

Low Voltage, Stereo DAC with Headphone Amp

Features

- 16-Pin TSSOP package
- 1.8 to 3.3 Volt supply
- 24-Bit conversion / 96 kHz sample rate
- 94 dB dynamic range at 3 V supply
- -85 dB THD+N at 1.8 V supply
- Low power consumption
- Digital volume control
 - -96 dB attenuation, 1 dB step size
- Digital bass and treble boost
 - -Selectable corner frequencies
 - -Up to 12 dB boost in 1 dB increments
- Peak signal limiting to prevent clipping
- De-emphasis for 32 kHz, 44.1 kHz, and 48 kHz
- Headphone amplifier
 - —up to 25 mW_{rms} power output into 16 Ω load*
 - -25 dB analog attenuation and mute
 - —Zero crossing click free level transitions
- ATAPI mixing functions

* 1 kHz sine wave at 3.3V supply

Description

The CS43L43 is a complete stereo digital-to-analog output system including interpolation, 1-bit D/A conversion, analog filtering, volume control, and a headphone amplifier, in a 16-pin TSSOP package.

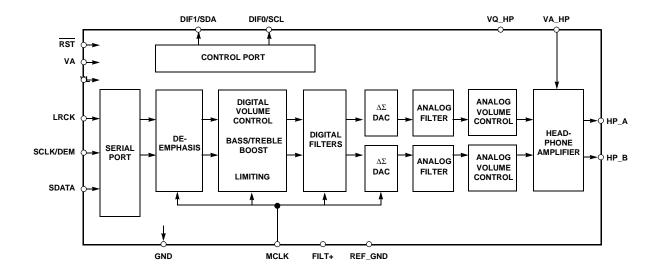
The CS43L43 is based on delta-sigma modulation, where the modulator output controls the reference voltage input to an ultra-linear analog low-pass filter. This architecture allows infinite adjustment of the sample rate between 2 kHz and 100 kHz simply by changing the master clock frequency.

The CS43L43 contains on-chip digital bass and treble boost, peak signal limiting and de-emphasis. The CS43L43 operates from a +1.8 V to +3.3 V supply and consumes only 16 mW of power with a 1.8 V supply. These features are ideal for portable CD, MP3 and MD players and other portable playback systems that require extremely low power consumption.

ORDERING INFORMATION

CS43L43-KZ -10 to 70 °C CDB43L43

16-pin TSSOP Evaluation Board



Preliminary Product Information

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1.0 CHARACTERISTICS AND SPECIFICATIONS

ANALOG CHARACTERISTICS ($T_A = 25$ °C; GND = 0 V; Logic "1" = VL = 1.8 V; Logic "0" = GND = 0 V; Full-Scale Output Sine Wave, 997 Hz, MCLK = 12.288 MHz, Measurement Bandwidth 10 Hz to 20 kHz, unless otherwise specified; Fs for Base-rate Mode = 48 kHz, SCLK = 3.072 MHz; Fs for High-Rate Mode = 96 kHz, SCLK = 6.144 MHz. Test load $R_I = 16 \Omega$, $C_I = 10 pF$ (See Figure 15))

				Bas	e-rate M	ode	Hig	h-Rate N	/lode	
ĺ	Parameter		Symbol	Min	Тур	Max	Min	Тур	Max	Unit
Headphone Outp	ut Dynamic Per	formance fo	r VA = VA	_ <i>HP</i> = 1	.8 V					
Dynamic Range		(Note 1)								
	18 to 24-Bit	unweighted		TBD	88	-	TBD	89	-	dB
		A-Weighted		TBD	91	-	TBD	92	-	dB
	16-Bit	unweighted		-	86	-	-	87	-	dB
		A-Weighted		-	89	-	-	90	-	dB
Total Harmonic Dis	stortion + Noise	(Note 1)	THD+N							
	18 to 24-Bit	0 dB		-	-82	TBD	-	-85	TBD	dB
		-20 dB		-	-68	-	-	-69	-	dB
		-60 dB		-	-28	-	-	-29	-	dB
	16-Bit	0 dB		-	-80	-	-	-83	-	dB
		-20 dB		-	-66	-	-	-67	-	dB
		-60 dB		-	-26	-	-	-27	-	dB
Interchannel Isolat	ion	(1 kHz)		-	66	-	-	66	-	dB
Headphone Outp	ut Dynamic Per	formance fo	r VA = VA	$_HP = 3$	3.0 V					
Dynamic Range.		(Note 1)								
	18 to 24-Bit.	unweighted		TBD	91	-	TBD	92	-	dB
		A-Weighted		TBD	93	-	TBD	94	-	dB
	16-Bit.	unweighted		-	89	-	-	90	-	dB
		A-Weighted		-	91	-	-	92	-	dB
Total Harmonic Dis	stortion + Noise.	(Note 1)	THD+N							
	18 to 24-Bit.	0 dB		-	-76	TBD	-	-73	TBD	dB
		-20 dB		-	-71	-	-	-72	-	dB
		-60 dB		-	-31	-	-	-32	-	dB
	16-Bit.	0 dB		-	-78	-	-	-78	-	dB
		-20 dB		-	-69	-	-	-70	-	dB
		-60 dB		-	-29	-	-	-30	-	dB
Interchannel Isolat	ion.	(1 kHz)		-	66	-	-	66	-	dB

Notes: 1. One-half LSB of triangular PDF dither is added to data.



ANALOG CHARACTERISTICS (Continued)

Parameters		Symbol	Min	Тур	Max	Units
Analog Output						
Full Scale Headphone Output Voltage			TBD	0.55 x VA	TBD	Vpp
Headphone Output Quiescent Voltage		V_{Q_HP}	-	0.5 x VA_HP	-	VDC
Interchannel Gain Mismatch			-	0.1	-	dB
Gain Drift			-	100	-	ppm/°C
Maximum Headphone Output	VA=VA_HP=1.8V	I _{HP}	-	31	-	mA
AC-Current	VA=VA_HP=3.0V		-	52	-	mA

			Base-rate Mode			High-Rate Mode			
Parameter	•	Symbol	Min	Тур	Max	Min	Тур	Max	Unit
Combined Digital and On-	chip Analog Filt	er Respoi	nse (No	ote 2)					
Passband to	(Note 3) o -0.05 dB corner		0	-	.4535	-	-	-	Fs
	to -0.1 dB corner to -3 dB corner		- 0	- -	- .4998	0 0	-	.4426 .4984	Fs Fs
Frequency Response 10 Hz	to 20 kHz (Note 4)		02	-	+.08	0	-	+0.11	dB
StopBand			.5465	-	-	.577	-	-	Fs
StopBand Attenuation	(Note 5)		50	-	-	55	-	-	dB
Group Delay		tgd	-	9/Fs	-	-	4/Fs	-	S
Passband Group Delay Dev	viation 0 - 40 kHz 0 - 20 kHz			- ±0.36/Fs	-	-	±1.39/Fs ±0.23/Fs	-	s s
De-emphasis Error (Relative to 1 kHz)	Fs = 32 kHz Fs = 44.1 kHz Fs = 48 kHz		- - -	- - -	+.2/1 +.05/14 +0/22		(Note 6)		dB dB dB

Notes: 2. Filter response is not tested but is guaranteed by design.

- 3. Response is clock dependent and will scale with Fs. Note that the response plots (Figures 7-14) have been normalized to Fs and can be de-normalized by multiplying the X-axis scale by Fs.
- 4. Referenced to a 1 kHz, full-scale sine wave.
- 5. For Base-Rate Mode, the measurement bandwidth is 0.5465 Fs to 3 Fs. For High-Rate Mode, the measurement bandwidth is 0.577 Fs to 1.4 Fs.
- 6. De-emphasis is not available in High-Rate Mode.



POWER AND THERMAL CHARACTERISTICS GND = 0 V (All voltages with respect to ground. All measurements taken with all zeros input and open outputs, unless otherwise specified.)

