



## FAN4174 / FAN4274

### Single and Dual, Rail-to-Rail I/O, CMOS Amplifier

#### Features

- 200  $\mu$ A Supply Current per Amplifier
- 3.7 MHz Bandwidth
- Output Swing to Within 10 mV of Either Rail
- Input Voltage Range Exceeds the Rails
- 3 V /  $\mu$ s Slew Rate
- 25 nV /  $\sqrt{\text{Hz}}$  Input Voltage Noise
- Replaces KM4170 and KM4270
- FAN4174 Competes with OPA340 and TLV2461; Available in a SOT23-5 Package
- FAN4274 Competes with OPA2340 and TLV2462; Available in MSOP-8 Package
- Fully Specified at +5 V Supplies

#### Applications

- Motor Control
- Portable / Battery-powered Applications
- PCMCIA, USB
- Mobile Communications, Cellular Phones, Pagers
- Notebooks and PDAs
- Sensor Interface
- A/D Buffer
- Active Filters
- Signal Conditioning
- Portable Test Instruments

#### Description

The FAN4174 (single) and FAN4274 (dual) are voltage feedback amplifiers with CMOS inputs that consume only 200  $\mu$ A of supply current per amplifier, while providing  $\pm 33$  mA of output short-circuit current. These amplifiers are designed to operate 5 V supplies. The common mode voltage range extends beyond the negative and positive rails.

The FAN4174 and FAN4274 are designed on a CMOS process and provide 3.7 MHz of bandwidth and 3 V /  $\mu$ s of slew rate at a supply voltage of 5 V.

These amplifiers operate and are reliable over a wide temperature range of -40°C to +125°C.

The combination of extended temperature operation, low power, rail-to-rail performance, low-voltage operation, and a tiny package optimize this amplifier family for use in many industrial, general-purpose, and battery-powered applications.

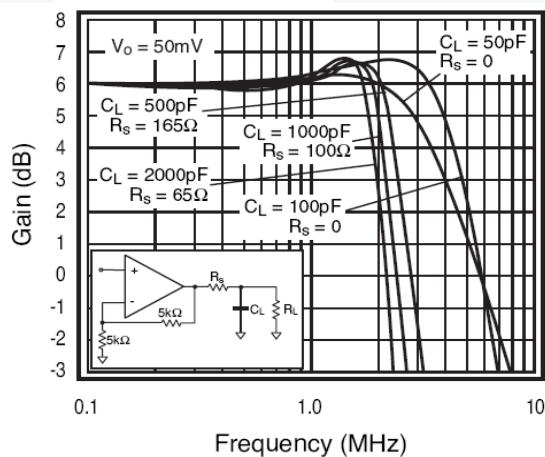


Figure 1. Frequency vs. Gain

#### Ordering Information

Part Number	Operating Temperature Range	Package	Packing Method
FAN4174IS5X	-40 to +125°C	5-Lead SOT23 Package	Tape and Reel (3000)
FAN4274IMU8X	-40 to +125°C	8-Lead Molded Small-Outline Package	Tape and Reel (3000)

## Typical Application

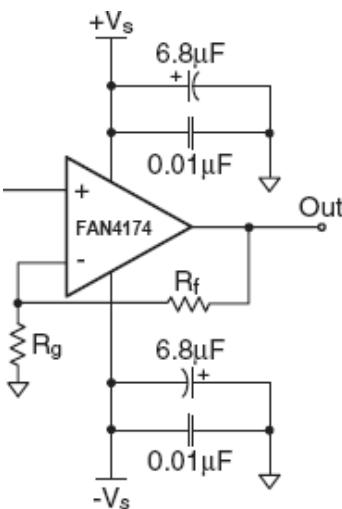


Figure 2. Typical Application Circuit

## Pin Configurations

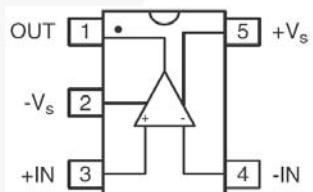


Figure 3. FAN4174 SOT23

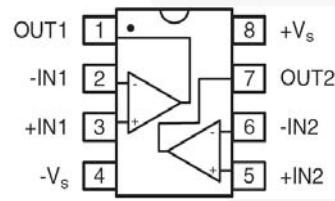


Figure 4. FAN4274 MSOP

## FAN4174 Pin Assignments

Pin #	Name	Description
1	OUT	Output
2	-Vs	Negative Supply
3	+IN	Positive Supply
4	-IN	Negative Input
5	+Vs	Positive Supply

## FAN4274 Pin Assignments

Pin #	Name	Description
1	OUT1	Output, Channel 1
2	-IN1	Negative Input, Channel 1
3	+IN1	Positive Input, Channel 1
4	-Vs	Negative Supply
5	+IN2	Positive Input, Channel 2
6	-IN2	Negative Input, Channel 2
7	OUT2	Output, Channel 2
8	+Vs	Positive Supply

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. Functional operation under any of these conditions is NOT implied. Performance and reliability are guaranteed only if operating conditions are not exceeded.

