

8-Bit

# SAL-XC886CLM

8-Bit Single Chip Microcontroller

Data Sheet V1.0 2010-05

Microcontrollers

Edition 2010-05

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## **8-Bit Single Chip Microcontroller**

SAL-XC886CLM

## 1 Summary of Features

The SAL-XC886 has the following features:

- High-performance XC800 Core
  - compatible with standard 8051 processor
  - two clocks per machine cycle architecture (for memory access without wait state)
  - two data pointers
- · On-chip memory
  - 12 Kbytes of Boot ROM
  - 256 bytes of RAM
  - 1.5 Kbytes of XRAM
  - 24/32 Kbytes of Flash (includes memory protection strategy)
- I/O port supply at 5.0 V and core logic supply at 2.5 V (generated by embedded voltage regulator)

(more features on next page)

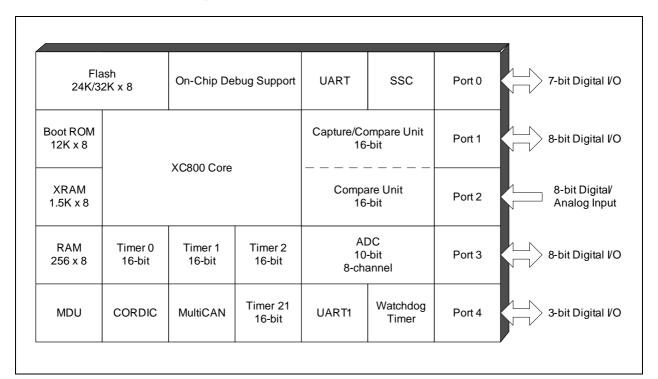


Figure 1 SAL-XC886 Functional Units



#### **Summary of Features**

#### Features: (continued)

- Power-on reset generation
- Brownout detection for core logic supply
- On-chip OSC and PLL for clock generation
  - PLL loss-of-lock detection
- Power saving modes
  - slow-down mode
  - idle mode
  - power-down mode with wake-up capability via RXD or EXINTO
  - clock gating control to each peripheral
- Programmable 16-bit Watchdog Timer (WDT)
- Six ports
  - Up to 48 pins as digital I/O
  - 8 pins as digital/analog input
- 8-channel, 10-bit ADC
- Four 16-bit timers
  - Timer 0 and Timer 1 (T0 and T1)
  - Timer 2 and Timer 21 (T2 and T21)
- Multiplication/Division Unit for arithmetic operations (MDU)
- Software libraries to support floating point and MDU calculations
- CORDIC Coprocessor for computation of trigonometric, hyperbolic and linear functions
- MultiCAN with 2 nodes, 32 message objects
- Capture/compare unit for PWM signal generation (CCU6)
- Two full-duplex serial interfaces (UART and UART1)
- Synchronous serial channel (SSC)
- On-chip debug support
  - 1 Kbyte of monitor ROM (part of the 12-Kbyte Boot ROM)
  - 64 bytes of monitor RAM
- Package:
  - PG-TQFP-48
- Temperature range T<sub>Δ</sub>:
  - SAL (-40 to 150 °C)



#### **Summary of Features**

#### **SAL-XC886 Variant Devices**

The SAL-XC886 product family features devices with different configurations and program memory sizes, to offer cost-effective solutions for different application requirements.

The list of SAL-XC886 device configurations are summarized in **Table 1**.

Table 1 Device Profile

Device Type	Sales Type	Program Memory (Kbytes)	CAN Module	LIN BSL Support	MDU Module
Flash	SAL-XC886-8FFA 5V	32	No	No	No
	SAL-XC886C-8FFA 5V	32	Yes	No	No
	SAL-XC886CM-8FFA 5V	32	Yes	No	Yes
	SAL-XC886LM-8FFA 5V	32	No	Yes	Yes
	SAL-XC886CLM-8FFA 5V	32	Yes	Yes	Yes
	SAL-XC886-6FFA 5V	24	No	No	No
	SAL-XC886C-6FFA 5V	24	Yes	No	No
	SAL-XC886CM-6FFA 5V	24	Yes	No	Yes
	SAL-XC886LM-6FFA 5V	24	No	Yes	Yes
	SAL-XC886CLM-6FFA 5V	24	Yes	Yes	Yes

Note: For variants with LIN BSL support, only LIN BSL is available regardless of the availability of the CAN module.

As this document refers to all the derivatives, some description may not apply to a specific product. For simplicity, all versions are referred to by the term SAL-XC886 throughout this document.

### **Ordering Information**

The ordering code for Infineon Technologies microcontrollers provides an exact reference to the required product. This ordering code identifies:

- The derivative itself, i.e. its function set, the temperature range, and the supply voltage
- The package and the type of delivery

For the available ordering codes for the SAL-XC886, please refer to your responsible sales representative or your local distributor.



## 2 General Device Information

**Chapter 2** contains the block diagram, pin configurations, definitions and functions of the SAL-XC886.

## 2.1 Block Diagram

The block diagram of the SAL-XC886 is shown in Figure 2.

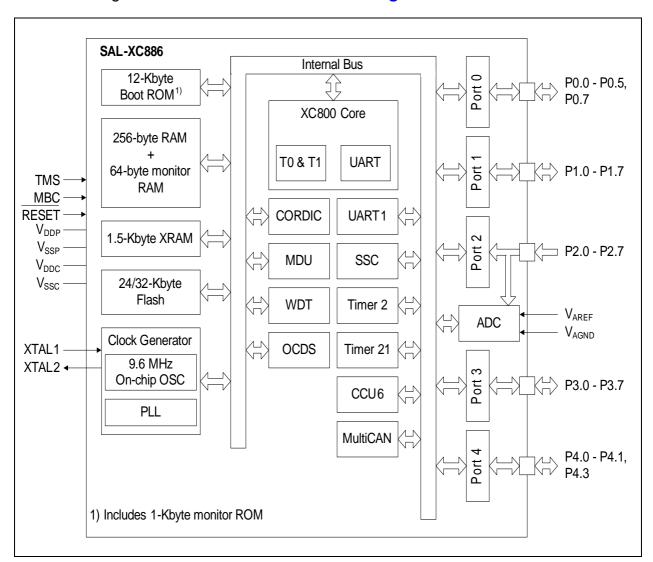


Figure 2 SAL-XC886 Block Diagram



## 2.2 Logic Symbol

The logic symbols of the SAL-XC886 are shown in Figure 3.

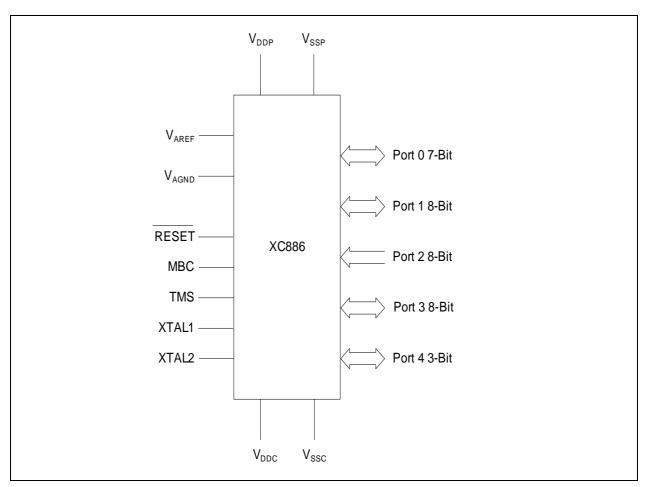


Figure 3 SAL-XC886 Logic Symbol



## 2.3 Pin Configuration

The pin configuration of the XC886, which is based on the PG-TQFP-48 package, is shown in Figure 4.

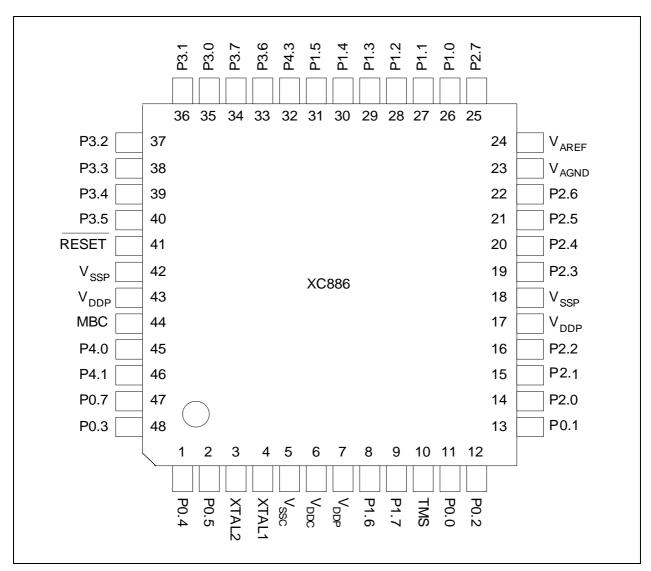


Figure 4 XC886 Pin Configuration, PG-TQFP-48 Package (top view)



## 2.4 Pin Definitions and Functions

The functions and default states of the SAL-XC886 external pins are provided in Table 2.

Table 2 Pin Definitions and Functions

Table 2	riii Deiiii		ı	Ι	
Symbol	Pin Number	Туре	Reset State	Function	
P0		I/O		Port 0 Port 0 is an 8-bit bidirectional general purpose I/O port. It can be used as alternate functions for the JTAG, CCU6, UART, UART1, Timer 2, Timer 21, MultiCAN and SSC.	
P0.0	11		Hi-Z	TCK_0 T12HR_1 CC61_1 CLKOUT_0 RXDO_1	JTAG Clock Input CCU6 Timer 12 Hardware Run Input Input/Output of Capture/Compare channel 1 Clock Output UART Transmit Data Output
P0.1	13		Hi-Z	TDI_0 T13HR_1 RXD_1 RXDC1_0 COUT61_1 EXF2_1	JTAG Serial Data Input CCU6 Timer 13 Hardware Run Input UART Receive Data Input MultiCAN Node 1 Receiver Input Output of Capture/Compare channel 1 Timer 2 External Flag Output
P0.2	12		PU	CTRAP_2 TDO_0 TXD_1 TXDC1_0	CCU6 Trap Input JTAG Serial Data Output UART Transmit Data Output/Clock Output MultiCAN Node 1 Transmitter Output
P0.3	48		Hi-Z	SCK_1 COUT63_1 RXDO1_0	SSC Clock Input/Output Output of Capture/Compare channel 3 UART1 Transmit Data Output



 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function	
P0.4	1		Hi-Z	MTSR_1	SSC Master Transmit Output/ Slave Receive Input
				CC62_1	Input/Output of Capture/Compare channel 2
				TXD1_0	UART1 Transmit Data Output/Clock Output
P0.5	2		Hi-Z	MRST_1	SSC Master Receive Input/Slave Transmit Output
				EXINT0_0	External Interrupt Input 0
				T2EX1_1	Timer 21 External Trigger Input
				RXD1_0	UART1 Receive Data Input
				COUT62_1	Output of Capture/Compare channel 2
P0.7	47		PU	CLKOUT_1	Clock Output

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 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function	
P1		I/O		Port 1  Port 1 is an 8-bit bidirectional general purpose I/O port. It can be used as alternate functions for the JTAG, CCU6, UART, Timer 0, Timer 1, Timer 2, Timer 21, MultiCAN and SSC.	
P1.0	26		PU	RXD_0 T2EX RXDC0_0	UART Receive Data Input Timer 2 External Trigger Input MultiCAN Node 0 Receiver Input
P1.1	27		PU	EXINT3 T0_1 TDO_1 TXD_0 TXDC0_0	External Interrupt Input 3 Timer 0 Input JTAG Serial Data Output UART Transmit Data Output/Clock Output MultiCAN Node 0 Transmitter Output
P1.2	28		PU	SCK_0	SSC Clock Input/Output
P1.3	29		PU	MTSR_0 TXDC1_3	SSC Master Transmit Output/Slave Receive Input MultiCAN Node 1 Transmitter Output
P1.4	30		PU	MRST_0 EXINT0_1 RXDC1_3	SSC Master Receive Input/ Slave Transmit Output External Interrupt Input 0 MultiCAN Node 1 Receiver Input
P1.5	31		PU	CCPOS0_1 EXINT5 T1_1 EXF2_0 RXDO_0	CCU6 Hall Input 0 External Interrupt Input 5 Timer 1 Input Timer 2 External Flag Output UART Transmit Data Output



 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function	
P1.6	8		PU	CCPOS1_1 T12HR_0	CCU6 Hall Input 1 CCU6 Timer 12 Hardware Run Input
				EXINT6_0 RXDC0_2 T21_1	·
P1.7	9		PU	CCPOS2_1 T13HR_0 T2_1 TXDC0_2	CCU6 Hall Input 2 CCU6 Timer 13 Hardware Run Input Timer 2 Input MultiCAN Node 0 Transmitter
					Output  .6 can be used as a software chip t for the SSC.



 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function		
P2		I		Port 2 Port 2 is an 8-bit general purpose input-only port. It can be used as alternate functions for the digital inputs of the JTAG and CCU6. It is also used as the analog inputs for the ADC.		
P2.0	14		Hi-Z	CCPOS0_0 EXINT1_0 T12HR_2 TCK_1 CC61_3 AN0	CCU6 Hall Input 0 External Interrupt Input 1 CCU6 Timer 12 Hardware Run Input JTAG Clock Input Input of Capture/Compare channel 1 Analog Input 0	
P2.1	15		Hi-Z	CCPOS1_0 EXINT2_0 T13HR_2 TDI_1 CC62_3 AN1	CCU6 Hall Input 1 External Interrupt Input 2 CCU6 Timer 13 Hardware Run Input JTAG Serial Data Input Input of Capture/Compare channel 2 Analog Input 1	
P2.2	16		Hi-Z	CCPOS2_0 CTRAP_1 CC60_3 AN2	CCU6 Hall Input 2 CCU6 Trap Input Input of Capture/Compare channel 0 Analog Input 2	
P2.3	19		Hi-Z	AN3	Analog Input 3	
P2.4	20		Hi-Z	AN4	Analog Input 4	
P2.5	21		Hi-Z	AN5	Analog Input 5	
P2.6	22		Hi-Z	AN6	Analog Input 6	
P2.7	25		Hi-Z	AN7	Analog Input 7	



 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function	
P3		I/O		I/O port. It ca	B-bit bidirectional general purpose in be used as alternate functions ART1, Timer 21 and MultiCAN.
P3.0	35		Hi-Z	CCPOS1_2 CC60_0 RXDO1_1	CCU6 Hall Input 1 Input/Output of Capture/Compare channel 0 UART1 Transmit Data Output
P3.1	36		Hi-Z	CCPOS0_2 CC61_2 COUT60_0 TXD1_1	CCU6 Hall Input 0 Input/Output of Capture/Compare channel 1 Output of Capture/Compare channel 0 UART1 Transmit Data Output/Clock Output
P3.2	37		Hi-Z	CCPOS2_2 RXDC1_1 RXD1_1 CC61_0	CCU6 Hall Input 2 MultiCAN Node 1 Receiver Input UART1 Receive Data Input Input/Output of Capture/Compare channel 1
P3.3	38		Hi-Z	COUT61_0 TXDC1_1	Output of Capture/Compare channel 1 MultiCAN Node 1 Transmitter Output
P3.4	39		Hi-Z	CC62_0 RXDC0_1 T2EX1_0	Input/Output of Capture/Compare channel 2 MultiCAN Node 0 Receiver Input Timer 21 External Trigger Input
P3.5	40		Hi-Z	COUT62_0 EXF21_0 TXDC0_1	Output of Capture/Compare channel 2 Timer 21 External Flag Output MultiCAN Node 0 Transmitter Output
P3.6	33		PD	CTRAP_0	CCU6 Trap Input



 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function	
P3.7	34		Hi-Z	EXINT4 COUT63_0	External Interrupt Input 4 Output of Capture/Compare channel 3

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 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function				
P4		I/O		Port 4 Port 4 is an 8-bit bidirectional general purp I/O port. It can be used as alternate functi for CCU6, Timer 0, Timer 1, Timer 21 and MultiCAN.  RXDC0_3 MultiCAN Node 0 Receiver				
P4.0	45		Hi-Z	RXDC0_3 CC60_1	MultiCAN Node 0 Receiver Input Output of Capture/Compare channel 0			
P4.1	46		Hi-Z	TXDC0_3 COUT60_1	MultiCAN Node 0 Transmitter Output Output of Capture/Compare channel 0			
P4.3	32		Hi-Z	EXF21_1 COUT63_2	Timer 21 External Flag Output Output of Capture/Compare channel 3			

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 Table 2
 Pin Definitions and Functions (cont'd)

Symbol	Pin Number	Туре	Reset State	Function
$\overline{V_{DDP}}$	7, 17, 43	_	_	I/O Port Supply (5.0 V) Also used by EVR and analog modules. All pins must be connected.
$\overline{V_{SSP}}$	18, 42	_	_	I/O Port Ground All pins must be connected.
$\overline{V_{DDC}}$	6	_	_	Core Supply Monitor (2.5 V)
$\overline{V_{ extsf{SSC}}}$	5	_	_	Core Supply Ground
$\overline{V_{AREF}}$	24	_	_	ADC Reference Voltage
$\overline{V_{AGND}}$	23	_	_	ADC Reference Ground
XTAL1	4	1	Hi-Z	External Oscillator Input (backup for on-chip OSC, normally NC)
XTAL2	3	0	Hi-Z	External Oscillator Output (backup for on-chip OSC, normally NC)
TMS	10	I	PD	Test Mode Select
RESET	41	I	PU	Reset Input
MBC <sup>1)</sup>	44	I	PU	Monitor & BootStrap Loader Control

<sup>1)</sup> An external pull-up device in the range of 4.7 k $\Omega$  to 100 k $\Omega$ . is required to enter user mode. Alternatively MBC can be tied to high if alternate functions (for debugging) of the pin are not utilized.

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## 3 Functional Description

**Chapter 3** provides an overview of the SAL-XC886 functional description.

#### 3.1 Processor Architecture

The SAL-XC886 is based on a high-performance 8-bit Central Processing Unit (CPU) that is compatible with the standard 8051 processor. While the standard 8051 processor is designed around a 12-clock machine cycle, the SAL-XC886 CPU uses a 2-clock machine cycle. This allows fast access to ROM or RAM memories without wait state. Access to the Flash memory, however, requires an additional wait state (one machine cycle). The instruction set consists of 45% one-byte, 41% two-byte and 14% three-byte instructions.

The SAL-XC886 CPU provides a range of debugging features, including basic stop/start, single-step execution, breakpoint support and read/write access to the data memory, program memory and Special Function Registers (SFRs).

Figure 5 shows the CPU functional blocks.

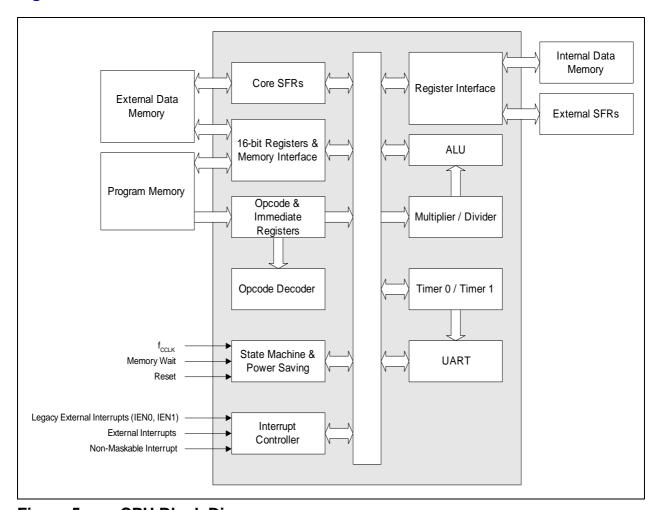


Figure 5 CPU Block Diagram



## 3.2 Memory Organization

The SAL-XC886 CPU operates in the following five address spaces:

- 12 Kbytes of Boot ROM program memory
- 256 bytes of internal RAM data memory
- 1.5 Kbytes of XRAM memory (XRAM can be read/written as program memory or external data memory)
- A 128-byte Special Function Register area
- 24/32 Kbytes of Flash program memory

**Figure 6** illustrates the memory address spaces of the 32-Kbyte Flash devices. For the 24-Kbyte Flash devices, the shaded banks are not available.

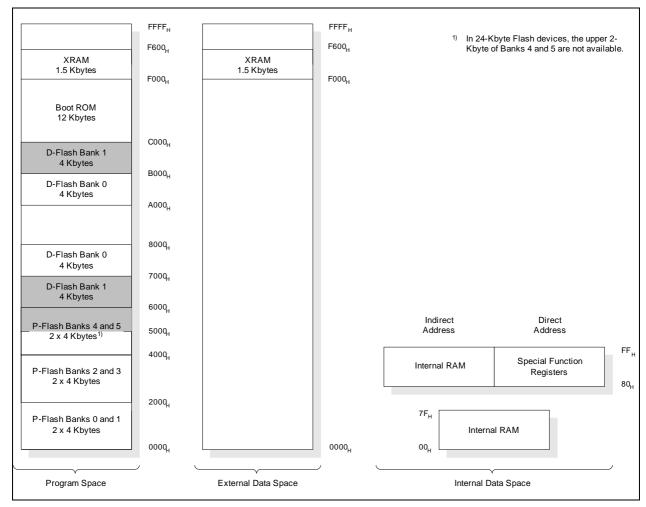


Figure 6 Memory Map of SAL-XC886 Flash Device



## 3.2.1 Memory Protection Strategy

The SAL-XC886 memory protection strategy includes:

- Read-out protection: The user is able to protect the contents in the Flash memory from being read
  - Flash protection is enabled by programming a valid password (8-bit non-zero value) via BSL mode 6.
- Flash program and erase protection.

## 3.2.1.1 Flash Memory Protection

As long as a valid password is available, all external access to the device, including the Flash, will be blocked.

For additional security, the Flash hardware protection can be enabled to implement a second layer of read-out protection, as well as to enable program and erase protection.

Flash hardware protection is available only for Flash devices and comes in two modes:

- Mode 0: Only the P-Flash is protected; the D-Flash is unprotected
- Mode 1: Both the P-Flash and D-Flash are protected

The selection of each protection mode and the restrictions imposed are summarized in **Table 3**.

Table 3 Flash Protection Modes

Flash Protection	Without hardware protection	With hardware protection					
Hardware Protection Mode	-	0	1				
Activation	Program a valid passv	vord via BSL mode 6					
Selection	Bit 4 of password = 0	Bit 4 of password = 1 MSB of password = 0	Bit 4 of password = 1 MSB of password = 1				
P-Flash contents can be read by	Read instructions in any program memory	Read instructions in the P-Flash	Read instructions in the P-Flash or D- Flash				
External access to P-Flash	Not possible	Not possible	Not possible				
P-Flash program and erase	Possible	Not possible	Not possible				
D-Flash contents can be read by	Read instructions in any program memory	Read instructions in any program memory	Read instructions in the P-Flash or D-Flash				



Table 3 Flash Protection Modes (cont'd)

Flash Protection	Without hardware protection	With hardware protection					
External access to D-Flash	Not possible	Not possible	Not possible				
D-Flash program	Possible	Possible	Not possible				
D-Flash erase	Possible	Possible, on condition that bit DFLASHEN in register MISC_CON is set to 1 prior to each erase operation	Not possible				

BSL mode 6, which is used for enabling Flash protection, can also be used for disabling Flash protection. Here, the programmed password must be provided by the user. A password match triggers an automatic erase of the protected P-Flash and D-Flash contents, including the programmed password. The Flash protection is then disabled upon the next reset.

Although no protection scheme can be considered infallible, the SAL-XC886 memory protection strategy provides a very high level of protection for a general purpose microcontroller.

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## 3.2.2 Special Function Register

The Special Function Registers (SFRs) occupy direct internal data memory space in the range  $80_H$  to  $FF_H$ . All registers, except the program counter, reside in the SFR area. The SFRs include pointers and registers that provide an interface between the CPU and the on-chip peripherals. As the 128-SFR range is less than the total number of registers required, address extension mechanisms are required to increase the number of addressable SFRs. The address extension mechanisms include:

- Mapping
- Paging

### 3.2.2.1 Address Extension by Mapping

Address extension is performed at the system level by mapping. The SFR area is extended into two portions: the standard (non-mapped) SFR area and the mapped SFR area. Each portion supports the same address range  $80_{\rm H}$  to FF<sub>H</sub>, bringing the number of addressable SFRs to 256. The extended address range is not directly controlled by the CPU instruction itself, but is derived from bit RMAP in the system control register SYSCON0 at address  $8F_{\rm H}$ . To access SFRs in the mapped area, bit RMAP in SFR SYSCON0 must be set. Alternatively, the SFRs in the standard area can be accessed by clearing bit RMAP. The SFR area can be selected as shown in **Figure 7**.

As long as bit RMAP is set, the mapped SFR area can be accessed. This bit is not cleared automatically by hardware. Thus, before standard/mapped registers are accessed, bit RMAP must be cleared/set, respectively, by software.

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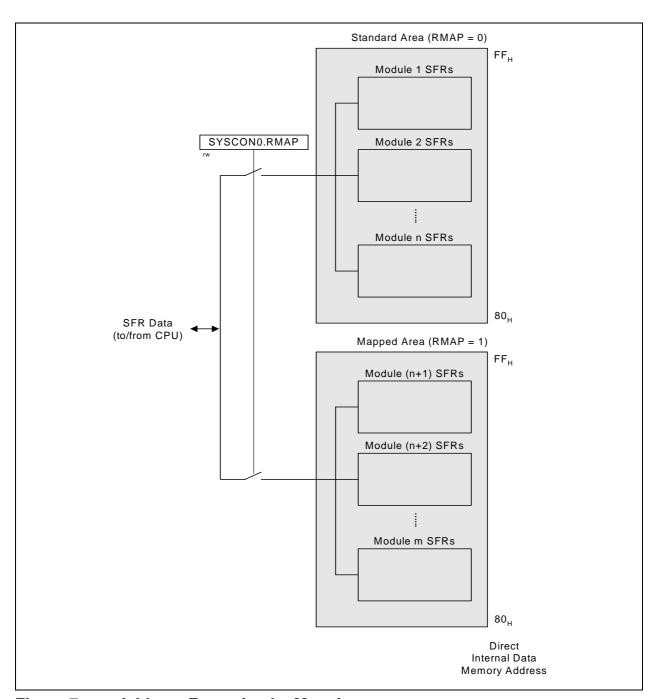


Figure 7 Address Extension by Mapping

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Reset Value: 04 L



#### **Functional Description**

# SYSCON0 System Control Register 0

7	6	5	4	3	2	1	0
	0	1	IMODE	0	1	0	RMAP
	r	I	rw.	r	r	<u> </u>	rw/

Field	Bits	Туре	Description
RMAP	0	rw	Interrupt Node XINTR0 Enable  O The access to the standard SFR area is enabled  The access to the mapped SFR area is enabled
1	2	r	Reserved Returns 1 if read; should be written with 1.
0	[7:5], 3,1	r	Reserved Returns 0 if read; should be written with 0.

Note: The RMAP bit should be cleared/set by ANL or ORL instructions.

