LT1072, LT1072HV 1.25-A HIGH-EFFICIENCY SWITCHING REGULATORS

SLVS071A - NOVEMBER 1989 - REVISED AUGUST 1995

- Wide Supply-Voltage Range: LT1072HV . . . 3 V to 60 V LT1072 . . . 3 V to 40 V
- Low Quiescent Current . . . 6 mA Typ
- Internal 1.25-A Switch
- Few External Parts Required
- Self-Protected Against Overloads
- Operates in Most Switching Configurations
- Low Shutdown-Mode Supply Current
- Floating Outputs in Flyback-Regulated Mode
- Can Be Externally Synchronized

AVAILABLE OPTIONS

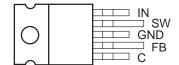
TJ	MAX INPUT VOLTAGE	KC PACKAGE	KV PACKAGE	P PACKAGE
0°C to	60 V	LT1072HVCKC	LT1072HVCKV	LT1072HVCP
100°C	40 V	LT1072CKC	LT1072CKV	LT1072CP
-40°C to	60 V	LT1072HVIKC	LT1072HVIKV	LT1072HVIP
125°C	40 V	LT1072IKC	LT1072IKV	LT1072IP

description

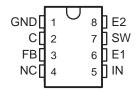
The LT1072 is a monolithic, high-efficiency switching configurations including: step-down (buck), step-up

regulator. It can be operated in all standard switching

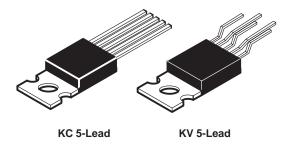
KC AND KV PACKAGE (KV Package Used for Illustration) (TOP VIEW)



P Package (TOP VIEW)



NC = No internal connection



(boost), flyback, forward, inverting, and Cuk[†]. A high-current, high-efficiency switch is included in the package along with all oscillator, control, and protection circuitry. Integration of all functions allows the LT1072 to be built in standard 5-terminal KC or a KV packages and the 8-terminal P package. This makes it extremely easy to use and provides reliable operation similar to that obtained with 3-terminal linear regulators.

The LT1072 operates with supply voltages from 3 V to 40 V. The LT1072HV, a high-voltage version of the LT1072, operates with supply voltages from 3 V to 60 V. These devices draw only 6 mA of quiescent current, deliver load power up to 20 W with no external power devices, and by utilizing current-mode switching techniques, provide excellent ac and dc input and output regulation.

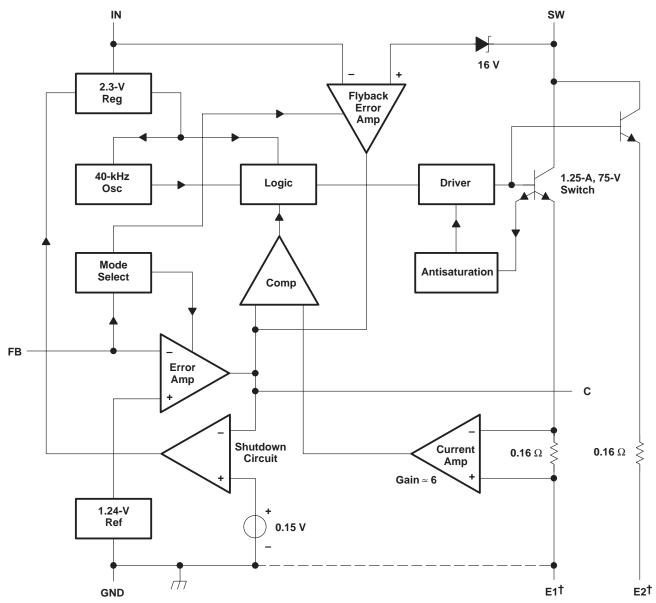
The LT1072 is much easier to use than the low-power control chips that are presently available and has many unique features that are not found on these chips. It uses an adaptive saturation-preventing switch drive to allow very-wide-ranging load currents with no loss in efficiency. An externally activated shutdown mode reduces total supply current to 50 μA typical for standby operation. Totally isolated and regulated outputs can be generated by using the optional flyback-regulation mode built into the LT1072 without using optocouplers or extra transformer windings.

