

# HIGH RIPPLE-REJECTION LOW DROPOUT HIGH OUTPUT CURRENT CMOS VOLTAGE REGULATOR

## S-1155 Series

The S-1155 Series, developed based on CMOS technology, is a positive voltage regulator with a super low dropout voltage, high output voltage accuracy, and low current consumption.

The S-1155 Series provides the very small dropout voltage and the large output current due to the built-in transistor with low on-resistance. The overcurrent protector prevents the load current from exceeding the capacitance of output transistor. The thermal shutdown circuit prevents damage caused by heat, the rush current control circuit limits the excessive rush current during start-up.

The ON/OFF circuit ensures longer battery life. Various capacitors, also small ceramic capacitors, can be used for this IC more than for the conventional regulator ICs which have CMOS technology.

The small package SOT-89-5 enables high-density mounting.

### ■ Features

- Output voltage : 1.0 to 5.0 V, selectable in 0.05 V step.
- Low equivalent series resistance capacitor : Ceramic capacitor of 4.7  $\mu$ F or more can be used as the I/O capacitor.
- Input voltage : 1.5 to 5.5 V
- High-accuracy output voltage :  $\pm 1.0\%$  (1.0 to 1.45 V output product :  $\pm 15$  mV)
- Low dropout voltage : 70 mV typ. (3.0 V output product, at  $I_{OUT} = 200$  mA)
- Low current consumption : During operation : 70  $\mu$ A typ., 90  $\mu$ A max. (3.0 V output product)  
During shutdown : 0.1  $\mu$ A typ., 1.0  $\mu$ A max.
- Output current : 500 mA (3.0 V output product, at  $V_{IN} \geq V_{OUT(S)} + 1.0$  V)<sup>\*1</sup>
- High ripple rejection : 70 dB typ. (at 1.0 kHz,  $V_{OUT} = 1.0$  V)
- Built-in overcurrent protection circuit : limits overcurrent of output transistor
- Built-in thermal shutdown circuit : prevents damage caused by heat
- Built-in rush current control circuit : limits excessive rush current during start-up
- Built-in power on/off circuit : Longer battery life
- Small package : SOT-89-5
- Lead-free product

\*1. Attention should be paid to the power dissipation of the package when the output current is large.

### ■ Applications

- Power supply for battery-powered devices
- Power supply for TV, notebook PCs and home electric appliances
- Constant-voltage power supply for portable equipment

### ■ Packages

Package Name	Drawing Code		
	Package	Tape	Reel
SOT-89-5	UP005-A	UP005-A	UP005-A

## ■ Block Diagram

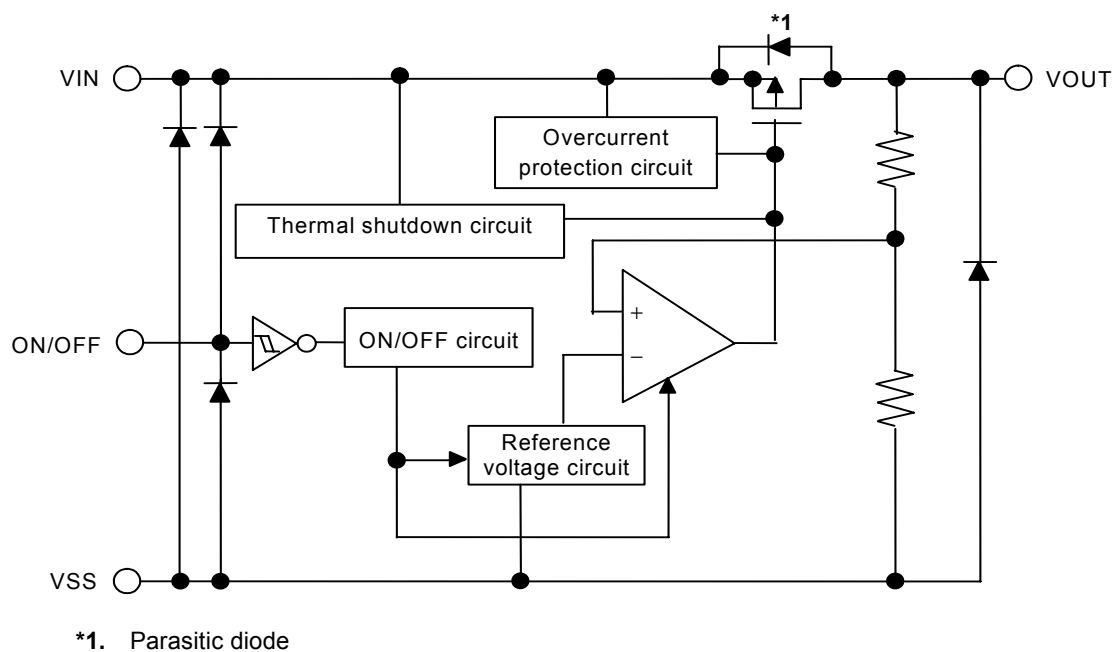
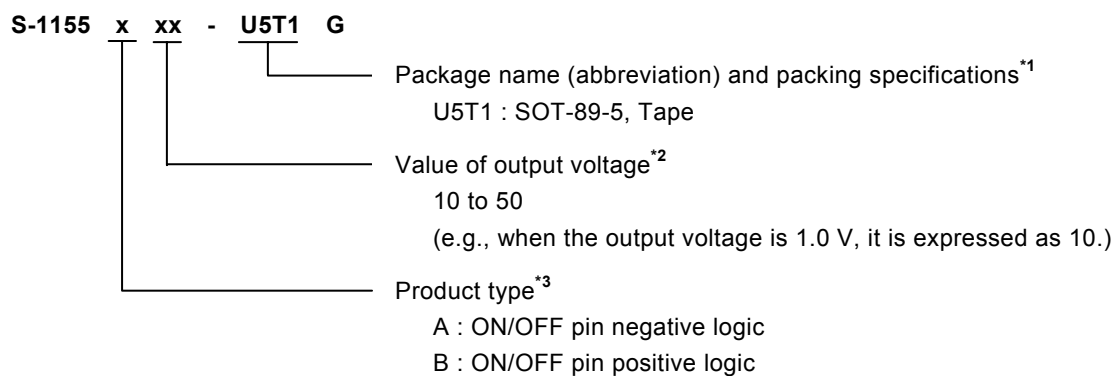


Figure 1

## ■ Product Name Structure

Users can select the product type, output voltage, and package type for the S-1155 Series. Refer to “**1. Product Name**” regarding the contents of product name, “**2. Product Name List**” regarding details of product name.

### 1. Product Name



\*1. Refer to the taping specifications.

\*2. If you request the product which has 0.05 step, contact our sales office.

\*3. Refer to “**3. Shutdown pin (ON/OFF pin)**” in “**■ Operation**”.

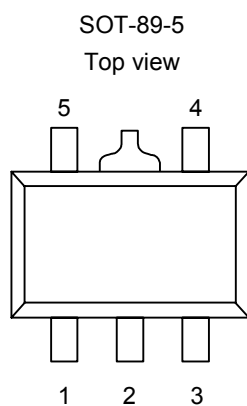
## 2. Product Name List

Table 1

Output Voltage	SOT-89-5
1.0 V $\pm$ 15 mV	S-1155B10-U5T1G
1.1 V $\pm$ 15 mV	S-1155B11-U5T1G
1.2 V $\pm$ 15 mV	S-1155B12-U5T1G
1.3 V $\pm$ 15 mV	S-1155B13-U5T1G
1.4 V $\pm$ 15 mV	S-1155B14-U5T1G
1.5 V $\pm$ 1.0%	S-1155B15-U5T1G
1.6 V $\pm$ 1.0%	S-1155B16-U5T1G
1.7 V $\pm$ 1.0%	S-1155B17-U5T1G
1.8 V $\pm$ 1.0%	S-1155B18-U5T1G
1.85 V $\pm$ 1.0%	S-1155B1J-U5T1G
1.9 V $\pm$ 1.0%	S-1155B19-U5T1G
2.0 V $\pm$ 1.0%	S-1155B20-U5T1G
2.1 V $\pm$ 1.0%	S-1155B21-U5T1G
2.2 V $\pm$ 1.0%	S-1155B22-U5T1G
2.3 V $\pm$ 1.0%	S-1155B23-U5T1G
2.4 V $\pm$ 1.0%	S-1155B24-U5T1G
2.5 V $\pm$ 1.0%	S-1155B25-U5T1G
2.6 V $\pm$ 1.0%	S-1155B26-U5T1G
2.7 V $\pm$ 1.0%	S-1155B27-U5T1G
2.8 V $\pm$ 1.0%	S-1155B28-U5T1G
2.85 V $\pm$ 1.0%	S-1155B2J-U5T1G
2.9 V $\pm$ 1.0%	S-1155B29-U5T1G
3.0 V $\pm$ 1.0%	S-1155B30-U5T1G
3.1 V $\pm$ 1.0%	S-1155B31-U5T1G
3.2 V $\pm$ 1.0%	S-1155B32-U5T1G
3.3 V $\pm$ 1.0%	S-1155B33-U5T1G
3.4 V $\pm$ 1.0%	S-1155B34-U5T1G
3.5 V $\pm$ 1.0%	S-1155B35-U5T1G
3.6 V $\pm$ 1.0%	S-1155B36-U5T1G
3.7 V $\pm$ 1.0%	S-1155B37-U5T1G
3.8 V $\pm$ 1.0%	S-1155B38-U5T1G
3.9 V $\pm$ 1.0%	S-1155B39-U5T1G
4.0 V $\pm$ 1.0%	S-1155B40-U5T1G
4.1 V $\pm$ 1.0%	S-1155B41-U5T1G
4.2 V $\pm$ 1.0%	S-1155B42-U5T1G
4.3 V $\pm$ 1.0%	S-1155B43-U5T1G
4.4 V $\pm$ 1.0%	S-1155B44-U5T1G
4.5 V $\pm$ 1.0%	S-1155B45-U5T1G
4.6 V $\pm$ 1.0%	S-1155B46-U5T1G
4.7 V $\pm$ 1.0%	S-1155B47-U5T1G
4.8 V $\pm$ 1.0%	S-1155B48-U5T1G
4.9 V $\pm$ 1.0%	S-1155B49-U5T1G
5.0 V $\pm$ 1.0%	S-1155B50-U5T1G

**Remark** Please contact our sales office for products with an output voltage other than those specified above or type A products.

