



16-Bit, Single Channel DIGITAL-TO-ANALOG CONVERTER With Internal Reference and Parallel Interface

FEATURES

- **LOW POWER: 150mW MAXIMUM**
- **+10V INTERNAL REFERENCE**
- **UNIPOLAR OR BIPOLAR OPERATION**
- **SETTLING TIME: 5 μ s to $\pm 0.003\%$ FSR**
- **16-BIT MONOTONICITY, -40°C TO +85°C**
- **± 10 V, ± 5 V OR +10V CONFIGURABLE VOLTAGE OUTPUT**
- **RESET TO MIN-SCALE OR MID-SCALE**
- **DOUBLE-BUFFERED DATA INPUT**
- **INPUT REGISTER DATA READBACK**
- **SMALL LOFP-48 PACKAGE**

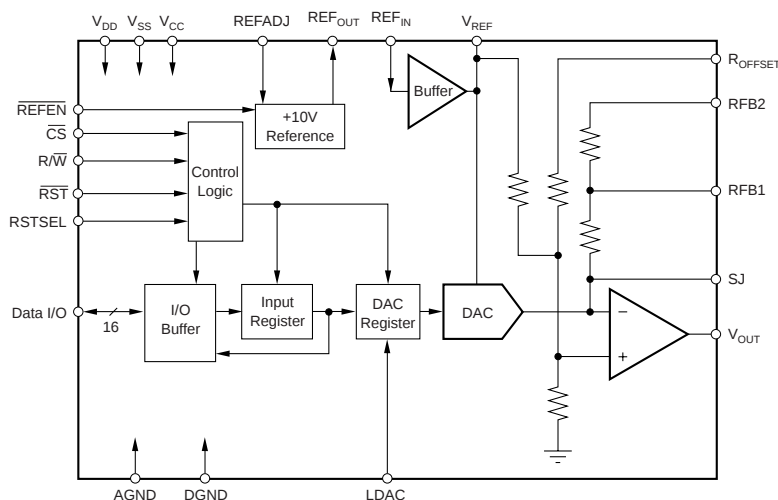
APPLICATIONS

- PROCESS CONTROL
- ATE PIN ELECTRONICS
- CLOSED-LOOP SERVO CONTROL
- MOTOR CONTROL
- DATA ACQUISITION SYSTEMS

DESCRIPTION

The DAC7741 is a 16-bit Digital-to-Analog Converter (DAC) which provides 16 bits of monotonic performance over the specified operating temperature range and offers a +10V, low-drift internal reference. Designed for automatic test equipment and industrial process control applications, the DAC7741 output swing can be configured in a $\pm 10V$, $\pm 5V$, or $+10V$ range. The flexibility of the output configuration allows the DAC7741 to provide both unipolar and bipolar operation by pin strapping. The DAC7741 includes a high-speed output amplifier with a maximum settling time of $5\mu s$ to $\pm 0.003\%$ FSR for a 20V full-scale change and only consumes 100mW (typical) of power.

The DAC7741 features a standard 16-bit parallel interface with double buffering to allow asynchronous updates of the analog output and data read-back to support data integrity verification prior to an update. A user-programmable reset control allows the DAC output to reset to min-scale (0000_H) or mid-scale (8000_H) overriding the DAC register values. The DAC7741 is available in a LQFP-48 package and four performance grades specified to operate from 0°C to +70°C and -40°C to +85°C.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

All trademarks are the property of their respective owners.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

V_{CC} to V_{SS}	-0.3V to +34V
V_{CC} to AGND	-0.3V to +17V
V_{SS} to AGND	-17V to +0.3V
AGND to DGND	-0.3V to +0.3V
REF _{IN} to AGND	0V to $V_{CC} - 1.4V$
V_{DD} to DGND	-0.3V to +6V
Digital Input Voltage to DGND	-0.3V to $V_{DD} + 0.3V$
Digital Output Voltage to DGND	-0.3V to $V_{DD} + 0.3V$
Operating Temperature Range	-40°C to +85°C
Storage Temperature Range	-65°C to +150°C
Junction Temperature	+150°C

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

PRODUCT	LINEARITY ERROR (LSB)	DIFFERENTIAL NONLINEARITY (LSB)	PACKAGE-LEAD	PACKAGE DESIGNATOR ⁽¹⁾	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	TRANSPORT MEDIA, QUANTITY
DAC7741Y "	±6 "	±4 "	LQFP-48 "	PT "	−40°C to +85°C "	DAC7741Y/250 DAC7741Y/2K	DAC7741Y "	Tape and Reel, 250 Tape and Reel, 2000
DAC7741YB "	±4 "	±2 "	LQFP-48 "	PT "	−40°C to +85°C "	DAC7741YB/250 DAC7741YB/2K	DAC7741YB "	Tape and Reel, 250 Tape and Reel, 2000
DAC7741YC "	±3 "	±1 "	LQFP-48 "	PT "	−40°C to +85°C "	DAC7741YC/250 DAC7741YC/2K	DAC7741YC "	Tape and Reel, 250 Tape and Reel, 2000
DAC7741YL "	±2 "	±1 "	LQFP-48 "	PT "	0°C to +70°C "	DAC7741YL/250 DAC7741YL/2K	DAC7741YL "	Tape and Reel, 250 Tape and Reel, 2000

All specifications at $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = +15V$, $V_{SS} = -15V$, $V_{DD} = +5V$, internal reference enabled, unless otherwise noted.



TEXAS
INSTRUMENTS
www.ti.com

2

ELECTRICAL CHARACTERISTICS (Cont.)

All specifications at $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = +15V$, $V_{SS} = -15V$, $V_{DD} = +5V$, internal reference enabled, unless otherwise noted.

PARAMETER	CONDITIONS	DAC7741Y			DAC7741YB			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
DYNAMIC PERFORMANCE Settling Time to $\pm 0.003\%$	20V Output Step $R_L = 5k\Omega$, $C_L = 200pF$, with external REF_{OUT} to REF_{IN} filter ⁽⁵⁾		3	5		*	*	μs
Digital Feedthrough Output Noise Voltage	at 10kHz		2 100			*	*	$\frac{nV \cdot s}{\sqrt{Hz}}$
DIGITAL INPUT V_{IH} V_{IL}	$ I_{IH} < 10\mu A$ $ I_{IL} < 10\mu A$	$0.7 \cdot V_{DD}$		$0.3 \cdot V_{DD}$	*		*	V V
DIGITAL OUTPUT V_{OH} V_{OL}	$I_{OH} = -0.8mA$ $I_{OL} = 1.6mA$	3.6		0.4	*		*	V V
POWER SUPPLY V_{DD} V_{CC} V_{SS} I_{DD} I_{CC} I_{SS} Power	Bipolar Operation Unipolar Operation Unloaded Unloaded No Load, Ext. Reference No Load, Int. Reference	+4.75 +11.4 -15.75 -15.75	+5.0 100 4 -2.5 85 100	+5.25 +15.75 -11.4 -4.75 6 150	* * * * * * * *	* * * * *	* * * * * * *	V V V V μA mA mA mW mW
TEMPERATURE RANGE Specified Performance		-40		+85	*		*	$^{\circ}C$

* Specifications same as grade to the left.

NOTES: (1) With minimum V_{CC}/V_{SS} requirements, internal reference enabled.

(2) Please refer to the "Theory of Operation" section for more information with respect to output voltage configurations.

(3) See Figure 7 for gain and offset adjustment connection diagrams when using the internal reference.

(4) The minimum value for REF_{IN} must be equal to the greater of $V_{SS} + 14V$ and $+4.75V$, where $+4.75V$ is the minimum voltage allowed.

(5) Reference low-pass filter values: $100k\Omega$, $1.0\mu F$ (See Figure 10).

All specifications at $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = +15V$, $V_{SS} = -15V$, $V_{DD} = +5V$, internal reference enabled, unless otherwise noted.

* Specifications same as grade to the left.

NOTES: (1) With minimum V_{CC}/V_{SS} requirements, internal reference enabled.

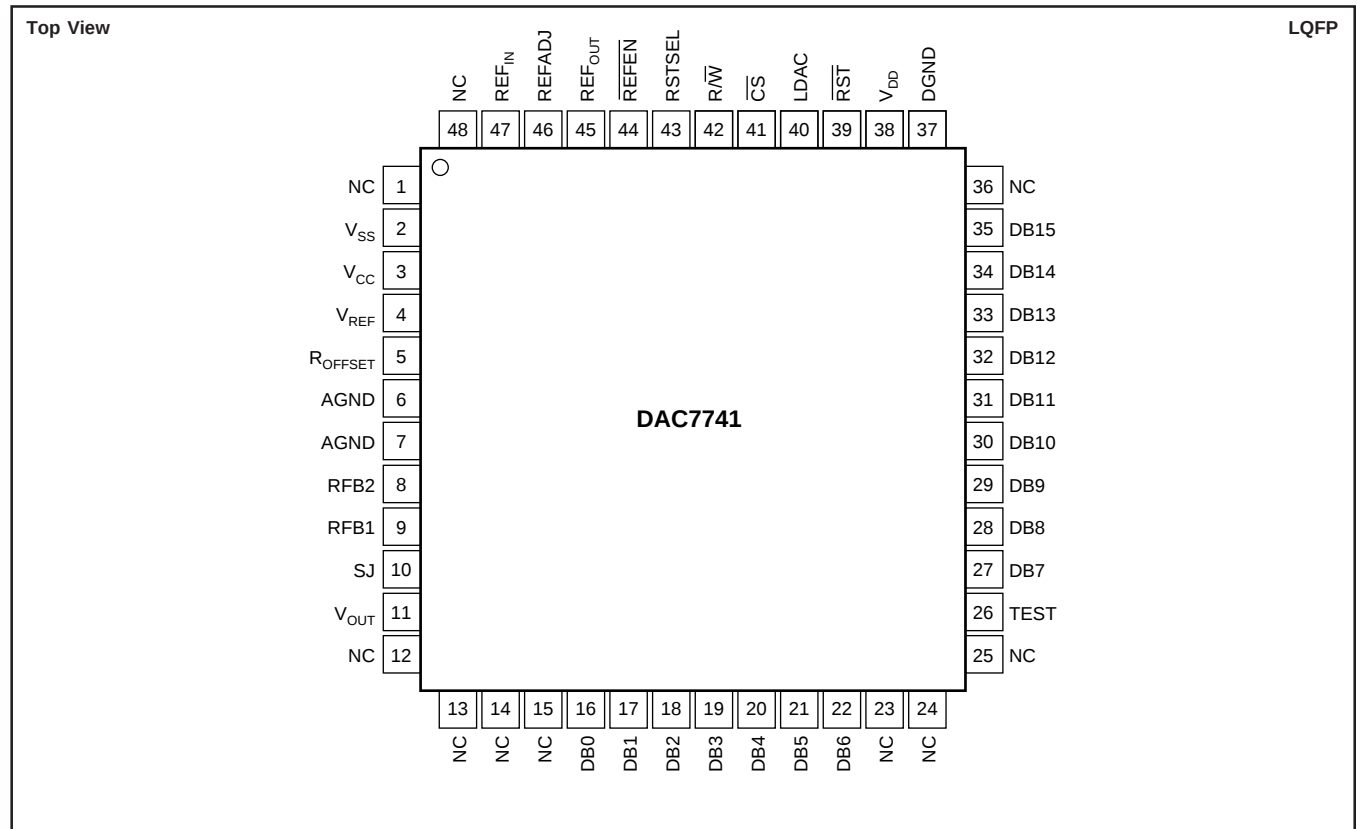
(2) Please refer to the "Theory of Operation" section for more information with respect to output voltage configurations.

