



18-BIT, 1.25-MSPS, PSEUDO-BIPOLAR, FULLY DIFFERENTIAL INPUT, MICROPOLAR SAMPLING ANALOG-TO-DIGITAL CONVERTER WITH PARALLEL INTERFACE, REFERENCE

FEATURES

- 1.25-MHz Sample Rate
- ± 1.5 LSB Typ, ± 2.5 LSB Max INL
- $\pm 0.8/-0.6$ LSB Typ, $\pm 1.5/-1$ LSB Max DNL
- 18-Bit NMC Ensured Over Temperature
- ± 0.5 -mV Offset Error
- ± 0.05 -PPM/ $^{\circ}$ C Offset Error Drift
- ± 0.1 %FSR Gain Error
- ± 0.5 -PPM/ $^{\circ}$ C Gain Error Drift
- 98.5dB SNR, -120db THD, 121dB SFDR
- Zero Latency
- Low Power: 235 mW Typ at 1.25 MSPS
- Pseudo-Bipolar Fully Differential Input Range: V_{ref} to $-V_{ref}$
- Onboard Reference with 6 PPM/ $^{\circ}$ C Drift
- Onboard Reference Buffer
- High-Speed Parallel Interface
- Wide Digital Supply 2.7 V to 5.25 V
- 8-/16-/18-Bit Bus Transfer

- 48-Pin 7x7 QFN Package

APPLICATIONS

- Medical Instruments
- Optical Networking
- Transducer Interface
- High Accuracy Data Acquisition Systems
- Magnetometers

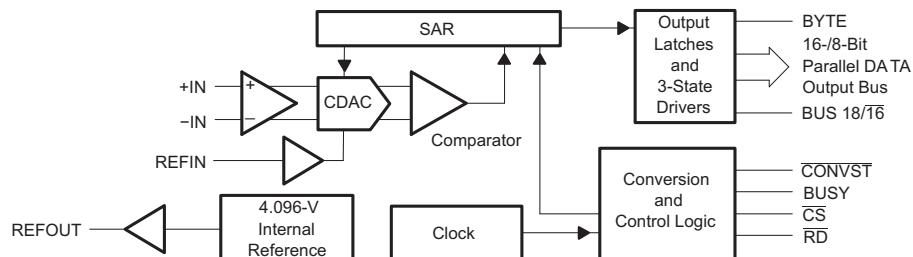
DESCRIPTION

The ADS8484 is an 18-bit, 1.25-MSPS A/C converter with an internal 4.096-V reference and a pseudo-bipolar, fully differential input. The device includes a 18-bit capacitor-based SAR A/D converter with inherent sample and hold. The ADS8484 offers a full 18-bit interface, a 16-bit option where data is read using two read cycles, or an 8-bit bus option using three read cycles.

The ADS8484 is available in a 48-lead 7x7 QFN package and is characterized over the industrial -40° C to 85° C temperature range.

HIGH SPEED SAR CONVERTER FAMILY

TYPE/SPEED	500 kHz	~600 kHz	750 kHz	1 MHz	1.25 MHz	2 MHz	3 MHz	4MHz
18-Bit Pseudo-Diff	ADS8383	ADS8381		ADS8481				
		ADS8380 (s)						
18-Bit Pseudo-Bipolar, Fully Diff		ADS8382 (s)		ADS8482	ADS8484			
16-Bit Pseudo-Diff	ADS8327	ADS8370 (s)	ADS8371	ADS8471	ADS8401	ADS8411		
	ADS8328	ADS8472 (s)			ADS8405	ADS8410 (s)		
16-Bit Pseudo-Bipolar, Fully Diff				ADS8472	ADS8402	ADS8412		ADS8422
					ADS8406	ADS8413 (s)		
14-Bit Pseudo-Diff					ADS7890 (s)		ADS7891	
12-Bit Pseudo-Diff				ADS7886				ADS7881



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION⁽¹⁾

MODEL	MAXIMUM INTEGRAL LINEARITY (LSB)	MAXIMUM DIFFERENTIAL LINEARITY (LSB)	NO MISSING CODES RESOLUTION (BIT)	PACKAGE TYPE	PACKAGE DESIGNATOR	TEMPERATURE RANGE	ORDERING INFORMATION	TRANSPORT MEDIA QTY.
ADS8484I	±4	–1 to +2	18	7x7 48 Pin QFN	RGZ	–40°C to 85°C	ADS8484IRGZT	Tape and reel 250
							ADS8484IRGZR	Tape and reel 1000
ADS8484IB	±2.5	–1 to +1.5	18	7x7 48 Pin QFN	RGZ	–40°C to 85°C	ADS8484IBRGZT	Tape and reel 250
							ADS8484IBRGZR	Tape and reel 1000

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

		VALUE	UNIT
Voltage	+IN to AGND	–0.4 to +VA + 0.1	V
	–IN to AGND	–0.4 to +VA + 0.1	V
	+VA to AGND	–0.3 to 7	V
	+VBD to BDGND	–0.3 to 7	V
	+VA to +VBD	–0.3 to 2.55	V
Digital input voltage to BDGND		–0.3 to +VBD + 0.3	V
Digital output voltage to BDGND		–0.3 to +VBD + 0.3	V
T _A	Operating free-air temperature range	–40 to 85	°C
T _{stg}	Storage temperature range	–65 to 150	°C
Junction temperature (T _J max)		150	°C
QFN package	Power dissipation	(T _J Max – T _A)/θ _{JA}	
	θ _{JA} thermal impedance	22	°C/W
Lead temperature, soldering	Vapor phase (60 sec)	215	°C
	Infrared (15 sec)	220	°C

(1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

SPECIFICATIONS
 $T_A = -40^\circ\text{C}$ to 85°C , $+\text{VA} = 5\text{ V}$, $+\text{VBD} = 3\text{ V}$ or 5 V , $V_{\text{ref}} = 4.096\text{ V}$, $f_{\text{SAMPLE}} = 1.25\text{ MSPS}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ANALOG INPUT						
Full-scale input voltage ⁽¹⁾		+IN – (–IN)	$-V_{\text{ref}}$		V_{ref}	V
Absolute input voltage		+IN	–0.2	$V_{\text{ref}} + 0.2$		V
		–IN	–0.2	$V_{\text{ref}} + 0.2$		
Common-mode input range			$(V_{\text{ref}})/2 - 0.2$	$(V_{\text{ref}})/2$	$(V_{\text{ref}})/2 + 0.2$	V
Input capacitance				65		pF
Input leakage current				1		nA
SYSTEM PERFORMANCE						
Resolution				18		Bits
No missing codes	ADS8484I		18			Bits
	ADS8484IB		18			
INL	Integral linearity ⁽²⁾	ADS8484I	–4	± 1.5	4	LSB
		ADS8484IB	–2.5	± 1.5	2.5	(18 bit) ⁽³⁾
DNL	Differential linearity	ADS8484I	–1	–0.6/0.8	2	LSB
		ADS8484IB	–1	–0.6/0.8	1.5	(18 bit)
	Offset error ⁽⁴⁾	ADS8484I	–2	± 1	2	mV
		ADS8484IB	–0.5	± 0.1	0.5	
	Offset error temperature drift	ADS8484I		± 0.05		ppm/°C
		ADS8484IB		± 0.05		
E_G	Gain error ⁽⁴⁾ (5)	ADS8484I	$V_{\text{ref}} = 4.096\text{ V}$	–0.1	± 0.035	%FS
		ADS8484IB	$V_{\text{ref}} = 4.096\text{ V}$	–0.1	± 0.035	%FS
	Gain error temperature drift	ADS8484I		± 0.5		ppm/°C
		ADS8484IB		± 0.5		
CMRR	Common-mode rejection ratio	At dc ($\pm 0.2\text{ V}$ around $V_{\text{ref}}/2$)		60		dB
		+IN – (–IN) = 1 Vpp at 1.25 MHz		55		
	Noise			30		$\mu\text{V RMS}$
	Power supply rejection ratio	At 1FFFFh output code		60		dB
SAMPLING DYNAMICS						
Conversion time				575	610	ns
Acquisition time			175	200		ns
Throughput rate					1.25	MHz
Aperture delay				4		ns
Aperture jitter				5		ps
Step response				150		ns
Over voltage recovery				150		ns

(1) Ideal input span, does not include gain or offset error.

(2) This is endpoint INL, not best fit.

(3) LSB means least significant bit

(4) Measured relative to an ideal full-scale input [+IN – (–IN)] of 8.192 V

(5) This specification does not include the internal reference voltage error and drift.

SPECIFICATIONS (Continued)

$T_A = -40^\circ\text{C}$ to 85°C , $+\text{VA} = 5\text{ V}$, $+\text{VBD} = 3\text{ V}$ or 5 V , $V_{\text{ref}} = 4.096\text{ V}$, $f_{\text{SAMPLE}} = 1.25\text{ MSPS}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
DYNAMIC CHARACTERISTICS							
THD	Total harmonic distortion ⁽¹⁾	ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 2 kHz	-115	dB		
		ADS8484IB		-120			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 20 kHz	-105			
		ADS8484IB		-110			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 100 kHz	-100			
		ADS8484IB		-103			
SNR	Signal-to-noise ratio ⁽¹⁾	ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 2 kHz	96	97	dB	
		ADS8484IB		97	98.5		
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 20 kHz	96	dB		
		ADS8484IB		98			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 100 kHz	95			
		ADS8484IB		97			
SINAD	Signal-to-noise + distortion ⁽¹⁾	ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 2 kHz	96	96	dB	
		ADS8484IB		97	98.5		
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 20 kHz	95	dB		
		ADS8484IB		97			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 100 kHz	93			
		ADS8484IB		95			
SFDR	Spurious free dynamic range ⁽¹⁾	ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 2 kHz	117	dB		
		ADS8484IB		121			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 20 kHz	107			
		ADS8484IB		113			
		ADS8484I	$V_{\text{IN}} = 8\text{ V}_{\text{pp}}$ at 100 kHz	102			
		ADS8484IB		105			
-3dB Small signal bandwidth				15		MHz	

(1) Calculated on the first nine harmonics of the input frequency.

SPECIFICATIONS (Continued)

$T_A = -40^\circ\text{C}$ to 85°C , $+VA = 5\text{ V}$, $+VBD = 3\text{ V}$ or 5 V , $V_{\text{ref}} = 4.096\text{ V}$, $f_{\text{SAMPLE}} = 1.25\text{ MSPS}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VOLTAGE REFERENCE INPUT					
V_{ref}	Reference voltage at REFIN	3.0	4.096	$+VA - 0.8$	V
	Reference resistance ⁽¹⁾		500		$\text{k}\Omega$
	Reference current drain	$f_s = 1.25\text{ MHz}$		1	mA
INTERNAL REFERENCE OUTPUT					
Internal reference start-up time	From 95% ($+VA$), with $1\text{-}\mu\text{F}$ storage capacitor			120	ms
V_{ref}	$I_O = 0$	4.081	4.096	4.111	V
Source current	Static load			10	μA
Line regulation	$+VA = 4.75\text{ V} \sim 5.25\text{ V}$		60		μV
Drift	$I_O = 0$		± 6		$\text{PPM}/^\circ\text{C}$
DIGITAL INPUT/OUTPUT					
Logic family – CMOS					
V_{IH}	$I_{IH} = 5\text{ }\mu\text{A}$	$+VBD - 1$	$+VBD + 0.3$		V
V_{IL}	$I_{IL} = 5\text{ }\mu\text{A}$	-0.3	0.8		
V_{OH}	$I_{OH} = 2$ TTL loads	$+VBD - 0.6$			
V_{OL}	$I_{OL} = 2$ TTL loads		0.4		
Data format – Two's Complement					
POWER SUPPLY REQUIREMENTS					
Power supply voltage	$+VBD$	2.7	3.3	5.25	V
	$+VA$	4.75	5	5.25	V
Supply current ⁽²⁾	$f_s = 1.25\text{ MHz}$		47	52	mA
Power dissipation ⁽²⁾	$f_s = 1.25\text{ MHz}$		235	260	mW
TEMPERATURE RANGE					
Operating free-air		-40		85	$^\circ\text{C}$

(1) Can vary $\pm 20\%$

(2) This includes only $+VA$ current. $+VBD$ current is typical 1 mA with 5-pF load capacitance on all output pins.

