Regarding the change of names mentioned in the document, such as Mitsubishi Electric and Mitsubishi XX, to Renesas Technology Corp.

The semiconductor operations of Hitachi and Mitsubishi Electric were transferred to Renesas Technology Corporation on April 1st 2003. These operations include microcomputer, logic, analog and discrete devices, and memory chips other than DRAMs (flash memory, SRAMs etc.) Accordingly, although Mitsubishi Electric, Mitsubishi Electric Corporation, Mitsubishi Semiconductors, and other Mitsubishi brand names are mentioned in the document, these names have in fact all been changed to Renesas Technology Corp. Thank you for your understanding. Except for our corporate trademark, logo and corporate statement, no changes whatsoever have been made to the contents of the document, and these changes do not constitute any alteration to the contents of the document itself.

Note: Mitsubishi Electric will continue the business operations of high frequency & optical devices and power devices.

Renesas Technology Corp. Customer Support Dept. April 1, 2003



MITSUBISHI MICROCOMPUTERS



M37905M4C-XXXFP, M37905M4C-XXXSP M37905M6C-XXXFP, M37905M6C-XXXSP M37905M8C-XXXFP, M37905M8C-XXXSP

16-BIT CMOS MICROCOMPUTER

DESCRIPTION

These are single-chip 16-bit microcomputers designed with high-performance CMOS silicon gate technology, being packaged in 64-pin plastic molded QFP or shrink plastic molded SDIP. These microcomputers support the 7900 Series instruction set, which are enhanced and expanded instruction set and are upper-compatible with the 7700/7751 Series instruction set.

The CPU of these microcomputers is a 16-bit parallel processor that can also be switched to perform 8-bit parallel processing. Also, the bus interface unit of these microcomputers enhances the memory access efficiency to execute instructions fast. Therefore, these microcomputers are suitable for office, business, and industrial equipment controller that require high-speed processing of large data. Also, they are suitable for motor-control equipment since each of

DISTINCTIVE FEATURES

them includes the motor control circuit.

Number of basic machine instructions	203
● Memory	
[M37905M4C-XXXFP, M37905M4C-XXXSP]	
ROM	32 Kbytes
RAM	1024 bytes
[M37905M6C-XXXFP, M37905M6C-XXXSP]	
ROM	48 Kbytes
RAM	3072 bytes

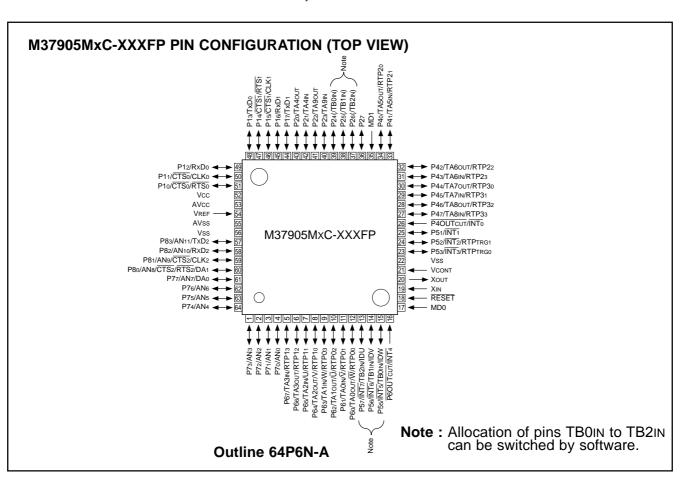
[M37905M8C-XXXFP, M37905M8C-XXXSP]
ROM 60 Kbytes
RAM
 ■ Instruction execution time
The fastest instruction at 20 MHz frequency 50 ns
● Single power supply 5 V ± 0.5 V
• Interrupts 8 external sources, 23 internal sources, 7 levels
Multi-functional 16-bit timer
(Three-phase motor drive waveform and Pulse motor drive wave-
form output are available.)
• Serial I/O (UART or Clock synchronous)
● 10-bit A-D converter12-channel inputs
● 8-bit D-A converter2-channel outputs
● 12-bit watchdog timer

APPLICATION

Control devices for office equipment such as copiers and facsimiles Control devices for industrial equipment such as communication and measuring instruments

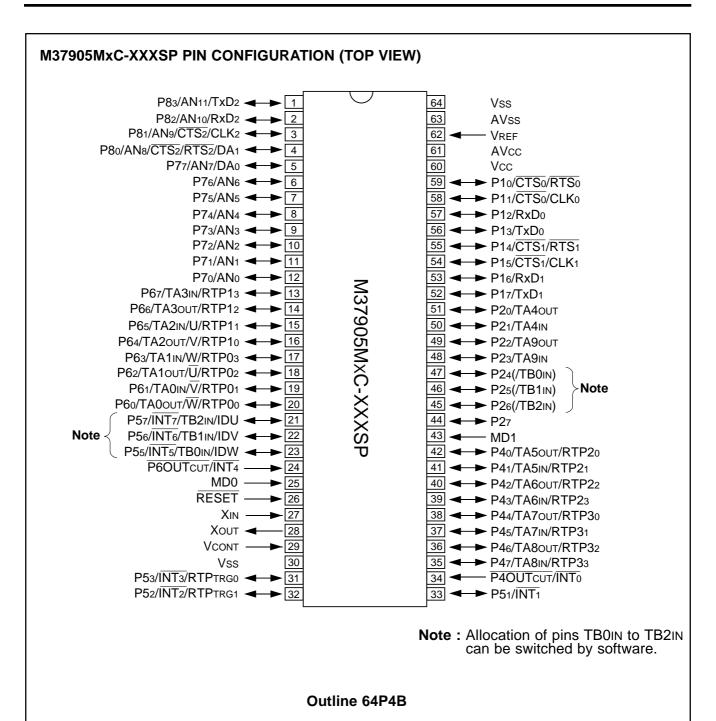
• Programmable input/output (ports P1, P2, P4, P5, P6, P7, P8) .. 50

Control devices for equipment, requiring motor control, such as inverter air conditioners and general-purpose inverters

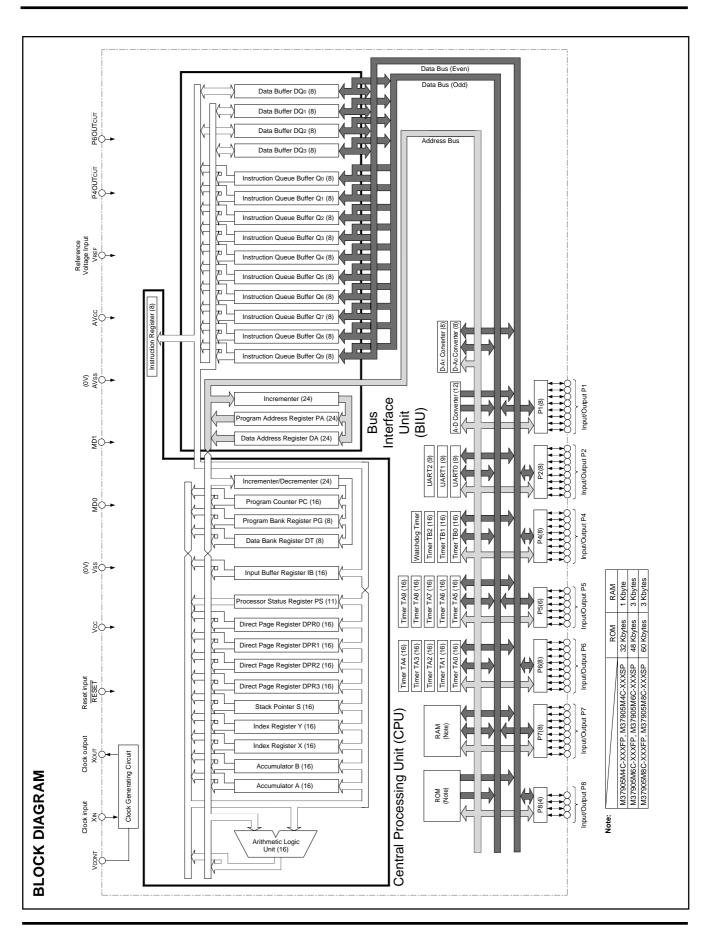














16-BIT CMOS MICROCOMPUTER

FUNCTIONS

Parameter		Functions	
Number of basic machine instructions		203	
Instruction execution time		50 ns (the fastest instruction at f(fsys) = 20 MHz)	
External clock input frequency f(XIN)		20 MHz (Max.)	
System clock frequency f(fsys)		20 MHz (Max.)	
Memory size	ROM	(Note 1)	
	RAM	(Note 1)	
Programmable input/output	P1, P2, P4, P6, P7	8-bit × 5	
ports	P5	6-bit X 1	
	P8	4-bit X 1	
Multi-functional timers	TA0-TA9	16-bit X 10	
	TB0-TB2	16-bit X 3	
Serial I/O UART0, UART1, and UART2		(UART or Clock synchronous serial I/O) X 3	
A-D converter		10-bit successive approximation method X 1 (12 channels)	
D-A converter		8-bit × 2	
Dead-time timer		8-bit X 3	
Watchdog timer		12-bit X 1	
Interrupts Maskable interrups		8 external sources, 20 internal sources. Each interrupt can be set to a priority level within the range of 0–7 by software.	
	Non-maskable interrups	3 internal sources	
Clock generating circuit		Incorporated (externally connected to a ceramic resonator or quartz-crystal resonator).	
PLL frequency multiplier		The following multiplication ratios are available: X2, X3, X4.	
Power supply voltage		5 V±0.5 V	
Power dissipation		125 mW (at f(fsys) = 20 MHz, Typ, ; the PLL frequency multiplier is inactive.)	
Ports' input/output	nput/Output withstand voltage	5 V	
characteristics	utput current	5 mA	
Memory expansion		Not available (single-chip mode only).	
Operating ambient temperatur	e range	−20 to 85 °C	
Device structure		CMOS high-performance silicon gate process	
Package		(Note 2)	

Notes 1:

ROM	M37905M4C-XXXFP, M37905M4C-XXXSP	32 Kbytes
	M37905M6C-XXXFP, M37905M6C-XXXSP	48 Kbytes
	M37905M8C-XXXFP, M37905M8C-XXXSP	60 Kbytes
RAM	M37905M4C-XXXFP, M37905M4C-XXXSP	1024 bytes
	M37905M6C-XXXFP, M37905M6C-XXXSP	3072 bytes
	M37905M8C-XXXFP, M37905M8C-XXXSP	3072 bytes

2:	Packages	M37905M4C-XXXFP, M37905M6C-XXXFP, M37905M8C-XXXFP	64-pin plastic molded QFP (64P6N-A)	
		M37905M4C-XXXSP, M37905M6C-XXXSP, M37905M8C-XXXSP	64-pin shrink plastic moldeds DIP (64P4B)	





16-BIT CMOS MICROCOMPUTER

PIN DESCRIPTION

Pin	Name	Input/ Output	Functions
Vcc, Vss	Power supply input	T -	Apply 5 V±0.5 V to Vcc, and 0 V to Vss.
MD0	MD0	Input	Connect this pin to Vss.
MD1	MD1	Input	Connect this pin to Vss.
RESET	Reset input	Input	The microcomputer is reset when "L" level is applies to this pin.
XIN	Clock input	Input	These are input and output pins of the internal clock generating circuit. Connect a
Хоит	Clock output	Output	ceramic resonator or quartz-crystal oscillator between pins XIN and XOUT. When an external clock is used, the clock source should be connected to pin XIN, and pin XOUT should be left open.
VCONT	Filter circuit connection	_	When using the PLL frequency multiplier, connect this pin to the filter circuit. When not using the PLL frequency multiplier, this pin should be left open.
AVcc, AVss	Analog power supply input	_	Power supply input pins for the A-D and D-A converters. Connect AVcc to Vcc, and AVss to Vss externally.
VREF	Reference voltage input	Input	This is the reference voltage input pin for the A-D and D-A converters.
P10-P17	I/O port P1	I/O	Port P1 is an 8-bit I/O port. This port has an I/O direction register, and each pin can be programmed for input or output. These pins enter the input mode ar reset. These pins also function as I/O pins of UARTO, 1.
P20-P27	I/O port P2	I/O	In addition to having the same functions as port P1, these pins function as I/O pins for timers A4 and A9. Also, they can be programmed to function as input pins for timers B0 to B2.
P40-P47	I/O port P4	I/O	In addition to having the same functions as port P1, these pins function as I/O pins for timers A5 to A8. Also, they function as output pins for motor drive waveform.
P51–P53, P55–P57	I/O port P5	I/O	In addition to having the same functions as port P1, these pins function as input pins for INT1 to INT3 and INT5 to INT7. Also, pins P55 to P57 function as input pins for timers B0 to B2 and as input pins for position data in the three-phase waveform mode; and pins P52 and P53 function as trigger-input pins in the pulse output port mode.
P60-P67	I/O port P6	I/O	In addition to having the same functions as port P1, these pins function as I/O pins for timers A0 to A3. Also, they function as motor drive waveform output pins.
P70-P77	I/O port P7	I/O	In addition to having the same functions as port P1, these pins function as input pins for the A-D converter. Also, P77 functions as an output pin for the D-A converter.
P80-P83	I/O port P8	I/O	In addition to having the same functions as port P1, these pins function as input pins for the A-D converter. Also, these pins function as I/O pins for UART2,and pin P80 functions as an output pin for the D-A converter.
P4OUTcut	P4OUTcuT input	Input	This pin has the function to forcibly place port P4 pins in the input mode. Also, this pin functions as an input pin for INTo; and this pin is used to input a signal, which forcibly cuts off a motor drive waveform output.
P6OUTcut	P6OUTcuT input	Input	This pin has the function to forcibly place port P6 pins in the input mode. Also, this pin functions as an input pin for INT4; and this pin is used to input a signal, which forcibly cuts off a motor drive waveform output.





16-BIT CMOS MICROCOMPUTER

BASIC FUNCTION BLOCKS

These microcomputers contain the following devices in the single chip: ROM, RAM, CPU, bus interface unit, and peripheral devices such as the interrupt control circuit, timers, serial I/O, A-D converter, D-A converter, I/O ports, clock generating circuit, etc.

MEMORY

Figures 1 (1) through (3) show the memory maps. The address space is 64 Kbytes from addresses 016 through FFFF16. This ad-

dress space is called "bank 016".

The internal ROM and RAM are allocated as shown in Figures 1 (1) through (3).

Addresses FFB416 through FFFF16 contain the RESET and the interrupt vector addresses, and the interrupt vectors are stored there. For details, refer to the section on interrupts.

Allocated to addresses 016 through FF16 are peripheral devices such as I/O ports, A-D converter, D-A converter, serial I/O, timers, interrupt control registers, etc. Figures 2 and 3 show the location of SFRs.

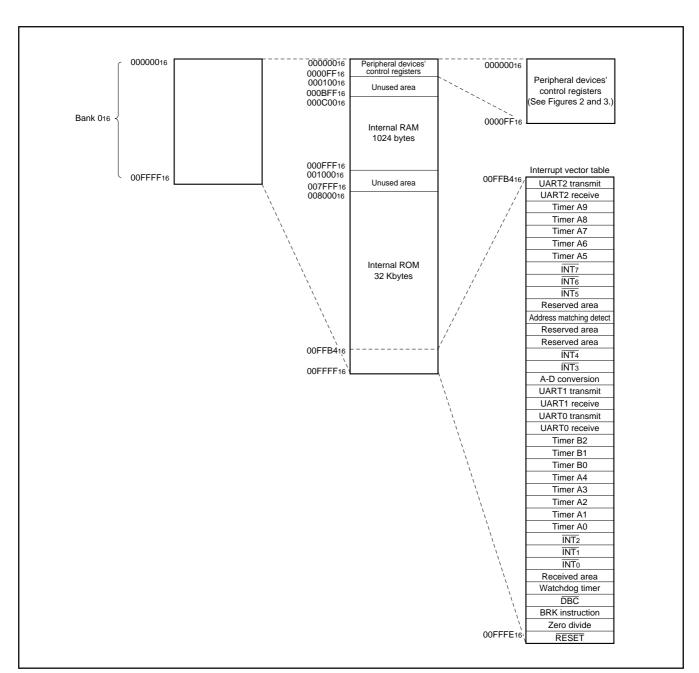


Fig. 1 (1) Memory map of M37905M4C-XXXFP/SP (Single-chip mode)





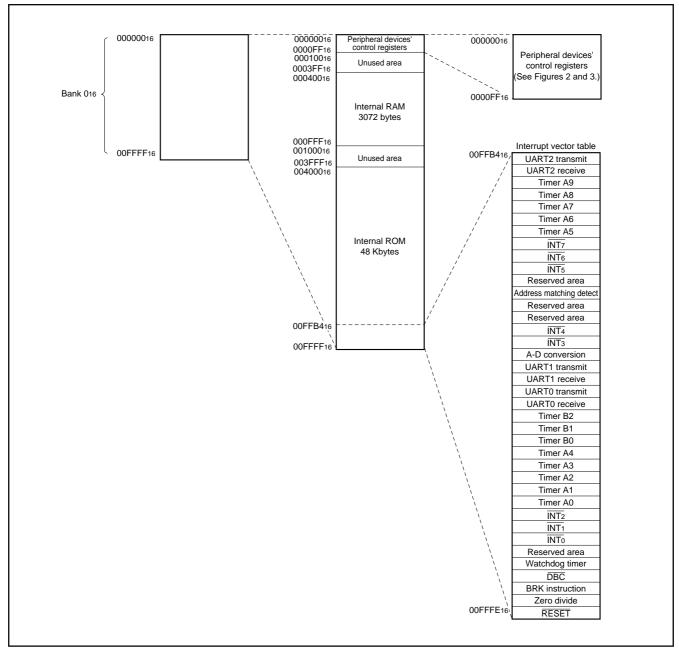


Fig. 1 (2) Memory map of M37905M6C-XXXFP/SP (Single-chip mode)



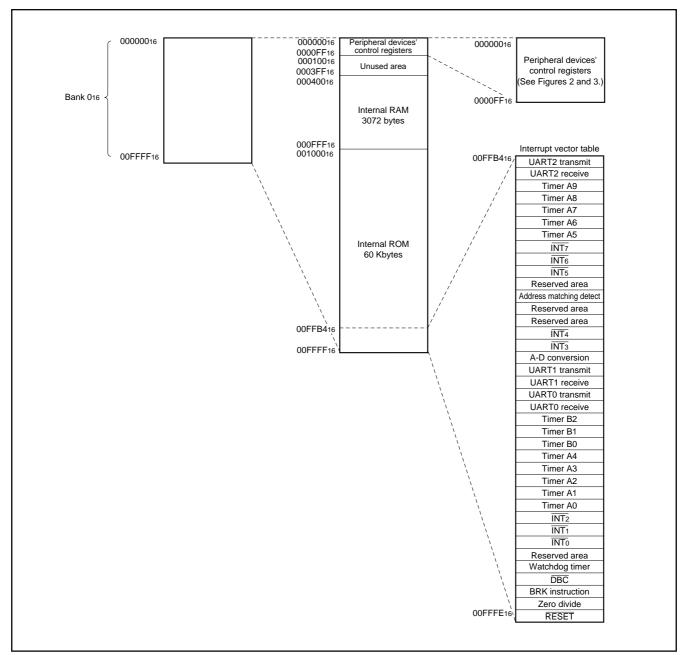


Fig. 1 (3) Memory map of M37905M8C-XXXFP/SP (Single-chip mode)



Fig. 2 Location of SFRs (1)



00008116 00008216	Reserved area (Note)	0000C016	
00008216	Reserved area (Note)	0000C116	
	Reserved area (Note) Reserved area (Note)	0000C116 0000C216	
00008316	Reserved area (Note)	0000C216	
	Reserved area (Note)	0000C316	Up-down register 1
	Reserved area (Note)	0000C516	op demiregieter i
	Reserved area (Note)	0000C616	
00008716	Reserved area (Note)	0000C716	Timer A5 register
00008816		0000C816	
00008916		0000C916	Timer A6 register
00008A16	Reserved area (Note)	0000CA16	Timer A7 register
00008B16		0000CB16	Timer A7 register
00008C16	Reserved area (Note)	0000CC16	Timer A8 register
00008D16		0000CD16	Timer //o register
00008E16	Reserved area (Note)	0000CE16	Timer A9 register
00008F16	Decembed area (Nets)	0000CF16	
00009016	Reserved area (Note)	0000D016	Timer A01 register
00009116	Peneryod area (Nete)	0000D116	<u> </u>
00009216	Reserved area (Note)	0000D216 0000D316	Timer A11 register
00009316		0000D316 0000D416	
00009416	External interrupt input read-out register	0000D416	Timer A21 register
	D-A control register	0000D316	Timer A5 mode register
00003016		0000D016	Timer A6 mode register
⊢	D-A register 0	0000D816	Timer A7 mode register
	D-A register 1	0000D916	Timer A8 mode register
00009A16		0000DA16	Timer A9 mode register
00009B16		0000DB16	A-D control register 2
00009C16		0000DC16	Comparator function select register 0
00009D16		0000DD16	Comparator function select register 1
00009E16		0000DE16	Comparator result register 0
00009F16		0000DF16	Comparator result register 1
-	Pulse output control register	0000E016	A-D register 8
0000A116		0000E116	D. Oglotor O
	Pulse output data register 0	0000E216	A-D register 9
0000A316	Dulas sutant data assists 4	0000E316	J
0000A416	Pulse output data register 1	0000E416	A-D register 10
0000A516 0000A616	Waveform output mode register	0000E516 0000E616	
F	Dead-time timer	0000E616	A-D register 11
	Three-phase output data register 0	0000E716	Reserved area (Note)
0000A816	Three-phase output data register 1	0000E916	Reserved area (Note)
	Position-data-retain function control register	0000EA16	Reserved area (Note)
0000AB16	39.0.0	0000EB16	Reserved area (Note)
⊢	Serial I/O pin control register	0000EC16	Reserved area (Note)
0000AD16		0000ED16	Reserved area (Note)
0000AE16	Port P2 pin function control register	0000EE16	Reserved area (Note)
0000AF16		0000EF16	Reserved area (Note)
	UART2 transmit/receive mode register	0000F016	
	UART2 band rate register (BRG2)	0000F116	UART2 transmit interrupt control register
0000B216	UART2 transmit buffer register	0000F216	UART2 receive interrupt control register
0000B316	•	0000F316	
<u> </u>	UART2 transmit/receive control register 0	0000F416	Times AE intersunt control as sister
0000B516	UART2 transmit/receive control register 1	0000F516	Timer A5 interrupt control register
0000B616 0000B716	UART2 receive buffer register	0000F616	Timer A6 interrupt control register Timer A7 interrupt control register
	Reserved area (Note)	0000F716 0000F816	Timer A7 interrupt control register Timer A8 interrupt control register
0000B816 0000B916	Neserved area (Note)	0000F816	Timer A9 interrupt control register
	Reserved area (Note)	0000F916	rimor no interrupt control register
	Reserved area (Note)	0000FB16	
⊢	Clock control register 0	0000FC16	
	Reserved area (Note)	0000FD16	INT5 interrupt control register
	Reserved area (Note)	0000FE16	INT6 interrupt control register
	Reserved area (Note)	0000FF16	
· · · · - L	, ,		

Fig. 3 Location of SFRs (2)





16-BIT CMOS MICROCOMPUTER

CENTRAL PROCESSING UNIT (CPU)

The CPU has 13 registers and is shown in Figure 4. Each of these registers is described below.

ACCUMULATOR A (A)

Accumulator A is the main register of the microcomputer. It consists of 16 bits and the low-order 8 bits can be used separately. Data length flag m determines whether the register is used as 16-bit register or as 8-bit register. It is used as a 16-bit register when flag m is "0" and as an 8-bit register when flag m is "1". Flag m is a part of the processor status register (PS) which is described later.

Data operations such as calculations, data transfer, input/output, etc., are executed mainly through accumulator A.

ACCUMULATOR B (B)

Accumulator B has the same functions as accumulator A, but the use of accumulator B requires more instruction bytes and execution cycles than accumulator A.

ACCUMULATOR E

Accumulator E is a 32-bit register and consists of accumulator A (low-order 16 bits) and accumulator B (high-order 16 bits). It is used for 32-bit data processing.

INDEX REGISTER X (X)

Index register X consists of 16 bits and the low-order 8 bits can be used separately. Index register length flag x determines whether the register is used as 16-bit register or as 8-bit register. It is used as a 16-bit register when flag x is "0" and as an 8-bit register when flag x is "1". Flag x is a part of the processor status register (PS) which is described later.

In index addressing modes in which register X is used as the index register, the contents of this address are added to obtain the real address.

Index register X functions as a pointer register which indicates an address of data table in instructions MVP, MVN, RMPA (Repeat MultiPly and Accumulate).

INDEX REGISTER Y (Y)

Index register Y consists of 16 bits and the low-order 8 bits can be used separately. The index register length flag x determines whether the register is used as 16-bit register or as 8-bit register. It is used as a 16-bit register when flag x is "0" and as an 8-bit register when flag x is "1". Flag x is a part of the processor status register (PS) which is described later.

In index addressing modes in which register Y is used as the index register, the contents of this address are added to obtain the real address.

Index register Y functions as a pointer register which indicates an address of data table in instructions MVP, MVN, RMPA (Repeat MultiPly and Accumulate).

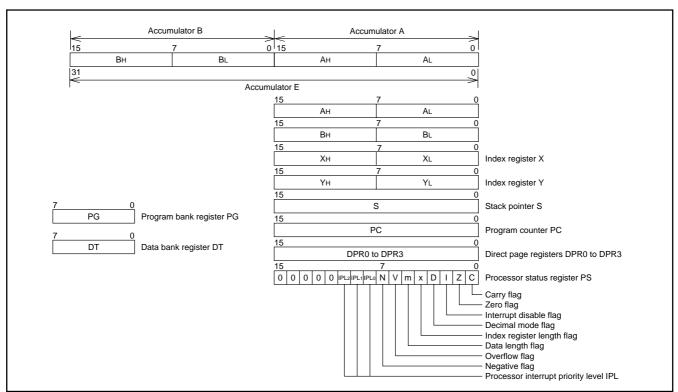


Fig. 4 Register structure



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STACK POINTER (S)

Stack pointer (S) is a 16-bit register. It is used during a subroutine call or interrupts. It is also used during stack, stack pointer relative, or stack pointer relative indirect indexed Y addressing mode.

PROGRAM COUNTER (PC)

Program counter (PC) is a 16-bit counter that indicates the low-order 16 bits of the next program memory address to be executed. There is a bus interface unit between the program memory and the CPU, so that the program memory is accessed through bus interface unit. This is described later.

PROGRAM BANK REGISTER (PG)

Program bank register is an 8-bit register that indicates the high-order 8 bits of the next program memory address to be executed. When a carry occurs by incrementing the contents of the program counter, the contents of the program bank register (PG) is increased by 1. Also, when a carry or borrow occurs after adding or subtracting the offset value to or from the contents of the program counter (PC) using the branch instruction, the contents of the program bank register (PG) is increased or decreased by 1, so that programs can be written without worrying about bank boundaries.

DATA BANK REGISTER (DT)

Data bank register (DT) is an 8-bit register. With some addressing modes, the data bank register (DT) is used to specify a part of the memory address. The contents of data bank register (DT) is used as the high-order 8 bits of a 24-bit address. Addressing modes that use the data bank register (DT) are direct indirect, direct indexed X indirect, direct indirect indexed Y, absolute, absolute bit, absolute indexed X, absolute indexed Y, absolute bit relative, and stack pointer relative indirect indexed Y.

DIRECT PAGE REGISTERS 0 through 3 (DPR0 through DPR3)

The direct page register is a 16-bit register. An addressing mode of which name includes 'direct' generates an address of data to be accessed, regarding the contents of this register as the base address. The 7900 Series has been expanded direct page registers up to 4 (DPR0 to DPR3), in comparison to the 7700 Series which has the single direct page register. Accordingly, the 7900 Series's direct addressing method which uses direct page registers differs from that of the 7700 Series. However, the conventional direct addressing method, using only DPR0, is still be selectable, in order to make use of the 7700 Series software property. For more details, refer to the section on the direct page.

PROCESSOR STATUS REGISTER (PS)

Processor status register (PS) is an 11-bit register. It consists of flags to indicate the result of operation and CPU interrupt levels. Branch operations can be performed by testing the flags C, Z, V, and N.

The details of each bit of the processor status register are described below.

1. Carry flag (C)

The carry flag contains the carry or borrow generated by the ALU after an arithmetic operation. This flag is also affected by shift and rotate instructions. This flag can be set and reset directly with the SEC and CLC instructions or with the SEP and CLP instructions.

2. Zero flag (Z)

The zero flag is set if the result of an arithmetic operation or data transfer is zero and reset if it is not. This flag can be set and reset directly with the SEP and CLP instructions.

3. Interrupt disable flag (I)

When the interrupt disable flag is set to "1", all interrupts except watchdog timer and software interrupts are disabled. This flag is set to "1" automatically when an interrupt is accepted. It can be set and reset directly with the SEI and CLI instructions or SEP and CLP instructions.

4. Decimal mode flag (D)

The decimal mode flag determines whether addition and subtraction are performed as binary or decimal. Binary arithmetic is performed when this flag is "0". If it is "1", decimal arithmetic is performed with each word treated as 2- or 4- digit decimal. Arithmetic operation is performed using four digits when data length flag m is "0" and with two digits when it is "1". Decimal adjust is automatically performed. (Decimal operation is possible only with the ADC and SBC instructions.) This flag can be set and reset with the SEP and CLP instructions.





16-BIT CMOS MICROCOMPUTER

5. Index register length flag (x)

The index register length flag determines whether index register X and index register Y are used as 16-bit registers or as 8-bit registers. The registers are used as 16-bit registers when flag x is "0" and as 8-bit registers when it is "1".

This flag can be set and reset with the SEP and CLP instructions.

6. Data length flag (m)

The data length flag determines whether the data length is 16-bit or 8-bit. The data length is 16 bits when flag m is "0" and 8 bits when it is "1". This flag can be set and reset with the SEM and CLM instructions or with the SEP and CLP instructions.

7. Overflow flag (V)

The overflow flag is valid when addition or subtraction is performed with a word treated as a signed binary number. If data length flag m is "0", the overflow flag is set when the result of addition or subtraction is outside the range between –32768 and +32767. If data length flag m is "1", the overflow flag is set when the result of addition or subtraction is outside the range between –128 and +127. It is reset in all other cases. The overflow flag can also be set and reset directly with the SEP, and CLV or CLP instructions.

