

FEATURES

Passive, wideband I/Q mixer
 RF and LO range: 6 GHz to 26.5 GHz
 Wide IF bandwidth of dc to 5 GHz
 Single-ended RF, LO, and IF
 Conversion loss: 9 dB (typical)
 Image rejection: 25 dBc (typical)
 Noise figure: 14 dB (typical)
 Input IP3 (downconverter): 24 dBm (typical)
 Input P1dB compression point (downconverter): 15 dBm (typical)
 Input IP2: 55 dBm (typical)
 LO to RF isolation: 40 dB (typical)
 LO to IF isolation: 40 dB (typical)
 RF to IF isolation: 20 dB (typical)
 Amplitude balance: ± 0.5 dB (typical)
 Phase balance (downconverter): $\pm 5^\circ$ (typical)
 RF return loss: 15 dB (typical)
 LO return loss: 15 dB (typical)
 IF return loss: 15 dB (typical)
 Exposed pad, 4 mm $\times$ 4 mm, 24-terminal, ceramic, LCC package

APPLICATIONS

Test and measurement instrumentation
 Military, aerospace, and defense applications
 Microwave point to point base stations

GENERAL DESCRIPTION

The HMC8191 is a passive, wideband, I/Q monolithic microwave integrated circuit (MMIC) mixer that can be used either as an image reject mixer for receiver operations or as a single-sideband upconverter for transmitter operations. With a radio frequency (RF) and local oscillator (LO) range of 6 GHz to 26.5 GHz, and an intermediate frequency (IF) bandwidth of dc to 5 GHz, the HMC8191 is ideal for applications requiring a wide frequency range, excellent RF performance, and a simple design with fewer components and a small printed circuit board (PCB) footprint. A single HMC8191 can replace multiple narrow-band mixers in a design.

The inherent I/Q architecture of the HMC8191 offers excellent image rejection and thereby eliminates the need for expensive filtering for unwanted sidebands. The mixer also provides

FUNCTIONAL BLOCK DIAGRAM

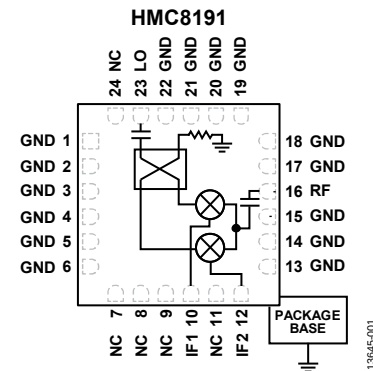


Figure 1.

excellent LO to RF and LO to IF isolation and reduces the effect of LO leakage to ensure signal integrity.

Being a passive mixer, the HMC8191 does not require any dc power sources. It offers a lower noise figure compared to an active mixer, ensuring superior dynamic range for high performance and precision applications.

The HMC8191 is fabricated on a gallium arsenide (GaAs) metal semiconductor field effect transistor (MESFET) process and uses Analog Devices, Inc. mixer cells and a 90-degree hybrid. The HMC8191 is available in a compact, 4 mm $\times$ 4 mm, 24-terminal leadless chip carrier (LCC) package and operates over a -40°C to $+85^\circ\text{C}$ temperature range. An evaluation board for the HMC8191 is also available from the Analog Devices website.

HMC8191* PRODUCT PAGE QUICK LINKS

Last Content Update: 07/01/2017

COMPARABLE PARTS

View a parametric search of comparable parts.

EVALUATION KITS

- HMC8191 Evaluation Board

DOCUMENTATION

Data Sheet

- HMC8191: 6 GHz to 26 GHz Wideband I/Q Mixer

REFERENCE MATERIALS

Press

- Analog Devices' Wideband RF Mixers Simplify Design and Significantly Reduce Time to Market for Industrial Applications

Technical Articles

- The Changing Landscape of Frequency Mixing Components

DESIGN RESOURCES

- HMC8191 Material Declaration
- PCN-PDN Information
- Quality And Reliability
- Symbols and Footprints

DISCUSSIONS

View all HMC8191 EngineerZone Discussions.

SAMPLE AND BUY

Visit the product page to see pricing options.

TECHNICAL SUPPORT

Submit a technical question or find your regional support number.

DOCUMENT FEEDBACK

Submit feedback for this data sheet.

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REVISION HISTORY

6/2017—Revision 0: Initial Version

SPECIFICATIONS

$T_A = 25^\circ\text{C}$, IF = 100 MHz, LO drive = 18 dBm, all measurements performed as downconverter with lower sideband selected, external 90° hybrid at the IF ports, and LO amplifier in line with lab bench LO source, unless otherwise noted.

Table 1.

Parameter	Symbol	Min	Typ	Max	Unit
RADIO FREQUENCY	RF	6		26.5	GHz
LOCAL OSCILLATOR FREQUENCY	f_{LO}	6		26.5	GHz
INTERMEDIATE FREQUENCY	IF	DC		5	GHz
LOCAL OSCILLATOR DRIVE LEVEL			18		dBm
RF PERFORMANCE AS DOWNCONVERTER					
Conversion Loss			9	11.5	dB
Image Rejection		20	25		dBc
Single Sideband Noise Figure	SSB NF		14		dB
Input Third-Order Intercept	IP3		24		dBm
Input 1 dB Compression Point	P1dB		15		dBm
Input Second-Order Intercept	IP2		55		dBm
Isolation					
RF to IF			20		dB
LO to RF		30	40		dB
LO to IF		27	40		dB
Amplitude Balance ¹			± 0.5		dB
Phase Balance ¹			± 5		Degrees
RF PERFORMANCE AS UPCONVERTER					
Conversion Loss			9		dB
Sideband Rejection			25		dBc
Input Third-Order Intercept	IP3		22		dBm
Input 1 dB Compression Point	P1dB		13		dBm
RETURN LOSS PERFORMANCE ¹					
RF			15		dB
LO			15		dB
IFx			15		dB

¹ Measurements taken without 90° hybrid at the IF ports.

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
RF Input Power	24 dBm
LO Input Power	24 dBm
IF Input Power	24 dBm
IF Source/Sink Current	3 mA
Continuous Power Dissipation, P_{DISS} ($T_A = 85^\circ\text{C}$, Derate 7.29 mW/ $^\circ\text{C}$ Above 85°C)	657 mW
Maximum Junction Temperature	175°C
Maximum Peak Reflow Temperature (MLS3)	260°C
Operating Temperature Range	-40°C to $+85^\circ\text{C}$
Storage Temperature Range	-65°C to $+150^\circ\text{C}$
Electrostatic Discharge Sensitivity	
Human Body Model	750 V
Field Induced Charged Device Model	1200 V

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to PCB design and operating environment. Careful attention to PCB thermal design is required.

Table 3. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
E-24-1 ¹	38.3	137	$^\circ\text{C}/\text{W}$

¹ Refer to JEDEC standard JESD51-2 for additional information on optimizing the thermal impedance (PCB with 4×4 vias).

ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

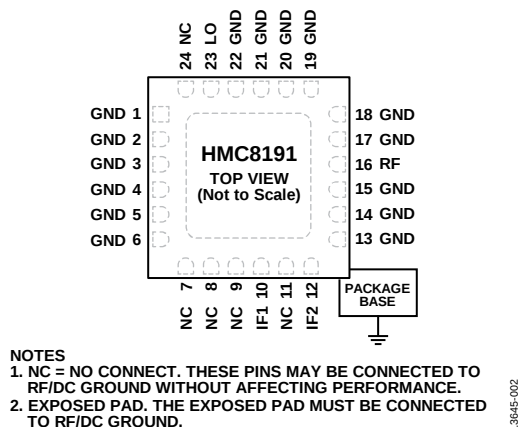
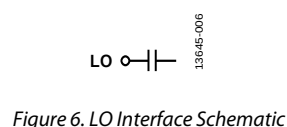
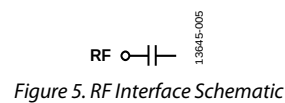
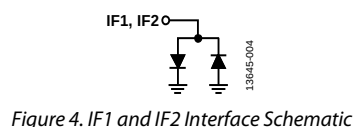
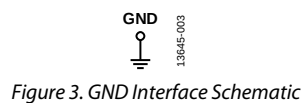


Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1 to 6, 13 to 15, 17 to 22	GND	Ground. These pins and the package bottom must be connected to RF/dc ground. See Figure 3 for the interface schematic.
7 to 9, 11, 24	NC	No Connect. These pins can be connected to RF/dc ground without affecting performance.
10, 12	IF1, IF2	First and Second Quadrature Intermediate Frequency Input/Output Pins. These pins are dc-coupled. For applications not requiring operation to dc, use an off-chip dc blocking capacitor. For operations to dc, these pins must not source/sink more than 3 mA of current; otherwise, the device may not function and may fail. See Figure 4 for the interface schematic.
16	RF	Radio Frequency Input/Output. This pin is ac-coupled and matched to 50 Ω . See Figure 5 for the interface schematic.
23	LO	Local Oscillator Input. This pin is ac-coupled and matched to 50 Ω . See Figure 6 for the interface schematic.
	EPAD	Exposed Pad. The exposed pad must be connected to RF/dc ground.

INTERFACE SCHEMATICS



TYPICAL PERFORMANCE CHARACTERISTICS

DOWNCONVERTER PERFORMANCE: IF = 100 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

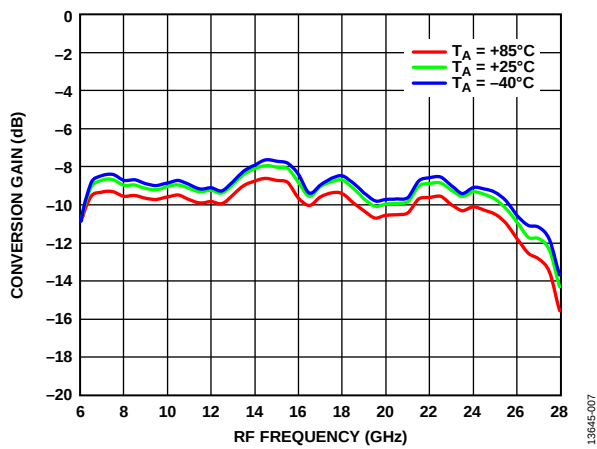


Figure 7. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

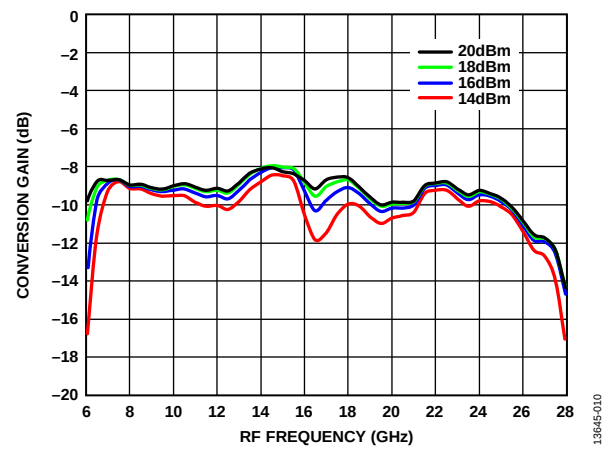


Figure 10. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

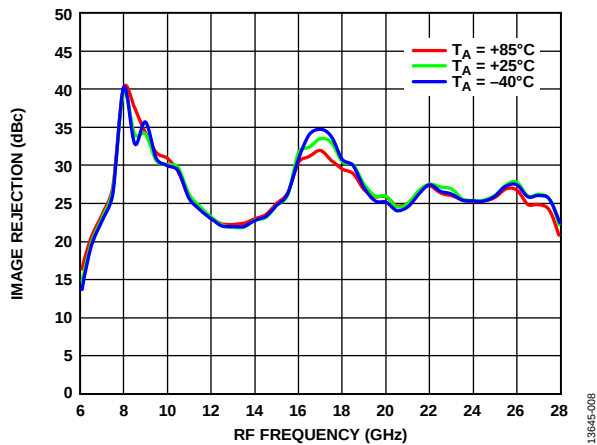


Figure 8. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

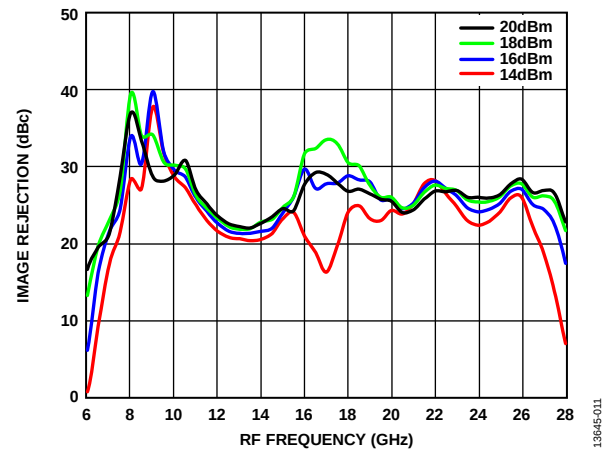


Figure 11. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

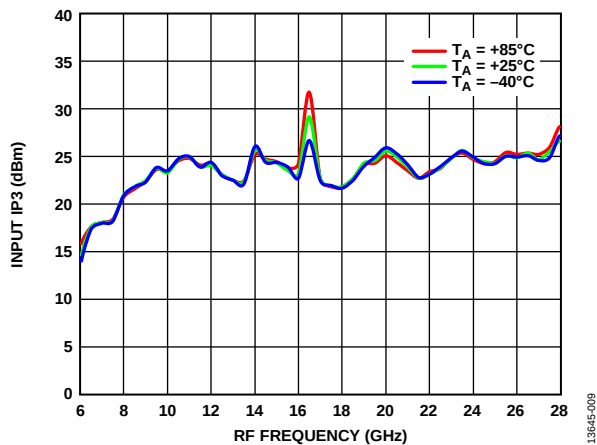


Figure 9. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

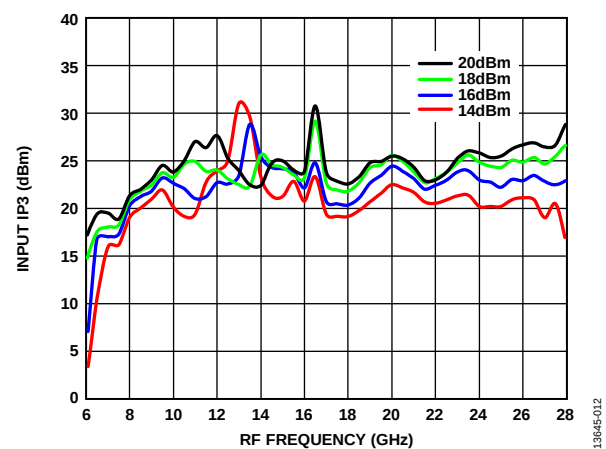


Figure 12. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

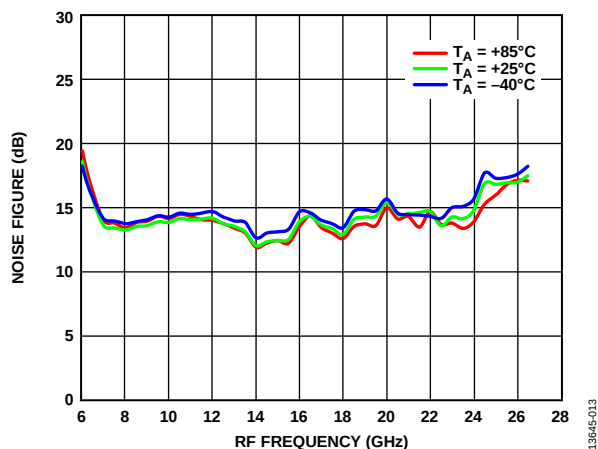


Figure 13. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

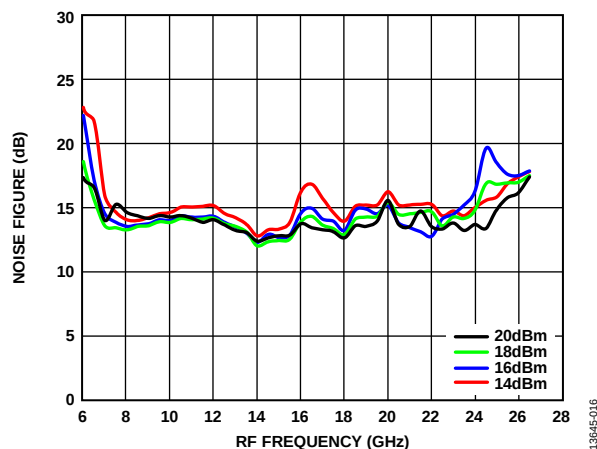


Figure 16. Noise Figure vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

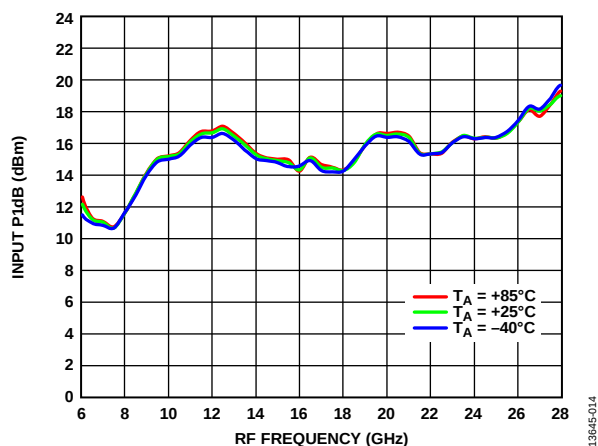


Figure 14. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

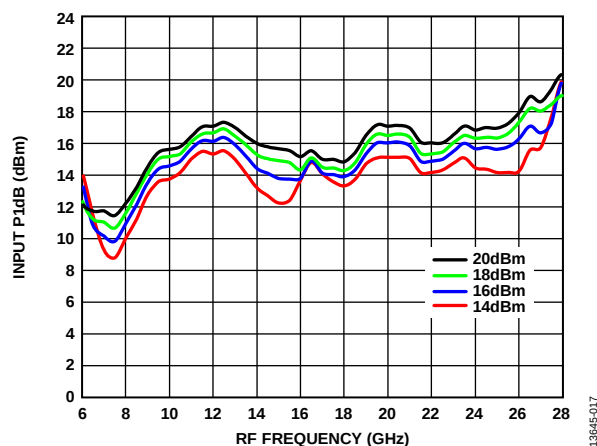


Figure 17. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

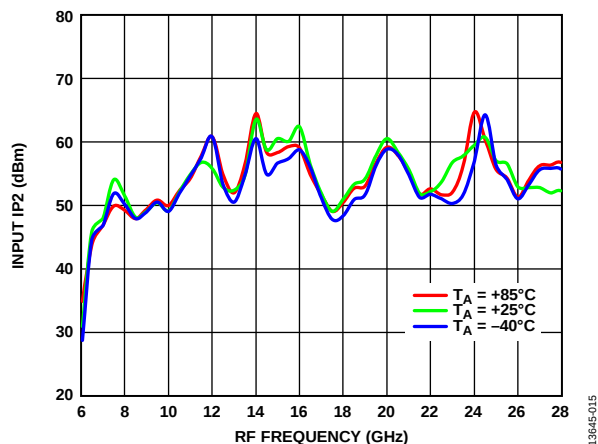


Figure 15. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

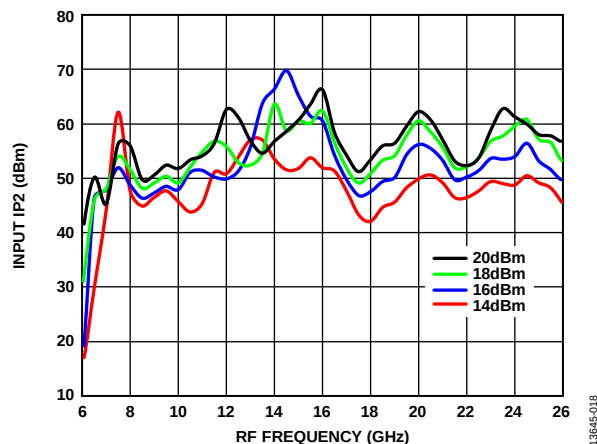


Figure 18. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

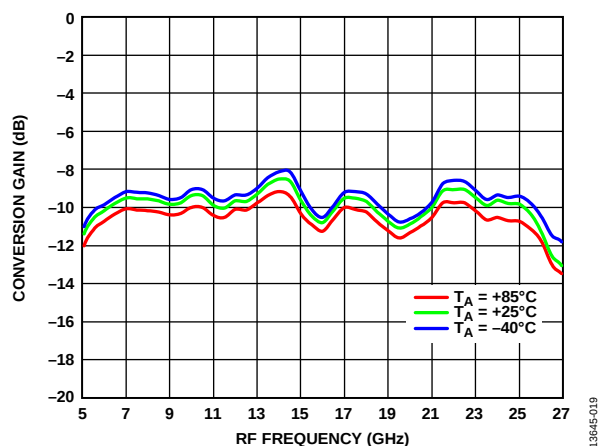
DOWNCONVERTER PERFORMANCE: IF = 2500 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