[†]A boost-buck-derived regulator circuit patented by Slobodan Cuk.



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Functional Block Diagram



All resistor values shown are nominal.

[†] Always connect E1 to ground when using the P package. The emitters (E1 and E2) are tied internally to ground on the KC and KV packages.

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absolute maximum ratings over operating virtual junction temperature range (unless otherwise noted)†

Supply voltage, V _{I(IN)} (see Note 1):	LT1072			. 40 V
(,				
Switch output voltage: LT1072				. 65 V
LT1072HV				. 75 V
Feedback input voltage, V _(FB) (transi	ent, 1 ms) .			±15 V
Continuous total dissipation			See Dissipation Rating Tables	1 and 2
Operating virtual-junction temperatur	e range, T _J :	LT1072C, LT1072I	HVC0°C t	o 125°C
		LT1072I, LT1072H	IVI −40°C t	o 125°C
Storage temperature range, T _{stq}			−65°C t	o 150°C
Lead temperature 1,6 mm (1/16 inch) from case fo	or 10 seconds		260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE 1 – FREE-AIR TEMPERATURE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 125°C POWER RATING
KC	2000 mW	16 mW/°C	400 mW
KV	2000 mW	16 mW/°C	400 mW
Р	1000 mW	8 mW/°C	200 mW

DISSIPATION RATING TABLE 2 - CASE TEMPERATURE

PACKAGE	$T_C \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _C = 70°C	T _C = 125°C POWER RATING
KC	20 W	250 mW/°C	6.25 W
KV	20 W	250 mW/°C	6.25 W

recommended operating conditions

		MIN	MAX	UNIT
Innut Voltage Vivi	LT1072C, LT1072I	3	40	.,
Input Voltage, VI(IN)	LT1072HVC, LT1072HVI	3	60	V
Virtual impation to report up. T.	LT1072C, LT1072HVC	0	100	°C
Virtual-junction temperature, TJ	LT1072I, LT1072HVI	-40	125	

NOTE 1: Minimum switch-on time for the LT1072 in current limit is \approx 0.7 μ s. This limits the maximum input voltage during short-circuit conditions, in the step-down and inverting modes only, to \approx 40 V. Normal (unshorted) conditions are not affected. If the LT1072 is being operated in the step-down or inverting mode at high input voltages and short-circuit conditions are expected, a resistor must be placed in series with the inductor.

LT1072, LT1072HV 1.25-A HIGH-EFFICIENCY SWITCHING REGULATORS

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electrical characteristics at specified virtual junction temperature, V_{IN} = 15 V, V_{FB} = V_{ref} with SW output open (unless otherwise noted)

reference section

	PARAMETER	TEST CONDIT	гюмѕ†	TJ [‡]	MIN	TYP§	MAX	UNIT
V .	Output voltage	Magazirod at EB input	V(0) - 0.11	25°C	1.224	1.244	1.264	V
v ref	V _{ref} Output voltage Mea	Measured at FB input,	V(C) = 0.6 V	Full range	1.214		1.274	V
	Input regulation	$V_{(IN)} = 3 V \text{ to MAX},$	$V_{(C)} = 0.6 V$	Full range			0.03	%/V