TIMING CHARACTERISTICS

All specifications typical at -40°C to 85°C , $+\text{VA} = +\text{VBD} = 5\text{ V}$ ⁽¹⁾ ⁽²⁾ ⁽³⁾

PARAMETER		MIN	TYP	MAX	UNIT
$t_{(\text{CONV})}$	Conversion time			610	ns
$t_{(\text{ACQ})}$	Acquisition time		175		ns
$t_{(\text{HOLD})}$	Sample capacitor hold time			15	ns
t_{pd1}	$\overline{\text{CONVST}}$ low to $\overline{\text{BUSY}}$ high			40	ns
t_{pd2}	Propagation delay time, end of conversion to $\overline{\text{BUSY}}$ low			15	ns
t_{pd3}	Propagation delay time, start of convert state to rising edge of $\overline{\text{BUSY}}$			25	ns
t_{w1}	Pulse duration, $\overline{\text{CONVST}}$ low		40		ns
t_{su1}	Setup time, $\overline{\text{CS}}$ low to $\overline{\text{CONVST}}$ low		20		ns
t_{w2}	Pulse duration, $\overline{\text{CONVST}}$ high		20		ns
	$\overline{\text{CONVST}}$ falling edge jitter			10	ps
t_{w3}	Pulse duration, $\overline{\text{BUSY}}$ signal low		$t_{(\text{ACQ})\text{min}}$		ns
t_{w4}	Pulse duration, $\overline{\text{BUSY}}$ signal high			610	ns
t_{h1}	Hold time, first data bus transition ($\overline{\text{RD}}$ low, or $\overline{\text{CS}}$ low for read cycle, or BYTE or BUS18/16 input changes) after $\overline{\text{CONVST}}$ low		40		ns
t_{d1}	Delay time, $\overline{\text{CS}}$ low to $\overline{\text{RD}}$ low		0		ns
t_{su2}	Setup time, $\overline{\text{RD}}$ high to $\overline{\text{CS}}$ high		0		ns
t_{w5}	Pulse duration, $\overline{\text{RD}}$ low		50		ns
t_{en}	Enable time, $\overline{\text{RD}}$ low (or $\overline{\text{CS}}$ low for read cycle) to data valid			20	ns
t_{d2}	Delay time, data hold from $\overline{\text{RD}}$ high		5		ns
t_{d3}	Delay time, BUS18/16 or BYTE rising edge or falling edge to data valid		10	20	ns
t_{w6}	Pulse duration, $\overline{\text{RD}}$ high		20		ns
t_{w7}	Pulse duration, $\overline{\text{CS}}$ high		20		ns
t_{h2}	Hold time, last $\overline{\text{RD}}$ (or $\overline{\text{CS}}$ for read cycle) rising edge to $\overline{\text{CONVST}}$ falling edge		50		ns
t_{pd4}	Propagation delay time, $\overline{\text{BUSY}}$ falling edge to next $\overline{\text{RD}}$ (or $\overline{\text{CS}}$ for read cycle) falling edge		0		ns
t_{d4}	Delay time, BYTE edge to BUS18/16 edge skew		0		ns
t_{su3}	Setup time, BYTE or BUS18/16 transition to $\overline{\text{RD}}$ falling edge		10		ns
t_{h3}	Hold time, BYTE or BUS18/16 transition to $\overline{\text{RD}}$ falling edge		10		ns
t_{dis}	Disable time, $\overline{\text{RD}}$ high ($\overline{\text{CS}}$ high for read cycle) to 3-stated data bus			20	ns
t_{d5}	Delay time, $\overline{\text{BUSY}}$ low to MSB data valid delay			0	ns
t_{d6}	Delay time, $\overline{\text{CS}}$ rising edge to $\overline{\text{BUSY}}$ falling edge		50		ns
t_{d7}	Delay time, $\overline{\text{BUSY}}$ falling edge to $\overline{\text{CS}}$ rising edge		50		ns
t_{su5}	BYTE transition setup time, from BYTE transition to next BYTE transition, or BUS18/16 transition setup time, from BUS18/16 to next BUS18/16 .		50		ns
$t_{\text{su(ABORT)}}$	Setup time from the falling edge of $\overline{\text{CONVST}}$ (used to start the valid conversion) to the next falling edge of $\overline{\text{CONVST}}$ (when $\text{CS} = 0$ and $\overline{\text{CONVST}}$ are used to abort) or to the next falling edge of $\overline{\text{CS}}$ (when $\overline{\text{CS}}$ is used to abort).		60	480	ns

(1) All input signals are specified with $t_r = t_f = 5\text{ ns}$ (10% to 90% of $+VBD$) and timed from a voltage level of $(V_{IL} + V_{IH})/2$.

(2) See timing diagrams.

(3) All timing are measured with 20-pF equivalent loads on all data bits and $\overline{\text{BUSY}}$ pins.

TIMING CHARACTERISTICS

All specifications typical at -40°C to 85°C , $+\text{VA} = 5\text{ V}$ $+\text{VBD} = 3\text{ V}$ ⁽¹⁾ ⁽²⁾ ⁽³⁾

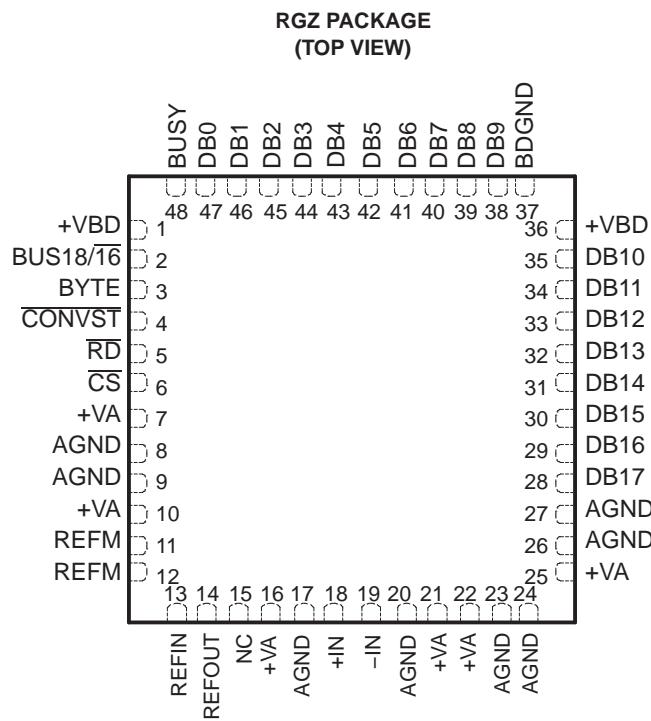
PARAMETER		MIN	TYP	MAX	UNIT
$t_{(\text{CONV})}$	Conversion time			610	ns
$t_{(\text{ACQ})}$	Acquisition time		175		ns
$t_{(\text{HOLD})}$	Sample capacitor hold time			15	ns
t_{pd1}	CONVST low to BUSY high			40	ns
t_{pd2}	Propagation delay time, end of conversion to BUSY low			15	ns
t_{pd3}	Propagation delay time, start of convert state to rising edge of BUSY			25	ns
t_{w1}	Pulse duration, CONVST low		40		ns
t_{su1}	Setup time, CS low to CONVST low		20		ns
t_{w2}	Pulse duration, CONVST high		20		ns
	CONVST falling edge jitter			10	ps
t_{w3}	Pulse duration, BUSY signal low		$t_{(\text{ACQ})\text{min}}$		ns
t_{w4}	Pulse duration, BUSY signal high			610	ns
t_{h1}	Hold time, first data bus transition (RD low, or CS low for read cycle, or BYTE or BUS18/16 input changes) after CONVST low		40		ns
t_{d1}	Delay time, CS low to RD low		0		ns
t_{su2}	Setup time, RD high to CS high		0		ns
t_{w5}	Pulse duration, RD low		50		ns
t_{en}	Enable time, RD low (or CS low for read cycle) to data valid			30	ns
t_{d2}	Delay time, data hold from RD high		5		ns
t_{d3}	Delay time, BUS18/16 or BYTE rising edge or falling edge to data valid		10	30	ns
t_{w6}	Pulse duration, RD high		20		ns
t_{w7}	Pulse duration, CS high		20		ns
t_{h2}	Hold time, last RD (or CS for read cycle) rising edge to CONVST falling edge		50		ns
t_{pd4}	Propagation delay time, BUSY falling edge to next RD (or CS for read cycle) falling edge		0		ns
t_{d4}	Delay time, BYTE edge to BUS18/16 edge skew		0		ns
t_{su3}	Setup time, BYTE or BUS18/16 transition to RD falling edge		10		ns
t_{h3}	Hold time, BYTE or BUS18/16 transition to RD falling edge		10		ns
t_{dis}	Disable time, RD high (CS high for read cycle) to 3-stated data bus			30	ns
t_{d5}	Delay time, BUSY low to MSB data valid delay			0	ns
t_{d6}	Delay time, CS rising edge to BUSY falling edge		50		ns
t_{d7}	Delay time, BUSY falling edge to CS rising edge		50		ns
t_{su5}	BYTE transition setup time, from BYTE transition to next BYTE transition, or BUS18/16 transition setup time, from BUS18/16 to next BUS18/16 .		50		ns
$t_{\text{su(ABORT)}}$	Setup time from the falling edge of CONVST (used to start the valid conversion) to the next falling edge of CONVST (when $\text{CS} = 0$ and CONVST are used to abort) or to the next falling edge of CS (when CS is used to abort).		70	480	ns

(1) All input signals are specified with $t_r = t_f = 5\text{ ns}$ (10% to 90% of $+\text{VBD}$) and timed from a voltage level of $(V_{IL} + V_{IH})/2$.

(2) See timing diagrams.

(3) All timing are measured with 20-pF equivalent loads on all data bits and BUSY pins.

PIN ASSIGNMENTS



NC – No internal connection

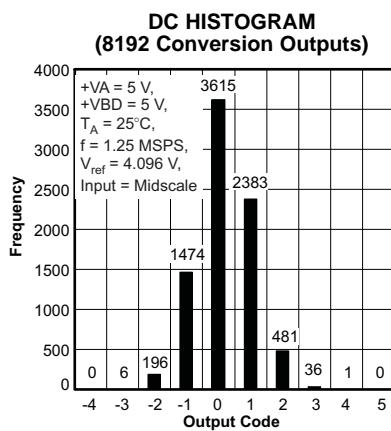
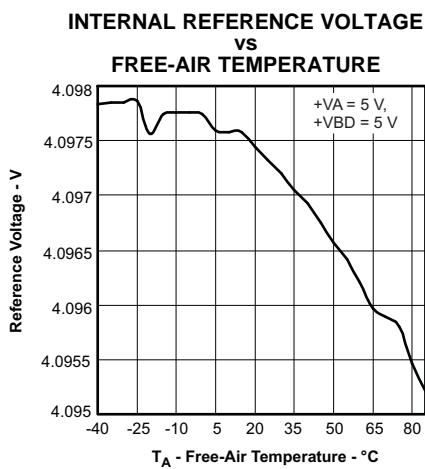
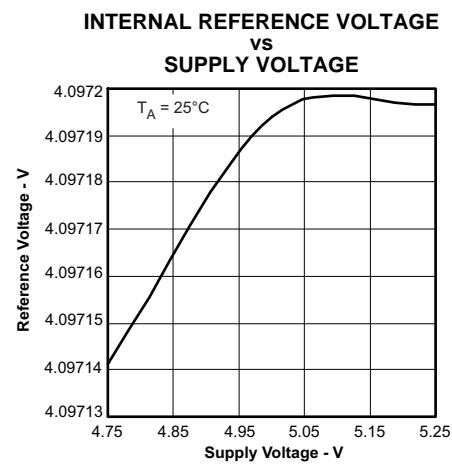
NOTE: The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

TERMINAL FUNCTIONS

NAME	NO	I/O	DESCRIPTION					
AGND	8, 9, 17, 20, 23, 24, 26, 27	–	Analog ground					
BDGND	37	–	Digital ground for bus interface digital supply					
BUSY	48	O	Status output. High when a conversion is in progress.					
BUS18/16	2	I	Bus size select input. Used for selecting 18-bit or 16-bit wide bus transfer. 0: Data bits output on the 18-bit data bus pins DB[17:0]. 1: Last two data bits D1[1:0] from 18-bit wide bus output on: a) the low byte pins DB[9:2] if BYTE = 0 b) the high byte pins DB[17:10] if BYTE = 1					
BYTE	3	I	Byte select input. Used for 8-bit bus reading. 0: No fold back 1: Low byte D[9:2] of the 16 most significant bits is folded back to high byte of the 16 most significant pins DB[17:10].					
CONVST	4	I	Convert start. The falling edge of this input ends the acquisition period and starts the hold period.					
CS	6	I	Chip select. The falling edge of this input starts the acquisition period.					
Data Bus			8-BIT BUS			16-BIT BUS		18-BIT BUS
			BYTE = 0	BYTE = 1	BYTE = 1	BYTE = 0	BYTE = 0	BYTE = 0
			BUS18/16 = 0	BUS18/16 = 0	BUS18/16 = 1	BUS18/16 = 0	BUS18/16 = 1	BUS18/16 = 0
DB17	28	O	D17 (MSB)	D9	All ones	D17 (MSB)	All ones	D17 (MSB)
DB16	29	O	D16	D8	All ones	D16	All ones	D16
DB15	30	O	D15	D7	All ones	D15	All ones	D15
DB14	31	O	D14	D6	All ones	D14	All ones	D14
DB13	32	O	D13	D5	All ones	D13	All ones	D13
DB12	33	O	D12	D4	All ones	D12	All ones	D12
DB11	34	O	D11	D3	D1	D11	All ones	D11
DB10	35	O	D10	D2	D0 (LSB)	D10	All ones	D10

TERMINAL FUNCTIONS (continued)

NAME	NO	I/O	DESCRIPTION					
DB9	38	O	D9	All ones	All ones	D9	All ones	D9
DB8	39	O	D8	All ones	All ones	D8	All ones	D8
DB7	40	O	D7	All ones	All ones	D7	All ones	D7
DB6	41	O	D6	All ones	All ones	D6	All ones	D6
DB5	42	O	D5	All ones	All ones	D5	All ones	D5
DB4	43	O	D4	All ones	All ones	D4	All ones	D4
DB3	44	O	D3	All ones	All ones	D3	D1	D3
DB2	45	O	D2	All ones	All ones	D2	D0 (LSB)	D2
DB1	46	O	D1	All ones	All ones	D1	All ones	D1
DB0	47	O	D0 (LSB)	All ones	All ones	D0 (LSB)	All ones	D0 (LSB)
–IN	19	I	Inverting input channel					
+IN	18	I	Noninverting input channel					
NC	15		No connection					
REFIN	13	I	Reference input					
REFOUT	14	O	Reference output. Add 1- μ F capacitor between the REFOUT pin and REFM pin when internal reference is used.					
REFM	11, 12	I	Reference ground					
RD	5	I	Synchronization pulse for the parallel output. When CS is low, this serves as output enable and puts the previous conversion results on the bus.					
+VA	7, 10, 16, 21, 22, 25	–	Analog power supplies, 5-V DC					
+VBD	1, 36	–	Digital power supply for bus					

TYPICAL CHARACTERISTICS

Figure 1.