Figure 19. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

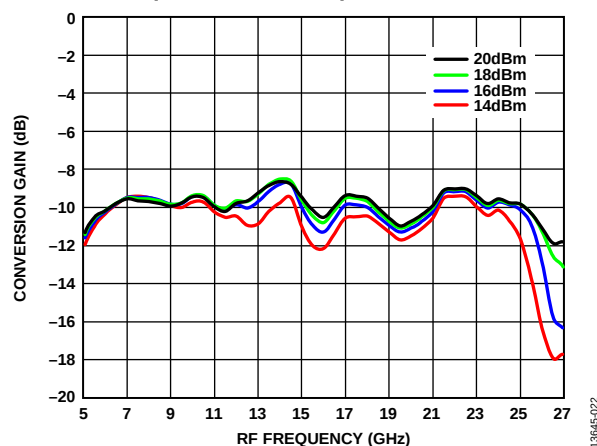


Figure 22. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

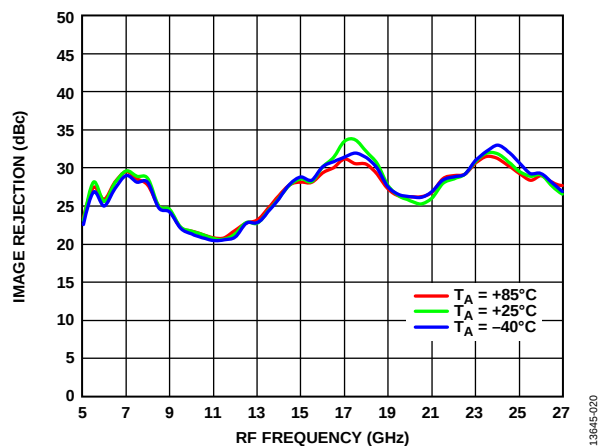


Figure 20. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

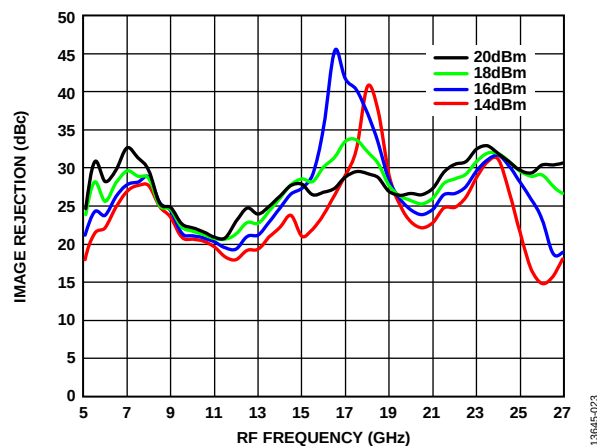


Figure 23. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

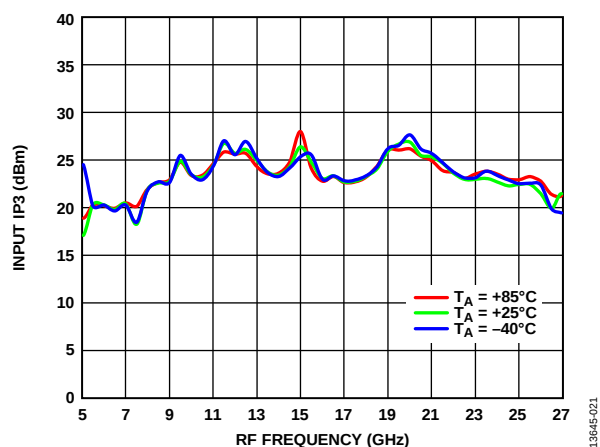


Figure 21. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

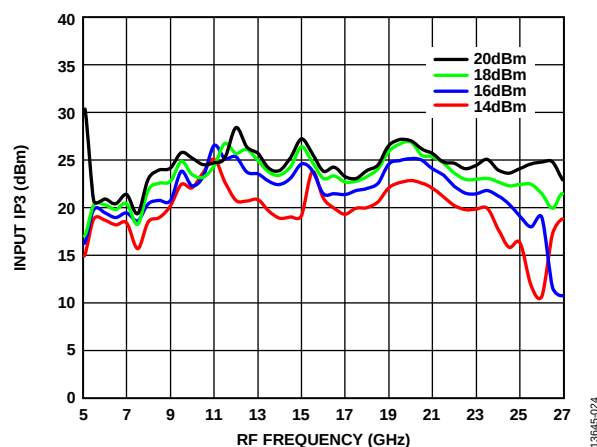


Figure 24. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

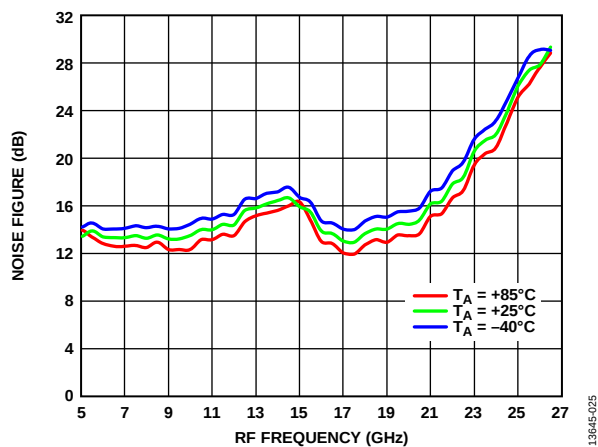


Figure 25. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

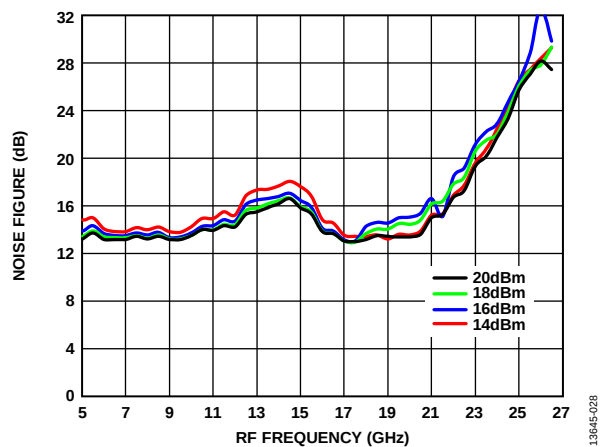


Figure 28. Noise Figure vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

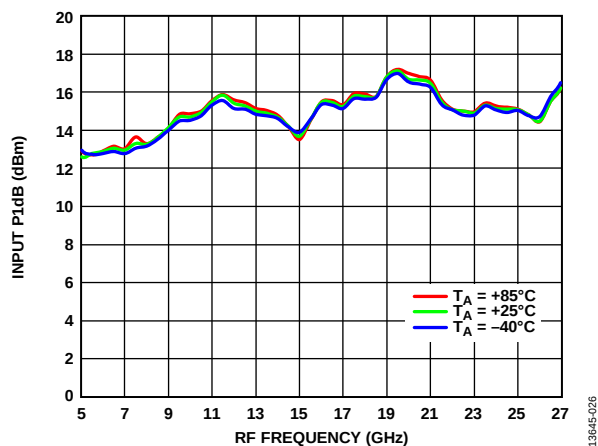


Figure 26. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

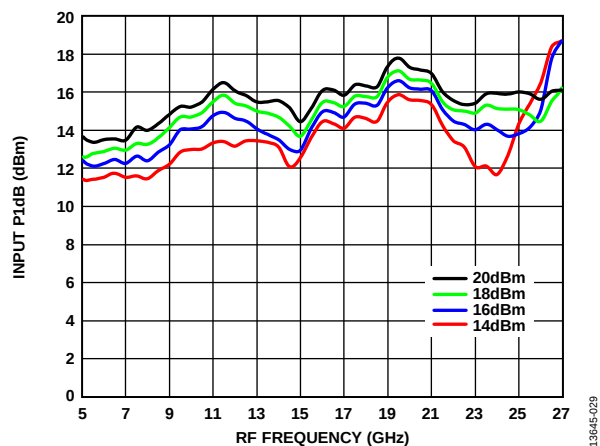


Figure 29. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

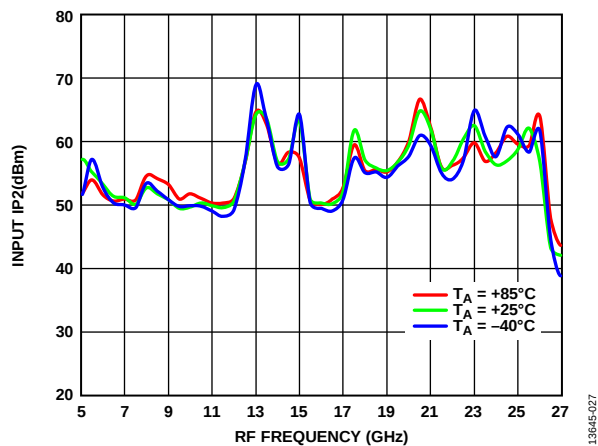


Figure 27. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

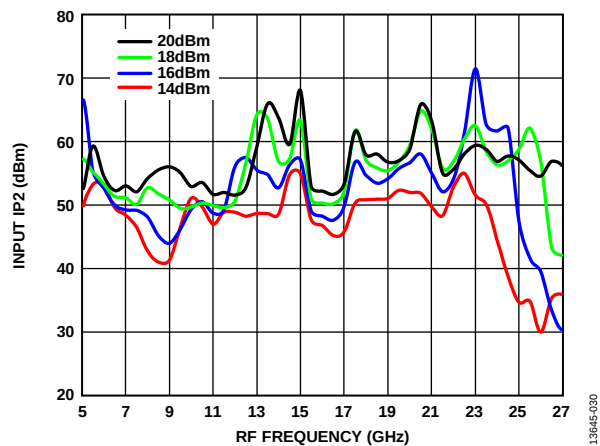


Figure 30. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

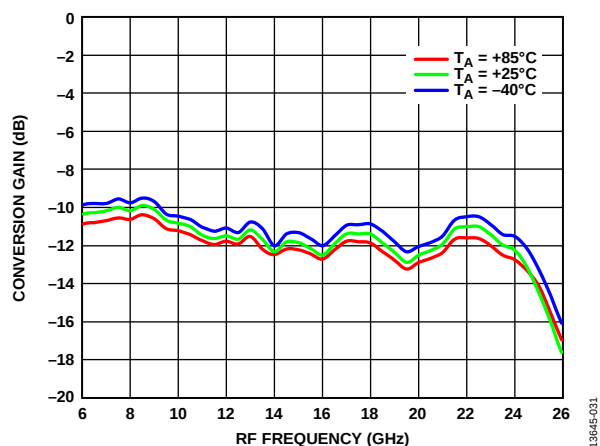
DOWNCONVERTER PERFORMANCE: IF = 5000 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

Figure 31. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

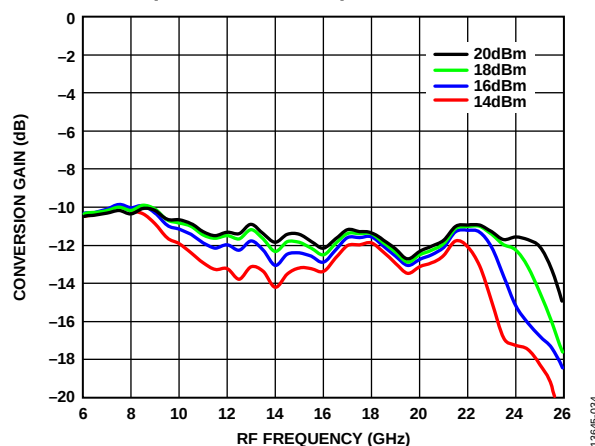


Figure 34. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

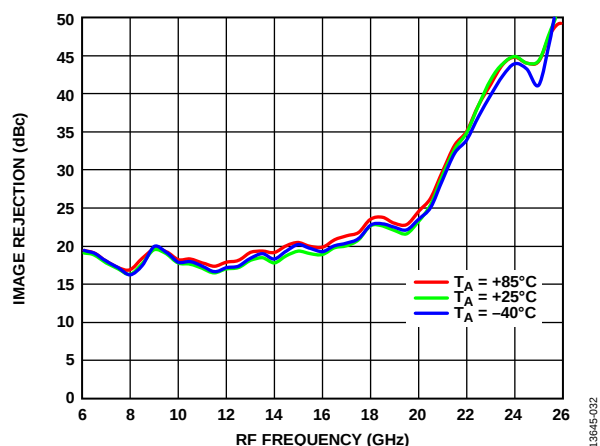


Figure 32. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

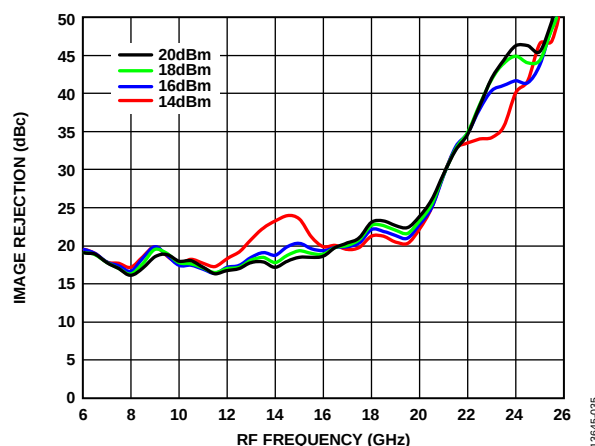


Figure 35. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

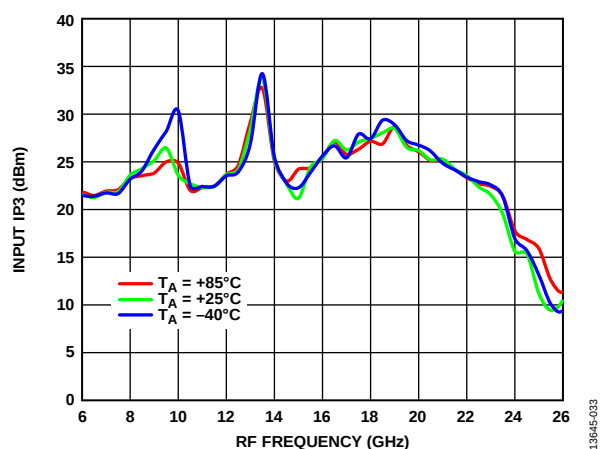


Figure 33. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

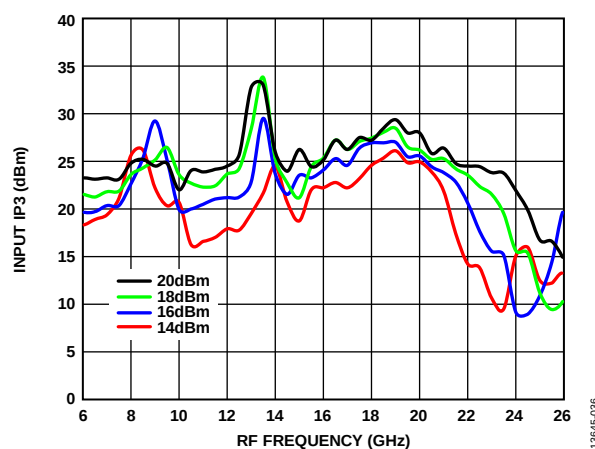


Figure 36. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

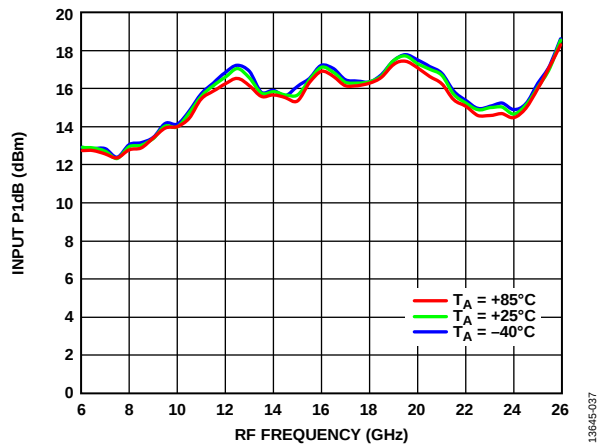


Figure 37. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

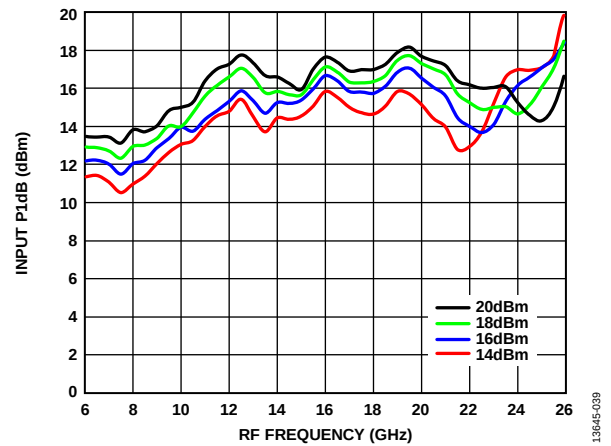


Figure 39. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

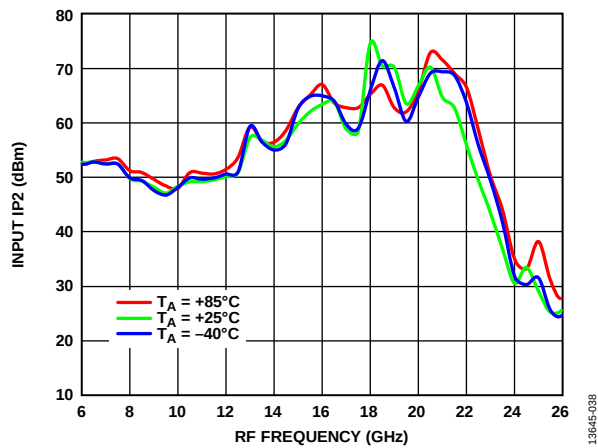


Figure 38. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

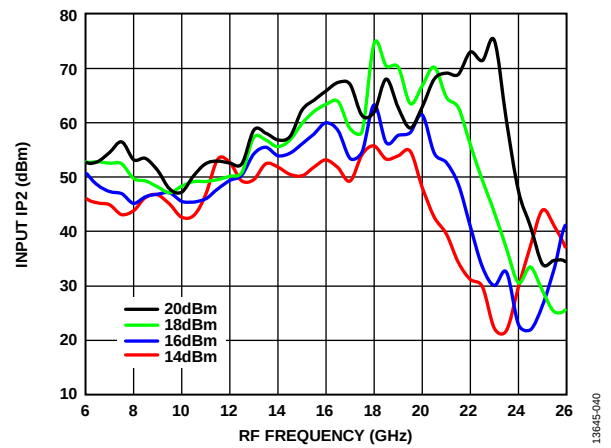


Figure 40. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

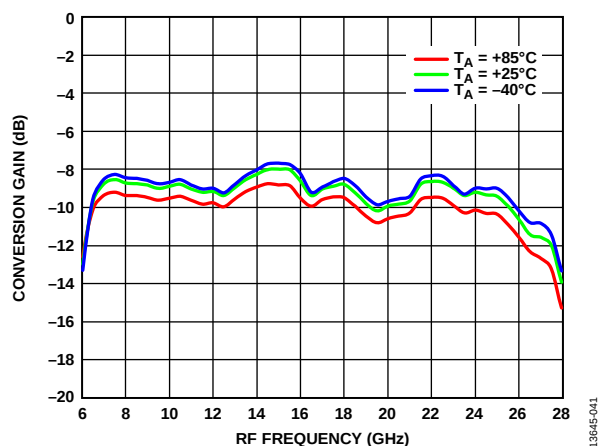
DOWNCONVERTER PERFORMANCE: IF = 100 MHz, UPPER SIDEBAND (LOW-SIDE LO)

Figure 41. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

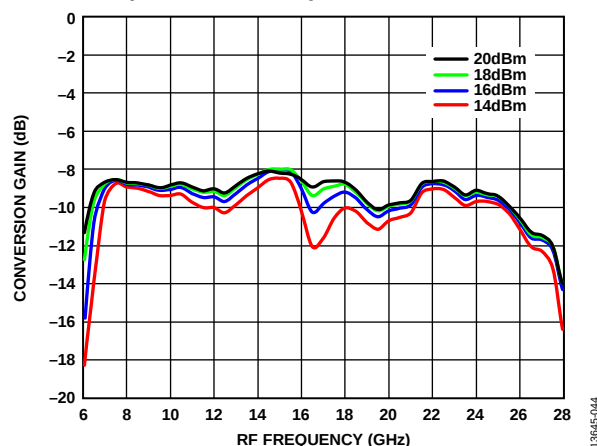
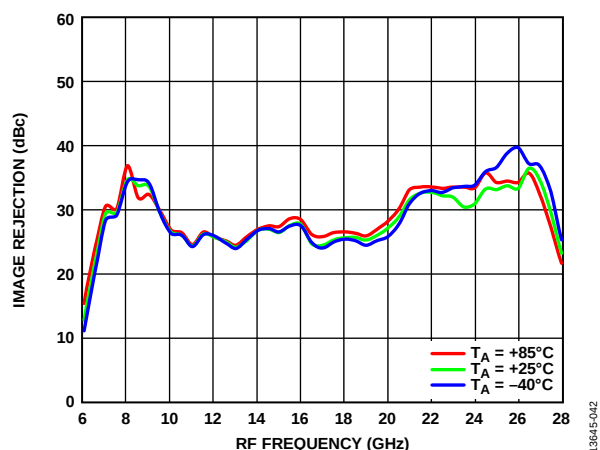
Figure 44. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 42. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