## 3.2.2.2 Address Extension by Paging

Address extension is further performed at the module level by paging. With the address extension by mapping, the SAL-XC886 has a 256-SFR address range. However, this is still less than the total number of SFRs needed by the on-chip peripherals. To meet this requirement, some peripherals have a built-in local address extension mechanism for increasing the number of addressable SFRs. The extended address range is not directly controlled by the CPU instruction itself, but is derived from bit field PAGE in the module page register MOD\_PAGE. Hence, the bit field PAGE must be programmed before accessing the SFR of the target module. Each module may contain a different number of pages and a different number of SFRs per page, depending on the specific requirement. Besides setting the correct RMAP bit value to select the SFR area, the user must also ensure that a valid PAGE is selected to target the desired SFR. A page inside the extended address range can be selected as shown in Figure 8.



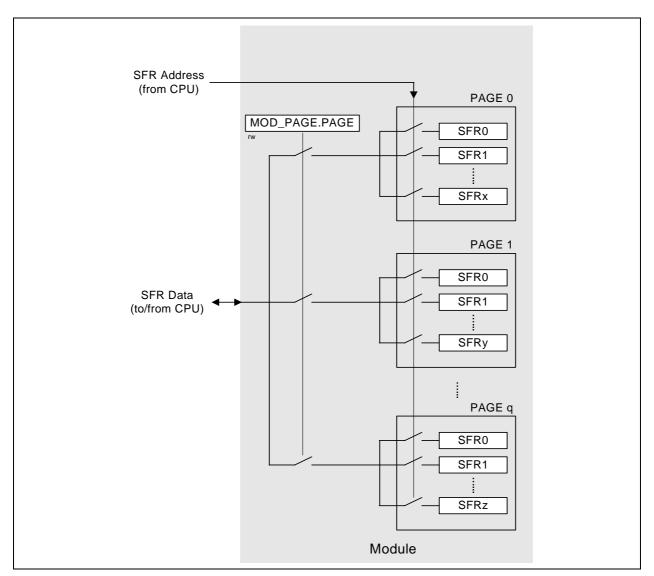


Figure 8 Address Extension by Paging

In order to access a register located in a page different from the actual one, the current page must be exited. This is done by reprogramming the bit field PAGE in the page register. Only then can the desired access be performed.

If an interrupt routine is initiated between the page register access and the module register access, and the interrupt needs to access a register located in another page, the current page setting can be saved, the new one programmed and the old page setting restored. This is possible with the storage fields STx (x = 0 - 3) for the save and restore action of the current page setting. By indicating which storage bit field should be used in parallel with the new page value, a single write operation can:

 Save the contents of PAGE in STx before overwriting with the new value (this is done in the beginning of the interrupt routine to save the current page setting and program the new page number); or



- Overwrite the contents of PAGE with the contents of STx, ignoring the value written to the bit positions of PAGE
  - (this is done at the end of the interrupt routine to restore the previous page setting before the interrupt occurred)

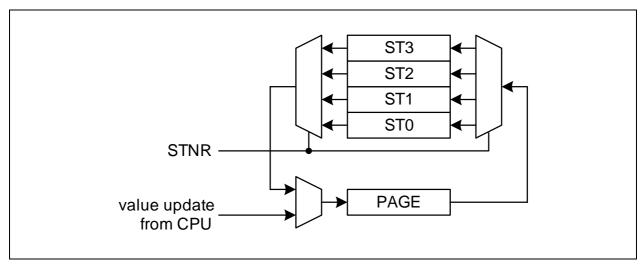


Figure 9 Storage Elements for Paging

With this mechanism, a certain number of interrupt routines (or other routines) can perform page changes without reading and storing the previously used page information. The use of only write operations makes the system simpler and faster. Consequently, this mechanism significantly improves the performance of short interrupt routines.

The SAL-XC886 supports local address extension for:

- Parallel Ports
- Analog-to-Digital Converter (ADC)
- Capture/Compare Unit 6 (CCU6)
- System Control Registers

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Reset Value: 00<sub>H</sub>



## **Functional Description**

The page register has the following definition:

# MOD\_PAGE Page Register for module MOD

7	6	5	4	3	2	1	0
0	P	ST	NR	0		PAGE	
V	۸/	,	۸/	r		r\A/	

Field	Bits	Туре	Description
PAGE	[2:0]	rw	Page Bits When written, the value indicates the new page. When read, the value indicates the currently active page.
STNR	[5:4]	W	Storage Number This number indicates which storage bit field is the target of the operation defined by bit field OP. If OP = 10 <sub>B</sub> , the contents of PAGE are saved in STx before being overwritten with the new value. If OP = 11 <sub>B</sub> , the contents of PAGE are overwritten by the contents of STx. The value written to the bit positions of PAGE is ignored.  OO ST0 is selected.
			01 ST1 is selected. 10 ST2 is selected. 11 ST3 is selected.



Field	Bits	Туре	Description
OP	[7:6]	W	<ul> <li>Operation</li> <li>OX Manual page mode. The value of STNR is ignored and PAGE is directly written.</li> <li>10 New page programming with automatic page saving. The value written to the bit positions of PAGE is stored. In parallel, the previous contents of PAGE are saved in the storage bit field STx indicated by STNR.</li> <li>11 Automatic restore page action. The value written to the bit positions PAGE is ignored and instead, PAGE is overwritten by the contents of the storage bit field STx indicated by STNR.</li> </ul>
0	3	r	Reserved Returns 0 if read; should be written with 0.

#### 3.2.3 Bit Protection Scheme

The bit protection scheme prevents direct software writing of selected bits (i.e., protected bits) using the PASSWD register. When the bit field MODE is  $11_B$ , writing  $10011_B$  to the bit field PASS opens access to writing of all protected bits, and writing  $10101_B$  to the bit field PASS closes access to writing of all protected bits. In both cases, the value of the bit field MODE is not changed even if PASSWD register is written with  $98_H$  or  $A8_H$ . It can only be changed when bit field PASS is written with  $11000_B$ , for example, writing  $D0_H$  to PASSWD register disables the bit protection scheme.

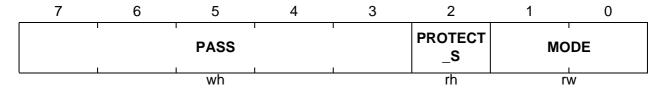
Note that access is opened for maximum 32 CCLKs if the "close access" password is not written. If "open access" password is written again before the end of 32 CCLK cycles, there will be a recount of 32 CCLK cycles. The protected bits include the N- and K-Divider bits, NDIV and KDIV; the Watchdog Timer enable bit, WDTEN; and the power-down and slow-down enable bits, PD and SD.



# 3.2.3.1 Password Register

## **PASSWD**

Password Register Reset Value: 07<sub>H</sub>



Field	Bits	Туре	Description
MODE	[1:0]	rw	Bit Protection Scheme Control Bits  O Scheme disabled - direct access to the protected bits is allowed.  11 Scheme enabled - the bit field PASS has to be written with the passwords to open and close the access to protected bits. (default)  Others:Scheme Enabled.  These two bits cannot be written directly. To change the value between 11 <sub>B</sub> and 00 <sub>B</sub> , the bit field PASS must be written with 11000 <sub>B</sub> ; only then, will the
PROTECT_S	2	rh	MODE[1:0] be registered.  Bit Protection Signal Status Bit This bit shows the status of the protection.  O Software is able to write to all protected bits.  Software is unable to write to any protected bits.
PASS	[7:3]	wh	Password Bits The Bit Protection Scheme only recognizes three patterns.  11000 <sub>B</sub> Enables writing of the bit field MODE.  10011 <sub>B</sub> Opens access to writing of all protected bits.  10101 <sub>B</sub> Closes access to writing of all protected bits



## 3.2.4 SAL-XC886 Register Overview

The SFRs of the SAL-XC886 are organized into groups according to their functional units. The contents (bits) of the SFRs are summarized in **Chapter 3.2.4.1** to **Chapter 3.2.4.14**.

Note: The addresses of the bitaddressable SFRs appear in bold typeface.

## 3.2.4.1 CPU Registers

The CPU SFRs can be accessed in both the standard and mapped memory areas (RMAP = 0 or 1).

Table 4 CPU Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	= 0 or 1		I		I		I	I	I		
81 <sub>H</sub>	SP Reset: 07 <sub>H</sub>	Bit Field				S	Р				
	Stack Pointer Register	Туре				r	W				
82 <sub>H</sub>	DPL Reset: 00 <sub>H</sub>	Bit Field	DPL7	DPL6	DPL5	DPL4	DPL3	DPL2	DPL1	DPL0	
	Data Pointer Register Low	Type	rw	rw	rw	rw	rw	rw	rw	rw	
83 <sub>H</sub>	DPH Reset: 00 <sub>H</sub>	Bit Field	DPH7	DPH6	DPH5	DPH4	DPH3	DPH2	DPH1	DPH0	
	Data Pointer Register High	Туре	rw	rw	rw	rw	rw	rw	rw	rw	
87 <sub>H</sub>	PCON Reset: 00 <sub>H</sub>	Bit Field	SMOD		0		GF1	GF0	0	IDLE	
	Power Control Register	Туре	rw		r		rw	rw	r	rw	
88 <sub>H</sub>	TCON Reset: 00H	Bit Field	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	
	Timer Control Register	Туре	rwh	rw	rwh	rw	rwh	rw	rwh	rw	
89 <sub>H</sub>	TMOD Reset: 00 <sub>H</sub> Timer Mode Register	Bit Field	GATE 1	T1S	T1	M	GATE 0	T0S	TO	)M	
		Туре	rw	rw	rw		rw	rw	rw		
8A <sub>H</sub>	TL0 Reset: 00 <sub>H</sub>	Bit Field	it Field VAL								
	Timer 0 Register Low	Туре				rv	vh				
8B <sub>H</sub>	TL1 Reset: 00 <sub>H</sub>	Bit Field				V	٩L				
	Timer 1 Register Low	Туре				rv	vh				
8C <sub>H</sub>	TH0 Reset: 00 <sub>H</sub>	Bit Field				V	٩L				
	Timer 0 Register High	Туре				rv	vh				
8D <sub>H</sub>	TH1 Reset: 00 <sub>H</sub>	Bit Field				V	٩L				
	Timer 1 Register High	Туре				rv	vh				
98 <sub>H</sub>	SCON Reset: 00H	Bit Field	SM0	SM1	SM2	REN	TB8	RB8	TI	RI	
	Serial Channel Control Register	Туре	rw	rw	rw	rw	rw	rwh	rwh	rwh	
99 <sub>H</sub>	SBUF Reset: 00 <sub>H</sub>	Bit Field				V	AL.				
	Serial Data Buffer Register	Туре				rv	vh				



Table 4 CPU Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
A2 <sub>H</sub>	EO Reset: 00 <sub>H</sub> Extended Operation Register	Bit Field		0		TRAP_ EN		0		
		Type		r		rw		r		rw
A8 <sub>H</sub>	IEN0 Reset: 00 <sub>H</sub>	Bit Field	EA	0	ET2	ES	ET1	EX1	ET0	EX0
	Interrupt Enable Register 0  IP Reset: 00H Interrupt Priority Register  IPH Reset: 00H Interrupt Priority High Register  PSW Reset: 00H Program Status Word Register	Туре	rw	r	rw	rw	rw	rw	rw	rw
B8 <sub>H</sub>		Bit Field	(	)	PT2	PS	PT1	PX1	PT0	PX0
	Interrupt Priority Register	Type	1	r	rw	rw	rw	rw	rw	rw
в9 <sub>Н</sub>		Bit Field	(	)	PT2H	PSH	PT1H	PX1H	PT0H	PX0H
	Interrupt Priority High Register	Туре	r		rw	rw	rw	rw	rw	rw
D0 <sub>H</sub>	· · · · ·	Bit Field	CY	AC	F0	RS1	RS0	OV	F1	Р
	Program Status Word Register	Туре	rwh	rwh	rw	rw	rw	rwh	rw	rh
E0 <sub>H</sub>	ACC Reset: 00H	Bit Field	ACC7	ACC6	ACC5	ACC4	ACC3	ACC2	ACC1	ACC0
	Accumulator Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
E8 <sub>H</sub>	IEN1 Reset: 00 <sub>H</sub> Interrupt Enable Register 1	Bit Field	ECCIP 3	ECCIP 2	ECCIP 1	ECCIP 0	EXM	EX2	ESSC	EADC
		Type	rw	rw	rw	rw	rw	rw	rw	rw
F0 <sub>H</sub>	B Reset: 00 <sub>H</sub>	Bit Field	В7	В6	B5	B4	В3	B2	B1	В0
	B Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
F8 <sub>H</sub>	IP1 Reset: 00 <sub>H</sub> Interrupt Priority 1 Register	Bit Field	PCCIP 3	PCCIP 2	PCCIP 1	PCCIP 0	PXM	PX2	PSSC	PADC
		Туре	rw	rw	rw	rw	rw	rw	rw	rw
F9 <sub>H</sub>	IPH1 Reset: 00 <sub>H</sub> Interrupt Priority 1 High Register	Bit Field	PCCIP 3H	PCCIP 2H	PCCIP 1H	PCCIP 0H	PXMH	PX2H	PSSC H	PADC H
		Type	rw	rw	rw	rw	rw	rw	rw	rw

# 3.2.4.2 MDU Registers

The MDU SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 5 MDU Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP = 1										
B0 <sub>H</sub>	MDUSTAT Reset: 00 <sub>H</sub> MDU Status Register	Bit Field	0					BSY	IERR	IRDY
		Туре	r					rh	rwh	rwh
B1 <sub>H</sub>	MDUCON Reset: 00 <sub>H</sub> MDU Control Register	Bit Field	IE	IR	RSEL	STAR T	OPCODE			
		Туре	rw	rw	rw	rwh	rw			
B2 <sub>H</sub>	MD0 Reset: 00 <sub>H</sub> MDU Operand Register 0	Bit Field	DATA DATA							
		Туре	rw							
B2 <sub>H</sub>	MR0 Reset: 00 <sub>H</sub> MDU Result Register 0	Bit Field	DATA							
		Туре	rh							



Table 5 MDU Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
B3 <sub>H</sub>	MD1 Reset: 00 <sub>H</sub>	Bit Field		·I		DA	TA		·	
	MDU Operand Register 1	Туре				r	W			
B3 <sub>H</sub>	MR1 Reset: 00 <sub>H</sub>	Bit Field				DA	λTA			
	MDU Result Register 1	Туре				r	h			
B4 <sub>H</sub>	MD2 Reset: 00 <sub>H</sub>	Bit Field				DA	ΛTA			
	MDU Operand Register 2	Туре				r	W			
B4 <sub>H</sub>	MR2 Reset: 00 <sub>H</sub>	Bit Field				DA	λTA			
	MDU Result Register 2	Туре				r	h			
B5 <sub>H</sub>	MD3 Reset: 00 <sub>H</sub>	Bit Field				DA	λTA			
	MDU Operand Register 3	Туре				r	W			
B5 <sub>H</sub>	MR3 Reset: 00 <sub>H</sub>	Bit Field				DA	ΛTA			
	MDU Result Register 3	Туре				r	h			
B6 <sub>H</sub>	MD4 Reset: 00 <sub>H</sub>	Bit Field				DA	ΛTA			
	MDU Operand Register 4	Туре				r	W			
B6 <sub>H</sub>	MR4 Reset: 00 <sub>H</sub>	Bit Field				DA	ΛTA			
	MDU Result Register 4	Туре				r	h			
В7 <sub>Н</sub>	MD5 Reset: 00 <sub>H</sub>	Bit Field				DA	ΛTA			
	MDU Operand Register 5	Туре				r	W			
В7 <sub>Н</sub>	MR5 Reset: 00 <sub>H</sub>	Bit Field				DA	λTA			
	MDU Result Register 5	Туре				r	'n			

## 3.2.4.3 CORDIC Registers

The CORDIC SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 6 CORDIC Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 1			•				•		
9A <sub>H</sub>	CD_CORDXL Reset: 00H	Bit Field				DA	ΓAL			
	CORDIC X Data Low Byte	Туре				r	N			
9B <sub>H</sub>	CD_CORDXH Reset: 00H	Bit Field				DA	ГАН			
	CORDIC X Data High Byte	Туре				r	N			
9C <sub>H</sub>	CD_CORDYL Reset: 00H	Bit Field				DA	ΓAL			
	CORDIC Y Data Low Byte	Туре				r	N			
9D <sub>H</sub>	CD_CORDYH Reset: 00H	Bit Field				DA	ГАН			
	CORDIC Y Data High Byte	Туре				r	N			
9E <sub>H</sub>	CD_CORDZL Reset: 00H	Bit Field				DA	ΓAL			
	CORDIC Z Data Low Byte	Туре				r	N			



 Table 6
 CORDIC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
9F <sub>H</sub>	CD_CORDZH Reset: 00H	Bit Field	DATAH							
	CORDIC Z Data High Byte	Туре	rw							
A0 <sub>H</sub>	CD_STATC Reset: 00 <sub>H</sub> CORDIC Status and Data	Bit Field	KEEP     KEEP     KEEP     DMAP     INT_E     EOC     ERRO       Z     Y     X     DMAP     INT_E     R						BSY	
	Control Register	Туре	rw	rw	rw	rw	rw	rwh	rh	rh
A1 <sub>H</sub>	CD_CON Reset: 00 <sub>H</sub> CORDIC Control Register	Bit Field	MPS X_USI ST_M ROTV MODE GN ODE EC					ST		
		Туре	rw rw rw rw					rwh		

## 3.2.4.4 System Control Registers

The system control SFRs can be accessed in the mapped memory area (RMAP = 0).

Table 7 SCU Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 0 or 1	•				•		•	•	
8F <sub>H</sub>	SYSCON0 Reset: 04 <sub>H</sub> System Control Register 0	Bit Field		0		IMOD E	0	1	0	RMAP
		Туре		r		rw	r	r	r	rw
RMAP =	= 0									
BF <sub>H</sub>	SCU_PAGE Reset: 00H	Bit Field	C	)P	ST	NR	0		PAGE	
	Page Register	Туре	١	W	١	N	r		rw	
RMAP =	= 0, PAGE 0									
вз <sub>Н</sub>	MODPISEL Reset: 00 <sub>H</sub> Peripheral Input Select Register	Bit Field	0	URRIS H	JTAGT DIS	JTAGT CKS	EXINT 2IS	EXINT 1IS	EXINT 0IS	URRIS
		Туре	r	rw	rw	rw	rw	rw	rw	rw
B4 <sub>H</sub>	IRCON0 Reset: 00 <sub>H</sub> Interrupt Request Register 0	Bit Field	0	EXINT 6	EXINT 5	EXINT 4	EXINT 3	EXINT 2	EXINT 1	EXINT 0
		Туре	r	rwh	rwh	rwh	rwh	rwh	rwh	rwh
B5 <sub>H</sub>	IRCON1 Reset: 00 <sub>H</sub> Interrupt Request Register 1	Bit Field	0	CANS RC2	CANS RC1	ADCS R1	ADCS R0	RIR	TIR	EIR
		Туре	r	rwh	rwh	rwh	rwh	rwh	rwh	rwh
B6 <sub>H</sub>	IRCON2 Reset: 00 <sub>H</sub> Interrupt Request Register 2	Bit Field		0		CANS RC3		0		CANS RC0
		Туре		r		rwh		r		rwh
В7 <sub>Н</sub>	EXICONO Reset: F0H	Bit Field	EXI	NT3	EXI	NT2	EXI	NT1	EXI	NT0
	External Interrupt Control Register 0	Туре	r	W	r	w	r	w	r	W
BA <sub>H</sub>	EXICON1 Reset: 3F <sub>H</sub>	Bit Field	(	0	EXI	NT6	EXI	NT5	EXI	NT4
	External Interrupt Control Register 1	Туре		r	r	w	r	w	r	w
ввн	NMICON Reset: 00 <sub>H</sub> NMI Control Register	Bit Field	0	NMI ECC	NMI VDDP	NMI VDD	NMI OCDS	NMI FLASH	NMI PLL	NMI WDT
		Туре	r	rw	rw	rw	rw	rw	rw	rw



## Table 7 SCU Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
всн	NMISR Reset: 00 <sub>H</sub> NMI Status Register	Bit Field	0	FNMI ECC	FNMI VDDP	FNMI VDD	FNMI OCDS	FNMI FLASH	FNMI PLL	FNMI WDT
		Туре	r	rwh	rwh	rwh	rwh	rwh	rwh	rwh
вD <sub>Н</sub>	BCON Reset: 00 <sub>H</sub>	Bit Field	BG	SEL	0	BRDIS		BRPRE		R
	Baud Rate Control Register	Туре	r	W	r	rw		rw		rw
BE <sub>H</sub>	BG Reset: 00 <sub>H</sub>	Bit Field				BR_V	ALUE			
	Baud Rate Timer/Reload Register	Туре				rv	vh			
E9 <sub>H</sub>	FDCON Reset: 00 <sub>H</sub> Fractional Divider Control	Bit Field	BGS	SYNE N	ERRS YN	EOFS YN	BRK	NDOV	FDM	FDEN
	Register	Type	rw	rw	rwh	rwh	rwh	rwh	rw	rw
EAH	FDSTEP Reset: 00 <sub>H</sub>	Bit Field				ST	ΈP			
	Fractional Divider Reload Register	Туре				r	W			
EBH	FDRES Reset: 00 <sub>H</sub>	Bit Field				RES	SULT			
	Fractional Divider Result Register	Туре				r	h			
RMAP =	: 0, PAGE 1									
вз <sub>Н</sub>	ID Reset: UU <sub>H</sub>	Bit Field			PRODID				VERID	
	Identity Register	Type			r				r	
B4 <sub>H</sub>	PMCON0 Reset: 00 <sub>H</sub> Power Mode Control Register 0	Bit Field	0	WDT RST	WKRS	WK SEL	SD	PD	V	'S
		Туре	r	rwh	rwh	rw	rw	rwh	r	W
B5 <sub>H</sub>	PMCON1 Reset: 00 <sub>H</sub> Power Mode Control Register 1	Bit Field	0	CDC_ DIS	CAN_ DIS	MDU_ DIS	T2_ DIS	CCU_ DIS	SSC_ DIS	ADC_ DIS
		Type	r	rw	rw	rw	rw	rw	rw	rw
B6 <sub>H</sub>	OSC_CON Reset: 08 <sub>H</sub> OSC Control Register	Bit Field		0		OSC PD	XPD	OSC SS	ORD RES	OSCR
		Туре		r		rw	rw	rw	rwh	rh
B7 <sub>H</sub>	PLL_CON Reset: 90 <sub>H</sub> PLL Control Register	Bit Field		NE	DIV		VCO BYP	OSC DISC	RESL D	LOCK
		Туре		r	w		rw	rw	rwh	rh
<sup>BA</sup> H	CMCON Reset: 10 <sub>H</sub> Clock Control Register	Bit Field	VCO SEL	KDIV	0	FCCF G		CLK	REL	
		Type	rw	rw	r	rw		n	W	
BB <sub>H</sub>	PASSWD Reset: 07 <sub>H</sub> Password Register	Bit Field			PASS			PROT ECT_S	МС	DE
		Type			wh			rh	r	W
всн	FEAL Reset: 00 <sub>H</sub>	Bit Field				ECCER	RADDR			
	Flash Error Address Register Low	Туре				r	h			
вD <sub>Н</sub>	FEAH Reset: 00 <sub>H</sub>	Bit Field				ECCER	RADDR			
	Flash Error Address Register High	Туре				r	h			



Table 7 SCU Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
BE <sub>H</sub>	COCON Reset: 00 <sub>H</sub> Clock Output Control Register	Bit Field		0	TLEN	COUT		СО	REL	
		Туре		r	rw	rw		r	W	
E9 <sub>H</sub>	MISC_CON Reset: 00 <sub>H</sub> Miscellaneous Control Register	Bit Field				0				DFLAS HEN
		Type				r				rwh
RMAP =	= 0, PAGE 3									
B3 <sub>H</sub>	XADDRH Reset: F0H	Bit Field			ADDRH					
	On-chip XRAM Address Higher Order	Туре				r	W			
B4 <sub>H</sub>	IRCON3 Reset: 00 <sub>H</sub> Interrupt Request Register 3	Bit Field		0	CANS RC5	CCU6 SR1	(	)	CANS RC4	CCU6 SR0
		Туре		r	rwh	rwh		r	rwh	rwh
в5 <sub>Н</sub>	IRCON4 Reset: 00 <sub>H</sub> Interrupt Request Register 4	Bit Field		0	CANS RC7	CCU6 SR3	(	)	CANS RC6	CCU6 SR2
		Type		r	rwh	rwh		r	rwh	rwh
в7 <sub>Н</sub>	MODPISEL1 Reset: 00 <sub>H</sub> Peripheral Input Select Register	Bit Field	EXINT 6IS		0	UR1	RIS	T21EX IS	JTAGT DIS1	JTAGT CKS1
	1	Type	rw		r	r	W	rw	rw	rw
ва <sub>Н</sub>	MODPISEL2 Reset: 00 <sub>H</sub>	Bit Field		(	0		T21IS	T2IS	T1IS	TOIS
	Peripheral Input Select Register 2	Туре			r		rw	rw	rw	rw
ввн	PMCON2 Reset: 00 <sub>H</sub> Power Mode Control Register 2	Bit Field	old 0 UART 1_DIS		T21_D IS					
		Type				r			rw	rw
BD <sub>H</sub>	MODSUSP Reset: 01 <sub>H</sub> Module Suspend Control	Bit Field		0		T21SU SP	T2SUS P	T13SU SP	T12SU SP	WDTS USP
	Register	Туре		r		rw	rw	rw	rw	rw

# 3.2.4.5 WDT Registers

The WDT SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 8 WDT Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 1									
ввН	WDTCON Reset: 00 <sub>H</sub> Watchdog Timer Control	Bit Field	(	)	WINB EN	WDTP R	0	WDTE N	WDTR S	WDTI N
	Register	Туре		r	rw	rh	r	rw	rwh	rw
всн	WDTREL Reset: 00 <sub>H</sub>	Bit Field				WDT	ΓREL			
	Watchdog Timer Reload Register	Туре				r	W			
вD <sub>Н</sub>	WDTWINB Reset: 00 <sub>H</sub>	Bit Field				WDT	WINB			
	Watchdog Window-Boundary Count Register	Туре		•		r	w		•	



Table 8 WDT Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
BE <sub>H</sub>	WDTL Reset: 00 <sub>H</sub>	Bit Field	d WDT							
	Watchdog Timer Register Low	Туре	rh							
BF <sub>H</sub>	WDTH Reset: 00 <sub>H</sub>	Bit Field	WDT							
	Watchdog Timer Register High	Туре	rh							

## 3.2.4.6 Port Registers

The Port SFRs can be accessed in the standard memory area (RMAP = 0).