Figure 2.

Figure 3.

TYPICAL CHARACTERISTICS (continued)

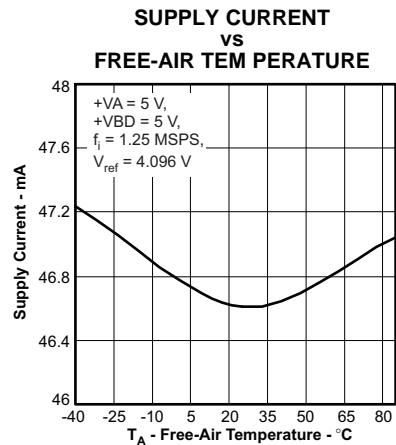


Figure 4.

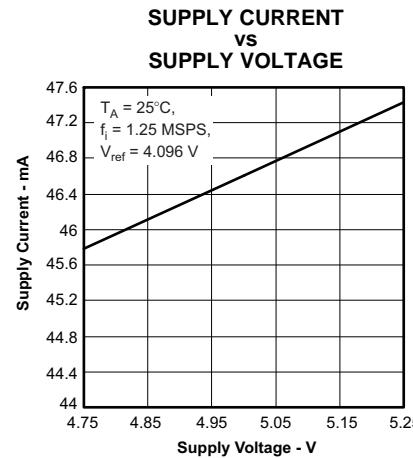


Figure 5.

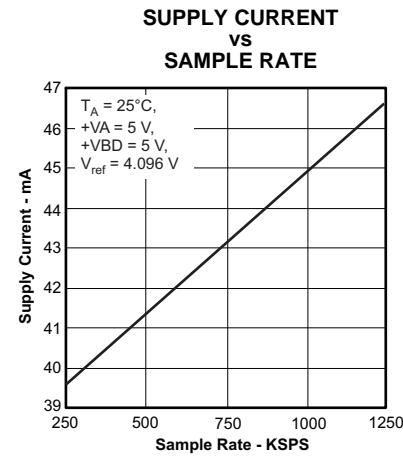


Figure 6.

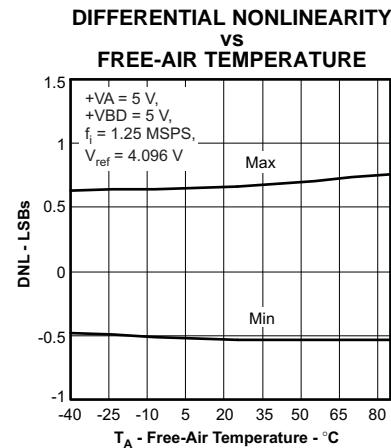


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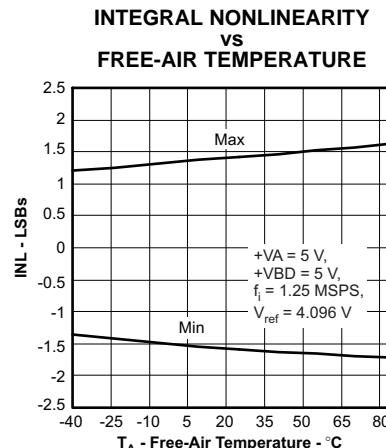


Figure 8.

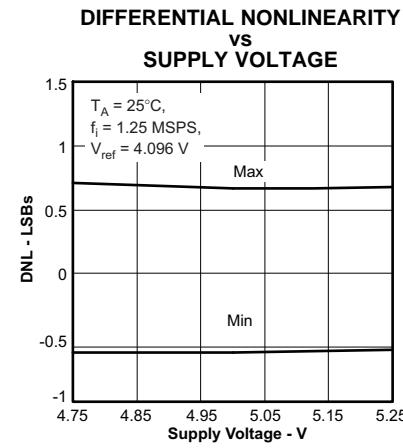


Figure 9.

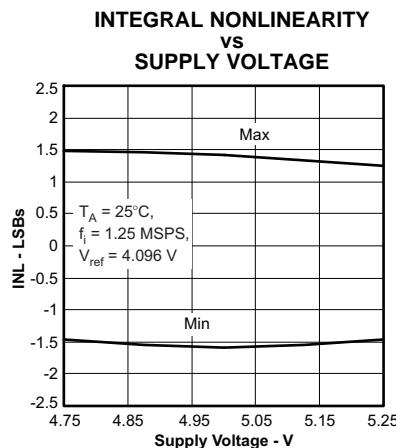


Figure 10.

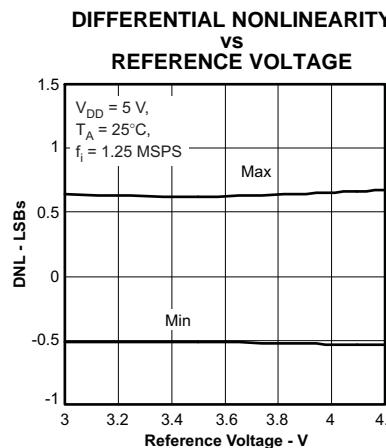


Figure 11.

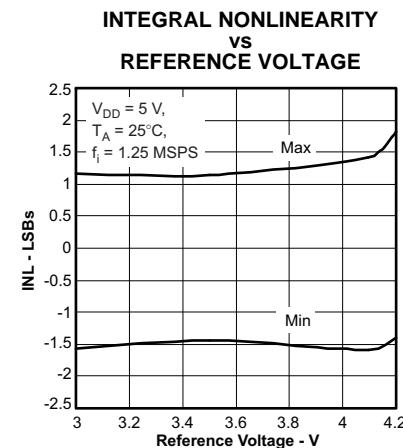


Figure 12.

TYPICAL CHARACTERISTICS (continued)

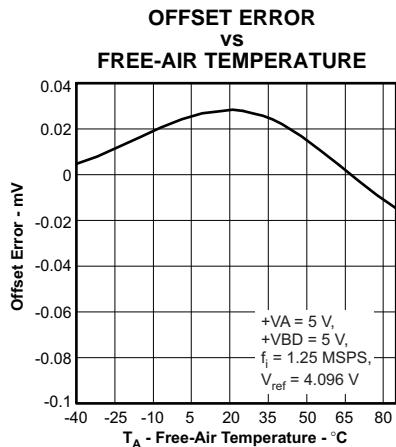


Figure 13.

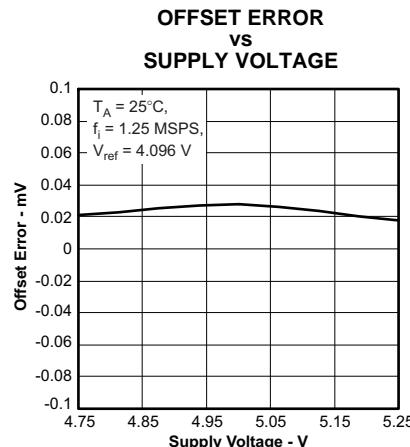


Figure 14.

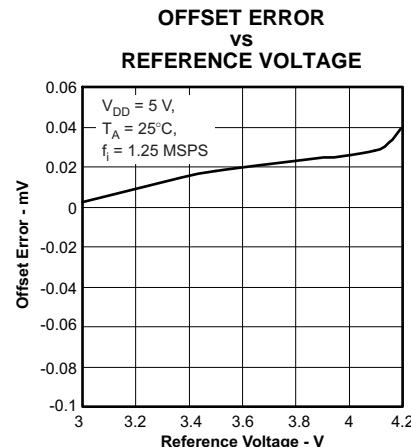


Figure 15.

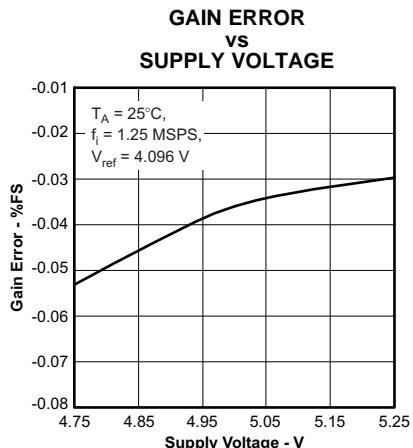


Figure 16.

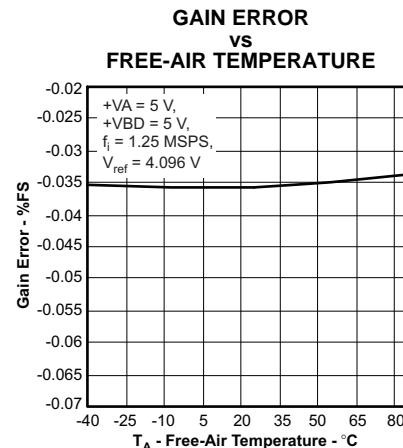


Figure 17.

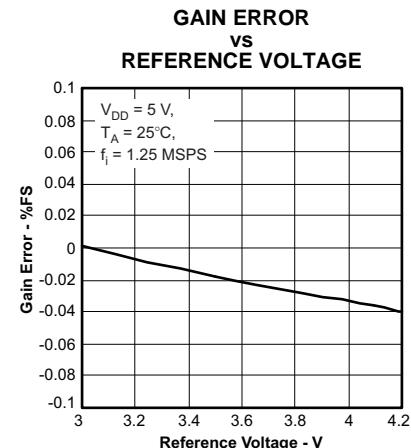


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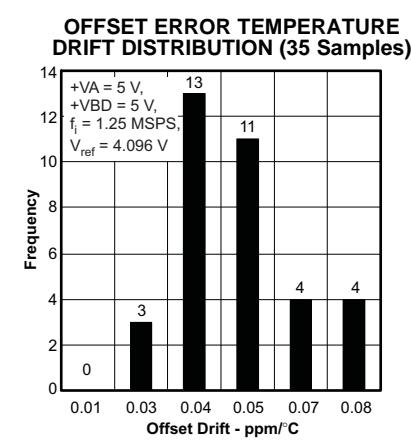


Figure 19.

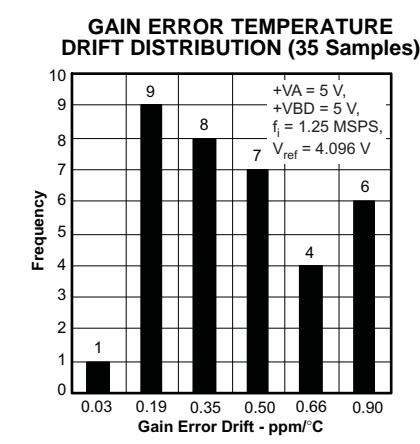


Figure 20.

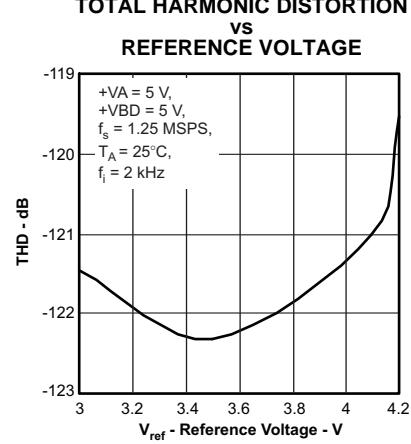


Figure 21.

