung** Tel: 886-7-213-7830

**Taiwan - Taipei** Tel: 886-2-2508-8600 Fax: 886-2-2508-0102 **Thailand - Bangkok** 

Tel: 66-2-694-1351 Fax: 66-2-694-1350

#### **EUROPE**

Austria - Wels

Tel: 43-7242-2244-39 Fax: 43-7242-2244-393

**Denmark - Copenhagen** Tel: 45-4450-2828

Fax: 45-4485-2829

France - Paris

Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

**Germany - Dusseldorf** Tel: 49-2129-3766400

**Germany - Munich** Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

**Germany - Pforzheim** Tel: 49-7231-424750

Italy - Milan

Tel: 39-0331-742611 Fax: 39-0331-466781

**Italy - Venice** Tel: 39-049-7625286

**Netherlands - Drunen** Tel: 31-416-690399 Fax: 31-416-690340

**Poland - Warsaw** Tel: 48-22-3325737

**Spain - Madrid** Tel: 34-91-708-08-90

Fax: 34-91-708-08-91 **Sweden - Stockholm** Tel: 46-8-5090-4654

**UK - Wokingham** Tel: 44-118-921-5800 Fax: 44-118-921-5820

03/25/14