

CMOS SINGLE CHIP 8-BIT MICROCONTROLLER with 4-Kbytes of FLASH

PRELIMINARY
MAY 1998

FEATURES

- 80C51 based architecture
- 4-Kbytes of on-chip Reprogrammable Flash Memory
- 128 x 8 RAM
- Two 16-bit Timer/Counters
- Full duplex serial channel
- Boolean processor
- Four 8-bit I/O ports, 32 I/O lines
- Memory addressing capability
 - 64K ROM and 64K RAM
- Program memory lock
 - Encrypted verify (32 bytes)
 - Lock bits (3)
- Power save modes:
 - Idle and power-down
- Six interrupt sources
- Most instructions execute in 0.3 μ s
- CMOS and TTL compatible
- Maximum speed: 40 MHz @ $V_{cc} = 5V$
- Packages available:
 - 40-pin DIP
 - 44-pin PLCC

GENERAL DESCRIPTION

The *ISSI* IS89C51 is a high-performance microcontroller fabricated using high-density CMOS technology. The CMOS IS89C51 is functionally compatible with the industry standard 80C52 microcontrollers.

The IS89C51 is designed with 4-Kbytes of Flash memory, 128x8 RAM; 32 programmable I/O lines; a serial I/O port for either multiprocessor communications, I/O expansion or full duplex UART; two 16-bit timer/counters; an six-source, two-priority-level, nested interrupt structure; and an on-chip oscillator and clock circuit. The IS89C51 can be expanded using standard TTL compatible memory.

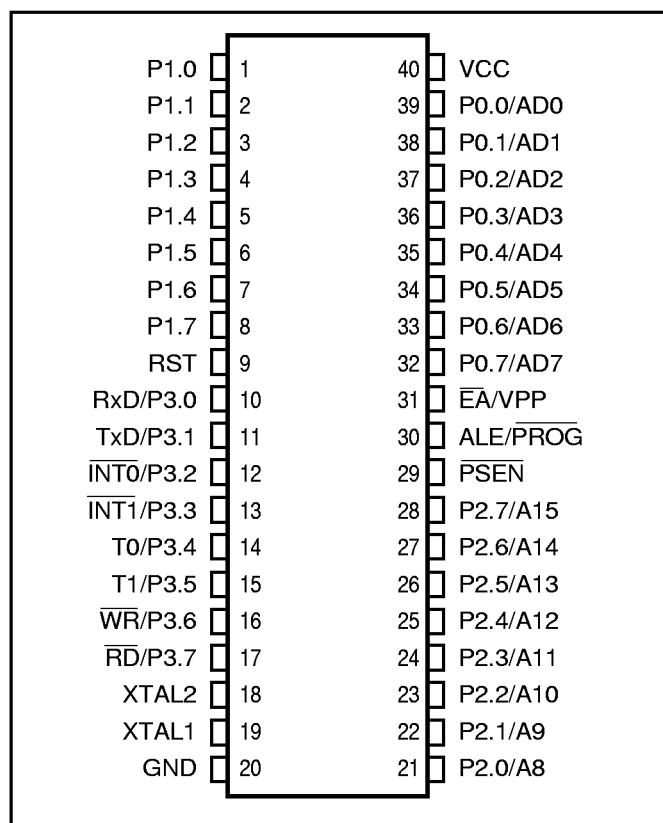


Figure 1. IS89C51 Pin Configuration: 40-pin PDIP

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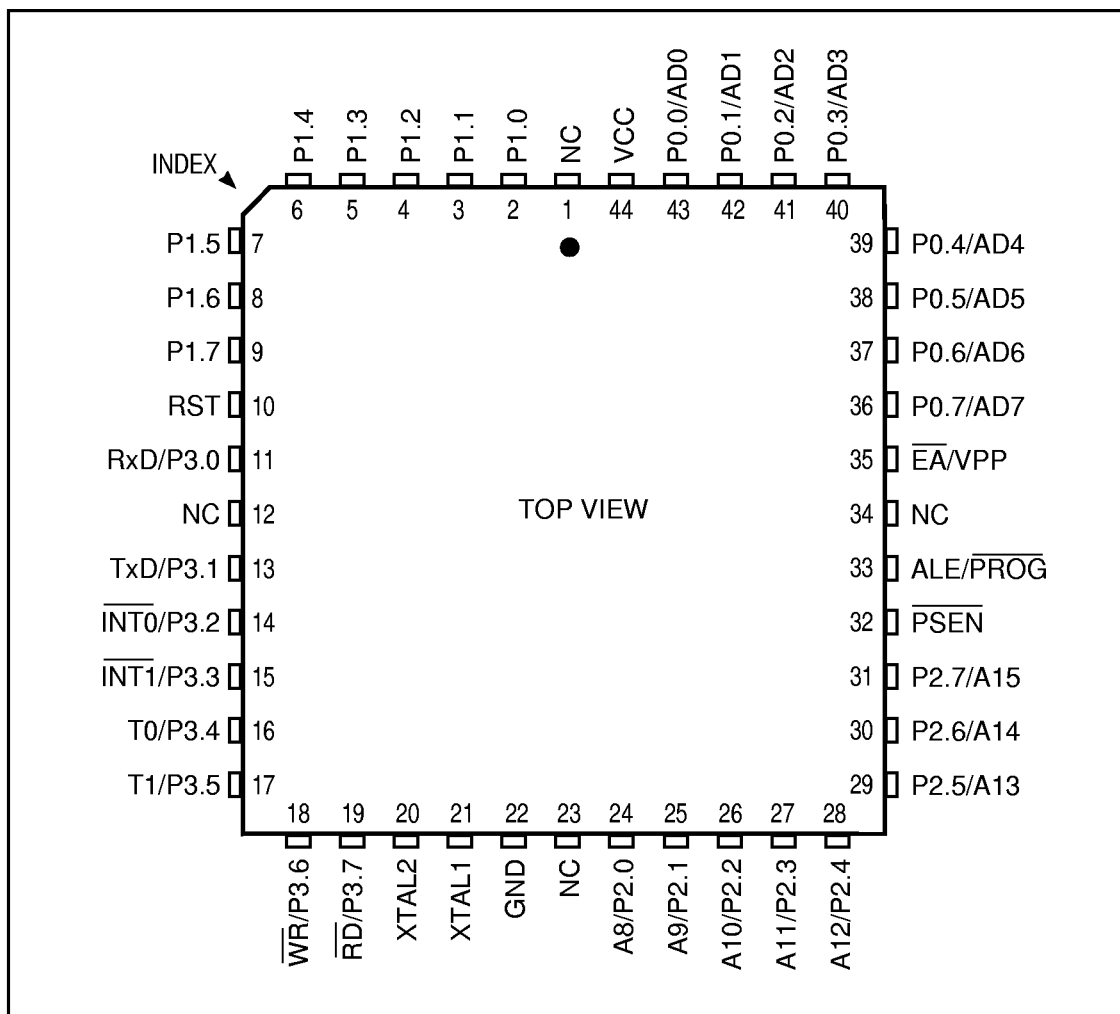


Figure 2. IS89C51 Pin Configuration: 44-pin PLCC

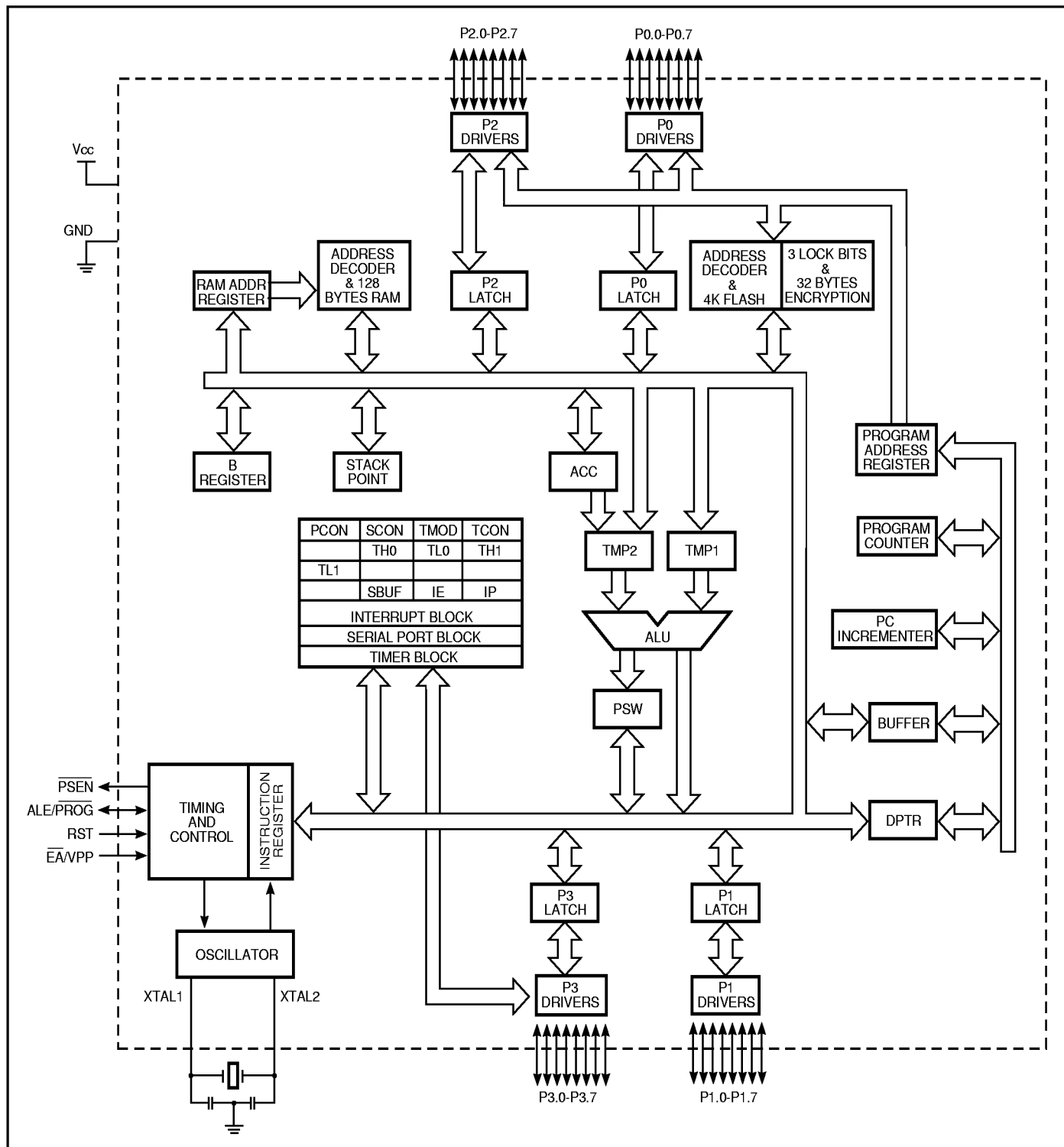


Figure 3. IS89C51 Block Diagram

Table 1. Detailed Pin Description

Symbol	PDIP	PLCC	I/O	Name and Function
$\overline{\text{ALE}}/\overline{\text{PROG}}$	30	33	I/O	Address Latch Enable: Output pulse for latching the low byte of the address during an address to the external memory. In normal operation, ALE is emitted at a constant rate of 1/6 the oscillator frequency, and can be used for external timing or clocking. Note that one ALE pulse is skipped during each access to external data memory. This pin is also the Program Pulse input ($\overline{\text{PROG}}$) during Flash programming.
$\overline{\text{EA}}/\text{V}_{\text{PP}}$	31	35	I	External Access enable: $\overline{\text{EA}}$ must be externally held low to enable the device to fetch code from external program memory locations 0000H to FFFFH. If $\overline{\text{EA}}$ is held high, the device executes from internal program memory unless the program counter contains an address greater than 0FFFFH. This also receives the 12V programming enable voltage (V_{PP}) during Flash programming.
P0.0-P0.7	39-32	43-36	I/O	Port 0: Port 0 is an 8-bit open-drain, bidirectional I/O port. Port 0 pins that have 1s written to them float and can be used as high-impedance inputs. Port 0 is also the multiplexed low-order address and data bus during accesses to external program and data memory. In this application, it uses strong internal pullups when emitting 1s. Port 0 also receives the code bytes during programmable memory programming and outputs the code bytes during program verification. External pullups are required during program verification.
P1.0-P1.7	1-8	2-9	I/O	Port 1: Port 1 is an 8-bit bidirectional I/O port with internal pullups. Port 1 pins that have 1s written to them are pulled high by the internal pullups and can be used as inputs. As inputs, Port 1 pins that are externally pulled low will source current because of the internal pullups. (See DC Characteristics: I_{IL}). The Port 1 output buffers can sink/source four TTL inputs. Port 1 also receives the low-order address byte during Flash programming and verification.
P2.0-P2.7	21-28	24-31	I/O	Port 2: Port 2 is an 8-bit bidirectional I/O port with internal pullups. Port 2 pins that have 1s written to them are pulled high by the internal pullups and can be used as inputs. As inputs, Port 2 pins that are externally pulled low will source current because of the internal pullups. (See DC Characteristics: I_{IL}). Port 2 emits the high order address byte during fetches from external program memory and during accesses to external data memory that used 16-bit addresses (MOVX @ DPTR). In this application, Port 2 uses strong internal pullups when emitting 1s. During accesses to external data memory that use 8-bit addresses ($\text{MOVX @ Ri [i = 0, 1]}$), Port 2 emits the contents of the P2 Special Function Register. Port 2 also receives the high-order bits and some control signals during Flash programming and verification. P2.6 and P2.7 are the control signals while the chip programs and erases.

Table 1. Detailed Pin Description (*continued*)

Symbol	PDIP	PLCC	I/O	Name and Function
P3.0-P3.7	10-17	11, 13-19	I/O	<p>Port 3: Port 3 is an 8-bit bidirectional I/O port with internal pullups. Port 3 pins that have 1s written to them are pulled high by the internal pullups and can be used as inputs. As inputs, Port 3 pins that are externally pulled low will source current because of the internal pullups. (See DC Characteristics: I_{IL}).</p> <p>Port 3 also serves the special features of the IS89C51, as listed below:</p> <p>RxD (P3.0): Serial input port.</p> <p>TxD (P3.1): Serial output port.</p> <p>INT0 (P3.2): External interrupt 0.</p> <p>INT1 (P3.3): External interrupt 1.</p> <p>T0 (P3.4): Timer 0 external input.</p> <p>T1 (P3.5): Timer 1 external input.</p> <p>WR (P3.6): External data memory write strobe.</p> <p>RD (P3.7): External data memory read strobe.</p>
$\overline{\text{PSEN}}$	29	32	O	<p>Program Store Enable: The read strobe to external program memory. When the device is executing code from the external program memory, $\overline{\text{PSEN}}$ is activated twice each machine cycle except that two $\overline{\text{PSEN}}$ activations are skipped during each access to external data memory. $\overline{\text{PSEN}}$ is not activated during fetches from internal program memory.</p>
RST	9	10	I	<p>Reset: A high on this pin for two machine cycles while the oscillator is running, resets the device. An internal MOS resistor to GND permits a power-on reset using only an external capacitor connected to VCC.</p>
XTAL 1	19	21	I	<p>Crystal 1: Input to the inverting oscillator amplifier and input to the internal clock generator circuits.</p>
XTAL 2	18	20	O	<p>Crystal 2: Output from the inverting oscillator amplifier.</p>
GND	20	22	I	<p>Ground: 0V reference.</p>
Vcc	40	44	I	<p>Power Supply: This is the power supply voltage for operation.</p>

OPERATING DESCRIPTION

The detail description of the IS89C51 included in this description are:

- Memory Map and Registers
- Timer/Counters
- Serial Interface
- Interrupt System
- Other Information
- Flash Memory

MEMORY MAP AND REGISTERS

Memory

The IS89C51 has separate address spaces for program and data memory. The program and data memory can be up to 64K bytes long. The lower 4K program memory can reside on-chip. Figure 4 shows a map of the IS89C51 program and data memory.

The IS89C51 has 128 bytes of on-chip RAM, plus numbers of special function registers. The lower 128 bytes can be accessed either by direct addressing or by indirect

addressing. Figure 5 shows internal data memory organization and SFR Memory Map.

The lower 128 bytes of RAM can be divided into three segments as listed below and shown in Figure 6.

1. **Register Banks 0-3:** locations 00H through 1FH (32 bytes). The device after reset defaults to register bank 0. To use the other register banks, the user must select them in software. Each register bank contains eight 1-byte registers R0-R7. Reset initializes the stack point to location 07H, and is incremented once to start from 08H, which is the first register of the second register bank.
2. **Bit Addressable Area:** 16 bytes have been assigned for this segment 20H-2FH. Each one of the 128 bits of this segment can be directly addressed (0-7FH). Each of the 16 bytes in this segment can also be addressed as a byte.
3. **Scratch Pad Area:** 30H-7FH are available to the user as data RAM. However, if the data pointer has been initialized to this area, enough bytes should be left aside to prevent SP data destruction.

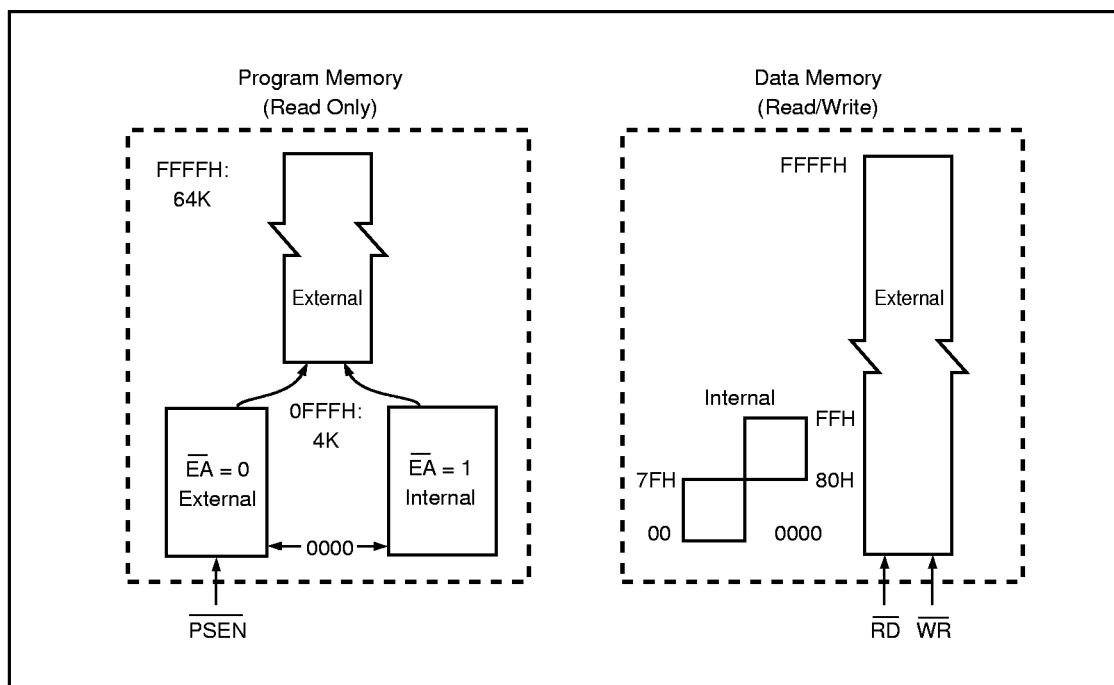


Figure 4. IS89C51 Program and Data Memory Structure

SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFR's) are located in upper 128 Bytes direct addressing area. The SFR Memory Map in Figure 5 shows that.