(3) See Figure 7 for gain and offset adjustment connection diagrams when using the internal reference.

(4) The minimum value for REF_{IN} must be equal to the greater of $V_{SS} + 14V$ and $+4.75V$, where $+4.75V$ is the minimum voltage allowed.

(5) Reference low-pass filter values: 100k Ω , 1.0 μ F (See Figure 10).

PIN CONFIGURATION



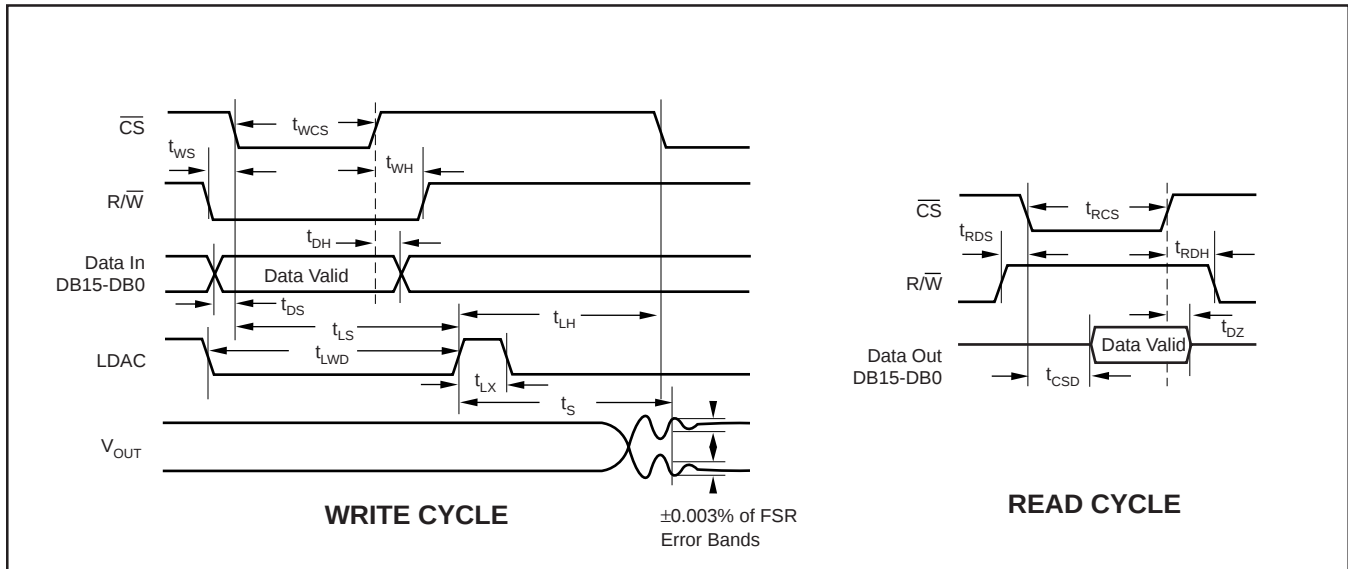
PIN DESCRIPTIONS

PIN	NAME	DESCRIPTION	PIN	NAME	DESCRIPTION
1	NC	No Connection	28	DB8	Data Bit 8
2	V _{SS}	Negative Analog Power Supply.	29	DB9	Data Bit 9
3	V _{CC}	Positive Analog Power Supply.	30	DB10	Data Bit 10
4	V _{REF}	Buffered Output from REF _{IN} , can be used to drive external devices. Internally, this pin directly drives the DAC circuitry.	31	DB11	Data Bit 11
5	R _{OFFSET}	Offsetting Resistor	32	DB12	Data Bit 12
6	AGND	Analog ground (Must be tied to analog ground)	33	DB13	Data Bit 13
7	AGND	Analog ground (Must be tied to analog ground)	34	DB14	Data Bit 14
8	RFB2	Feedback Resistor 2, used to configure DAC output range.	35	DB15	Data Bit 15 (MSB)
9	RFB1	Feedback Resistor 1, used to configure DAC output range.	36	NC	No Connection
10	SJ	Summing Junction of the Output Amplifier	37	DGND	Digital Ground
11	V _{OUT}	DAC Voltage Output	38	V _{DD}	Digital Power Supply
12	NC	No Connection	39	R _{ST}	V _{OUT} reset; active LOW, depending on the state of RSTSEL, the DAC register is either reset to mid-scale or min-scale.
13	NC	No Connection	40	LDAC	DAC register load control, rising edge triggered. Data is loaded from the input register to the DAC register.
14	NC	No Connection	41	CS	Chip Select, active LOW
15	NC	No Connection	42	R/W	Enabled by CS, controls data read (HIGH) and write (LOW) from or to the input register.
16	DB0	Data Bit 0 (LSB)	43	RSTSEL	Reset Select; determines the action of RST. If HIGH, RST will reset the DAC register to mid-scale. If LOW, RST will reset the DAC register to min-scale.
17	DB1	Data Bit 1	44	REFEN	Enables internal +10V reference (REF _{OUT}), active LOW.
18	DB2	Data Bit 2	45	REF _{OUT}	Internal Reference Output
19	DB3	Data Bit 3	46	REFADJ	Internal Reference Trim. (Acts as a gain adjustment input when the internal reference is used.)
20	DB4	Data Bit 4	47	REF _{IN}	Reference Input
21	DB5	Data Bit 5	48	NC	No Connection
22	DB6	Data Bit 6			
23	NC	No Connection			
24	NC	No Connection			
25	NC	No Connection			
26	TEST	Reserved, Connect to DGND			
27	DB7	Data Bit 7			

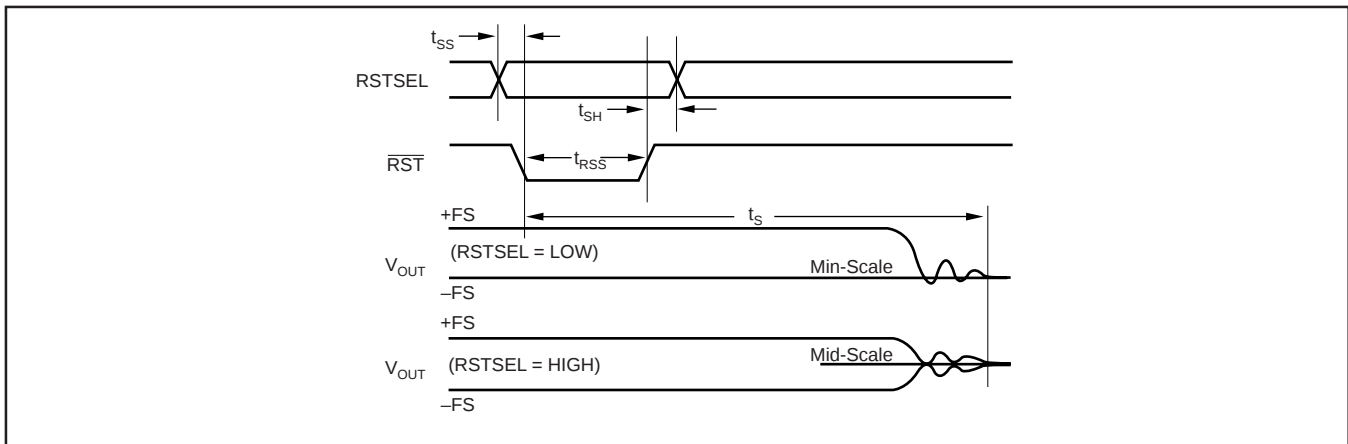
TIMING CHARACTERISTICS

PARAMETER	DESCRIPTION	DAC7741Y			UNITS
		MIN	TYP	MAX	
t_{RCS}	$\overline{CS}$ LOW for Read	100			ns
t_{RDS}	$R/\overline{W}$ HIGH to $\overline{CS}$ LOW	10			ns
t_{RDH}	$R/\overline{W}$ HIGH after $\overline{CS}$ HIGH	10			ns
t_{DZ}	$\overline{CS}$ HIGH to Data Bus High Impedance	10		70	ns
t_{CSD}	$\overline{CS}$ LOW to Data Bus Valid		85	100	ns
t_{WCS}	$\overline{CS}$ LOW for Write	30			ns
t_{WS}	$R/\overline{W}$ LOW to $\overline{CS}$ LOW	10			ns
t_{WH}	$R/\overline{W}$ LOW after $\overline{CS}$ HIGH	10			ns
t_{LS}	$\overline{CS}$ LOW to LDAC HIGH	40			ns
t_{LH}	$\overline{CS}$ LOW after LDAC HIGH	0			ns
t_{LX}	LDAC HIGH	30			ns
t_{DS}	Data Valid to $\overline{CS}$ LOW	0			ns
t_{DH}	Data Valid after $\overline{CS}$ HIGH	20			ns
t_{LWD}	LDAC LOW	40			ns
t_{SS}	RSTSEL Valid Before $\overline{RST}$ LOW	0			ns
t_{SH}	RSTSEL Valid After $\overline{RST}$ HIGH	10			ns
t_{RSS}	$\overline{RST}$ LOW	30			ns
t_S	Voltage Output Settling Time			5	μ s

