

# Switchmode Driver for 3- $\phi$ Brushless DC Motors

## FEATURES

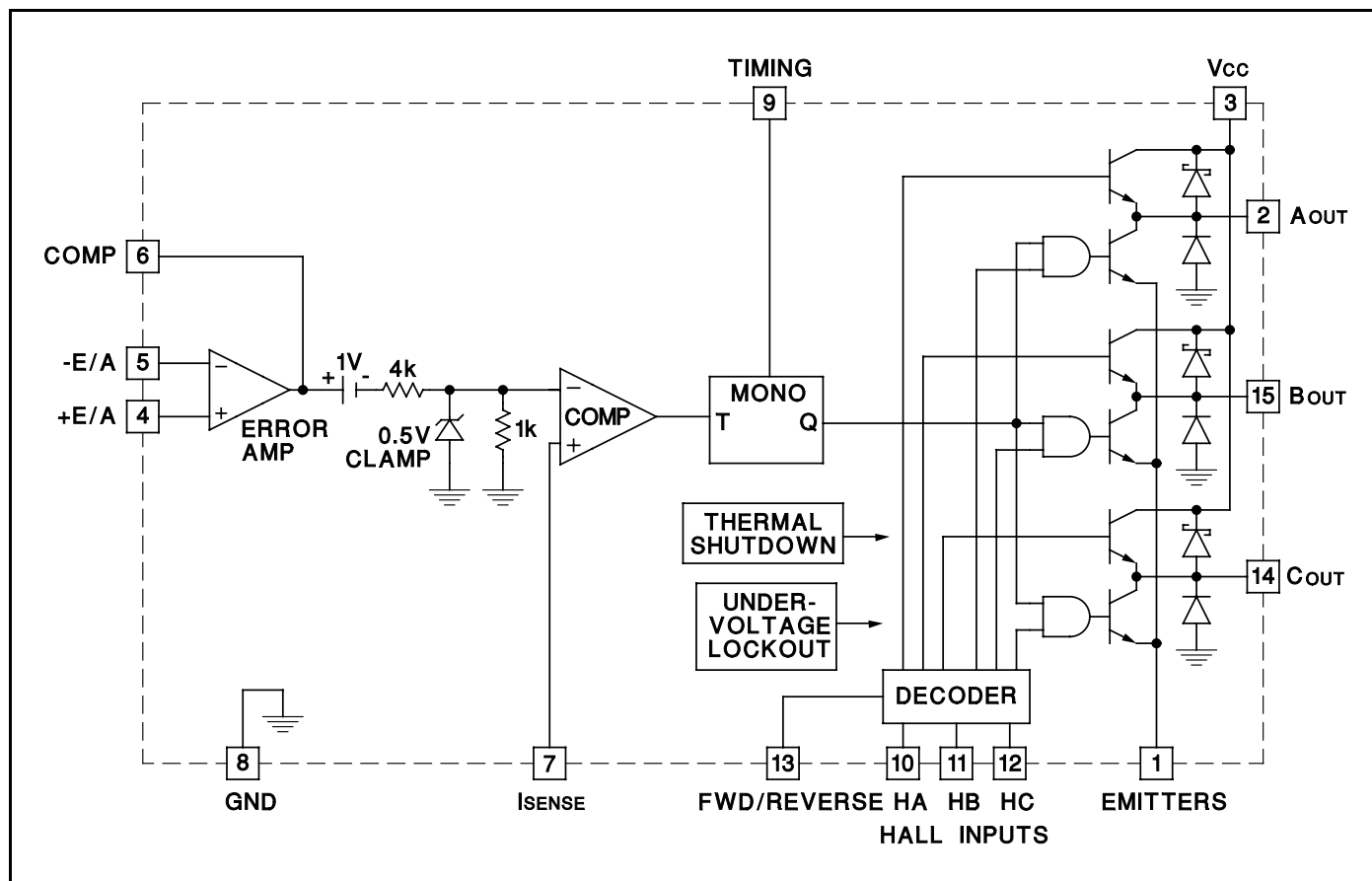
- 2A Continuous, 3A Peak Output Current
- 8V to 40V Operation
- Internal High Gain Amplifier for Velocity Control Applications
- TTL Compatible Hall Inputs
- Mask Programmable Decode Logic
- Pulse-by-Pulse Current Limiting
- Internal Thermal Shutdown Protection
- Under-Voltage Lockout
- Available in SP Hermetic Package