Not all of the addresses are occupied. Unoccupied addresses are not implemented on the chip. Read accesses to these addresses in general return random data, and write accesses have no effect.

User software should not write 1s to these unimplemented locations, since they may be used in future microcontrollers to invoke new features. In that case, the reset or inactive values of the new bits will always be 0, and their active values will be 1.

The functions of the SFRs are outlined in the following sections, and detailed in Table 2.

Accumulator (ACC)

ACC is the Accumulator register. The mnemonics for Accumulator-specific instructions, however, refer to the Accumulator simply as A.

B Register (B)

The B register is used during multiply and divide operations. For other instructions it can be treated as another scratch pad register.

Program Status Word (PSW). The PSW register contains program status information.

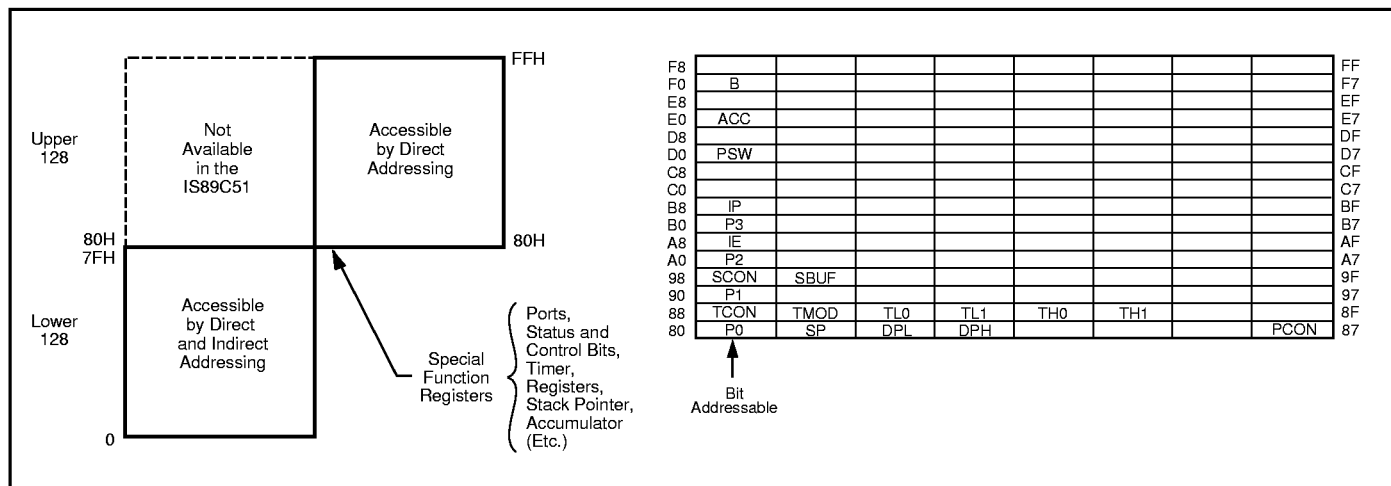


Figure 5. Internal Data Memory and SFR Memory Map

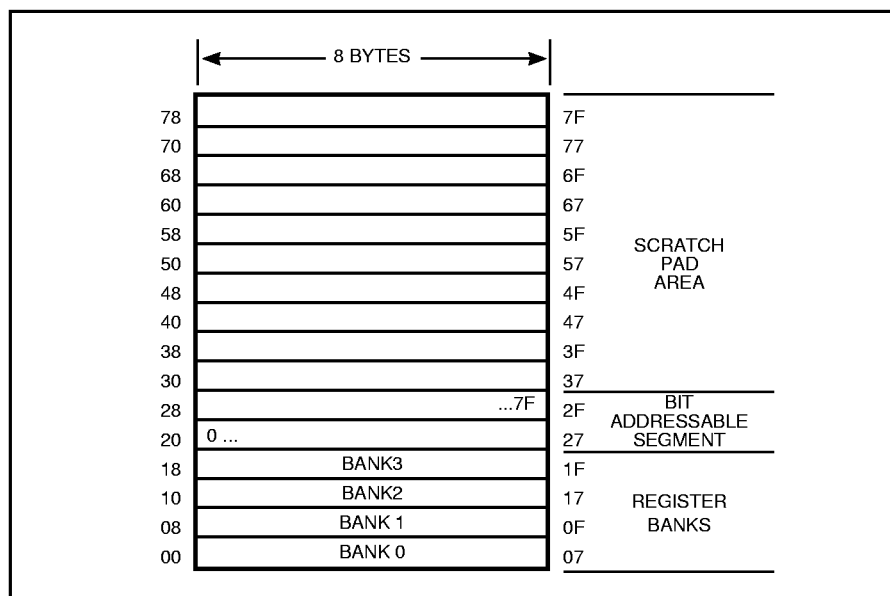


Figure 6. Lower 128 Bytes of Internal RAM

SPECIAL FUNCTION REGISTERS

(Continued)

Stack Pointer (SP)

The Stack Pointer Register is eight bits wide. It is incremented before data is stored during PUSH and CALL executions. While the stack may reside anywhere in on-chip RAM, the Stack Pointer is initialized to 07H after a reset. This causes the stack to begin at location 08H.

Data Pointer (DPTR)

The Data Pointer consists of a high byte (DPH) and a low byte (DPL). Its function is to hold a 16-bit address. It may be manipulated as a 16-bit register or as two independent 8-bit registers.

Ports 0 To 3

P0, P1, P2, and P3 are the SFR latches of Ports 0, 1, 2, and 3, respectively.

Serial Data Buffer (SBUF)

The Serial Data Buffer is actually two separate registers, a transmit buffer and a receive buffer register. When data is moved to SBUF, it goes to the transmit buffer, where it is held for serial transmission. (Moving a byte to SBUF initiates the transmission.) When data is moved from SBUF, it comes from the receive buffer.

Timer Registers

Register pairs (TH0, TL0) and (TH1, TL1) are the 16-bit Counter registers for Timer/Counters 0 and 1, respectively.

Control Registers

Special Function Registers IP, IE, TMOD, TCON, SCON, and PCON contain control and status bits for the interrupt system, the Timer/Counters, and the serial port. They are described in later sections of this chapter.

Table 2: Special Function Registers

Symbol	Description	Direct Address	Bit Address, Symbol, or Alternative Port Function								Reset Value
ACC ⁽¹⁾	Accumulator	E0H	E7	E6	E5	E4	E3	E2	E1	E0	00H
B ⁽¹⁾	B register	F0H	F7	F6	F5	F4	F3	F2	F1	F0	00H
DPH	Data pointer (DPTR) high	83H									00H
DPL	Data pointer (DPTR) low	82H									00H
IE ⁽¹⁾	Interrupt enable	A8H	AF EA	AE —	AD —	AC ES	AB ET1	AA EX1	A9 ET0	A8 EX0	0XX00000B
IP ⁽¹⁾	Interrupt priority	B8H	BF —	BE —	BD —	BC PS	BB PT1	BA PX1	B9 PT0	B8 PX0	XXX00000B
P0 ⁽¹⁾	Port 0	80H	87 P0.7 AD7	86 P0.6 AD6	85 P0.5 AD5	84 P0.4 AD4	83 P0.3 AD3	82 P0.2 AD2	81 P0.1 AD1	80 P0.0 AD0	FFH
P1 ⁽¹⁾	Port 1	90H	97 P1.7	96 P1.6	95 P1.5	94 P1.4	93 P1.3	92 P1.2	91 P1.1	90 P1.0	FFH
P2 ⁽¹⁾	Port 2	A0H	A7 P2.7 AD15	A6 P2.6 AD14	A5 P2.5 AD13	A4 P2.4 AD12	A3 P2.3 AD11	A2 P2.2 AD10	A1 P2.1 AD9	A0 P2.0 AD8	FFH
P3 ⁽¹⁾	Port 3	B0H	B7 P3.7 <u>RD</u>	B6 P3.6 <u>WR</u>	B5 P3.5 T1	B4 P3.4 T0	B3 P3.3 <u>INT1</u>	B2 P3.2 <u>INT0</u>	B1 P3.1 TXD	B0 P3.0 RXD	FFH
PCON	Power control	87H	SMOD	—	—	—	GF1	GF0	PD	IDL	0XXX0000B
PSW ⁽¹⁾	Program status word	D0H	D7 CY	D6 AC	D5 F0	D4 RS1	D3 RS0	D2 OV	D1 —	D0 P	00H
SBUF	Serial data buffer	99H									XXXXXXXXB
SCON ⁽¹⁾	Serial controller	98H	9F SM0	9E SM1	9D SM2	9C REN	9B TB8	9A RB8	99 TI	98 RI	00H
SP	Stack pointer	81H									07H
TCON ⁽¹⁾	Timer control	88H	8F TF1	8E TR1	8D TF0	8C TR0	8B IE1	8A IT1	89 IE0	88 IT0	00H
TMOD	Timer mode	89H	GATE	C/ <u>T</u>	M1	M0	GATE	C/ <u>T</u>	M1	M0	00H
TH0	Timer high 0	8CH									00H
TH1	Timer high 1	8DH									00H
TL0	Timer low 0	8AH									00H
TL1	Timer low 1	8BH									00H

Notes:

1. Denotes bit addressable.

The detail description of each bit is as follows:

PSW:

Program Status Word. Bit Addressable.

7	6	5	4	3	2	1	0
CY	AC	F0	RS1	RS0	OV	—	P

Register Description:

CY	PSW.7	Carry flag.
AC	PSW.6	Auxiliary carry flag.
F0	PSW.5	Flag 0 available to the user for general purpose.
RS1	PSW.4	Register bank selector bit 1. ⁽¹⁾
RS0	PSW.3	Register bank selector bit 0. ⁽¹⁾
OV	PSW.2	Overflow flag.
—	PSW.1	Usable as a general purpose flag
P	PSW.0	Parity flag. Set/Clear by hardware each instruction cycle to indicate an odd/even number of "1" bits in the accumulator.

Note:

1. The value presented by RS0 and RS1 selects the corresponding register bank.

RS1	RS0	Register Bank	Address
0	0	0	00H-07H
0	1	1	08H-0FH
1	0	2	10H-17H
1	1	3	18H-1FH

PCON:

Power Control Register. Not Bit Addressable.

7	6	5	4	3	2	1	0
SMOD	—	—	—	GF1	GF0	PD	IDL

Register Description:

SMOD	Double baud rate bit. If Timer 1 is used to generate baud rate and SMOD=1, the baud rate is doubled when the serial port is used in modes 1, 2, or 3.
—	Not implemented, reserve for future use. ⁽¹⁾
—	Not implemented, reserve for future use. ⁽¹⁾
—	Not implemented, reserve for future use. ⁽¹⁾
GF1	General purpose flag bit.
GF0	General purpose flag bit.
PD	Power-down bit. Setting this bit activates power-down mode.
IDL	Idle mode bit. Setting this bit activates idle mode. If 1s are written to PD and IDL at the same time, PD takes precedence.

Note:

1. User software should not write 1s to reserved bits. These bits may be used in future products to invoke new features.

IE:

Interrupt Enable Register. Bit Addressable.

7	6	5	4	3	2	1	0
EA	—	—	ES	ET1	EX1	ET0	EX0

Register Description:

EA	IE.7	Disable all interrupts. If EA=0, no interrupt will be acknowledged. If EA=1, each interrupt source is individually enabled or disabled by setting or clearing its enable bit.
—	IE.6	Not implemented, reserve for future use. ⁽⁵⁾
—	IE.5	Not implemented, reserve for future use. ⁽⁵⁾
ES	IE.4	Enable or disable the serial port interrupt.
ET1	IE.3	Enable or disable the Timer 1 overflow interrupt.
EX1	IE.2	Enable or disable External Interrupt 1.
ET0	IE.1	Enable or disable the Timer 0 overflow interrupt.
EX0	IE.0	Enable or disable External Interrupt 0.

Note: To use any of the interrupts in the 80C51 Family, the following three steps must be taken:

1. Set the EA (enable all) bit in the IE register to 1.
2. Set the corresponding individual interrupt enable bit in the IE register to 1.
3. Begin the interrupt service routine at the corresponding Vector Address of that interrupt (see below).

Interrupt Source	Vector Address
IE0	0003H
TF0	000BH
IE1	0013H
TF1	001BH
RI & TI	0023H

4. In addition, for external interrupts, pins $\overline{\text{INT0}}$ and $\overline{\text{INT1}}$ (P3.2 and P3.3) must be set to 1, and depending on whether the interrupt is to be level or transition activated, bits IT0 or IT1 in the TCON register may need to be set to 0 or 1.
ITX = 0 level activated (X = 0, 1)
ITX = 1 transition activated
5. User software should not write 1s to reserved bits. These bits may be used in future products to invoke new features.

IP:**Interrupt Priority Register. Bit Addressable.**

7	6	5	4	3	2	1	0
—	—	—	PS	PT1	PX1	PT0	PX0

Register Description:		
—	IP.7	Not implemented, reserve for future use ⁽³⁾
—	IP.6	Not implemented, reserve for future use ⁽³⁾
—	IP.5	Not implemented, reserve for future use ⁽³⁾
PS	IP.4	Defines Serial Port interrupt priority level
PT1	IP.3	Defines Timer 1 interrupt priority level
PX1	IP.2	Defines External Interrupt 1 priority level
PT0	IP.1	Defines Timer 0 interrupt priority level
PX0	IP.0	Defines External Interrupt 0 priority level

Notes:

1. In order to assign higher priority to an interrupt the corresponding bit in the IP register must be set to 1. While an interrupt service is in progress, it cannot be interrupted by a lower or same level interrupt.
2. Priority within level is only to resolve simultaneous requests of the same priority level. From high-to-low, interrupt sources are listed below:
IE0
TF0
IE1
TF1
RI or TI
3. User software should not write 1s to reserved bits. These bits may be used in future products to invoke new features.

TCON:**Timer/Counter Control Register. Bit Addressable**

7	6	5	4	3	2	1	0
TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0

Register Description:		
TF1	TCON.7	Timer 1 overflow flag. Set by hardware when the Timer/Counter 1 overflows. Cleared by hardware as processor vectors to the interrupt service routine.
TR1	TCON.6	Timer 1 run control bit. Set/Cleared by software to turn Timer/Counter 1 ON/OFF.
TF0	TCON.5	Timer 0 overflow flag. Set by hardware when the Timer/Counter 0 overflows. Cleared by hardware as processor vectors to the interrupt service routine.
TR0	TCON.4	Timer 0 run control bit. Set/Cleared by software to turn Timer/Counter 0 ON/OFF.
IE1	TCON.3	External Interrupt 1 edge flag. Set by hardware when the External Interrupt edge is detected. Cleared by hardware when interrupt is processed.
IT1	TCON.2	Interrupt 1 type control bit. Set/Cleared by software specify falling edge/low level triggered External Interrupt.
IE0	TCON.1	External Interrupt 0 edge flag. Set by hardware when the External Interrupt edge is detected. Cleared by hardware when interrupt is processed.
IT0	TCON.0	Interrupt 0 type control bit. Set/Cleared by software specify falling edge/low level triggered External Interrupt.

TMOD:

Timer/Counter Mode Control Register.
Not Bit Addressable.

Timer 1				Timer 0			
GATE	C/T	M1	M0	GATE	C/T	M1	M0
GATE	When TRx (in TCON) is set and GATE=1, TIMER/COUNTERx will run only while $\overline{\text{INTx}}$ pin is high (hardware control). When GATE=0, TIMER/COUNTERx will run only while TRx=1 (software control).						
C/T	Timer or Counter selector. Cleared for Timer operation (input from internal system clock). Set for Counter operation (input from Tx input pin).						
M1	Mode selector bit. ⁽¹⁾						
M0	Mode selector bit. ⁽¹⁾						

Note 1:

M1	M0	Operating Mode
0	0	Mode 0. (13-bit Timer)
0	1	Mode 1. (16-bit Timer/Counter)
1	0	Mode 2. (8-bit auto-load Timer/Counter)
1	1	Mode 3. (Splits Timer 0 into TL0 and TH0. TL0 is an 8-bit Timer/Counter controller by the standard Timer 0 control bits. TH0 is an 8-bit Timer and is controlled by Timer 1 control bits.)
1	1	Mode 3. (Timer/Counter 1 stopped).

SCON:

Serial Port Control Register. Bit Addressable.

7	6	5	4	3	2	1	0
SM0	SM1	SM2	REN	TB8	RB8	TI	RI

Register Description:

SM0	SCON.7	Serial port mode specifier. ⁽¹⁾
SM1	SCON.6	Serial port mode specifier. ⁽¹⁾
SM2	SCON.5	Enable the multiprocessor communication feature in mode 2 and 3. In mode 2 or 3, if SM2 is set to 1 then RI will not be activated if the received 9th data bit (RB8) is 0. In mode 1, if SM2=1 then RI will not be activated if valid stop bit was not received. In mode 0, SM2 should be 0.
REN	SCON.4	Set/Cleared by software to Enable/Disable reception.
TB8	SCON.3	The 9th bit that will be transmitted in mode 2 and 3. Set/Cleared by software.
RB8	SCON.2	In modes 2 and 3, RB8 is the 9th data bit that was received. In mode 1, if SM2=0, RB8 is the stop bit that was received. In mode 0, RB8 is not used.
TI	SCON.1	Transmit interrupt flag. Set by hardware at the end of the eighth bit time in mode 0, or at the beginning of the stop bit in the other modes. Must be cleared by software.
RI	SCON.0	Receive interrupt flag. Set by hardware at the end of the eighth bit time in mode 0, or halfway through the stop bit time in the other modes (except see SM2). Must be cleared by software.