TIMING DIAGRAMS

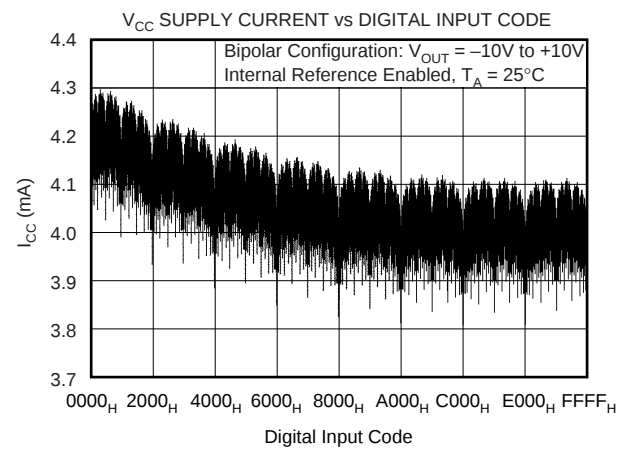
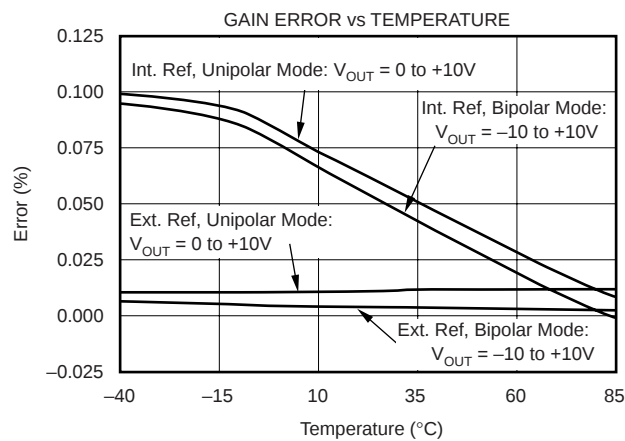
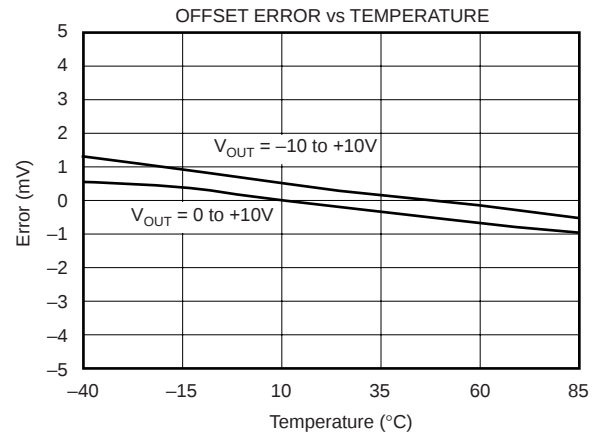
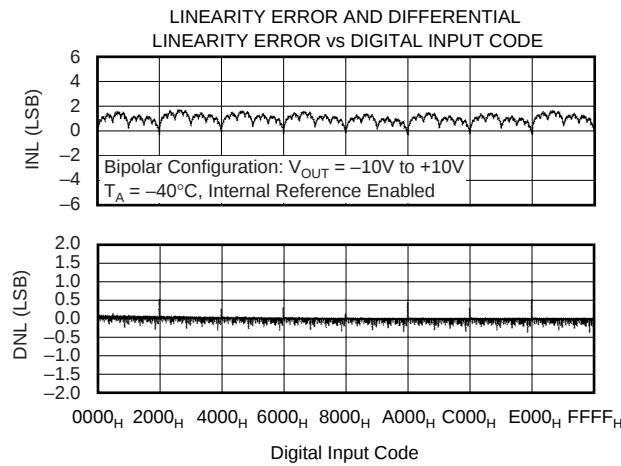
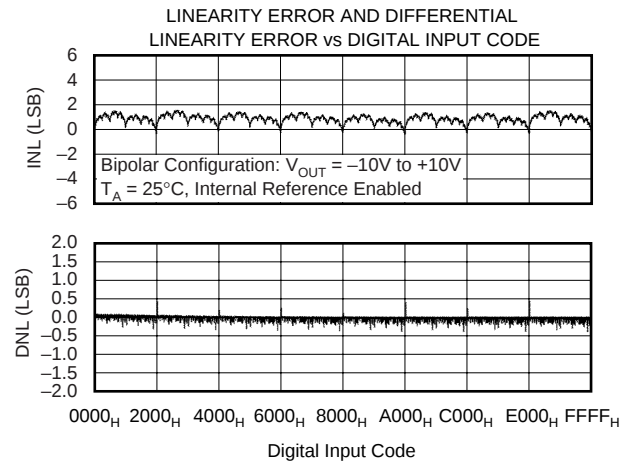
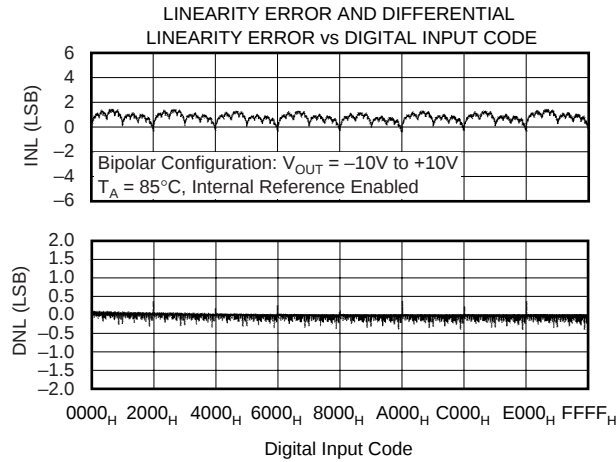


RESET TIMING



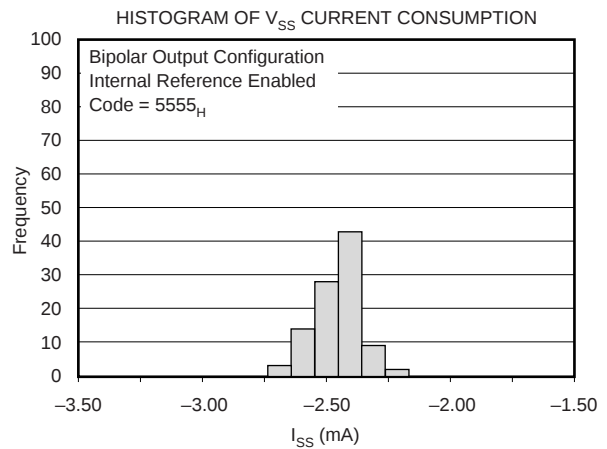
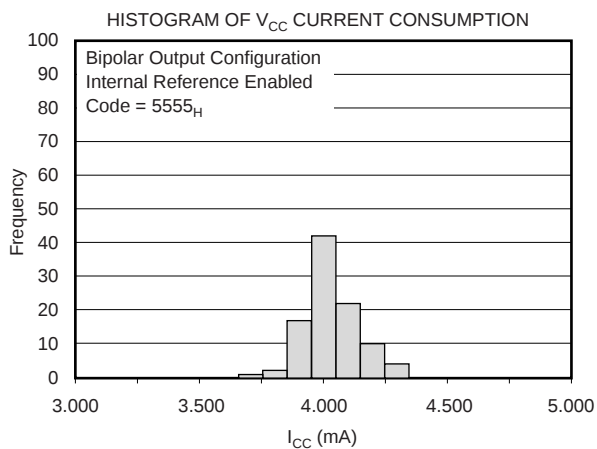
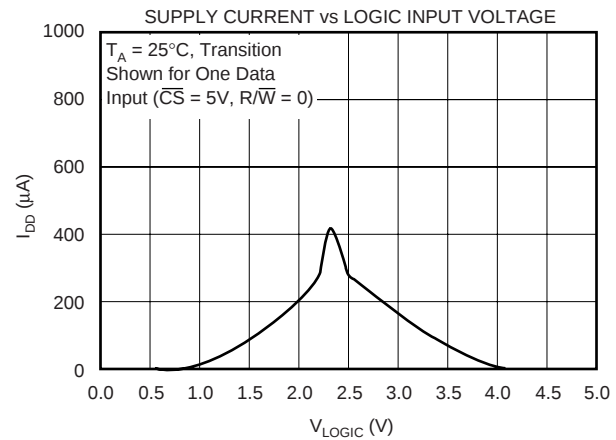
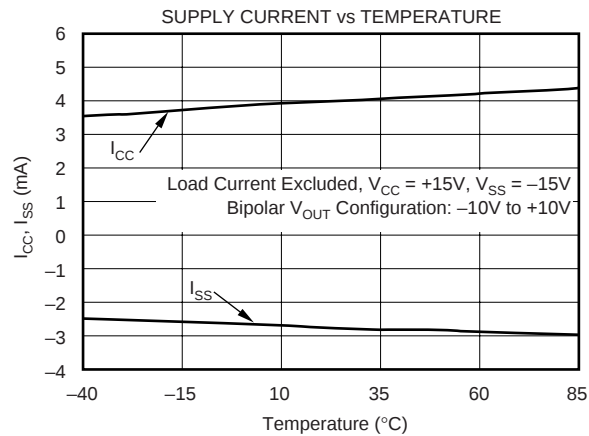
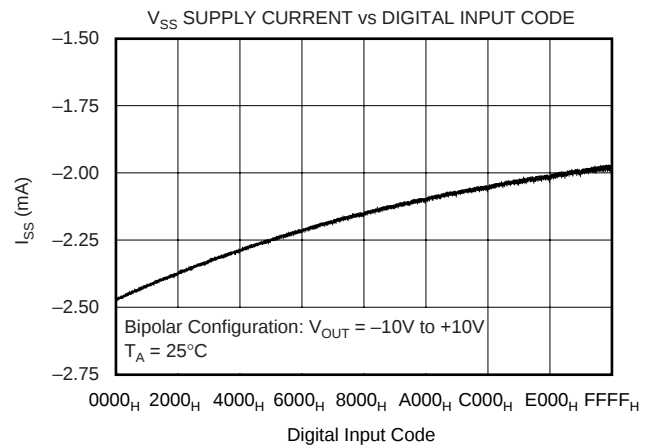
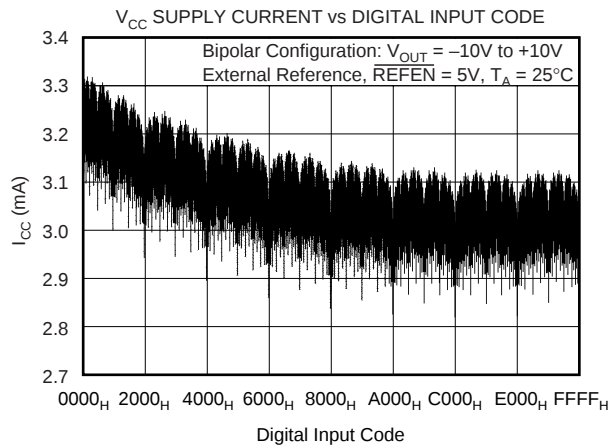
TYPICAL CHARACTERISTICS

$T_A = +25^\circ\text{C}$ (unless otherwise noted)



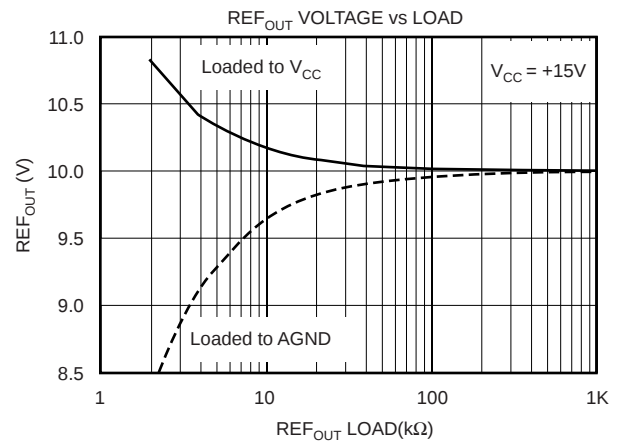
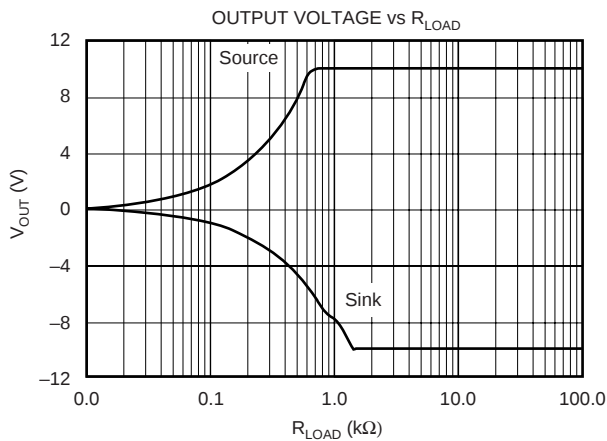
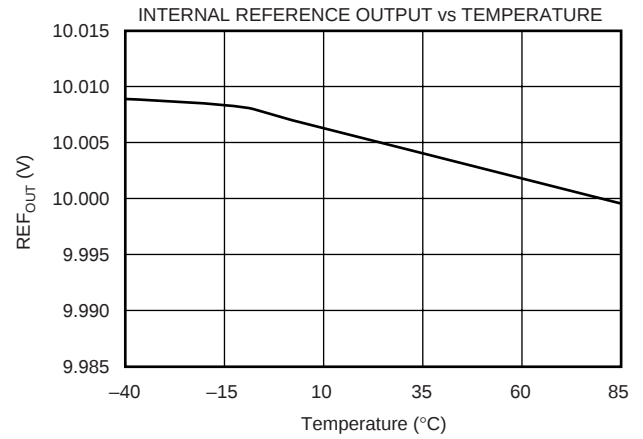
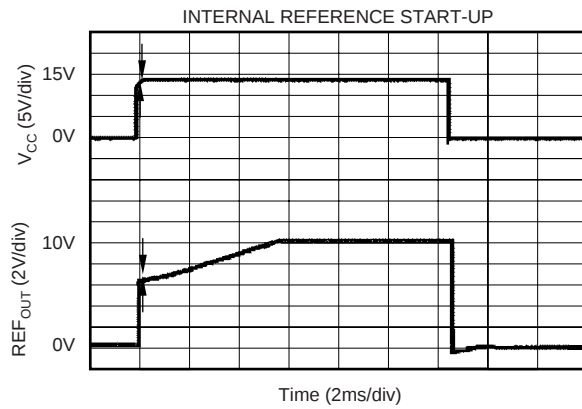
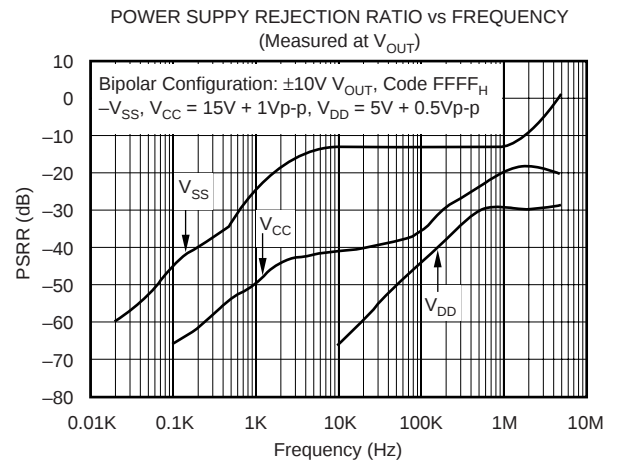
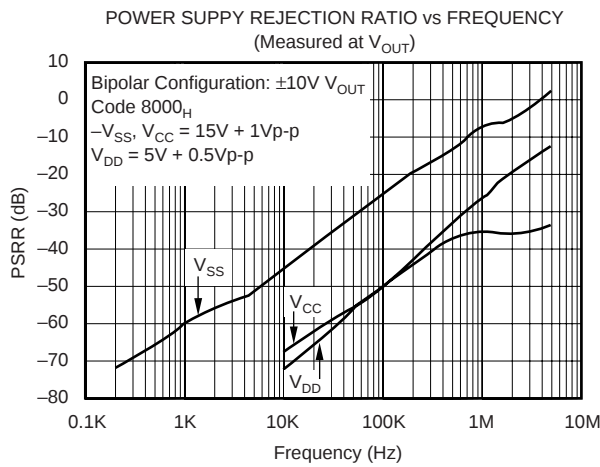
TYPICAL CHARACTERISTICS (Cont.)