error amplifier section

	PARAMETER	TEST CONDI	TIONST	TJ‡	MIN	TYP§	MAX	UNIT	
L/==>	Foodback input current	V(=>) - V		25°C		350	750	n A	
I(FB)	Feedback input current	V(FB) = V _{ref}		Full range			1100	nA	
	Transconductance	Al(a) - +25 !! A		25°C	3000	4400	6000	μmho	
	Transconductance	$\Delta I(C) = \pm 25 \mu A$		Full range	2400		7000	μππο	
	Source current	V(a) = 4.5.V	V/ 0.9.V/	25°C	150	200	350		
	Source current	$V_{(C)} = 1.5 \text{ V},$	$V_{(FB)} = 0.8 V$	Full range	120		400	μΑ	
	Sink current	V 45V	V 45V	\\\\1.E.\\	25°C	150	200	350	
	Sink current	$V_{(C)} = 1.5 \text{ V},$	$V_{(FB)} = 1.5 V$	Full range	120		400	μΑ	
V- (-)	Output valtage	High state,	V _(FB) = 1 V	25°C	1.8		2.3	V	
VO(C)	Output voltage	Low state,	V _(FB) = 1.5 V	25°C	0.25	0.38	0.52	V	
Ay	Voltage amplication	$V_{(C)} = 0.7 \text{ V to } 1.4 \text{ V}$		Full range	500	800	2000	V/V	
V/==>	Control throughold voltage	Potential 0		25°C	0.8	0.9	1.08	V	
V(TO)(C)	Control threshold voltage	Duty cycle = 0		Full range	0.6		1.25	V	

flyback amplifier section

	PARAMETER	TES	ST CONDITIONS [†]		T _J ‡	MIN	TYP§	MAX	UNIT
V _{T(FB)}	Flyback threshold voltage	I _(FB) = 50 μA			25°C	0.4	0.45	0.54	V
	Flyback threshold	L(ED) = 50 !! A	B) = 50 μ A, $I_{(C)}$ = -1 to + 1 μ A, $V_{(C)}$		25°C	15	16.3	17.6	V
	voltage	$I_{(FB)} = 50 \mu A,$	$I(C) = -1 \text{ to } + 1\mu A$	v(C) = 0.0 v	Full range	14		18	V
	Change in flyback reference	$I_{(FB)} = 0.05 \text{ to 1 mA},$	$I(C) = -1 \text{ to } +1 \mu\text{A},$	V _(C) = 0.6 V	25°C	4.5	6.8	8.5	V
	Flyback reference input regulation	$I_{(FB)} = 50 \mu A,$ $I_{(C)} = -1 \text{ to } +1 \mu A,$	$V_{(IN)} = 3 \text{ V to MAX},$ $V_{(C)} = 0.6 \text{ V}$		25°C		0.01	0.03	%/V
	Transconductance	$I_{(FB)} = 50 \mu A$	$\Delta I_{(C)} \le \pm 10 \mu\text{A}$		25°C	150	300	500	μmho
	Sink or source current	$V_{(C)} = 1.5 V,$	I _(FB) = 50 μA,	Source	Full range	15	32	50	
	Silik of Source current	$V(SW) = VZ + V(IN) \pm$	1 V	Sink	Full range	25	40	70	μΑ

[†] For conditions shown as MIN or MAX, use the appropriate value specified under the recommended operating conditions.



[‡] Full range virtual junction temperature is 0°C to 100°C for LT1072C and LT1072HVC and -40°C to 125°C for LT1072I and LT1072HVI.

[§] All typical values are $T_A = 25$ °C.

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electrical characteristics at specified virtual junction temperature, V_{IN} = 15 V, V_{FB} = V_{ref} with SW output open (unless otherwise noted)

output section

	PARAMETER	Т	EST CONDITIONS†		TJ‡	MIN	TYP§	MAX	UNIT
\/ (2.2.) (2.1.)	Curitale has alcdown valtage	$V_{(FB)} = 1.5 V,$	$V_{(IN)} = 3 V \text{ to MAX},$	LT1072	Full rongs	65			V
V(BR)(SW)	Switch breakdown voltage	$I_{(SW)} = 5 \text{ mA}$ LT1072H		LT1072HV	Full range	75			l
ron	Switch on-state resistance	$V_{(FB)} = 0.8 V,$	I(SW) = 1.25 mA		Full range		0.6	1	Ω
	Control-to-switch transconductance				25°C		2		mho
	.,		Duty cycle ≤ 50%		≥25°C	1.25		3	
I(SW)(lim)	Switch current limit	V _(FB) = 0.8 V, See Note 2	Duty cycle ≤ 50%	Duty cycle ≤ 50%		1.25		3.5	Α
		Duty cycle = 80%			Full range	1		2.5	
ΔI(IN)/ΔI(SW)	Input current increase during switch turn-on	V _(FB) = 0.8 V			25°C		25	35	mA/A
4	Fraguesay				25°C	35	40	45	lel I=
T	Frequency				Full range	33		47	kHz
	Maximum duty cycle	V _(FB) = 1 V			25°C	90%	92%	97%	
t _d	Flyback sense delay time				25°C		1.5		μs