Note 1:

SM0	SM1	MODE	Description	Baud Rate
0	0	0	Shift register	Fosc/12
0	1	1	8-bit UART	Variable
1	0	2	9-bit UART	Fosc/64 or Fosc/32
1	1	3	9-bit UART	Variable

TIMER/COUNTERS

The IS89C51 has two 16-bit Timer/Counter registers: Timer 0 and Timer 1. All two can be configured to operate either as Timers or event Counters.

As a Timer, the register is incremented every machine cycle. Thus, the register counts machine cycles. Since a machine cycle consists of 12 oscillator periods, the count rate is 1/12 of the oscillator frequency.

As a Counter, the register is incremented in response to a 1-to-0 transition at its corresponding external input pin, T0 and T1. The external input is sampled during S5P2 of every machine cycle. When the samples show a high in one cycle and a low in the next cycle, the count is incremented. The new count value appears in the register during S3P1 of the cycle following the one in which the transition was detected. Since two machine cycles (24 oscillator periods) are required to recognize a 1-to-0 transition, the maximum count rate is 1/24 of the oscillator frequency. There are no restrictions on the duty cycle of the external input signal, but it should be held for at least one full machine cycle to ensure that a given level is sampled at least once before it changes.

In addition to the Timer or Counter functions, Timer 0 and Timer 1 have four operating modes: 13-bit timer, 16-bit timer, 8-bit auto-reload, split timer.

Timer 0 and Timer 1

The Timer or Counter function is selected by control bits C/\bar{T} in the Special Function Register TMOD. These two Timer/Counters have four operating modes, which are selected by bit pairs (M1, M0) in TMOD. Modes 0, 1, and 2 are the same for both Timer/Counters, but Mode 3 is different. The four modes are described in the following sections.

Mode 0:

Both Timers in Mode 0 are 8-bit Counters with a divide-by-32 prescaler. Figure 7 shows the Mode 0 operation as it applies to Timer 1.

In this mode, the Timer register is configured as a 13-bit register. As the count rolls over from all 1s to all 0s, it sets the Timer interrupt flag TF1. The counted input is enabled to the Timer when $TR1 = 1$ and either $GATE = 0$ or $\overline{INT1} = 1$. Setting $GATE = 1$ allows the Timer to be controlled by external input $\overline{INT1}$, to facilitate pulse width measurements. $TR1$ is a control bit in the Special Function Register TCON. Gate is in TMOD.

The 13-bit register consists of all eight bits of TH1 and the lower five bits of TL1. The upper three bits of TL1 are indeterminate and should be ignored. Setting the run flag (TR1) does not clear the registers.

Mode 0 operation is the same for Timer 0 as for Timer 1, except that $TR0$, $TF0$ and $\overline{INT0}$ replace the corresponding Timer 1 signals in Figure 7. There are two different GATE bits, one for Timer 1 (TMOD.7) and one for Timer 0 (TMOD.3).

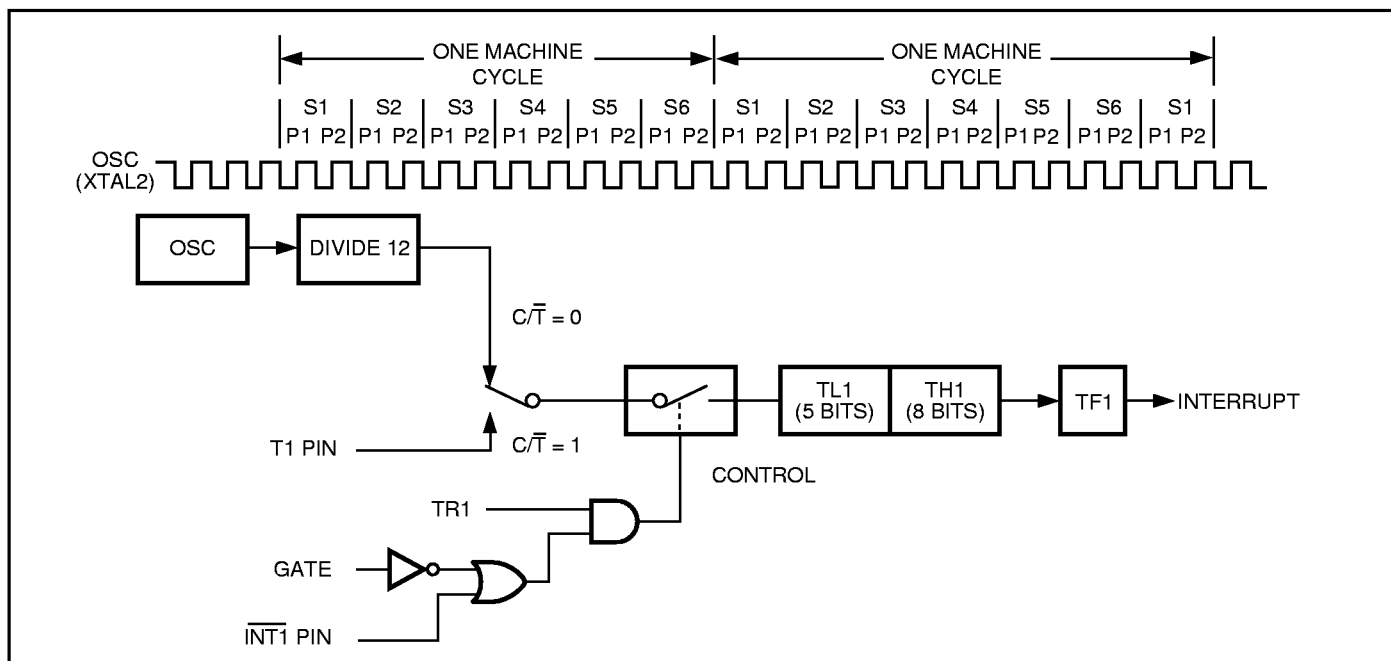


Figure 7. Timer/Counter 1 Mode 0: 13-Bit Counter

Mode 1:

Mode 1 is the same as Mode 0, except that the Timer register is run with all 16 bits. The clock is applied to the combined high and low timer registers (TL1/TH1). As clock pulses are received, the timer counts up: 0000H, 0001H, 0002H, etc. An overflow occurs on the FFFFH-to-0000H overflow flag. The timer continues to count. The overflow flag is the TF1 bit in TCON that is read or written by software (see Figure 8).

Mode 2:

Mode 2 configures the Timer register as an 8-bit Counter (TL1) with automatic reload, as shown in Figure 9. Overflow from TL1 not only sets TF1, but also reloads TL1 with the contents of TH1, which is preset by software. The reload leaves the TH1 unchanged. Mode 2 operation is the same for Timer/Counter 0.

Mode 3:

Timer 1 in Mode 3 simply holds its count. The effect is the same as setting TR1 = 0. Timer 0 in Mode 3 establishes TL0 and TH0 as two separate counters. The logic for Mode 3 on Timer 0 is shown in Figure 10. TL0 uses the Timer 0 control bits: C/\overline{T} , GATE, TR0, $\overline{INT0}$, and TF0. TH0 is locked into a timer function (counting machine cycles) and over the use of TR1 and TF1 from Timer 1. Thus, TH0 now controls the Timer 1 interrupt.

Mode 3 is for applications requiring an extra 8-bit timer or counter. With Timer 0 in Mode 3, the IS89C51 can appear to have three Timer/Counters. When Timer 0 is in Mode 3, Timer 1 can be turned on and off by switching it out of and into its own Mode 3. In this case, Timer 1 can still be used by the serial port as a baud rate generator or in any application not requiring an interrupt.

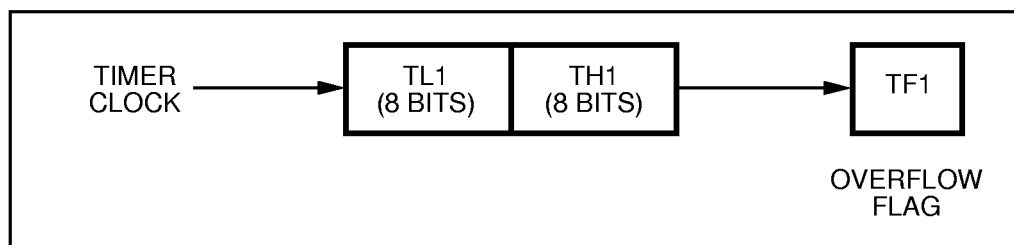


Figure 8. Timer/Counter 1 Mode 1: 16-Bit Counter

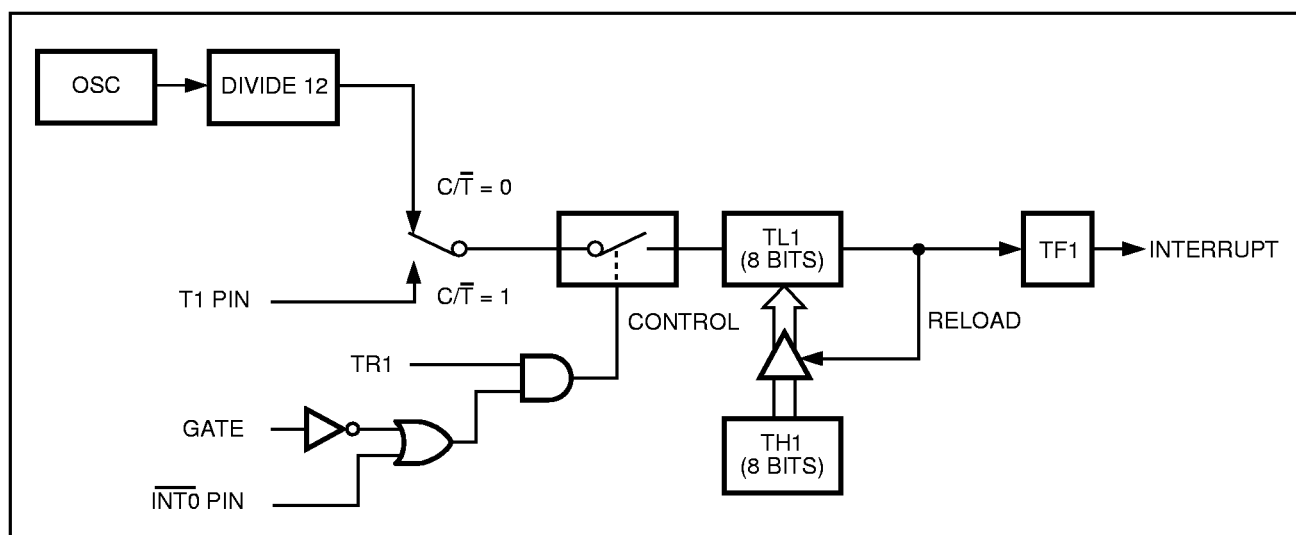


Figure 9. Timer/Counter 1 Mode 2: 8-Bit Auto-Reload

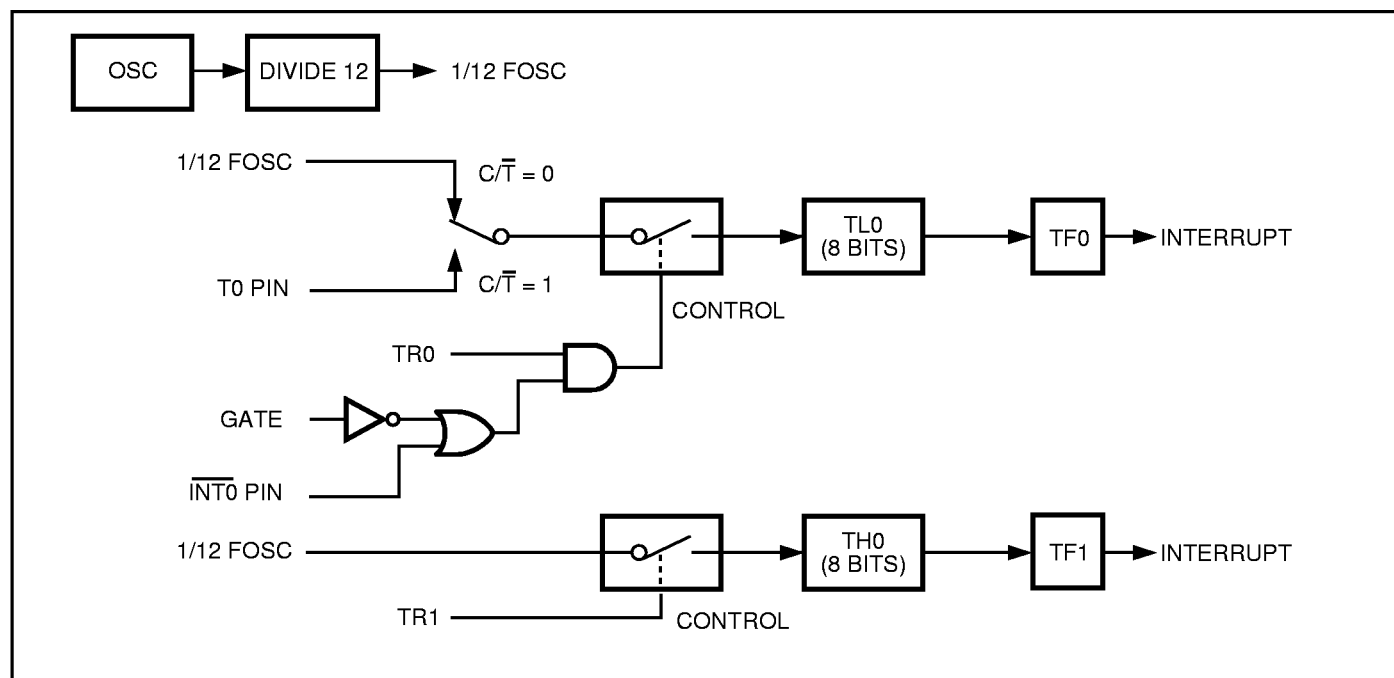


Figure 10. Timer/Counter 0 Mode 3: Two 8-Bit Counters

Timer Set-Up

Tables 3 through 6 give TMOD values that can be used to set up Timers in different modes.

It assumes that only one timer is used at a time. If Timers 0 and 1 must run simultaneously in any mode, the value in TMOD for Timer 0 must be ORed with the value shown for Timer 1 (Tables 5 and 6).

For example, if Timer 0 must run in Mode 1 GATE (external control), and Timer 1 must run in Mode 2 COUNTER, then the value that must be loaded into TMOD is 69H (09H from Table 3 ORed with 60H from Table 6).

Moreover, it is assumed that the user is not ready at this point to turn the timers on and will do so at another point in the program by setting bit TRx (in TCON) to 1.

Table 3. Timer/Counter 0 Used as a Timer

Mode	Timer 0 Function	TMOD	
		Internal Control ⁽¹⁾	External Control ⁽²⁾
0	13-Bit Timer	00H	08H
1	16-Bit Timer	01H	09H
2	8-Bit Auto-Reload	02H	0AH
3	Two 8-Bit Timers	03H	0BH

Table 4. Timer/Counter 0 Used as a Counter

Mode	Timer 0 Function	TMOD	
		Internal Control ⁽¹⁾	External Control ⁽²⁾
0	13-Bit Timer	04H	0CH
1	16-Bit Timer	05H	0DH
2	8-Bit Auto-Reload	06H	0EH
3	One 8-Bit Counter	07H	0FH

Notes:

1. The Timer is turned ON/OFF by setting/clearing bit TR0 in the software.
2. The Timer is turned ON/OFF by the 1 to 0 transition on INT0 (P3.2) when TR0 = 1 (hardware control).

Table 5. Timer/Counter 1 Used as a Timer

Mode	Timer 1 Function	TMOD	
		Internal Control ⁽¹⁾	External Control ⁽²⁾
0	13-Bit Timer	00H	80H
1	16-Bit Timer	10H	90H
2	8-Bit Auto-Reload	20H	A0H
3	Does Not Run	30H	B0H

Table 6. Timer/Counter 1 Used as a Counter

Mode	Timer 1 Function	TMOD	
		Internal Control ⁽¹⁾	External Control ⁽²⁾
0	13-Bit Timer	40H	C0H
1	16-Bit Timer	50H	D0H
2	8-Bit Auto-Reload	60H	E0H
3	Not Available	—	—

Notes:

1. The Timer is turned ON/OFF by setting/clearing bit TR1 in the software.
2. The Timer is turned ON/OFF by the 1-to-0 transition on INT1 (P3.3) when TR1 = 1 (hardware control).

SERIAL INTERFACE

The Serial port is full duplex, which means it can transmit and receive simultaneously. It is also receive-buffered, which means it can begin receiving a second byte before a previously received byte has been read from the receive register. (However, if the first byte still has not been read when reception of the second byte is complete, one of the bytes will be lost.) The serial port receive and transmit registers are both accessed at Special Function Register SBUF. Writing to SBUF loads the transmit register, and reading SBUF accesses a physically separate receive register.

The serial port can operate in the following four modes:

Mode 0:

Serial data enters and exits through RXD. TXD outputs the shift clock. Eight data bits are transmitted/received, with the LSB first. The baud rate is fixed at 1/12 the oscillator frequency (see Figure 11).

Mode 1:

Ten bits are transmitted (through TXD) or received (through RXD): a start bit (0), eight data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in Special Function Register SCON. The baud rate is variable (see Figure 12).

Mode 2:

Eleven bits are transmitted (through TXD) or received (through RXD): a start bit (0), eight data bits (LSB first), a programmable ninth data bit, and a stop bit (1). On transmit, the ninth data bit (TB8 in SCON) can be assigned the value of 0 or 1. Or, for example, the parity bit (P, in the PSW) can be moved into TB8. On receive, the ninth data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/32 or 1/64 the oscillator frequency (see Figure 13).

Mode 3:

Eleven bits are transmitted (through TXD) or received (through RXD): a start bit (0), eight data bits (LSB first), a programmable ninth data bit, and a stop bit (1). In fact, Mode 3 is the same as Mode 2 in all respects except the baud rate, which is variable in Mode 3 (see Figure 14).

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

Multiprocessor Communications

Modes 2 and 3 have a special provision for multiprocessor communications. In these modes, nine data bits are received, followed by a stop bit. The ninth bit goes into RB8; then comes a stop bit. The port can be programmed such that when the stop bit is received, the serial port interrupt is activated only if RB8 = 1. This feature is enabled by setting bit SM2 in SCON.

The following example shows how to use the serial interrupt for multiprocessor communications. When the master processor must transmit a block of data to one of several slaves, it first sends out an address byte that identifies the target slave. An address byte differs from a data byte in that the ninth bit is 1 in an address byte and 0 in a data byte. With SM2 = 1, no slave is interrupted by a data byte. An address byte, however, interrupts all slaves, so that each slave can examine the received byte and see if it is being addressed. The addressed slave clears its SM2 bit and prepares to receive the data bytes that follows. The slaves that are not addressed set their SM2 bits and ignore the data bytes.