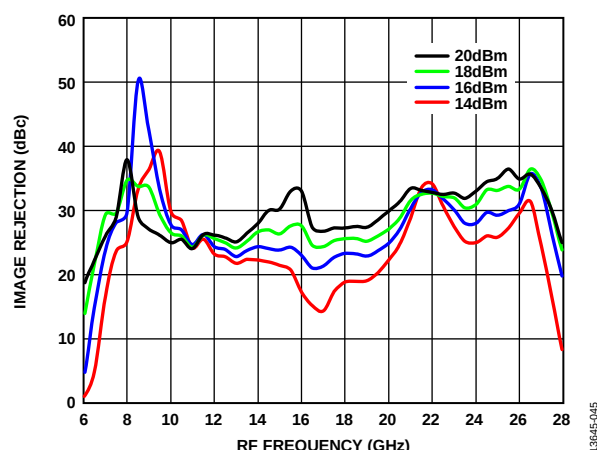
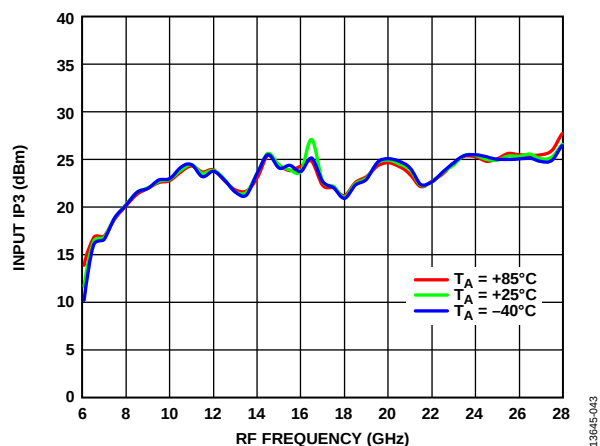
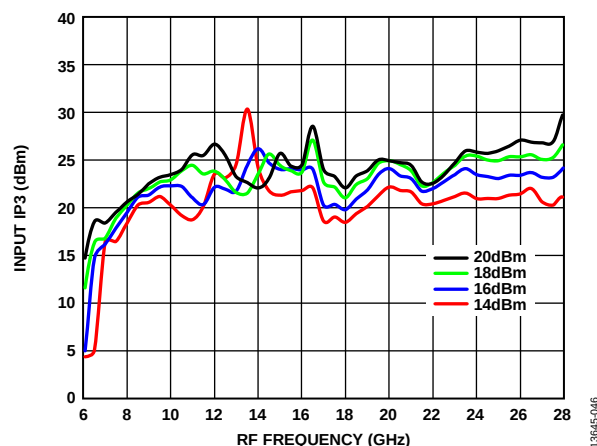
Figure 45. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 43. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 46. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

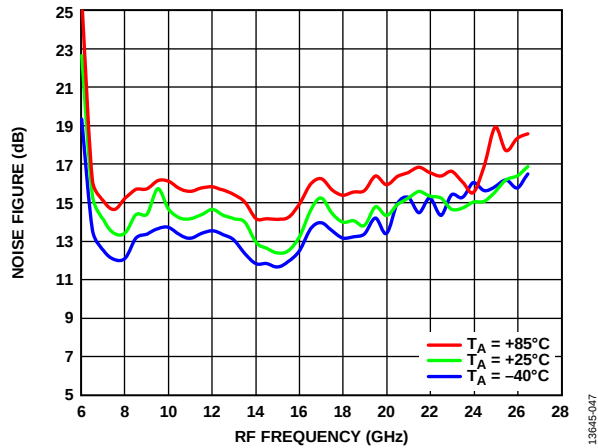


Figure 47. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

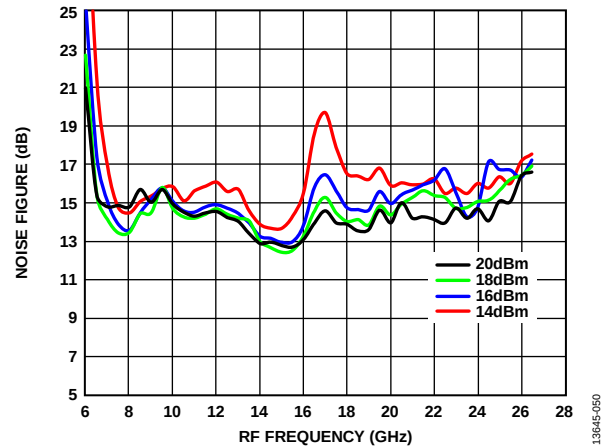


Figure 50. Noise Figure vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

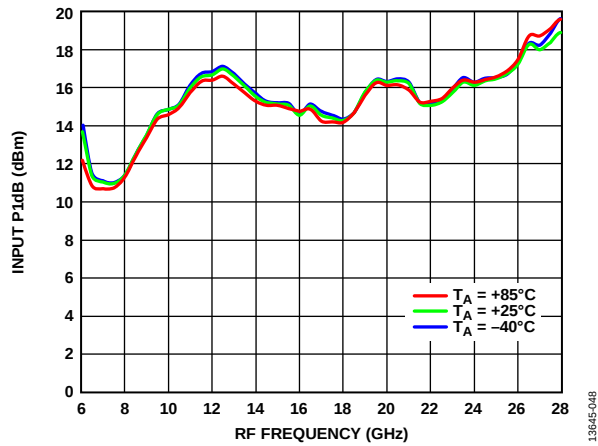


Figure 48. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

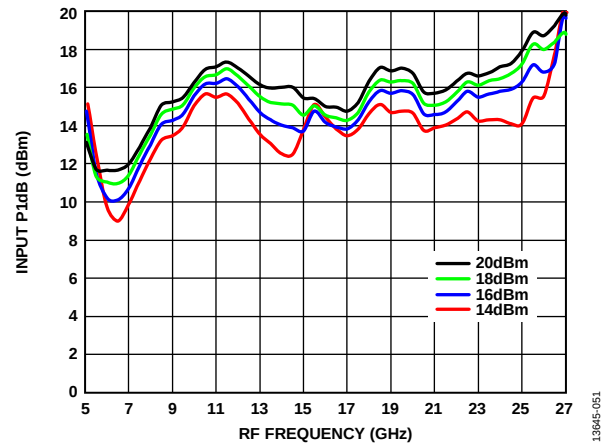


Figure 51. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

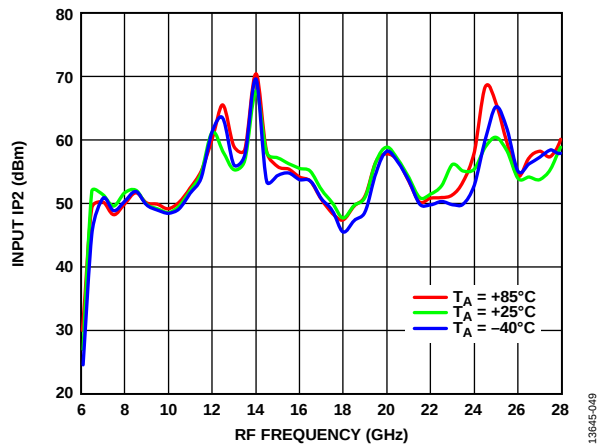


Figure 49. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

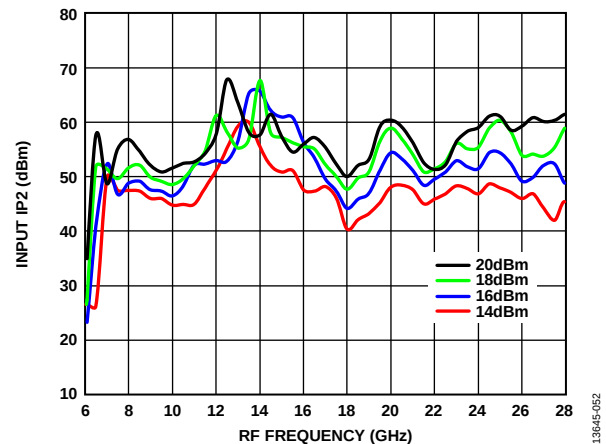


Figure 52. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

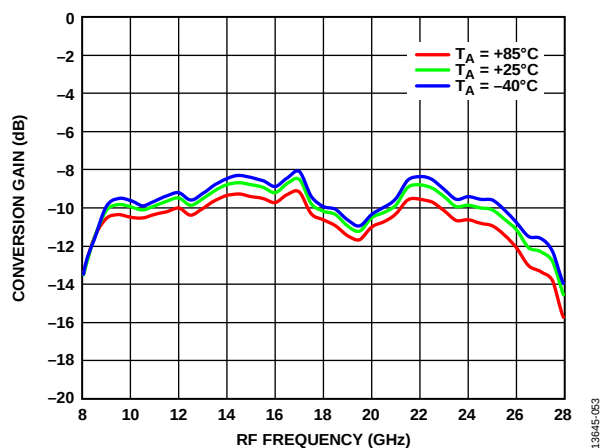
DOWNCONVERTER PERFORMANCE: IF = 2500 MHz, UPPER SIDEBAND (LOW-SIDE LO)

Figure 53. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

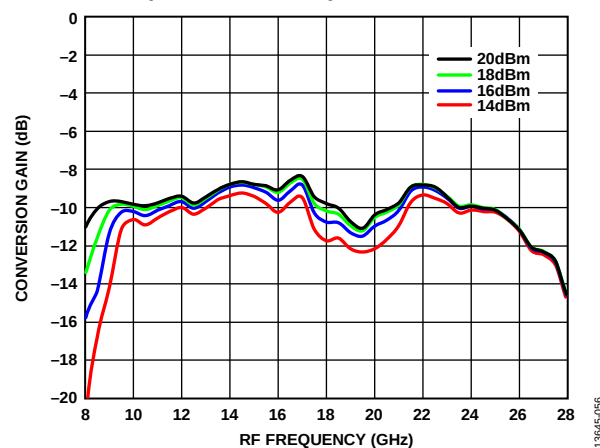
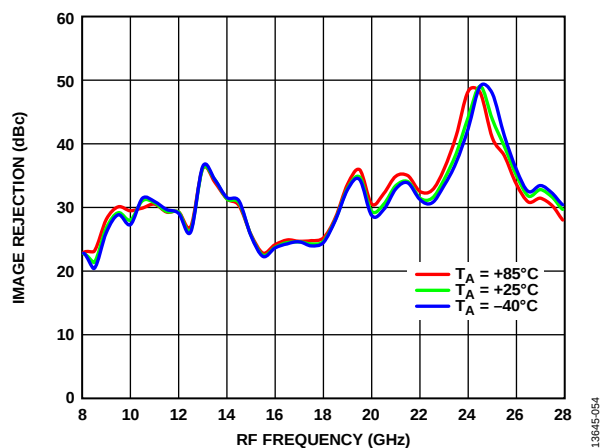
Figure 56. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 54. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

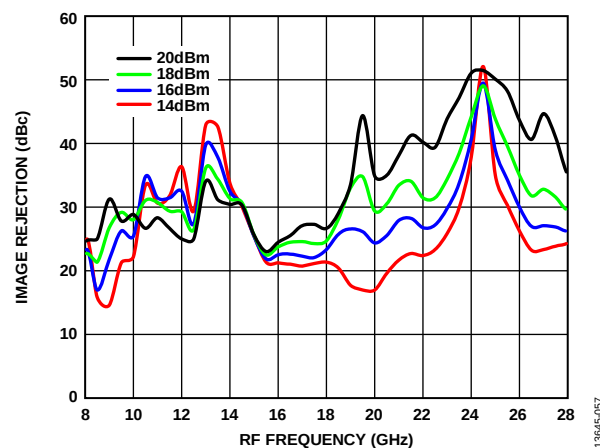
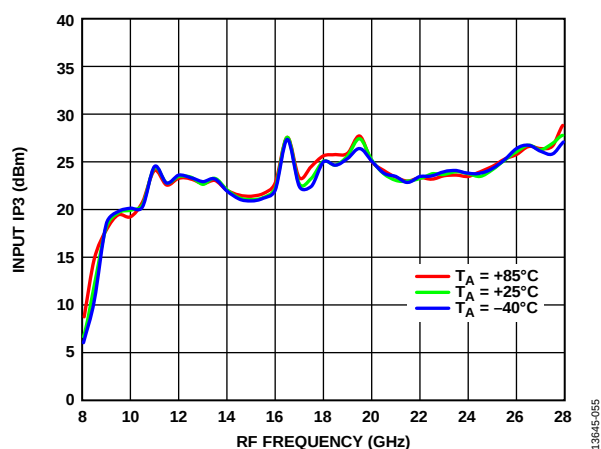
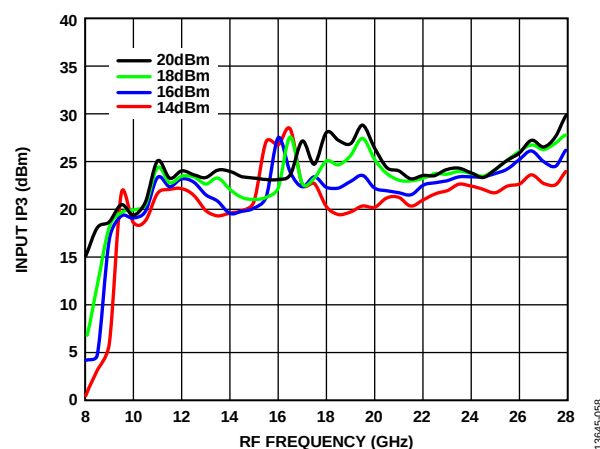
Figure 57. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 55. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 58. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

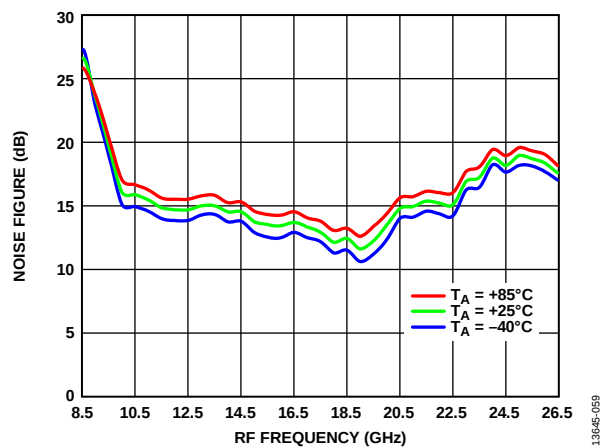


Figure 59. Noise Figure vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

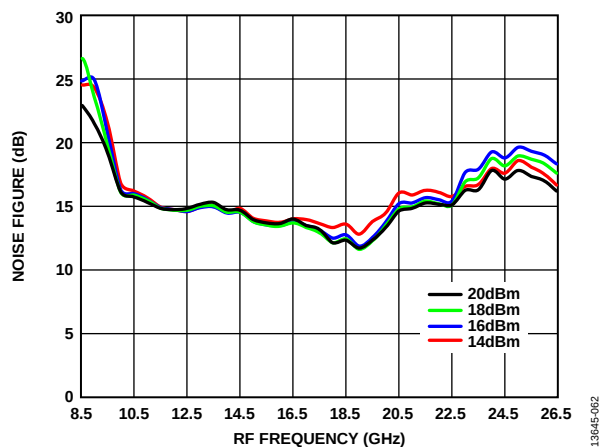


Figure 62. Noise Figure vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

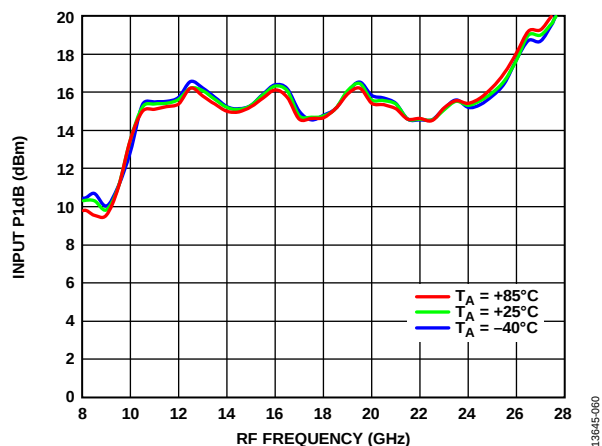


Figure 60. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

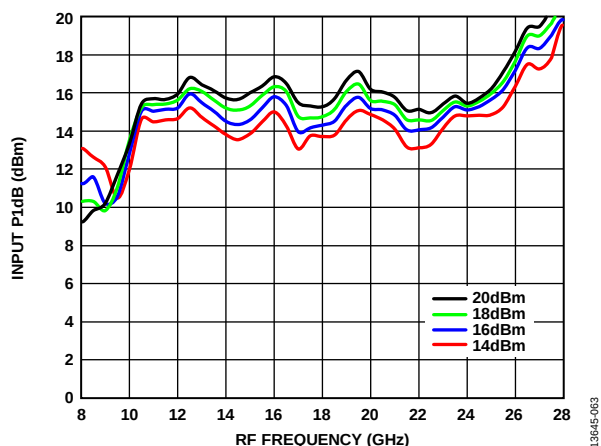


Figure 63. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

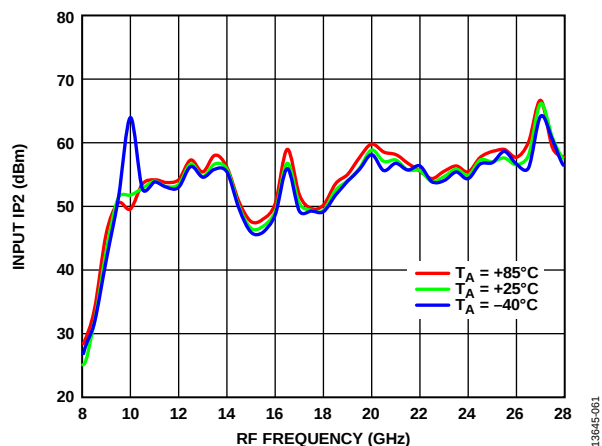


Figure 61. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

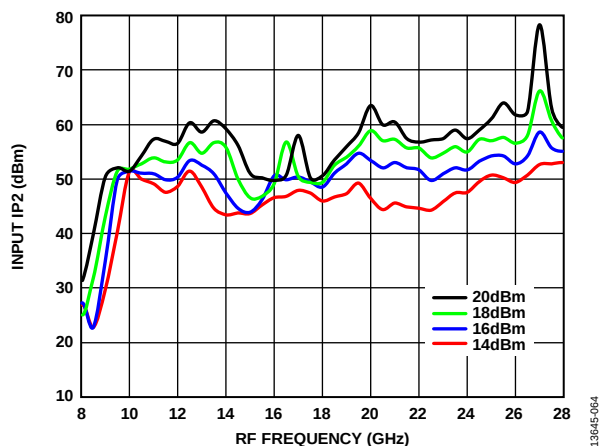


Figure 64. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

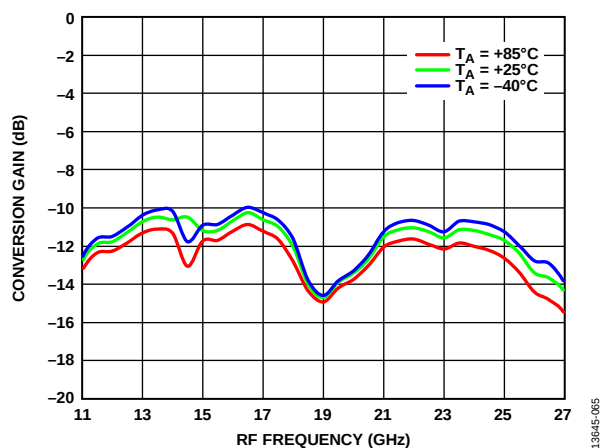
DOWNCONVERTER PERFORMANCE: IF = 5000 MHz, UPPER SIDEBAND (LOW-SIDE LO)

Figure 65. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

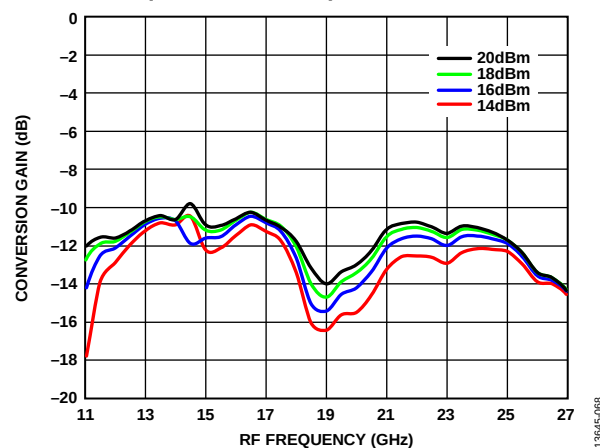
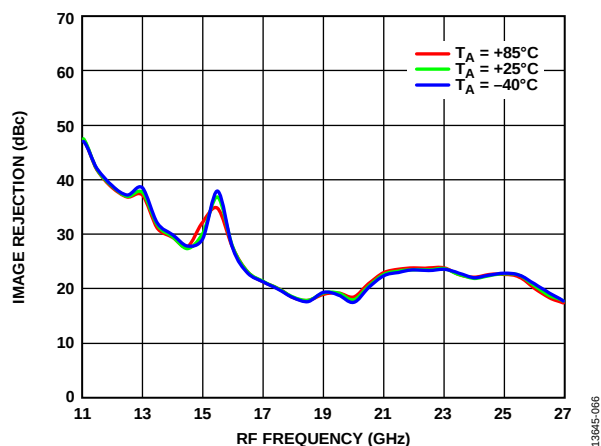
Figure 68. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 66. Image Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

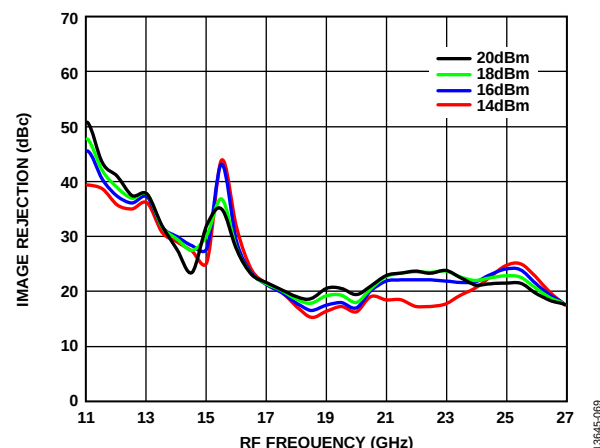
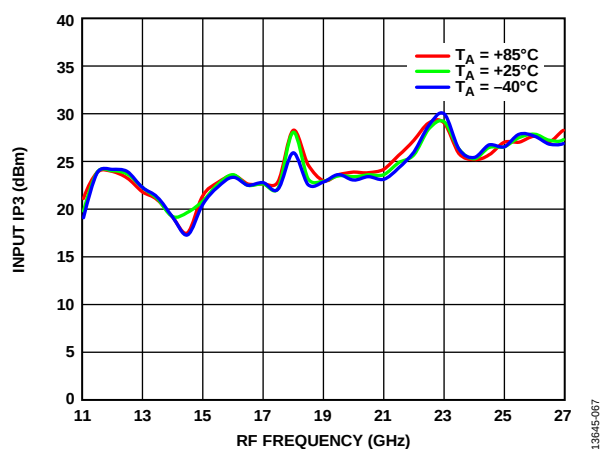
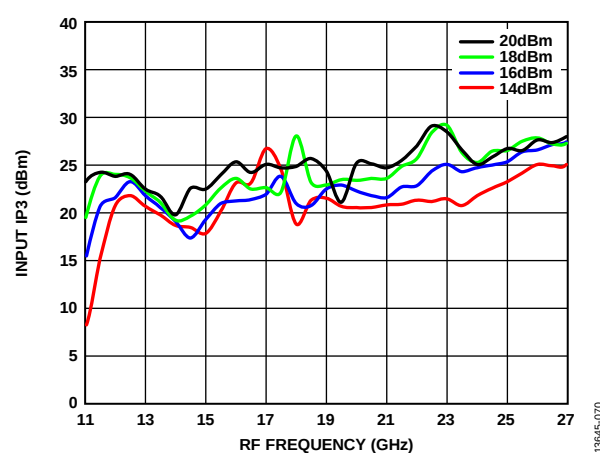
Figure 69. Image Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 67. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 70. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