 Table 9
 Port Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 0			•	•		•			
B2 <sub>H</sub>	PORT_PAGE Reset: 00H	Bit Field	C	)P	ST	NR	0		PAGE	
	Page Register	Туре	\	N	V	N	r		rw	
RMAP =	= 0, PAGE 0				•		•			
80 <sub>H</sub>	P0_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Data Register	Туре	rw	rw						
86 <sub>H</sub>	P0_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Direction Register	Туре	rw	rw						
90 <sub>H</sub>	P1_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Data Register	Туре	rw	rw						
91 <sub>H</sub>	P1_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Direction Register	Туре	rw	rw						
92 <sub>H</sub>	P5_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Data Register	Туре	rw	rw						
93 <sub>H</sub>	P5_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Direction Register	Туре	rw	rw						
A0 <sub>H</sub>	P2_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Data Register	Туре	rw	rw						
A1 <sub>H</sub>	P2_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Direction Register	Туре	rw	rw						
во <sub>Н</sub>	P3_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Data Register	Туре	rw	rw						
В1 <sub>Н</sub>	P3_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Direction Register	Туре	rw	rw						
C8H	P4_DATA Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Data Register	Туре	rw	rw						
C9H	P4_DIR Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Direction Register	Туре	rw	rw						



 Table 9
 Port Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 0, PAGE 1	<u> </u>		l	<u> </u>	<u> </u>	<u> </u>	<u> </u>		
80 <sub>H</sub>	P0_PUDSEL Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
86 <sub>H</sub>	P0_PUDEN Reset: C4 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1_PUDSEL Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
91 <sub>H</sub>	P1_PUDEN Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
92 <sub>H</sub>	P5_PUDSEL Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
93 <sub>H</sub>	P5_PUDEN Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
A0 <sub>H</sub>	P2_PUDSEL Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
A1 <sub>H</sub>	P2_PUDEN Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
во <sub>Н</sub>	P3_PUDSEL Reset: BF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
B1 <sub>H</sub>	P3_PUDEN Reset: 40H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
C8H	P4_PUDSEL Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Pull-Up/Pull-Down Select Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
C9H	P4_PUDEN Reset: 04 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Pull-Up/Pull-Down Enable Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
RMAP =	= 0, PAGE 2	.1			I.	I.	I.	I.	I	
80 <sub>H</sub>	P0_ALTSEL0 Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Alternate Select 0 Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
86 <sub>H</sub>	P0_ALTSEL1 Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Alternate Select 1 Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1_ALTSEL0 Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Alternate Select 0 Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
91 <sub>H</sub>	P1_ALTSEL1 Reset: 00 <sub>H</sub> P1 Alternate Select 1 Register	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
		Туре	rw	rw	rw	rw	rw	rw	rw	rw
92 <sub>H</sub>	P5_ALTSEL0 Reset: 00 <sub>H</sub> P5 Alternate Select 0 Register	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	2 3.001 0 1.09.0101	Type	rw	rw	rw	rw	rw	rw	rw	rw



Table 9 Port Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
93 <sub>H</sub>	P5_ALTSEL1 Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Alternate Select 1 Register	Туре	rw							
во <sub>Н</sub>	P3_ALTSEL0 Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Alternate Select 0 Register	Туре	rw							
В1 <sub>Н</sub>	P3_ALTSEL1 Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Alternate Select 1 Register	Type	rw							
C8 <sub>H</sub>	P4_ALTSEL0 Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Alternate Select 0 Register	Type	rw							
C9 <sub>H</sub>	P4_ALTSEL1 Reset: 00H	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Alternate Select 1 Register	Type	rw							
RMAP =	= 0, PAGE 3									
80 <sub>H</sub>	P0_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P0 Open Drain Control Register	Type	rw							
90 <sub>H</sub>	P1_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P1 Open Drain Control Register	Type	rw							
92 <sub>H</sub>	P5_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P5 Open Drain Control Register	Туре	rw							
во <sub>Н</sub>	P3_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Open Drain Control Register	Туре	rw							
C8 <sub>H</sub>	P4_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P4 Open Drain Control Register	Туре	rw							

# 3.2.4.7 ADC Registers

The ADC SFRs can be accessed in the standard memory area (RMAP = 0).

Table 10 ADC Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	0										
D1 <sub>H</sub>	ADC_PAGE Reset: 00H	Bit Field	О	P	ST	NR	0		PAGE		
	Page Register	Туре	١	V	١	V	r		rw		
RMAP =	: 0, PAGE 0								100		
CAH	ADC_GLOBCTR Reset: 30H	Bit Field	ANON	DW	C <sup>-</sup>	ГС		(			
	Global Control Register	Туре	rw	rw	r	W			ſ		
СВН	ADC_GLOBSTR Reset: 00 <sub>H</sub> Global Status Register	Bit Field	(	)		CHNR		0	SAMP LE	BUSY	
		Туре		r		rh		r	rh	rh	
ccH	ADC_PRAR Reset: 00 <sub>H</sub> Priority and Arbitration Register	Bit Field	ASEN 1	ASEN 0	0	ARBM	CSM1	PRIO1	PRIO0		
		Туре	rw	rw	r	rw	rw	rw	rw rw rv		



 Table 10
 ADC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0				
CDH	ADC_LCBR Reset: B7H	Bit Field		BOL	IND1			BOL	JND0					
	Limit Check Boundary Register	Туре		r	W			r	w					
CEH	ADC_INPCR0 Reset: 00H	Bit Field				S <sup>-</sup>	тс		BOUNDO  rw  ETRSELO  rw  RESRSE  rw  RESRSE  rw  RESRSE  rw  RESRSE  rw  RESRSE  rw  RESRSE					
	Input Class 0 Register	Туре				r	W							
CF <sub>H</sub>	ADC_ETRCR Reset: 00 <sub>H</sub> External Trigger Control	Bit Field	SYNE N1	SYNE N0		ETRSEL1	ĺ		ETRSELO	)				
	Register	Туре	rw	rw		rw			rw					
RMAP =	= 0, PAGE 1													
CAH	ADC_CHCTR0 Reset: 00H	Bit Field	0		LCC		(	)	RESI	RSEL				
	Channel Control Register 0	Type	r		rw			r	r	W				
СВН	ADC_CHCTR1 Reset: 00H	Bit Field	0		LCC		(	כ	RESI	RSEL				
	Channel Control Register 1	Туре	r	rw		rw		rw		rw		r	r	W
CCH	ADC_CHCTR2 Reset: 00H	Bit Field	0	LCC		LCC		LCC		LCC		כ	RESI	RSEL
	Channel Control Register 2	Type	r		rw			r	r	W				
CDH	ADC_CHCTR3 Reset: 00H	Bit Field	0		LCC		(	כ	RESI	RSEL				
	Channel Control Register 3	Туре	r		rw			r	r	W				
CEH	ADC_CHCTR4 Reset: 00H	Bit Field	0		LCC		(		RESRSEL rw RESRSEL rw RESRSEL rw					
	Channel Control Register 4	Туре	r		rw			r	r	W				
CF <sub>H</sub>	ADC_CHCTR5 Reset: 00H	Bit Field	0		LCC		(	rw RESRSE		RSEL				
	Channel Control Register 5	Type	r		rw			r	RESRSI					
D2 <sub>H</sub>	ADC_CHCTR6 Reset: 00H	Bit Field	0		LCC		(	כ	RESI	RSEL				
	Channel Control Register 6	Type	r		rw			r	r	W				
D3 <sub>H</sub>	ADC_CHCTR7 Reset: 00 <sub>H</sub>	Bit Field	0		LCC		(	)	RESI	RSEL				
	Channel Control Register 7	Туре	r		rw			r	r	W				
RMAP =	= 0, PAGE 2													
CAH	ADC_RESR0L Reset: 00H	Bit Field	RES	SULT	0	VF	DRC		CHNR					
	Result Register 0 Low	Туре	r	h	r	rh	rh		rh					
СВН	ADC_RESR0H Reset: 00H	Bit Field				RES	SULT							
	Result Register 0 High	Type				r	h							
CC <sub>H</sub>	ADC_RESR1L Reset: 00H	Bit Field	RES	SULT	0	VF	DRC		CHNR					
	Result Register 1 Low	Туре	rh r rh rh		rh		rh							
CDH	ADC_RESR1H Reset: 00H	Bit Field	Bit Field RESULT		Field RESULT									
	Result Register 1 High	Туре				r	h							
CEH	ADC_RESR2L Reset: 00H	Bit Field	RES	SULT 0 VF DRC		CHNR								
	Result Register 2 Low	Туре	rh r rh			rh	rh							
CF <sub>H</sub>	ADC_RESR2H Reset: 00H	Bit Field				RES	SULT							
	Result Register 2 High	Туре				r	h							
D2 <sub>H</sub>	ADC_RESR3L Reset: 00H	Bit Field	RES	SULT	0 VF		DRC		CHNR					
	Result Register 3 Low	Туре	r	h .	r	rh	rh		rh					



# Table 10 ADC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
	ADC_RESR3H Reset: 00 <sub>H</sub>	Bit Field	•				ULT		<u> </u>	
D3 <sub>H</sub>	Result Register 3 High									
DMAD -	 = 0, PAGE 3	Туре				- 1	h			
	ADC_RESRA0L Reset: 00 <sub>H</sub>	Bit Field		RESULT		VF	DRC		CHNR	
CAH	Result Register 0, View A Low	Type		rh		rh	rh		rh	
CB	ADC RESPANH Paget: 00	Bit Field		111		L	ULT		111	
СВН	ADC_RESRA0H Reset: 00 <sub>H</sub> Result Register 0, View A High									
CC	ADC_RESRA1L Reset: 00H	Type Bit Field		RESULT		VF	h DRC		CHNR	
сс <sub>Н</sub>	Result Register 1, View A Low	Туре		rh		rh	rh		rh	
CD	ADC RESPAIN Paget: 00	Bit Field		111			L ''' SULT		111	
CDH	ADC_RESRA1H Reset: 00 <sub>H</sub> Result Register 1, View A High						h			
C.E.	ADC BESDAGE Books 00	Type Bit Field		RESULT		VF	DRC		CHNR	
CEH	ADC_RESRA2L Reset: 00 <sub>H</sub> Result Register 2, View A Low									
	ADC DECDAOU Breet 00	Type		rh		rh	rh		rh	
CF <sub>H</sub>	ADC_RESRA2H Reset: 00 <sub>H</sub> Result Register 2, View A High	Bit Field					BULT			
D0	ADO DEODAGI. Barati 00	Type	rh				DRC CHNR			
D2 <sub>H</sub>	ADC_RESRA3L Reset: 00 <sub>H</sub> Result Register 3, View A Low	Bit Field		RESULT		VF				
	4D0 DE0D40U D	Type		rh		rh		rh rh		
D3 <sub>H</sub>	ADC_RESRA3H Reset: 00 <sub>H</sub> Result Register 3, View A High	Bit Field					SULT			
		Туре				r	h			
	= 0, PAGE 4	I					I			I
CA <sub>H</sub>	ADC_RCR0 Reset: 00 <sub>H</sub> Result Control Register 0	Bit Field	VFCT R	WFR	0	IEN		0		DRCT R
		Туре	rw	rw	r	rw		r		rw
СВ <sub>Н</sub>	ADC_RCR1 Reset: 00 <sub>H</sub> Result Control Register 1	Bit Field	VFCT R	WFR	0	IEN		0		DRCT R
		Туре	rw	rw	r	rw		r		rw
cc <sub>H</sub>	ADC_RCR2 Reset: 00 <sub>H</sub> Result Control Register 2	Bit Field	VFCT R	WFR	0	IEN		0		DRCT R
		Туре	rw	rw	r	rw		r		rw
CDH	ADC_RCR3 Reset: 00 <sub>H</sub> Result Control Register 3	Bit Field	VFCT R	WFR	0	IEN		0		DRCT R
		Туре	rw	rw	r	rw		r		rw
CEH	ADC_VFCR Reset: 00H	Bit Field		(	0		VFC3	VFC2	VFC1	VFC0
	Valid Flag Clear Register	Туре			r		W	w	w	w
RMAP =	= 0, PAGE 5									
CA <sub>H</sub>	ADC_CHINFR Reset: 00 <sub>H</sub> Channel Interrupt Flag Register	Bit Field	CHINF 7	CHINF 6	CHINF 5	CHINF 4	CHINF 3	CHINF 2	CHINF 1	CHINF 0
		Туре	rh	rh	rh	rh	rh	rh	rh	rh
СВН	ADC_CHINCR Reset: 00 <sub>H</sub> Channel Interrupt Clear Register	Bit Field	CHINC 7	CHINC 6	CHINC 5	CHINC 4	CHINC 3	CHINC 2	CHINC 1	CHINC 0
		Туре	w	w	w	w	w	w	w	w



 Table 10
 ADC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0		
сс <sub>Н</sub>	ADC_CHINSR Reset: 00 <sub>H</sub> Channel Interrupt Set Register	Bit Field	CHINS 7	CHINS 6	CHINS 5	CHINS 4	CHINS 3	CHINS 2	CHINS 1	CHINS 0		
		Туре	W	w	W	w	W	W	w	w		
CDH	ADC_CHINPR Reset: 00 <sub>H</sub> Channel Interrupt Node Pointer	Bit Field	CHINP 7	CHINP 6	CHINP 5	CHINP 4	CHINP 3	CHINP 2	CHINP 1	CHINP 0		
	Register	Type	rw									
CEH	ADC_EVINFR Reset: 00 <sub>H</sub> Event Interrupt Flag Register	Bit Field	EVINF 7	EVINF 6	EVINF 5	EVINF 4	(	)	EVINF 1	EVINF 0		
		Type	rh	rh	rh	rh		r	rh	rh		
CF <sub>H</sub>	ADC_EVINCR Reset: 00 <sub>H</sub> Event Interrupt Clear Flag	Bit Field	EVINC 7	EVINC 6	EVINC 5	EVINC 4	(	)	EVINC 1	EVINC 0		
	Register	Туре	w	w	w	w		r	w	w		
D2 <sub>H</sub>	ADC_EVINSR Reset: 00 <sub>H</sub> Event Interrupt Set Flag Register	Bit Field	EVINS 7	EVINS 6	EVINS 5	EVINS 4	(	)	EVINS 1	EVINS 0		
		Туре	W	W	W	W		r				
D3 <sub>H</sub>	ADC_EVINPR Reset: 00 <sub>H</sub> Event Interrupt Node Pointer	Bit Field	EVINP 7	EVINP 6	EVINP 5	EVINP 4	(	)	EVINP 1	EVINP 0		
	Register	Type	rw	rw	rw	rw		r	rw	rw		
RMAP =	= 0, PAGE 6											
CAH	ADC_CRCR1 Reset: 00 <sub>H</sub>	Bit Field	CH7	CH6	CH5	CH4		(	)			
	Conversion Request Control Register 1	Туре	rwh	rwh	rwh	rwh		-	r			
СВН	ADC_CRPR1 Reset: 00H	Bit Field	CHP7	CHP6	CHP5	CHP4		(	)			
	Conversion Request Pending Register 1	Туре	rwh	rwh	rwh	rwh		-	r			
cc <sub>H</sub>	ADC_CRMR1 Reset: 00 <sub>H</sub> Conversion Request Mode	Bit Field	Rsv	LDEV	CLRP ND	SCAN	ENSI	ENTR	0	ENGT		
	Register 1	Туре	r	W	W	rw	rw	rw	r	rw		
CDH	ADC_QMR0 Reset: 00 <sub>H</sub> Queue Mode Register 0	Bit Field	CEV	TREV	FLUS H	CLRV	0	ENTR	0	ENGT		
		Туре	W	w	W	w	r	rw	r	rw		
CEH	ADC_QSR0 Reset: 20 <sub>H</sub> Queue Status Register 0	Bit Field	Rsv	0	EMPT Y	EV	(	)	FI	LL		
		Туре	r	r	rh	rh		r rh				
CF <sub>H</sub>	ADC_Q0R0 Reset: 00H	Bit Field	EXTR	ENSI	RF	V	0	REQCHNR				
	Queue 0 Register 0	Туре	rh	rh	rh	rh	r	rh				
D2 <sub>H</sub>	ADC_QBUR0 Reset: 00H	Bit Field	EXTR	ENSI	RF	V	0	F	REQCHN	2		
	Queue Backup Register 0	Туре	rh	rh	rh	rh	r	rh				
D2 <sub>H</sub>	ADC_QINR0 Reset: 00H	Bit Field	EXTR	ENSI	RF	(	)	F	REQCHN	3		
	Queue Input Register 0	Туре	w	w	w		r		w			



## 3.2.4.8 Timer 2 Registers

The Timer 2 SFRs can be accessed in the standard memory area (RMAP = 0).

Table 11 T2 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	= 0			•	•			•	•		
C0H	T2_T2CON Reset: 00 <sub>H</sub> Timer 2 Control Register	Bit Field	TF2	EXF2	(	)	EXEN 2	TR2	C/T2	CP/ RL2	
		Туре	rwh	rwh		r	rw	rwh	rw	rw	
C1 <sub>H</sub>	T2_T2MOD Reset: 00 <sub>H</sub> Timer 2 Mode Register	Bit Field	T2RE GS	T2RH EN	EDGE SEL	PREN		T2PRE		DCEN	
		Туре	rw	rw	rw	rw	rw	rw	rw	rw	
C2 <sub>H</sub>	T2_RC2L Reset: 00H	Bit Field				R	C2				
	Timer 2 Reload/Capture Register Low	Туре				rv	vh				
СЗН	T2_RC2H Reset: 00H	Bit Field				R	C2				
	Timer 2 Reload/Capture Register High	Туре				rv	vh				
C4 <sub>H</sub>	T2_T2L Reset: 00H	Bit Field				TH	IL2				
	Timer 2 Register Low	Туре				rv	vh				
C5 <sub>H</sub>	T2_T2H Reset: 00H	Bit Field				TH	L2				
	Timer 2 Register High	Туре				rv	vh	า			

## 3.2.4.9 Timer 21 Registers

The Timer 21 SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 12 T21 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	= 1									•	
C0H	T21_T2CON Reset: 00 <sub>H</sub> Timer 2 Control Register	Bit Field	TF2	EXF2	(	)	EXEN 2	TR2	C/T2	CP/ RL2	
		Туре	rwh	rwh	ı	r	rw	rwh	rw	rw	
C1 <sub>H</sub>	T21_T2MOD Reset: 00 <sub>H</sub> Timer 2 Mode Register	Bit Field	T2RE GS	T2RH EN	EDGE SEL	PREN		T2PRE		DCEN	
		Туре	rw	rw	rw	rw	rw	rw	rw	rw	
C2 <sub>H</sub>	T21_RC2L Reset: 00H	Bit Field				R	C2				
	Timer 2 Reload/Capture Register Low	Туре				rv	vh				
C3 <sub>H</sub>	T21_RC2H Reset: 00H	Bit Field				R	C2				
	Timer 2 Reload/Capture Register High	Туре				rv	vh				
C4 <sub>H</sub>	T21_T2L Reset: 00H	Bit Field				TH	L2				
	Timer 2 Register Low	Туре				rv	vh	h			



Table 12 T21 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
C5 <sub>H</sub>	T21_T2H Reset: 00 <sub>H</sub>	Bit Field				TH	IL2			
	Timer 2 Register High	Туре				rv	vh			

## 3.2.4.10 CCU6 Registers

The CCU6 SFRs can be accessed in the standard memory area (RMAP = 0).

Table 13 CCU6 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0		
RMAP =	= 0	I	I		I		I	I		I		
A3 <sub>H</sub>	CCU6_PAGE Reset: 00H	Bit Field	С	)P	ST	NR	0		PAGE			
	Page Register	Туре	\	N	\	N	r		rw			
RMAP =	= 0, PAGE 0											
9A <sub>H</sub>	CCU6_CC63SRL Reset: 00H	Bit Field				CC6	3SL					
	Capture/Compare Shadow Register for Channel CC63 Low	Туре				r	w					
9B <sub>H</sub>	CCU6_CC63SRH Reset: 00 <sub>H</sub>	Bit Field				CC6	3SH					
	Capture/Compare Shadow Register for Channel CC63 High	Туре				r	w					
9CH	CCU6_TCTR4L Reset: 00 <sub>H</sub> Timer Control Register 4 Low	Bit Field	T12 STD	T12 STR		0	DT RES	T12 RES				
		Туре	W	W		r	W	W	W	w		
9DH	CCU6_TCTR4H Reset: 00 <sub>H</sub> Timer Control Register 4 High	Bit Field	T13 STD	T13 STR		0		T13 RES	T13 T13R T13			
		Туре	W	W		r		W	W	W		
9E <sub>H</sub>	CCU6_MCMOUTSL Reset: 00 <sub>H</sub> Multi-Channel Mode Output Shadow	Bit Field	STRM CM	0			MCI	MPS				
	Register Low	Туре	W	r			r	W				
<sup>9F</sup> H	CCU6_MCMOUTSH Reset: 00 <sub>H</sub> Multi-Channel Mode Output Shadow	Bit Field	STRH P	0		CURHS			EXPHS			
	Register High	Туре	W	r		rw			rw			
<sup>A4</sup> H	CCU6_ISRL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	RT12 PM	RT12 OM	RCC6 2F	RCC6 2R	RCC6 1F	RCC6 1R	RCC6 0F	RCC6 0R		
	Reset Register Low	Туре	W	W	W	W	W	W	W	W		
<sup>A5</sup> H	CCU6_ISRH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	RSTR	RIDLE	RWH E	RCHE	0	RTRP F	RT13 PM	RT13 CM		
	Reset Register High	Туре	w	w	w	w	r	W	W	w		
A6 <sub>H</sub>	CCU6_CMPMODIFL Reset: 00 <sub>H</sub> Compare State Modification Register	Bit Field	0	MCC6 3S		0		MCC6 2S	MCC6 1S	MCC6 0S		
	Low	Туре	r	W		r		W	W	w		
A7 <sub>H</sub>	CCU6_CMPMODIFH Reset: 00 <sub>H</sub> Compare State Modification Register	Bit Field	0	MCC6 3R		0		MCC6 2R	MCC6 1R	MCC6 0R		
	High	Туре	r	w		r		w	w	w		



# Table 13 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
FAH	CCU6_CC60SRL Reset: 00H	Bit Field				CC6	iosl			
	Capture/Compare Shadow Register for Channel CC60 Low	Туре				rv	vh			
FBH	CCU6_CC60SRH Reset: 00H	Bit Field				CC6	i0SH			
	Capture/Compare Shadow Register for Channel CC60 High	Туре				rv	vh			
FC <sub>H</sub>	CCU6_CC61SRL Reset: 00H	Bit Field				CC6	1SL			
	Capture/Compare Shadow Register for Channel CC61 Low	Туре				rv	vh			
FDH	CCU6_CC61SRH Reset: 00H	Bit Field				CC6	1SH			
	Capture/Compare Shadow Register for Channel CC61 High	Туре				rv	vh			
FE <sub>H</sub>	CCU6_CC62SRL Reset: 00 <sub>H</sub> Capture/Compare Shadow Register	Bit Field				CC6	S2SL			
	for Channel CC62 Low	Туре				rv	vh			
FFH	CCU6_CC62SRH Reset: 00H	Bit Field				CC6	2SH			
	Capture/Compare Shadow Register for Channel CC62 High	Туре				rv	vh			
RMAP =	= 0, PAGE 1									
9A <sub>H</sub>	CCU6_CC63RL Reset: 00H	Bit Field				CC6	3VL			
	Capture/Compare Register for Channel CC63 Low	Туре				r	h			
9B <sub>H</sub>	CCU6_CC63RH Reset: 00H	Bit Field				CC6	3VH			
	Capture/Compare Register for Channel CC63 High	Туре				r	h			
9CH	CCU6_T12PRL Reset: 00H	Bit Field				T12	PVL			
	Timer T12 Period Register Low	Туре				rv	vh			
9D <sub>H</sub>	CCU6_T12PRH Reset: 00 <sub>H</sub> Timer T12 Period Register High	Bit Field				T12	PVH			
	Timer 112 Feriod Register Flight	Туре				rv	vh			
9E <sub>H</sub>	CCU6_T13PRL Reset: 00 <sub>H</sub> Timer T13 Period Register Low	Bit Field				T13	PVL			
	Timer Te Fende Register 2011	Туре				rv	vh			
<sup>9F</sup> H	CCU6_T13PRH Reset: 00 <sub>H</sub> Timer T13 Period Register High	Bit Field				T13	PVH			
		Туре				rv	vh			
<sup>A4</sup> H	CCU6_T12DTCL Reset: 00 <sub>H</sub> Dead-Time Control Register for	Bit Field				D	ГМ			
	Timer T12 Low	Туре				r	W			
A5 <sub>H</sub>	CCU6_T12DTCH Reset: 00 <sub>H</sub> Dead-Time Control Register for	Bit Field	0	DTR2	DTR1	DTR0	0	DTE2	DTE1	DTE0
	Timer T12 High	Туре	r	rh	rh	rh	r	rw	rw	rw
A6 <sub>H</sub>	CCU6_TCTR0L Reset: 00 <sub>H</sub> Timer Control Register 0 Low	Bit Field	СТМ	CDIR	STE1	T12R	T12 PRE		T12CLK	
		Туре	rw	rh	rh	rh	rw		rw	
A7 <sub>H</sub>	CCU6_TCTR0H Reset: 00 <sub>H</sub> Timer Control Register 0 High	Bit Field	(	)	STE1	T13R	T13 PRE		T13CLK	
		Туре		r	rh	rh	rw		rw	
FA <sub>H</sub>	CCU6_CC60RL Reset: 00H	Bit Field				CC6	60VL			
	Capture/Compare Register for Channel CC60 Low	Туре				r	h			



# Table 13 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3 2 1 0				
FB <sub>H</sub>	CCU6_CC60RH Reset: 00 <sub>H</sub>	Bit Field				CC6	0VH				
''	Capture/Compare Register for Channel CC60 High	Туре				r	h				
FC <sub>H</sub>	CCU6_CC61RL Reset: 00H	Bit Field				CC6	1VL				
	Capture/Compare Register for Channel CC61 Low	Туре				r	h				
FD <sub>H</sub>	CCU6_CC61RH Reset: 00H	Bit Field				CC6	1VH				
	Capture/Compare Register for Channel CC61 High	Туре				r	h				
FE <sub>H</sub>	CCU6_CC62RL Reset: 00H	Bit Field				CC6	2VL				
	Capture/Compare Register for Channel CC62 Low	Туре				r	h				
FF <sub>H</sub>	CCU6_CC62RH Reset: 00H	Bit Field				CC6	2VH				
	Capture/Compare Register for Channel CC62 High	Туре				r	h				
RMAP =	= 0, PAGE 2	ı									
9A <sub>H</sub>	CCU6_T12MSELL Reset: 00 <sub>H</sub>	Bit Field		MSI	EL61			MSE	EL60		
	T12 Capture/Compare Mode Select Register Low	Туре		r	W			rw MSEL62			
9B <sub>H</sub>	CCU6_T12MSELH Reset: 00H	Bit Field	DBYP		HSYNC			MSEL62			
	T12 Capture/Compare Mode Select Register High	Туре	rw		rw			rw			
9CH	CCU6_IENL Reset: 00H	Bit Field	ENT1	ENT1	ENCC	ENCC	ENCC	ENCC ENCC ENCC E			
	Capture/Compare Interrupt Enable Register Low		2 PM	2 OM	62F	62R	61F	61R	60R		
		Туре	rw	rw	rw	rw	rw	rw rw rw r			
9D <sub>H</sub>	CCU6_IENH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Enable	Bit Field	EN STR	EN IDLE	EN WHE	EN CHE	0	EN TRPF	ENT1 3PM	ENT1 3CM	
	Register High	Туре	rw	rw	rw	rw	r	rw	rw	rw	
9E <sub>H</sub>	CCU6_INPL Reset: 40 <sub>H</sub>	Bit Field	INP	CHE	INPO	CC62	INPO	CC61	INPO	CC60	
	Capture/Compare Interrupt Node Pointer Register Low	Туре	r	W	r	W	r	W	r	W	
9F <sub>H</sub>	CCU6_INPH Reset: 39 <sub>H</sub>	Bit Field	(	)	INP	T13	INP	T12	INP	ERR	
	Capture/Compare Interrupt Node Pointer Register High	Туре		r	r	W	r	w	r	W	
<sup>A4</sup> H	CCU6_ISSL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	ST12 PM	ST12 OM	SCC6 2F	SCC6 2R	SCC6 1F	SCC6 1R	SCC6 0F	SCC6 0R	
	Set Register Low	Туре	W	w	w	W	w	w	W	W	
A5 <sub>H</sub>	CCU6_ISSH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	SSTR	SIDLE	SWHE	SCHE	SWH C	STRP F	ST13 PM	ST13 CM	
	Set Register High	Туре	w	w	w	w	w	w	w	w	
A6 <sub>H</sub>	CCU6_PSLR Reset: 00 <sub>H</sub> Passive State Level Register	Bit Field	PSL63	0			PSL				
	j ,	Туре	rwh	r			rwh				
<sup>A7</sup> H	CCU6_MCMCTR Reset: 00 <sub>H</sub> Multi-Channel Mode Control Register	Bit Field	(	)	SW	SYN	0 SWSEL				
	,	Туре		r I		w I	r rw				
FA <sub>H</sub>	CCU6_TCTR2L Reset: 00 <sub>H</sub> Timer Control Register 2 Low	Bit Field	0	T13	TED					T12 SSC	
		Туре	r	r	W		rw		rw	rw	