Symbol	Parameter		Min.	Max.	Unit
$V_{CC}$	Supply Voltage		0	6	V
$V_{IN}$	Input Voltage Range		$-V_S - 0.5$	$+V_S + 0.5$	V
$T_J$	Junction Temperature			+150	°C
$T_{STG}$	Storage Temperature		-65	+150	°C
$T_L$	Lead Soldering, 10 Seconds			+300	°C
$\Theta_{JA}$	Thermal Resistance <sup>(1)</sup>	5-Lead SOT23 8-Lead MSOP		256 206	°C/W

**Note:**

1. Package thermal resistance JEDEC standard, multi-layer test boards, still air.

## Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter		Min.	Max.	Unit
$+V_S$	Supply Voltage		2.30	5.25	V
$T_A$	Operating Temperature Range		-40	+125	°C

## Electrical Specifications at +2.7 V

$V_S=+2.7$  V,  $G=2$ ,  $R_L=10$  k $\Omega$  to  $V_S/2$ ,  $R_F=5$  k $\Omega$ ; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
<b>Frequency Domain Response</b>						
UGBW	-3 dB Bandwidth	G=+1		4.0		MHz
BW <sub>SS</sub>				2.5		MHz
GBWP	Gain Bandwidth Product			4		MHz
<b>Time Domain Response</b>						
t <sub>R</sub> , t <sub>F</sub>	Rise and Fall Time	V <sub>O</sub> =1.0 V Step		300		ns
OS	Overshoot	V <sub>O</sub> =1.0 V Step		5		%
SR	Slew Rate	V <sub>O</sub> =3 V Step, G=-1		3		V/ $\mu$ s
<b>Distortion and Noise Response</b>						
HD2	2nd Harmonic Distortion	V <sub>O</sub> =1 V <sub>PP</sub> , 10 kHz		-66		dBc
HD3	3rd Harmonic Distortion	V <sub>O</sub> =1 V <sub>PP</sub> , 10 kHz		-67		dBc
THD	Total Harmonic Distortion	V <sub>O</sub> =1 V <sub>PP</sub> , 10 kHz		0.1		%
e <sub>n</sub>	Input Voltage Noise			26		nV/ $\sqrt$ Hz
X <sub>TALK</sub>	Crosstalk (FAN4274)	100 kHz		-100		dB
<b>DC Performance</b>						
V <sub>IO</sub>	Input Offset Voltage <sup>(2)</sup>		-6	0	+6	mV
dV <sub>IO</sub>	Average Drift			2.1		$\mu$ V/ $^{\circ}$ C
I <sub>bn</sub>	Input Bias Current			5		pA
PSRR	Power Supply Rejection Ratio <sup>(2)</sup>	DC	50	73		dB
A <sub>OL</sub>	Open-loop Gain	DC		98		dB
I <sub>S</sub>	Supply Current per Amplifier <sup>(2)</sup>			200	300	$\mu$ A
<b>Input Characteristics</b>						
R <sub>IN</sub>	Input Resistance			10		G $\Omega$
C <sub>IN</sub>	Input Capacitance			1.4		pF
CMIR	Input Common Mode Voltage Range			-0.3 to 2.8		V
CMRR	Common Mode Rejection Ratio <sup>(2)</sup>	FAN4174	DC, V <sub>CM</sub> =0 V to 2.2 V	50	65	dB
		FAN4274	DC, V <sub>CM</sub> =0 V to 2.2 V	50	65	
<b>Output Characteristics</b>						
V <sub>O</sub>	Output Voltage Swing <sup>(2)</sup>	R <sub>L</sub> =10 k $\Omega$ to $V_S/2$	0.03	0.01 to 2.69	2.65	V
		R <sub>L</sub> =1 k $\Omega$ to $V_S/2$		0.05 to 2.55		
I <sub>SC</sub>	Short-Circuit Output Current			+34/-12		mA
V <sub>S</sub>	Power Supply Operating Range			2.5 to 5.5		V

### Note:

2. 100% tested at 25 $^{\circ}$ C.

## Electrical Specifications at +5 V

$V_S=+5$  V,  $G=2$ ,  $R_L=10$  k $\Omega$  to  $V_S/2$ ,  $R_F=5$  k $\Omega$ ; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
<b>Frequency Domain Response</b>						
UGBW	-3dB Bandwidth	$G=+1$ , $T_A=85^\circ\text{C}$		3.7		MHz
		$G=+1$ , $T_A=125^\circ\text{C}$		3.2		
BW <sub>SS</sub>				2.3		MHz
		$T_A=85^\circ\text{C}$		3.7		
GBWP	Gain Bandwidth Product	$T_A=125^\circ\text{C}$		3.2		MHz
<b>Time Domain Response</b>						
t <sub>R</sub> , f <sub>F</sub>	Rise and Fall Time	$V_O=1.0$ V Step		300		ns
OS	Overshoot	$V_O=1.0$ V Step		5		%
SR	Slew Rate	$V_O=3$ V Step, $G=-1$		3		V/ $\mu$ s
<b>Distortion and Noise Response</b>						
HD2	2nd Harmonic Distortion	$V_O=1$ V <sub>PP</sub> , 10 kHz		-80		dBc
HD3	3rd Harmonic Distortion	$V_O=1$ V <sub>PP</sub> , 10 kHz		-80		dBc
THD	Total Harmonic Distortion	$V_O=1$ V <sub>PP</sub> , 10 kHz		0.02		%
e <sub>n</sub>	Input Voltage Noise			25		nV/ $\sqrt{\text{Hz}}$
X <sub>TALK</sub>	Crosstalk (FAN4274)	100 kHz		-100		dB
<b>DC Performance</b>						
V <sub>IO</sub>	Input Offset Voltage <sup>(3)</sup>		-8	0	+8	mV
dV <sub>IO</sub>	Average Drift			2.9		$\mu\text{V}/^\circ\text{C}$
I <sub>bn</sub>	Input Bias Current			5		pA
PSRR	Power Supply Rejection Ratio <sup>(3)</sup>	DC	50	73		dB
A <sub>OL</sub>	Open-loop Gain	DC		102		dB
I <sub>S</sub>	Supply Current per Amplifier <sup>(3)</sup>		200	300		$\mu\text{A}$
<b>Input Characteristics</b>						
R <sub>IN</sub>	Input Resistance			10		G $\Omega$
C <sub>IN</sub>	Input Capacitance			1.2		pF
CMIR	Input Common Mode Voltage Range			-0.3 to 5.1		V
CMRR	Common Mode Rejection Ratio <sup>(3)</sup>	DC, $V_{CM}=0$ V to $V_S$	58	73		dB
<b>Output Characteristics</b>						
V <sub>O</sub>	Output Voltage Swing <sup>(3)</sup>	$R_L=10$ k $\Omega$ to $V_S/2$	0.03	0.01 to 4.99	4.95	V
		$R_L=1$ k $\Omega$ to $V_S/2$		0.1 to 4.9		
I <sub>SC</sub>	Short-Circuit Output Current			±33		mA
V <sub>S</sub>	Power Supply Operating Range			2.5 to 5.5		V

**Note:**

3. 100% tested at 25°C.

