

TC74HC423AP, TC74HC423AF

Dual Retriggerable Monostable Multivibrator

The TC74HC423A is a high speed CMOS MONOSTABLE MULTIVIBRATOR fabricated with silicon gate C²MOS technology.

It achieves the high speed operation similar to equivalent LSTTL while maintaining the CMOS low power dissipation.

There are two trigger inputs, \overline{A} input (negative edge), and B input (positive edge). These inputs are valid for a slow rise/fall time signal ($t_r = t_f = 1\text{ s}$) as they are schmitt trigger inputs.

After triggering, the output stays in a MONOSTABLE state for a time period determined by the external resistor and capacitor (R_x, C_x). A low level at the \overline{CLR} input breaks this state. In the MONOSTABLE state, if a new trigger is applied, it extends the MONOSTABLE period (retrigger mode).

Limitations for C_x and R_x are:

External capacitor, C_x : No limit

External resistor, R_x : $V_{CC} = 2.0\text{ V}$ more than $5\text{ k}\Omega$

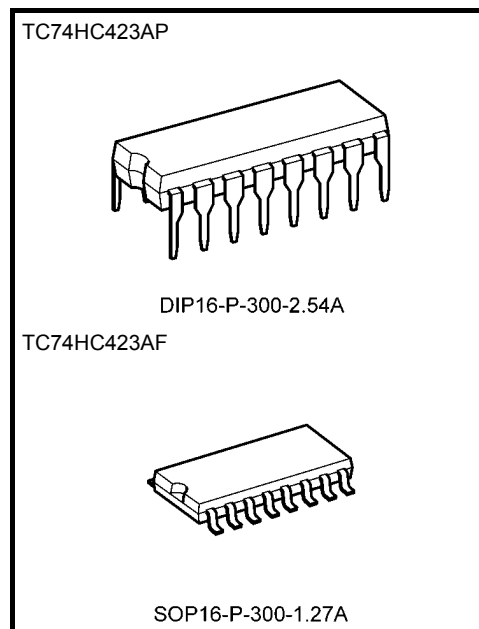
$V_{CC} \geq 3.0\text{ V}$ more than $1\text{ k}\Omega$

All inputs are equipped with protection circuits against static discharge or transient excess voltage.

Features (Note)

- High speed: $t_{pd} = 25\text{ ns}$ (typ.) at $V_{CC} = 5\text{ V}$
- Low power dissipation
 - Standby state: $I_{CC} = 4\text{ }\mu\text{A}$ (max) at $T_a = 25^\circ\text{C}$
 - Active state: $I_{CC} = 700\text{ }\mu\text{A}$ (max) at $V_{CC} = 5\text{ V}$
- High noise immunity: $V_{NIH} = V_{NIL} = 28\% V_{CC}$ (min)
- Output drive capability: 10 LSTTL loads
- Symmetrical output impedance: $|I_{OH}| = I_{OL} = 4\text{ mA}$ (min)
- Balanced propagation delays: $t_{pLH} \approx t_{pHL}$
- Wide operating voltage range: $V_{CC}(\text{opr}) = 2\sim 6\text{ V}$
- Pin and function compatible with 74LS423

Note: In the case of using only one circuit, \overline{CLR} should be tied to GND, $R_x/C_x \cdot C_x \cdot Q \cdot \overline{Q}$ should be tied to OPEN, the other inputs should be tied to V_{CC} or GND.

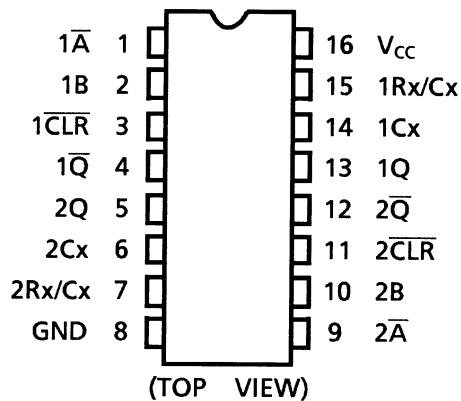


Weight

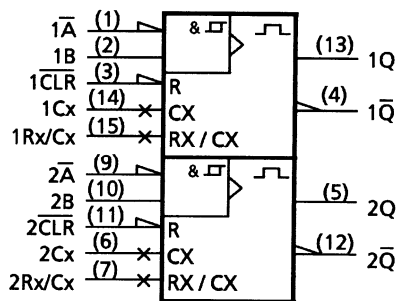
DIP16-P-300-2.54A : 1.00 g (typ.)

