



# TPD12S015 HDMI Companion Chip With Step-Up DC-DC, I<sup>2</sup>C Level Shifter, and High-Speed ESD Clamps for Portable Applications

## 1 Features

- HDMI 1.3 and HDMI 1.4 Data Rate
- HDMI High-Speed Differential Signals –3-dB Bandwidth Exceeds 6.4 Gbps
- Excellent Matching Capacitance (0.05 pF) in Each Differential Signal Pair
- Internal Boost Converter to Generate 5 V From a 2.3-V to 5.5-V Battery Voltage
- HDMI Minimum Current Limit and Short-Circuit Protection at 5VOUT Pin
- Flexible Power-Saving Modes Through Separate Control Pins
- Auto-Direction Sensing Level Shifting in the CEC, SDA, and SCL Lines Drive up to 750-pF Load
- Seamless Type C and Type D Connector Routing With Flow-Through Pin Mapping
- IEC 61000-4-2 (Level 4) System Level ESD Compliance
- Integrated I<sub>OFF</sub> and Backdrive Current Protection
- Space-Saving 1.6-mm × 2.8-mm DSBGA (YFF) Package

## 2 Applications

- Smart Phones
- Multimedia Phones
- Digital Camcorders
- Digital Still Cameras
- Portable Game Consoles

## 3 Description

The TPD12S015 device is an integrated HDMI ESD solution. The device pin mapping matches the HDMI Type C and Type D connector with four differential pairs. This device offers eight low-capacitance ESD clamps, allowing HDMI 1.3 or 1.4 data rates. The integrated ESD clamps and resistors provide good matching between each differential signal pair, which allows an advantage over discrete ESD clamp solutions where variations between ESD clamps degrade the differential signal quality.

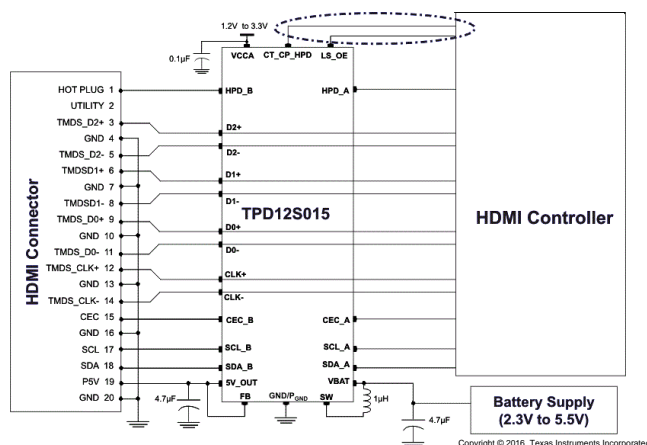
The TPD12S015 provides a regulated 5-V output (5VOUT) for sourcing the HDMI power line. The regulated 5-V output supplies up to 55 mA to the HDMI receiver. The control of 5VOUT and the hot plug detect (HPD) circuitry is independent of the LS\_OE control signal and is controlled by the CT\_CP\_HPDP pin. This independent control enables the detection scheme (5VOUT + HPD) to be active before enabling the HDMI link.

There are three noninverting, bidirectional translation circuits for the SDA, SCL, and CEC lines. Each have a common power rail (V<sub>CCA</sub>) on the A side from 1.1 V to 3.6 V. On the B side, the SCL\_B and SDA\_B each have an internal 1.75-kΩ pullup connected to the regulated 5-V rail (5VOUT). The SCL and SDA pins meet the I<sup>2</sup>C specification and drive up to 750-pF loads. The CEC\_B pin has an internal 27-kΩ pullup to an internal 3.3-V supply.

The HPD\_B port has a glitch filter to avoid false detection due to the bouncing while inserting the HDMI plug.

The TPD12S015 provides IEC61000-4-2 (Level 4) ESD protection. This device is offered in a space-saving 1.6-mm × 2.8-mm wafer-level chip scale package [DSBGA (YFF)] with a 0.4-mm pitch.

**Typical System Diagram**



**Device Information<sup>(1)</sup>**

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPD12S015	DSBGA (28)	1.56 mm × 2.76 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.



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## 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision E (June 2013) to Revision F	Page
<ul style="list-style-type: none"> <li>Added ESD Ratings table, Thermal Information table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section</li> </ul>	<b>1</b>

Changes from Revision D (April 2012) to Revision E	Page
<ul style="list-style-type: none"> <li>Updated test I<sub>OH</sub> and I<sub>OL</sub> test conditions for V<sub>OHA</sub>, V<sub>OLA</sub>, and V<sub>OHB</sub></li> <li>Updated test I<sub>OH</sub> and I<sub>OL</sub> test conditions for V<sub>OHA</sub>, V<sub>OLA</sub>, and V<sub>OHB</sub></li> </ul>	<b>9</b>

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**Changes from Revision C (November 2010) to Revision D**
**Page**

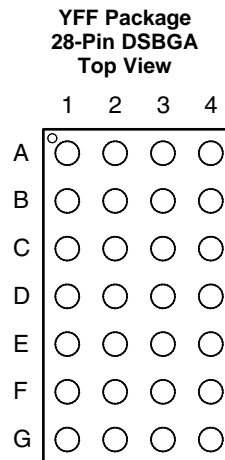
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- Changed  $V_{IH}$  MAX value for CT\_CP\_HPD, LS\_OE parameter from  $V_{CCA}$  to 3.6. .... [6](#)
- 

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**Changes from Revision B (July 2010) to Revision C**
**Page**

- 
- Added Type D connector specification to "FEATURES" ..... [1](#)
  - Added Type D connector specification to "DESCRIPTION/ORDERING INFORMATION" ..... [1](#)
-

## 5 Pin Configuration and Functions



For package dimensions, see the [Mechanical, Packaging, and Orderable Information](#) section.

### Pin Functions

PIN		TYPE	DESCRIPTION
NAME	NO.		
5VOUT	F1	Pwr O	DC-DC output. The 5-V power pin can supply 55-mA regulated current to the HDMI receiver. Separate DC-DC converter control pin CT_CP_HPDP disables the DC-DC converter when operating at low-power mode.
CEC_A	B2	I/O	System-side CEC bus I/O. This pin is bidirectional and referenced to $V_{CCA}$ .
CEC_B	D3	I/O	HDMI-side CEC bus I/O. This pin is bidirectional and referenced to the 3.3-V internal supply.
CLK–	G4	ESD	High-speed ESD clamp: provides ESD protection to the high-speed HDMI differential data lines.
CLK+	F4		
CT_CP_HPDP	D1	Ctrl	DC-DC Enable. Enables the DC-DC converter and HPD circuitry when CT_CP_HPDP = H. The CT_CP_HPDP is referenced to $V_{CCA}$ .
D0–	E4	ESD	High-speed ESD clamp: provides ESD protection to the high-speed HDMI differential data lines.
D0+	D4		
D1–	C4		
D1+	B4		
D2–	A4		
D2+	A3		
FB	E1	I	Feedback input. This pin is a feedback control pin for the DC-DC converter. It must be connected to 5VOUT.
GND	B3, C3, D2, E2	—	Device ground
HPD_A	C2	O	System-side output for the hot plug detect. This pin is unidirectional and is referenced to $V_{CCA}$ .
HPD_B	G3	I	HDMI-side input for the hot plug detect. This pin is unidirectional and is referenced to 5VOUT.
LS_OE	A1	Ctrl	Level shifter enable. This pin is referenced to $V_{CCA}$ . Enables level shifters and LDO when OE = H.
P <sub>GND</sub>	G1	—	DC-DC converter ground. This pin should be tied externally to the system GND plane. See <a href="#">Layout Guidelines</a> .
SCL_A	B1	I/O	System-side input and output for I <sup>2</sup> C bus. This pin is bidirectional and referenced to $V_{CCA}$ .
SCL_B	E3	I/O	HDMI-side input and output for I <sup>2</sup> C bus. This pin is bidirectional and referenced to 5VOUT.

**Pin Functions (continued)**

PIN		TYPE	DESCRIPTION
NAME	NO.		
SDA_A	C1	I/O	System-side input and output for I <sup>2</sup> C bus. This pin is bidirectional and referenced to V <sub>CCA</sub> .
SDA_B	F3	I/O	HDMI-side input and output for I <sup>2</sup> C bus. This pin is bidirectional and referenced to 5VOUT.
SW	F2	I	Switch input. This pin is the inductor input for the DC-DC converter.
V <sub>BAT</sub>	G2	Supply	Battery supply. This voltage is typically 2.3 V to 5.5 V.
V <sub>CCA</sub>	A2	Supply	System-side supply. This voltage is typically 1.2 V to 3.3 V from the core microcontroller.

**Table 1. YFF Package Pin Mapping**

	1	2	3	4
<b>A</b>	LS_OE	V <sub>CCA</sub>	D2+	D2–
<b>B</b>	SCL_A	CEC_A	GND	D1+
<b>C</b>	SDA_A	HPD_A	GND	D1–
<b>D</b>	CT_CP_HPD	GND	CEC_B	D0+
<b>E</b>	FB	GND	SCL_B	D0–
<b>F</b>	5VOUT	SW	SDA_B	CLK+
<b>G</b>	P <sub>GND</sub>	V <sub>BAT</sub>	HPD_B	CLK–

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
V <sub>CCA</sub>	Supply voltage			4	V
V <sub>BAT</sub>	Supply voltage		–0.3	6.5	V
V <sub>I</sub>	Input voltage	SCL_A, SDA_A, CEC_A, CT_CP_HPDP, LS_OE	–0.3	4	V
		SCL_B, SDA_B, CEC_B, D, CLK	–0.3	6	
V <sub>O</sub>	Voltage applied to any output in the high-impedance or power-off state <sup>(2)</sup>	SCL_A, SDA_A, CEC_A, HPD_A	–0.3	4	V
		SCL_B, SDA_B, CEC_B	–0.3	6	
	Voltage applied to any output in the high or low state <sup>(2)</sup>	SCL_A, SDA_A, CEC_A, HPD_A	–0.3	V <sub>CCA</sub> + 0.3	
		SCL_B, SDA_B, CEC_B	–0.5	6	
I <sub>IK</sub>	Input clamp current	V <sub>I</sub> < 0		–50	mA
I <sub>OK</sub>	Output clamp current	V <sub>O</sub> < 0		–50	mA
I <sub>OUTMAX</sub>	Continuous current through 5VOUT or GND			±100	mA
T <sub>stg</sub>	Storage temperature		–65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) The input and output voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2500	V
		Pins E4, D4, C4, B4, A4, A3, G4, F4, D3, G3, E3, F3 F1, and E1	±15000	
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000	
		IEC 61000-4-2 contact discharge	±8000	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over recommended operating free-air temperature range (unless otherwise noted)

			SUPPLY	MIN	NOM	MAX	UNIT
V <sub>CCA</sub>	Supply voltage			1.1		3.6	V
V <sub>BAT</sub>	Supply voltage			2.3		5.5	V
V <sub>IH</sub>	High-level input voltage	SCL_A, SDA_A, CEC_A	V <sub>CCA</sub> = 1.1 V to 3.6 V	0.7 × V <sub>CCA</sub>		V <sub>CCA</sub>	V
		CT_CP_HPDP, LS_OE		1		3.6	
		SCL_B, SDA_B	5VOUT = 5 V	0.7 × 5VOUT		5VOUT	
		CEC_B		0.7 × 3.3 (internal)		3.3 (internal)	
		HPD_B		2.4		5VOUT	