shutdown section

	PARAMETER	TEST CONDITIONS [†]	TJ‡	MIN	TYP§	MAX	UNIT
I _{off(IN)}	Shutdown mode input current	$V_{(IN)} = 3 \text{ V to MAX}, V_{(C)} = 0.05 \text{ V}$	25°C		100	250	μΑ
\/(=0\(0)	Control throubold voltage	Vann – 2 V to MAY	25°C	100	150	250	mV
V(TO)(C)	Control threshold voltage	V(IN) = 3 V to MAX	Full range	50		300	IIIV

total device

	PARAMETER	TEST COND	OITIONS [†]	TJ [‡]	MIN	TYP§	MAX	UNIT
V _{I(min)(IN)}	Minimum input voltage			Full range		2.6	3	V
I _I (IN)	Input current	$V_{(IN)} = 3 V \text{ to MAX},$	V(C) = 0.6 V	25°C		6	9	mA

[†] For conditions shown as MIN or MAX, use the appropriate value specified under the recommended operating conditions.

NOTE 2: For duty cycles between 50% and 80%, minimum switch output current is given by I_{(SW)(lim)} = 0.833 (2-duty cycle).



[‡] Full range virtual junction temperature is 0°C to 100°C for LT1072C and LT1072HVC and -40°C to 125°C for LT1072I and LT1072HVI. § All typical values are $T_A = 25$ °C.

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theory of operation

The LT1072 is a current-mode switcher. This means that the switch duty cycle is directly controlled by switch current rather than by output voltage. Referring to the functional block diagram, the switch is turned on at the start of each oscillator cycle. It is turned off when the switch current reaches a predetermined level. Control of output voltage is obtained by using the output of a voltage-sensing error amplifier to set the current trip level. This technique has several advantages. First, it has immediate response to input-voltage variations, which is unlike ordinary switchers that have poor input transient response. Second, it reduces the 90° phase shift at midfrequencies in the energy-storage inductor. This greatly simplifies closed-loop frequency compensation under widely varying input-voltage or output-load conditions. Finally, it allows simple pulse-by-pulse current limiting to provide maximum switch protection under output overload or short conditions. A low-dropout internal regulator provides a 2.3-V supply for all internal circuitry on the LT1072. This low-dropout design allows input voltage to vary from 3 V to 60 V with virtually no change in device performance. A 40-kHz oscillator is the basic clock for all internal timing. It turns on the output switch via the logic and driver circuitry. Special adaptive antisaturation circuitry detects the onset of saturation in the power switch and adjusts driver current instantaneously to limit switch saturation. This minimizes driver dissipation and provides very rapid turn off of the switch.

A 1.2-V band-gap reference biases the positive input of the error amplifier. The negative input is brought out for output-voltage sensing. This feedback terminal has a second function when pulled low with an external resistor. It programs the LT1072 to disconnect the main error-amplifier output and connects the output of the flyback amplifier to the comparator input. The LT1072 will then regulate the value of the flyback pulse with respect to the supply voltage. This flyback pulse is directly proportional to output voltage in the traditional transformer-coupled flyback-topology regulator. By regulating the amplitude of the flyback pulse, the output voltage can be regulated with no direct connection between input and output. The output is fully floating up to the breakdown voltage of the transformer windings. Multiple floating outputs are easily obtained with additional windings. A special delay network inside the LT1072 ignores the leakage inductance spike at the leading edge of the flyback pulse to improve output regulation.