			Bas	e-rate M	ode	
Parameters		Symbol	Min	Тур	Max	Units
Power Supplies						
Power Supply Current-	VA=1.8V	I _A	-	7.3	-	mA
Normal Operation	VA_HP=1.8V	I _{A_HP}	-	1.5	-	mA
	VL=1.8V	$I_D_L^-$	-	4	-	μΑ
Power Supply Current-	VA=1.8V	I _A	-	TBD	-	μA
Power Down Mode (Note 7)	VA_HP=1.8V	I _{A_HP}	-	TBD	-	μA
	VL=1.8V	$I_D_L^-$	-	TBD	-	μΑ
Power Supply Current-	VA=3.0V	I _A	-	10.5	-	mA
Normal Operation	VA_HP=3.0V	I _{A_HP}	-	1.5	-	mA
	VL=3.0V	$I_D_L^-$	-	9.3	-	μΑ
Power Supply Current-	VA=3.0V	I _A	-	TBD	-	μΑ
Power Down Mode (Note 7)	VA_HP=3.0V	I_{A_HP}	-	TBD	-	μΑ
	VL=3.0V	$I_D_L^-$	-	TBD	-	μΑ
Total Power Dissipation-	All Supplies=1.8V		-	16	TBD	mW
Normal Operation	All Supplies=3.0V		-	36	TBD	mW
Max Headphone Power Dissipation						
With Full-scale Output	VA_HP=1.8V		-	TBD	-	mW
and 16ohm Load	VA_HP=3.0V		-	TBD	-	mW
Package Thermal Resistance		θ_{JA}	-	75	-	°C/Watt
Power Supply Rejection Ratio (Note 8)	1 kHz	PSRR	-	60	-	dB
	60 Hz		-	40	-	dB

Notes: 7. Power Down Mode is defined as $\overline{RST} = LO$ with all clocks and data lines held static.

^{8.} Valid with the recommended capacitor values on FILT+ and VQ_HP as shown in Figure 5. Increasing the capacitance will also increase the PSRR. NOTE: Care should be taken when selecting capacitor type, as any leakage current in excess of 1.0 μA will cause degradation in analog performance.



DIGITAL CHARACTERISTICS (T_A = 25°C; VL = 1.7V - 3.6V; GND = 0 V)

Parameters	Symbol	Min	Тур	Max	Units
High-Level Input Voltage	V_{IH}	0.7 x VL	-	-	V
Low-Level Input Voltage	V _{IL}	-	-	0.3 x VL	V
Input Leakage Current	l _{in}	-	-	±10	μΑ
Input Capacitance		-	8	-	pF

ABSOLUTE MAXIMUM RATINGS (GND = 0V; all voltages with respect to ground.)

Parameters	Symbol	Min	Max	Units
DC Power Supplies: Positive Analog	VA	-0.3	4.0	V
Headphone	VA_HP	-0.3	4.0	V
Digital I/O	VL	-0.3	4.0	V
Input Current, Any Pin Except Supplies	I _{in}	-	±10	mA
Digital Input Voltage	V _{IND}	-0.3	VL + 0.4	V
Ambient Operating Temperature (power applied)	T _A	-55	125	°C
Storage Temperature	T _{stg}	-65	150	°C

WARNING: Operation at or beyond these limits may result in permanent damage to the device. Normal operation is not guaranteed at these extremes.

RECOMMENDED OPERATING CONDITIONS (GND = 0V; all voltages with respect to ground.)

Parameters	Symbol	Min	Тур	Max	Units
Ambient Temperature	T _A	-10	-	70	°C
DC Power Supplies: Positive Analog	VA	1.7	-	3.6	V
Headphone (Note 9)	VA_HP	0.9	-	3.6	V
Digital I/O	VL	1.7	-	3.6	V

^{9.} To prevent clipping the outputs, VA_HP_{MIN} is limited by the Full-Scale Output Voltage V_{FS_HP} , where VA_HP must be 200 mV greater than V_{FS_HP} . However, if distortion is not a concern, VA_HP may be as low as 0.9 V at any time.

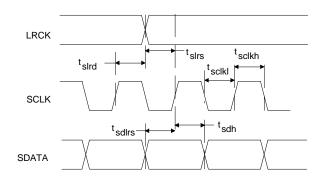


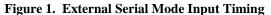
SWITCHING CHARACTERISTICS ($T_A = -10$ to 70° C; VA = 1.7V - 3.6V; Inputs: Logic 0 = GND, Logic 1 = VL, CL = 20pF)

Parameters	Symbol	Min	Тур	Max	Units
Input Sample Rate Base Rate Mode	Fs	2	-	50	kHz
High Rate Mode	Fs	50	-	100	kHz
MCLK Pulse Width High MCLK/LRCK = 1024		7	-	-	ns
MCLK Pulse Width High MCLK/LRCK = 1024		7	-	-	ns
MCLK Pulse Width High MCLK/LRCK = 768		10	-	-	ns
MCLK Pulse Width High MCLK/LRCK = 768		10	-	-	ns
MCLK Pulse Width High MCLK/LRCK = 512		15	-	1	ns
MCLK Pulse Width Low MCLK/LRCK = 512		15	-	-	ns
MCLK Pulse Width High MCLK / LRCK = 384 or 192		25	-	1	ns
MCLK Pulse Width Low MCLK / LRCK = 384 or 192		25	-	1	ns
MCLK Pulse Width High MCLK / LRCK = 256 or 128		35	-	1	ns
MCLK Pulse Width Low MCLK / LRCK = 256 or 128		35	-	1	ns
External SCLK Mode			-		
LRCK Duty Cycle (External SCLK only)		40	50	60	%
SCLK Pulse Width Low	t _{sclkl}	20	-	-	ns
SCLK Pulse Width High	t _{sclkh}	20	-	-	ns
SCLK Period Base Rate Mode	t _{sclkw}	1 (128)Fs	-	-	ns
High Rate Mode	t _{sclkw}	1 (64)Fs	-	-	ns
SCLK rising to LRCK edge delay	t _{slrd}	20	-	1	ns
SCLK rising to LRCK edge setup time	t _{slrs}	20	-	1	ns
SDATA valid to SCLK rising setup time	t _{sdlrs}	20	-	1	ns
SCLK rising to SDATA hold time	t _{sdh}	20	-	1	ns
Internal SCLK Mode					
LRCK Duty Cycle (Internal SCLK only) (Note 10)		-	50	-	%
SCLK Period	t _{sclkw}	1 SCLK	-	-	ns
SCLK rising to LRCK edge	t _{sclkr}	-	tsclkw 2	-	μs
SDATA valid to SCLK rising setup time	t _{sdlrs}	1 (512)Fs + 10	-	-	ns
SCLK rising to SDATA hold time Base Rate Mode	t _{sdh}	1 (512)Fs + 15	-	-	ns
High Rate Mode	t _{sdh}	1 (384)Fs + 15	-	1	ns

Notes: 10. In Internal SCLK Mode, the duty cycle must be 50% \pm /- 1/2 MCLK Period.







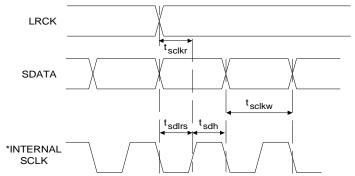
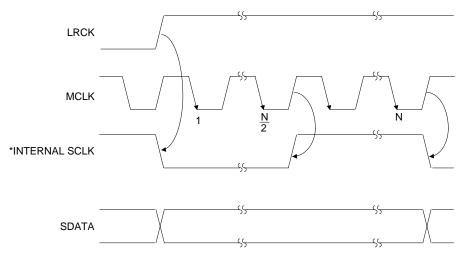


Figure 2. Internal Serial Mode Input Timing *The SCLK pulses shown are internal to the CS43L43.



* The SCLK pulses shown are internal to the CS43L43.