## DESCRIPTION

The UC3620 is a brushless DC motor driver capable of decoding and driving all 3 windings of a 3-phase brushless DC motor. In addition, an on-board current comparator, oscillator, and high gain Op-Amp provide all necessary circuitry for implementing a high performance, chopped mode servo amplifier. Full protection, including thermal shutdown, pulse-by-pulse current limiting, and under-voltage lockout aid in the simple implementation of reliable designs. Both conducted and radiated EMI have been greatly reduced by limiting the output dv/dt to 150V/ $\mu$ s for any load condition.

The UC1620SP is characterized for operation over the full military temperature range of -55°C to +125°C, while the UC3620SP is characterized for 0°C to +70°C.

## BLOCK DIAGRAM



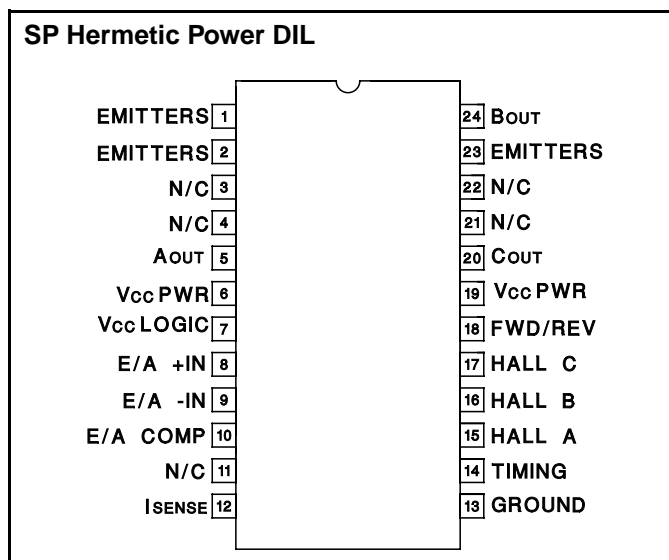
## ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage, $V_{CC}$	40V
Output Current, Source or Sink	
Non-Repetitive ( $t = 100\mu\text{sec}$ ), $I_O$	3A
Repetitive (80% on - 20% off; $t_{ON} = 10\text{ms}$ )	2.5A
DC Operation	2A
Analog Inputs	-0.3 to $+V_{CC}$
Logic Inputs	-0.3 to $+V_{CC}$
Total Power Dissipation (at $T_{CASE} = 75^\circ\text{C}$ )	
for SP Package (Note 2)	15W
Storage Temperature	$-65^\circ\text{C}$ to $+150^\circ\text{C}$
Operating Junction Temperature	$-55^\circ\text{C}$ to $+150^\circ\text{C}$

Note 1: All voltages are with respect to ground. Currents are positive into, negative out of the specified terminal.

Note 2: Consult Packaging Section of Databook for thermal limitations and considerations of package.

## CONNECTION DIAGRAMS



**ELECTRICAL CHARACTERISTICS:** Unless otherwise stated, these specifications apply for  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$  for 3620;  $T_A = 55^\circ\text{C}$  to  $+125^\circ\text{C}$  for UC1620;  $V_{CC} = 20\text{V}$ ,  $R_T = 20\text{k}\Omega$ ,  $R_T = 10\text{k}\Omega$ ,  $C_T = -2.2\text{nF}$ .  $T_A = T_J$ .