## ■ Pin Configuration



**Figure 2**

**Table 2**

Pin No.	Symbol	Description
1	ON/OFF	Shutdown pin
2	VSS	GND pin
3	NC <sup>*1</sup>	No connection
4	VIN	Input voltage pin
5	VOUT	Output voltage pin

<sup>\*1</sup>. The NC pin is electrically open.

The NC pin can be connected to VIN or VSS.

## ■ Absolute Maximum Ratings

Table 3

(Ta = 25°C unless otherwise specified)

Parameter	Symbol	Absolute Maximum Rating	Unit
Input voltage	V <sub>IN</sub>	V <sub>SS</sub> – 0.3 to V <sub>SS</sub> + 6	V
	V <sub>ON/OFF</sub>	V <sub>SS</sub> – 0.3 to V <sub>IN</sub> + 0.3	V
Output voltage	V <sub>OUT</sub>	V <sub>SS</sub> – 0.3 to V <sub>IN</sub> + 0.3	V
Power dissipation	P <sub>D</sub>	1000 <sup>*1</sup>	mW
Operating ambient temperature	T <sub>opr</sub>	–40 to +85	°C
Storage temperature	T <sub>stg</sub>	–40 to +125	°C

\*1. When mounted on board

### [Mounted board]

- (1) Board size : 114.3 mm × 76.2 mm × t1.6 mm  
(2) Name : JEDEC STANDARD51-7

**Caution** The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

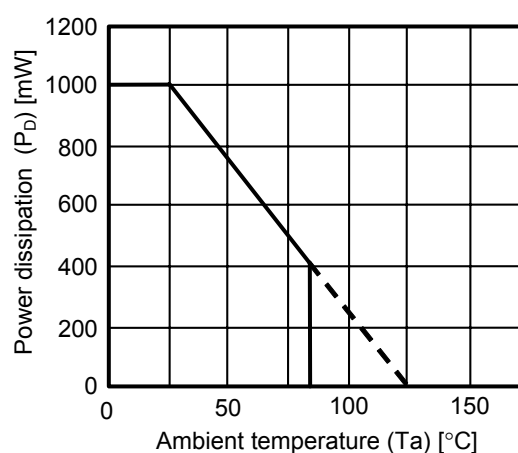


Figure 3 Power Dissipation of Package (When Mounted on Board)

Table 4

Condition	Power Dissipation	Thermal Resistance Value (θj–a)
SOT-89-5 (When mounted on board)	1000 mW	100°C/W

## ■ Electrical Characteristics

Table 5 (1 / 2)

(Ta = 25°C unless otherwise specified)

Parameter	Symbol	Conditions		Min.	Typ.	Max.	Unit	Test Circuit
Output voltage <sup>*1</sup>	V <sub>OUT(E)</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, I <sub>OUT</sub> = 100 mA	1.0 V ≤ V <sub>OUT(S)</sub> < 1.5 V	V <sub>OUT(S)</sub> − 0.015	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> + 0.015	V	1
			1.5 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	V <sub>OUT(S)</sub> × 0.99	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.01	V	1
		V <sub>IN</sub> = 5.5 V, I <sub>OUT</sub> = 100 mA	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	V <sub>OUT(S)</sub> × 0.99	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.01	V	1
Output current <sup>*2</sup>	I <sub>OUT</sub>	V <sub>IN</sub> ≥ V <sub>OUT(S)</sub> + 1.0 V	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	500 <sup>*5</sup>	—	—	mA	3
		V <sub>IN</sub> = 5.5 V	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	500 <sup>*5</sup>	—	—	mA	3
Dropout voltage <sup>*3</sup>	V <sub>drop</sub>	I <sub>OUT</sub> = 200 mA	1.0 V ≤ V <sub>OUT(S)</sub> < 1.1 V	0.5	0.54	0.58	V	1
			1.1 V ≤ V <sub>OUT(S)</sub> < 1.2 V	—	0.44	0.48	V	1
			1.2 V ≤ V <sub>OUT(S)</sub> < 1.3 V	—	0.34	0.38	V	1
			1.3 V ≤ V <sub>OUT(S)</sub> < 1.4 V	—	0.24	0.28	V	1
			1.4 V ≤ V <sub>OUT(S)</sub> < 1.5 V	—	0.14	0.18	V	1
			1.5 V ≤ V <sub>OUT(S)</sub> < 2.6 V	—	0.10	0.15	V	1
			2.6 V ≤ V <sub>OUT(S)</sub> ≤ 5.0 V	—	0.07	0.10	V	1
Line regulation	$\frac{\Delta V_{OUT1}}{\Delta V_{IN} \bullet V_{OUT}}$	V <sub>OUT(S)</sub> + 0.5 V ≤ V <sub>IN</sub> ≤ 5.5 V, I <sub>OUT</sub> = 100 mA	1.0 V ≤ V <sub>OUT(S)</sub> < 3.6 V	—	0.05	0.2	%/V	1
			3.6 V ≤ V <sub>OUT(S)</sub> ≤ 4.8 V	—	2.5	10	mV	1
		5.3 V ≤ V <sub>IN</sub> ≤ 5.5 V, I <sub>OUT</sub> = 100 mA	4.8 V < V <sub>OUT(S)</sub> ≤ 5.0 V	—	2.5	10	mV	1
Load regulation	ΔV <sub>OUT2</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, 1 mA ≤ I <sub>OUT</sub> ≤ 200 mA	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	—	15	30	mV	1
		V <sub>IN</sub> = 5.5 V, 1 mA ≤ I <sub>OUT</sub> ≤ 200 mA	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	—	15	30	mV	1
Output voltage temperature coefficient <sup>*4</sup>	$\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}$	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, I <sub>OUT</sub> = 100 mA, −40 ≤ Ta ≤ 85°C	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	—	±130	—	ppm/°C	1
		V <sub>IN</sub> = 5.5 V, I <sub>OUT</sub> = 100 mA, −40 ≤ Ta ≤ 85°C	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	—	±130	—	ppm/°C	1
Current consumption during operation	I <sub>SS1</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, ON/OFF pin = ON, no load, 1.0 V ≤ V <sub>OUT(S)</sub> < 1.8 V		—	90	110	μA	2
		V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, ON/OFF pin = ON, no load, 1.8 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V		—	70	90	μA	2
		V <sub>IN</sub> = 5.5 V, ON/OFF pin = ON, no load, 4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V		—	70	90	μA	2
Current consumption during shutdown	I <sub>SS2</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, ON/OFF pin = OFF, no load	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	—	0.1	1.0	μA	2
		V <sub>IN</sub> = 5.5 V, ON/OFF pin = OFF, no load	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	—	0.1	1.0	μA	2
Input voltage	V <sub>IN</sub>	—		1.5	—	5.5	V	—

**Table 5 (2 / 2)**

(Ta = 25°C unless otherwise specified)

Parameter	Symbol	Conditions		Min.	Typ.	Max.	Unit	Test Circuit
Shutdown pin input voltage “H”	V <sub>SH</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, R <sub>L</sub> = 1.0 kΩ, determined by V <sub>OUT</sub> output level	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	1.0	–	–	V	4
		V <sub>IN</sub> = 5.5 V, R <sub>L</sub> = 1.0 kΩ, determined by V <sub>OUT</sub> output level	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	1.0	–	–	V	4
Shutdown pin input voltage “L”	V <sub>SL</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, R <sub>L</sub> = 1.0 kΩ, determined by V <sub>OUT</sub> output level	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	–	–	0.3	V	4
		V <sub>IN</sub> = 5.5 V, R <sub>L</sub> = 1.0 kΩ, determined by V <sub>OUT</sub> output level	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	–	–	0.3	V	4
Shutdown pin input current “H”	I <sub>SH</sub>	V <sub>IN</sub> = 5.5 V, V <sub>ON/OFF</sub> = 5.5 V		–0.1	–	0.1	μA	4
Shutdown pin input current “L”	I <sub>SL</sub>	V <sub>IN</sub> = 5.5 V, V <sub>ON/OFF</sub> = 0 V		–0.1	–	0.1	μA	4
Ripple rejection	RR	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, f = 1 kHz, ΔV <sub>rip</sub> = 0.5 Vrms, I <sub>OUT</sub> = 100 mA	1.0 V ≤ V <sub>OUT(S)</sub> < 1.2 V	–	70	–	dB	5
			1.2 V ≤ V <sub>OUT(S)</sub> ≤ 3.0 V	–	65	–	dB	5
			3.0 V < V <sub>OUT(S)</sub> ≤ 4.5 V	–	60	–	dB	5
		V <sub>IN</sub> = 5.5 V, f = 1 kHz, ΔV <sub>rip</sub> = 0.5 Vrms, I <sub>OUT</sub> = 100 mA	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	–	60	–	dB	5
Short-circuit current	I <sub>SHORT</sub>	V <sub>IN</sub> = V <sub>OUT(S)</sub> + 1.0 V, ON/OFF pin = ON, V <sub>OUT</sub> = 0 V	1.0 V ≤ V <sub>OUT(S)</sub> ≤ 4.5 V	–	200	–	mA	3
		V <sub>IN</sub> = 5.5 V, ON/OFF pin = ON, V <sub>OUT</sub> = 0 V	4.5 V < V <sub>OUT(S)</sub> ≤ 5.0 V	–	200	–	mA	3
Thermal shutdown detection temperature	T <sub>SD</sub>	Junction temperature		–	150	–	°C	–
Thermal shutdown release temperature	T <sub>SR</sub>	Junction temperature		–	120	–	°C	–

\*1. V<sub>OUT(S)</sub>: Specified output voltage

V<sub>OUT(E)</sub>: Actual output voltage

The output voltage when fixing I<sub>OUT</sub> (= 100 mA) and inputting V<sub>OUT(S)</sub> + 1.0 V or 5.5 V

\*2. The output current at which the output voltage becomes 95% of V<sub>OUT(E)</sub> after gradually increasing the output current.