N equals MCLK divided by SCLK



SWITCHING CHARACTERISTICS - CONTROL PORT - TWO-WIRE MODE

 $(T_A = 25 \, ^{\circ}\text{C}; \, VL = 1.7 \text{V} - 3.6 \text{V}; \, \text{Inputs: Logic 0 = GND, Logic 1 = VL, C}_L = 30 \, \text{pF})$

Parameter	Symbol	Min	Max	Unit
Two-Wire Mode (Note 11)		1	1	•
SCL Clock Frequency	f _{scl}	-	100	kHz
RST Rising Edge to Start	t _{irs}	500	-	ns
Bus Free Time Between Transmissions	t _{buf}	4.7	-	μs
Start Condition Hold Time (prior to first clock pulse)	t _{hdst}	4.0	-	μs
Clock Low time	t _{low}	4.7	-	μs
Clock High Time	t _{high}	4.0	-	μs
Setup Time for Repeated Start Condition	t _{sust}	4.7	-	μs
SDA Hold Time from SCL Falling (Note 12)	t _{hdd}	0	-	μs
SDA Setup time to SCL Rising	t _{sud}	250	-	ns
Rise Time of SCL	t _{rc}	-	25	ns
Fall Time of SCL	t _{fc}	-	25	ns
Rise Time SDA	t _{rd}	-	1	μs
Fall Time of SDA	t _{fd}	-	300	ns
Setup Time for Stop Condition	t _{susp}	4.7	-	μs

- Notes: 11. The Two-Wire Mode is compatible with the I^2C protocol.
 - 12. Data must be held for sufficient time to bridge the transition time, t_{fc} , of SCL.

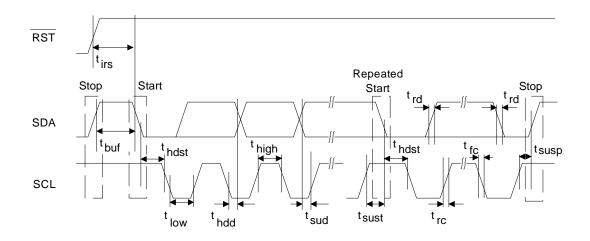


Figure 4. Control Port Timing - Two-Wire Mode



2.0 TYPICAL CONNECTION DIAGRAM

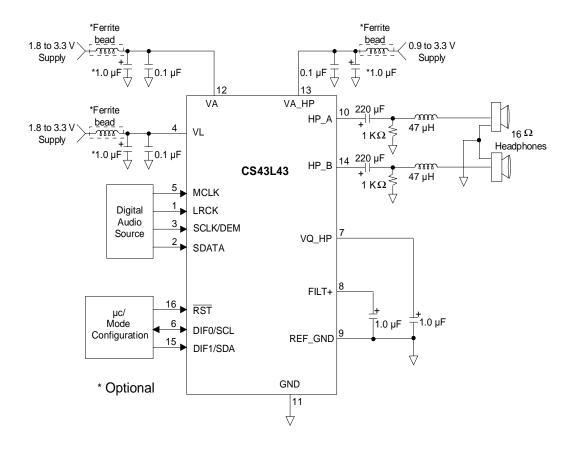


Figure 5. Typical Connection Diagram



3.0 REGISTER QUICK REFERENCE

Addr	Function	7	6	5	4	3	2	1	0
0h	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
	default	0	0	0	0	0	0	0	0
1h	Power and Muting Control	AMUTE	SZC1	SZC0	POR	Reserved	Reserved	PDN	Reserved
	default	1	1	0	1	0	0	1	0
2h	Channel A Analog Attenuation Control	VOLA7	VOLA6	VOLA5	VOLA4	VOLA3	VOLA2	VOLA1	VOLA0
	default	0	0	0	0	0	0	0	0
3h	Channel B Analog Attenuation Control	VOLB7	VOLB6	VOLB5	VOLB4	VOLB3	VOLB2	VOLB1	VOLB0
	default	0	0	0	0	0	0	0	0
4h	Channel A Digital Volume Control	DVOLA7	DVOLA6	DVOLA5	DVOLA4	DVOLA3	DVOLA2	DVOLA1	DVOLA0
	default	0	0	0	0	0	0	0	0
5h	Channel B Digital Volume Control	DVOLB7	DVOLB6	DVOLB5	DVOLB4	DVOLB3	DVOLB2	DVOLB1	DVOLB0
	default	0	0	0	0	0	0	0	0
6h	Tone Control	BB3	BB2	BB1	BB0	TB3	TB2	TB1	TB0
	default	0	0	0	0	0	0	0	0
7h	Mode Control	BBCF1	BBCF0	TBCF1	TBCF0	A=B	DEM1	DEM0	VCBYP
	default	0	0	0	0	0	0	0	0
8h	Limiter Attack Rate	ARATE7	ARATE6	ARATE5	ARATE4	ARATE3	ARATE2	ARATE1	ARATE0
	default	0	0	1	0	0	0	0	0
9h	Limiter Release Rate	RRATE7	RRATE6	RRATE5	RRATE4	RRATE3	RRATE2	RRATE1	RRATE0
	default	0	0	0	1	0	0	0	0
Ah	Volume and Mixing Control	TC1	TC0	TC_EN	LIM_EN	ATAPI3	ATAPI2	ATAPI1	ATAPI0
	default	0	0	0	0	1	0	0	1
Bh	Mode Control 2	MCLKDIV	Reserved	Reserved	Reserved	Reserved	DIF2	DIF1	DIF0
	default	0	0	0	0	0	0	0	0



4.0 REGISTER DESCRIPTION

4.1 POWER AND MUTING CONTROL (ADDRESS 01H)

7	6	5	4	3	2	1	0
AMUTE	SZC1	SZC0	POR	RESERVED	RESERVED	PDN	RESERVED
1	1	0	1	0	0	1	0

4.1.1 AUTO-MUTE (AMUTE)

Default = 1

0 - Disabled

1 - Enabled

Function:

The Digital-to-Analog converter output will mute following the reception of 8192 consecutive audio samples of static 0 or -1. A single sample of non-static data will release the mute. Detection and muting is done independently for each channel. The quiescent voltage on the output will be retained and the Mute Control pin will go active during the mute period. The muting function is affected, similar to volume control changes, by the Soft and Zero Cross bits in the Power and Muting Control register.

4.1.2 SOFT RAMP AND ZERO CROSS CONTROL (SZC)

Default = 10

00 - Immediate Change

01 - Zero Cross Digital and Analog

10 - Ramped Digital and Analog

11 - Reserved

Function:

Immediate Change

When Immediate Change is selected all level changes will take effect immediately in one step.

Zero Cross Digital and Analog

Zero Cross Enable dictates that signal level changes, either by attenuation changes or muting, will occur on a signal zero crossing to minimize audible artifacts. The requested level change will occur after a timeout period of 512 sample periods (10.7 ms at 48 kHz sample rate) if the signal does not encounter a zero crossing. The zero cross function is independently monitored and implemented for each channel.

Ramped Digital and Analog

Soft Ramp allows digital level changes, both muting and attenuation, to be implemented by incrementally ramping, in 1/8 dB steps, from the current level to the new level at a rate of 1dB per 8 left/right clock periods. Analog level changes will occur in 1 dB steps on a signal zero crossing. The analog level change will occur after a timeout period of 512 sample periods (10.7 ms at 48 kHz sample rate) if the signal does not encounter a zero crossing. The zero cross function is independently monitored and implemented for each channel.



4.1.3 POPGUARD® TRANSIENT CONTROL (POR)

Default - 1

0 - Disabled

1 - Enabled

Function:

The Popguard® Transient Control allows the quiescent voltage to slowly ramp to and from 0 volts to the quiescent voltage during power-on or power-off when this feature is enabled. Please see section 6.5 for implementation details.