## Typical Performance Characteristics

$V_S = +2.7$  V,  $G = 2$ ,  $R_L = 10$  k $\Omega$  to  $V_S/2$ ,  $R_F = 5$  k $\Omega$ ; unless otherwise noted.

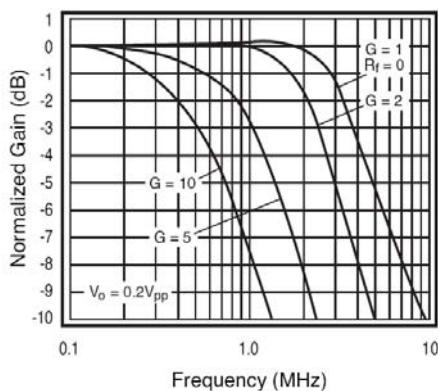


Figure 5. Non-Inverting Frequency Response (+5 V)

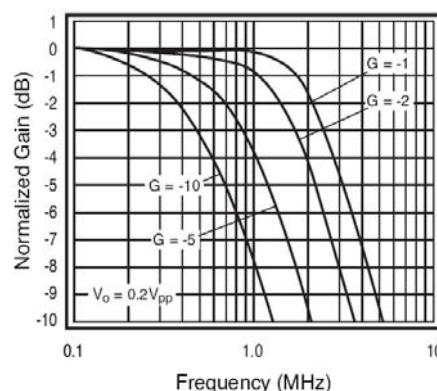


Figure 6. Inverting Frequency Response (+5 V)

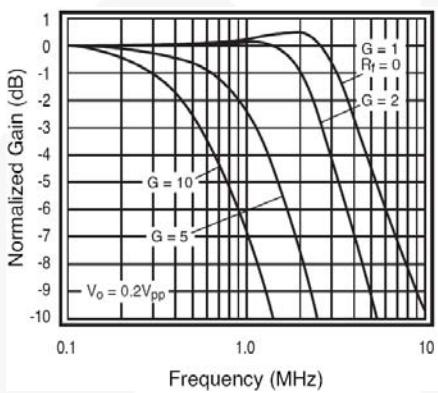


Figure 7. Non-Inverting Frequency Response

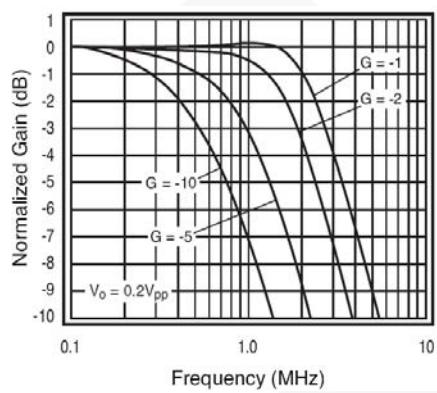


Figure 8. Inverting Frequency Response

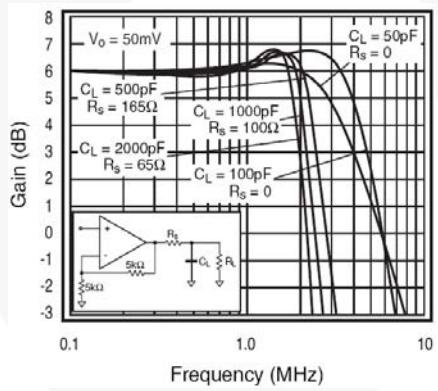


Figure 9. Frequency Response vs.  $C_L$

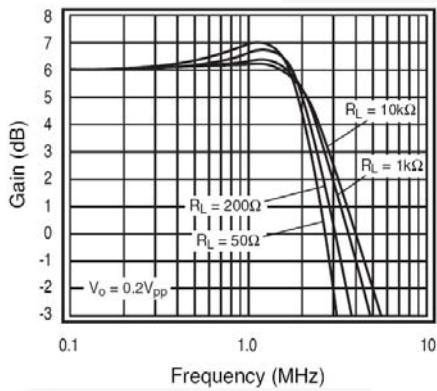


Figure 10. Frequency Response vs.  $R_L$

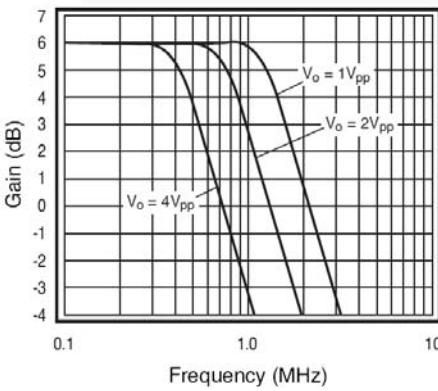


Figure 11. Large Signal Frequency Response (+5 V)

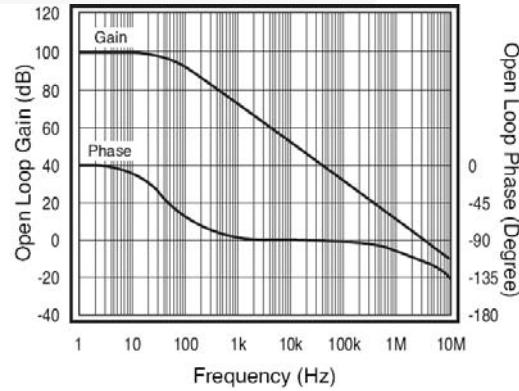


Figure 12. Open-loop Gain and Phase vs. Frequency

## Typical Performance Characteristic

$V_S=+2.7$  V,  $G=2$ ,  $R_L=10$  k $\Omega$  to  $V_S/2$ ,  $R_F=5$  k $\Omega$ ; unless otherwise noted.

