DP83905

DP83905 AT/LANTIC(TM) AT Local Area Network Twisted-Pair Interface
Controller



Literature Number: SNLS087A



PRELIMINARY

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DP83905 AT/LANTIC™ AT Local Area Network Twisted-Pair **Interface Controller**

General Description

The AT/LANTIC AT Local Area Network Twisted-pair Interface Controller is a CMOS VLSI device designed for easy implementation of CSMA/CD local area networks.

Unique to the AT/LANTIC is the integration of the entire bus interface for PC-AT® ISA (Industry Standard Architecture) bus based systems. Hardware and software selectable options allow the AT/LANTIC's bus interface to be configured software compatible to either an NE2000 or Ethercard PLUS16TM. All bus drivers and control logic are integrated to reduce board cost and area.

Supported network interfaces include 10BASE5 or 10BASE2 Ethernet via an external transciever connected to its AUI port, and Twisted-pair Ethernet (10BASE-T) using the on-board transceiver. The AT/LANTIC provides the Ethernet Media Access Control (MAC), Encode-Decode (ENDEC) with an AUI interface, and 10BASE-T transceiver functions in accordance with the IEEE 802.3 standards.

The AT/LANTIC's integrated 10BASE-T transceiver fully complies with the IEEE standard. This functional block incorporates the receiver, transmitter, collision, heartbeat, loopback, jabber, and link integrity blocks as defined in the standard. The transceiver when combined with equalization resistors, transmit/receive filters, and pulse transformers provides a complete physical interface from the AT/LANTIC Controller's ENDEC module and the twisted pair medium.

Features

- Controller and integrated bus interface solution for IEEE 802.3, 10BASE5, 10BASE2, and 10BASE-T
- Software compatible with Novell®'s NE2000/Plus industry standard Ethernet Adapters
- Selectable buffer memory size
- No external bus logic or drivers
- Integrated controller, ENDEC, and transceiver
- Full IEEE 802.3 AUI interface
- Single 5V supply

10BASE-T TRANSCEIVER MODULE:

- Integrates transceiver functionality:
 - Transmitter and receiver functions
 - Collision detect, heartbeat and jabber
 - Selectable link integrity test or link disable
 - Polarity Detection/Correction

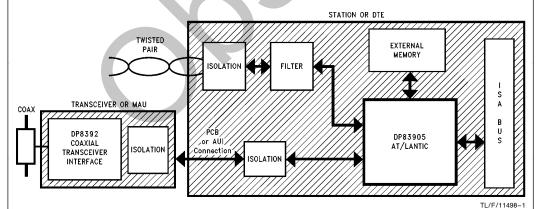
ENDEC MODULE:

- 10 Mbit/s Manchester encoding/decoding
- Squelch on receive and collision pairs

MAC/CONTROLLER MODULE:

- Software compatible with DP8390, DP83901, DP83902
- Efficient buffer management implementation

1.0 System Diagram



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General Description (Continued)

The integrated ENDEC module allows Manchester encoding and decoding via a differential transceiver and phase lock loop decoder at 10 Mbit/sec. Also included are a collision detect translator and diagnostic loopback capability. The ENDEC module interfaces directly to the transceiver module, and also provides a fully IEEE compliant AUI (Attachment Unit Interface) for connection to other media transceivers.

The Media Access Control function which is provided by the Network Interface Control module (NIC) provides simple and efficient packet transmission and reception control by means of off-board memory which can be accessed either through an I/O port or mapped into the system memory.

AT/LANTIC Controller provides a comprehensive solution for 10BASE-T IEEE 802.3 networks. Due to the inherent constraints of CMOS processing, isolation is required at the AUI differential signal interface for 10BASE5 and 10BASE2 applications.

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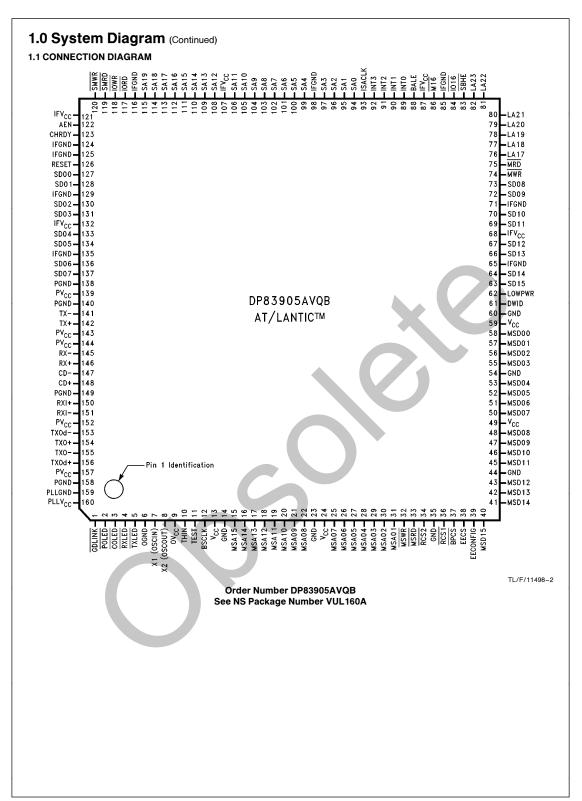
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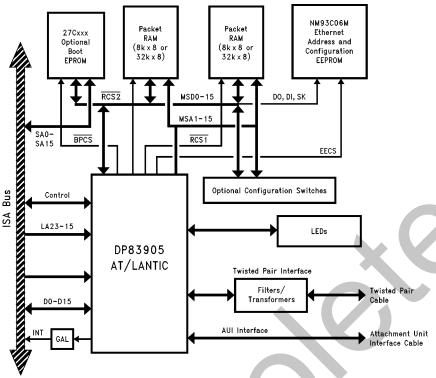
2.0 Piı	2.0 Pin Description					
Pin No.	Pin Name	Type*	Description			
ISA BUS IN	TERFACE PIN	S				
94–97 99–106 108–115	SA0-SA1 9	I TTL	LATCHED ADDRESS BUS: Low-order 20 bits of the system's 24 bit address bus. These lines are enabled onto the bus, by the system, when BALE is high and are latched when BALE returns low. These bits are used to decode accesses to the AT/LANTIC Controller's I/O map and to the boot PROM. In addition they are used to decode accesses to the AT/LANTIC Controller's memory in shared memory mode.			
76-82	LA17-LA23	I TTL	UNLATCHED ADDRESS BUS: High order 7 bits of the 24-bit system address bus. These lines are valid on the falling edge of BALE. These bits are used to decode accesses to the AT/LANTIC Controller's memory in shared memory mode.			
127, 128, 130, 131, 133, 134, 136, 137, 73, 72, 70, 69, 67, 66, 64, 63	SD0-SD15	1/O 3SH	SYSTEM DATA BUS: 16-bit system data bus. Used to transfer data between the system and the AT/LANTIC Controller.			
88	BALE	I TTL	BUS ADDRESS LATCH ENABLE: This signal indicates when the system address lines are valid.			
83	SBHE	I TTL	SYSTEM BUS HIGH ENABLE: This signal indicates that the system expects a transfer on the upper byte lane.			
86	M16	OCH	16-BIT MEMORY TRANSFER: In 16-bit shared memory mode this signal indicates that the AT/LANTIC Controller has decoded an address within the 128 kbyte space that it occupies part of.			
84	IO16	O OCH	16-BIT I/O TRANSFER: In I/O mode this signal indicates that the AT/LANTIC Controller is responding to a 16-bit I/O access by driving 16-bits of data on the bus.			
74	MWR	I TTL	MEMORY WRITE STROBE: Strobe from system to write to AT/LANTIC Controller's memory map. <i>This pin should be connected to allow the CHRDY fix in 16-bit I/O mode to operate correctly.</i> (See Section 6.0)			
75	MRD	I TTL	MEMORY READ STROBE: Strobe from system to read from AT/LANTIC Controller's memory map. This pin should be connected to allow the CHRDY fix in 16-bit I/O mode to operate correctly. (See Section 6.0)			
119, 120	SMRD & SMWR	I TTL	LOW MEMORY STROBES: In Memory mode these signals strobe memory transfers in the same manner as MRD and MWR except that these signals only occur if the access is to the lowest 1 Megabyte. This partial address decode means that these signals can be used in an 8-bit slot to properly decode an access to this area. The AT/LANTIC Controller will use MRD and MWR in 16-bit Memory mode and will use SMRD and SMWR in Memory mode when DWID is low (8-bit mode). SMRD and SMWR are also used to access the BOOT PROM.			
118	IOWR	TTL	I/O WRITE STROBE: Strobe from system to write to the AT/LANTIC Controller's I/O map.			
117	IORD	I TTL	I/O READ STROBE: Strobe from system to read from the AT/LANTIC Controller's I/O map.			
126	RESET	I TTL	RESET: This signal is output by the system to reset all devices on the bus.			

2.0 Pi	2.0 Pin Description (Continued)				
Pin No.	Pin Name	Type*	Description		
ISA BUS IN	ITERFACE PINS (C	ontinued)			
123	CHRDY	O OCH	CHANNEL READY: This signal is used to insert wait states into system accesses.		
122	AEN	I TTL	DMA ACTIVE: This signal indicates that the system's DMA controller has control of the bus.		
89-92	INT0-3	O 3SH	INTERRUPT REQUEST: The operation of these 4 outputs is determined by the Configuration registers. They can either be used to directly drive the interrupt lines or used as a 3-bit code with a strobe to generate up to 8 interrupts.		
61	DWID	I MOS	DATA WIDTH: This input specifies whether the AT/LANTIC Controller is interfacing to an 8- or 16-bit ISA bus. When high it is in 16-bit mode. It has an internal pull-down resistor.		
93	ISACLK	I TTL	ISA CLOCK: Clock from ISA bus. This signal is only required if CHRDY timing has to be altered, by changing the CHRDY bit of Configuration Register B.		
NETWORK	INTERFACE PINS	,			
TXO+, TXOd- TPI are resistively combined external to the chip to produce a differential output		TWISTED PAIR TRANSMIT OUTPUTS: These high drive CMOS level outputs are resistively combined external to the chip to produce a differential output signal with equalization to compensate for Intersymbol Interference (ISI) on the twisted pair medium.			
150, 151	RXI+,RXI-	I TPI	TWISTED PAIR RECEIVE INPUTS: These inputs feed a differential amplifier which passes valid data to the ENDEC module.		
141, 142	TX- TX+	O AUI	AUI TRANSMIT OUTPUT: Differential driver which sends the encoded data to the transceiver. The outputs are source followers which require 270Ω pull-down resistors.		
145, 146	RX- RX+	I AUI	AUI RECEIVE INPUT: Differential receive input pair from the transceiver.		
147, 148	CD- CD+	I AUI	AUI COLLISION INPUT: Differential collision pair input from the transceiver.		
5	TXLED	O LED	TRANSMIT: An open-drain active low output. It is asserted for approximately 50 ms whenever the AT/LANTIC Controller transmits data in either AUI or TPI modes.		
4	RXLED	O LED	RECEIVE: An open-drain active low output. It is asserted for approximately 50 ms whenever receive data is detected in either AUI or TPI mode.		
3	COLED	O LED	COLLISION: An open-drain active low output. It is asserted for approximately 50 ms whenever the AT/LANTIC Controller detects a collision in either AUI or TPI modes.		
1	GDLNK	O LED	GOOD LINK: An open-drain active low output. This pin operates as an output to display link integrity status if this function has not been disabled by the GDLNK bit in Configuration Register B. This output is off if the AT/LANTIC Controller is in AUI mode or if link testing is enabled and the link integrity is bad (i.e. the twisted pair link has been broken). This output is on if the AT/LANTIC Controller is in Twisted Pair Interface (TPI) mode, link integrity checking is enabled and the link integrity is good (i.e. the twisted pair link has not been broken) or if the link testing is disabled.		
2	POLED	O LED	POLARITY: An open-drain active low output. This signal is normally inactive. When the TPI module detects seven consecutive link pulses or three consecutive received packets with reversed polarity POLED is asserted.		

2.0 P	2.0 Pin Description (Continued)				
Pin No.	Pin Name	Type*	Description		
NETWOR	K INTERFACE P	INS (Contir	nued)		
7	X1 (OSCIN)	I XTAL	CRYSTAL OR EXTERNAL OSCILLATOR INPUT		
8	X2 (OSCOUT)	O XTAL	CRYSTAL FEEDBACK OUTPUT: Used in crystal connections only. Should be left completely unconnected when using an oscillator module.		
10	THIN	O DCDC	THIN CABLE: This output is high if AT/LANTIC Controller is configured for thin cable. It can be used to enable the DC-DC converter required by the thin ethernet configuration.		
EXTERNA	AL MEMORY SU	PPORT			
58-50	MSD0-7, CA0-7, DO, DI, SK	I/O, I, O MOS	MEMORY SUPPORT DATA BUS—CONFIGURATION REGISTER A INPUT EEPROM SIGNALS: MSD0-7: When RESET is inactive these pins can be used to access external memory and boot PROM. CA0-7: When RESET is active Configuration Register A is loaded with the data value on these pins. If the user puts an external pull-up on any of these pins then the corresponding register bit is set to a 1. If the pin is left unconnected then the register bit is 0. DO, DI, SK: When RESET goes from an active to an inactive level AT/LANTIC Controller will read the contents of an EEPROM, using these signals, and load the contents into internal registers. These internal registers will then be mapped into the space taken up by the PROM in the NE2000 and Ethercard PLUS16. After the EEPROM read operation has completed these pins will revert to MSD0-2 (D0 = MSD0, DI = MSD1, SK = MSD2).		
48–45 43-40	MSD8-15 or CB0-7	I/O, I MOS	MEMORY SUPPORT DATA BUS—CONFIGURATION REGISTER B INPUT: MSD8-15: When RESET is inactive these pins can be used to access external memory. CB0-7: When RESET is active Configuration Register B is loaded with the data value on these pins. If the user puts an external pull-up on any of these pins then the corresponding register bit is set to a 1. If the pin is left unconnected then the register bit is 0.		
31–25, 22	MSA1-8 or CC0-7	O, I MOS	MEMORY SUPPORT ADDRESS BUS—CONFIGURATION REGISTER C INPUT: MSA1-8: When RESET is inactive these pins drive the memory support address bus. CC0-7: When RESET is active Configuration Register C is loaded with the data value on these pins. If the user puts an external pull-up on any of these pins then the corresponding register bit is set to a 1. If the pin is left unconnected then the register bit is 0.		
21–15	MSA9-15	O MOS	MEMORY SUPPORT ADDRESS BUS: MSA9–15: When RESET is inactive these pins drive the memory support address bus. When the memory is only 8 bits wide A0 will appear on A13, in compatible mode, and on A15, in non-compatible mode.		
33	MSRD	O MOS	MEMORY SUPPORT BUS READ: Strobes data from the external RAM into the AT/LANTIC Controller via the memory support data bus.		
32	MSWR	O MOS	MEMORY SUPPORT BUS WRITE: Strobes data from the AT/LANTIC Controller into the external RAM via the memory support data bus.		
37	BPCS	O MOS	BOOT PROM CHIP SELECT: Selects the boot PROM on the memory support data bus.		

2.0 Pin	Descripti	2.0 Pin Description (Continued)					
Pin No.	Pin Name	Type*	Description				
EXTERNAL N	MEMORY SUP	PORT (Co					
36	RCS1	O MOS	RAM CHIP SELECT 1: Drives the chip select of the external RAM on the lower half of the memory support data bus.				
34	RCS2	O MOS	RAM CHIP SELECT 2: Drives the chip select of the external RAM on the upper half of the memory support data bus.				
38	EECS	O MOS	EEPROM CHIP SELECT: Strobes data from the EEPROM onto the memory support data bus.				
39	EECONFIG	I TTL	CONFIGURE FROM EEPROM: When this pin is tied high the AT/LANTIC Controller loads the configuration from an EEPROM.				
12	BSCLK	I TTL	INTERNAL BUS CLOCK: This controls the speed of the NIC core if it is not running off of an internal clock (see Configuration Register C). This pin should be tied to ground if it is unused.				
LOW POWER	SUPPORT						
62	LOWPWR	I TTL	LOW POWER: Instructs AT/LANTIC Controller to enter its low power mode, as detailed in Section 4.5. Should be tied to ground for normal operation.				
TEST SUPPO	RT						
			EST: This input is only used for test mode. It should be left unconnected as it has n internal pull-down resistor which will enable correct operation.				
POWER SUP	PLY PINS						
160 PLLV _{CC}			PLL 5V SUPPLY PINS: This pin supplies 5V to the AT/LANTIC's analog PLL inside the ENDEC block. To maximize data recovery it is recommended that analog layout and decoupling rules be applied between this pin and PLLGND.				
159	PLLGND		PLL NEGATIVE (GROUND) SUPPLY PINS				
157, 152, 144, 143, 139	PV _{CC}		PHYSICAL MEDIA 5V SUPPLY PINS: These pins supply 5V to the AT/LANTIC's analog physical media interface circuitry.				
158, 149, 140, 138	PGND		PHYSICAL LAYER NEGATIVE (GROUND) SUPPLY PINS: These pins are the ground to the AT/LANTIC's analog physical media interface circuitry.				
9	OV _{CC}		OSCILLATOR 5V SUPPLY PINS: This pin supplies 5V to the AT/LANTIC's oscillator and LED circuitry.				
6	OGND		OSCILLATOR NEGATIVE (GROUND) SUPPLY PINS: This pin is the ground to the AT/LANTIC's oscillator and LED circuitry.				
59, 49, 24, 13	V _{CC}		POSITIVE 5V SUPPLY PINS: These pins supply power to the AT/LANTIC Controller's logic.				
60, 54, 44, 35, 23, 14	GND		NEGATIVE (GROUND) SUPPLY PINS: These are the supply pins for the AT/LANTIC Controller's logic. It is suggested that decoupling capacitors be connected between the V_{CC} and GND pins. It is essential to provide a path to ground for the GND pins with the lowest possible impedance.				
132, 121, 107, 87, 68	IFV _{CC}		INTERFACE POSITIVE 5V SUPPLY PINS: These pins supply power to the AT/LANTIC Controller's ISA interface.				
135, 129, IFGND 125, 124, 116, 98, 85, 71, 65			INTERFACE NEGATIVE (GROUND) SUPPLY PINS: These are the supply pins for the AT/LANTIC Controller's ISA interface. It is suggested that decoupling capacitors be connected between the IFV $_{\rm CC}$ and IFGND pins. It is essential to provide a path to ground for the IFGND pins with the lowest possible impedance.				

3.0 Simplified Application Diagram



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4.0 Functional Description

The AT/LANTIC Controller is a highly integrated and configurable Ethernet controller making it suitable for most Ethernet applications. The AT/LANTIC Controller integrates the functions of the following blocks:

- DP8390 Ethernet Controller Core and Media Access Control logic.
- ISA Bus Interface containing all logic required to connect the DP8390 core to a packet buffer RAM and the ISA bus.
- Media Interface which includes an Encoder/Decoder block with an AUI (Attachment Unit Interface) and a 10BASE-T Twisted Pair Interface.

4.1 BUS INTERFACE BLOCK

The AT/LANTIC Controller's Bus interface block provides the circuitry to interface the Ethernet controller logic, and the external packet buffer RAM to an ISA (Industry Standard Architecture) Bus. The bus interface provides several configuration modes which offer various different features depending on the designer's specific design requirements. The possible modes are:

- 1. 16-Bit or 8-Bit Shared Memory Compatible Mode
- 2. 16-Bit or 8-Bit Shared Memory Enhanced Mode
- 3. 16-Bit or 8-Bit I/O Port Compatible Mode
- 4. 16-Bit or 8-Bit I/O Port Enhanced Mode

This section describes the function of each of these modes.

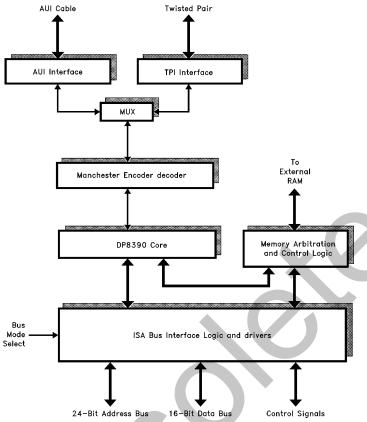


FIGURE 1. Block Diagram of AT/LANTIC Controller

DETERMINING 8- OR 16-BIT WIDE DATA

AT/LANTIC Controller can treat the system data bus and all internal data busses as 8 or 16 bits wide. 8- or 16-bit mode is determined by the DWID pin. For an adapter card this bit can be used to automatically detect if the card has been plugged into an 8- or 16-bit slot. If this pin is connected to a $V_{\mbox{\scriptsize DD}}$ on the upper connector it will be high when plugged into a 16-bit slot, enabling 16-bit mode, and floating when plugged into an 8-bit slot. When floating the internal pull-down resistor will enable 8-bit mode.

SHARED MEMORY ARCHITECTURE

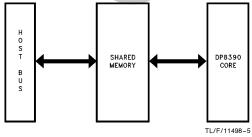


FIGURE 2. Shared Memory

In this mode the AT/LANTIC Controller's internal memory map, using external RAM devices, is mapped into the host system's memory map. Both the AT/LANTIC Controller and the host system can directly access this memory. The AT/LANTIC Controller controls the arbitration for this memory area, giving priority to its internal accesses. It also has an internal FIFO to allow for any latency on internal transfers introduced by system accesses. If a system access occurs while an internal access is current the AT/LANTIC Controller will insert wait states into the system cycle until the internal transfer is complete.

TL/F/11498-4

In this mode the AT/LANTIC Controller's internal registers are accessed within the system's I/O map. The address within this I/O map is set by Configuration Register A. The user programs the address of the shared memory within the host systems memory map by writing to a register in AT/LANTIC Controller. The memory is not accessible by the user until after this register has been programmed.

There are two basic Shared Memory modes, compatible mode, and non-compatible mode, as described in the following text.

Shared Memory Compatible Mode I/O Address Mapping

The shared memory is at an address decided by the Address Decode Register and the base I/O address of AT/LANTIC Controller is configured in Configuration Register A. At that address the following structure appears.

Addr	D7-0	
00	Control 1	
01	AT detect	(Read only)
02	Unused	
03	Unused	
04	Unused	
05	Control 2	
06	Unused	
07	Unused	
08	Node addr 0	(Read only)
09	Node addr 1	(Read only)
0A	Node addr 2	(Read only)
0B	Node addr 3	(Read only)
0C	Node addr 4	(Read only)
0D	Node addr 5	(Read only)
0E	05h	(Read only)
0F	Checksum	(Read only)
10 to	NIC	
1F	registers	
	(a)	

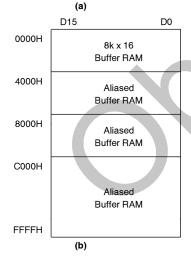


FIGURE 3. Shared Memory Mode a) Register Mapping and b) NIC Core Memory Map

The AT Detect Register indicates whether AT/LANTIC Controller is in an 8- or 16-bit slot. The least significant bit of this register is set high when AT/LANTIC Controller is in 16-bit mode and low in 8-bit mode. Addresses 08H to 10H are

specified as the PROM space for compatibility with the Ethercard PLUS16. This is actually an array of 8-bit registers which are loaded from an external EEPROM after AT/LANTIC Controller is initialized by a reset pulse. The user should program the EEPROM to contain these values.

The 8k words of memory can be accessed directly by the host system in the same manner as any other memory. Typically the programmer would remove data from this buffer using a "MOV" or "MOVSW" instruction.

8-BIT SHARED MEMORY COMPATIBLE MODE

In this mode the I/O map remains the same. The NIC core can still operate in 16-bit mode, if bit 6 of Control Register 2 is set high and the full 16 kbytes of RAM are still available. However, only 8-bit system accesses are allowed. If bit 6 of Control register 2 is low the NIC core must operate in 8-bit mode and only 8k of memory is available. The NIC Core data width is set by the WTS bit in the Data Configuration Register.

A low cost card, using only one 8 kbyte RAM, can be designed. If the DWID pin is left unconnected, or tied to GND, then the AT/LANTIC Controller will always operate in 8-bit mode, regardless of the slot the board is in.

If DWID is low the address bits of Control Register 2 should not be written to as they have no effect. In this mode the address comparator assumes that SA19 is to be compared to a logic high, with the other address comparisons programmed into Control Register 1.

SHARED MEMORY NON-COMPATIBLE MODE

These modes are similar to the compatible mode. The difference is that they map a full 64 kbytes of RAM into the PC's memory address space. The I/O map remains the same.

I/O PORT ARCHITECTURE

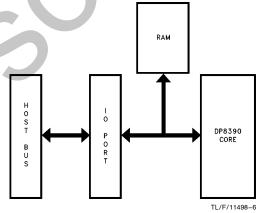


FIGURE 4. I/O Port

This is the architecture used by Novell's NE2000. In this mode the AT/LANTIC Controller's internal memory map is accessed byte or word at a time, via a port within the system's I/O space. AT/LANTIC Controller is programmed by the user to control the transfers between its internal memory and the I/O port.