Additionally, the overflow flag is set when a result of unsigned/signed division exceeds the length of the register where the result is to be stored; the flag is also set when the addition result is outside range of –2147483648 to +2147483647 in the RMPA operation.

8. Negative flag (N)

The negative flag is set when the result of arithmetic operation or data transfer is negative (If data length flag m is "0", data's bit 15 is "1". If data length flag m is "1", data's bit 7 is "1".) It is reset in all other cases. It can also be set and reset with the SEP and CLP instructions.

9. Processor interrupt priority level (IPL)

The processor interrupt priority level (IPL) consists of 3 bits and determines the priority of processor interrupts from level 0 to level 7. Interrupt is enabled when the interrupt priority of the device requesting interrupt (set using the interrupt control register) is higher than the processor interrupt priority. When an interrupt is enabled, the current processor interrupt priority level is saved in a stack and the processor interrupt priority level is replaced by the interrupt priority level of the device requesting the interrupt. Refer to the section on interrupts for more details.

Note: Fix bits 11 to 15 of the processor status register (PS) to "0".





16-BIT CMOS MICROCOMPUTER

BANK

In order to effectively use the integrated hardware on the chip, this CPU core uses an address generating method with a 24-bit address split into high-order 8 bits and low-order 16 bits. In other words, the 64 Kbytes specified by the low-order 16 bits are one unit (referred to as "bank"), and the address space is divided into 256 banks (016 to FF16) specified by the high-order 8 bits.

In the program area on the address space, the bank is specified by the program bank register (PG), and the address in the bank is specified by the program counter (PC).

As for each bank boundary, when an overflow has occurred in PC, the contents of PG are incremented by 1. When a borrow has occurred in PC, the contents of PG are decremented by 1. Under the normal conditions, therefore, programming without concern for the bank boundaries is possible. Furthermore, as for the data area on the address space, the bank is specified by the data bank register (DT), and the address in the bank is specified by the operation result by using the various addressing modes (Note).

Note: Some addressing modes directly specify a bank.

DIRECT PAGE

The internal memory and control registers for internal peripheral devices, etc. are assigned to bank 016 (addresses 016 to FFFF16). The direct page and direct addressing modes have been provided for the effective access to bank 016. In the 7900 Series, two types of direct addressing modes are available: the conventional direct addressing mode which uses only DPR0, as in the 7700 Series, and the expanded direct addressing mode, which uses up to 4 direct page registers as selected by the user. The addressing mode is selected according to the contents of bit 1 of the processor mode register 1. This bit 1 is cleared to "0" at reset. (In other words, the conventional direct addressing mode is selected.) However, once this bit 1 has been set to "1" by software, this bit cannot be cleared to "0" again, except by reset. That is to say, when one of these two direct addressing modes has been selected just after reset, the selected addressing mode cannot be switched to another one while the program is running.

■ Conventional direct addressing mode

The direct page area consists of 256-byte space. Its bank address is "0016", and the base address of its low-order 16-bit address is specified by the contents of the direct page register 0 (DPR0). In this conventional direct addressing modes, a value (1 byte) just after an instruction code is regarded as an offset value for the DPR0 contents, and the CPU accesses each address in the direct page area.

■ Expanded direct addressing mode

The direct page area consists of four 64-byte spaces. Their bank address is "0016", and the four base addresses of their low-order 16-bit addresses are respectively specified by the contents of four direct page registers. In this expanded direct addressing mode, a value (1 byte) just after an instruction code is regarded as follows:

- High-order 2 bits: regarded as a selection field for DPR0 to DPR3.
- Low-order 6 bits: regarded as an offset value for the selected direct page register.

Then, the CPU accesses each address in each direct page area:

Refer to "7900 Series Software Manual" for details concerning the various addressing modes which use the direct page area.

Instruction Set

The CPU core of the 7900 Series has an expanded instruction set based on the existing 7700/7751 Series' CPU core. In addition, its source code (mnemonic) has the complete upper compatibility with the 7700 Series instruction set.

For details concerning addressing modes and instruction set, refer to "7900 Series Software Manual".





16-BIT CMOS MICROCOMPUTER

BUS INTERFACE UNIT

Data transfer between the central processing unit (CPU) and internal memory, internal peripheral devices is always performed via the bus interface unit (BIU), which is located between the CPU and the internal buses.

Figure 5 shows the BIU and the bus structure. The CPU and BIU are connected by a dedicated bus, and any transfer between the CPU and BIU is controlled by this dedicated bus.

On the other hand, data transfer between the BIU and internal pe-

ripheral devices uses the following internal common buses: 32-bit code bus, 16-bit data bus, 24-bit address bus, and control signals. The bus control method where the code bus and the data bus separate out (hereafter, this method is referred to as the separate code/data bus method) is employed in order to improve data transfer capabilities. As a result, the internal memory is connected to both the code bus and the data bus, and registers of all other internal peripheral devices are connected only to the data bus.

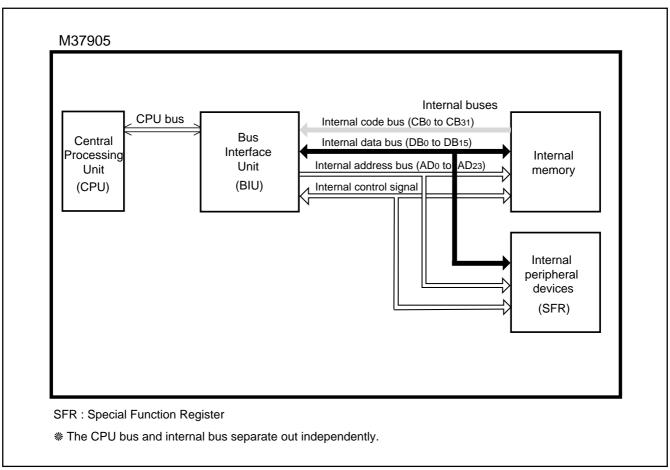


Fig. 5 BIU and bus structure



16-BIT CMOS MICROCOMPUTER

BIU structure

The BIU consists of four registers shown in Figure 6. Table 1 lists the functions of each register.

Table 1. Functions of each register

Name	Functions
Program address register	Indicates a storage address for an instruction to be next taken into an instruction queue buffer.
Instruction queue buffer	Temporarily stores an instruction which has been taken from a memory. Consists of 10 bytes.
Data address register	Indicates an address where data will be next read from or written to.
Data buffer	Temporarily stores data which has been read from internal memory or internal peripheral devices by the BIU; or temporarily stores data which is to be written to internal memory or internal peripheral devices by the CPU. Consists of 32 bits.

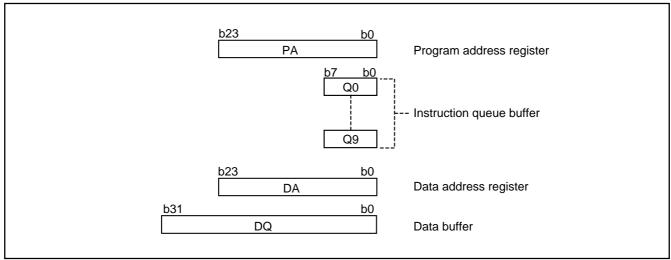


Fig. 6 Register structure of BIU



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BIU Functions

(1) Instruction prefetch

The BIU has ten instruction queue buffers; each buffer consists of 1 byte. When there is an opening in the bus and the instruction queue buffer, an instruction code is read from the program memory (in other words, the memory where a program is stored) and prefetched into an instruction queue buffer. The prefetched instruction code is transferred from the BIU to the CPU, in response to a request from the CPU, via a dedicated bus.

When a branch occurs as a result of a branch instruction (JMP, BRA, etc.), subroutine call, or interrupt, the contents of the instruction queue buffer are initialized and the BIU reads a new instruction from the branch destination address.

Note that the operations of the BIU instruction prefetch also differ depending on the store addresses for instructions. The store addresses for instructions to be prefetched are categorized as listed in Table 2.

(2) Data read operation

When executing an instruction for reading data from the internal memory or internal peripheral devices, at first, the CPU informs the BIU's data address register of the address where the data has been located

Next, the BIU reads the above data from the specified address, passes it to the data buffer, and then, transfers it to the CPU.

(3) Data write operation

When executing an instruction for writing data into the internal memory or internal peripheral devices, at first, the CPU informs the BIU's data address register of the address where the data has been located.

Next, the BIU passes the above data to the data buffer register, and then, writes it into the specified address.

(4) Bus cycle

In order for the BIU to execute the above operations (1) through (3), the 24-bit address bus, 32-bit code bus, 16-bit data bus and internal control signals must be appropriately controlled during data transfer between the BIU and internal memory or internal peripheral devices. This operation is called "bus cycle". The bus cycle is affected by the lengh of data to be transferred (byte, word, or double-word) at data access.

Table 2. Store addresses for instructions to be prefetched

	Low-order 3 bits of store address for instruction			
	AD2 (A2)	AD1 (A1)	AD ₀ (A ₀)	
Even address	Χ	Х	0	
4-byte boundary	Χ	0	0	
8-byte boundary	0	0	0	

X: 0 or 1

Figures 7 and 8 show the bus cycle waveform examples for instruction prefetch and data access.

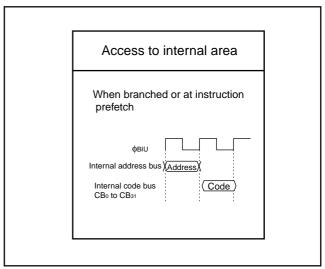


Fig. 7 Bus cycle waveform example for instruction prefetch





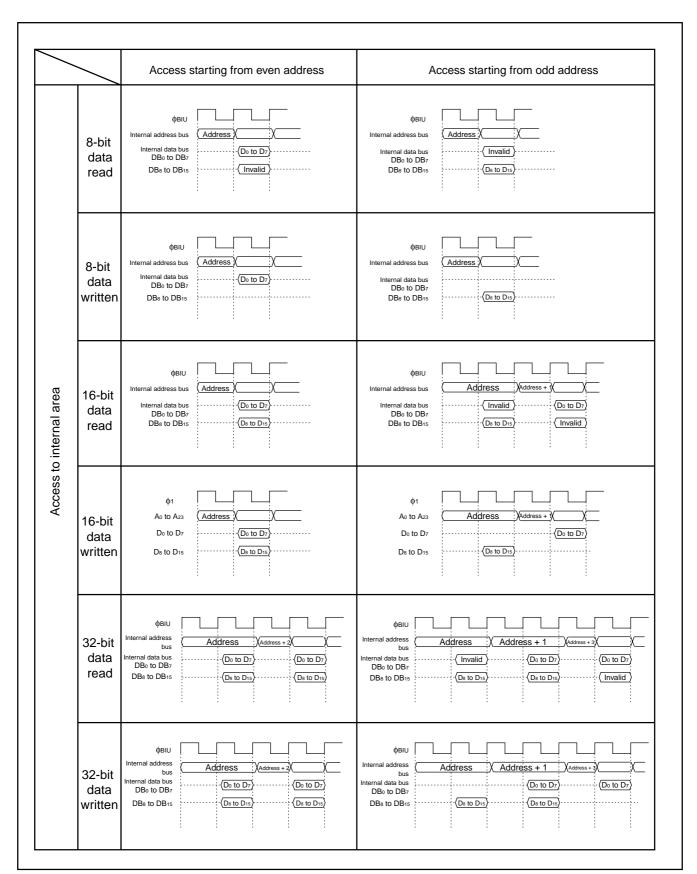


Fig. 8 Bus cycle waveform example for data access (access to internal area)



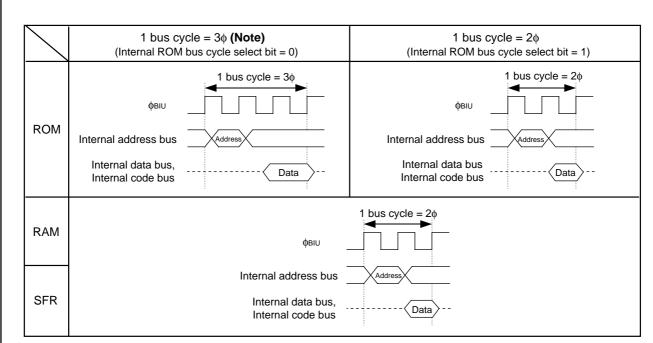


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Number of bus cycles

Figure 9 shows the bus cycle waveform at access to the internal area. Bit 7 of the processor mode register 1 (address 5F16), which is shown in Figure 10, selects the number of bus cycles for the internal

ROM: 3ϕ or 2ϕ . (This bit 7 is the internal ROM bus cycle select bit.) The internal RAM, SFRs (internal peripheral devices' control registers) are always accessed with 1 bus cycle = 2ϕ .



Note: When reprogramming the internal flash memory in the CPU reprogramming mode, select the bus cycle = 3φ.

Fig. 9 Bus cycle waveform at access to internal area

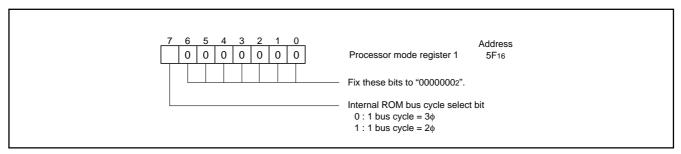


Fig. 10 Bit configuration of processor mode register 1



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PROCESSOR MODES

This microcomputer is dedicated to the single-chip mode. Therefore, be sure to connect pin MD0 to Vss, and be sure to fix the processor mode bits (bits 1 and 0 of the processor mode register 0, address 5E₁₆), which is shown in Figure 11, to "002".

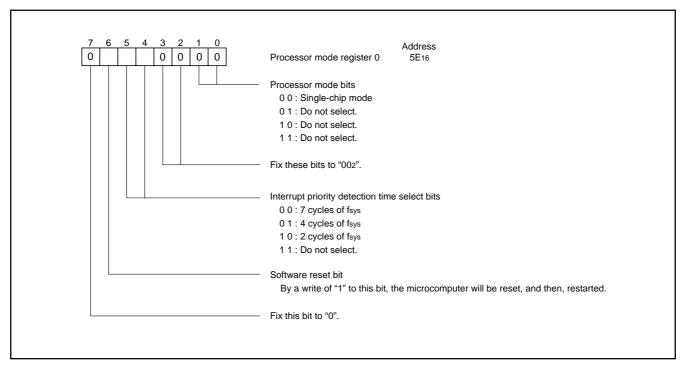


Fig. 11 Bit configuration of processor mode register 0



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INTERRUPTS

Table 3 shows the interrupt sources and the corresponding interrupt vector addresses. Reset is also handled as an interrupt source in this section, too.

DBC and BRK instruction are interrupts used only for debugging. Therefore, do not use these interrupts.

Interrupts other than reset, watchdog timer, zero divide, and address matching detection all have interrupt control registers. Table 4 shows the addresses of the interrupt control registers and Figure 13 shows the bit configuration of the interrupt control register.

The interrupt request bit is automatically cleared by the hardware during reset or when processing an interrupt. Also, interrupt request bits except for that of a watchdog timer interrupt can be cleared by software.

An $\overline{\text{INTi}}$ (i = 0 to 7) interrupt request is generated by an external input.

INTi is an external interrupt; whether to cause an interrupt at the input level (level sense) or at the edge (edge sense) can be selected with the level/edge select bit. Furthermore, the polarity of the interrupt input can be selected with the polarity select bit.

When using the following pins as external interrupt input pins, be sure to clear the direction registers of the corresponding multiplexed ports to "0": pins P51/ $\overline{INT1}$, P52/ $\overline{INT2}$, P53/ $\overline{INT3}$, P55/ $\overline{INT5}$, P56/ $\overline{INT6}$, and P57/ $\overline{INT7}$.

When the external interrupt input read register (address 9516), which is shown in Figure 12, is read out, the status of pins $\overline{\text{INT0}}$ through $\overline{\text{INT7}}$ can directly be read.

Timer and UART interrupts are described in the respective section. The priority of interrupts when multiple interrupt requests are caused simultaneously is partially fixed by hardware, but, it can also be adjusted by software as shown in Figure 14.

The hardware priority is fixed as the following: reset > watchdog timer > other interrupts

Table 3. Interrupt sources and interrupt vector addresses

UART2 transmit 00FFB416 00FFB516 UART2 receive 00FFB616 00FFB716 Timer A9 00FFB816 00FFB916 Timer A8 00FFBA16 00FFBB16 Timer A7 00FFBC16 00FFBD16 Timer A6 00FFBE16 00FFB16 Timer A5 00FFC016 00FFC316 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD316 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD916 UART0 transmit 00FFDA16 00FFDB16
Timer A9 00FFB816 00FFB916 Timer A8 00FFBA16 00FFBB16 Timer A7 00FFBC16 00FFBD16 Timer A6 00FFBE16 00FFB16 Timer A5 00FFC016 00FFC116 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
Timer A8 00FFBA16 00FFBB16 Timer A7 00FFBC16 00FFBD16 Timer A6 00FFBE16 00FFB16 Timer A5 00FFC016 00FFC116 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
Timer A7 00FFBC16 00FFBD16 Timer A6 00FFBE16 00FFBF16 Timer A5 00FFC016 00FFC116 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
Timer A6 00FFBE16 00FFBF16 Timer A5 00FFC016 00FFC116 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
Timer A5 00FFC016 00FFC116 INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
INT7 external interrupt 00FFC216 00FFC316 INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
INT6 external interrupt 00FFC416 00FFC516 INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
INT5 external interrupt 00FFC616 00FFC716 Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
Address matching detection interrupt 00FFCA16 00FFCB16 INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
INT4 external interrupt 00FFD016 00FFD116 INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
INT3 external interrupt 00FFD216 00FFD316 A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
A-D conversion 00FFD416 00FFD516 UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
UART1 transmit 00FFD616 00FFD716 UART1 receive 00FFD816 00FFD916
UART1 receive 00FFD816 00FFD916
UART0 transmit 00FFDA16 00FFDB16
UART0 receive 00FFDC16 00FFDD16
Timer B2 00FFDE16 00FFDF16
Timer B1 00FFE016 00FFE116
Timer B0 00FFE216 00FFE316
Timer A4 00FFE416 00FFE516
Timer A3 00FFE616 00FFE716
Timer A2 00FFE816 00FFE916
Timer A1 00FFEA16 00FFEB16
Timer A0 00FFEC16 00FFED16
INT2 external interrupt 00FFEE16 00FFEF16
INT1 external interrupt 00FFF016 00FFF116
INTo external interrupt 00FFF216 00FFF316
Watchdog timer 00FFF616 00FFF716
DBC (Do not select.) 00FFF816 00FFF916
Break instruction (Do not select.) 00FFFA16 00FFFB16
Zero divide 00FFFC16 00FFFD16
Reset 00FFFE16 00FFFF16

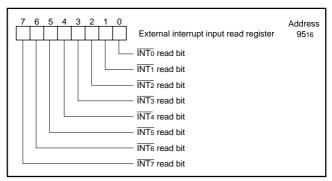


Fig. 12 Bit configuration of external interrupt input read register



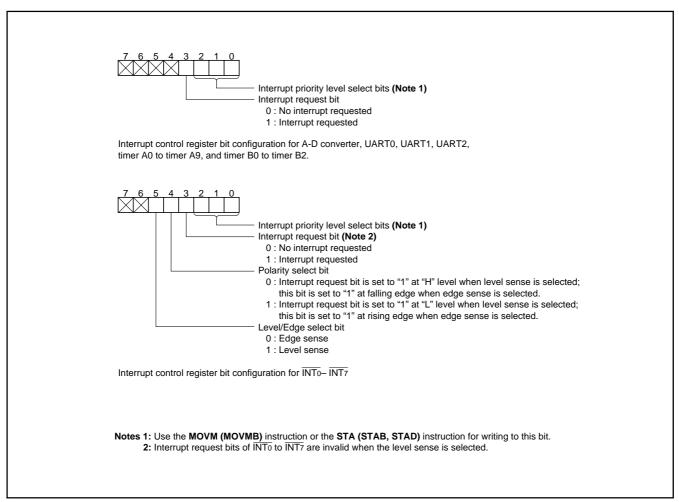


Fig. 13 Bit configuration of interrupt control register



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Table 4. Addresses of interrupt control registers

Table 4. Addresses of Interrupt control registers					
Interrupt control registers	Addresses				
INT3 interrupt control register	00006E16				
INT4 interrupt control register	00006F16				
A-D interrupt control register	00007016				
UART0 transmit interrupt control register	00007116				
UART0 receive interrupt control register	00007216				
UART1 transmit interrupt control register	00007316				
UART1 receive interrupt control register	00007416				
Timer A0 interrupt control register	00007516				
Timer A1 interrupt control register	00007616				
Timer A2 interrupt control register	00007716				
Timer A3 interrupt control register	00007816				
Timer A4 interrupt control register	00007916				
Timer B0 interrupt control register	00007A16				
Timer B1 interrupt control register	00007B16				
Timer B2 interrupt control register	00007C16				
INTo interrupt control register	00007D16				
INT ₁ interrupt control register	00007E16				
INT2 interrupt control register	00007F16				
UART2 transmit interrupt control register	0000F116				
UART2 receive interrupt control register	0000F216				
Timer A5 interrupt control register	0000F516				
Timer A6 interrupt control register	0000F616				
Timer A7 receive control register	0000F716				
Timer A8 interrupt control register	0000F816				
Timer A9 interrupt control register	0000F916				
INTs interrupt control register	0000FD16				
INT6 interrupt control register	0000FE16				
INT7 interrupt control register	0000FF16				

Interrupts caused by the address matching detection and when dividing by zero are software interrupts and are not included in Figure 14.

Other interrupts previously mentioned are A-D converter, UART, etc. interrupts. The priority of these interrupts can be changed by changing the priority level in the corresponding interrupt control register by software

Figure 15 shows a diagram of the interrupt priority detection circuit. When an interrupt is caused, each interrupt device compares its own priority with the priority from above and if its own priority is higher, then it sends the priority below and requests the interrupt. If the priorities are the same, the one above has priority.

This comparison is repeated to select the interrupt with the highest priority among the interrupts that are being requested. Finally the selected interrupt is compared with the processor interrupt priority level (IPL) contained in the processor status register (PS) and the request is accepted if it is higher than IPL and the interrupt disable flag I is "0". The request is not accepted if flag I is "1". The reset and watchdog timer interrupts are not affected by the interrupt disable flag I.

When an interrupt is accepted, the contents of the processor status register (PS) is saved to the stack and the interrupt disable flag I is set to "1".

Furthermore, the interrupt request bit of the accepted interrupt is cleared to "0" and the processor interrupt priority level (IPL) in the

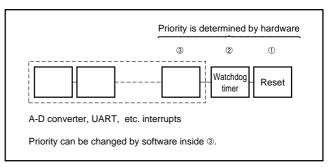


Fig. 14 Interrupt priority

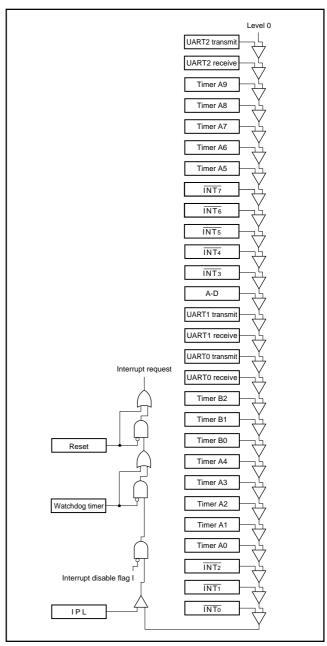


Fig. 15 Interrupt priority detection





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processor status register (PS) is replaced by the priority level of the accepted interrupt.

Therefore, multi-level priority interrupts are possible by resetting the interrupt disable flag I to "0" and enable further interrupts.

For reset, watchdog timer, zero divide, and address match detection interrupts, which do not have an interrupt control register, the processor interrupt level (IPL) is set as shown in Table 5.

The interrupt request bit and the interrupt priority level of each interrupt source are sampled and latched at each operation code fetch cycle while fsys is "H". However, no sampling pulse is generated until the cycles whose number is selected by software has passed, even if the next operation code fetch cycle is generated. The detection of an interrupt which has the highest priority is performed during that time.

As shown in Figure 16, there are three different interrupt priority detection time from which one is selected by software. After the selected time has elapsed, the highest priority is determined and is processed after the currently executing instruction has been completed.

The time is selected with bits 4 and 5 of the processor mode register 0 (address 5E₁₆) shown in Figure 11. Table 6 shows the relationship between these bits and the number of cycles. After a reset, the processor mode register 0 is initialized to "0016." Therefore, the longest time is automatically set, however, the shortest time must be selected by software.

Table 5. Value loaded in processor interrupt level (IPL) during an interrupt

Interrupt types	Setting value		
Reset	0		
Watchdog timer	7		
Zero divide	Not change value of IPL.		
Address matching detection	Not change value of IPL.		

Table 6. Relationship between interrupt priority detection time select bit and number of cycles

Priority detection time select bit		Number of avalog (Nata)		
Bit 5	Bit 4	Number of cycles (Note)		
0	0	7 cycles of fsys		
0	1	4 cycles of fsys		
1	0	2 cycles of fsys		

Note: For system clock fsys, refer to the section on the clock generating circuit.

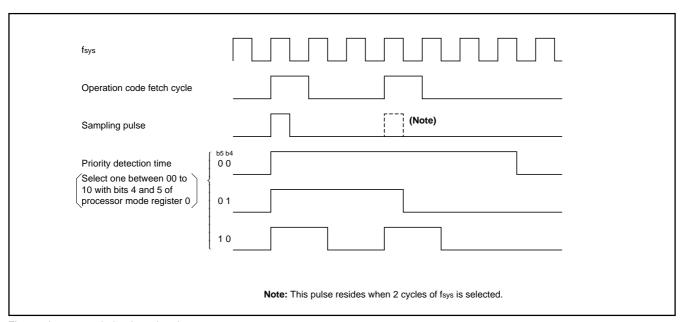


Fig. 16 Interrupt priority detection time





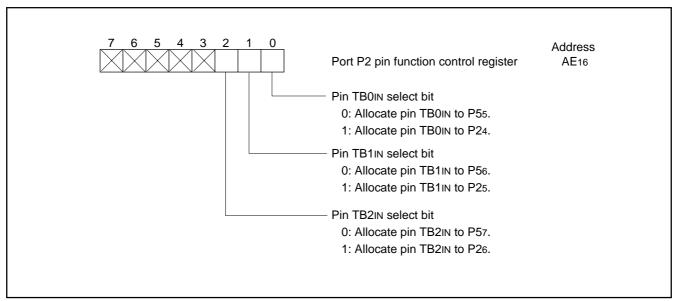


Fig. 17 Bit configuration of port P2 pin function control register



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TIMER

There are eight 16-bit timers. They are divided by type into timer A (10) and timer B (3).

The timer I/O pins are multiplexed with I/O pins for ports P2, P4, P5 and P6. To use these pins as timer input pins, the port direction register bit corresponding to the pin must be cleared to "0" to specify input mode.

TIMER A

Figure 18 shows a block diagram of timer A.

Timer A has four modes: timer mode, event counter mode, one-shot pulse mode, and pulse width modulation mode. The mode is selected with bits 0 and 1 of the timer Ai mode register (i = 0 to 9). Each of these modes is described below.

Figure 19 shows the bit configuration of the timer A clock division select register. Timers A0 to A9 use the count source which has been selected by bits 0 and 1 of this register.

(1) Timer mode [00]

Figure 20 shows the bit configuration of the timer Ai mode register in the timer mode. Bits 0, 1 and 5 of the timer Ai mode register must be "0" in timer mode. The timer A's count source is selected by bits 6 and 7 of the timer Ai mode register and the contents of the timer A clock division select register. (See Table 7.)

The counting of the selected clock starts when the count start bit is "1" and stops when it is "0".

Figure 21 shows the bit configuration of the count start bit. The counter is decremented, an interrupt is caused and the interrupt request bit in the timer Ai interrupt control register is set when the contents becomes 000016. At the same time, the contents of the reload register is transferred to the counter and count is continued.

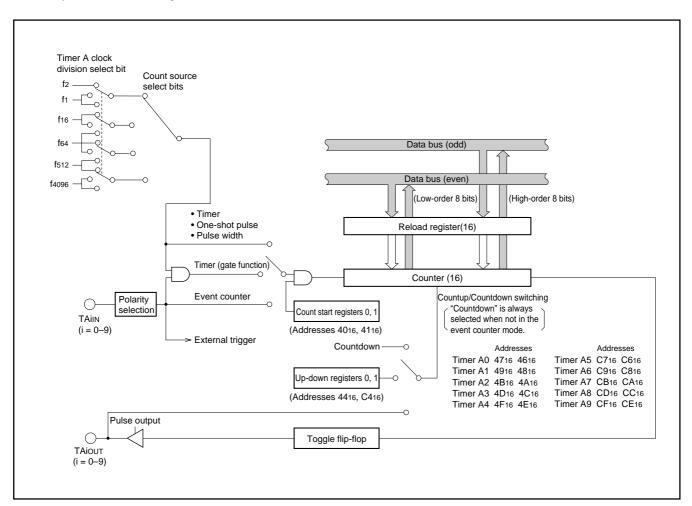


Fig. 18 Block diagram of timer A



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When bit 2 of the timer Ai mode register is "1", the output is generated from TAiout pin. The output is toggled each time the contents of the counter reaches to 000016. When the contents of the count start bit is "0", "L" is output from TAiout pin.

When bit 2 is "0", TAiouT can be used as a normal port pin. When bit 4 is "0", TAiiN can be used as a normal port pin.

When bit 4 is "1", counting is performed only while the input signal from the TAilN pin is "H" or "L" as shown in Figure 22. Therefore, this can be used to measure the pulse width of the TAilN input signal. Whether to count while the input signal is "H" or while it is "L" is determined by bit 3. If bit 3 is "1", counting is performed while the TAilN pin input signal is "H" and if bit 3 is "0", counting is performed while it is "L".

Note that, the duration of "H" or "L" on the TAilN pin must be 2 or more cycles of the timer count source.

When data is written to timer Ai register with timer Ai halted, the same data is also written to the reload register and the counter.

When data is written to timer Ai which is busy, the data is written to the reload register, but not to the counter. The new data is reloaded from the reload register to the counter at the next reload time and counting continues. The contents of the counter can be read at any time.

When the value set in the timer Ai register is n, the timer frequency division ratio is 1/(n+1).