Table 13 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0		
FB <sub>H</sub>	CCU6_TCTR2H Reset: 00H	Bit Field			0		T13F	RSEL	T12F	RSEL		
	Timer Control Register 2 High	Туре			r		r	W	r	W		
FC <sub>H</sub>	CCU6_MODCTRL Reset: 00 <sub>H</sub> Modulation Control Register Low	Bit Field	MCM EN	0			T12M	ODEN				
		Туре	rw	r			r	W				
FD <sub>H</sub>	CCU6_MODCTRH Reset: 00 <sub>H</sub> Modulation Control Register High	Bit Field	ECT1 3O	0			T13M	ODEN				
		Туре	rw	r			r	w				
FE <sub>H</sub>	CCU6_TRPCTRL Reset: 00 <sub>H</sub> Trap Control Register Low	Bit Field			0			TRPM 2	TRPM 1	TRPM 0		
		Туре		1	r			rw	rw	rw		
FFH	CCU6_TRPCTRH Reset: 00 <sub>H</sub> Trap Control Register High	Bit Field	TRPP EN	TRPE N13			TRI	PEN				
		Туре	rw	rw			r	rw  MCMP  rh				
	= 0, PAGE 3	1	ı	1	1			rw MCMP				
<sup>9A</sup> H	CCU6_MCMOUTL Reset: 00 <sub>H</sub> Multi-Channel Mode Output Register	Bit Field	0	R			MC	MCMP				
	Low	Type	r	rh			r	rh				
9B <sub>H</sub>	CCU6_MCMOUTH Reset: 00H	Bit Field	(	0		CURH		EXPH				
	Multi-Channel Mode Output Register High	Туре		r		rh		rh  EXPH  rh  1 ICC61 ICC60 ICC  R F R				
9CH	CCU6_ISL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	T12 PM	T12 OM	ICC62 F	ICC62 R	ICC61 F	ICC61 ICC60 F F				
	Register Low	Туре	rh	rh	rh	rh	rh	rh	rh			
9DH	CCU6_ISH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status Register High	Bit Field	STR	IDLE	WHE	CHE	TRPS	TRPF	T13 PM	T13 CM		
	Register riigii	Туре	rh	rh	rh	rh	rh	rh	rh	rh		
9E <sub>H</sub>	CCU6_PISEL0L Reset: 00 <sub>H</sub> Port Input Select Register 0 Low	Bit Field	IST	RP	ISC	C62	ISC	C61	ISC	C60		
	Fort input Select Negister o Low	Туре	r	w	r	W	r	W	r	W		
9F <sub>H</sub>	CCU6_PISEL0H Reset: 00 <sub>H</sub> Port Input Select Register 0 High	Bit Field	IST1	2HR	ISP	OS2	ISP	OS1	ISP	OS0		
		Туре	r	W	r	W	r	W	r	W		
<sup>A4</sup> H	CCU6_PISEL2 Reset: 00 <sub>H</sub> Port Input Select Register 2	Bit Field			(	0			IST1	3HR		
		Туре				r			r	W		
FA <sub>H</sub>	CCU6_T12L Reset: 00 <sub>H</sub> Timer T12 Counter Register Low	Bit Field				T12	CVL					
	_	Туре					wh					
FB <sub>H</sub>	CCU6_T12H Reset: 00 <sub>H</sub> Timer T12 Counter Register High	Bit Field				T12	CVH	CVH				
		Туре					h					
FC <sub>H</sub>	CCU6_T13L Reset: 00 <sub>H</sub> Timer T13 Counter Register Low	Bit Field					CVL					
		Туре					vh	h				
FDH	CCU6_T13H Reset: 00 <sub>H</sub> Timer T13 Counter Register High	Bit Field					CVH					
		Туре				rv	vh					



Table 13 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
FE <sub>H</sub>	CCU6_CMPSTATL Reset: 00 <sub>H</sub> Compare State Register Low	Bit Field	0	CC63 ST	CC POS2	CC POS1	CC POS0	CC62 ST	CC61 ST	CC60 ST
		Туре	r	rh	rh	rh	rh	rh	rh	rh
FF <sub>H</sub>	CCU6_CMPSTATH Reset: 00 <sub>H</sub> Compare State Register High	Bit Field	T13IM	COUT 63PS	COUT 62PS	CC62 PS	COUT 61PS	CC61 PS	COUT 60PS	CC60 PS
		Туре	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh

## 3.2.4.11 UART1 Registers

The UART1 SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 14 UART1 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0			
RMAP =	:1			•		•	•	•					
C8H	SCON Reset: 00 <sub>H</sub>	Bit Field	SM0	SM1	SM2	REN	TB8	RB8	TI	RI			
	Serial Channel Control Register	Туре	rw	rw	rw	rw	rw	rwh	rwh	rwh			
C9H	SBUF Reset: 00 <sub>H</sub>	Bit Field				V	AL						
	Serial Data Buffer Register	Туре				rv	vh						
CAH	BCON Reset: 00 <sub>H</sub>	Bit Field	0						BRPRE R				
	Baud Rate Control Register	Туре	r rw						rw rw				
СВН	BG Reset: 00 <sub>H</sub>	Bit Field				BR_V	ALUE						
	Baud Rate Timer/Reload Register	Туре				rv	vh						
ССН	FDCON Reset: 00 <sub>H</sub>	Bit Field			0			NDOV	FDM	FDEN			
	Fractional Divider Control Register	Туре			r			rwh	rw	rw			
CDH	FDSTEP Reset: 00 <sub>H</sub>	Bit Field				ST	ΈP						
	Fractional Divider Reload Register	Туре	rw										
CEH	FDRES Reset: 00 <sub>H</sub>	Bit Field				RES	SULT						
	Fractional Divider Result Register	Туре				r	h						



## 3.2.4.12 SSC Registers

The SSC SFRs can be accessed in the standard memory area (RMAP = 0).

 Table 15
 SSC Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =		I		1				1		
A9 <sub>H</sub>	SSC_PISEL Reset: 00H	Bit Field			0			CIS	SIS	MIS
	Port Input Select Register	Туре			r			rw	rw	rw
$AA_H$	SSC_CONL Reset: 00H	Bit Field	LB	PO	PH	НВ		В	М	
	Control Register Low Programming Mode	Туре	rw	rw	rw	rw		r	W	
AAH	SSC_CONL Reset: 00H	Bit Field		(	)			В	SC .	
	Control Register Low Operating Mode	Туре			r		rh			
AB <sub>H</sub>	SSC_CONH Reset: 00 <sub>H</sub> Control Register High Programming Mode	Bit Field	EN	MS	0	AREN	BEN	PEN	REN	TEN
		Туре	rw	rw	r	rw	rw	rw	rw	rw
AB <sub>H</sub>	SSC_CONH Reset: 00H	Bit Field	EN	MS	0	BSY	BE	PE	RE	TE
	Control Register High Operating Mode	Туре	rw	rw	r	rh	rwh	rwh	rwh	rwh
AC <sub>H</sub>	SSC_TBL Reset: 00H	Bit Field	TB_VALUE							
	Transmitter Buffer Register Low	Туре				r	N			
$AD_H$	SSC_RBL Reset: 00H	Bit Field				RB_V	ALUE			
	Receiver Buffer Register Low	Туре	rh							
ΑΕ <sub>Η</sub>	SSC_BRL Reset: 00H	Bit Field				BR_V	ALUE			
	Baud Rate Timer Reload Register Low	Туре	rw							
AF <sub>H</sub>	SSC_BRH Reset: 00H	Bit Field				BR_V	ALUE			
	Baud Rate Timer Reload Register High	Туре	rw							

## 3.2.4.13 MultiCAN Registers

The MultiCAN SFRs can be accessed in the standard memory area (RMAP = 0).

Table 16 CAN Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	: 0									
D8 <sub>H</sub>	ADCON Reset: 00 <sub>H</sub>	Bit Field	V3	V2	V1	V0	AU	IAD	BSY	RWEN
	CAN Address/Data Control Register	Туре	rw	rw	rw	rw	rw		rh	rw
D9 <sub>H</sub>	ADL Reset: 00 <sub>H</sub>	Bit Field	CA9	CA8	CA7	CA6	CA5	CA4	CA3	CA2
	CAN Address Register Low	Туре	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh
DA <sub>H</sub>	ADH Reset: 00 <sub>H</sub>	Bit Field	·	(	)		CA13	CA12	CA11	CA10
	CAN Address Register High	Туре			r		rwh	rwh	rwh	rwh



Table 16 CAN Register Overview (cont'd)

Addr	Register Nam	ne	Bit	7	6	5	4	3	2	1	0	
DB <sub>H</sub>		Reset: 00 <sub>H</sub>	Bit Field		CD							
	CAN Data Register	r 0	Туре				rv	vh				
DCH		Reset: 00 <sub>H</sub>	Bit Field	Field CD								
	CAN Data Register	CAN Data Register 1		rwh								
DDH	DATA2	Reset: 00 <sub>H</sub>	Bit Field	Bit Field CD								
	CAN Data Register	r 2	Туре	rwh								
DEH	DATA3 Reset: 00 <sub>H</sub>		Bit Field	Field CD								
	CAN Data Register	r 3	Туре				rv	vh				

### 3.2.4.14 OCDS Registers

The OCDS SFRs can be accessed in the mapped memory area (RMAP = 1).

Table 17 OCDS Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	= 1									
E9 <sub>H</sub>	H MMCR2 Reset: 1U <sub>H</sub> Monitor Mode Control 2	Bit Field	STMO DE	EXBC	DSUS P	MBCO N	ALTDI	MMEP	MMOD E	JENA
	Register	Туре	rw	rw	rw	rwh	rw	rwh	rh	rh
F1 <sub>H</sub>	MMCR Reset: 00 <sub>H</sub> Monitor Mode Control Register	Bit Field	MEXIT _P	MEXIT	0	MSTE P	MRAM S_P	MRAM S	TRF	RRF
		Туре	W	rwh	r	rw	W	rwh	rh	rh
F2 <sub>H</sub>	MMSR Reset: 00 <sub>H</sub> Monitor Mode Status Register	Bit Field	MBCA M	MBCIN	EXBF	SWBF	HWB3 F	HWB2 F	HWB1 F	HWB0 F
		Туре	rw	rwh	rwh	rwh	rwh	rwh	rwh	rwh
F3 <sub>H</sub>	F3 <sub>H</sub> MMBPCR Reset: 00 <sub>H</sub> Breakpoints Control Register		SWBC	HW	VB3C HWB2C		B2C	HWB1 C	HWB0C	
		Туре	rw rw		rw		rw	rw		
F4 <sub>H</sub>	MMICR Reset: 00 <sub>H</sub> Monitor Mode Interrupt Control	Bit Field	DVEC T	DRET R	COMR ST	MSTS EL	MMUI E_P	MMUI E	RRIE_ P	RRIE
	Register	Туре	rwh	rwh	rwh	rh	W	rw	w	rw
F5 <sub>H</sub>	MMDR Reset: 00 <sub>H</sub>	Bit Field	MMRR							
	Monitor Mode Data Transfer Register Receive	Туре				r	h			
F6 <sub>H</sub>	HWBPSR Reset: 00 <sub>H</sub> Hardware Breakpoints Select	Bit Field		0	BPSEL BPSE			SEL		
	Register	Туре		r		w		r	W	
F7 <sub>H</sub>	HWBPDR Reset: 00 <sub>H</sub>	Bit Field				HWI	ЗРхх			
	Hardware Breakpoints Data Register	Туре	rw							
EBH	MMWR1 Reset: 00 <sub>H</sub>	Bit Field				MM	WR1			
	Monitor Work Register 1	Туре				r	W			



## Table 17 OCDS Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
ECH	MMWR2 Reset: 00 <sub>H</sub>	Bit Field	MMWR2							
	Monitor Work Register 2	Туре				r	w			

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### 3.3 Flash Memory

The Flash memory provides an embedded user-programmable non-volatile memory, allowing fast and reliable storage of user code and data. It is operated from a single 2.5 V supply from the Embedded Voltage Regulator (EVR) and does not require additional programming or erasing voltage. The sectorization of the Flash memory allows each sector to be erased independently.

#### **Features**

- In-System Programming (ISP) via UART
- In-Application Programming (IAP)
- Error Correction Code (ECC) for dynamic correction of single-bit errors
- Background program and erase operations for CPU load minimization
- Support for aborting erase operation
- Minimum program width<sup>1)</sup> of 32-byte for D-Flash and 64-byte for P-Flash
- 1-sector minimum erase width
- 1-byte read access
- Flash is delivered in erased state (read all zeros)
- Operating supply voltage: 2.5 V ± 7.5 %
- Read access time:  $3 \times t_{CCLK} = 150 \text{ ns}^{2}$
- Program time:  $248256 / f_{SYS} = 3.1 \text{ ms}^{3)}$
- Erase time:  $9807360 / f_{SYS} = 123 \text{ ms}^{3)}$

<sup>1)</sup> P-Flash: 64-byte wordline can only be programmed once, i.e., one gate disturb allowed. D-Flash: 32-byte wordline can be programmed twice, i.e., two gate disturbs allowed.

<sup>2)</sup> Values shown here are typical values.  $f_{\rm sys}$  = 80 MHz ± 7.5% ( $f_{\rm CCLK}$  = 20 MHz ± 7.5 %) is the maximum frequency range for Flash read access.

<sup>3)</sup> Values shown here are typical values.  $f_{\rm sys} = 80 \, \rm MHz \pm 7.5\%$  is the only frequency range for Flash programming and erasing.  $f_{\rm sysmin}$  is used for obtaining the worst case timing.



**Table 18** shows the Flash data retention and endurance targets.

Table 18 Flash Data Retention and Endurance (Operating Conditions apply)

Retention	Endurance <sup>1)</sup>	Si	Remarks	
		<i>T</i> <sub>A</sub> = -40 to 125 °C	<i>T</i> <sub>A</sub> = 125 to 150 °C	
Program F	lash			•
20 years	1,000 cycles	up to 32 Kbytes <sup>2)</sup>		for 32-Kbyte Variant
20 years	1,000 cycles	up to 24 Kbytes <sup>2)</sup>		for 24-Kbyte Variant
Data Flash				•
20 years	1,000 cycles <sup>3)</sup>	4 Kbytes	1 Kbyte	
5 years	10,000 cycles <sup>3)</sup>	1 Kbyte	256 bytes	
2 years	70,000 cycles <sup>3)</sup>	512 bytes	128 bytes	
2 years	100,000 cycles <sup>3)</sup>	128 bytes 32 bytes		

- 1) One cycle refers to the programming of all wordlines in a sector and erasing of sector. The Flash endurance data specified in **Table 18** is valid only if the following conditions are fulfilled:
  - the maximum number of erase cycles per Flash sector must not exceed 100,000 cycles.
  - the maximum number of erase cycles per Flash bank must not exceed 300,000 cycles.
  - the maximum number of program cycles per Flash bank must not exceed 2,500,000 cycles.
- 2) If no Flash is used for data, the Program Flash size can be up to the maximum Flash size available in the device variant. Having more Data Flash will mean less Flash is available for Program Flash.
- 3) For  $T_A$  = 125 to 150°C, refers to programming of second 8 bytes (bytes 8 to 15) per WL.

#### 3.3.1 Flash Bank Sectorization

The SAL-XC886 product family offers Flash devices with either 24 Kbytes or 32 Kbytes of embedded Flash memory. Each Flash device consists of Program Flash (P-Flash) and Data Flash (D-Flash) bank(s) with different sectorization shown in **Figure 10**. Both types can be used for code and data storage. The label "Data" neither implies that the D-Flash is mapped to the data memory region, nor that it can only be used for data storage. It is used to distinguish the different Flash bank sectorizations.

The 32-Kbyte Flash device consists of 6 P-Flash and 2 D-Flash banks, while the 24-Kbyte Flash device consists of also of 6 P-Flash banks but with the upper 2 banks only 2 Kbytes each, and only 1 D-Flash bank.

The P-Flash banks are always grouped in pairs. As such, the P-Flash banks are also sometimes referred to as P-Flash bank pair. Each sector in a P-Flash bank is grouped with the corresponding sector from the other bank within a bank pair to form a P-Flash bank pair sector.

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Sector 2: 128-byte	Sector 9: 128-byte
Sector 1: 128-byte	Sector 8: 128-byte
	Sector 7: 128-byte
	Sector 6: 128-byte
	Sector 5: 256-byte
	Sector 4: 256-byte
	Sector 3: 512-byte
Sector 0: 3.75-Kbyte	Sector 2: 512-byte
	Sector 1: 1-Kbyte
	Sector 0: 1-Kbyte
P-Flash	D-Flash

Figure 10 Flash Bank Sectorization

The internal structure of each Flash bank represents a sector architecture for flexible erase capability. The minimum erase width is always a complete sector, and sectors can be erased separately or in parallel. Contrary to standard EPROMs, erased Flash memory cells contain 0s.

The D-Flash bank is divided into more physical sectors for extended erasing and reprogramming capability; even numbers for each sector size are provided to allow greater flexibility and the ability to adapt to a wide range of application requirements.

#### 3.3.2 Parallel Read Access of P-Flash

To enhance system performance, the P-Flash banks are configured for parallel read to allow two bytes of linear code to be read in 4 x CCLK cycles, compared to 6 x CCLK cycles if serial read is performed. This is achieved by reading two bytes in parallel from a P-Flash bank pair within the 3 x CCLK cycles access time and storing them in a cache. Subsequent read from the cache by the CPU does not require a wait state and can be completed within 1 x CCLK cycle. The result is the average instruction fetch time from the P-Flash banks is reduced and thus, the MIPS (Mega Instruction Per Second) of the system is increased.

However, if the parallel read feature is not desired due to certain timing constraints, it can be disabled by calling the parallel read disable subroutine.

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### 3.3.3 Flash Programming Width

For the P-Flash banks, a programmed wordline (WL) must be erased before it can be reprogrammed as the Flash cells can only withstand one gate disturb. This means that the entire sector containing the WL must be erased since it is impossible to erase a single WL.

For the D-Flash bank, the same WL can be programmed twice before erasing is required as the Flash cells are able to withstand two gate disturbs. This means if the number of data bytes that needs to be written is smaller than the 32-byte minimum programming width, the user can opt to program this number of data bytes (x; where x can be any integer from 1 to 31) first and program the remaining bytes (32 - x) later. Hence, it is possible to program the same WL, for example, with 16 bytes of data two times (see Figure 11)

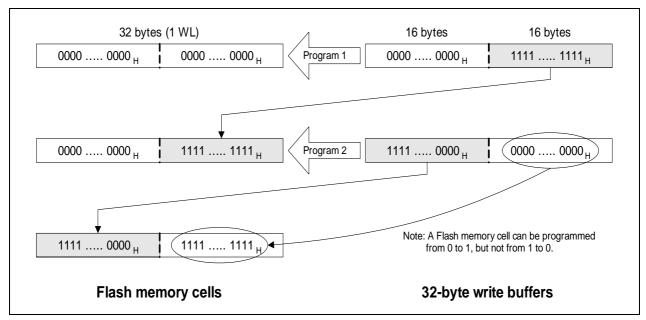


Figure 11 D-Flash Programming

Note: When programming a D-Flash WL the second time, the previously programmed Flash memory cells (whether 0s or 1s) should be reprogrammed with 0s to retain its original contents and to prevent "over-programming".

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### 3.4 Interrupt System

The XC800 Core supports one non-maskable interrupt (NMI) and 14 maskable interrupt requests. In addition to the standard interrupt functions supported by the core, e.g., configurable interrupt priority and interrupt masking, the SAL-XC886 interrupt system provides extended interrupt support capabilities such as the mapping of each interrupt vector to several interrupt sources to increase the number of interrupt sources supported, and additional status registers for detecting and determining the interrupt source.

### 3.4.1 Interrupt Source

Figure 12 to Figure 16 give a general overview of the interrupt sources and nodes, and their corresponding control and status flags.

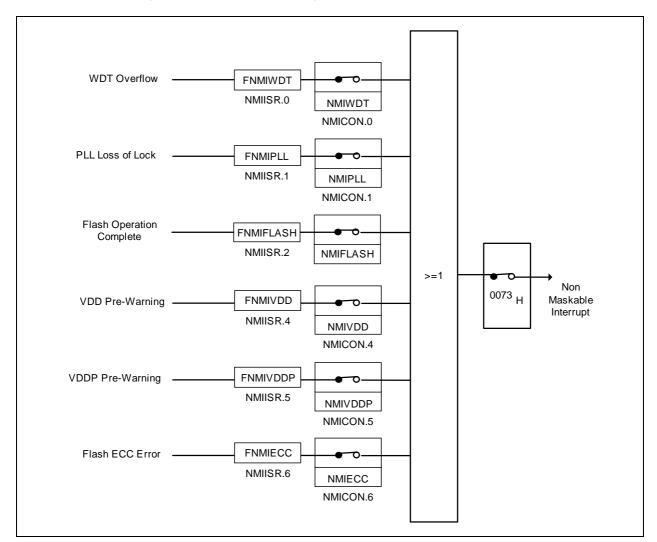


Figure 12 Non-Maskable Interrupt Request Sources



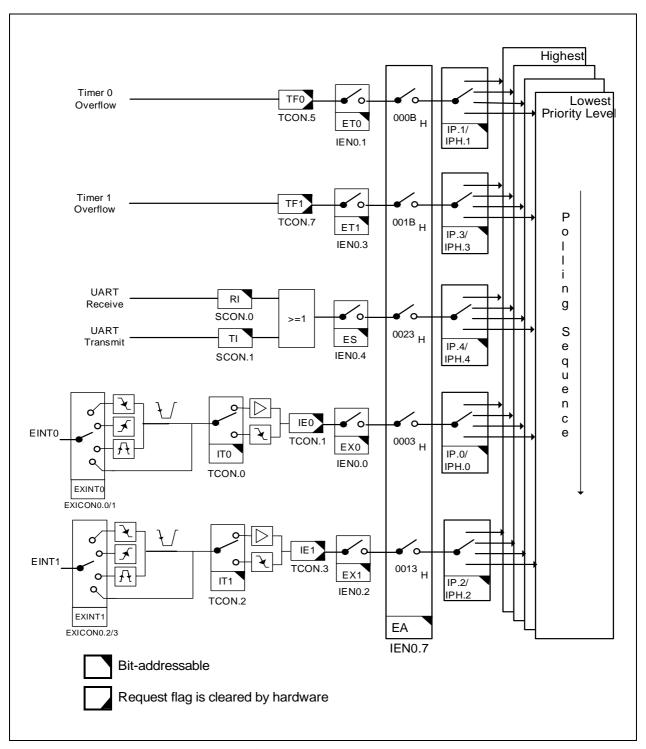


Figure 13 Interrupt Request Sources (Part 1)



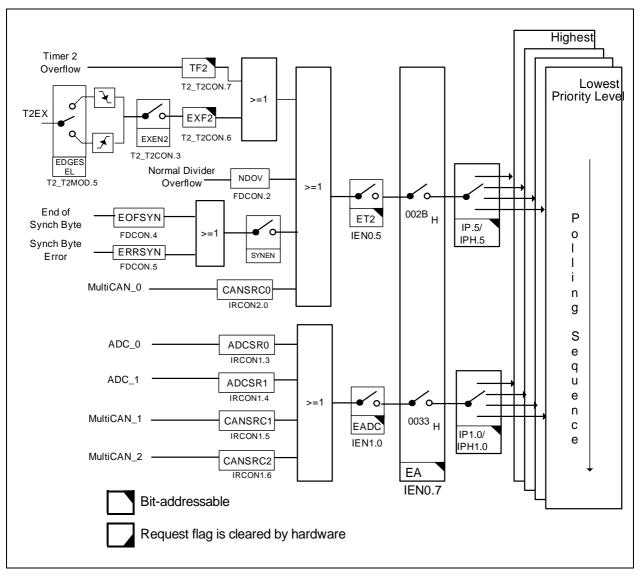


Figure 14 Interrupt Request Sources (Part 2)



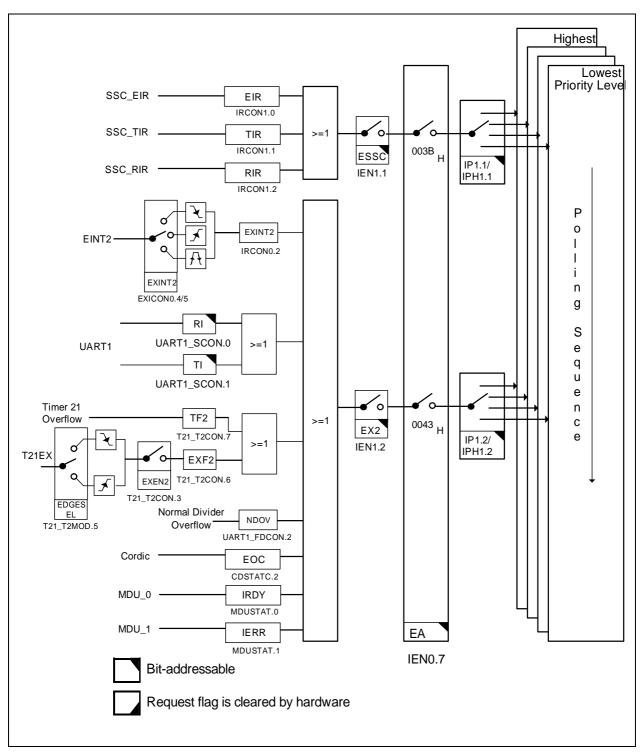


Figure 15 Interrupt Request Sources (Part 3)



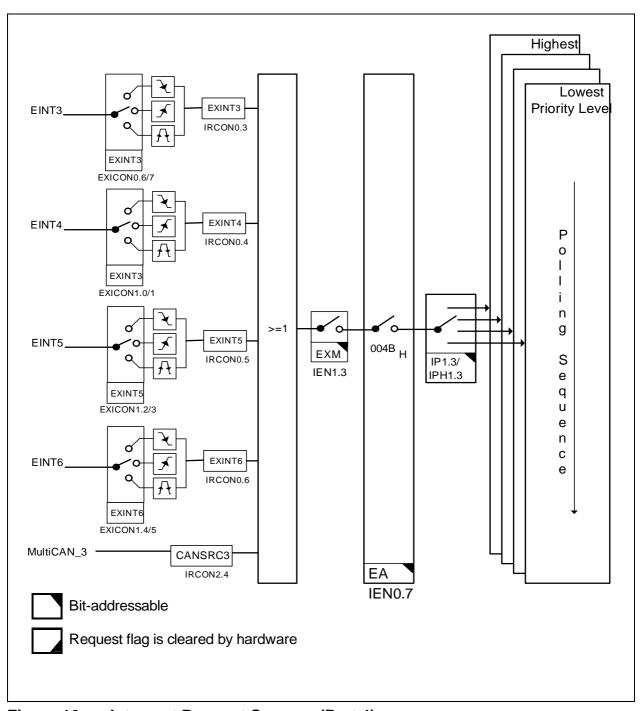


Figure 16 Interrupt Request Sources (Part 4)



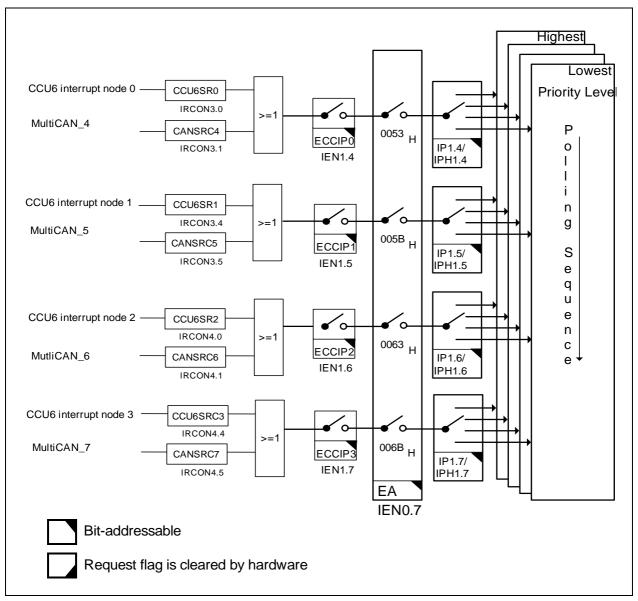


Figure 17 Interrupt Request Sources (Part 5)

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### 3.4.2 Interrupt Source and Vector

Each interrupt event source has an associated interrupt vector address for the interrupt node it belongs to. This vector is accessed to service the corresponding interrupt node request. The interrupt service of each interrupt source can be individually enabled or disabled via an enable bit. The assignment of the SAL-XC886 interrupt sources to the interrupt vector address and the corresponding interrupt node enable bits are summarized in **Table 19**.