4.1.4 POWER DOWN (PDN)

Default = 1

0 - Disabled

1 - Enabled

Function:

The entire device will enter a low-power state whenever this function is enabled, and the contents of the control registers are retained in this mode. The power-down bit defaults to 'enabled' on power-up and must be disabled before normal operation will begin.



4.2 CHANNEL A ANALOG ATTENUATION CONTROL (ADDRESS 02H) (VOLA)

4.3 CHANNEL B ANALOG ATTENUATION CONTROL (ADDRESS 03H) (VOLB)

7	6	5	4	3	2	1	0
VOLx7	VOLx6	VOLx5	VOLx4	VOLx3	VOLx2	VOLx1	VOLx0
0	0	0	0	0	0	0	0

Default = 0 dB (No attenuation)

Function:

The Analog Attenuation Control operates independently from the Digital Volume Control. The Analog Attenuation Control registers allow the user to attenuate the headphone output signal in 1 dB increments from 0 to -25 dB, using the analog volume control. Attenuation settings are decoded as shown in Table 1, using a 2's complement code. The volume changes are implemented as dictated by the Soft and Zero Cross bits in the Power and Muting Control register. All volume settings greater than zero are interpreted as zero.

Binary Code	Decimal Value	Volume Setting
00000000	0	0 dB
11110110	-10	-10 dB
11110001	-15	-15 dB

Table 1. Example Analog Volume Settings

4.4 CHANNEL A DIGITAL VOLUME CONTROL (ADDRESS 04H) (DVOLA)

4.5 CHANNEL B DIGITAL VOLUME CONTROL (ADDRESS 05H) (DVOLB)

7	6	5	4	3	2	1	0
DVOLx7	DVOLx6	DVOLx5	DVOLx4	DVOLx3	DVOLx2	DVOLx1	DVOLx0
0	0	0	0	0	0	0	0

Default = 0 dB (No attenuation)

Function:

The Digital Volume Control allows the user to alter the signal level in 1 dB increments from +18 to -96 dB, using the Digital Volume Control. Volume settings are decoded as shown in Table 2, using a 2's complement code. The volume changes are implemented as dictated by the Soft and Zero Cross bits in the Power and Muting Control register. All volume settings less than - 96 dB are equivalent to muting the channel via the ATAPI bits (See Section 4.10.4). NOTES: Setting this register to values greater than +18 dB will cause distortion in the audio outputs.

Binary Code	Decimal Value	Volume Setting
00001010	12	+12 dB
00000111	7	+7 dB
00000000	0	0 dB
11000100	-60	-60 dB
10100110	-90	-90 dB

Table 2. Example Digital Volume Settings



4.6 TONE CONTROL (ADDRESS 06H)

7	6	5	4	3	2	1	0
BB3	BB2	BB1	BB0	TB3	TB2	TB1	TB0
0	0	0	0	0	0	0	0

4.6.1 BASS BOOST LEVEL (BB)

Default = 0 dB (No Bass Boost)

Function:

The level of the shelving bass boost filter is set by Bass Boost Level. The level can be adjusted in 1 dB increments from 0 to +12 dB of boost. Boost levels are decoded as shown in Table 3. Levels above +12 dB are interpreted as +12 dB.

Binary Code	Decimal Value	Boost Setting
0000	0	0 dB
0010	2	+2 dB
1010	6	+6 dB
1001	9	+9 dB
1100	12	+12 dB

Table 3. Example Bass Boost Settings

4.6.2 TREBLE BOOST LEVEL (TB)

Default = 0 dB (No Treble Boost)

Function:

The level of the shelving treble boost filter is set by Treble Boost Level. The level can be adjusted in 1 dB increments from 0 to +12 dB of boost. Boost levels are decoded as shown in Table 4. Levels above +12 dB are interpreted as +12 dB. NOTE: Treble Boost is not available in High-Rate Mode.

Binary Code	Decimal Value	Boost Setting
0000	0	0 dB
0010	2	+2 dB
1010	6	+6 dB
1001	9	+9 dB
1100	12	+12 dB

Table 4. Example Treble Boost Settings



4.7 MODE CONTROL (ADDRESS 07H)

7	6	5	4	3	2	1	0
BBCF1	BBCF0	TBCF1	TBCF0	A=B	DEM1	DEM0	VCBYP
0	0	0	0	0	0	0	0

4.7.1 BASS BOOST CORNER FREQUENCY (BBCF)

Default = 00

00 - 50 Hz

01 - 100 Hz

10 - 200 Hz

11 - Reserved

Function:

The bass boost corner frequency is user selectable as shown above.

4.7.2 TREBLE BOOST CORNER FREQUENCY (TBCF)

Default = 00

00 - 2 kHz

01 - 4 kHz

10 - 7 kHz

11 - Reserved

Function:

The treble boost corner frequency is user selectable as shown above. NOTE: Treble Boost is not available in High-Rate Mode.

4.7.3 CHANNEL A VOLUME = CHANNEL B VOLUME (A=B)

Default = 0

0 - Disabled

1 - Enabled

Function:

The HP_A and HP_B volume levels are independently controlled by the A and the B Channel Volume Control Bytes when this function is disabled. The volume on both HP_A and HP_B are determined by the A Channel Attenuation and Volume Control Bytes and the B Channel Bytes are ignored when this function is enabled.

4.7.4 DE-EMPHASIS CONTROL (DEM)

Default = 00

00 - Disabled

01 - 44.1 kHz

10 - 48 kHz

11 - 32 kHz

Function:

Selects the appropriate digital filter to maintain the standard 15 μ s/50 μ s digital de-emphasis filter response at 32, 44.1 or 48 kHz sample rates. (See Figure 27) NOTE: De-emphasis is not available in High-Rate Mode.



4.7.5 DIGITAL VOLUME CONTROL BYPASS (VCBYP)

Default = 0

0 - Disabled

1 - Enabled

Function:

When this function is enabled the digital volume control section is bypassed. This disables the digital volume control, muting, bass boost, treble boost, limiting and ATAPI functions. The analog attenuation control will remain functional.

4.8 LIMITER ATTACK RATE (ADDRESS 08H) (ARATE)

7	6	5	4	3	2	1	0
ARATE7	ARATE6	ARATE5	ARATE4	ARATE3	ARATE2	ARATE1	ARATE0
0	0	1	0	0	0	0	0

Default = 20h - 1 LRCK's per 1/8 dB

Function:

The limiter attack rate is user selectable. The rate is a function of sampling frequency, Fs, and the value in the Limiter Attack Rate register. Rates are calculated using the function RATE = 32/{value}. Where {value} is the decimal value in the Limiter Attack Rate register and RATE is in LRCK's per 1/8 dB of change. NOTE: A value of zero in this register is not recommended, as it will induce erratic behavior of the limiter. Use the LIM_EN bit to disable the limiter function (see Section 4.10.3).

Binary Code	Decimal Value	LRCK's per 1/8 dB
0000001	1	32
00010100	20	1.6
00101000	40	0.8
00111100	60	0.53
01011010	90	0.356

Table 5. Example Limiter Attack Rate Settings



4.9 LIMITER RELEASE RATE (ADDRESS 09H) (RRATE)

7	6	5	4	3	2	1	0
RRATE7	RRATE6	RRATE5	RRATE4	RRATE3	RRATE2	RRATE1	RRATE0
0	0	0	1	0	0	0	0

Default = 10h - 32 LRCK's per 1/8 dB

Function:

The limiter release rate is user selectable. The rate is a function of sampling frequency, Fs, and the value in Limiter Release Rate register. Rates are calculated using the function RATE = 512/{value}. Where {value} is the decimal value in the Limiter Release Rate register and RATE is in LRCK's per 1/8 dB of change. NOTE: A value of zero in this register is not recommended, as it will induce erratic behavior of the limiter. Use the LIM_EN bit to disable the limiter function (see Section 4.10.3).