SOP16-P-300-1.27A : 0.18 g (typ.)

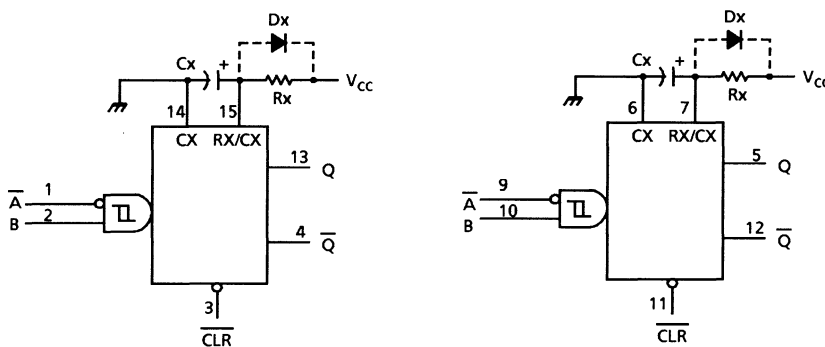
Pin Assignment



IEC Logic Symbol



Block Diagram (Note 1)(Note 2)



Note 1: Cx, Rx, Dx are external capacitor, resistor, and diode, respectively.

Note 2: External clamping diode, Dx;

The external capacitor is charged to V_{CC} level in the wait state, i.e. when no trigger is applied.

If the supply voltage is turned off, Cx is discharges mainly through the internal (parasitic) diode. If Cx is sufficiently large and V_{CC} drops rapidly, there will be some possibility of damaging the IC through inrush current or latch-up. If the capacitance of the supply voltage filter is large enough and V_{CC} drops slowly, the inrush current is automatically limited and damage to the IC is avoided.

The maximum value of forward current through the parasitic diode is ± 20 mA.







In the case of a large Cx, the limit of fall time of the supply voltage is determined as follows:

$$t_f \geq (V_{CC} - 0.7) Cx / 20 \text{ mA}$$

(t_f is the time between the supply voltage turn off and the supply voltage reaching $0.4 V_{CC}$.)

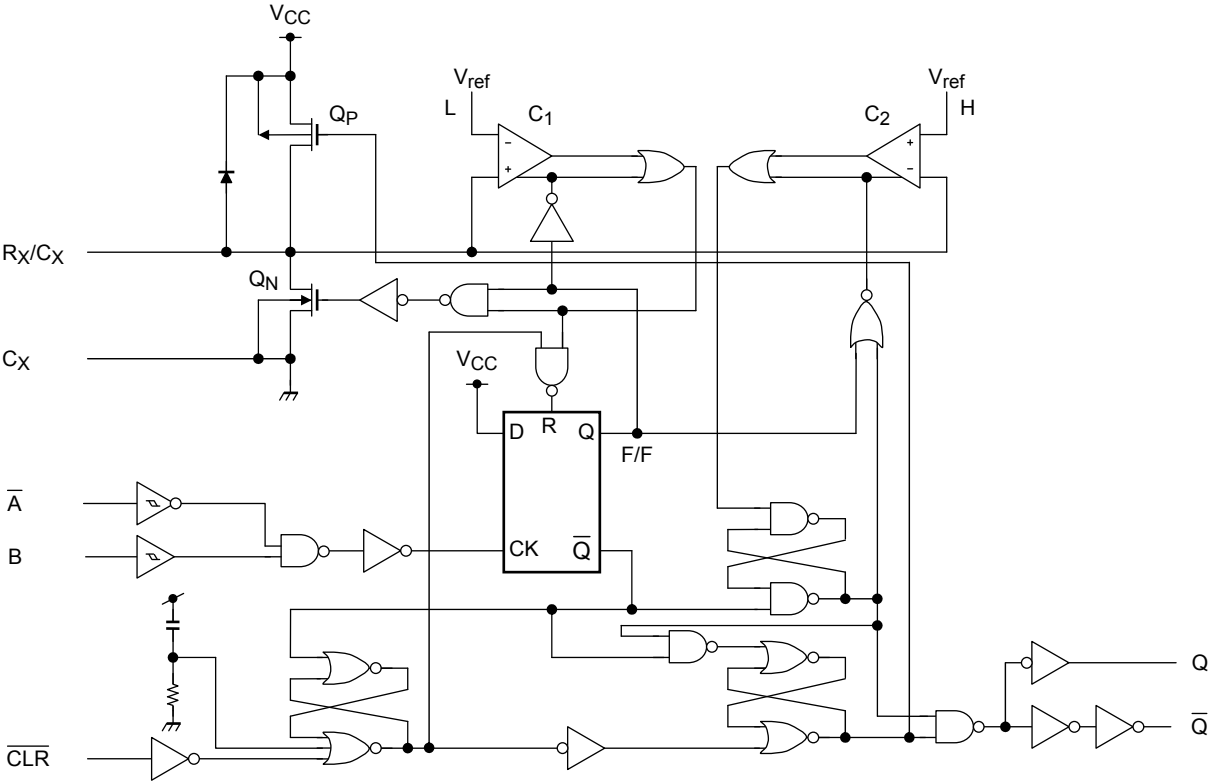
In the event a system does not satisfy the above condition, an external clamping diode (Dx) is needed to protect the IC from inrush current.