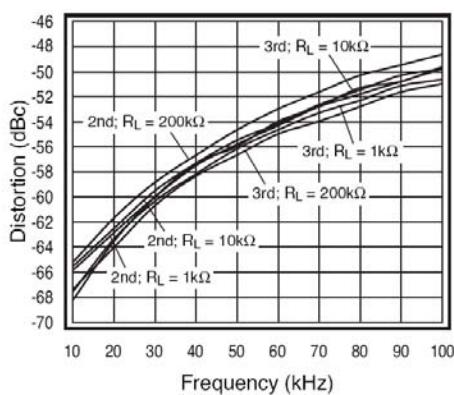


Figure 13. 2nd and 3rd Harmonic Distortion

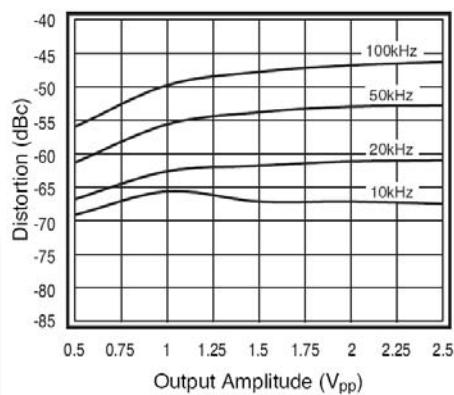


Figure 14. 2nd Harmonic Distortion vs.  $V_O$

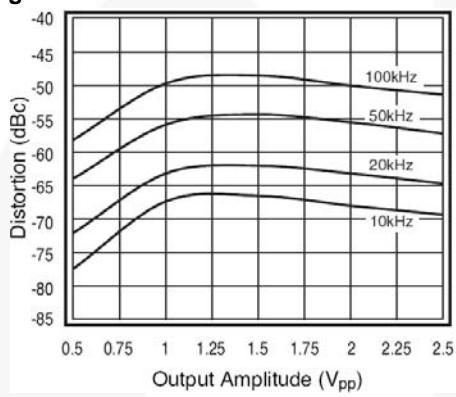


Figure 15. 3rd Harmonic Distortion vs.  $V_O$

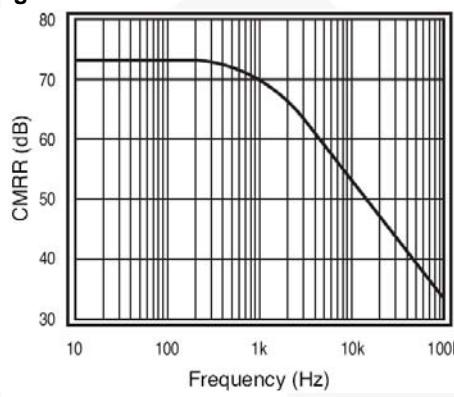


Figure 16. CMRR  $V_S=5$  V

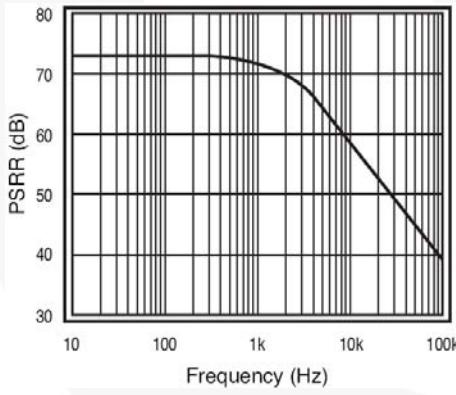


Figure 17. PSRR  $V_S=5$  V

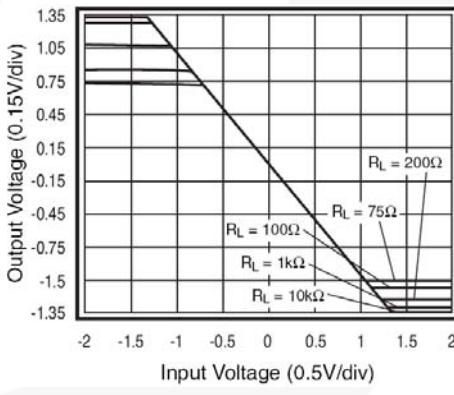


Figure 18. Output Swing vs. Load

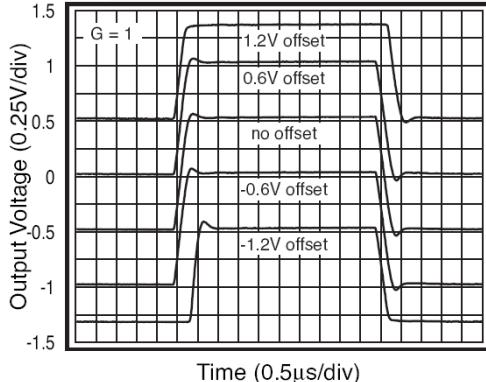


Figure 19. Pulse Response vs. Common-Mode Voltage

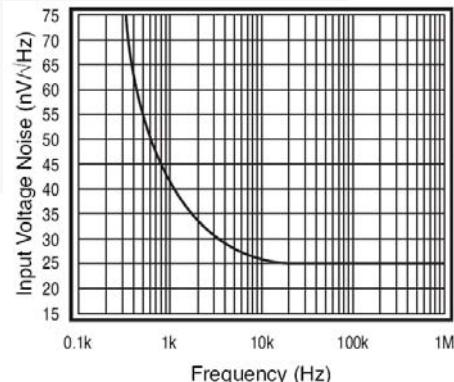


Figure 20. Input Voltage Noise

## Application Information

### General Description

The FAN4174 amplifier includes single-supply, general-purpose, voltage-feedback amplifiers, fabricated on a bi-CMOS process. The family features a rail-to-rail input and output and is unity gain stable. The typical non-inverting circuit schematic is shown in Figure 21.