SM2 has no effect in Mode 0 but can be used to check the validity of the stop bit in Mode 1. In a Mode 1 reception, if SM2 = 1, the receive interrupt is not activated unless a valid stop bit is received.

Baud Rates

The baud rate in Mode 0 is fixed as shown in the following equation.

$$\text{Mode 0 Baud Rate} = \frac{\text{Oscillator Frequency}}{12}$$

The baud rate in Mode 2 depends on the value of the SMOD bit in Special Function Register PCON. If SMOD = 0 (the value on reset), the baud rate is 1/64 of the oscillator frequency. If SMOD = 1, the baud rate is 1/32 of the oscillator frequency, as shown in the following equation.

$$\text{Mode 2 Baud Rate} = \frac{2^{\text{SMOD}}}{64} \times (\text{Oscillator Frequency})$$

In the IS89C51, the Timer 1 overflow rate determines the baud rates in Modes 1 and 3.

Using the Timer 1 to Generate Baud Rates

When Timer 1 is the baud rate generator, the baud rates in Modes 1 and 3 are determined by the Timer 1 overflow rate and the value of SMOD according to the following equation.

$$\text{Mode 1, 3 Baud Rate} = \frac{2^{\text{SMOD}}}{32} \times (\text{Timer 1 Overflow Rate})$$

The Timer 1 interrupt should be disabled in this application. The Timer itself can be configured for either timer or counter operation in any of its three running modes. In the most typical applications, it is configured for timer operation in auto-reload mode (high nibble of TMOD = 0010B). In this case, the baud rate is given by the following formula.

$$\text{Mode 1,3 Baud Rate} = \frac{2^{\text{SMOD}}}{32} \times \frac{\text{Oscillator Frequency}}{12 \times [256 - (\text{TH1})]}$$

Programmers can achieve very low baud rates with Timer 1 by leaving the Timer 1 interrupt enabled, configuring the Timer to run as a 16-bit timer (high nibble of TMOD = 0001B), and using the Timer 1 interrupt to do a 16-bit software reload.

Table 7 lists commonly used baud rates and how they can be obtained from Timer 1.

Table 7. Commonly Used Baud Rates Generated by Timer 1

Baud Rate	fosc	SMOD	Timer 1		
			C/T	Mode	Reload Value
Mode 0 Max: 1 MHz	12 MHz	X	X	X	X
Mode 2 Max: 375K	12 MHz	1	X	X	X
Modes 1, 3: 62.5K	12 MHz	1	0	2	FFH
19.2K	11.059 MHz	1	0	2	FDH
9.6K	11.059 MHz	0	0	2	FDH
4.8K	11.059 MHz	0	0	2	FAH
2.4K	11.059 MHz	0	0	2	F4H
1.2K	11.059 MHz	0	0	2	E8H
137.5	11.986 MHz	0	0	2	1DH
110	6 MHz	0	0	2	72H
110	12 MHz	0	0	1	FEEBH

More About Mode 0

Serial data enters and exits through RXD. TXD outputs the shift clock. Eight data bits are transmitted/received, with the LSB first. The baud rate is fixed at 1/12 the oscillator frequency.

Figure 11 shows a simplified functional diagram of the serial port in Mode 0 and associated timing.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal at S6P2 also loads a 1 into the ninth position of the transmit shift register and tells the TX Control block to begin a transmission. The internal timing is such that one full machine cycle will elapse between "write to SBUF" and activation of SEND.

SEND transfer the output of the shift register to the alternate output function line of P3.0, and also transfers SHIFT CLOCK to the alternate output function line of P3.1. SHIFT CLOCK is low during S3, S4, and S5 of every machine cycle, and high during S6, S1, and S2. At S6P2 of every machine cycle in which SEND is active, the contents of the transmit shift register are shifted one position to the right.

As data bits shift out to the right, 0s come in from the left. When the MSB of the data byte is at the output position of the shift register, the 1 that was initially loaded into the ninth position is just to the left of the MSB, and all positions to the left of that contain 0s. This condition flags the TX Control block to do one last shift, then deactivate SEND and set TI. Both of these actions occur at S1P1 of the tenth machine cycle after "write to SBUF."

Reception is initiated by the condition REN = 1 and RI = 0. At S6P2 of the next machine cycle, the RX Control unit writes the bits 11111110 to the receive shift register and activates RECEIVE in the next clock phase.

RECEIVE enables SHIFT CLOCK to the alternate output function line of P3.1. SHIFT CLOCK makes transitions at S3P1 and S6P1 of every machine cycle. At S6P2 of every machine cycle in which RECEIVE is active, the contents of the receive shift register are shifted one position to the left. The value that comes in from the right is the value that was sampled at the P3.0 pin at S5P2 of the same machine cycle.

As data bits come in from the right, 1s shift out to the left. When the 0 that was initially loaded into the right-most position arrives at the left-most position in the shift register, it flags the RX Control block to do one last shift and load SBUF. At S1P1 of the tenth machine cycle after the write to SCON that cleared RI, RECEIVE is cleared and RI is set.

More About Mode 1

Ten bits are transmitted (through TXD), or received (through RXD): a start bit (0), eight data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in SCON. In the IS89C51 the baud rate is determined by the Timer 1 overflow rate.

Figure 12 shows a simplified functional diagram of the serial port in Mode 1 and associated timings for transmit and receive.

Transmission is initiated by any instruction that uses SBUF as a destination register.

The "write to SBUF" signal also loads a 1 into the ninth bit position of the transmit shift register and flags the TX control unit that a transmission is requested. Transmission actually commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. Thus, the bit times are synchronized to the divide-by-16 counter, not to the "write to SBUF" signal.

The transmission begins when SEND is activated, which puts the start bit at TXD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TXD. The first shift pulse occurs one bit time after that.

As data bits shift out to the right, 0s are clocked in from the left. When the MSB of the data byte is at the output position of the shift register, the 1 that was initially loaded into the ninth position is just to the left of the MSB, and all positions to the left of that contain 0s. This condition flags the TX Control unit to do one last shift, then deactivate SEND and set TI. This occurs at the tenth divide-by-16 rollover after "write to SBUF".

Reception is initiated by a 1-to-0 transition detected at RXD. For this purpose, RXD is sampled at a rate of 16 times the established baud rate. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written into the input shift register. Resetting the divide-by-16 counter aligns its rollovers with the boundaries of the incoming bit times.

The 16 states of the counter divide each bit time into 16th. At the seventh, eighth, and ninth counter states of each bit time, the bit detector samples the value of RXD. The value accepted is the value that was seen in at least two of the three samples. This is done to reject noise. In order to reject false bits, if the value accepted during the first bit time is not 0, the receive circuits are reset and the unit continues looking for another 1-to-0 transition. If the start bit is valid, it is shifted into the input shift register, and reception of the rest of the frame proceeds.

As data bits come in from the right, 1s shift to the left. When the start bit arrives at the leftmost position in the shift register, (which is a 9-bit register in Mode 1), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI. The signal to load SBUF and RB8 and to set RI is generated if, and only if, the following conditions are met at the time the final shift pulse is generated.

- 1) RI = 0 and
- 2) Either SM2 = 0, or the received stop bit = 1

If either of these two conditions is not met, the received frame is irretrievably lost. If both conditions are met, the stop bit goes into RB8, the eight data bits go into SBUF, and RI is activated. At this time, whether or not the above conditions are met, the unit continues looking for a 1-to-0 transition in RXD.

More About Modes 2 and 3

Eleven bits are transmitted (through TXD), or received (through RXD): a start bit (0), eight data bits (LSB first), a programmable ninth data bit, and a stop bit (1). On transmit, the ninth data bit (TB8) can be assigned the value of 0 or 1. On receive, the ninth data bit goes into RB8 in SCON. The baud rate is programmable to either 1/32 or 1/64 of the oscillator frequency in Mode 2. Mode 3 may have a variable baud rate generated from Timer 1.

Figures 13 and 14 show a functional diagram of the serial port in Modes 2 and 3. The receive portion is exactly the same as in Mode 1. The transmit portion differs from Mode 1 only in the ninth bit of the transmit shift register.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal also loads TB8 into the ninth bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. Thus, the bit times are synchronized to the divide-by-16 counter, not to the "write to SBUF" signal.

The transmission begins when SEND is activated, which puts the start bit at TXD. One bit timer later, DATA is activated, which enables the output bit of the transmit shift register to TXD. The first shift pulse occurs one bit time after that. The first shift clocks a 1 (the stop bit) into the ninth bit position of the shift register. Thereafter, only 0s are clocked in. Thus, as data bits shift out to the right, 0s are clocked in from the left. When TB8 is at the output position of the shift register, then the stop bit is just to the left of TB8, and all positions to the left of that contain 0s. This condition flags the TX Control unit to do one last shift, then deactivate SEND and set TI. This occurs at the eleventh divide-by-16 rollover after "write to SBUF".

Reception is initiated by a 1-to-0 transition detected at RXD. For this purpose, RXD is sampled at a rate of 16 times the established baud rate. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written to the input shift register.

At the seventh, eighth, and ninth counter states of each bit time, the bit detector samples the value of RXD. The value accepted is the value that was seen in at least two of the three samples. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit continues looking for another 1-to-0 transition. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame proceeds.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in Modes 2 and 3 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI. The signal to load SBUF and RB8 and to set RI is generated if, and only if, the following conditions are met at the time the final shift pulse is generated:

- 1) RI = 0, and
- 2) Either SM2 = 0 or the received ninth data bit = 1

If either of these conditions is not met, the received frame is irretrievably lost, and RI is not set. If both conditions are met, the received ninth data bit goes into RB8, and the first eight data bits go into SBUF. One bit time later, whether the above conditions were met or not, the unit continues looking for a 1-to-0 transition at the RXD input.

Note that the value of the received stop bit is irrelevant to SBUF, RB8, or RI.

Table 8. Serial Port Setup

Mode	SCON	SM2Variation
0	10H	Single Processor Environment (SM2 = 0)
1	50H	
2	90H	
3	D0H	
0	NA	Multiprocessor Environment (SM2 = 1)
1	70H	
2	B0H	
3	F0H	