## Recommended Operating Conditions (continued)

over recommended operating free-air temperature range (unless otherwise noted)

		SUPPLY	MIN	NOM	MAX	UNIT
V <sub>IL</sub>	Low-level input voltage	SCL_A, SDA_A, CEC_A	V <sub>CCA</sub> = 1.1 V to 3.6 V	–0.5	0.082 × V <sub>CCA</sub>	V
		CT_CP_HPDP, LS_OE		–0.5	0.4	
		SCL_B, SDA_B	5VOUT = 5 V	–0.5	0.3 × 5VOUT	
		CEC_B		–0.5	0.3 × V <sub>3P3</sub>	
		HPD_B		0	0.8	
V <sub>ILC</sub>	Low-level input voltage (contention)	SCL_A, SDA_A, CEC_A	V <sub>CCA</sub> = 1.1 V to 3.6 V	–0.5	0.065 × V <sub>CCA</sub>	V
V <sub>OL</sub> – V <sub>ILC</sub>	Delta between V <sub>OL</sub> and V <sub>ILC</sub>	SCL_A, SDA_A, CEC_A	V <sub>CCA</sub> = 1.1 V to 3.6 V	0.1 × V <sub>CCA</sub>		V
T <sub>A</sub>	Operating free-air temperature			–40	85	°C

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		TPD12S015	UNIT
		YFF (DSBGA)	
		28 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	63	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	0.4	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	9.2	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	1.6	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	9.1	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics: I<sub>CC</sub>

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
I <sub>CCA</sub>	Standby	V <sub>CCA</sub>	I/O = High			2	μA
	Active					15	
I <sub>CCB</sub>	Standby	V <sub>BAT</sub>	CT_CP_HPDP=L, LS_OE=L, HPD_B=L		2		μA
	DC-DC and HPD active		CT_CP_HPDP=H, LS_OE=L, HPD_B=L		30	50	
	DC-DC, HPD, DDC, CEC active		CT_CP_HPDP=H LS_OE=H, HPD_B=L, I/O =H		225	300	

## 6.6 Electrical Characteristics: High-Speed ESD Lines: Dx, CLK

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
I <sub>OFF</sub>	Current from IO port to supply pins	V <sub>CC</sub> = 0 V, V <sub>IO</sub> = 3.3 V			0.01	0.5	μA
V <sub>DL</sub>	Diode forward voltage	I <sub>D</sub> = 8 mA,	Lower clamp diode		0.85	1	V
R <sub>DYN</sub>	Dynamic resistance	I = 1 A	D, CLK		1		Ω
C <sub>IO</sub>	IO capacitance	V <sub>IO</sub> = 2.5 V	D, CLK		1.3		pF
V <sub>BR</sub>	Break-down voltage	I <sub>IO</sub> = 1 mA		9		12	V

## 6.7 Electrical Characteristics: DC-DC Converter

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>BAT</sub>	Input voltage range		2.3		5.5	V
5VOUT	Total DC output voltage	Includes voltage references, DC load and line regulations, process and temperature	4.9	5	5.13	V
TOVA	Total output voltage accuracy	Includes voltage references, DC load and line regulations, transient load and line regulations, ripple, process and temperature	4.8	5	5.3	V
V <sub>O_Ripple</sub>	Output voltage ripple, loaded	I <sub>O</sub> = 65 mA			20	mVp-p
F <sub>clk</sub>	Internal operating frequency	V <sub>BAT</sub> = 2.3 V to 5.5 V		3.5		MHz
t <sub>start</sub>	Start-up time	From CT_CP_HPD input to 5-V power output 90% point			300	μs
I <sub>O</sub>	Output current	V <sub>BAT</sub> = 2.3 V to 5.5 V	55			mA
	Reverse leakage current V <sub>O</sub>	CT_CP_HPD = L, V <sub>O</sub> = 5.5 V			2.5	μA
	Leakage current from battery to V <sub>O</sub>	CT_CP_HPD = L			5	μA
V <sub>BATUV</sub>	Undervoltage lockout threshold	Falling		2		V
		Rising		2.1		
V <sub>OVC</sub>	Input overvoltage threshold	Falling		5.9		V
		Rising		6		
	Line transient response	V <sub>BAT</sub> = 3.6 V, a pulse of 217-Hz 600 mVp-p square wave, I <sub>O</sub> = 20/65 mA		±25	±50	mVpk
	Load transient response	V <sub>BAT</sub> = 3.6 V, I <sub>O</sub> = 5 to 65 mA, pulse of 10 μs, t <sub>r</sub> = t <sub>f</sub> = 0.1 μs		50		mVpk
I <sub>DD</sub> (idle)	Power supply current from V <sub>BAT</sub> to DC-DC, enabled, unloaded	I <sub>O</sub> = 0 mA		30	50	μA
I <sub>DD</sub> (disabled)	Power supply current from V <sub>BAT</sub> , DC-DC Disabled, Unloaded	V <sub>BAT</sub> = 2.3 V to 5.5 V, I <sub>O</sub> = 0 mA, CT_CP_HPD Low			2	μA
I <sub>DD</sub> (system off)	Power supply current from V <sub>BAT</sub> , V <sub>CCA</sub> = 0 V	V <sub>CCA</sub> = 0 V			5	μA
I <sub>inrush</sub> (start-up)	Inrush current, average over T <sub>startup</sub> time	V <sub>BAT</sub> = 2.3 V to 5.5 V, I <sub>O</sub> = 65 mA		100		mA
T <sub>SD</sub>	Thermal shutdown	Increasing junction temperature		140		°C
	Thermal shutdown hysteresis	Decreasing junction temperature		20		
I <sub>SC</sub>	Short-circuit current limit from output	5-Ω short to GND			500	mA

## 6.8 Electrical Characteristics: Passive Components

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TYP	UNIT
L <sub>IN</sub>	External inductor, 0805 footprint	1	μH
C <sub>IN</sub>	Input capacitor, 0603 footprint	4.7	μF
C <sub>OUT</sub>	Output capacitor, 0603 footprint	4.7	μF
C <sub>VCCA</sub>	Input capacitor, 0402 footprint	0.1	μF



## 6.9 Electrical Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A/x\_B Ports)

 $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  unless otherwise specified

PARAMETER		TEST CONDITIONS	$V_{CCA}$	MIN	TYP	MAX	UNIT
$V_{OHA}$		$I_{OH} = -10\ \mu\text{A}$ , $V_I = V_{IH}$	1.1 V to 3.6 V	$V_{CCA} \times 0.8$			V
$V_{OLA}$		$I_{OL} = 10\ \mu\text{A}$ , $V_I = V_{IL}$	1.1 V to 3.6 V		$V_{CCA} \times 0.17$		V
$V_{OHB}$		$I_{OH} = -10\ \mu\text{A}$ , $V_I = V_{IH}$		$5V_{OUT} \times 0.9$			V
$V_{OLB}$		$I_{OL} = 3\ \text{mA}$ , $V_I = V_{IL}$				0.4	V
$\Delta V_T$ hysteresis	SDx_A ( $V_{T+} - V_{T-}$ )		1.1 V to 3.6 V		40		mV
	SDx_B ( $V_{T+} - V_{T-}$ )		1.1 V to 3.6 V		400		
$R_{PU}$	(Internal pullup)	SCL_A, SDA_A, Internal pullup connected to $V_{CCA}$ rail			10		k $\Omega$
		SCL_B, SDA_B, Internal pullup connected to 5-V rail			1.75		
$I_{PULLUPAC}$	Transient boosted pullup current (rise time accelerator)	SCL_B, SDA_B, Internal pullup connected to 5-V rail			15		mA
$I_{OFF}$	A port	$V_{CCA} = 0\ \text{V}$ , $V_I$ or $V_O = 0$ to 3.6 V	0 V			$\pm 5$	$\mu\text{A}$
	B port	$5V_{OUT} = 0\ \text{V}$ , $V_I$ or $V_O = 0$ to 5.5 V	0 V to 3.6 V			$\pm 5$	
$I_{OZ}$	B port	$V_O = V_{CCO}$ or GND	1.1 V to 3.6 V			$\pm 5$	$\mu\text{A}$
	A port	$V_I = V_{CCI}$ or GND	1.1 V to 3.6 V			$\pm 5$	

## 6.10 Electrical Characteristics: Voltage Level Shifter: CEC Lines (x\_A/x\_B Ports)

 $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  unless otherwise specified

PARAMETER		TEST CONDITIONS	$V_{CCA}$	MIN	TYP	MAX	UNIT
$V_{OHA}$		$I_{OH} = -10\ \mu\text{A}$ , $V_I = V_{IH}$	1.1 V to 3.6 V	$V_{CCA} \times 0.8$			V
$V_{OLA}$		$I_{OL} = 10\ \mu\text{A}$ , $V_I = V_{IL}$	1.1 V to 3.6 V		$V_{CCA} \times 0.17$		V
$V_{OHB}$		$I_{OH} = -10\ \mu\text{A}$ , $V_I = V_{IH}$		$V_{3P3} \times 0.9$			V
$V_{OLB}$		$I_{OL} = 3\ \text{mA}$ , $V_I = V_{IL}$				0.4	V
$\Delta V_T$ hysteresis	CEC_A ( $V_{T+} - V_{T-}$ )		1.1 V to 3.6 V		40		mV
	CEC_B ( $V_{T+} - V_{T-}$ )		1.1 V to 3.6 V		300		
$R_{PU}$	(Internal pullup)	CEC_A, Internal pullup connected to $V_{CCA}$ rail			10		k $\Omega$
		CEC_B, Internal pullup connected to internal 3.3-V rail			26		
$I_{OFF}$	A port	$V_{CCA} = 0\ \text{V}$ , $V_I$ or $V_O = 0$ to 3.6 V	0 V			$\pm 5$	$\mu\text{A}$
	B port	$5V_{OUT} = 0\ \text{V}$ , $V_I$ or $V_O = 0$ to 5.5 V	0 V to 3.6 V			$\pm 1.8$	
$I_{OZ}$	B port	$V_O = V_{CCO}$ or GND	1.1 V to 3.6 V			$\pm 5$	$\mu\text{A}$
	A port	$V_I = V_{CCI}$ or GND	1.1 V to 3.6 V			$\pm 5$	

## 6.11 Electrical Characteristics: Voltage Level Shifter: HPD Line (x\_A/x\_B Ports)

 $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  unless otherwise specified

PARAMETER		TEST CONDITIONS	$V_{CCA}$	MIN	TYP	MAX	UNIT
$V_{OHA}$		$I_{OH} = -3\ \text{mA}$ , $V_I = V_{IH}$	1.1 V to 3.6 V	$V_{CCA} \times 0.7$			V
$V_{OLA}$		$I_{OL} = 3\ \text{mA}$ , $V_I = V_{IL}$	1.1 V to 3.6 V		$V_{CCA} \times 0.17$		V
$\Delta V_T$ hysteresis		HPD_B ( $V_{T+} - V_{T-}$ )	1.1 V to 3.6 V		700		mV
$R_{PD}$	(Internal pulldown)	HPD_B, Internal pulldown connected to GND			11		k $\Omega$
$I_{OFF}$	A port	$V_O = V_{CCO}$ or GND	0 V			$\pm 5$	$\mu\text{A}$
$I_{OZ}$	A port	$V_I = V_{CCI}$ or GND	3.6 V			$\pm 5$	$\mu\text{A}$