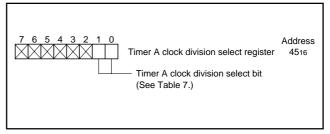


Fig. 19 Bit configuration of timer A clock division select register

Table 7. Relationship between timer A clock division select bits, clock source select bits, and count source

Clock source select bits (bits 7 and 6 at addresses 5616 to 5A16) (bits 7 and 6 at addresses	Timer A clock division select bits (bits 1 and 0 at address 4516)			
D616 to DA16)	00	01	10	11
0 0	f2	f1	f1	
0 1	f16	f16	f64	Do not
1 0	f64	f64	f512	select.
1 1	f512	f4096	f4096	

Note: Timers A0 to A9 use the same clock, which is selected by the timer A clock division select bits.

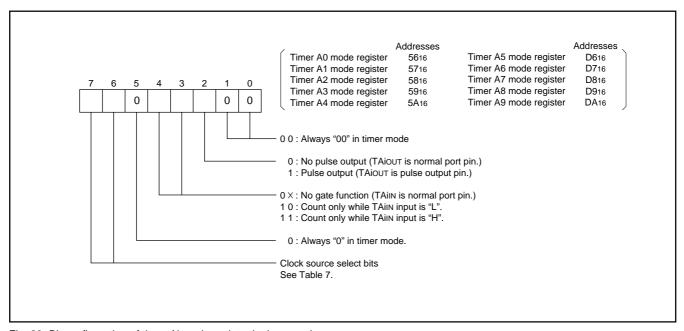


Fig. 20 Bit configuration of timer Ai mode register in timer mode



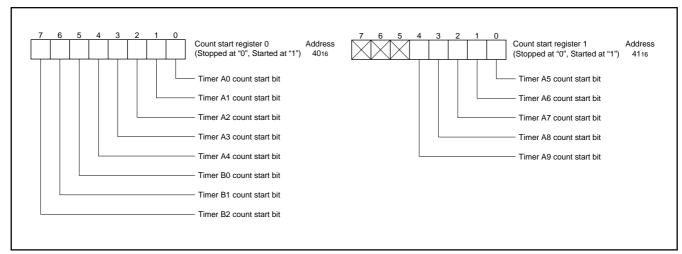


Fig. 21 Bit configuration of count start register

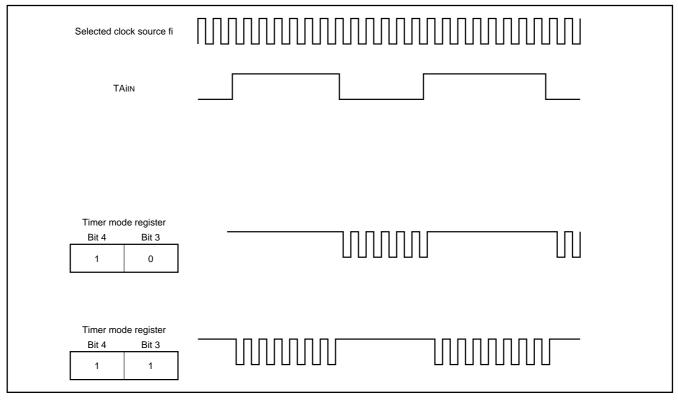


Fig. 22 Count waveform when gate function is available



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(2) Event counter mode [01]

Figure 23 shows the bit configuration of the timer Ai mode register in the event counter mode. In event counter mode, bit 0 of the timer Ai mode register must be "1" and bits 1 and 5 must be "0".

The input signal from the TAiIN pin is counted when the count start bit shown in Figure 21 is "1" and counting is stopped when it is "0". Count is performed at the fall of the input signal when bit 3 is "0" and at the rise of the signal when it is "1".

In event counter mode, whether to increment or decrement the count can be selected with the up-down bit or the input signal from the TAIOUT pin.

When bit 4 of the timer Ai mode register is "0", the up-down bit is used to determine whether to increment or decrement the count (decrement when the bit is "0" and increment when it is "1"). Figure 24 shows the bit configuration of the up-down register.

When bit 4 of the timer Ai mode register is "1", the input signal from the TAiOUT pin is used to determine whether to increment or decrement the count. However, note that bit 2 must be "0" if bit 4 is "1". It is because if bit 2 is "1", TAiOUT pin becomes an output pin to output pulses.

The count is decremented when the input signal from the TAioUT pin is "L" and incremented when it is "H". Determine the level of the input signal from the TAioUT pin before a valid edge is input to the TAiIN pin.

An interrupt request signal is generated and the interrupt request bit in the timer Ai interrupt control register is set when the counter reaches 000016 (decrement count) or FFFF16 (increment count). At the same time, the contents of the reload register is transferred to the counter and the count is continued.

When bit 2 is "1", each time the counter reaches 000016 (decrement

count) or FFFF16 (increment count), the waveform's polarity is reversed and is output from TAiOUT pin.

If bit 2 is "0", TAiOUT pin can be used as a normal port pin.

However, if bit 4 is "1" and the TAioUT pin is used as an output pin, the output from the pin changes the count direction. Therefore, bit 4 must be "0" unless the output from the TAioUT pin is to be used to select the count direction.

Data write and data read are performed in the same way as for timer mode. That is, when data is written to timer Ai halted, it is also written to the reload register and the counter. When data is written to timer Ai which is busy, the data is written to the reload register, but not to the counter. The counter is reloaded with new data from the reload register at the next reload time. The counter can be read at any time.

In event counter mode, whether to increment or decrement the counter can also be determined by supplying two kinds of pulses of which phases differ by 90° to timer A2, A3, A4, A7, A8 or A9. There are two types of two-phase pulse processing operations. One uses timers A2, A3, A7, and A8 and the other uses timers A4 and A9. In both processing operations, two pulses described above are input to the TAjout (j = 2 to 4, 7 to 9) pin and TAjiN pin respectively.

When timers A2, A3, A7, and A8 are used, as shown in Figure 25, the count is incremented when a rising edge is input to the TAkIN (k=2, 3, 7, 8) pin after the level of TAkout pin changes from "L" to "H", and when the falling edge is input, the count is decremented.

For timers A4 and A9, as shown in Figure 26, when a phase-related pulse with a rising edge input to the TAIIN (I = 4, 9) pin is input after the level of TAIOUT pin changes from "L" to "H", the count is incremented at the respective rising edge and falling edge of the TAIOUT pin and TAIIN pin.

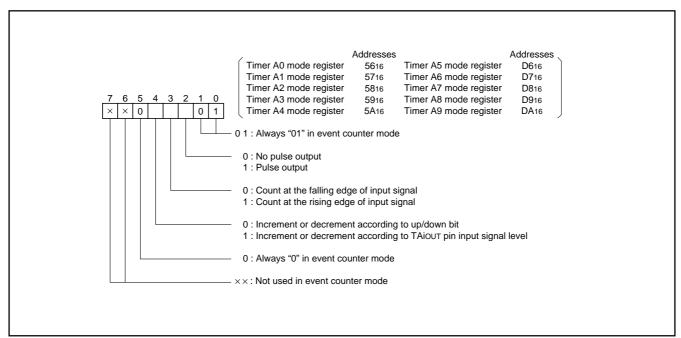


Fig. 23 Bit configuration of timer Ai mode register in event counter mode



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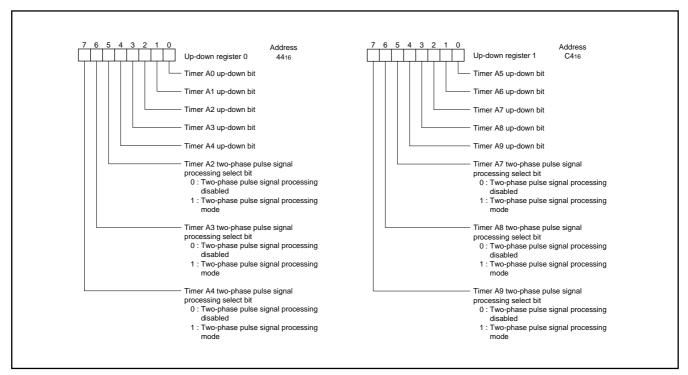


Fig. 24 Bit configuration of up-down register

When a phase-related pulse with a falling edge input to the TAkout pin is input after the level of TAIn pin changes from "H" to "L", the count is decremented at the respective rising edge and falling edge of the TAIn pin and TAIOUT pin. When performing this two-phase pulse signal processing, bits 0 and 4 of timer Aj mode register must be set to "1" and bits 1, 2, 3, and 5 must be "0". Bits 6 and 7 are ignored. (See Figure 27.) Note that bits 5, 6, and 7 of the up-down register 0 (address 4416) are the two-phase pulse signal processing select bits for timers A2, A3, and A4, respectively. Also, bits 5, 6, and 7 of the up-down register 1 (address C416) are the two-phase pulse signal processing select bits for timers A7, A8, and A9, respectively. Each timer operates in normal event counter mode when the corresponding bit is "0" and performs two-phase pulse signal processing when it is "1".

Count is started by setting the count start bit to "1". Data write and read are performed in the same way as for normal event counter mode. Note that the direction register of the input port must be set to input mode because two kinds of pulse signals, described above, are input. Also, there can be no pulse output in this mode.

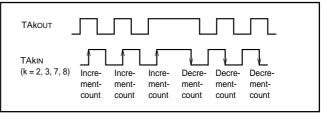


Fig. 25 Two-phase pulse processing operation of timer A2, A3, A7, A8

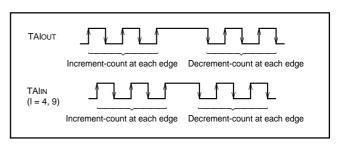


Fig. 26 Two-phase pulse processing operation of timers A4 and A9

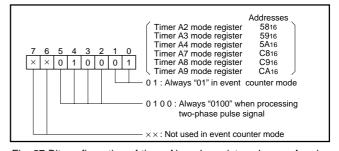


Fig. 27 Bit configuration of timer Aj mode register when performing two-phase pulse signal processing in event counter mode





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(3) One-shot pulse mode [10]

Figure 28 shows the bit configuration of the timer Ai mode register in the one-shot pulse mode. In one-shot pulse mode, bit 0 and bit 5 must be "0" and bit 1 and bit 2 must be "1".

The trigger is enabled when the count start bit is "1". The trigger can be generated by software or it can be input from the TAilN pin. Software trigger is selected when bit 4 is "0" and the input signal from the TAilN pin is used as the trigger when it is "1".

Bit 3 is used to determine whether to trigger at the fall of the trigger signal or at the rise. The trigger is at the fall of the trigger signal when bit 3 is "0" and at the rise of the trigger signal when it is "1".

Software trigger is generated by setting "1" to a bit in the one-shot start register. Each bit corresponds to each timer.

Figure 29 shows the bit configuration of the one-shot start register. As shown in Figure 30, when a trigger signal is received, the counter counts the clock selected by bits 6 and 7 and the contents of the timer A clock division select register. (Set Table 7.)

If the contents of the counter is not 000016, the TAioUT pin goes "H" when a trigger signal is received. The count direction is decrement. When the counter reaches 000116, the TAioUT pin goes "L" and count is stopped. The contents of the reload register is transferred to the counter. At the same time, an interrupt request signal is generated and the interrupt request bit in the timer Ai interrupt control register is set. This is repeated each time a trigger signal is received.

The output pulse width is

pulse frequency of the selected clock
× (counter's value at the time of trigger).

If the count start flag is "0", TAiouT goes "L". Therefore, the value corresponding to the desired pulse width must be written to timer Ai before setting the timer Ai count start bit.

As shown in Figure 31, a trigger signal can be received before the operation for the previous trigger signal is completed. In this case, the contents of the reload register is transferred to the counter by the trigger and then that value is decremented.

Except when retriggering while operating, the contents of the reload register are not transferred to the counter by triggering.

When retriggering, there must be at least one timer count source cycle before a new trigger can be issued.

Data write is performed in the same way as for timer mode.

When data is written in timer Ai halted, it is also written to the reload register and the counter.

When data is written to timer Ai which is busy, the data is written to the reload register, but not to the counter. The counter is reloaded with new data from the reload register at the next reload time.

Undefined data is read when timer Ai is read.

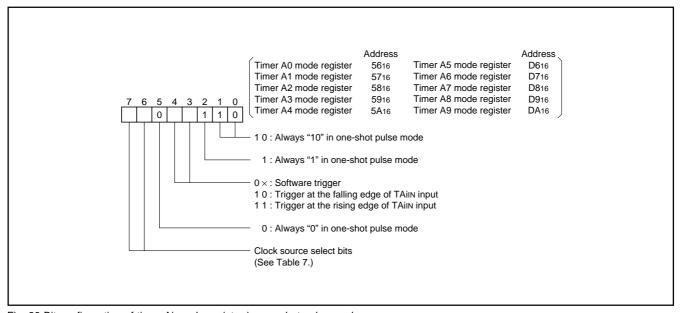


Fig. 28 Bit configuration of timer Ai mode register in one-shot pulse mode



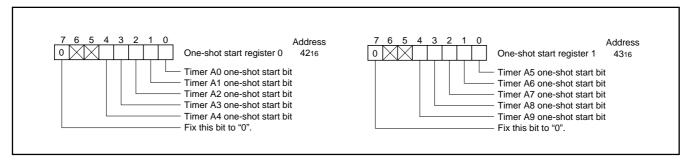


Fig. 29 Bit configuration of one-shot start register

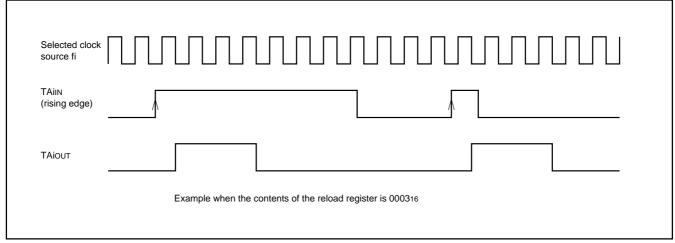


Fig. 30 Pulse output example when external rising edge is selected

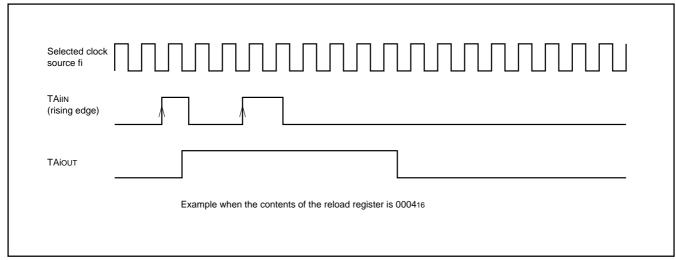


Fig. 31 Example when trigger is re-issued during pulse output





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(4) Pulse width modulation mode [11]

Figure 32 shows the bit configuration of the timer Ai mode register in the pulse width modulation mode. In pulse width modulation mode, bits 0, 1, and 2 must be set to "1".

Bit 5 is used to determine whether to perform 16-bit length pulse width modulator or 8-bit length pulse width modulator. 16-bit length pulse width modulator is selected when bit 5 is "0" and 8-bit length pulse width modulator is selected when it is "1". The 16-bit length pulse width modulator is described first.

The pulse width modulator can be started with a software trigger or with an input signal from a TAilN pin (external trigger).

The software trigger mode is selected when bit 4 is "0".

Pulse width modulator is started and a pulse is output from TAiouT when the count start bit is set to "1".

The external trigger mode is selected when bit 4 is "1".

Pulse width modulation starts when a trigger signal is input from the TAIIN pin when the count start bit is "1". Whether to trigger at the fall or rise of the trigger signal is determined by bit 3. The trigger is at the fall of the trigger signal when bit 3 is "0" and at the rise when it is "1". When data is written to timer Ai with the pulse width modulator halted, it is written to the reload register and the counter.

Then when the count start bit is set to "1" and a software trigger or an external trigger is issued to start modulation, the waveform shown in Figure 33 is output continuously.

Once modulation is started, triggers are not accepted. If the value in the reload register is m, the duration "H" of pulse is

$$\frac{1}{\text{selected clock frequency}} \times m$$

and the output pulse period is

$$\frac{1}{\text{selected clock frequency}} \times (2^{16} - 1).$$

An interrupt request signal is generated and the interrupt request bit in the timer Ai interrupt control register is set at each fall of the output

The width of the output pulse is changed by updating timer data. The update can be performed at any time. The output pulse width is changed at the rise of the pulse after data is written to the timer.

The contents of the reload register are transferred to the counter just before the rise of the next pulse so that the pulse width is changed from the next output pulse.

Undefined data is read when timer Ai is read.

The 8-bit length pulse width modulator is described next.

The 8-bit length pulse width modulator is selected when the timer Ai mode register bit 5 is "1".

The reload register and the counter are both divided into 8-bit halves.

The low-order 8 bits function as a prescaler and the high-order 8 bits function as the 8-bit length pulse width modulator. The prescaler counts the clock selected by bits 6, 7, and the contents of the timer A clock division select register. (See Table 7.) A pulse is generated when the counter reaches 000016 as shown in Figure 34. At the same time, the contents of the reload register is transferred to the counter and count is continued.

Therefore, if the low-order 8 bits of the reload register are n, the period of the generated pulse is

$$\frac{1}{\text{selected clock frequency}} \times (n+1).$$

The high-order 8 bits function as an 8-bit length pulse width modulator using this pulse as input. The operation is the same as for 16-bit length pulse width modulator except that the length is 8 bits. If the high-order 8 bits of the reload register are m, the duration "H" of pulse is

$$\frac{1}{\text{selected clock frequency}} \times (n+1) \times m.$$

And the output pulse period is

$$\frac{1}{\text{selected clock frequency}} \times (n+1) \times (2^8-1).$$

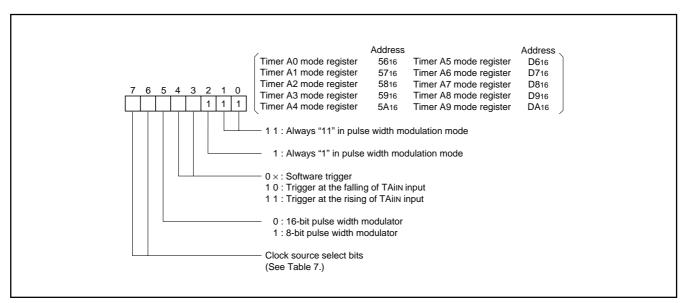


Fig. 32 Bit configuration of timer Ai mode register in pulse width modulation mode





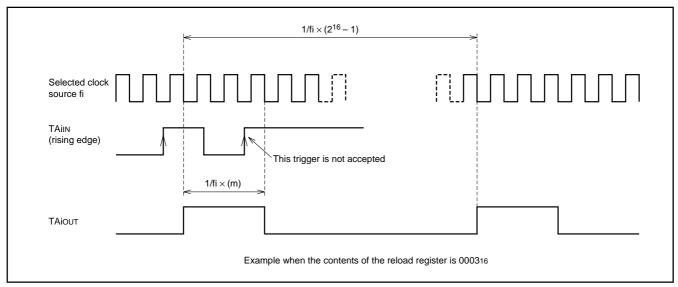


Fig. 33 16-bit length pulse width modulator output pulse example

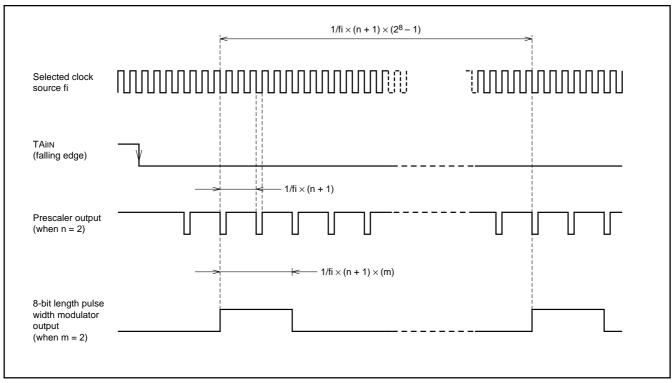


Fig. 34 8-bit length pulse width modulator output pulse example



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TIMER B

Figure 35 shows a block diagram of timer B.

Timer B has three modes: timer mode, event counter mode, and pulse period measurement/pulse width measurement mode. The mode is selected with bits 0 and 1 of the timer Bi mode register (i=0 to 2). Each of these modes is described below.

(1) Timer mode [00]

Figure 36 shows the bit configuration of the timer Bi mode register in the timer mode. Bits 0 and 1 of the timer Bi mode register must always be "0" in timer mode.

Bits 6 and 7 are used to select the clock source. The counting of the selected clock starts when the count start bit is "1" and stops when "0".

As shown in Figure 21, the timer Bi's count start bits are allocated at

the same address to which some of the timer Ai's count start bits are allocated. (In other words, they are allocated in the count start register 0.) The count is decremented, an interrupt occurs, and the interrupt request bit in the timer Bi interrupt control register is set when the contents becomes 000016. At the same time, the contents of the reload register is stored in the counter and count is continued.

Timer Bi does not have a pulse output function or a gate function like timer A.

When data is written to timer Bi halted, it is written to the reload register and the counter. When data is written to timer Bi which is busy, the data is written to the reload register, but not to the counter. The new data is reloaded from the reload register to the counter at the next reload time and counting continues.

The contents of the counter can be read at any time.

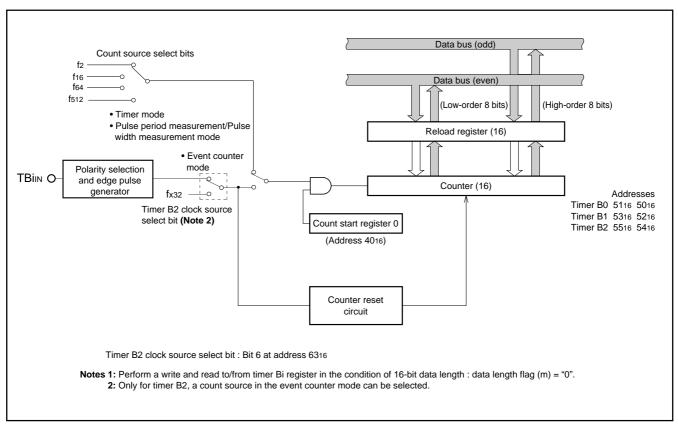


Fig. 35 Block diagram of timer B



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(2) Event counter mode [01]

Figure 37 shows the bit configuration of the timer Bi mode register in the event counter mode. In event counter mode, bit 0 in the timer Bi mode register must be "1" and bit 1 must be "0".

The input signal from the TBiIN pin is counted when the count start bit is "1" and counting is stopped when it is "0".

Count is performed at the fall of the input signal when bits 2 and 3 are "0" and at the rise of the input signal when bit 3 is "0" and bit 2 is "1" 1 "

When bit 3 is "1" and bit 2 is "0", count is performed at the rise and fall of the input signal.

Only for timer B2, when the timer B2 clock source select bit of the particular function select register 1 (bit 6 at address 6316) = "1" in the event counter mode, fx32 can be selected. (When this bit is "0", an input signal from pin TB2 $\rm IN$ becomes the count source as described above.) For the bit configuration of the particular function select register 1, refer to the section on the standby function.

Also, the pin position where pin TBiIN is to be allocated can be selected by the pin TBiIN select bit (bit 0, 1, or 2 at address AE₁₆; the port P2 pin function control register).

(3) Pulse period measurement/Pulse width measurement mode [10]

Figure 38 shows the bit configuration of the timer Bi mode register in the pulse period measurement/pulse width measurement mode.

In the pulse period measurement/pulse width measurement mode, bit 0 must be "0" and bit 1 must be "1". Bits 6 and 7 are used to select the clock source. The selected clock is counted when the count start bit is "1" and counting stops when it is "0".

The pulse period measurement mode is selected when bit 3 is "0". In the pulse period measurement mode, the selected clock is counted during the interval starting at the fall of the input signal from the TBiIN pin to the next fall or at the rise of the input signal to the next rise; the result is stored in the reload register. In this case, the reload register acts as a buffer register.

When bit 2 is "0", the clock is counted from the fall of the input signal to the next fall. When bit 2 is "1", the clock is counted from the rise of the input signal to the next rise.

In the case of counting from the fall of the input signal to the next fall, counting is performed as follows. As shown in Figure 39, when the fall of the input signal from TBiIN pin is detected, the contents of the counter is transferred to the reload register. Next, the counter is cleared and count is started from the next clock. When the fall of the next input signal is detected, the contents of the counter is transferred to the reload register once more, the counter is cleared, and the count is started. The period from the fall of the input signal to the next fall is measured in this way.

After the contents of the counter is transferred to the reload register, an interrupt request signal is generated and the interrupt request bit in the timer Bi interrupt control register is set. However, no interrupt request signal is generated when the contents of the counter is transferred first to the reload register after the count start bit is set to "1".

When bit 3 is "1", the pulse width measurement mode is selected. The pulse width measurement mode is the same as the pulse period measurement mode except that the clock is counted from the fall of the TBiIN pin input signal to the next rise or from the rise of the input signal to the next fall as shown in Figure 40.

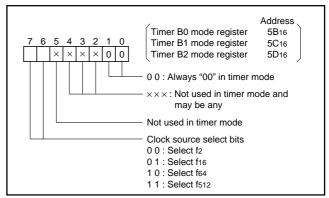


Fig. 36 Bit configuration of timer Bi mode register in timer mode

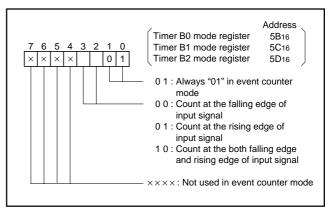


Fig. 37 Bit configuration of timer Bi mode register in event counter mode

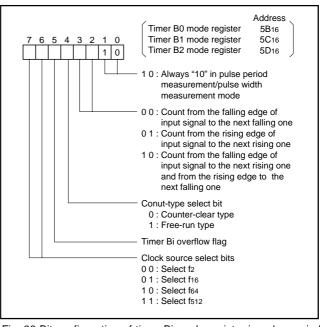


Fig. 38 Bit configuration of timer Bi mode register in pulse period measurement/pulse width measurement mode





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When timer Bi is read, the contents of the reload register is read. Note that in this mode, the interval between the fall of the TBiIN pin input signal to the next rise or from the rise to the next fall must be at least two cycles of the timer count source.

Timer Bi overflow flag which is bit 5 of timer Bi mode register is set to "1" when the timer Bi counter reaches 000016, which indicates that a pulse width or pulse period is longer than that which can be measured by a 16-bit length.

This flag is cleared by writing data to the corresponding timer Bi mode register. This flag is set to "1" at reset.

In these modes, the count type can be selected by the count-type select bit (bit 4 of the timer Bi mode register). When this bit = "1", the free-run type is selected; in this case, even when a valid edge is input to pin TBiIN, the contents of the counter will not be cleared to "000016", and counting will continue. However, when a valid edge is input, an interrupt-requesting signal will be generated.

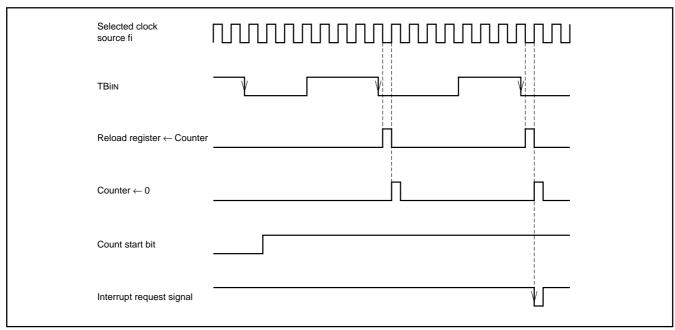


Fig. 39 Pulse period measurement mode operation (example of measuring the interval between the falling edge to next falling one)

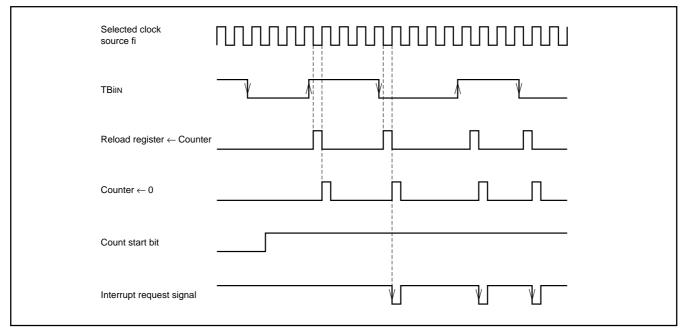


Fig. 40 Pulse width measurement mode operation



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TIMER FUNCTION FOR MOTOR CONTROL

Three-phase motor drive waveform or pulse motor drive waveform can be output by using plural internal timers As. These modes are explained bellow.

Three-phase motor drive waveform output mode (three-phase waveform mode)

Three-phase waveform mode using timers A0, A1, A2 and A3 is selected by setting the waveform output select bits of the waveform output mode register (bits 2 through 0 at address A616, Figure 41) to "1002".

There are two types of the three-phase waveform mode: three-phase mode 0 and three-phase mode 1. Bit 4 of the waveform output mode register selects either mode. In the three-phase waveform mode, set the corresponding timer mode registers of timers A0, A1, and A2 to select the one-shot pulse mode with the rising edge of an external trigger valid; set the timer mode register of timer A3 to select the timer mode.

Figure 43 shows the block diagram in the three-phase waveform mode. The three-phase waveform mode outputs six waveforms, positive waveforms (U, V, W phases) and negative waveforms (\overline{U} , \overline{V} , \overline{W} phases), from the respective ports with "L" level active.

Timer A2 controls U and \overline{U} phases; timer A1 does V and \overline{V} phases and timer A0 does W and \overline{W} phases. Timer A3 controls these one-shot pulses' periods of timers A2, A1 and A0.

In the waveform output, a short circuit prevention time can be set to prevent "L" level of positive waveform outputs (U, V, W phases) from overlapping with "L" level of their negative waveform outputs $(\overline{U},\,\overline{V},\,\overline{W})$ phases). The short circuit prevention time can be set with three 8-bit dead-time timers, sharing one reload register. The dead-time timer operates as a one-shot timer. It's start trigger is selected from the following two types: both the rising and falling edges of timers A0 to A2's one-shot pulses or their falling edges. Additionally, bit 6 of the waveform output mode register (address A616) controls this selection. When that bit is "0", both the rising and falling edges are selected; when that bit is "1", the falling edges are selected.