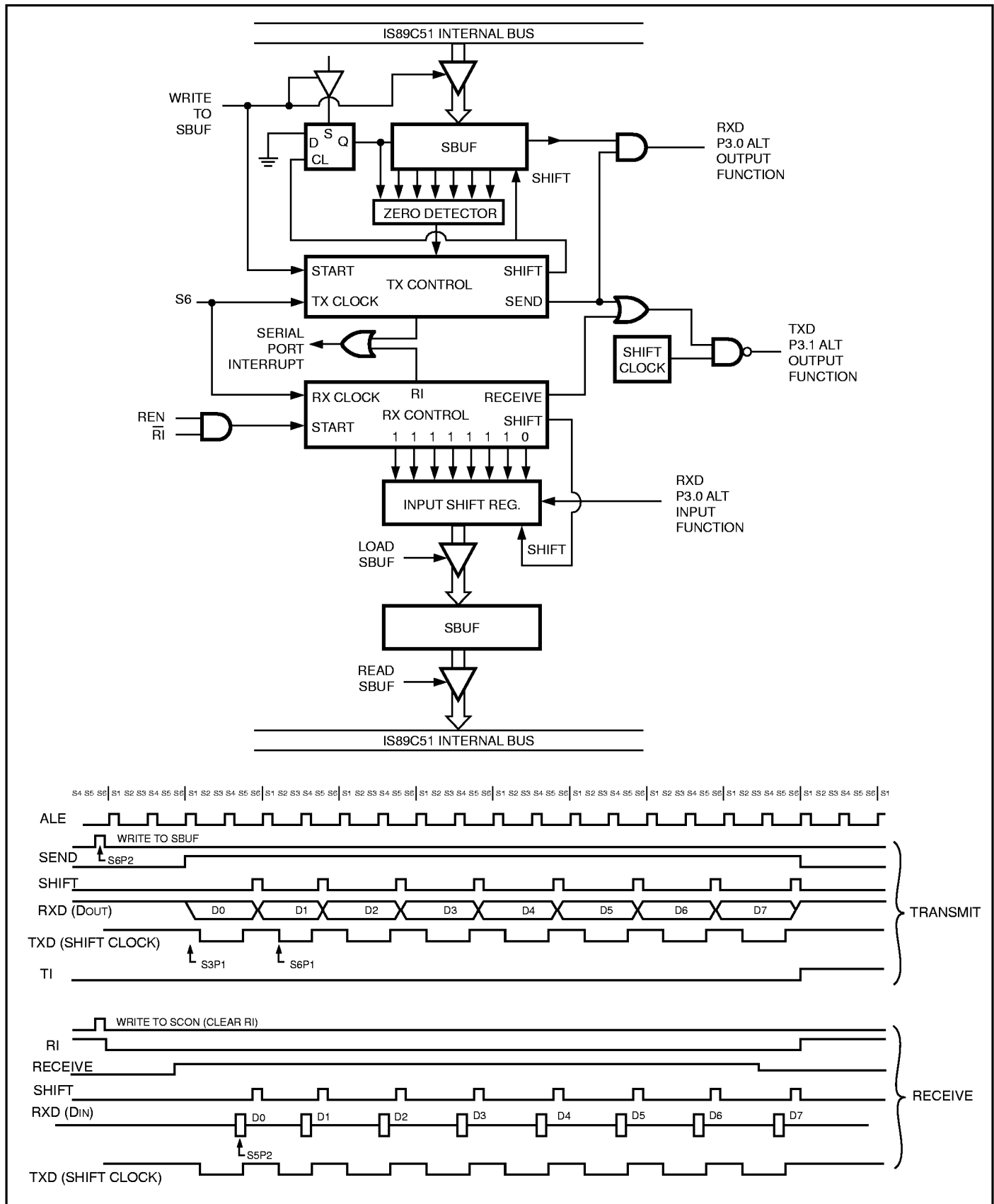


Figure 11. Serial Port Mode 0

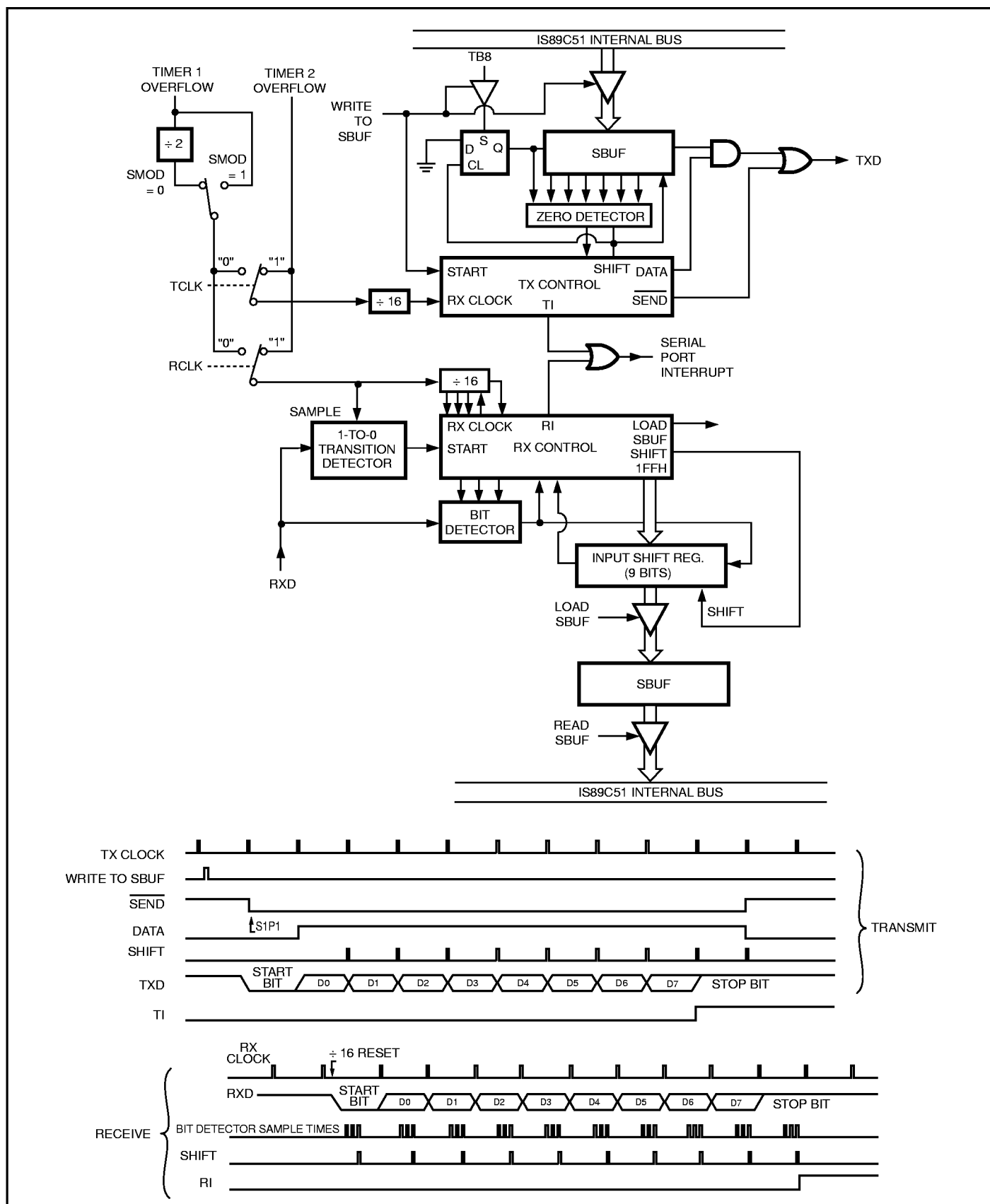


Figure 12. Serial Port Mode 1

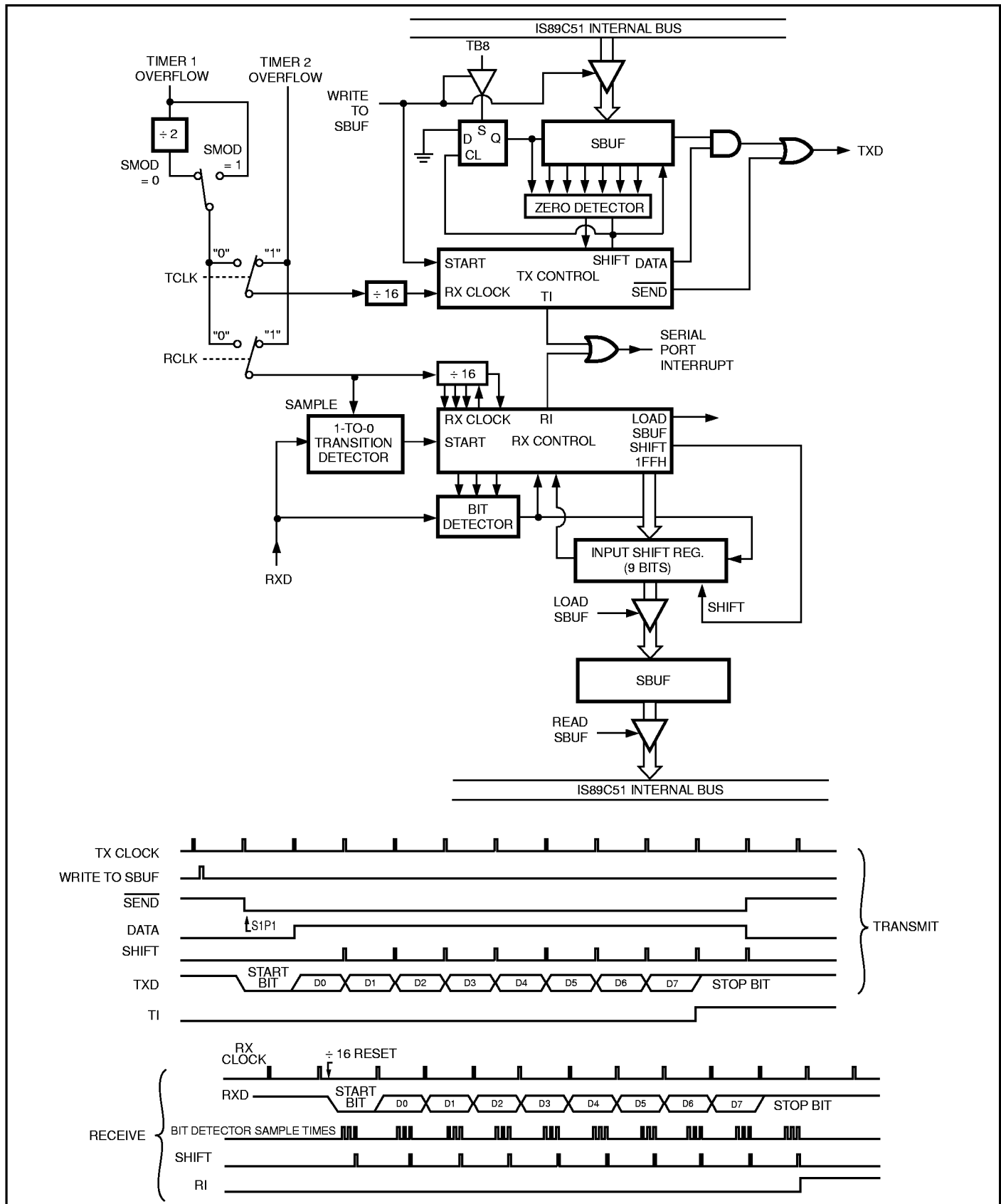


Figure 13. Serial Port Mode 2

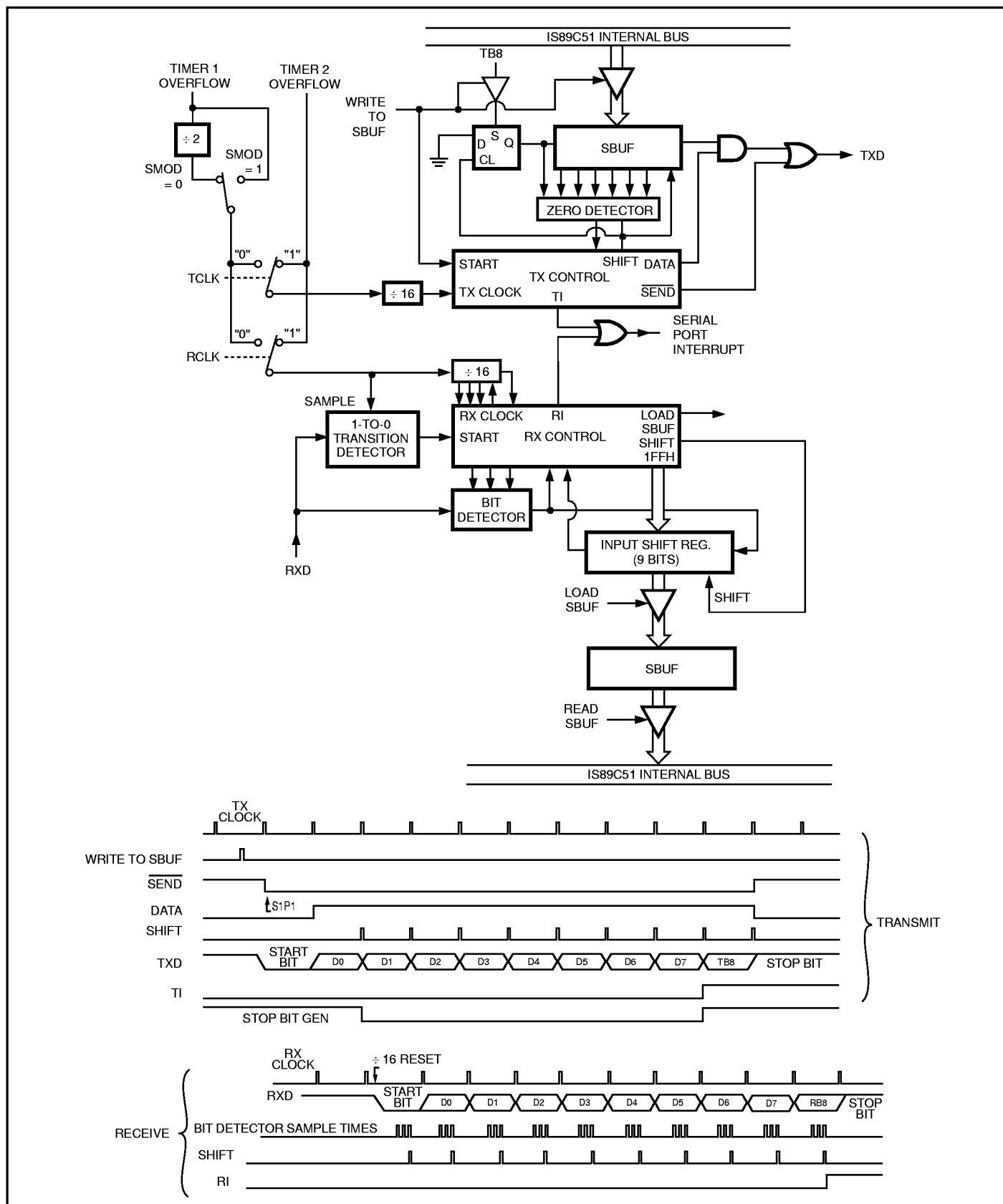


Figure 14. Serial Port Mode 3

INTERRUPT SYSTEM

The IS89C51 provides six interrupt sources: two external interrupts, two timer interrupts, and a serial port interrupt. These are shown in Figure 15.

The External Interrupts $\overline{INT0}$ and $\overline{INT1}$ can each be either level-activated or transition-activated, depending on bits IT0 and IT1 in Register TCON. The flags that actually generate these interrupts are the IE0 and IE1 bits in TCON. When the service routine is vectored, hardware clears the flag that generated an external interrupt only if the interrupt was transition-activated. If the interrupt was level-activated, then the external requesting source (rather than the on-chip hardware) controls the request flag.

The Timer 0 and Timer 1 Interrupts are generated by TF0 and TF1, which are set by a rollover in their respective Timer/Counter registers (except for Timer 0 in Mode 3). When a timer interrupt is generated, the on-chip hardware clears the flag that generated it when the service routine is vectored to.

The Serial Port Interrupt is generated by the logical OR of RI and TI. Neither of these flags is cleared by hardware when the service routine is vectored to. In fact, the service routine normally must determine whether RI or TI generated the interrupt, and the bit must be cleared in software.

All of the bits that generate interrupts can be set or cleared by software, with the same result as though they had been set or cleared by hardware. That is, interrupts can be generated and pending interrupts can be canceled in software.

Each of these interrupt sources can be individually enabled or disabled by setting or clearing a bit in Special Function Register IE (interrupt enable) at address 0A8H. As well as individual enable bits for each interrupt source, there is a global enable/disable bit that is cleared to disable all interrupts or set to turn on interrupts (see SFR IE).

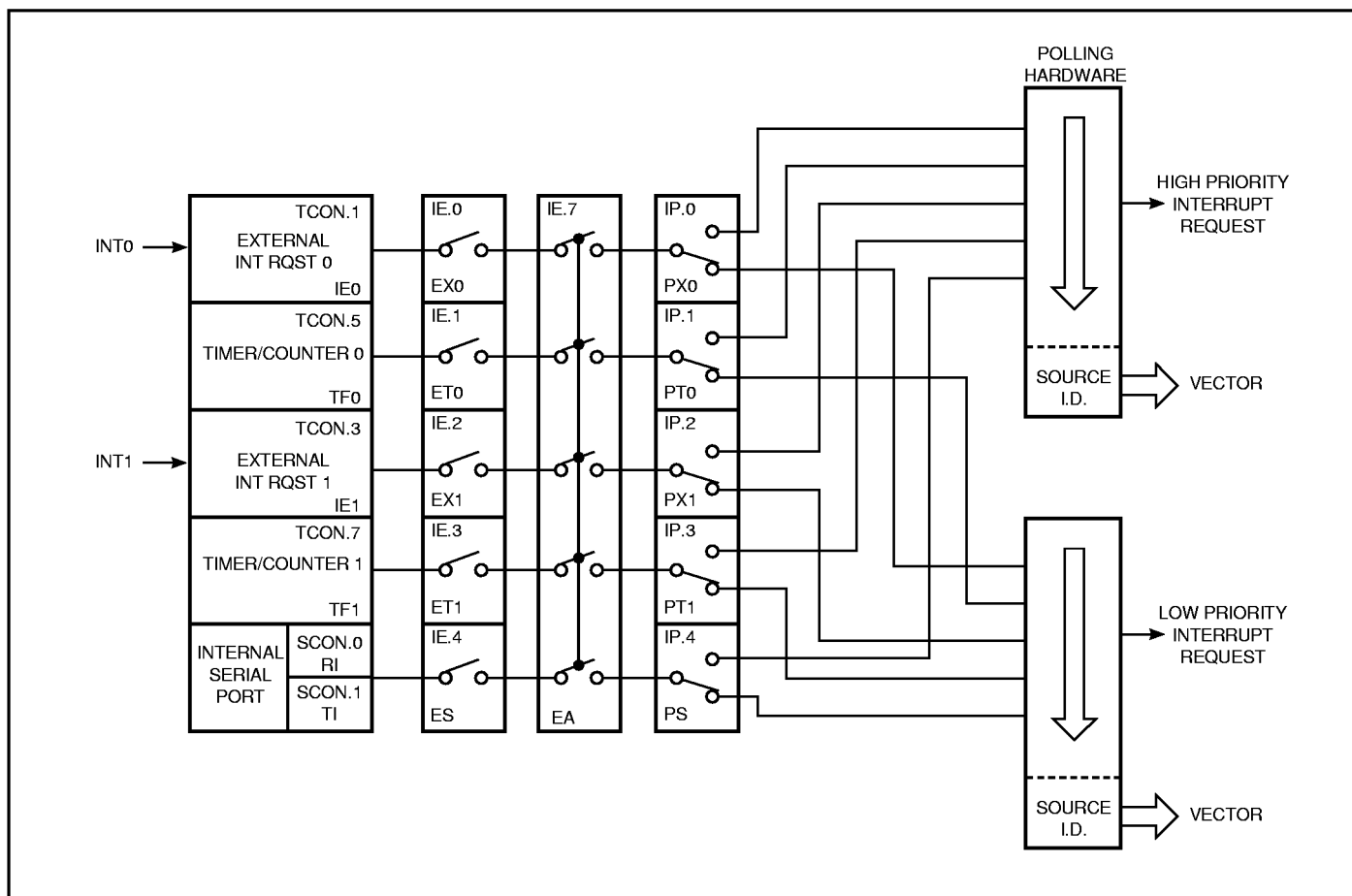


Figure 15. Interrupt System

Priority Level Structure

Each interrupt source can also be individually programmed to one of two priority levels by setting or clearing a bit in Special Function Register IP (interrupt priority) at address 0B8H. IP is cleared after a system reset to place all interrupts at the lower priority level by default. A low-priority interrupt can be interrupted by a high-priority interrupt but not by another low-priority interrupt. A high-priority interrupt can not be interrupted by any other interrupt source.

If two requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus, within each priority level there is a second priority structure determined by the polling sequence, as follows:

	Source	Priority Within Level
1.	IE0	(Highest)
2.	TF0	
3.	IE1	
4.	TF1	
5.	RI + TI	(Lowest)

Note that the "priority within level" structure is only used to resolve *simultaneous requests of the same priority level*.

How Interrupts Are Handled

The interrupt flags are sampled at S5P2 of every machine cycle. The samples are polled during the following machine cycle. If one of the flags was in a set condition at S5P2 of the preceding cycle, the polling cycle will find it and the

interrupt system will generate an LCALL to the appropriate service routine, provided this hardware generated LCALL is not blocked by any of the following conditions:

1. An interrupt of equal or higher priority level is already in progress.
2. The current (polling) cycle is not the final cycle in the execution of the instruction in progress.
3. The instruction in progress is RETI or any write to the IE or IP registers.

Any of these three conditions will block the generation of the LCALL to the interrupt service routine. Condition 2 ensures that the instruction in progress will be completed before vectoring to any service routine. Condition 3 ensures that if the instruction in progress is RETI or any access to IE or IP, then at least one more instruction will be executed before any interrupt is vectored to.

The polling cycle is repeated with each machine cycle, and the values polled are the values that were present at S5P2 of the previous machine cycle. If an active interrupt flag is not being serviced because of one of the above conditions and is not still active when the blocking condition is removed, the denied interrupt will not be serviced. In other words, the fact that the interrupt flag was once active but not serviced is not remembered. Every polling cycle is new. The polling cycle/LCALL sequence is illustrated in Figure 16.

Note that if an interrupt of higher priority level goes active prior to S5P2 of the machine cycle labeled C3 in Figure 16, then in accordance with the above rules it will be serviced during C5 and C6, without any instruction of the lower priority routine having been executed.

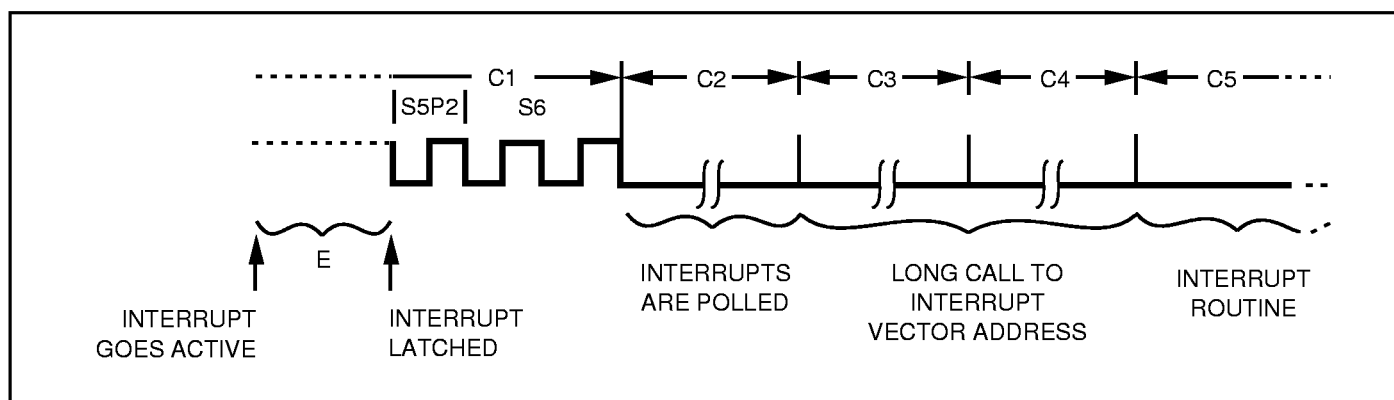


Figure 16. Interrupt Response Timing Diagram

Thus, the processor acknowledges an interrupt request by executing a hardware-generated LCALL to the appropriate servicing routine. In some cases it also clears the flag that generated the interrupt, and in other cases it does not. It never clears the Serial Port flags. This must be done in the user's software. The processor clears an external interrupt flag (IE0 or IE1) only if it was transition-activated. The hardware-generated LCALL pushes the contents of the Program Counter onto the stack (but it does not save the PSW) and reloads the PC with an address that depends on the source of the interrupt being serviced, as shown in the following table.

Interrupt Source	Interrupt Request Bits	Cleared by Hardware	Vector Address
$\overline{\text{INT0}}$	IE0	No (level) Yes (trans.)	0003H
Timer 0	TF0	Yes	000BH
$\overline{\text{INT1}}$	IE1	No (level) Yes (trans.)	0013H
Timer 1	TF1	Yes	001BH
Serial Port	RI, TI	No	0023H
System Reset	RST		0000H

Execution proceeds from that location until the RETI instruction is encountered. The RETI instruction informs the processor that this interrupt routine is no longer in progress, then pops the top two bytes from the stack and reloads the Program Counter. Execution of the interrupted program continues from where it left off.

Note that a simple RET instruction would also have returned execution to the interrupted program, but it would have left the interrupt control system thinking an interrupt was still in progress.

Interrupt	Flag	SFR Register and Bit Position
External 0	IE0	TCON.1
External 1	IE1	TCON.3
Timer 1	TF1	TCON.7
Timer 0	TF0	TCON.5
Serial Port	TI	SCON.1
Serial Port	RI	SCON.0

When an interrupt is accepted, the following action occurs:

1. The current instruction completes operation.
2. The PC is saved on the stack.
3. The current interrupt status is saved internally.
4. Interrupts are blocked at the level of the interrupts.
5. The PC is loaded with the vector address of the ISR (interrupts service routine).
6. The ISR executes.

The ISR executes and takes action in response to the interrupt. The ISR finishes with RETI (return from interrupt) instruction. This retrieves the old value of the PC from the stack and restores the old interrupt status. Execution of the main program continues where it left off.

External Interrupts

The external sources can be programmed to be level-activated or transition-activated by setting or clearing bit IT1 or IT0 in Register TCON. If ITx = 0, external interrupt x is triggered by a detected low at the $\overline{\text{INTx}}$ pin. If ITx = 1, external interrupt x is edge-triggered. In this mode if successive samples of the $\overline{\text{INTx}}$ pin show a high in one cycle and a low in the next cycle, interrupt request flag IEx in TCON is set. Flag bit IEx then requests the interrupt.