Truth Table

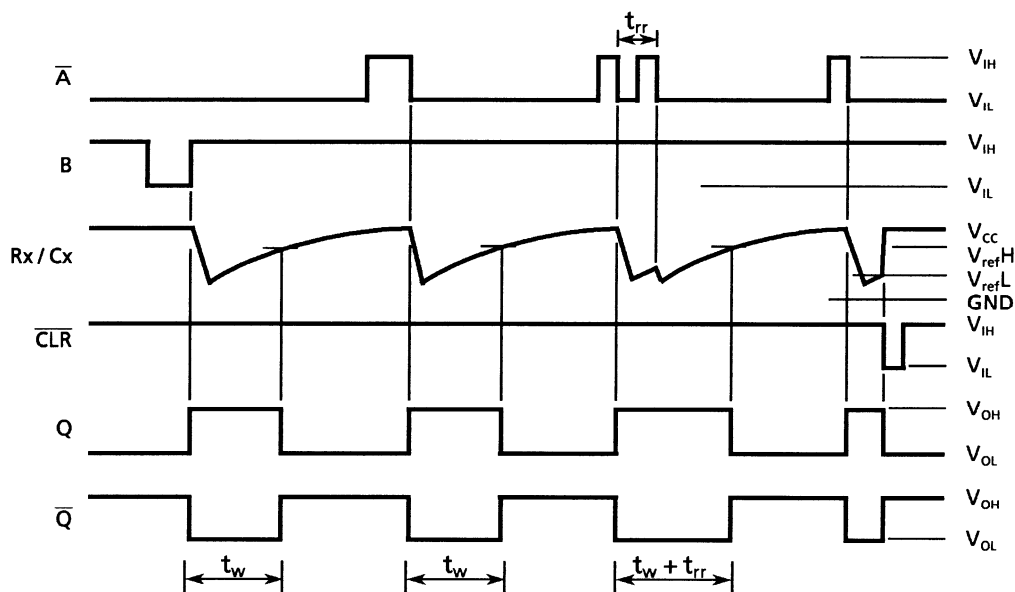
Inputs			Outputs		Note
\bar{A}	B	\overline{CLR}	Q	\bar{Q}	
	H	H			Output Enable
X	L	H	L	H	Inhibit
H	X	H	L	H	Inhibit
L		H			Output Enable
X	X	L	L	H	Reset

X: Don't care

System Diagram



Timing Chart



Functional Description

(1) Stand-by state

The external capacitor C_x is fully charged to V_{CC} in the stand-by state. That means, before triggering, the Q_P and Q_N transistors which are connected to the Rx/Cx node are in the off state. Two comparators that relate to the timing of the output pulse, and two reference voltage supplies turn off. The total supply current is only leakage current.

(2) Trigger operation

Trigger operation is effective in either of the following two cases. First, the condition where the \bar{A} input is low, and the B input has a rising signal; second, where the B input is high, and the \bar{A} input has a falling signal.