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### 6.12 Electrical Characteristics: LS\_OE, CT\_CP\_HP

 $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  unless otherwise specified

PARAMETER	TEST CONDITIONS	$V_{CCA}$	MIN	TYP	MAX	UNIT
$I_I$	$V_I = V_{CCA}$ or GND	1.1 V to 3.6 V			$\pm 12$	$\mu\text{A}$

### 6.13 Electrical Characteristics: I/O Capacitance

 $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  unless otherwise specified

PARAMETER	TEST CONDITIONS	$V_{CCA}$	MIN	TYP	MAX	UNIT
$C_I$ Control inputs	$V_I = 1.89\text{ V}$ or GND	1.1 V to 3.6 V		7.1	8.5	pF
$C_{IO}$	A port	$V_O = 1.89\text{ V}$ or GND		8.3	9.5	pF
	B port	$V_O = 5.0\text{ V}$ or GND		15	16.5	

### 6.14 Switching Characteristics

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$C_L$	Bus load capacitance (B side)			750	pF
	Bus load capacitance (A side)			15	

### 6.15 Switching Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A & x\_B ports); $V_{CCA} = 1.2\text{ V}$

 $V_{CCA} = 1.2\text{ V}$ 

PARAMETER	PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PHL}$ Propagation delay	A to B	DDC Channels Enabled		344		ns
	B to A			335		
$t_{PLH}$ Propagation delay	A to B	DDC Channels Enabled		452		ns
	B to A			178		
$t_f$	A port fall time	DDC Channels Enabled		138		ns
	B port fall time			83		
$t_r$	A port rise time	DDC Channels Enabled		194		ns
	B port rise time			92		
$f_{MAX}$ Maximum switching frequency		DDC Channels Enabled	400			kHz

### 6.16 Switching Characteristics: Voltage Level Shifter: CEC Line (x\_A & x\_B ports); $V_{CCA} = 1.2\text{ V}$

 $V_{CCA} = 1.2\text{ V}$ 

PARAMETER	PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	A to B	CEC Channels Enabled		445		ns
	B to A			337		
$t_{PLH}$	A to B			13		$\mu\text{s}$
	B to A			0.266		
$t_f$	A port fall time	CEC Channels Enabled		140		ns
	B port fall time			96		
$t_r$	A port rise time	CEC Channels Enabled		202		ns
	B port rise time			15		

### 6.17 Switching Characteristics: Voltage Level Shifter: HPD Line (x\_A & x\_B ports); $V_{CCA} = 1.2\text{ V}$

 $V_{CCA} = 1.2\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	B to A	CEC Channels Enabled		10		$\mu\text{s}$
$t_{PLH}$		B to A			9		
$t_f$	A port fall time	A Port	CEC Channels Enabled		0.67		ns
$t_r$	A port rise time	A Port	CEC Channels Enabled		0.74		ns

### 6.18 Switching Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A & x\_B ports); $V_{CCA} = 1.5\text{ V}$

 $V_{CCA} = 1.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	A to B	DDC Channels Enabled		335		ns
		B to A			265		
$t_{PLH}$		A to B			438		
		B to A			169		
$t_f$	A port fall time	A Port	DDC Channels Enabled		110		ns
	B port fall time	B Port			83		
$t_r$	A port rise time	A Port	DDC Channels Enabled		190		ns
	B port rise time	B Port			92		
$f_{MAX}$	Maximum switching frequency		DDC Channels Enabled	400			kHz

### 6.19 Switching Characteristics: Voltage Level Shifter: CEC Line (x\_A & x\_B ports); $V_{CCA} = 1.5\text{ V}$

 $V_{CCA} = 1.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	A to B	CEC Channels Enabled		437		ns
		B to A			267		
$t_{PLH}$		A to B			13		$\mu\text{s}$
		B to A			0.264		
$t_f$	A port fall time	A Port	CEC Channels Enabled		110		ns
	B port fall time	B Port			96		
$t_r$	A port rise time	A Port	CEC Channels Enabled		202		ns
	B port rise time	B Port			15		$\mu\text{s}$

### 6.20 Switching Characteristics: Voltage Level Shifter: HPD Line (x\_A & x\_B ports); $V_{CCA} = 1.5\text{ V}$

 $V_{CCA} = 1.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	B to A	CEC Channels Enabled		10		$\mu\text{s}$
$t_{PLH}$		B to A			9		
$t_f$	A port fall time	A Port	CEC Channels Enabled		0.47		ns
$t_r$	A port rise time	A Port	CEC Channels Enabled		0.51		ns

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**6.21 Switching Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A & x\_B ports);  $V_{CCA} = 1.8\text{ V}$** 
 $V_{CCA} = 1.8\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	DDC Channels Enabled	334		ns	
		B to A		229			
t <sub>PLH</sub>		A to B		431			
		B to A		169			
t <sub>f</sub>	A port fall time	A Port	DDC Channels Enabled	94		ns	
	B port fall time	B Port		83			
t <sub>r</sub>	A port rise time	A Port	DDC Channels Enabled	191		ns	
	B port rise time	B Port		92			
f <sub>MAX</sub>	Maximum switching frequency		DDC Channels Enabled	400			kHz

**6.22 Switching Characteristics: Voltage Level Shifter: CEC Line (x\_A & x\_B ports);  $V_{CCA} = 1.8\text{ V}$** 
 $V_{CCA} = 1.8\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	CEC Channels Enabled		441		ns
		B to A			231		
t <sub>PLH</sub>		A to B			13		μs
		B to A			0.26		
t <sub>f</sub>	A port fall time	A Port	CEC Channels Enabled		94		ns
	B port fall time	B Port			96		
t <sub>r</sub>	A port rise time	A Port	CEC Channels Enabled		201		ns
	B port rise time	B Port			15		μs

**6.23 Switching Characteristics: Voltage Level Shifter: HPD Line (x\_A & x\_B ports);  $V_{CCA} = 1.8\text{ V}$** 
 $V_{CCA} = 1.8\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	B to A	CEC Channels Enabled		10		$\mu\text{s}$
		B to A			9		
$t_f$	A port fall time	A Port	CEC Channels Enabled		0.41		ns
$t_r$	A port rise time	A Port	CEC Channels Enabled		0.45		ns

**6.24 Switching Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A & x\_B ports);  $V_{CCA} = 2.5\text{ V}$** 
 $V_{CCA} = 2.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	DDC Channels Enabled		330		ns
		B to A			182		
t <sub>PLH</sub>		A to B			423		
		B to A			166		
t <sub>f</sub>	A port fall time	A Port	DDC Channels Enabled		79		ns
	B port fall time	B Port			83		
t <sub>r</sub>	A port rise time	A Port	DDC Channels Enabled		188		ns
	B port rise time	B Port			92		
f <sub>MAX</sub>	Maximum switching frequency		DDC Channels Enabled	400			kHz

## 6.25 Switching Characteristics: Voltage Level Shifter: CEC Line (x\_A & x\_B ports); $V_{CCA} = 2.5\text{ V}$

 $V_{CCA} = 2.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	CEC Channels Enabled		454		ns
		B to A			184		
t <sub>PLH</sub>		A to B			13		μs
		B to A			0.255		
t <sub>f</sub>	A port fall time	A Port	CEC Channels Enabled		79		ns
	B port fall time	B Port			96		
t <sub>r</sub>	A port rise time	A Port	CEC Channels Enabled		194		ns
	B port rise time	B Port			15		μs

## 6.26 Switching Characteristics: Voltage Level Shifter: HPD Line (x\_A & x\_B ports); $V_{CCA} = 2.5\text{ V}$

 $V_{CCA} = 2.5\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	B to A	CEC Channels Enabled		10		$\mu\text{s}$
		B to A			9		
$t_f$	A port fall time	A Port	CEC Channels Enabled		0.37		ns
$t_r$	A port rise time	A Port	CEC Channels Enabled		0.39		ns

## 6.27 Switching Characteristics: Voltage Level Shifter: SCL, SDA Lines (x\_A & x\_B ports); $V_{CCA} = 3.3\text{ V}$

 $V_{CCA} = 3.3\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	DDC channels enabled	323		ns	
		B to A		158			
t <sub>PLH</sub>		A to B		421			
		B to A		162			
t <sub>f</sub>	A port fall time	A Port	DDC channels enabled	71		ns	
	B port fall time	B Port		84			
t <sub>r</sub>	A port rise time	A Port	DDC channels enabled	188		ns	
	B port rise time	B Port		92			
f <sub>MAX</sub>	Maximum switching frequency		DDC channels enabled	400		kHz	

## 6.28 Switching Characteristics: Voltage Level Shifter: CEC Line (x\_A & x\_B ports); $V_{CCA} = 3.3\text{ V}$

 $V_{CCA} = 3.3\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay	A to B	CEC channels enabled		450		ns
		B to A			160		
t <sub>PLH</sub>		A to B			13		μs
		B to A			0.251		
t <sub>f</sub>	A port fall time	A Port	CEC channels enabled		71		ns
	B port fall time	B Port			96		
t <sub>r</sub>	A port rise time	A Port	CEC channels enabled		194		ns
	B port rise time	B Port			15		μs

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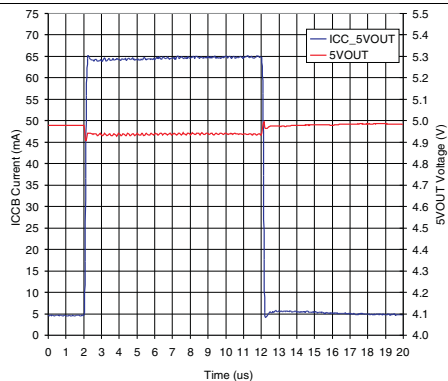
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### 6.29 Switching Characteristics: Voltage Level Shifter: HPD Line (x\_A & x\_B ports); $V_{CCA} = 3.3\text{ V}$

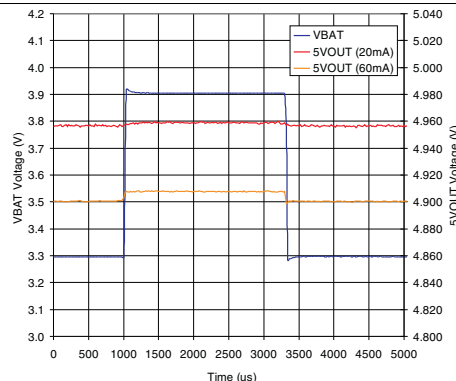
 $V_{CCA} = 3.3\text{ V}$ 

PARAMETER		PINS	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{PLH}$	Propagation delay	B to A	CEC channels enabled		10		$\mu\text{s}$
$t_{PLH}$		B to A			9		
$t_f$	A port fall time	A Port	CEC channels enabled		0.35		ns
$t_r$	A port rise time	A Port	CEC channels enabled		0.37		ns