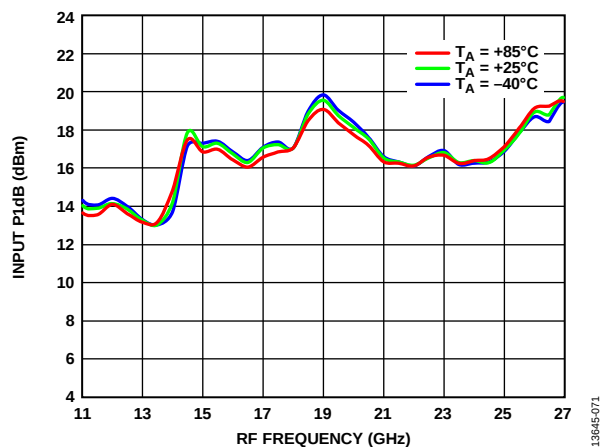


Figure 71. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

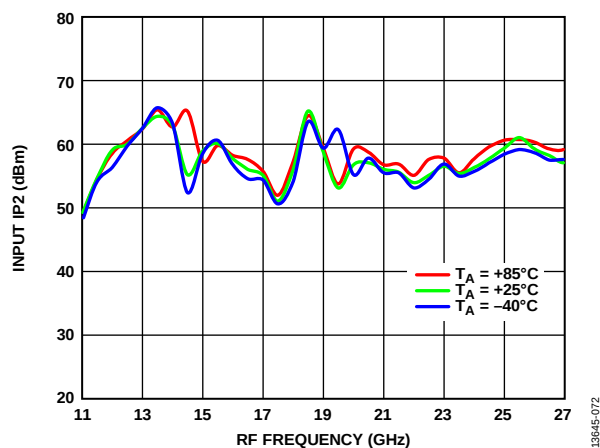


Figure 72. Input IP2 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

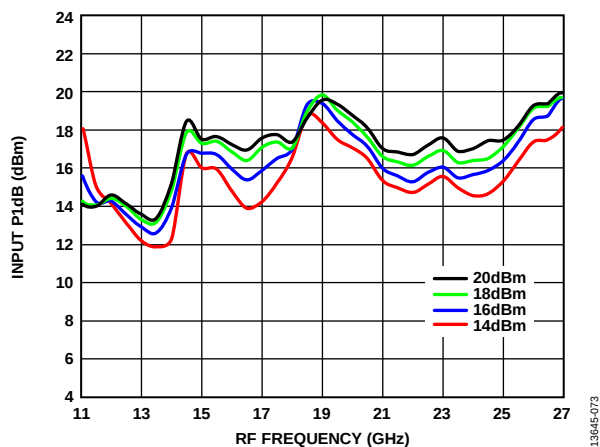


Figure 73. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

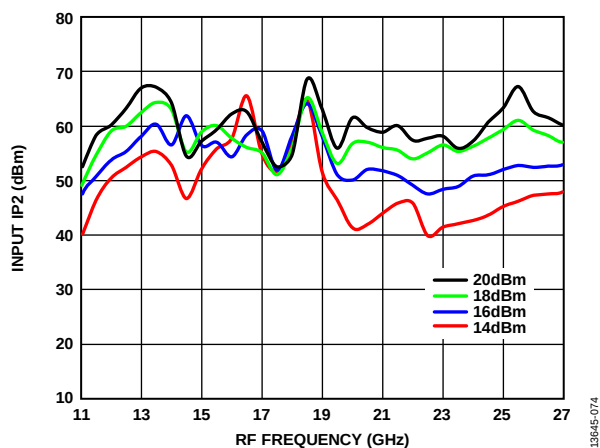


Figure 74. Input IP2 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 100 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

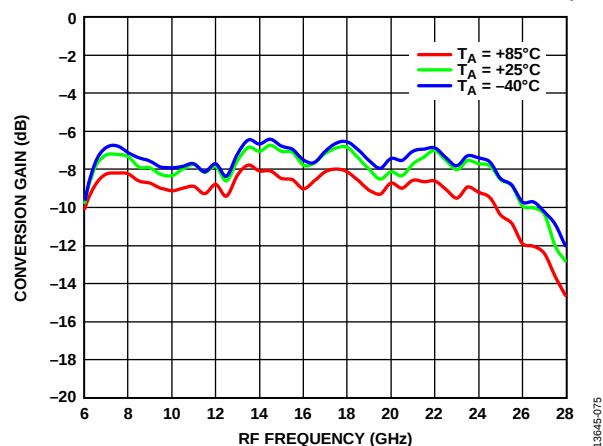


Figure 75. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

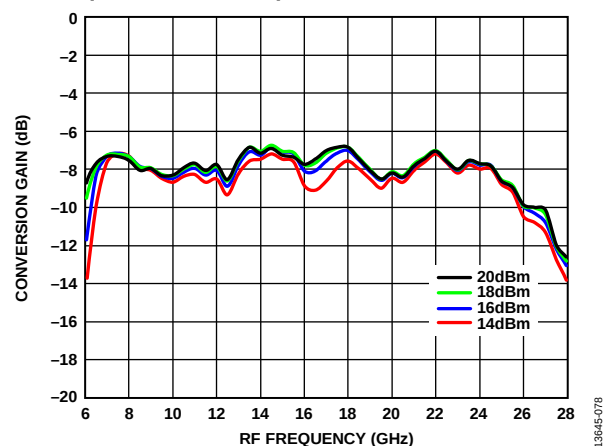
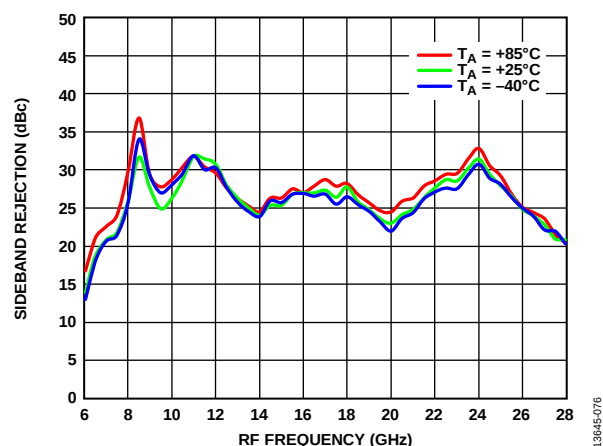
Figure 78. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 76. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

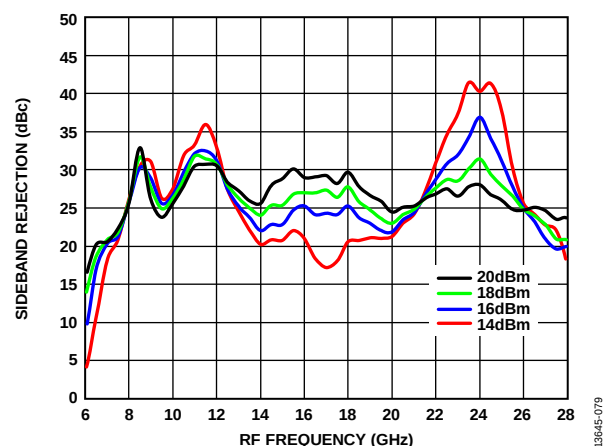
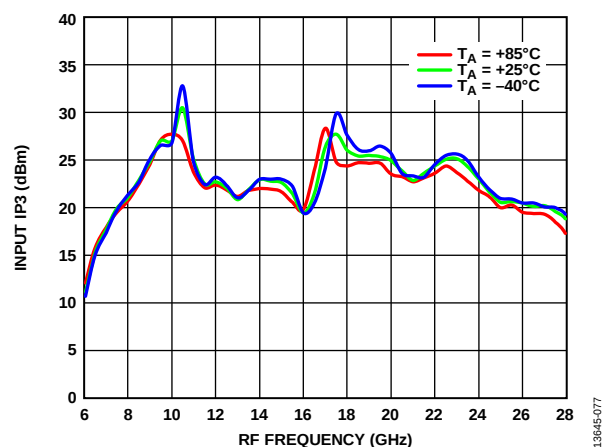
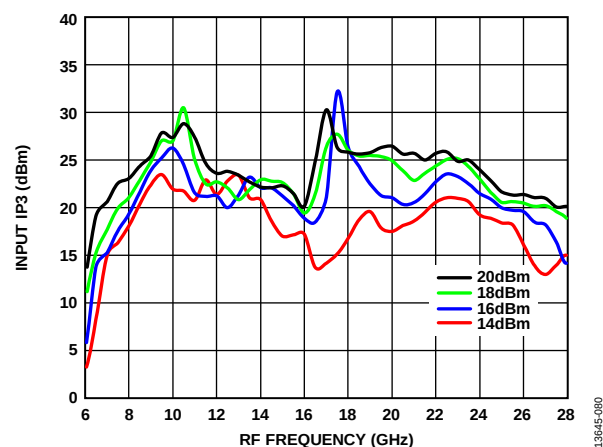
Figure 79. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 77. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 80. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

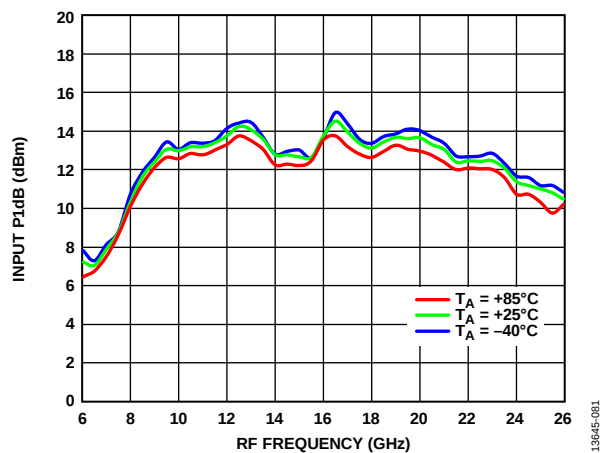


Figure 81. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

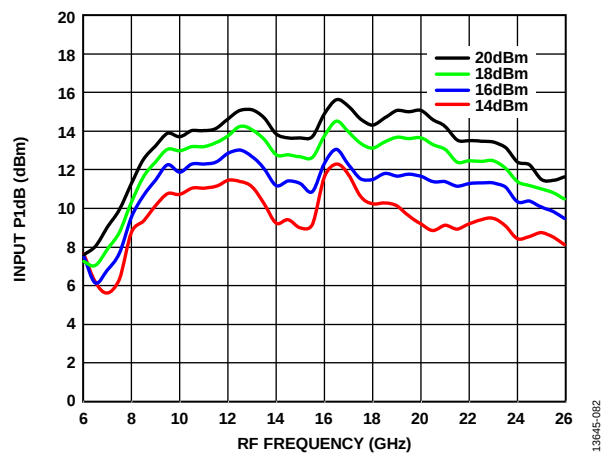


Figure 82. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 2500 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

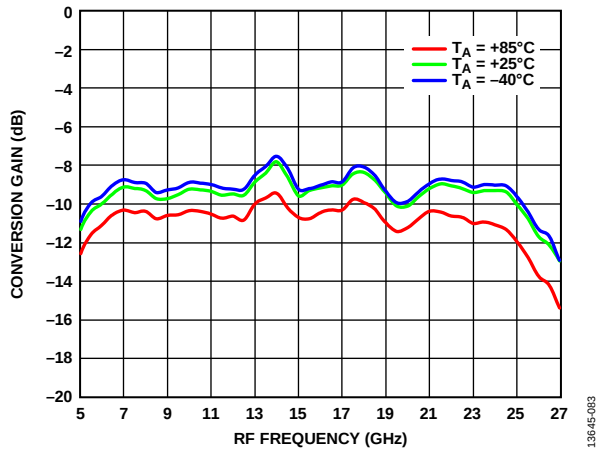


Figure 83. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

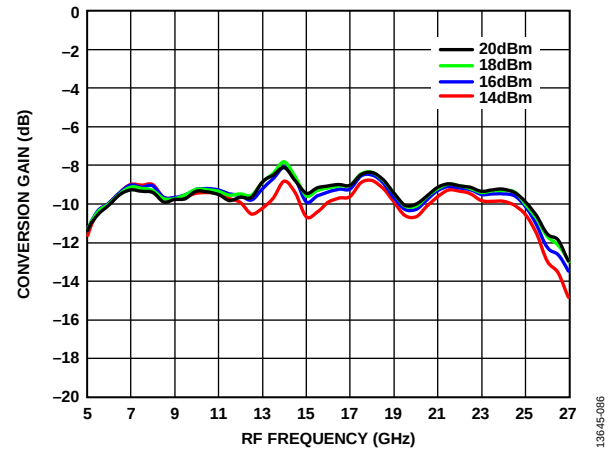
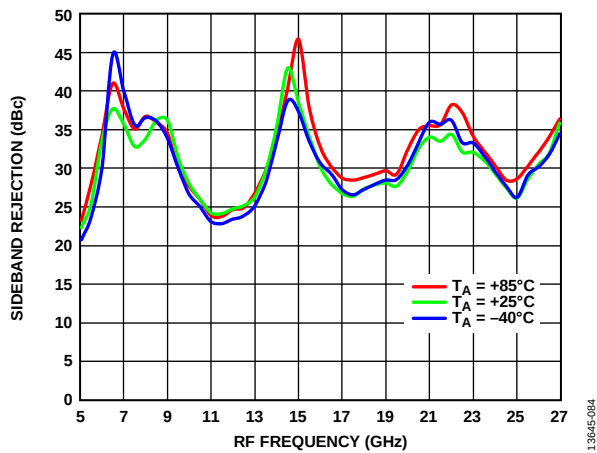
Figure 86. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 84. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

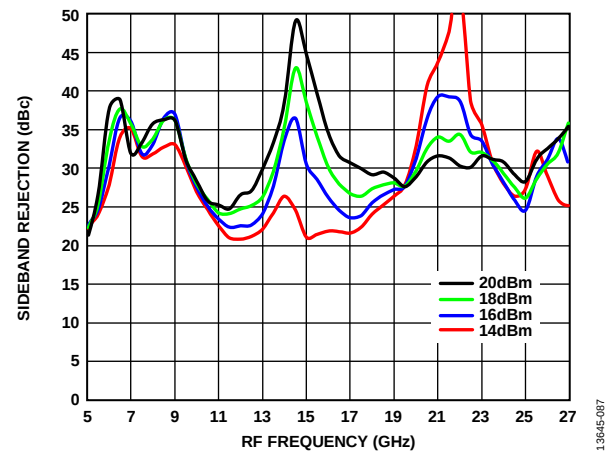
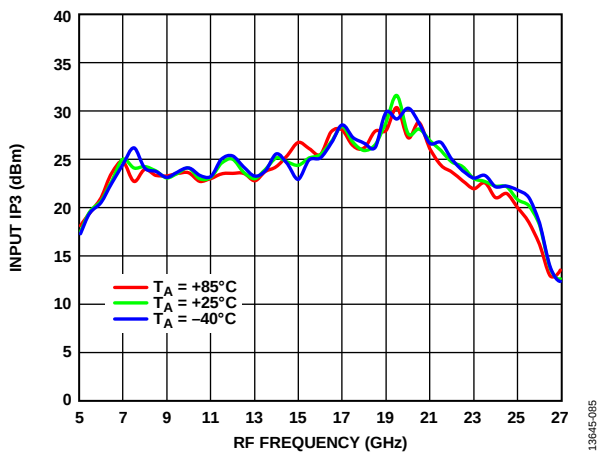
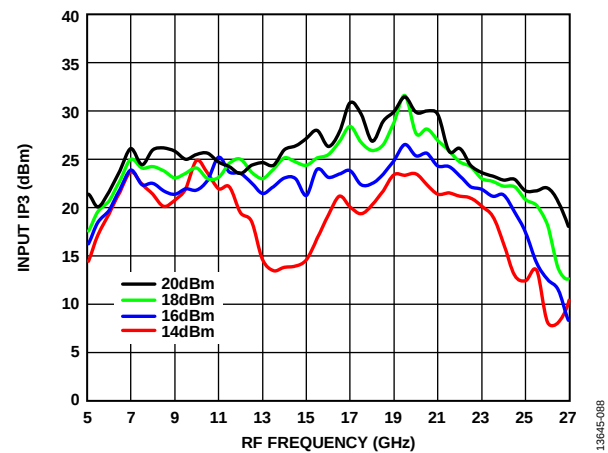
Figure 87. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 85. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 88. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

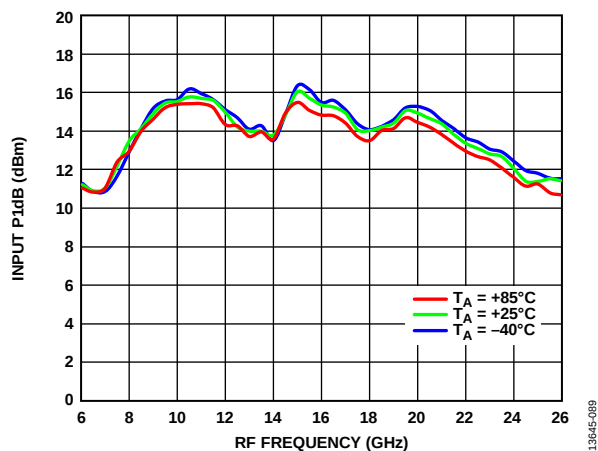


Figure 89. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

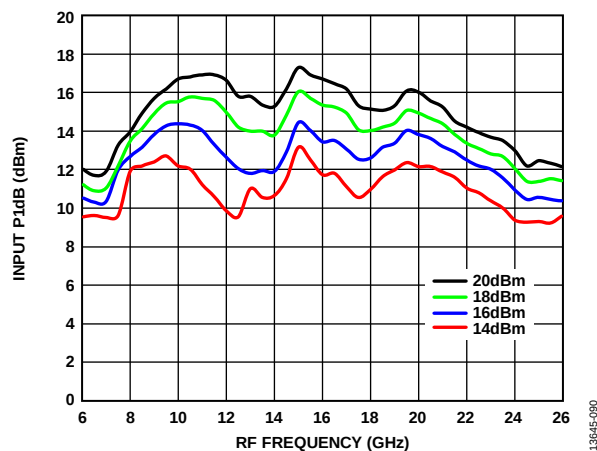


Figure 90. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 5000 MHz, LOWER SIDEBAND (HIGH-SIDE LO)

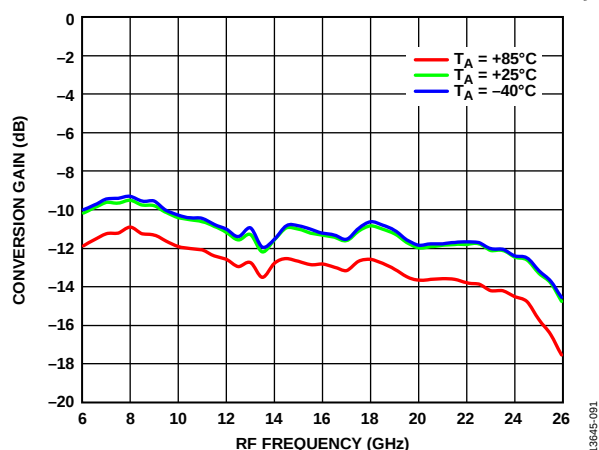


Figure 91. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

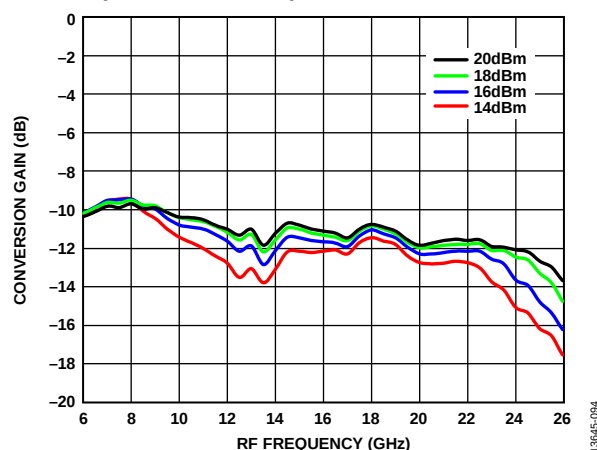
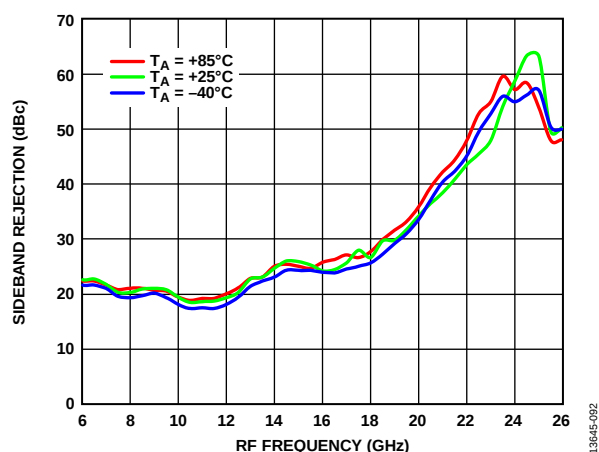
Figure 94. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 92. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