The error signal developed at the comparator input is brought out externally. This terminal (C) has four different functions. It is used for frequency compensation, current limit adjustment, soft starting, and total regulator shutdown. During normal regulator operation, this terminal sits at a voltage between 0.9 V (low output current) and 2 V (high output current). The error amplifiers are current-output (g_m) types, so this voltage can be externally clamped for adjusting current limit. Likewise, a capacitor-coupled external clamp will provide soft start. Switch duty cycle goes to zero if the C terminal is pulled to ground through a diode. This places the LT1072 in an idle mode. Pulling the C terminal below 0.15 V causes total regulator shutdown, with only 50- μ A supply current for shutdown-circuitry biasing.

In the P package, the emitters of the power transistors are brought out separately from the ground terminal. This eliminates errors due to ground-terminal voltage drops and allows the user to reduce the switch-current limit (2:1) by leaving the second emitter (E2) disconnected. The first emitter (E1) should always be connected to the ground terminal. Note that switch on-state resistance doubles when E2 is left open, so efficiency will suffer somewhat when switch currents exceed 100 mA. Also, note that chip dissipation will actually *increase* with E2 open during normal load operations, even though dissipation in current-limit mode will *decrease*.



Table of Graphs

				FIGU	RE
Ром	Maximum output power	vs	Input voltage	1	
V _{ref}	Reference voltage	vs	Junction temperature	2	<u>?</u>
f	Switching frequency	vs	Junction temperature	2	2
	Reference voltage change	vs	Input voltage	3	3
I _{FB}	Feedback input current	vs	Junction temperature	4	-
9 _m	Error amplifier transconductance	vs	Junction temperature	5	;
9m	Error amplifier transconductance	vs	Frequency	6	,
	Error amplifier phase shift	vs	Frequency	6	;
IC	Control current	vs	Control voltage	7	,
V _{T(FB)}	Normal/flyback mode threshold voltage	vs	Junction temperature	8	}
I _{FB}	Feedback input current	vs	Junction temperature	8	}
V _Z	Flyback reference voltage	vs	Junction temperature	9	,
t _d	Flyback sense delay time	vs	Junction temperature	10	,
IO(SW)	Switch output current (with switch off)	vs	Switch voltage	11	
	Driver base current	vs	Switch output current	12	<u>, </u>
V _{sat(SW)}	Switch saturation voltage	vs	Switch output current	13	3
I _{O(SW)}	Switch output current limit	vs	Duty cycle	14	-
	Maximum duty cycle	vs	Junction temperature	15	;
I _{IN}	Shutdown-mode input current	vs	Control voltage	16	;
I _{IN}	Shutdown-mode input current	vs	Input voltage	17	,
V _{T(C)}	Shutdown-mode control threshold voltage	vs	Junction temperature	18	3
IT(C)	Shutdown-mode control threshold current	vs	Junction temperature	18	3
V _{FB}	Feedback input voltage at normal/flyback mode threshold	vs	Feedback input current	19	,
	Minimum input voltage	vs	Junction temperature	20)
I _{IN}	Input current (SW output open)	vs	Junction temperature	21	
IN	Input current	vs	Input voltage	22	

Table of Application Circuits

APPLICATION	FIGURE
Totally isolated converter	23
Boost converter (5 V to 12 V)	24