Table 19 Interrupt Vector Addresses

Interrupt Source	Vector Address	Assignment for SAL- XC886	Enable Bit	SFR
NMI	0073 <sub>H</sub>	Watchdog Timer NMI	NMIWDT	NMICON
		PLL NMI	NMIPLL	
		Flash NMI	NMIFLASH	
		VDDC Prewarning NMI	NMIVDD	
		VDDP Prewarning NMI	NMIVDDP	
		Flash ECC NMI	NMIECC	
XINTR0	0003 <sub>H</sub>	External Interrupt 0	EX0	IEN0
XINTR1	000B <sub>H</sub>	Timer 0	ET0	
XINTR2	0013 <sub>H</sub>	External Interrupt 1	EX1	
XINTR3	001B <sub>H</sub>	Timer 1	ET1	
XINTR4	0023 <sub>H</sub>	UART	ES	
XINTR5	002B <sub>H</sub>	T2	ET2	
		UART Fractional Divider (Normal Divider Overflow)		
		MultiCAN Node 0		
		LIN		



 Table 19
 Interrupt Vector Addresses (cont'd)

Interrupt Source	Vector Address	Assignment for SAL- XC886	Enable Bit	SFR
XINTR6	0033 <sub>H</sub>	MultiCAN Nodes 1 and 2	EADC	IEN1
		ADC[1:0]		
XINTR7	003B <sub>H</sub>	SSC	ESSC	
XINTR8	0043 <sub>H</sub>	External Interrupt 2	EX2	
		T21		
		CORDIC		
		UART1		
		UART1 Fractional Divider (Normal Divider Overflow)		
		MDU[1:0]		
XINTR9 004B <sub>H</sub>		004B <sub>H</sub> External Interrupt 3		
		External Interrupt 4		
		External Interrupt 5		
		External Interrupt 6		
		MultiCAN Node 3		
XINTR10	0053 <sub>H</sub>	CCU6 INP0	ECCIP0	
		MultiCAN Node 4		
XINTR11	005B <sub>H</sub>	CCU6 INP1	ECCIP1	
		MultiCAN Node 5		
XINTR12	0063 <sub>H</sub>	CCU6 INP2	ECCIP2	
		MultiCAN Node 6		
XINTR13	006B <sub>H</sub>	CCU6 INP3	ECCIP3	
		MultiCAN Node 7		



### 3.4.3 Interrupt Priority

An interrupt that is currently being serviced can only be interrupted by a higher-priority interrupt, but not by another interrupt of the same or lower priority. Hence, an interrupt of the highest priority cannot be interrupted by any other interrupt request.

If two or more requests of different priority levels are received simultaneously, the request of the highest priority is serviced first. If requests of the same priority are received simultaneously, then an internal polling sequence determines which request is serviced first. Thus, within each priority level, there is a second priority structure determined by the polling sequence shown in **Table 20**.

Table 20 Priority Structure within Interrupt Level

	T
Source	Level
Non-Maskable Interrupt (NMI)	(highest)
External Interrupt 0	1
Timer 0 Interrupt	2
External Interrupt 1	3
Timer 1 Interrupt	4
UART Interrupt	5
Timer 2,UART Normal Divider Overflow, MultiCAN, LIN Interrupt	6
ADC, MultiCAN Interrupt	7
SSC Interrupt	8
External Interrupt 2, Timer 21, UART1, UART1 Normal Divider Overflow, MDU, CORDIC Interrupt	9
External Interrupt [6:3], MultiCAN Interrupt	10
CCU6 Interrupt Node Pointer 0, MultiCAN interrupt	11
CCU6 Interrupt Node Pointer 1, MultiCAN Interrupt	12
CCU6 Interrupt Node Pointer 2, MultiCAN Interrupt	13
CCU6 Interrupt Node Pointer 3, MultiCAN Interrupt	14



#### 3.5 Parallel Ports

The SAL-XC886 has 34 port pins organized into five parallel ports, Port 0 (P0) to Port 4 (P4). Each pin has a pair of internal pull-up and pull-down devices that can be individually enabled or disabled. Ports P0, P1, P3 and P4 are bidirectional and can be used as general purpose input/output (GPIO) or to perform alternate input/output functions for the on-chip peripherals. When configured as an output, the open drain mode can be selected. Port P2 is an input-only port, providing general purpose input functions, alternate input functions for the on-chip peripherals, and also analog inputs for the Analog-to-Digital Converter (ADC).

#### **Bidirectional Port Features**

- Configurable pin direction
- Configurable pull-up/pull-down devices
- Configurable open drain mode
- Transfer of data through digital inputs and outputs (general purpose I/O)
- Alternate input/output for on-chip peripherals

#### **Input Port Features**

- Configurable input driver
- Configurable pull-up/pull-down devices
- Receive of data through digital input (general purpose input)
- Alternate input for on-chip peripherals
- Analog input for ADC module



Figure 18 shows the structure of a bidirectional port pin.

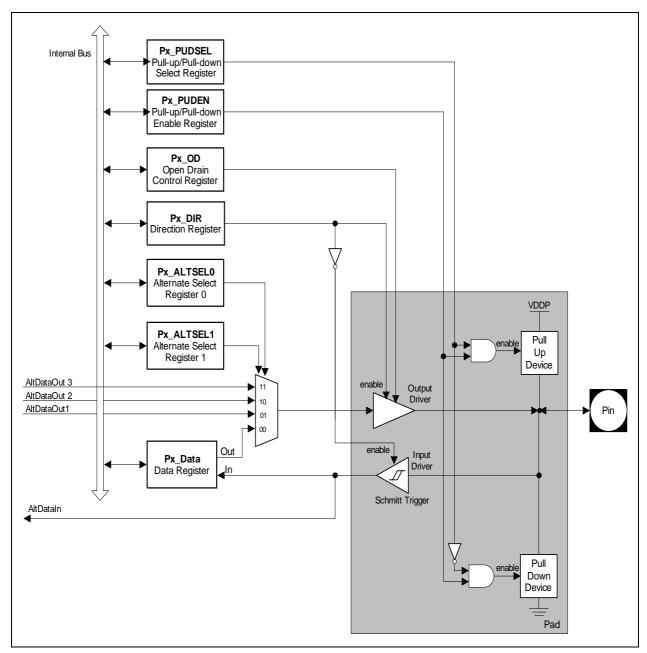


Figure 18 General Structure of Bidirectional Port



Figure 19 shows the structure of an input-only port pin.

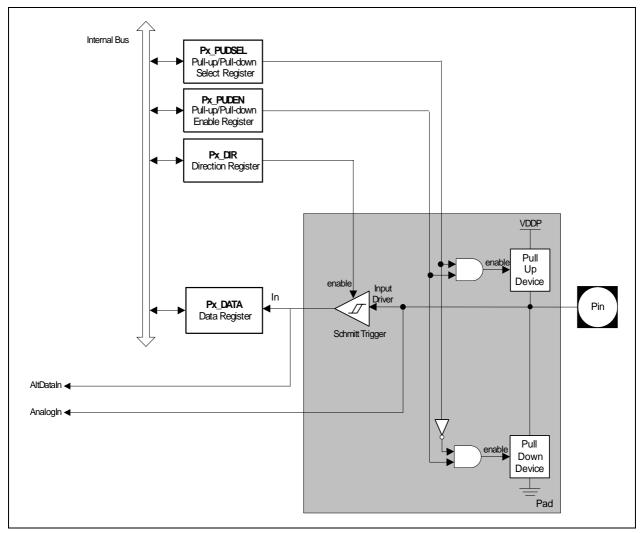


Figure 19 General Structure of Input Port



### 3.6 Power Supply System with Embedded Voltage Regulator

The SAL-XC886 microcontroller requires two different levels of power supply:

- 5.0 V for the Embedded Voltage Regulator (EVR) and Ports
- 2.5 V for the core, memory, on-chip oscillator, and peripherals

**Figure 20** shows the SAL-XC886 power supply system. A power supply of 5.0 V must be provided from the external power supply pin. The 2.5 V power supply for the logic is generated by the EVR. The EVR helps to reduce the power consumption of the whole chip and the complexity of the application board design.

The EVR consists of a main voltage regulator and a low power voltage regulator. In active mode, both voltage regulators are enabled. In power-down mode, the main voltage regulator is switched off, while the low power voltage regulator continues to function and provide power supply to the system with low power consumption.

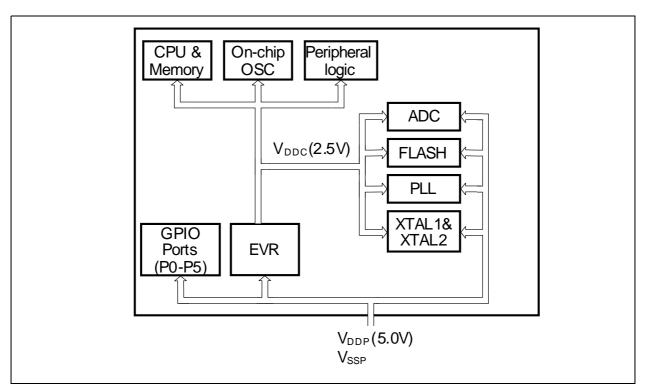


Figure 20 SAL-XC886 Power Supply System

#### **EVR Features**

- Input voltage (V<sub>DDP</sub>): 5.0 V
- Output voltage  $(V_{DDC})$ : 2.5 V ± 7.5%
- Low power voltage regulator provided in power-down mode
- $V_{\rm DDC}$  and  $V_{\rm DDP}$  prewarning detection
- V<sub>DDC</sub> brownout detection



#### 3.7 Reset Control

The SAL-XC886 has five types of reset: power-on reset, hardware reset, watchdog timer reset, power-down wake-up reset, and brownout reset.

When the SAL-XC886 is first powered up, the status of certain pins (see **Table 22**) must be defined to ensure proper start operation of the device. At the end of a reset sequence, the sampled values are latched to select the desired boot option, which cannot be modified until the next power-on reset or hardware reset. This guarantees stable conditions during the normal operation of the device.

In order to power up the system properly, the external reset pin RESET must be asserted until  $V_{\rm DDC}$  reaches 0.9\* $V_{\rm DDC}$ . The delay of external reset can be realized by an external capacitor at  $\overline{\rm RESET}$  pin. This capacitor value must be selected so that  $V_{\rm RESET}$  reaches 0.4 V, but not before  $V_{\rm DDC}$  reaches 0.9\*  $V_{\rm DDC}$ .

A typical application example is shown in Figure 21. The  $V_{\rm DDP}$  capacitor value is 100 nF while the  $V_{\rm DDC}$  capacitor value is 220 nF. The capacitor connected to RESET pin is 100 nF.

Typically, the time taken for  $V_{\rm DDC}$  to reach  $0.9^*V_{\rm DDC}$  is less than 50  $\mu \rm s$  once  $V_{\rm DDP}$  reaches 2.3V. Hence, based on the condition that 10% to 90%  $V_{\rm DDP}$  (slew rate) is less than 500  $\mu \rm s$ , the RESET pin should be held low for 500  $\mu \rm s$  typically. See Figure 22.

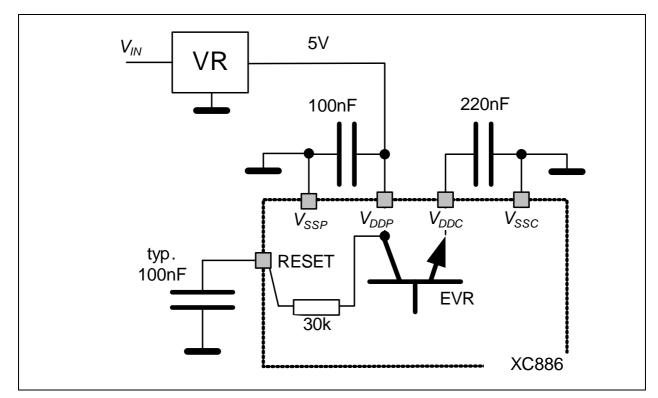


Figure 21 Reset Circuitry



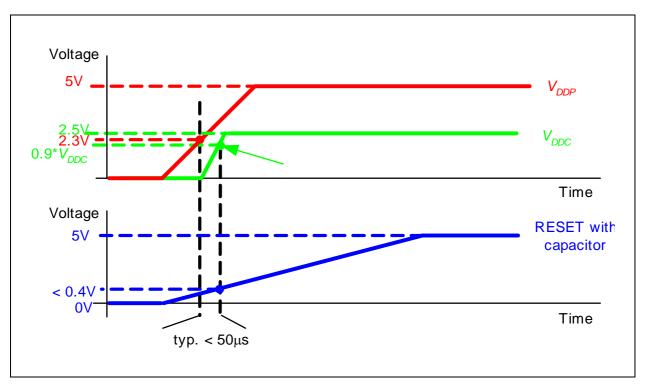


Figure 22  $V_{\text{DDP}}$ ,  $V_{\text{DDC}}$  and  $V_{\text{RESET}}$  during Power-on Reset

The second type of reset in SAL-XC886 is the hardware reset. This reset function can be used during normal operation or when the chip is in power-down mode. A reset input pin RESET is provided for the hardware reset.

The Watchdog Timer (WDT) module is also capable of resetting the device if it detects a malfunction in the system.

Another type of reset that needs to be detected is a reset while the device is in power-down mode (wake-up reset). While the contents of the static RAM are undefined after a power-on reset, they are well defined after a wake-up reset from power-down mode.



#### 3.7.1 Module Reset Behavior

**Table 21** lists the functions of the SAL-XC886 and the various reset types that affect these functions. The symbol "■" signifies that the particular function is reset to its default state.

Table 21 Effect of Reset on Device Functions

Module/ Function	Wake-Up Reset	Watchdog Reset	Hardware Reset	Power-On Reset	Brownout Reset
CPU Core					
Peripherals					
On-Chip Static RAM	Not affected, Reliable	Not affected, Reliable	Not affected, Reliable	Affected, un- reliable	Affected, un- reliable
Oscillator, PLL		Not affected			
Port Pins					
EVR	The voltage regulator is switched on	Not affected			
FLASH					
NMI	Disabled	Disabled			

# 3.7.2 Booting Scheme

When the SAL-XC886 is reset, it must identify the type of configuration with which to start the different modes once the reset sequence is complete. Thus, boot configuration information that is required for activation of special modes and conditions needs to be applied by the external world through input pins. After power-on reset or hardware reset, the pins MBC, TMS and P0.0 collectively select the different boot options. **Table 22** shows the available boot options in the SAL-XC886.

Table 22 SAL-XC886 Boot Selection

MBC	TMS	P0.0	Type of Mode	PC Start Value
1	0	Χ	User Mode <sup>1)</sup> ; on-chip OSC/PLL non-bypassed	0000 <sub>H</sub>
0	0	Χ	BSL Mode; on-chip OSC/PLL non-bypassed <sup>2)</sup>	0000 <sub>H</sub>
0	1	0	OCDS Mode; on-chip OSC/PLL non-bypassed	0000 <sub>H</sub>
1	1	0	User (JTAG) Mode <sup>3)</sup> ; on-chip OSC/PLL non-bypassed (normal)	0000 <sub>H</sub>



- 1) BSL mode is automatically entered if no valid password is installed and data at memory address 0000H equals zero.
- 2) OSC is bypassed in MultiCAN BSL mode
- 3) Normal user mode with standard JTAG (TCK,TDI,TDO) pins for hot-attach purpose.

Note: The boot options are valid only with the default set of UART and JTAG pins.

#### 3.8 Clock Generation Unit

The Clock Generation Unit (CGU) allows great flexibility in the clock generation for the SAL-XC886. The power consumption is indirectly proportional to the frequency, whereas the performance of the microcontroller is directly proportional to the frequency. During user program execution, the frequency can be programmed for an optimal ratio between performance and power consumption. Therefore the power consumption can be adapted to the actual application state.

#### **Features**

- Phase-Locked Loop (PLL) for multiplying clock source by different factors
- PLL Base Mode
- Prescaler Mode
- PLL Mode
- Power-down mode support

The CGU consists of an oscillator circuit and a PLL. In the SAL-XC886, the oscillator can be from either of these two sources: the on-chip oscillator (10 MHz) or the external oscillator (4 MHz to 12 MHz). The term "oscillator" is used to refer to both on-chip oscillator and external oscillator, unless otherwise stated. After the reset, the on-chip oscillator will be used by default. The external oscillator can be selected via software. In addition, the PLL provides a fail-safe logic to perform oscillator run and loss-of-lock detection. This allows emergency routines to be executed for system recovery or to perform system shut down.

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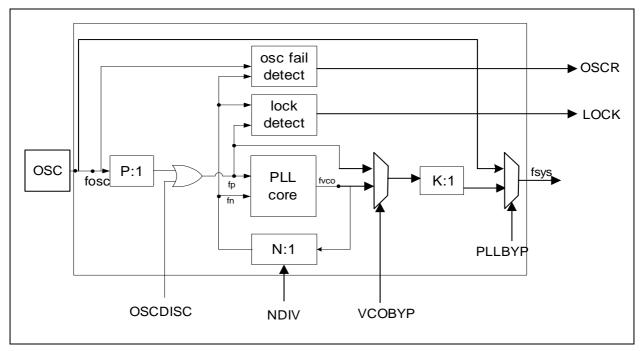


Figure 23 CGU Block Diagram

#### **PLL Base Mode**

When the oscillator is disconnected from the PLL, the system clock is derived from the VCO base (free running) frequency clock (Table 24) divided by the K factor.

$$f_{SYS} = f_{VCObase} \times \frac{1}{K}$$

(3.1)

## **Prescaler Mode (VCO Bypass Operation)**

In VCO bypass operation, the system clock is derived from the oscillator clock, divided by the P and K factors.

$$f_{SYS} = f_{OSC} \times \frac{1}{P \times K}$$

(3.2)



#### **PLL Mode**

The system clock is derived from the oscillator clock, multiplied by the N factor, and divided by the P and K factors. Both VCO bypass and PLL bypass must be inactive for this PLL mode. The PLL mode is used during normal system operation.

$$f_{SYS} = f_{OSC} \times \frac{N}{P \times K}$$

(3.3)

## **System Frequency Selection**

For the SAL-XC886, the value of P is fixed to 1. In order to obtain the required fsys, the value of N and K can be selected by bits NDIV and KDIV respectively for different oscillator inputs. The output frequency must always be configured for 80 MHz. Table 23 provides examples on how  $f_{\rm sys}$  = 80 MHz can be obtained for the different oscillator sources.

Table 23 System frequency ( $f_{sys}$  = 80 MHz)

Oscillator	Fosc	N	Р	K	Fsys
On-chip	10 MHz	16	1	2	80 MHz
External	8 MHz	20	1	2	80 MHz
	5 MHz	32	1	2	80 MHz
	4 MHz	40	1	2	80 MHz



Table 24 shows the VCO range for the SAL-XC886.

Table 24 VCO Range

$f_{\sf VCOmin}$	$f_{\sf VCOmax}$	$f_{ extsf{VCOFREEmin}}$	$f_{\sf VCOFREEmax}$	Unit
150	200	20	80	MHz
100	150	10	80	MHz

#### 3.8.1 Recommended External Oscillator Circuits

The oscillator circuit, a Pierce oscillator, is designed to work with both, an external crystal oscillator or an external stable clock source. It basically consists of an inverting amplifier and a feedback element with XTAL1 as input, and XTAL2 as output.

When using a crystal, a proper external oscillator circuitry must be connected to both pins, XTAL1 and XTAL2. The crystal frequency can be within the range of 4 MHz to 12 MHz. Additionally, it is necessary to have two load capacitances  $C_{\rm X1}$  and  $C_{\rm X2}$ , and depending on the crystal type, a series resistor  $R_{\rm X2}$ , to limit the current. A test resistor  $R_{\rm Q}$  may be temporarily inserted to measure the oscillation allowance (negative resistance) of the oscillator circuitry.  $R_{\rm Q}$  values are typically specified by the crystal vendor. The  $C_{\rm X1}$  and  $C_{\rm X2}$  values shown in **Figure 24** can be used as starting points for the negative resistance evaluation and for non-productive systems. The exact values and related operating range are dependent on the crystal frequency and have to be determined and optimized together with the crystal vendor using the negative resistance method. Oscillation measurement with the final target system is strongly recommended to verify the input amplitude at XTAL1 and to determine the actual oscillation allowance (margin negative resistance) for the oscillator-crystal system.

When using an external clock signal, the signal must be connected to XTAL1. XTAL2 is left open (unconnected).

The oscillator can also be used in combination with a ceramic resonator. The final circuitry must also be verified by the resonator vendor. **Figure 24** shows the recommended external oscillator circuitries for both operating modes, external crystal mode and external input clock mode.



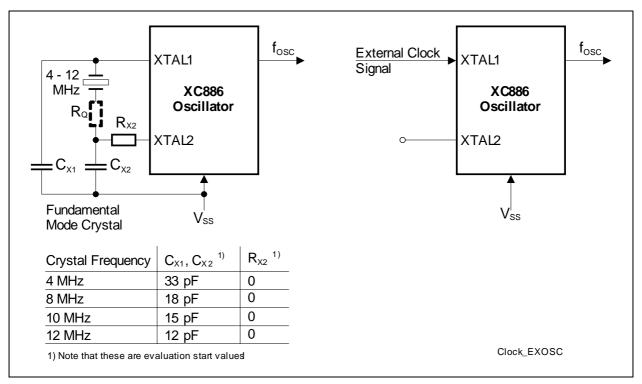


Figure 24 External Oscillator Circuitry

Note: For crystal operation, it is strongly recommended to measure the negative resistance in the final target system (layout) to determine the optimum parameters for the oscillator operation. Please refer to the minimum and maximum values of the negative resistance specified by the crystal supplier.



## 3.8.2 Clock Management

The CGU generates all clock signals required within the microcontroller from a single clock,  $f_{\rm sys}$ . During normal system operation, the typical frequencies of the different modules are as follow:

- CPU clock: CCLK, SCLK = 20 MHz
- Fast clock (used by MultiCAN): FCLK = 20 or 40 MHz
- Peripheral clock: PCLK = 20 MHz
- Flash Interface clock: CCLK2 = 40 MHz and CCLK = 20 MHz

In addition, different clock frequencies can be output to pin CLKOUT (P0.0 or P0.7). The clock output frequency, which is derived from the clock output divider (bit COREL), can further be divided by 2 using toggle latch (bit TLEN is set to 1). The resulting output frequency has a 50% duty cycle. **Figure 25** shows the clock distribution of the SAL-XC886.

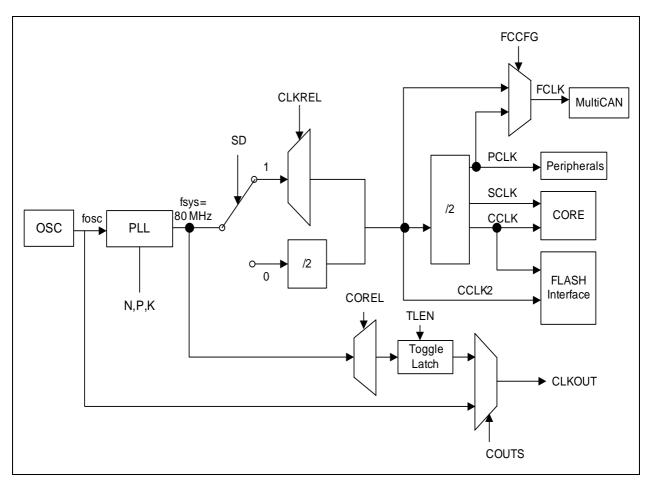


Figure 25 Clock Generation from  $f_{\text{sys}}$ 



For power saving purposes, the clocks may be disabled or slowed down according to **Table 25**.

Table 25 System frequency ( $f_{\text{sys}} = 80 \text{ MHz}$ )

Power Saving Mode	Action
Idle	Clock to the CPU is disabled.
Slow-down	Clocks to the CPU and all the peripherals are divided by a common programmable factor defined by bit field CMCON.CLKREL.
Power-down	Oscillator and PLL are switched off.