\*3. V<sub>drop</sub> = V<sub>IN1</sub> – (V<sub>OUT3</sub> × 0.98)

V<sub>OUT3</sub> is the output voltage when V<sub>IN</sub> = V<sub>OUT(S)</sub> + 1.0 V or 5.5 V, and I<sub>OUT</sub> = 200 mA.

V<sub>IN1</sub> is the input voltage at which the output voltage becomes 98% of V<sub>OUT3</sub> after gradually decreasing the input voltage.

\*4. The change in temperature [mV/°C] is calculated using the following equation.

$$\frac{\Delta V_{OUT}}{\Delta T_a} [\text{mV}/^\circ\text{C}]^*1 = V_{OUT(S)} [\text{V}]^*2 \times \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} [\text{ppm}/^\circ\text{C}]^*3 \div 1000$$

\*1. Change in temperature of the output voltage

\*2. Specified output voltage

\*3. Output voltage temperature coefficient

\*5. The output current can be at least this value.

Due to limitation of the package power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation of the package when the output current is large.

This specification is guaranteed by design.



## ■ Test Circuits

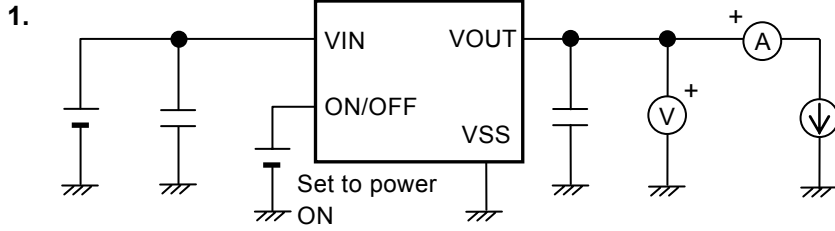


Figure 4

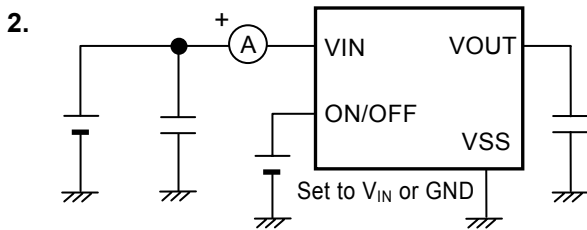


Figure 5

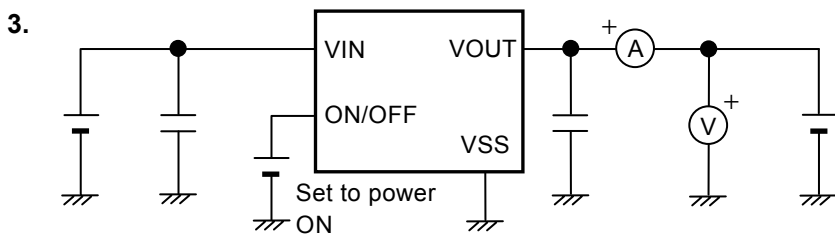


Figure 6

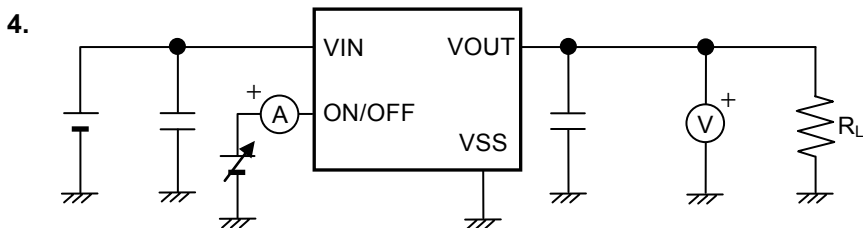


Figure 7

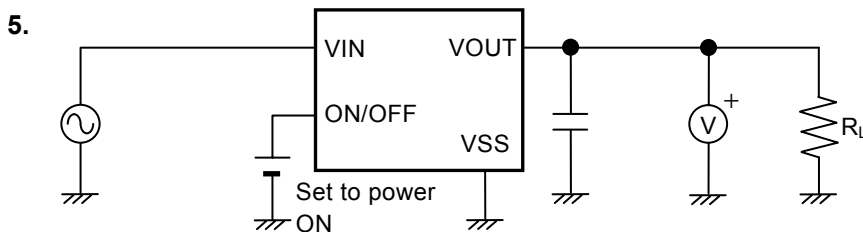
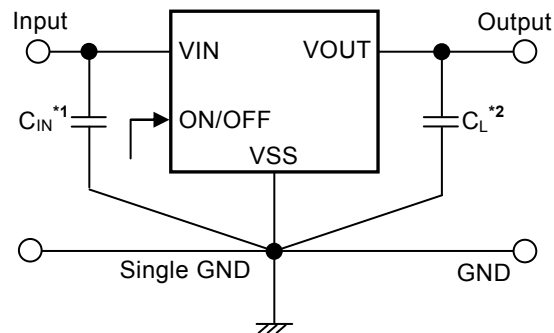


Figure 8

## ■ Standard Circuit



\*1.  $C_{IN}$  is a capacitor for stabilizing the input.

\*2. Ceramic capacitor of 4.7  $\mu\text{F}$  or more can be used as  $C_L$ .

Figure 9

**Caution** The above connection diagram and constant will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constant.

## ■ Application Conditions

Input capacitor ( $C_{IN}$ ) : 4.7  $\mu\text{F}$  or more

Output capacitor ( $C_L$ ) : 4.7  $\mu\text{F}$  or more (ceramic capacitor)

- Caution**
1. Set input capacitor ( $C_{IN}$ ) and output capacitor ( $C_L$ ) as  $C_{IN} = C_L$ .
  2. Generally, series regulator may oscillate depending on the external components. Confirm that no oscillation occurs in the application for which the above capacitors are used.

## ■ Selection of Input and Output Capacitors ( $C_{IN}$ , $C_L$ )

The S-1155 Series requires an output capacitor between the VOUT and VSS pin for phase compensation. Operation is stabilized by a ceramic capacitor with an output capacitance of 4.7  $\mu\text{F}$  or more over the entire temperature range. When using an OS capacitor, a tantalum capacitor, or an aluminum electrolytic capacitor, the capacitance must be 4.7  $\mu\text{F}$  or more.

The values of output overshoot and undershoot, which are transient response characteristics, vary depending on the value of output capacitor.

The required value of capacitance for the input capacitor differs depending on the application.

Set the value for input capacitor ( $C_{IN}$ ) and output capacitor ( $C_L$ ) as follows.

$$C_{IN} \geq 4.7 \mu\text{F}$$

$$C_L \geq 4.7 \mu\text{F}$$

$$C_{IN} = C_L$$

**Caution** The S-1155 Series may oscillate if setting the value as  $C_{IN} \geq 4.7 \mu\text{F}$ ,  $C_L \geq 4.7 \mu\text{F}$ ,  $C_{IN} < C_L$ . Define the values by sufficient evaluation including the temperature characteristics under the usage condition.

## ■ Explanation of Terms

### 1. Low dropout voltage regulator

This IC's voltage regulator has the low dropout voltage due to its built-in transistor with low on-resistance.

### 2. Output voltage ( $V_{OUT}$ )

The accuracy of the output voltage is ensured at  $\pm 1.0\%$  or  $\pm 15 \text{ mV}^2$  under the specified conditions of fixed input voltage<sup>\*1</sup>, fixed output current, and fixed temperature.

\*1. Differs depending on the product.

\*2. When  $V_{OUT} < 1.5 \text{ V}$  :  $\pm 15 \text{ mV}$ , When  $1.5 \text{ V} \leq V_{OUT}$  :  $\pm 1.0\%$

**Caution** If the above conditions change, the output voltage value may vary and exceed the accuracy range of the output voltage. See “■ Electrical Characteristics” and “■ Characteristics (Typical Data)” for details.

### 3. Line regulation $\left( \frac{\Delta V_{OUT1}}{\Delta V_{IN} \bullet V_{OUT}} \right)$

Indicates the dependency of the output voltage against the input voltage. That is, the value shows how much the output voltage changes due to a change in the input voltage after fixing output current constant.

### 4. Load regulation ( $\Delta V_{OUT2}$ )

Indicates the dependency of the output voltage against the output current. That is, the value shows how much the output voltage changes due to a change in the output current after fixing output current constant.