$T_A = +25^\circ\text{C}$ (unless otherwise noted)



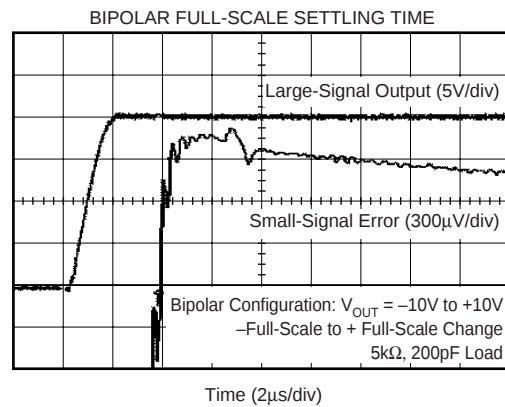
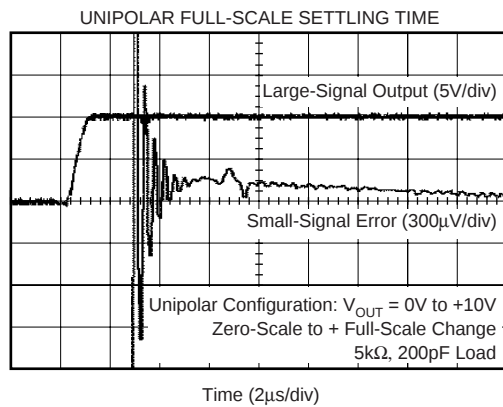
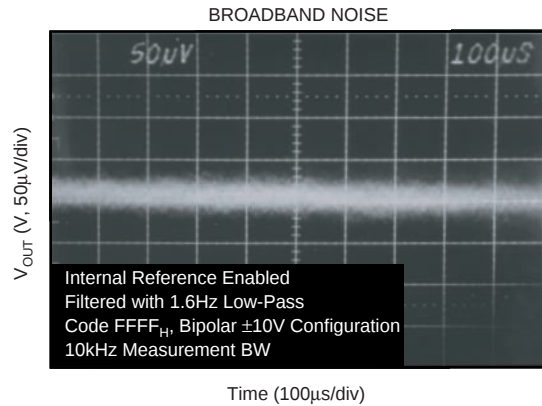
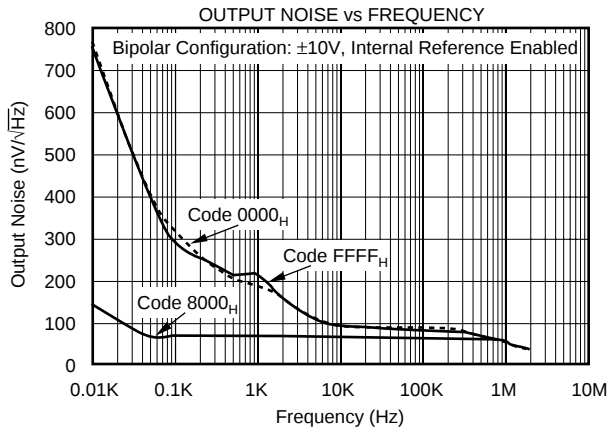
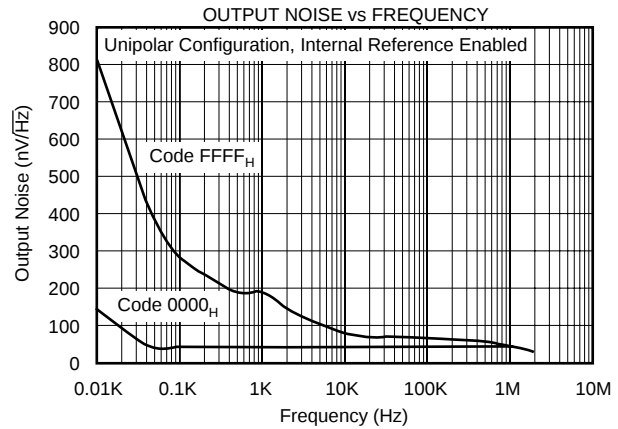
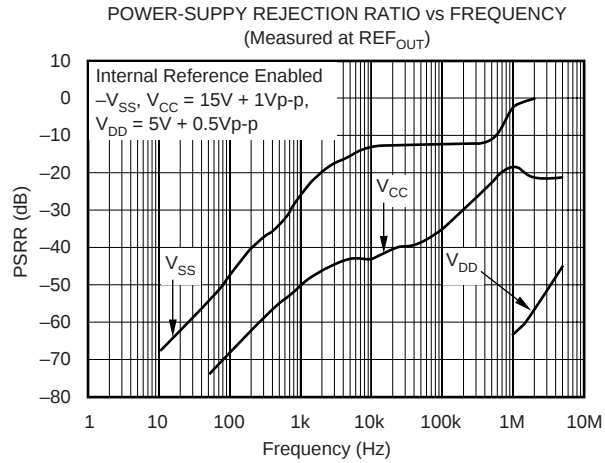
TYPICAL CHARACTERISTICS (Cont.)

$T_A = +25^\circ\text{C}$ (unless otherwise noted)



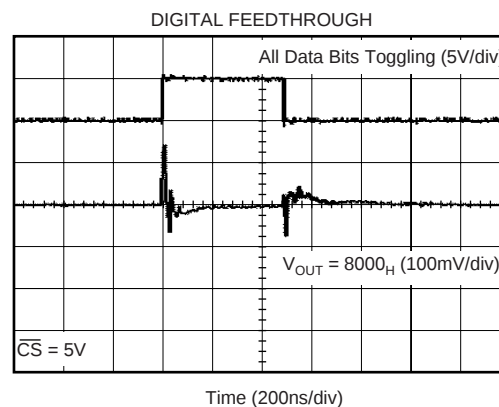
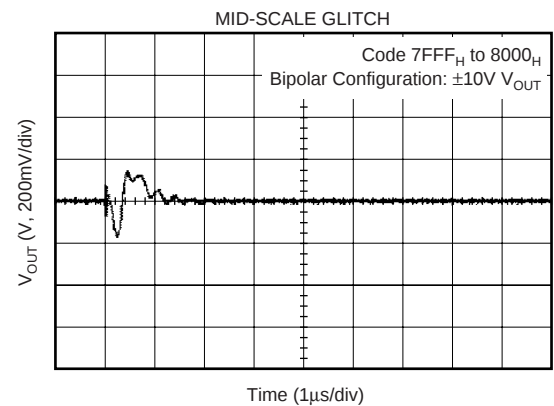
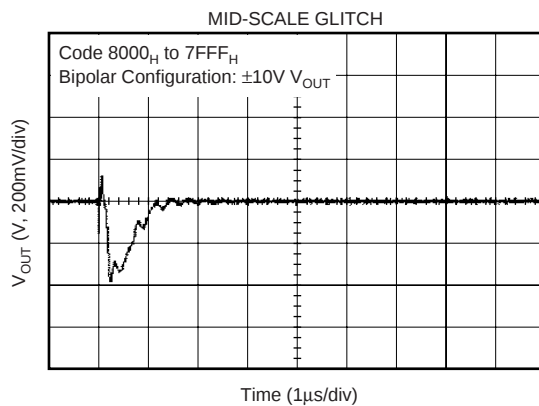
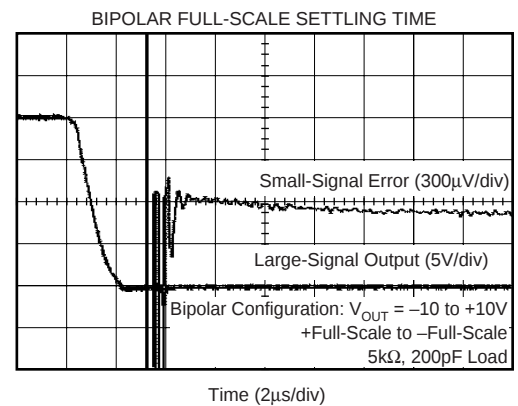
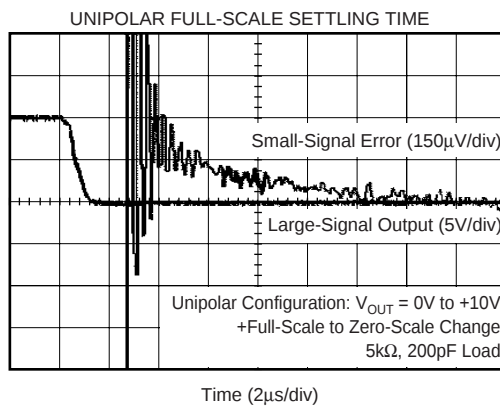
TYPICAL CHARACTERISTICS (Cont.)

$T_A = +25^\circ\text{C}$ (unless otherwise noted)



TYPICAL CHARACTERISTICS (Cont.)

$T_A = +25^\circ\text{C}$ (unless otherwise noted)



THEORY OF OPERATION

The DAC7741 is a voltage output, 16-bit DAC with a +10V built-in internal reference. The architecture is an R-2R ladder configuration with the three MSBs segmented, followed by an operational amplifier that serves as a buffer. The output buffer is designed to allow user-configurable output adjustments, giving the DAC7741 output voltage ranges of 0V to +10V, -5V to +5V, or -10V to +10V. Please refer to Figures 2, 3, and 4 for pin configuration information.