Since the external interrupt pins are sampled once each machine cycle, an input high or low should hold for at least 12 oscillator periods to ensure sampling. If the external interrupt is transition-activated, the external source has to hold the request pin high for at least one machine cycle, and then hold it low for at least one machine cycle to ensure that the transition is seen so that interrupt request flag IEx will be set. IEx will be automatically cleared by the CPU when the service routine is called.

If the external interrupt is level-activated, the external source has to hold the request active until the requested interrupt is actually generated. Then the external source must deactivate the request before the interrupt service routine is completed, or else another interrupt will be generated.

Response Time

The $\overline{\text{INT0}}$ and $\overline{\text{INT1}}$ levels are inverted and latched into the interrupt flags IE0 and IE1 at S5P2 of every machine cycle. Similarly, the Serial Port flags RI and TI are set at S5P2. The values are not actually polled by the circuitry until the next machine cycle.

The Timer 0 and Timer 1 flags, TF0 and TF1, are set at S5P2 of the cycle in which the timers overflow. The values are then polled by the circuitry in the next cycle.

If a request is active and conditions are right for it to be acknowledged, a hardware subroutine call to the requested service routine will be the next instruction executed. The call itself takes two cycles. Thus, a minimum of three complete machine cycles elapsed between activation of an external interrupt request and the beginning of execution of the first instruction of the service routine. Figure 16 shows response timings.

A longer response time results if the request is blocked by one of the three previously listed conditions. If an interrupt of equal or higher priority level is already in progress, the additional wait time depends on the nature of the other interrupt's service routine. If the instruction in progress is not in its final cycle, the additional wait time cannot be more than three cycles, since the longest instructions (MUL and DIV) are only four cycles long. If the instruction in progress is RETI or an access to IE or IP, the additional wait time cannot be more than five cycles (a maximum of one more cycle to complete the instruction in progress, plus four cycles to complete the next instruction if the instruction is MUL or DIV).

Thus, in a single-interrupt system, the response time is always more than three cycles and less than nine cycles.

Single-Step Operation

The IS89C51 interrupt structure allows single-step execution with very little software overhead. As previously noted, an interrupt request will not be serviced while an interrupt of equal priority level is still in progress, nor will it be serviced after RETI until at least one other instruction has been executed. Thus, once an interrupt routine has been entered, it cannot be re-entered until at least one instruction of the interrupted program is executed. One way to use this feature for single-step operation is to program one of the external interrupts (for example, $\overline{\text{INT0}}$) to be level-activated. The service routine for the interrupt will terminate with the following code:

```
JNB    P3.2,$    ;Wait Here Till  $\overline{\text{INT0}}$  Goes High
JB     P3.2,$    ;Now Wait Here Till it Goes Low
RETI                               ;Go Back and Execute One
                                   Instruction
```

If the $\overline{\text{INT0}}$ pin, which is also the P3.2 pin, is held normally low, the CPU will go right into the External Interrupt 0 routine and stay there until $\overline{\text{INT0}}$ is pulsed (from low-to-high-to-low). Then it will execute RETI, go back to the task program, execute one instruction, and immediately re-enter the External Interrupt 0 routine to await the next pulsing of P3.2. One step of the task program is executed each time P3.2 is pulsed.

OTHER INFORMATION

Reset

The reset input is the RST pin, which is the input to a Schmitt Trigger.

A reset is accomplished by holding the RST pin high for at least two machine cycles (24 oscillator periods), *while the oscillator is running*. The CPU responds by generating an internal reset, with the timing shown in Figure 18.

The external reset signal is asynchronous to the internal clock. The RST pin is sampled during State 5 Phase 2 of every machine cycle. The port pins will maintain their current activities for 19 oscillator periods after a logic 1 has been sampled at the RST pin; that is, for 19 to 31 oscillator periods after the external reset signal has been applied to the RST pin.

The internal reset algorithm writes 0s to all the SFRs except the port latches, the Stack Pointer, and SBUF. The port latches are initialized to FFH, the Stack Pointer to 07H, and SBUF is indeterminate. Table 9 lists the SFRs and their reset values.

Then internal RAM is not affected by reset. On power-up the RAM content is indeterminate.

Table 9. Reset Values of the SFR's

SFR Name	Reset Value
PC	0000H
ACC	00H
B	00H
PSW	00H
SP	07H
DPTR	0000H
P0-P3	FFH
IP	XXX00000B
IE	0XX00000B
TMOD	00H
TCON	00H
TH0	00H
TL0	00H
TH1	00H
TL1	00H
SCON	00H
SBUF	Indeterminate
PCON	0XXX0000B

Power-on Reset

For the IS89C51, the external resistor can be removed because the RST pin has an internal pulldown. The capacitor value can then be reduced to 1 μF (see Figure 17).

When power is turned on, the circuit holds the RST pin high for an amount of time that depends on the value of the capacitor and the rate at which it charges. To ensure a good reset, the RST pin must be high long enough to allow the oscillator time to start-up (normally a few msec) plus two machine cycles.

Note that the port pins will be in a random state until the oscillator has start and the internal reset algorithm has written 1s to them.

With this circuit, reducing V_{CC} quickly to 0 causes the RST pin voltage to momentarily fall below 0V. However, this voltage is internally limited and will not harm the device.

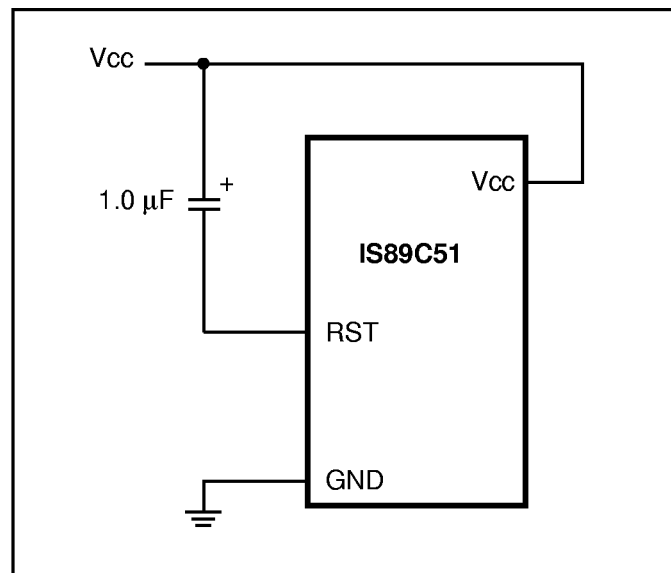


Figure 17. Power-On Reset Circuit

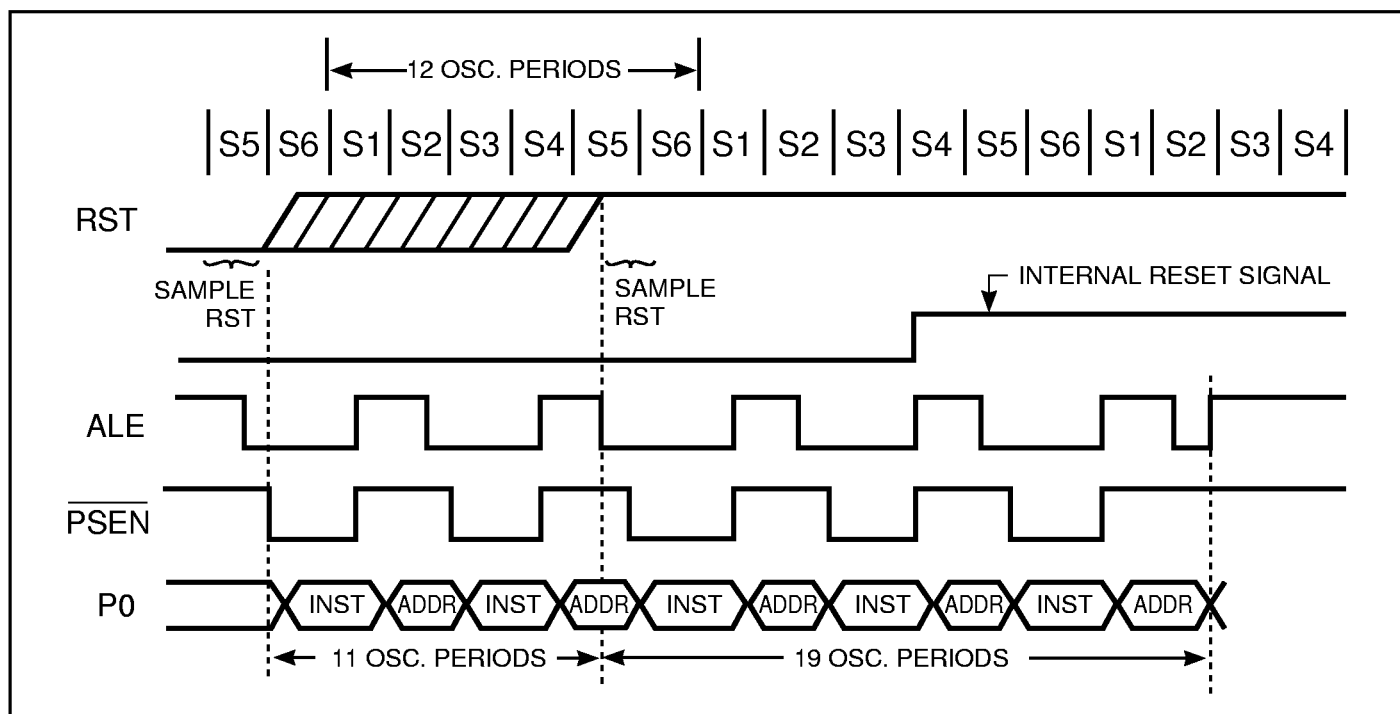


Figure 18. Reset Timing

Power-Saving Modes of Operation

The IS89C51 has two power-reducing modes. Idle and Power-down. The input through which backup power is supplied during these operations is V_{CC} . Figure 19 shows the internal circuitry which implements these features. In the Idle mode ($IDL = 1$), the oscillator continues to run and the Interrupt, Serial Port, and Timer blocks continue to be clocked, but the clock signal is gated off to the CPU. In Power-down ($PD = 1$), the oscillator is frozen. The Idle and Power-down modes are activated by setting bits in Special Function Register PCON.

Idle Mode

An instruction that sets PCON.0 is the last instruction executed before the Idle mode begins. In the Idle mode, the internal clock signal is gated off to the CPU, but not to the Interrupt, Timer, and Serial Port functions. The CPU status is preserved in its entirety; the Stack Pointer, Program Counter, Program Status Word, Accumulator, and all other registers maintain their data during Idle. The port pins hold the logical states they had at the time Idle was activated. ALE and \overline{PSEN} hold at logic high levels.

There are two ways to terminate the Idle. Activation of any enabled interrupt will cause PCON.0 to be cleared by hardware, terminating the Idle mode. The interrupt will be serviced, and following RETI the next instruction to be executed will be the one following the instruction that put the device into Idle.

The flag bits GF0 and GF1 can be used to indicate whether an interrupt occurred during normal operation or during an Idle. For example, an instruction that activates Idle can also set one or both flag bits. When Idle is terminated by an interrupt, the interrupt service routine can examine the flag bits.

The other way of terminating the Idle mode is with a hardware reset. Since the clock oscillator is still running, the hardware reset must be held active for only two machine cycles (24 oscillator periods) to complete the reset.

The signal at the RST pin clears the IDL bit directly and asynchronously. At this time, the CPU resumes program execution from where it left off; that is, at the instruction following the one that invoked the Idle Mode. As shown in Figure 18, two or three machine cycles of program execution may take place before the internal reset algorithm takes control. On-chip hardware inhibits access to the internal RAM during this time, but access to the port pins is not inhibited. To eliminate the possibility of unexpected outputs at the port pins, the instruction following the one that invokes Idle should not write to a port pin or to external data RAM.

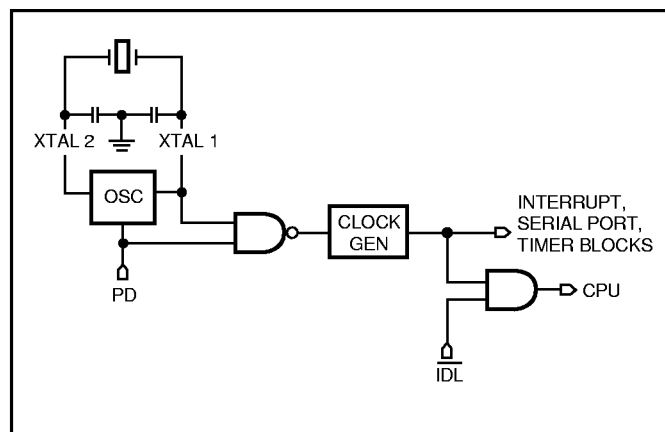


Figure 19. Idle and Power-Down Hardware

Power-down Mode

An instruction that sets PCON.1 is the last instruction executed before Power-down mode begins. In the Power-down mode, the on-chip oscillator stops. With the clock frozen, all functions are stopped, but the on-chip RAM and Special function Registers are held. The port pins output the values held by their respective SFRs. ALE and \overline{PSEN} output lows.

In the Power-down mode of operation, V_{CC} can be reduced to as low as 2V. However, V_{CC} must not be reduced before the Power-down mode is invoked, and V_{CC} must be restored to its normal operating level before the Power-down mode is terminated. The reset that terminates Power-down also frees the oscillator. The reset should not be activated before V_{CC} is restored to its normal operating level and must be held active long enough to allow the oscillator to restart and stabilize (normally less than 10 msec).

The only exit from Power-down is a hardware reset. Reset redefines all the SFRs but does not change the on-chip RAM.

Table 10. Status of the External Pins During Idle and Power-down Modes.

Mode	Memory	ALE	PSEN	PORT 0	PORT 1	PORT 2	PORT 3
Idle	Internal	1	1	Data	Data	Data	Data
Idle	External	1	1	Float	Data	Address	Data
Power-down	Internal	0	0	Data	Data	Data	Data
Power-down	External	0	0	Float	Data	Data	Data

On-Chip Oscillators

The on-chip oscillator circuitry of the IS89C51 is a single stage inverter, intended for use as a crystal-controlled, positive reactance oscillator. In this application the crystal is operated in its fundamental response mode as an inductive reactance in parallel resonance with capacitance external to the crystal (Figure 20). Examples of how to drive the clock with external oscillator are shown in Figure 21.

The crystal specifications and capacitance values (C1 and C2 in Figure 20) are not critical. 20 pF to 30 pF can be used in these positions at a 12 MHz to 24 MHz frequency with good quality crystals. (For ranges greater than 24 MHz refer to Figure 22.) A ceramic resonator can be used in place of the crystal in cost-sensitive applications. When a ceramic resonator is used, C1 and C2 are normally selected to be of somewhat higher values. The manufacturer of the ceramic resonator should be consulted for recommendation on the values of these capacitors.

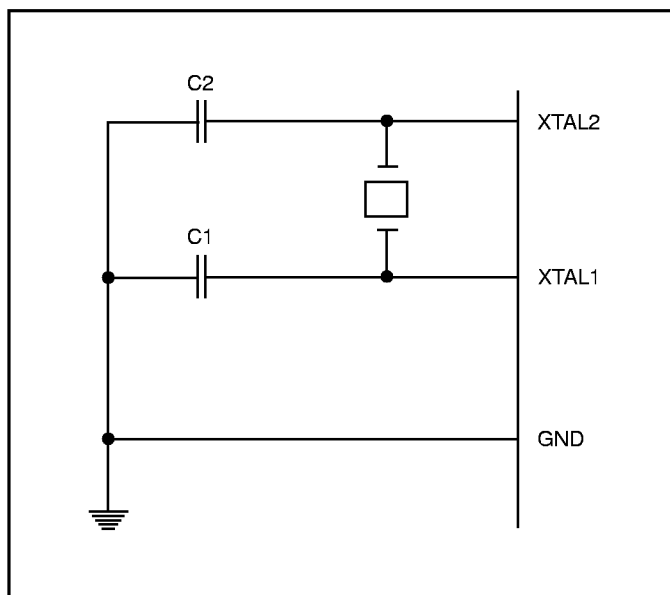


Figure 20. Oscillator Connections

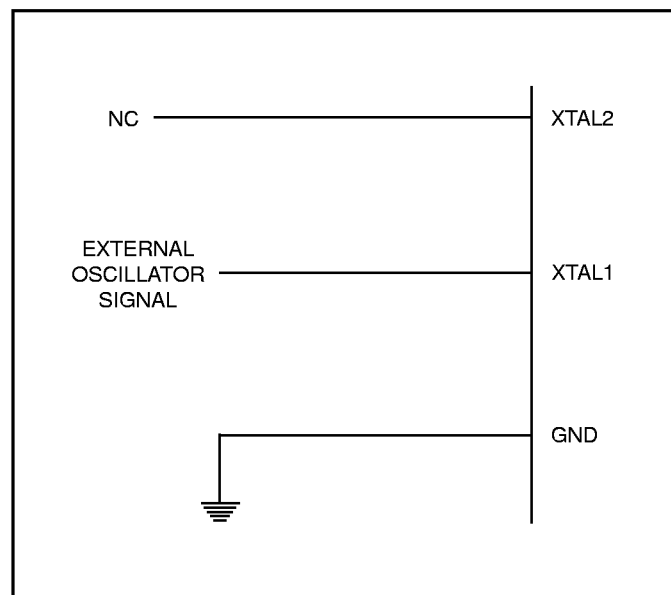


Figure 21. External Clock Drive Configuration

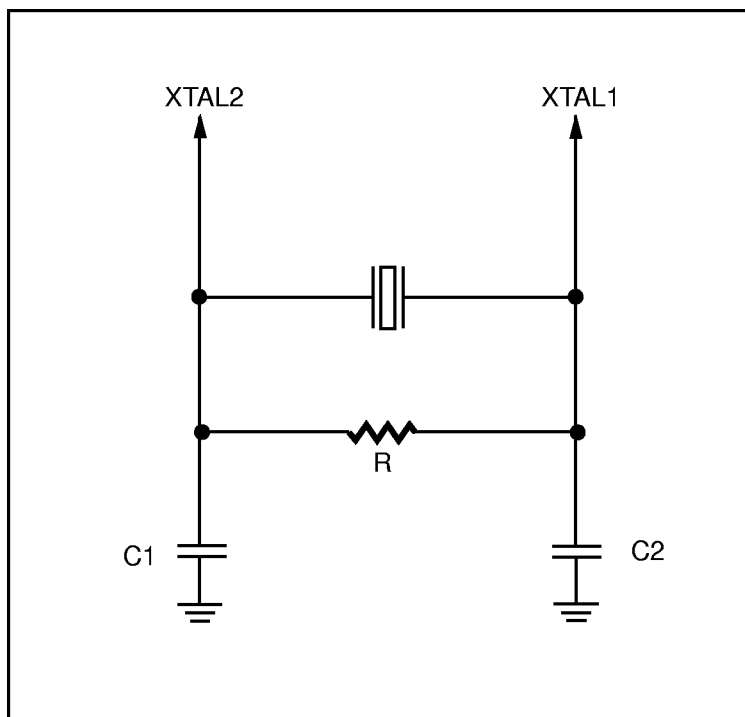


Figure 22. Oscillator Connections for High Speed (> 24 MHz)

Note:

When the frequency is higher than 24 MHz, please refer to Table 11 for recommended values of C1, C2, and R.