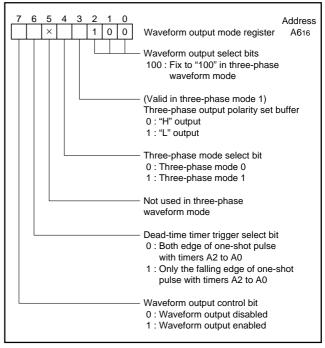


Fig. 41 Bit configuration of waveform output mode register in threephase waveform mode

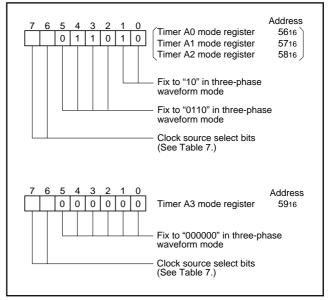


Fig. 42 Bit configuration of timer A0, A1, A2, mode register and timer A3 mode register in three-phase waveform mode



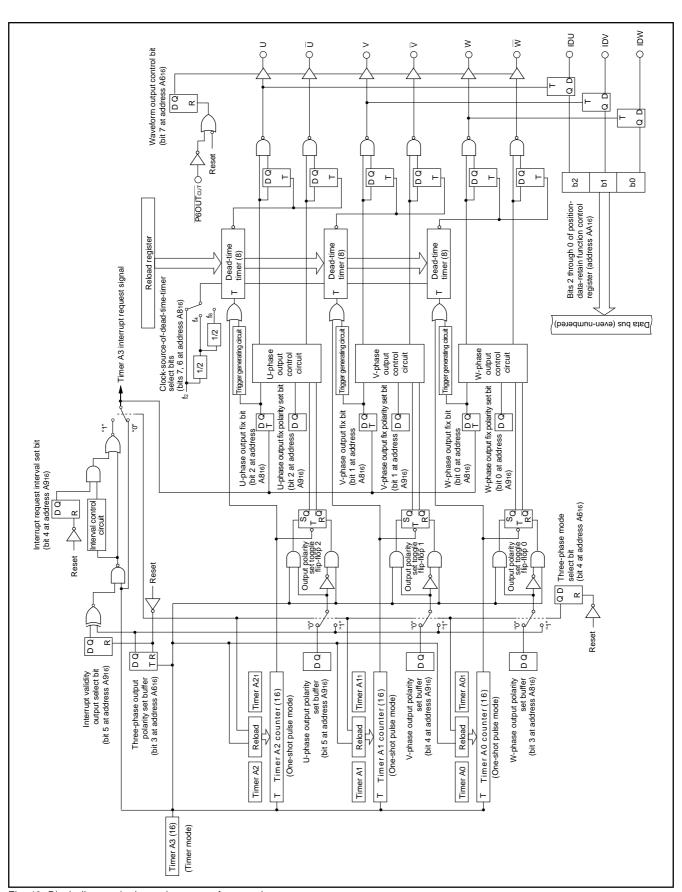


Fig. 43 Block diagram in three-phase waveform mode





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When writing data to the dead-time timer (address A716), the data is written to the reload register shared by three dead-time timers. When the dead-time timers catch the start trigger from the respective timers, the reload register contents are transferred to its counter and the dead-time timer decrements with the clock source selected by bits 6 and 7 of three-phase output data register 0 (address A816). Additionally, this timer can accept another trigger before completion of the preceding trigger operation. In this case, after transferring the reload register contents to the dead-time timer at acceptance of the trigger, the value is decremented.

The dead-time timer operates as a one-shot pulse timer. Accordingly, this timer starts pulse output when the trigger is caught, and finishes pulse output and stops operation when its contents become "0016", and waits next trigger.

In the three-phase waveform mode, setting bit 7 of the waveform output mode register (address A616) to "1" makes positive waveforms (U, V, W phases) and their negative waveforms (\overline{U} , \overline{V} , \overline{W} phases) output from the respective ports. When this bit is "0", their ports are floating. This bit is cleared to "0" by inputting a falling edge to pin $\overline{P6OUTcut}$, by reset, or by executing instructions.

Additionally, setting bits 2 through 0 of the three-phase output data register 0 (address A816) to "1" makes the corresponding waveform outputs fixed. Whether the outputs are fixed to "H" or "L" is selected by bits 2 through 0 of the three-phase output data register 1 (address A916). Clearing these bits to "0" makes the corresponding waveform outputs fixed to "H"; setting these bits to "1" makes the outputs fixed to "L".

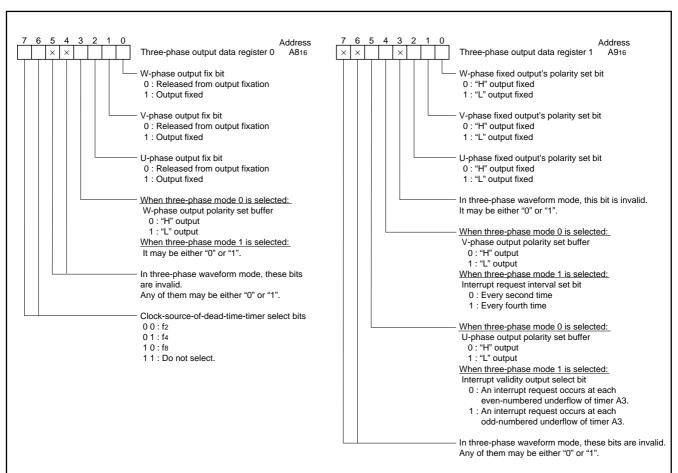


Fig. 44 Bit configuration of three-phase output data registers 1 and 0 in three-phase waveform mode



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Three-phase mode 0

In selecting three-phase waveform mode, three-phase mode 0 is selected by setting bit 4 of the waveform output mode register (address A616) to "0".

The output polarity of three-phase waveform depends on the output polarity set toggle flip-flop. The positive waveform of the three-phase waveform is "H" output when the toggle flip-flop is "0"; it is "L" output when the toggle flip-flop is "1". (Three-phase waveform is output as a negative waveform.)

Each output polarity set toggle flip-flop has the output polarity set buffer shown in Figure 44. When the timer A3's counter contents become 000016, the contents of output polarity set buffer are set into the output polarity set toggle flip-flop. After that, the polarity of the contents of output polarity set toggle flip-flop are reversed each time completion of one-shot pulse of timer (timers A2 to A0) corresponding to each phase.

Figure 45 shows an example of U-phase waveform and the output operation is explained. Three-phase mode 0 becomes valid when writing "0" to the U-phase output polarity set buffer (bit 5 at address A916) and actuating the timer A3. When the counter of timer A3 becomes 000016, the timer A3 interrupt request signal occurs and the timer A2 simultaneously starts one-shot pulse output. At this time, the contents of U-phase output polarity set buffer, "0" in this case, are set into the output polarity set toggle flip-flop 2.

When the one-shot pulse output of timer A2 is completed, the contents of output polarity set toggle flip-flop 2 is reversed from "0" to "1". Simultaneously, the one-shot pulse of the 8-bit dead-time timer is output for ensuring time not to overlap "L" levels of U phase waveform and its negative \overline{U} phase waveform.

The U-phase waveform output keeps "H" level from the start until the one-shot pulse output of the dead-time timer is completed, even if the contents of output polarity set toggle flip-flop 2 are reversed from "0" to "1" owing to the timer A2's one-shot pulse output. When the one-shot pulse output of the dead-time timer is completed, "1" of output polarity set toggle flip-flop 2 which has been reversed becomes valid and the U phase waveform changes to "L" level.

Then, write "1" to the U-phase output polarity set buffer (bit 5 at address A916) before the counter of timer A3 becomes 000016.

After that, when the counter of timer A3 becomes 000016, the timer A2 starts one-shot pulse output. Simultaneously, the contents of U-phase output polarity set buffer, "1" in this case, are set into the output polarity set toggle flip-flop 2 and the U phase waveform remains "L" level.

When the one-shot pulse output of timer A2 is completed, the contents of output polarity set toggle flip-flop 2 is reversed from "1" to "0". Simultaneously, the one-shot pulse output of the dead-time timer starts.

When the contents of output polarity set toggle flip-flop 2 are reversed from "1" to "0", the U-phase waveform changes its output level from "L" to "H" without waiting for completion of the one-shot pulse output of the dead-time timer.

U-phase waveform is generated by repeating the operation above. The way to generate \overline{U} -phase waveform, which is the negative phase of U-phase, is the same as that for U-phase waveform except that the contents of output polarity set toggle flip-flop 2 are treated as the reversed signal from the case of U-phase waveform.

In this way, U-phase waveform and \overline{U} -phase waveform, having the negative phase of U-phase, are output from the pins so that their "L" levels do not overlap each other. The width of "L" level can be also modified by changing the value of timer A2 or A3.

V-, W-phase waveform and \overline{V} -, \overline{W} -phase waveform, having their negative phase, are similarly output according to the corresponding timer operation.

The explanation above is an example of three-phase waveform generating due to an triangular wave modulation. Three-phase waveform due to a saw-tooth-wave modulation can also be generated by fixing each beginning level of phases.

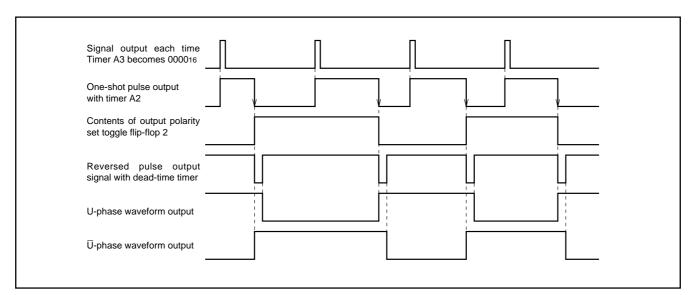


Fig. 45 U-phase waveform output example in three-phase mode 0 (triangular wave modulation)





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Three-phase mode 1

In selecting three-phase waveform mode, three-phase mode 1 is selected by setting bit 4 of the waveform output mode register (address A616) to "1".

In this mode, each of timers A0 to A2 can have two timer registers and the contents of those registers are alternately reloaded into the counter each time the counter of timer A3 becomes 000016.

The interrupt request normally occurs when the counter of timer A3 becomes 000016. However, this occurrence interval can be switched between "every second time" and "every fourth time." Bit 4 of the pulse output data register 1 (address A916) selects that.

Additionally, an even-numbered or odd-numbered timer A3's underflow can be used as the occurrence factor of timer A3 interrupt request. Bit 5 of the three-phase output data register 1 (address A916) selects that.

When the timer A3's counter contents become 000016, the contents of three-phase output polarity set buffer are set into the output polarity set toggle flip-flop on which the output polarity of three-phase waveform depends. The contents of three-phase output polarity set buffer are reversed after that operation.

The polarity of the contents of output polarity set toggle flip-flop is reversed each time completion of one-shot pulse of timer (timers A2 to A0) corresponding to each phase.

Figure 46 shows an example of U-phase waveform and the output operation is explained.

Write "0" to the three-phase-output-polarity set buffer (bit 3 at address A616). Clear the interrupt request interval set bit (bit 4 at address A916) to "0" so that the timer A3 interrupt request occurs at every second time. Additionally, clear the interrupt validity output select bit (bit 5 at address A916) so that the timer A3 interrupt request occurs at an each even-numbered underflow of timer A3.

After the procedure above, three-phase mode 1 starts operation when actuating timer A3.

When the counter of timer A3 becomes 000016, the timer A3 interrupt request occurs and timer A2 simultaneously starts one-shot pulse output. At this time, the contents of three-phase output polarity set buffer, "0" in this case, are set into the output polarity set toggle flipflop 2. The contents of three-phase output polarity set buffer are reversed from "0" to "1" after that operation.

When the timer A2 counter counts the value written into the timer A2 and the one-shot pulse output of timer A2 is completed, the contents of output polarity set toggle flip-flop 2 are reversed from "0" to "1". Simultaneously, the one-shot pulse of the 8-bit dead-time timer is output for ensuring time, so that "L" levels of U- and $\overline{\text{U}}$ -phase waveforms do not overlap.

The U-phase waveform output keeps "H" level from the start until the one-shot pulse output of the dead-time timer is completed, even if the contents of output polarity set toggle flip-flop 2 are reversed from "0" to "1" owing to the timer A2's one-shot pulse output.

When the one-shot pulse output of the dead-time timer is completed, "1" of output polarity set toggle flip-flop 2 which has been reversed becomes valid and the U-phase waveform changes to "L" level.

Then, when the counter of timer A3 becomes 000016, the timer A2 counter counts the value written into timer A2 and timer A2 starts one-shot pulse output. Simultaneously, the contents of three-phase output polarity set buffer are set into the output polarity set toggle flip-flop 2. However, the U-phase waveform remains "L" level, be-

cause the value is the same ("1").

The contents of three-phase output polarity set buffer are reversed from "1" to "0" after that operation.

When the one-shot pulse output of timer A2 is completed, the contents of output polarity set toggle flip-flop 2 is reversed from "1" to "0". Simultaneously, the one-shot pulse output of the dead-time timer starts.

When the contents of output polarity set toggle flip-flop 2 is reversed from "1" to "0", the U-phase waveform changes its output level from "L" to "H" without waiting for completion of the one-shot pulse output of the dead-time timer.

U-phase waveform is generated by repeating the operation above. The way to generate \overline{U} -phase waveform, which is the negative phase of U-phase, is the same as that for U-phase waveform except that the contents of output polarity set toggle flip-flop 2 is treated as the reversed signal from the case of U-phase waveform.

In this way, U-phase waveform and \overline{U} -phase waveform, having the negative phase of U-phase, are output from the pins so that their "L" levels do not overlap each other. The width of "L" level can be also modified by changing the value of timers A2, A21, or A3.

V-, W-phase waveform and \overline{V} -, \overline{W} -phase waveform, having their negative phase, are similarly output according to the corresponding timer operation.





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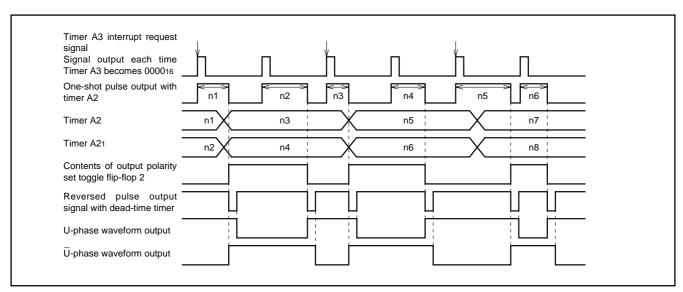


Fig. 46 U-phase waveform output example in three-phase mode 1 (triangular wave modulation)

Position-data-retain function

The three-phase waveform mode has the function to retain the input data of the corresponding pin (IDU, IDV, IDW) at an edge of a positive waveform (U, V, or W phase). Whether to retain the data at a falling edge or rising one is selected by bit 3 of the position-data-retain function control register (address AA16).

Retain data can be read out by bits 2 through 0 of the position-dataretain function control register (address AA16).

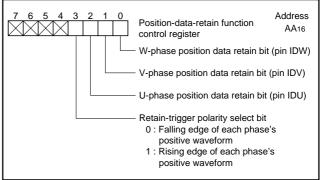


Fig. 47 Bit configuration of position-data-retain function control register



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PULSE OUTPUT PORT MODE 0

Figure 48 shows the block diagram in the pulse output port mode 0. This mode has an 8-bit pulse output port. The waveform output select bits (bits 0 to 2) of waveform output mode register (address A616) select use of pulse output port. The 8-bit pulse output port can also be divided into "4 bits and 4 bits" or "6 bits and 2 bits", with the pulse output mode select bit (bit 3) of waveform output mode register (address A616); each of them can be individually controlled. Set timers A3 and A0 to the timer mode because they are used in the pulse output port mode 0. Figure 50 shows the bit configuration of timer A3 and A0 mode registers in the pulse output port mode 0. Timers A3 and A0 start count when setting the corresponding timer count start bit to "1", and they stop it when clearing that bit to "0".

Each bit using timer A0 as a trigger can also be controlled by an input trigger from pin RTPTRG0. This control is selected by the pulse output trigger select bits of the three-phase output data register 0 (bits 7 and 6 at address A816). Also, this externally-input trigger can be selected from the following three types: falling edges, rising edges, and falling and rising edges.

The reversed content of the pulse output data bit can be output to each pulse output port by the pulse output polarity select bit of the three-phase output data register 1 (bit 3 at address A916). When the pulse output polarity select bit = "0", the content of the pulse output data bit is output as it is; when the pulse output polarity select bit = "1", the reversed content is output.

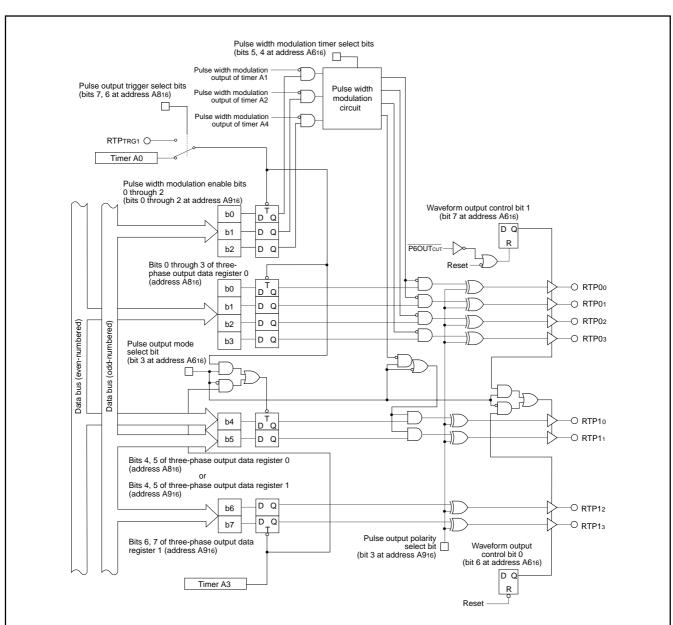


Fig. 48 Block diagram in pulse output port mode 0





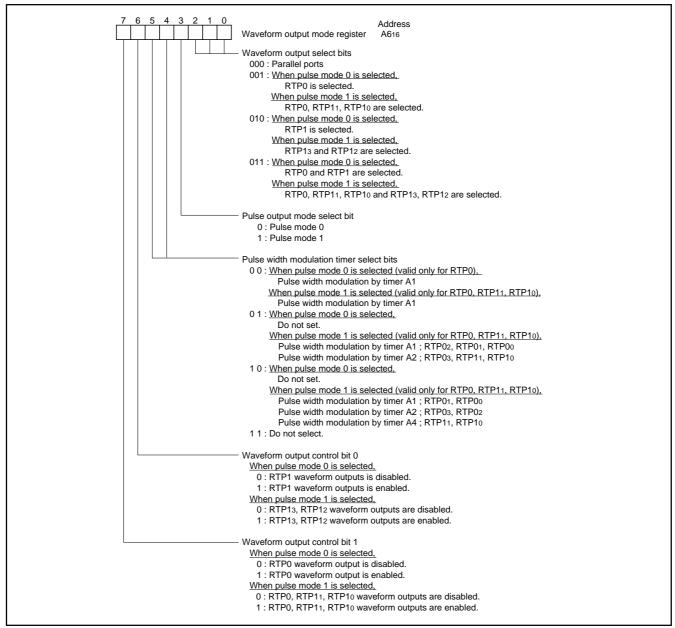


Fig. 49 Bit configuration of waveform output mode register in pulse output port mode 0

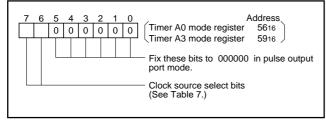


Fig. 50 Bit configuration of timer A3 and A0 mode registers in pulse output port mode 0



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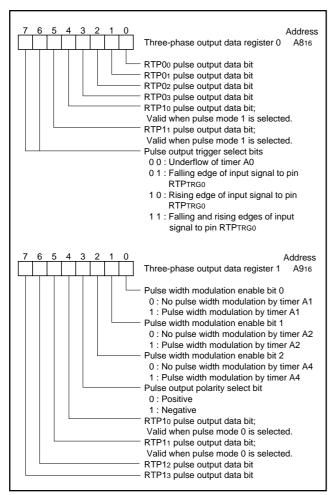


Fig. 51 Bit configuration of three-phase output data registers 1 and 0 in pulse output port mode 0

Pulse mode 0

This mode divides a pulse output port into 4 bits and 4 bits and individually controls them.

When setting the pulse output mode select bit to "0", and setting bits 2 and 1 to "0" and bit 0 to "1" of the waveform output select bits, four of RTP03, RTP02, RTP01, and RTP00 become the pulse output ports (RTP0 selected).

When setting the pulse output mode select bit to "0", and setting bits 2 and 0 to "0" and bit 1 to "1" of the waveform output select bits, four of RTP13, RTP12, RTP11, RTP10 become the pulse output ports (RTP1 selected).

When setting the pulse output mode select bit to "0", and setting bit 2 to "0" and bits 1 and 0 to "1" of the waveform output select bits, the following two groups become the pulse output ports:

- •Four of RTP13, RTP12, RTP11, RTP10
- •Four of RTP03, RTP02, RTP01, RTP00.

Each time the contents of timer A3 counter become 000016, the contents of three-phase output data register 1 (address A916)'s high order 4 bits, (bits 7 to 4), which corresponding to RTP13, RTP12, RTP11, RTP10, are output from ports.

Each time the contents of timer A0 counter become 000016, the contents of three-phase output data register 0 (address A816)'s low order 4 bits (bits 3 to 0), which corresponding to RTP03, RTP02, RTP01, RTP00, are output from ports.

When writing "0" to the specified one of the pulse output data bits, "L" level is output from the pulse output port when the contents of the corresponding timer counter become 000016; when writing "1" to it, "H" level is output from the pulse output port.

In the case that an input trigger of pin RTPTRG0 is selected, the data written to the pulse output data bit is output from the corresponding pulse output port by this selected trigger.

Additionally, pulse width modulation can be applied for RTP03, RTP02, RTP01, and RTP00. Because timer A1 is used for pulse width modulation, actuate timer A1 in the pulse width modulation mode. When any of pulse output data bits is "1", the pulse to which pulse width modulation has been applied is output from the pulse output port when the contents of timer A0 counter become 000016.

The pulse width modulation using timer A1 can be applied by setting the pulse width modulation enable bit 0 of the three-phase output data register 1 (bit 0 at address A916) to "1" and the pulse width modulation timer select bits of the waveform output mode register (bits 5 and 4 at address A616) to "00". Figure 52 shows example waveforms in pulse mode 0.

In ports selecting pulse mode 0, output of RTP13, RTP12, of RTP11 and RTP10 is controlled by the waveform output control bit 0 (bit 6) of waveform output mode register; output of RTP03, RTP02, RTP01 and RTP00 is done by the waveform output control bit 1 (bit 7).

When setting the waveform output control bit to "1", waveform is output from the corresponding port. When clearing that bit to "0", waveform output from the corresponding port stops, and the port becomes floating. The waveform output control bits are cleared to "0" by reset or by executing instructions.

Also, the waveform output control bit 1 can be cleared to "0" by inputting a falling edge to pin P6OUTcuT.





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Pulse mode 1

This mode divides a pules output port into 6 bits and 2bits and individually control them.

When setting the pulse output mode select bit to "1", and setting bits 2 and 1 to "0" and bit 0 to "1" of the waveform output select bits, the following become the pulse output ports:

•Six of RTP11, RTP10, RTP03, RTP02, RTP01, RTP00

When setting the pulse output mode select bit to "1", and setting bits 2 and 0 to "0" and bit 1 to "1" of the waveform output select bits, two of RTP13, RTP12 become the pulse output ports.

When setting the pulse output mode select bit to "1", and setting bit 2 to "0" and bits 1 and 0 to "1" of the waveform output select bits, the following two groups become the pulse output ports:

- •Two of RTP13, RTP12
- •Six of RTP11, RTP10, RTP03, RTP02, RTP01, RTP00.

Each time the contents of timer A3 counter become 000016, the contents of three-phase output data register 1 (address A916)'s bits 7 and 6, which corresponding to RTP13 and RTP12, are output from ports.

Each time the contents of timer A0 counter become 000016, the contents of three-phase output data register 0 (address A816)s' low order 6 bits (bits 5 to 0), which corresponding to RTP11, RTP10, RTP03, RTP02, RTP01, RTP00, are output from ports.

Whether to control these pulse output ports by an underflow of timer A0 or an input edge to pin RTPTRG0 is selected by the pulse output trigger select bits of the three-phase output data register 0 (bits 7 and 6 at address A816).

Additionally, pulse width modulation can be applied to RTP11, RTP10, RTP03, RTP02, RTP01, and RTP00. The pulse width modulation timer select bits of the waveform output mode register (bits 5 and 4 at address A616) select the type of pulse width modulation from the following:

- (1) When the pulse width modulation timer select bits = "00", the common modulation to six of RTP11, RTP10, RTP03, RTP02, RTP01, RTP00 is selected. For this modulation, since timer A1 is necessary, be sure to actuate this timer in the pulse width modulation mode.
- (2) When the pulse width modulation timer select bits = "01", the modulation to the following two groups is selected; one consists of RTP11, RTP10, RTP03, and the other consists of RTP02, RTP01, RTP00. For this modulation, since timers A1 and A2 are necessary, be sure to actuate these timers in the pulse width modulation mode.
- (3) When the pulse width modulation timer select bits = "10", the modulation to the following three groups is selected; one consists of RTP11, RTP10, and another consists of RTP03, RTP02, and the other consists of RTP01, RTP00. For this modulation, since timers A1, A2, and A4 are necessary, be sure to actuate these timers in the pulse width modulation mode.

Additionally, at this time, be sure to set the corresponding pulse width modulation enable bit of the three-phase output data register 1 (bits 2 through 0 at address A916) to "1" in order to enable the pulse width modulation.

The other operations are the same as that of pulse mode 0. Figure 53 shows example waveforms in pulse mode 1.

In ports selecting pulse mode 1, output of RTP13 and RTP12 is con-

trolled by the waveform output control bit 0 (bit 6) of the waveform output mode register; output of RTP11, RTP10, RTP03, RTP02, RTP01 and RTP00 is done by the waveform output control bit 1 (bit 7)

When setting the waveform output control bit to "1", waveform is output from the corresponding port. When clearing that bit to "0", waveform output from the corresponding port stops, and the port becomes floating. These waveform output control bits 0, 1 are cleared to "0" by reset, by inputting a falling edge to pin $\overline{\text{P6OUTcut}}$, or by executing instructions.





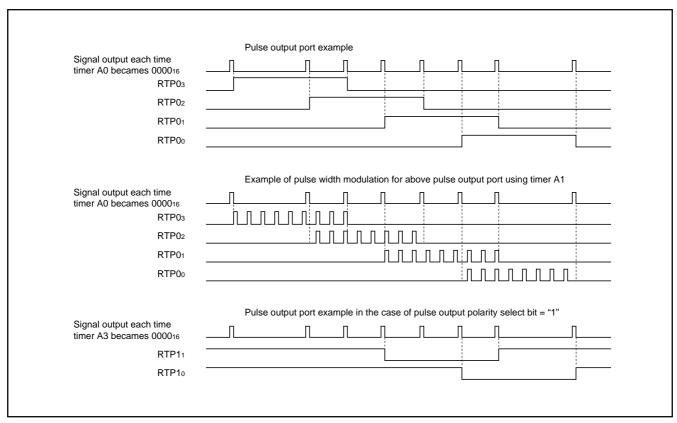


Fig. 52 Example waveforms in pulse mode 0

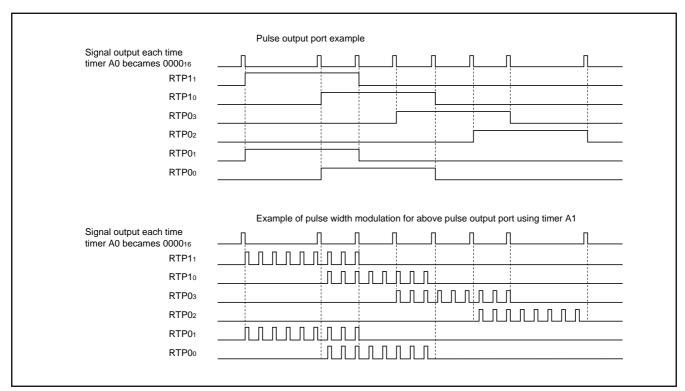


Fig. 53 Example waveforms in pulse mode 1





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PULSE OUTPUT PORT MODE 1

Figure 54 shows the block diagram in the pulse output port mode 1. This mode has an 8-bit pulse output port. The waveform output select bits (bits 0 to 2) of the pulse output control register (address A016) select use of pulse output the port. The 8-bit pulse output port can also be divided into "4 bits and 4 bits" or "6 bits and 2 bits", with the pulse output mode select bit (bit 3) of the pulse output control register (address A016); each of them can be individually controlled. Set timers A8 and A5 to the timer mode because they are used in the pulse output port mode 1. Figure 56 shows the bit configuration of timer A8 and A5 mode registers in the pulse output port mode 1. Timers A8 and A5 start count when setting the corresponding timer count start bit to "1", and they stop it when clearing that bit to "0".

Each bit using timer A5 as a trigger can also be controlled by an input trigger from pin RTPTRG1. This control is selected by the pulse output trigger select bits of the pulse output data register 0 (bits 7 and 6 at address A216). Also, this externally-input trigger can be selected from the following three types: falling edges, rising edges, and falling and rising edges.

The reversed content of the pulse output data bit can be output to each pulse output port by the pulse output polarity select bit of the pulse output data register 1 (bit 3 at address A416). When the pulse output polarity select bit = "0", the content of the pulse output data bit is output as it is; when the pulse output polarity select bit = "1", the reversed content is output.