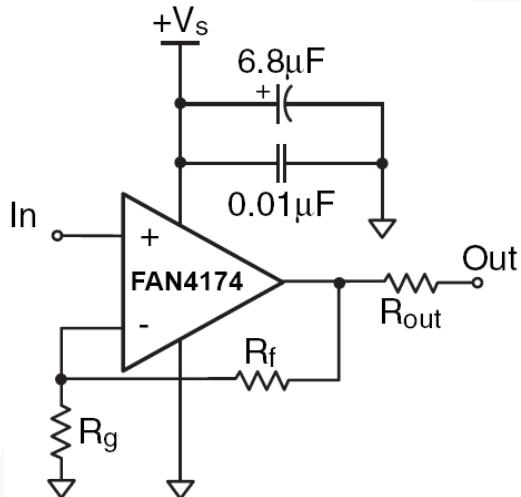


Figure 21. Typical Non-inverting Configuration

### Input Common Mode Voltage

The common mode input range extends to 300 mV below ground and to 100 mV above  $V_S$  in single supply operation. Exceeding these values does not cause phase reversal; however, if the input voltage exceeds the rails by more than 0.5 V, the input ESD devices begin to conduct. The output stays at the rail during this overdrive condition. If the absolute maximum input  $V_{IN}$  (700 mV beyond either rail) is exceeded, externally limit the input current to  $\pm 5$  mA, as shown in Figure 22.

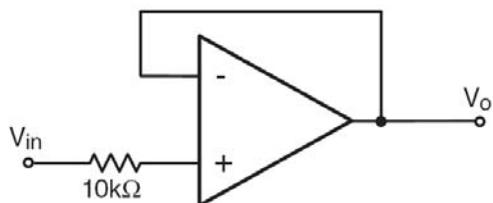


Figure 22. Circuit for Input Current Protection

### Power Dissipation

The maximum internal power dissipation allowed is directly related to the maximum junction temperature. If the maximum junction temperature exceeds 150°C, performance degradation occurs. If the maximum junction temperature exceeds 150°C for an extended time, device failure may occur.

### Overdrive Recovery

Overdrive of an amplifier occurs when the output and/or input ranges are exceeded. The recovery time varies based on whether the input or output is overdriven and by how much the range is exceeded. The FAN4174 typically recovers in less than 500 ns from an overdrive condition. Figure 23 shows the FAN4174 amplifier in an overdriven condition.

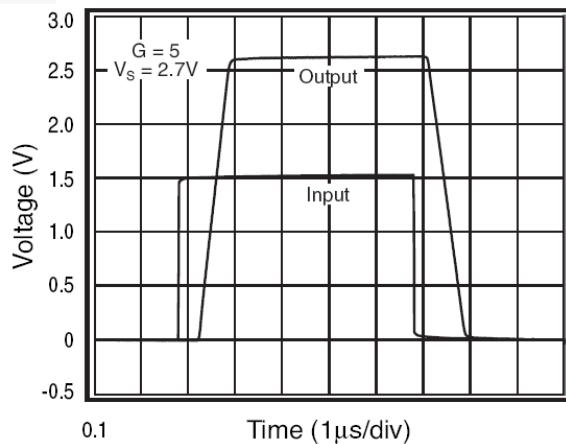


Figure 23. Overdrive Recovery

### Driving Capacitive Loads

Figure 9 illustrates the response of the FAN4174 amplifier. A small series resistance ( $R_S$ ) at the output of the amplifier, illustrated in Figure 24, improves stability and settling performance.  $R_S$  values in Figure 9 achieve maximum bandwidth with less than 2 dB of peaking. For maximum flatness, use a larger  $R_S$ . Capacitive loads larger than 500 pF require the use of  $R_S$ .

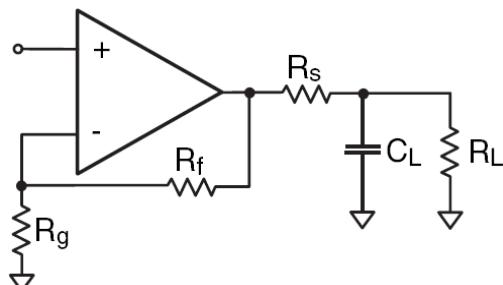


Figure 24. Typical Topology for Driving a Capacitive Load

Driving a capacitive load introduces phase-lag into the output signal, which reduces phase margin in the amplifier. The unity gain follower is the most sensitive configuration. In a unity gain follower configuration, the FAN4174 amplifier requires a 300 Ω series resistor to drive a 100 pF load.

## Layout Considerations

General layout and supply bypassing play major roles in high-frequency performance. Fairchild evaluation boards help guide high-frequency layout and aid in device testing and characterization. Follow the steps below as a basis for high-frequency layout:

1. Include 6.8  $\mu$ F and 0.01  $\mu$ F ceramic capacitors.
2. Place the 6.8  $\mu$ F capacitor within 19.05 mm (0.75 inches) of the power pin.
3. Place the 0.01  $\mu$ F capacitor within 2.54 mm (0.1 inches) of the power pin.
4. Remove the ground plane under and around the part, especially near the input and output pins, to reduce parasitic capacitance.

Minimize all trace lengths to reduce series inductances.

Refer to the evaluation board layouts shown in Figure 27 through Figure 30 for more information.