The digital input is a parallel word made up of the 16-bit DAC code, which is then loaded into the DAC register using the LDAC input pin. The converter can be powered from $\pm 12\text{V}$ to $\pm 15\text{V}$ dual analog supplies and a +5V logic supply. The device offers a reset function, which immediately sets the DAC output voltage and DAC register to min-scale (code 0000_H) or mid-scale (code 8000_H). The data I/O and reset functions are discussed in more detail in the following sections.

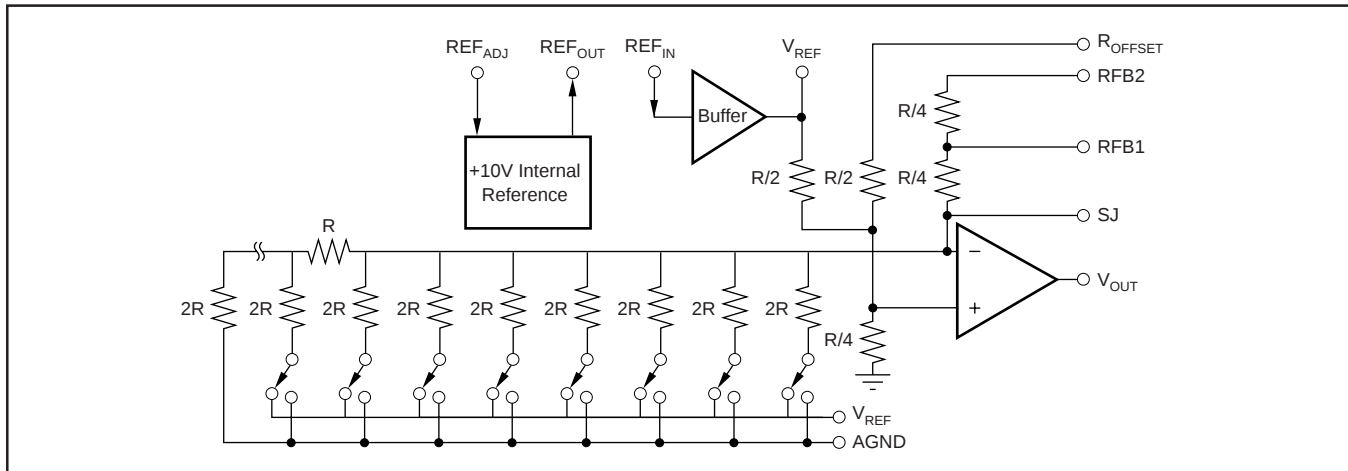


FIGURE 1. DAC7741 Architecture.

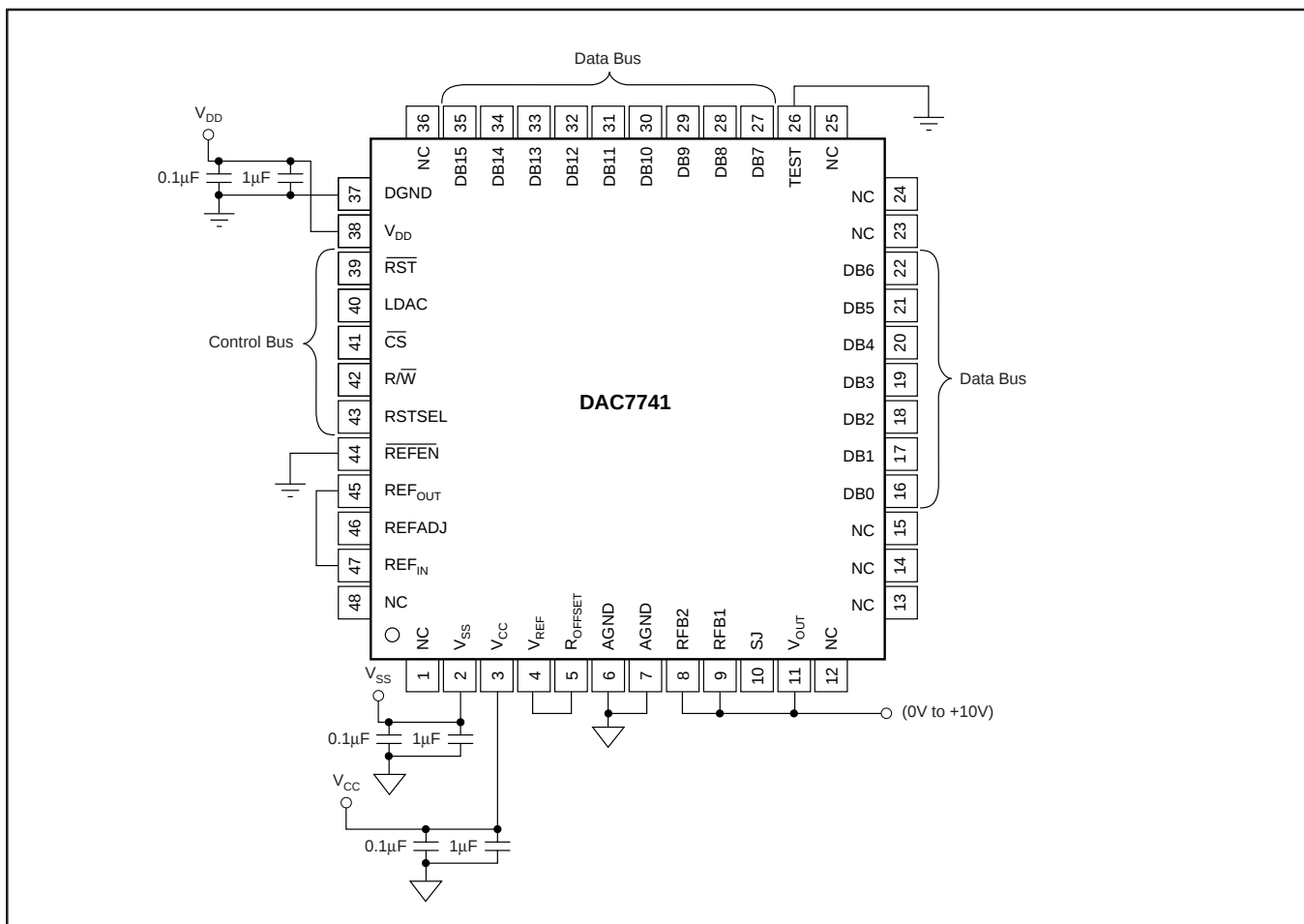


FIGURE 2. Basic Operation: $V_{OUT} = 0$ to +10V.

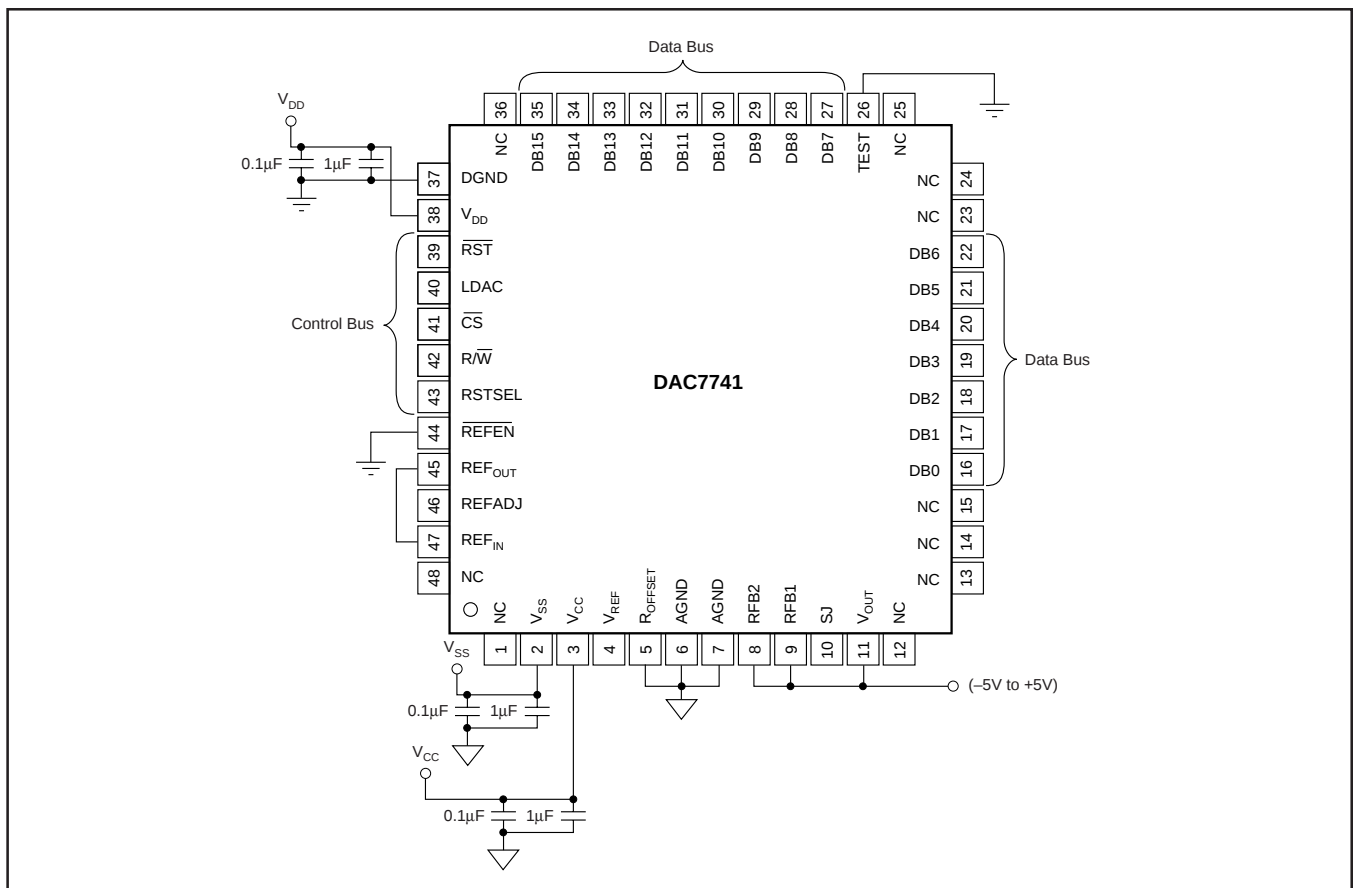


FIGURE 3. Basic Operation: $V_{OUT} = -5\text{V to } +5\text{V}$.