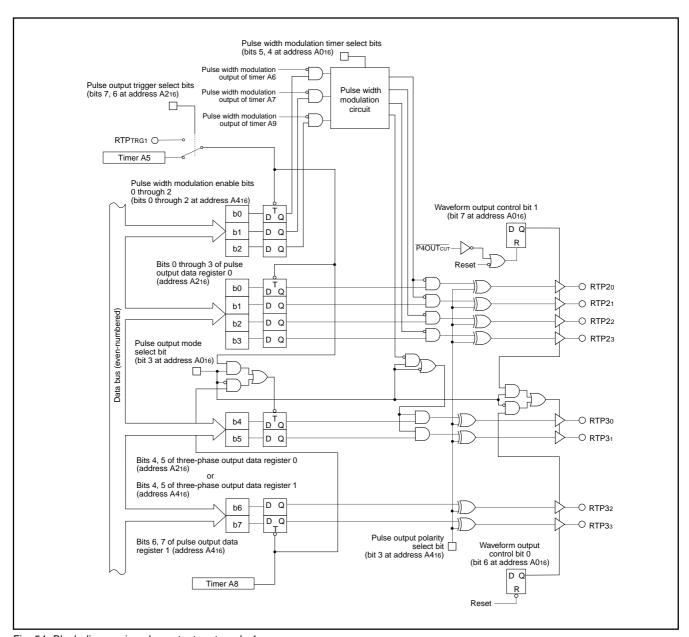


Fig. 54 Block diagram in pulse output port mode 1



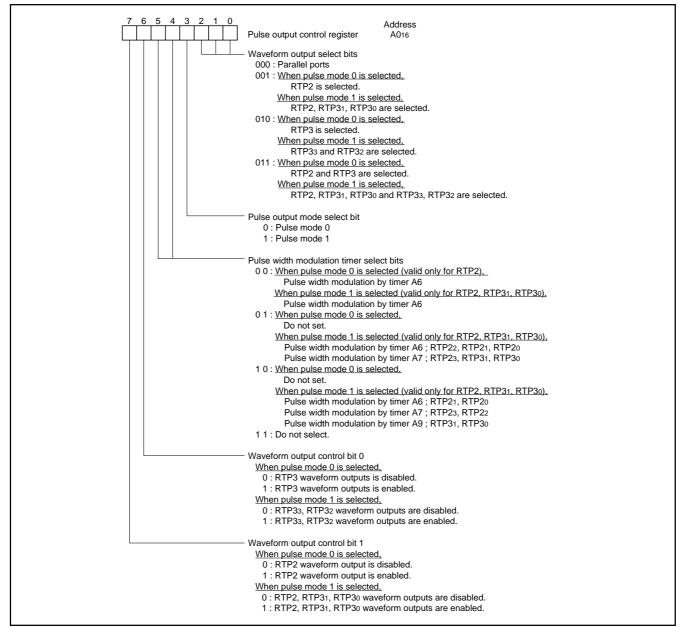


Fig. 55 Bit configuration of pulse output control register in pulse output port mode 1

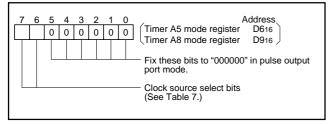


Fig. 56 Bit configuration of timer A8 and A5 mode registers in pulse output port mode 1





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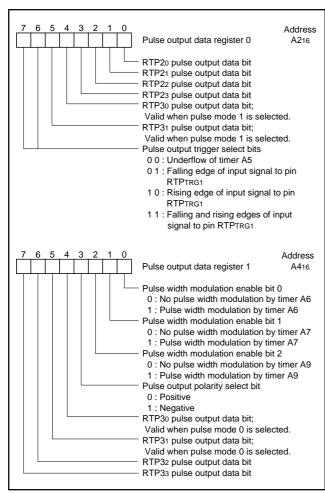


Fig. 57 Bit configuration of pulse output data registers 1 and 0 in pulse output port mode 1

Pulse mode 0

This mode divides a pulse output port into 4 bits and 4 bits and individually controls them.

When setting the pulse output mode select bit to "0", and setting bits 2 and 1 to "0" and bit 0 to "1" of the waveform output select bits, four of RTP23, RTP22, RTP21, and RTP20 become the pulse output ports (RTP2 selected).

When setting the pulse output mode select bit to "0", and setting bits 2 and 0 to "0" and bit 1 to "1" of the waveform output select bits, four of RTP33, RTP32, RTP31, RTP30 become the pulse output ports (RTP3 selected).

When setting the pulse output mode select bit to "0", and setting bit 2 to "0" and bits 1 and 0 to "1" of the waveform output select bits, the following two groups become the pulse output ports:

- •Four of RTP33, RTP32, RTP31, RTP30
- •Four of RTP23, RTP22, RTP21, RTP20.

Each time the contents of timer A8 counter become 000016, the contents of the pulse output data register 1 (address A416)'s high-order 4 bits (bits 7 to 4), which corresponding to RTP33, RTP32, RTP31, RTP30, are output from ports.

Each time the contents of timer A5 counter become 000016, the contents of pulse output data register 0 (address A216)'s low-order 4 bits (bits 3 to 0), which corresponding to RTP23, RTP22, RTP21, RTP20, are output from ports.

When writing "0" to the specified one of the pulse output data bits, "L" level is output from the pulse output port when the contents of the corresponding timer counter become 000016; when writing "1" to it, "H" level is output from the pulse output port.

In the case that an input trigger of pin RTPTRG1 is selected, the data written to the pulse output data bit is output from the corresponding pulse output port by this selected trigger.

Additionally, pulse width modulation can be applied for RTP23, RTP22, RTP21, and RTP20. Because timer A6 is used for pulse width modulation, actuate timer A6 in the pulse width modulation mode. When any of pulse output data bits is "1", the pulse to which pulse width modulation has been applied is output from the pulse output port when the contents of timer A5 counter become 000016.

The pulse width modulation using timer A6 can be applied by setting the pulse width modulation enable bit 0 of the pulse output data register 1 (bit 0 at address A416) to "1" and the pulse width modulation timer select bits of the pulse output control register (bits 5 and 4 at address A016) to "00". Figure 58 shows example waveforms in pulse mode 0.

In ports selecting pulse mode 0, output of RTP33, RTP32, RTP31 and RTP30 is controlled by the waveform output control bit 0 (bit 6) of pulse output control register; output of RTP23, RTP22, RTP21 and RTP20 is done by the waveform output control bit 1 (bit 7).

When setting the waveform output control bit to "1", waveform is output from the corresponding port. When clearing that bit to "0", waveform output from the corresponding port stops, and the port becomes floating. The waveform output control bits are cleared to "0" by reset or by executing instructions.

Also, the waveform output control bit 1 can be cleared to "0" by inputting a falling edge to pin P4OUTcuT.





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Pulse mode 1

This mode divides a pules output port into 6 bits and 2 bits and individually control them.

When setting the pulse output mode select bit to "1", and setting bits 2 and 1 to "0" and bit 0 to "1" of the waveform output select bits, the following become the pulse output ports:

•Six of RTP31, RTP30, RTP23, RTP22, RTP21, RTP20

When setting the pulse output mode select bit to "1", and setting bits 2 and 0 to "0" and bits 1 to "1" of the waveform output select bits, two of RTP33, RTP32 become the pulse output ports.

When setting the pulse output mode select bit to "1", and setting bit 2 to "0" and bits 1 and 0 to "1" of the waveform output select bits, the following two groups become the pulse output ports:

- •Two of RTP33, RTP32
- •Six of RTP31, RTP30, RTP23, RTP22, RTP21, RTP20

Each time the contents of timer A8 counter become 000016, the contents of pulse output data register 1 (address A416)'s bits 7 and 6, which corresponding to RTP33 and RTP32, are output from ports. Each time the contents of timer A5 counter become 000016, the contents of pulse output data register 0 (address A216)'s low-order 6 bits (bits 5 to 0), which corresponding to RTP31, RTP30, RTP23, RTP22, RTP21, RTP20, are output from ports.

Whether to control these pulse output ports by an underflow of timer A5 or an input edge to pin RTPTRG1 is selected by the pulse output trigger select bit of the pulse output data register 0 (bits 7 and 6 at address A216).

Additionally, pulse width modulation can be applied to RTP31, RTP30, RTP23, RTP22, RTP21, and RTP20. The pulse width modulation timer select bits of the pulse output control register (bits 5 and 4 at address A016) select the type of pulse width modulation from the following:

- (1) When the pulse width modulation timer select bits = "00", the common modulation to six of RTP31, RTP30, RTP23, RTP22, RTP21, RTP20 is selected. For this modulation, since timer A6 is necessary, be sure to actuate this timer in the pulse width modulation mode.
- (2) When the pulse width modulation timer select bits = "01", the modulation to the following two groups is selected; one consists of RTP31, RTP30, RTP23, and the other consists of RTP22, RTP21, RTP20. For this modulation, since timers A6 and A7 are necessary, be sure to actuate these timers in the pulse width modulation mode.
- (3) When the pulse width modulation timer select bits = "10", the modulation to the following three groups is selected; one consists of RTP31, RTP30, and another consists of RTP23, RTP22, and the other consists of RTP21, RTP20. For this modulation, since timers A6, A7, and A9 are necessary, be sure to actuate these timers in the pulse width modulation mode.

Additionally, at this time, be sure to set the corresponding pulse width modulation enable bit of the pulse output data register 1 (bits 2 through 0 at address A416) to "1" in order to enable the pulse width modulation.

The other operations are the same as that of pulse mode 0. Figure 59 shows example waveforms in pulse mode 1.

In ports selecting pulse mode 1, output of RTP33 and RTP32 is controlled by the waveform output control bit 0 (bit 6) of the pulse output

control register; output of RTP31, RTP30, RTP23, RTP22, RTP21, and RTP20 is done by the waveform output control bit 1 (bit 7).

When setting the waveform output control bit to "1", waveform is output from the corresponding port. When clearing that bit to "0", waveform output from the corresponding port stops, and the port becomes floating. These waveform output control bits 0, 1 are cleared to "0" by reset, by inputting a falling edge to pin P4OUTcut, or by executing instructions.





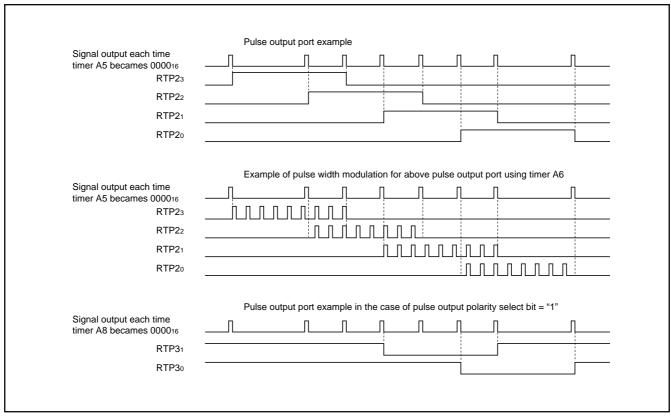


Fig. 58 Example waveforms in pulse mode 0

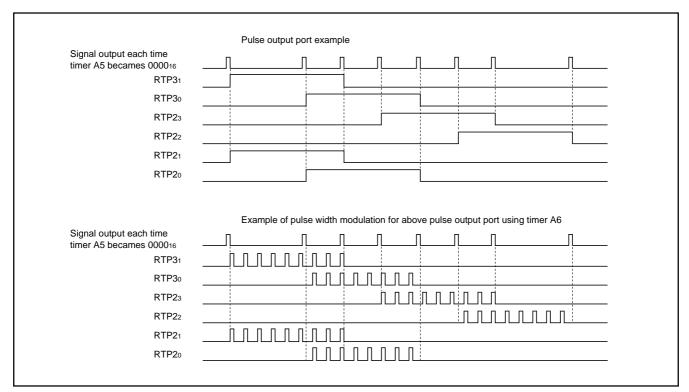


Fig. 59 Example waveforms in pulse mode 1





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SERIAL I/O PORTS

Two independent serial I/O ports are provided. Figure 60 shows a block diagram of the serial I/O ports.

Bits 0 through 2 of the UARTi(i = 0, 1, 2) transmit/receive mode register shown in Figure 61 are used to determine whether to use ports P1 and P8 as programmable I/O ports, clock synchronous serial I/O ports, or asynchronous (UART) serial I/O ports which use start and

stop bits.

Figures 62 and 63 show the block diagrams of the receiver/transmitter.

Figure 64 shows the bit configuration of the UARTi transmit/receive control register

Each communication method is described below.

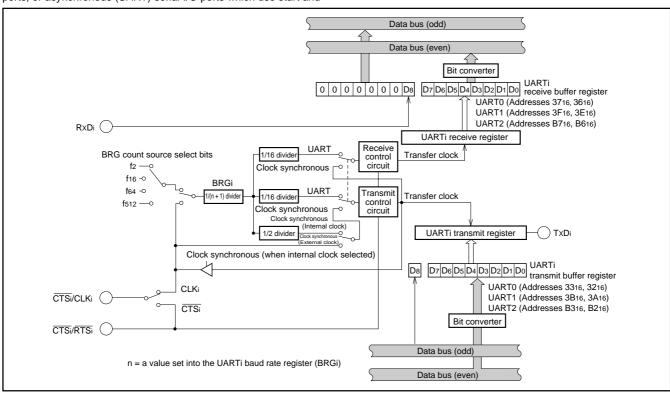


Fig. 60 Block diagram of serial I/O port

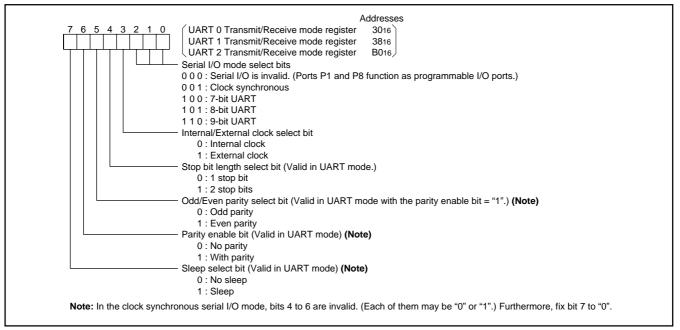


Fig. 61 Bit configuration of UARTi transmit/receive mode register



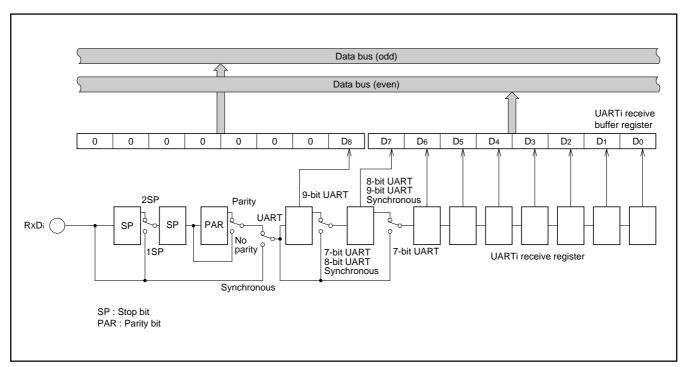


Fig. 62 Block diagram of receiver

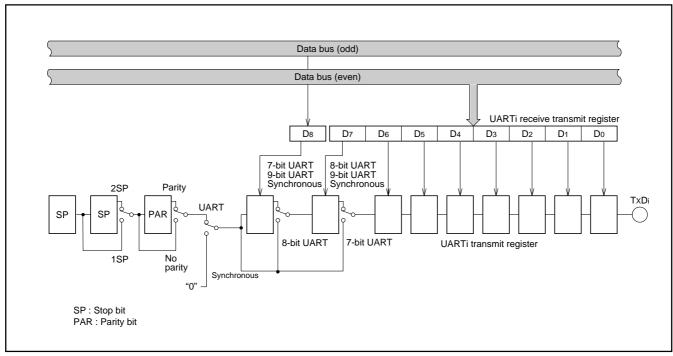


Fig. 63 Block diagram of transmitter



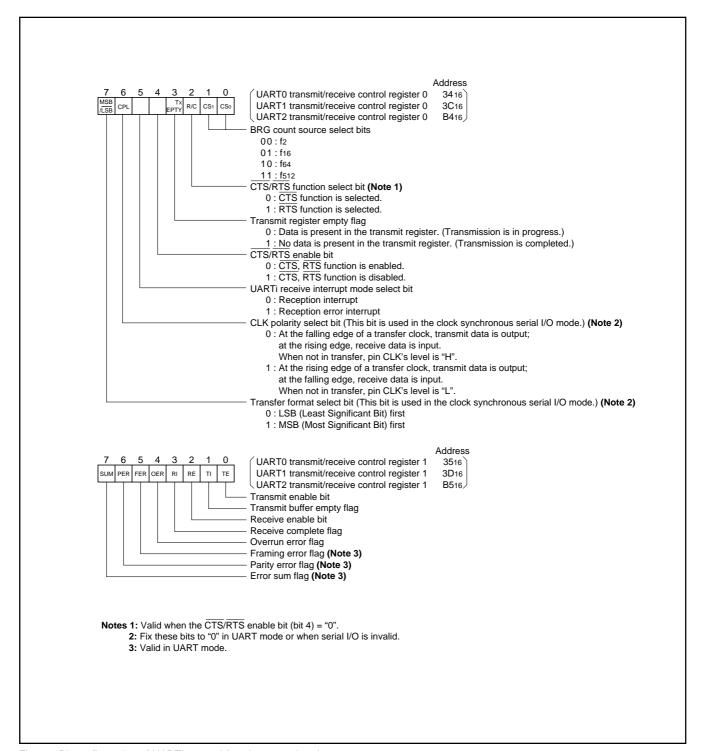


Fig. 64 Bit configuration of UARTi transmit/receive control register



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CLOCK SYNCHRONOUS SERIAL COMMUNI-CATION

A case where communication is performed between two clock synchronous serial I/O ports as shown in Figure 65 will be described. (The transmission side will be denoted by subscript j and the receiving side will be denoted by subscript k.)

Bit 0 of the UARTj transmit/receive mode register and UARTk transmit/receive mode register must be set to "1" and bits 1 and 2 must be "0". The length of the transmission data is fixed at 8 bits.

Bit 3 of the UARTj transmit/receive mode register of the clock sending side is cleared to "0" to select the internal clock. Bit 3 of the UARTk transmit/receive mode register of the clock receiving side is set to "1" to select the external clock. Bits 4, 5 and 6 are ignored in clock synchronous mode. Bit 7 must always be "0".

The clock source is selected by bit 0 (CS0) and bit 1 (CS1) of the clock-sending-side UARTj transmit/receive control register 0. As shown in Figure 60, the selected clock is divided by (n + 1), then by 2, is passed through a transmission control circuit, and is output as transmission clock CLKj. Therefore, when the selected clock is fi,

Bit Rate = $fi/\{(n + 1) \times 2\}$

On the clock receiving side, the CSo and CS1 bits of the UARTk transmit/receive control register 0 are ignored because an external clock is selected.

Both of UART0 and UART1 can use $\overline{\text{CTS}}$ and $\overline{\text{RTS}}$ functions.

Bit 4 of the UARTi transmit/receive control register 0 is used to determine whether to use $\overline{\text{CTS}}$ or $\overline{\text{RTS}}$ signal. Bit 4 must be "0" when $\overline{\text{CTS}}$ or $\overline{\text{RTS}}$ signal is used. Bit 4 must be "1" when $\overline{\text{CTS}}$ and $\overline{\text{RTS}}$ signals are not used. When $\overline{\text{CTS}}$ and $\overline{\text{RTS}}$ signals are not used, $\overline{\text{CTS}}/\overline{\text{RTS}}$ pin can be used as a normal port pin.

When using pin CTS/RTS, :

- If bit 2 of the UARTi transmit/receive control register 0 is cleared to "0", CTS input is selected.
- If bit 2 is set to "1", RTS output is selected.

The case using \overline{CTS} and \overline{RTS} signals are explained below. As shown in Figure 72, bits 2, 3 and 5 of the serial I/O pin control register can determine whether port pins P13, P17 and P83 are used as pins TxDi or as port pins. When bits 2, 3 and 5 are "0", P13, P17 and P83 function as pins TxDi; when bits 2, 3 and 5 are "1", P13, P17 and P83 function as port pins. Therefore, in the input-only system where pins TxDi are not used, pins TxDi can function as port pins.

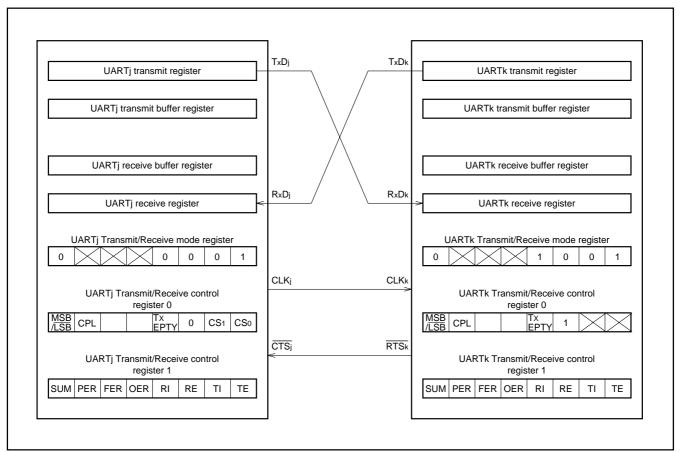


Fig. 65 Clock synchronous serial communication



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Transmission

Transmission is started when bit 0 (TEj flag: transmit enable bit) of UARTj transmit/receive control register 1 is "1", bit 1 (Tlj flag) of one is "0", and $\overline{\text{CTSj}}$ input is "L". The Tlj flag indicates whether the transmit buffer register is empty or not. It is cleared to "0" when data is written in the transmit buffer register; it is set to "1" when the contents of the transmit buffer register is transferred to the transmit register and the transmit buffer register becomes empty.

When all of the transmit conditions are satisfied, the transmit data in the transmit buffer register are transferred to the transmit register, and transmission starts. As shown in Figure 66, data is output from T_xD_j pin each time when transmission clock CLK_j changes from "H" to "L". (In the clock synchronous serial I/O mode, the polarity of a transfer clock can be changed. For details, refer to the section on the selection of the transfer clock polarity.) The data is output from the least significant bit.

When the transmit register becomes empty after the contents has been transmitted, data is transferred automatically from the transmit buffer register to the transmit register if the next transmission start condition is satisfied. The next transmission is performed succeedingly. Once transmission has started, the TEj flag, TIj flag, and $\overline{\text{CTS}_{ij}}$ signals are ignored until data transmission completes. Therefore, transmission is not interrupted when $\overline{\text{CTS}_{ij}}$ input is changed to "H" during transmission.

The transmission start condition indicated by TEj flag, TIj flag, and $\overline{\text{CTSj}}$ is checked while the TENDj signal (shown in Figure 66) is "H". Therefore, data can be transmitted continuously if the next transmission data is written in the transmit buffer register and TIj flag is cleared to "0" before the TENDj signal goes "H".

Bit 3 (TXEPTYj flag) of UARTj transmit/receive control register 0 changes to "1" at the next cycle just after the TENDj signal goes "H" and changes to "0" when transmission starts. Therefore, this flag can be used to determine whether data transmission has completed. When the TIj flag changes from "0" to "1", the interrupt request bit in the UARTj transmit interrupt control register is set to "1".

Receive

When bit 2 (REk flag) of the UARTk transmit/receive control register 1 is set to "1", reception becomes enabled. In this case, when the CLKk signal is input, the receive operation starts simultaneously with this signal.

The \overline{RTSk} output is "H" when the REk flag is "0". When the REk flag is set to "1", the \overline{RTSk} output becomes "L". This informs the transmitter side that reception becomes enabled. When the receive operation starts, the \overline{RTSk} output automatically becomes "H".

When the receive operation starts, the receiver takes data from pin RxDk each time when the transmit clock (CLKj) turns from "L" to "H". Simultaneously with reception, the contents of the receiver register is shifted bit by bit.

(Note that, in the clock synchronous serial communication, the polarity of a transfer clock can be inverted. For details, refer to the section on the polarity of the transfer clock.) When an 8-bit data is received, the contents of the receive register is transferred to the receive buffer register and bit 3 (Rlk flag) of UARTk transmit/receive control register 1 is set to "1". In other words, the setting "1" to the Rlk flag indicates that the receive buffer register contains the received data. At this time, if the low-order byte of the UARTk receive buffer register is

read out, the RTSk output turns back to "L". This indicates that the next data reception becomes enabled. Bit 4 (OERk flag) of UARTk transmit/receive control register 1 is set to "1" when the next data is transferred from the receive register to the receive buffer register while Rlk flag is "1", and indicates that the next data was transferred to the receive register before the contents of the receive buffer register was read. (In other words, this indicates that an overrun error has occurred.) Rlk flag is automatically cleared to "0" when the low-order byte of the receive buffer register is read or when the REk flag is cleared to "0". The OERk flag is cleared when the REk flag is cleared. Bit 5 (FERk flag), bit 6 (PERk flag), and bit 7 (SUMk flag) are ignored in clock synchronous mode.

As shown in Figure 60, with clock synchronous serial communication, data cannot be received unless the transmitter is operating because the receive clock is created from the transmission clock. Therefore, the transmitter must be operating even when there is no need to sent data from UARTk to UARTj.





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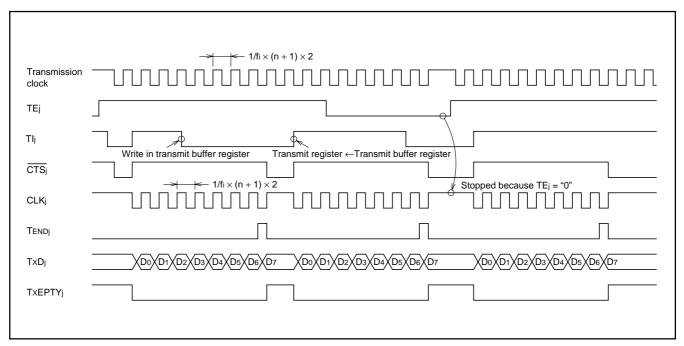


Fig. 66 Clock synchronous serial I/O timing

Interrupt request at completion of reception

When the Rlk flag changes from "0" to "1", in other words, when the receive operation is completed, the interrupt request bit of the UARTk receive interrupt control register can be set to "1".

The timing when this interrupt request bit is to be set to "1" can be selected from the following:

- Each reception
- When an error occurs at reception

If bit 5 of the UARTk transmit/receive control register 0 (UART receive interrupt mode select bit) is cleared to "0", the interrupt request bit is set to "1" at each reception. If bit 5 is set to "1", the interrupt request bit is set to "1" only when an error occurs. (In the clock synchronous serial communication, only when an overrun error occurs, the interrupt request bit is set to "1".)

Polarity of transfer clock

In the clock synchronous serial communication, by bit 6 of the UARTj transmit/receive control register 0 (CPL), the polarity of a transfer clock can be selected.

As shown in Figure 67, when bit 6 = "0", the polarity is as follows:

- In transmission, transmit data is output at the falling edge of CLKj.
- In reception, receive data is input at the rising edge of CLKk.
- When not in transfer, CLKi is at "H" level.

When bit 6 = "1", the polarity is as follows:

- In transmission, transmit data is output at the rising edge of CLKj.
- In reception, receive data is input at the rising edge of CLKk.
- When not in transfer, CLKi is at "L" level.





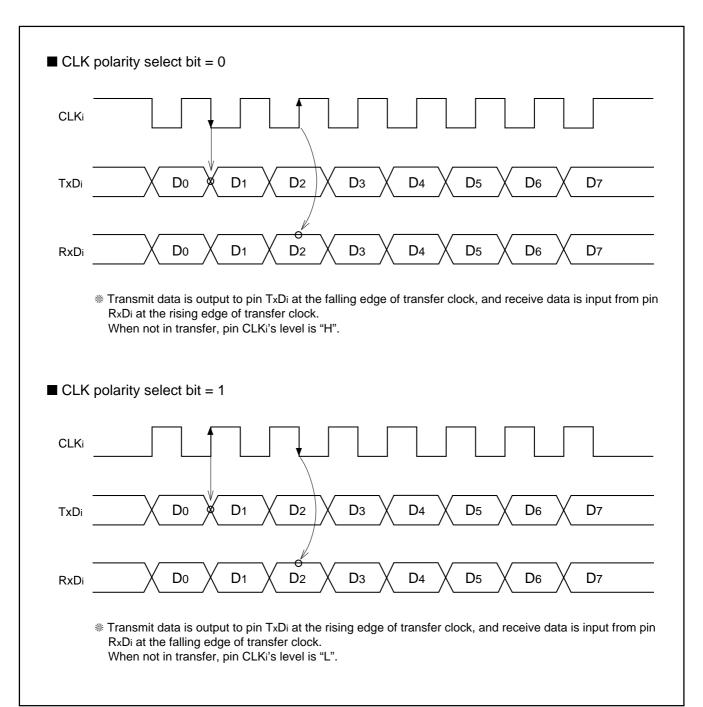


Fig. 67 Polarity of transfer clock



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Selection of transfer format

In clock synchronous serial communication, transfer format can be selected by bit 7 of the transmit/receive control register 0. When bit 7 is "0", transfer format is LSB first. When bit 7 is "1", transfer format is MSB first.

This function is realized by changing connection relation between

the transmit buffer register and the receive buffer register when writing transmit data to the transmit buffer register or reading receive data from the receive buffer register. Accordingly, the transmitter's operation is the same in both transfer formats.

Figure 68 shows the connection relation.

Bit 7 in transmit/receive control register 0	Write to transmit buffer register		Read from receive buffer register	
	Data bus	Transmit buffer register	Data bus DB7 ←	Receive buffer register
0 (LSB first)	DB6 ———	—> D6 —> D5	DB6 <	D6 D5
	DB4 ———	→ D4	DB4 ← DB3 ←	D4
	DB2 ——		DB2 <	—— D2
	DB1 ———	→ D1 → D0	DB1 ← DB0 ←	—— D1 —— D0
	Data bus	Transmit buffer register	Data bus	Receive buffer register
	DB7 \	D7 D6	DB ₇	D7
1 (MSB first)	DB ₅	D5 D4	DB ₅	D5
,	DB3	D3	DB3	D3
	DB2 /// DB1	D2 D1	DB ₂	D2 D1
	DBo /	₩ Do	DB₀ V	\ D0

Fig. 68 Connection relation between transmit buffer register, receive buffer register, and data bus

Precautions for clock synchronous serial communication

In the clock synchronous serial communication, the separate function for $\overline{\text{CTSi/RTSi}}$ cannot be selected. Furthermore, when an internal clock is selected, $\overline{\text{RTS}}$ output is undefined. Therefore, do not use the $\overline{\text{RTS}}$ function.

Before transmit operation is performed, be sure to clear bits 2, 3 and 5 of the serial I/O pin control register (address AC16) to "00".



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ASYNCHRONOUS SERIAL COMMUNICATION

Asynchronous serial communication can be performed using 7-, 8-, or 9-bit length data. The operation is the same for all data lengths. The following is the description for 8-bit asynchronous communication. With 8-bit asynchronous communication, bit 0 of UARTi transmit/receive mode register is "1", bit 1 is "0", and bit 2 is "1".

Bit 3 is used to select an internal clock or an external clock. If bit 3 is "0", an internal clock is selected and if bit 3 is "1", then external clock is selected. If an internal clock is selected, bit 0 (CS0) and bit 1 (CS1) of UARTi transmit/receive control register 0 are used to select the clock source. When an internal clock is selected for asynchronous serial communication, the CLKi pin can be used as a normal I/O pin. The selected internal or external clock is divided by (n+1), then by 16, and is passed through a control circuit to create the UART transmission clock or UART receive clock.

Therefore, the transmission speed can be changed by changing the contents (n) of the bit rate generator. If the selected clock is an internal clock Pfi or an external clock fext,

Bit 4 is the stop bit length select bit to select 1 stop bit or 2 stop bits. Bit 5 is a select bit of odd parity or even parity.