When evaluating only one channel, complete the following on the unused channel:

1. Ground the non-inverting input.
2. Short the output to the inverting input.

## Evaluation Board Information

The following evaluation boards are available to aid in the testing and layout of this device:

Board	Description	Product
KEB002	Single Channel, Dual Supply, 5 and 6-Lead SOT23	FAN4174IS5X
KEB010	Dual Channel, Dual Supply 8-Lead MSOP	FAN4274IMU8X

Evaluation board schematics are shown in Figure 25 and Figure 26; layouts are shown in Figure 27 through Figure 30.

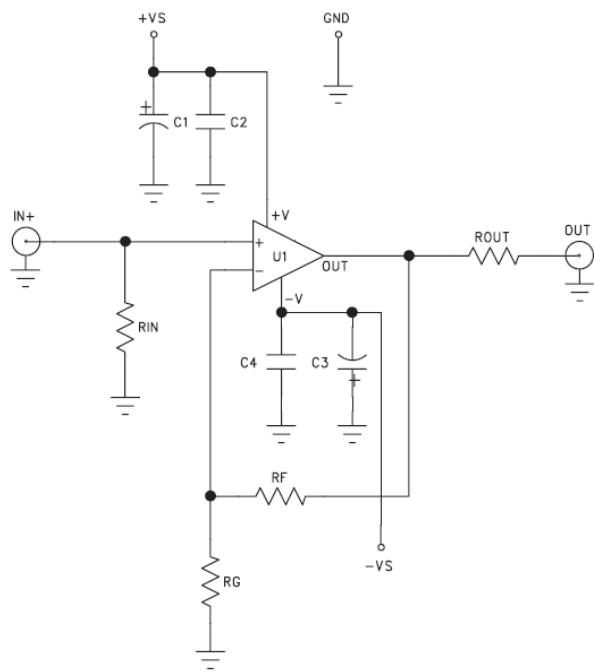


Figure 25. FAN4174 Evaluation Board Schematic (KEV002)

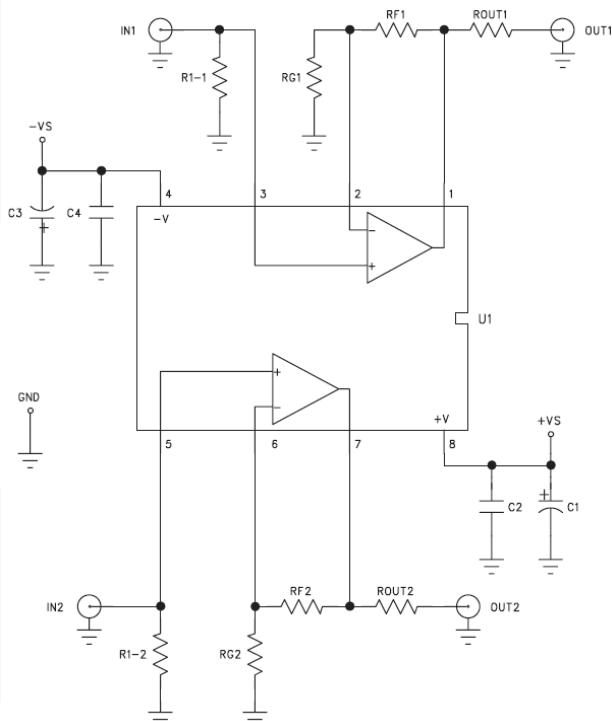


Figure 26. FAN4274 Evaluation Board Schematic (KEB010)

### Board Layout Information

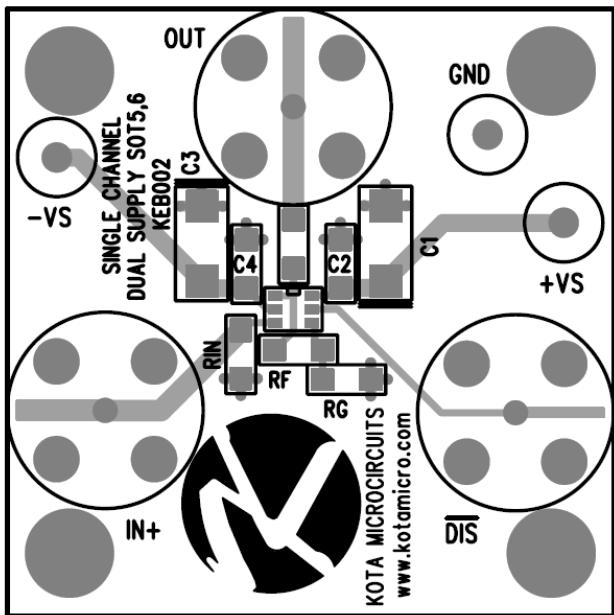


Figure 27. KEB002 (Top Side)

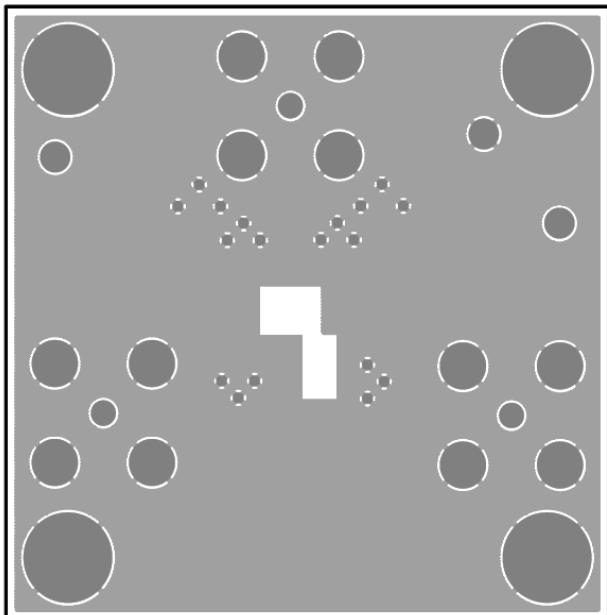


Figure 28. KEB002 (Bottom Side)