PARAMETER	TEST CONDITIONS	UC3620			UC1620			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
Error Amplifier Section								
Input Offset Voltage			1.5	10		1.5	10	mV
Input Bias Current			-.25	-2.0		-.25	-2.0	μA
Input Offset Current			15	250		15	250	nA
Common Mode Range	V <sub>CC</sub> = 8V to 40V	0		V <sub>IN-2</sub>	0		V <sub>IN-2</sub>	V
Open Loop Gain	ΔV <sub>COMP</sub> = 1V to 4V	80	100		75	100		dB
Unity Gain Bandwidth	T <sub>J</sub> = 25°C, Note 2		0.8			0.8		MHz
Output Sink Current	V <sub>COMP</sub> = 1V		2			2		mA
Output Source Current	V <sub>COMP</sub> = 4V		8			8		mA
Current Sense Section								
Input Bias Current			-2.0	-5		-2.0	-5	μA
Internal Clamp		.425	0.5	.575	.405	0.5	.595	V
Divider Gain		.180	0.2	.220	.170	0.2	.230	V/V
Internal Offset Voltage		.8	1.0	1.2	.75	1.0	1.25	V
Timing Section								
Output Off Time		18	20	22	17	20	23	μs
Upper Mono Threshold			5.0			5.0		V
Lower Mono Threshold			2.0			2.0		V
Decoder Section								
High-Level Input Voltage		2.2			2.5			V
Low-Level Input Voltage				0.8			0.8	V
High-Level Input Current				10			10	μA
Low-Level Input Current		-10			-10			μA
Output Section								
Output Leakage Current	V <sub>CC</sub> = 40V			500			1500	μA
VF1 Schottky Diode	I <sub>O</sub> = 2A		1.5	2.0		1.5	2.0	V
VF1 Substrate Diode	I <sub>O</sub> = 2A		2.2	3.0		2.2	3.0	V
Total Output Voltage Drop	I <sub>O</sub> = 2A, Note 3		3.0	3.6		3.0	3.6	V

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PARAMETER	TEST CONDITIONS	UC3620			UC1620			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
Output Section (cont.)								
Output Rise Time	RL = 44Ω		150			150		ns
Output Fall Time	RL = 44Ω		150			150		ns
Under Voltage Lockout								
Startup Threshold				8.0			8.0	V
Threshold Hysteresis			0.5			0.5		V
Thermal Shutdown								
Junction Temperature	Note 2	150		180	150		180	°C
Total Standby Current								
Supply Current			32	55		32	55	mA

Note 2: These parameters, although guaranteed over the recommended operating conditions, are not 100% tested in production.

Note 3: The total voltage drop is defined as the sum of both top and bottom side driver.

**TABLE 1**

STEP	FWD/REV	Ha	Hb	Hc	AOUT	BOUT	COUT
1	1	1	0	1	H	L	O
2	1	1	0	0	H	O	L
3	1	1	1	0	O	H	L
4	1	0	1	0	L	H	O
5	1	0	1	1	L	O	H
6	1	0	0	1	O	L	H
1	0	1	0	1	L	H	O
2	0	1	0	0	L	O	H
3	0	1	1	0	O	L	H
4	0	0	1	0	H	L	O
5	0	0	1	1	H	O	L
6	0	0	0	1	O	H	L

H = HIGH OUTPUT

L = LOW OUTPUT

O = OPEN OUTPUT

## CIRCUIT DESCRIPTION

The UC3620 is designed for implementation of a complete 3- $\phi$  brushless DC servo drive using a minimum number of external components. Below is a functional description of each major circuit feature.

### DECODER

Table 1 shows the decoding scheme used in the UC3620 to decode and drive each of three high current totem pole output stages. A forward/reverse signal, pin 13, is used to provide direction. At any point in time, one driver is sourcing, one driver is sinking, and the remaining driver is off or tri-stated. Pulse width modulation is accomplished by turning the sink driver off during the monostable reset time, producing a fixed off-time chop mode. Controlled output rise and fall times help reduce electrical switching noise while maintaining relatively small switching losses.

Hall lines require pull-up resistors.