In this mode the AT/LANTIC Controller's internal registers and the memory access port are accessed within the system's I/O map. The address within this I/O map is set by Configuration Register A.

16-BIT I/O PORT COMPATIBLE MODE I/O ADDRESS MAPPING

This mode is compatible with Novell's NE2000. The base I/O address of the AT/LANTIC Controller is configured by Configuration Register A (either upon power up or by software writing to this register). At that address the following structure appears.

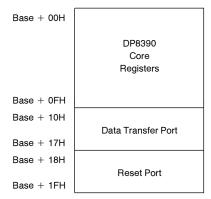
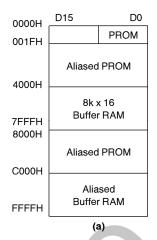


FIGURE 5. I/O Port Mode Register I/O Map

The registers within this area are 8 bits wide, but the data transfer port is 16 bits wide. The AT/LANTIC Controller's registers can be programmed to control the passing of data between its internal memory and the data transfer port. By accessing the data transfer port (using I/O instructions) the user can transfer data to or from the AT/LANTIC Controller's internal memory. The AT/LANTIC Controller's internal memory map is as shown in *Figure 6*.

AT/LANTIC Controller actually has a 64k address range but only does partial decoding on these devices. The PROM data is mirrored at all decodes up to 4000H and the entire map is repeated at 8000H. To access either the PROM or the RAM the user must initiate a Remote DMA transfer between the I/O port and memory.

On a remote read the AT/LANTIC Controller moves data from its internal memory map to the I/O port and the host system reads it by using an "INW" or "INSW" instruction from the I/O address of the data transfer port. If the system attempts to read the port before AT/LANTIC Controller has written the next word of data to it AT/LANTIC Controller will insert wait states into the system cycle, using the CHRDY



	D15	D0
1EH	00	57H
1CH	00	57H
	00	• RESERVED
0AH	00	E'net Address 5
08H	00	E'net Address 4
06H	00	E'net Address 3
04H	00	E'net Address 2
02H	00	E'net Address 1
00H	00	E'net Address 0
		(b)

FIGURE 6. a) NIC Core's Memory Map b) 16-Bit Prom Map

line. AT/LANTIC Controller will not begin the next memory read until the previous word of data has been read.

On a remote write the system writes data to the I/O port, using an "OUTW" or "OUTSW" instruction, and AT/LANTIC Controller moves it to its buffer memory. If the system attempts to write to the port before AT/LANTIC Controller has moved the data to memory AT/LANTIC Controller will insert wait states into the system cycle, using the CHRDY line. AT/LANTIC Controller will not begin the next memory write until a new word has been written to the I/O port.

Addresses 00H to 1FH are specified as the PROM space for compatibility with the NE2000. This is actually an array of 8-bit registers which are loaded from an external EEPROM after AT/LANTIC Controller is initialized by an ISA RESET. They should contain the same data as the PROM did in the NE2000 and in the same format. As can be seen the PROM registers are only 8-bits wide. To transfer the data out the user must initiate a 16-bit DMA read transfer and discard the most significant byte of data on each transfer.

At address 00H of the PROM is a six byte Ethernet address for this node. The upper two addresses of the PROM store contain bytes which identify whether the AT/LANTIC Controller is in 8- or 16-bit mode. For 16-bit mode these bytes both contain the value 57H, for 8-bit mode they both contain 42H

8-BIT I/O PORT COMPATIBLE MODE

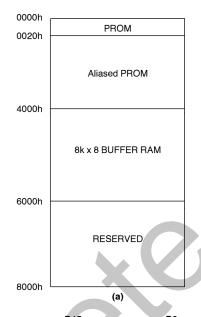
This mode is compatible with the 8-bit mode offered by Novell's NE2000. The NE2000 automatically detects whether it is in an 8- or 16-bit slot and configures itself appropriately. As explained in the previous paragraphs, the user can determine whether the board is in 8- or 16-bit mode by reading the PROM. In 8-bit mode only 8 kbytes of RAM are addressable, as in the 8-bit mode of the NE2000. The I/O map is the same as the 16-bit mode, the memory map is shown in Figure 7. Again the PROM has only a partial decode, so is mirrored at all addresses up to 4000H. The PROM still occupies 32 bytes of address space, although it only has 16 bytes of data, as the data at all odd address locations is merely a mirror of the data at the previous even address location. The RAM is mirrored at 6000H and the entire map mirrored at 8000H.

A low cost card, using only one 8 kbyte RAM, can be designed. If the DWID pin is left unconnected, or tied to GND, then the AT/LANTIC Controller will always operate in 8-bit mode, regardless of the slot the board is in.

I/O PORT NON-COMPATIBLE MODE

This mode is similar to Novell's NE2000, but this mode allows the user to use the full 64 kbytes of address space except for an initial page for the PROM. The memory map for this board is shown in $Figure \ \delta$. The memory map is the same for both 8- and 16-bit modes. Although the PROM store occupies 256 bytes, it is only 16 bytes long. The entire map is mirrored at 8000H.

A low cost card, using only one 8 kbyte RAM, can be designed. If the DWID pin is left unconnected, or tied to GND, then the AT/LANTIC Controller will always operate in 8-bit mode, regardless of the slot the board is in.



	D15	D0
1EH	42H	42H
1CH	42H	42H
	RESERVED	RESERVED
оан	E'net Address 5	E'net Address 5
08H	E'net Address 4	E'net Address 4
06H	E'net Address 3	E'net Address 3
04H	E'net Address 2	E'net Address 2
02H	E'net Address 1	E'net Address 1
00H	E'net Address 0	E'net Address 0
	(I	o)

FIGURE 7. a) 8-Bit NIC Core's Memory Map b) 8-Bit PROM Map

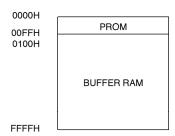


FIGURE 8. I/O Port Enhanced Mode DP8390 Core Memory Map

4.2 POWER ON RESET OPERATION

The AT/LANTIC Controller configures itself after a Reset signal is applied. To be recognized as a valid Power-On-Reset the Reset signal must be active for at least 415 μ s. *Figure 9* shows how the RESET circuitry operates.

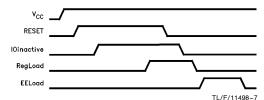


FIGURE 9. RESET Operation

The ISA standard determines that within 500 ns of RESET going active all devices should enter the appropriate reset condition. The AT/LANTIC Controller will generate the internal signal IOinactive after RESET has been active for 415 ns, which will disable all outputs and cause RESET to be the only input monitored. The AT/LANTIC Controller will not respond to a RESET pulse of shorter duration than this. An internal timer continues to monitor the amount of time RESET is active. After 415 μs it is considered a valid Power-On-Reset and an internal signal called RegLoad is generated

When a Power-On-Reset occurs the AT/LANTIC Controller latches in the values on the configuration pins and uses these to configure the internal registers and options. Internally these pins contain pull-down resistors, which are enabled when lOinactive goes active. If any pins are unconnected they default to a logic zero. The internal pull-down resistor has a high resistance to allow the external pull-up resistors to be of a high value. This limits the current taken by the memory support bus. The suggested external resistor value is $10~\mathrm{k}\Omega$. The configuration registers are loaded from the memory support bus when RESET goes inactive if RegLoad is active. The internal pull-down resistors are enabled onto the bus until RegLoad has gone inactive.

A Power-On-Reset also causes the AT/LANTIC Controller to load the internal PROM store from the EEPROM, which can take up to 320 $\mu_{\rm S}$. This occurs after RegLoad has gone inactive. The AT/LANTIC Controller will be inaccessible during this time. If EECONFIG is held high the configuration data loaded on the falling edge of RESET will be overwritten with data read from the serial EEPROM. Regardless of the level on EECONFIG the PROM store will always be loaded with data from the serial EEPROM during the time specified as EELoad.

4.3 EEPROM OPERATION

The AT/LANTIC Controller uses an NM93C06, or EEPROM with compatible timings. The NM93C06 is a 256-bit device, arranged as 16 words each 16 bits wide. The programmed contents of the EEPROM is shown in *Figure 10*.

Mapping EEPROM Into PROM Space

Data is read from the EEPROM at boot time and stored in registers within the AT/LANTIC Controller. While this operation takes place the AT/LANTIC Controller can not be ac-

	D15	D0		
0FH	73H	Config. C		
0EH	Config B	Config. A		
	•	•		
	•	•		
	•	•		
08H	42H	42H		
07H	57H	57H		
	•	•		
	•	•		
	•	•		
03H	Reserved	Reserved		
	(Checksum)	(Board Type)		
02H	E'net Address 5	E'net Address 4		
01H	E'net Address 3	E'net Address 2		
00H	E'net Address 1	E'net Address 0		

Note 1: The contents of locations 03H and 04H differ between I/O Mode and Shared Memory Mode. The Shared Memory Mode values are shown in parentheses. For compatibility with both modes default to the shared memory mode values.

Note 2: Programming 73H into the upper address is not absolutely required but is strongly recommended for future compatibility of manufacturing pro-

FIGURE 10. EEPROM Programming Map

cessed by the system. These registers are mapped into the space traditionally occupied by the PROM in the NE2000 or the EtherCard PLUS16. The size and format of this data read is determined by the mode of operation.

SHARED MEMORY MODE

In this mode, program the EEPROM to contain the node's Ethernet address in the first six bytes, a byte identifying the type of board AT/LANTIC Controller is emulating in byte 7 and a checksum byte in byte 8. The two's complement sum of these eight bytes should equal FFH.

In this Mode the AT/LANTIC Controller reads the first 4 words from the EEPROM and maps them into the I/O map at the appropriate address.

I/O PORT MODE

In this mode, program the EEPROM to contain the node's Ethernet address in the first six bytes. The user should then program 5757H and 4242H into the subsequent bytes. The AT/LANTIC Controller will decide which of these values should be loaded into the PROM store depending on the DWID pin. (The data width is programmed in this mode by setting the WTS bit in the Data Configuration Register and setting the DWID pin for the proper mode.) If some other numerical values are preferred to indicate the mode then they can be programmed at this location in the EEPROM and AT/LANTIC Controller will put them at the correct address.

In this mode the AT/LANTIC Controller reads the first 7 words from the EEPROM and maps them into the memory map at the appropriate address. If in 16-bit mode it also

reads the next word in the EEPROM and appends this. If in 8-bit mode it skips a word, then reads and appends the next word

Storing and Loading Configuration from EEPROM

If the EECONFIG pin is high during boot up the AT/LANTIC Controller's configuration is read from the EEPROM, before the PROM data is read. The configuration data is stored within the upper two words of the EEPROM's address space. Configuration Registers A and B are located in the lower of these words, Register C in the lower byte of the upper word, as shown in Figure 10.

To write this configuration into the EEPROM the user must follow the routine specified in the pseudo code below. This operation will work regardless of the level on EECONFIG. The EELOAD bit of Configuration Register B being set starts the EEPROM write process. Care should be taken not to accidently set the GDLINK bit and therefore disable link integrity checking. The next 3 writes to this register load the values that will be stored in the configuration register (note that the last 2 of these writes do not have to follow the normal practice of preceding a write to this register with a read to this address). The AT/LANTIC Controller will then commence the EEPROM write. The write has been completed when the EELOAD bit goes to zero. This loading procedure should be followed exactly and interrupts should be disabled until it has completed, to prevent any accidental accesses to the AT/LANTIC Controller.

```
EEPROM_LOAD()

{
    DISABLE_INTERRUPTS();
    value = READ(CONFIG_B);
    value = value AND 1 GDLINK;
    value = value OR EELOAD;
    WRITE(CONFIG_B, value);
    READ(CONFIG_B);
    WRITE(CONFIG_B, config_for_A);
    WRITE(CONFIG_B, config_for_C);
    WRITE(CONFIG_B, config_for_C);
    while (value AND EELOAD)
    {
        value = READ(CONFIG_B);
        WAIT();
    }
    ENABLE_INTERRUPTS();
}
```

4.4 JUMPERED AND JUMPERLESS OPERATION SUPPORT

The AT/LANTIC Controller supports several options that enable the implementation of either a "jumpered" or "jumperless" power on configuration when installed into a standard PC compatible's ISA bus. A wide range of options are provided to ensure that the AT/LANTIC Controller can be configured by an end user to function in all possible PC-AT system configurations. Several types of configuration options can be implemented examples including:

Full jumper options: All programmable options are selected by utilizing jumpers on the board. Option selection requires no special software. An example of this is shown in the Figure 11.

- I/O address jumpers only: All other options configurable via software. This option simplifies installation while maximizing compatibility.
- 3. Jumperless: Special scheme provides contention-free I/O address selection.

The AT/LANTIC Controller's Configuration Registers are the key to providing the ability to implement various configuration options. These registers are configured by the same method in shared memory and I/O port modes, 8- or 16-bit modes. The bit definitions of these registers are provided in Section 5. All three registers are configured by hardware selection during the Power-On-Reset of the system. Two of these registers can be configured via software (the Mode Configuration Registers A and B). The third register (Hardware Configuration Register C) is only configured during reset

The following table indicates *most* of the AT/LANTIC Controller options that a designer may like to have user configurable. (This list does not represent the complete list. For the full list see the Configuration register descriptions in Section 5.)

TABLE I. Some Configuration Options for AT/LANTIC Controller

Option	Selections			
I/O Base Address	0300H Software 0240H 0280H	02C0H 0320H 0340H 0360H		
Interrupt No.	4 Interrupts	8 Interrupts		
Boot PROM Address	Disabled 0C000H 0C400H 0C800H	0CC00H 0D000H 0D400H 0D800H 0DC00H		
Boot PROM Size	None 16k	32k 64k		
Media Selection	Twisted Pair AUI Port	Thin Ethernet		
Architecture	I/O Mode	Shared RAM Mode		
Bus Timing Options	IOCHRDY Mode	MEM16 Mode		

The three basic options are described below. Because of the variety of programmable options there are a number of variations possible, only a few typical examples will be discussed.

FULLY JUMPERED OPERATION

This option is shown in Figure 11. In this configuration most options are selected by jumpers on the AT/LANTIC Controller's memory bus. For this option all configuration options are set upon power-on by the AT/LANTIC Controller as described in Section 4.2. Accessing the configuration registers is unnecessary and the EEPROM need only contain the

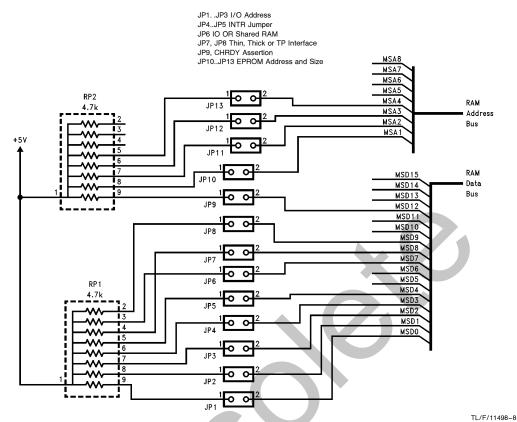


FIGURE 11. Example of Jumper Configuration

Ethernet ID address (Configuration Register B bit 7 should be set to disable EEPROM configuration mode, and Configuration Register C bit 7 could be set to disable software configuration completely).

MINIMAL JUMPERS

The AT/LANTIC Controller's configuration registers provide the capability to enable software to configure various options (some may be hardwired). For the one option that is not easily configured on the ISA bus is the I/O address options. The reason for this is that the I/O locations must first be known by the software in order for the software (usually a device driver) to access the AT/LANTIC Controller. However, upon power up, in order to access a register to configure the I/O base address to avoid conflicts some default location must be given (typically set in hardware on the memory bus). It is possible that this default location conflicts with an already installed device. If this is the case then one possible solution, is to provide a jumper option for only the I/O Addresses. A similar situation exists for the boot PROM memory addresses.

In this application all options except the I/O address and the boot PROM are hardwired on the memory bus to a default setting. After power up software can change the con-

figuration to avoid conflicts on these settings. The advantage of this approach is that for most systems the default I/O address setting is the correct one and no installation will be required in this case. This approach minimizes any compatibility issues.

NO JUMPERS

The conflicts possible in the I/O base selection can be overcome by a special mode for software configuration of the I/O base address. By using this mode, and by using the configuration storage capability of the EEPROM a fully software configurable design on the ISA bus can be realized without address conflict problems.

This mode is invoked by having the AT/LANTIC Controller default to jumperless software configuration option in the I/O base selection. This mode enables configuration register A to be mapped to address location 278H which is defined to be a printer port's data register. If software writes to this location four consecutive times on the fourth write the AT/LANTIC Controller will load the data written into the I/O address bits of Configuration Register A. This data should set the I/O base address to a known conflict-free value. The AT/LANTIC Controller can now be configured and operated

at the desired base I/O address. If desired the configuration software could change the EEPROM content to the new values eliminating the need to reconfigure upon each power up. Alternately the software could leave the EEPROM alone and execute the configuration using the printer port's data register upon each power up. This configuration scheme will only work once after each power-up. Therefore the user cannot enable the AT/LANTIC Controller from reserved mode, change it back into reserved mode, and enable it again. A power-on reset must occur between the first time it is enabled from the reserved mode and the second.

A second consideration is the location of the boot PROM in the system memory map, which also has the same conflict and programming considerations as the I/O address selection. However the solution is different, primarily because the boot PROM must be configured before power up. This is because during normal usage of the boot PROM the PC's BIOS will look for the ROM immediately after reset, not allowing configuration software to first select the boot PROM addressing prior to usage.

To configure the boot PROM without jumpers the configuration software must first power up the AT/LANTIC Controller, configure the EEPROM to the desired location, then hardware reset the AT/LANTIC Controller. After the reset the AT/LANTIC Controller's EEPROM will load in the desired boot PROM configuration automatically during the reset. Now after reset when the PC scans for the boot PROM, the ROM will be correctly mapped in the memory space enabling the network boot operation to proceed.

Ethernet Cable Configuration

AT/LANTIC Controller offers the choice of all the possible Ethernet cabling options, that is Ethernet (10BASE5), Thin Ethernet (10BASE2) and Twisted-pair Ethernet (10BASE-T). The type of cabling used is controlled by Configuration Register B. AT/LANTIC Controller also supplies a THIN output signal which can be used to disable/enable an external DC-DC converter which is required for 10BASE2.

4.5 LOW POWER OPERATION

The AT/LANTIC Controller has a low power support mode that can be used to disable the Ethernet port and conserve power. It should be noted that the device is not operational in this mode and requires to be initialized after exiting this mode.

The power and ground pins to the AT/LANTIC Controller are split up into two groups, interface and core. By switching the power off to the core logic while still powering the interface logic the AT/LANTIC Controller can be powered down without crashing the ISA bus. The LOWPWR pin should be driven high to indicate that the device is about to go into low power then the power to the V_{DD} pins should be switched off. The same signal that is used to drive the LOWPWR pin can be used to drive a p-channel load switch to disable power to the core. This switch must have a very low on resistance to minimize the voltage difference between the V_{CC} and the IFV $_{CC}$. All devices on the memory support bus should also be powered from the V_{CC} supply.

4.6 BOOT PROM OPERATION

The AT/LANTIC Controller supports an optional boot PROM, the address and size of which can be set in Configuration Register C. This boot PROM can be any 8 bits wide storage device implemented with a non-volatile technology. Write cycles to this device can be enabled and disabled by programming Configuration Register B. This can be used to prevent unwanted write cycles to certain devices, such as a Flash EEPROM. It should be noted that the address pins for the boot PROM should be connected directly to the ISA bus. The AT/LANTIC Controller supplies the chip select to the device and buffers the data onto and from the ISA bus, so the memory support data bus should be connected to the boot PROM's data pins.

4.7 DP8390 CORE (NETWORK INTERFACE CONTROLLER)

The DP8390 Core logic, *Figure 12*, contains the Serializer/Deserializer which is controlled by the Protocol PLA, DMA Control, FIFO, Address Comparator, Multicast Hashing Register. The DP8390 core implements all of the IEEE 802.3 Media access control functions for the AT/LANTIC Controller, and interfaces to the internal ENDEC (on the left of the block diagram) and also interfaces to the Bus Interface and memory support bus via a number of address, data and control signal (and the right side of the block diagram). The following sections describe the functions of the DP8390 core

Receive Deserializer

The Receive Deserializer is activated when the input signal Carrier Sense is asserted to allow incoming bits to be shifted into the shift register by the receive clock. The serial receive data is also routed to the CRC generator/checker. The Receive Deserializer includes a synch detector which detects the SFD (Start of Frame Delimiter) to establish where byte boundaries within the serial bit stream are located. After every eight receive clocks, the byte wide data is transferred to the 16-byte FIFO and the Receive Byte Count is incremented. The first six bytes after the SFD are checked for valid comparison by the Address Recognition Logic. If the Address Recognition Logic does not recognize the packet, the FIFO is cleared.

CRC Generator/Checker

During transmission, the CRC logic generates a local CRC field for the transmitted bit sequence. The CRC encodes all fields after the synch byte. The CRC is shifted out MSB first following the last transmit byte. During reception the CRC logic generates a CRC field from the incoming packet. This local CRC is serially compared to the incoming CRC appended to the end of the packet by the transmitting node. If the local and received CRC match, a specific pattern will be generated and decoded to indicate no data errors. Transmission errors result in different patterns and are detected, resulting in rejection of a packet.

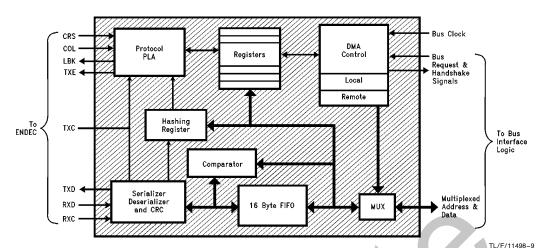


FIGURE 12. DP8390 Controller Core Simplified Block Diagram

Transmit Serializer

The Transmit Serializer reads parallel data from the FIFO and serializes it for transmission. The serializer is clocked by the transmit clock generated internally. The serial data is also shifted into the CRC generator/checker. At the beginning of each transmission, the Preamble and Synch Generator append 62 bits of 1,0 preamble and a 1,1 synch pattern. After the last data byte of the packet has been serialized the 32-bit FCS (Frame Check Sequence) field is shifted directly out of the CRC generator. In the event of a collision the Preamble and Synch generator is used to generate a 32-bit JAM pattern of all 1's.

Comparator-address Recognition Logic

The address recognition logic compares the Destination Address Field (first 6 bytes of the received packet) to the Physical address registers stored in the Address Register Array. If any one of the six bytes does not match the pre-programmed physical address, the Protocol Control Logic rejects the packet. All multicast destination addresses are filtered using a hashing technique. (See register description.) If the multicast address indexes a bit that has been set in the filter bit array of the Multicast Address Register Array the packet is accepted, otherwise it is rejected by the Protocol Control Logic. Each destination address is also checked for all 1's which is the reserved broadcast address.

FIFO and Packet Data Operations

OVERVIEW

To accommodate the different rates at which data comes from (or goes to) the network and goes to (or comes from) the packet buffer memory, the AT/LANTIC Controller contains a 16-byte FIFO for buffering data between the media and the buffer RAM located on the memory support bus. The FIFO threshold is programmable, allowing filling (or emptying) the FIFO at different rates. When the FIFO has filled to its programmed threshold, the local DMA channel transfers these bytes (or words) into local memory (via the

memory bus). It is crucial that the local DMA is given access to the bus within a minimum bus latency time; otherwise a FIFO underrun (or overrun) occurs.

FIFO underruns or overruns are caused when a local DMA request is issued while an ISA bus access is current and the ISA cycle takes longer to complete than the local DMA's tolerable latency. This tolerable latency depends on the FIFO threshold, whether it is in byte or word wide mode and the speed of the DMA clock (BSCLK frequency). Note that this refers to standard ISA cycles NOT those where the CHRDY is deasserted extending the cycle.

FIFO THRESHOLD DETECTION

To assure that there is no overwriting of data in the FIFO, the FIFO logic flags a FIFO overrun as the 13th byte is written into the FIFO, effectively shortening the FIFO to 13 bytes. The FIFO logic also operates differently in Byte Mode and in Word Mode. In Byte Mode, a threshold is indicated when the n+1 byte has entered the FIFO; thus, with an 8 byte threshold, the AT/LANTIC Controller issues a request to the buffer RAM when the 9th byte has entered the FIFO, making the effective threshold 9 bytes. For Word Mode, the request is not generated until the n+2 bytes have entered the FIFO. Thus, with a 4 word threshold (equivalent to 8 byte threshold), a request to the buffer RAM is issued when the 10th byte has entered the FIFO, making the effective threshold 10 bytes.