TYPICAL CHARACTERISTICS (continued)

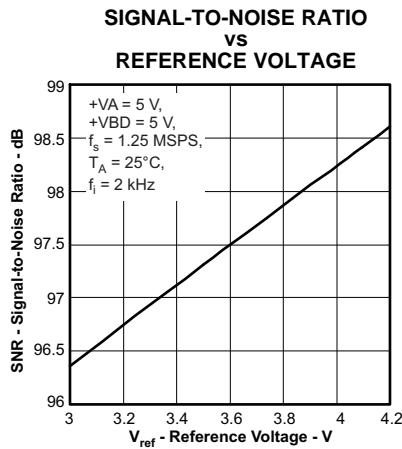


Figure 22.

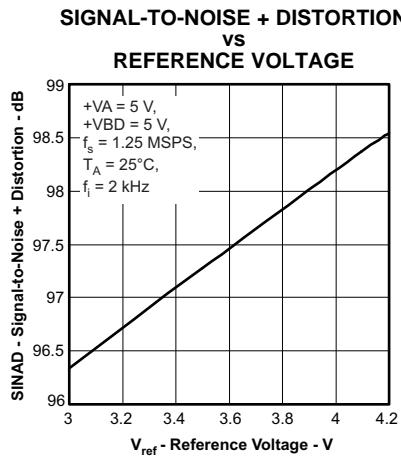


Figure 23.

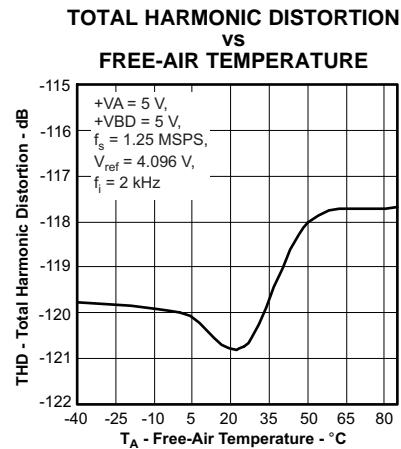


Figure 24.

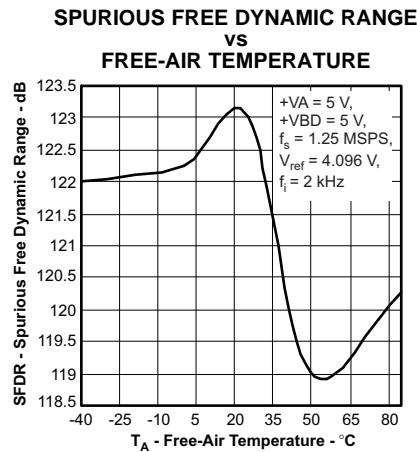


Figure 25.

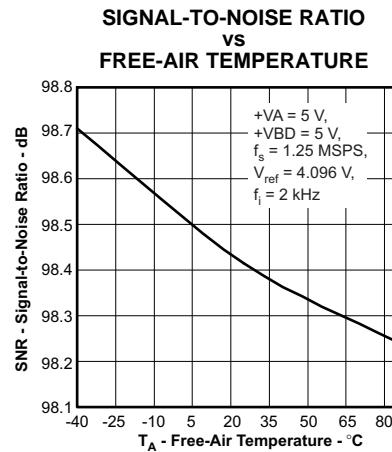
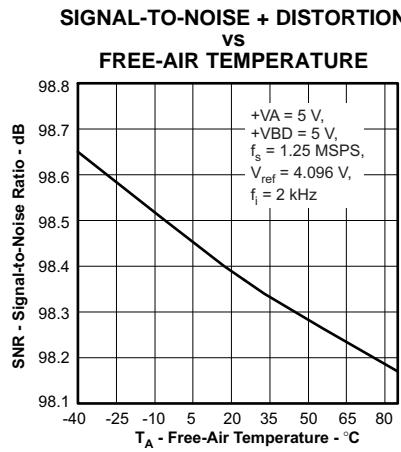
Figure 26.
DNL

Figure 27.

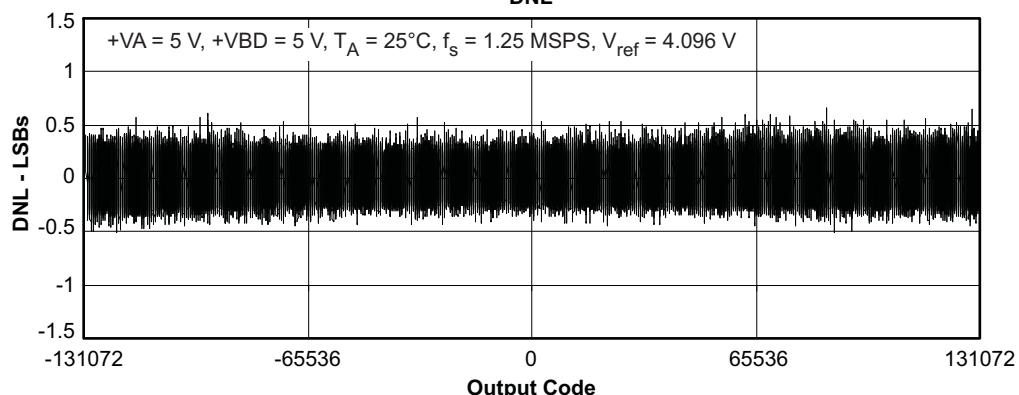


Figure 28.

TYPICAL CHARACTERISTICS (continued)

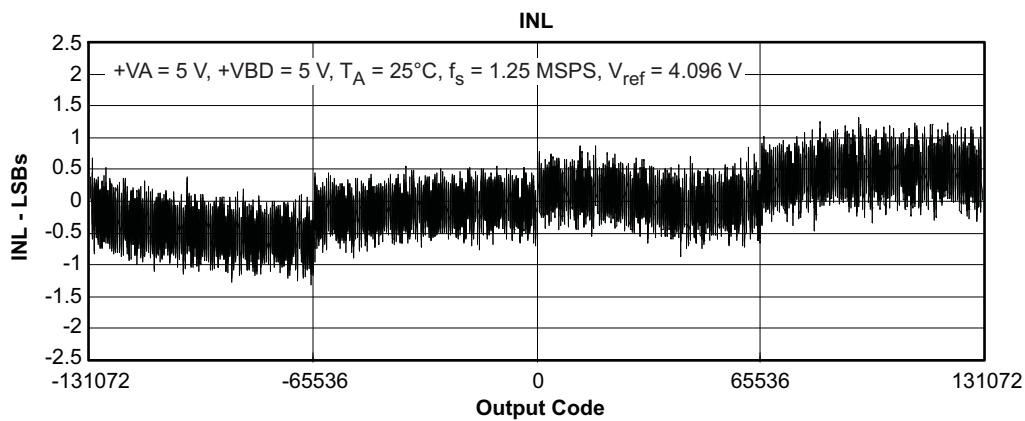


Figure 29.

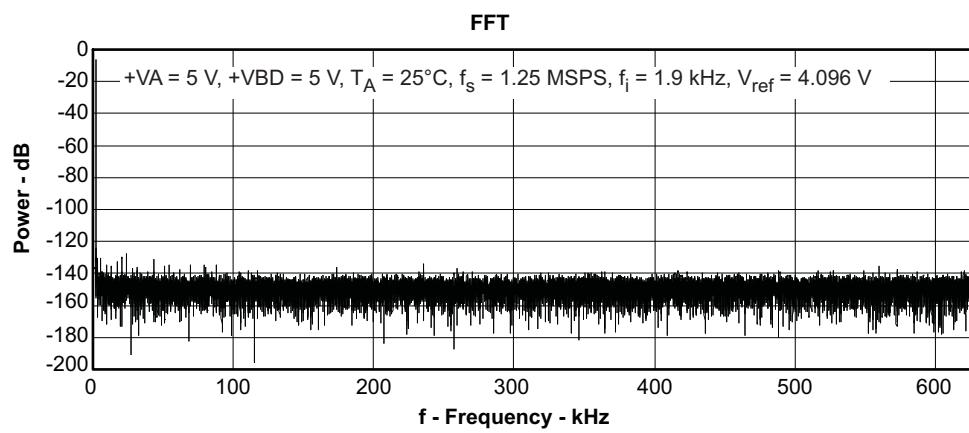
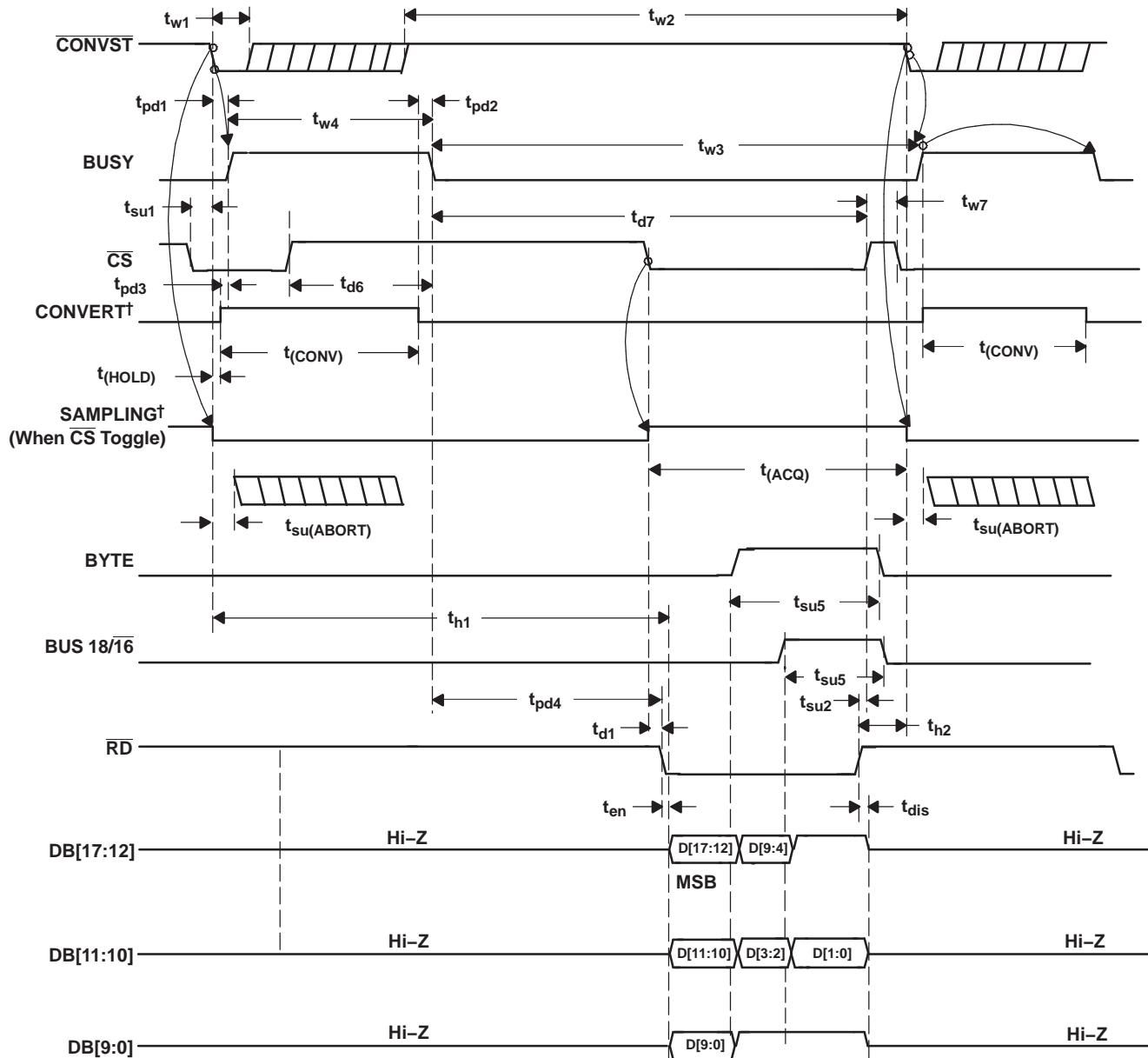


Figure 30.