### 5. Dropout voltage ( $V_{drop}$ )

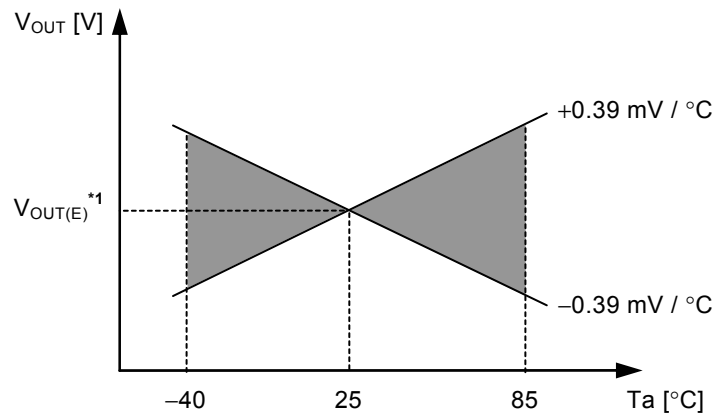
Indicates the difference between the output voltage and the input voltage  $V_{IN1}$ , which is the input voltage ( $V_{IN}$ ) when; decreasing input voltage  $V_{IN}$  gradually until the output voltage has dropped to the value of 98% of output voltage  $V_{OUT3}$ , which is at  $V_{IN} = V_{OUT(S)} + 1.0 \text{ V}$  or  $5.5 \text{ V}$ .

$$V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$$

## 6. Temperature coefficient of output voltage $\left( \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} \right)$

The shadowed area in **Figure 10** is the range where  $V_{OUT}$  varies in the operating temperature range when the temperature coefficient of the output voltage is  $\pm 130 \text{ ppm}/^\circ\text{C}$ .

Ex. S-1155B30 Typ.



\*1.  $V_{OUT(E)}$  is the value of the output voltage measured at  $25^\circ\text{C}$ .

**Figure 10**

A change in the temperature of the output voltage [ $\text{mV}/^\circ\text{C}$ ] is calculated using the following equation.

$$\frac{\Delta V_{OUT}}{\Delta T_a} [\text{mV} / ^\circ\text{C}]^*1 = V_{OUT(S)} [\text{V}]^*2 \times \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} [\text{ppm} / ^\circ\text{C}]^*3 \div 1000$$

- \*1. Change in temperature of output voltage
- \*2. Specified output voltage
- \*3. Output voltage temperature coefficient

## ■ Operation

### 1. Basic operation

Figure 11 shows the block diagram of the S-1155 Series.

The error amplifier compares the reference voltage ( $V_{ref}$ ) with  $V_{fb}$ , which is the output voltage resistance-divided by feedback resistors  $R_s$  and  $R_f$ . It supplies the gate voltage necessary to maintain the constant output voltage which is not influenced by the input voltage and temperature change, to the output transistor.

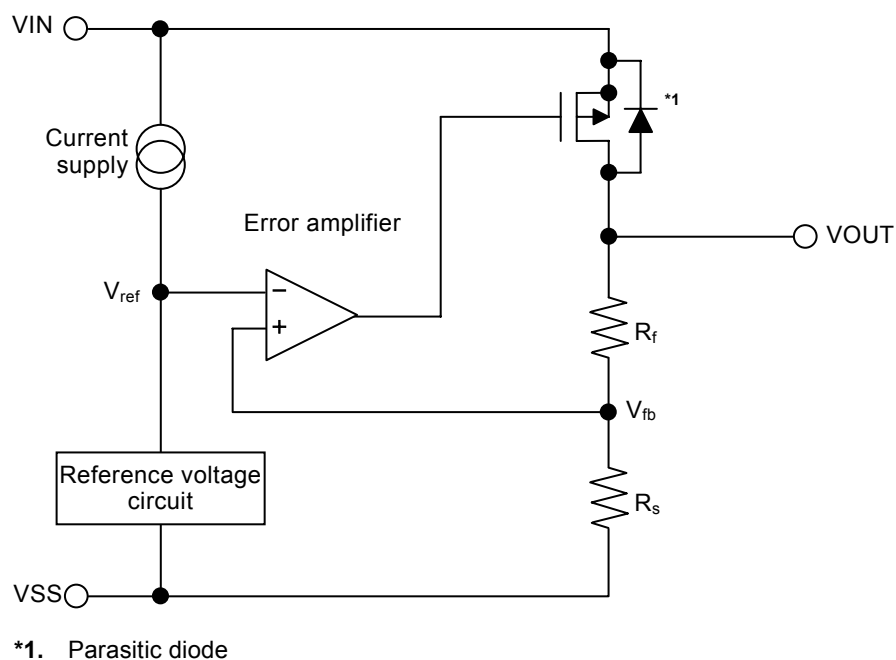


Figure 11

### 2. Output transistor

In the S-1155 Series, a low on-resistance P-channel MOS FET is used as the output transistor.

Be sure that  $V_{OUT}$  does not exceed  $V_{IN} + 0.3$  V to prevent the voltage regulator from being damaged due to inverse current which flows, because of a parasitic diode between the VIN and VOUT pin, when the potential of  $V_{OUT}$  became higher than  $V_{IN}$ .

### 3. Shutdown pin (ON/OFF pin)

This pin starts and stops the regulator.

When the ON/OFF pin is set to the shutdown level, the entire internal circuit stops operating, and the built-in P-channel MOS FET output transistor between the VIN and VOUT pin is turned off, in order to reduce the current consumption significantly. The VOUT pin is set to the  $V_{SS}$  level by the internal dividing resistor of several hundred  $k\Omega$  between the VOUT and VSS pin.

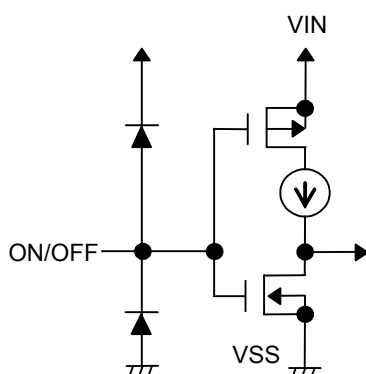
Note that the current consumption increases when a voltage of 0.3 V to 1.0 V is applied to the ON/OFF pin.

The ON/OFF pin is configured as shown in **Figure 12**. Since the ON/OFF pin is neither pulled down nor pulled up internally, do not use it in the floating state.

When not using the ON/OFF pin, connect it to the VSS pin in the product A type, and connect it to the VIN pin in B type.

**Table 6**

Logic Type	ON/OFF Pin	Internal Circuits	VOUT Pin Voltage	Current Consumption
A	"L": Power on	Operate	Set value	$I_{SS1}$
A	"H": Power off	Stop	$V_{SS}$ level	$I_{SS2}$
B	"L": Power off	Stop	$V_{SS}$ level	$I_{SS2}$
B	"H": Power on	Operate	Set value	$I_{SS1}$



**Figure 12**

### 4. Overcurrent protection circuit

The S-1155 Series has an overcurrent protection circuit having the characteristics shown in **(1) Output Voltage vs. Output Current (When load current is increased) ( $T_a = 25^\circ\text{C}$ )** in "■ Characteristics (Typical Data)", in order to protect the output transistor against an excessive output current and short circuiting between the VOUT and VSS pin. The current ( $I_{SHORT}$ ) when the output pin is short-circuited is internally set at approx. 200 mA (Typ.), and the initial value is restored for the output voltage, if releasing a short circuit once.

**Caution** This overcurrent protection circuit does not work as for thermal protection. If this IC long keeps short circuiting inside, pay attention to the conditions of input voltage and the load current so that, under the usage conditions including short circuit, the loss of the IC will not exceed power dissipation of the package.

## 5. Thermal shutdown circuit

The S-1155 Series has a thermal shutdown circuit to protect the device from damage due to overheat. When the junction temperature rises to 150°C (Typ.), the thermal shutdown circuit operates to stop regulating. When the junction temperature drops to 120°C (Typ.), the thermal shutdown circuit is released to restart regulating.

Due to self-heating of the S-1155 Series, if the thermal shutdown circuit starts operating, it stops regulating so that the output voltage drops. When regulation stops, the S-1155 does not itself generate heat so that the IC's temperature drops. When the temperature drops, the thermal shutdown circuit is released to restart regulating, thus this IC generates heat again. Repeating this procedure makes waveform of the output voltage pulse-like form. Stop/Restart of regulation continues unless decreasing either/both of the input voltage and the output voltage in order to reduce the internal current consumption, or decreasing the ambient temperature.

Table 7

Thermal Shutdown Circuit	VOUT Pin Voltage
Operation : 150°C (Typ.)*1	V <sub>SS</sub> level
Release : 120°C (Typ.)*1	Set value

\*1. Junction temperature

## 6. Rush current control circuit

The S-1155 Series has a rush current control circuit to control the rush current generated during power-on or when the ON/OFF pin is ON. This circuit limits the rush current (400 mA, Typ.) immediately after power-on or from the ON/OFF pin is ON until the specified time (100 μs min.) which is set internally.

**Caution** The junction temperature drops to 120°C (Typ.) by the operation of thermal shutdown circuit, after stopping regulation, the circuit is released to restart regulation; in this case, note that the period to limit rush current may become shorter (10 μs min.).