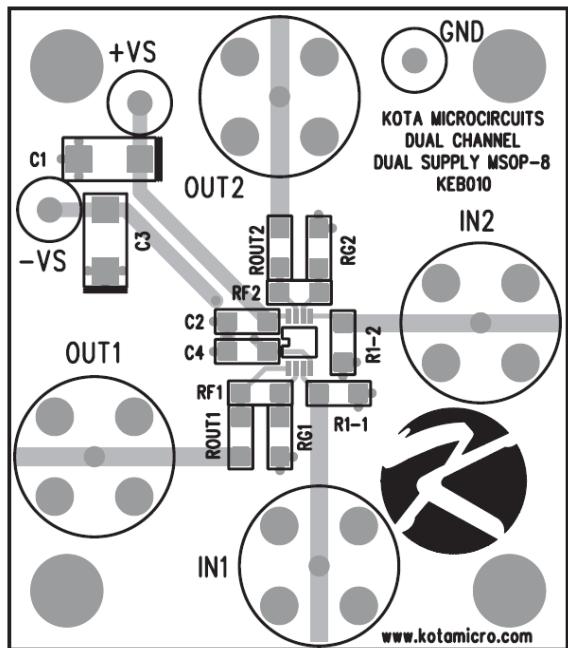


Figure 29. KEB010 (Top Side)

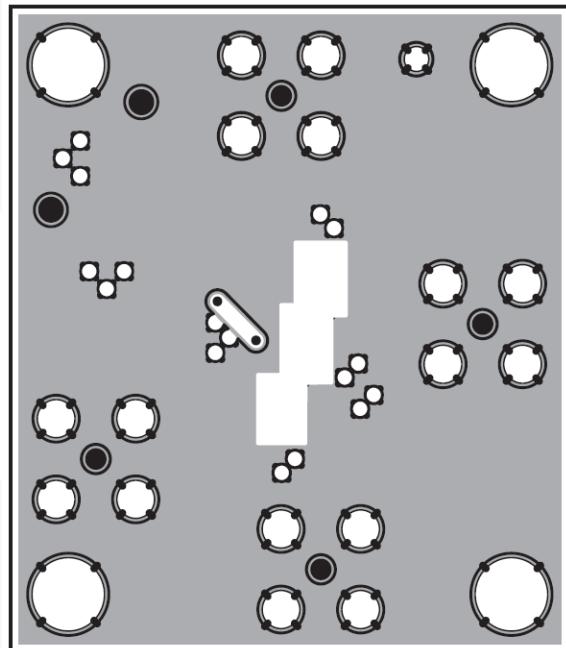


Figure 30. KEB010 (Bottom Side)