TYPICAL CHARACTERISTICS (continued)

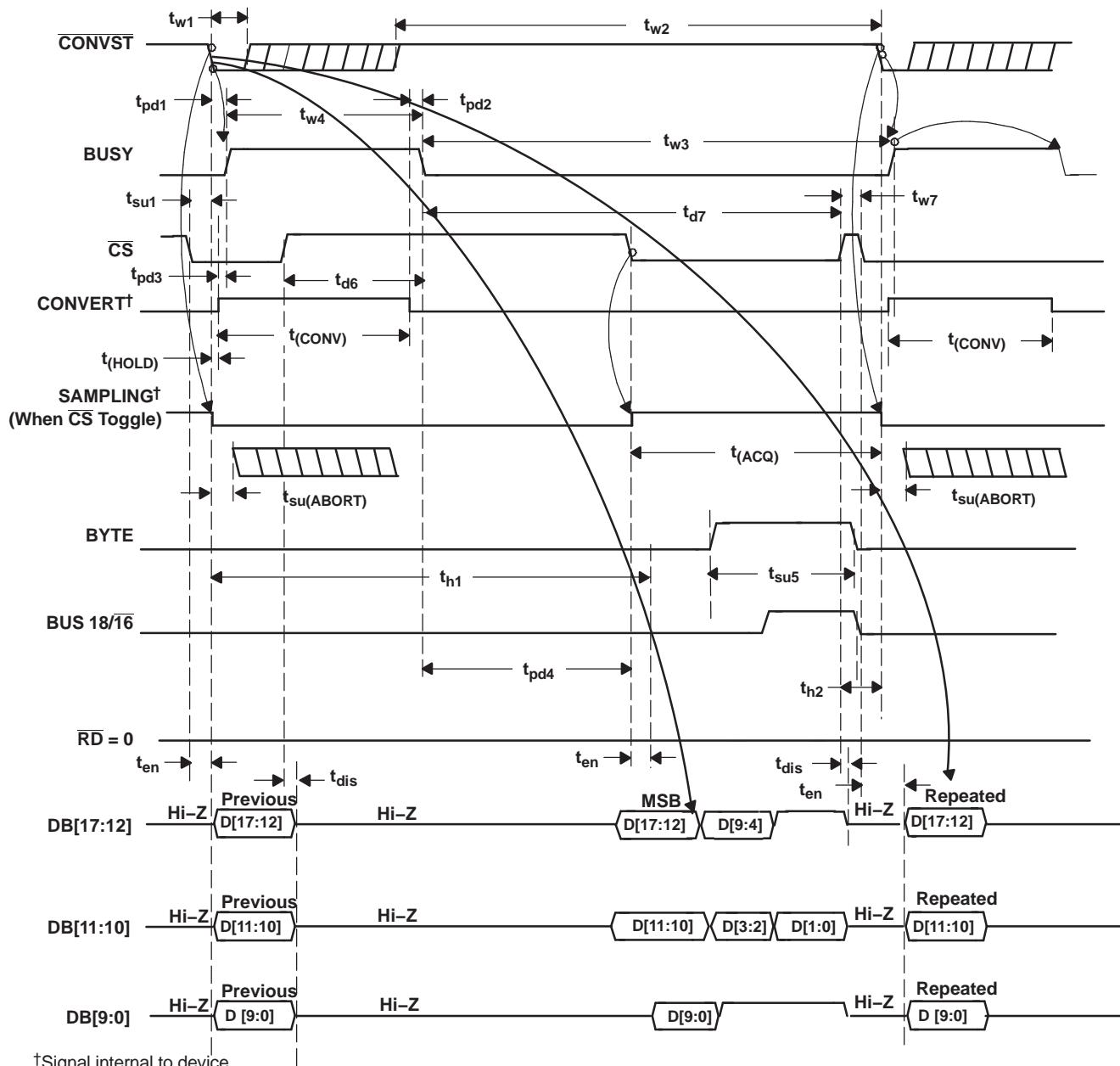
TIMING DIAGRAMS



[†]Signal internal to device

Figure 31. Timing for Conversion and Acquisition Cycles With \overline{CS} and \overline{RD} Toggling

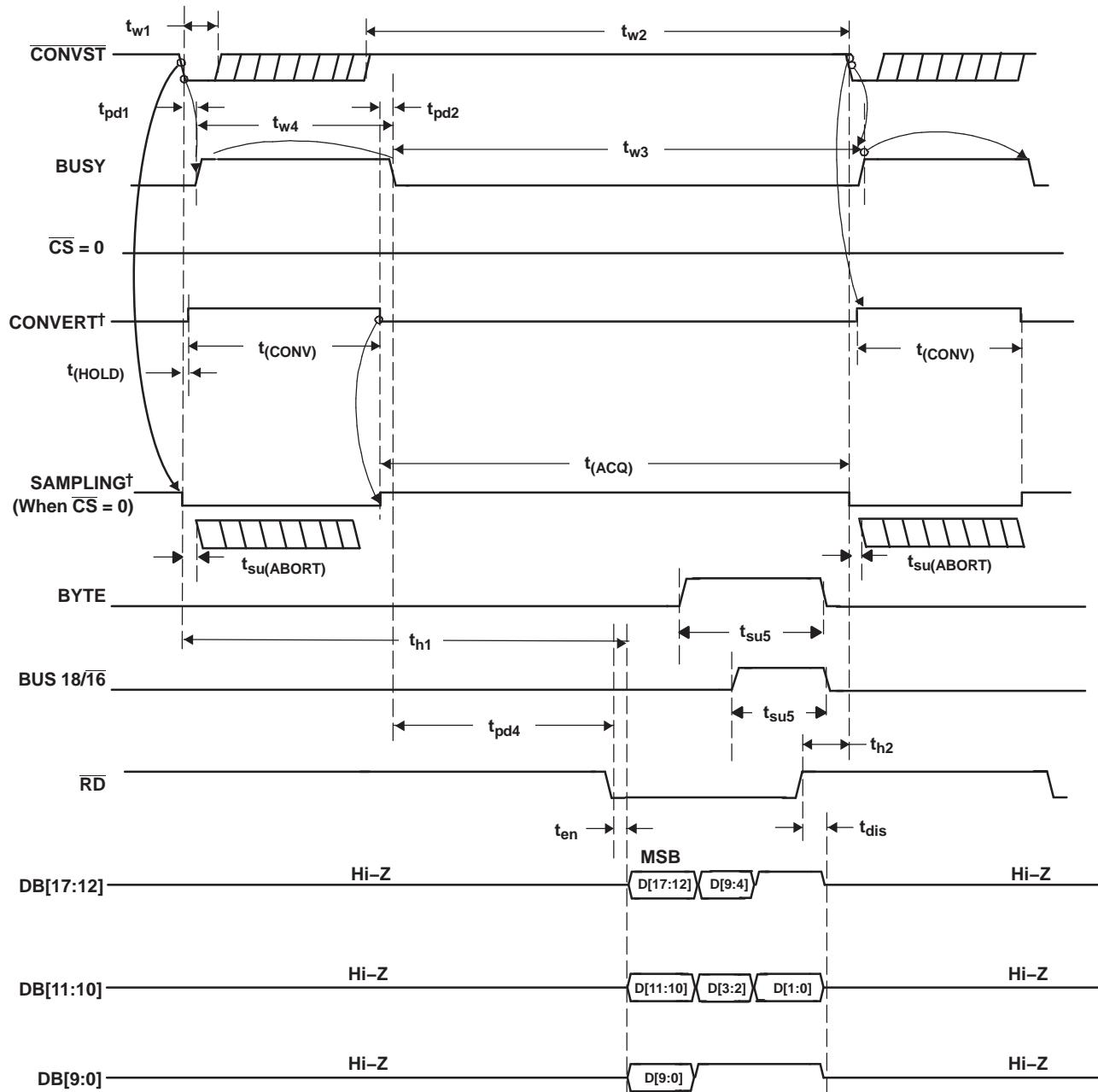
TYPICAL CHARACTERISTICS (continued)



[†]Signal internal to device

Figure 32. Timing for Conversion and Acquisition Cycles With \overline{CS} Toggling, \overline{RD} Tied to BDGND

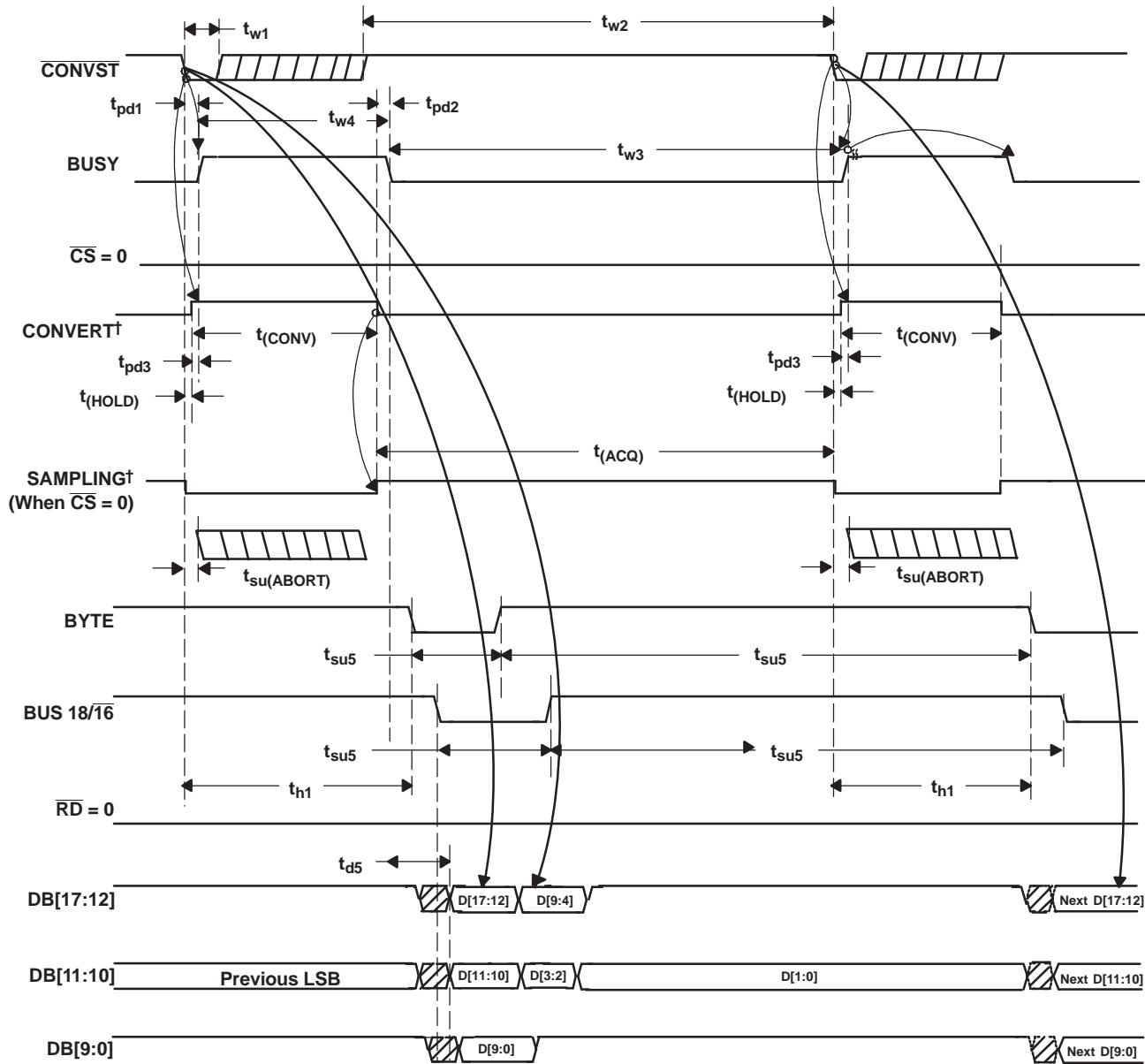
TYPICAL CHARACTERISTICS (continued)



†Signal internal to device

Figure 33. Timing for Conversion and Acquisition Cycles With \overline{CS} Tied to BDGND, \overline{RD} Toggling

TYPICAL CHARACTERISTICS (continued)



†Signal internal to device

Figure 34. Timing for Conversion and Acquisition Cycles With \overline{CS} and \overline{RD} Tied to BDGND - Auto Read

TYPICAL CHARACTERISTICS (continued)

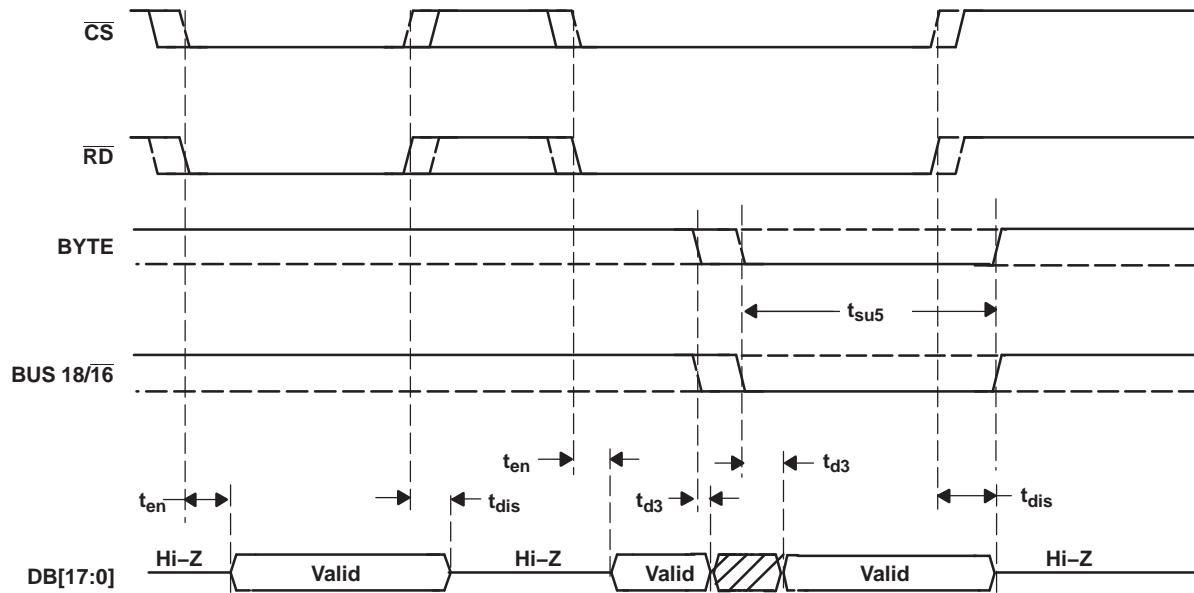


Figure 35. Detailed Timing for Read Cycles

APPLICATION INFORMATION

MICROCONTROLLER INTERFACING

ADS8484 to 8-Bit Microcontroller Interface

Figure 36 shows a parallel interface between the ADS8484 and a typical microcontroller using the 8-bit data bus. The BUSY signal is used as a falling-edge interrupt to the microcontroller.