Binary Code	Decimal Value	LRCK's per 1/8 dB
0000001	1	512
00010100	20	25
00101000	40	12
00111100	60	8
01011010	90	5

Table 6. Example Limiter Release Rate Settings

4.10 VOLUME AND MIXING CONTROL (ADDRESS 0AH)

7	6	5	4	3	2	1	0
TC1	TC0	TC_EN	LIM_EN	ATAPI3	ATAPI2	ATAPI1	ATAPI0
0	0	0	0	1	0	0	1

4.10.1 TONE CONTROL MODE (TC)

Default = 00

00 - All settings are taken from user registers

01 - 12 dB of Bass Boost at 100 Hz and 6 dB of Treble Boost at 7 kHz

10 - 8 dB of Bass Boost at 100 Hz and 4 dB of Treble Boost at 7 kHz

11 - 4 dB of Bass Boost at 100 Hz and 2 dB of Treble Boost at 7 kHz

Function:

The Tone Control Mode bits determine how the Bass Boost and Treble Boost features are configured. The user defined settings from the Bass and Treble Boost Level and Corner Frequency registers are used when these bits are set to '00'. Alternately, one of three pre-defined settings may be used.

4.10.2 TONE CONTROL ENABLE (TC EN)

Default = 0

0 - Disabled

1 - Enabled

Function:

The Bass Boost and Treble Boost features are active when this function is enabled.



4.10.3 PEAK SIGNAL LIMITER ENABLE (LIM EN)

Default = 0

0 - Disabled

1 - Enabled

Function:

The CS43L43 will limit the maximum signal amplitude to prevent clipping when this function is enabled. Peak Signal Limiting is performed by first decreasing the Bass and Treble Boost Levels. If the signal is still clipping, then the digital attenuation is increased. The attack rate is determined by the Limiter Attack Rate register.

Once the signal has dropped below the clipping level, the attenuation is decreased back to the user selected level and then, the Bass Boost is increased back to the user selected level. The release rate is determined by the Limiter Release Rate register. NOTE: The A=B bit should be set to '1' for optimal limiter performance.

4.10.4 ATAPI CHANNEL MIXING AND MUTING (ATAPI)

 $Default = 1001 - HP_A = L, HP_B = R (Stereo)$

Function:

The CS43L43 implements the channel mixing functions of the ATAPI CD-ROM specification. Refer to Table 7 and Figure 28 for additional information. NOTE: All mixing functions occur prior to the digital volume control.

ATAPI3	ATAPI2	ATAPI1	ATAPI0	HP_A	HP_B
0	0	0	0	MUTE	MUTE
0	0	0	1	MUTE	R
0	0	1	0	MUTE	L
0	0	1	1	MUTE	[(L+R)/2]
0	1	0	0	R	MUTE
0	1	0	1	R	R
0	1	1	0	R	L
0	1	1	1	R	[(L+R)/2]
1	0	0	0	L	MUTE
1	0	0	1	L	R
1	0	1	0	L	L
1	0	1	1	L	[(L+R)/2]
1	1	0	0	[(L+R)/2]	MUTE
1	1	0	1	[(L+R)/2]	R
1	1	1	0	[(L+R)/2]	L
1	1	1	1	[(L+R)/2]	[(L+R)/2]

Table 7. ATAPI Decode



4.11 MODE CONTROL 2 (ADDRESS 0BH)

7	6	5	4	3	2	1	0
MCLKDIV	RESERVED	RESERVED	RESERVED	RESERVED	DIF2	DIF1	DIF0
0	0	0	0	0	0	0	0

4.11.1 MASTER CLOCK DIVIDE ENABLE (MCLKDIV)

Default = 0

0 - Disabled

1 - Enabled

Function:

The MCLKDIV bit enables a circuit which divides the externally applied MCLK signal by 2 prior to all other internal circuitry. NOTE: Internal SCLK is not available when this function is enabled.

4.11.2 DIGITAL INTERFACE FORMAT (DIF)

Default = 000 - Format 0 (2S , up to 24-bit data, 64 x Fs Internal SLCK)

Function:

The required relationship between the Left/Right clock, serial clock and serial data is defined by the Digital Interface Format and the options are detailed in Figures 16-22. NOTE: Internal SCLK is not available when MCLKDIV is enabled.

DIF2	DIF1	DIF0	DESCRIPTION	Format	FIGURE
0	0	0	I ² S, up to 24-bit data, 64 x Fs Internal SLCK	0	16
0	0	1	I ² S, up to 24-bit data, 32 x Fs Internal SLCK	1	17
0	1	0	Left Justified, up to 24-bit data,	2	18
0	1	1	Right Justified, 24-bit data	3	19
1	0	0	Right Justified, 20-bit data	4	20
1	0	1	Right Justified, 16-bit data	5	21
1	1	0	Right Justified, 18-bit data	6	22
1	1	1	Identical to Format 1	1	17

Table 8. Digital Interface Format



5.0 PIN DESCRIPTION

		•	-	1		
Left/Right Clock	LRCK	\Box	l 16		RST	Reset
Serial Data	SDATA	\Box 2	2 15		DIF1/SDA	DIF1/SDA
Serial Clock/DEM	SCLK/DEM	\Box 3	3 14		HP_B	Headphone Output B
Interface Power	VL	\Box	1 13		VA_HP	Headphone Amp Power
Master Clock	MCLK		5 12		VA	Analog Power
DIF0/SCL	DIF0/SCL	\Box 6	3 11		GND	Ground
HP Quiescent Voltage	VQ_HP		7 10		HP_A	Headphone Output A
Reference Ground	REF_GND	□ 8	9		FILT+	Positive Voltage Reference

LRCK	1	Left/Right Clock (<i>Input</i>) - The Left/Right clock determines which channel is currently being input on the serial audio data input, SDATA. The frequency of the Left/Right clock must be at the input sample rate. Audio samples in Left/Right sample pairs will be simultaneously output from the digital-to-analog converter whereas Right/Left pairs will exhibit a one sample period difference. The required relationship between the Left/Right clock, serial clock and serial data is defined by the Mode Control 2 (0Bh) register when in Control Port Mode or by the DIF1-0 pins when in Stand-Alone mode. The options are detailed in Figures 16-26.
SDATA	2	Serial Audio Data (<i>Input</i>) - Two's complement MSB-first serial data is input on this pin. The data is clocked into SDATA via the serial clock and the channel is determined by the Left/Right clock. The required relationship between the Left/Right clock, serial clock and serial data is defined by the Mode Control 2 (0Bh) register when in Control Port Mode or by the DIF1-0 pins when in Stand-Alone mode. The options are detailed in Figures 16-26.
SCLK	3	Serial Clock (<i>Input</i>) - Clocks the individual bits of the serial data into the SDATA pin. The required relationship between the Left/Right clock, serial clock and serial data is defined by the Mode Control 2 (0Bh) register when in Control Port Mode or by the DIF1-0 pins when in Stand-Alone mode. The options are detailed in Figures 16-26. The CS3L43 supports both internal and external serial clock generation modes. The Internal Serial Clock Mode eliminates possible clock interference from an external SCLK. Use of the Internal Serial Clock Mode is always preferred. <i>Internal Serial Clock Mode</i> - In the Internal Serial Clock Mode, the serial clock is internally derived and synchronous with the master clock and left/right clock. The SCLK/LRCK frequency ratio is either 32, 48, or 64 depending upon the Mode Control 2 (0Bh) register when in Control Port Mode or the DIF1-0 pins when in Stand-Alone mode as shown in Figures 16-26. Operation in this mode is identical to operation with an external serial clock synchronized with LRCK. <i>External Serial Clock Mode</i> - The CS3L43 will enter the External Serial Clock Mode whenever 16 low to high transitions are detected on the SCLK pin during any phase of the LRCK period. The device will revert to Internal Serial Clock Mode if no low to high transitions are detected on the SCLK pin for 2 consecutive periods of LRCK.