After a trigger becomes effective, comparators $C1$ and $C2$ start operating, and Q_N is turned on. The external capacitor discharges through Q_N . The voltage level of the Rx/Cx node drops. If the Rx/Cx voltage level falls to the internal reference voltage V_{refL} , the output of $C1$ becomes low. The flip-flop is then reset and Q_N turns off. At that moment $C1$ stops but $C2$ continues operating.

After Q_N turns off, the voltage at the Rx/Cx starts rising at a rate determined by the time constant of external capacitor C_x and resistor R_x .

Upon the triggering, output Q becomes high, following some delay time of the internal F/F and gates. It stays high even if the voltage of Rx/Cx changes from falling to rising. When Rx/Cx reaches the internal reference voltage V_{refH} , the output of $C2$ becomes low, the output Q goes low and $C2$ stops its operation. That means, after triggering, when the voltage level of the Rx/Cx reaches V_{refH} , the IC returns to its MONOSTABLE state.

With large values of C_x and R_x , and ignoring the discharge time of the capacitor and internal delays of the IC, the width of the output pulse, t_w (OUT), is as follows:

$$t_w (\text{OUT}) = 1.0 \cdot C_x \cdot R_x$$

(3) Retrigger operation

When a new trigger is applied to input \bar{A} or B while in the MONOSTABLE state, it is effective only if the IC is charging C_x . The voltage level of the Rx/Cx node then falls to V_{refL} level again. Therefore the Q output stays high if the next trigger comes in before the time period set by C_x and R_x .

If the 2nd trigger is very close to previous trigger, such as an occurrence during the discharge cycle, it will have no effect.

The minimum time for a trigger to be effective 2nd trigger, t_{rr} (min), depends on V_{CC} and C_x .

(4) Reset operation

In normal operation, \overline{CLR} input is held high. If \overline{CLR} is low, a trigger has no effect because the Q output is held low and the trigger control F/F is reset. Also, Q_P turns on and C_x is charged rapidly to V_{CC} .

This means if \overline{CLR} input is set low, the IC goes into a wait state.

Absolute Maximum Ratings (Note 1)

Characteristics	Symbol	Rating	Unit
Supply voltage range	V_{CC}	$-0.5 \sim 7$	V
DC input voltage	V_{IN}	$-0.5 \sim V_{CC} + 0.5$	V
DC output voltage	V_{OUT}	$-0.5 \sim V_{CC} + 0.5$	V
Input diode current	I_{IK}	± 20	mA
Output diode current	I_{OK}	± 20	mA
DC output current	I_{OUT}	± 25	mA
DC V_{CC} /ground current	I_{CC}	± 50	mA
Power dissipation	P_D	500 (DIP) (Note 2)/180 (SOP)	mW
Storage temperature	T_{stg}	$-65 \sim 150$	$^{\circ}\text{C}$

Note 1: Exceeding any of the absolute maximum ratings, even briefly, lead to deterioration in IC performance or even destruction.

Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings and the operating ranges.

Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc.).

Note 2: 500 mW in the range of $T_a = -40 \sim 65^{\circ}\text{C}$. From $T_a = 65$ to 85°C a derating factor of $-10 \text{ mW}/^{\circ}\text{C}$ shall be applied until 300 mW.

Operating Ranges (Note 1)

Characteristics	Symbol	Rating	Unit
Supply voltage	V_{CC}	2~6	V
Input voltage	V_{IN}	$0 \sim V_{CC}$	V
Output voltage	V_{OUT}	$0 \sim V_{CC}$	V
Operating temperature	T_{opr}	$-40 \sim 85$	$^{\circ}\text{C}$
Input rise and fall time ($\overline{\text{CLR}}$ only)	t_r, t_f	0~1000 ($V_{CC} = 2.0 \text{ V}$) 0~500 ($V_{CC} = 4.5 \text{ V}$) 0~400 ($V_{CC} = 6.0 \text{ V}$)	ns
External capacitor	C_x	No limitation (Note 2)	F
External resistor	R_x	$\geq 5 \text{ k}$ ($V_{CC} = 2.0 \text{ V}$) (Note 2) $\geq 1 \text{ k}$ ($V_{CC} \geq 3.0 \text{ V}$) (Note 2)	Ω

Note 1: The operating ranges must be maintained to ensure the normal operation of the device. Unused inputs must be tied to either V_{CC} or GND.