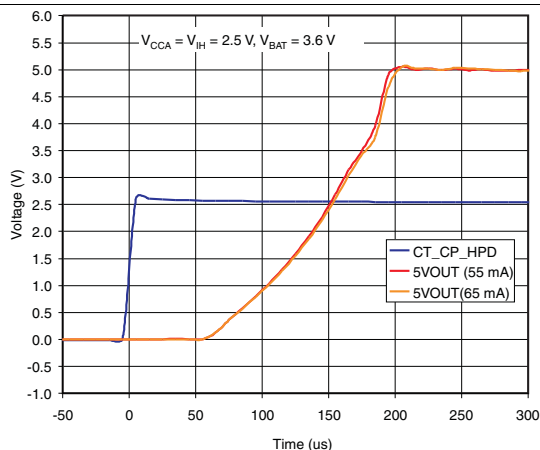
## 6.30 Typical Characteristics



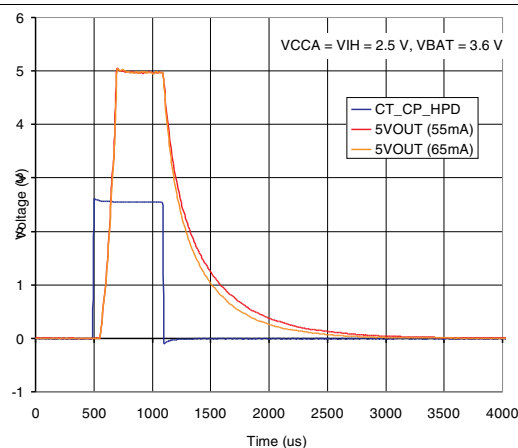
**Figure 1. Load Transient Response**



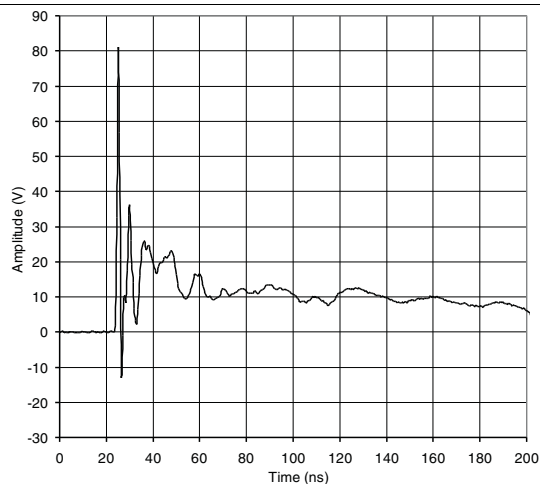
**Figure 2. Line Transient Response**



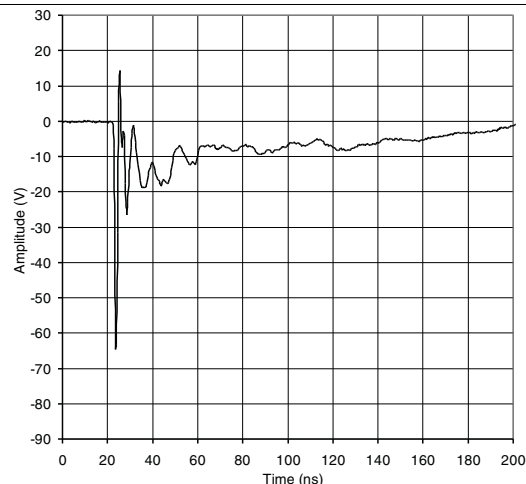
**Figure 3.  $t_{START}$**



**Figure 4. DC-DC Start-Up and Shutdown**

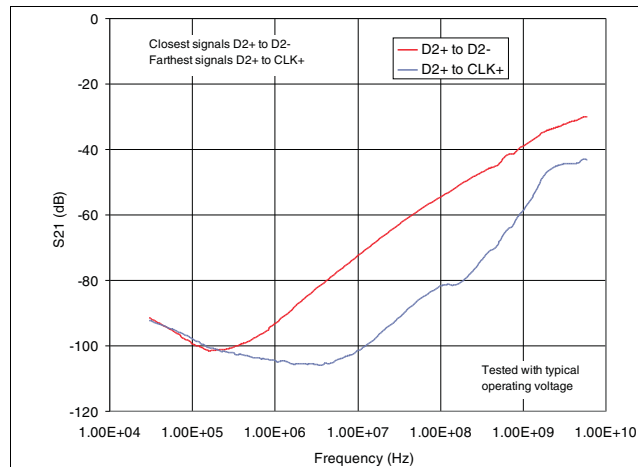


**Figure 5. IEC Clamping Waveforms 8-kV Contact (IEC ESD Pins)**

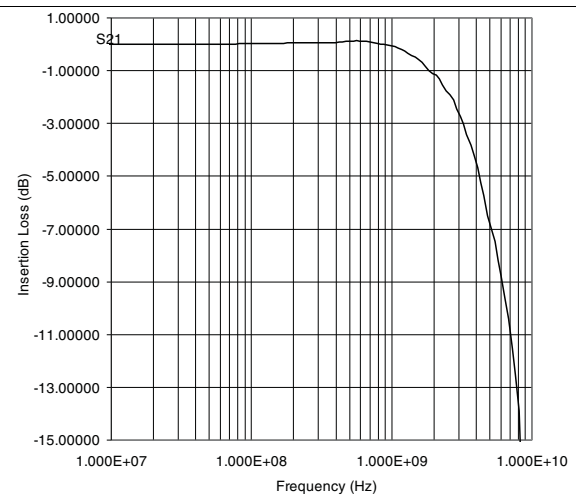


**Figure 6. IEC Clamping Waveforms -8-kV Contact (IEC ESD Pins)**

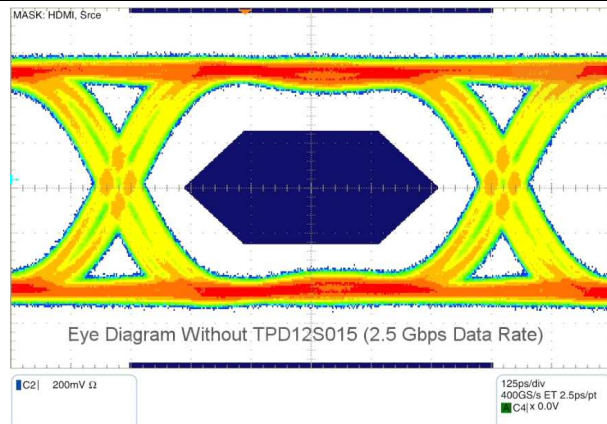
## Typical Characteristics (continued)



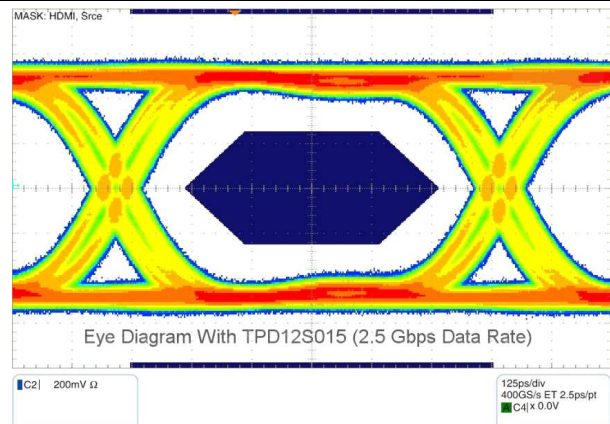
**Figure 7. Channel-to-Channel Crosstalk**



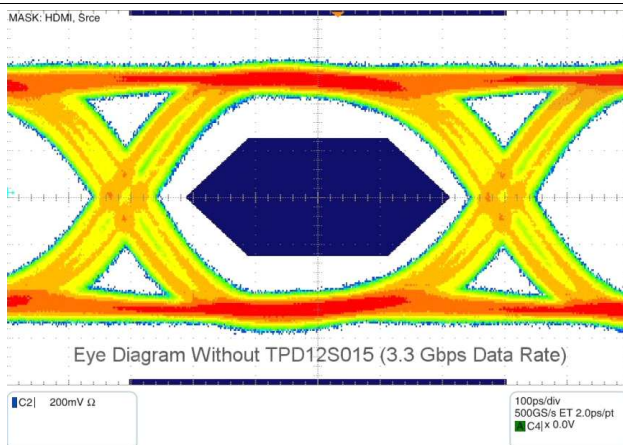
**Figure 8. Insertion Loss Data Line to GND**



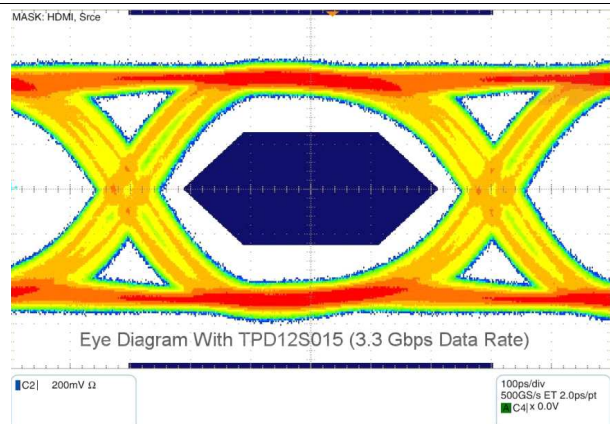
**Figure 9. Eye Diagram Performance on a Test Board for the D+, D- Lines at 2.5 Gbps**



**Figure 10. Eye Diagram Performance on a Test Board for the D+, D- Lines at 2.5 Gbps**



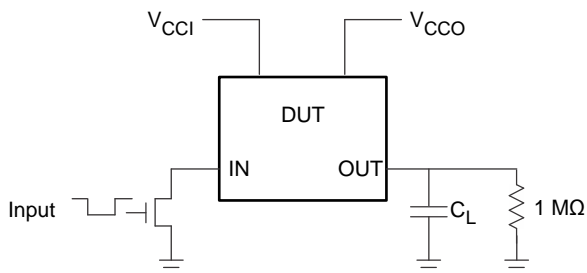
**Figure 11. Eye Diagram Performance on a Test Board for the D+, D- Lines at 3.3 Gbps**



**Figure 12. Eye Diagram Performance on a Test Board for the D+, D- Lines at 3.3 Gbps**



## 7 Parameter Measurement Information

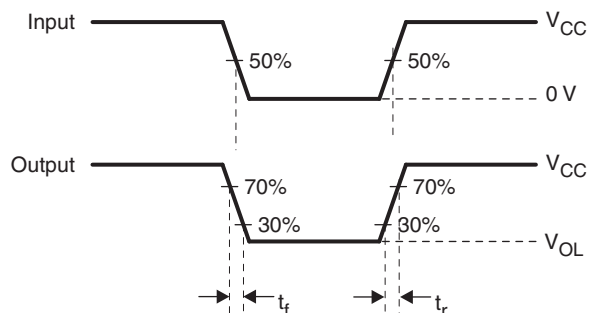


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**Figure 13. Test Circuit**

**Table 2. Design Parameters**

PIN	C <sub>L</sub>
DDC, CEC (A side)	750 pF
DDC, CEC, HPD (B side)	15 pF



- A.  $R_T$  termination resistance should be equal to  $Z_{OUT}$  of pulse generators.
- B.  $C_L$  includes probe and jig capacitance.
- C. All input pulses are supplied by generators having the following characteristics:  $PRR \leq 10$  MHz,  $Z_O = 50 \Omega$ , slew rate  $\geq 1$  V/ns.
- D. The outputs are measured one at a time, with one transition per measurement.
- E.  $t_{PLH}$  and  $t_{PHL}$  are the same as  $t_{pd}$ .