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## 3.9 Power Saving Modes

The power saving modes of the SAL-XC886 provide flexible power consumption through a combination of techniques, including:

- Stopping the CPU clock
- Stopping the clocks of individual system components
- Reducing clock speed of some peripheral components
- Power-down of the entire system with fast restart capability

After a reset, the active mode (normal operating mode) is selected by default (see Figure 26) and the system runs in the main system clock frequency. From active mode, different power saving modes can be selected by software. They are:

- Idle mode
- Slow-down mode
- Power-down mode

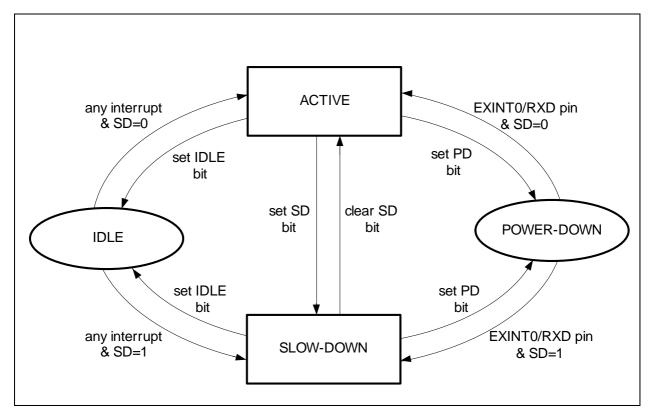


Figure 26 Transition between Power Saving Modes



## 3.10 Watchdog Timer

The Watchdog Timer (WDT) provides a highly reliable and secure way to detect and recover from software or hardware failures. The WDT is reset at a regular interval that is predefined by the user. The CPU must service the WDT within this interval to prevent the WDT from causing an SAL-XC886 system reset. Hence, routine service of the WDT confirms that the system is functioning properly. This ensures that an accidental malfunction of the SAL-XC886 will be aborted in a user-specified time period.

In debug mode, the WDT is default suspended and stops counting. Therefore, there is no need to refresh the WDT during debugging.

#### **Features**

- 16-bit Watchdog Timer
- Programmable reload value for upper 8 bits of timer
- Programmable window boundary
- Selectable input frequency of  $f_{PCLK}/2$  or  $f_{PCLK}/128$
- Time-out detection with NMI generation and reset prewarning activation (after which a system reset will be performed)

The WDT is a 16-bit timer incremented by a count rate of  $f_{\rm PCLK}/2$  or  $f_{\rm PCLK}/128$ . This 16-bit timer is realized as two concatenated 8-bit timers. The upper 8 bits of the WDT can be preset to a user-programmable value via a watchdog service access in order to modify the watchdog expire time period. The lower 8 bits are reset on each service access. Figure 27 shows the block diagram of the WDT unit.

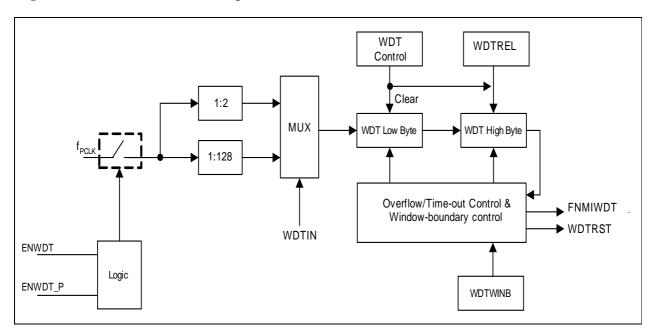


Figure 27 WDT Block Diagram



If the WDT is not serviced before the timer overflow, a system malfunction is assumed. As a result, the WDT NMI is triggered (assert FNMIWDT) and the reset prewarning is entered. The prewarning period lasts for  $30_{\rm H}$  count, after which the system is reset (assert WDTRST).

The WDT has a "programmable window boundary" which disallows any refresh during the WDT's count-up. A refresh during this window boundary constitutes an invalid access to the WDT, causing the reset prewarning to be entered but without triggering the WDT NMI. The system will still be reset after the prewarning period is over. The window boundary is from  $0000_{\rm H}$  to the value obtained from the concatenation of WDTWINB and  $00_{\rm H}$ .

After being serviced, the WDT continues counting up from the value (<WDTREL> \* 2<sup>8</sup>). The time period for an overflow of the WDT is programmable in two ways:

- The input frequency to the WDT can be selected to be either  $f_{\rm PCLK}/2$  or  $f_{\rm PCLK}/128$
- The reload value WDTREL for the high byte of WDT can be programmed in register WDTREL

The period,  $P_{\rm WDT}$ , between servicing the WDT and the next overflow can be determined by the following formula:

$$P_{WDT} = \frac{2^{(1+WDTIN \times 6)} \times (2^{16} - WDTREL \times 2^{8})}{f_{PCLK}}$$

(3.4)

If the Window-Boundary Refresh feature of the WDT is enabled, the period  $P_{\rm WDT}$  between servicing the WDT and the next overflow is shortened if WDTWINB is greater than WDTREL, see **Figure 28**. This period can be calculated using the same formula by replacing WDTREL with WDTWINB. For this feature to be useful, WDTWINB cannot be smaller than WDTREL.

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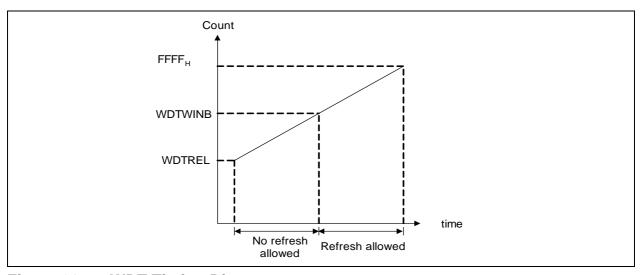


Figure 28 WDT Timing Diagram

**Table 26** lists the possible watchdog time ranges that can be achieved using a certain module clock. Some numbers are rounded to 3 significant digits.

Table 26 Watchdog Time Ranges

Reload value In WDTREL	Prescaler for $f_{\sf PCLK}$			
	2 (WDTIN = 0)	128 (WDTIN = 1)		
	20 MHz	20 MHz		
FF <sub>H</sub>	25.6 μs	1.64 ms		
$\frac{FF_{H}}{7F_{H}}$ $00_{H}$	3.30 ms	211 ms		
00 <sub>H</sub>	6.55 ms	419 ms		



## 3.11 Multiplication/Division Unit

The Multiplication/Division Unit (MDU) provides fast 16-bit multiplication, 16-bit and 32-bit division as well as shift and normalize features. It has been integrated to support the SAL-XC886 Core in real-time control applications, which require fast mathematical computations.

#### **Features**

- Fast signed/unsigned 16-bit multiplication
- Fast signed/unsigned 32-bit divide by 16-bit and 16-bit divide by 16-bit operations
- 32-bit unsigned normalize operation
- 32-bit arithmetic/logical shift operations

Table 27 specifies the number of clock cycles used for calculation in various operations.

**Table 27** MDU Operation Characteristics

Operation	Result	Remainder	No. of Clock Cycles used for calculation
Signed 32-bit/16-bit	32-bit	16-bit	33
Signed 16-bit/16bit	16-bit	16-bit	17
Signed 16-bit x 16-bit	32-bit	-	16
Unsigned 32-bit/16-bit	32-bit	16-bit	32
Unsigned 16-bit/16-bit	16-bit	16-bit	16
Unsigned 16-bit x 16-bit	32-bit	-	16
32-bit normalize	-	-	No. of shifts + 1 (Max. 32)
32-bit shift L/R	-	-	No. of shifts + 1 (Max. 32)



## 3.12 CORDIC Coprocessor

The CORDIC Coprocessor provides CPU with hardware support for the solving of circular (trigonometric), linear (multiply-add, divide-add) and hyperbolic functions.

#### **Features**

- Modes of operation
  - Supports all CORDIC operating modes for solving circular (trigonometric), linear (multiply-add, divide-add) and hyperbolic functions
  - Integrated look-up tables (LUTs) for all operating modes
- Circular vectoring mode: Extended support for values of initial X and Y data up to full range of [-2<sup>15</sup>,(2<sup>15</sup>-1)] for solving angle and magnitude
- Circular rotation mode: Extended support for values of initial Z data up to full range of  $[-2^{15},(2^{15}-1)]$ , representing angles in the range  $[-\pi,((2^{15}-1)/2^{15})\pi]$  for solving trigonometry
- Implementation-dependent operational frequency of up to 80 MHz
- Gated clock input to support disabling of module
- 16-bit accessible data width
  - 24-bit kernel data width plus 2 overflow bits for X and Y each
  - 20-bit kernel data width plus 1 overflow bit for Z
  - With KEEP bit to retain the last value in the kernel register for a new calculation
- 16 iterations per calculation: Approximately 41 clock-cycles or less, from set of start (ST) bit to set of end-of-calculation flag, excluding time taken for write and read access of data bytes.
- Twos complement data processing
  - Only exception: X result data with user selectable option for unsigned result
- X and Y data generally accepted as integer or rational number; X and Y must be of the same data form
- Entries of LUTs are 20-bit signed integers
  - Entries of atan and atanh LUTs are integer representations (S19) of angles with the scaling such that  $[-2^{15},(2^{15}-1)]$  represents the range  $[-\pi,((2^{15}-1)/2^{15})\pi]$
  - Accessible Z result data for circular and hyperbolic functions is integer in data form of S15
- Emulated LUT for linear function
  - Data form is 1 integer bit and 15-bit fractional part (1.15)
  - Accessible Z result data for linear function is rational number with fixed data form of S4.11 (signed 4Q16)
- Truncation Error
  - The result of a CORDIC calculation may return an approximation due to truncation of LSBs
  - Good accuracy of the CORDIC calculated result data, especially in circular mode
- Interrupt
  - On completion of a calculation

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- Interrupt enabling and corresponding flag

#### 3.13 UART and UART1

The SAL-XC886 provides two Universal Asynchronous Receiver/Transmitter (UART and UART1) modules for full-duplex asynchronous reception/transmission. Both are also receive-buffered, i.e., they can commence reception of a second byte before a previously received byte has been read from the receive register. However, if the first byte still has not been read by the time reception of the second byte is complete, one of the bytes will be lost.

#### **Features**

- Full-duplex asynchronous modes
  - 8-bit or 9-bit data frames, LSB first
  - Fixed or variable baud rate
- Receive buffered
- Multiprocessor communication
- Interrupt generation on the completion of a data transmission or reception

The UART modules can operate in the four modes shown in Table 28.

Table 28 UART Modes

Operating Mode	Baud Rate
Mode 0: 8-bit shift register	$f_{PCLK}/2$
Mode 1: 8-bit shift UART	Variable
Mode 2: 9-bit shift UART	$f_{\text{PCLK}}/32 \text{ or } f_{\text{PCLK}}/64^{1)}$
Mode 3: 9-bit shift UART	Variable

<sup>1)</sup> For UART1 module, the baud rate is fixed at f<sub>PCLK</sub>/64.

There are several ways to generate the baud rate clock for the serial port, depending on the mode in which it is operating. In mode 0, the baud rate for the transfer is fixed at  $f_{\rm PCLK}/2$ . In mode 2, the baud rate is generated internally based on the UART input clock and can be configured to either  $f_{\rm PCLK}/32$  or  $f_{\rm PCLK}/64$ . For UART1 module, only  $f_{\rm PCLK}/64$  is available. The variable baud rate is set by the underflow rate on the dedicated baud-rate generator. For UART module, the variable baud rate alternatively can be set by the overflow rate on Timer 1.

#### 3.13.1 Baud-Rate Generator

Both UART modules have their own dedicated baud-rate generator, which is based on a programmable 8-bit reload value, and includes divider stages (i.e., prescaler and



fractional divider) for generating a wide range of baud rates based on its input clock  $f_{PCLK}$ , see **Figure 29**.

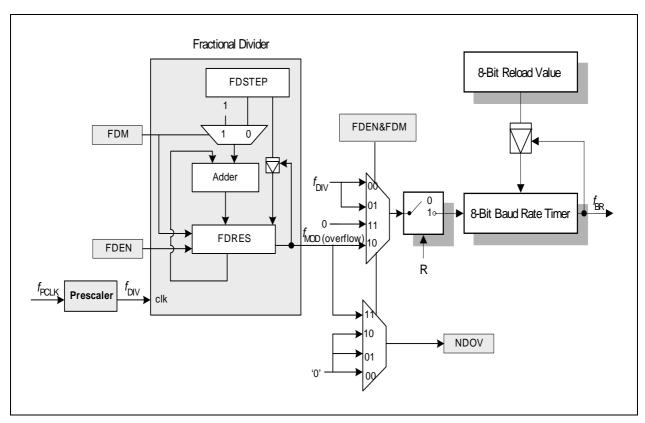


Figure 29 Baud-rate Generator Circuitry

The baud rate timer is a count-down timer and is clocked by either the output of the fractional divider ( $f_{MOD}$ ) if the fractional divider is enabled (FDCON.FDEN = 1), or the output of the prescaler ( $f_{DIV}$ ) if the fractional divider is disabled (FDEN = 0). For baud rate generation, the fractional divider must be configured to fractional divider mode (FDCON.FDM = 0). This allows the baud rate control run bit BCON.R to be used to start or stop the baud rate timer. At each timer underflow, the timer is reloaded with the 8-bit reload value in register BG and one clock pulse is generated for the serial channel.

Enabling the fractional divider in normal divider mode (FDEN = 1 and FDM = 1) stops the baud rate timer and nullifies the effect of bit BCON.R. See **Section 3.14**.

The baud rate  $(f_{BR})$  value is dependent on the following parameters:

- Input clock  $f_{PCLK}$
- Prescaling factor (2<sup>BRPRE</sup>) defined by bit field BRPRE in register BCON
- Fractional divider (STEP/256) defined by register FDSTEP
   (to be considered only if fractional divider is enabled and operating in fractional divider mode)
- 8-bit reload value (BR\_VALUE) for the baud rate timer defined by register BG

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The following formulas calculate the final baud rate without and with the fractional divider respectively:

baud rate = 
$$\frac{f_{PCLK}}{16 \times 2^{BRPRE} \times (BR\_VALUE + 1)} \quad \text{where } 2^{BRPRE} \times (BR\_VALUE + 1) > 1$$

(3.5)

baud rate = 
$$\frac{f_{PCLK}}{16 \times 2^{BRPRE} \times (BR\_VALUE + 1)} \times \frac{STEP}{256}$$

(3.6)

The maximum baud rate that can be generated is limited to  $f_{\rm PCLK}/32$ . Hence, for a module clock of 20 MHz, the maximum achievable baud rate is 0.625 MBaud.

Standard LIN protocol can support a maximum baud rate of 20 kHz, the baud rate accuracy is not critical and the fractional divider can be disabled. Only the prescaler is used for auto baud rate calculation. For LIN fast mode, which supports the baud rate of 20 kHz to 115.2 kHz, the higher baud rates require the use of the fractional divider for greater accuracy.

**Table 29** lists the various commonly used baud rates with their corresponding parameter settings and deviation errors. The fractional divider is disabled and a module clock of 20 MHz is used.

Table 29 Typical Baud rates for UART with Fractional Divider disabled

Baud rate	Prescaling Factor (2BRPRE)	Reload Value (BR_VALUE + 1)	Deviation Error
19.2 kBaud	1 (BRPRE=000 <sub>B</sub> )	65 (41 <sub>H</sub> )	0.16 %
9600 Baud	1 (BRPRE=000 <sub>B</sub> )	130 (82 <sub>H</sub> )	0.16 %
4800 Baud	2 (BRPRE=001 <sub>B</sub> )	130 (82 <sub>H</sub> )	0.16 %
2400 Baud	4 (BRPRE=010 <sub>B</sub> )	130 (82 <sub>H</sub> )	0.16 %

The fractional divider allows baud rates of higher accuracy (lower deviation error) to be generated. **Table 30** lists the resulting deviation errors from generating a baud rate of 115.2 kHz, using different module clock frequencies. The fractional divider is enabled (fractional divider mode) and the corresponding parameter settings are shown.

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Table 30 Deviation Error for UART with Fractional Divider enabled

$f_{ t PCLK}$	Prescaling Factor (BRPRE)	Reload Value (BR_VALUE + 1)	STEP	Deviation Error
20 MHz	1	10 (A <sub>H</sub> )	230 (E6 <sub>H</sub> )	+0.03 %
10 MHz	1	5 (5 <sub>H</sub> )	230 (E6 <sub>H</sub> )	+0.03 %
6.67 MHz	1	3 (3 <sub>H</sub> )	212 (D4 <sub>H</sub> )	-0.16 %
5 MHz	1	2 (2 <sub>H</sub> )	189 (BD <sub>H</sub> )	+0.14 %

## 3.13.2 Baud Rate Generation using Timer 1

In UART modes 1 and 3 of UART module, Timer 1 can be used for generating the variable baud rates. In theory, this timer could be used in any of its modes. But in practice, it should be set into auto-reload mode (Timer 1 mode 2), with its high byte set to the appropriate value for the required baud rate. The baud rate is determined by the Timer 1 overflow rate and the value of SMOD as follows:

Mode 1, 3 band rate= 
$$\frac{2^{\text{SMOD}} \times f_{\text{PCLK}}}{32 \times 2 \times (256 - \text{TH1})}$$

(3.7)

# 3.14 Normal Divider Mode (8-bit Auto-reload Timer)

Setting bit FDM in register FDCON to 1 configures the fractional divider to normal divider mode, while at the same time disables baud rate generation (see Figure 29). Once the fractional divider is enabled (FDEN = 1), it functions as an 8-bit auto-reload timer (with no relation to baud rate generation) and counts up from the reload value with each input clock pulse. Bit field RESULT in register FDRES represents the timer value, while bit field STEP in register FDSTEP defines the reload value. At each timer overflow, an overflow flag (FDCON.NDOV) will be set and an interrupt request generated. This gives an output clock  $f_{\text{MOD}}$  that is 1/n of the input clock  $f_{\text{DIV}}$ , where n is defined by 256 - STEP.

The output frequency in normal divider mode is derived as follows:

$$f_{MOD} = f_{DIV} \times \frac{1}{256 - STEP}$$

(3.8)



#### 3.15 LIN Protocol

The UART module can be used to support the Local Interconnect Network (LIN) protocol for both master and slave operations. The LIN baud rate detection feature, which consists of the hardware logic for Break and Synch Byte detection, provides the capability to detect the baud rate within LIN protocol using Timer 2. This allows the UART to be synchronized to the LIN baud rate for data transmission and reception.

Note: The LIN baud rate detection feature is available for use only with UART. To use UART1 for LIN communication, software has to be implemented to detect the Break and Synch Byte.

LIN is a holistic communication concept for local interconnected networks in vehicles. The communication is based on the SCI (UART) data format, a single-master/multiple-slave concept, a clock synchronization for nodes without stabilized time base. An attractive feature of LIN is self-synchronization of the slave nodes without a crystal or ceramic resonator, which significantly reduces the cost of hardware platform. Hence, the baud rate must be calculated and returned with every message frame.

The structure of a LIN frame is shown in **Figure 30**. The frame consists of the:

- Header, which comprises a Break (13-bit time low), Synch Byte (55<sub>H</sub>), and ID field
- Response time
- Data bytes (according to UART protocol)
- Checksum

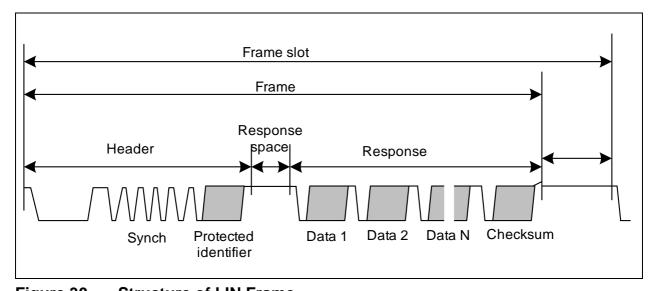


Figure 30 Structure of LIN Frame



#### 3.15.1 LIN Header Transmission

LIN header transmission is only applicable in master mode. In the LIN communication, a master task decides when and which frame is to be transferred on the bus. It also identifies a slave task to provide the data transported by each frame. The information needed for the handshaking between the master and slave tasks is provided by the master task through the header portion of the frame.

The header consists of a break and synch pattern followed by an identifier. Among these three fields, only the break pattern cannot be transmitted as a normal 8-bit UART data. The break must contain a dominant value of 13 bits or more to ensure proper synchronization of slave nodes.

In the LIN communication, a slave task is required to be synchronized at the beginning of the protected identifier field of frame. For this purpose, every frame starts with a sequence consisting of a break field followed by a synch byte field. This sequence is unique and provides enough information for any slave task to detect the beginning of a new frame and be synchronized at the start of the identifier field.

Upon entering LIN communication, a connection is established and the transfer speed (baud rate) of the serial communication partner (host) is automatically synchronized in the following steps:

STEP 1: Initialize interface for reception and timer for baud rate measurement

STEP 2: Wait for an incoming LIN frame from host

STEP 3: Synchronize the baud rate to the host

STEP 4: Enter for Master Request Frame or for Slave Response Frame

Note: Re-synchronization and setup of baud rate are always done for **every** Master Request Header or Slave Response Header LIN frame.



## 3.16 High-Speed Synchronous Serial Interface

The High-Speed Synchronous Serial Interface (SSC) supports full-duplex and half-duplex synchronous communication. The serial clock signal can be generated by the SSC internally (master mode), using its own 16-bit baud-rate generator, or can be received from an external master (slave mode). Data width, shift direction, clock polarity and phase are programmable. This allows communication with SPI-compatible devices or devices using other synchronous serial interfaces.

#### **Features**

- Master and slave mode operation
  - Full-duplex or half-duplex operation
- · Transmit and receive buffered
- Flexible data format
  - Programmable number of data bits: 2 to 8 bits
  - Programmable shift direction: LSB or MSB shift first
  - Programmable clock polarity: idle low or high state for the shift clock
  - Programmable clock/data phase: data shift with leading or trailing edge of the shift clock
- Variable baud rate
- Compatible with Serial Peripheral Interface (SPI)
- Interrupt generation
  - On a transmitter empty condition
  - On a receiver full condition
  - On an error condition (receive, phase, baud rate, transmit error)

Data is transmitted or received on lines TXD and RXD, which are normally connected to the pins MTSR (Master Transmit/Slave Receive) and MRST (Master Receive/Slave Transmit). The clock signal is output via line MS\_CLK (Master Serial Shift Clock) or input via line SS\_CLK (Slave Serial Shift Clock). Both lines are normally connected to the pin SCLK. Transmission and reception of data are double-buffered.

Figure 31 shows the block diagram of the SSC.



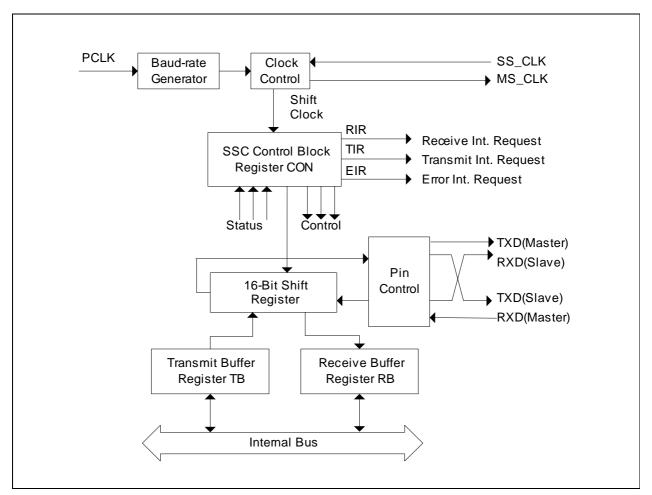


Figure 31 SSC Block Diagram



#### 3.17 Timer 0 and Timer 1

Timer 0 and Timer 1 can function as both timers or counters. When functioning as a timer, Timer 0 and Timer 1 are incremented every machine cycle, i.e. every 2 input clocks (or 2 PCLKs). When functioning as a counter, Timer 0 and Timer 1 are incremented in response to a 1-to-0 transition (falling edge) at their respective external input pins, T0 or T1.

Timer 0 and 1 are fully compatible and can be configured in four different operating modes for use in a variety of applications, see **Table 31**. In modes 0, 1 and 2, the two timers operate independently, but in mode 3, their functions are specialized.

Table 31 Timer 0 and Timer 1 Modes

Mode	Operation
0	13-bit timer The timer is essentially an 8-bit counter with a divide-by-32 prescaler. This mode is included solely for compatibility with Intel 8048 devices.
1	16-bit timer The timer registers, TLx and THx, are concatenated to form a 16-bit counter.
2	8-bit timer with auto-reload The timer register TLx is reloaded with a user-defined 8-bit value in THx upon overflow.
3	Timer 0 operates as two 8-bit timers The timer registers, TL0 and TH0, operate as two separate 8-bit counters. Timer 1 is halted and retains its count even if enabled.



#### 3.18 Timer 2 and Timer 21

Timer 2 and Timer 21 are 16-bit general purpose timers (THL2) that are fully compatible and have two modes of operation, a 16-bit auto-reload mode and a 16-bit one channel capture mode, see **Table 32**. As a timer, the timers count with an input clock of PCLK/12 (if prescaler is disabled). As a counter, they count 1-to-0 transitions on pin T2. In the counter mode, the maximum resolution for the count is PCLK/24 (if prescaler is disabled).

Table 32	Timer 2 Modes
Mode	Description
Auto-reload	<ul> <li>Up/Down Count Disabled</li> <li>Count up only</li> <li>Start counting from 16-bit reload value, overflow at FFFF<sub>H</sub></li> <li>Reload event configurable for trigger by overflow condition only, or by negative/positive edge at input pin T2EX as well</li> <li>Programmble reload value in register RC2</li> <li>Interrupt is generated with reload event</li> </ul>
	<ul> <li>Up/Down Count Enabled</li> <li>Count up or down, direction determined by level at input pin T2EX</li> <li>No interrupt is generated</li> <li>Count up <ul> <li>Start counting from 16-bit reload value, overflow at FFFF<sub>H</sub></li> <li>Reload event triggered by overflow condition</li> <li>Programmble reload value in register RC2</li> </ul> </li> <li>Count down <ul> <li>Start counting from FFFF<sub>H</sub>, underflow at value defined in register RC2</li> <li>Reload event triggered by underflow condition</li> <li>Reload value fixed at FFFF<sub>H</sub></li> </ul> </li> </ul>
Channel capture	<ul> <li>Count up only</li> <li>Start counting from 0000<sub>H</sub>, overflow at FFFF<sub>H</sub></li> <li>Reload event triggered by overflow condition</li> <li>Reload value fixed at 0000<sub>H</sub></li> <li>Capture event triggered by falling/rising edge at pin T2EX</li> <li>Captured timer value stored in register RC2</li> <li>Interrupt is generated with reload or capture event</li> </ul>



## 3.19 Capture/Compare Unit 6

The Capture/Compare Unit 6 (CCU6) provides two independent timers (T12, T13), which can be used for Pulse Width Modulation (PWM) generation, especially for AC-motor control. The CCU6 also supports special control modes for block commutation and multi-phase machines.

The timer T12 can function in capture and/or compare mode for its three channels. The timer T13 can work in compare mode only.

The multi-channel control unit generates output patterns, which can be modulated by T12 and/or T13. The modulation sources can be selected and combined for the signal modulation.