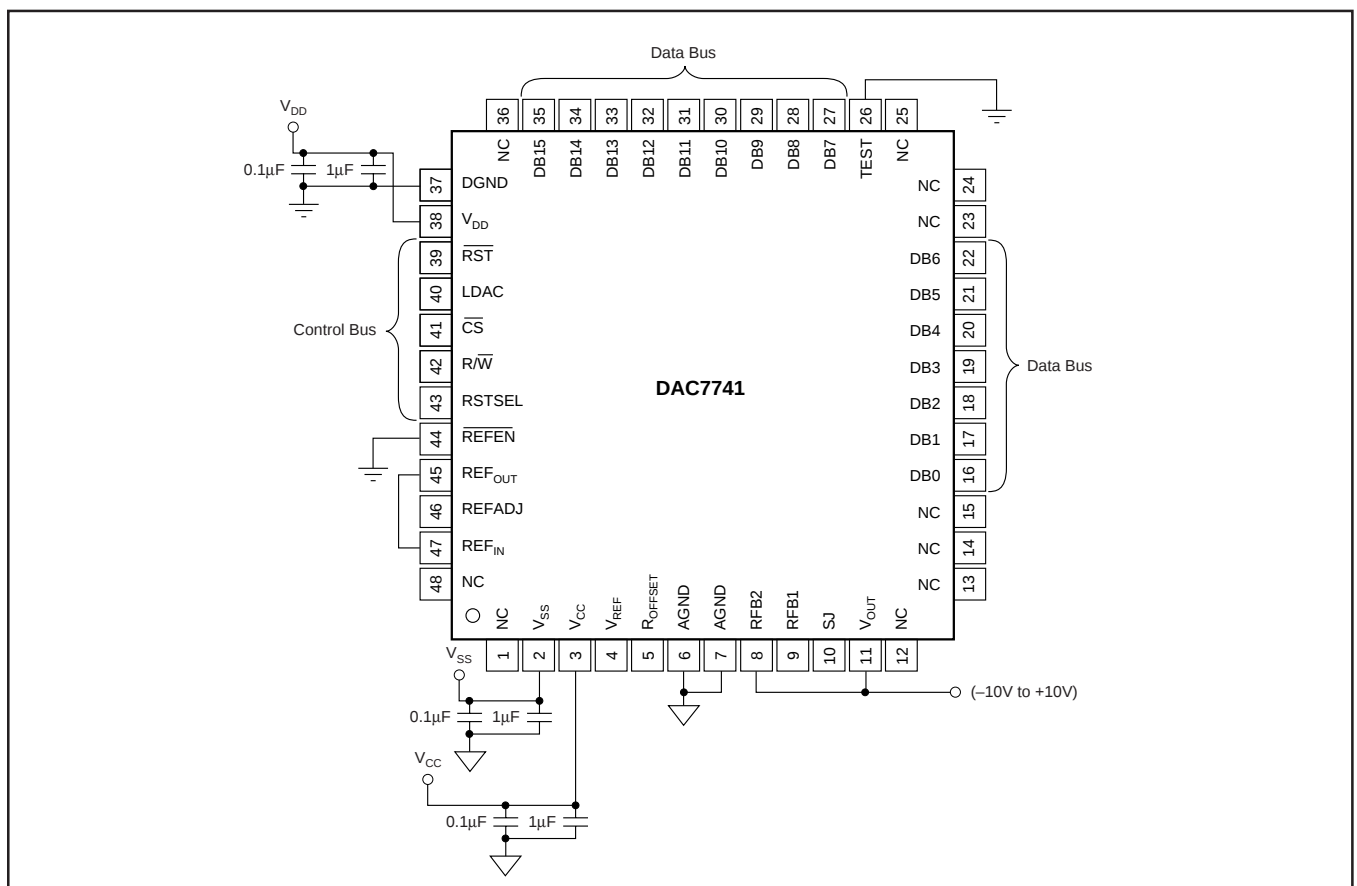


FIGURE 4. Basic Operation: $V_{OUT} = -10\text{V to } +10\text{V}$.

ANALOG OUTPUTS

The output amplifier can swing to within 1.4V of the supply rails, specified over the -40°C to $+85^{\circ}\text{C}$ temperature range. This allows for a $\pm 10\text{V}$ DAC voltage output operation from $\pm 12\text{V}$ supplies with a typical 5% tolerance.

When the DAC7741 is configured for a unipolar, 0V to 10V output, a negative voltage supply is required. This is due to internal biasing of the output stage. Please refer to the "Electrical Characteristics" table for more information.

The minimum and maximum voltage output values are dependent upon the output configuration implemented and reference voltage applied to the DAC7741. Please note that V_{SS} (the negative power supply) must be in the range of -4.75V to -15.75V for unipolar operation. The voltage on V_{SS} sets several bias points within the converter and is required in all modes of operation. If V_{SS} is not in one of these two configurations, the bias values may be in error and proper operation of the device is not ensured.

Supply sequence is important in establishing the correct startup of the DAC. The digital supply (V_{DD}) needs to establish correct bias conditions before the analog supplies (V_{CC} , V_{SS}) are brought up. If the digital supply cannot be brought up first, it must come up before either analog supply (V_{CC} or V_{SS}), with the preferred sequence of: V_{SS} (device substrate), V_{DD} then V_{CC} .

REFERENCE INPUTS

The DAC7741 provides a built-in $+10\text{V}$ voltage reference and on-chip buffer to allow external component reference drive. To use the internal reference, $\overline{\text{REFEN}}$ must be LOW, enabling the reference circuitry of the DAC7741 (see Table I) and the REF_{OUT} pin must be connected to REF_{IN} . This is the input to the on-chip reference buffer. The buffers output is provided at

REFEN	ACTION
1	Internal Reference disabled; $\text{REF}_{\text{OUT}} = \text{HIGH Impedance}$
0	Internal Reference enabled; $\text{REF}_{\text{OUT}} = +10\text{V}$

TABLE I. $\overline{\text{REFEN}}$ Action.

the V_{REF} pin. In this configuration, V_{REF} is used to setup the DAC7741 output amplifier into one of three voltage output modes as discussed earlier. V_{REF} can also be used to drive other system components requiring an external reference.

The internal reference of the DAC7741 can be disabled when use of an external reference is desired. When using an external reference, the reference input, REF_{IN} , can be any voltage between 4.75V (or $V_{SS} + 14\text{V}$, whichever is greater) and $V_{CC} - 1.4\text{V}$.

DIGITAL INTERFACE

Table III shows the data format for the DAC7741 and Table II illustrates the basic control logic of the device. The interface consists of a chip select input ($\overline{\text{CS}}$), read/write control input ($\text{R}/\overline{\text{W}}$), data inputs (DB0-DB15) and a load DAC input (LDAC). An asynchronous reset input ($\overline{\text{RST}}$) which is active low, is provided to simplify start-up conditions, periodic resets, or emergency resets to a known state, depending on the status of the reset select (RSTSEL) signal. The DAC code is provided via a 16-bit parallel interface, as shown in Table II. The input word makes up the DAC code to be loaded into the data input register of the device. The data is latched into the input register on rising $\overline{\text{CS}}$ and is loaded into the DAC register upon reception of a rising edge on the LDAC input. This action updates the analog output, V_{OUT} , to the desired value. LDAC inputs of multiple DAC7741 devices can be connected when a synchronized update of numerous DAC outputs is desired. Please refer to the timing section for more detailed data I/O information.

DIGITAL INPUT	ANALOG OUTPUT	
	Unipolar Configuration	Bipolar Configuration
	Unipolar Straight Binary	Bipolar Offset Binary
0x0000	Zero (0V)	–Full-Scale ($-V_{\text{REF}}$ or $-V_{\text{REF}}/2$)
0x0001	Zero + 1LSB	–Full-Scale + 1LSB
:	:	:
0x8000	1/2 Full-Scale	Bipolar Zero
0x8001	1/2 Full-Scale + 1LSB	Bipolar Zero + 1LSB
:	:	:
0xFFFF	Full-Scale ($V_{\text{REF}} - 1\text{LSB}$)	+Full-Scale ($+V_{\text{REF}} - 1\text{LSB}$ or $+V_{\text{REF}}/2 - 1\text{LSB}$)

TABLE III. DAC7741 Data Format.

CONTROL STATUS					COMMAND		
$\text{R}/\overline{\text{W}}$	$\overline{\text{CS}}$	$\overline{\text{RST}}$	RSTSEL	LDAC	Input Register	DAC Register	Mode
L	L	H	X	H, L, $\downarrow$	Write	Hold	Write Data to Input Register
X	H	H	X	$\uparrow$	Hold	Write	Update DAC register with data from input register.
L	L	H	X	$\uparrow$	Transparent	Write	Write DAC register directly from data bus
H	L	H	X	H, L, $\downarrow$	Read	Hold	Read data in input register.
X	H	H	X	H, L, $\downarrow$	Hold	Hold	No Change
X	X	L	L	X	Reset to Min-Scale	Reset to Min-Scale	Reset to Input and DAC Register (0000 _H) Min-Scale
X	X	L	H	X	Reset to Mid-Scale	Reset to Mid-Scale	Reset to Input and DAC Register (8000 _H) Mid-Scale

TABLE II. DAC7741 Logic Truth Table.

DAC RESET

The $\overline{\text{RST}}$ and RSTSEL inputs control the reset of the analog output. The reset command is level triggered by a low signal on $\overline{\text{RST}}$. Once $\overline{\text{RST}}$ is LOW, the DAC output will begin settling to the mid-scale or min-scale code depending on the state of the RSTSEL input. A HIGH value on RSTSEL will cause V_{OUT} to reset to the mid-scale code (8000H) and a LOW value will reset V_{OUT} to min-scale (0000H). A change in the state of the RSTSEL input while $\overline{\text{RST}}$ is LOW will cause a corresponding change in the reset command selected internally and consequently change the output value of V_{OUT} of the DAC. Note that a valid reset signal also resets the input register of the DAC to the value specified by the state of RSTSEL .

GAIN AND OFFSET CALIBRATION

The architecture of the DAC7741 is designed in such a way as to allow for easily configurable offset and gain calibration using a minimum of external components. The DAC7741 has built-in feedback resistors and output amplifier summing points brought out of the package in order to make the absolute calibration possible. Figures 5 and 6 illustrate the relationship of offset and gain adjustments for the DAC7741 in a unipolar configuration and in a bipolar configuration, respectively.

When calibrating the DAC output, offset should be adjusted first to avoid first order interaction of adjustments. In unipolar mode, the DAC7741 offset is adjusted from code 0000H and for either bipolar mode, offset adjustments are made at code 8000H. Gain adjustment can then be made at code FFFFH for each configuration, where the output of the DAC should be at +10V for the 0V to +10V – 1LSB or $\pm 10\text{V}$ output range and +5V – 1LSB for the $\pm 5\text{V}$ output range. Figure 7 shows the generalized external offset and gain adjustment circuitry using potentiometers.

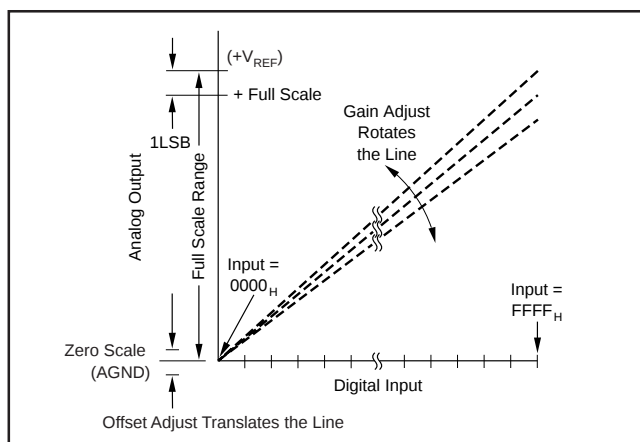


FIGURE 5. Relationship of Offset and Gain Adjustments for $V_{\text{OUT}} = 0\text{V}$ to +10V Output Configuration.

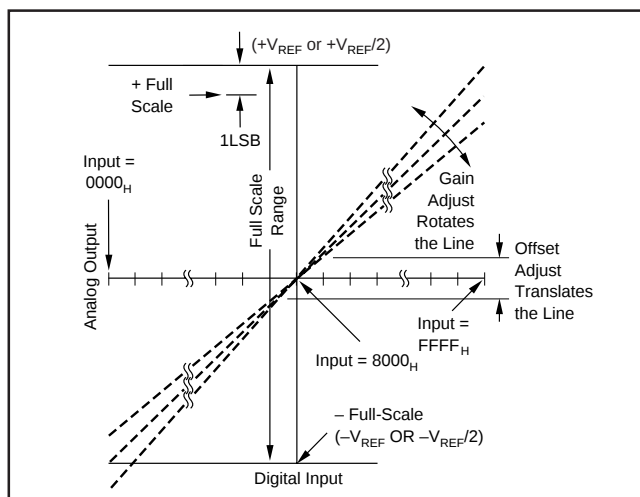


FIGURE 6. Relationship of Offset and Gain Adjustments for $V_{\text{OUT}} = -10\text{V}$ to +10V Output Configuration. (Same theory applies for $V_{\text{OUT}} = -5\text{V}$ to +5V).