TOLERABLE LATENCY CALCULATION

To prevent a FIFO **overrun** a byte (or word) of data must be **removed** from the FIFO before the 13th byte is written. Therefore the worst case tolerable latency is the time from the effective threshold being reached to the time the 13th byte is written minus the time taken to load the first byte (or word) of data to the FIFO during a local DMA burst (8 BSCLKs).

tolerable latency = ((overrun - effective) threshold

- × time to transfer byte on network)
- time to fill 1st FIFO location

For the case of a 4 word threshold using a 20 MHz BSCLK: tolerable latency = ((13 - 10) imes 800) - (8 imes 50) ns = 2 μ s

To prevent a FIFO **underrun** a byte (or word) of data must be **added** from the FIFO before the last byte is removed. Therefore the worst case tolerable latency is the time from the effective threshold being reached to the time the last byte is removed minus the time taken to load the first byte (or word) of data to the FIFO during a local DMA burst (8 BSCLKs).

tolerable latency = (threshold

- × time to transfer byte on network)
- time to fill 1st FIFO location

For the case of a 4 word threshold using a 20 MHz BSCLK: tolerable latency = $(4 \times 800) - (8 \times 50)$ ns

tolerable latency = (4 \times 800) - (8 \times 50) ns = 2.8 μ s

The worst case latency, either overrun or underrun, ultimately limits the overall latency that the AT/LANTIC Controller can tolerate. If the standard ISA cycles are shorter than the worst case latency then no FIFO overruns or underruns will occur.

BEGINNING OF RECEIVE

At the beginning or reception, the AT/LANTIC Controller stores entire Address field of each incoming packet in the FIFO to determine whether the packet matches its Physical Address Registers or maps to one of its Multicast Registers. This causes the FIFO to accumulate 8 bytes.

Furthermore, there are some synchronization delays in the DMA PLA. Thus, the actual time that a request to access the buffer RAM is asserted from the time the Start of Frame Delimiter (SFD) is detected is 7.8 μs . This operation affects the bus latencies at 2 byte and 4 byte thresholds during the first receive request since the FIFO must be filled to 8 bytes (or 4 words) before issuing a request to the buffer RAM.

END OF RECEIVE

When the end of a packet is detected by the ENDEC module, the AT/LANTIC Controller enters its end of packet processing sequence, emptying its FIFO and writing the status information at the beginning of the packet. The AT/LANTIC Controller holds onto the memory bus for the entire sequence. The longest time that local DMA will hold the buffer RAM occurs when a packet ends just as the AT/LANTIC Controller performs its last FIFO burst. The AT/LANTIC Controller, in this case, performs a programmed burst transfer followed by flushing the remaining bytes in the FIFO, and completed by writing the header information to the buffer memory. The following steps occur during this sequence.

- 1. AT/LANTIC Controller issues request to access the RAM because the FIFO threshold has been reached.
- During the burst, packet ends, resulting in the request being extended.
- AT/LANTIC Controller flushes remaining bytes from FIFO.
- AT/LANTIC Controller performs internal processing to prepare for writing the header.
- 5. AT/LANTIC Controller writes 4-byte (2-word) header
- AT/LANTIC Controller de-asserts access to the buffer RAM.

BEGINNING OF TRANSMIT

Before transmitting, the AT/LANTIC Controller performs a prefetch from memory to load the FIFO. The number of bytes prefetched is the programmed FIFO threshold. The next request to the buffer RAM is not issued until after the AT/LANTIC Controller actually begins transmitting data, i.e., after SFD.

READING THE FIFO

If the FIFO is read during normal operation the AT/LANTIC Controller will "hang" the ISA bus by deasserting CHRDY and never asserting it. The FIFO should only be read during loopback diagnostics, when it will operate normally.

PROTOCOL PLA

The Protocol PLA is responsible for implementing the IEEE 802.3 protocol, including collision recovery with random backoff. The Protocol PLA also formats packets during transmission and strips preamble and synch during reception

DMA AND BUFFER CONTROL LOGIC

The DMA and Buffer Control Logic is used to control two 16-bit DMA channels. During reception, the Local DMA stores packets in a receive buffer ring, located in buffer memory. During transmission the Local DMA uses programmed pointer and length registers to transfer a packet from local buffer memory to the FIFO.

A second DMA channel is used when the AT/LANTIC Controller is used in I/O Port mode. This DMA is used as a slave DMA to transfer data between the local buffer memory and the host system. The Local DMA and Remote DMA are internally arbitrated, with the Local DMA channel having highest priority. Both DMA channels use a common external bus clock to generate all required bus timing. External arbitration is performed with a standard bus request, bus acknowledge handshake protocol.

In the shared memory mode the Remote DMA is not used, because in this mode the system has direct read/write access to the buffer RAM.

4.8 TWISTED PAIR INTERFACE MODULE

The TPI consists of five main logical functions:

- a) The Receiver/Smart Squelch, responsible for determining when valid data is present on the differential receive inputs (RXI±) and receiving the data.
- b) The Collision function checks for simultaneous transmission and reception of data on the TXO \pm and RXI \pm pins.
- c) The Link Detector/Generator checks the integrity of the cable connecting the two twisted pair MAUs.
- d) The Jabber disables the transmitter if it attempts to transmit a longer than legal packet.
- e) The TX Driver and Pre-emphasis transmits Manchester encoded data to the twisted pair network via the summing resistors and transformer/filter.

Receiver and Smart Squelch

The AT/LANTIC Controller implements an intelligent receive squelch on the RXI \pm differential inputs to ensure that impulse noise on the receive inputs will not be mistaken for a valid signal.

The squelch circuitry employs a combination of amplitude and timing measurements to determine the validity of data on the twisted pair inputs. There are two voltage level options for the smart squelch. One mode, 10BASE-T mode, uses levels that meet the 10BASE-T specification. The second mode, reduced squelch mode, uses a lower squelch threshold level, and can be used in longer cable applications where smaller signal levels may be applied. The squelch level mode can be selected in the AT/LANTIC Controller configuration registers.

Figure 14 shows the operation of the smart squelch in 10BASE-T mode.

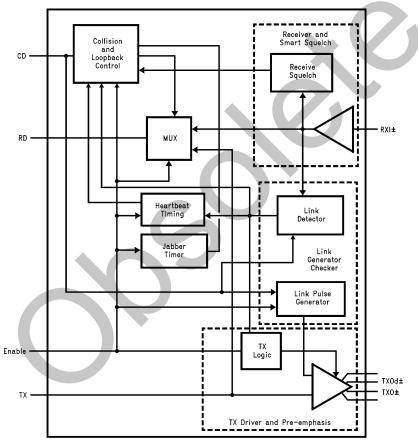


FIGURE 13. Twisted Pair Interface Module Block Diagram

TL/F/11498-10

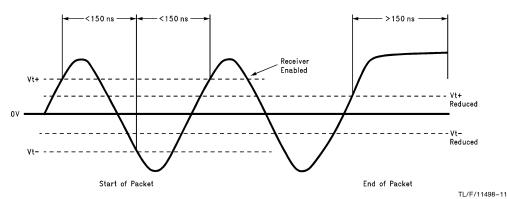


FIGURE 14. Twisted Pair Squelch Waveform

The signal at the start of packet is checked by the smart squelch and any pulses not exceeding the squelch level (either positive or negative, depending upon polarity) will be rejected. Once this first squelch level is overcome correctly the opposite squelch level must then be exceeded within 150 ns later. Finally the signal must exceed the original squelch level within a further 150 ns to ensure that the input waveform will not be rejected. The checking procedure results in the loss of typically three bits at the beginning of each packet.

Only after all these conditions have been satisfied will a control signal be generated to indicate to the remainder of the circuitry that valid data is present. At this time the smart squelch circuitry is reset.

In the reduced squelch mode the operation is identical except that the lower squelch levels shown in *Figure 14* are used.

Valid data is considered to be present until either squelch level has not been generated for a time longer than 150 ns, indicating End of Packet. Once good data has been detected the squelch levels are reduced to minimize the effect of noise causing premature End of Packet detection.

Collision

A collision is detected by the TPI module when the receive and transmit channels are active simultaneously. If the TPI is receiving when a collision is detected it is reported to the controller immediately. If, however, the TPI is transmitting when a collision is detected the collision is not reported until seven bits have been received while in the collision state. This prevents a collision being reported incorrectly due to noise on the network. The signal to the controller remains for the duration of the collision.

Approximately 1 μs after the transmission of each packet a signal called the Signal Quality Error (SQE) consisting of typically 10 cycles of 10 MHz is generated. This 10 MHz signal, also called the Heartbeat, ensures the continued functioning of the collision circuitry.

Link Detector/Generator

The link generator is a timer circuit that generates a link pulse as defined by the 10 Base-T specification that will be generated by the transmitter section. The pulse which is 100 ns wide is transmitted on the TXO+ output, every 16 ms, in the absence of transmit data.

The pulse is used to check the integrity of the connection to the remote MAU. The link detection circuit checks for valid pulses from the remote MAU and if valid link pulses are not received the link detector will disable the transmit, receive and collision detection functions.

The $\overline{\text{GDLNK}}$ output can directly drive a LED to show that there is a good twisted pair link. For normal conditions the LED will be on. The link integrity function can be disabled by setting the $\overline{\text{GDLNK}}$ bit of Configuration Register B.

Jabber

The jabber timer monitors the transmitter and disables the transmission if the transmitter is active for greater than 26 ms. The transmitter is then disabled for the whole time that the Endec module's internal transmit enable is asserted. This signal has to be deasserted for approximately 750 ms (the unjab time) before the Jabber re-enables the transmit outputs.

Transmitter

The transmitter consists of four signals, the true and compliment Manchester encoded data (TXO \pm) and these signals delayed by 50 ns (TXOd \pm)

These four signals are resistively combined TXO+ with TXOd- and TXO- with TXOd+. This is known as digital pre-emphasis and is required to compensate for the twisted pair cable which acts like a low pass filter causing greater attenuation to the 10 MHz (50 ns) pulses of the Manchester encoded waveform than the 5 MHz (100 ns) pulses.

An example of how these signals are combined is shown in the following diagram.



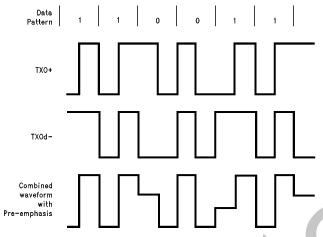


FIGURE 15. Typical Summed Transmit Waveform

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The signal with pre-emphasis shown above is generated by resistively combining TXO+ and TXOd-. This signal along with its compliment is passed to the transmit filter.

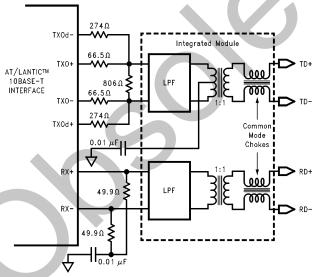
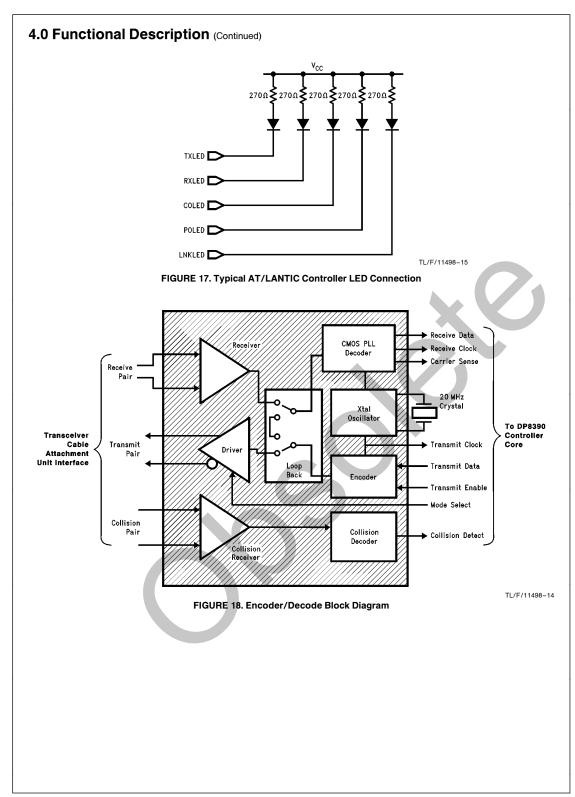


FIGURE 16. External Circuitry to Connect AT/LANTIC Controller to Twisted Pair Cable



Status Information

Status information is provided by the AT/LANTIC Controller on the RXLED, $\overline{\text{TXLED}}$, $\overline{\text{COLED}}$ and $\overline{\text{POL}}$ outputs as described in the pin description table. These outputs are suitable for driving status LED's as shown in *Figure 17*. All outputs are open drain.

Recommended integrated Filter-Transformer-choke modules:

- 1. Pulse Engineering PE65424
- 2. Valor FL1012 or FL1030.

4.9 ENCODER/DECODER (ENDEC) MODULE

The ENDEC consists of four main logical blocks:

- a. The oscillator generates the 10 MHz transmit clock signal for system timing.
- b. The Manchester encoder accepts NRZ data from the controller, encodes the data to Manchester, and transmits the data differentially to the transceiver, through the differential transmit driver.
- c. The Manchester decoder receives Manchester data from the transceiver, converts it to NRZ data and clock pulses, and sends it to the controller.
- d. The collision translator indicates to the controller the presence of a valid 10 MHz collision signal to the PLL.

Oscillator

The oscillator is controlled by a 20 MHz parallel resonant crystal connected between X1 and X2 or by an external clock on X1. The 20 MHz output of the oscillator is divided by 2 to generate the 10 MHz transmit clock for the controller. The oscillator also provides internal clock signals to the encoding and decoding circuits.

CRYSTAL OPERATION

If the crystal used with the internal oscillator circuit is not properly selected, the AT/LANTIC Controller oscillator may not reliably start oscillation under all conditions.

If this occurs, it could be deceiving to a designer, since his prototypes may work fine. However, when the designer does qualification testing or starts production, he may encounter a higher than expected board yield loss due to the oscillator not starting. The AT/LANTIC Controller's oscillator circuit clocks the Encoder-Decoder logic. The AT/LANTIC Controller's oscillator also clocks the twisted pair interface block. If the oscillator does not start, the AT/LANTIC Controller will not be able to transmit or receive.

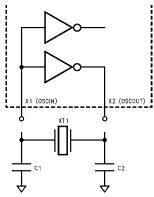
If a crystal is connected to the AT/LANTIC Controller, it is recommended that the circuit shown in *Figure 19* be used and that the components used meet the following:

Crystal XT1: AT cut parallel resonant crystal

Series Resistance: ≤25Ω
Specified Load Capacitance: ≤20 pF
Accuracy: 0.005% (50 ppm)
Typical Load: 50 μW-75 μW

The recommended values for capacitors C1 and C2 are 26 pF minus the board capacitance on that pin. Therefore if both X1 and X2 have 4 pF of board capacitance, then a 22 pF capacitor should be used.

According to the IEEE 802.3 standard, the entire oscillator circuit (crystal and amplifier) must be accurate to 0.01%. When using a crystal, the X2 pin is not guaranteed to provide a TTL compatible logic output, and should not be used



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FIGURE 19. Crystal Connection to AT/LANTIC Controller (see text for component values)

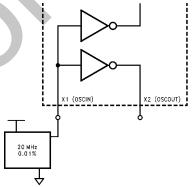
to drive external standard logic. If additional logic needs to be driven, then an external oscillator should be used, as described in the following section.

OSCILLATOR MODULE OPERATION

If the designer wishes to use a crystal clock oscillator, one that provides the following should be employed:

1. TTL or CMOS output with a 0.01% frequency tolerance 2. 40%-60% duty cycle

The circuit is shown in *Figure 20*. When using a clock oscillator it is recommended that the designer connect the oscillator output to the X1 pin and leave the X2 pin floating.



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FIGURE 20. AT/LANTIC Controller Connection for Oscillator Module

Manchester Encoder and Differential Driver

The differential transmit pair, on the secondary of the employed transformer, drives up to 50 meters of twisted pair AUI cable. These outputs are source followers which require two 270Ω pull-down resistors to ground (see *Figure 21*).

The AT/LANTIC Controller allows full-step to be compatible with IEEE 802.3. Transmit + and Transmit - are equal in the idle state, providing zero differential voltage to operate with transformer coupled loads.

Manchester Decoder

The decoder consists of a differential receiver and a PLL to separate a Manchester encoded data stream into internal

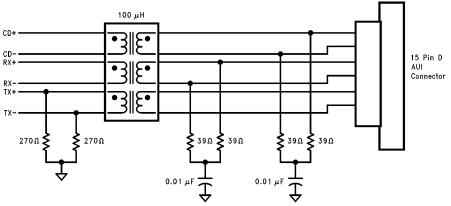


FIGURE 21. Connection from AT/LANTIC Controller's AUI Port to the AUI Connector

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clock signals and data. The differential input must be externally terminated with two 39Ω resistors connected in series if the standard 78Ω transceiver drop cable is used, in thin Ethernet applications, these resistors are optional. To prevent noise from falsely triggering the decoder, a squelch circuit at the input rejects signals with levels less than -175~mV. Data becomes valid typically within 6 bit times. The AT/LANTIC Controller may tolerate bit jitter up to 20 ns in the received data. The decoder detects the end of a frame when no more mid-bit transitions are detected.

Collision Translator

When in AUI Mode, the Ethernet transceiver (DP8392 CTI) detects a collision, it generates a 10 MHz signal to the differential collision inputs (CD±) of the AT/LANTIC Controller. When these inputs are detected active, the AT/LANTIC Controller uses this signal to back off its current transmission and reschedule another one.

In this mode the $\overline{\text{COLED}}$ output will indicate when the CD \pm lines are active during activity on the network. This means it will correctly indicate any collision on the network, but will not be lit for heartbeat or if there is no cable connected.

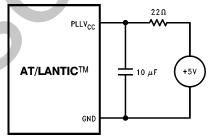
The collision differential inputs are terminated the same way as the differential receive inputs. The squelch circuitry is also similar, rejecting pulse levels less than -175 mV.

PLL V_{CC} Power Supply Consideration

The PLL V_{CC} pin is the +5V power supply for the phase lock loop (PLL) of the ST-NIC ENDEC unit. Since this is an

analog circuit, excessive noise on the PLL V_{CC} pin can affect the performance of the PLL. This noise, if in the 10 kHz–400 kHz range, can reduce the jitter performance of the ENDEC, resulting in missing packets or CRC errors.

If the power supply noise is causing significant packet reception error, a low pass filter could be added to reduce the power supply noise and hence improve the jitter performance. Standard analog design techniques should be utilized when laying out the power supply traces on the board. If the digital power supply is used, it may be desirable to add a one pole RC filter (designed to have a cut-off frequency of 1 kHz) as shown in *Figure 4* to improve the jitter performance. The PLL V_{CC} only draws 3 mA-4 mA so the voltage across the resister is less than 90 mV, which will not affect the PLL's operation.



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FIGURE 22. Filtering Power Supply Noise

5.0 Register Descriptions

5.1 CONFIGURATION REGISTERS

These registers are used to configure the operation of the AT/LANTIC Controller typically after power up. These registers control the configuration of bus interface, setting options like interrupt selection, I/O base address, and other specific modes.

MODE CONFIGURATION REGISTER A

To prevent any accidental writes of this register it is "hidden" behind a previously unused register. Register 0AH in the AT/LANTIC Controller's Page 0 of registers was previously reserved on a read. Now Configuration Register A can be read at that address and can be written to by following a read to 0AH with a write to 0AH. If any other AT/LANTIC Controller register accesses take place between the read and the write then the write to 0AH will access the Remote Byte Count Register 0.

7	6	5	4	3	2	1	0	
MEMIO	FREAD	INT2	INT1	INT0	IOAD2	IOAD1	IOAD0	

Bits	Symbols	Function					
0-2	IOAD0- IOAD2	I/O ADDRESS: These three bits determine the base I/O address of the AT/LANTIC Controller, within the system's I/O map. The AT/LANTIC Controller occupies 20H bytes of the system's address space. 0 0 0 0300H 0 0 1 Software (Note 1) 0 1 0 0240H 0 1 1 0280H 1 0 0 02C0H 1 0 1 0320H 1 1 0 0340H 1 1 1 0360H Note 1: When 001 is selected the AT/LANTIC controller will not respond to any I/O Addresses, but will allow 4 consecutive writes to 278H to write these three bits of this register. This sequence will only operate once after a power-on reset. This mode allows the AT/LANTIC Controller to be configured via software without conflicting with other peripherals.					
3-5	INTO- INT2	INTERRUPT LINE USED: There are two interrupt modes which can be enabled by setting bit 5 of Configuration Register C to the appropriate level. DIRECT DRIVE MODE: In this mode an interrupt output pin will be driven active on a valid interrupt condition. Only one pin may be driven, the other three will remain at TRI-STATE®. The pin driven is determined by the value in this register. Bit 5 Bit 4 Bit 3 Interrupt X 0 0 INT0 X 0 1 INT1 X 1 0 INT2 X 1 1 INT3 CODED OUTPUT MODE: In this mode INT3 is the active interrupt output while pins INT0 to INT2 are programmable outputs reflecting the values on bits 3 to 5.					
6	FREAD	FAST READ: When this bit is set high the AT/LANTIC Controller, in I/O mode, will begin the next port fetch before the current IORD has completed. In slow ISA systems this may cause the data in the port to be overwritten before the ISA cycle has been completed.					
7	MEMIO	MEMORY OR I/O MODE: If this bit is set high then the AT/LANTIC Controller is in shared memory mode. If it is set low it is in I/O mode.					

Mode Configuration Register B

To prevent any accidental writes of this register it is "hidden" behind a previously unused register. Register 0BH in the AT/LANTIC Controller's Page 0 of registers was previously reserved on a read. Now Configuration Register B can be read at that address and can be written to by following a read to 0BH with a write to 0BH. If any other AT/LANTIC Controller register accesses take place between the read and the write then the write to 0BH will access the Remote Byte Count Register 1. Care should be taken when writing to this register as GDLINK and BE are not simple read/write bits, e.g., the user cannot change the physical layer by reading B, or-ing the returned value with the bits to be set, and writing this value to B. This could inadvertently disable link integrity generation and clear a bus error indication before it was noted.

7	6	5	4	3	2	1	0
EELOAD	BPWR	BE	CHRDY	IO16CON	GDLINK	PHYS1	PHYS0

Bits	Symbols	Function			
0-1	PHYS0- PHYS1	PHYSICAL LAYER INTERFACE: These 2 bits determine which type of physical interface the AT/LANTIC Controller is using. The 2 TPI interfaces use twisted pair outputs and inputs, while the other 2 interfaces use the AUI outputs and inputs. In 10BASE5 mode the THIN output pin is driven low, in 10BASE2 mode it is driven high. This can be used to enable the DC-DC converter required by the 10BASE2 specification to provide electrical isolation. The Non spec TPI mode is a twisted pair mode with reduced receive squelch levels. This allows the use of longer cable lengths than specified in the twisted pair specification, or use of cable with higher losses. 0 0 TPI (10BASE-T Compatible Squelch Level) 1 Thin Ethernet (10BASE2) 1 0 Thick Ethernet (10BASE5) (AUI Port) 1 TPI (Reduced Squelch Level)			
2	GDLNK	GOOD LINK: When a 1 is written to this bit the link test pulse generation and integrity checking is disabled. When this bit is read it will indicate link status, reflecting the value shown on the LED output. It is 0 if the AT/LANTIC Controller is in AUI mode or if link testing is enabled and the link integrity is bad (i.e., the twisted pair link has been broken). It is 1 if the AT/LANTIC Controller is in TPI mode, link integrity checking is enabled and the link integrity is good (i.e., the twisted pair link has not been broken) or if the link testing is disabled.			
3	IO16CON	IO16 CONTROL: When this bit is set high the AT/LANTIC Controller generates IO16 after IORD or IOWR go active. If low this output is generated only on address decode.			
4	CHRDY	CHRDY FROM IORD OR IOWR OR FROM BALE: When this bit is low the AT/LANTIC Controller will generate CHARDY after the command strobe. When high it will generate it after BALE goes high.			
5	BE	BUS ERROR: This bit shows that the AT/LANTIC Controller has detected a bus error condition. This will go high if the AT/LANTIC Controller attempts to insert wait states into a system access and the system terminates the cycle without inserting the wait states. Writing a one to this bit clears it to zero. Writing a zero has no effect.			
_ 6	BPWR	BOOT PROM WRITE: When this bit is low no write cycles are generated to the boot PROM.			
7	EELOAD	EEPROM LOAD: Writing a 1 to this bit enables the EEPROM load algorithm as detailed in Section 4. This bit should not be configured to be high, either from switches or an EEPROM.			

Hardware Configuration Register C

This register is configured during a RESET and can not be accessed by software.