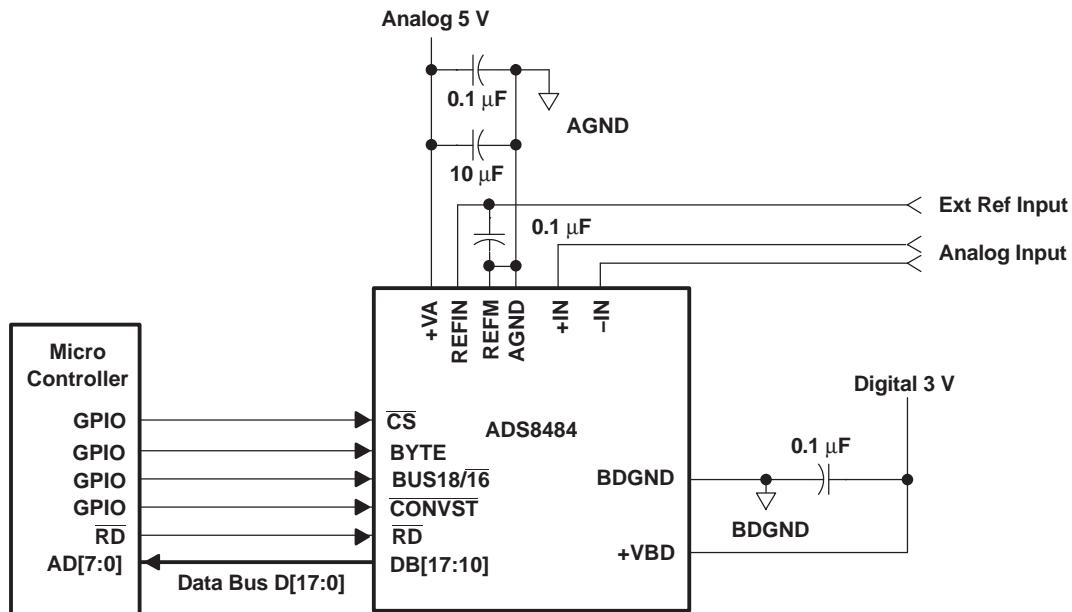


Figure 36. ADS8484 Application Circuitry

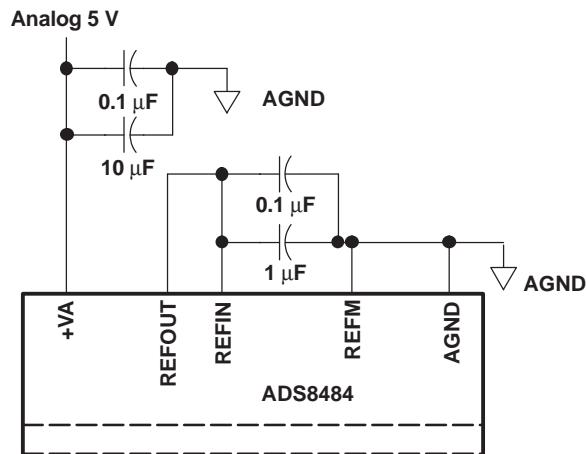


Figure 37. ADS8484 Using Internal Reference

PRINCIPLES OF OPERATION

The ADS8484 is a high-speed successive approximation register (SAR) analog-to-digital converter (ADC). The architecture is based on charge redistribution which inherently includes a sample/hold function. See [Figure 36](#) for the application circuit for the ADS8484.

The conversion clock is generated internally. The conversion time of 610 ns is capable of sustaining a 1.25-MHz throughput.

The analog input is provided to two input pins: +IN and –IN. When a conversion is initiated, the differential input on these pins is sampled on the internal capacitor array. While a conversion is in progress, both inputs are disconnected from any internal function.

REFERENCE

The ADS8484 can operate with an external reference with a range from 3.0 V to 4.2 V. The reference voltage on the input pin 13 (REFIN) of the converter is internally buffered. A clean, low noise, well-decoupled reference voltage on this pin is required to ensure good performance of the converter. A low noise band-gap reference like the REF3240 can be used to drive this pin. A 0.1- μ F decoupling capacitor is required between REFIN and REFM pins (pin 13 and pin 12) of the converter. This capacitor should be placed as close as possible to the pins of the device. Designers should strive to minimize the routing length of the traces that connect the terminals of the capacitor to the pins of the converter. An RC network can also be used to filter the reference voltage. A 100- Ω series resistor and a 0.1- μ F capacitor, which can also serve as the decoupling capacitor can be used to filter the reference voltage.

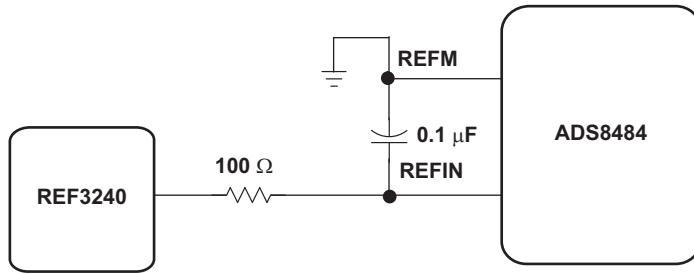


Figure 38. ADS8484 Using External Reference

The ADS8484 also has limited low pass filtering capability built into the converter. The equivalent circuitry on the REFIN input is as shown in [Figure 39](#).

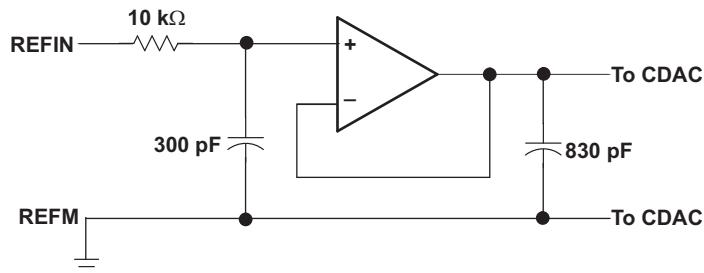


Figure 39. Simplified Reference Input Circuit

The REFM input of the ADS8484 should always be shorted to AGND. A 4.096-V internal reference is included. When internal reference is used, pin 14 (REFOUT) is connected to pin 13 (REFIN) with an 0.1- μ F decoupling capacitor and 1- μ F storage capacitor between pin 14 (REFOUT) and pins 11 and 12 (REFM) (see [Figure 37](#)). The internal reference of the converter is double buffered. If an external reference is used, the second buffer provides isolation between the external reference and the CDAC. This buffer is also used to recharge all of the capacitors of the CDAC during conversion. Pin 14 (REFOUT) can be left unconnected (floating) if external reference is used.

ANALOG INPUT

When the converter enters the hold mode, the voltage difference between the +IN and –IN inputs is captured on the internal capacitor array. Both +IN and –IN input has a range of -0.2 V to $V_{\text{ref}} + 0.2\text{ V}$. The input span [+IN – (–IN)] is limited to $-V_{\text{ref}}$ to V_{ref} .

The input current on the analog inputs depends upon a number of factors: sample rate, input voltage, and source impedance. Essentially, the current into the ADS8484 charges the internal capacitor array during the sample period. After this capacitance has been fully charged, there is no further input current. The source of the analog input must be able to charge the input capacitance (65 pF) to an 18-bit settling level within the acquisition time (175 ns) of the device. When the converter goes into the hold mode, the input impedance is greater than 1 G Ω .

Care must be taken regarding the absolute analog input voltage. To maintain the linearity of the converter, the +IN and –IN inputs and the span [+IN – (–IN)] must be within the limits specified. Outside of these ranges, the converter's linearity may not meet specifications. To minimize noise, low bandwidth input signals with low-pass filters are used.

Care must be taken to ensure that the output impedance of the sources driving the +IN and –IN inputs are matched. If this is not observed, the two inputs could have different settling times. This may result in offset error, gain error, and linearity error which varies with temperature and input voltage.

The analog input to the converter needs to be driven with a low noise, high-speed op-amp like the THS4031. An RC filter is recommended at the input pins to low-pass filter the noise from the source. The input to the converter is a uni-polar input voltage in the range 0 to V_{ref} . The THS4031 can be used in the source follower configuration to drive the converter.

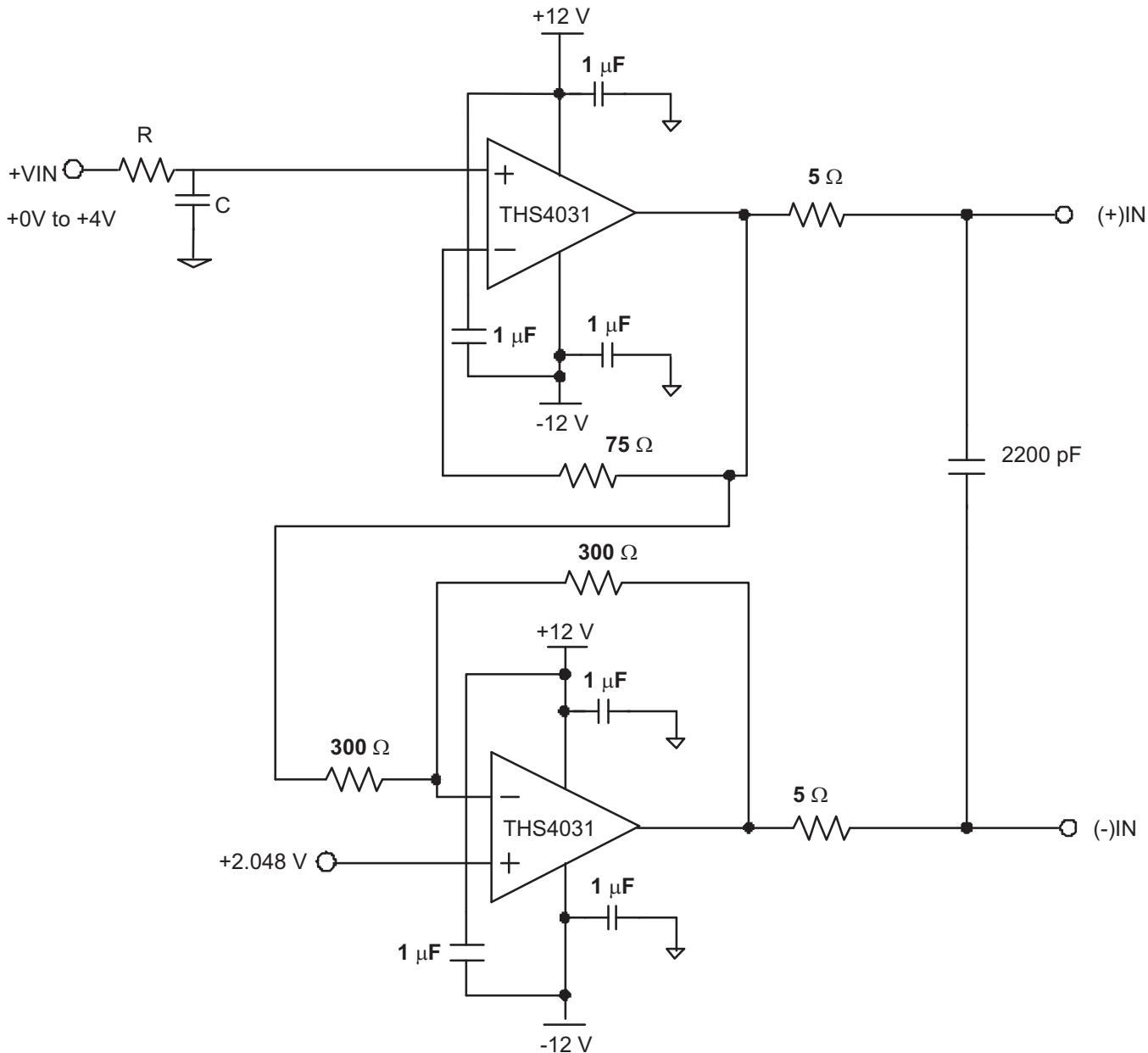


Figure 40. Single-Ended Input, Differential Output Configuration

In systems, where the input is differential, the THS4031 can be used in the inverting configuration with an additional DC bias applied to its $+$ input so as to keep the input to the ADS8484 within its rated operating voltage range. The DC bias can be derived from the REF3220 or the REF3240 reference voltage ICs. The input configuration shown below is capable of delivering better than 97dB SNR and -103db THD at an input frequency of 100 kHz. In case band-pass filters are used to filter the input, care should be taken to ensure that the signal swing at the input of the band-pass filter is small so as to keep the distortion introduced by the filter minimal. In such cases, the gain of the circuit shown below can be increased to keep the input to the ADS8484 large to keep the SNR of the system high. Note that the gain of the system from the $+$ input to the output of the THS4031 in such a configuration is a function of the gain of the AC signal. A resistor divider can be used to scale the output of the REF3220 or REF3240 to reduce the voltage at the DC input to THS4031 to keep the voltage at the input of the converter within its rated operating range.