### CURRENT SENSING

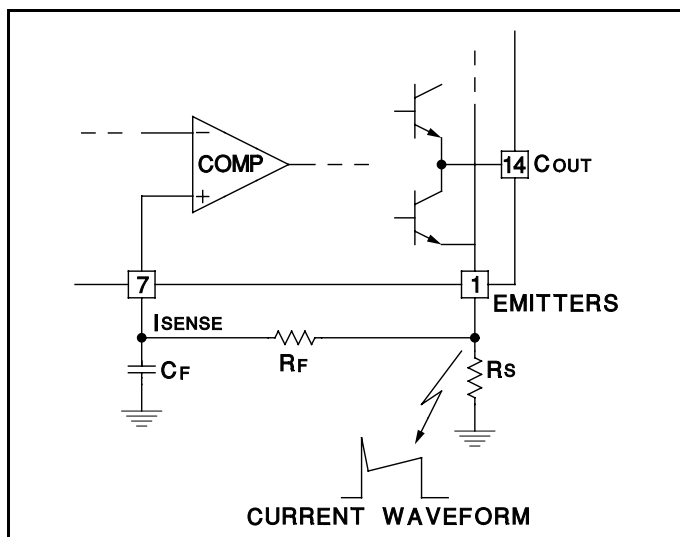
Referring to Figure 1, emitter current is sensed across  $R_S$  and fed back through a low pass filter to the current sense pin 7. This filter is required to eliminate false triggering of the monostable due to leading edge current spikes. Actual filter values, although somewhat dependent on external loads, will generally be in the  $1\text{k}\Omega$  and  $1000\text{pF}$  range.

### TIMING

An R-C time constant on pin 9 is used by the monostable to generate a fixed off time at the outputs according to the formula:

$$T_{OFF} = .916RTCT$$

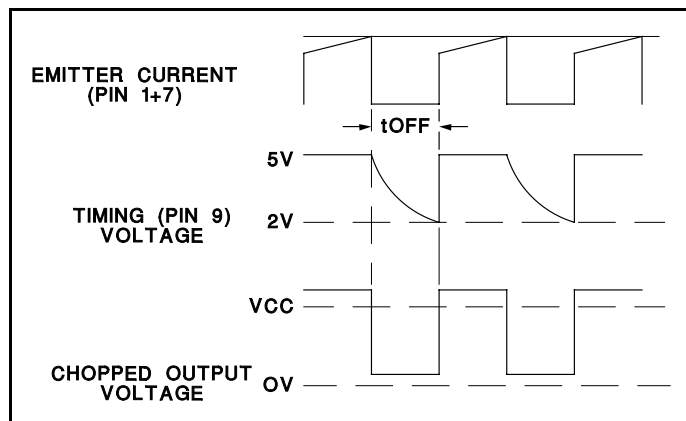
As the peak current in the emitters approaches the value at the minus (-) input of the on-board comparator, the monostable is triggered, causing the outputs to be turned



**Figure 1.** Current Sense Filter

## CIRCUIT DESCRIPTION (cont.)

off. On time is determined by the amount of time required for motor current to increase to the value required to re-trip the monostable. A timing sequence of these events is shown in Figure 2.



**Figure 2.** Chopped Mode Timing Diagram

## CURRENT LIMIT

Since peak current is being controlled at all times by the internal comparator, a simple voltage clamp at its negative (-) input will limit peak current to a maximum value. A fixed 0.5V internal clamp has been included on the UC3620, and any current spike in the output which generates a sensed voltage greater than 0.5V will immediately shut down the outputs. Actual peak current values may be programmed by selecting the appropriate value of  $R_s$  according to the formula:

$$R_s = \frac{0.5}{I_{CURRENTLIMIT}}$$

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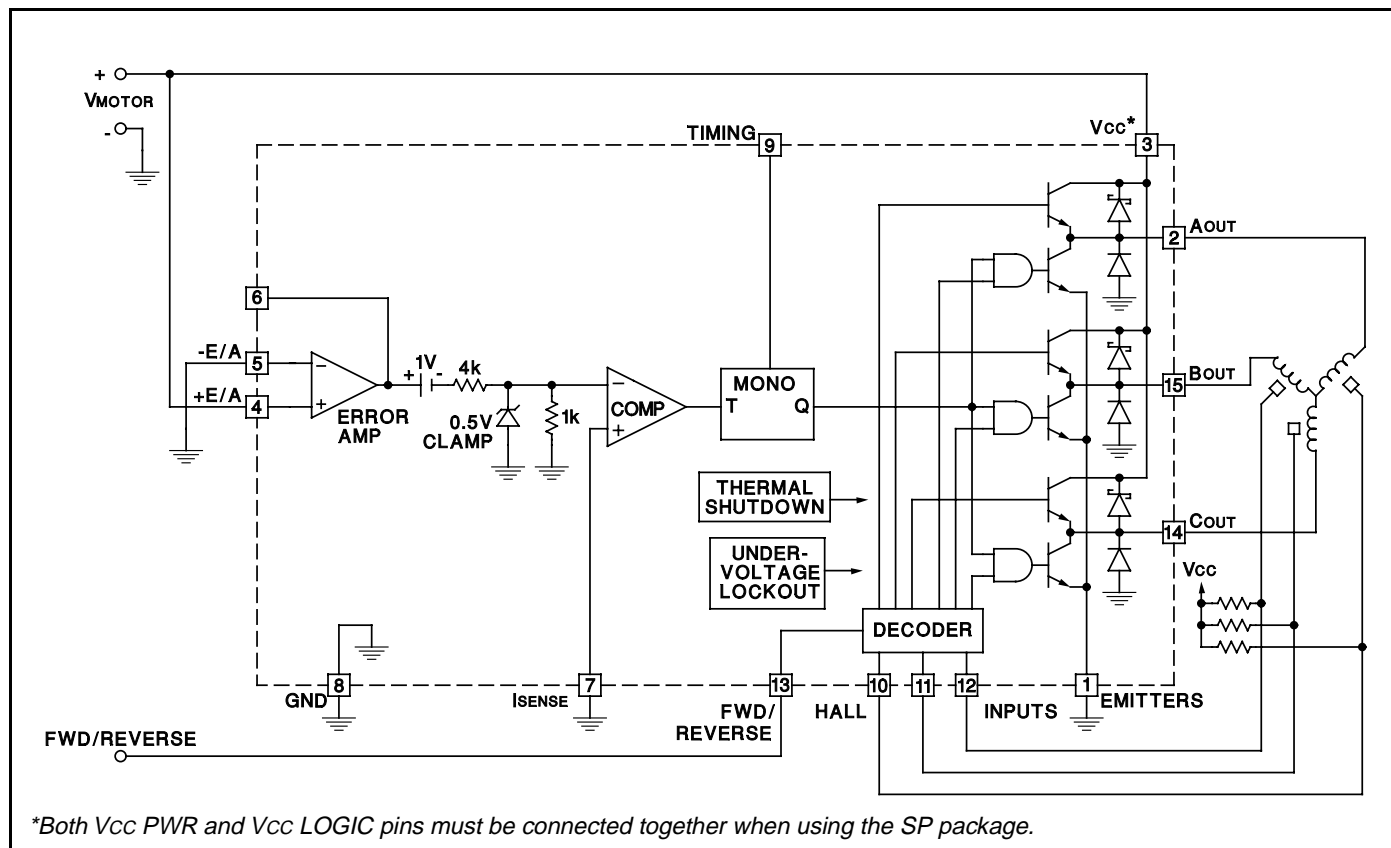
## ERROR AMPLIFIER LIMIT

A high performance, on-board error amplifier is included to facilitate implementing closed loop motor control. Error voltage generation and loop compensation are easily accomplished by appropriately configuring the gain and feedback of this amplifier. To provide a larger dynamic signal range at the output of the error amplifier, a divide by 5 resistor network is used to reduce the error signal level before applying to the internal comparator. In addition, a one volt offset has been introduced at the output of the error amplifier to guarantee control down to zero current in the output stages. Since this offset is divided by the open loop gain of the feedback loop, it has virtually no effect on closed loop performance.