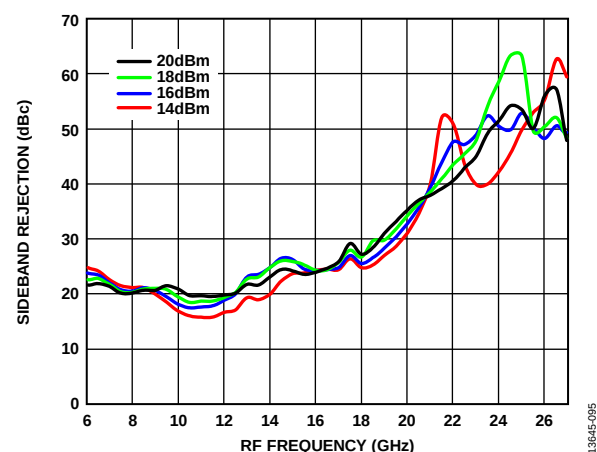
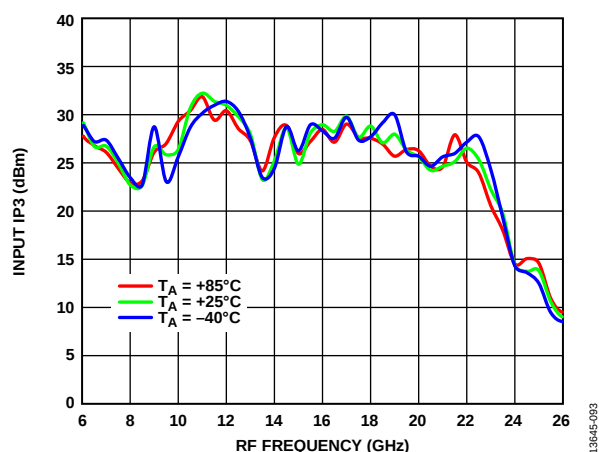
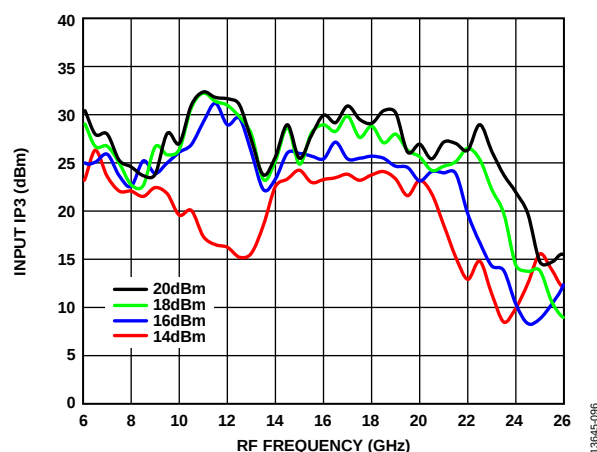
Figure 95. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 93. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 96. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

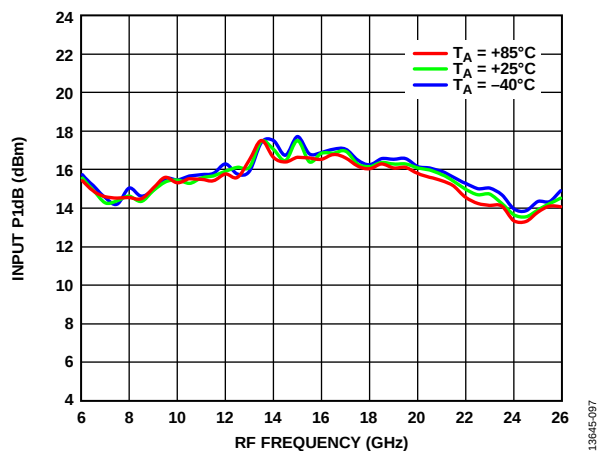


Figure 97. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

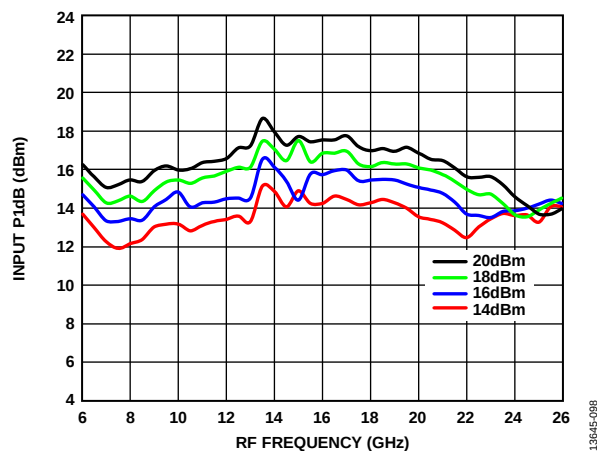


Figure 98. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 100 MHz, UPPER SIDEBAND (LOW-SIDE LO)

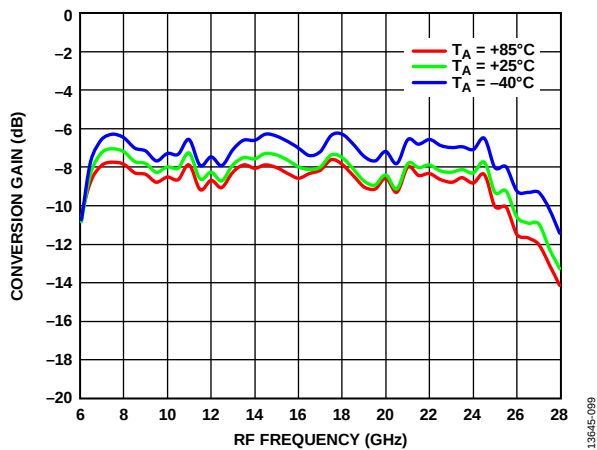


Figure 99. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

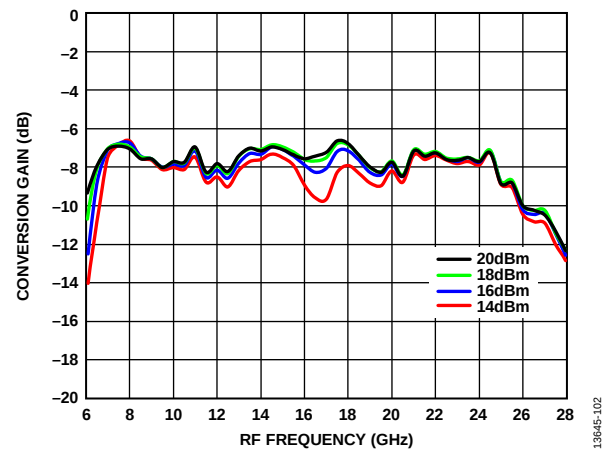
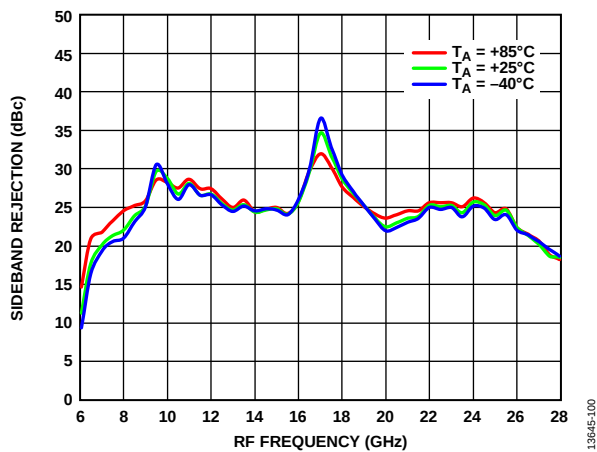
Figure 102. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 100. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

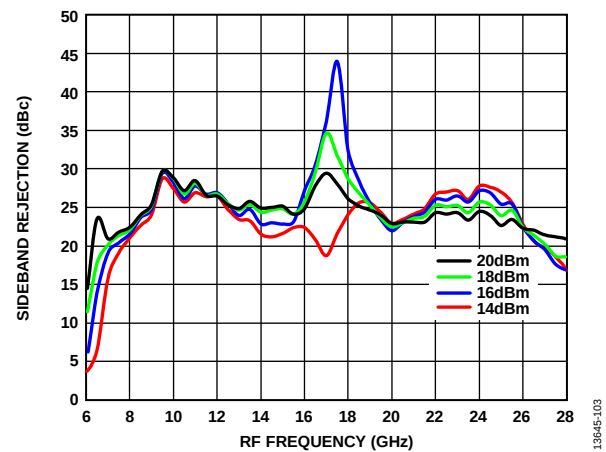
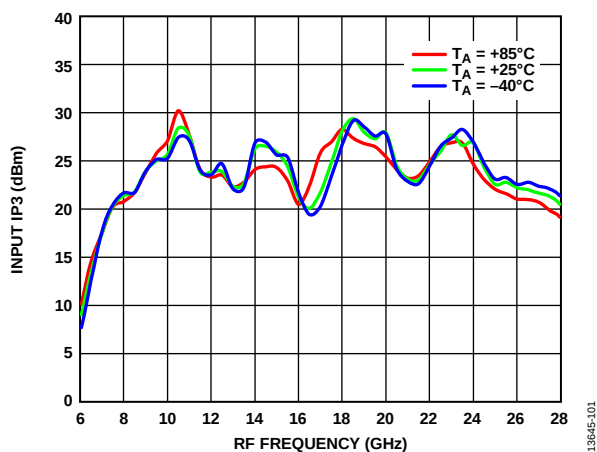
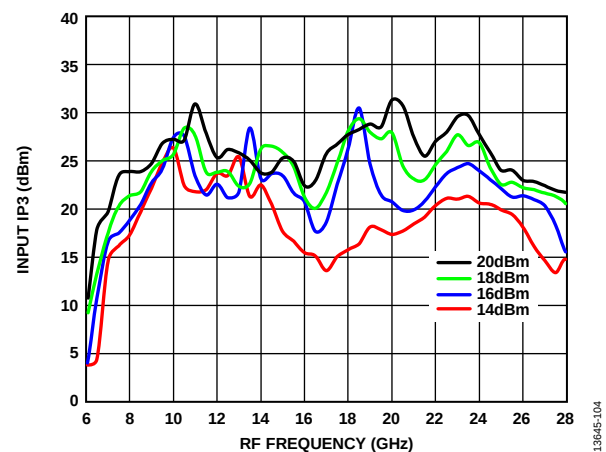
Figure 103. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 101. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 104. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

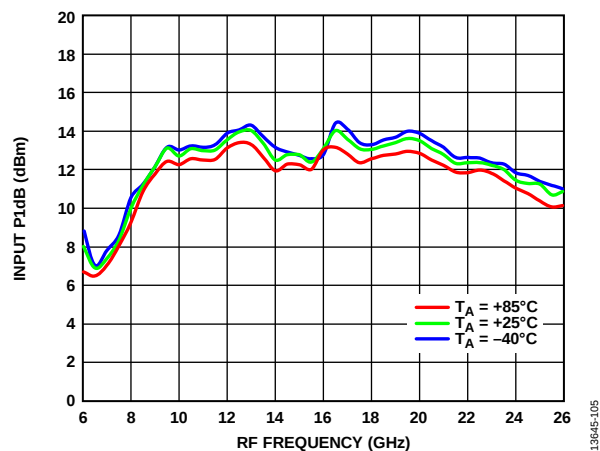


Figure 105. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

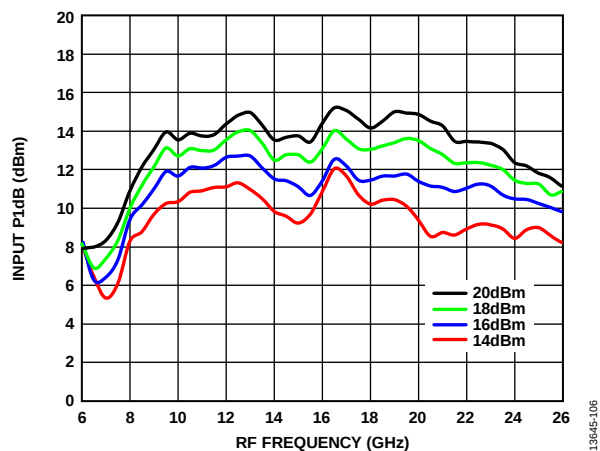


Figure 106. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 2500 MHz, UPPER SIDEBAND (LOW-SIDE LO)

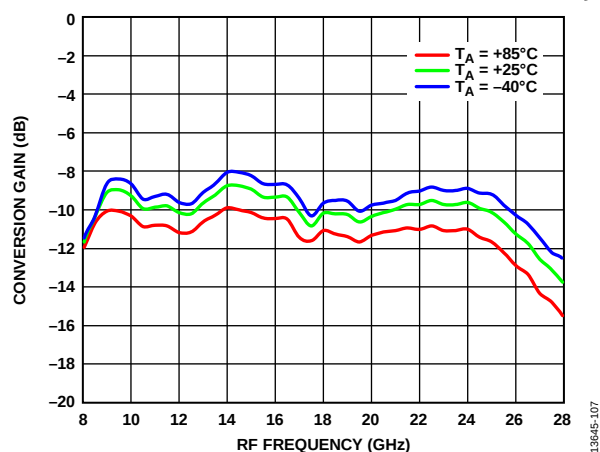


Figure 107. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

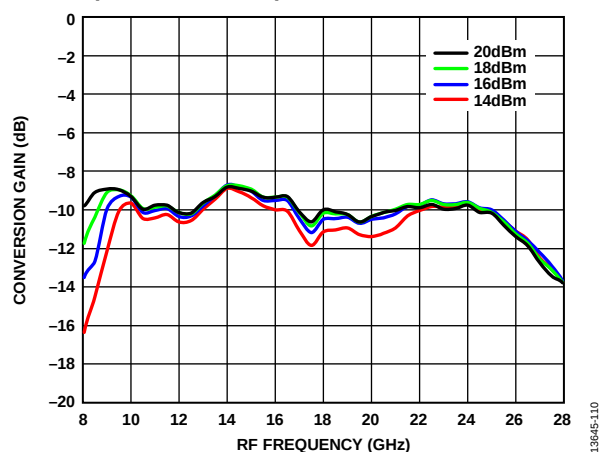
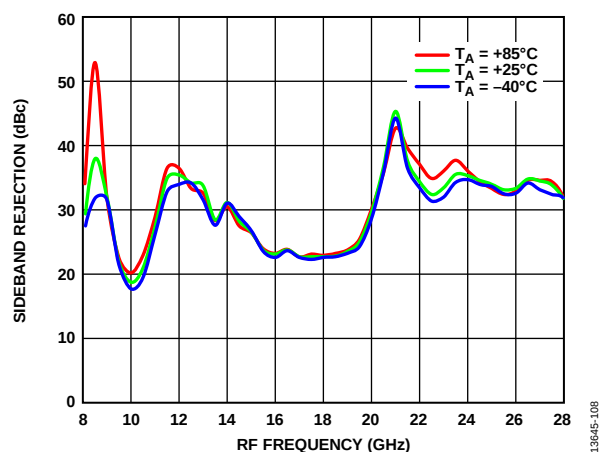
Figure 110. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 108. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

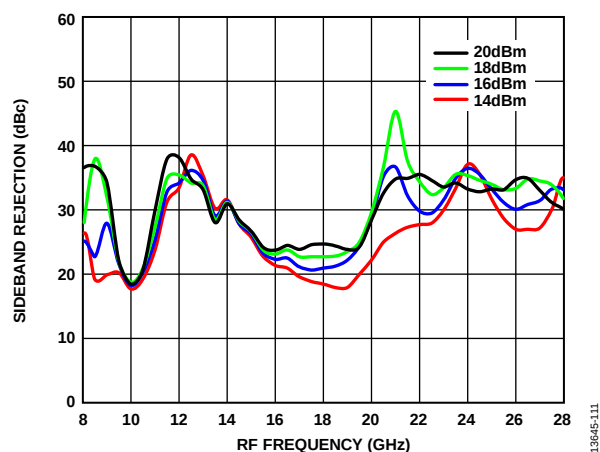
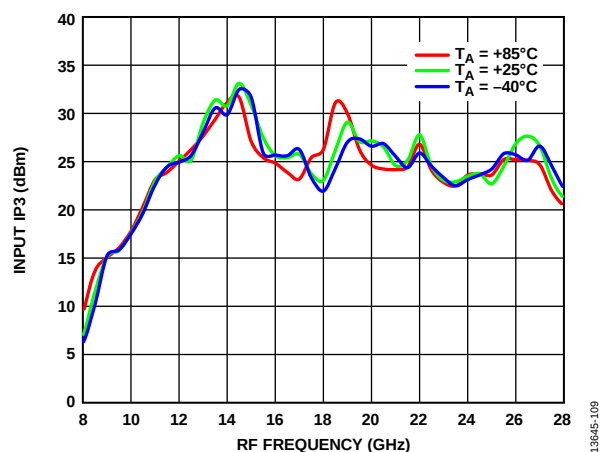
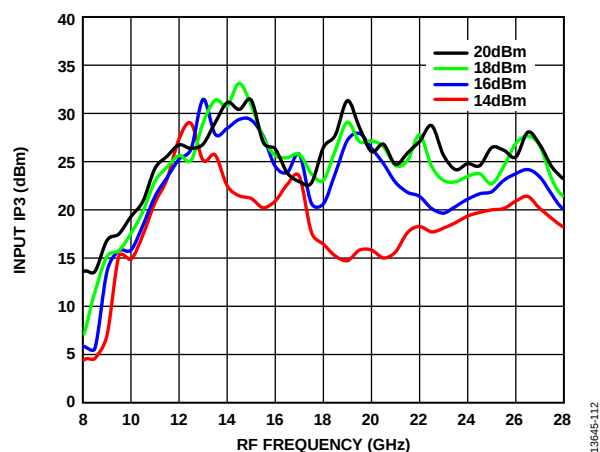
Figure 111. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 109. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 112. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

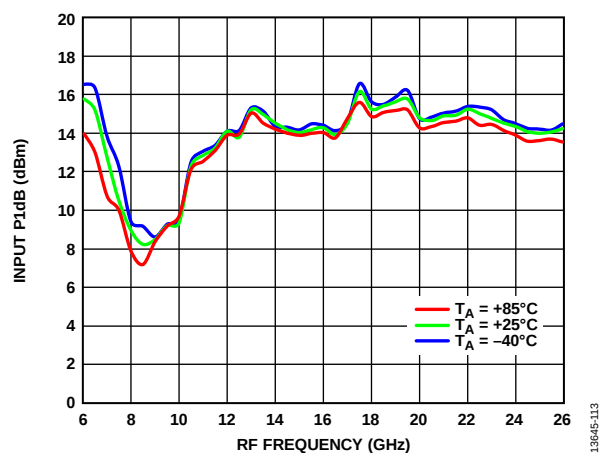


Figure 113. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

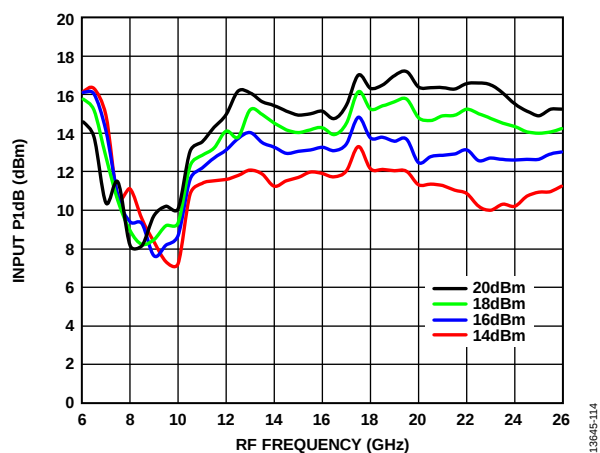


Figure 114. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

UPCONVERTER PERFORMANCE: IF = 5000 MHz, UPPER SIDEBAND (LOW-SIDE LO)

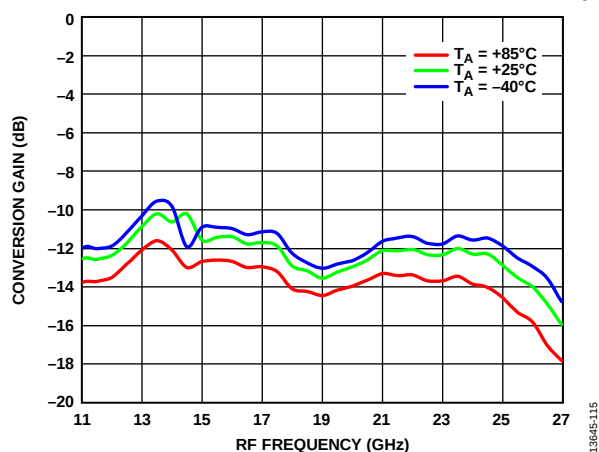


Figure 115. Conversion Gain vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