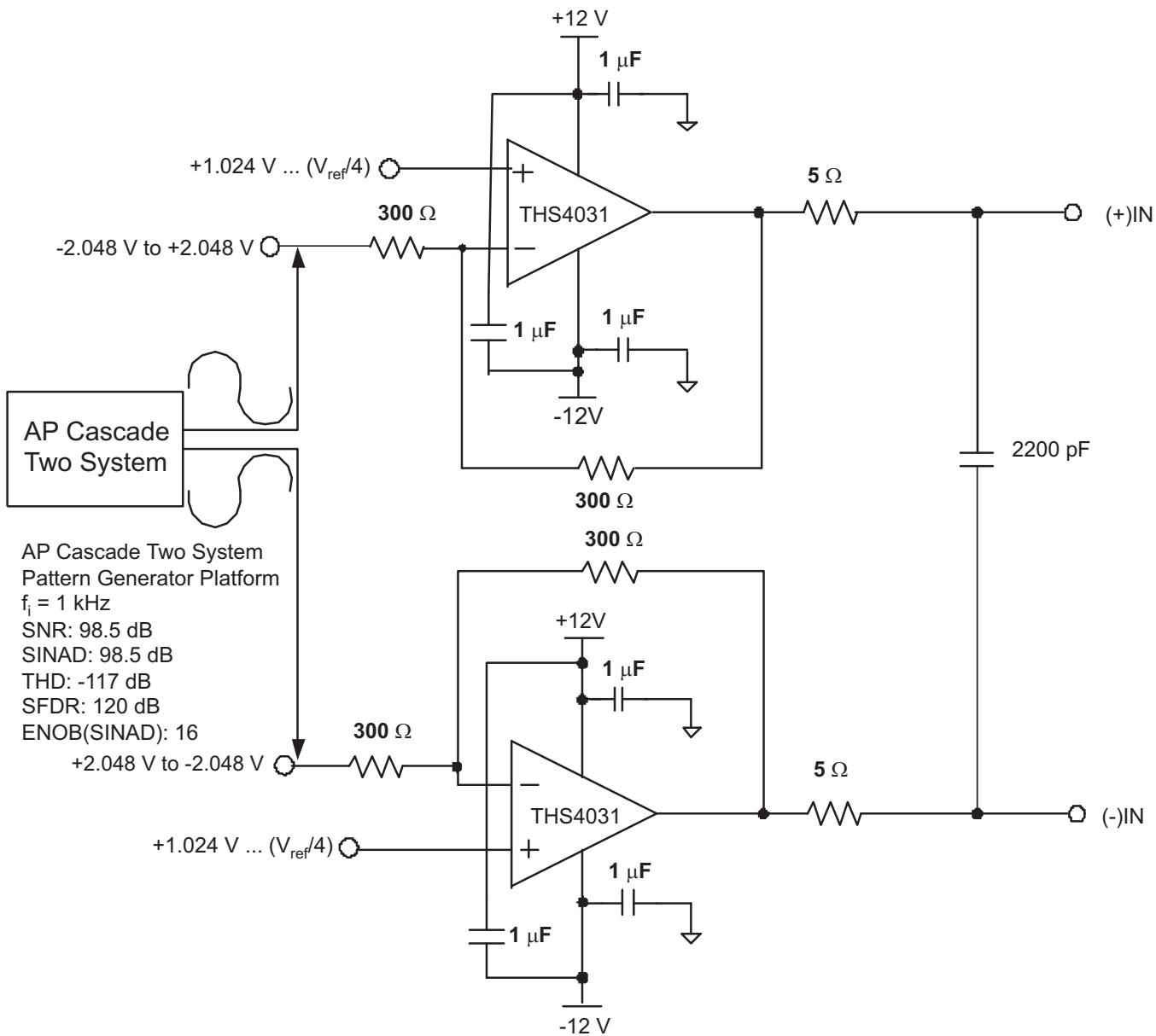


Figure 41. Differential Input, Differential Output Configuration

DIGITAL INTERFACE

Timing and Control

See the timing diagrams in the specifications section for detailed information on timing signals and their requirements.

The ADS8484 uses an internal oscillator generated clock which controls the conversion rate and in turn the throughput of the converter. No external clock input is required.

Conversions are initiated by bringing the CONVST pin low for a minimum of 20 ns (after the 20 ns minimum requirement has been met, the CONVST pin can be brought high), while CS is low. The ADS8484 switches from the sample to the hold mode on the falling edge of the CONVST command. A clean and low jitter falling edge of this signal is important to the performance of the converter. The BUSY output is brought high immediately following CONVST going low. BUSY stays high throughout the conversion process and returns low when the conversion has ended.

Sampling starts t_{pd} ns before the falling edge of the BUSY signal when CS is tied low or starts with the falling edge of CS when BUSY is low.

Both RD and CS can be high during and before a conversion with one exception (CS must be low when CONVST goes low to initiate a conversion). Both the RD and CS pins are brought low in order to enable the parallel output bus with the conversion.

Reading Data

The ADS8484 outputs full parallel data in two's complement format as shown in [Table 1](#). The parallel output is active when CS and RD are both low. There is a minimal quiet zone requirement around the falling edge of CONVST. This is 50 ns prior to the falling edge of CONVST and 40 ns after the falling edge. No data read should be attempted within this zone. Any other combination of CS and RD sets the parallel output to 3-state. BYTE and BUS18/16 are used for multiword read operations. BYTE is used whenever lower bits on the bus are output on the higher byte of the bus. BUS18/16 is used whenever the last two bits on the 18-bit bus are output on either bytes of the higher 16-bit bus. Refer to [Table 1](#) for ideal output codes.

Table 1. Ideal Input Voltages and Output Codes

DESCRIPTION	ANALOG VALUE	DIGITAL OUTPUT TWO'S COMPLEMENT	
		BINARY CODE	HEX CODE
Full scale range	$+V_{ref}$		
Least significant bit (LSB)	$2 \times (+V_{ref})/262144$		
+Full scale	$(+V_{ref}) - 1 \text{ LSB}$	01 1111 1111 1111 1111	1FFF
Midscale	0 V	00 0000 0000 0000 0000	00000
Midscale – 1 LSB	0 V – 1 LSB	11 1111 1111 1111 1111	3FFF
Zero	$-V_{ref}$	10 0000 0000 0000 0000	20000

The output data is a full 18-bit word (D17–D0) on DB17–DB0 pins (MSB–LSB) if both BUS18/16 and BYTE are low.

The result may also be read on an 16-bit bus by using only pins DB17–DB2. In this case two reads are necessary: the first as before, leaving both BUS18/16 and BYTE low and reading the 16 most significant bits (D17–D2) on pins DB17–DB2, then bringing BUS18/16 high while holding BYTE low. When BUS18/16 is high, the lower two bits (D1–D0) appear on pins DB3–DB2.

The result may also be read on an 8-bit bus for convenience. This is done by using only pins DB17–DB10. In this case three reads are necessary: the first as before, leaving both BUS18/16 and BYTE low and reading the 8 most significant bits on pins DB17–DB10, then bringing BYTE high while holding BUS18/16 low. When BYTE is high, the medium bits (D9–D2) appear on pins DB17–DB10. The last read is done by bringing BUS18/16 high while holding BYTE high. When BUS18/16 is high, the lower two bits (D1–D0) appear on pins DB11–DB10. The last read cycle is not necessary if only the first 16 most significant bits are of interest.

All of these multiword read operations can be performed with multiple active RD (toggling) or with RD held low for simplicity. This is referred to as the AUTO READ operation.

Table 2. Conversion Data Read Out

BYTE	BUS18/16	DATA READ OUT				
		PINS DB17–DB12	PINS DB11–DB10	PINS DB9–DB4	PINS DB3–DB2	PINS DB1–DB0
High	High	All One's	D1–D0	All One's	All One's	All One's
Low	High	All One's	All One's	All One's	D1–D0	All One's
High	Low	D9–D4	D3–D2	All One's	All One's	All One's
Low	Low	D17–D12	D11–D10	D9–D4	D3–D2	D1–D0

RESET

On power-up, internal POWER-ON RESET circuitry generates the reset required for the device. The first three conversions after power-up are used to load factory trimming data for a specific device to assure high accuracy of the converter. The results of the first three conversions are invalid and should be discarded.

The device can also be reset through the use of the combination of \overline{CS} and \overline{CONVST} . Since the BUSY signal is held at high during the conversion, either one of these conditions triggers an internal self-clear reset to the converter.

- Issue a \overline{CONVST} when \overline{CS} is low and the internal convert state is high. The falling edge of \overline{CONVST} starts a reset.
- Issue a \overline{CS} (select the device) while the internal convert state is high. The falling edge of \overline{CS} causes a reset.

Once the device is reset, all output latches are cleared (set to zeroes) and the BUSY signal is brought low. A new sampling period is started at the falling edge of the BUSY signal immediately after the instant of the internal reset.

LAYOUT

For optimum performance, care must be taken with the physical layout of the ADS8484 circuitry.

As the ADS8484 offers single-supply operation, it is often used in close proximity with digital logic, microcontrollers, microprocessors, and digital signal processors. The more digital logic present in the design and the higher the switching speed, the more difficult it is to achieve good performance from the converter.

The basic SAR architecture is sensitive to glitches or sudden changes on the power supply, reference, ground connections and digital inputs that occur just prior to latching the output of the analog comparator. Thus, driving any single conversion for an n-bit SAR converter, there are at least n windows in which large external transient voltages can affect the conversion result. Such glitches might originate from switching power supplies, nearby digital logic, or high power devices.

The degree of error in the digital output depends on the reference voltage, layout, and the exact timing of the external event.

On average, the ADS8484 draws very little current from an external reference as the reference voltage is internally buffered. If the reference voltage is external and originates from an op amp, make sure that it can drive the bypass capacitor or capacitors without oscillation. A 0.1- μ F capacitor is recommended from pin 13 (REFIN) directly to pin 12 (REFM). REFM and AGND must be shorted on the same ground plane under the device.

The AGND and BDGND pins should be connected to a clean ground point. In all cases, this should be the analog ground. Avoid connections which are too close to the grounding point of a microcontroller or digital signal processor. If required, run a ground trace directly from the converter to the power supply entry point. The ideal layout consists of an analog ground plane dedicated to the converter and associated analog circuitry.

As with the AGND connections, +VA should be connected to a 5-V power supply plane or trace that is separate from the connection for digital logic until they are connected at the power entry point. Power to the ADS8484 should be clean and well bypassed. A 0.1- μ F ceramic bypass capacitor should be placed as close to the device as possible. See [Table 3](#) for the placement of the capacitor. In addition, a 1- μ F to 10- μ F capacitor is recommended. In some situations, additional bypassing may be required, such as a 100- μ F electrolytic capacitor or even a Pi filter made up of inductors and capacitors-all designed to essentially low-pass filter the 5-V supply, removing the high frequency noise.