Note 2: The maximum allowable values of C_x and R_x are a function of leakage of capacitor C_x , the leakage of TC74HC423A, and leakage due to board layout and surface resistance.

Susceptibility to externally induced noise signals may occur for $R_x > 1 \text{ M}\Omega$.

Electrical Characteristics
DC Characteristics

Characteristics	Symbol	Test Condition		Ta = 25°C			Ta = -40~85°C		Unit
				V _{CC} (V)	Min	Typ.	Max	Min	Max
High-level input voltage	V _{IH}	—		2.0 4.5 6.0	1.50 3.15 4.20	— — —	— — —	1.50 3.15 4.20	V
Low-level input voltage	V _{IL}	—		2.0 4.5 6.0	— — —	— — —	0.50 1.35 1.80	— — —	V
High-level output voltage (Q, \bar{Q})	V _{OH}	V _{IN} = V _{IH} or V _{IL}	I _{OH} = -20 μ A	2.0 4.5 6.0	1.9 4.4 5.9	2.0 4.5 6.0	— — —	1.9 4.4 5.9	V
			I _{OH} = -4 mA	4.5	4.18	4.31	—	4.13	
			I _{OH} = -5.2 mA	6.0	5.68	5.80	—	5.63	
Low-level output voltage (Q, \bar{Q})	V _{OL}	V _{IN} = V _{IH} or V _{IL}	I _{OL} = 20 μ A	2.0 4.5 6.0	— — —	0.0 0.0 0.0	0.1 0.1 0.1	— — —	V
			I _{OL} = 4 mA	4.5	—	0.17	0.26	—	
			I _{OL} = 5.2 mA	6.0	—	0.18	0.26	—	
Input leakage current	I _{IN}	V _{IN} = V _{CC} or GND		6.0	—	—	±0.1	—	±1.0 μ A
Rx/Cx terminal off-state current	I _{IN}	V _{IN} = V _{CC} or GND		6.0	—	—	±0.1	—	±1.0 μ A
Quiescent supply current	I _{CC}	V _{IN} = V _{CC} or GND		6.0	—	—	4.0	—	40.0 μ A
Active-state supply current (Note)	I _{CC}	V _{IN} = V _{CC} or GND		2.0	—	45	200	—	260 μ A
		Rx/Cx = 0.5 V _{CC}		4.5	—	400	500	—	650 μ A
				6.0	—	0.7	1.0	—	1.3 mA

Note: Per circuit

Timing Requirements (input: $t_r = t_f = 6 \text{ ns}$)

Characteristics	Symbol	Test Condition	Ta = 25°C		Ta = -40 ~ 85°C		Unit
			V _{CC} (V)	Typ.	Limit	Limit	
Minimum pulse width	t_W (L) t_W (H)	—	2.0	—	75	95	ns
			4.5	—	15	19	
			6.0	—	13	16	
Minimum clear pulse width	t_W (L)	—	2.0	—	75	95	ns
			4.5	—	15	19	
			6.0	—	13	16	
Minimum clear removal time	t_{rem}	—	2.0	—	5	5	ns
			4.5	—	5	5	
			6.0	—	5	5	
Minimum retrigger time	t_{rr}	Rx = 1 k Ω Cx = 100 pF	2.0	325	—	—	ns
			4.5	108	—	—	
			6.0	78	—	—	
		Rx = 1 k Ω Cx = 0.01 μ F	2.0	5.0	—	—	μ s
			4.5	1.4	—	—	
			6.0	1.2	—	—	

AC Characteristics (C_L = 15 pF, V_{CC} = 5 V, Ta = 25°C, input: $t_r = t_f = 6 \text{ ns}$)

Characteristics	Symbol	Test Condition	Min	Typ.	Max	Unit
Output transition time	t_{TLH}	—	—	4	8	ns
	t_{THL}					
Propagation delay time (\bar{A} , B-Q, \bar{Q})	t_{PLH}	—	—	25	36	ns
	t_{PHL}					
Propagation delay time (\bar{CL} -Q, \bar{Q})	t_{PLH}	—	—	16	27	ns
	t_{PHL}					

AC Characteristics ($C_L = 50 \text{ pF}$, input: $t_r = t_f = 6 \text{ ns}$)