In the odd parity mode, the parity bit is adjusted so that the sum of 1s in the data and parity bit is always odd.

In the even parity mode, the parity bit is adjusted so that the sum of the 1s in the data and parity bit is always even.

Bit 6 is the parity bit select bit which indicates whether to add parity bit or not.

Bits 4 to 6 must be set or reset according to the data format used in the communicating devices.

Bit 7 is the sleep select bit. The sleep mode is described later.

Figure 72 shows the bit configuration of the serial I/O pin control register. By bits 0, 1 and 4 of the serial I/O pin control register ($\overline{CTSi}/\overline{RTS}$) separate select bits), the function of the $\overline{CTS}/\overline{RTS}$ pin can be separated into two functions, and each function can be assigned to two different pins. When each of bits 0, 1 and 4 = "1", the above separation is performed. When each of bits 0, 1 and 4 = "0", no separation is performed.

Table 8 lists the selection methods of the CTS/RTS function.

Bit Rate = (fi or fEXT) / $\{(n+1)\times 16\}$

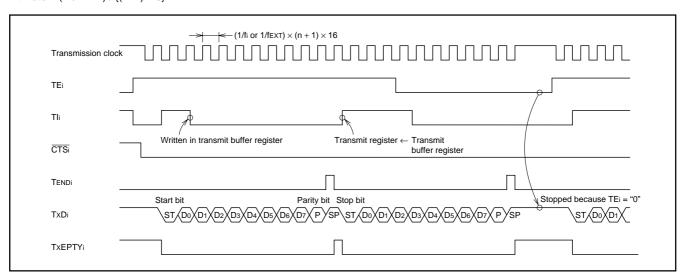


Fig. 69 Transmit timing example when 8-bit asynchronous communication with parity and 1 stop bit selected

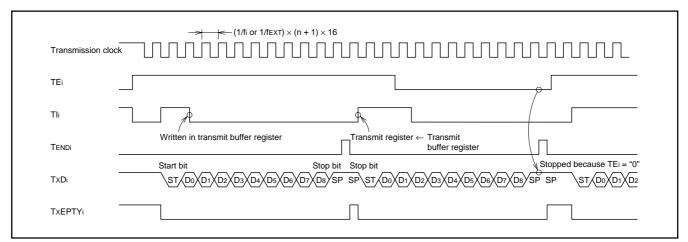


Fig. 70 Transmit timing example when 9-bit asynchronous communication with no parity and 2 stop bits selected





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Transmission

Transmission is started when bit 0 (TEi flag transmit enable flag) of UARTi transmit/receive control register 1 is "1", bit 1 (Tli flag) is "0", and $\overline{\text{CTSi}}$ input (in other words, transmit enable signal input from receiver) is "L". The Tli flag indicates whether the transmit buffer is empty or not. It is cleared to "0" when data is written in the transmit buffer; it is set to "1" when the contents of the transmit buffer register is transferred to the transmit register.

When all of the transmission conditions are satisfied, transmit data is transferred to the transmit register, and transmit operation starts. As shown in Figures 69 and 70, data is output from the TxDi pin with the stop bit or parity bit specified by bits 4 through 6 of UARTi transmit/receive mode register. The data is output from the least significant bit.

When the transmit register becomes empty after the contents has been transmitted, data is transferred automatically from the transmit buffer register to the transmit register if the next transmit start condition is satisfied. Then, the next transmission is performed succeedingly.

Once transmission has started, the TEi flag, Tli flag, and $\overline{\text{CTSi}}$ signal are ignored until data transmission is completed.

Therefore, transmission does not stop until it completes event if, during transmission, the TEi flag is cleared to "0" or CTSi input is set to "1"

The transmission start condition indicated by TEi flag, Tli flag, and $\overline{\text{CTSi}}$ is checked while the TENDi signal shown in Figure 69 is "H". Therefore, data can be transmitted continuously if the next transmission data is written in the transmit buffer register and Tli flag is cleared to "0" before the TENDi signal goes "H".

Bit 3 (TXEPTYi flag) of UARTi transmit/receive control register 0 changes to "1" at the next cycle just after the TENDi signal goes "H" and changes to "0" when transmission starts. Therefore, this flag can be used to determine whether data transmission is completed.

When the Tli flag changes from "0" to "1", the interrupt request bit of the UARTi transmit interrupt control register is set to "1".

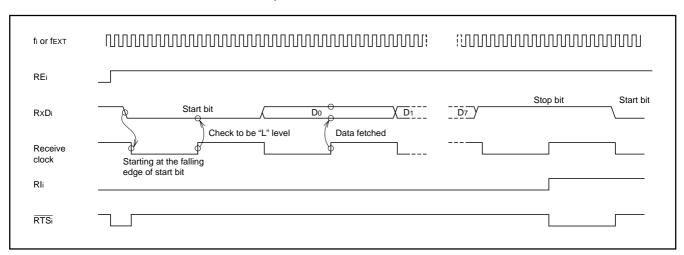


Fig. 71 Receive timing example when 8-bit asynchronous communication with no parity and 1 stop bit selected



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Table 8. Selection methods of CTS/RTS function

CTS/RTS	CTSi/RTSi	CTS/RTS	Functions							
enable bit	separate select bit	function select bit	Pin P10/CTS0/RTS0	Pin P11/CTS0/CLK0	Pin P14/CTS1/RTS1	Pin P15/CTS1/CLK1	Pin P80/CTS2/RTS2	Pin P81/CTS2/CLK2		
0 0	0	0	CTS ₀	P11 or CLKo	CTS ₁	P15 or CLK1	CTS2	P81 or CLK2		
	1	RTSo	P11 or CLKo	RTS ₁	P15 or CLK1	RTS2	P81 or CLK2			
	1	×	RTS ₀	CTSo (Notes 1, 2)	RTS ₁	CTS1 (Notes 1 and 2)	RTS2	CTS2 (Notes 1, 2)		
1	×	×	P10	P11 or CLKo	P14	P15 or CLK1	P80	P81 or CLK2		

X: It may be "0" or "1".

Notes 1: When using the CTS function, be sure to clear the corresponding bit of the port P1 and port P8 direction registers to "0".

2: When CTSi and RTSi has been separated, the CLKi pin cannot be used. Therefore, in the clock synchronous serial communication, CTSi and RTSi cannot be separated. Also, when CTSi and RTSi are separated in UART mode, be sure to select an internal clock.

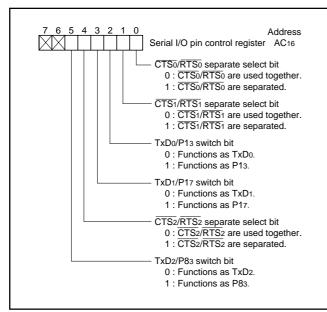


Fig. 72 Bit configuration of serial I/O pin control register

Receive

Receive is enabled when bit 2 (REi flag) of UARTi transmit/receive control register 1 is set to "1." As shown in Figure 71, the frequency divider circuit (1/16) at the receiving side begin to work when a start bit arrives and the data is received.

If \overline{RTSi} output is selected by setting bit 2 of UARTi transmit/receive control register 0 to "1", the \overline{RTSi} output is "H" when the REi flag is "0". When the REi flag changes to "1", the \overline{RTSi} output goes "L" to inform the receiver that reception has become enabled. When the receive operation starts, the \overline{RTSi} output automatically becomes "H". The entire transmission data bits are received when the start bit passes the final bit of the receive block shown in Figure 62. At this point, the contents of the receive register is transferred to the receive buffer register and bit 3 (Rli flag) of UARTi transmit/receive control register 1 is set to "1." In other words, the Rli flag indicates that the receive buffer register contains data when it is set to "1." At this time, when the low-order byte of the UARTk receive buffer register is read out, \overline{RTSi} output goes back to "L" to indicate that the register is ready to receive the next data.

Bit 4 (OERi flag) of UARTi transmit/receive control register 1 is set to "1" when the next data is transferred from the receive register to the receive buffer register while the RIi flag is "1", in other words, when an overrun error occurs. If the OERi flag is "1", it indicates that the next data has been transferred to the receive buffer register before the contents of the receive buffer register has been read.

Bit 5 (FERi flag) is set to "1" when the number of stop bits is less than required (framing error).

Bit 6 (PERi flag) is set to "1" when a parity error occurs.

Bit 7 (SUMi flag) is set to "1" when either the OERi flag, FERi flag, or the PERi flag is set to "1." Therefore, the SUMi flag can be used to determine whether there is an error.

The setting of the Rli flag, OERi flag, FERi flag, and the PERi flag is performed while transferring the contents of the receive register to the receive buffer register.





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The FERi, PERi, and SUMi flags are cleared to "0" when reading the low-order byte of the receive buffer register or when writing "0" to the REi flag.

The OERi flag is cleared to "0" when writing "0" to the REi flag.

Interrupt request at completion of reception

When the RIk flag changes from "0" to "1", in other words, when the receive operation is completed, the interrupt request bit of the UARTk receive interrupt control register can be set to "1".

The timing when this interrupt request bit is to be set to "1" can be selected from the following:

- Each reception
- · When an error occurs at reception

If bit 5 of the UARTk transmit/receive control register 0 (UART receive interrupt mode select bit) is cleared to "0", the interrupt request bit is set to "1" at each reception. If bit 5 is set to "1", the interrupt request bit is set to "1" only when an error occurs. (In the clock asynchronous serial communication, when an overrun error, framing error, or parity error occurs, the interrupt request bit is set to "1".)

Sleep mode

The sleep mode is used to communicate only between certain microcomputers when multiple microcomputers are connected through serial I/O.

The microcomputer enters the sleep mode when bit 7 of UARTi transmit/receive mode register is set to "1".

The operation of the sleep mode for an 8-bit asynchronous communication is described below.

When sleep mode is selected, the contents of the receive register is not transferred to the receive buffer register if bit 7 (bit 6 if 7-bit asynchronous communication and bit 8 if 9-bit asynchronous communication) of the received data is "0". Also the Rli, OERi, FERi, PERi, and the SUMi flags are unchanged. Therefore, the interrupt request bit of the UARTi receive interrupt control register is also unchanged. Normal receive operation takes place when bit 7 of the received data is "1".

The following is an example of how the sleep mode can be used. The main microcomputer first sends data: bit 7 is "1" and bits 0 to 6 are set to the address of the subordinate microcomputer to be communicated with. Then all subordinate microcomputers receive this data. Each subordinate microcomputer checks the received data, clears the sleep bit to "0" if bits 0 through 6 are its own address and sets the sleep bit to "1" if not. Next, the main microcomputer sends data with bit 7 cleared. Then the microcomputer which cleared the sleep bit to "1" will receive the data, but the microcomputers which set the sleep bit to "1" will not. In this way, the main microcomputer is able to communicate only with the designated microcomputer.

Precautions for clock asynchronous (UART) serial communication

When $\overline{\text{CTSi}}$ and $\overline{\text{RTSi}}$ are separated, pin CLKi cannot be used. Therefore, when $\overline{\text{CTSi}}$ and $\overline{\text{RTSi}}$ are separated in UART mode, be sure to select an internal clock.

Before transmit operation is performed, be sure to clear bits 2, 3 and 5 of the serial I/O pin control register (address AC16) to "00".





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A-D CONVERTER

The A-D converter is a 10-bit successive approximation converter. The use of A-D converter or the use of comparator can be selected for each A-D input pin. The contents of the comparator function select register specify it.

Figure 73 shows a block diagram of the A-D converter.

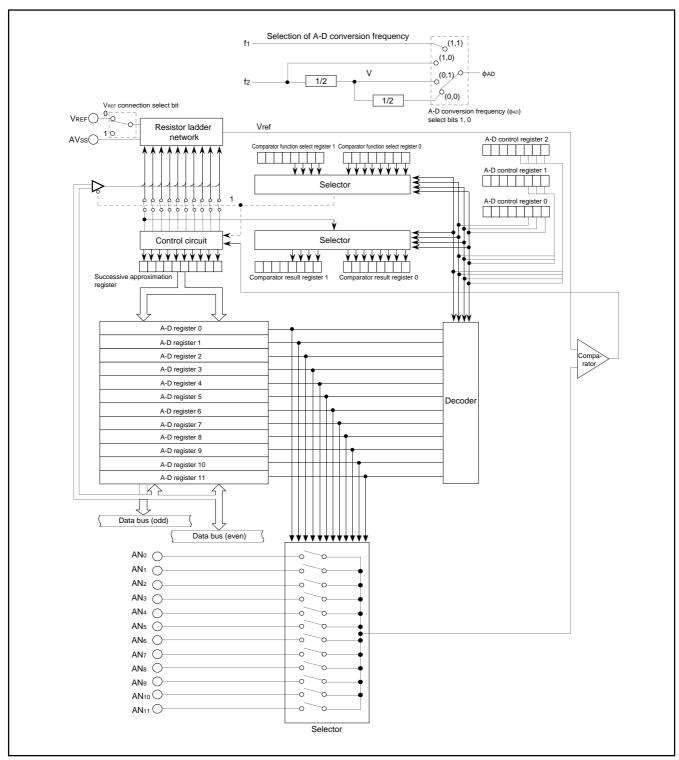


Fig. 73 Block diagram of A-D converter





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Figure 74 shows the bit configuration of the comparator function select register 0 (address DC16), and Figure 75 shows that of the comparator function select register 1 (address DD16). Each of bits 7 to 0 corresponds to its own channel, respectively. Each channel can be selected as either an A-D converter or a comparator. When the bit is "0", the channel corresponding to it functions as a 10-bit or an 8-bit A-D converter. When the bit is "1", the channel functions as a comparator.

When selecting an A-D converter, an input voltage to a selected analog input pin is A-D converted and the result is stored into one of these A-D registers.

When selecting a comparator, D-A conversion is performed to the value of which high-order 8 bits are the value stored in an even address of the A-D converter and of which low-order 2 bits are "102." Then, this D-A converted value is compared with the voltage supplied to an analog input pin. After the comparison, when the voltage supplied to an analog input pin is higher, "1" is stored into the comparator result register 0 (address DE16) shown in Figure 76, or the comparator result register 1 (Address DF16) shown in Figure 77. When it is lower, "0" is stored into that of these register.

Be sure to perform only read to the A-D register of which channel is selected as an A-D converter, and perform only write to the A-D register of which channel is selected as a comparator. Additionally, do not write to the comparator function select registers 0, 1 and the A-D register while an A-D converter or a comparator is operating.

Port direction register's bits corresponding to pins to be A-D converted must be "0" (input mode) because analog input ports are multiplexed with ports P7 and P8.

Figure 78 shows the bit configuration of the A-D control register 0 (address 1E₁₆), Figure 79 shows that of the A-D control register 1 (address 1F₁₆), and Figure 80 shows that of the A-D control register 2 (address DB₁₆).

The operation clock of the A-D converter, ϕ AD, is selected by the following bits: bit 7 of the A-D control register 0 and bit 4 of the A-D control register 1.

When bit 4 of the A-D control register 1 = "0", ϕ AD is selected as follows:

- if bit 7 of the A-D control register 0 = 0, $\phi AD = f^2/4$.
- if bit 7 of the A-D control register 0 = "1", $\phi AD = f2/2$.

When bit 4 of the A-D control register 1 = "1", ϕ AD is selected as follows:

- if bit 7 of the A-D control register 0 = "0", $\phi AD = f2$.
- if bit 7 of the A-D control register 0 = "1", $\phi AD = f1$.

Note that the highest frequency, $\phi AD = f1$, can be selected only in the 8-bit resolution mode.

 ϕ AD during A-D conversion must be 250 kHz or more because the comparator uses a capacity coupling amplifier.

Bit 3 of A-D control register 1 is used to select whether to regard the conversion result as 10-bit or as 8-bit data. The conversion result is regarded as 10-bit data when bit 3 is "1" and as 8-bit data when bit 3 is "0".

When the conversion result is used as 10-bit data, the low-order 8 bits of the conversion result are stored in the even address of the corresponding A-D register and the high-order 2 bits are stored in bits 0 and 1 at the odd address of the corresponding A-D register. Bits 2 to 7 of the A-D register odd address are "0000002" when read. When the conversion result is used as 8-bit data, the high-order 8

bits of the 10-bit A-D conversion result are stored in even address of the corresponding A-D register. In this case, the value at the A-D register's odd address is "0016" when read.

Whether to connect the reference voltage input (VREF) with the ladder network or not depends on bit 5 of the A-D control register 1. The VREF pin is connected when bit 5 is "0" and is disconnected when bit 5 is "1" (High impedance state).

When A-D or D-A conversion is not performed, current from the VREF pin to the ladder network can be cut off by disconnecting ladder network from the VREF pin.

Before starting A-D conversion, wait for 1 μ s or more after clearing bit 5 to "0".





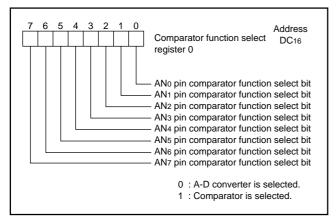


Fig. 74 Bit configuration of comparator function select register 0

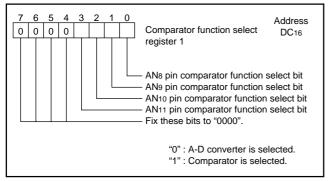


Fig. 75 Bit configuration of comparator function select register 1

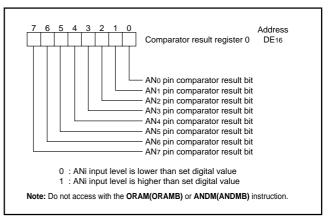


Fig. 76 Bit configuration of comparator result register 0

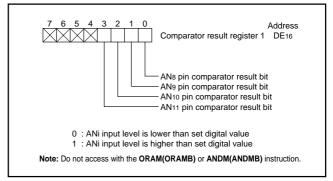


Fig. 77 Bit configuration of comparator result register 1





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Operation mode

The operation mode is selected by bits 3 and 4 of A-D control register 0 and bit 2 of A-D control register 1. The available operation modes are one-shot, repeat, single sweep, repeat sweep 0, and repeat sweap 1. Note that, as for pins AN8 through AN11, only one-shot and repeat modes can be selected. Either an A-D converter or a comparator can be selected respectively for each pin in the following 5 modes. The following description applies to the case where the bit of the comparator function select register 0/1 is "0" and an A-D converter is selected. It also applies to a comparator's operation except that an A-D conversion is changed to a comparator operation and the result of the comparison is stored into the comparator result register 0/1.

(1) One-shot mode

One-shot mode is selected when bits 3 and 4 of A-D control register 0 are "0". The A-D conversion pins are selected with bits 0 to 2 of A-D control register 0 and bits 0 to 3 of A-D control register 2. A-D

conversion or comparator operation is started when bit 6 of A-D control register 0 (A-D conversion start bit) is set to "1".

When the ANi (i = 11 through 0) comparator function select bit of the comparator function select register 0/1 = "0" and bit 3 of the A-D control register 1 = "1", A-D conversion ends 59 ϕ AD cycles after, and the interrupt request bit of the A-D conversion interrupt control register is set to "1". At the same time, bit 6 of the A-D control register 0 (A-D conversion start bit) is cleared to "0" and this A-D conversion stops. The result of A-D conversion is stored into the A-D register corresponding to the selected pin.

When the ANi (i = 11 through 0) comparator function select bit of the comparator function select register 0/1 = "1", a comparator operation ends 14 ϕ AD cycles after, and the interrupt request bit of the A-D conversion interrupt control register is set to "1". At the same time, bit 6 of the A-D control register 0 (A-D conversion start bit) is cleared to "0" and the comparator operation stops. The result of the comparison is stored into the bits of the comparator result register corresponding to the selected pin.

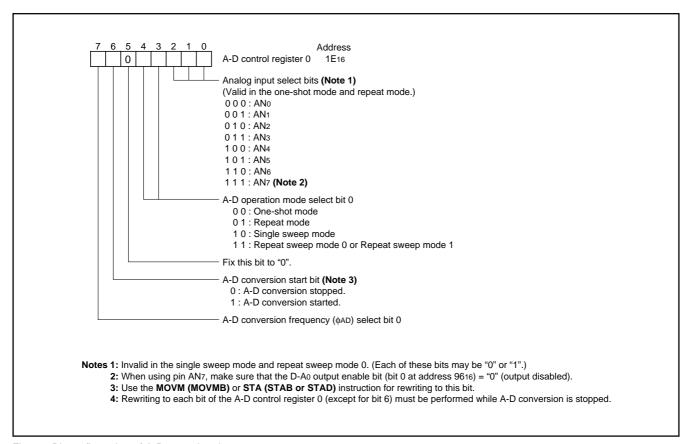


Fig. 78 Bit configuration of A-D control register 0



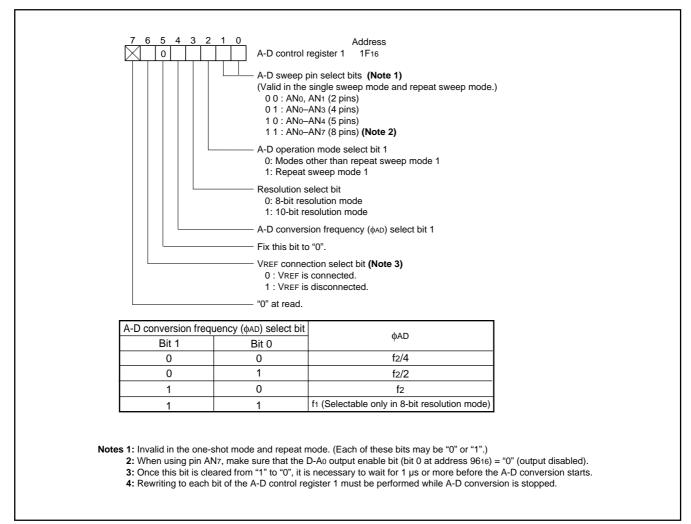


Fig. 79 Bit configuration of A-D control register 1

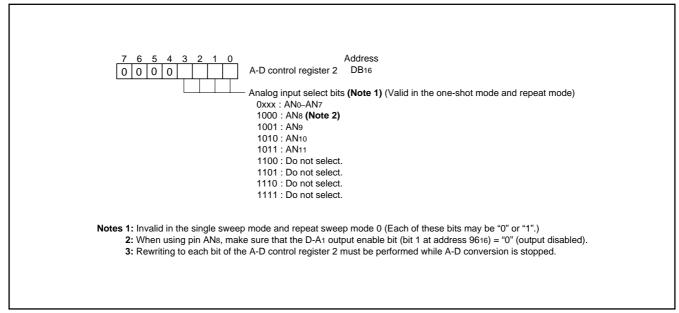


Fig. 80 Bit configuration of A-D control register 2





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(2) Repeat mode

Repeat mode is selected when bit 3 of the A-D control register 0 = "1" and bit 4 = "0".

The operation of this mode is the same as the operation of one-shot mode except that when A-D conversion for the selected pin is complete and the result is stored in the A-D register, conversion does not stop, but is repeated.

No interrupt request is generated in this mode. Furthermore, the A-D conversion start bit is not cleared.

The contents of the A-D register can be read at any time.

Be sure not to write to the A-D register corresponding to the pins selected for a comparator during operation.

(3) Single sweep mode

Single sweep mode is selected when bit 3 of the A-D control register 0 = 0 and bit 4 = 1.

In the single sweep mode, the number of analog input pins to be swept can be selected. Analog input pins are selected by bits 1 and 0 of the A-D control register 1 (address 1F16). Two pins, four pins, or five pins can be selected as analog input pins, depending on the contents of these bits.

A-D conversion is performed only for selected input pins. After A-D conversion is performed for input of ANo pin, the conversion result is stored in A-D register 0, and in the same way, A-D conversion is performed for selected pins one after another. After A-D conversion is performed for all selected pins, the sweep is stopped.

A-D conversion is started when bit 6 of the A-D control register 0 (A-D conversion start bit) is set to "1". When A-D conversion for all selected pins end, the interrupt request bit of the A-D conversion interrupt control register is set to "1". At the same time, A-D conversion start bit is cleared to "0" and A-D conversion stops.

(4) Repeat sweep mode 0

Repeat sweep mode 0 is selected when bit 3 of the A-D control register 0 = "1" and bit 4 = "1".

The difference from the single sweep mode is that A-D conversion does not stop after conversion for all selected pins, but repeats again from the ANo pin. The repeat is performed among the selected pins. Also, no interrupt request is generated. Furthermore, the A-D convension start bit is not cleared. The contents of the A-D register can be read at any time.

Be sure not to write to the A-D register corresponding to the pins selected for a comparator during operation.

(5) Repeat sweep mode 1

Repeat sweep mode 1 is selected when bit 3 of the A-D control register 0 = "1" and bit 4 = "1", and bit 2 of the A-D control register 1 = "1". The differences from the repeat sweap mode 0 are as follows:

- the A-D conversion for one unselected pin is performed each time when A-D conversion for selected pins is completed, and the A-D conversion is repeated once again from ANo pin.
- the number of analog input pins to be swept.

The analog input pins to be repeatedly swept are selected with bits 1 and 0 of the A-D control register 1. The contents of these pins are used to select one pin, two pins, three pins or four pins.

The unselected pins are converted starting from the pin next to the

pins selected as repeat sweep pins. No interrupt request is generated. Furthermore, the A-D conversion start bit is not cleared. The contents of the A-D register can be read at any time. Be sure not to write to the A-D register, corresponding to the pins selected for a comparator, during operation.

Precaution for A-D conversion interrupts

Clear the interrupt request bit of the A-D conversion interrupt control register (bit 3 at address 7016) before using an A-D conversion interrupt. It is because this interrupt request bit is undefined just after reset.





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D-A CONVERTER

Two independent D-A converters are included in this microcomputer, and each D-A converter adopts an 8-bit R-2R method. Figure 81 shows the block diagram of the D-A converter, and Figure 82 shows the bit configuration of the D-A control register (address 9616).

D-A conversion is performed by writing a value to the corresponding D-A register i. Whether to output the analog voltage or not is determined by bits 0 and 1 of the D-A control register. When any of bits 0 and 1 = "1", the corresponding pin (D-Ao or D-A1) outputs the analog voltage.

This analog voltage (V) is determined according to value n. ("n" = decimal number. This has been set in the D-A register.)

V = VREF X n/256 (n = 0 to 255)

VREF : Reference voltage

The contents of the corresponding D-A output enable bit and D-A register are cleared to "0" at reset.

An external buffer is necessary when connecting a low impedance load with the D-A converter. It is because that a D-A output pin does not include a buffer.

Pin D-Ai (i = 0, 1) is multiplexed with I/O port pins, analog input pins, and external interrupt input pins. When a D-Ai output enable bit = "1" (in other words, output is enabled.), however, the corresponding pin cannot function as another I/O pin, which is multiplexed with pin D-Ai.

Also, when not using the D-A converter, be sure to clear the contents of the corresponding D-A output enable bit and D-A register to "0".

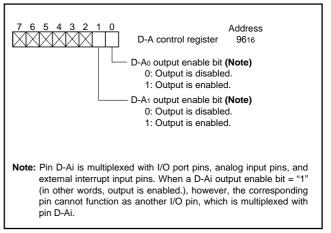


Fig. 82 Bit configuration of D-A control register

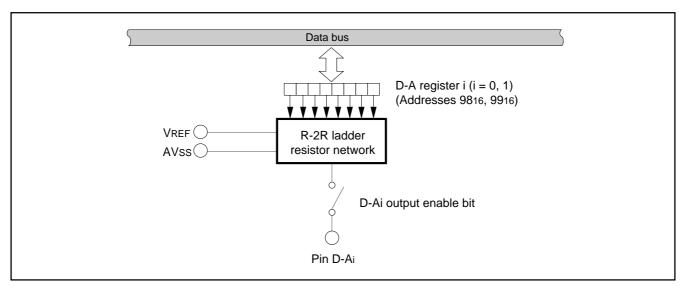


Fig. 81 Block diagram of D-A converter



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WATCHDOG TIMER

The watchdog timer is used to detect unexpected execution sequence caused by software runaway and others. Figure 83 shows the block diagram of the watchdog timer.

The watchdog timer consists of a 12-bit binary counter.

The watchdog timer counts clock Wf32, which is obtained by dividing the peripheral devices' clock f2 by 16; or clock Wf512, which is obtained by doing it by 256. Bit 0 of the watchdog timer frequency select register (watchdog timer frequency select bit) shown in Figure 84 selects which clock is to be counted.

Wf512 is selected when this bit 0 is "0", and Wf32 is selected when bit 0 is "1". Bit 0 is cleared to "0" after reset.

FFF16 is set in the watchdog timer when "L" level voltage is applied to pin RESET, STP instruction is executed, data is written to the watchdog timer register (address 6016), or the most significant bit of the watchdog timer becomes "0".

After FFF16 is set in the watchdog timer, when the watchdog timer counts Wf32 or Wf512 by 2048 counts, the most significant bit of the watchdog timer becomes "0", the watchdog timer interrupt request bit is set to "1", and FFF16 is set again in the watchdog timer.

In program coding, make sure that data is written in the watchdog timer before the most significant bit of the watchdog timer becomes "0". If this routine is not executed owing to unexpected program execution or others, the most significant bit of the watchdog timer be-

comes "0" and an interrupt is generated.

The microcomputer can generate a reset pulse by writing "1" to bit 6 (software reset bit) of processor mode register 0 in an interrupt routine and can be restarted.

The watchdog timer can also be used to return from the **STP** state, where a clock has stopped its operation owing to the **STP** instruction execution. For details, refer to the sections on the clock generating circuit and standby function.

The watchdog timer stops its operation in the following cases, and at this time, input to the watchdog timer is disabled:

- When the external area is accessed in the hold state
- In the wait mode
- In the stop mode

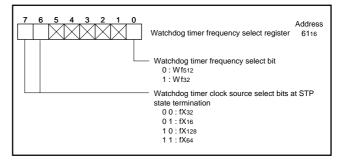


Fig. 84 Bit configuration of watchdog timer frequency select register

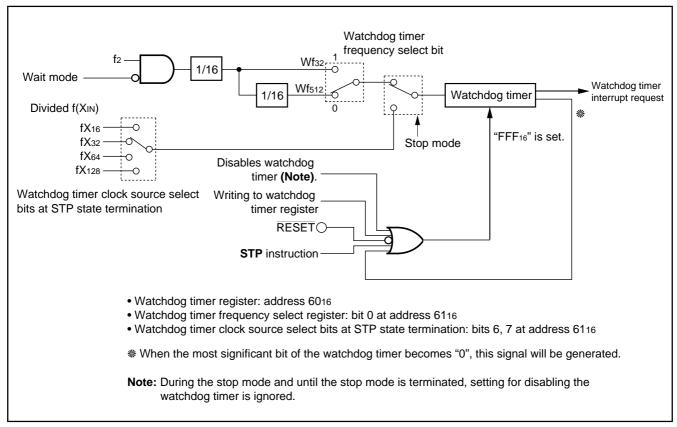


Fig. 83 Block diagram of watchdog timer



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How to disable watchdog timer

When not using the watchdog timer, it can be disabled. When the watchdog timer is disabled, it's operation stops and no watchdog timer interrupt has been generated.