## ■ Precautions

- Wiring patterns for the VIN, VOUT and GND pins should be designed so that the impedance is low. When mounting an output capacitor between the VOUT and VSS pin ( $C_L$ ) and a capacitor for stabilizing the input between the VIN and VSS pin ( $C_{IN}$ ), the distance from the capacitors to these pins should be as short as possible.
- Note that the output voltage may increase when a series regulator is used at low load current (1.0 mA or less).
- Note that the output voltage may generally increase due to the leakage current from a driver when a series regulator is used at high temperature.
- Generally series regulator may oscillate, depending on the selection of external parts. The following conditions are recommended for this IC. However, be sure to perform sufficient evaluation under the actual usage conditions for selection, including the temperature characteristics. Regarding the equivalent series resistance ( $R_{ESR}$ ) for the output capacitor, refer to **(5) Example of Equivalent Series Resistance vs. Output Current Characteristics ( $T_a = 25^\circ\text{C}$ )** in “■ Reference Data”.

Input capacitor ( $C_{IN}$ ) : 4.7  $\mu\text{F}$  or more

Output capacitor ( $C_L$ ) : 4.7  $\mu\text{F}$  or more

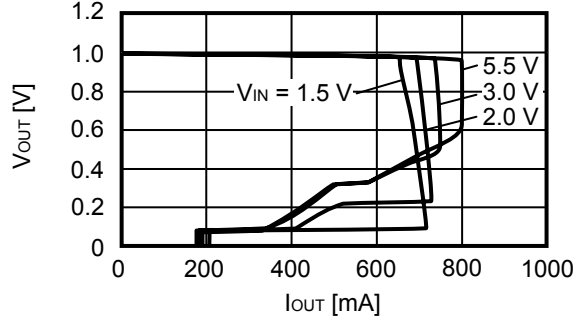
- The voltage regulator may oscillate when the impedance of the power supply is high and the input capacitance is small, or an input capacitor is not connected. Note that the voltage regulator may oscillate when the value of the input capacitor is greater than that of the output capacitor.
- Ringing may occur when these three conditions below are satisfied. Before selecting an input capacitor, be sure to evaluate sufficiently under the actual usage conditions, including the temperature characteristics.
  - The power supply inductance is high.
  - The load current is 100 mA or more.
  - The difference between the input voltage and the output voltage is close to the value of dropout voltage.
- If the output capacitance is small, power supply's fluctuation and the characteristics of load fluctuation become worse. Sufficiently evaluate the output voltage's fluctuation with the actual device.
- A momentary overshoot may be output when the power supply suddenly increases, and the output capacitance is small. It is therefore important to sufficiently evaluate the output voltage at power application in actual device.
- The application conditions for the input voltage, output voltage, and load current should not exceed power dissipation of the package.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- In determining the output current, attention should be paid to the output current value specified in **Table 5** in “■ Electrical Characteristics” and footnote \*5 of the table.
- SII claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.



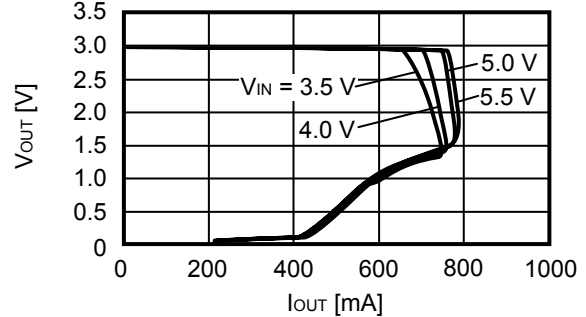
## ■ Characteristics (Typical Data)

### (1) Output Voltage vs. Output Current (when load current is increased) ( $T_a = 25^\circ\text{C}$ )

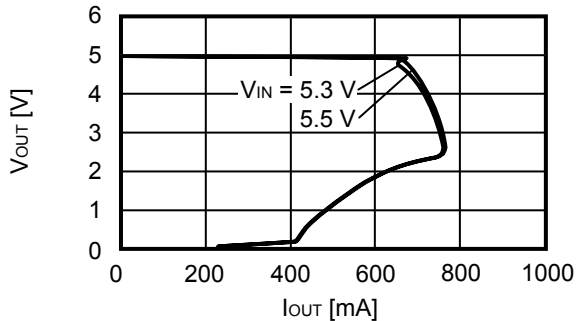
S-1155B10



S-1155B30



S-1155B50

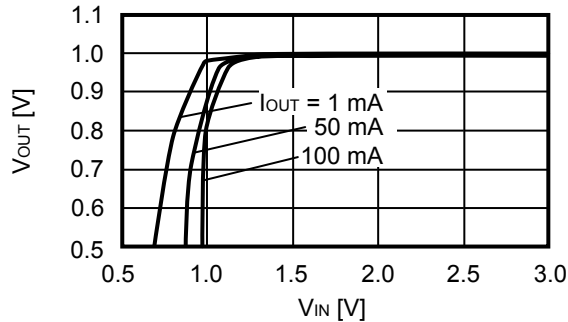


**Remark** In determining the output current, attention should be paid to the following.

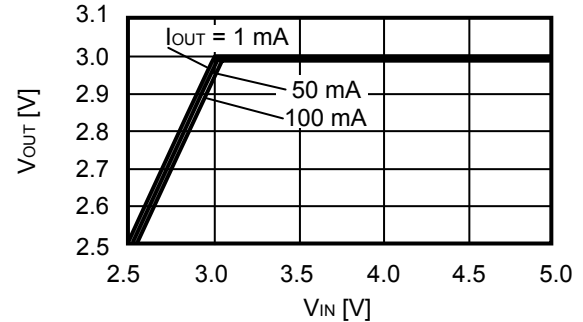
1. The minimum value of output current in **Table 5** and footnote \*5 in "■ Electrical Characteristics"
2. Power dissipation of the package

### (2) Output Voltage vs. Input Voltage ( $T_a = 25^\circ\text{C}$ )

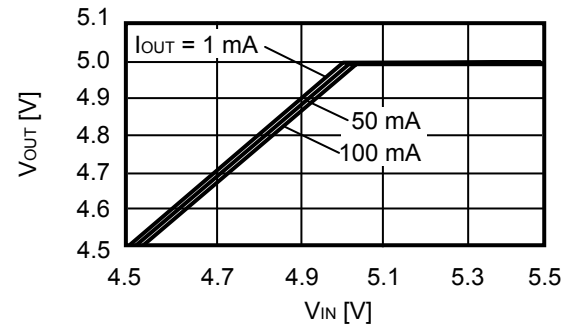
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S-1155B30

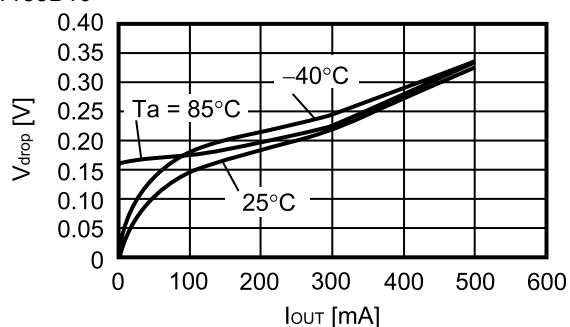


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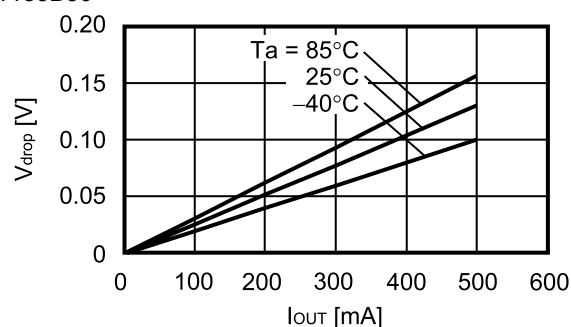


**(3) Dropout Voltage vs. Output Current**

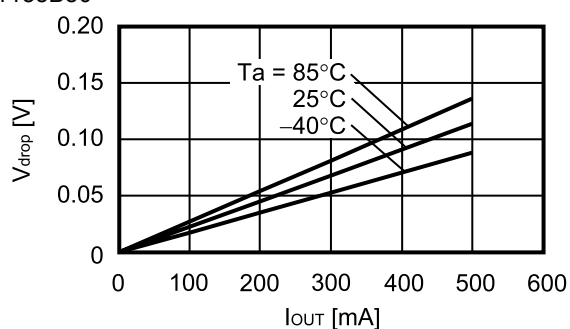
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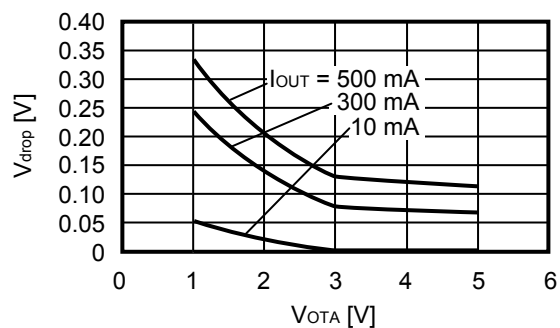
S-1155B30