Characteristics	Symbol	Test Condition	V_{CC} (V)	$T_a = 25^\circ\text{C}$			$T_a = -40 \sim 85^\circ\text{C}$		Unit
				Min	Typ.	Max	Min	Max	
Output transition time	t_{TLH} t_{THL}	—	2.0	—	30	75	—	95	ns
			4.5	—	8	15	—	19	
			6.0	—	7	13	—	16	
Propagation delay time (\bar{A} , B-Q, \bar{Q})	t_{pLH} t_{pHL}	—	2.0	—	102	210	—	265	ns
			4.5	—	29	42	—	53	
			6.0	—	22	36	—	45	
Propagation delay time (\bar{CL} -Q, \bar{Q})	t_{pLH} t_{pHL}	—	2.0	—	68	160	—	200	ns
			4.5	—	20	32	—	40	
			6.0	—	16	27	—	34	
Output pulse width	tw_{OUT}	$C_x = 28 \text{ pF}$ $R_x = 6 \text{ k}\Omega$ ($V_{CC} = 2 \text{ V}$) $R_x = 2 \text{ k}\Omega$ ($V_{CC} = 4.5 \text{ V}, 6 \text{ V}$)	2.0	—	700	2000	—	2500	ns
			4.5	—	250	400	—	500	
			6.0	—	210	340	—	425	
		$C_x = 0.01 \text{ }\mu\text{F}$ $R_x = 10 \text{ k}\Omega$	2.0	90	110	130	90	130	μs
			4.5	95	105	115	95	115	
			6.0	95	105	115	95	115	
		$C_x = 0.1 \text{ }\mu\text{F}$ $R_x = 10 \text{ k}\Omega$	2.0	0.9	1.0	1.2	0.9	1.2	ms
			4.5	0.9	1.0	1.1	0.9	1.1	
			6.0	0.9	1.0	1.1	0.9	1.1	
Output pulse width error between circuits (in same package)	Δtw_{OUT}	—	—	—	± 1	—	—	—	%
Input capacitance	C_{IN}	—	—	—	5	10	—	10	pF
Power dissipation capacitance	C_{PD} (Note)	—	—	—	162	—	—	—	pF

Note: C_{PD} is defined as the value of the internal equivalent capacitance which is calculated from the operating current consumption without load.

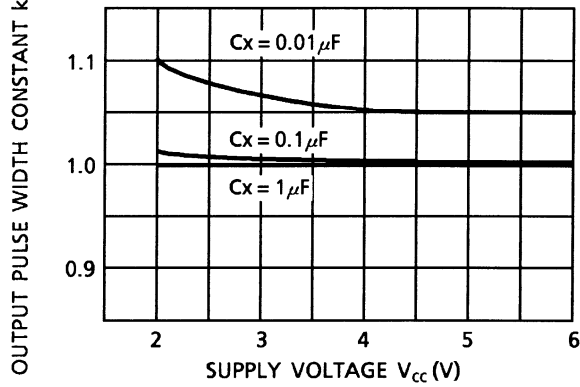
Average operating current can be obtained by the equation:

$$I_{CC}(\text{opr}) = C_{PD} \cdot V_{CC} \cdot f_{IN} + I_{CC}' \cdot \text{duty}/100 + I_{CC}/2 \text{ (per circuit)}$$

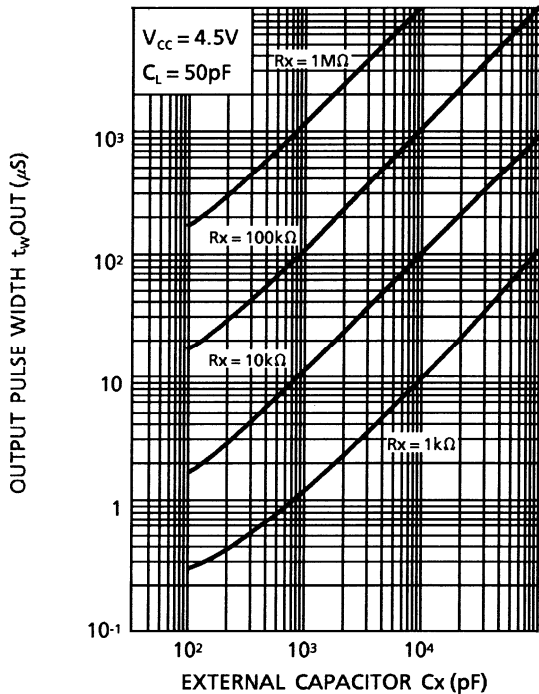
(I_{CC}' : active supply current)