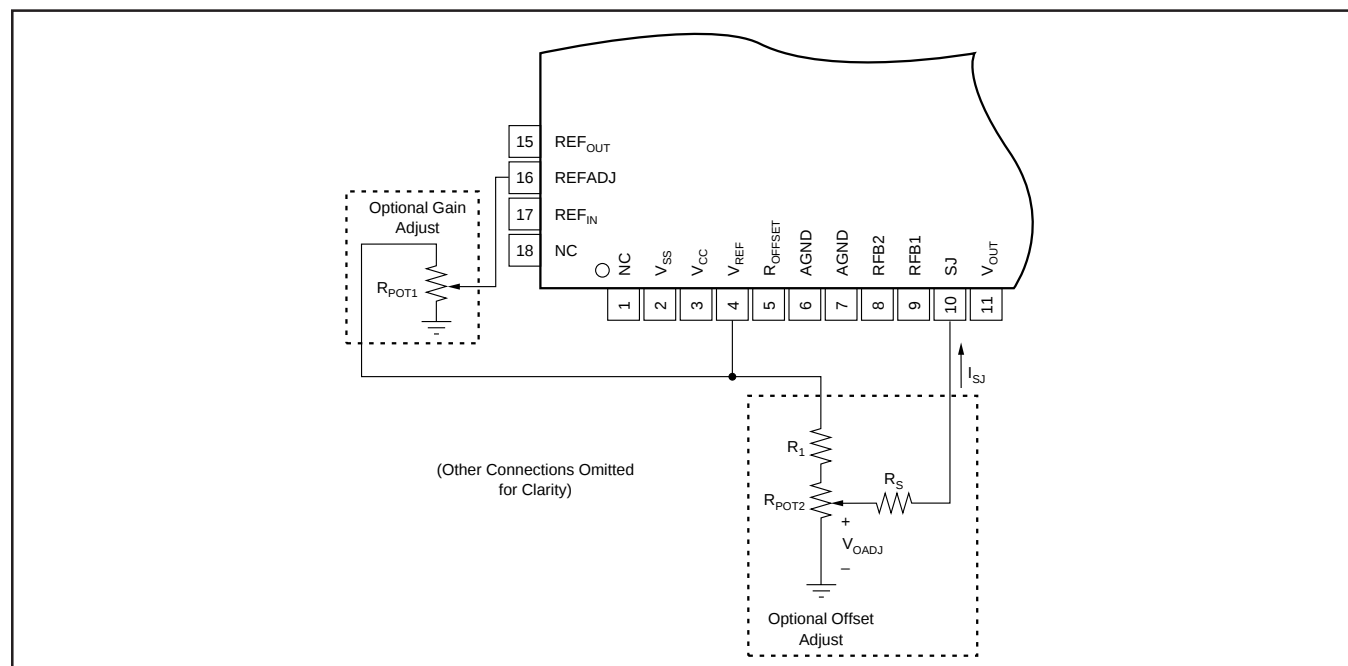


FIGURE 7. Generalized External Calibration Circuitry for Gain and Symmetrical Offset Adjustment.

OFFSET ADJUSTMENT

Offset adjustment is accomplished by introducing a small current into the summing junction (SJ) of the DAC7741. The voltage at SJ, or V_{SJ} , is dependent on the output configuration of the DAC7741. See Table IV for the required pin strapping for a given configuration and the nominal values of V_{SJ} for each output range.

REFERENCE CONFIGURATION	OUTPUT CONFIGURATION	PIN STRAPPING			$V_{SJ}^{(1)}$
		R _{OFFSET}	RFB1	RFB2	
Internal Reference	0V to +10V	to V _{REF}	to V _{OUT}	to V _{OUT}	+5V
	-10V to +10V	NC	NC	to V _{OUT}	+3.333V
	-5V to +5V	to AGND	to V _{OUT}	to V _{OUT}	+1.666V
External Reference	0V to V _{REF}	to V _{REF}	to V _{OUT}	to V _{OUT}	V _{REF} /2
	-V _{REF} to V _{REF}	NC	NC	to V _{OUT}	V _{REF} /3
	-V _{REF} /2 to V _{REF} /2	to AGND	to V _{OUT}	to V _{OUT}	V _{REF} /6

NOTE: (1) Voltage measured at V_{SJ} for a given configuration.

TABLE IV. Nominal V_{SJ} vs. V_{OUT} and Reference Configuration.

The current level required to adjust the DAC7741 offset can be created by using a potentiometer divider as shown in Figure 7. Another alternative is to use a unipolar DAC in order to apply a voltage, V_{OAdj} , to the resistor R_S . A $\pm 2\mu A$ current range applied to SJ will ensure offset adjustment coverage of the $\pm 0.1\%$ maximum offset specification of the DAC7741.

When in a unipolar configuration ($V_{SJ} = 5V$), only a single resistor, R_S , is needed for symmetrical offset adjustment with a 0V to 10V V_{OAdj} range. When in one of the two bipolar configurations, V_{SJ} is either +3.333V ($\pm 10V$ range) or +1.666V ($\pm 5V$ range), and circuit values chosen to match those given in Table V will provide symmetrical offset adjust. Please refer to Figure 7 for component configuration.

OUTPUT CONFIGURATION	R _{POT2}	R ₁	R _S	I _{SJ} RANGE	NOMINAL OFFSET ADJUSTMENT
0V to +10V	10K	0	2.5M	$\pm 2\mu A$	$\pm 25mV$
-10V to +10V	10K	5K	1.5M	$\pm 2.2\mu A$	$\pm 55mV$
-5V to +5V	10K	20K	1M	$\pm 1.7\mu A$	$\pm 21mV$

TABLE V. Recommended External Component Values for Symmetrical Offset Adjustment ($V_{REF} = 10V$).

Figure 8 illustrates the typical and minimum offset adjustment ranges provided by forcing a current at SJ for a given output voltage configuration.

GAIN ADJUSTMENT

When using the internal reference of the DAC7741, gain adjustment is performed by adjusting the internal reference voltage via the reference adjust pin, REFADJ. The effect of a reference voltage change on the gain of the DAC output can be seen in the generic equation (for unipolar configuration):

$$V_{OUT} = V_{REFIN} \cdot (N/65536)$$

where N is represented in decimal format and ranges from 0 to 65535.

REFADJ can be driven by a low impedance voltage source such as a unipolar, 0V to +10V DAC or a potentiometer (less

than 100k Ω) as shown in Figure 7. Since the input impedance of REFADJ is typically 50k Ω , the smaller the resistance of the potentiometer, the more linear the adjustment will be. A 10k Ω potentiometer is suggested if linearity of the reference adjustment is of concern.

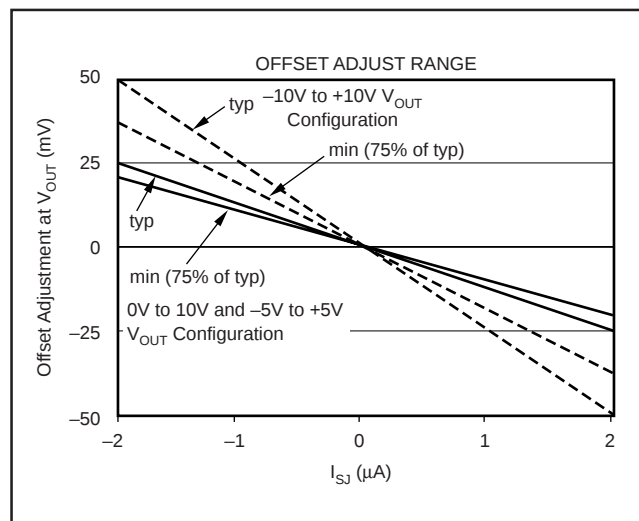


FIGURE 8. Offset Adjustment Transfer Characteristic.

When the DAC7741 internal reference is not used, gain adjustments can be made via trimming the external reference applied to the DAC at REF_{IN}. This can be accomplished through using a potentiometer, unipolar DAC, or other means of precision voltage adjustment to control the voltage presented to the DAC7741 by the external reference. Figure 9 and Table VI summarize the range of adjustment of the internal reference via REFADJ.

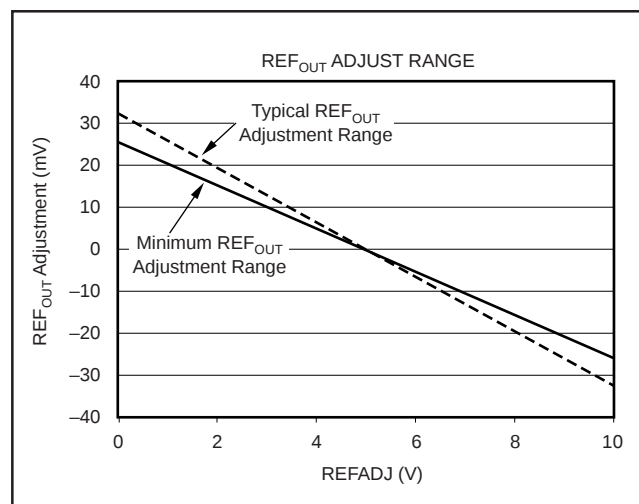


FIGURE 9. Internal Reference Adjustment Transfer Characteristic.

VOLTAGE AT REFADJ	REF _{OUT} VOLTAGE
REFADJ = 0V	10V + 25mV (min)
REFADJ = 5V or NC ⁽¹⁾	10V
REFADJ = 10V	10V - 25mV (max)

NOTE: "NC" is "Not Connected"

TABLE VI. Minimum Internal Reference Adjustment Range.

NOISE PERFORMANCE

Increased noise performance of the DAC output can be achieved by filtering the voltage reference input to the DAC7741. Figure 10 shows a typical internal reference filter schematic. A low-pass filter applied between the REF_{OUT} and REF_{IN} pins can increase noise immunity at the DAC and output amplifier. The REF_{OUT} pin can source a maximum of 50µA so care should be taken in order to avoid overloading the internal reference output.

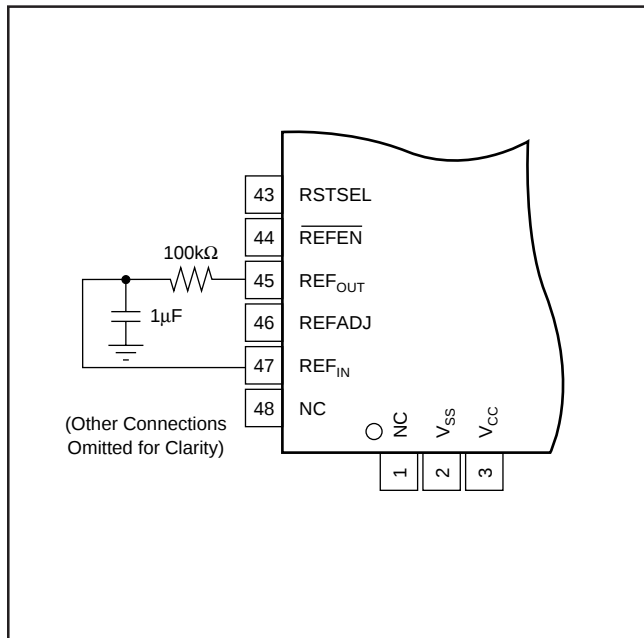


FIGURE 10. Internal Reference Filter.