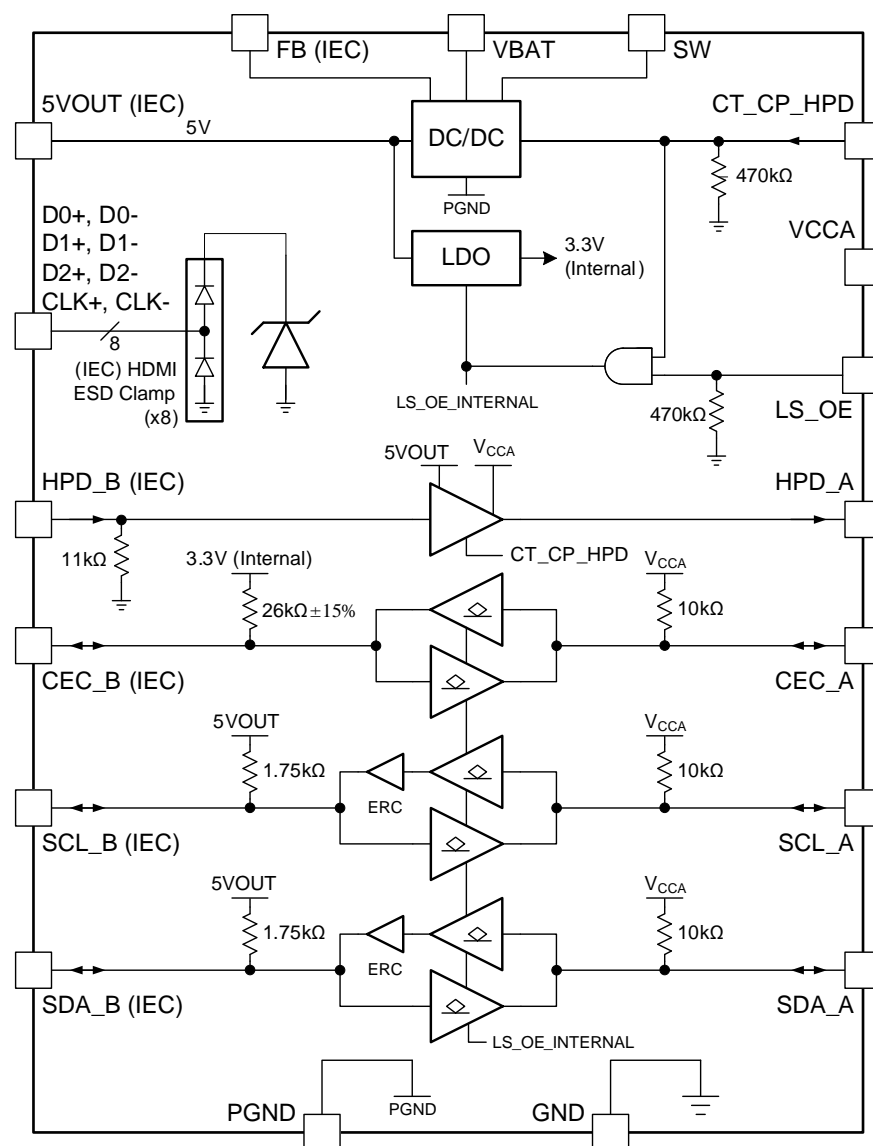
**Figure 14. Test Circuit and Voltage Waveforms**

## 8 Detailed Description

### 8.1 Overview

The TPD12S015 is an integrated interface solution for HDMI 1.3 and 1.4 interfaces, for both portable and non-portable electronics applications. The device has a boost DC-DC converter that uses the 2.3-V to 5.5-V internal power supply and outputs regulated 5-V standard compliant power supply to the cable. This power supply output has current limit and short-circuit protection function. There are bidirectional level-shifting and signal-conditioning circuits on CEC, SCL, and SDA with pullup resistors integrated to minimize the external passive discrete component use. There is also a unidirectional level shifter for HPD signal that translates the 5-V HPD down to  $V_{CCA}$  level. The HPD\_B port has a glitch filter to avoid false detection due to the bouncing while inserting the HDMI plug. For the eight TMDS lines, there are high-speed ESD diodes on each line to make sure that the system pass 8-kV contact ESD.

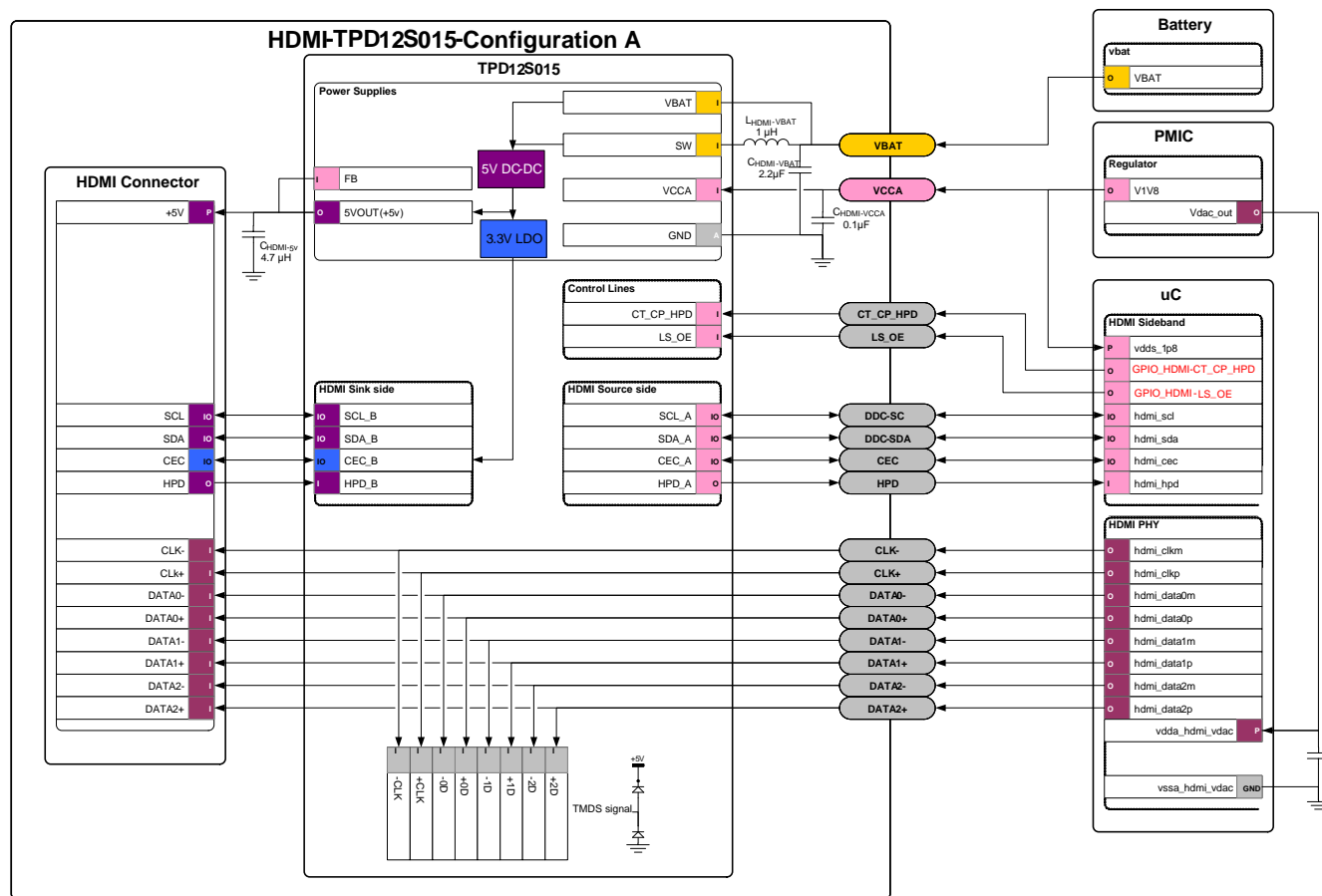
### 8.2 Functional Block Diagrams



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**Figure 15. Circuit Block Diagram**

## Functional Block Diagrams (continued)



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Figure 16. System-Level Block Diagram

## 8.3 Feature Description

### 8.3.1 Rise-Time Accelerators

The HDMI cable side of the DDC lines incorporates rise-time accelerators to support the high-capacitive load on the HDMI cable side. The rise-time accelerator boosts the cable side DDC signal, independent to which side of the bus is releasing the signal.

### 8.3.2 Internal Pullup Resistor

The TPD12S015 has incorporated all the required pullup and pulldown resistors at the interface pins. The system is designed to work properly with no external pullup resistors on the DDC, CEC, and HPD lines. For proper system operation, no external resistors are placed at the A and B ports. If there are internal pullups at the host processor, they must be disabled.

### 8.3.3 Undervoltage Lockout

The undervoltage lockout circuit prevents the DC-DC converter from malfunctioning at low input voltages and from excessive discharge of the battery. It disables the output stage of the converter once the falling  $V_{IN}$  trips the undervoltage lockout threshold  $V_{BATUV}$ . The undervoltage lockout threshold  $V_{BATUV}$  for falling  $V_{IN}$  is typically 2 V. The device starts operation once the rising  $V_{IN}$  trips undervoltage lockout threshold  $V_{BATUV}$  again at typical 2.1 V.

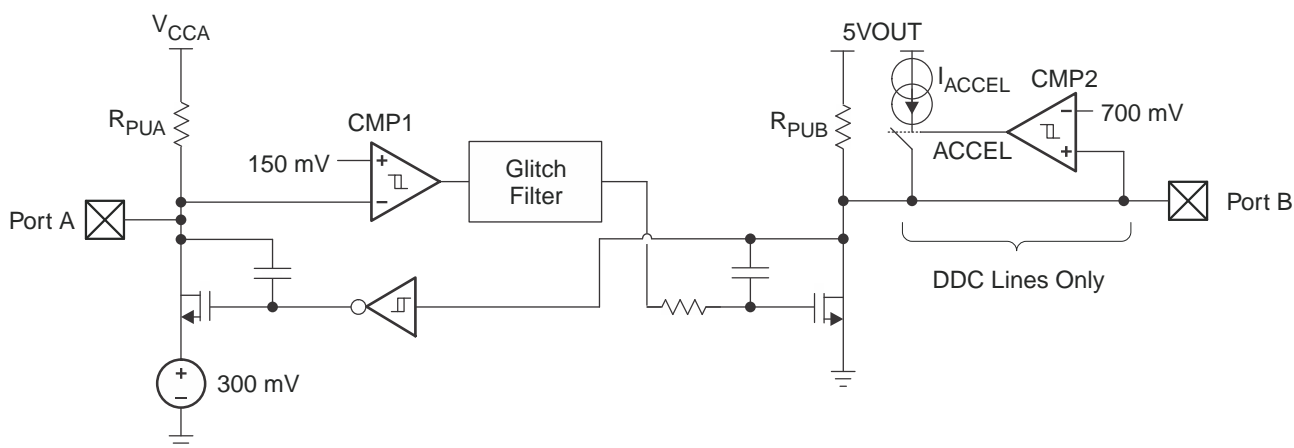
## Feature Description (continued)

### 8.3.4 Soft Start

The DC-DC converter has an internal soft-start circuit that controls the ramp-up of the output voltage. The output voltage reaches its nominal value within  $t_{\text{Start}}$  of typically 250  $\mu\text{s}$  after CT\_CP\_HPD pin has been pulled to high level. The output voltage ramps up from 5% to its nominal value within  $t_{\text{Ramp}}$  of 300  $\mu\text{s}$ . This limits the inrush current in the converter during start-up and prevents possible input voltage drops when a battery or high-impedance power source is used. During soft start, the switch current limit is reduced to 300 mA until the output voltage reaches  $V_{\text{IN}}$ . Once the output voltage trips this threshold, the device operates with its nominal current limit ILIMF.