S-1155B50

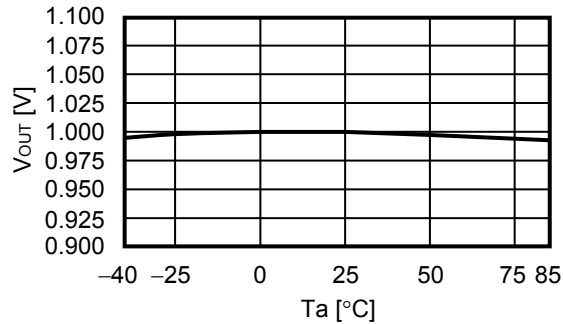


**(4) Dropout Voltage vs. Set Output Voltage**

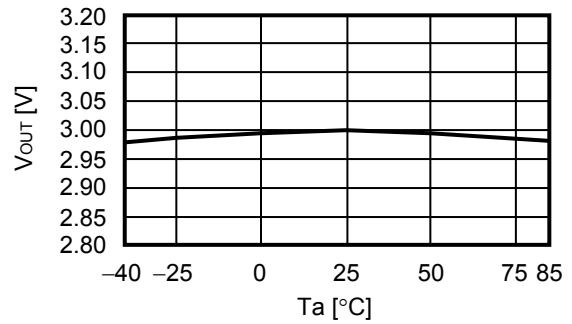


(5) Output Voltage vs. Ambient Temperature

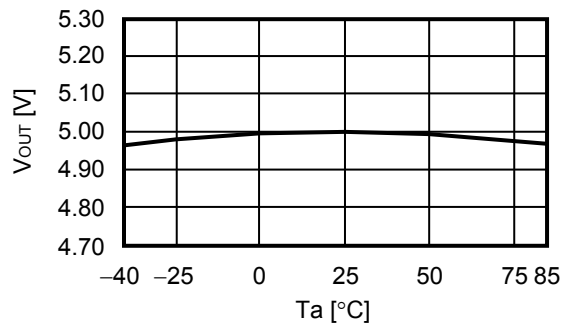
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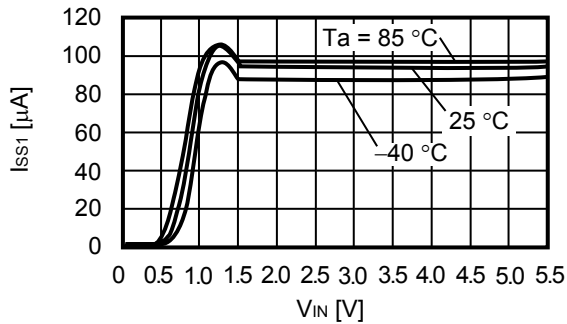


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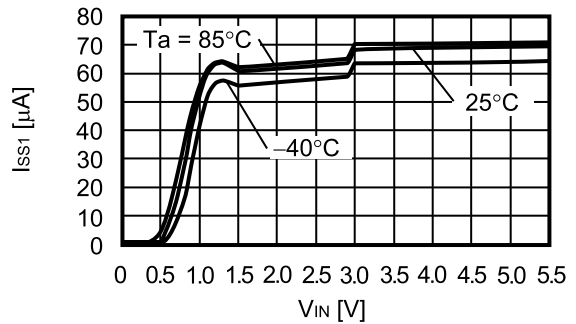


(6) Current Consumption vs. Input Voltage

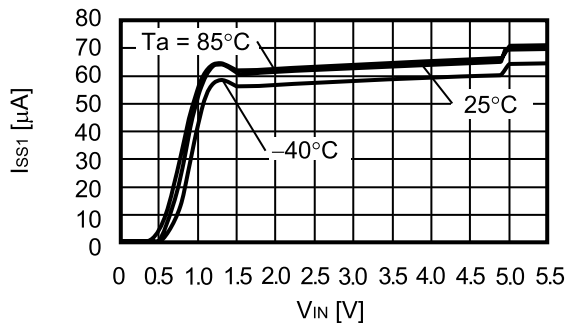
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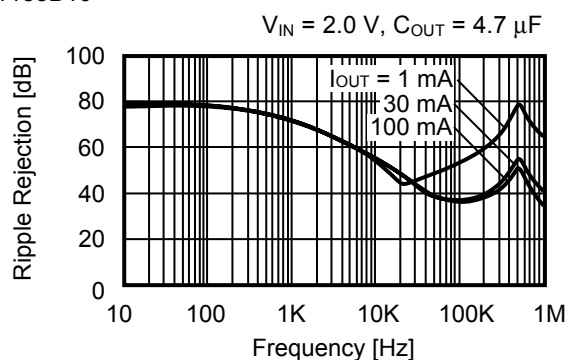


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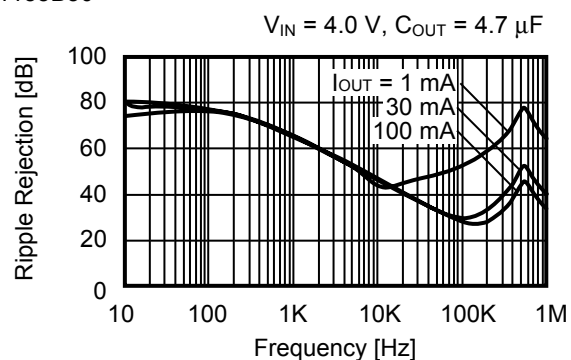


**(7) Ripple Rejection ( $T_a = 25^\circ\text{C}$ )**

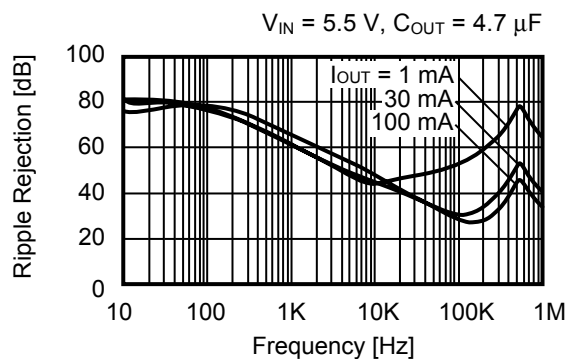
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S-1155B30



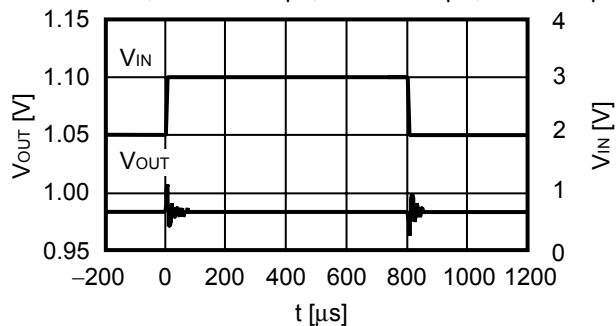
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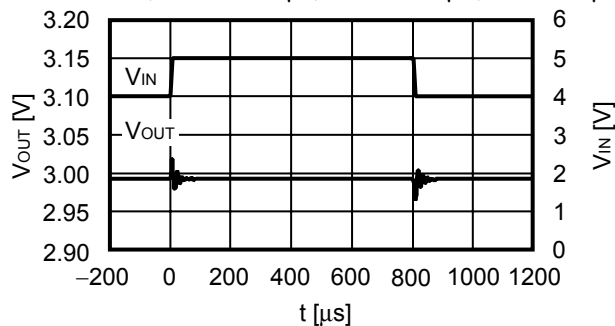
## ■ Reference Data

### (1) Characteristics of Input Transient Response ( $T_a = 25^\circ\text{C}$ )

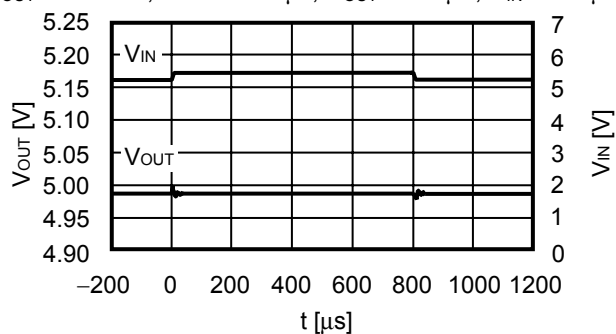
S-1155B10

 $I_{OUT} = 100\text{ mA}$ ,  $t_r = t_f = 5.0\text{ }\mu\text{s}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ 

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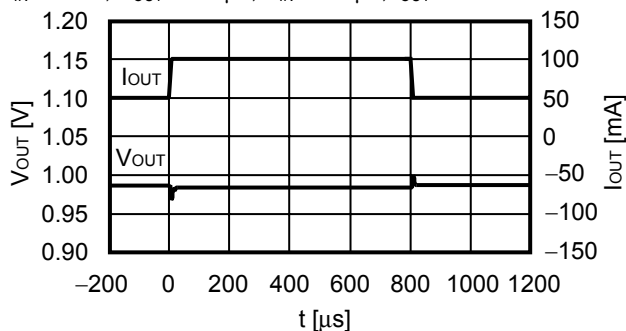
 $I_{OUT} = 100\text{ mA}$ ,  $t_r = t_f = 5.0\text{ }\mu\text{s}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ 

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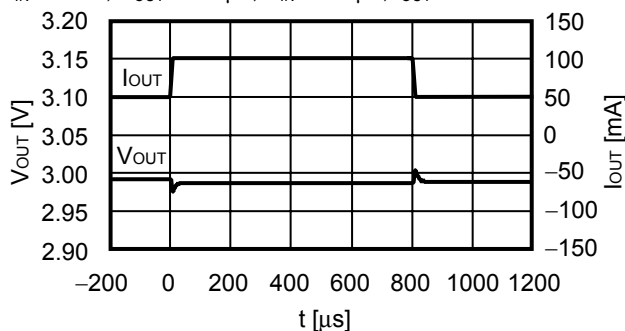
 $I_{OUT} = 100\text{ mA}$ ,  $t_r = t_f = 5.0\text{ }\mu\text{s}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ 