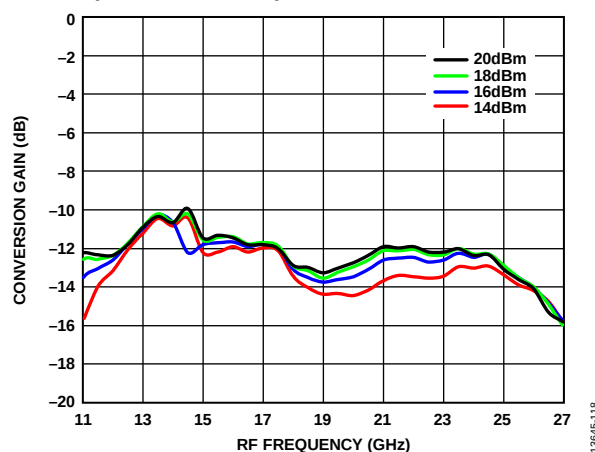
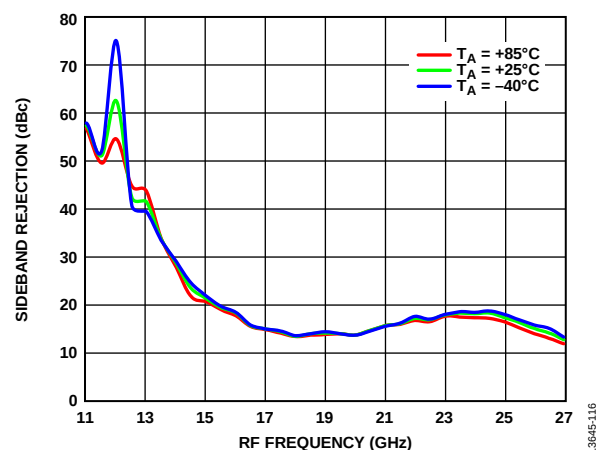
Figure 118. Conversion Gain vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 116. Sideband Rejection vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

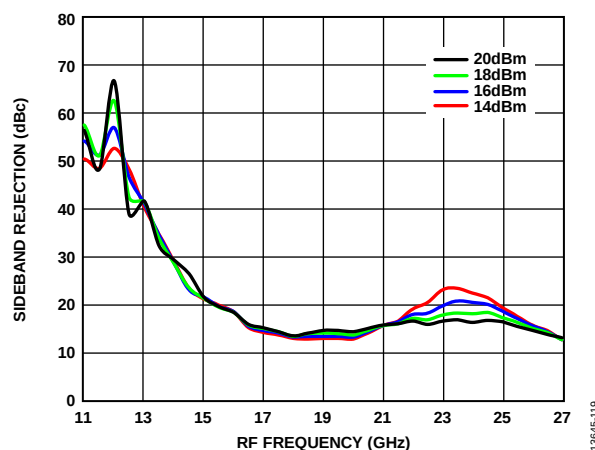
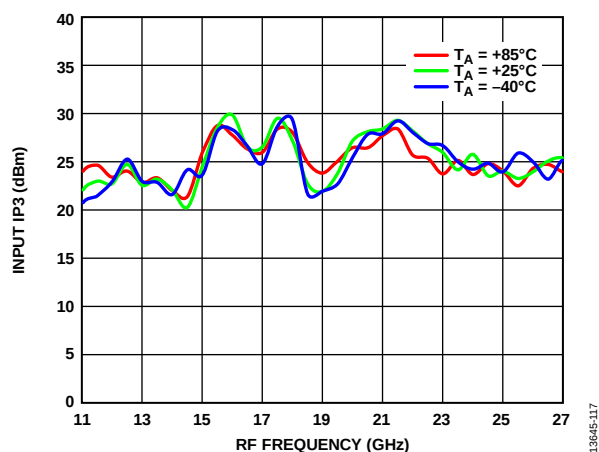
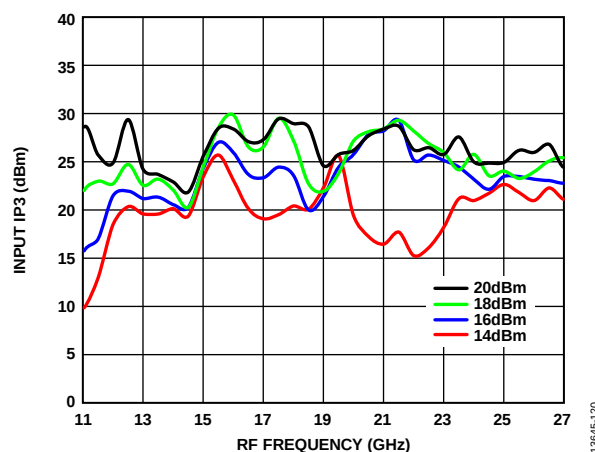
Figure 119. Sideband Rejection vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$ 

Figure 117. Input IP3 vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

Figure 120. Input IP3 vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

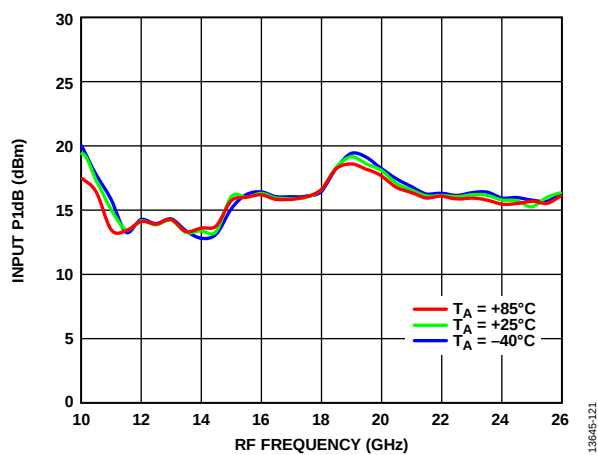


Figure 121. Input P1dB vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm

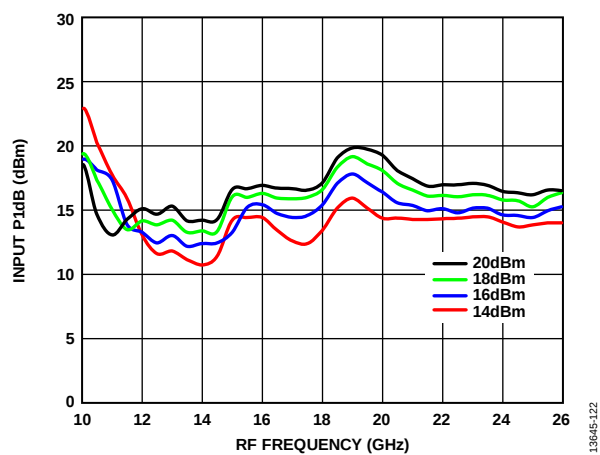


Figure 122. Input P1dB vs. RF Frequency at Various LO Drives, $T_A = 25^\circ\text{C}$

ISOLATION AND RETURN LOSS

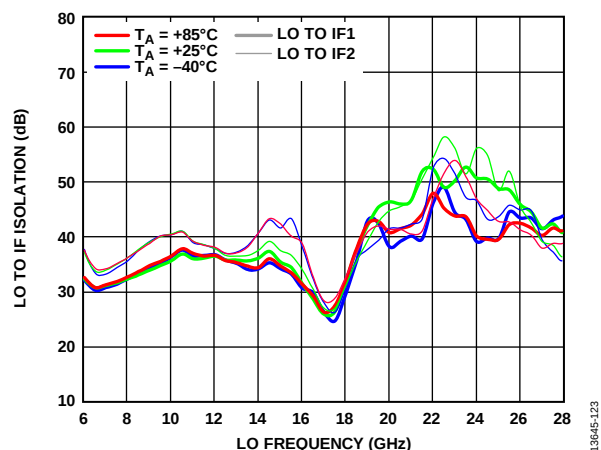


Figure 123. LO to IF Isolation vs. LO Frequency at Various Temperatures, IF = 100 MHz, LO Drive = 18 dBm

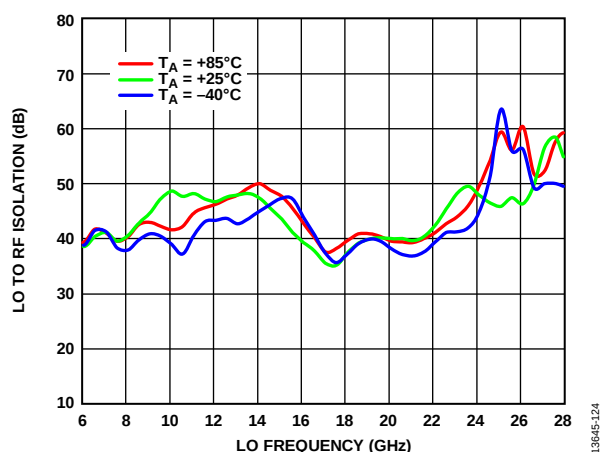


Figure 124. LO to RF Isolation vs. LO Frequency at Various Temperatures, IF = 100 MHz, LO Drive = 18 dBm

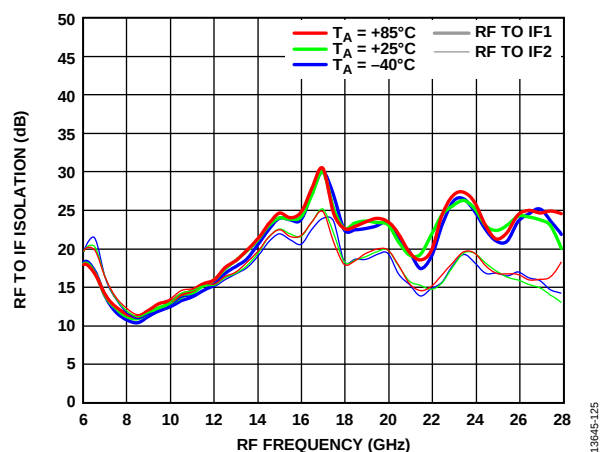


Figure 125. RF to IF Isolation vs. RF Frequency at Various Temperatures, IF = 100 MHz, LO Drive = 18 dBm

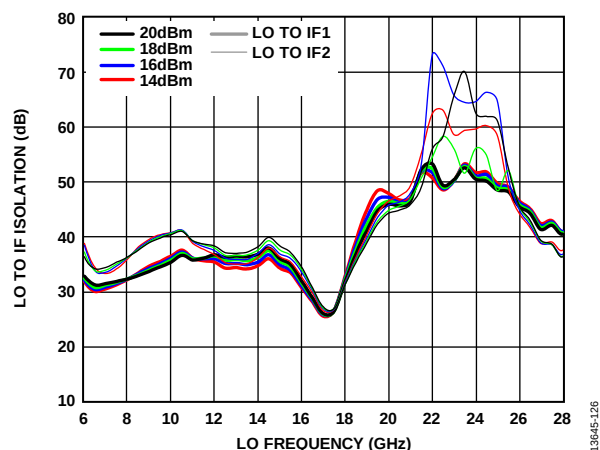


Figure 126. LO to IF Isolation vs. LO Frequency at Various LO Drives, IF = 100 MHz, TA = 25°C

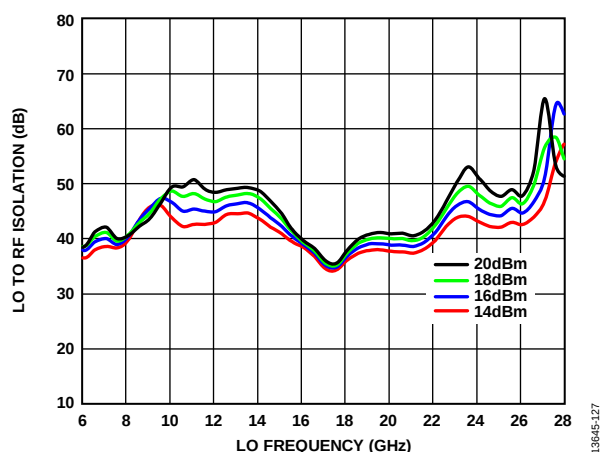


Figure 127. LO to RF Isolation vs. LO Frequency at Various LO Drives, IF = 100 MHz, TA = 25°C

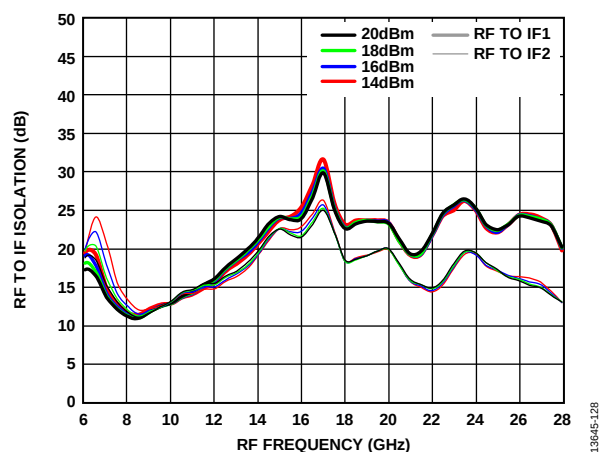


Figure 128. RF to IF Isolation vs. RF Frequency at Various LO Drives, IF = 100 MHz, TA = 25°C

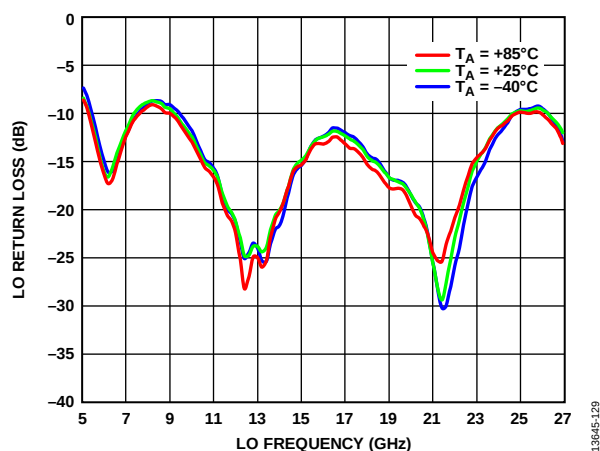


Figure 129. LO Return Loss vs. LO Frequency at Various Temperatures, LO Drive = 18 dBm

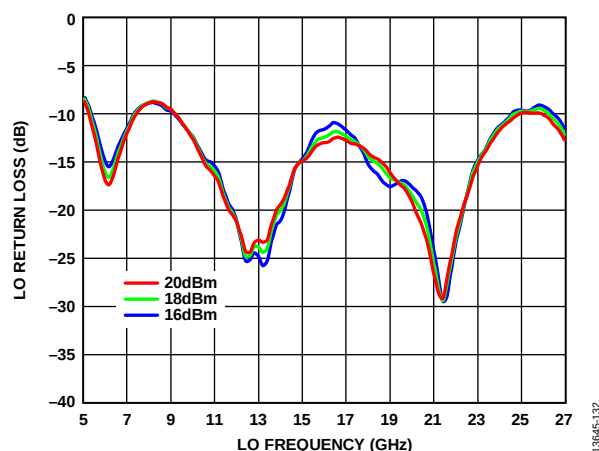


Figure 132. LO Return Loss vs. LO Frequency at Various LO Drives

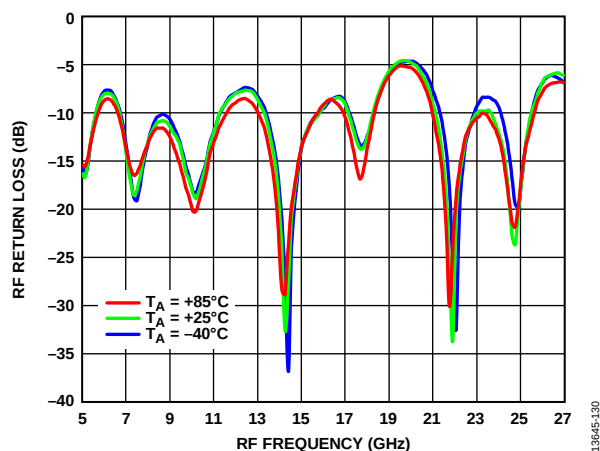


Figure 130. RF Return Loss vs. RF Frequency at Various Temperatures, LO Frequency = 16 GHz, LO Drive = 18 dBm

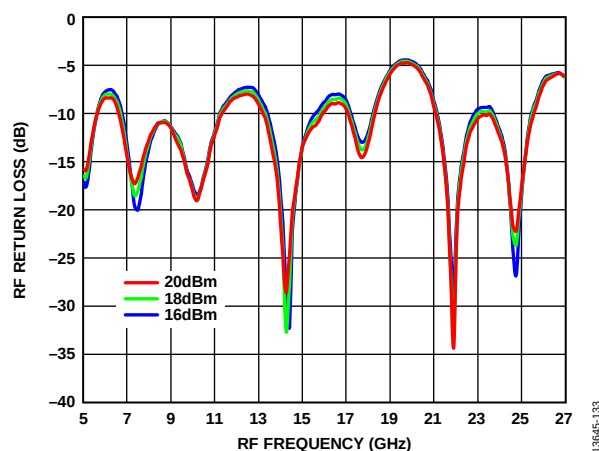


Figure 133. RF Return Loss vs. RF Frequency at Various LO Drives, LO Frequency = 16 GHz

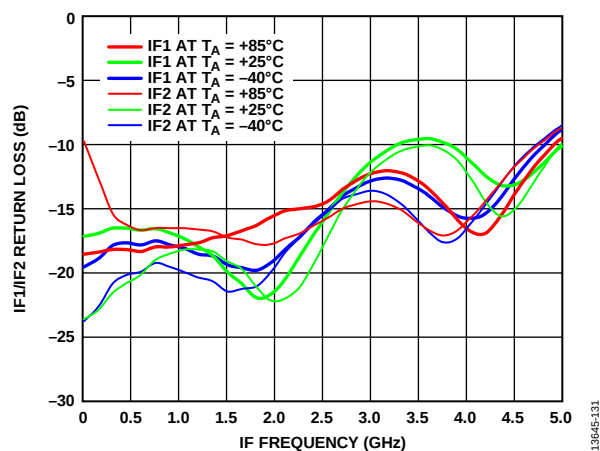


Figure 131. IF1/IF2 Return Loss vs. IF Frequency at Various Temperatures, LO Frequency = 16 GHz, LO Drive = 18 dBm

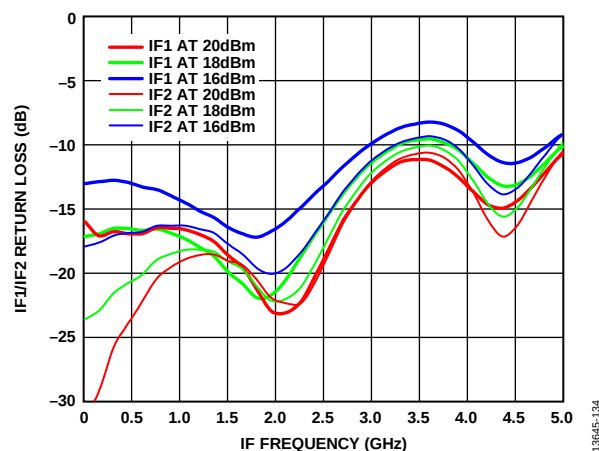


Figure 134. IF1/IF2 Return Loss vs. IF Frequency at Various LO Drives, LO Frequency = 16 GHz

IF BANDWIDTH PERFORMANCE: DOWNCONVERTER, LOWER SIDEBAND (HIGH-SIDE LO)

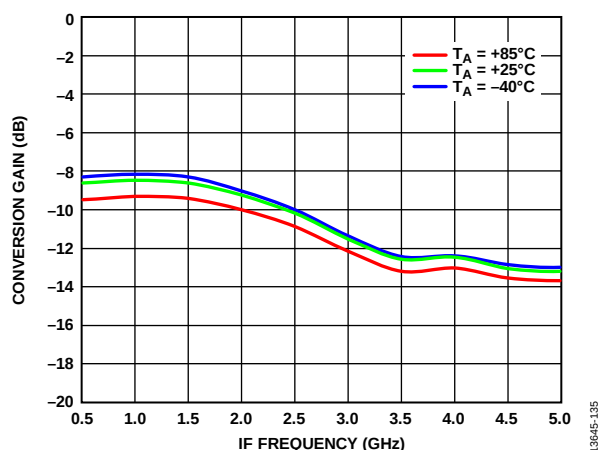


Figure 135. Conversion Gain vs. IF Frequency at Various Temperatures, LO Drive = 18 dBm at 16 GHz

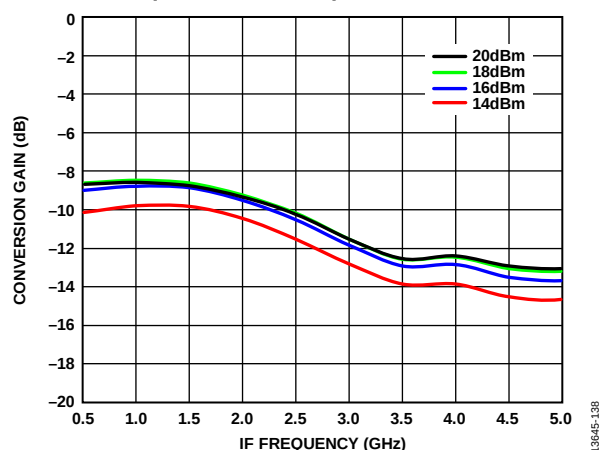
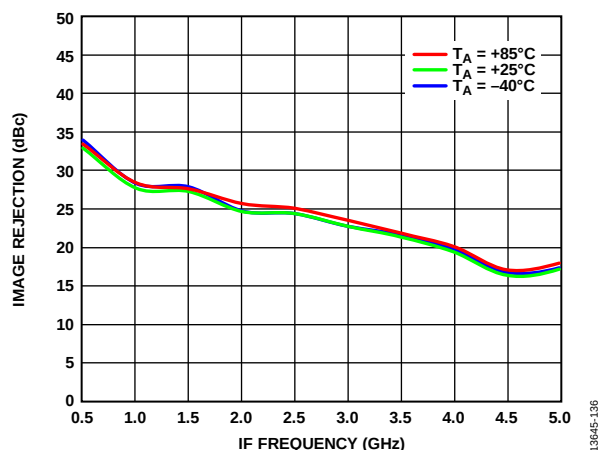
Figure 138. Conversion Gain vs. IF Frequency at Various LO Drives, LO Frequency = 16 GHz, $T_A = 25^\circ\text{C}$ 