DEM

3 **De-emphasis Control** (*Input*) - When using Internal Serial Clock Mode, Pin 3 is available for de-emphasis control and selects the 44.1 kHz de-emphasis filter, see Table 9 and Figure 30. When using External Serial Clock Mode, de-emphasis control is not available.NOTE: De-emphasis is not available in High-Rate Mode.

External		
DEMO	DESCRIPTION	
0	Disabled	
1	44.1 kHz	

Table 9. Stand Alone De-Emphasis Control

VL 4 Interface Power (Input) - Digital interface power supply. Typically 1.8 to 3.3 VDC.

MCLK

Master Clock (Input) - The master clock frequency must be either 256x, 384x, 512x, 768x or 1024x the input sample rate in Base Rate Mode (BRM) and 128x, 192x, 256x or 384x the input sample rate in High Rate Mode (HRM). Note that some multiplication factors require setting the MCLKDIV bit (see Section 4.11.1). Tables 10 and 11 illustrates several standard audio sample rates and the required master clock frequencies.

Samula Data		MCLK	(MHz)					
Sample Rate		HRM						
(kHz)	128x	192x	256x*	384x*				
32	4.0960	6.1440	8.1920	12.2880				
44.1	5.6448	8.4672	11.2896	16.9344				
48	6.1440	9.2160	12.2880	18.4320				
64	8.1920	12.2880	16.3840	24.5760				
88.2	11.2896	16.9344	22.5792	33.8688				
96	12.2880	18.4320	24.5760	36.8640				

^{*} Requires MCLKDIV bit = 1 in Mode Control 2 register (address 0Bh).

Table 10. HRM Common Clock Frequencies

Camarla Bata			MCLK (MHz)		
Sample Rate (kHz)			BRM		
(KHZ)	256x	384x	512x	768x*	1024x*
32	8.1920	12.2880	16.3840	24.5760	32.7680
44.1	11.2896	16.9344	22.5792	32.7680	45.1584
48	12.2880	18.4320	24.5760	36.8640	49.1520

^{*} Requires MCLKDIV bit = 1 in Mode Control 2 register (address 0Bh).

Table 11. BRM Common Clock Frequencies

DIF0 and DIF1 (Stand-Alone Mode)

6 & 15 Digital Interface Format (Input) - The required relationship between the Left/Right clock, serial clock and serial data is defined by the Digital Interface Format and the options are detailed in Figures 23-26

DIF1	DIF0	DESCRIPTION	FORMAT	FIGURE
0	0	I ² S, up to 24-bit data	0	23
0	1	Left Justified, up to 24-bit data	1	24
1	0	Right Justified, 24-bit Data	2	25
1	1	Right Justified, 16-bit Data	3	26

Table 12. Digital Interface Format - DIF1 and DIF0 (Stand-Alone Mode)

SCL (Control Port Mode) 6 Serial Control Interface Clock (Input) - Clocks the serial control data into or out of SDA/CDIN.



VQ_HP	7	Headphone Quiescent Voltage (<i>Output</i>) - Filter connection for internal headphone amp quiescent reference voltage. A capacitor must be connected from VQ_HP to analog ground, as shown in Figure 5. VQ_HP is not intended to supply external current. VQ_HP has a typical source impedance of 250 k Ω and any current drawn from this pin will alter device performance.
REF_GND	8	Reference Ground (<i>Input</i>) - Ground reference for the internal sampling circuits. Must be connected to analog ground.
FILT+	9	Positive Voltage Reference ($Output$) - Positive reference for internal sampling circuits. An external capacitor is required from FILT+ to analog ground, as shown in Figure 5. The recommended value will typically provide 60 dB of PSRR at 1 kHz and 40 dB of PSRR at 60 Hz. FILT+ is not intended to supply external current. FILT+ has a typical source impedance of 250 k Ω and any current drawn from this pin will alter device performance.
HP_A and HP_B	10 & 14	4 Headphone Outputs (<i>Output</i>) - The full scale analog headphone output level is specified in the Analog Characteristics specifications table.
GND	11	Ground (Input) - Ground Reference.
VA	12	Analog Power (Input) - Analog power supply. Typically 1.8 to 3.3 VDC.
VA_HP	13	Headphone Amp Power (Input) - Headphone amplifier power supply. Typically 0.9 to 3.3 VDC.
SDA (Control Port Mode)	15	Serial Control Data I/O (Input/Output) - In Two-Wire mode, SDA is a data I/O line.
RST	16	Reset (<i>Input</i>) - The device enters a low power mode and all internal registers are reset to their default settings, including the control port, when low. See "Recommended Power-up Sequence" on page 25.



6.0 APPLICATIONS

6.1 Grounding and Power Supply Decoupling

As with any high resolution converter, the CS43L43 requires careful attention to power supply and grounding arrangements to optimize performance. Figure 5 shows the recommended power arrangement with VA, VA_HP and VL connected to clean supplies. Decoupling capacitors should be located as close to the device package as possible. If desired, all supply pins may be connected to the same supply, but a decoupling capacitor should still be used on each supply pin.

6.2 Clock Modes

The CS43L43 operates in one of two clocking modes. Base Rate Mode supports input sample rates up to 50 kHz while High Rate Mode supports input sample rates up to 100 kHz, see Table 10 and 11. All clock modes use 64x oversampling.

6.3 De-Emphasis

The CS43L43 includes on-chip digital de-emphasis. Figure 27 shows the de-emphasis curve for Fs equal to 44.1 kHz. The frequency response of the de-emphasis curve will scale proportionally with changes in sample rate, Fs.

The de-emphasis feature is included to accommodate older audio recordings that utilize pre-emphasis equalization as a means of noise reduction.

6.4 Recommended Power-up Sequence

- 1. Hold \overline{RST} low until the power supply, master, and left/right clocks are stable. In this state, the control port is reset to its default settings and VQ_HP will remain low.
- 2. Bring RST high. The device will remain in a low power state with VQ_HP low and will initiate the Stand-Alone power-up sequence. The control port will be accessible at this time. If Control Port operation is desired, write the CP_EN bit prior to the

completion of the Stand-Alone power-up sequence, approximately 1024 LRCK cycles. Writing this bit will halt the Stand-Alone power-up sequence and initialize the control port to its default settings. The desired register settings can be loaded while keeping the PDN bit set to 1.

3. If Control Port Mode is selected via the CP_EN bit, set the PDN bit to 0 which will initiate the power-up sequence, which requires approximately $50 \,\mu\text{S}$ when the POR bit is set to 0. If the POR bit is set to 1, see Section 6.5 for total power-up timing.

6.5 Popguard® Transient Control

The CS43L43 uses Popguard® technology to minimize the effects of output transients during power-up and power-down. This technique minimizes the audio transients commonly produced by single-ended, single-supply converters when it is implemented with external DC-blocking capacitors connected in series with the audio outputs.

When the device is initially powered-up, the audio outputs, HP_A and HP_B, are clamped to GND. Following a delay of approximately 1000 sample periods, each output begins to ramp toward the quiescent voltage. Approximately 10,000 left/right clock cycles later, the outputs reach V_{Q_HP} and audio output begins. This gradual voltage ramping allows time for the external DC-blocking capacitor to charge to the quiescent voltage, minimizing the power-up transient.

To prevent transients at power-down, the device must first enter its power-down state. When this occurs, audio output ceases and the internal output buffers are disconnected from HP_A and HP_B. In their place, a soft-start current sink is substituted which allows the DC-blocking capacitors to slowly discharge. Once this charge is dissipated, the power to the device may be turned off and the system is ready for the next power-on.



To prevent an audio transient at the next power-on, the DC-blocking capacitors must fully discharge before turning off the power or exiting the power-down state. If full discharge does not occur, a transient will occur when the audio outputs are initially clamped to GND. The time that the device must remain in the power-down state is related to the value of the DC-blocking capacitance and the output load. For example, with a 220 μ F capacitor and a 16 ohm load on the headphone outputs, the minimum power-down time will be approximately 0.4 seconds.