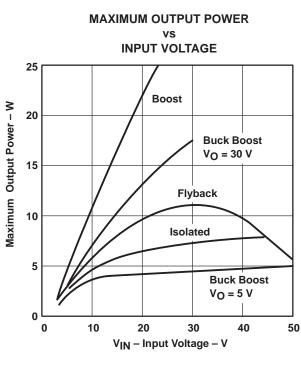


Figure 1

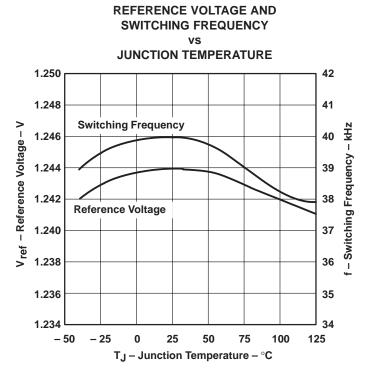
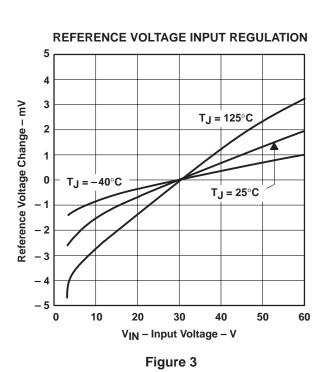
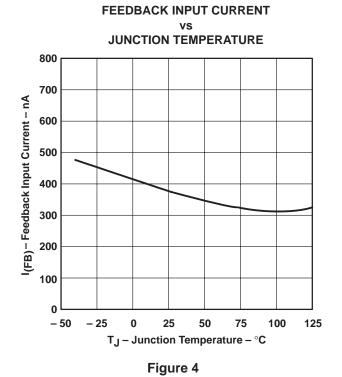
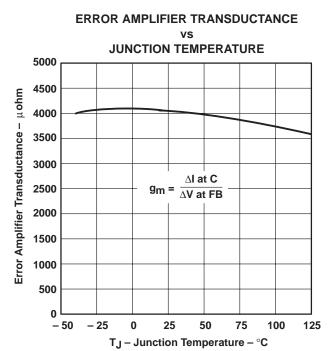


Figure 2









ERROR AMPLIFIER TRANSCONDUCTANCE AND PHASE SHIFT

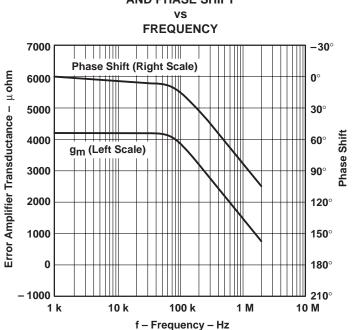
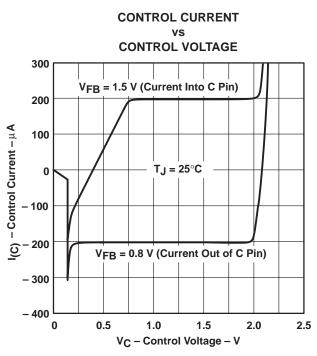


Figure 5





NORMAL-FLYBACK-MODE THRESHOLD VOLTAGE AND FEEDBACK INPUT CURRENT

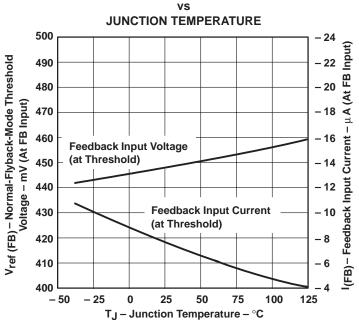


Figure 7 Figure 8

FLYBACK MODE REFERENCE VOLTAGE **JUNCTION TEMPERATURE** 23 22 Flyback Mode Reference Voltage - V R_{Feedback} = 500 Ω 21 20 19 $R_{Feedback} = 1 k\Omega$ 18 17 R_{Feedback} = 10 k Ω 16 15 - 50 - 25 0 25 50 75 100 125 T_J - Junction Temperature - °C

Figure 9

SWITCH OUTPUT CURRENT

SWITCH VOLTAGE AND INPUT VOLTAGE (WITH SWITCH OFF) 1000 900 O(SW) - Switch Output Current - μA 800 V_{IN} = 55 V 700 V_{IN} = 40 V $V_{IN} = 15 V$ 600 500 $V_{IN} = 3 V$ 400 300 200 100 0 10 20 30 40 50 60 70 80 90

Switch Voltage - V

Figure 11

FLYBACK SENSE DELAY TIME
vs
JUNCTION TEMPERATURE

2.2

string and string and

Figure 10

25

T_J - Junction Temperature - °C

50

75

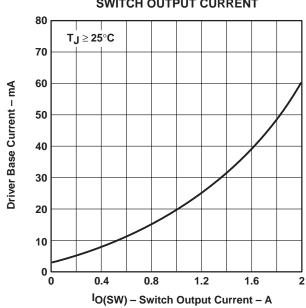
100

125

- 50

- 25





[†] Average input current is found by multiplying driver base by duty cycle plus quiescent current.