Table 3. Power Supply Decoupling Capacitor Placement

POWER SUPPLY PLANE	CONVERTER ANALOG SIDE	CONVERTER DIGITAL SIDE
SUPPLY PINS		
Pin pairs that require shortest path to decoupling capacitors	(7,8), (9,10), (16,17), (20,21), (22,23), (25,26)	(36,37)
Pins that require no decoupling	24, 26	1

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
ADS8484IBRGZT	NRND	Production	VQFN (RGZ) 48	250 SMALL T&R	Yes	Call TI	Level-2-260C-1 YEAR	-40 to 85	ADS8484I B
ADS8484IBRGZT.A	NRND	Production	VQFN (RGZ) 48	250 SMALL T&R	Yes	Call TI	Level-2-260C-1 YEAR	-40 to 85	ADS8484I B

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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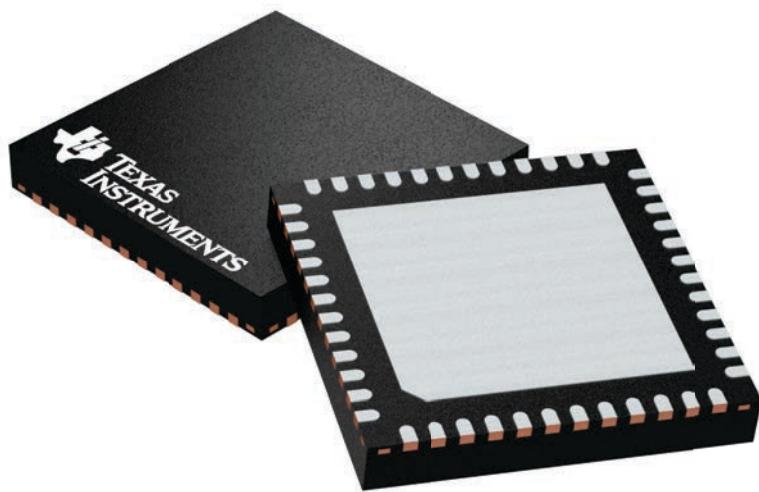
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

7 x 7, 0.5 mm pitch

PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

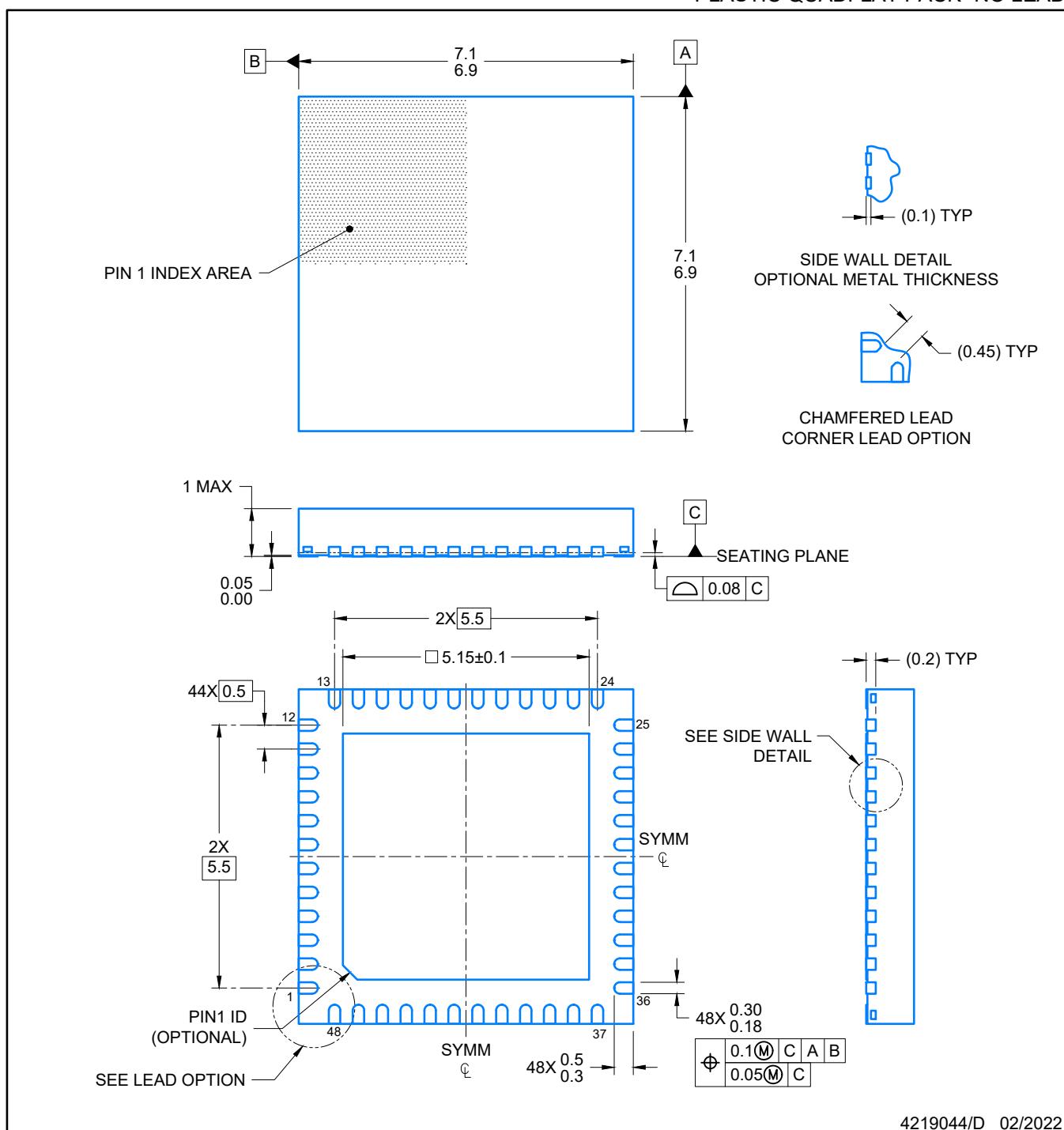
4224671/A

PACKAGE OUTLINE

VQFN - 1 mm max height

RGZ0048A

PLASTIC QUADFLAT PACK- NO LEAD



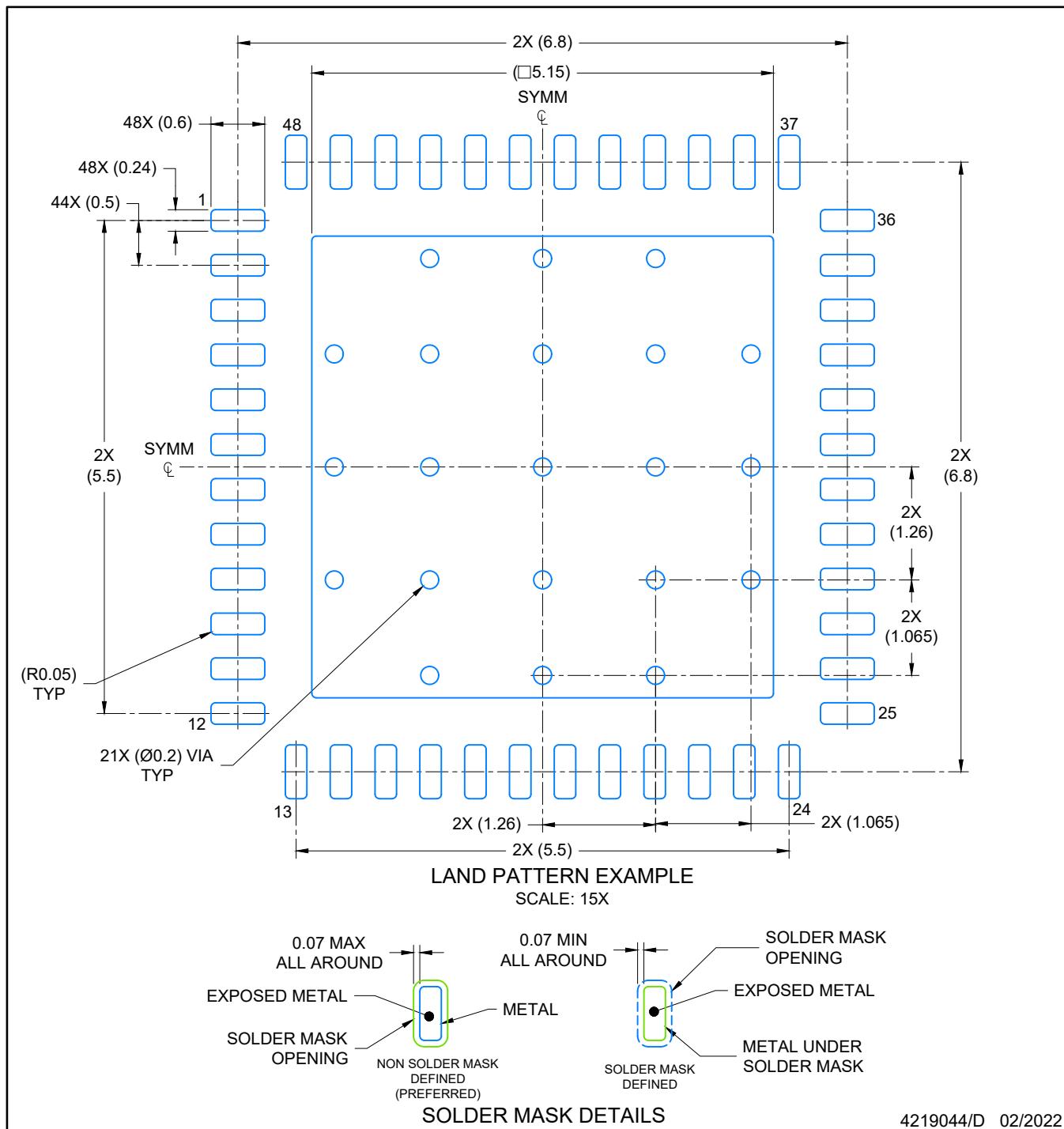
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

VQFN - 1 mm max height

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NOTES: (continued)

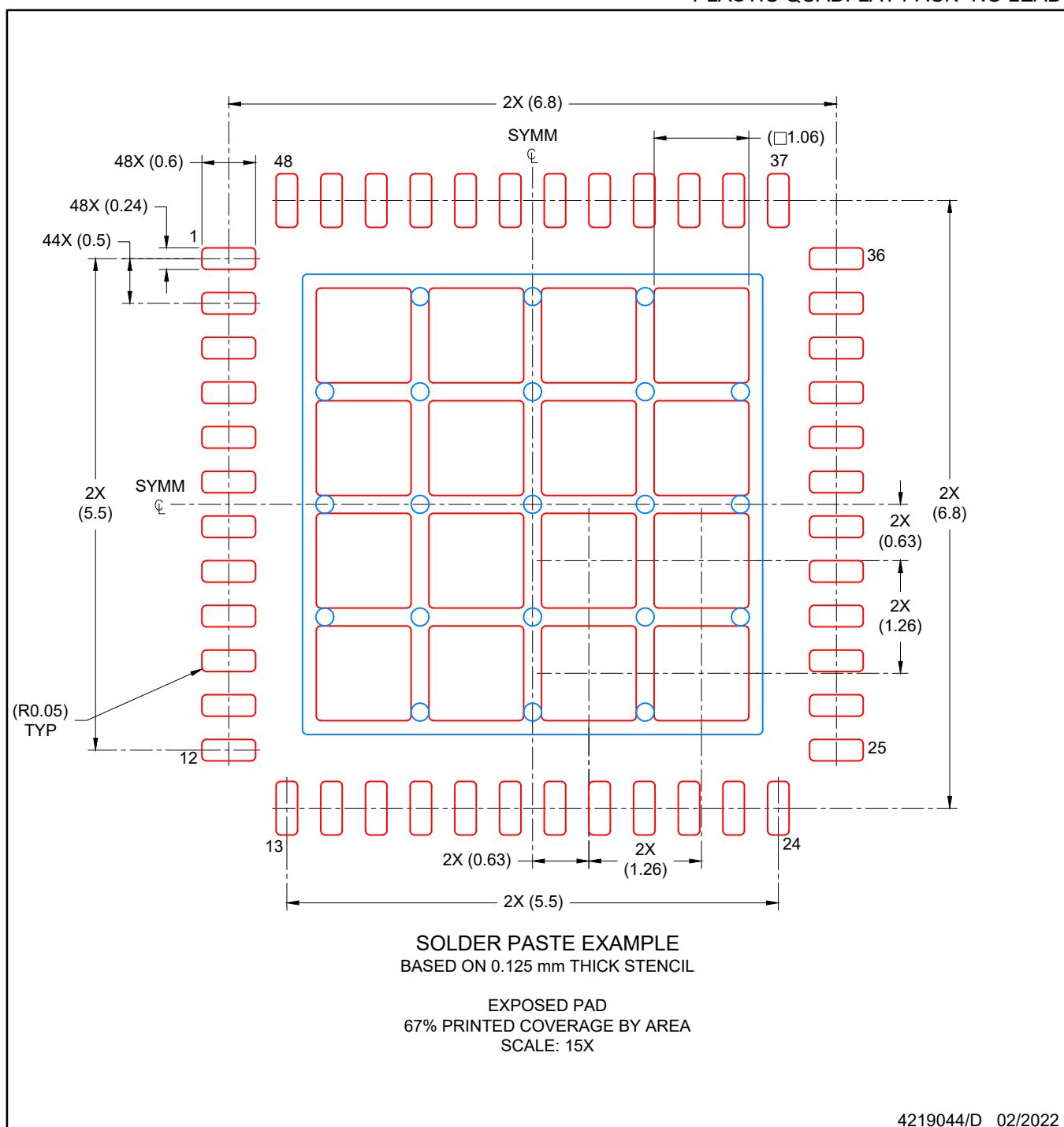
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

VQFN - 1 mm max height

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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