7	6	5	4	3	2	1	0
SOFEN	CLKSEL	INTMOD	COMP	BPS3	BPS2	BPS1	BPS0

Bits	Symbols						Function	
0-3	BPS0-3	within	the sele	cted me	emory a	rea AT/LAN	at which boot PROM begins and the size. When the system reads ITIC Controller reads the data in through MSD0-7 and drives it onto lid addresses and sizes:	
		Bit 3	Bit 2	Bit 1	Bit 0	Address	Size (I/O / Shared Mem.)	
		0	0	0	X	Χ	No boot PROM	
		0	0	1	0	0C000H	8k/16k	
		0	0	1	1	0C400H	8k/16k	
		0	1	0	0	0C800H	8k/16k	
		0	1	0	1	0CC00H	8k/16k	
		0	1	1	0	0D000H	8k/16k	
		0	1	1	1	0D400H	8k/16k	
		1	0	0	0	0D800H	8k/16k	
		1	0	0	1	0DC00H	8k/16k	
		1	0	1	0	0C000H	32k/32k	
		1	0	1	1	0C800H	32k/32k	
		1	1	0	0	0D000H	32k/32k	
		1	1	0	1	0D800H	32k/32k	
		1	1	1	0	0C000H	64k/64k	
		1	1	1	1	0D000H	64k/64k	
4	COMP	the Et	COMPATIBLE: This bit determines if the AT/LANTIC Controller's memory and I/O maps are compatible with the EtherCard PLUS and Novell boards or if they use the full 64k address space available to the NIC. A low level indicates compatible mode.					
5	INTMOD		INTERRUPT MODE: When this bit is low the AT/LANTIC Controller is in Direct Drive interrupt mode. When it is high Coded Output interrupt mode is used.					
6	CLKSEL		CLOCK SELECT: If this bit is low the NIC core is clocked by the 20 MHz. If this bit is high the NIC core is clocked by the signal on the BSCLK pin.					
7	SOFEN	softwa	are. If thi uration t	s bit is s	et high	then the cor	withen the user can program configuration registers A and B in infiguration registers are not accessible. If EECONFIG is high, the erwritten by the configuration from the EEPROM even if this bit is	

5.2 SHARED MEMORY MODE CONTROL REGISTERS

The following tables describe the functionality of the two control registers and the 8/16 detection registers.

Shared Memory AT Detect Register (Read only)

7	6	5	4	3	2	1	0
х	х	х	х	х	х	х	ATDET

Bits	Symbols	Function
D0	ATDET	ATDET: This bit shows the value on the DWID pin and can be read by software to determine whether the AT/LANTIC Controller is operating in an 8- or 16-bit slot. When this bit is read as a 1 the AT/LANTIC Controller
		is in a 16-bit slot (PC-AT system bus) and when read as a 0 it is in an 8-bit slot.

Shared Memory Control Register 1

7	6	5	4	3	2	1	0
RESET	MEME	A18	A17	A16	A15	A14	A13

Bits	Symbols	Function
D0-D5	A13-A18	A13–18: Lower part of the address register used to determine the position of the AT/LANTIC Controller's memory within the system memory map.
D6	MEME	MEMORY ENABLE: Enables external memory accesses when held high. This bit will power up low, so the user must program the base memory address and set this bit high to enable the memory into the system's memory map.
D7	RESET	RESET: Resets NIC core of AT/LANTIC Controller.

Shared Memory Control Register 2

7	6	5	4	3	2	1	0
8/16	MEMW	Unused	LA23	LA22	LA21	LA20	LA19

Bits	Symbols	Function
D0-D4	LA19-LA23	LA19–23: Upper part of the address register used to determine the position of the AT/LANTIC Controller's memory within the system memory map.
D5		UNUSED
D6	MEMW	MEMORY WIDTH: Sets width of external memory. When set low external memory is accessed as byte wide, so only 8 kbytes of memory are available. When set high external memory is accessed as word wide, so 16 kbytes are available. In non-compatible mode up to 64 kbytes of external memory is allowed when this bit is set high, or 32 kbytes when low. When bit 7 is set high this bit must also be set high.
D7	8/16	8/16-BIT: Allows 16-bit system accesses to external memory when set high. When low only 8-bit accesses are allowed. When high the generation of the M16 output is allowed.

5.3 NIC CORE REGISTERS

All registers are 8-bit wide and mapped into two pages which are selected in the Command Register (PS0, PS1). Pins SA0–SA3 are used to address registers within each page. Page 0 registers are those registers which are commonly accessed during AT/LANTIC Controller operation while page 1 registers are used primarily for initialization. The registers are partitioned to avoid having to perform two write/read cycles to access commonly used registers.

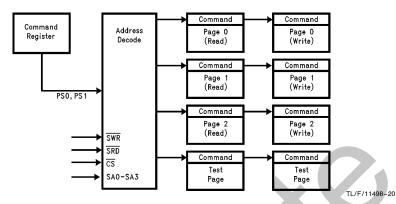


FIGURE 23. NIC Core Register Mapping

Register Assignments

Page 0 Address Assignments (PS1 = 0, PS0 = 0)

SA0-SA3	RD	WR
00H	Command (CR)	Command (CR)
01H	Current Local DMA Address 0 (CLDA0)	Page Start Register (PSTART)
02H	Current Local DMA Address 1 (CLDA1)	Page Stop Register (PSTOP)
03H	Boundary Pointer (BNRY)	Boundary Pointer (BNRY)
04H	Transmit Status Register (TSR)	Transmit Page Start Address (TPSR)
05H	Number of Collisions Register (NCR)	Transmit Byte Count Register 0 (TBCR0)
06H	FIFO (FIFO)	Transmit Byte Count Register 1 (TBCR1)
07H	Interrupt Status Register (ISR)	Interrupt Status Register (ISR)
08H	Current Remote DMA Address 0 (CRDA0)	Remote Start Address Register 0 (RSAR0)
09H	Current Remote DMA Address 1 (CRDA1)	Remote Start Address Register 1 (RSAR1)
0AH	Reserved	Remote Byte Count Register 0 (RBCR0)
0BH	Reserved	Remote Byte Count Register 1 (RBCR1)
0CH	Receive Status Register (RSR)	Receive Configuration Register (RCR)
0DH	Tally Counter 0 (Frame Alignment Errors) (CNTR0)	Transmit Configuration Register (TCR)
0EH	Tally Counter 1 (CRC Errors) (CNTR1)	Data Configuration Register (DCR)
0FH	Tally Counter 2 Missed Packet Errors) (CNTR2)	Interrupt Mask Register (IMR)

Page 1 Address Assignments (PS1 = 0, PS0 = 1)

SA0-SA3	RD	WR		
00H	Command (CR)	Command (CR)		
01H	Physical Address	Physical Address		
	Register 0 (PAR0)	Register 0 (PA R0)		
02H	Physical Address	Physical Address		
	Register 1 (PAR1)	Register 1 (PA R1)		
03H	Physical Address	Physical Address		
	Register 2 (PAR2)	Register 2 (PAR2)		
04H	Physical Address	Physical Address		
	Register 3 (PAR3)	Register 3 (PAR3)		
05H	Physical Address	Physical Address		
	Register 4 (PAR4)	Register 4 (PAR4)		
06H	Physical Address	Physical Address		
	Register 5 (PAR5)	Register 5 (PAR5)		
07H	Current Page	Current Page		
	Register (CURR)	Register (CURR)		
08H	Multicast Address	Multicast Address		
	Register 0 (MAR0)	Register 0 (MAR0)		
09H	Multicast Address	Multicast Address		
	Register 1 (MAR1)	Register 1 (MAR1)		
0AH	Multicast Address	Multicast Address		
	Register 2 (MAR2)	Register 2 (MAR2)		
0BH	Multicast Address	Multicast Address		
	Register 3 (MAR3)	Register 3 (MAR3)		
0CH	Multicast Address	Multicast Address		
	Register 4 (MAR4)	Register 4 (MAR4)		
0DH	Multicast Address	Multicast Address		
	Register 5 (MAR5)	Register 5 (MAR5)		
0EH	Multicast Address	Multicast Address		
	Register 6 (MAR6)	Register 6 (MAR6)		
0FH	Multicast Address	Multicast Address		
	Register 7 (MAR7)	Register 7 (MAR7)		

Page 2 Address Assignments (PS1 = 1, PS0 = 0)

SA0-SA3	RD	WR
00H	Command (CR)	Command (CR)
01H	Page Start Register (PSTART)	Current Local DMA Address 0 (CLDA0)
02H	Page Stop Register (PSTOP)	Current Local DMA Address 1 (CLDA1)
03H	Remote Next Packet Pointer	Remote Next Packet Pointer
04H	Transmit Page Start Address (TPSR)	Reserved
05H	Local Next Packet Pointer	Local Next Packet Pointer
06H	Address Counter (Upper)	Address Counter (Upper)
07H	Address Counter (Lower)	Address Counter (Lower)
08H	Reserved	Reserved
09H	Reserved	Reserved
0AH	Reserved	Reserved
0BH	Reserved	Reserved
0CH	Receive Configuration Register (RCR)	Reserved
ODH	Transmit Configuration Register (TCR)	Reserved
OEH	Data Configuration Register (DCR)	Reserved
0FH	Interrupt Mask Register (IMR)	Reserved

Note: Page 2 registers should only be accessed for diagnostic purposes. They should not be modified during normal operation.

Page 3 should never be modified.

COMMAND REGISTER (CR) 00H (READ/WRITE)

The Command Register is used to initiate transmissions, enable or disable Remote DMA operations and to select register pages. To issue a command the microprocessor sets the corresponding bit(s) (RD2, RDI, RD0, TXP). Further commands may be overlapped, but with the following rules: (1) If a transmit command overlaps with a remote DMA operation, bits RD0, RD1, and RD2 must be maintained for the remote DMA command when setting the TXP bit. Note, if a remote DMA command is re-issued when giving the transmit command, the DMA will complete immediately if the remote byte count register has not been re-initialized. (2) If a remote DMA operation overlaps a transmission, RD0, RD1, and RD2 may be written with the desired values and a "0" written to the TXP bit. Writing a "0" to this bit has no effect. (3) A remote write DMA may not overlap remote read operation or visa versa. Either of these operations must either complete or be aborted before the other operation may start. Bits PS1, PS0, RD2, and STP may be set any time.

7	6	5	4	3	2	1	0
PS1	PS0	RD2	RD1	RD0	TXP	STA	STP

Bits	Symbols	Description					
D0	STP	STOP: Software reset command, takes the controller offline, no packets will be received or transmitted. Any reception or transmission in progress will continue to completion before entering the reset state. To exit this state, the STP bit must be reset and the STA bit must be set high. To perform a software reset, this bit should be set high. The software reset has executed only when indicated by the RST bit in the ISR being set to at 1. STP powers up high. Note: If the AT/LANTIC Controller has previously been in start mode and the STP is set, both the STP and STA bits will remain set.					
D1	STA	START: This bit is used to activate the NIC Core after either power up, or when the NIC Core has been placed in a reset mode by software command or error. STA powers up low.					
D2	TXP	TRANSMIT PACKET: This bit must be set to initiate transmission of a packet. TXP is internally reset either after the transmission is completed or aborted. This bit should be set only after the Transmit Byte Count and Transmit Page Start registers have been programmed.					
D3-D5	RD0-RD2	REMOTE DMA COMMAND: These three encoded bits control operation of the Remote DMA channel. RD2 can be set to abort any Remote DMA command in progress. The Remote Byte Count Registers should be cleared when a Remote DMA has been aborted. The Remote Start Addresses are not restored to the starting address if the Remote DMA is aborted.					
		RD2	RD1	RD0			
		0	0	0	Not Allowed		
		0	0	1	Remote Read		
		0	1	0	Remote Write		
		0	1	1	Send Packet		
		1	X	X	Abort/Complete Remote DMA (Note 1)		
D6, D7	PS0, PS1	address PS1 0 0	SELECT: 1 ses RA0-3 PS0 0 1 0	Register Page 0 Register Page 1 Register Page 2	its select which register page is to be accessed with		
		1	1	Reserved			

Note 1: If a remote DMA operation is aborted and the remote byte count has not decremented to zero, the data transfer port should be read, for a remote read or send packet, or written to, for a remote write. This is required to ensure future correct operation.

INTERRUPT STATUS REGISTER (ISR) 07H (READ/WRITE)

This register is accessed by the host processor to determine the cause of an interrupt. Any interrupt can be masked in the Interrupt Mask Register (IMR). Individual interrupt bits are cleared by writing a "1" into the corresponding bit of the ISR. The valid interrupt output is active as long as any unmasked signal is set, and will not go low until all unmasked bits in this register have been cleared. The ISR must be cleared after power up by writing it with all 1's.

7	6	5	4	3	2	1	0
RST	RDC	CNT	OVW	TXE	RXE	PTX	PRX

Bits	Symbols	Description
D0	PRX	PACKET RECEIVED: Indicates packet received with no errors.
D1	PTX	PACKET TRANSMITTED: Indicates packet transmitted with no errors.
D2	RXE	RECEIVE ERROR: Indicates that a packet was received with one or more of the following errors: —CRC Error —Frame Alignment Error —FIFO Overrun —Missed Packet
D3	TXE	TRANSMIT ERROR: Set when packet transmitted with one or more of the following errors: —Excessive Collisions —FIFO Underrun
D4	OVW	OVERWRITE WARNING: Set when receive buffer ring storage resources have been exhausted. (Local DMA has reached Boundary Pointer)
D5	CNT	COUNTER OVERFLOW: Set when MSB of one or more of the Network Tally Counters has been set.
D6	RDC	REMOTE DMA COMPLETE: Set when Remote DMA operation has been completed.
D7	RST	RESET STATUS: Set when AT/LANTIC Controller enters reset state and cleared when a Start Command is issued to the CR. This bit is also set when a Receive Buffer Ring overflow occurs and is cleared when one or more packets have been removed from the ring. Writing to this bit has no effect. Note: This bit does not generate an interrupt, it is merely a status indicator.

INTERRUPT MASK REGISTER (IMR) 0FH (WRITE)

The Interrupt Mask Register is used to mask interrupts. Each interrupt mask bit corresponds to a bit in the Interrupt Status Register (ISR). If an interrupt mask bit is set, an interrupt will be issued whenever the corresponding bit in the ISR is set. If any bit in the IMR is set low, an interrupt will not occur when the bit in the ISR is set. **The IMR powers up all zeros.**

7	6	5	4	3	2	1	0
_	RDCE	CNTE	OVWE	TXEE	RXEE	PTXE	PRXE

Bits	Symbols	Description
D0	PRXE	PACKET RECEIVED INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when packet received
D1	PTXE	PACKET TRANSMITTED INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when packet is transmitted
D2	RXEE	RECEIVE ERROR INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when packet received with error
D3	TXEE	TRANSMIT ERROR INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when packet transmission results in error
D4	OVWE	OVERWRITE WARNING INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when Buffer Management Logic lacks sufficient buffers to store incoming packet
D5	CNTE	COUNTER OVERFLOW INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when MSB of one or more of the Network Statistics counters has been set
D6	RDCE	DMA COMPLETE INTERRUPT ENABLE 0: Interrupt Disabled 1: Enables Interrupt when Remote DMA transfer has been completed
D7	reserved	reserved

DATA CONFIGURATION REGISTER (DCR) 0EH (WRITE)

This Register is used to program the AT/LANTIC Controller for 8- or 16-bit memory interface, select byte ordering in 16-bit applications and establish FIFO thresholds. The DCR must be initialized prior to loading the Remote Byte Count Registers. LAS is set on power up.

7	6	5	4	3	2	1	0
_	FT1	FT0	ARM	LS	LAS	BOS	WTS

Bits	Symbols	Description
D0	WTS	WORD TRANSFER SELECT 0: Selects byte-wide DMA transfers 1: Selects word-wide DMA transfers ;WTS establishes byte or word transfers for both Remote and Local DMA transfers Note: When word-wide mode is selected, up to 32k words are addressable; A0 remains low.
D1	BOS	BYTE ORDER SELECT 0: MS byte placed on AD15-AD8 and LS byte on AD7-AD0. (32xxx, 80x86) 1: MS byte placed on AD7-AD0 and LS byte on AD15-AD8. (680x0) :lgnored when WTS is low
D2	LAS	LONG ADDRESS SELECT 0: Dual 16-bit DMA mode 1: Single 32-bit DMA mode ;When LAS is high, the contents of the Remote DMA registers RSAR0, 1 are issued as A16-A31 Power up high
D3	LS	LOOPBACK SELECT 0: Loopback mode selected. Bits D1 and D2 of the TCR must also be programmed for Loopback operation 1: Normal Operation
D4	ARM	AUTO-INITIALIZE REMOTE 0: Send Command not executed, all packets removed from Buffer Ring under program control 1: Send Command executed, Remote DMA auto-initialized to remove packets from Buffer Ring Note: Send Command cannot be used with 680x0 byte processors.
D5 and D6	FT0 and FT1	FIFO THRESHOLD SELECT: Encoded FIFO threshold. Establishes point at which the memory bus is requested when filling or emptying the FIFO. During reception, the FIFO threshold indicates the number of bytes (or words) the FIFO has filled serially from the network before the FIFO is emptied onto the memory bus. Note: FIFO threshold setting determines the DMA burst length. Receive Thresholds FT1 FT0 Word Wide Byte Wide 0 0 1 Word 2 Bytes 0 1 2 Words 4 Bytes 1 0 4 Words 8 Bytes 1 1 6 Words 12 Bytes During transmission, the FIFO threshold indicates the number of bytes (or words) the FIFO has filled from the Local DMA before being transferred to the memory. Thus, the transmission threshold is 13 bytes less the received threshold.

TRANSMIT CONFIGURATION REGISTER (TCR) ODH (WRITE)

The transmit configuration establishes the actions of the transmitter section of the AT/LANTIC Controller during transmission of a packet on the network. **LB1 and LB0 which select loopback mode power up as 0.**

7	6	5	4	3	2	1	0
_	_	_	OFST	ATD	LB1	LB0	CRC

Bits	Symbols	Description						
D0	CRC	IHIBIT CRC 0: CRC appended by transmitter 1: CRC inhibited by transmitter loopback mode CRC can be enabled or disabled to test the CRC logic						
D1 and D2	LB0 and LB1	ENCODED LOOPBACK CONTROL: These encoded configuration bits set the type of loopback that is to be performed. Note that loopback in mode 2 places the ENDEC Module in loopback mode and that D3 of the DCR must be set to zero for loopback operation. LB1 LB0 Mode 0 0 Normal Operation (LPBK = 0) Mode 1 0 1 Internal NIC Module Loopback (LPBK = 0) Mode 2 1 0 Internal ENDEC Module Loopback (LPBK = 1) Mode 3 1 External Loopback (LPBK = 0)						
D3	ATD	AUTO TRANSMIT DISABLE: This bit allows another station to disable the AT/LANTIC Controller's transmitter by transmission of a particular multicast packet. The transmitter can be re-enabled by resetting this bit or by reception of a second particular multicast packet. 0: Normal Operation 1: Reception of multicast address hashing to bit 62 disables transmitter, reception of multicast address hashing to bit 63 enables transmitter.						
D4	OFST	COLLISION OFFSET ENABLE: This bit modifies the backoff algorithm to allow prioritization of nodes. 0: Backoff Logic implements normal algorithm. 1: Forces Backoff algorithm modification to 0 to 2 ^{min(3 + n, 10)} slot times for first three collisions, then follows standard backoff. (For the first three collisions, the station has higher average backoff delay making a low priority mode.)						
D5	reserved	reserved						
D6	reserved	reserved						
D7	reserved	reserved						

TRANSMIT STATUS REGISTER (TSR) 04H (READ)

This register records events that occur on the media during transmission of a packet. It is cleared when the next transmission is initiated by the host. All bits remain low unless the event that corresponds to a particular bit occurs during transmission. Each transmission should be followed by a read of this register. The contents of this register are not specified until after the first transmission.

	7	6	5	4	3	2	1	0
ĺ	OWC	CDH	FU	CRS	ABT	COL	_	PTX

Bits	Symbols	Description
D0	PTX	PACKET TRANSMITTED: Indicates transmission without error. (No excessive collisions or FIFO underrun)(ABT = "0", FU = "0")
D1	reserved	reserved
D2	COL	TRANSMIT COLLIDED: Indicates that the transmission collided at least once with another station on the network. The number of collisions is recorded in the Number of Collisions Registers (NCR).
D3	ABT	TRANSMIT ABORTED: Indicates the AT/LANTIC Controller aborted transmission because of excessive collisions. (Total number of transmissions including original transmission attempt equals 16.)
D4	CRS	CARRIER SENSE LOST: This bit is set when carrier is lost during transmission of the packet. Transmission is not aborted on loss of carrier.
D5	FU	FIFO UNDERRUN: If the AT/LANTIC Controller cannot gain access of the bus before the FIFO empties, this bit is set. Transmission of the packet will be aborted.
D6	CDH	CD HEARTBEAT: Failure of the transceiver to transmit a collision signal after transmission of a packet will set this bit. The Collision Detect (CD) heartbeat signal must commence during the first 6.4 μ s of the Interframe Gap following a transmission. In certain collisions, the CD Heartbeat bit will be set even though the transceiver is not performing the CD heartbeat test.
D7	OWC	OUT OF WINDOW COLLISION: Indicates that a collision occurred after a slot time (51.2 μs). Transmissions rescheduled as in normal collisions.

RECEIVE CONFIGURATION REGISTER (RCR) 0CH (WRITE)

This register determines operation of the AT/LANTIC Controller during reception of a packet and is used to program what types of packets to accept.

7	6	5	4	3	2	1	0
_	_	MON	PRO	AM	AB	AR	SEP

Bits	Symbols	Description
D0	SEP	SAVE ERRORED PACKETS 0: Packets with receive errors are rejected. 1: Packets with receive errors are accepted. Receive errors are CRC and Frame Alignment errors.
D1	AR	ACCEPT RUNT PACKETS: This bit allows the receiver to accept packets that are smaller than 64 bytes. The packet must be at least 8 bytes long to be accepted as a runt. 0: Packets with fewer than 64 bytes rejected. 1: Packets with fewer than 64 bytes accepted.
D2	AB	ACCEPT BROADCAST: Enables the receiver to accept a packet with an all 1's destination address. 0: Packets with broadcast destination address rejected. 1: Packets with broadcast destination address accepted.
D3	AM	ACCEPT MULTICAST: Enables the receiver to accept a packet with a multicast address, all multicast addresses must pass the hashing array. 0: Packets with multicast destination address not checked. 1: Packets with multicast destination address checked.
D4	PRO	PROMISCUOUS PHYSICAL: Enables the receiver to accept all packets with a physical address. 0: Physical address of node must match the station address programmed in PAR0-PAR5. 1: All packets with physical addresses accepted.
D5	reserved	reserved (program to 0)
D6	reserved	reserved
D7	reserved	reserved

Note: D2 and D3 are "OR'd" together, i.e., if D2 and D3 are set the AT/LANTIC Controller will accept broadcast and multicast addresses as well as its own physical address. To establish full promiscuous mode, bits D2, D3, and D4 should be set. In addition the multicast hashing array must be set to all 1's in order to accept all multicast addresses.

RECEIVE STATUS REGISTER (RSR) 0CH (READ)

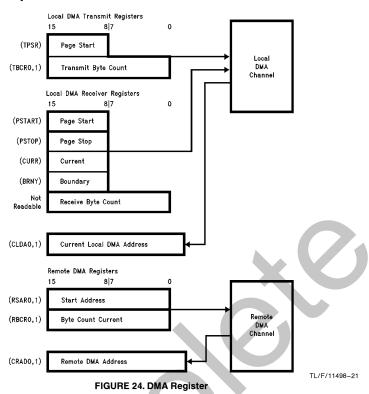
This register records status of the received packet, including information on errors and the type of address match, either physical or multicast. The contents of this register are written to buffer memory by the DMA after reception of a good packet. If packets with errors are to be saved the receive status is written to memory at the head of the erroneous packet if an erroneous packet is received. If packets with errors are to be rejected the RSR will not be written to memory. The contents will be cleared when the next packet arrives. CRC errors, Frame Alignment errors and missed packets are counted internally by the AT/LAN-TIC Controller which relinquishes the Host from reading the RSR in real time to record errors for Network Management Functions. The contents of this register are not specified until after the first reception.

7	6	5	4	3	2	1	0
DFR	DIS	PHY	MPA	FO	FAE	CRC	PRX

Bits	Symbols	Description
D0	PRX	PACKET RECEIVED INTACT: Indicates packet received without error. (Bits CRC, FAE, FO, and MPA are zero for the received packet.)
D1	CRC	CRC ERROR: Indicates packet received with CRC error. Increments Tally Counter (CNTR1). This bit will also be set for Frame Alignment errors.
D2	FAE	FRAME ALIGNMENT ERROR: Indicates that the incoming packet did not end on a byte boundary and the CRC did not match at last byte boundary. Increments Tally Counter (CNTR0).
D3	FO	FIFO OVERRUN: This bit is set when the FIFO is not serviced causing overflow during reception. Reception of the packet will be aborted.
D4	MPA	MISSED PACKET: Set when packet intended for node cannot be accepted by SNIC because of a lack of receive buffers or if the controller is in monitor mode and did not buffer the packet to memory. Increments Tally Counter (CNTR2).
D5	PHY	PHYSICAL/MULTICAST ADDRESS: Indicates whether received packet had a physical or multicast address type. 0: Physical Address Match 1: Multicast/Broadcast Address Match
D6	DIS	RECEIVER DISABLED: Set when receiver disabled by entering Monitor mode. Reset when receiver is re-enabled when exiting Monitor mode.
D7	DFR	DEFERRING: Set when internal Carrier Sense or Collision signals are generated in the ENDEC module. If the transceiver has asserted the CD line as a result of the jabber, this bit will stay set indicating the jabber condition.