Figure 12



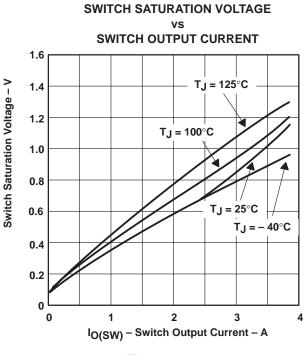


Figure 13

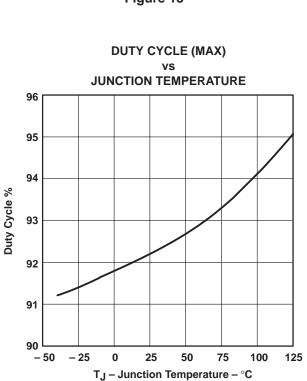


Figure 15

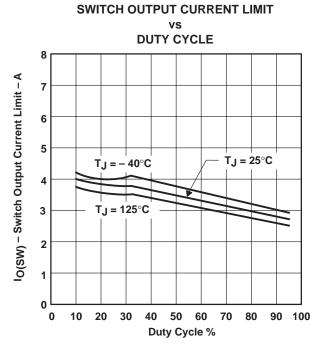


Figure 14

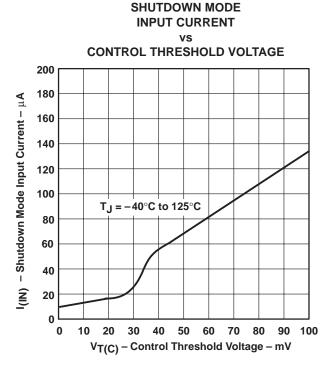
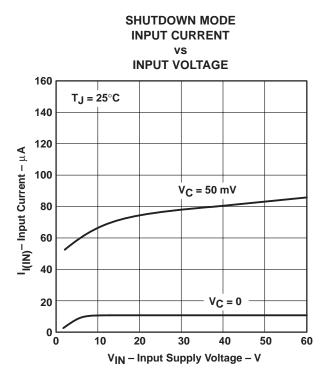


Figure 16



SHUTDOWN MODE
CONTROL THRESHOLD VOLTAGE AND CURRENT
vs

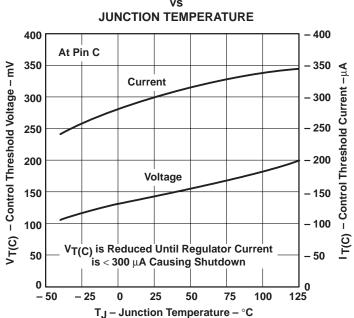
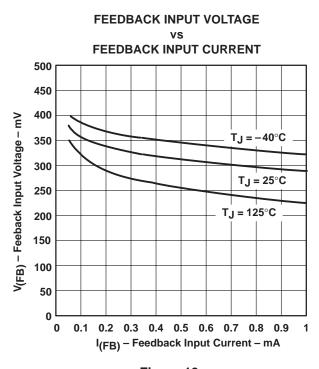


Figure 17

Figure 18



MINIMUM INPUT VOLTAGE vs JUNCTION TEMPERATURE 2.9 Switch Current = 1.25 V V_I(min)(IN) - Minimum Input Voltage - V 2.8 2.7 Switch Current = 0 2.6 2.5 - 50 - 25 50 75 100 125 T_J - Junction Temperature - °C

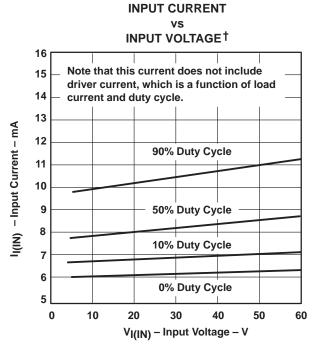
Figure 19

Figure 20

INPUT CURRENT ٧S **JUNCTION TEMPERATURE** (SW OUTPUT OPEN) 11 $V_{T(C)} = 0.6 V$ 10 9 II(IN) - Input Current - mA 8 $V_{IN} = 60 V$ 7 6 $V_{IN} = 3 V$ 5 4 3 2 - 50 - 25 0 25 50 75 100 125