### (2) Characteristics of Load Transient Response ( $T_a = 25^\circ\text{C}$ )

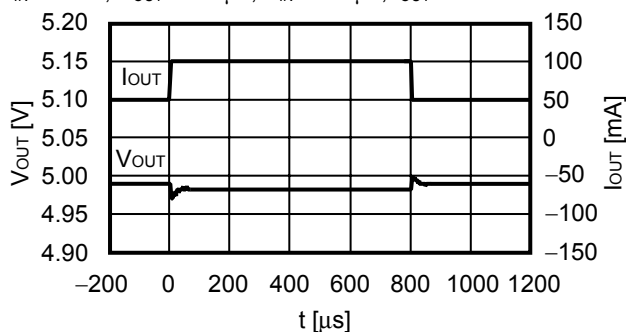
S-1155B10

 $V_{IN} = 2.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 50 \leftrightarrow 100\text{ mA}$ 

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 $V_{IN} = 4.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 50 \leftrightarrow 100\text{ mA}$ 

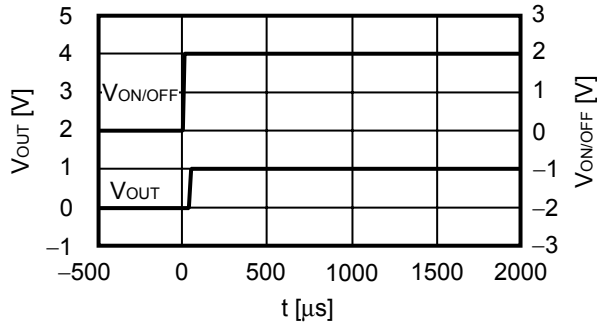
S-1155B50

 $V_{IN} = 5.5\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 50 \leftrightarrow 100\text{ mA}$ 

**(3) Transient Response Characteristics of ON/OFF Pin ( $T_a = 25^\circ\text{C}$ )**

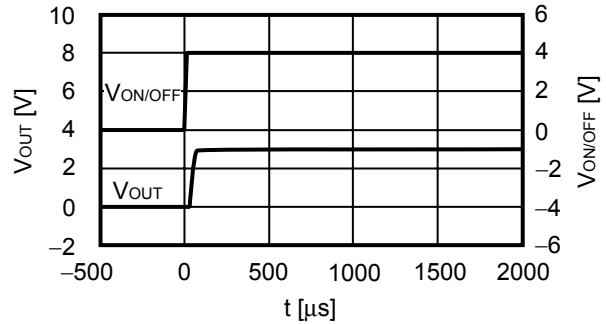
S-1155B10

$V_{IN} = 2.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 100\text{ mA}$



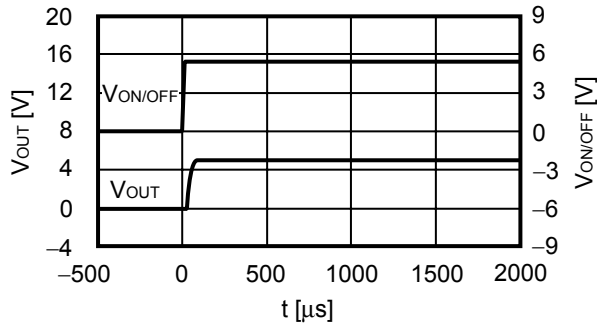
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$V_{IN} = 4.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 100\text{ mA}$



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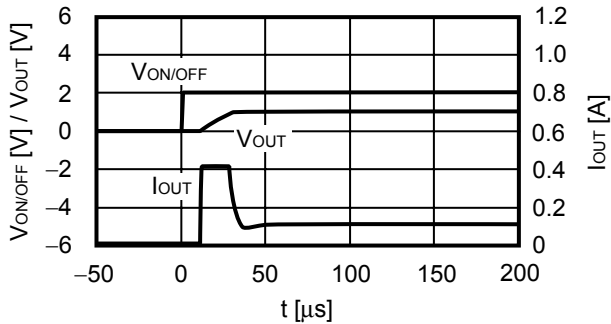
$V_{IN} = 5.5\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 100\text{ mA}$



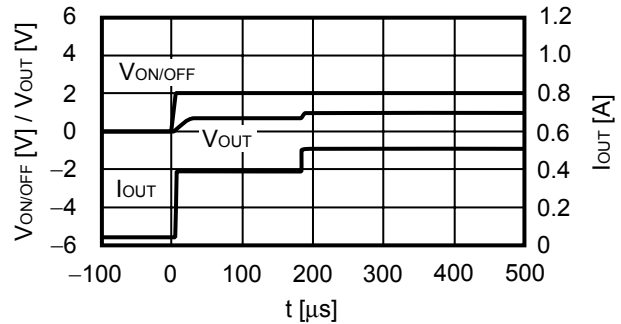
**(4) Characteristics of Rush Current ( $T_a = 25^\circ\text{C}$ )**

S-1155B10

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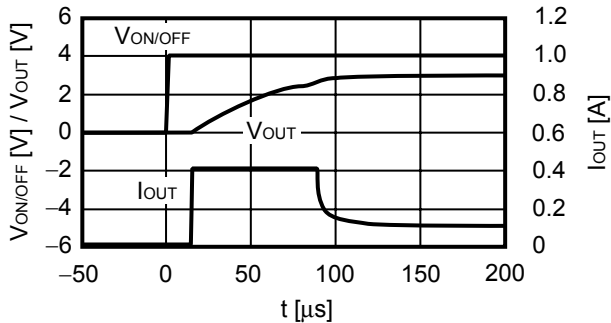


$V_{IN} = 2.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 500\text{ mA}$

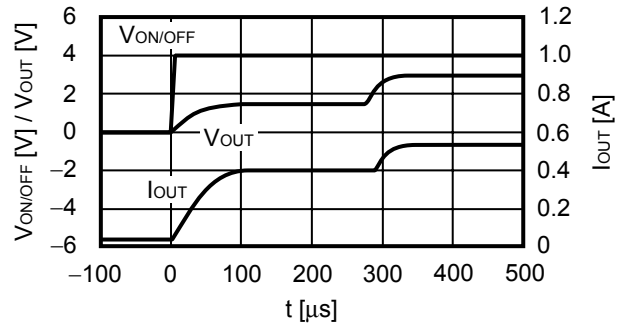


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$V_{IN} = 4.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 100\text{ mA}$

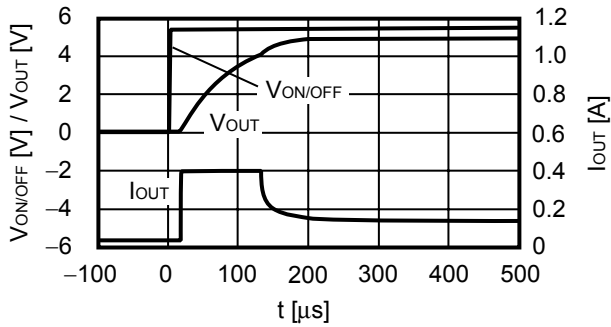


$V_{IN} = 4.0\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 500\text{ mA}$

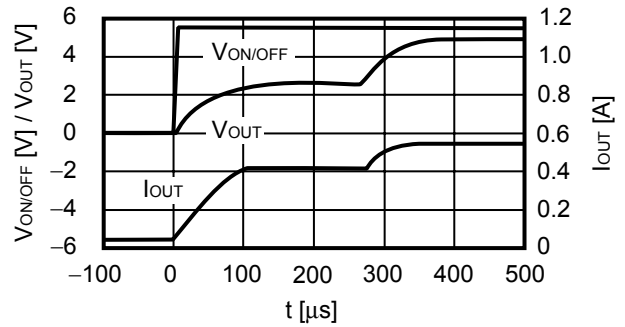


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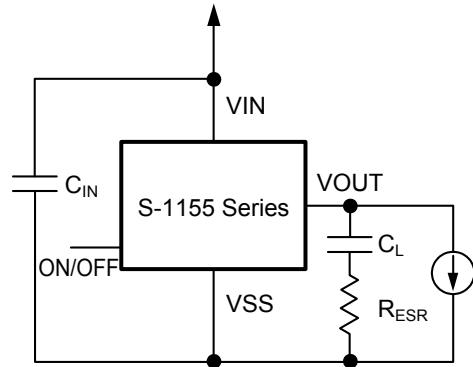
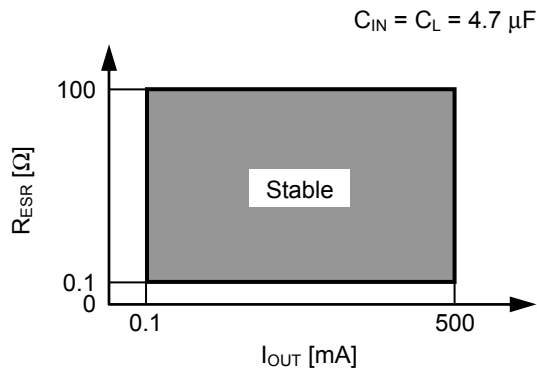
$V_{IN} = 5.5\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 100\text{ mA}$



$V_{IN} = 5.5\text{ V}$ ,  $C_{OUT} = 4.7\text{ }\mu\text{F}$ ,  $C_{IN} = 4.7\text{ }\mu\text{F}$ ,  $I_{OUT} = 500\text{ mA}$

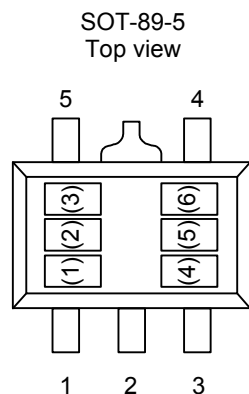


(5) Example of Equivalent Series Resistance vs. Output Current Characteristics ( $T_a = 25^\circ\text{C}$ )