Use of the Mute Control function on the line outputs is recommended for designs requiring the absolute minimum in extraneous clicks and pops. Also, use of the Mute Control function can enable the system designer to achieve idle channel noise/signal-to-noise ratios. which are only limited by the external mute circuit. See the CDB43L43 datasheet for a suggested mute circuit

7.0 CONTROL PORT INTERFACE

The control port is used to load all the internal settings. The operation of the control port may be completely asynchronous with the audio sample rate. However, to avoid potential interference problems, the control port pins should remain static if no operation is required.

In Control Port Mode, SDA is a bi-directional data line. Data is clocked into and out of the part by the clock, SCL, with the clock to data relationship as shown in Figure 4. The 7-bit address field must be 0010000. The eighth bit of the address byte is the R/W bit (high for a read, low for a write). If the operation is a write, the next byte is the Memory Address Pointer, MAP, which selects the register to be read or written. The MAP is then followed by the data to be written. If the operation is a read, then the contents of the register pointed to by the MAP will be output after the chip address.

The CS43L43 has MAP auto increment capability, enabled by the INCR bit in the MAP register. If INCR is 0, then the MAP will stay constant for successive writes. If INCR is set to 1, then MAP will auto increment after each byte is written, allowing block reads or writes of successive registers.

The Two-Wire control port mode is compatible with the I^2C protocol.





7.1 MEMORY ADDRESS POINTER (MAP)

7	6	5	4	3	2	1	0
INCR	Reserved	Reserved	Reserved	MAP3	MAP2	MAP1	MAP0
0	0	0	0	0	0	0	0

7.1.1 INCR (AUTO MAP INCREMENT ENABLE)

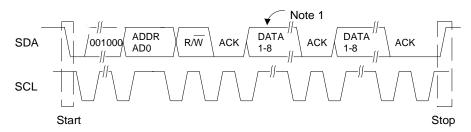
Default = '0'

0 - Disabled

1 - Enabled

7.1.2 MAPO-3 (MEMORY ADDRESS POINTER)

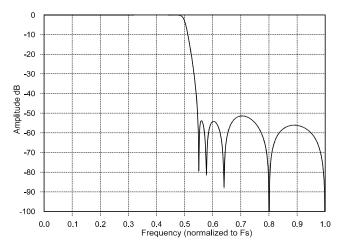
Default = '0000'



Note: If operation is a write, this byte contains the Memory Address Pointer, MAP.

Figure 6. Control Port Timing, Two-Wire Mode





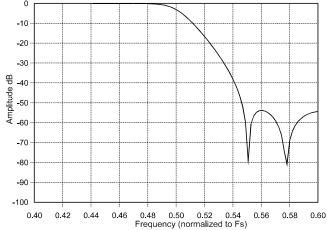


Figure 7. Base-Rate Stopband Rejection

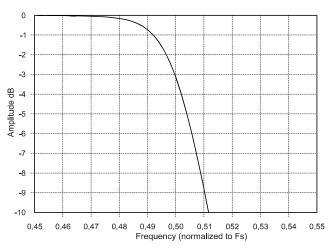


Figure 8. Base-Rate Transition Band

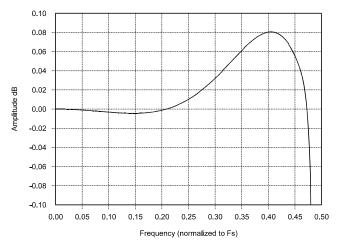


Figure 9. Base-Rate Transition Band (Detail)

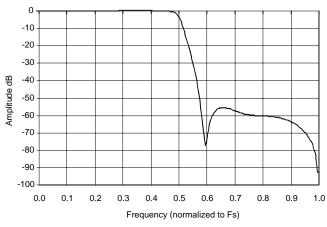


Figure 10. Base-Rate Passband Ripple

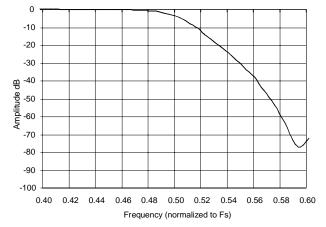


Figure 11. High-Rate Stopband Rejection

Figure 12. High-Rate Transition Band



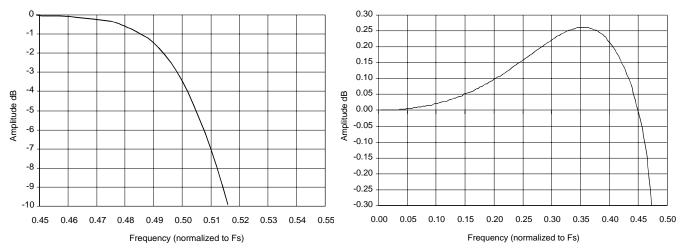


Figure 13. High-Rate Transition Band (Detail)

Figure 14. High-Rate Passband Ripple

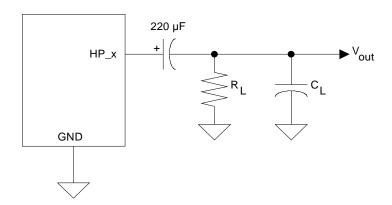
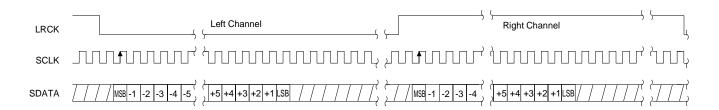


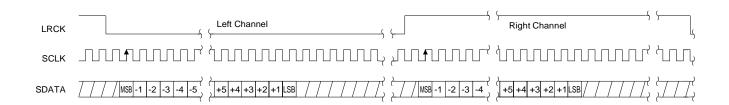
Figure 15. Output Test Load





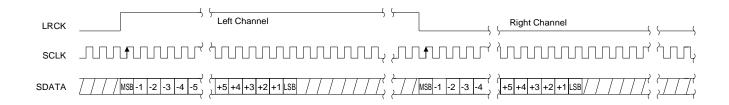
Internal SCLK Mode	External SCLK Mode
I^2 S, Up to 24-Bit data and INT SCLK = 64 Fs if	I ² S, up to 24-Bit Data
MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
I^2 S, Up to 24-Bit data and INT SCLK = 48 Fs if	
MCLK/LRCK = 384 or 192	

Figure 16. CS43L43 Control Port Mode - Serial Audio Format 0 (I²S)



Internal SCLK Mode	External SCLK Mode
I ² S, 16-Bit data and INT SCLK = 32 Fs if	I ² S, up to 24-Bit Data
MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
I^2 S, Up to 24-Bit data and INT SCLK = 48 Fs if	
MCLK/LRCK = 384 or 192	

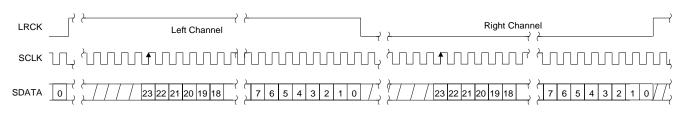
Figure 17. CS43L43 Control Port Mode - Serial Audio Format 1 (I²S)



Internal SCLK Mode	External SCLK Mode
• •	Left Justified, up to 24-Bit Data Data Valid on Rising Edge of SCLK

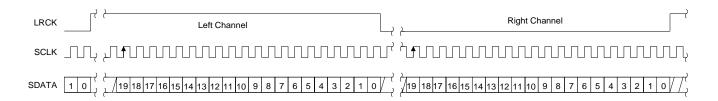
Figure 18. CS43L43 Control Port Mode - Serial Audio Format 2





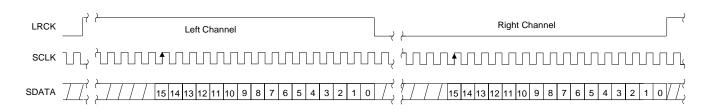
Internal SCLK Mode	External SCLK Mode
Right Justified, 24-Bit Data	Right Justified, 24-Bit Data
INT SCLK = 64 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 48 Cycles per LRCK Period