(duty: %)

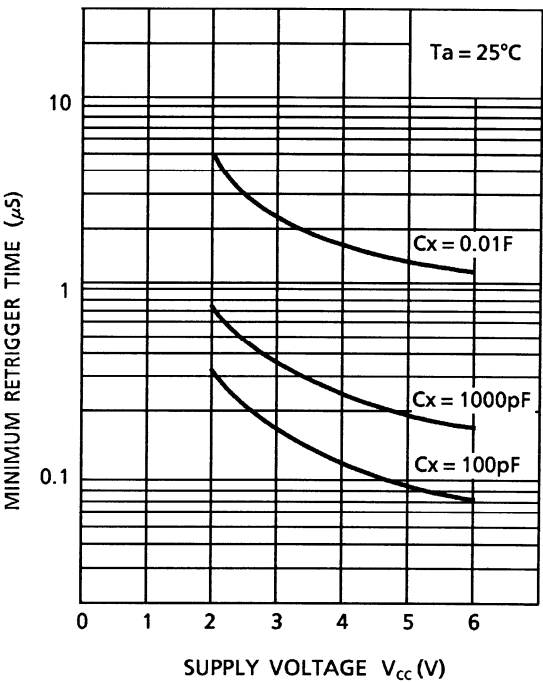
Output Pulse Width Constant K – Supply Voltage (typical)
(EXTERNAL RESISTOR (Rx) = 10kΩ : $t_{WOUT} = K \cdot Cx \cdot Rx$)



$t_{WOUT} - Cx$ Characteristics (typ.)



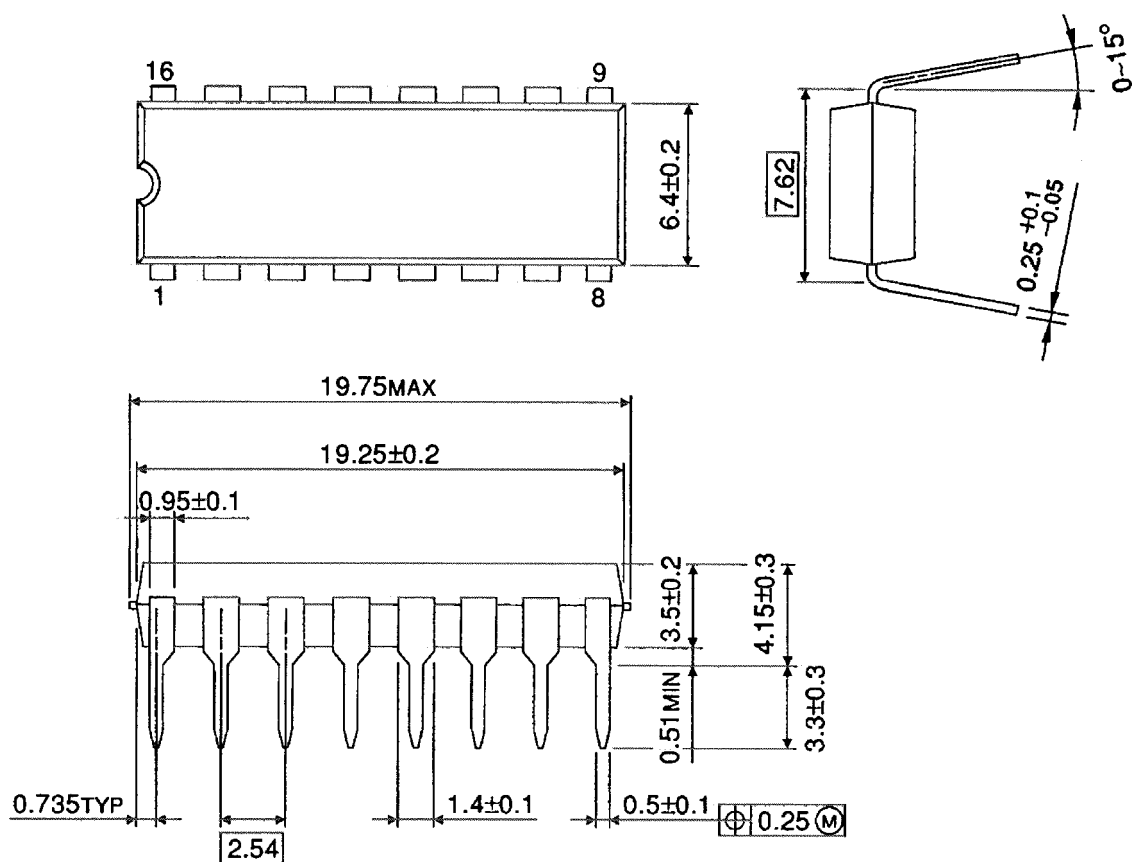
$t_{rr} - V_{CC}$ Characteristics (typ.)