#### **Timer T12 Features**

- Three capture/compare channels, each channel can be used either as a capture or as a compare channel
- Supports generation of a three-phase PWM (six outputs, individual signals for highside and lowside switches)
- 16-bit resolution, maximum count frequency = peripheral clock frequency
- Dead-time control for each channel to avoid short-circuits in the power stage
- Concurrent update of the required T12/13 registers
- Generation of center-aligned and edge-aligned PWM
- Supports single-shot mode
- Supports many interrupt request sources
- Hysteresis-like control mode

#### **Timer T13 Features**

- One independent compare channel with one output
- 16-bit resolution, maximum count frequency = peripheral clock frequency
- Can be synchronized to T12
- Interrupt generation at period-match and compare-match
- Supports single-shot mode

#### **Additional Features**

- Implements block commutation for Brushless DC-drives
- Position detection via Hall-sensor pattern
- Automatic rotational speed measurement for block commutation
- Integrated error handling
- Fast emergency stop without CPU load via external signal (CTRAP)
- Control modes for multi-channel AC-drives
- Output levels can be selected and adapted to the power stage

The block diagram of the CCU6 module is shown in Figure 32.



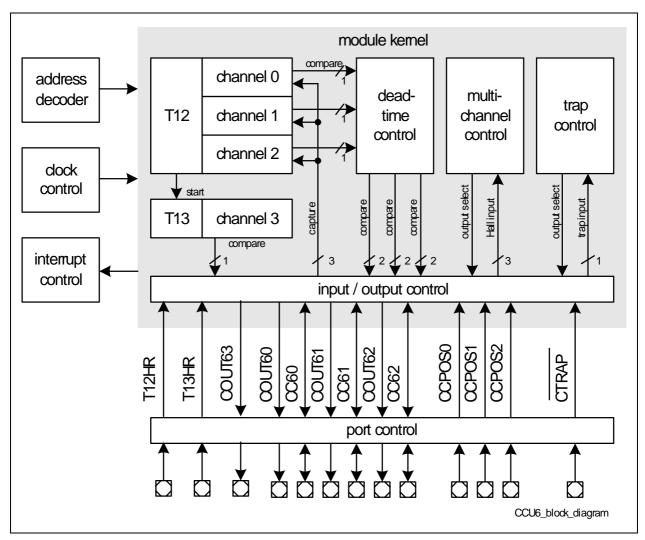


Figure 32 CCU6 Block Diagram



### 3.20 Controller Area Network (MultiCAN)

The MultiCAN module contains two Full-CAN nodes operating independently or exchanging data and remote frames via a gateway function. Transmission and reception of CAN frames is handled in accordance to CAN specification V2.0 B active. Each CAN node can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

Both CAN nodes share a common set of message objects, where each message object may be individually allocated to one of the CAN nodes. Besides serving as a storage container for incoming and outgoing frames, message objects may be combined to build gateways between the CAN nodes or to setup a FIFO buffer.

The message objects are organized in double chained lists, where each CAN node has it's own list of message objects. A CAN node stores frames only into message objects that are allocated to the list of the CAN node. It only transmits messages from objects of this list. A powerful, command driven list controller performs all list operations.

The bit timings for the CAN nodes are derived from the peripheral clock ( $f_{\text{CAN}}$ ) and are programmable up to a data rate of 1 MBaud. A pair of receive and transmit pins connects each CAN node to a bus transceiver.

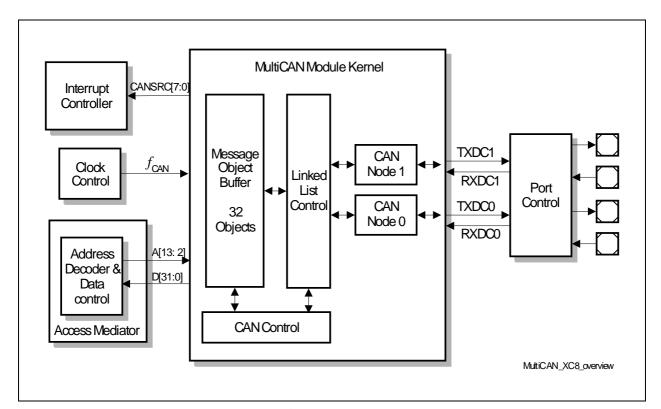


Figure 33 Overview of the MultiCAN

#### **Features**

Compliant to ISO 11898.



- CAN functionality according to CAN specification V2.0 B active.
- Dedicated control registers are provided for each CAN node.
- A data transfer rate up to 1 MBaud is supported.
- Flexible and powerful message transfer control and error handling capabilities are implemented.
- Advanced CAN bus bit timing analysis and baud rate detection can be performed for each CAN node via the frame counter.
- Full-CAN functionality: A set of 32 message objects can be individually
  - allocated (assigned) to any CAN node
  - configured as transmit or receive object
  - setup to handle frames with 11-bit or 29-bit identifier
  - counted or assigned a timestamp via a frame counter
  - configured to remote monitoring mode
- Advanced Acceptance Filtering:
  - Each message object provides an individual acceptance mask to filter incoming frames.
  - A message object can be configured to accept only standard or only extended frames or to accept both standard and extended frames.
  - Message objects can be grouped into 4 priority classes.
  - The selection of the message to be transmitted first can be performed on the basis of frame identifier, IDE bit and RTR bit according to CAN arbitration rules.
- Advanced Message Object Functionality:
  - Message Objects can be combined to build FIFO message buffers of arbitrary size, which is only limited by the total number of message objects.
  - Message objects can be linked to form a gateway to automatically transfer frames between 2 different CAN buses. A single gateway can link any two CAN nodes. An arbitrary number of gateways may be defined.
- Advanced Data Management:
  - The Message objects are organized in double chained lists.
  - List reorganizations may be performed any time, even during full operation of the CAN nodes.
  - A powerful, command driven list controller manages the organization of the list structure and ensures consistency of the list.
  - Message FIFOs are based on the list structure and can easily be scaled in size during CAN operation.
  - Static Allocation Commands offer compatibility with TwinCAN applications, which are not list based.
- Advanced Interrupt Handling:
  - Up to 8 interrupt output lines are available. Most interrupt requests can be individually routed to one of the 8 interrupt output lines.
  - Message postprocessing notifications can be flexibly aggregated into a dedicated register field of 64 notification bits.

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## 3.21 Analog-to-Digital Converter

The SAL-XC886 includes a high-performance 10-bit Analog-to-Digital Converter (ADC) with eight multiplexed analog input channels. The ADC uses a successive approximation technique to convert the analog voltage levels from up to eight different sources. The analog input channels of the ADC are available at Port 2.

### **Features**

- Successive approximation
- 8-bit or 10-bit resolution
   (TUE of ± 1 LSB and ± 2 LSB, respectively)
- Eight analog channels
- Four independent result registers
- Result data protection for slow CPU access (wait-for-read mode)
- Single conversion mode
- Autoscan functionality
- Limit checking for conversion results
- Data reduction filter (accumulation of up to 2 conversion results)
- Two independent conversion request sources with programmable priority
- Selectable conversion request trigger
- Flexible interrupt generation with configurable service nodes
- Programmable sample time
- Programmable clock divider
- Cancel/restart feature for running conversions
- Integrated sample and hold circuitry
- Compensation of offset errors
- Low power modes

# 3.21.1 ADC Clocking Scheme

A common module clock  $f_{\rm ADC}$  generates the various clock signals used by the analog and digital parts of the ADC module:

- $f_{ADCA}$  is input clock for the analog part.
- $f_{\rm ADCI}$  is internal clock for the analog part (defines the time base for conversion length and the sample time). This clock is generated internally in the analog part, based on the input clock  $f_{\rm ADCA}$  to generate a correct duty cycle for the analog components.
- f<sub>ADCD</sub> is input clock for the digital part.

The internal clock for the analog part  $f_{\rm ADCI}$  is limited to a maximum frequency of 10 MHz. Therefore, the ADC clock prescaler must be programmed to a value that ensures  $f_{\rm ADCI}$  does not exceed 10 MHz. The prescaler ratio is selected by bit field CTC in register



GLOBCTR. A prescaling ratio of 32 can be selected when the maximum performance of the ADC is not required.

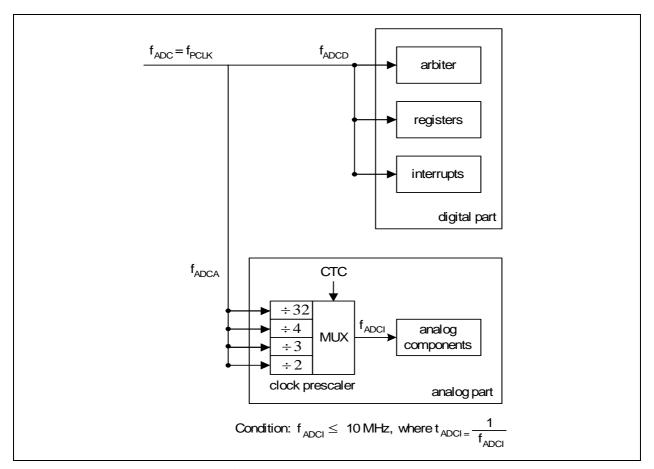


Figure 34 ADC Clocking Scheme

For module clock  $f_{ADC}$  = 20 MHz, the analog clock  $f_{ADCI}$  frequency can be selected as shown in **Table 33**.

Table 33  $f_{ADCI}$  Frequency Selection

${\sf Module\ Clock} f_{\sf ADC}$	СТС	Prescaling Ratio	Analog Clock $f_{ADCI}$
20 MHz	00 <sub>B</sub>	÷ 2	10 MHz
	01 <sub>B</sub>	÷ 3	6.67 MHz
	10 <sub>B</sub>	÷ 4	5 MHz
	11 <sub>B</sub> (default)	÷ 32	625 kHz

As  $f_{\rm ADCI}$  cannot exceed 10 MHz, bit field CTC should not be set to  $00_{\rm B}$  when  $f_{\rm ADC}$  is more than 20 MHz. During slow-down mode where  $f_{\rm ADC}$  may be reduced to 10 MHz, 5 MHz etc., CTC can be set to  $00_{\rm B}$  as long as the divided analog clock  $f_{\rm ADCI}$  does not exceed

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10 MHz. However, it is important to note that the conversion error could increase due to loss of charges on the capacitors, if  $f_{\rm ADC}$  becomes too low during slow-down mode.

# 3.21.2 ADC Conversion Sequence

The analog-to-digital conversion procedure consists of the following phases:

- Synchronization phase (t<sub>SYN</sub>)
- Sample phase (t<sub>S</sub>)
- Conversion phase
- Write result phase  $(t_{WR})$

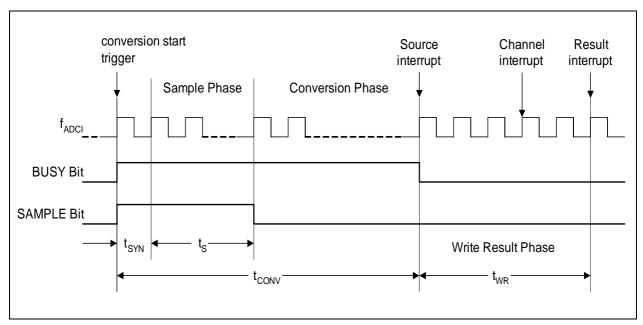


Figure 35 ADC Conversion Timing

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## 3.22 On-Chip Debug Support

The On-Chip Debug Support (OCDS) provides the basic functionality required for the software development and debugging of XC800-based systems.

The OCDS design is based on these principles:

- Use the built-in debug functionality of the XC800 Core
- Add a minimum of hardware overhead
- Provide support for most of the operations by a Monitor Program
- Use standard interfaces to communicate with the Host (a Debugger)

#### **Features**

- Set breakpoints on instruction address and on address range within the Program Memory
- Set breakpoints on internal RAM address range
- Support unlimited software breakpoints in Flash/RAM code region
- Process external breaks via JTAG and upon activating a dedicated pin
- Step through the program code

The OCDS functional blocks are shown in **Figure 36**. The Monitor Mode Control (MMC) block at the center of OCDS system brings together control signals and supports the overall functionality. The MMC communicates with the XC800 Core, primarily via the Debug Interface, and also receives reset and clock signals.

After processing memory address and control signals from the core, the MMC provides proper access to the dedicated extra-memories: a Monitor ROM (holding the code) and a Monitor RAM (for work-data and Monitor-stack).

The OCDS system is accessed through the JTAG<sup>1)</sup>, which is an interface dedicated exclusively for testing and debugging activities and is not normally used in an application. The dedicated MBC pin is used for external configuration and debugging control.

Note: All the debug functionality described here can normally be used only after SAL-XC886 has been started in OCDS mode.

<sup>1)</sup> The pins of the JTAG port can be assigned to either the primary port (Port 0) or either of the secondary ports (Ports 1 and 2/Port 5).

User must set the JTAG pins (TCK and TDI) as input during connection with the OCDS system.



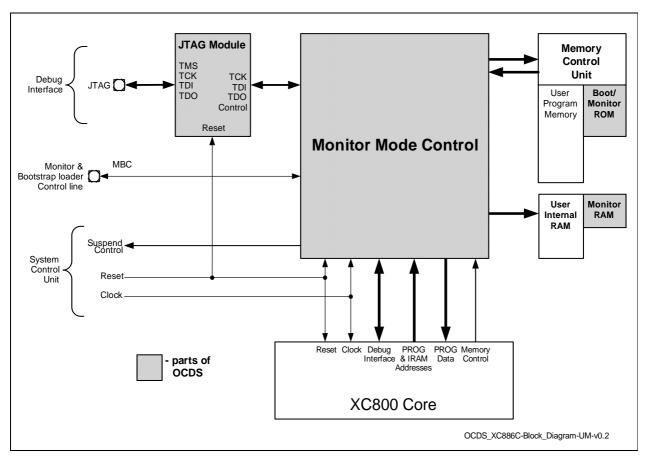


Figure 36 OCDS Block Diagram

# 3.22.1 JTAG ID Register

This is a read-only register located inside the JTAG module, and is used to recognize the device(s) connected to the JTAG interface. Its content is shifted out when INSTRUCTION register contains the IDCODE command (opcode  $04_{\rm H}$ ), and the same is also true immediately after reset.

The JTAG ID register contents for the SAL-XC886 Flash devices are given in Table 34.

Table 34 JTAG ID Summary

Device Type	Device Name	JTAG ID
Flash	SAL-XC886*-8FF	1012 0083 <sub>H</sub>
	SAL-XC886*-6FF	1012 5083 <sub>H</sub>

Note: The asterisk (\*) above denotes all possible device configurations.



# 3.23 Chip Identification Number

The SAL-XC886 identity (ID) register is located at Page 1 of address  $B3_H$ . The value of ID register is  $09_H$ . However, for easy identification of product variants, the Chip Identification Number, which is an unique number assigned to each product variant, is available. The differentiation is based on the product, variant type and device step information.

Two methods are provided to read a device's chip identification number:

- In-application subroutine, GET\_CHIP\_INFO
- Bootstrap loader (BSL) mode A

Table 35 lists the chip identification numbers of available SAL-XC886 device variants.

Table 35 Chip Identification Number

Product Variant	Chip Identification Number		
	AB-Step	AB-Step	AC-Step
XC886CLM-8FFA 5V	-	09900102 <sub>H</sub>	0B900102 <sub>H</sub>
XC886LM-8FFA 5V	-	09900122 <sub>H</sub>	0B900122 <sub>H</sub>
XC886CLM-6FFA 5V	-	09951502 <sub>H</sub>	0B951502 <sub>H</sub>
XC886LM-6FFA 5V	-	09951522 <sub>H</sub>	0B951522 <sub>H</sub>
XC886CM-8FFA 5V	-	09980102 <sub>H</sub>	0B980102 <sub>H</sub>
XC886C-8FFA 5V	-	09980142 <sub>H</sub>	0B980142 <sub>H</sub>
XC886-8FFA 5V	-	09980162 <sub>H</sub>	0B980162 <sub>H</sub>
XC886CM-6FFA 5V	-	099D1502 <sub>H</sub>	0B9D1502 <sub>H</sub>
XC886C-6FFA 5V	-	099D1542 <sub>H</sub>	0B9D1542 <sub>H</sub>
XC886-6FFA 5V	-	099D1562 <sub>H</sub>	0B9D1562 <sub>H</sub>



### 4 Electrical Parameters

**Chapter 4** provides the characteristics of the electrical parameters which are implementation-specific for the SAL-XC886.

#### 4.1 General Parameters

The general parameters are described here to aid the users in interpreting the parameters mainly in **Section 4.2** and **Section 4.3**.

### 4.1.1 Parameter Interpretation

The parameters listed in this section represent partly the characteristics of the SAL-XC886 and partly its requirements on the system. To aid interpreting the parameters easily when evaluating them for a design, they are indicated by the abbreviations in the "Symbol" column:

#### • CC

These parameters indicate **C**ontroller **C**haracteristics, which are distinctive features of the SAL-XC886 and must be regarded for a system design.

#### SR

These parameters indicate **S**ystem **R**equirements, which must be provided by the microcontroller system in which the SAL-XC886 is designed in.



### 4.1.2 Absolute Maximum Rating

Maximum ratings are the extreme limits to which the SAL-XC886 can be subjected to without permanent damage.

Table 36 Absolute Maximum Rating Parameters

Parameter	Symbol	Lim	it Values	Unit	Notes	
		min.	max.			
Ambient temperature	$T_{A}$	-40	150	°C	under bias	
Storage temperature	$T_{ST}$	-65	150	°C	1)	
Junction temperature	$T_{J}$	-40	160	°C	under bias <sup>1)</sup>	
$\begin{tabular}{lll} \hline & Voltage on power supply pin with \\ & respect to $V_{\rm SS}$ \\ \hline \end{tabular}$	$V_{DDP}$	-0.5	6	V	1)	
Voltage on any pin with respect to $V_{\rm SS}$	$V_{IN}$	-0.5	$V_{\rm DDP}$ + 0.5 or max. 6	V	whichever is lower <sup>1)</sup>	
Input current on any pin during overload condition	$I_{IN}$	-10	10	mA	1)	
Absolute sum of all input currents during overload condition	$\Sigma  I_{IN} $	_	50	mA	1)	

<sup>1)</sup> Not subjected to production test, verified by design/characterization.

Note: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During absolute maximum rating overload conditions  $(V_{IN} > V_{DDP})$  or  $V_{IN} < V_{SS}$  the voltage on  $V_{DDP}$  pin with respect to ground  $(V_{SS})$  must not exceed the values defined by the absolute maximum ratings.



# 4.1.3 Operating Conditions

The following operating conditions must not be exceeded in order to ensure correct operation of the SAL-XC886. All parameters mentioned in the following table refer to these operating conditions, unless otherwise noted.

**Table 37 Operating Condition Parameters** 

Parameter	Symbol	Limit Values		Unit	Notes/
		min.	max.		Conditions
Digital power supply voltage	$V_{DDP}$	4.5	5.5	V	5V Device
Digital ground voltage	$V_{SS}$	0		V	
Digital core supply voltage	$V_{DDC}$	2.3	2.7	V	
System Clock Frequency <sup>1)</sup>	$f_{\sf SYS}$	74.0	86.0	MHz	
Ambient temperature	$T_{A}$	-40	150	°C	SAL-XC886

<sup>1)</sup>  $f_{SYS}$  is the PLL output clock. During normal operating mode, CPU clock is  $f_{SYS}$  / 4. Please refer to **Figure 25** for detailed description.

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### 4.2 DC Parameters

The electrical characteristics of the DC Parameters are detailed in this section.

# 4.2.1 Input/Output Characteristics

Table 38 provides the characteristics of the input/output pins of the SAL-XC886.

 Table 38
 Input/Output Characteristics (Operating Conditions apply)

Parameter	Symbol		Limit	Values	Unit	Test Conditions
			min.	max.		
$V_{\rm DDP}$ = 5 V Range						
Output low voltage	$V_{OL}$	CC	_	1.0	V	I <sub>OL</sub> = 15 mA
			_	1.0	V	$I_{\rm OL}$ = 5 mA, current into all pins > 60 mA
			_	0.4	V	$I_{\rm OL}$ = 5 mA, current into all pins $\leq$ 60 mA
Output high voltage	$V_{OH}$	CC	V <sub>DDP</sub> - 1.0	_	V	I <sub>OH</sub> = -15 mA
			V <sub>DDP</sub> - 1.0	_	V	$I_{\rm OH}$ = -5 mA, current from all pins > 60 mA
			V <sub>DDP</sub> - 0.4	_	V	$I_{\rm OH}$ = -5 mA, current from all pins $\leq$ 60 mA
Input low voltage on port pins (all except P0.0 & P0.1)	$V_{ILP}$	SR	_	$0.3  imes V_{DDP}$	V	CMOS Mode
Input low voltage on P0.0 & P0.1	$V_{ILP0}$	SR	-0.2	$0.3  imes V_{ m DDP}$	V	CMOS Mode
Input low voltage on RESET pin	$V_{ILR}$	SR	_	$0.3  imes V_{ m DDP}$	V	CMOS Mode
Input low voltage on TMS pin	$V_{ILT}$	SR	_	$0.3  imes V_{ m DDP}$	V	CMOS Mode
Input high voltage on port pins (all except P0.0 & P0.1)	$V_{IHP}$	SR	$0.7  imes V_{ m DDP}$	_	V	CMOS Mode
Input high voltage on P0.0 & P0.1	$V_{IHP0}$	SR	$0.7  imes V_{ extsf{DDP}}$	$V_{DDP}$	V	CMOS Mode



Table 38 Input/Output Characteristics (Operating Conditions apply) (cont'd)

Parameter	Symbol		Limit	Values	Unit	Test Conditions	
				max.			
Input high voltage on RESET pin	$V_{IHR}$	SR	$0.7 imes V_{ extsf{DDP}}$	_	V	CMOS Mode	
Input high voltage on TMS pin	$V_{IHT}$	SR	$0.75  imes V_{ extsf{DDP}}$	_	V	CMOS Mode	
Input Hysteresis on port pins	HYSP	CC	$\begin{matrix} 0.07 \times \\ V_{\rm DDP} \end{matrix}$	_	V	CMOS Mode <sup>1)</sup>	
Input Hysteresis on XTAL1	HYSX	CC	$0.07  imes V_{ m DDC}$	_	V	1)	
Input low voltage at XTAL1	$V_{ILX}$	SR	V <sub>SS</sub> - 0.5	$0.3  imes V_{ extsf{DDC}}$	V		
Input high voltage at XTAL1	$V_{IHX}$	SR	$0.7  imes V_{ extsf{DDC}}$	$V_{\rm DDC}$ + 0.5	V		
Pull-up current	$I_{PU}$	SR	_	-10	μΑ	$V_{IHP,min}$	
			-150	_	μΑ	$V_{ILP,max}$	
Pull-down current	$I_{PD}$	SR	_	10	μΑ	$V_{ILP,max}$	
			150	_	μΑ	$V_{IHP,min}$	
Input leakage current	$I_{OZ1}$	CC	-2	2	μΑ	$0 < V_{IN} < V_{DDP},$ $T_A \le 150^{\circ}C^{2)}$	
Input current at XTAL1	$I_{ILX}$	CC	-10	10	μΑ		
Overload current on any pin	$I_{OV}$	SR	-5	5	mA	3)	
Absolute sum of overload currents	$\Sigma  I_{OV} $	SR	_	25	mA	3)	
Voltage on any pin during $V_{\rm DDP}$ power off	$V_{PO}$	SR	_	0.3	V	4)	
Maximum current per pin (excluding $V_{\rm DDP}$ and $V_{\rm SS}$ )	$I_{M}$ SR	SR	_	15	mA		
Maximum current for all pins (excluding $V_{\rm DDP}$ and $V_{\rm SS}$ )	$\Sigma  I_{M} $	SR	_	90	mA		



Table 38 Input/Output Characteristics (Operating Conditions apply) (cont'd)

Parameter	Symbol		Limit '	Values	Unit	Test Conditions
			min.	max.		
	$I_{MVDDP}$	SR	_	120	mA	3)
Maximum current out of $V_{\rm SS}$	$I_{MVSS}$	SR	_	120	mA	3)

- Not subjected to production test, verified by design/characterization. Hysteresis is implemented to avoid meta stable states and switching due to internal ground bounce. It cannot be guaranteed that it suppresses switching due to external system noise.
- 2) An additional error current ( $I_{INJ}$ ) will flow if an overload current flows through an adjacent pin. TMS pin and RESET pin have internal pull devices and are not included in the input leakage current characteristic.
- 3) Not subjected to production test, verified by design/characterization.
- 4) Not subjected to production test, verified by design/characterization. However, for applications with strict low power-down current requirements, it is mandatory that no active voltage source is supplied at any GPIO pin when  $V_{\rm DDP}$  is powered off.

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# 4.2.2 Supply Threshold Characteristics

Table 39 provides the characteristics of the supply threshold in the SAL-XC886.

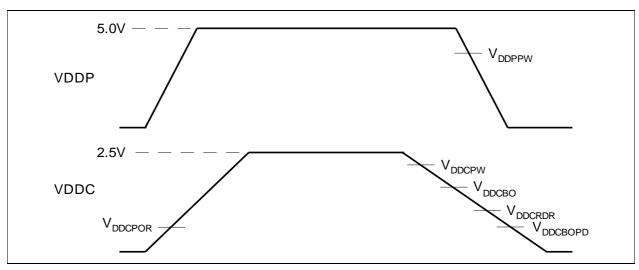


Figure 37 Supply Threshold Parameters

**Table 39** Supply Threshold Parameters (Operating Conditions apply)

	•	•	_	• •	• /	
Parameters	Symbol		I	ues	Unit	
			min.	typ.	max.	
$\overline{V_{\rm DDC}}$ prewarning voltage <sup>1)</sup>	$V_{DDCPW}$	CC	2.2	2.3	2.4	V
$\overline{V_{\mathrm{DDC}}}$ brownout voltage in active $\mathrm{mode^{1)}}$	$V_{DDCBO}$	СС	2.0	2.1	2.2	V
RAM data retention voltage	$V_{\mathtt{DDCRDR}}$	CC	0.9	1.0	1.1	V
$\overline{V_{\mathrm{DDC}}}$ brownout voltage in power-down mode <sup>2)</sup>	$V_{\rm DDCBOPD}$	CC	1.3	1.5	1.7	V
$\overline{V_{\text{DDP}}}$ prewarning voltage <sup>3)</sup>	$V_{DDPPW}$	CC	3.4	4.0	4.6	V
Power-on reset voltage <sup>2)4)</sup>	$V_{DDCPOR}$	CC	1.3	1.5	1.7	V

<sup>1)</sup> Detection is disabled in power-down mode.

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<sup>2)</sup> Detection is enabled in both active and power-down mode.

<sup>3)</sup> Detection is enabled for external power supply of 5.0V.

<sup>4)</sup> The reset of EVR is extended by 300 μs typically after the VDDC reaches the power-on reset voltage.



#### 4.2.3 ADC Characteristics

The values in the table below are given for an analog power supply between 4.5 V to 5.5 V. All ground pins ( $V_{\rm SS}$ ) must be externally connected to one single star point in the system. The voltage difference between the ground pins must not exceed 200mV.