Note: Following coding applies to CRC and FAE bits

FAE	CRC	Type of Error
0	0	No Error (Good CRC and ≤6 Dribble Bits
0	1	CRC Error
1	0	Illegal, wil not occur
1	1	Frame Alignment Error and CRC Error



Note: In the figure above, registers are shown as 8- or 16-bits wide. Although some registers are 16-bit internal registers, all registers are accessed as 8-bit registers. Thus the 16-bit Transmit Byte Count Register is broken into two 8-bit registers, TBCR0 and TBCR1. Also TPSR, PSTART, PSTOP, CURR and BNRY only check or control the upper 8 bits of address information on the bus. Thus they are shifted to positions 15–8 in the diagram above.

5.4 DP8390 Core DMA Registers

The DMA Registers are partitioned into groups; Transmit, Receive and Remote DMA Registers. The Transmit registers are used to initialize the Local DMA Channel for transmission of packets while the Receive Registers are used to initialize the Local DMA Channel for packet Reception. The Page Stop, Page Start, Current and Boundary Registers are used by the Buffer Management Logic to supervise the Receive Buffer Ring. The Remote DMA Registers are used to initialize the Remote DMA.

Transmit DMA Registers

TRANSMIT PAGE START REGISTER (TPSR)

This register points to the assembled packet to be transmitted. Only the eight higher order addresses are specified since all transmit packets are assembled on 256-byte page boundaries. The bit assignment is shown below. The values placed in bits D7-D0 will be used to initialize the higher order address (A8-A15) of the Local DMA for transmission. The lower order bits (A7-A0) are initialized to zero.

Bit Assignment

TRANSMIT BYTE COUNT REGISTER 0,1 (TBCR0, TBCR1)

These two registers indicate the length of the packet to be transmitted in bytes. The count must include the number of bytes in the source, destination, length and data fields. The maximum number of transmit bytes allowed is 64 kbytes. The AT/LANTIC Controller will not truncate transmissions longer than 1500 bytes. The bit assignment is shown below:

		-	-	4	-		-	-
TBCR1	L15	L14	L13	L12	L11	L10	L9	L8
TBCR0				4				

Local DMA Receive Registers

PAGE START STOP REGISTERS (PSTART, PSTOP)

The Page Start and Page Stop Registers program the starting and stopping address of the Receive Buffer Ring. Since

the AT/LANTIC Controller uses fixed 256-byte buffers aligned on page boundaries only the upper eight bits of the start and stop address are specified.

PSTART, PSTOP bit assignment

	_ 7	6	5	4	3	2	1	0
PSTART PSTOP	A15	A14	A13	A12	A11	A10	A9	A8

BOUNDARY (BNRY) REGISTER

This register is used to prevent overflow of the Receive Buffer Ring. Buffer management compares the contents of this register to the next buffer address when linking buffers together. If the contents of this register match the next buffer address the Local DMA operation is aborted.

	7	6	5	4	3	2	1	0
BNRY	A15	A14	A13	A12	A11	A10	А9	A8

CURRENT PAGE REGISTER (CURR)

This register is used internally by the Buffer Management Logic as a backup register for reception. CURR contains the address of the first buffer to be used for a packet reception and is used to restore DMA pointers in the event of receive errors. This register is initialized to the same value as PSTART and should not be written to again unless the controller is Reset.

	7	6	5	4	3	2	1	0
CURR	A15	A14	A13	A12	A11	A10	A9	A8

CURRENT LOCAL DMA REGISTER 0,1 (CLDA0,1)

These two registers can be accessed to determine the current Local DMA Address.

	7	6	5	4	3	2	1	0
CLDA1	A15	A14	A13	A12	A11	A10	А9	A8
	7	6	5	4	3	2	1	0

Remote DMA Registers

REMOTE START ADDRESS REGISTERS (RSAR0,1)

Remote DMA operations are programmed via the Remote Start Address (RSAR0,1) and Remote Byte Count (RBCR0,1) registers. The Remote Start Address is used to point to the start of the block of data to be transferred and the Remote Byte Count is used to indicate the length of the block (in bytes).

	7	6	5	4	3	2	1	0
RSAR1	A15	A14	A13	A12	A11	A10	A9	A8
RSAR0		-	5		-		_	

REMOTE BYTE COUNT REGISTERS (RCB0,1)

	7	6	5	4	3	2	1	0
RBCR1	A15	A14	A13	A12	A11	A10	A9	A8
	7	6	5	4	3	2	1	0

Notes:

RSAR0 programs the start address bits A0-A7.

RSAR1 programs the start address bits A8-A15.

Address incremented by two for word transfers, and by one for byte transfers. Byte count decremented by two for word transfers and by one for byte transfers.

RBCR0 programs LSB byte count.

RBCR1 programs MSB byte count.

CURRENT REMOTE DMA ADDRESS (CRDA0, CRDA1)

The Current Remote DMA Registers contain the current address of the Remote DMA. The bit assignment is shown below:

	7	6	5	4	3	2	1	0
CRDA1	A15	A14	A13	A12	A11	A10	A9	A8
				,				
	7	6	5	4	3	2	1	0

Physical Address Registers (PAR0-PAR5)

The physical address registers are used to compare the destination address of incoming packets for rejecting or accepting packets. Comparisons are performed on a bytewide basis. The bit assignment shown below relates the sequence in PARO-PAR5 to the bit sequence of the received packet.

	D7	D6	D5	D4	D3	D2	D1	D0
PAR0	DA7	DA6	DA5	DA4	DA3	DA2	DA1	DA0
PAR1	DA15	DA14	DA13	DA12	DA11	DA10	DA9	DA8
PAR2	DA23	DA22	DA21	DA20	DA19	DA18	DA17	DA16
PARC	DA31	DA30	DA29	DA28	DA27	DA26	DA25	DA24
PAR4	DA39	DA38	DA37	DA36	DA35	DA34	DA33	DA32
PAR5	DA47	DA46	DA45	DA44	DA43	DA42	DA41	DA40

Destination Address							S	ource	
P/S	DA0	DA1	DA2	DA3		DA46	DA47	SA0	

Note: P/S = Preamble, Synch DA0 = Physical/Multicast Bit

Multicast Address Registers (MAR0-MAR7)

The multicast address registers provide filtering of multicast addresses hashed by the CRC logic. All destination addresses are fed through the CRC logic and as the last bit of the destination address enters the CRC, the 6 most significant bits of the CRC generator are latched. These 6 bits are then decoded by a 1 of 64 decode to index a unique filter bit (FB0-63) in the multicast address registers. If the filter bit selected is set, the multicast packet is accepted. The system designer would use a program to determine which filter

bits to set in the multicast registers. All multicast filter bits that correspond to multicast address accepted by the node are then set to one. To accept all multicast packets all of the registers are set to all ones.

Note: Although the hashing algorithm does not guarantee pertect filtering of multicast address, it will pertectly filter up to 64 multicast addresses if these addresses are chosen to map into unique locations in the multi-

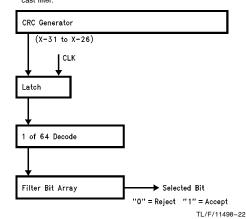


FIGURE 25. Multicast Addressing

6.0 Operation of AT/LANTIC Controller

This section details the operation of the AT/LANTIC Controller. The operations discussed are packet reception and transmission, bus operations, and loopback diagnostics.

6.1 TRANSMIT/RECEIVE PACKET ENCAPSULATION/ DECAPSULATION

A standard IEEE 802.3 packet consists of the following fields: preamble, Start of Frame Delimiter (SFD), destination address, source address, length, data, and Frame Check Sequence (FCS). The typical format is shown in the figure following. The packets are Manchester encoded and decoded by the ENDEC module and transferred serially to the NIC module using NRZ data with a clock, All fields are of fixed length except for the data field. The AT/LANTIC Controller generates and appends the preamble, SFD and FCS field during transmission. The Preamble and SFD fields are stripped during reception. (The CRC is passed through to buffer memory during reception.)

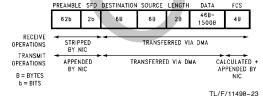


FIGURE 26. Ethernet Packet

PREAMBLE AND START OF FRAME DELIMITER (SFD)

The Manchester encoded alternating 1,0 preamble field is used by the ENDEC to acquire bit synchronization with an incoming packet. When transmitted each packet contains 62 bits of alternating 1,0 preamble. Some of this preamble will be lost as the packet travels through the network. The preamble field is stripped by the NIC module. Byte alignment is performed with the Start of Frame Delimiter (SFD) pattern which consists of two consecutive 1's. The AT/LAN-TIC Controller does not treat the SFD pattern as a byte, it detects only the two-bit pattern. This allows any preceding preamble within the SFD to be used for phase locking.

Destination Address

The destination address indicates the destination of the packet on the network and is used to filter unwanted packets from reaching a node. There are three types of address formats supported by the AT/LANTIC: physical, multicast and broadcast. The physical address is a unique address that corresponds only to a single node. All physical addresses have an MSB of "0". These addresses are compared to the internally stored physical address registers. Each bit in the destination address must match in order for the AT/LANTIC Controller to accept the packet. Multicast addresses begin with an MSB of "1". The AT/LANTIC Controller filters multicast addresses using a standard hashing algorithm that maps all multicast addresses into a 6-bit value. This 6-bit value indexes a 64-bit array that filters the value. If the address consists of all 1's it is a broadcast address, indicating that the packet is intended for all nodes. A promiscuous mode allows reception of all packets: the destination address is not required to match any filters. Physical, broadcast, multicast, and promiscuous address modes can be selected.

Source Address

The source address is the physical address of the node that sent the packet. Source addresses cannot be multicast or broadcast addresses. This field is simply passed to buffer remove.

Length Field

The 2-byte length field indicates the number of bytes that are contained in the data field of the packet. This field is not interpreted by the AT/LANTIC Controller.

Data Field

The data field consists of anywhere from 46 to 1500 bytes. Messages longer than 1500 bytes need to be broken into multiple packets. Messages shorter than 46 bytes will require appending a pad to bring the data field to the minimum length of 46 bytes. If the data field is padded, the number of valid data bytes is indicated in the length field. The AT/LANTIC Controller does not strip or append pad bytes for short packets, or check for oversize packets.

FCS Field

The Frame Check Sequence (FCS) is a 32-bit CRC field calculated and appended to a packet during transmission to allow detection of errors when a packet is received. During reception, error free packets result in a specific pattern in

the CRC generator. Packets with improper CRC will be rejected. The AUTODIN II ($X^{32}+X^{26}+X^{23}+X^{22}+X^{16}+X^{12}+X^{11}+X^{10}+X^8+X^7+X^5+X^4+X^2+X^1+1$) polynomial is used for the CRC calculations.

6.2 BUFFER MEMORY ACCESS CONTROL (DMA)

The buffer memory control capabilities of the AT/LANTIC Controller greatly simplify the use of the AT/LANTIC Controller in typical configurations. The local DMA channel transfers data between the FIFO and memory. On transmission, the packet is DMA'd from memory to the FIFO in bursts. Should a collision occur (up to 15 times), the packet is re-transmitted with no processor intervention. On reception, packets are moved via DMA from the FIFO to the receive buffer ring (as explained below).

A Remote DMA channel is also provided on the AT/LANTIC Controller to accomplish transfers between a buffer memory and an internal Data Port when using the AT/LANTIC Controller in I/O Mode. This Remote DMA channel is not used when the AT/LANTIC Controller is used in a shared Memory mode. In this second mode the buffer memory is dual ported, and directly mapped into the system memory. In this mode the system CPU directly accesses the RAM under software control to transfer packet data.

The following sections describe the operation of the Local DMA channel for packet reception which is used in both modes. For Shared Memory mode the description of the Remote DMA does not apply.

For reference an example configuration using the AT/LAN-TIC Controller is shown in *Figure 27*.

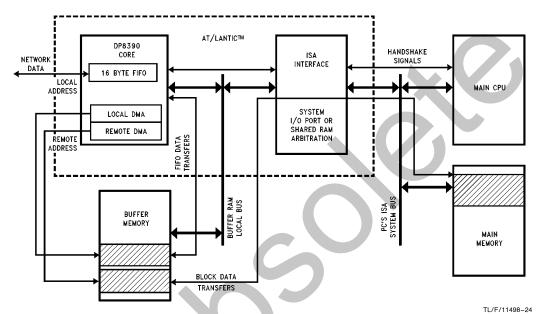


FIGURE 27. AT/LANTIC Controller Bus Architecture

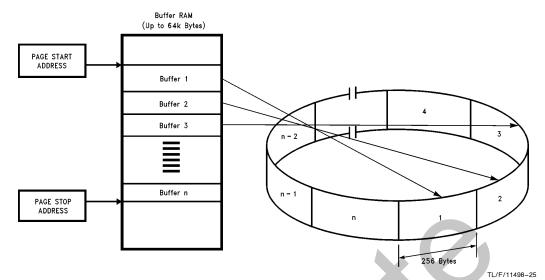


FIGURE 28. AT/LANTIC Controller Receiver Buffer Ring

6.3 PACKET RECEPTION

The Local DMA receive channel uses a Buffer Ring Structure comprised of a series of contiguous fixed length 256 byte (128 word) buffers for storage of received packets. The location of the Receive Buffer Ring is programmed in two registers, a Page Start and a Page Stop Register. Ethernet packets consist of a distribution of shorter link control packets and longer data packets, the 256 byte buffer length provides a good compromise between short packets and longer packets to most efficiently use memory. In addition these buffers provide memory resources for storage of back-toback packets in loaded networks. The assignment of buffers for storing packets is controlled by Buffer Management Logic in the AT/LANTIC Controller. The Buffer Management Logic provides three basic functions: linking receive buffers for long packets, recovery of buffers when a packet is rejected, and recirculation of buffer pages that have been

At initialization, a portion of the 64 kbyte (or 32 kword) address space is reserved for the receive buffer ring. Two eight bit registers, the Page Start Address Register (PSTART) and the Page Stop Address Register (PSTOP) define the physical boundaries of where the buffers reside. The AT/LANTIC Controller treats the list of buffers as a logical ring; whenever the DMA address reaches the Page Stop Address, the DMA is reset to the Page Start Address.

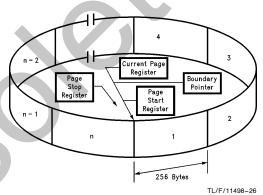


FIGURE 29. Buffer Ring at Initialization

Initialization of the Buffer Ring

Two static registers and two working registers control the operation of the Buffer Ring. These are the Page Start Register, Page Stop Register (both described previously), the Current Page Register and the Boundary Pointer Register. The Current Page Register points to the first buffer used to store a packet and is used to restore the DMA for writing status to the Buffer Ring or for restoring the DMA address in the event of a Runt packet, a CRC, or Frame Alignment error. The Boundary Register points to the first packet in the

Ring not yet read by the host. If the local DMA address ever reaches the Boundary, reception is aborted. The Boundary Pointer is also used to initialize the Remote DMA for removing a packet and is advanced when a packet is removed. A simple analogy to remember the function of these registers is that the Current Page Register acts as a Write Pointer and the Boundary Pointer acts as a Read Pointer.

Note 1: At initialization, the Page Start Register value should be loaded into both the Current Page Register and the Boundary Pointer Register.

Note 2: The Page Start Register mut not be initalized to 00H.

Beginning of Reception

When the first packet begins arriving the AT/LANTIC Controller begins storing the packet at the location pointed to by the Current Page Register. An offset of 4 bytes is saved in this first buffer to allow room for storing receive status corresponding to this packet.

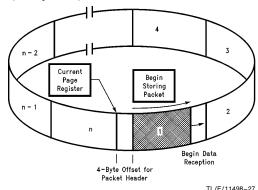


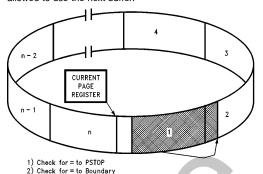
FIGURE 30. Received Packet Enters the Buffer Pages

Linking Receive Buffer Pages

If the length of the packet exhausts the first 256 byte buffer, the DMA performs a forward link to the next buffer to store the remainder of the packet. For a maximal length packet the buffer logic will link six buffers to store the entire packet. Buffers cannot be skipped when linking, a packet will always be stored in contiguous buffers. Before the next buffer can be linked, the Buffer Management Logic performs two conparisons. The first comparison tests for equality between the DMA address of the next buffer and the contents of the Page Stop Register. If the buffer address equals the Page Stop Register, the buffer management logic will restore the DMA to the first buffer in the Receive Buffer Ring value programmed in the Page Start Address Register. The second comparison tests for equality between the DMA address of the next buffer address and the contents of the Boundary Pointer Register. If the two values are equal the reception is aborted. The Boundary Pointer Register can be used to protect against overwriting any area in the receive buffer ring that has not yet been read. When linking buffers, buffer management will never cross this pointer, effectively avoiding any overwrites. If the buffer address does not match either the Boundary Pointer or Page Stop Address, the link to the next buffer is performed.

Linking Buffers

Before the DMA can enter the next contiguous 256 byte buffer, the address is checked for equality to PSTOP and to the Boundary Pointer. If neither are reached, the DMA is allowed to use the next buffer.



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FIGURE 31. Linking Receive Buffer Pages

Buffer Ring Overflow

If the Buffer Ring has been filled and the DMA reaches the Boundary Pointer Address, reception of the incoming packet will be aborted by the AT/LANTIC Controller. Thus, the packets previously received and still contained in the Ring will not be destroyed.

In heavily loaded networks which cause overflows of the Receive Buffer Ring, the AT/LANTIC Controller may disable the local DMA and suspend further receptions even if the Boundary register is advanced beyond the Current register. In the event that the AT/LANTIC Controller should encounter a receive buffer overflow, it is necessary to implement the following routine. A receive buffer overflow is indicated by the AT/LANTIC Controller's assertion of the overflow bit (OVW) in the Interrupt Status Register (ISR).

If this routine is not adhered to, the AT/LANTIC Controller may act in an unpredictable manner. It should also be noted that it is not permissible to service an overflow interrupt by continuing to empty packets from the receive buffer without implementing the prescribed overflow routine. A flow chart of the AT/LANTICTM Controller's overflow routine can be found in *Figure 32*.

Note: It is necessary to define a variable in the driver, which will be called "Resend".

- Read and store the value of the TXP bit in the AT/LAN-TIC Controller's Command Register.
- Issue the STOP command to the AT/LANTIC Controller. This is accomplished by setting the STP bit in the AT/LANTIC Controller's Command Register. Writing 21 H to the Command Register will stop the AT/LANTIC Controller
- 3. Wait for at least 1.6 ms. Since the AT/LANTIC Controller will complete any transmission or reception that is in progress, it is necessary to time out for the maximum possible duration of an Ethernet transmission or reception. By waiting 1.6 ms this is achieved with some guard band added. Previously, it was recommended that the RST bit of the Interrupt Status Register be polled to insure that the pending transmission or reception is completed. This bit is not a reliable indicator and subsequently should be ignored.

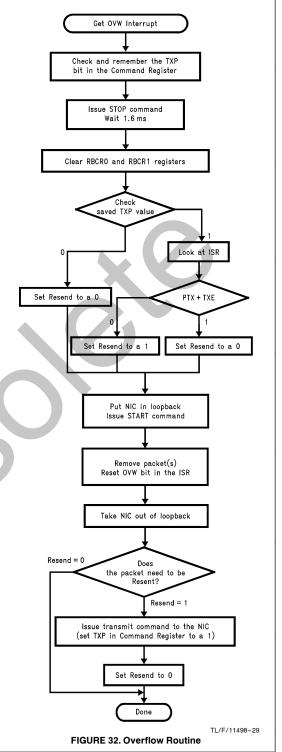
- 4. Clear the AT/LANTIC Controller's Remote Byte Count registers (RBCR0 and RBCR1).
- Read the stored value of the TXP bit from step 1, above.

If this value is a 0, set the "Resend" variable to a 0 and iump to step 6.

If this value is a 1, read the AT/LANTIC Controller's Interrupt Status Register. If either the Packet Transmitted bit (PTX) or Transmit Error bit (TXE) is set to a 1, set the "Resend" variable to a 0 and jump to step 6. If neither of these bits is set, place a 1 in the "Resend" variable and jump to step 6.

This step determines if there was a transmission in progress when the stop command was issued in step 2. If there was a transmission in progress, the AT/LANTIC Controller's ISR is read to determine whether or not the packet was recognized by the AT/LANTIC Controller. If neither the PTX nor TXE bit was set, then the packet will essentially be lost and re-transmitted only after a time-out takes place in the upper level software. By determining that the packet was lost at the driver level, a transmit command can be reissued to the AT/LANTIC Controller once the overflow routine is completed (as in step 11). Also, it is possible for the AT/LANTIC Controller to defer indefinitely, when it is stopped on a busy network. Step 5 also alleviates this problem. Step 5 is essential and should not be omitted from the overflow routine, in order for the AT/LANTIC Controller to operate correctly.

- Place the AT/LANTIC Controller in either mode 1 or mode 2 loopback. This can be accomplished by setting bits D2 and D1, of the Transmit Configuration Register, to 0,1 or 1,0 respectively.
- Issue the START command to the AT/LANTIC Controller. This can be accomplished by Writing 22H to the Command Register. This is necessary to activate the AT/LANTIC Controller's Remote DMA channel.
- 8. Remove one or more packets from the receive buffer ring.
- Reset the overwrite warning (OVW, overflow) bit in the Interrupt Status Register.
- Take the AT/LANTIC Controller out of loopback. This is done by Writing the Transmit Configuration Register with the value it contains during normal operation. (Bits D2 and D1 should both be programmed to 0.)
- 11. If the "Resend" variable is set to a 1, reset the "Resend" variable and reissue the transmit command. This is done by writing a value of 26H to the Command Register. If the "Resend" variable is 0, nothing needs to be done.
- Note 1: If Remote DMA is not being used, the AT/LANTIC Controller does not need to be started before packets can be removed from the receive buffer ring. Hence, step 8 could be done before step 7, eliminating or reducing the time spent polling in step 5.
- Note 2: When the AT/LANTIC Controller is in STOP mode, the Missed Packet Tally counter is disabled.



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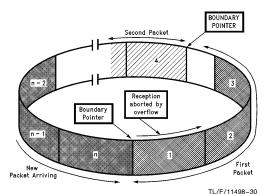


FIGURE 33. Received Packet Aborted if it Hits Boundary

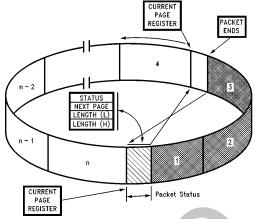
Enabling the AT/LANTIC Controller on an Active Network

After the AT/LANTIC Controller has been initialized the procedure for disabling and then re-enabling the AT/LANTIC Controller on the network is similar to handling Receive Buffer Ring overflow as described previously.

- Program Command Register for page 0 (Command Register = 21H)
- 2. Initialize Data Configuration Register (DCR)
- 3. Clear Remote Byte Count Registers (RBCR0, RBCR1) if using Remote DMA
- 4. Initialize Receive Configuration Register (RCR)
- 5. Place the AT/LANTIC Controller in LOOPBACK mode 1 or 2 (Transmit Configuration Register = 02H or 04H)
- Initialize Receive Buffer Ring: Boundary Pointer (BNDRY), Page Start (PSTART), and Page Stop (PSTOP)
- 7. Clear Interrupt Status Register (ISR) by writing 0FFH to it
- 8. Initialize Interrupt Mask Register (IMR)
- 9. Program Command Register for page 1 (Command Register = 61H)
 - i) Initialize Physical Address Registers (PAR0-PAR5)
 - ii) Initialize Multicast Address Registers (MAR0– MAR7)
 - iii) Initialize CURRENT pointer
- Put AT/LANTIC Controller in START mode (Command Register = 22H). The local receive DMA is still not active since the AT/LANTIC Controller is in LOOPBACK.
- Initialize the Transmit Configuration for the intended value. The AT/LANTIC Controller is now ready for transmission and reception.

End of Packet Operations

At the end of the packet the AT/LANTIC Controller determines whether the received packet is to be accepted or rejected. It either branches to a routine to store the Buffer Header or to another routine that recovers the buffers used to store the packet.



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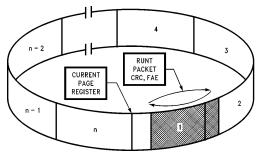
FIGURE 34. Termination of Received Packet—Packet Accepted

Successful Reception

If the packet is successfully received, the DMA is restored to the first buffer used to store the packet (pointed to by the Current Page Register). The DMA then stores the Receive Status, a Pointer to where the next packet will be stored (Buffer 4) and the number of received bytes. Note that the remaining bytes in the last buffer are discarded and reception of the next packet begins on the next empty 256-byte buffer boundary. The Current Page Register is then initialized to the next available buffer in the Buffer Ring. (The location of the next buffer had been previously calculated and temporarily stored in an internal scratchpad register.)