## Physical Dimensions

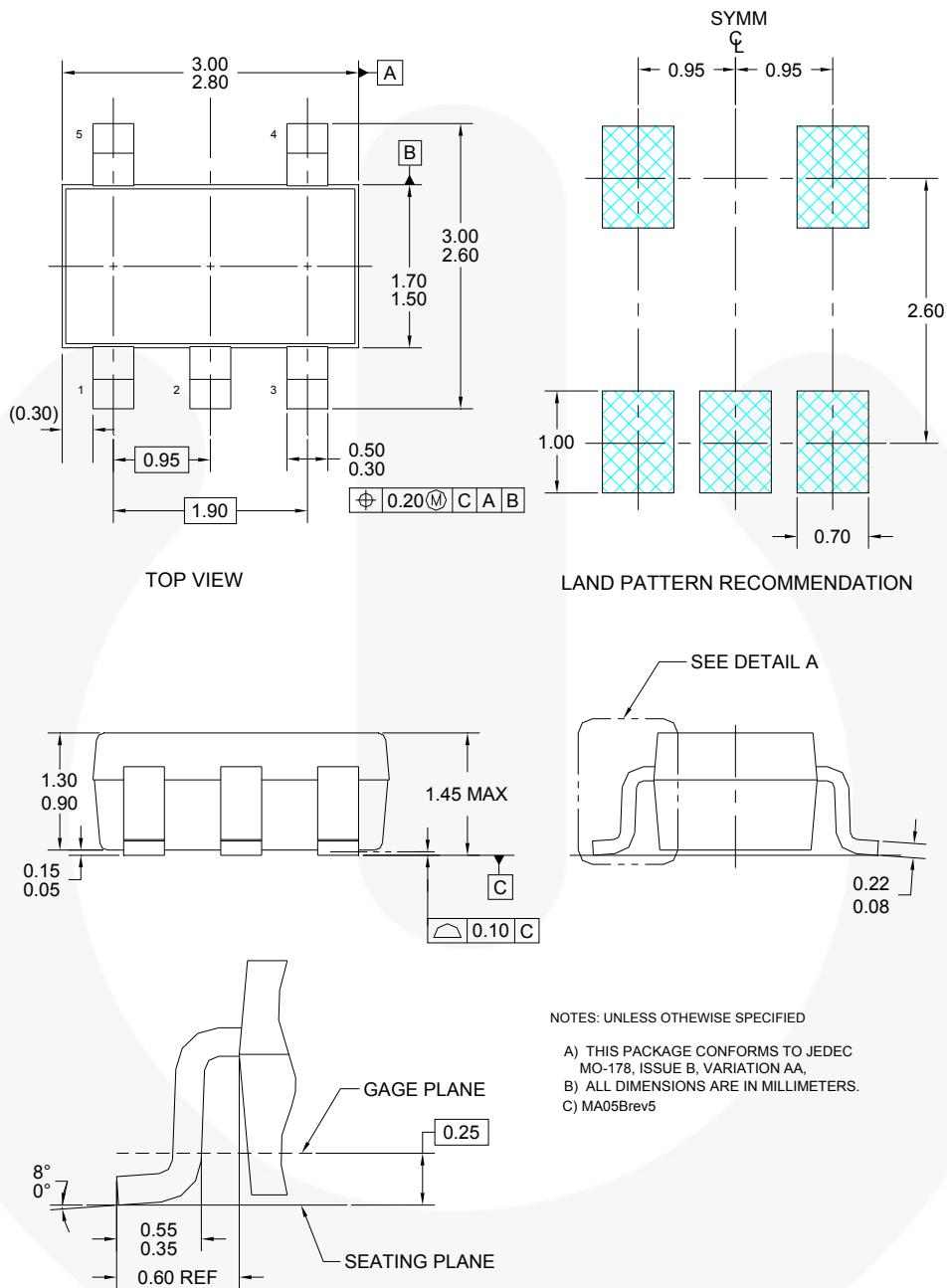


Figure 31. 5-Lead SOT-23 Package

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## Physical Dimensions

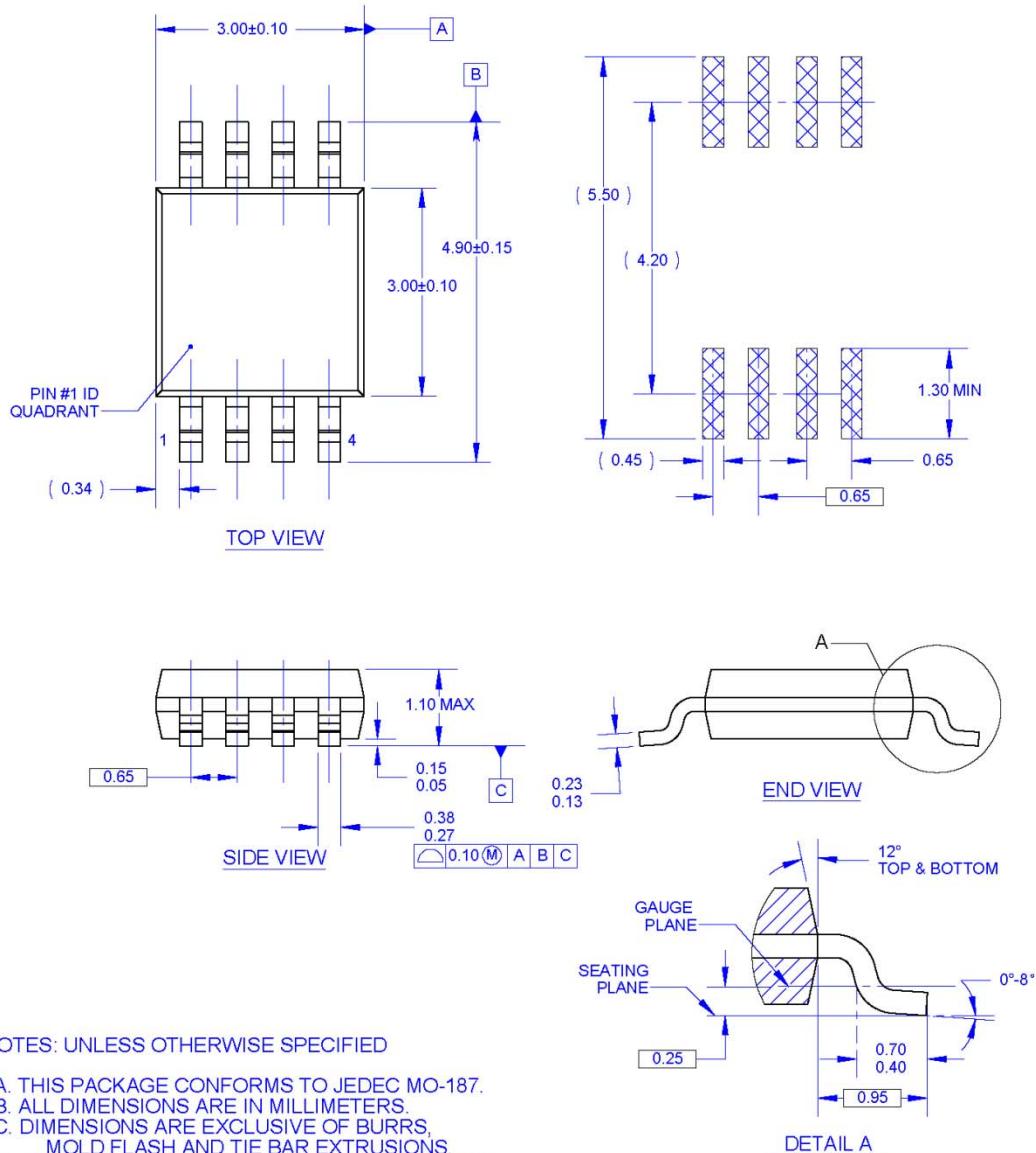


Figure 32. 8-Lead Molded Small Outline Package (MSOP)

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FACT®  
FAST®  
FastvCore™  
FETBench™  
FPS™  
F-PFS™  
FRFET®  
Global Power Resource™  
GreenBridge™  
Green FPS™  
Green FPS™ e-Series™  
Gmax™  
GTO™  
IntelliMAX™  
ISOPLANAR™  
Making Small Speakers Sound Louder and Better™  
MegaBuck™  
MICROCOUPLER™  
MicroFET™  
MicroPak™  
MicroPak™  
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QS™  
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SmartMax™  
SMART START™  
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STEALTH™  
SuperFET®  
SuperSOT™-3  
SuperSOT™-6  
SuperSOT™-8  
SupreMOS®  
SyncFET™

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TinyBoost™  
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