Table 40 ADC Characteristics (Operating Conditions apply;  $V_{\text{DDP}}$  = 5V Range)

Parameter	Symbol		Lir	nit Val	ues	Unit	Test Conditions/
			min.	typ.	max.		Remarks
Analog reference voltage	$V_{AREF}$	SR	V <sub>AGND</sub> + 1	$V_{DDP}$	V <sub>DDP</sub> + 0.05	V	1)
Analog reference ground	$V_{AGND}$	SR	V <sub>SS</sub> - 0.05	$V_{SS}$	V <sub>AREF</sub>	V	1)
Analog input voltage range	$V_{AIN}$	SR	$V_{AGND}$	_	$V_{AREF}$	V	
ADC clocks	$f_{ADC}$		_	20	25.8	MHz	module clock <sup>1)</sup>
	$f_{ADCI}$		_	_	10	MHz	internal analog clock <sup>1)</sup> See <b>Figure 34</b>
Sample time	$t_{\rm S}$	CC	$(2 + INPCR0.STC) \times t_{ADCI}$			μS	1)
Conversion time	$t_{C}$	CC	See Se	ection	4.2.3.1	μS	1)
Total unadjusted	TUE	CC	_	_	1	LSB	8-bit conversion <sup>2)</sup>
error			_	_	2	LSB	10-bit conversion <sup>2)</sup>
Differential Nonlinearity	/EA <sub>DNL</sub>	CC	_	1	_	LSB	10-bit conversion <sup>1)</sup>
Integral Nonlinearity	/EA <sub>INL</sub>	CC	_	1	_	LSB	10-bit conversion <sup>1)</sup>
Offset	/EA <sub>OFF</sub>	CC	_	1	_	LSB	10-bit conversion <sup>1)</sup>
Gain	$/EA_{GAIN}$	CC	_	1	_	LSB	10-bit conversion <sup>1)</sup>
Overload current coupling factor for	$K_{OVA}$	CC	_	_	1.0 x 10 <sup>-4</sup>	_	$I_{\rm OV} > 0^{1)3)}$
analog inputs			_	_	1.5 x 10 <sup>-3</sup>	_	$I_{\rm OV} < 0^{1)3)}$



Table 40 ADC Characteristics (Operating Conditions apply;  $V_{\text{DDP}}$  = 5V Range) (cont'd)

Parameter	Symbol		Liı	mit Val	ues	Unit	Test Conditions/	
			min.	typ.	max.		Remarks	
Overload current coupling factor for	$K_{OVD}$	CC	_	_	5.0 x 10 <sup>-3</sup>	_	$I_{\rm OV} > 0^{1)3)}$	
digital I/O pins			_	_	1.0 x 10 <sup>-2</sup>	_	$I_{\rm OV} < 0^{1)3)}$	
Switched capacitance at the reference voltage input	$C_{AREFSW}$	CC	_	10	20	pF	1)4)	
Switched capacitance at the analog voltage inputs	$C_{AINSW}$	CC	_	5	7	pF	1)5)	
Input resistance of the reference input	$R_{AREF}$	CC	_	1	2	kΩ	1)	
Input resistance of the selected analog channel	R <sub>AIN</sub>	CC	_	1	1.5	kΩ	1)	

- 1) Not subjected to production test, verified by design/characterization
- 2) TUE is tested at  $V_{\rm AREF}$  = 5.0 V,  $V_{\rm AGND}$  = 0 V,  $V_{\rm DDP}$  = 5.0 V.
- 3) An overload current  $(I_{\text{OV}})$  through a pin injects a certain error current  $(I_{\text{INJ}})$  into the adjacent pins. This error current adds to the respective pin's leakage current  $(I_{\text{OZ}})$ . The amount of error current depends on the overload current and is defined by the overload coupling factor  $K_{\text{OV}}$ . The polarity of the injected error current is inverse compared to the polarity of the overload current that produces it. The total current through a pin is  $|I_{\text{TOT}}| = |I_{\text{OZ1}}| + (|I_{\text{OV}}| \times K_{\text{OV}})$ . The additional error current may distort the input voltage on analog inputs.
- 4) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead of this, smaller capacitances are successively switched to the reference voltage.
- 5) The sampling capacity of the conversion C-Network is pre-charged to  $V_{\rm AREF}/2$  before connecting the input to the C-Network. Because of the parasitic elements, the voltage measured at ANx is lower than  $V_{\rm AREF}/2$ .

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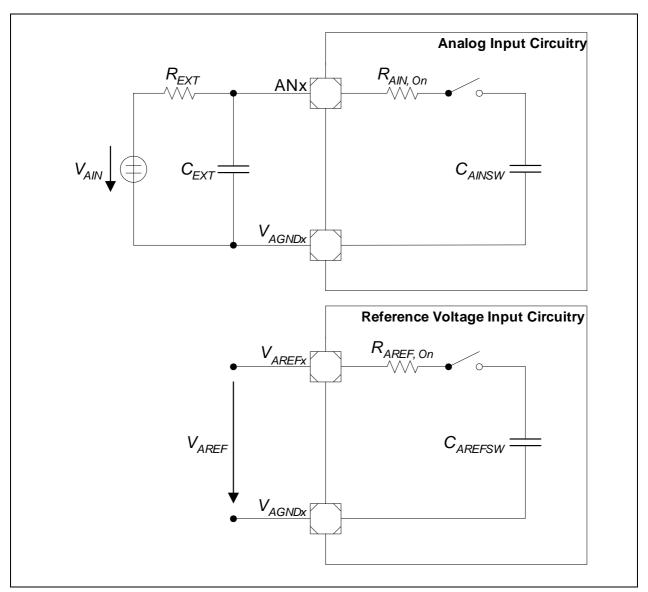


Figure 38 ADC Input Circuits



# 4.2.3.1 ADC Conversion Timing

Conversion time,  $t_{\rm C} = t_{\rm ADC} \times$  (1 + r × (3 + n + STC)), where r = CTC + 2 for CTC = 00<sub>B</sub>, 01<sub>B</sub> or 10<sub>B</sub>, r = 32 for CTC = 11<sub>B</sub>, CTC = Conversion Time Control (GLOBCTR.CTC), STC = Sample Time Control (INPCR0.STC), n = 8 or 10 (for 8-bit and 10-bit conversion respectively),  $t_{\rm ADC} = 1/f_{\rm ADC}$ 



### 4.2.4 Power Supply Current

Table 41 and Table 42 provide the characteristics of the power supply current in the SAL-XC886.

Table 41 Power Supply Current Parameters (Operating Conditions apply;  $V_{\rm DDP}$  = 5V range)

Parameter	Symbol	Limit	Values	Unit	<b>Test Condition</b>	
		typ. <sup>1)</sup>	max. <sup>2)</sup>			
$V_{\rm DDP}$ = 5V Range						
Active Mode	$I_{DDP}$	22.9	29.9	mA	3)	
Idle Mode	$I_{DDP}$	17.8	23.7	mA	4)	
Active Mode with slow-down enabled	$I_{DDP}$	12.0	16.6	mA	5)	
Idle Mode with slow-down enabled	$I_{DDP}$	10.0	14.2	mA	6)	

- 1) The typical  $I_{\rm DDP}$  values are periodically measured at  $T_{\rm A}$  = + 25 °C and  $V_{\rm DDP}$  = 5.0 V.
- 2) The maximum  $I_{\rm DDP}$  values are measured under worst case conditions ( $T_{\rm A}$  = + 150 °C and  $V_{\rm DDP}$  = 5.5 V).
- 3)  $I_{\text{DDP}}$  (active mode) is measured with: CPU clock and input clock to all peripherals running at 20 MHz(set by on-chip oscillator of 10 MHz and NDIV in PLL\_CON to 1001<sub>B</sub>),  $\overline{\text{RESET}} = V_{\text{DDP}}$ , no load on ports.
- 4)  $I_{\text{DDP}}$  (idle mode) is measured with: CPU clock disabled, watchdog timer disabled, input clock to all peripherals enabled and running at 20 MHz,  $\overline{\text{RESET}} = V_{\text{DDP}}$ , no load on ports.
- 5)  $I_{\text{DDP}}$  (active mode with slow-down mode) is measured with: <u>CPU</u> clock and input clock to all peripherals running at 2.5 MHz by setting CLKREL in CMCON to 0110<sub>B</sub>, RESET =  $V_{\text{DDP}}$ , no load on ports.
- 6) $I_{\rm DDP}$  (idle mode with slow-down mode) is measured with: CPU clock disabled, watchdog timer disabled, input clock to all peripherals enabled and running at 2.5 MHz by setting CLKREL in CMCON to 0110<sub>B</sub>,  $\overline{\rm RESET} = V_{\rm DDP}$ , no load on ports.

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Table 42 Power Down Current (Operating Conditions apply;  $V_{\text{DDP}} = 5V$  range)

Parameter	Symbol	Limit Values		Unit	Test Condition
		typ. <sup>1)</sup>	typ. <sup>1)</sup> max. <sup>2)</sup>		
$V_{DDP}$ = 5V Range		•		•	•
Power-Down Mode	$I_{PDP}$	1	10	μА	$T_{\rm A}$ = + 25 °C <sup>3)4)</sup>
		-	30	μΑ	$T_{\rm A} = + 85  {}^{\circ}{\rm C}^{4)5)}$

<sup>1)</sup> The typical  $I_{\rm PDP}$  values are measured at  $V_{\rm DDP}$  = 5.0 V.

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<sup>2)</sup> The maximum  $I_{\rm PDP}$  values are measured at  $V_{\rm DDP}$  = 5.5 V.

<sup>3)</sup> $I_{PDP}$  has a maximum value of 500  $\mu$ A at  $T_A$  = + 150 °C.

<sup>4)</sup>  $I_{\text{PDP}}$  is measured with:  $\overline{\text{RESET}} = V_{\text{DDP}}$ ,  $V_{\text{AGND}} = V_{\text{SS}}$ , RXD/INT0 =  $V_{\text{DDP}}$ ; rest of the ports are programmed to be input with either internal pull devices enabled or driven externally to ensure no floating inputs.

<sup>5)</sup> Not subjected to production test, verified by design/characterization.



#### 4.3 AC Parameters

The electrical characteristics of the AC Parameters are detailed in this section.

### 4.3.1 Testing Waveforms

The testing waveforms for rise/fall time, output delay and output high impedance are shown in Figure 39, Figure 40 and Figure 41.

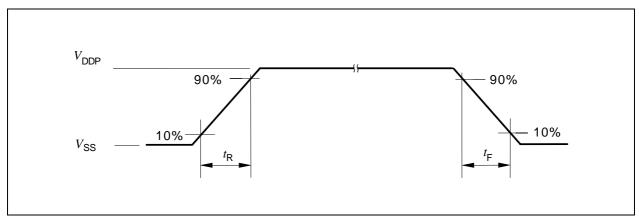


Figure 39 Rise/Fall Time Parameters

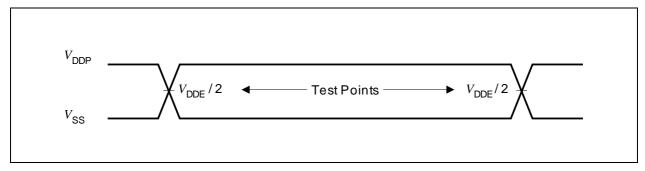


Figure 40 Testing Waveform, Output Delay

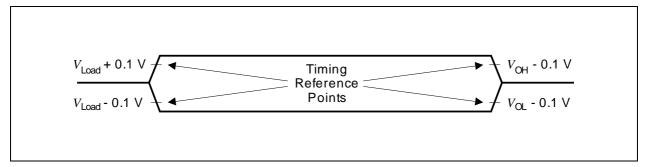


Figure 41 Testing Waveform, Output High Impedance



# 4.3.2 Output Rise/Fall Times

Table 43 provides the characteristics of the output rise/fall times in the SAL-XC886.

Table 43 Output Rise/Fall Times Parameters (Operating Conditions apply)

Parameter	Symbol		Limit Values												Test Conditions	
		min.	max.													
$\overline{V_{\text{DDP}}}$ = 5V Range		•	•	•												
Rise/fall times	t <sub>R</sub> , t <sub>F</sub>	_	10	ns	20 pF. <sup>1)2)3)</sup>											

- 1) Rise/Fall time measurements are taken with 10% 90% of pad supply.
- 2) Not all parameters are 100% tested, but are verified by design/characterization and test correlation.
- 3) Additional rise/fall time valid for  $C_{\rm L}$  = 20pF 100pF @ 0.125 ns/pF.

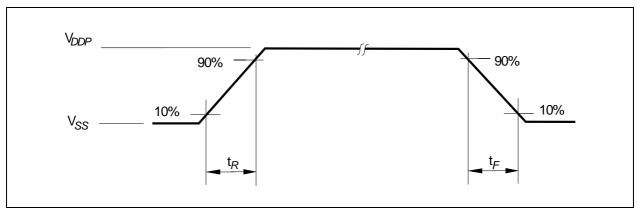


Figure 42 Rise/Fall Times Parameters



# 4.3.3 Power-on Reset and PLL Timing

**Table 47** provides the characteristics of the power-on reset and PLL timing in the SAL-XC886.

Table 44 Power-On Reset and PLL Timing (Operating Conditions apply)

Parameter	Symbol		Limit Values			Unit	Test Conditions
			min.	typ.	max.		
Pad operating voltage	$V_{PAD}$	CC	2.3	_	_	V	1)
On-Chip Oscillator start-up time	$t_{OSCST}$	CC	_	_	500	ns	1)
Flash initialization time	$t_{\sf FINIT}$	CC	_	160	_	μS	1)
RESET hold time	$t_{RST}$	SR	_	500	_	μS	$V_{\rm DDP}$ rise time $(10\% - 90\%) \le 500 \mu s^{1)2)}$
PLL lock-in in time	$t_{LOCK}$	CC	_	_	200	μS	1)
PLL accumulated jitter	$D_{P}$		_	_	0.7	ns	1)3)

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>2)</sup>  $\overline{\text{RESET}}$  signal has to be active (low) until  $V_{\text{DDC}}$  has reached 90% of its maximum value (typ. 2.5 V).

<sup>3)</sup> PLL lock at 80 MHz using a 4 MHz external oscillator. The PLL Divider settings are K = 2, N = 40 and P = 1.



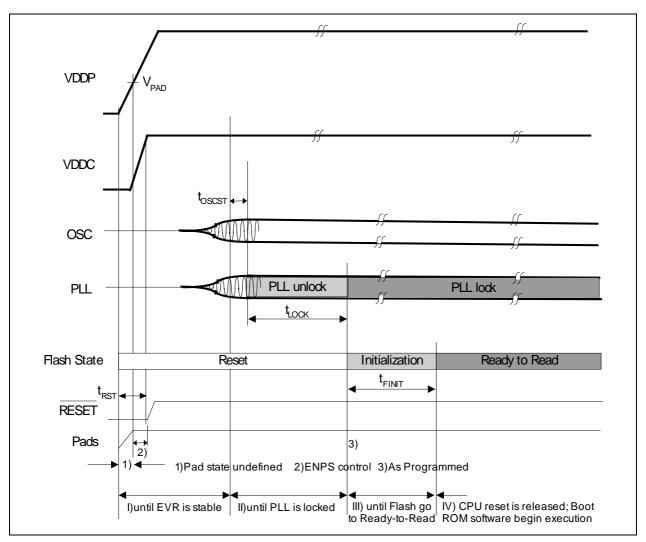


Figure 43 Power-on Reset Timing



# 4.3.4 On-Chip Oscillator Characteristics

Table 45 provides the characteristics of the on-chip oscillator in the SAL-XC886.

**Table 45** On-chip Oscillator Characteristics (Operating Conditions apply)

Parameter	Sym	bol	Lir	nit Val	ues	Unit	Test Conditions
			min.	typ.	max.		
Nominal frequency	$f_{NOM}$	CC	9.75	10	10.25	MHz	under nominal conditions <sup>1)</sup>
Long term frequency deviation	$\Delta f_{LT}$	CC	0	_	6.0	%	with respect to $f_{\rm NOM}$ , over lifetime and temperature (125°C to 150°C), for one given device after trimming
			-5.0	_	5.0	%	with respect to $f_{\rm NOM}$ , over lifetime and temperature (-10°C to 125°C), for one given device after trimming
			-6.0	_	0	%	with respect to $f_{\text{NOM}}$ , over lifetime and temperature (-40°C to -10°C), for one given device after trimming
Short term frequency deviation	$\Delta f_{ST}$	CC	-1.0	_	1.0	%	within one LIN message (<10 ms 100 ms)

<sup>1)</sup> Nominal condition:  $V_{\rm DDC}$  = 2.5 V,  $T_{\rm A}$  = + 25°C.



#### 4.3.5 External Clock Drive XTAL1

**Table 46** shows the parameters that define the external clock supply for SAL-XC886. These timing parameters are based on the direct XTAL1 drive of clock input signals. They are not applicable if an external crystal or ceramic resonator is considered.

Table 46 External Clock Drive Characteristics (Operating Conditions apply)

Parameter	ameter Symbol			t Values	Unit	Test Conditions	
			Min.	Max.			
Oscillator period	$t_{\rm osc}$	SR	100	250	ns	1)2)	
High time	<i>t</i> <sub>1</sub>	SR	25	-	ns	2)3)	
Low time	$t_2$	SR	25	-	ns	2)3)	
Rise time	$t_3$	SR	-	20	ns	2)3)	
Fall time	$t_4$	SR	-	20	ns	2)3)	

- 1) The clock input signals with 45-55% duty cycle are used.
- 2) Not all parameters are 100% tested, but are verified by design/characterization and test correlation.
- 3) The clock input signal must reach the defined levels  $V_{\rm ILX}$  and  $V_{\rm IHX}$ .

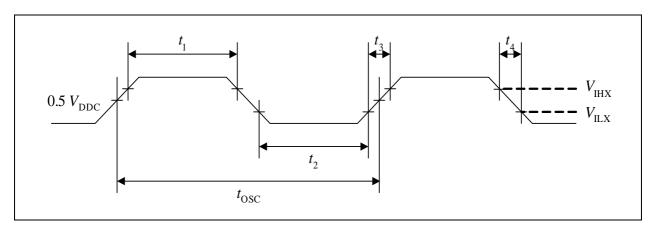


Figure 44 External Clock Drive XTAL1



# 4.3.6 JTAG Timing

Table 47 provides the characteristics of the JTAG timing in the SAL-XC886.

Table 47 TCK Clock Timing (Operating Conditions apply; CL = 50 pF)

Parameter	Sym	Symbol		nits	Unit	<b>Test Conditions</b>
			min	max		
TCK clock period	$t_{TCK}$	SR	50	-	ns	1)
TCK high time	<i>t</i> <sub>1</sub>	SR	20	_	ns	1)
TCK low time	$t_2$	SR	20	-	ns	1)
TCK clock rise time	$t_3$	SR	-	4	ns	1)
TCK clock fall time	$t_4$	SR	-	4	ns	1)

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

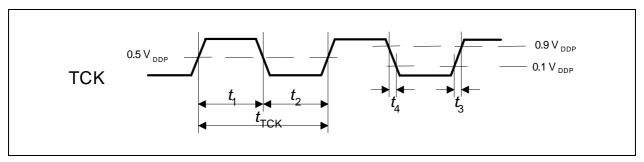


Figure 45 TCK Clock Timing

Table 48 JTAG Timing (Operating Conditions apply; CL = 50 pF)

Parameter	Symbol		Lir	nits	Unit	Test	
			min	max		Conditions	
TMS setup to TCK	<i>t</i> <sub>1</sub>	SR	8	-	ns	1)	
TMS hold to TCK	$t_2$	SR	24	-	ns	1)	
TDI setup to TCK  _√	$t_1$	SR	11	-	ns	1)	
TDI hold to TCK	$t_2$	SR	24	-	ns	1)	
TDO valid output from TCK	$t_3$	CC	-	27	ns	1)	

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Table 48 JTAG Timing (Operating Conditions apply; CL = 50 pF) (cont'd)

Parameter	Symbol		Lir	nits	Unit	Test	
			min	max		Conditions	
TDO high impedance to valid output from TCK	$t_4$	CC	-	35	ns	1)	
TDO valid output to high impedance from TCK	<i>t</i> <sub>5</sub>	CC	-	27	ns	1)	

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

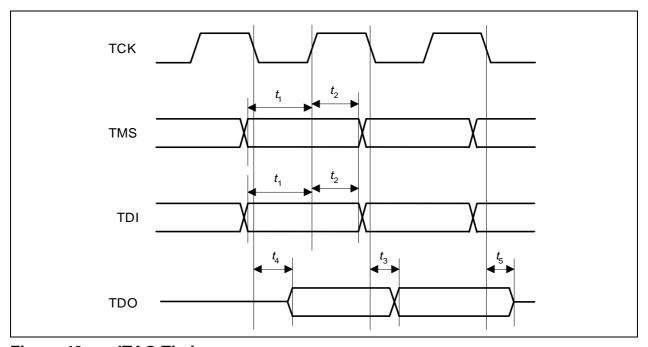


Figure 46 JTAG Timing

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# 4.3.7 SSC Master Mode Timing

Table 49 provides the characteristics of the SSC timing in the SAL-XC886.

Table 49 SSC Master Mode Timing (Operating Conditions apply; CL = 50 pF)

Parameter	Syn	nbol	Limit	t Values	Unit	Test
			min.	max.		Conditions
SCLK clock period	$t_0$	CC	2*T <sub>SSC</sub>	_	ns	1)2)
MTSR delay from SCLK	$t_1$	CC	0	8	ns	2)
MRST setup to SCLK	$t_2$	SR	24	_	ns	2)
MRST hold from SCLK	$t_3$	SR	0	_	ns	2)

<sup>1)</sup> $T_{SSCmin} = T_{CPU} = 1/f_{CPU}$ . When  $f_{CPU} = 20$  MHz,  $t_0 = 100$ ns.  $T_{CPU}$  is the CPU clock period.

<sup>2)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

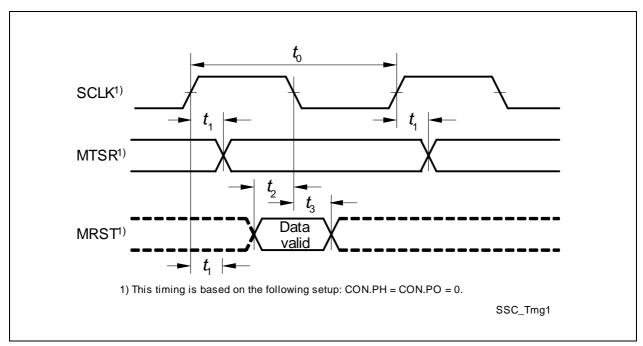


Figure 47 SSC Master Mode Timing

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### **Package and Quality Declaration**

# 5 Package and Quality Declaration

**Chapter 5** provides the information of the SAL-XC886 package and reliability section.

### 5.1 Package Parameters

**Table 50** provides the thermal characteristics of the package used in SAL-XC886.

Table 50 Thermal Characteristics of the Packages

Parameter	Symbol	Lim	nit Values	Unit	Notes	
	Min.		Max.			
PG-TQFP-48	1	1	1	•	-	
Thermal resistance junction case	$R_{TJC}$ CC	-	11.6	K/W	1)2)	
Thermal resistance junction lead	$R_{TJL}$ CC	-	33.2	K/W	1)2)	

<sup>1)</sup> The thermal resistances between the case and the ambient  $(R_{\mathsf{TCA}})$ , the lead and the ambient  $(R_{\mathsf{TLA}})$  are to be combined with the thermal resistances between the junction and the case  $(R_{\mathsf{TJC}})$ , the junction and the lead  $(R_{\mathsf{TJL}})$  given above, in order to calculate the total thermal resistance between the junction and the ambient  $(R_{\mathsf{TJA}})$ . The thermal resistances between the case and the ambient  $(R_{\mathsf{TCA}})$ , the lead and the ambient  $(R_{\mathsf{TLA}})$  depend on the external system (PCB, case) characteristics, and are under user responsibility.

- a) simply adding only the two thermal resistances (junction lead and lead ambient), or
- b) by taking all four resistances into account, depending on the precision needed.
- 2) Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

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The junction temperature can be calculated using the following equation:  $T_J = T_A + R_{TJA} \times P_D$ , where the  $R_{TJA}$  is the total thermal resistance between the junction and the ambient. This total junction ambient resistance  $R_{TJA}$  can be obtained from the upper four partial thermal resistances, by



#### **Package and Quality Declaration**

# 5.2 Package Outline

Figure 48 shows the package outlines of the SAL-XC886.

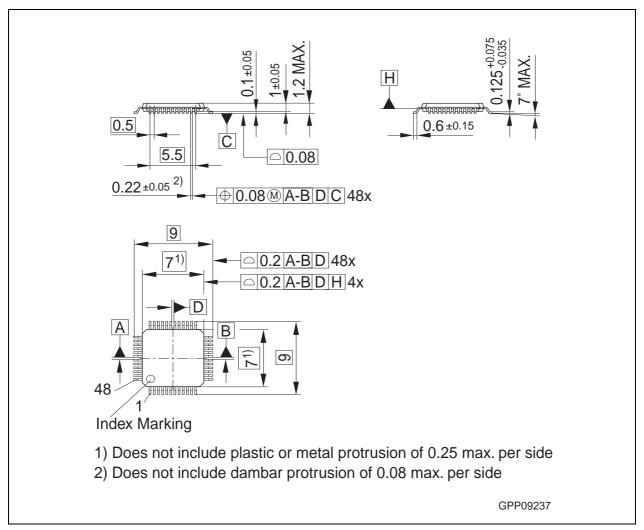


Figure 48 PG-TQFP-48 Package Outline

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# **Package and Quality Declaration**

# 5.3 Quality Declaration

Table 51 shows the characteristics of the quality parameters in the SAL-XC886.

Table 51 Quality Parameters

Parameter	Symbol	Limit	Values		Unit	Notes
		Min.	Тур.	Max.		
Operation Lifetime	$t_{OP}$	-	-	500	hours	$T_{\rm A} = 150^{\circ}{\rm C}^{2)}$
when the device is used at the five stated $T_A^{-1}$		-	-	1000	hours	$T_{\rm A} = 140^{\circ}{\rm C}^{2)}$
at the live stated TA		-	-	2000	hours	$T_{\rm A} = 125^{\circ}{\rm C}^{2)}$
		-	-	10000	hours	$T_{\rm A} = 85^{\circ}{\rm C}^{2)}$
		-	-	1500	hours	$T_{\rm A} = -40^{\circ}{\rm C}^{2)}$
Operation Lifetime	t <sub>OP2</sub>	-	-	18000	hours	$T_{\rm A} = 108^{\circ}{\rm C}^{2)}$
when the device is used at the two stated $T_A^{1)}$		-	-	130000	hours	$T_{\rm A} = 27^{\circ}{\rm C}^{2)}$
Weighted Average Temperature <sup>3)</sup>	$T_{WA}$	-	107	-	°C	for 15000 hours <sup>2)</sup>
ESD susceptibility according to Human Body Model (HBM) for all pins (except $V_{\rm DDC}$ )	$V_{HBM}$	-	-	2000	V	Conforming to EIA/JESD22- A114-B <sup>2)</sup>
ESD susceptibility according to Charged Device Model (CDM) pins	$V_{CDM}$	-	-	750	V	Conforming to JESD22-C101-C <sup>2)</sup>

<sup>1)</sup> This lifetime refers only to the time when the device is powered-on.

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<sup>2)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>3)</sup> This parameter is derived based on the Arrhenius model.

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