### 8.3.5 DDC and CEC Level-Shifting Circuit Operation

The TPD12S015 enables DDC translation from  $V_{\text{CCA}}$  (system side) voltage levels to 5-V (HDMI cable side) voltage levels without degradation of system performance. The TPD12S015 contains two bidirectional open-drain buffers specifically designed to support up-translation or down-translation between the low voltage,  $V_{\text{CCA}}$  side DDC-bus, and the 5-V DDC-bus. The port B I/Os are overvoltage tolerant to 5.5 V even when the device is unpowered. After power up and with the LS\_OE and CT\_CP\_HPD pins high, a low level on port A (below approximately  $V_{\text{ILC}} = 0.08 \times V_{\text{CCA}}$  V) turns the corresponding port B driver (either SDA or SCL) on and drives port B down to  $V_{\text{OLB}}$  V. When port A rises above approximately  $0.10 \times V_{\text{CCA}}$  V, the port B pulldown driver is turned off and the internal pullup resistor pulls the pin high. When port B falls first and goes below  $0.3 \times 5\text{VOUT}$ , a CMOS hysteresis input buffer detects the falling edge, turns on the port A driver, and pulls port A down to approximately  $V_{\text{OLA}} = 0.16 \times V_{\text{CCA}}$  V. The port B pulldown is not enabled unless the port A voltage goes below  $V_{\text{ILC}}$ . If the port A low voltage goes below  $V_{\text{ILC}}$ , the port B pulldown driver is enabled until port A rises above  $(V_{\text{ILC}} + \Delta V_{\text{T-HYSTA}})$ , then port B, if not externally driven LOW, continues to rise being pulled up by the internal pullup resistor.



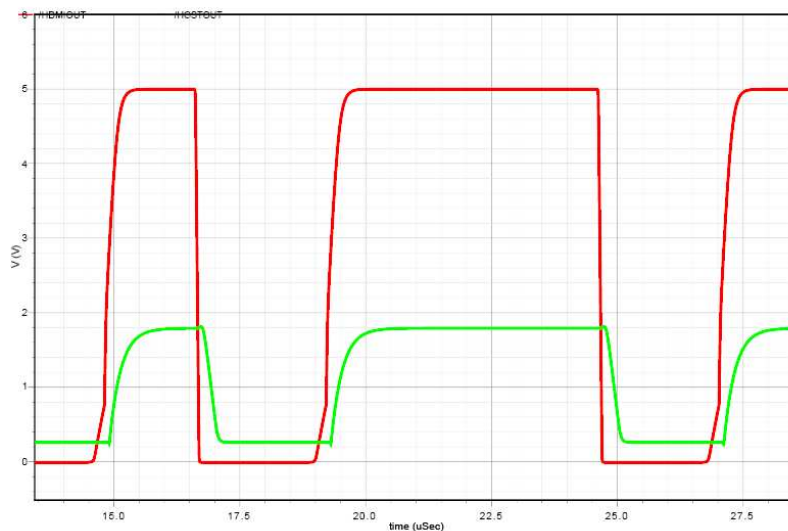
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**Figure 17. DDC and CEC Level Shifter Block Diagram**

### 8.3.6 DDC and CEC Level Shifting Operation When $V_{\text{CCA}} = 1.8 \text{ V}$

- The threshold of CMP1 is approximately 150 mV  $\pm$  the 40 mV of total hysteresis.
- The comparator trips for a falling waveform at approximately 130 mV
- The comparator trips for a rising waveform at approximately 170 mV
- To be recognized as a zero, the level at Port A must first go below 130 mV ( $V_{\text{ILC}}$  in spec) and then stay below 170 mV ( $V_{\text{ILA}}$  in spec)
- To be recognized as a one, the level at A must first go above 170 mV and then stay above 130 mV
- $V_{\text{ILC}}$  is set to 110 mV to give some margin to the 130 mV
- $V_{\text{ILA}}$  is set to 140 mV to give some margin to the 170 mV
- $V_{\text{IHA}}$  is set to 70% of  $V_{\text{CCA}}$  to be consistent with standard CMOS levels

## Feature Description (continued)



**Figure 18. DDC and CEC Level-Shifting Operation (B to A Direction)**

### 8.3.7 CEC Level-Shifting Operation

The CEC level-shifting function operates in the same manner as the DDC lines except that the CEC line does not need the rise time accelerator function.

## 8.4 Device Functional Modes

### 8.4.1 Enable

The DC-DC converter is enabled when the CT\_CP\_HPD is set to high. At first, the internal reference is activated and the internal analog circuits are settled. Afterwards, the soft start is activated and the output voltage is ramped up. The output voltage reaches its nominal value in typically 250  $\mu$ s after the device has been enabled. The CT\_CP\_HPD input can be used to control power sequencing in a system with various DC-DC converters. The CT\_CP\_HPD pin can be connected to the output of another converter, to drive the EN pin high and getting a sequencing of supply rails. With CT\_CP\_HPD = GND, the DC-DC enters shutdown mode.

### 8.4.2 Power Save Mode

The TPD12S015 integrates a power save mode to improve efficiency at light load. In power save mode, the converter only operates when the output voltage trips below a set threshold voltage. It ramps up the output voltage with several pulses and goes into power save mode once the output voltage exceeds the set threshold voltage. The PFM mode is left and PWM mode entered in case the output current can not longer be supported in PFM mode.

## 9 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

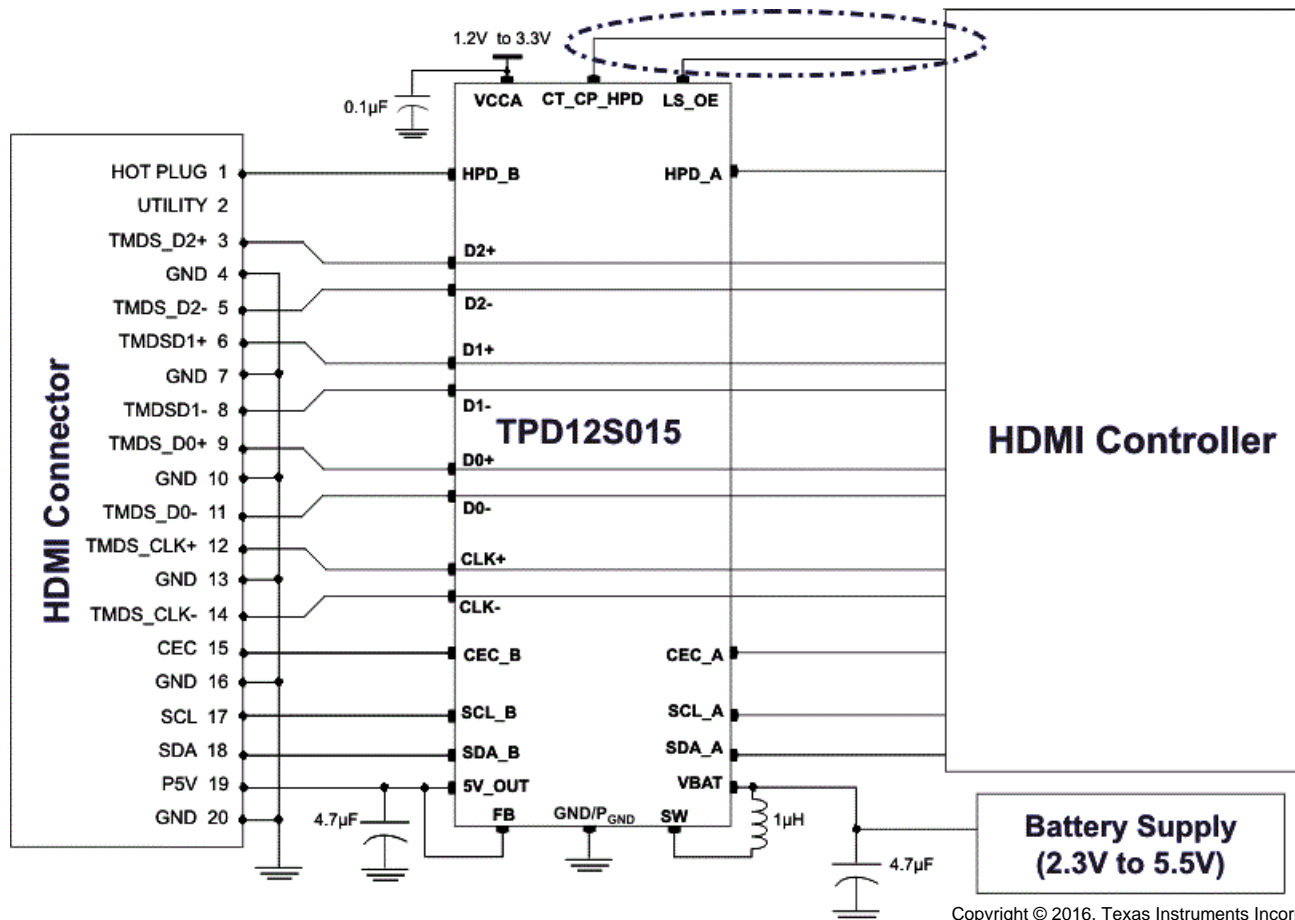
### 9.1 Application Information

The TPD12S015 is an integrated solution for HDMI 1.3 and 1.4 interface. The device has a boost converter on the power supply, signal conditioning circuits on CEC, SCL, SDA, HPD lines, and ESD protection on the TMDS lines. To get the best performance, see [Design Requirements](#), [Detailed Design Procedure](#), and [Application Curves](#).

### 9.2 Typical Applications

#### 9.2.1 TPD12S015 Controlled by Two GPIOs from Controller

Some HDMI controller chips may have two GPIOs to control the HDMI interface chip. Figure 19 shows how TPD12S015 is used in this situation.



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**Figure 19. TPD12S015 Controlled by Two GPIOs from Controller Schematic**

#### 9.2.1.1 Design Requirements

Table 3 lists the known system parameters for an HDMI 1.4 application.

**Table 3. Design Parameters**

DESIGN PARAMETER	VALUE
5V_OUT DC current	55 mA
CEC_A, HPD_A, SCL_A, SDA_A voltage level	V <sub>CCA</sub>
HDMI data rate per TMDS signal pair	3.4 Gbps
Required IEC 61000-4-2 ESD Protection	±8-kV Contact

### 9.2.1.2 Detailed Design Procedure

#### 9.2.1.2.1 Inductor Selection

To make sure that the TPD12S015 devices can operate, an inductor must be connected between pin V<sub>BAT</sub> and pin SW. A boost converter normally requires two main passive components for storing energy during the conversion. A boost inductor and a storage capacitor at the output are required. To select the boost inductor, TI recommends keeping the possible peak inductor current below the current limit threshold of the power switch in the chosen configuration. The highest peak current through the inductor and the switch depends on the output load, the input (V<sub>BAT</sub>), and the output voltage (5VOUT). Estimation of the maximum average inductor current can be done using [Equation 1](#).

$$I_{L\_MAX} \approx I_{OUT} \times \frac{V_{OUT}}{\eta \times V_{IN}} \quad (1)$$

For example, for an output current of 55 mA at 5VOUT, approximately 150 mA of average current flows through the inductor at a minimum input voltage of 2.3 V.