LAYOUT

A precision analog component requires careful layout, adequate bypassing, and clean, well-regulated power supplies. The DAC7741 offers separate digital and analog supplies, as it will often be used in close proximity with digital logic, microcontrollers, microprocessors, and digital signal processors. The more digital logic present in the design and the higher the switching speed, the more important it will become to separate the analog and digital ground and supply planes at the device.

Since the DAC7741 has both analog and digital ground pins, return currents can be better controlled and have less effect on the DAC output error. Ideally, AGND would be connected directly to an analog ground plane and DGND to the digital ground plane. The analog ground plane would be separate from the ground connection for the digital components until they were connected at the power entry point of the system.

The voltages applied to V_{CC} and V_{SS} should be well regulated and low noise. Switching power supplies and dc/dc converters will often have high-frequency glitches or spikes riding on the output voltage. In addition, digital components can create similar high-frequency spikes as their internal logic switches states. This noise can easily couple into the DAC output voltage through various paths between the power connections and analog output.

In addition, a 1µF to 10µF bypass capacitor in parallel with a 0.1µF bypass capacitor is strongly recommended for each supply input. In some situations, additional bypassing may be required, such as a 100µF electrolytic capacitor or even a "Pi" filter made up of inductors and capacitors—all designed to essentially low-pass filter the analog supplies, removing any high frequency noise components.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
DAC7741YB/250	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	-40 to 85	DAC7741Y B
DAC7741YB/250.A	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	-40 to 85	DAC7741Y B
DAC7741YC/250	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	-40 to 85	DAC7741Y C
DAC7741YC/250.A	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	-40 to 85	DAC7741Y C
DAC7741YL/250	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	0 to 70	DAC7741Y L
DAC7741YL/250.A	Active	Production	LQFP (PT) 48	250 SMALL T&R	Yes	Call TI	Level-3-260C-168 HR	0 to 70	DAC7741Y L

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



LQFP - 1.6 mm max height

Technical drawing of a square electronic component, likely a microchip, showing dimensions and feature callouts.

Top View Dimensions:

- Overall width: 9.2
- Overall height: 9.2
- Distance from top edge to center of top pins: 8.8
- Distance from left edge to center of left pins: 7.2
- Distance from right edge to center of right pins: 6.8
- Distance from bottom edge to center of bottom pins: 7.2
- Distance from left edge to center of left pins: 6.8

Pin Dimensions:

- Pin pitch: 0.5
- Pin width: 0.27
- Pin thickness: 0.17
- Pin length: 48X
- Pin spacing: 44X
- Pin width: 4X
- Pin thickness: 5.5

Feature Callouts:

- SEE DETAIL A:** Points to a circular feature on the top surface.
- 1.6 MAX:** Dimension for the circular feature.
- SEATING PLANE:** Dimension for the bottom surface.
- 0.1 C:** Dimension for the bottom surface.
- 0.25 GAGE PLANE:** Dimension for the bottom surface.
- 0.75 0.45:** Dimension for the bottom surface.
- 1.45 1.35:** Dimension for the bottom surface.
- 0.05 MIN:** Dimension for the bottom surface.
- 0° - 7°:** Angle for the bottom surface.

Detail A: A circular feature on the top surface, shown in a separate view.

NOTES:

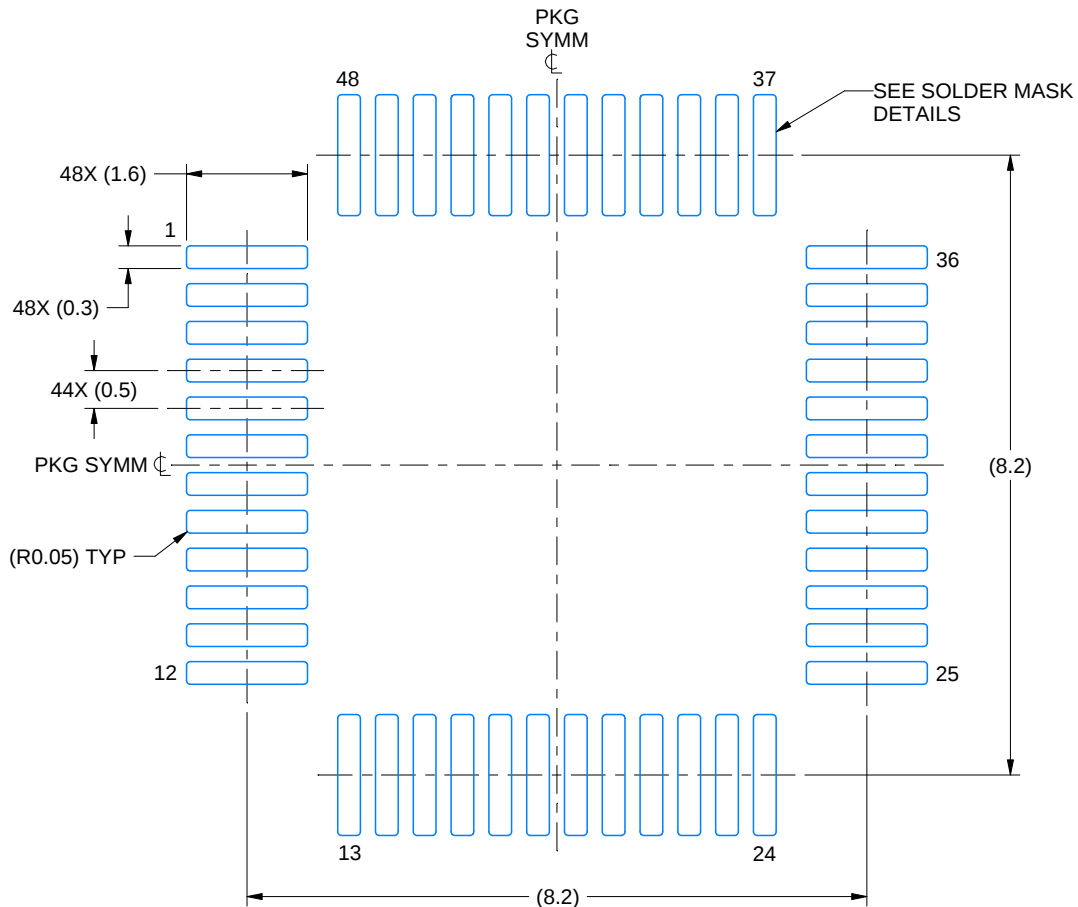
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC registration MS-026.
4. This may also be a thermally enhanced plastic package with leads connected to the die pads.

EXAMPLE BOARD LAYOUT

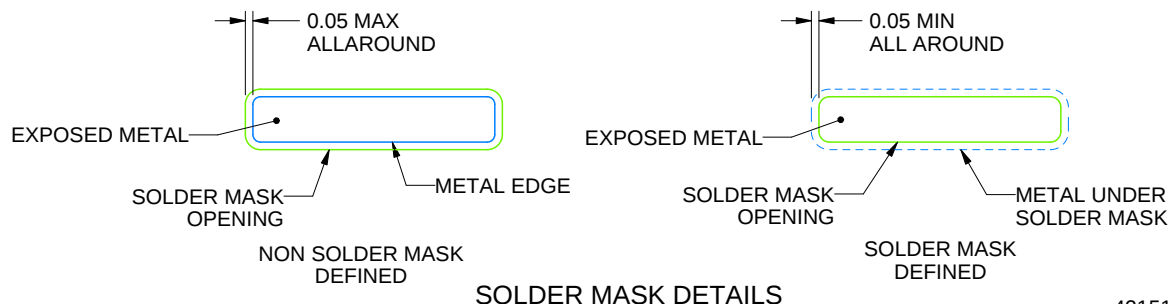
PT0048A

LQFP - 1.6 mm max height

LOW PROFILE QUAD FLATPACK



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE 10.000



SOLDER MASK DETAILS

4215159/B 11/2023

NOTES: (continued)

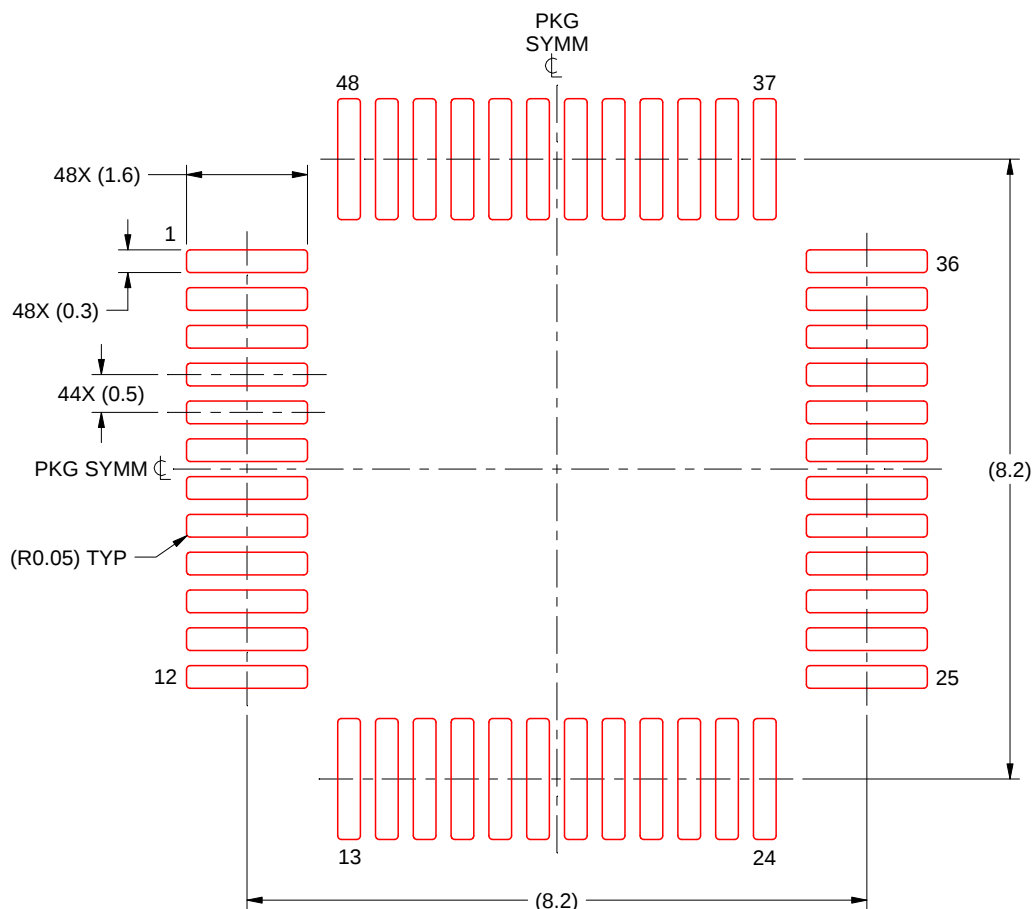
5. Publication IPC-7351 may have alternate designs.
6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PT0048A

LQFP - 1.6 mm max height

LOW PROFILE QUAD FLATPACK



SOLDER PASTE EXAMPLE
BASED ON 0.1 mm THICK STENCIL
SCALE: 10X

4215159/B 11/2023

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you fully indemnify TI and its representatives against any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#), [TI's General Quality Guidelines](#), or other applicable terms available either on [ti.com](https://www.ti.com) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products. Unless TI explicitly designates a product as custom or customer-specified, TI products are standard, catalog, general purpose devices.

TI objects to and rejects any additional or different terms you may propose.

Copyright © 2025, Texas Instruments Incorporated

Last updated 10/2025