Package Dimensions

DIP16-P-300-2.54A

Unit : mm

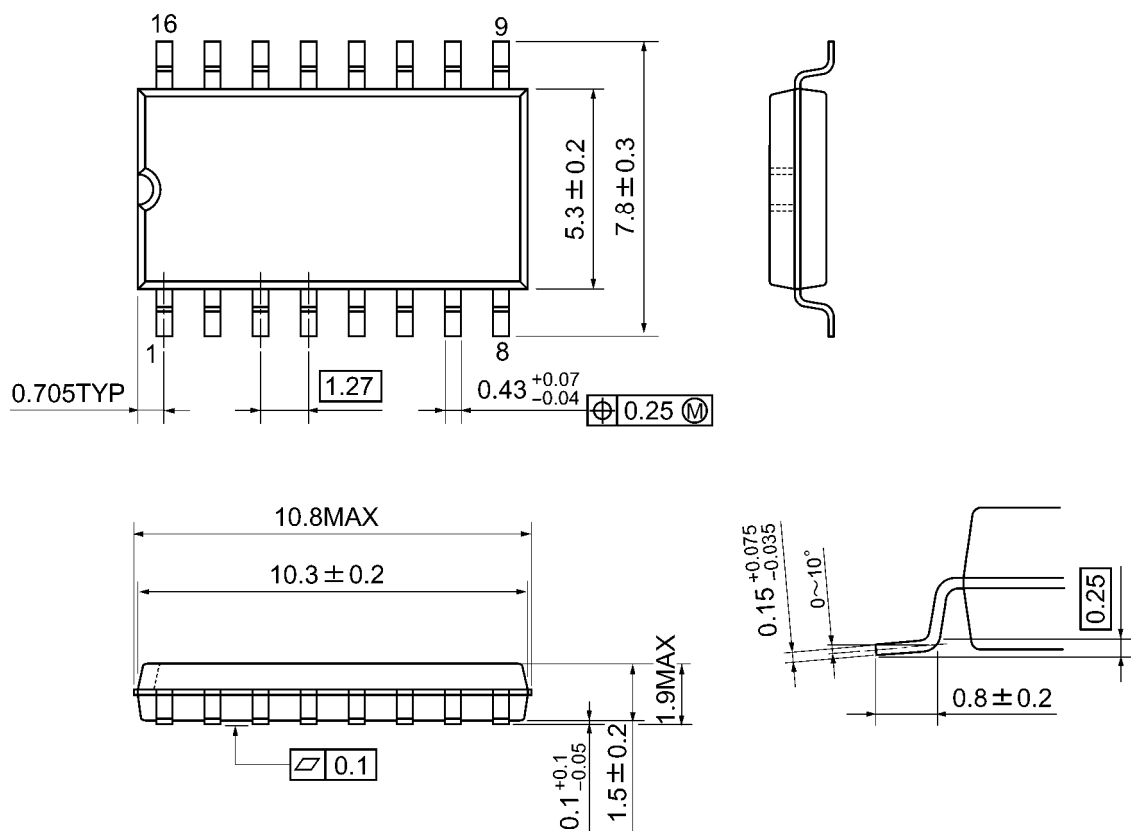


Weight: 1.00 g (typ.)

Package Dimensions

SOP16-P-300-1.27A

Unit: mm



Weight: 0.18 g (typ.)

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