The second parameter for choosing the inductor is the desired current ripple in the inductor. Normally, it is advisable to work with a ripple of less than 20% of the average inductor current. A smaller ripple reduces the magnetic hysteresis losses in the inductor, as well as output voltage ripple and EMI. But in the same way, regulation time at load changes rises. In addition, a larger inductor increases the total system size and cost. With these parameters, it is possible to calculate the value of the minimum inductance by using [Equation 2](#).

$$L_{MIN} \approx \frac{V_{IN} \times (V_{OUT} - V_{IN})}{\Delta I_L \times f \times V_{OUT}}$$

where

- f is the switching frequency
  - ΔI<sub>L</sub> is the ripple current in the inductor, that is, 20% × I<sub>L</sub>
- (2)

With this calculated value and the calculated currents, it is possible to choose a suitable inductor. In typical applications, TI recommends an inductance of 1 μH, even if [Equation 2](#) yields something lower. Take care so that load transients and losses in the circuit can lead to higher currents as estimated in [Equation 3](#). Also, the losses in the inductor caused by magnetic hysteresis losses and copper losses are a major parameter for total circuit efficiency.

With the chosen inductance value, the peak current for the inductor in steady-state operation can be calculated. [Equation 3](#) shows how to calculate the peak current I.

$$I_{L(peak)} = \frac{V_{IN} \times D}{2 \times f \times L} + \frac{I_{OUT}}{(1-D) \times \eta}$$

where

- $D = \frac{V_{OUT} - V_{IN}}{V_{OUT}}$
- (3)

This would be the critical value for the current rating for selecting the inductor. Also consider that load transients and error conditions may cause higher inductor currents.

### 9.2.1.2.2 Input Capacitor

Because of the nature of the boost converter having a pulsating input current, a low-ESR input capacitor is required to prevent large voltage transients that can cause misbehavior of the device or interferences with other circuits in the system. TI recommends at least a 1.2-μF input capacitor to improve transient behavior of the regulator and EMI behavior of the total power supply circuit. TI recommends placing a ceramic capacitor as close as possible to the  $V_{IN}$  and GND pins; to improve the input noise filtering, it is better to use a 4.7-μF capacitor.

### 9.2.1.2.3 Output Capacitor

For the output capacitor, TI recommends using small ceramic capacitors placed as close as possible to the  $V_{OUT}$  and GND pins of the IC. If, for any reason, the application requires the use of large capacitors that cannot be placed close to the IC, TI recommends using a smaller ceramic capacitor in parallel to the large one. This small capacitor must be placed as close as possible to the  $V_{OUT}$  and GND pins of the IC. Use Equation 4 to estimate the recommended minimum output capacitance.

$$C_{min} = \frac{I_{OUT} \times (V_{OUT} - V_{IN})}{f \times \Delta V \times V_{OUT}}$$

where

- $f$  is the switching frequency
  - $\Delta V$  is the maximum allowed ripple
- (4)

With a chosen ripple voltage of 10 mV, a minimum effective capacitance of 2.7 μF is needed. The total ripple is larger due to the ESR of the output capacitor. This additional component of the ripple can be calculated using Equation 5.

$$\Delta V_{ESR} = I_{OUT} \times R_{ESR}$$
(5)

A capacitor with a value in the range of the calculated minimum must be used. This is required to maintain control loop stability. There are no additional requirements regarding minimum ESR. There is no upper limit for the output capacitance value. Larger capacitors cause lower output voltage ripple as well as lower output voltage drop during load transients.

#### NOTE

Ceramic capacitors have a DC Bias effect, which have a strong influence on the final effective capacitance needed. Therefore the right capacitor value has to be chosen very carefully.

Package size and voltage rating in combination with material are responsible for differences between the rated capacitor value and the effective capacitance. The minimum effective capacitance value should be 1.2 μF, but the preferred value is about 4.7 μF.

**Table 4. Passive Components: Recommended Minimum Effective Values**

COMPONENT	MIN	TARGET	MAX	UNIT
$C_{IN}$	1.2	4.7	6.5	μF
$C_{OUT}$	1.2	4.7	10	μF
$L_{IN}$	0.7	1	1.3	μH

### 9.2.1.2.4 CEC, HPD, SCL, and SDA Level-Shifting Function

To accommodate for the lower logic levels of some processors' control lines, level shifters are needed to translate the interface voltage down to  $V_{CCA}$ , the voltage level used by the processor. The TPD12S015 has bidirectional level shifters on CEC, SCL, and SDA lines to support the two-way communication. The pullup resistors are integrated to minimize the number of external components. For HPD line, only one way of hot-plug indication is needed, the level shifter is unidirectional. There is a built-in HPD\_B pulldown resistor to keep the voltage level low on the connector side when nothing is attached. Apart from the signal level translation, the rise-time accelerators on the connector side increases the load driving capability.



### 9.2.1.2.5 ESD

To get the best ESD performance on the interface side pins, high-performance ESD diodes are needed. The TPD12S015's ESD diodes on D0+, D0-, D1+, D1-, D2+, D2-, CLK+, CLK-, SCL\_B, SDA\_B, CEC\_B, HPD\_B, 5VOUT, and FB ensure passing 8-kV contact IEC, the highest level ESD. Signal integrity on TMDS lines is also a design concern that must be evaluated to meet the HDMI 1.3 or 1.4 data rate. With the typical I/O capacitance of 1.3 pF and a bandwidth above 3 GHz, Figure 11 shows that TPD12S015's ESD structure has enough margin to meet the data rate requirement of HDMI 1.3 or 1.4.

### 9.2.1.2.6 Ground Offset Consideration

Ground offset between the TPD12S015 ground and the ground of devices on port A of the TPD12S015 must be avoided. The reason for this cautionary remark is that a CMOS or NMOS open-drain capable of sinking 3 mA of current at 0.4 V has an output resistance of 133  $\Omega$  or less. Such a driver shares enough current with the port A output pulldown of the TPD12S015 to be seen as a LOW as long as the ground offset is zero. If the ground offset is greater than 0 V, then the driver resistance must be less. Because VILC can be as low as 90 mV at cold temperatures and the low end of the current distribution, the maximum ground offset must not exceed 50 mV. Bus repeaters that use an output offset are not interoperable with the port A of the TPD12S015 as their output LOW levels are not recognized by the TPD12S015 as a LOW. If the TPD12S015 is placed in an application where the VIL of port A of the TPD12S015 does not go below its VILC it pulls port B LOW initially when port A input transitions LOW but the port B returns HIGH, so it does not reproduce the port A input on port B. Such applications must be avoided. Port B is interoperable with all I<sup>2</sup>C bus slaves, masters, and repeaters.

### 9.2.1.3 Application Curves

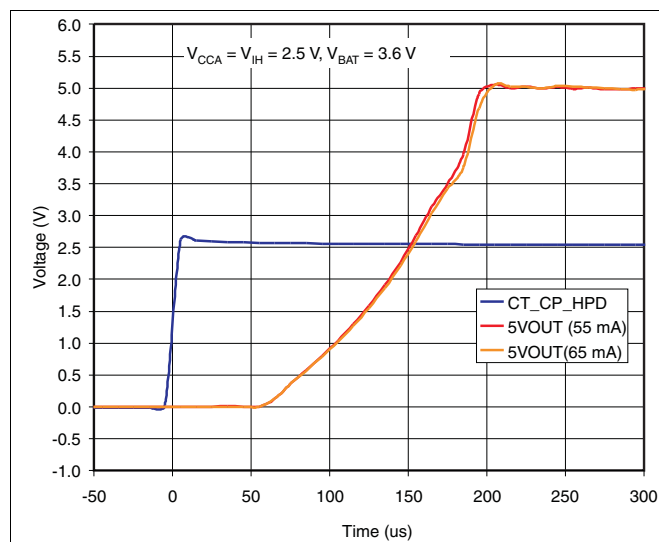


Figure 20.  $t_{START}$

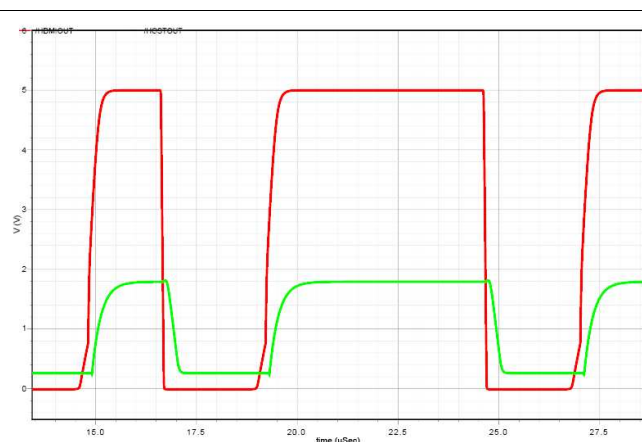
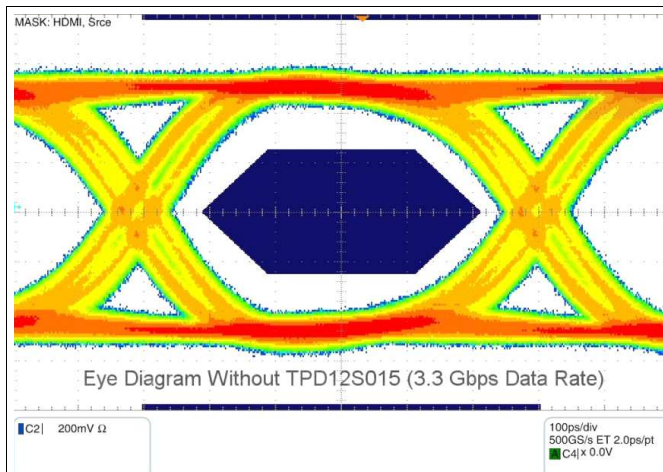


Figure 21. DDC and CEC Level Shifting Operation (B to A Direction)