## ■ Marking Specifications

### SOT-89-5



(1) to (3) : Product abbreviation (refer to **Product Name vs. Product Code**)  
(4) to (6) : Lot number

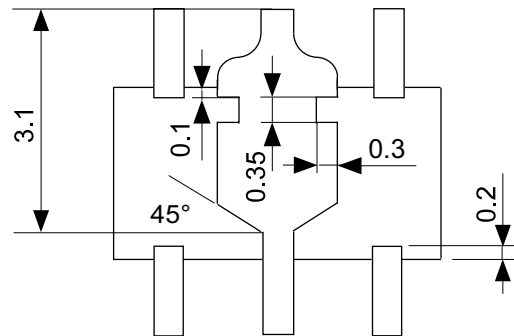
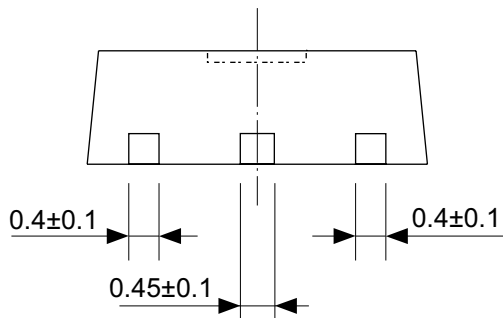
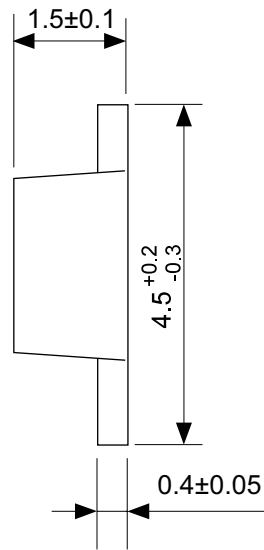
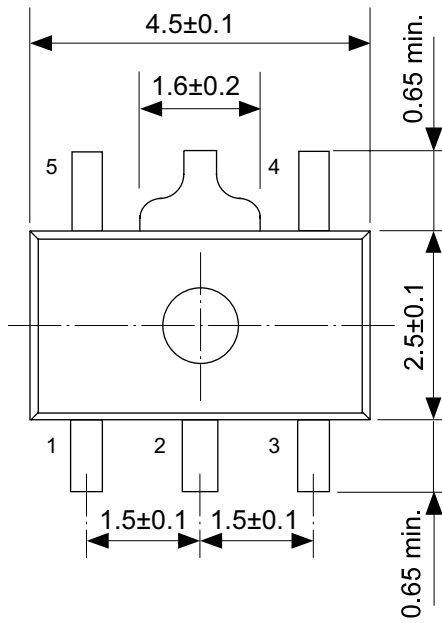
### Product Name vs. Product Code

Product Name	Product Code		
	(1)	(2)	(3)
S-1155B10-U5T1G	S	K	A
S-1155B11-U5T1G	S	K	B
S-1155B12-U5T1G	S	K	C
S-1155B13-U5T1G	S	K	D
S-1155B14-U5T1G	S	K	E
S-1155B15-U5T1G	S	K	F
S-1155B16-U5T1G	S	K	G
S-1155B17-U5T1G	S	K	H
S-1155B18-U5T1G	S	K	I
S-1155B1J-U5T1G	S	K	J
S-1155B19-U5T1G	S	K	K
S-1155B20-U5T1G	S	K	L
S-1155B21-U5T1G	S	K	M
S-1155B22-U5T1G	S	K	N
S-1155B23-U5T1G	S	K	O
S-1155B24-U5T1G	S	K	P
S-1155B25-U5T1G	S	K	Q
S-1155B26-U5T1G	S	K	R
S-1155B27-U5T1G	S	K	S
S-1155B28-U5T1G	S	K	T
S-1155B2J-U5T1G	S	K	U
S-1155B29-U5T1G	S	K	V

Product Name	Product Code		
	(1)	(2)	(3)
S-1155B30-U5T1G	S	K	W
S-1155B31-U5T1G	S	K	X
S-1155B32-U5T1G	S	K	Y
S-1155B33-U5T1G	S	K	Z
S-1155B34-U5T1G	S	L	A
S-1155B35-U5T1G	S	L	B
S-1155B36-U5T1G	S	L	C
S-1155B37-U5T1G	S	L	D
S-1155B38-U5T1G	S	L	E
S-1155B39-U5T1G	S	L	F
S-1155B40-U5T1G	S	L	G
S-1155B41-U5T1G	S	L	H
S-1155B42-U5T1G	S	L	I
S-1155B43-U5T1G	S	L	J
S-1155B44-U5T1G	S	L	K
S-1155B45-U5T1G	S	L	L
S-1155B46-U5T1G	S	L	M
S-1155B47-U5T1G	S	L	N
S-1155B48-U5T1G	S	L	O
S-1155B49-U5T1G	S	L	P
S-1155B50-U5T1G	S	L	Q

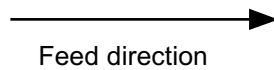
**Remark** Please contact our sales office for products with an output voltage other than those specified above or type A products.



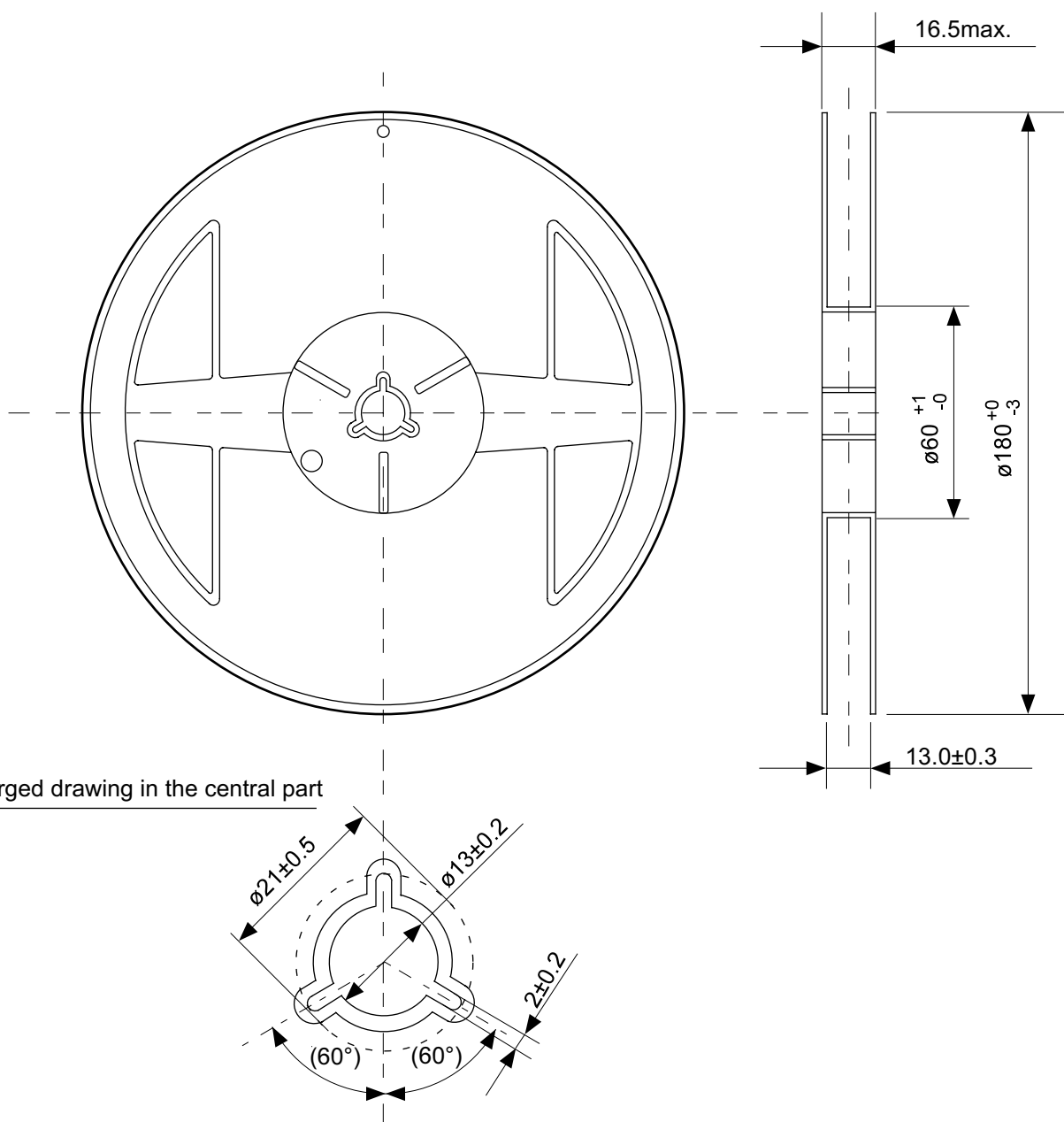


No. UP005-A-P-SD-1.1

TITLE	SOT895-A-PKG Dimensions
No.	UP005-A-P-SD-1.1
SCALE	
UNIT	mm
Seiko Instruments Inc.	



TITLE	SOT895-A-Carrier Tape
No.	UP005-A-C-SD-1.1
SCALE	
UNIT	mm
Seiko Instruments Inc.	



No. UP005-A-R-SD-1.1

TITLE	SOT895-A-Reel		
No.	UP005-A-R-SD-1.1		
SCALE		QTY.	1,000
UNIT	mm		
Seiko Instruments Inc.			

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