Figure 136. Image Rejection vs. IF Frequency at Various Temperatures, LO Drive = 18 dBm at 16 GHz

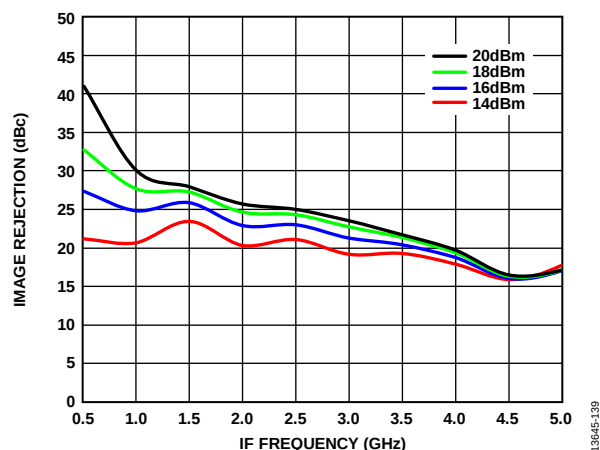
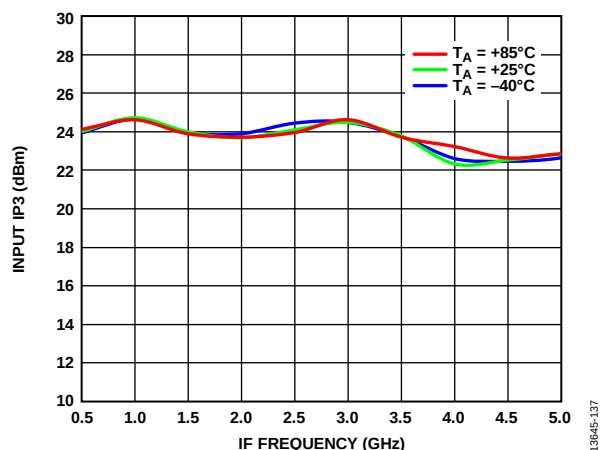
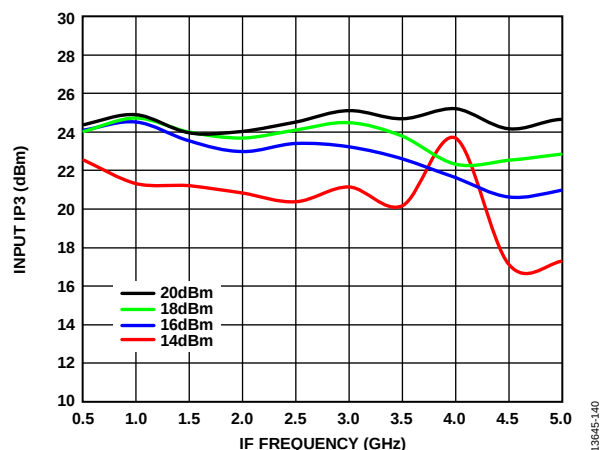
Figure 139. Image Rejection vs. IF Frequency at Various LO Drives, LO Frequency = 16 GHz, $T_A = 25^\circ\text{C}$ 

Figure 137. Input IP3 vs. IF Frequency at Various Temperatures, LO Drive = 18 dBm at 16 GHz

Figure 140. Input IP3 vs. IF Frequency at Various LO Drives, LO Frequency = 16 GHz, $T_A = 25^\circ\text{C}$

AMPLITUDE AND PHASE IMBALANCE PERFORMANCE: DOWNCONVERTER, LOWER SIDEBAND (HIGH-SIDE LO)

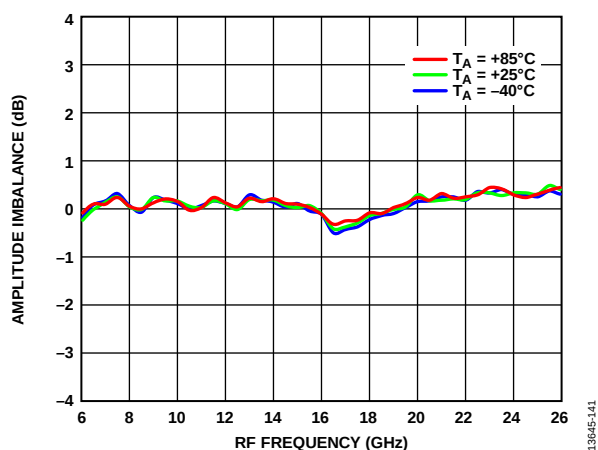


Figure 141. Amplitude Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz

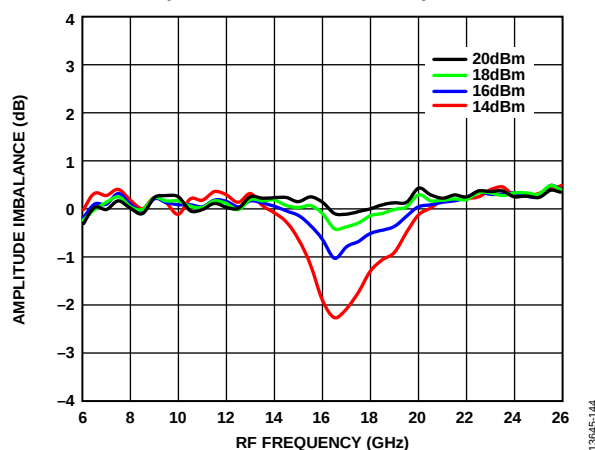


Figure 144. Amplitude Imbalance vs. RF Frequency at Various LO Drives, IF = 100 MHz, $T_A = 25^\circ\text{C}$

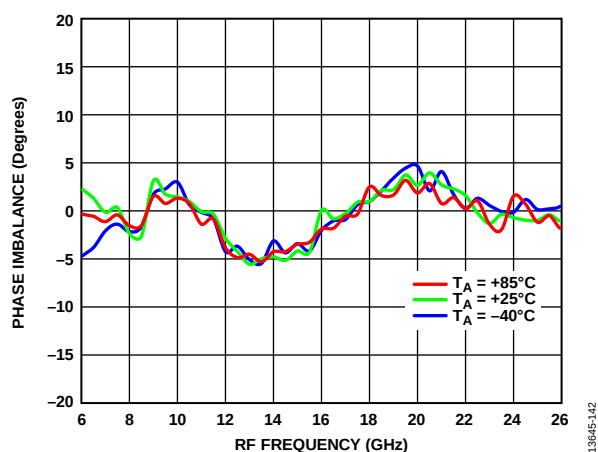


Figure 142. Phase Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz

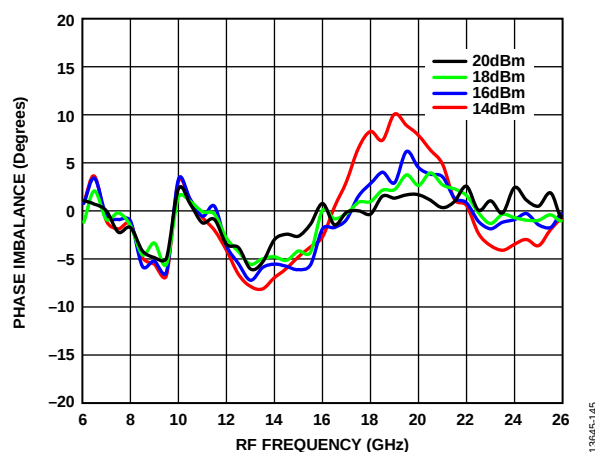


Figure 145. Phase Imbalance vs. RF Frequency at Various LO Drives, IF = 100 MHz, $T_A = 25^\circ\text{C}$

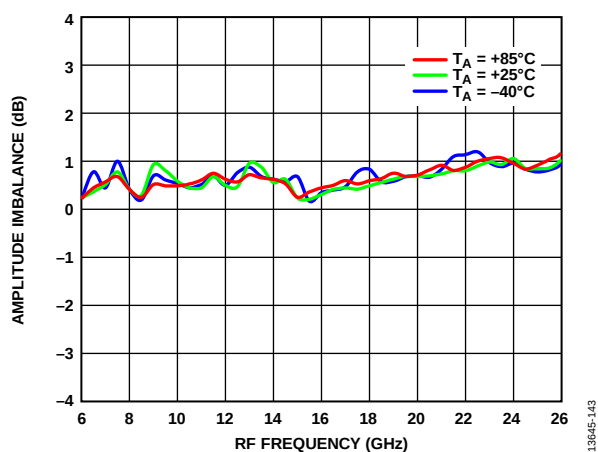


Figure 143. Amplitude Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 2500 MHz

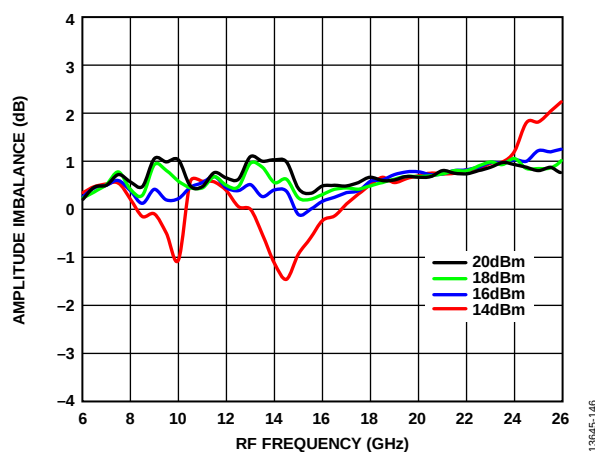


Figure 146. Amplitude Imbalance vs. RF Frequency at Various LO Drives, IF = 2500 MHz, $T_A = 25^\circ\text{C}$

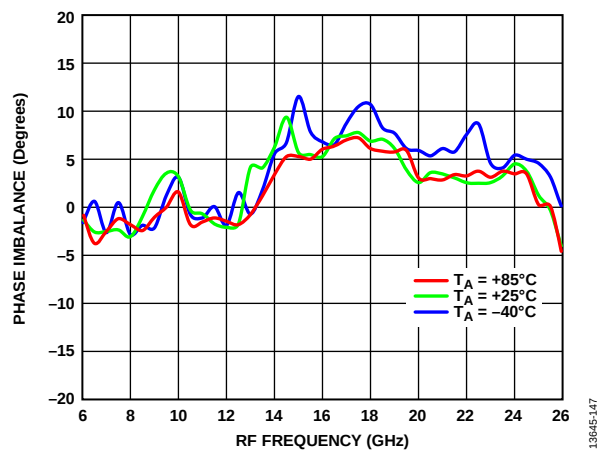


Figure 147. Phase Imbalance vs. RF Frequency at Various Temperatures,
LO Drive = 18 dBm, IF = 2500 MHz

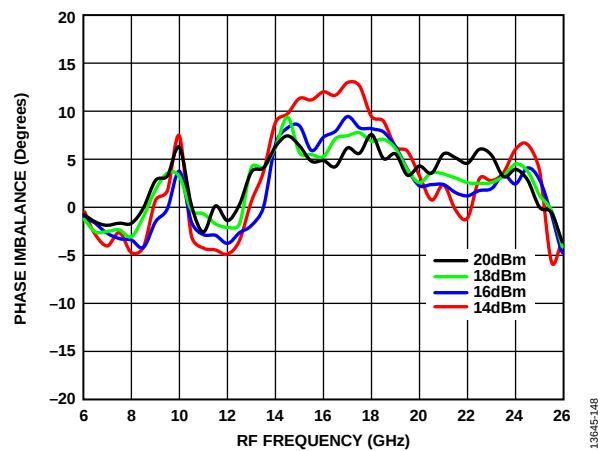


Figure 148. Phase Imbalance vs. RF Frequency at Various LO Drives,
IF = 2500 MHz, $T_A = 25^\circ\text{C}$

AMPLITUDE AND PHASE IMBALANCE PERFORMANCE: DOWNCONVERTER, UPPER SIDEBAND (LOW-SIDE LO)

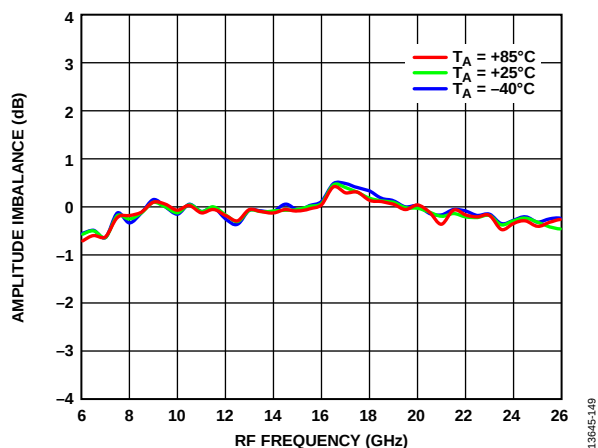


Figure 149. Amplitude Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz

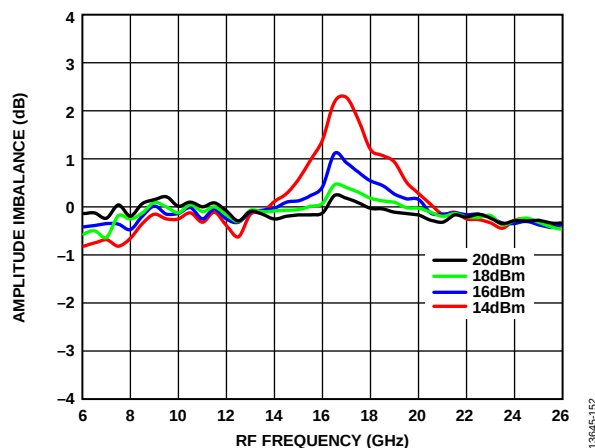


Figure 152. Amplitude Imbalance vs. RF Frequency at Various LO Drives, IF = 100 MHz, $T_A = 25^\circ\text{C}$

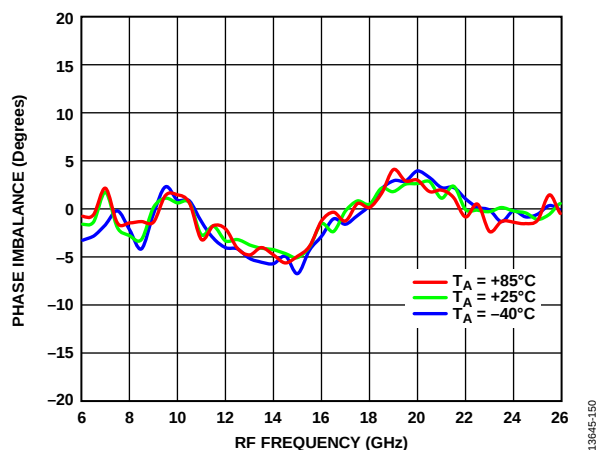


Figure 150. Phase Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 100 MHz

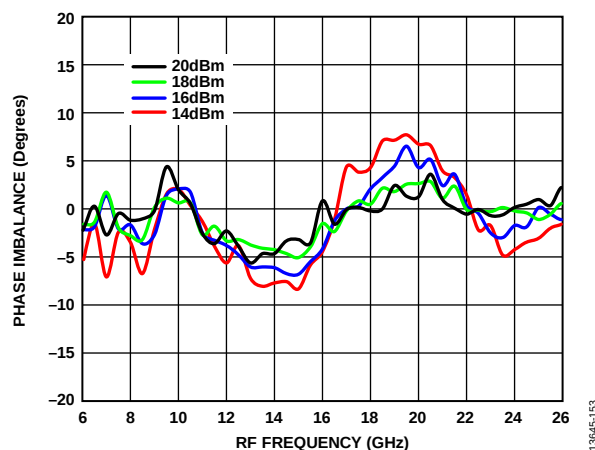


Figure 153. Phase Imbalance vs. RF Frequency at Various LO Drives, IF = 100 MHz, $T_A = 25^\circ\text{C}$

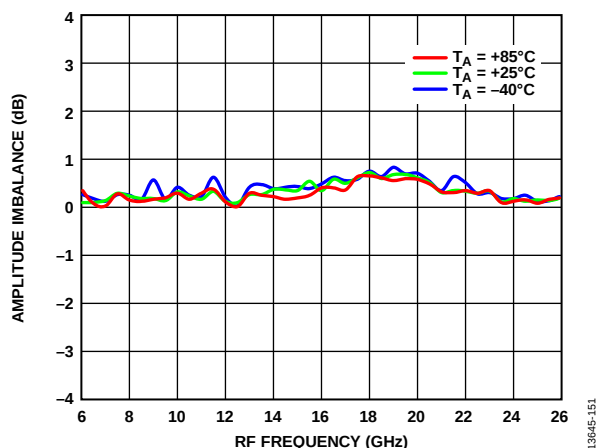


Figure 151. Amplitude Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 2500 MHz

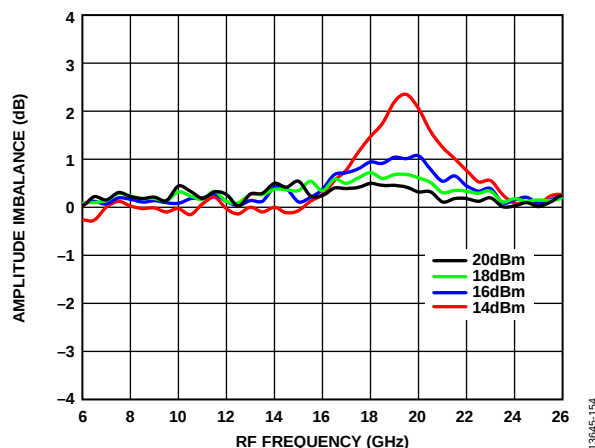


Figure 154. Amplitude Imbalance vs. RF Frequency at Various LO Drives, IF = 2500 MHz, $T_A = 25^\circ\text{C}$

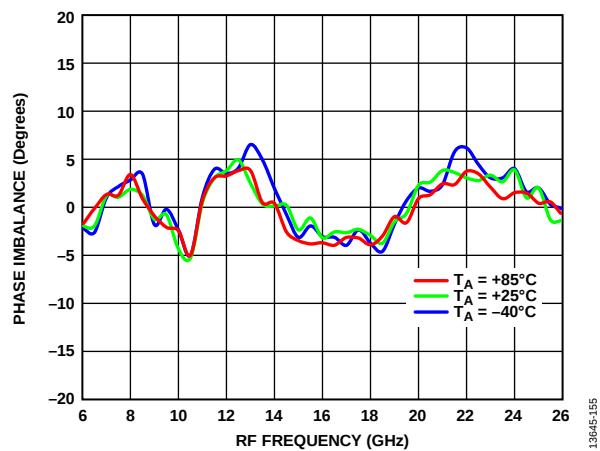


Figure 155. Phase Imbalance vs. RF Frequency at Various Temperatures, LO Drive = 18 dBm, IF = 2500 MHz

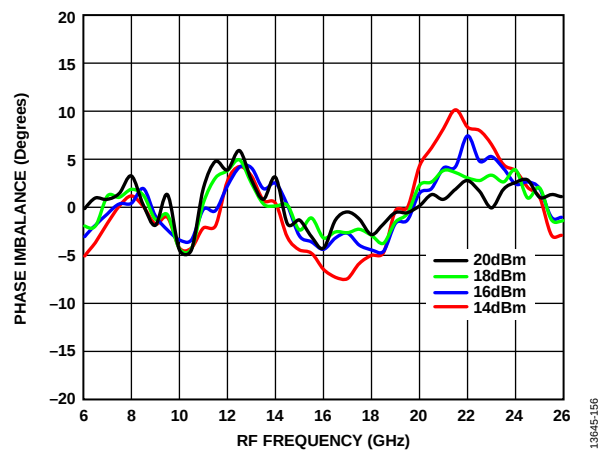


Figure 156. Phase Imbalance vs. RF Frequency at Various LO Drives, IF = 2500 MHz, $T_A = 25^\circ\text{C}$

SPURIOUS AND HARMONICS PERFORMANCE

N/A means not applicable.

LO Harmonics Isolation

LO power = 18 dBm, $T_A = 25^\circ\text{C}$, and all values are in dBc below the input LO level measured at the RF port.

Table 5. $N \times \text{LO}$ Spur at RF Output

LO Frequency (GHz)	$N \times \text{LO}$ Spur at RF Port			
	1	2	3	4
6	37	47	57	37
8	40	52	53	40
10	46	61	62	46
12	47	68	79	47
14	46	68	72	46
16	39	77	N/A	39
18	37	78	N/A	37
20	39	60	N/A	39
22	41	55	N/A	40
24	46	N/A	N/A	46
26	45	N/A	N/A	45

Downconverter $M \times N$ Spurious Outputs

Mixer spurious products are measured in dBc from the IF output power level, unless otherwise specified. Spur values are $(M \times \text{RF}) - (N \times \text{LO})$.