Buffer Recovery for Rejected Packets

If the packet is a runt packet or contains CRC or Frame Alignment errors, it is rejected. The buffer management logic resets the DMA back to the first buffer page used to store the packet (pointed to by CURR), recovering all buffers that had been used to store the rejected packet. This operation will not be performed if the AT/LANTIC Controller is programmed to accept either runt packets or packets with CRC or Frame Alignment errors. The received CRC is always stored in buffer memory after the last byte of received data for the packet.



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FIGURE 35. Termination of Receive Packet—Packet Reject

Error Recovery

If the packet is rejected as shown, the DMA is restored by the AT/LANTIC Controller by reprogramming the DMA starting address pointed to by the Current Page Register.

Storage Format for Received Packets

The following diagrams describe the format for how received packets are placed into memory by the local DMA channel. These modes are selected in the Data Configuration Register.

AD15	AD8	AD7	AD0
Next Packet	Pointer	Recei	ve Status
Receive Byte	Count 1	Receive B	Byte Count 0
Byte 2	2	В	yte 1

BOS=0, WTS =1 in Data Configuration Register. This format is used with Series 32xxx, or 808xx processors.

AD15	AD8	AD7	AD0
Next Packet Po	inter	Receiv	e Status
Receive Byte Co	ount 0	Receive B	yte Count 1
Byte 1		Ву	rte 2

BOS=1, WTS =1 in Data Configuration Register. This format is used with 680x0 type processors. (**Note:** The Receiver Count ordering remains the same for BOS=0 or 1.)

Receive Status
Next Packet Pointer
Receive Byte Count 0
Receive Byte Count 1
Byte 0
Byte 1

 ${\rm BOS}={\rm 0,~WTS}={\rm 0}$ in Data Configuration Register. This format is used with general 8-bit processors.

6.4 PACKET TRANSMISSION

The Local DMA is also used during transmission of a packet. Three registers control the DMA transfer during transmission, a Transmit Page Start Address Register (TPSR) and the Transmit Byte Count Registers (TBCR0,1). When the AT/LANTIC Controller receives a command to transmit the packet pointed to by these registers, buffer memory data will be moved into the FIFO as required during transmission. The AT/LANTIC Controller will generate and append the preamble, synch and CRC fields.

General Transmit Packet Format

Transmit	Destination Address	6 Bytes
Byte	Source Address	6 Bytes
Count	Type/Length	2 Bytes
TBCR0, 1	Data	≥46 Bytes
	Pad (If data < 46 Bytes)	

Transmit Packet Assembly

The AT/LANTIC Controller requires a contiguous assembled packet with the format shown. The transmit byte count includes the Destination Address, Source Address, Length Field and Data. It does not include preamble and CRC.

When transmitting data smaller than 46 bytes, the packet must be padded to a minimum size of 64 bytes. The programmer is responsible for adding and stripping pad bytes.

The packets are placed in the buffer RAM by the system. In I/O Mode the system programs the NIC Core's Remote DMA to mode the data from the data port to the RAM handshaking with system transfers loading the I/O data port. In Shared Memory Mode the packets are written directly to the RAM by system using standard memory transfer instructions (MOV).

For I/O mode the data transfer must be 16 bits (1 word) when in 16-bit mode, and 8 bits when the AT/LANTIC Controller is set in 8-bit mode. The data width is selected by setting the WTS bit in the Data Configuration Register and setting the DWID pin for the proper mode.

In Shared Memory mode data transfer can be accomplished by using either 8- or 16-bit data transfer instructions, because this mode responds to 8/16-bit data signalling on the ISA bus. In this mode Shared Memory Control Register 2-bit 6 sets the bus interface data width, and the NIC Core's data width is set by the WTS bit in the Data Configuration Register

Transmission

Prior to transmission, the TPSR (Transmit Page Start Register) and TBCR0, TBCR1 (Transmit Byte Count Registers) must be initialized. To initiate transmission of the packet the TXP bit in the Command Register is set. The Transmit Status Register (TSR) is cleared and the AT/LANTIC Controller begins to prefetch transmit data from memory (unless the AT/LANTIC Controller is currently receiving). If the interframe gap has timed out the AT/LANTIC Controller will begin transmission.

Conditions Required to Begin Transmission

In order to transmit a packet, the following three conditions must be met:

- 1. The Interframe Gap Timer has timed out the first 6.4 μs of the Interframe Gap
- At least one byte has entered the FIFO. (This indicates that the burst transfer has been started)
- 3. If a collision had been detected then before transmission the packet time must have timed out.

In typical systems the AT/LANTIC Controller prefetches the first burst of bytes before the 6.4 μs timer expires. The time during which AT/LANTIC Controller transmits preamble can also be used to load the FIFO.

Note: If carrier sense is asserted before a byte has been loaded into the FIFO, the AT/LANTIC Controller will become a receiver.

Collision Recovery

During transmission, the Buffer Management logic monitors the transmit circuitry to determine if a collision has occurred. If a collision is detected, the Buffer Management logic will reset the FIFO and restore the Transmit DMA pointers for retransmission of the packet. The COL bit will be set in the TSR and the NCR (Number of Collisions Register) will be incremented. If 15 retransmissions each result in a collision the transmission will be aborted and the ABT bit in the TSR will be set.

Note: NCR reads as zeroes if excessive collisions are encountered.

Transmit Packet Assembly Format

The following diagrams describe the format for how packets must be assembled prior to transmission for different byte ordering schemes. The various formats are selected in the Data Configuration Register.

D15 D	8 D7	D0		
Destination Address	1 Dest	ination Address 0		
Desitination Address	3 Dest	ination Address 2		
Desitination Address	5 Dest	ination Address 4		
Source Address 1	So	Source Address 0		
Source Address 3	So	urce Address 2		
Source Address 5	So	urce Address 4		
Type/Length 1	Т	Type Length 0		
Data 1		Data 0		

BOS = 1, WTS = 1 in Data Configuration Register.

This format is used with Series 32xxx, or 808xx processors.

D15 D8	D7 D0		
Destination Address 0	Destination Address 1		
Desitination Address 2	Destination Address 3		
Desitination Address 4	Destination Address 5		
Source Address 0	Source Address 1		
Source Address 2	Source Address 3		
Source Address 4	Source Address 5		
Type/Length 0	Type Length 1		
Data 0	Data 1		

BOS = 1, WTS = 1 in Data Configuration Register.

This format is used with 680x0 type processors.

D7	D0
Destination Address 0	
Destination Address 1	
Destination Address 2	
Destination Address 3	
Destination Address 4	
Destination Address 5	
Source Address 0	
Source Address 1	
Source Address 2	
Source Address 3	
Source Address 4	
Source Address 5	

BOS = 0, WTS = 0 in a Data Configuration Register.

This format is used with 8-bit processors.

Note: All examples above will result in a transmission of a packet in order of DA0, DA1, DA3 . . . bits within each byte will be transmitted least significant bit first

DA = Destination Address

6.5 LOOPBACK DIAGNOSTICS

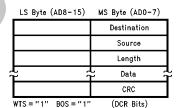
Three forms of local loopback are provided on the AT/LANTIC Controller. The user has the ability to loopback through the deserializer on the controller, through the ENDEC module or tranceiver. Because of the half duplex architecture of the AT/LANTIC Controller, loopback testing is a special mode of operation with the following restrictions:

Restrictions during Loopback

The FIFO is split into two halves, one half is used for transmission the other for reception. Only 8-bit fields can be fetched from memory so two tests are required for 16-bit systems to verify integrity of the entire data path. During loopback the maximum latency to obtain access to the buffer memory is 2.0 μ s. Systems that wish to use the loopback test yet do not meet this latency can limit the loopback packet to 7 bytes without experiencing underflow. Only the last 8 bytes of the loopback packet are retained in the FIFO. The last 8 bytes can be read through the FIFO register which will advance through the FIFO to allow reading the receive packet sequentially.

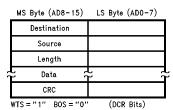
Destination Address	= 6 bytes Station Physical Address
Source Address	= 6 bytes Station Physical Address
Length	2 bytes
Data	= 46 to 1500 bytes
CRC	Appended by AT/LANTIC Controller if CRC = 0 in TCF

When in word-wide mode with Byte Order Select set, the loopback packet must be assembled in the even byte locations as shown below. (The loopback only operated with byte wide transfers.)



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When in word-wide mode with Byte Order Select low, the following format must be used for the loopback packet.



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Note: When using loopback in word mode 2n bytes must be programmed in the TBCR0, 1. When n = actual number of bytes assembled in even or odd location.

To initiate a loopback the user first assembles the loopback packet then selects the type of loopback using the Transmit Configuration register bits LB0, LB1. The transmit configuration register must also be set to enable or disable CRC generation during transmission. The user then issues a normal transmit command to send the packet. During loopback the receiver checks for an address match and if CRC bit in the TCR is set, the receiver will also check the CRC. The last 8 bytes of the loopback packet are buffered and can read out of the FIFO using FIFO read port.

Loopback Modes

MODE1: Loopback through the AT/LANTIC Controller Module (LB1 = 0, LB0 = 1): If this loopback is used, the AT/LANTIC Controller Module's serializer is connected to the deserializer.

MODE 2: Loopback through the ENDEC Module (LB1 = 1, LB0 = 0): If the loopback is to be performed through the SNI, the AT/LANTIC Controller provides a control (LPBK) that forces the ENDEC module to loopback all signals.

MODE 3: Loopback to the external coax interface or twisted pair interface module (LB1 = 1, LB0 = 1). Packets can be transmitted to the cable in loopback mode to check all of the transmit and receive paths and the cable itself. If, in twisted pair mode, there is a link fail the transmitter will be disabled which could give misleading results in Mode 3. The link integrity should be checked, by reading Configuration Register B. before this test.

Note: Collision and Carrier Sense can be generated by the ENDEC module and are masked by the NIC module, it is not possible to go directly between the loopback modes, it is necessary to return to normal operation (00H) when changing modes.

Reading the Loopback Packet

The last eight bytes of a received packet can be examined by 8 consecutive reads of the FIFO register. The FIFO pointer is increment after the rising edge of the CPU's read strobe by internally synchronizing and advancing the pointer. This may take up to four bus clock cycles, if the pointer has not been incremented by the time the CPU reads the FIFO register again, the AT/LANTIC Controller will insert wait states.

Note: The FIFO may only be read during Loopback. Reading the FIFO at any other time will cause the AT/LANTIC Controller to malfunction.

Alignment of the Received Packet in the FIFO

Reception of the packet in the FIFO begins at location zero, after the FIFO pointer reaches the last location in the FIFO, the pointer wraps to the top of the FIFO overwriting the previously received data.

This process is continued until the last byte is received. The AT/LANTIC Controller then appends the received byte count in the next two locations of the FIFO. The contents of the Upper Byte Count are also copied to the next FIFO location. The number of bytes used in the loopback packet determined the alignment of the packet in the FIFO. The alignment for a 64-byte packet is shown below.

FIFO Location	FIFO Contents		
0	Lower Byte Count	\rightarrow	First Byte Read
1	Upper Byte count	\rightarrow	Second Byte Read
2	Upper Byte Count		•
3	Last Byte		•
4	CRC1		•
5	CRC2		•
6	CRC3		•
7	CRC4	\rightarrow	Last Byte Read

For the following alignment in the FIFO the packet length should be (N \times 8) +5 Bytes. Note that if the CRC bit in the TCR is set, CRC will not be appended by the transmitter. If the CRC is appended by the transmitter, the 1st four bytes, bytes N-3 to N, correspond to the CRC.

FIFO Location	FIFO Contents		
0	Byte N-4	\rightarrow	First Byte Read
1	Byte N-3 (CRC1)	\rightarrow	Second Byte Read
2	Byte N-2 (CRC2)		•
3	Byte N-1 (CRC3)		•
4	Byte N (CRC4)		•
5	Lower Byte Count		•
6	Upper Byte Count		•
7	Upper Byte Count	\rightarrow	Last Byte Read

Loopback Tests

Loopback capabilities are provided to allow certain tests to be performed to validate operation of the AT/LANTIC Controller prior to transmitting and receiving packets on a live network. Typically these tests may be performed during power up of a node. The diagnostic provides support to verify the following:

- Verify integrity of data path. Received data is checked against transmitted data.
- Verify CRC logic's capability to generate good CRC on transmit, verify CRC on receive (good or bad CRC).
- 3. Verify that the Address Recognition Logic can
 - a. Recognize address match packets
 - b. Reject packets that fail to match an address

Loopback Operation in the AT/LANTIC Controller

Loopback is a modified form of transmission using only half of the FIFO. This places certain restrictions on the use of

loopback testing. When loopback mode is selected in the TCR, the FIFO is spilt. A packet should be assembled in memory with programming of TPSR and TBCR0, TBCR1 registers. When the transmit command is issued the following operations occur:

TRANSMITTER ACTIONS

- Data is transferred from memory by the DMA until the FIFO is filled. For each transfer TBCR0 and TBCR1 are decremented. (Subsequent burst transfers are initiated when the number of bytes in the FIFO drops below the programmed threshold.)
- 2. The AT/LANTIC Controller generates 56 bits of preamble followed by an 8-bit synch pattern.
- 3. Data transferred from FIFO to serializer.
- 4. If CRC = 1 in TCR, no CRC calculated by AT/LANTIC Controller, the last byte transmitted is the last byte from the FIFO (allows software CRC to be appended). If CRC = 0, AT/LANTIC Controller calculates and appends four bytes of CRC.
- 5. At end of Transmission PTX bit set in ISR.

RECEIVER ACTIONS

- 1. Wait for synch, all preamble stripped.
- Store packet in FIFO, increment receive byte count for each incoming byte.
- If CRC = 0 in TRC, receiver checks incoming packet for CRC errors. If CRC = 1 in TCR, receiver does not check CRC errors, CRC error bit always set in RSR (for address matching packets).
- 4. At end of receive, receive byte count written into FIFO, receive status register is updated. The PRX bit is typically set in the RSR even if the address does not match. If CRC errors are forced, the packet must match the address filters in order for the CRC error bit in the RS to be set.

EXAMPLES

The following examples show what results can be expected from a properly operating AT/LANTIC Controller during loopback. The restrictions and results of each type of loopback are listed for reference. The loopback tests are divided into two sets of tests. One to verify the data path, CRC generation and byte count through all three paths. The second set of tests uses internal loopback to verify the receiver's CRC checking and address recognition. For all of the tests the DCR was programmed to 40H.

Path	TCR	RCR	TSR	RSR	ISR
AT/LANTIC	02	1F	53	02	02
Controller			(Note 1)	(Note 2)	(Note 3)
Internal					

- Note 1: Since carrier sense and collision detect are generated in the ENDEC module. They are blocked during internal loopback, carrier and CD heartbeat are not seen and the CRS and CDH bits are set.
- Note 2: CRC errors are always indicated by receiver if CRC is appended by the transmitter.
- Note 3: Only the PTX bit in the ISR is set, the PRX bit is only set if status is written to memory. In loopback this action does not occur and the PRx bit remains 0 for all loopback modes.
- Note 4: All values are hex.

Path	TCR	RCR	TSR	RSR	ISR
AT/LANTIC Controller Internal	04	1F	43 (Note 1)	02	02

Note 1: CDH is set, CRS is not set since it is generated by the external

Path	TCR	RCR	TSR	RSR	ISR
AT/LANTIC	06	1F	03	02	02
Controller			(Note 1)		(Note 2)
External					

Note 1: CDH and CRS should not be set. The TSR however, could also contain 01H,03H,07H and a variety of other values depending on whether collisions were encountered or the packet was deferred.

Note 2: Will contain 08H if packet is not transmittable.

Note 3: During external loopback the AT/LANTIC Controller is now exposed to network traffic, it is therefore possible for the contents of both the Receive portion of the FIFO and the RSR to be corrupted by any other packet on the network. Thus in a live network the contents of the FIFO and RSR should not be depended on. The AT/LANTIC Controller will still abide by the standard CSMA/CD protocol in external loopback mode (i.e. the network will not be disturbed by the loopback packet).

Note 4: All values are her

CRC and Address Recognition

The next three tests exercise the address recognition logic and CRC. These tests should be performed using internal loopback only so that the AT/LANTIC Controller is isolated from interference from the network. These tests also require the capability to generate CRC in software.

The address recognition logic cannot be directly tested. The CRC and FAE bits in the RSR are only set if the address in the packet matches the address filters. If errors are expected to be set and they are not set, the packet has been rejected on the basis of an address mismatch. The following sequence of packets will test the address recognition logic. The DCR should be set to 40H, the TCR should be set to 03H with a software generated CRC.

Packet Contents		Packet Contents F		
Test Address		CRC	RSR	
Test A	Matching	Good	01 (Note 1)	
Test B	Matching	Bad	02 (Note 2)	
Test C			01	

Note 1: Status will read 21H if multicast address used.

Note 2: Status will read 22H if multicast address used.

Note 3: In test A, the RSR is set up. In test B the address is found to match since the CRC is flagged as bad. Test C proves that the address recognition logic can distinguish a bad address and does not notify the RSR of the bad CRC. The receiving CRC is proven to work in test A and test B.

Note 4: All values are hex.

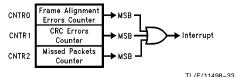


FIGURE 36. Tally Counters

Network Management Functions

Network management capabilities are required for maintenance and planning of a local area network. The AT/LANTIC Controller supports the minimum requirement for network management in hardware, the remaining requirements can be met with software. Software alone can not track during reception of packets: CRC errors, Frame Alignment errors, and missed packets, *Figure 36*.

Since errored packets can be rejected, the status associated with these packets is lost unless the CPU can access the Receive Status Register before the next packer arrives. In situations where another packet arrives very quickly, the CPU may have no opportunity to do this. The AT/LANTIC Controller counts the number of packets with CRC errors and Frame Alignment errors. 8-bit counters have been selected to reduce overhead. The counters will generate interrupts whenever their MSBs are set so that a software routine can accumulate the network statistics and reset the counter before overflow occurs. The counters are sticky so that when they reach a count of 192 (COH) counting is halted. An additional counter is provided to count the number of packets the AT/LANTIC Controller misses due to buffer overflow or being offline.

The structure of the counters is shown in Figure 36.

Additional information required for network management is available in the Receive and Transmit Status Registers. Transmit status is available after each transmission for information regarding events during transmission.

Typically, the following statistics might be gathered in software:

Traffic: Frames Sent OK

Frames Received OK

Multicast Frames Received

Packets Lost Due to Lack of Resources

Retries/Packet

Errors: CRC Errors

Alignment Errors
Excessive Collisions
Packet with Length Errors

Heartbeat Failure

6.6 MEMORY ARBITRATION AND BUS OPERATION

The AT/LANTIC Controller will always operate as a slave device on its peripheral interface to the ISA bus. However on the memory bus, the AT/LANTIC Controller operates in three possible modes:

- 1. Bus Master of Local Packet Buffer RAM
- Bus Slave when accessed by the CPU via the Bus Interface
- 3. Idle, when no activity is occurring

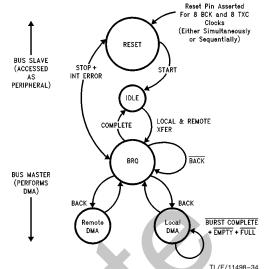


FIGURE 37. DP8390 Core Bus States

Upon power-up the AT/LANTIC Controller is in an indeterminate state. After receiving a hardware reset the AT/LANTIC Controller is a bus slave in the Reset State, the receiver and transmitter are both disabled in this state. The reset state can be re-entered under four conditions, soft reset (Stop Command), register reset (reset port in I/O mode, bit in Control Register 1 in shared memory mode), hard reset (RESET input) or an error that shuts down the receiver or transmitter (FIFO underflow or overflow, receive buffer ring overflow).

After initialization of registers, the AT/LANTIC Controller is issued a Start command and the AT/LANTIC Controller enters Idle state. Until the DMA is required the AT/LANTIC Controller remains in idle state.

The idle state is exited and the AT/LANTIC Controller will drive the local memory bus when a request from the FIFO in the DP8390 (NIC) core causes the memory bus interface logic to issue a read or write operation, such as when the AT/LANTIC Controller is transmitting or receiving data.

In I/O mode the NIC Core's Remote DMA also requests access from the memory bus. When software programs an I/O mode data transfer between the CPU and the buffer RAM, the Remote DMA controls this request.

In Shared Memory Mode, the memory bus is accessed via the CPU interface directly.

All Local DMA transfers are burst transfers, the DMA will transfer an exact burst of bytes programmed in the Data Configuration Register (DCR) then relinquish the memory bus. If there are remaining bytes in the FIFO the next burst will not be initiated until the FIFO threshold is exceeded.

I/O Mode Operation

In I/O mode the AT/LANTIC Controller transfers data to and from the packet buffer RAM by utilizing the Remote DMA logic which is programmed by the main system CPU to transfer data through the AT/LANTIC Controller's internal data port register.

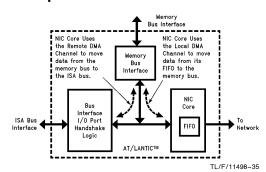


FIGURE 38. I/O Operation: All Data Transfers and Arbitration is Controlled by the NIC Core

INTERLEAVED LOCAL/REMOTE OPERATION

When in I/O mode the remote DMA is used to transfer data to/from the main system. If a remote DMA transfer is initiated or in progress when a packet is being received or transmitted, the Remote DMA transfers will be interrupted for higher priority Local DMA transfers. When the Local DMA transfer is completed the Remote DMA will rearbitrate for the bus and continue its transfers.

If the FIFO requires service while a remote DMA is in progress the Local DMA burst is appended to the Remote Transfer. When switching from a local transfer to a remote transfer there is a break to allow the CPU to fairly contend for the bus.

REMOTE DMA BI-DIRECTIONAL PORT

The Remote DMA transfers data between the local buffer memory and the internal bidirectional port (memory to I/O transfer).

This transfer is arbitrated on a transfer by transfer basis versus the burst transfer mode used for Local DMA transfers. This bidirectional port is integrated onto the AT/LAN-TIC Controller, and is read/written by the host. All transfers through this port are asynchronous. At any one time transfers are limited to one direction, either from the port to local buffer memory (Remote Write) or from local buffer memory to the port (Remote Read).

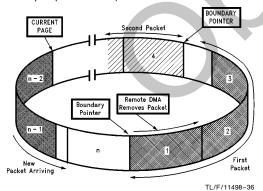


FIGURE 39. 1st Received Packet Removed by Remote DMA

I/O MODE REMOVING PACKETS FROM RING

Network activity is isolated on a local bus, where the AT/LANTIC Controller's local DMA channel performs burst transfers between the buffer memory and the AT/LANTIC Controller's FIFO. The Remote DMA transfers data between the buffer memory and the host memory via the internal bidirectional I/O port. The Remote DMA provides local addressing capability and is used as a slave DMA by the host. The host system reads the I/O port to transfer data between the system and I/O port. The AT/LANTIC Controller allows Local and Remote DMA operations to be interleaved.

Packets are removed from the ring using the Remote DMA. When using the Remote DMA the Send Packet command can be used. This programs the Remote DMA to automatically remove the received packet pointed to by the Boundary Pointer. At the end of the transfer, the AT/LANTIC Controller moves the Boundary Pointer, freeing additional buffers for reception. The Boundary Pointer can also be moved manually by programming the Boundary Register. Care should be taken to keep the Boundary Pointer at least one buffer behind the Current Page Pointer.

Data transfer by the Remote DMA to the integrated I/O data port is dependent on whether the AT/LANTIC Controller is set into 8-bit mode or 16-bit mode. In 8-bit mode all transfers are 8 bits (1 byte) wide. When in 16-bit mode all transfers are 16 bits (1 word) wide. The data width is selected by setting the WTS bit in the Data Configuration Register and setting the DWID pin for the proper mode.

The following is a suggested method for maintaining the Receive Buffer Ring pointers if in shared memory mode or if remote read is used in I/O mode.

- At initialization, set up a software variable (next_pkt) to indicate where the next packet will be read. At the beginning of each Remote Read DMA operation, the value of next_pkt will be loaded into RSAR0 and RSAR1.
- 2. When initializing the AT/LANTIC Controller set:

```
BNDRY=PSTART
CURR = PSTART + 1
next_pkt = PSTART + 1
```

 After a packet is DMAed from the Receive Buffer Ring, the Next Page Pointer (second byte in the AT/LANTIC Controller buffer header is used to update BNDRY and next_pkt.

```
next_pkt = Next Page Pointer
BNDRY = Next Page Pointer - 1
If BNDRY < PSTART then BNDRY = PSTOP - 1
```

Note the size of the Receive Buffer Ring is reduced by one 256 byte buffer, this will not, however, impede the operation of the AT/LANTIC Controller. The advantage of this scheme is that it easily differentiates between buffer full and buffer empty: it is full if BNDRY = CURR; empty when BNDRY = CURR-1. If, in I/O mode, send packet is used to empty the buffer ring this scheme cannot be used. BNDRY must be initialized equal to CURR, or the first executed send packet will not return data from the received packet, which will be written at CURR. The Overwrite Warning bit of the Interrupt Status Register must be used in this mode to differentiate between buffer full and buffer empty.