Figure 19. CS43L43 Control Port Mode - Serial Audio Format 3



Internal SCLK Mode	External SCLK Mode
Right Justified, 20-Bit Data	Right Justified, 20-Bit Data
INT SCLK = 64 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 40 Cycles per LRCK Period

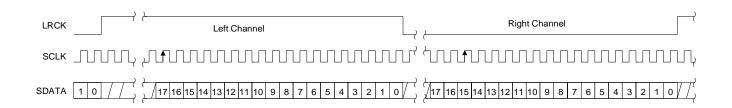
Figure 20. CS43L43 Control Port Mode - Serial Audio Format 4



Internal SCLK Mode	External SCLK Mode
Right Justified, 16-Bit Data	Right Justified, 16-Bit Data
INT SCLK = 32 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 32 Cycles per LRCK Period

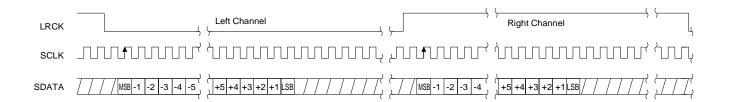
Figure 21. CS43L43 Control Port Mode - Serial Audio Format 5





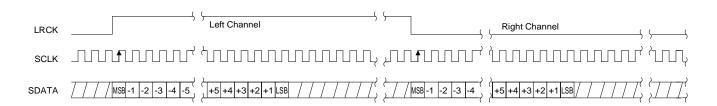
Internal SCLK Mode	External SCLK Mode
Right Justified, 18-Bit Data	Right Justified, 18-Bit Data
INT SCLK = 64 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 36 Cycles per LRCK Period

Figure 22. CS43L43 Control Port Mode - Serial Audio Format 6



Internal SCLK Mode	External SCLK Mode
I ² S, 16-Bit data and INT SCLK = 32 Fs if	I ² S, up to 24-Bit Data
	Data Valid on Rising Edge of SCLK
I ² S, Up to 24-Bit data and INT SCLK = 48 Fs if	
MCLK/LRCK = 384 or 192	

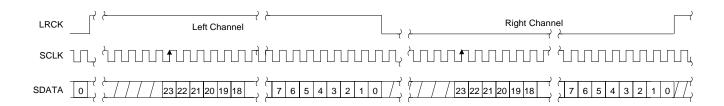
Figure 23. CS43L43 Stand-Alone Mode - Serial Audio Format 0 (I²S)



Internal SCLK Mode	External SCLK Mode
· •	Left Justified, up to 24-Bit Data Data Valid on Rising Edge of SCLK

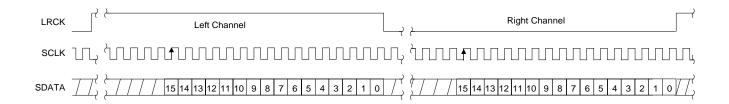
Figure 24. CS43L43 Stand-Alone Mode - Serial Audio Format 1





Internal SCLK Mode	External SCLK Mode
Right Justified, 24-Bit Data	Right Justified, 24-Bit Data
INT SCLK = 64 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 48 Cycles per LRCK Period

Figure 25. CS43L43 Stand-Alone Mode - Serial Audio Format 2



Internal SCLK Mode	External SCLK Mode
Right Justified, 16-Bit Data	Right Justified, 16-Bit Data
INT SCLK = 32 Fs if MCLK/LRCK = 512, 256 or 128	Data Valid on Rising Edge of SCLK
INT SCLK = 48 Fs if MCLK/LRCK = 384 or 192	SCLK Must Have at Least 32 Cycles per LRCK Period

Figure 26. CS43L43 Stand-Alone Mode - Serial Audio Format 3



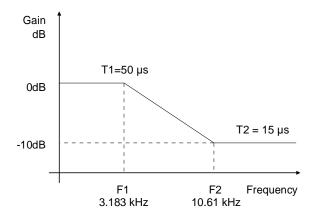


Figure 27. De-Emphasis Curve

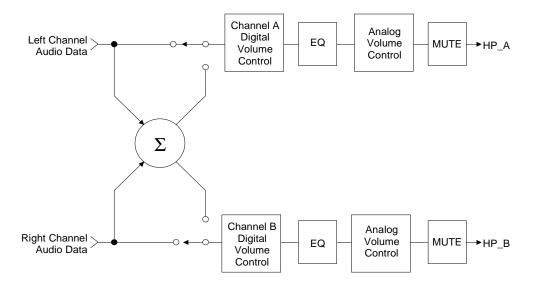


Figure 28. ATAPI Block Diagram



8.0 PARAMETER DEFINITIONS

Total Harmonic Distortion + Noise (THD+N)

The ratio of the rms value of the signal to the rms sum of all other spectral components over the specified bandwidth (typically 10Hz to 20kHz), including distortion components. Expressed in decibels.

Dynamic Range

The ratio of the full scale rms value of the signal to the rms sum of all other spectral components over the specified bandwidth. Dynamic range is a signal-to-noise measurement over the specified bandwidth made with a -60 dBFS signal. 60 dB is then added to the resulting measurement to refer the measurement to full scale. This technique ensures that the distortion components are below the noise level and do not effect the measurement. This measurement technique has been accepted by the Audio Engineering Society, AES17-1991, and the Electronic Industries Association of Japan, EIAJ CP-307.

Interchannel Isolation

A measure of crosstalk between the left and right channels. Measured for each channel at the converter's output with all zeros to the input under test and a full-scale signal applied to the other channel. Units in decibels.

Interchannel Gain Mismatch

The gain difference between left and right channels. Units in decibels.

Gain Error

The deviation from the nominal full scale analog output for a full scale digital input.

Gain Drift

The change in gain value with temperature. Units in ppm/°C.

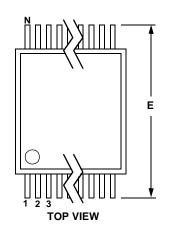
9.0 REFERENCES

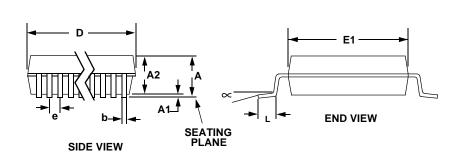
- 1) "How to Achieve Optimum Performance from Delta-Sigma A/D & D/A Converters" by Steven Harris. Paper presented at the 93rd Convention of the Audio Engineering Society, October 1992.
- 2) CDB4343 Evaluation Board Datasheet
- 3) "The I²C-Bus Specification: Version 2.0" Philips Semiconductors, December 1998. http://www.semiconductors.philips.com



10.0 PACKAGE DIMENSIONS

16L SSOP PACKAGE DRAWING





	INCHES			MILLIMETERS			NOTE
DIM	MIN	NOM	MAX	MIN	NOM	MAX	
Α			0.043			1.10	
A1	0.002		0.006	0.05		0.15	
A2	0.033	0.035	0.037	0.85	0.90	0.95	
b	0.008		0.012	0.19		0.30	2,3
D		0.197			5.00		1
Е		0.252			6.40		
E1	0.169	0.173	0.177	4.30	4.40	4.50	1
е		0.026			0.65		
Ĺ	0.020	0.024	0.028	0.50	0.60	0.70	
~	0°		8°	0°		8°	

JEDEC #: MO-150

- Notes: 1. "D" and "E1" are reference datums and do not included mold flash or protrusions, but do include mold mismatch and are measured at the parting line, mold flash or protrusions shall not exceed 0.20 mm per side.
 - 2. Dimension "b" does not include dambar protrusion/intrusion. Allowable dambar protrusion shall be 0.13 mm total in excess of "b" dimension at maximum material condition. Dambar intrusion shall not reduce dimension "b" by more than 0.07 mm at least material condition.
 - 3. These dimensions apply to the flat section of the lead between 0.10 and 0.25 mm from lead tips.



• Notes •

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