Setting for disabling the watchdog timer is possible by writing "7916" and "5016" to the particular function select register 2 (address 6416) sequentially with the following instructions:

- MOVMB/STAB instruction, or
- MOVM/STA instruction (m = 1)

If any method other than above has been adopted in order to access (in other words, read/write) the particular function select register 2, the watchdog timer will not be disabled until reset operation is performed. (Also, reset is the only one method to remove the setting for disabling the watchdog timer.)

Moreover, this setting for disabling the watchdog timer is ignored at return from the STP mode, and the watchdog timer operates. (For details, refer to the section on the standby function.)





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INPUT/OUTPUT PINS

Ports P1, P2, and P4 through P8 all have the direction register, and each bit can be programmed for input or output. A pin becomes an output pin when the corresponding bit of direction register is "1", and an input pin when it is "0".

Also, each bit of the port P6 direction register can be cleared to "0" by inputting a falling edge to pin P6OUTcuT or by executing instructions. Each bit of the port P4 direction register can be cleared to "0" by inputting a falling edge to pin P4OUTcuT or by executing instructions.

When a pin is programmed for output, the data is written to its port latch and it is output to the output pin. When a pin is programmed for output, the contents of the port latch is read instead of the value of the pin. Accordingly, a previously output value can be read correctly even when the output "H" voltage is lowered or the output "L" voltage is raised owing to an external load, etc.

A pin programmed as an input pin is in the flooting state, and the value input to the pin can be read. When a pin is programmed as an input pin, the data is written only in the port latch and the pin remains floating.

Each of Figures 85 and 86 shows the block diagram for each port pin.

When using a port pin as an internal peripheral device's input pin, clear the corresponding port direction register's bit to "0". When using a port pin as an internal peripheral device's output pin, the port direction register's bit may be "0" or "1".





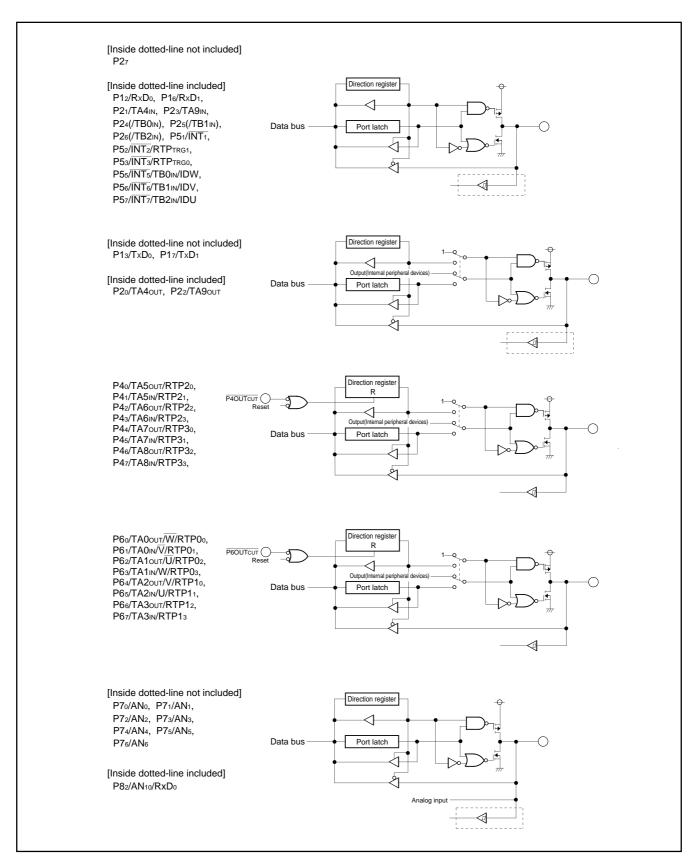


Fig. 85 Block diagram for each port pin (1)





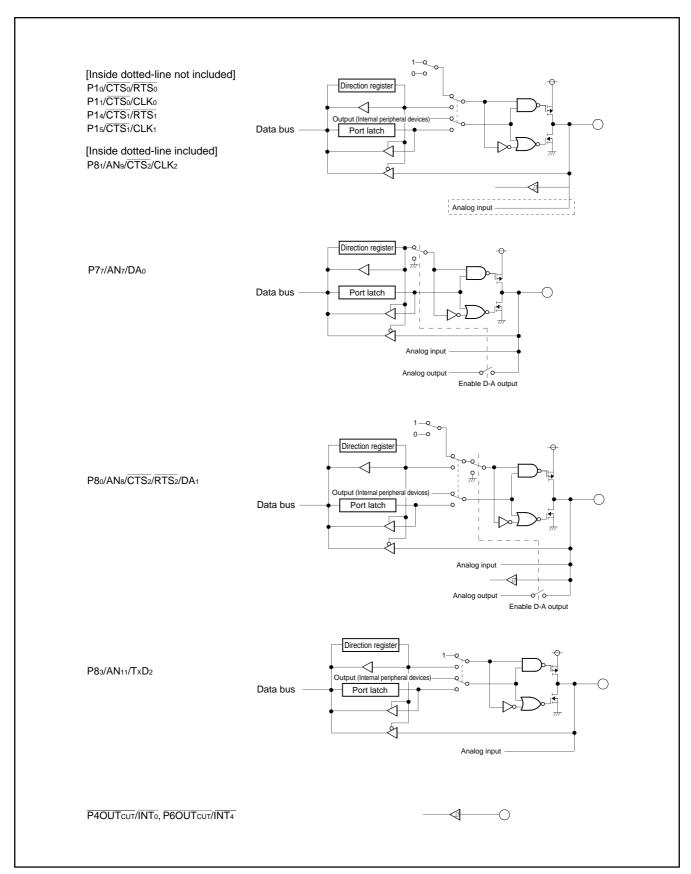


Fig. 86 Block diagram for each port pin (2)



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RESET CIRCUIT

While the power source voltage satisfies the recommended operating condition, reset state is removed if pin $\overline{\text{RESET}}$'s level returns from the stabilized "L" level to the "H" level. As a result, program execution starts from the reset vector address. This reset vector address is expressed as shown below:

- A23 to A16 = 0016
- A15 to A8 = Contents at address FFFF16
- A7 to A0 = Contents at address FFFE16

Figures 87 and 88 show the microcomputer internal register's status at reset, and Figure 89 shows an operation example of the reset circuit. Apply "L" level voltage to pin $\overline{\text{RESET}}$ for a period (10 μ s or more) under the following conditions:

- Pin Vcc's level satisfies the recommended operating condition.
- · Oscillator's operation has been stabilized.

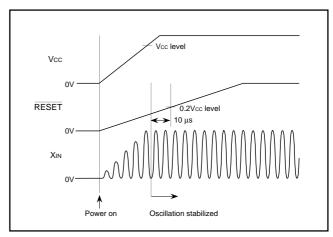


Fig. 89 Operation example of reset circuit (Note that proper evaluation is necessary in the system development stage.)

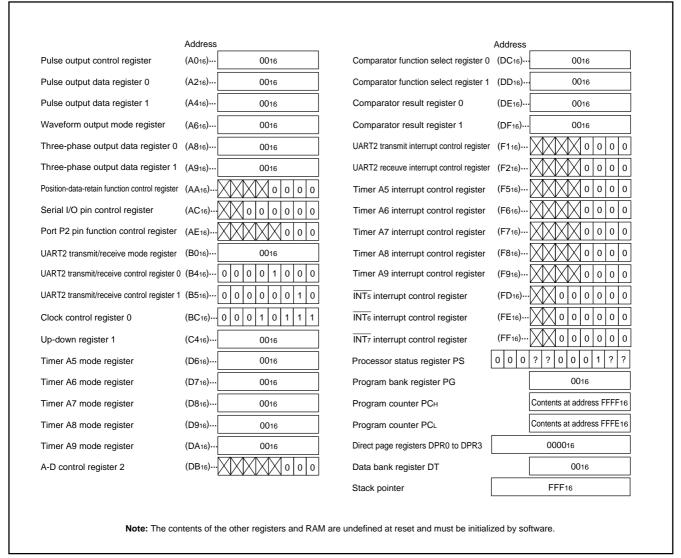


Fig. 88 Microcomputer internal register's status at reset (2)





Deat D4 discribes as sisten	Address	Processor and an address 0	Address
Port P1 direction register	(0516) 0016	Processor mode register 0	(5E ₁₆) 0 0 0 0 1 0 0 0
Port P2 direction register	(0816) 0016	Processor mode register 1	(5F ₁₆) 0 0 0 0 0 0 0 1
Port P4 direction register	(0C ₁₆) 00 ₁₆	Watchdog timer	(60 ₁₆) FFF ₁₆
Port P5 direction register	(0D ₁₆) 0 0 0 0 0 0	Watchdog timer frequency select registe	(61 ₁₆) 0 0 X X X 0
Port P6 direction register	(1016)	0 Particular function select register 0	(62 ₁₆) 0 0 0
Port P7 direction register	(1116)	Particular function select register 1	(63 ₁₆) 0 0 0 (Note 2)
Port P8 direction register	(1416)	0 Debug control register 0	(66 ₁₆) 1 (Note 2)
A-D control register 0	(1E ₁₆) 0 0 0 0 0 ?	? Debug control register 1	(67 ₁₆) 0 0 0 0 0 Note 2
A-D control register 1	(1F ₁₆) 0 0 0 0 0	? INT3 interrupt control register	(6E ₁₆)
UART 0 transmit/receive mode register	(3016) 0016	INT4 interrupt control register	(6F ₁₆)
UART 1 transmit/receive mode register	(3816) 0016	A-D conversion interrupt control registe	er (70 ₁₆) ? 0 0 0
UART 0 transmit/receive control register 0	(3416) 0 0 0 0 1 0 0	0 UART 0 transmit interrupt control registe	ır (71 ₁₆)
UART 1 transmit/receive control register 0	(3C ₁₆) 0 0 0 0 1 0 0	0 UART 0 receive interrupt control register	(7216)
UART 0 transmit/receive control register 1	(3516) 0 0 0 0 0 1	0 UART 1 transmit interrupt control registe	r (7316) 0 0 0 0
UART 1 transmit/receive control register 1	(3D ₁₆) 0 0 0 0 0 1	0 UART 1 receive interrupt control register	(7416)
Count start register 0	(4016) 0016	Timer A0 interrupt control registe	er (75 ₁₆)
Count start register 1	(4116)	0 Timer A1 interrupt control registe	er (76 ₁₆)
One-shot start register 0	(4216)	Timer A2 interrupt control registe	er (77 ₁₆)
One-shot start register 1	(4316)	0 Timer A3 interrupt control registe	er (78 ₁₆)
Up-down register 0	(4416) 0016	Timer A4 interrupt control registe	er (79 ₁₆)
Timer A clock frequency select register	(4516)	0 Timer B0 interrupt control registe	er (7A ₁₆)
Timer A0 mode register	(5616) 0016	Timer B1 interrupt control registe	er (7B ₁₆)
Timer A1 mode register	(5716) 0016	Timer B2 interrupt control registe	er (7C ₁₆)
Timer A2 mode register	(5816) 0016	INTo interrupt control register	(7D ₁₆) 0 0 0 0 0 0
Timer A3 mode register	(59 ₁₆) 0016	INT1 interrupt control register	(7E ₁₆) 0 0 0 0 0 0
Timer A4 mode register	(5A ₁₆) 0016	INT2 interrupt control register	(7F ₁₆) 0 0 0 0 0 0
Timer B0 mode register	(5B ₁₆) 0 0 ? 0 0 0	0 D-A control register	(9616)
Timer B1 mode register	(5C ₁₆) 0 0 ? 0 0 0	0 D-A register 0	(9816) 0016
Timer B2 mode register	(5D ₁₆) 0 0 ? 0 0 0	0 D-A register 1	(9916) 0016

Fig. 87 Microcomputer internal register's status at reset (1)



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OSCILLATION CIRCUIT

An oscillation circuit locates between pins XIN and XOUT, and Figure 90 shows a circuit example with an external ceramic resonator or quartz crystal oscillator. The constants such as capacitance etc. depend on a resonator/oscillator. Therefore, for these constants, adopt the resonator/oscillator manufacturer's recommended values.

Figure 91 shows a circuit example with an external clock source. When an external clock is input, be sure to leave pin XOUT open. Also, in this case, when the external clock input select bit (bit 1 of the particular function select register 0; See Figure 95.) is set to "1", the oscillation circuit stops it's operation and resumes the current dissipation. Moreover, this bit has another function, which selects the return condition from the stop mode. For details, refer to the section on the standby function.

On the other hand, the PLL (Phase Locked Loop) frequency multiplier (hereafter, referred to as PLL circuit.) is included, also. This PLL circuit uses a clock input from pin XIN and generates a multiplicated clock. When using the PLL circuit, be sure to connect pin VCONT with an external filter circuit. (See Figure 92.) When not using the PLL circuit, be sure to leave pin VCONT open.

When not using the PLL circuit, be sure to clear the PLL circuit operation enable bit (bit 1 of the clock control register 0; See Figure 94.), so that the PLL circuit will stop its operation.

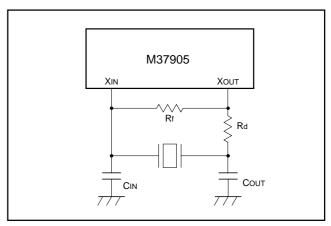


Fig. 90 Circuit example with external ceramic resonator or quartz-crystal oscillator

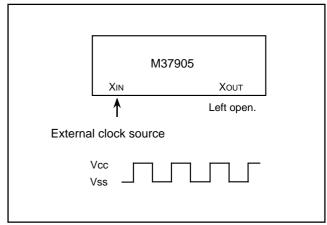


Fig. 91 Circuit example with external clock source

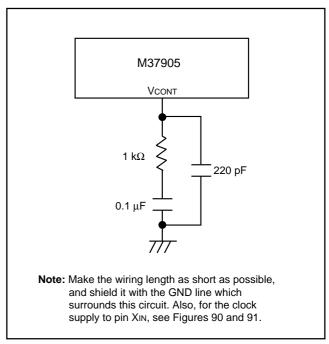


Fig. 92 Circuit example of connection with pin V_{CONT} when PLL circuit used





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CLOCK GENERATING CIRCUIT

Figure 93 shows the block diagram of the clock generating circuit. The clock generating circuit consists of the clock oscillation circuit, PLL frequency multiplier (PLL circuit), system clock switch circuit, peripheral devices' clock switch circuit, clock divider, standby control circuit, etc. As control registers for the clock generating circuit, also, the clock control register 0 (address BC16), particular function select register 0 (address 6216) are provided. (See Figures 94 and 95.) As shown in Figure 93, clocks used in the CPU, BIU, peripheral devices, watchdog timer (in other words, clocks ϕ CPU, ϕ BIU, f1 to f4096, Wf32, Wf512) are made from system clock fsys. System clock fsys can

and fPLL (in other words, an output clock generated by the PLL circuit.

The PLL circuit's operation, system clock (fsys) selection, and division ratio selection for peripheral devices' clocks (f1 to f4096) are controlled by the clock control register. The following describes about

be selected between fXIN (in other words, a clock input from pin XIN)

Bit 1 of the clock control register 0 (the PLL circuit operation enable bit) selects the PLL circuit's operation (inactive/active). When this bit is set to "1", pin VCONT will becomes valid, and the PLL circuit will be active. At reset, the PLL circuit operation enable bit becomes "1". (In this case, the PLL circuit is active.) When not using the PLL circuit, be sure to clear the PLL circuit operation enable bit to "0" (inactive). At the **STP** instruction execution, the PLL circuit is inactive, and pin VCONT is invalid, regardless of this bit 1's status.

Bits 2 and 3 of the clock control register (the PLL multiplication ratio select bits) select the ratio of fPLL/fXIN. The PLL multiplication ratio

must be set so that the frequency of fPLL must be in the range from 10 MHz to 20 MHz. At reset, the PLL multiplication ratio select bits become "0,1" (\times 2). The change of the PLL multiplication ratio must be performed while input clock fXIN is selected as the system clock. (In this case, bit 5 of the clock control register 0 = "0".) After that, be sure to wait that the operation-stabilizing time of the PLL circuit has passed, and switch the system clock to fPLL. (In other words, set bit 5 to "1".) Note that, after reset, the PLL multiplication ratio select bits are allowed to be changed only once.

Bit 5 of the clock control register 0 is the system clock select bit, and input clock fXIN is selected as the system clock when bit 5 = "0". On the other hand, when bit 5 = "1", fPLL is selected. At reset, the system clock select bit becomes "0". When selecting fPLL, be sure that the PLL circuit's operation has fully been stabilized, and then, set the system clock select bit to "1". Also, when the PLL circuit operation enable bit is cleared to "0" (the PLL circuit is inactive.), the system clock select bit will automatically be cleared to "0". Note that a value of "1" cannot be written to the system clock select bit while the PLL circuit operation enable bit = "0".

Table 9 lists the fsys selection.

Bits 6 and 7 of the clock control register 0 are the peripheral devices' clock select bits 0, 1, and these bits select the division ratio of (f1 to f4096)/(fsvs).

Table 10 lists the internal peripheral devices' operation clock frequency. At reset, these bits become "0, 0".

Table 9. f_{sys} selection

these control.

Table 3. Isys selection				
System clock select bit	PLL circuit operation enable bit	PLL multiplication ratio select bits	System of	clock fsys
(Bit 5)	(Bit 1)	(Bits 3, 2) (Note)	Clock source	Frequency (Note)
0			fXIN	f(XIN)
		01 (X 2)	fPLL	f(XIN) X 2
1	1	10 (X 3)	fPLL	f(XIN) × 3
		11 (X 4)	fPLL	f(XIN) X 4

Note: The PLL multiplication ratio must be set so that the frequency of fPLL must be in the range from 10 MHz to 20 MHz. f(XIN) means the frequency of the input clock from pin XIN (fXIN). After reset, the PLL multiplication ratio select bits are allowed to be changed only once.

Table 10. Internal peripheral devices' operation clock frequency

Internal peripheral devices'	Peripheral devices' clock select bits 1, 0 (bits 7, 6)						
operation clock	0 0	0 1 (Note)	1 0	11			
f1	fsys	fsys	fsys/2				
f2	fsys/2	fsys	fsys/4				
f16	fsys/16	fsys/8	fsys/32	Do not select.			
f64	fsys/64	fsys/32	fsys/128	Do not select.			
f512	fsys/512	fsys/256	fsys/1024				
f4096	fsys/4096	fsys/2048	fsys/8192				

Note: When selecting the peripheral devices' clock select bits 1, 0 = "012", be sure that system clock fsys does not exceed 10 MHz.





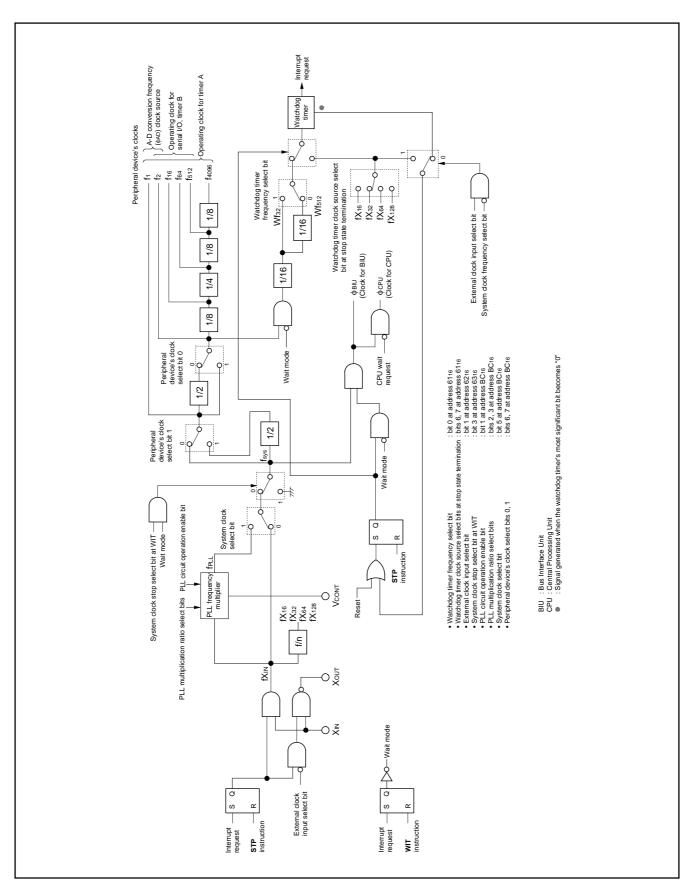


Fig. 93 Block diagram of clock generating circuit





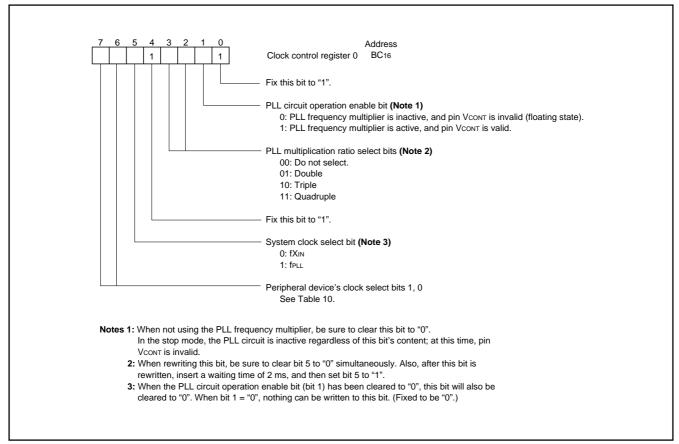


Fig. 94 Bit configuration of clock control register 0

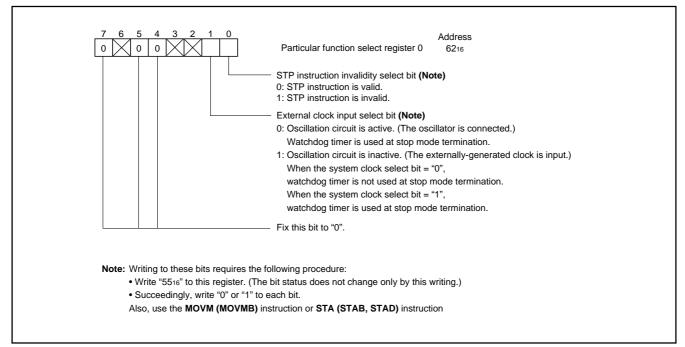


Fig. 95 Bit configuration of particular function select register 0



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STANDBY FUNCTION

The standby function provides the stop (hereafter called STP) and the wait (hereafter called WIT) mode. These modes are used to save the power dissipation of the system, by making oscillation or system clock inactive in the case that the CPU needs not be active.

The microcomputer enters the STP or WIT mode by executing the STP or WIT instruction, and either mode is terminated by acceptance of an interrupt request or reset.

To terminate the STP or WIT mode by an interrupt request, the interrupt to be used for termination of the STP or WIT mode must be enabled in advance to execution of the STP or WIT instruction. The interrupt priority level of this interrupt needs to be higher than the processor interrupt priority level (IPL) of the routine where the STP or WIT instruction will be executed.

Figures 95 shows the bit configuration of the particular function select register 0, Figure 96 shows the bit configuration of the particular function select register 1, and Figure 97 shows the bit configuration of the watchdog timer frequency select register. Setting the STP instruction invalidity select bit (bit 0 of the particular function select register 0) to "1" invalidates the STP instruction, and the STP instruction will be ignored. Since the above bit is cleared to "0" after reset is removed, however, the STP instruction is valid.

The STP- or the WIT-instruction-execution status bit (bit 0 or 1 of the particular function select register 1) is set to "1" by the execution of the STP or the WIT instruction, and so, after the STP or WIT mode has been terminated, each bit will indicate that the STP or WIT instruction has been executed. Accordingly, each of these bits must be cleared to "0" by software at termination of the STP or the WIT mode. Table 11 explains the microcomputer's operation in the STP and WIT modes.

STP mode

The execution of the STP instruction makes the oscillation circuit and PLL circuit inactive. It also makes the following inactive: input clock fXIN, system clock fsys, ϕ BIU, ϕ CPU, and peripheral devices' clocks f1 to f4096, Wf32 and Wf512 with the "L" state, and divide clocks fX16 to fX128 with the "H" state. In the watchdog timer, "FFF16" is automatically set. As shown in Figure 93, any one of divide clocks fX16 to fX128, which is selected by the watchdog timer clock source select bits at STP termination (bits 6 and 7 of the watchdog timer frequency select register), becomes the watchdog timer's clock source.

In the STP mode, the A-D converter and watchdog timer, which uses peripheral devices' clocks f1 to f4096, Wf32 and Wf512, are inactive. At this time, timers A and B can be active only in the event counter mode, and serial I/O communication is active while an external clock is selected.

The STP mode is terminated by acceptance of an interrupt request or reset, and the oscillation circuit and PLL circuit restart their operations. Input clock fXIN, system clock fsys, and peripheral devices' clocks f1 to f4096, Wf32 and Wf512 are also supplied.

When the STP mode is terminated by reset, supply of ϕBIU and ϕCPU starts immediately after the oscillation circuit and PLL circuit restart their operations. Therefore, the reset input must be raised "H" after the operation-stabilizing time for these circuits has passed.

The following two modes are available in order to terminate the STP mode by an interrupt:

- (1) The watchdog timer is used in order to measure the period from the operation restart of the oscillation circuit and PLL circuit until the supply start of ϕ BIU and ϕ CPU.
- (2) The supply of φΒΙU and φCPU is started immediately after the operation restart of the oscillation circuit and PLL circuit.

Table 11. Microcomputer's operation in STP and WIT modes

	System clock			Opera	ations of fund	tion while W	TT, STP modes
Mode	stop select bit at WIT	Oscillation circuit	PLL circuit	fsys, φ1, f1 to f4096	Wf32, Wf512	фВІИ, фСРИ	Peripheral devices using f1 to f4096, Wf32, Wf512
STP	-	Inactive	Inactive	Inactive ("L")	Inactive ("L")	Inactive ("L")	Timers A, B: Operation is enabled only in the event counter mode. Serial I/O: Operation is enabled only while an external clock is selected. A-D converter: Inactive. (Watchdog timer: Inactive)
	"0"	Active (Note 1)	Active (Note 2)	Active	Inactive ("L")	Inactive ("L")	Timers A, B, Serial I/O, A-D converter: Operation is enabled (Watchdog timer: Inactive)
WIT	"1"	Active (Note 1)	Active (Note 2)	Inactive ("L")	Inactive ("L")	Inactive ("L")	Timers A, B: Operation is enabled only in the event counter mode. Serial I/O: Operation is enabled only while an external clock is selected. A-D converter: Inactive. (Watchdog timer: Inactive)

Notes 1: When the external clock input select bit = "1", the oscillation circuit is inactive. Also, clock input from pin XIN is allowed.

2: When the PLL circuit operation enable bit = "0", the PLL circuit is inactive.





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When the external clock input select bit (bit 1 of the particular function select register 0) = "0" or the system clock select bit (bit 5 of the clock control register 0) = "1", the watchdog timer will start counting down with one of the above divide clocks, fX16 to fX128, after the oscillation circuit and PLL circuit have been restarted their operations owing to an interrupt. The most significant bit of the watchdog timer reaching "0", supply of ϕ BIU and ϕ CPU restarts.

On the other hand, when the external clock input select bit = "1" and the system clock select bit = "0", supply of ϕ BIU and ϕ CPU will restart immediately after the oscillation circuit and PLL circuit have been restarted their operations owing to an interrupt. (In actual fact, after the selected one of the above divide clocks, fX16 to fX128, has been changed from "H" to "L", this supply will restart.)

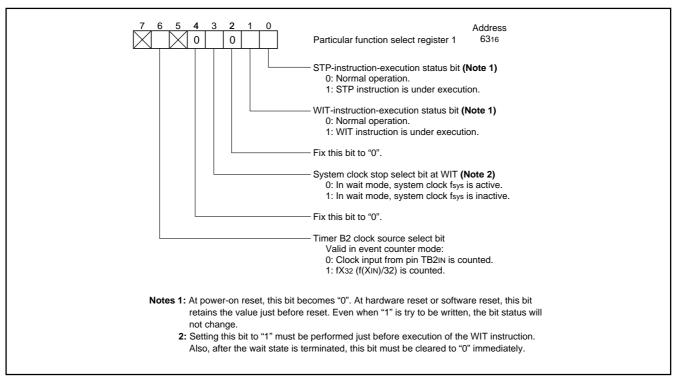


Fig. 96 Bit configuration of particular function select register 1

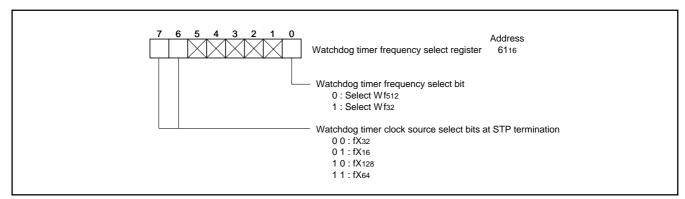


Fig. 97 Bit configuration of watchdog timer frequency select register





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WIT mode

When the WIT instruction is executed with the system clock stop select bit at WIT (bit 3 of the particular function select register 1 in Figure 93) being "0", ϕ_{BIU} , ϕ_{CPU} , and divide clocks Wf32 and Wf512 are inactive with the "L" state. However, the oscillation circuit, PLL circuit, input clock fXIN, system clock fsys, ϕ_{1} , and peripheral devices' clocks f1 to f4096 remain active. Therefore, BIU and CPU are inactive, where as timers A and B, serial I/O, and the A-D converter, which use the peripheral devices' clocks f1 to f4096, are still active. Note that the watchdog timer is inactive.

On the other hand, when the WIT instruction is executed with the system clock stop select bit at WIT being "1", the oscillation circuit, PLL circuit, and input clock fXIN are active, while system clock fsys, ϕ BIU, ϕ CPU, and peripheral devices' clocks are inactive. As a result, the A-D converter and watchdog timer, which use peripheral devices' clocks f1 to f4096, Wf32 and Wf512, become inactive. At this time, timers A and B are active only in the event counter mode, and serial I/O communication is active only while an external clock is selected. If the internal peripheral devices are not used in the WIT mode, the latter is better because the current dissipation is more saved. Note that the system clock stop select bit at WIT needs to be set to "1" immediately before execution of the WIT instruction and cleared to "0" immediately after the WIT mode is terminated.