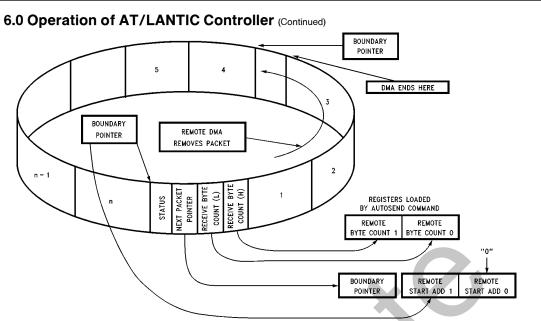


FIGURE 40. Remote DMA Autoinitialization from Buffer Ring

TL/F/11498-37

I/O MODE REMOTE DMA COMMANDS

The Remote DMA channel is used in the I/O Mode to both assemble packets for transmission, and to remove received packets from the Receive Buffer Ring. It may also be used for moving blocks of data or commands between host memory and local buffer memory. (In Shared Memory Mode, the Remote DMA should be disabled, and not used. Packet transfer to/from the system is accomplished by normal CPU read/write operations.)

There are three modes of Remote DMA operation: Remote Write, Remote Read, or Send Packet.

Two register pairs are used to control the Remote DMA, a Remote Start Address (RSAR0, RSAR1) and a Remote Byte Count (RBCR0, RBCRI) register pair. The Start Address Register pair points to the beginning of the block to be moved while the Byte Count Register pair is used to indicate the number of bytes to be transferred. Full handshake logic is provided to move data between local buffer memory and a bidirectional I/O port.

Remote Write: A Remote Write transfer is used to move a block of data from the host into local buffer memory. The Remote DMA will read data from the I/O port and sequentially write it to local buffer memory beginning at the Remote Start Address. The DMA Address will be incremented and the Byte Counter will be decremented after each transfer. The DMA is terminated when the Remote Byte Count Register reaches a count of zero.

Remote Read: A Remote Read transfer is used to move a block of data from local buffer memory to the host. The Remote DMA will sequentially read data from the local buffer memory, beginning at the Remote Start Address, and write data to the I/O port. The DMA Address will be incremented and the Byte Counter will be decremented after each transfer. The DMA is terminated when the Remote Byte Count Register reaches zero.

Send Packet Command: The Remote DMA channel can be automatically initialized to transfer a single packet from the Receive Buffer Ring. The CPU begins this transfer by issuing a "Send Packet" Command. The DMA will be initialized to the value of the Boundary Pointer Register and the Remote Byte Count Register pair (RBCR0, RBCR1) will be initialized to the value of the Receive Byte Count fields found in the Buffer Header of each packet. After the data is transferred, the Boundary Pointer is advanced to allow the buffers to be used for new receive packets. The Remote Read will terminate when the Byte Count equals zero. The Remote DMA is then prepared to read the next packet from the Receive Buffer Ring. If the DMA pointer crosses the Page Stop Register, it is reset to the Page Start Address. This allows the Remote DMA to remove packets that have wrapped around to the top of the Receive Buffer Ring.

Note 1: In order for the AT/LANTIC Controller to correctly execute the Send Packet command, the upper Remote Byte Count Register (RBCR1) must first be loaded with 0FH.

Note 2: The Send Packet command cannot be used with 680x0 type processors.

I/O MODE READ TIMING

- The DMA reads a word from local buffer memory and writes the word into the internal latch, increments the DMA address and decrements the byte count (RBCR0,1).
- Internally a request line is asserted to enable the system to read the port. If the system reads this port before the data has been written, then the system is sent a wait signal to wait until the data has been written to the port. Once written the system's read is allowed to complete.
- The system reads the port, the read strobe for the port is used as an acknowledge to the Remote DMA and it goes back to step 1.

Steps 1-3 are repeated until the remote DMA is complete (i.e. the byte count has gone to zero).

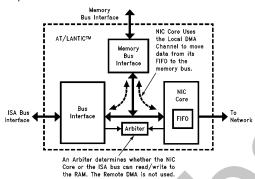
Note that in order for the Remote DMA to transfer a word from memory to the latch, it must arbitrate access to the local buffer RAM. After each word is transferred to the internal latch, access to the RAM is relinquished. If a Local DMA is in progress, the Remote DMA is held off until the local DMA is complete.

I/O MODE WRITE TIMING

A Remote Write operation transfers data from the I/O port to the local buffer RAM. The system transfers a byte-word to the latch via IOWR. This write strobe is detected by the AT/LANTIC Controller and the byte/word is transferred to local buffer memory. The Remote DMA holds off further transfers into the latch until the current byte/word has been transferred from the latch.

- AT/LANTIC Controller awaits data to be written by the system. System writes byte/word into latch.
- Remote DMA reads contents of port and writes byte/ word to local buffer memory, increments address and decrements byte count (RBCR0,1).
- 3. Go back to step 1.

Steps 1-3 are repeated until the remote DMA is complete.



TL/F/11498-38

FIGURE 41. Shared Memory Mode the ISA Bus Directly Access the RAM

Shared Memory Mode Operation

In shared memory mode the AT/LANTIC Controller transfers data to or from the packet buffer RAM directly from or to the ISA bus. The buffer RAM is mapped into system memory with the AT/LANTIC Controller doing all address decoding, synchronization and handshaking.

INTERLEAVED SHARED/LOCAL OPERATION

When in shared memory mode the local DMA is used to transfer data to or from the FIFO in the NIC core and ultimately the network. If a local DMA transfer is in progress when a shared memory access occurs the system is sent a wait state signal until the local DMA has been completed. If the shared memory access begins first then it will be completed before any local DMA is allowed.

SHARED MEMORY HOST DATA TRANSFER

In Shared Memory Mode the system reads data from the RAM directly, usually using memory string move instructions. The memory is enabled by setting D6 of Shared Memory Control Register 1. The base address of the memory is programmed by writing to the Control Registers.

If DWID is low only Control Register 1 is used to program base address, so the memory must exist in the lower 1 Mbytes of system memory. A19 is always Compared to a 1 when DWID is low. The A13–18 bits are compared to the address lines, if there is 8k of memory. A13 is not compared in 16k mode, A13–14 are not compared in 32k mode (8-bit non-compatible) and A13–15 are not compare in 64k mode (16-bit non-compatible).

If DWID is high both Control Registers must be programmed to set the base address, so the memory can exist anywhere in up to 16 Gbytes of system memory. LA19 can be either 1 or 0. The same limited decode, as detailed above, also occurs depending on the memory size.

SHARED MEMORY READ TIMING

The system executes a normal memory read cycle which the AT/LANTIC Controller will complete immediately, if idle, or insert wait states into if local DMA is current. The byte or word of data is fetched from the buffer RAM via the memory support bus.

SHARED MEMORY WRITE TIMING

The system executes a normal memory write cycle which the AT/LANTIC Controller will complete immediately, if idle, or insert wait states into if local DMA is current. The byte or word of data is written to the buffer RAM via the memory support bus.

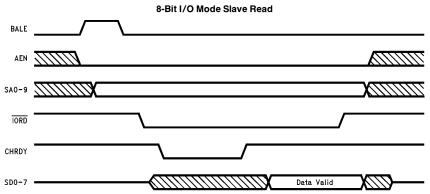
6.7 FUNCTIONAL BUS TIMING

This section describes the bus cycles that the AT/LANTIC Controller performs. These timings can be subdivided into 3 basic categories:

- ISA I/O Access: There are register accesses in both modes, and I/O data accesses in I/O mode.
- 2. Shared RAM ISA Accesses: These are the timing for the ISA bus accesses through the AT/LANTIC Controller to the memory bus and buffer RAM.
- 3. Boot PROM ISA Accesses: These are the timing for the ISA bus accesses through the AT/LANTIC Controller to the memory bus and boot PROM.
- Local and I/O RAM Accesses: This is the timing of the Local DMA, accesses from the NIC Core FIFO to the RAM, and the Remote DMA accesses to the RAM over the memory bus.

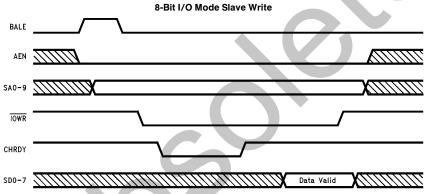
ISA Bus I/O Accesses

The AT/LANTIC Controller is designed to directly interface to the ISA bus (PC-AT backplane bus). The CPU can read or write any internal registers. All register accesses are byte wide. The functional timing for AT/LANTIC Controller accesses are shown in the following pages.



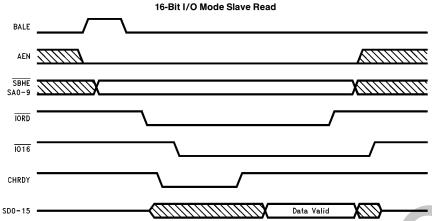
TL/F/11498-39

This is the type of cycle used to read from a register or, in 8-bit I/O mode, from a data transfer port. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0-9 and an $\overline{\text{IORD}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The data will always appear on SD0-7.



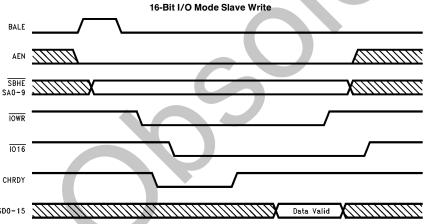
TL/F/11498-40

This is the type of cycle used to write to a register or, in 8-bit I/O mode, to a data transfer port. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0-9 and an $\overline{\text{IOWR}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The data will always be taken from SD0-7.



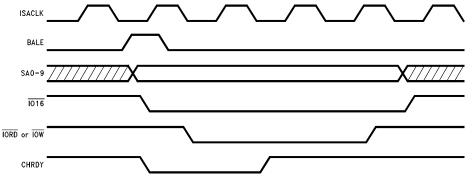
TL/F/11498-41

This is the type of cycle used to read from a data transfer port in 16-bit I/O mode. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0-9 and an $\overline{\text{IORD}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. $\overline{\text{IO16}}$ is generated, when an address within the AT/LANTIC Controller's data transfer port is decoded, to indicate to the system that this is a 16-bit transfer. If the IO16CON bit in Configuration Register B is low then it will be a straight decode of the SA0-9 lines. If that bit is high the $\overline{\text{IO16}}$ output will be generated after $\overline{\text{IORD}}$ goes active. $\overline{\text{SBHE}}$ must be low, to indicate that this is a 16-bit transfer, and the address should be even, SA0 low. The data will appear on SD0-15.



TL/F/11498-42

This is the type of cycle used to write to a data transfer port in 16-bit I/O mode. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0-9 and an $\overline{\text{IOWR}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. $\overline{\text{IO16}}$ is generated, when an address within the AT/LANTIC Controller's data transfer port is decoded, to indicate to the system that this is a 16-bit transfer. If the IO16CON bit in Configuration Register B is low then it will be a straight decode of the SA0-9 lines. If that bit is high the $\overline{\text{IO16}}$ output will be generated after $\overline{\text{IOWR}}$ goes active. $\overline{\text{SBHE}}$ must be low, to indicate that this is a 16-bit transfer, and the address should be even, SA0 low. The data will be taken from SD0-15.



16-Bit I/O Cycle with CHRDY Fix

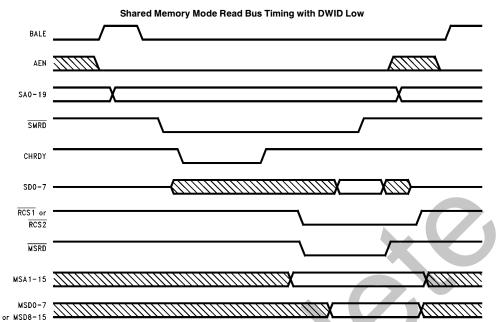
TI /F/11498-63

Some Chips and Technologies and VLSI Technologies PC-AT chip sets have timing requirements in 16-bit I/O cycles that cannot be achieved by the default AT/LANTIC cycle, described on the previous page. When that cycle is executed with these chip sets the system does not recognize the CHRDY signal and does not insert wait states. The system executes a standard cycle and deasserts IORD or IOW even if CHRDY is still deasserted. The AT/LANTIC recognizes if this situation has occurred, asserts CHRDY and sets a bus error bit in Configuration Register B to flag this error. Thus the user can test any new system to see if this error occurs and then take some remedial action. There are two ways of overcoming this problem, which are implemented by various board vendors. The AT/LANTIC supports both methods to allow the user to decide. Either fix can be selected by software, by writing to Configuration Register B.

The first fix is enabled by setting the IO16-bit of Configuration Register B. In normal operation any time a valid address exists on SA0-9 IO16 is generated. Delaying IO16 until after the IORD or IOW can cure the problem on non-compliant machines. The theory is that the system is fooled into thinking an 8-bit peripheral is responding, since IO16 is not generated for the valid address, and accepts 8-bit I/O cycle timings for CHRDY. It then rechecks IO16 after the IORD or IOW strobe and correctly determines it is a 16-bit peripheral. If a system did not recheck IO16 it would generate 2 8-bit cycles instead of 1 16-bit cycle. The AT/LANTIC would interpret each 8-bit access as a 16-bit transfer and decrement it's DMA byte count by 2. Eventually the system would attempt to access the data transfer port when the AT/LANTIC had finished transferring data and CHRDY would be deasserted indefinitely. To prevent misoperation, this fix should only be implemented on systems that require it.

The above figure shows the second fix to the problem with non-compliant machines. It is enabled by setting the CHRDY bit of Configuration Register B. This approach works on the theory that CHRDY deassertion is not fast enough and should be faster. In fact, it must be deasserted before the IORD or IOW strobe to operate correctly in some machines. All of the signals shown above are the same as a normal 16-bit I/O cycle, except CHRDY. BALE goes active and the address becomes valid after a falling edge of ISACLK. This causes the AT/LANTIC to generate IO16 if the address decodes to the data transfer port. BALE goes inactive after the next rising edge of ISACLK and IORD or IOW is asserted after the following falling edge. Normally CHRDY would be deasserted after the IORD or IOW strobe, if the AT/LANTIC was not ready. With this fix implemented CHRDY is deasserted as soon as the address becomes valid and BALE is active. If a memory cycle is in operation, instead of an I/O, CHRDY is asserted after the command strobe (MRD, MWR, SMRD or SMWR). If the address becomes invalid CHRDY is asserted. To prevent CHRDY being asserted for the half clock between BALE going inactive and IORD or IOW going active the AT/LANTIC holds CHRDY asserted as long as ISACLK is high between these signals. If the delay between the falling edge of ISACLK and the falling edge of IORD or IOW is too great, there may be a period where CHRDY is not held deasserted. This should not cause a problem. To prevent misoperation, this fix should only be implemented on systems that require it.

ISA Bus Shared Memory Access Timing



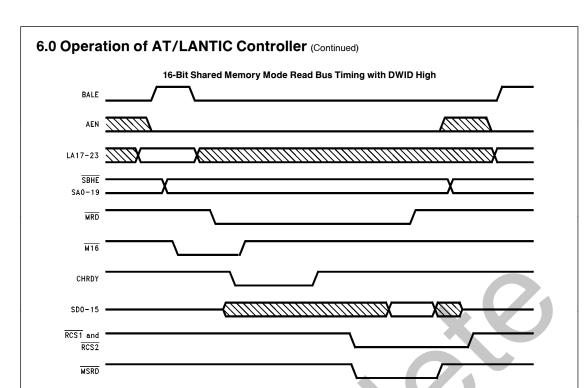
L/F/11498-43

This is the type of cycle used to read from buffer RAM in shared memory mode when DWID is low. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the address on SA0–19 matches Control Register 1 and an SMRD. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. If the memory width bit in Control Register 2 is low then 8 kbytes of RAM are accessible, so only RCS1 is used to strobe data and the data is always on MSD0–7. If this bit is high 16 kbytes of RAM are accessible, so both chip selects and byte lanes are used. If the memory address is even RCS1 and MSD0–7 are used, if odd RCS2 and MSD8–15 are used. System data is always output on SD0–7.

Shared Memory Mode Write Bus Timing with DWID Low BALE AEN SA0-19 SMWR CHRDY SD0-7 RCS1 or RCS2 MSWR MSA1-15 MSD0-7 or MSD0-7 or MSD0-15

TL/F/11498-44

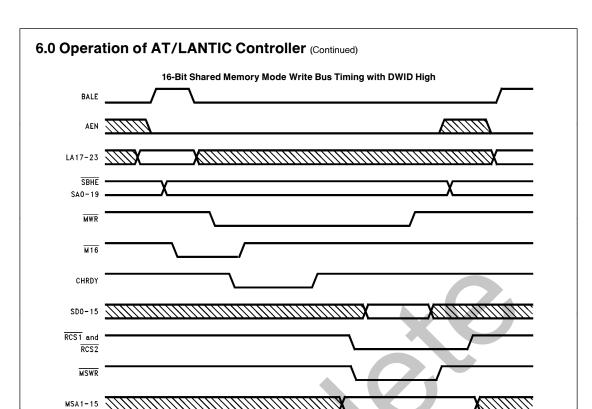
This is the type of cycle used to write to buffer RAM in shared memory mode when DWID is low. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the address on SA0–19 matches Control Register 1 and an $\overline{\text{SMWR}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. If the memory width bit in Control Register 2 is low then 8 kbytes of RAM are accessible, so only $\overline{\text{RCS1}}$ is used to strobe data and the data is always on MSD0–7. If this bit is high 16 kbytes of RAM are accessible, so both chip selects and byte lanes are used. If the memory address is even $\overline{\text{RCS1}}$ and MSD0–7 are used, if odd $\overline{\text{RCS2}}$ and MSD8–15 are used. System data is always taken from SD0–7.



L/F/11498-45

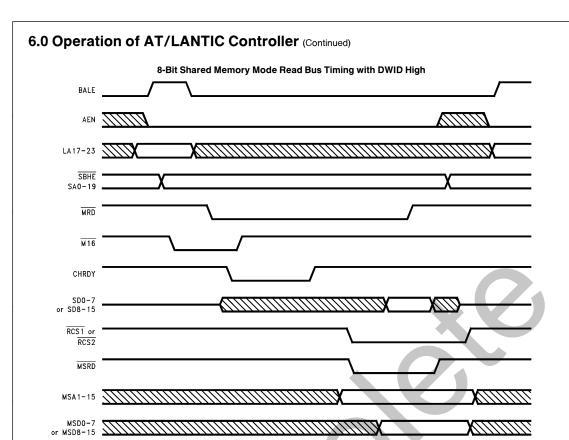
This is the type of cycle used to read 16 bits from buffer RAM is shared memory mode when DWID is high. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on LA17–23, SA0–19 and a MRD. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The 8/16-bit in Control Register 2 must be set to allow generation of M16. This will be generated whenever the LA17–23 lines match the corresponding values in Control Register 2. It will therefore be generated for a full 128 kbytes of address space, although the AT/LANTIC Controller will occupy less than that. It may be preferable to only set the 8/16-bit for the duration of a transfer from the buffer RAM. The AT/LANTIC Controller will also compare the address line programmed in Control Register 1 before allowing accesses to buffer RAM and therefore do a complete decode. The system indicates that this is a 16-bit transfer by asserting SBHE and accessing an even address, SA0 low. The full 16 bits of data bus are used on both system and memory support busses.

MSA1-15



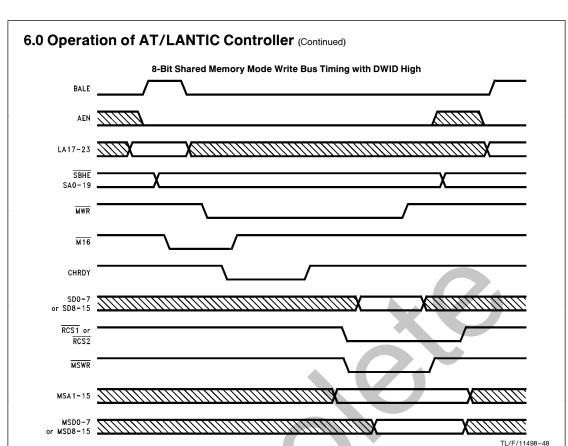
ΓL/F/11498-46

This is the type of cycle used to write 16 bits to buffer RAM in shared memory mode when DWID is high. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on LA17–23, SA0–19 and a $\overline{\text{MWR}}$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The 8/16-bit in Control Register 2 must be set to allow generation of $\overline{\text{M16}}$. This will be generated whenever the LA17–23 lines match the corresponding values in Control Register 2. It will therefore be generated for a full 128 kbytes of address space, although the AT/LANTIC Controller will occupy less than that. It may be preferable to only set the 8/16-bit for the duration of a transfer to the buffer RAM. The AT/LANTIC Controller will also compare the address line programmed in Control Register 1 before allowing accesses to buffer RAM and therefore do a complete decode. The system indicates that this is a 16-bit transfer by asserting $\overline{\text{SBHE}}$ and accessing an even address, SA0 low. The full 16 bits of data bus are used on both system and memory support busses.



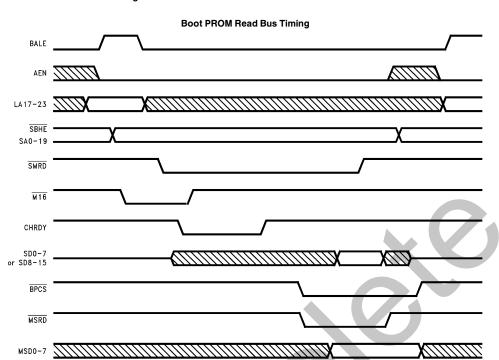
L/F/11498-47

This is the type of cycle used to read 8 bits from buffer RAM in shared memory mode when DWID is high. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on LA17–23, SA0–19 and a MRD. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The 8/16-bit in Control Register 2 must be set to allow generation of M16. This will be generated whenever the LA17–23 lines match the corresponding values in Control Register 2. It will therefore be generated for a full 128 kbytes of address space, although the AT/LANTIC Controller will occupy less than that. It may be preferable to only set the 8/16-bit for the duration of a transfer from the buffer RAM. The AT/LANTIC Controller will also compare the address line programmed in Control Register 1 before allowing accesses to buffer RAM and therefore do a complete decode. The system indicates that this is an 8-bit transfer by not asserting SBHE for an even address, SA0 low, or by accessing an odd address, SA0 high. If the 8/16-bit is low the AT/LANTIC Controller will only drive data onto SD0–7. Even addresses will use RCS1 and MSD0–7, odd addresses will use RCS2 and MSD8–15. If 8/16-bit is high the AT/LANTIC Controller can drive either SD0–7 or SD8–15. Even addresses are fetched using RCS1 and MSD0–7 and driven onto SD0–15.



This is the type of cycle used to write 8 bits to buffer RAM in shared memory mode when DWID is high. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on LA17–23, SA0–19 and a $\overline{\text{IORD}}$. If AEN is high the cycle will be ignored, CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. The 8/16-bit in Control Register 2 must be set to allow generation of $\overline{\text{M16}}$. This will be generated whenever the LA17–23 lines match the corresponding values in Control Register 2. It will therefore be generated for a full 128 kbytes of address space, although the AT/LANTIC Controller will occupy less than that. It may be preferable to only set the 8/16-bit for the duration of a transfer to the buffer RAM. The AT/LANTIC Controller will also compare the address line programmed in Control Register 1 before allowing accesses to buffer RAM and therefore do a complete decode. The system indicates that this is an 8-bit transfer by not asserting $\overline{\text{SBHE}}$ for an even address, SA0 low, or by accessing an odd address, SA0 high, If the 8/16-bit is low the AT/LANTIC Controller will only read data from SD0–7. Even addresses will use $\overline{\text{RCS1}}$ and MSD0–7, odd addresses will use $\overline{\text{RCS2}}$ and MSD8–15. If 8/16-bit is high the AT/LANTIC Controller can read from either SD0–7 or SD8–15. Even addresses are read from SD0–7 and written to RAM using $\overline{\text{RCS1}}$ and MSD0–7. Odd addresses are read from SD0–7 and written to RAM using $\overline{\text{RCS1}}$ and MSD0–7. Odd addresses are read from SD0–7 and written to RAM using $\overline{\text{RCS2}}$ and MSD0–15.

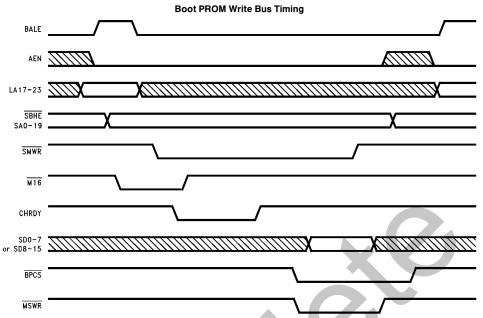
ISA Bus Boot PROM Access Timing



TL/F/11498-49

This is the type of cycle used to read the boot PROM. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0–19 and a \$\overline{SMRD}\$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. \$\overline{M16}\$ is only generated if the AT/LANTIC Controller is 1) in shared memory mode AND 2) DWID is high AND 3)8/16-bit in Control Register 2 is high AND 4) the LA17–23 lines match the corresponding values in Control Register 2. The data will normally be driven onto SD0–7. However, if \$\overline{M16}\$ is generated and the access is to an odd address the data will be driven onto SD8–15. The data will always be taken from MSD0–7.