Table 11. Recommended Value for C1, C2, R

	Frequency Range	
	4 MHz - 24 MHz	30 MHz - 40 MHz
C1	20 pF-30 pF	3 pF-10 pF
C2	20 pF-30 pF	3 pF-10 pF
R	Not Apply	6.2K-10K

Program Memory Lock System

The program lock system, when programmed, protects the code against software piracy. The IS89C51 has a two-level program lock system (see Table 12) and a 32-byte encryption table. No matter what lock bit is, the user submits the encryption table with his or her code in verify mode. The lock-bits and encryption array are programmed in the same manner as the program memory.

Encryption Array

The Encryption Array is a 32-byte Flash memory array that are normally unprogrammed (all 1's). Every time that a byte is addressed during verify, five address lines are used to select a byte of the Encryption Array. This byte is then exclusive-NOR'ed (XNOR) with the code byte, creating an Encryption verify byte. The algorithm, with the array in the unprogrammed state (all 1's), will return the code in its original form.

When using the Encryption Array, one important factor needs to be considered. If a code byte has the value 0FFH, verifying the byte will produce the encryption byte value. If a large block (> 32 bytes) of code is left unprogrammed, a verification routine will display the contents of the encryption array. For this reason, all unused code bytes should be programmed with some value other than 0FFH, and not all of them the same value. This will ensure maximum program protection.

Table 12. Program Lock Bits

	LB1	LB2	LB3	Protection Type
1	U	U	U	No Program Lock Features enabled. (Code verify will still be encrypted by the Encryption Array if Programmed.)
2	P	U	U	MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory, \overline{EA} is sampled and latched on Reset and further programming of the Flash is disabled.
3	P	P	U	Same as 2, also verify is disabled.
4	P	P	P	Same as 3, also external execution is disabled.

FLASH MEMORY







Programming the IS89C51

The IS89C51 is normally shipped with the on-chip Flash memory array in the erased state (i.e., contents = FFH) and ready to be programmed. The IS89C51 is programmed byte-by-byte in programming mode. Before the on-chip flash code memory can be re-programmed, the entire memory array must be erased electrically.

Programming Interface

Every code byte in the Flash array can be written and the entire array can be erased using the appropriate combination of control signals. The write operation cycles is self-timed once initiated, will automatically time itself to completion. The programming interface is shown in Table 13 and Figures 23 and 24.

Table 13. Flash Programming Mode

Mode	RST	PSEN	ALE/PROG	EA/V _{PP}	P2.6	P2.7	P3.6	P3.7
Program Code Data	H	L		12V	L	H	H	H
Verify Code Data	H	L	H	H	L	L	H	H
Program Encryption Array Address 0-1F	H	L		12V	L	H	L	H
Program Lock Bit 1	H	L		12V	H	H	H	H
Program Lock Bit 2	H	L		12V	H	H	L	L
Program Lock Bit 3	H	L		12V	H	L	H	L
Read Signature Byte	H	L	H	H	L	L	L	L
Chip Erase	H	L		12V	H	L	L	L

Programming Algorithm

Before programming the IS89C51, the control signals, the address, data should be setup according to the programming mode table and programming interface. To program the IS89C51, the following sequence should be followed:

1. Insert the desired memory location on the address bus.
2. Insert the appropriate data byte on the data bus.
3. Active the correct combination of control signals.
4. Raise \overline{EA} / V_{PP} to 12V.
5. Pulse ALE / \overline{PROG} once to program a byte in the Flash array, Encryption array or the lock bits.
6. Set \overline{EA} / V_{PP} to 5 V and verify data. If the data is correct then execute step 7, otherwise execute steps 1-6.
7. Repeat steps 1 through 6, changing the address and data for the entire array or until the end of the object file is reached.

Program Verify

If lock bits LB1, LB2 and LB3 and the Encryption Array have not been programmed, the programmed code data can be read back via the address and data lines for verification. The lock bits and encryption array cannot be verified directly. Verification of lock bits is achieved by observing that their features are enabled.

To verify the data after all addresses are programmed completely, power down the IS89C51 and then reapply power. The programmed data can then be verified by applying the verify signals to the device.

Erasing Chip

All Flash memory cells must be programmed to '00' (include encryption array and lock bits) before the chip is erased. The entire Flash array is erased electrically by using the proper combination of control signals and by holding ALE/ \overline{PROG} low for t_{GLGHE} duration (See Table 14. Flash Programming and Verification Characteristics for t_{GLGHE} value.) After the chip is erased, the code array, encryption array, and lock bits are written with all "1"s. If any Flash memory cell is not '1' (including encryption array and lock bits), repeat the chip erase again. The chip erase operation must be executed successfully before the code memory can be re-programmed.

Reading the Signature Bytes

The signature bytes are read by the same procedure as a normal verification of locations 030H, 031H and 032H, except that P3.6 and P3.7 need to be pulled to a logic low. The values returned are:

Signature Location	Value
(030H)	D5H indicates manufactured by ISSI
(031H)	51H indicates IS89C51
(032H)	FFH indicates programming voltage is 12V

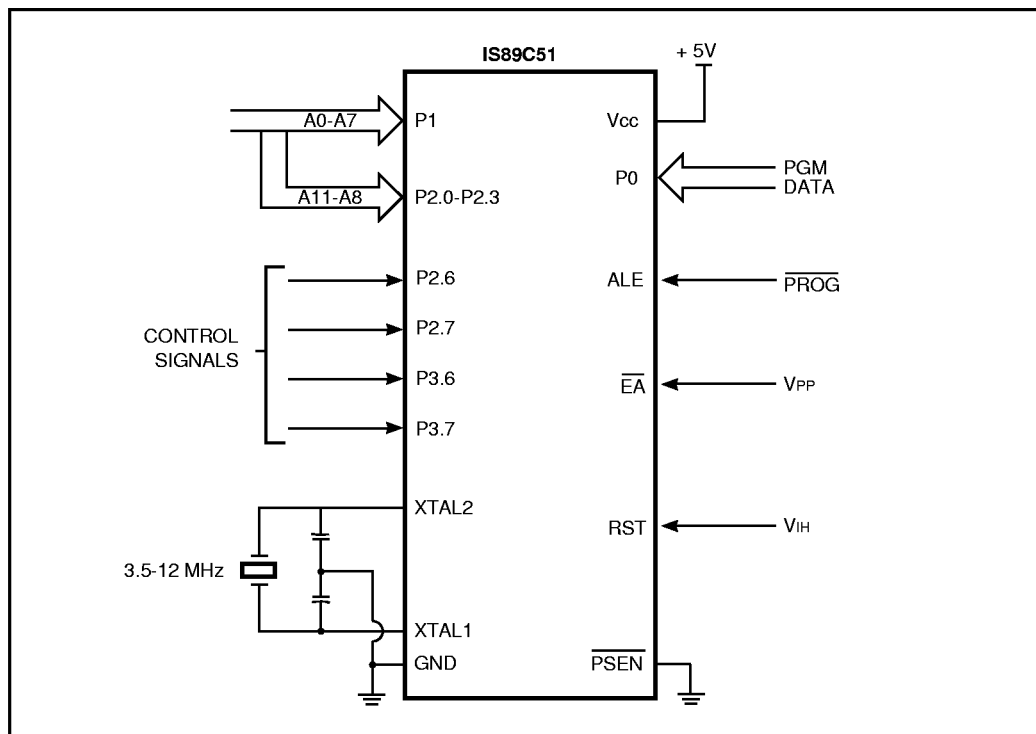


Figure 23. Programming the Flash Memory

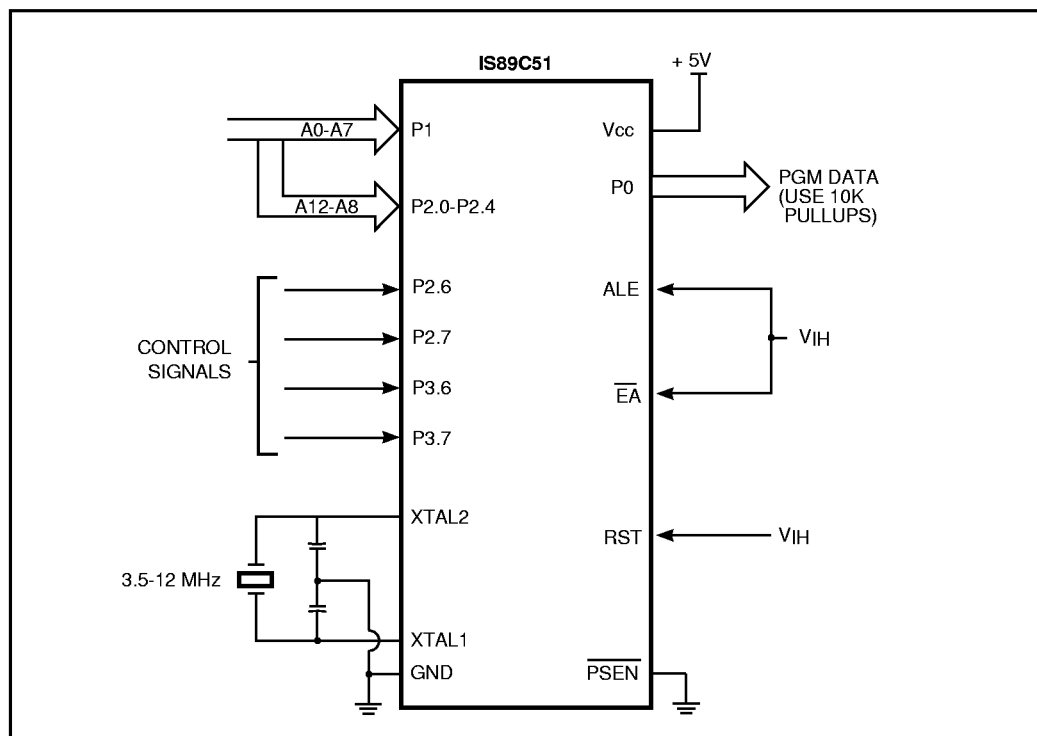


Figure 24. Verifying the Flash Memory

FLASH PROGRAMMING AND VERIFICATION CHARACTERISTICS AND WAVEFORMS

Table 14. Flash Programming and Verification Characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
V _{PP}	Programming Supply Voltage	11.5	12.5	V
1/t _{CLCL}	Oscillator Frequency	3.5	12	MHz
t _{AVGL}	Address Setup to $\overline{\text{PROG}}$ Low	48	—	t _{CLCL}
t _{GHAX}	Address Hold after $\overline{\text{PROG}}$	48	—	t _{CLCL}
t _{DVGL}	Data Setup to $\overline{\text{PROG}}$ Low	48	—	t _{CLCL}
t _{GHDX}	Data Hold after $\overline{\text{PROG}}$	48	—	t _{CLCL}
t _{EHSH}	COND ENABLE to V _{PPH}	48	—	t _{CLCL}
t _{SHGL}	V _{PPH} Setup to $\overline{\text{PROG}}$ Low	10	—	μs
t _{GHSL}	V _{PPH} Hold after $\overline{\text{PROG}}$	10	—	μs
t _{GLGH}	$\overline{\text{PROG}}$ Pulse Width	120	—	μs
t _{GHGL}	$\overline{\text{PROG}}$ High to $\overline{\text{PROG}}$ Low	10	—	μs
t _{AVQV}	Address to Data Valid	—	48	t _{CLCL}
t _{ELQV}	COND ENABLE to Data Valid	—	48	t _{CLCL}
t _{EHQZ}	Data Float after COND DISABLE	—	48	t _{CLCL}
t _{GLGHE}	Erase $\overline{\text{PROG}}$ Pulse Width	200	—	ms
t _{ELPL}	COND DISABLE to Power Low	0	—	ns
t _{PLPH}	Power Off Time	10	—	μs
t _{PHEH}	Power On to V _{PPL}	10	—	ms
t _{EHVH}	V _{PPL} to COND ENABLE	0	—	ns

Note:

1. T_A = 21°C to 27°C, V_{CC} = 5.0 ± 10%.
2. COND ENABLE and COND DISABLE are generated and depend on the control signals on pins P2.6, P2.7, P3.6 and P3.7. The signals set the device in to and out of different processing conditions, such as programming condition, erasing condition, and verify condition.

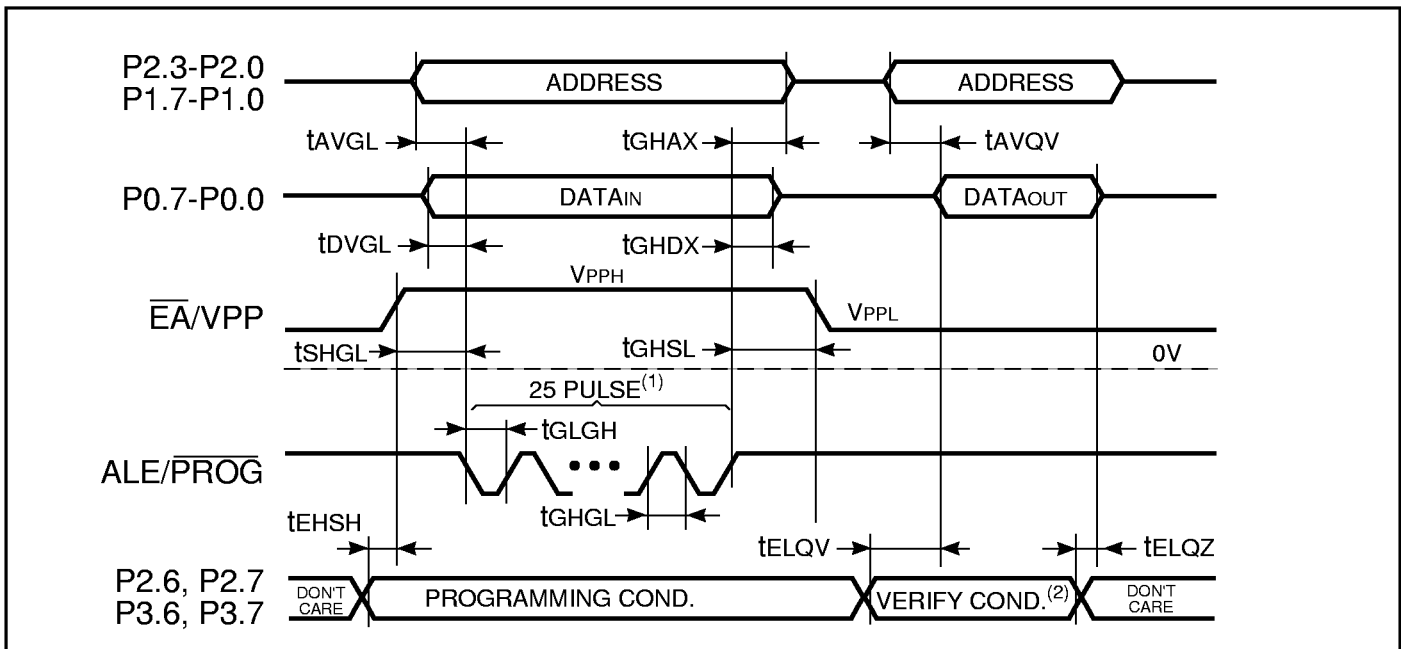


Figure 25. Flash Memory Programming and Verification Timing Waveform

Notes:

1. One pulse for the main code array, 25 pulses for the encryption array and lock bits.
2. This verify condition is using at main code verification.
3. Power off waveform not shown.

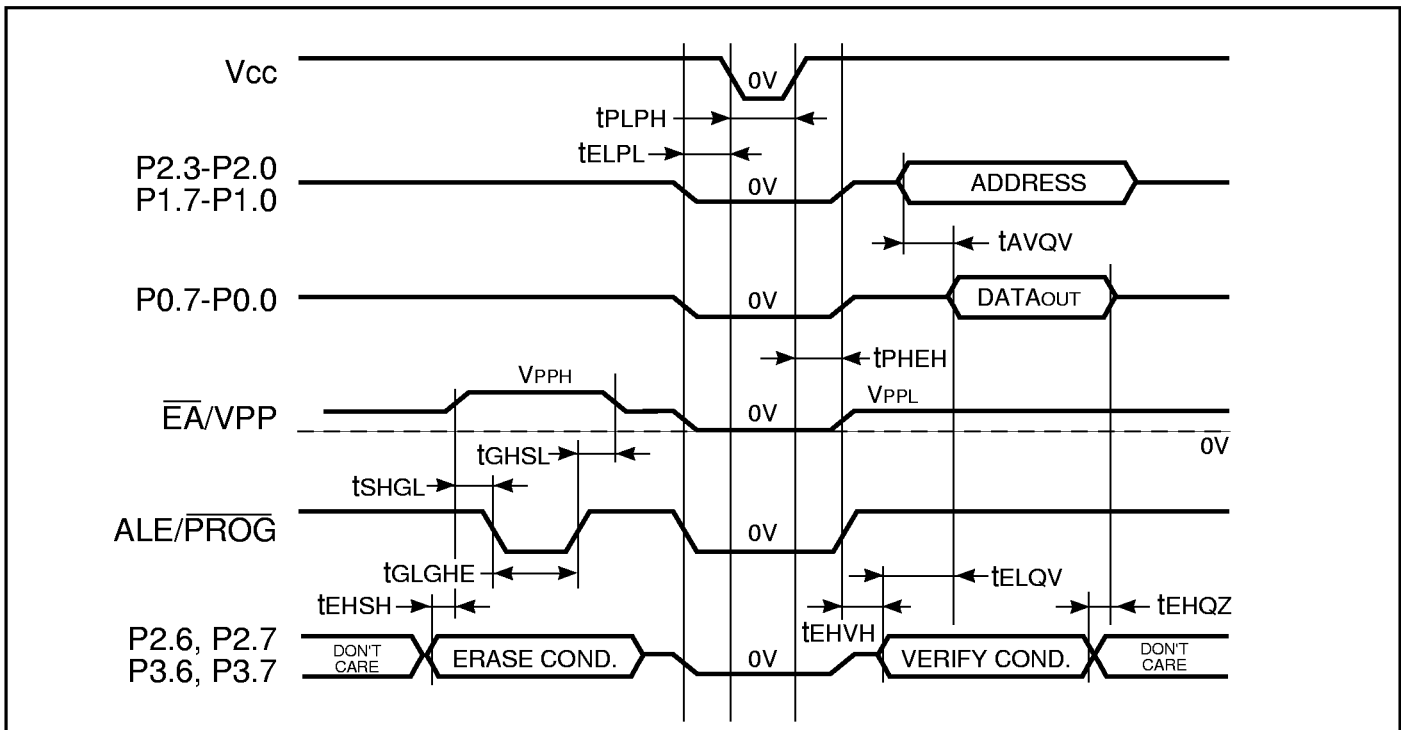


Figure 26. Flash Memory Erase Timing Waveform

Note:

1. The power off and power on waveform can be used in programming or erasing.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

Symbol	Parameter	Value	Unit
V _{TERM}	Terminal Voltage with Respect to GND ⁽²⁾	–2.0 to +7.0	V
T _{BIAS}	Temperature Under Bias ⁽³⁾	0 to +70	°C
T _{STG}	Storage Temperature	–65 to +125	°C
P _T	Power Dissipation	1.5	W

Note:

1. Stress greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.
2. Minimum DC input voltage is –0.5V. During transitions, inputs may undershoot to –2.0V for periods less than 20 ns. Maximum DC voltage on output pins is V_{CC} + 0.5V which may overshoot to V_{CC} + 2.0V for periods less than 20 ns.
3. Operating temperature is for commercial products only defined by this specification.