Figure 21

T_J – Junction Temperature – °C

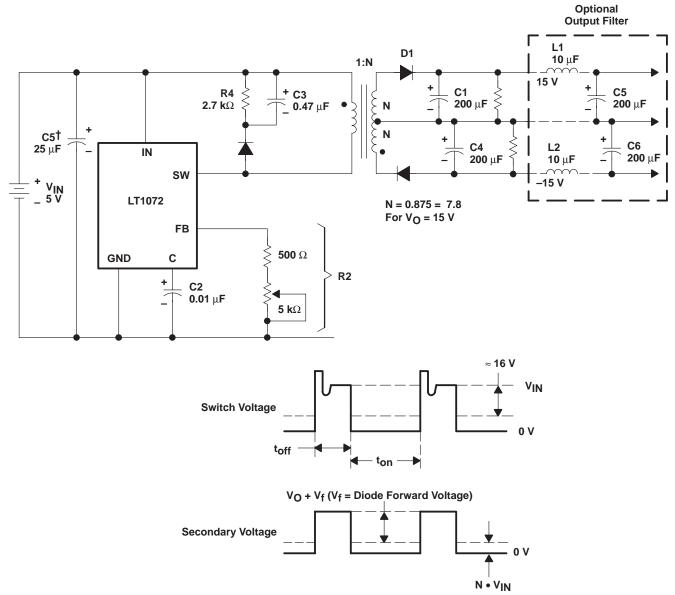


[†] Under very low output current conditions, duty cycle for most circuits will approach 10% or less.

Figure 22



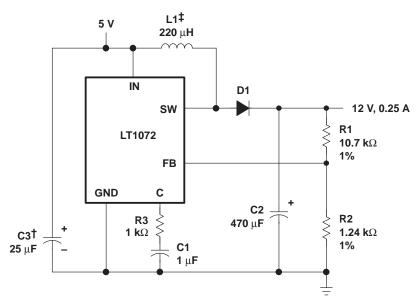
APPLICATION INFORMATION



[†] Capacitors are required if input leads \geq 2 inches.

Figure 23. Totally Isolated Converter

APPLICATION INFORMATION



 $^{\ ^{\}mbox{\dag}}$ Capacitor is required if input leads ≥ 2 inches. $\mbox{\dag}$ Pulse Engineering 52626

Figure 24. Boost Converter (5 V to 12 V)



PACKAGE OPTION ADDENDUM

30-Mar-2005

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
LT1072CP	OBSOLETE	PDIP	Р	8	TBD	Call TI	Call TI
LT1072IP	OBSOLETE	PDIP	Р	8	TBD	Call TI	Call TI

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

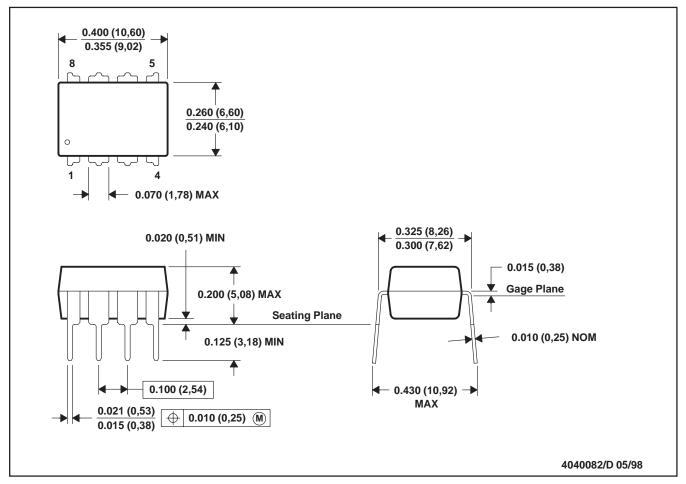
(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001

For the latest package information, go to http://www.ti.com/sc/docs/package/pkg_info.htm

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