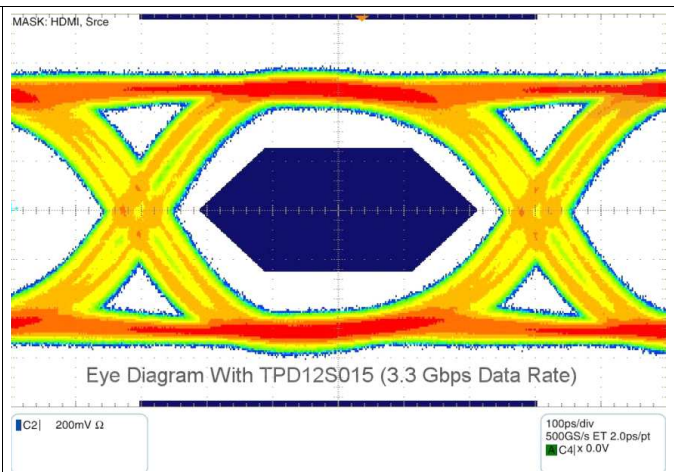
## TPD12S015

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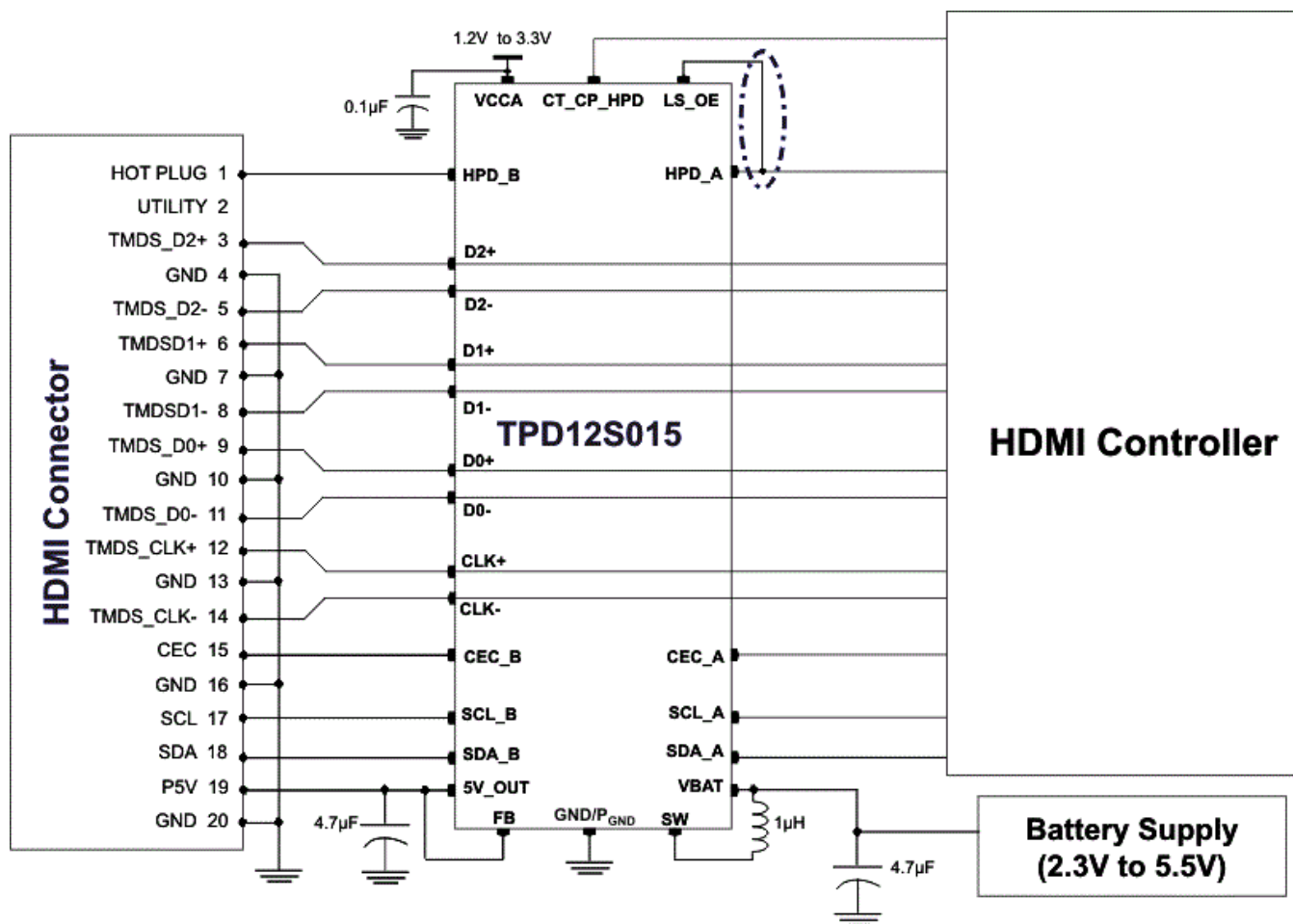
**Figure 22. Eye Diagram Performance on a Test Board for the D+ and D– Lines at 3.3 Gbps**



**Figure 23. Eye Diagram Performance on a Test Board for the D+ and D– Lines at 3.3 Gbps**

## 9.2.2 TPD12S015 Controlled by One GPIO from Controller

Some HDMI driver chips may have only one GPIO(CT\_CP\_HPDP) available. In this situation, LE\_OE pin is tied to HPD\_A instead. Figure 24 shows how TPD12S015 is used in this situation.



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Figure 24. TPD12S015 Controlled by One GPIO from Controller Schematic

### 9.2.2.1 Design Requirements

See [Design Requirements](#).

### 9.2.2.2 Detailed Design Procedure

See [Detailed Design Procedure](#).

### 9.2.2.3 Application Curves

See [Application Curves](#).

## 10 Power Supply Recommendations

See [Detailed Design Procedure](#) for detailed power supply recommendations.

## 11 Layout

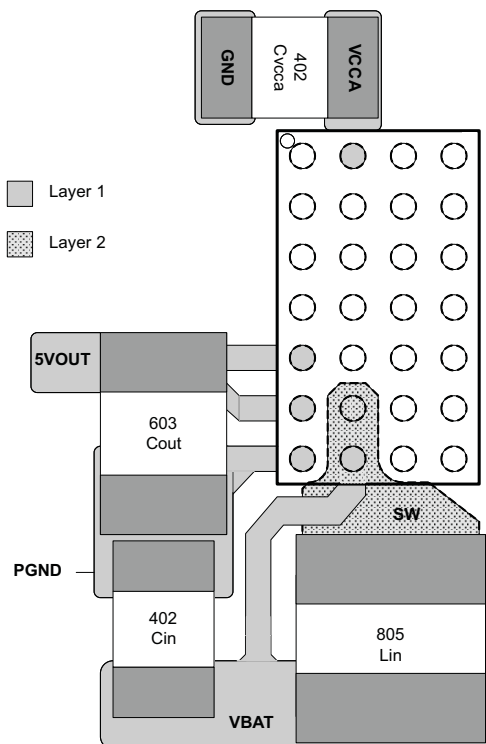
### 11.1 Layout Guidelines

For proper operation, follow these layout and design guidelines:

- Place the TPD12S015 as close to the connector as possible. This allows it to remove the energy associated with ESD strike before it reaches the internal circuitry of the system board.
- Place power line capacitors and inductors close to the pins with wide traces to allow enough current to flow through with less trace parasitics.
- Ensure that there is enough metallization for the GND pad. A sufficient current path enables safe discharge of all the energy associated with the ESD strike.
- The critical routing paths for HDMI interface are the high-speed TMDS lines. Make sure to match the lengths of the differential pair. Maintain constant trace width after to avoid impedance mismatches in the transmission lines. Maximize differential pair-to-pair spacing when possible.

For more layout information, see [TPD12S015 PCB Layout Guidelines](#).

### 11.2 Layout Example



**Figure 25. Board Layout (DC-DC Components) (Top View)**

List of components:

- $L_{IN}$  = MURATA LQM21PN1R0MC0 or  $L_{IN}$  = Toko MDT2010-CN1R0
- $C_{IN}$  = MURATA GRM188R60J225ME19 (2.2  $\mu$ F, 6.3 V, 0603, X5R) or MURATA GRM188R60J475ME19 (4.7  $\mu$ F, 6.3 V, 0603, X5R)
- $C_{OUT}$  = MURATA GRM188R60J475ME19 (4.7  $\mu$ F, 6.3 V, 0603, X5R)
- $C_{VCCA}$  = MURATA GRM155R60J104MA01 (0.1  $\mu$ F, 6.3 V, 0402, X5R)

## 12 Device and Documentation Support

### 12.1 Documentation Support

#### 12.1.1 Related Documentation

For related documentation see the following:

[TPD12S015 PCB Layout Guidelines](#) (SLVA430)

### 12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.3 Community Resource

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### 12.4 Trademarks

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### 12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPD12S015YFFR	ACTIVE	DSBGA	YFF	28	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	PN015	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPD12S015YFFR	DSBGA	YFF	28	3000	180.0	8.4	1.73	2.93	0.81	4.0	8.0	Q1



## TAPE AND REEL BOX DIMENSIONS

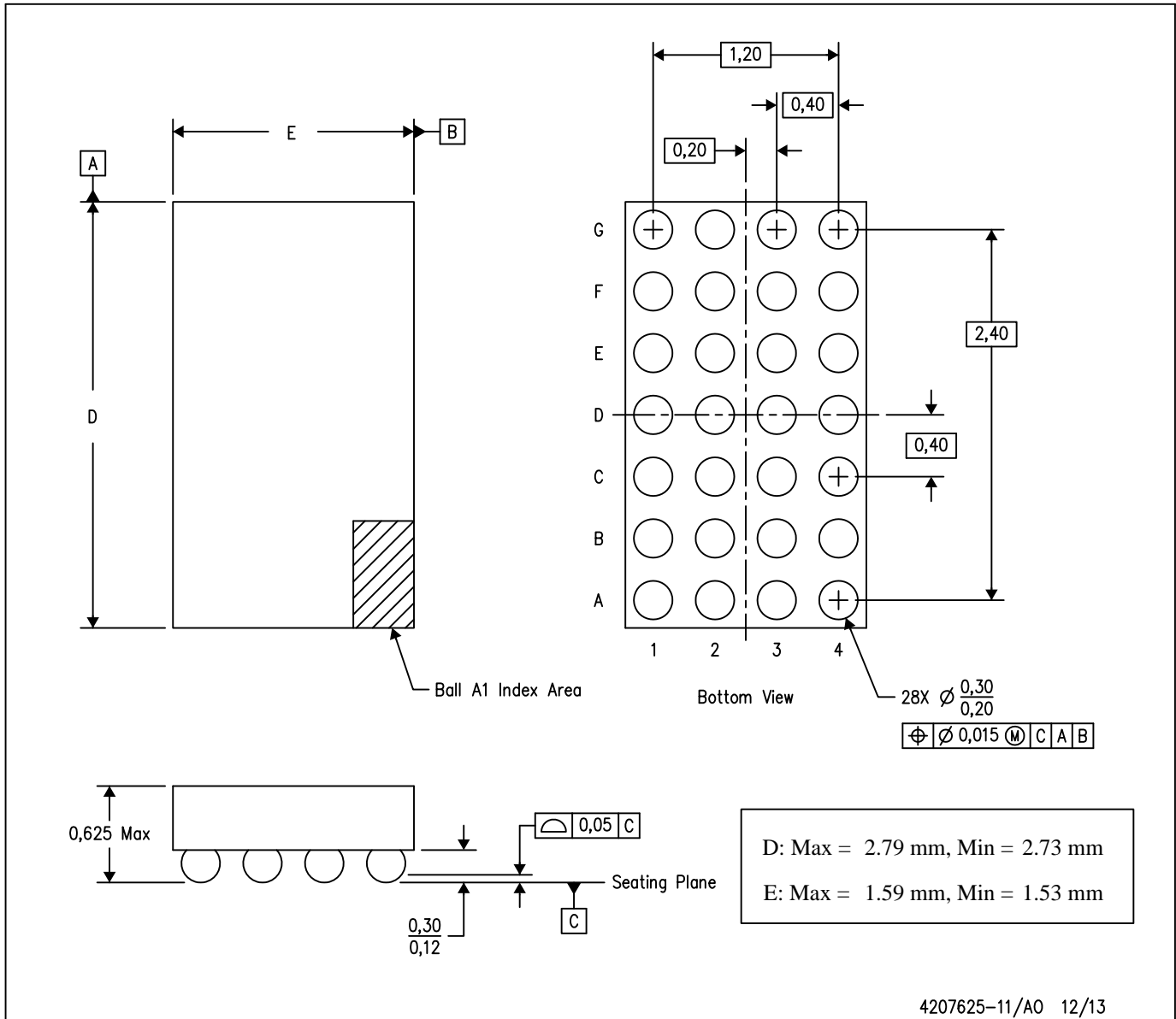


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPD12S015YFFR	DSBGA	YFF	28	3000	182.0	182.0	20.0

YFF (R-XBGA-N28)

DIE-SIZE BALL GRID ARRAY



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. NanoFree™ package configuration.

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