The WIT state is terminated by acceptance of an interrupt request, and then, supply of ϕBIU and ϕCPU will restart. Since the oscillation circuit, PLL circuit, and clock input fXIN are active in the WIT mode, an interrupt processing can be executed just after the WIT mode termination.





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POWER SAVING FUNCTION

The following functions can save the power dissipation of the whole system.

(1) Inactive system clock in wait mode

In the wait mode, if the internal peripheral devices need not to operate, when the system clock stop select bit at WIT (bit 3 of the particular function select register 1) = "1", both of system clock fsys and peripheral devices' clock are inactive, and the power dissipation can be saved

For details, refer to the section on the standby function.

(2) Inactive oscillation circuit

When an externally-generated stable clock is input to pin XIN, the power dissipation can be saved if both of the following conditions are met:

- the external clock input select bit (bit 1 of the particular function select register 0) = "1".
- the oscillation driver circuit between pins XIN and XOUT is inactive. At this time, the output level at pin XOUT is fixed to "H". When not using fPLL, also, the supply of ϕ BIU and ϕ CPU restarts just after the microcomputer returns from the stop mode, owing to an interrupt request occurrence. Therefore, an instruction can be executed just after the termination of the stop mode. For details, refer to the section on the clock generating circuit and standby function.

(3) Disconnection from pin VREF

When not using the A-D converter, by setting the VREF connection select bit (bit 6 of the A-D control register 1) to "1", the resistor ladder network of the A-D converter will be disconnected from the reference voltage input pin (VREF). In this case, no current flows from pin VREF to the resistor ladder network, and the power dissipation can be saved. Note that, after the VREF connection select bit changes from "1" (VREF disconnected) to "0" (VREF connected), be sure that a waiting time of 1 μs or more has passed before the A-D conversion starts. For details, refer to the section on the A-D converter.





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DEBUG FUNCTION

When the CPU fetches an instruction code, an interrupt request will be generated if a selected condition is satisfied, as a resultant of comparison between a specified address and the start address where the instruction code is stored (the contents of PG and PC). The decision whether this condition is satisfied or not is called address matching detection, and the interrupt generated by this detection is called an address matching detection interrupt. (For interrupt vector addresses, refer to the section on interrupts.)

In the address matching detection, a non-maskable interrupt routine is proceeded without execution of the original instruction which has been allocated to the target address.

The debug function provides the following two modes:

- the address matching detection mode, which is used to avoid the area where program exists or modify a program.
- the out-of-address-area detection mode, which is used to detect a program runaway.

Figure 98 shows the block diagram of the debug function. Figures 99 and 100 show the bit configurations of the debug control registers 0, 1, and address compare registers 0,1, respectively.

The detect condition select bits of the debug control register 0 can select one condition between the following 4 conditions. When the selected address condition is satisfied, an address matching detection interrupt request will be generated:

(1) Address matching detection 0

The contents of PG and PC match with the address which has been set in the address compare register 0.

(2) Address matching detection 1

The contents of PG and PC match with the address which has been set in the address compare register 1.

(3) Address matching detection 2

The contents of PG and PC match with the address which has been set in either of the address compare register 0 or address compare register 1.

(4) Out-of-address-area detection

The contents of PG and PC are less than the address which has been set in the address compare register 0 or larger than the address which has been set in the address compare register 1. By setting the detect enable bit of the debug control register 0 to "1", an address matching detection interrupt request will be generated if any one of the above address conditions is satisfied. Clearing the detect enable bit to "0" generates no interrupt request even if any of the above address conditions is satisfied.

The address compare register access enable bit of the debug control register 1 must be set to "1" by the instruction just before the access operation (read/write). Then, this bit must be cleared to "0" (disabled) by the next instruction. While this bit = "0", the address compare registers 0, 1 cannot be accessed.

The address-matching-detection 2 decision bit of the debug control register 1 decides, whether the address which has been set in the address compare register 0 or 1 matches with the contents of PG, PC, when the address matching detection 2 is selected. The contents of this bit is invalid when address matching detection 0 or 1 is selected.

In order to use the debug function to avoid the area where program exists or modify a program, perform the necessary processing within an address matching interrupt routine. As a result, the contents of PG, PC, PS at acceptance of an address matching detection interrupt request (i.e. the address at which an address matching detection condition is satisfied) have been pushed onto the stack. If a return destination address after the interrupt processing is to be altered, rewrite the contents of the stack, and then return by the RTI instruction

To use the debug function to detect a program runaway, set an address area where no program exists into the address compare registers 0 and 1 by using the out-of-address-area detection. When the CPU fetches instruction codes from this address area and executes them, an address matching detection interrupt request will be generated.

The above debug function cannot be evaluated by a debugger, so that the debug function must not be used while a debugger is running.

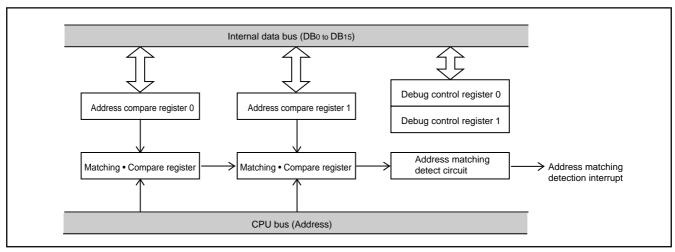


Fig. 98 Block diagram of debug function





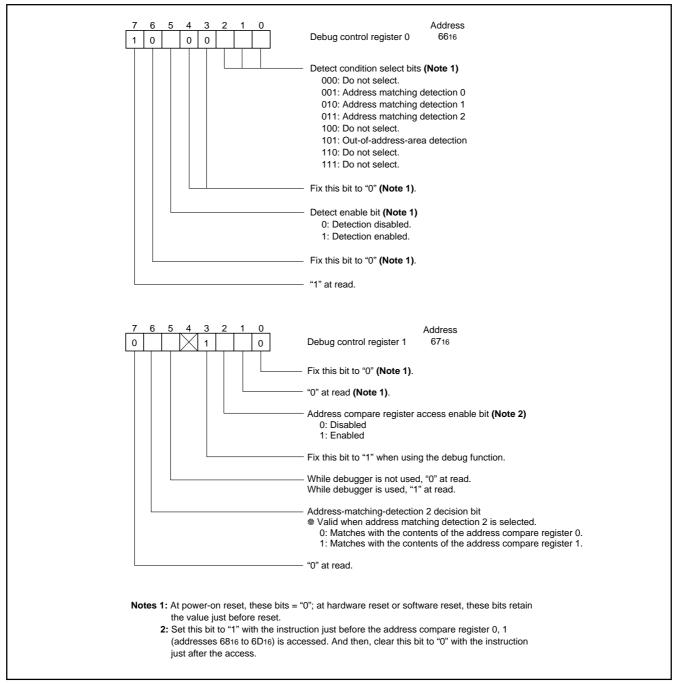


Fig. 99 Bit configuration of debug control register 0, 1

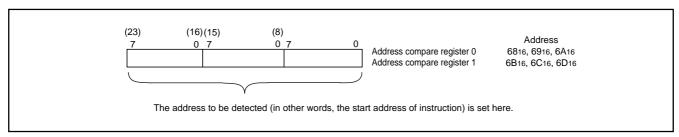


Fig. 100 Bit configuration of address compare register 0, 1



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ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Ratings	Unit
Vcc	Power source voltage	-0.3 to 6.5	V
AVcc	Analog power source voltage	-0.3 to 6.5	V
VI	Input voltage P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83, P40UTcuT, P60UTcuT, Vcont, Vref, XIN, RESET, MD0, MD1	-0.3 to Vcc+0.3	V
Vo	Output voltage P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83, XOUT	-0.3 to Vcc+0.3	V
Pd	Power dissipation	300	mW
Topr	Operating ambient temperature	-20 to 85	°C
Tstg	Storage temperature	-40 to 150	°C

RECOMMENDED OPERATING CONDITIONS (Vcc = 5 V, Ta = -20 to 85 °C, unless otherwise noted)

	_		Limits		
Symbol	Parameter		Тур.	Max.	Unit
Vcc	Power source voltage	4.5	5.0	5.5	V
AVcc	Analog power source voltage		Vcc		V
Vss	Power source voltage		0		V
AVss	Analog power source voltage		0		V
VIH	High-level Input voltage P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83, P40UTcut, P60UTcut, XIN, RESET, MD0, MD1	0.8 Vcc		Vcc	V
VIL	Low-level Input voltage P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83, P40UTcut, P60UTcut, XIN, RESET, MD0, MD1	0		0.2 VCC	V
IOH(peak)	High-level peak output current P10–P17, P20–P27, P55–P57, P60–P67, P70–P77			-10	mA
IOH(avg)	High-level average output current P10–P17, P20–P27, P55–P57, P60–P67, P70–P77			-5	mA
IOL(peak)	Low-level peak output current P10–P17, P20–P27, P51–P53, P55–P57, P70–P77			10	mA
IOL(peak)	Low-level peak output current P40–P47, P60–P67			20	mA
IOL(avg)	Low-level average output current P10–P17, P20–P27, P51–P53, P55–P57, P70–P77			5	mA
IOL(avg)	Low-level average output current P40-P47, P60-P67			15	mA
f(XIN)	External clock input frequency (Note 1)			20	MHz
f(fsys)	System clock frequency			20	MHz

Notes 1: When using the PLL frequency multiplier, be sure that $f(f_{sys}) = 20$ MHz or less.



^{2:} The average output current is the average value of an interval of 100 ms.

^{3:} The sum of IOL(peak) must be 110 mA or less, the sum of IOH(peak) must be 80 mA or less.



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DC ELECTRICAL CHARACTERISTICS (Vcc = 5 V, Vss = 0 V, Ta = -20 to 85 °C, f(fsys) = 20 MHz, unless otherwise noted)

Symbol	Parameter		Test conditions		Limits			Unit
Symbol			iesi (rest conditions		Тур.	Max.	Unit
Voh	High-level output voltag	ge P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83	IOH = -10 mA		3			V
VoL	Low-level output voltag	e P10–P17, P20–P27, P40–P47, P51–P53, P55–P57, P60–P67, P70–P77, P80–P83	IOL = 10 mA				2	V
VT+ —VT-	TB0IN-TE CTS2, CL	N9IN, TA0OUT—TA9OUT, B2IN, INTO—INT7, CTS0, CTS1, LK0, CLK1, CLK2, RxD0, RxD1, PTRG0, RTPTRG1, P4OUTCUT, UT			0.4		1	V
VT+ —VT-	Hysteresis RESET				0.5		1.5	V
VT+ —VT-	Hysteresis XIN				0.1		0.3	V
Іін	High-level input current	P10-P17, P20-P27, P40-P47, P51-P53, P55-P57, P60-P67, P70-P77, P80-P83, P4OUTCUT, P6OUTCUT, XIN, RESET, MD0, MD1	VI = 5.0 V				5	μΑ
liL	Low-level input current	P10-P17, P20-P27, P40-P47, P51-P53, P55-P57, P60-P67, P70-P77, P80-P83, P40UTCUT, P6OUTCUT, XIN, RESET, MD0, MD1	VI = 0 V				-5	μΑ
VRAM	RAM hold voltage		When clock is in	nactive.	2			V
Icc	Power source current		Output-only pins are open, and the other pins are con- nected to Vss or			25	50	mA
			Vcc. An external square-waveform clock is input. (Pin Xout is open.) The				1	μA
			PLL frequency multiplier is inactive.	Ta = 85 °C when clock is inactive.			20	





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A-D CONVERTER CHARACTERISTICS

(VCC = AVCC = 5 V ± 0.5 V, Vss = AVss = 0 V, Ta = -20 to 85 °C, unless otherwise noted)

	Б.,	To do so Pilos			Limits		
Symbol	Parameter Test conditions		Min.	Тур.	Max.	Unit	
	Resolution	VREF = VCC	A-D converter			10	Bits
			Comparator			1 256 VREF	V
			10-bit resolution mode			± 3	LSB
	Absolute accuracy	VREF = VCC	8-bit resolution mode			± 2	LSB
			Comparater			± 40	mV
RLADDER	Ladder resistance	VREF = VCC		5			kΩ
			10-bit resolution mode	5.9			μs
tCONV	Conversion time	f(fsys) ≤ 20 MHz	8-bit resolution mode	2.45 (Note)			
			Comparater	0.7 (Note)			
VREF	Reference voltage			2.7		Vcc	V
VIA	Analog input voltage			0		VREF	V

Note: This is applied when A-D conversion freguency $(\phi AD) = f1 (\phi)$.

D-A CONVERTER CHARACTERISTICS

(VCC = 5 V, VSS = AVSS = 0 V, VREF = 5 V, Ta = -20 to 85 °C, unless otherwise noted)

Symbol	Deservator	Test conditions		1.121		
	Parameter		Min.	Тур.	Max.	Unit
	Resolution				8	Bits
	Absolute accuracy				± 1.0	%
tsu	Set time				3	μs
Ro	Output resistance		2	3.5	4.5	kΩ
IVREF	Reference power source input current	(Note)			3.2	mA

Note: The test conditions are as follows:

- One D-A converter is used.
- The D-A register value of the unused D-A converter is "0016."
- The reference power source input current for the ladder resistance of the A-D converter is excluded.

RESET INPUT

Reset input timing requirements (Vcc = 5 V \pm 0.5 V, Vss = 0V, Ta = -20 to 85 °C, unless otherwise noted)

Courselle sel	Paramotor		Limits			
Symbol	Symbol Parameter	Min.	Тур.	Max.	Unit	
tw(RESETL)	RESET input low-level pulse width	10			μs	







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PERIPHERAL DEVICE INPUT/OUTPUT TIMING

(VCC = 5 V±0.5 V, VSS = 0 V, Ta = -20 to 85 °C, f(fsys) = 20 MHz unless otherwise noted)

* For limits depending on f(fsys), their calculation formulas are shown below. Also, the values at f(fsys) = 20 MHz are shown in ().

Timer A input (Count input in event counter mode)

Symbol	Parameter		Limits		
			Max.	Unit	
tc(TA)	TAil input cycle time	80		ns	
tw(TAH)	TAin input high-level pulse width	40		ns	
tw(TAL)	TAin input low-level pulse width	40		ns	

Timer A input (Gating input in timer mode)

Currely al	Devenuetes		Lin	nits	Linit
Symbol	Parameter	Min.	Max.	Unit	
tc(TA)	TAilN input cycle time	f(fsys) ≤ 20 MHz	$\frac{16 \times 10^9}{\text{f(fsys)}} (800)$		ns
tw(TAH)	TAilN input high-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns
tw(TAL)	TAilN input low-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns

Note : The TAilN input cycle time requires 4 or more cycles of a count source. The TAilN input high-level pulse width and the TAilN input low-level pulse width respectively require 2 or more cycles of a count source. The limits in this table are applied when the count source = f2 at f(fsys) ≤ 20 MHz.

Timer A input (External trigger input in one-shot pulse mode)

Course and	Deservator		Lin	nits	I lait
Symbol	Parameter		Min.	Max.	Unit
tc(TA)	TAilN input cycle time	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns
tw(TAH)	TAilN input high-level pulse width		80		ns
tw(TAL)	TAilN input low-level pulse width		80		ns

Timer A input (External trigger input in pulse width modulation mode)

Symbol	Demonstra		Limits		
	Parameter	Min.	Max.	Unit	
tw(TAH)	TAin input high-level pulse width	80		ns	
tw(TAL)	TAin input low-level pulse width	80		ns	

Timer A input (Up-down input and Count input in event counter mode)

Symbol	Parameter	Lim	1.1	
		Min.	Max.	Unit
tc(UP)	TAiout input cycle time	2000		ns
tw(UPH)	TAiout input high-level pulse width	1000		ns
tw(UPL)	TAiout input low-level pulse width	1000		ns
tsu(UP-Tin)	TAiout input setup time	400		ns
th(TIN-UP)	TAiout input hold time	400		ns



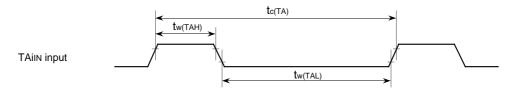


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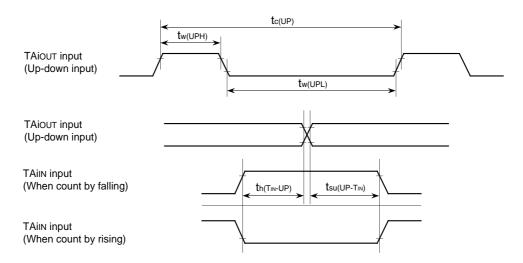
Timer A input (Two-phase pulse input in event counter mode)

Symbol	Dans and an		Limits		
	Parameter	Min.	Max.	Unit	
tc(TA)	TAjın input cycle time	800		ns	
tsu(TAjin-TAjout)	TAjın input setup time	200		ns	
tsu(TAjout-TAjin)	TAjout input setup time	200		ns	

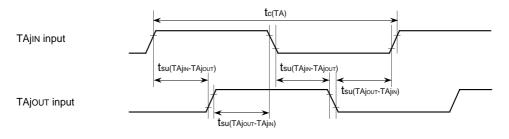
- · Gating input in timer mode
- Count input in event counter mode
- External trigger input in one-shot pulse mode
- External trigger input in pulse width modulation mode



• Up-down and Count input in event counter mode



• Two-phase pulse input in event counter mode



Test conditions

- \bullet Vcc = 5 V \pm 0.5 V, Ta = –20 to 85 $^{\circ}\text{C}$
- Input timing voltage: VIL = 1.0 V, VIH = 4.0 V



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Timer B input (Count input in event counter mode)

0	Parameter	Lin	1.1-21	
Symbol		Min.	Max.	Unit
tc(TB)	TBilN input cycle time (one edge count)	80		ns
tw(TBH)	TBilN input high-level pulse width (one edge count)	40		ns
tw(TBL)	TBiln input low-level pulse width (one edge count)	40		ns
tc(TB)	TBiln input cycle time (both edge count)	160		ns
tw(TBH)	TBilN input high-level pulse width (both edge count)	80		ns
tw(TBL)	TBiln input low-level pulse width (both edge count)	80		ns

Timer B input (Pulse period measurement mode)

Corrects at	Parameter		Lin	I India	
Symbol			Min.	Max.	Unit
tc(TB)	TBilN input cycle time	f(fsys) ≤ 20 MHz	$\frac{16 \times 10^9}{\text{f(fsys)}} (800)$		ns
tw(TBH)	TBilN input high-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns
tw(TBL)	TBilN input low-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns

Note: The TBin input cycle time requires 4 or more cycles of a count source. The TBin input high-level pulse width and the TBin input low-level pulse width respectively require 2 or more cycles of a count source. The limits in this table are applied when the count source = f2 at f(fsys) ≤ 20 MHz.

Timer B input (Pulse width measurement mode)

O make at	Symbol Parameter		Lin	1.111	
Symbol			Min.	Max.	Unit
tc(TB)	TBilN input cycle time	f(fsys) ≤ 20 MHz	$\frac{16 \times 10^9}{\text{f(fsys)}} (800)$		ns
tw(TBH)	TBilN input high-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns
tw(TBL)	TBilN input low-level pulse width	f(fsys) ≤ 20 MHz	$\frac{8 \times 10^9}{\text{f(fsys)}} (400)$		ns

Note: The TBiIN input cycle time requires 4 or more cycles of a count source. The TBiIN input high-level pulse width and the TBiIN input low-level pulse width respectively require 2 or more cycles of a count source. The limits in this table are applied when the count source = f_2 at $f(f_{sys}) \le 20$ MHz.

Serial I/O

O l		Lin		
Symbol	Parameter	Min.	Max.	Unit
tc(CK)	CLKi input cycle time	200		ns
tw(CKH)	CLKi input high-level pulse width	100		ns
tw(CKL)	CLKi input low-level pulse width	100		ns
td(C-Q)	TxDi output delay time		80	ns
th(C-Q)	TxDi hold time	0		ns
tsu(D-C)	RxDi input setup time	20		ns
th(C-D)	RxDi input hold time	90		ns

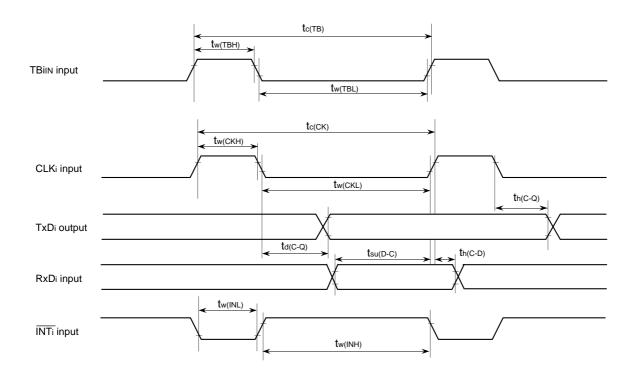




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External interrupt (INTi) input

0	Symbol Parameter	Lim	1.1.29	
Symbol		Min.	Max.	Unit
tw(INH)	ĪNTi input high-level pulse width	250		ns
tw(INL)	ĪNTi input low-level pulse width	250		ns



Test conditions

 \bullet Vcc = 5 V \pm 0.5 V, Ta = –20 to 85 $^{\circ}C$

• Input timing voltage : VIL = 1.0 V, VIH = 4.0 V

• Output timing voltage : VoL = 0.8 V, VoH = 2.0 V, CL = 50 pF



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External clock input

Timing Requirements (VCC = 5 V±0.5 V, Vss = 0 V, Ta = -20 to 85 °C, f(XIN) = 20 MHz, unless otherwise noted)

Symbol	Parameter	Lir	1.121	
		Min.	Max.	Unit
tc	External clock input cycle time	50		ns
tw(half)	External clock input pulse width with half input-voltage	0.45 tc	0.55 tc	ns
tw(H)	External clock input high-level pulse width	0.5 tc - 8		ns
tw(L)	External clock input low-level pulse width	0.5 tc - 8		ns
tr	External clock input rise time		8	ns
tf	External clock input fall time		8	ns

External clock input



Test conditions

- Vcc = 5 V \pm 0.5 V, Ta = -20 to 85 °C
- Input timing voltage : VIL = 1.0 V, VIH = 4.0 V ($t_{W(H)}$, $t_{W(L)}$, t_r , t_f)
- Output timing voltage : 2.5 V (tc, tw(half))

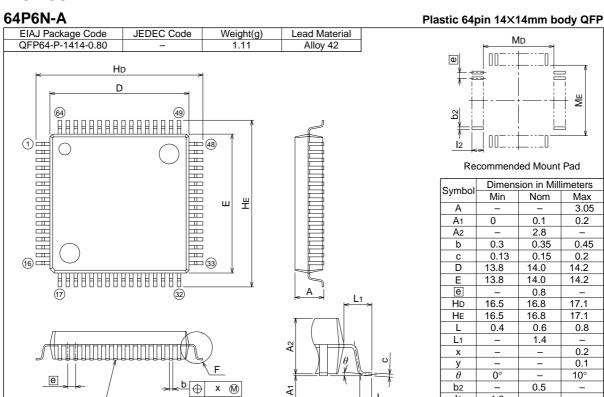


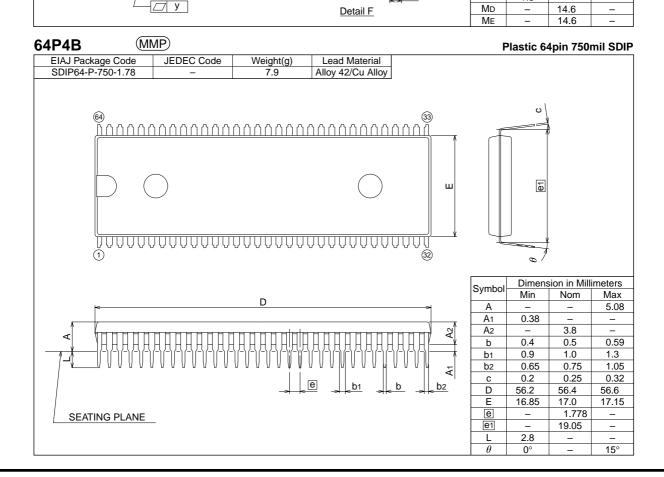
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1.3

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PACKAGE OUTLINE





Notice: This is not a final specification change. Some parametric limits are subject to change.

M37905M4C-XXXFP, M37905M4C-XXXSP M37905M6C-XXXFP, M37905M6C-XXXSP M37905M8C-XXXFP, M37905M8C-XXXSP

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Rev. No.	Revision Description	Rev. date
1.0	First Edition. The following are released. • DESCRIPTION	000630
2.0	 Revised Points Refer to "Corrections and Supplementary Explanation for M37905MxC-XXXFP/SP Datasheet (Rev.A)". Added Sections BLOCK DIAGRAM, BASIC FUNCTION BLOCKS, MEMORY, CENTRAL PROCESSING UNIT (CPU), BUS INTERFACE UNIT, PROCESSOR MODES, INTERRUPTS, TIMER, TIMER FUNCTION FOR MOTOR CONTROL, PULSE OUTPUT PORT MODE 0/1, SERIAL I/O PORTS, A-D CONVERTER, D-A CONVERTER, WATCHDOG TIMER, INPUT/OUTPUT PINS, RESET CIRCUIT, OSCILLATION CIRCUIT, CLOCK GENERATING CIRCUIT, STANDBY FUNCTION, POWER SAVING FUNCTION, DEBUG FUNCTION, ELECTRICAL CHARACTERISTICS, PACKAGE OUTLINE 	010301
3.0	Refer to "Corrections and Explanation for M37905MxC-XXXFP/SP Datasheet (Rev.B)". Note: ★ represents the new information added in Rev.3.0.	010702

Corrections and Supplementary Explanation for M37905MxC-XXXFP/SP Datasheet (Rev.B) No.1

	Page	Error	Correction							
	Page 1, DISTINCTIVE FEATURES	8-bit <u>A-D</u> converter	8-bit <u>D-A</u> converter							
	Page 1, APPLICATION	Control devices for equipment required for motor control such as inverter	Control devices for equipment, requiring motor control, such as inverter •••••							
	Page 1, M37905MxC-XXXFP PIN CONFIGURA- TION (TOP VIEW)	 ◆ Pin No.23; P5₃/INT₃/RTPTRG0/XCOUT ◆ Pin No.24; P5₂/INT₂/RTPTRG1/XCIN ◆ Pin No.50; P1₁/CTS0/CLK0 	 Pin No.23; P5₃/INT₃/RTPTRG0 Pin No.24; P5₂/INT₂/RTPTRG1 Pin No.50; P1₁/CTSo/CLKo 							
	Page 2, M37905MxC-XXXSP PIN CONFIGURA- TION (TOP VIEW)	 ◆ Pin No.28; X_{ONT} ◆ Pin No.31; P5₃/INT₃/RTP_{TRG0}/X_{COUT} ◆ Pin No.32; P5₂/INT₂/RTP_{TRG1}/X_{CIN} 	 ◆ Pin No.28; Xout ◆ Pin No.31; P5₃/INT₃/RTPTRG0 ◆ Pin No.32; P5₂/INT₂/RTPTRG1 							
	Page 4, Item 1	External main-clock input frequency f(XIN)	External clock input frequency f(XIN) * "External sub-clock input frequency f(XIN)" is deleted.							
	Item 2 Item 3: Clock generating circuit	System clock <u>input</u> frequency f(f _{sys}) <u>2 circuits</u> incorporated	System clock frequency f(fsys) Incorporated							
	Item 4: Power dissipation	•••• (at $\underline{f(X_{IN})} = 20 \text{ MHz}, ••••)$	•••• (at <u>f(fsys)</u> = 20 MHz, ••••)							
*	Page 43, Figure 43	Position-data-retain function control register Retained trigger's polarity select bit	Position-data-retain function control register Retain-trigger polarity select bit							
*	Page 70, Figure 79	3: ••••, it is necessary to wait for <u>1 s</u> or more before ••••	3: ••••, it is necessary to wait for <u>1 μs</u> or more before ••••							
*	Page 78, Figure 88	<u>Comparison</u> result register 0 (DE16)••• <u>Comparison</u> result register 1 (DF16)•••	Comparator result register 0 (DE16)••• Comparator result register 1 (DF16)•••							
*	Page 81, • Left column Line 7 • Right column Lines 5, 10, 21	•••• the clock control register ••••	•••• the clock control register <u>0</u> ••••							
*	Page 90, RECOMMENDED OPERATING CONDITIONS	Symbol Parameter Limits Unit Min. Typ. Max. VIL Low-level input voltage 0 0.2 Vcc VmA	Symbol Parameter Limits Min. Typ. Max. VIL Low-level input voltage 0 0.2 Vcc V							
*	Page 91, DC ELECTRICAL CHARACTERISTICS	(VCC = 5V, •••••, f(fsys) = 20 MHz)	(Vcc = 5V, ••••, f(fsys) = 20 MHz, unless otherwise noted)							

Corrections and Supplementary Explanation for M37905MxC-XXXFP/SP Datasheet (Rev.B) No.2

	Page	Error						Correction										
*	Page 91,	Symbol Parameter				Limits			Unit	Symbol	Par	arameter		Limits Unit				
	DC ELECTRICAL CHARACTERISTICS	Symbol S	<i>y</i>		**		Гур.	Max.	<u> </u>	<u>- </u>	<u> </u>		≈ <i>≈</i>	Min.	Typ. I	Max.	_	
		Vol	Low-level output voltage	P10-P17, P70-P74,	•••,			2	٧	VoL	Low-level output voltage	P10-P17, P70-P7 <u>7</u> ,	•••,			2	V	
		i											İ		1			
		Symbol Parameter Test co			nditions		imits	Max.	Unit	Symbol	Parameter	Test coi	nditions		Limits Typ. I		Unit	
		* * *				=		8	*						│			
		Vol	ROM hold voltage	When clo	ock ••••	2		50	٧	Vol	ROM hold voltage	When cl	ock ••••	2			V	
		Symbol Parameter Test conditions Limits Unit					Unit	Symbol Parameter Test conditions Limits Uni							— Unit			
		, , , , , , , , , , , , , , , , , , ,				¥ IVIIN.	тур.	wax.	<u> </u>		 			*	Min. Typ. Max.		_	
		Icc	Power source current		f(fsys) = 20 MHz.		25		mA	Icc	Power source current	Output- only pins	f(fsys) = 20 MHz		25 (50	mA	
									_					+	\Box		_	
									_								_	