IF = 100 MHz, RF = 6000 MHz, LO = 6100 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0	N/A	-4	+15	+18	+35	N/A
	1	+18	N/A	+23	+31	+41	+57
	2	+57	+49	+43	+53	+54	+72
	3	+87	+72	+62	+59	+63	+63
	4	+84	+86	+88	+80	+78	+80
	5	N/A	+84	+85	+88	+90	+88

IF = 100 MHz, RF = 16000 MHz, LO = 16100 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0	N/A	9	N/A	N/A	N/A	N/A
	1	28	N/A	N/A	N/A	N/A	N/A
	2	N/A	87	75	84	N/A	N/A
	3	N/A	N/A	86	75	90	N/A
	4	N/A	N/A	N/A	88	95	88
	5	N/A	N/A	N/A	N/A	87	96

IF = 100 MHz, RF = 26000 MHz, LO = 26100 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0		7	N/A	N/A	N/A	N/A
	1	21	N/A	46	N/A	N/A	N/A
	2	N/A	84	69	85	N/A	N/A
	3	N/A	N/A	82	83	82	N/A
	4	N/A	N/A	N/A	81	92	81
	5	N/A	N/A	N/A	N/A	81	94

IF = 2500 MHz, RF = 6000 MHz, LO = 8500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0		-2	+26	+22	0	0
	1	+7	N/A	+21	+41	0	0
	2	+73	+71	+65	+75	+86	0
	3	+85	+77	+71	+77	+85	+85
	4	+83	+85	+90	+93	+88	+85
	5	0	+84	+88	+93	+91	+88

IF = 2500 MHz, RF = 16000 MHz, LO = 18500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0	N/A	1	N/A	N/A	N/A	N/A
	1	13	N/A	40	N/A	N/A	N/A
	2	N/A	78	65	79	N/A	N/A
	3	N/A	N/A	85	88	84	N/A
	4	N/A	N/A	N/A	90	87	N/A
	5	N/A	N/A	N/A	85	89	87

IF = 2500 MHz, RF = 26000 MHz, LO = 28500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		$N \times \text{LO}$					
		0	1	2	3	4	5
$M \times \text{RF}$	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	11	N/A	N/A	N/A	N/A	N/A
	2	N/A	82	73	N/A	N/A	N/A
	3	N/A	N/A	85	79	N/A	N/A
	4	N/A	N/A	N/A	85	86	N/A
	5	N/A	N/A	N/A	N/A	83	87

IF = 5000 MHz, RF = 6000 MHz, LO = 11000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × RF	0	N/A	4	23	N/A	N/A	N/A
	1	7	N/A	30	N/A	N/A	N/A
	2	73	92	68	84	N/A	N/A
	3	86	82	83	83	82	N/A
	4	82	87	92	87	86	N/A
	5	N/A	83	88	88	84	83

IF = 5000 MHz, RF = 16000 MHz, LO = 21000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × RF	0	N/A	3	N/A	N/A	N/A	N/A
	1	11	N/A	40	N/A	N/A	N/A
	2	N/A	83	85	N/A	N/A	N/A
	3	N/A	N/A	85	82	N/A	N/A
	4	N/A	N/A	82	81	84	N/A
	5	N/A	N/A	N/A	82	88	80

IF = 5000 MHz, RF = 26000 MHz, LO = 31000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	5	N/A	N/A	N/A	N/A	N/A
	2	N/A	58	43	N/A	N/A	N/A
	3	N/A	N/A	66	67	N/A	N/A
	4	N/A	N/A	N/A	79	78	N/A
	5	N/A	N/A	N/A	N/A	82	75

Upconverter M × N Spurious Outputs

Mixer spurious products are measured in dBc from the RF output power level, unless otherwise specified. Spur values are (M × IF) – (N × LO).

IF = 100 MHz, RF = 6000 MHz, LO = 6100 GHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	-2	+10	+16	+18	+32
	1	+56	N/A	+16	+14	+31	+33
	2	+93	+46	+45	+44	+51	+56
	3	+92	+51	+62	+49	+57	+62
	4	N/A	-2	+10	+16	+18	+32
	5	+56	N/A	+16	+14	+31	+33

IF = 100 MHz, RF = 16000 MHz, LO = 16100 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0		2	37	N/A	N/A	N/A
	1	62	N/A	47	N/A	N/A	N/A
	2	94	58	79	N/A	N/A	N/A
	3	95	80	80	N/A	N/A	N/A
	4	95	83	78	N/A	N/A	N/A
	5	93	81	80	N/A	N/A	N/A

IF = 100 MHz, RF = 26000 MHz, LO = 26100 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0		4	N/A	N/A	N/A	N/A
	1	58	N/A	N/A	N/A	N/A	N/A
	2	93	57	N/A	N/A	N/A	N/A
	3	92	80	N/A	N/A	N/A	N/A
	4	91	82	N/A	N/A	N/A	N/A
	5	90	82	N/A	N/A	N/A	N/A

IF = 2500 MHz, RF = 6000 MHz, LO = 8500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	4	11	14	50	0
	1	14	0	15	30	43	58
	2	83	81	70	66	82	68
	3	71	90	78	69	80	75
	4	90	92	86	82	82	78
	5	85	91	89	84	82	82

IF = 2500 MHz, RF = 16000 MHz, LO = 18500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	N/A	27	N/A	N/A	N/A
	1	15	N/A	52	N/A	N/A	N/A
	2	92	70	78	N/A	N/A	N/A
	3	87	88	83	N/A	N/A	N/A
	4	89	87	83	N/A	N/A	N/A
	5	88	93	80	N/A	N/A	N/A

IF = 2500 MHz, RF = 26000 MHz, LO = 28500 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	12	N/A	N/A	N/A	N/A	N/A
	1	50	61	N/A	N/A	N/A	N/A
	2	70	66	N/A	N/A	N/A	N/A
	3	86	82	N/A	N/A	N/A	N/A
	4	83	83	N/A	N/A	N/A	N/A
	5	12	N/A	N/A	N/A	N/A	N/A

IF = 5000 MHz, RF = 6000 MHz, LO = 11000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	6	23	34	N/A	N/A
	1	5	N/A	17	47	70	N/A
	2	80	91	65	79	75	N/A
	3	81	89	85	83	84	72
	4	80	86	91	83	82	75
	5	77	83	89	85	82	82

IF = 5000 MHz, RF = 16000 MHz, LO = 21000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	-3	N/A	N/A	N/A	N/A
	1	+1	N/A	+62	N/A	N/A	N/A
	2	+81	+72	+74	N/A	N/A	N/A
	3	+82	+87	+81	N/A	N/A	N/A
	4	+79	+89	+81	N/A	N/A	N/A
	5	+76	+88	+84	+65	N/A	N/A

IF = 5000 MHz, RF = 26000 MHz, LO = 31000 MHz,
RF power = -10 dBm, LO power = 18 dBm, and $T_A = 25^\circ\text{C}$.

		N × LO					
		0	1	2	3	4	5
M × IF	0	N/A	+3	N/A	N/A	N/A	N/A
	1	-7	N/A	N/A	N/A	N/A	N/A
	2	+41	+42	N/A	N/A	N/A	N/A
	3	+77	+78	N/A	N/A	N/A	N/A
	4	+76	+81	N/A	N/A	N/A	N/A
	5	+72	+84	+64	N/A	N/A	N/A

THEORY OF OPERATION

The HMC8191 is a passive, wideband, I/Q MMIC mixer that can be used either as an image reject mixer for receiver operations, or as a single-sideband upconverter for transmitter operations. With an RF and LO range of 6 GHz to 26.5 GHz, and an IF bandwidth of dc to 5 GHz, the HMC8191 is ideal for applications requiring wide frequency range, excellent RF performance, and a simple design with fewer components and a small PCB footprint. A single HMC8191 can replace multiple narrow-band mixers in a design.

The inherent I/Q architecture of the HMC8191 offers excellent image rejection and thereby eliminates the need for expensive filtering for unwanted sidebands. The double balanced architecture of the mixer also provides excellent LO to RF isolation and LO to IF isolation, and reduces the effect of LO leakage to ensure signal integrity.

Because the HMC8191 is a passive mixer, the HMC8191 does not require any dc power sources. It offers a lower noise figure compared to an active mixer, ensuring superior dynamic range for high performance and precision applications.

The HMC8191 is fabricated on a GaAs MESFET process and uses Analog Devices mixer cells and a 90° hybrid. The HMC8191 is available in a compact, 4 mm × 4 mm, 24-terminal LCC package and operates over a –40°C to +85°C temperature range. An evaluation board for the HMC8191 is also available from the Analog Devices website.

For both upconversion and downconversion, an external 90° hybrid is required. See the Applications Information section for details to interface with an external 90° hybrid.

APPLICATIONS INFORMATION

Figure 157 shows the typical application circuit for the HMC8191. To select the appropriate sideband, an external 90° hybrid is needed. For applications not requiring operation to dc, use an off-chip dc blocking capacitor. For applications that require the LO signal at the output to be suppressed, use a bias tee or RF feed as shown in Figure 157. Ensure that the source or sink current used for LO suppression is <3 mA for each IF port to

prevent damage to the device. The common-mode voltage for each IF port is 0 V.

To select the upper sideband, connect the IF1 pin to the 90° port of the hybrid, and connect the IF2 pin to the 0° port of the hybrid. To select the lower sideband, connect IF1 to the 0° port of the hybrid and IF2 to the 90° port of the hybrid.

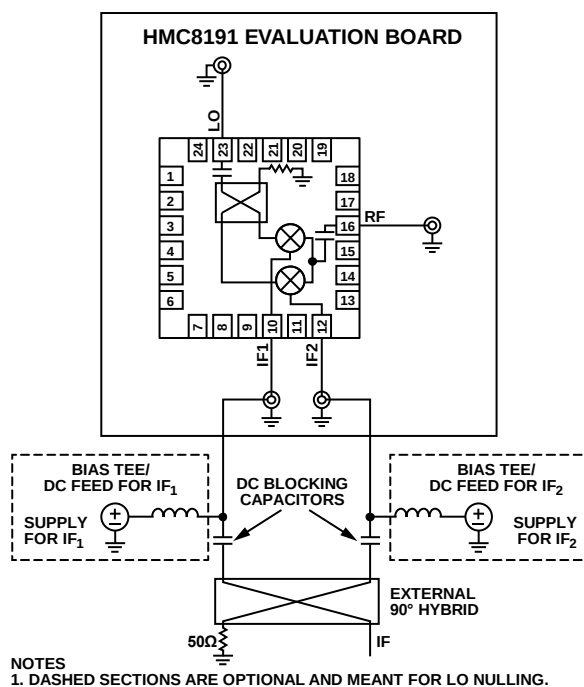


Figure 157. Typical Application Circuit

RF AND LO PERFORMANCE ABOVE 26 GHz

Figure 158 and Figure 159 shows the RF performance above 26 GHz for both upconversion and downconversion. The data was taken at an IF frequency of 100 MHz and LO power at 18 dBm.

Note that this performance is typical and not guaranteed.

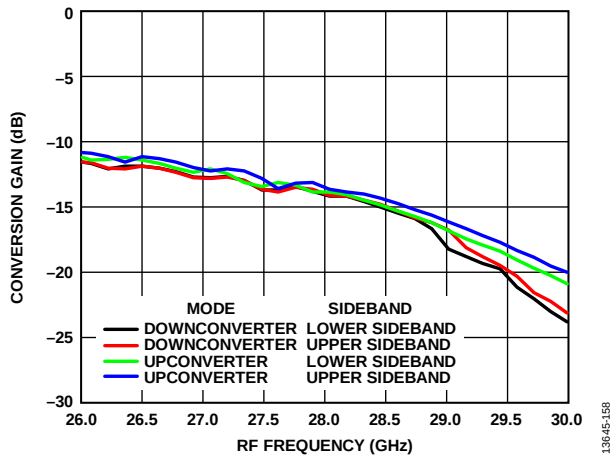


Figure 158. Conversion Gain vs. RF Frequency above 26 GHz at $T_A = 25^\circ\text{C}$ for Upper and Lower Sidebands, Upconversion and Downconversion, LO Drive = 18 dBm, IF = 100 MHz, External Hybrid Not Calibrated

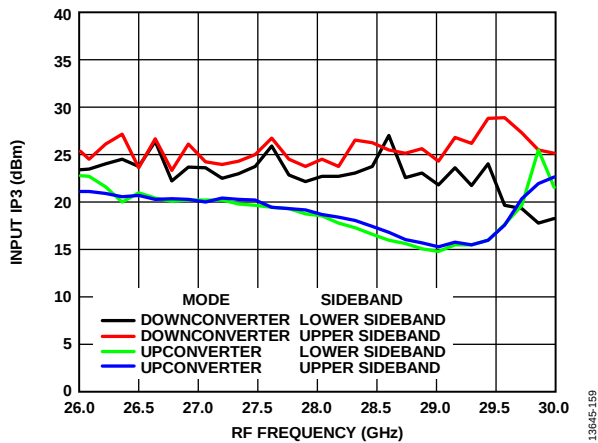


Figure 159. Input IP3 vs. RF Frequency above 26 GHz at $T_A = 25^\circ\text{C}$ for Upper and Lower Sidebands, Upconversion and Downconversion, LO Drive = 18 dBm, IF = 100 MHz, External Hybrid Not Calibrated

IF BANDWIDTH ABOVE 5 GHz

Figure 160, Figure 161, and Figure 162 show the IF performance above 5 GHz. The data for these figures has been taken in upconverter configuration at LO frequency and power of 16 GHz and 18 dBm, respectively.

Note that this performance is typical and not guaranteed.

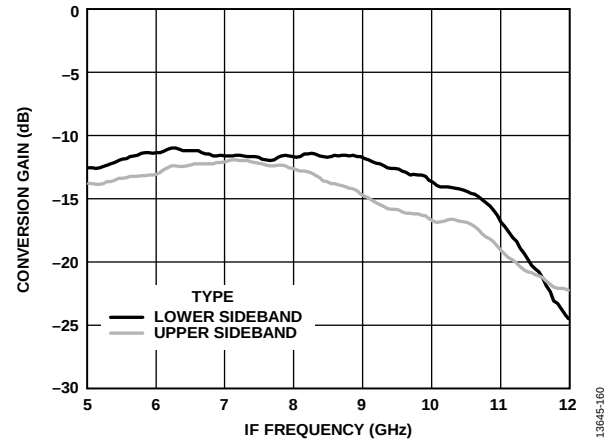


Figure 160. Conversion Gain vs. IF Frequency above 5 GHz at $T_A = 25^\circ\text{C}$ for Upper and Lower Sidebands, LO Drive = 18 dBm at 16 GHz, Calibration to the Connector of the Evaluation Board

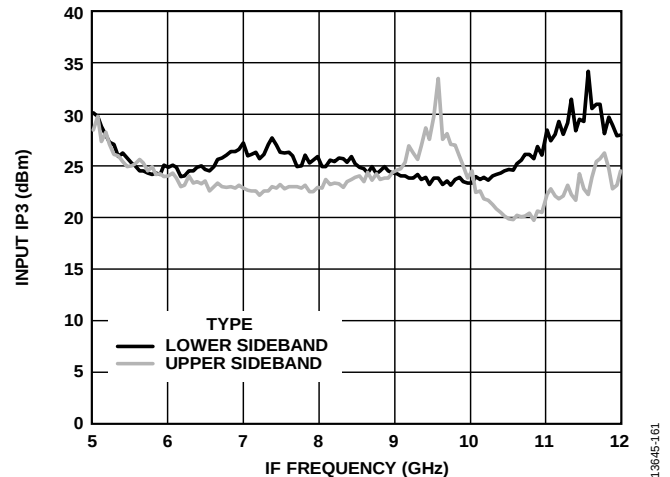


Figure 161. Input IP3 vs. IF Frequency above 5 GHz at $T_A = 25^\circ\text{C}$ for Upper and Lower Sidebands, LO Drive = 18 dBm at 16 GHz, Calibration to the Connector of the Evaluation Board

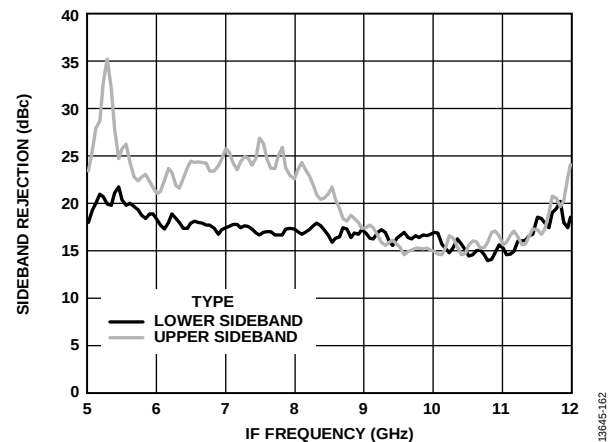


Figure 162. Sideband Rejection vs. IF Frequency above 5 GHz at $T_A = 25^\circ\text{C}$ for Upper and Lower Sidebands, LO Drive = 18 dBm at 16 GHz, Calibration to the Connector of the Evaluation Board

SOLDERING INFORMATION AND RECOMMENDED LAND PATTERN

Figure 163 shows the recommended land pattern for the HMC8191. The HMC8191 is contained in a 4 mm × 4 mm 24-terminal, ceramic, LCC package, with an exposed ground pad (EPAD). This pad is internally connected to the ground of the chip. To minimize thermal impedance and ensure electrical performance, solder the pad to the low impedance ground plane on the PCB. It is recommended that the ground planes on all layers under the pad be stitched together with vias, to further reduce thermal impedance.

The land pattern on the HMC8191 evaluation board provides a simulated thermal resistance (θ_{JA}) of 38.3°C/W.

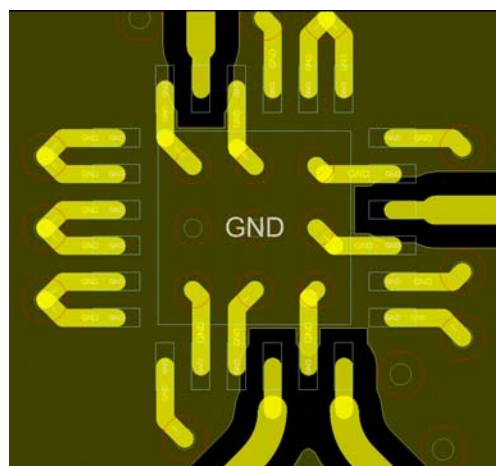


Figure 163. Evaluation Board Layout for the HMC8191 Package

EVALUATION BOARD INFORMATION

The EV1HMC8191LC4 evaluation board PCB used in the application must use RF circuit design techniques. Signal lines must have a 50 Ω impedance and connect the package ground leads and exposed pad directly to the ground plane similar to the setup shown in Figure 164. Use a sufficient number of via holes to connect the top and bottom ground planes.

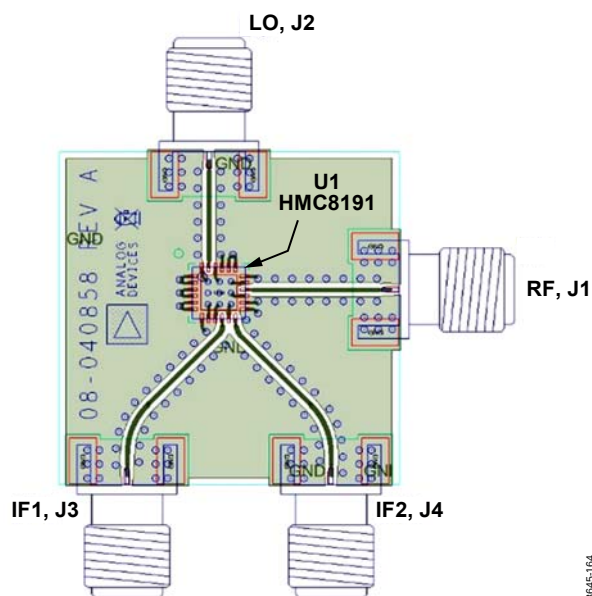


Figure 164. EV1HMC8191LC4 Evaluation Board PCB, Top Layer

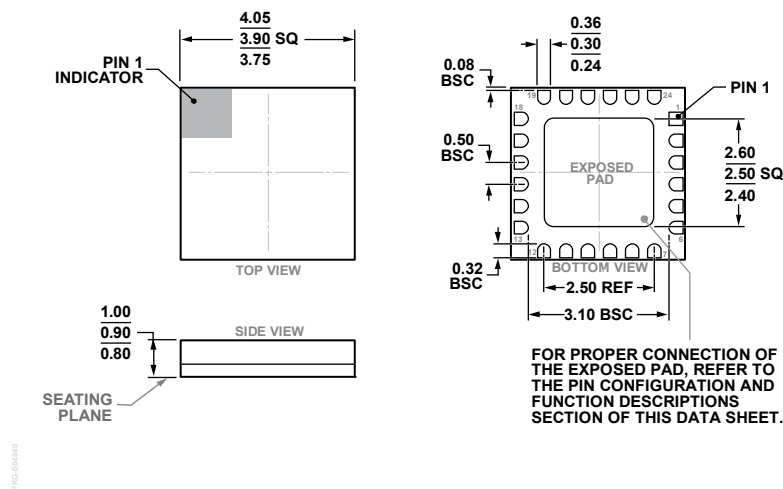
Table 6. Bill of Materials for the EV1HMC8191LC4¹ Evaluation Board PCB

Quantity	Reference Designator	Description	Manufacturer	Part Number
1		PCB, EV1HMC8191LC4 ²	Analog Devices	08_040858a
4	J1 to J4	2.92 mm SMA connectors, SRI Connector Gage	SRI Connector Gage Co.	25-146-1000-92
1	U1	Device under test, HMC8191	Analog Devices	HMC8191

¹ Reference this number when ordering the evaluation board PCB.

² Circuit board material: Rogers 4350.

OUTLINE DIMENSIONS



ORDERING GUIDE

Model ¹	Temperature Range	Package Body Material	Lead Finish	Package Description	MSL Rating ²	Package Option	Package Marking ³
HMC8191LC4	−40°C to +85°C	Alumina Ceramic	Gold over Nickel	24-Terminal LCC	MSL3	E-24-1	H8191 XXXX
HMC8191LC4TR	−40°C to +85°C	Alumina Ceramic	Gold over Nickel	24-Terminal LCC	MSL3	E-24-1	H8191 XXXX
HMC8191LC4TR-R5	−40°C to +85°C	Alumina Ceramic	Gold over Nickel	24-Terminal LCC	MSL3	E-24-1	H8191 XXXX
EV1HMC8191LC4				Evaluation PCB Assembly			

¹ The HMC8191LC4, HMC8191LC4TR, and the HMC8191LC4TR-R5 are RoHS Compliant Parts.

² See the Absolute Maximum Ratings section.

³ The HMC8191LC4, HMC8191LC4TR, and the HMC8191LC4TR-R5 have a four-digit lot number.