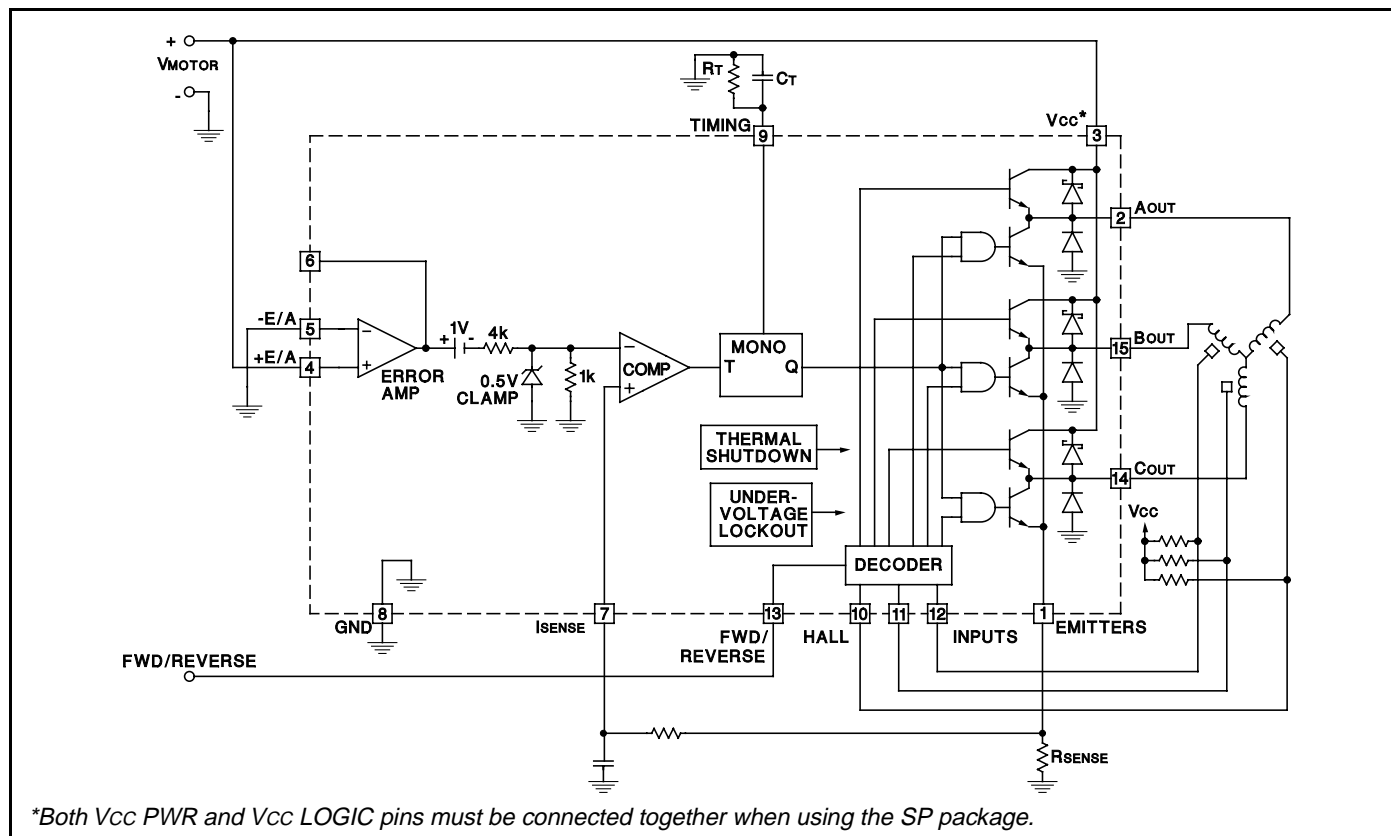
## PROTECTION FUNCTIONS

Protective functions including under-voltage lockout, peak current limiting, and thermal shutdown, provide an extremely rugged device capable of surviving under many types of fault conditions. Under-voltage lockout guarantees the outputs will be off or tri-stated until  $V_{CC}$  is sufficient for proper operation of the chip. Current limiting limits the peak current for a stalled or shorted motor, whereas thermal shutdown will tri-state the outputs if a temperature above 150°C is reached.