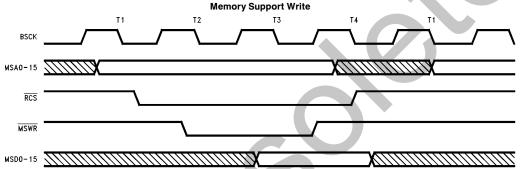
TL/F/11498-50

This is the type of cycle used to write to the boot PROM. These accesses are entirely asynchronous, with the AT/LANTIC Controller responding when it decodes the correct address on SA0-19 and a \$\overline{SMWR}\$. If AEN is high the cycle will be ignored. CHRDY is deasserted if the AT/LANTIC Controller is not ready to respond and asserted when ready. If it is ready immediately CHRDY is not deasserted. \$\overline{M16}\$ is only generated if the AT/LANTIC Controller is 1) in shared memory mode AND 2) DWID is high AND 3) 8/16-bit in Control Register 2 is high AND 4) the LA17-23 lines match the corresponding values in Control Register 2. The data will normally be taken from SD0-7. However, if \$\overline{M16}\$ is generated and the access is to an odd address the data will be taken from SD8-15. The data will always be driven onto MSD0-7. The BPWR bit of Configuration Register B must be high to allow write cycles to the boot PROM.

RAM Access Timing

TL/F/11498-51

This is a memory read cycle executed by the AT/LANTIC Controller's internal DMA. This is used to either load the data transfer port, during a Remote Read in I/O mode, or to load the FIFO, for a transmission on the network, in both modes. This transfer is synchronized to BSCLK, which can be either driven from the 20 MHz input on X1 or by the BSCLK input. This is selected by the CLKSEL bit in Configuration Register C. If there is 8 kbytes of RAM only RCS1 is used, if 16 kbytes are available RCS1 and RCS2 are used.



TL/F/11498-52

This is a memory write cycle executed by the AT/LANTIC Controller's internal DMA. This is used to either write from the data transfer port, during a Remote Write in I/O mode, or to empty the FIFO, during a reception from the network, in both modes. This transfer is synchronized to BSCLK, which can be either driven from the 20 MHz input on X1 or by the BSCLK input. This is selected by the CLKSEL bit in Configuration Register C. If there is 8 kbytes of RAM only RCS1 is used, if 16 kbytes are available RCS1 and RCS2 are used.

7.0 Preliminary Electrical Characteristics

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Operating Conditions

Preliminary DC Specifications

Symbol	Description	Conditions	Min	Max	Units
SUPPLY CU	RRENT				
Icc	Average Active (Transmitting/Receiving) Supply Current	X1 = 20 MHz Clock V _{IN} = Switching		100	mA
ICCIDLE	Average Idle Supply Current	X1 = 20 MHz Clock $V_{\text{IN}} = V_{\text{CC}} \text{ or GND}$		80	mA
I _{CCLP}	Low Power Supply Current	X1 = Undriven $V_{IN} = V_{CC} = Undriven$		35	μΑ
TTL INPUTS				•	
V _{IL}	Maximum Low Level Input Voltage			0.8	٧
V _{IH}	Minimum High Level Input Voltage		2.0		٧
I _{IN}	Input Current	$V_{I} = V_{CC}$ or GND	-1.0	+1.0	μΑ
SH TRI-ST	ATE HIGH DRIVE I/O				
V_{OH}	Minimum High Level Output Voltage	$I_{OH} = -3 \text{ mA}$	2.4		٧
V_{OL}	Maximum Low Level Output Voltage	$I_{OL} = 24 \text{ mA}$		0.5	٧
V_{IL}	Maximum Low Level Input Voltage			0.8	٧
V_{IH}	Minimum High Level Input Voltage		2.0		٧
I _{IN}	Input Current	$V_I = V_{CC}$ or GND	-1.0	+1.0	μΑ
loz	Maximum TRI-STATE Output Leakage Current	$V_{OUT} = V_{CC}$ or GND	-10.0	+10.0	μΑ
OS INPUT	S, OUTPUTS AND I/O				
V_{OH}	Minimum High Level Output Voltage	$I_{OH}=-20~\mu A$	V _{CC} - 0.1		٧
V _{OL}	Maximum Low Level Output Voltage	$I_{OL} = 20 \mu A$		0.1	٧
V_{IL}	Maximum Low Level Input Voltage			0.8	٧
V_{IH}	Minimum High Level Input Voltage		2.0		٧
V_{ILD}	Maximum Low Level Input Voltage DWID			1.0	٧
V_{IHD}	Minimum High Level Input Voltage DWID		4.0		V
I _{IN}	Input Current	$V_I = V_{CC}$ or GND	-1.0	+1.0	μΑ
I _{IND}	Input Current TEST, DWID Pull Down Resister	$V_I = V_{CC}$		2000	μΑ
I _{IN2}	Input Current TEST, MSD0-7	$V_I = V_{CC}$ or GND		2000	μΑ
I _{IN3}	Input Current TEST, MSD8-15, MSA1-8	$V_I = V_{CC}$ or GND RESET = Active		2000	μΑ
loz	Maximum TRI-STATE Output Leakage Current	$V_{OUT} = V_{CC}$ or GND	-10.0	+10.0	μΑ

7.0 Preliminary Electrical Characteristics (Continued) Preliminary DC Specifications (Continued)

Symbol	Description	Conditions	Min	Max	Units
OCH OPEN	COLLECTOR HIGH DRIVE OUTPUT				
V _{OL}	Maximum Low Level Output Voltage	$I_{OL} = 24 \text{ mA}$		0.5	V
LED DRIVE	R OUTPUT				
V _{OL}	Maximum Low Level Output Voltage	$I_{OL} = 16 \text{ mA}$		0.5	V
THIN DRIV	ER OUTPUT				
V _{OH}	Minimum High Level Output Voltage	$I_{OH} = -8 \text{ mA}$	2.4		V
V _{OL}	Maximum Low Level Output Voltage	$I_{OL} = 2 \text{ mA}$		0.5	V
OSCILLAT	OR PINS (X1 AND X2)				
V _{IH}	X1 Input High Voltage	X1 is Connected to an Oscillator	2.0		V
V _{IL}	X1 Input Low Voltage	X1 is Connected to an Oscillator		0.8	V
losc	X1 Input Current	X1 is Connected to an Oscillator V _{IN} = V _{CC} or GND	X	1	mA
AUI					
V _{OD}	Differential Output Voltage (TX±)	78Ω Termination and 270Ω from Each to GND (Note 1)	± 550	±1200	mV
V _{OB}	Differential Idle Output Voltage Imbalance (TX±)	78Ω Termination and 270Ω from Each to GND (Note 1)	Ту	/pical: 40 m	iV
V _U	Undershoot Voltage (TX±)	78Ω Termination and 270Ω from Each to GND (Note 1)	Typical: 80 mV		iV
V _{DS}	Diff. Squelch Threshold (RX±, CD±)		-175	-300	mV
V _{CM}	Diff. Input Common Mode Voltage (RX±, CD±)	(Note 1)	0	5.25	V
TPI					
R _{TOL}	TXOd ±, TXO ± Low Level Output Resistance	I _{OL} = 25 mA		15	Ω
R _{TOH}	TXOd±, TXO± High Level Output Resistance	OH = -25 mA		15	Ω
V _{SRON1}	Receive Threshold Turn-On Voltage 10BASE-T Mode		±300	± 585	mV
V _{SRON2}	Receive Threshold Turn-On Voltage Reduced Threshold		±175	±300	mV
V _{SROFF}	Receive Threshold Turn-Off Voltage	(Note 1)	± 175	±300	mV
V _{DIFF}	Differential Mode Input Voltage Range	V _{CC} = 5.0 V (Note 1)	-3.1	+3.1	V

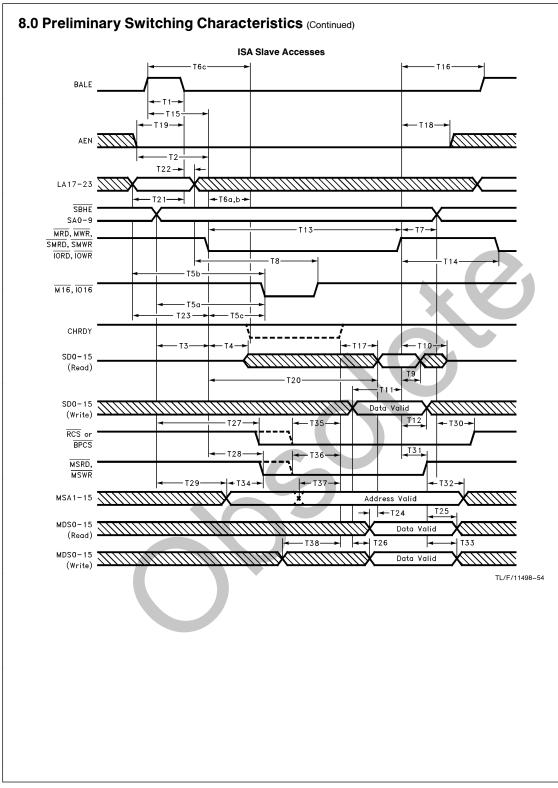
Note 1: These parameters are not guaranteed by production testing.

8.0 Preliminary Switching Characteristics

TL/F/11498-53

Symbol	nbol Description		8-Bit Transfers		16-Bit Transfers	
		Min	Max	Min	Max	
T1	MSA1 – 15 Valid before RCS Asserted (Note 1)		30		30	ns
T2	MSA1 – 15 Valid before MSRD – WR Asserted	20		20		ns
T3	MSRD – WR Width	70		70		ns
T4	RCS and MSA1 – 15 Valid to MSWR Deasserted (Note 1)	105		105		ns
T5	MSA1 – 15 Valid after MSRD – WR Deasserted	10		10		ns
Т6	RCS Held after MSRD – WR Deasserted (Note 1)	10		10		ns
T7	RCS and MSA1 – 15 Valid to MSD0 – 15 Valid (Note 1)		100		100	ns
T8	Read Data Hold from MSRD Deasserted	0		0		ns
Т9	Write Data Set-Up to MSWR Deasserted	50		50		ns
T10	Write Data Held from MSWR Deasserted	10		10		ns
T11	Time Between Transfers	4		4		bcyc
T12	Minimum Bus Clock High Time (bch)	20		20		ns
T13	Minimum Bus Clock Low Time (bcl)	20		20		ns
T14	Minimum Bus Clock Cycle Time (bcyc)	50		50		ns

Note 1: In 8-bit mode \overline{RCS} refers to $\overline{RCS1}$ only. In 16-bit mode \overline{RCS} refers to both $\overline{RCS1}$ and $\overline{RCS2}$.



ISA Slave Accesses

Symbol	Description	8-Bit Transfers		16-Bit Transfers		Units
		Min	Max	Min	Max	
T1	BALE Width	20		20		ns
T2	AEN Valid before Command Strobe Active	60		60		ns
T3a	SBHE and SA0-9 Valid before IORD, IOWR Asserted	40		20		ns
T3b	SA0-9 Valid before MRD, MWR Asserted	32		32		ns
T4a	IORD, MRD Asserted to SD0-15 Driven (Note 3)	0		0		ns
T5a	SBHE and SA0-9 Valid before IO16 Valid (Notes 1, 9)				60	ns
T5b	LA17-23 Valid to M16 Valid (Note 1)				55	ns
T5c	SBHE and SA0-9 Valid and IORD or IOWR Active before IO16 Valid (Notes 1, 10)				50	ns
T6a	IORD, IOWR Asserted to CHRDY Deasserted (Notes 2, 5)		100		50	ns
T6b	MRD, MWR Asserted to CHRDY Deasserted (Note 2)		45		45	ns
T6c	BALE Asserted and SA0-9 Valid to CHRDY Deasserted (Notes 2, 4)		60		60	ns
	IORD, IOWR Deasserted before SBHE and SA0-9 Invalid	15		15		ns
T8a	SBHE and SA0-9 Invalid to IO16 Invalid	0		0		ns
T8b	LA17-23 Invalid to M16 Invalid (Note 1)			0		ns
Т9	IORD, MRD Deasserted to SD0-15 Read Data Invalid (Note 3)	0		0		ns
T10	IORD, MRD Deasserted to SD0-15 Floating (Note 3)		45		45	ns
T11a	D0-15 Write Data Valid to IOWR Deasserted (Note 3)	60		20		ns
T12	IOWR, MWR Deasserted to SD0-15 Write Data Invalid (Note 3)	20		20		ns
T13a	IORD, IOWR Active Width (Note 8)	300		140		ns
T14a	IORD, IOWR Inactive Width	85		85		ns
T14b	SMRD, SMWR, MRD, MWR Inactive Width					ns
T15	BALE Asserted before MRD, MWR Asserted			25		ns
T16	MRD, MWR Deasserted before Next BALE Asserted			20		ns
T17	CHRDY Asserted to SD0-15 I/O Read Data Valid (Notes 2, 3, 6)		60		60	ns
T18	IORD, IOWR Deasserted before AEN Invalid	25		25		ns
T19	AEN Valid before BALE Deasserted	50		50		ns
T20	IORD Asserted to SD0-15 Read Data Valid (Notes 3 and 7)		150		90	ns
T21	LA17-23 Valid before BALE Deasserted			40		ns

ISA Slave Accesses

Symbol	Description	1	Bit sfers	16-Bit Transfers		Units
			Max	Min	Max	
T22	BALE Deasserted before LA17-23 Invalid			0		ns
T23	LA17-23 Valid before MRD, MWR Asserted			40		ns
T24	Read Data Valid on MSD0-15 to Valid on SD0-15		70		70	ns
T25	MSRD Deasserted to MSD0-15 Read Data Invalid (Note 3)	0		0		ns
T26	Write Data Valid on SD0-15 to Valid on MSD0-15		65		65	ns
T27	SA0-19 Valid to RCS or BPCS Asserted (Note 11)		55		55	ns
T28a	MRD Asserted to MSRD Asserted		60		60	ns
T28b	MWR Asserted to MSWR Asserted		120		120	ns
T29	SA0-19 Valid to MSA1-15 Valid		60		60	ns
T30	SA0-19 Invalid to RCS or BPCS Deasserted (Note 11)	0		0		ns
T31	MRD, SMRD Deasserted to MSRD Deasserted	0		0		ns
T32	MSWR Deasserted to MA1 – 15 Invalid	10		10		ns
T33	MSWR Deasserted to MSD0-15 Invalid (Note 3)	0		0		ns
T34	MSA1-15 valid before MSWR Asserted	20		20		ns
T35a	RCS Asserted to CHRDY Asserted (Note 11, 12)	80		80		ns
T35b	BPCS Asserted to CHRDY Asserted (Note 13)	175		175		ns
T36a	MSRD, MSWR Asserted to CHRDY Asserted (Note 11)	15		15		ns
T36b	MSRD, MSWR Asserted to CHRDY Asserted (Note 13)	150		150		ns
T37	MSA1-15 Valid to CHRDY Asserted (Note 11)	75		75		ns
T38a	Driving Data from SD0-15 on to MSD0-15 to CHRDY Asserted (Note 11)	60		60		ns
T38b	Driving Data from SD0-15 on to MSD0-15 to CHRDY Asserted (Note 13)	260		260		ns

Note 1: $\overline{\text{M16}}$, $\overline{\text{IO16}}$ are only asserted for 16-bit transfers.

Note 2: CHRDY is only deasserted if the NIC core can not service the access immediately. It is held deasserted until the NIC core is ready, causing the system to insert wait states.

Note 3: On 8-bit transfers only 8 bits of MSD0-15 and D0-7 are driven.

Note 4: This is the early CHRDY timing, required by some machines, where CHRDY is referenced to BALE. In this mode of operation, under certain circumstances, CHRDY will be asserted for cycles which are not for this device, i.e. memory cycles or I/O cycles where SA0-9 match our address before reaching their valid state. In such a case the time to assert CHRDY, from MRD, MWR or SA0-9 invalid, will be the same as the deassertion time specified.

Note 5: This is the standard CHRDY timing where CHRDY is asserted after $\overline{\text{IORD}}$ or $\overline{\text{IOWR}}$.

Note 6: Read data valid is referenced to CHRDY when wait states have been inserted.

Note 7: If no wait states are inserted read data valid can be measured from $\overline{\text{IORD}}$.

Note 8: This is a minimum timing with no additional wait states.

Note 9: This is the standard $\overline{\text{IO16}}$ timing where $\overline{\text{IO16}}$ is asserted after a valid address decode.

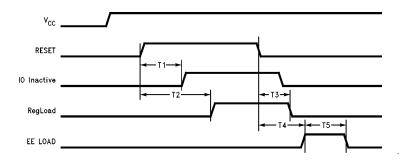
Note 10: This is the late $\overline{\text{IO16}}$ timing, required by some machines, where $\overline{\text{IO16}}$ is asserted after a valid address decode and $\overline{\text{IORD}}$ or $\overline{\text{IOWR}}$ going active.

Note 11: This is a timing for a RAM access.

Note 12: RCS refers to RCS1 and RCS2. Depending on the mode of operation either or both can be asserted. See the Functional Bus Timing section for a further explanation.

Note 13: This is a timing for a Boot PROM access.

RESET Timing



TL/F/11498-55

Symbol	Description	Min	Max	Units
T1	RESET Asserted Until IO Inactive Asserted (Note 1)	400		ns
T2	RESET Asserted Until RegLoad State Entered (Note 2)	415		μs
Т3	RESET Deasserted Until RegLoad Deasserted (Note 3)	100		ns
T4	RESET Deasserted Until EELOAD State Entered (Note 4)	0		μs
TS	EELoad Width (Note 4)		320	μs

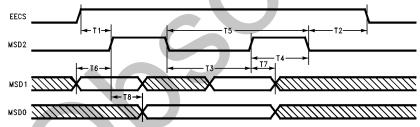
Note 1: I/O inactive is not an external signal. It is used here to indicate the length of time RESET must be active before the AT/LANTIC Controller recognizes it, begins to drive the ISA outputs to their inactive state and ignores ISA inputs except RESET.

Note 2: RegLoad is not an external signal. It is used here to indicate the length of time RESET must be active before the AT/LANTIC Controller begins contiguring. When IOinactive goes active the internal pull-down resistors on the memory support buses are enabled.

Note 3: If RegLoad is high the values on the memory support buses are latched into the configuration registers when RESET is deasserted. The pull-down resistors on this bus are enabled until RegLoad is deasserted.

Note 4: EELoad is not an external signal, it is used here to indicate when the EEPROM store is loading.

Serial EEPROM Timing

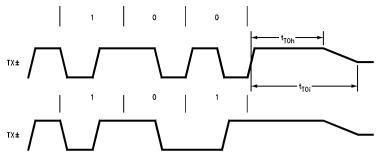


TL/F/11498-56

Symbol	Description	Min	Max	Units
T1	EECS Setup to SK	150		ns
T2	EECS Hold after SK	250		ns
T3	MSD2 Low Time	450		ns
T4	MSD2 High Time	450		ns
T5	MSD2 Clock Period (Note 1)	1		μs
T6	Data In Setup to MSD2 High	100		ns
T7	Data In Hold from MSD2 High	100		ns
Т8	Data Out Valid from MSD2 High		500	ns

Note 1: Derived from Crystal Oscillator Tolerance $= \pm 0.01\%$.

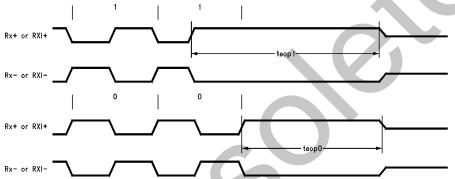
AUI Transmit Timing (End of Packet)



TL/F/11498-57

Symbol	Description	Min	Max	Units
t _{TOh}	Transmit Output High before Idle	200		ns
t _{TOI}	Transmit Output Idle Time	8000		ns

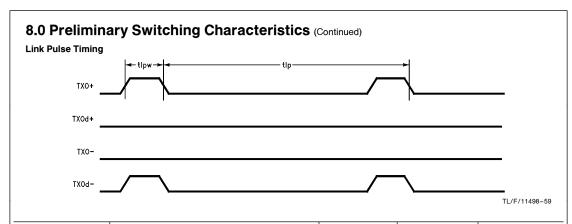
AUI/TPI Receive Timing (End of Packet)



TL/F/11498-58

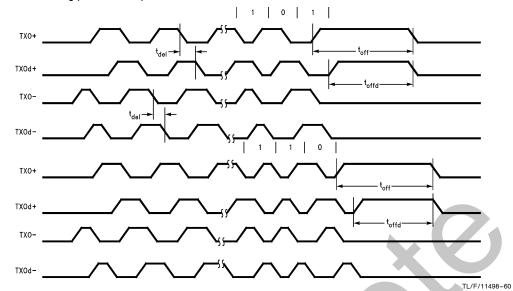
Symbol	Description	Min	Max	Units
t _{eop1}	Receive End of Packet Hold Time after Logic "1" (Note 1)	225		ns
t _{eop0}	Receive End of Packet Hold Time after Logic "0" (Note 1)	225		ns

Note 1: This parameter is guaranteed by design and is not tested.



Symbol	Description	Min	Max	Units
t _{lp}	Time between Link Output Pulses	8	24	ns
t _{lpw}	Link Integrity Output Pulse Width	80	130	ns

TPI Transmit Timing (End of Packet)



Symbol	Description	Min	Max	Units
t _{del}	Pre-Emphasis Output Delay (TXO \pm to TXOd \pm) (Note 1)	46	54	ns
t _{Off}	Transmit Hold Time at End of Packet (TXO±) (Note 1)	250		ns
t _{Offd}	Transmit Hold Time at End of Packet (TXOd±) (Note 1)	200		ns

Note 1: This parameter is guaranteed by design and is not tested.

9.0 AC Timing Test Conditions

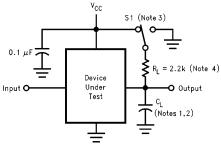
Input Pulse Levels (TTL/CMOS) GND to 3.0V
Input Rise and Fall Times (TTL/CMOS) 5 ns
Input and Output Reference Levels 1.3V

(TTL/CMOS)

Output Load (See Figure Below)

All specifications are valid only if the mandatory isolation is employed and all differential signals are taken to be at the AUI side of the pulse transformer.

Output Load (See Figure Below)



TL/F/11498-61

Note 1: Load Capacitance used depends on output type (includes scope and jig capacitance): For 3SL, MOS, TPI, AUI: $C_L=50\ pF$.

For 3SL, MOS, TPI, AUI: $C_L = 50 \text{ pF}$. For 3SH, OCH: $C_L = 240 \text{ pF}$.

Note 2: Specifications which measure delays from an active state to a high impedance state are not guaranteed by production test, but are characterized using 70 pF, and are correlated to determine true driver turn-off time by eliminating inherent R-C delay times in measurements.

Note 3: S1 = Open for timing test for push pull outputs.

S1 = V_{CC} for V_{OL} test.

 $S1 = GND \text{ for } V_{OH} \text{ test.}$

S1 = V_{CC} for High Impedance to active low and active low to High Impedance measurements.

= GND for High Impedance to active high and active high to High Impedance measurements.

Note 4: Pull-up load for CHRDY = 1 k Ω .

 $\begin{aligned} \text{IO16} &= 300\Omega.\\ \text{M16} &= 300\Omega. \end{aligned}$

Pin Capacitance $T_A = 25^{\circ}C$, f = 1 MHz

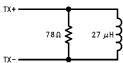
Symbol	Parameter	Тур	Units
C _{IN}	Input Capacitance	7	pF
C _{OUT}	Output Capacitance	10	pF

DERATING FACTOR

Output timing is measured with a purely capacitive load of 50 pF, or 240 pF. The following correction factor can be used for other loads (**Note:** This factor is preliminary):

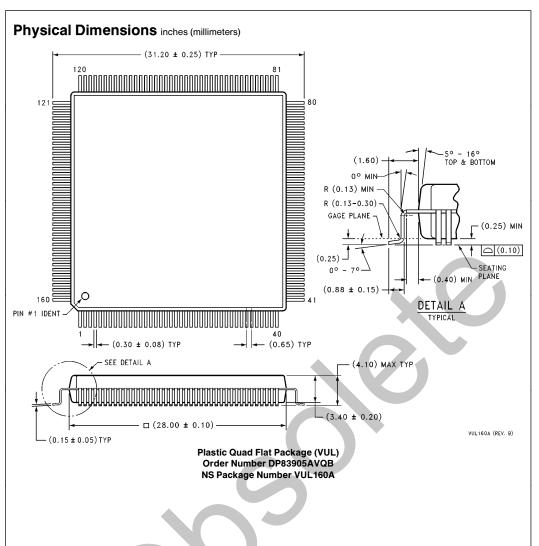
Derating for 3SL, MOS = ~ 0.05 ns/pF Derating for 3SH, OCL, TPI = ~ 0.03 ns/pF

AUI Transmit Test Load



TI /F/11498-6

Note: In the above diagram, the TX+ and TX- signals are taken from the AUI side of the isolation (pulse transformer). The pulse transformer used for all testing is a 100 μ H ±0.1% Pulse Engineering PE64103.



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