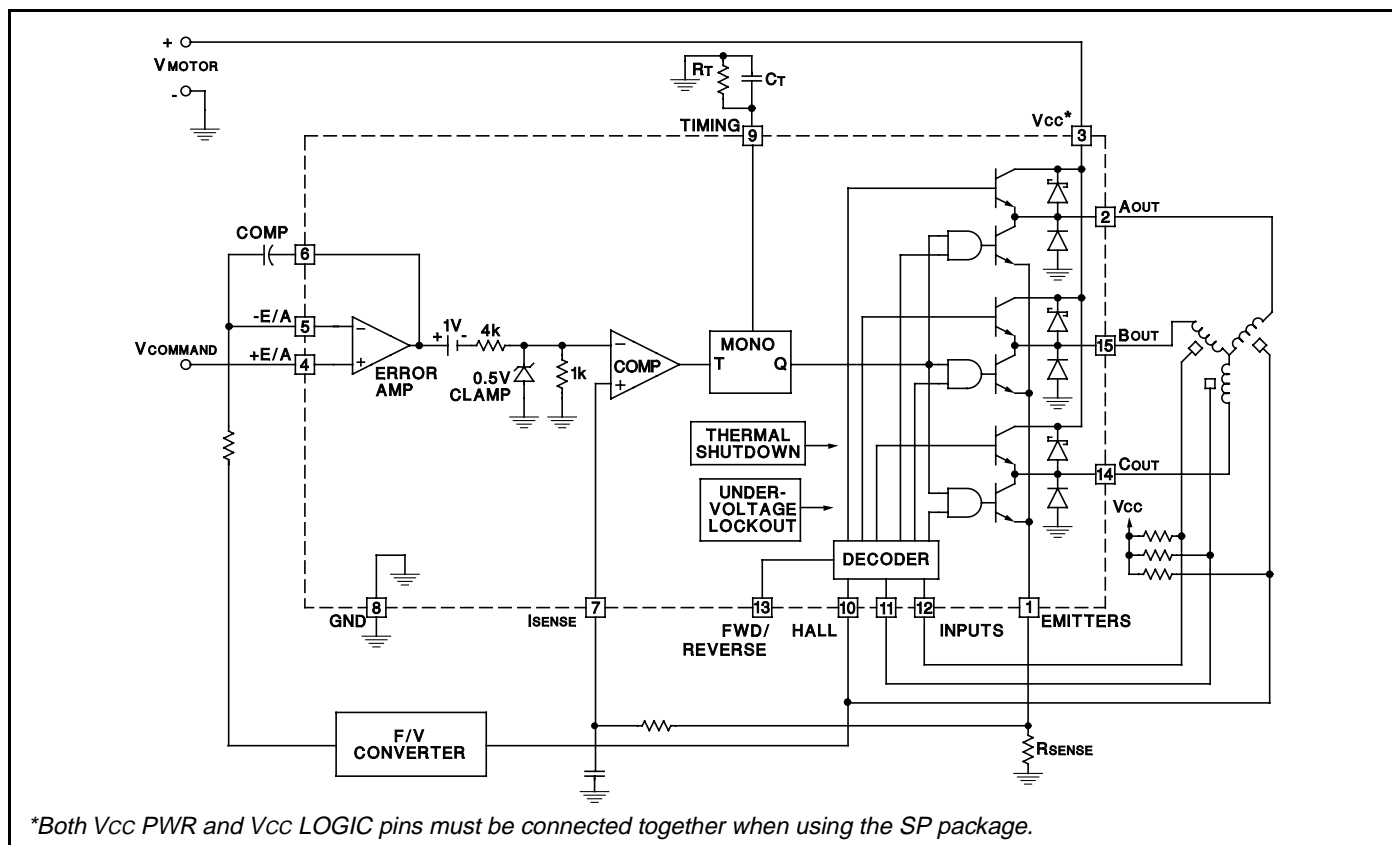
## TYPICAL APPLICATIONS



### 3- $\phi$ Brushless DC Open Loop Motor Drive



### 3- $\phi$ Brushless DC Open Loop Motor with Current Limit at 2A.



**Closed Loop Speed Control Servo**

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