OPERATING RANGE⁽¹⁾

Range	Ambient Temperature	V _{CC}	Oscillator Frequency
Commercial	0°C to +70°C	5V ± 10%	3.5 to 40 MHz

Note:

1. Operating ranges define those limits between which the functionality of the device is guaranteed.

DC CHARACTERISTICS(T_A = 0°C to 70°C; V_{CC} = 5V ± 10%; GND = 0V)

Symbol	Parameter	Test conditions	Min	Max	Unit
V _{IL}	Input low voltage (All except \overline{EA})		-0.5	0.2V _{CC} - 0.1	V
V _{IL1}	Input low voltage (\overline{EA})		-0.5	0.2V _{CC} - 0.3	V
V _{IH}	Input high voltage (All except XTAL 1, RST)		0.2V _{CC} + 0.9	V _{CC} + 0.5	V
V _{IH1}	Input high voltage (XTAL 1)		0.7V _{CC}	V _{CC} + 0.5	V
V _{SCH+}	RST positive schmitt-trigger threshold voltage		0.7V _{CC}	V _{CC} + 0.5	V
V _{SCH-}	RST negative schmitt-trigger threshold voltage		0	0.2V _{CC}	V
V _{OL} ⁽¹⁾	Output low voltage (Ports 1, 2, 3)	I _{OL} = 100 μA	—	0.3	V
		I _{OL} = 1.6 mA	—	0.45	V
		I _{OL} = 3.5 mA	—	1.0	V
V _{OL1} ⁽¹⁾	Output low voltage (Port 0, ALE, \overline{PSEN})	I _{OL} = 200 μA	—	0.3	V
		I _{OL} = 3.2 mA	—	0.45	V
		I _{OL} = 7.0 mA	—	1.0	V
V _{OH}	Output high voltage (Ports 1, 2, 3, ALE, \overline{PSEN})	I _{OH} = -10 μA V _{CC} = 4.5V-5.5V	0.9V _{CC}	—	V
		I _{OL} = -25 μA	0.75V _{CC}	—	V
		I _{OL} = -60 μA	2.4	—	V
V _{OH1}	Output high voltage (Port 0, ALE, \overline{PSEN})	I _{OH} = -80 μA V _{CC} = 4.5V-5.5V	0.9V _{CC}	—	V
		I _{OH} = -300 μA	0.75V _{CC}	—	V
		I _{OH} = -800 μA	2.4	—	V
I _{IL}	Logical 0 input current (Ports 1, 2, 3)	V _{IN} = 0.45V	—	-80	μA
I _{LI}	Input leakage current (Port 0)	0.45V < V _{IN} < V _{CC}	-10	+10	μA
I _{TL}	Logical 1-to-0 transition current (Ports 1, 2, 3)	V _{IN} = 2.0V	—	-650	μA
R _{RST}	RST pulldown resistor		50	300	KΩ

Note:1. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:Maximum I_{OL} per port pin: 10 mAMaximum I_{OL} per 8-bit port

Port 0: 26 mA

Ports 1, 2, 3: 15 mA

Maximum total I_{OL} for all output pins: 71 mAIf I_{OL} exceeds the test condition, V_{OL} may exceed the related specification.

Pins are not guaranteed to sink greater than the listed test conditions.

POWER SUPPLY CHARACTERISTICS

Symbol	Parameter	Test conditions	Min	Max	Unit
I_{CC}	Power supply current ⁽¹⁾	$V_{CC} = 5.0V$			
	Active mode	12 MHz	—	20	mA
		16 MHz	—	26	mA
		20 MHz	—	32	mA
		24 MHz	—	38	mA
		32 MHz	—	50	mA
		40 MHz	—	62	mA
	Idle mode	12 MHz	—	5	mA
		16 MHz	—	6	mA
		20 MHz	—	7.6	mA
		24 MHz	—	9	mA
		32 MHz	—	12	mA
		40 MHz	—	15	mA
	Power-down mode	$V_{CC} = 5V$	—	50	μA

Note:

1. See Figures 27, 28, 29, and 30 for I_{CC} test conditions.

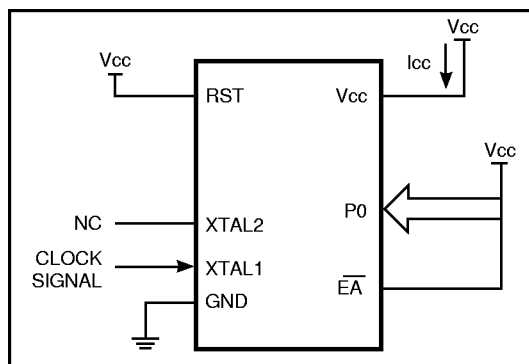


Figure 27. Active Mode

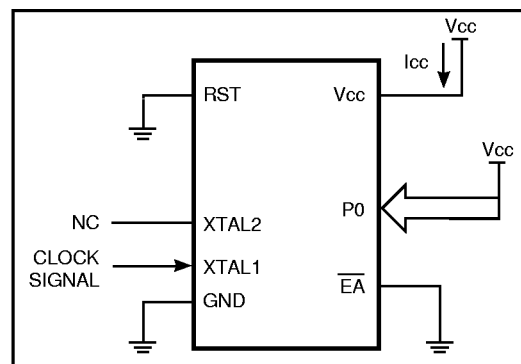


Figure 28. Idle Mode

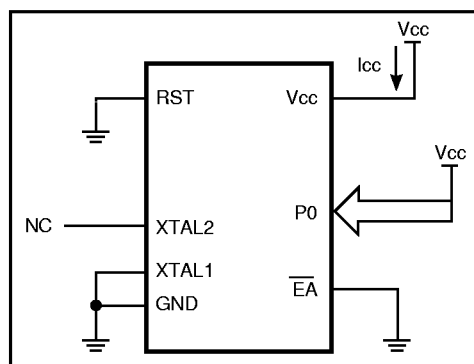


Figure 29. Power-down Mode

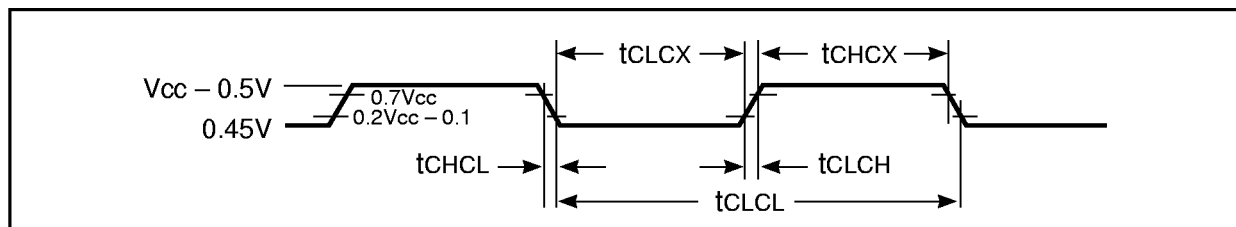


Figure 30. Icc Test Conditions

Note:

1. Clock signal waveform for Icc tests in active and idle mode ($t_{CLCH} = t_{CHCL} = 5 \text{ ns}$)

AC CHARACTERISTICS

($T_A = 0^\circ\text{C TO } 70^\circ\text{C}$; $V_{CC} = 5\text{V} \pm 10\%$; $\text{GND} = 0\text{V}$; C_L for Port 0, ALE and $\overline{\text{PSEN}}$ Outputs = 100 pF;

C_L for Other Outputs = 80 pF)

EXTERNAL MEMORY CHARACTERISTICS

Symbol	Parameter	24 MHz Clock		40 MHz Clock		Variable Oscillator (3.5 - 40 MHz)		Unit
		Min	Max	Min	Max	Min	Max	
$1/t_{CLCL}$	Oscillator frequency	—	—	—	—	3.5	40	MHz
t_{LHLL}	ALE pulse width	68	—	35	—	$2t_{CLCL}-15$	—	ns
t_{AVLL}	Address valid to ALE low	26	—	10	—	$t_{CLCL}-15$	—	ns
t_{LLAX}	Address hold after ALE low	31	—	15	—	$t_{CLCL}-10$	—	ns
t_{LLIV}	ALE low to valid instr in	—	147	—	80	—	$4t_{CLCL}-20$	ns
t_{LLPL}	ALE low to $\overline{\text{PSEN}}$ low	31	—	15	—	$t_{CLCL}-10$	—	ns
t_{PLPH}	$\overline{\text{PSEN}}$ pulse width	110	—	60	—	$3t_{CLCL}-15$	—	ns
t_{PLIV}	$\overline{\text{PSEN}}$ low to valid instr in	—	105	—	55	—	$3t_{CLCL}-20$	ns
t_{PXIX}	Input instr hold after $\overline{\text{PSEN}}$	0	—	0	—	0	—	ns
t_{PXIZ}	Input instr float after $\overline{\text{PSEN}}$	—	37	—	20	—	$t_{CLCL}-5$	ns
t_{AVIV}	Address to valid instr in	—	188	—	105	—	$5t_{CLCL}-20$	ns
t_{PLAZ}	$\overline{\text{PSEN}}$ low to address float	—	10	—	10	—	10	ns
t_{RLRH}	$\overline{\text{RD}}$ pulse width	230	—	130	—	$6t_{CLCL}-20$	—	ns
t_{WLWH}	$\overline{\text{WR}}$ pulse width	230	—	130	—	$6t_{CLCL}-20$	—	ns
t_{RLDV}	$\overline{\text{RD}}$ low to valid data in	—	157	—	90	—	$4t_{CLCL}-10$	ns
t_{RHDX}	Data hold after $\overline{\text{RD}}$	0	—	0	—	0	—	ns
t_{RHDZ}	Data float after $\overline{\text{RD}}$	—	78	—	45	—	$2t_{CLCL}-5$	ns
t_{LLDV}	ALE low to valid data in	—	282	—	165	—	$7t_{CLCL}-10$	ns
t_{AVDV}	Address to valid data in	—	323	—	190	—	$8t_{CLCL}-10$	ns
t_{LLWL}	ALE low to $\overline{\text{RD}}$ or $\overline{\text{WR}}$ low	105	145	55	95	$3t_{CLCL}-20$	$3t_{CLCL}+20$	ns
t_{AVWL}	Address to $\overline{\text{RD}}$ or $\overline{\text{WR}}$ low	146	—	80	—	$4t_{CLCL}-20$	—	ns
t_{QVWX}	Data valid to $\overline{\text{WR}}$ transition	26	—	10	—	$t_{CLCL}-15$	—	ns
t_{WHQX}	Data hold after $\overline{\text{WR}}$	31	—	15	—	$t_{CLCL}-10$	—	ns
t_{RLAZ}	$\overline{\text{RD}}$ low to address float	—	0	—	0	—	0	ns
t_{WHLH}	$\overline{\text{RD}}$ or $\overline{\text{WR}}$ high to ALE high	26	57	10	40	$t_{CLCL}-15$	$t_{CLCL}+15$	ns

SERIAL PORT TIMING: SHIFT REGISTER MODE

Symbol	Parameter	24 MHz Clock		40 MHz Clock		Variable Oscillator (3.5 - 40 MHz)		Unit
		Min	Max	Min	Max	Min	Max	
txLXL	Serial port clock cycle time	490	510	290	310	$12t_{CLCL}-10$	$12t_{CLCL}+10$	ns
tQVXH	Output data setup to clock rising edge	406	—	240	—	$10t_{CLCL}-10$	—	ns
txHQX	Output data hold after clock rising edge	73	—	40	—	$2t_{CLCL}-10$	—	ns
txHDX	Input data hold after clock rising edge	0	—	0	—	0	—	ns
txHDV	Clock rising edge to input data valid	—	417	—	250	—	$10t_{CLCL}$	ns

EXTERNAL CLOCK DRIVE CHARACTERISTICS

Symbol	Parameter	Min	Max	Unit
$1/t_{CLCL}$	Oscillator Frequency	3.5	40	MHz
tCHCX	High time	10	—	ns
tCLCX	Low time	10	—	ns
tCLCH	Rise time	—	10	ns
tCHCL	Fall time	—	10	ns

TIMING WAVEFORMS

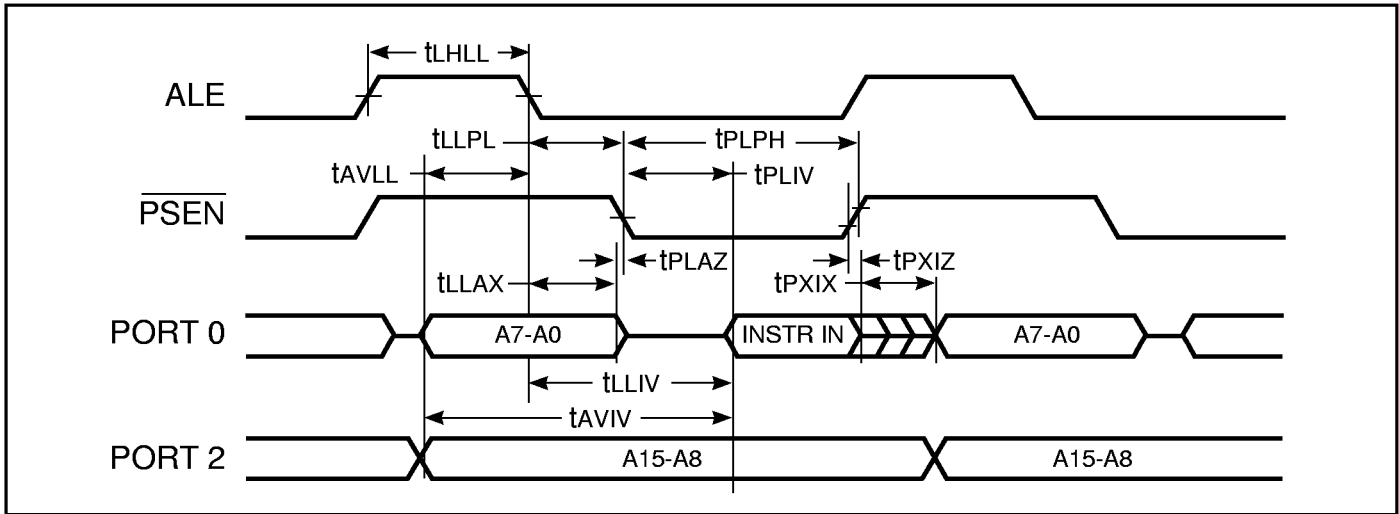


Figure 31. External Program Memory Read Cycle

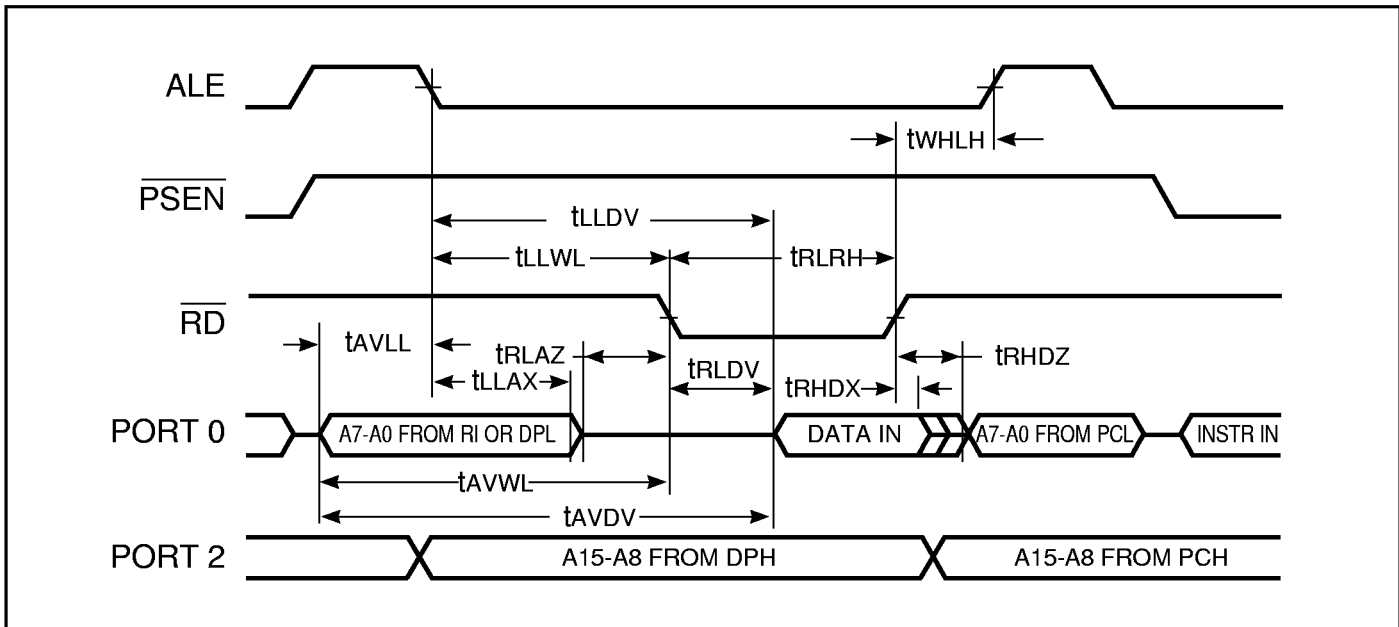


Figure 32. External Data Memory Read Cycle

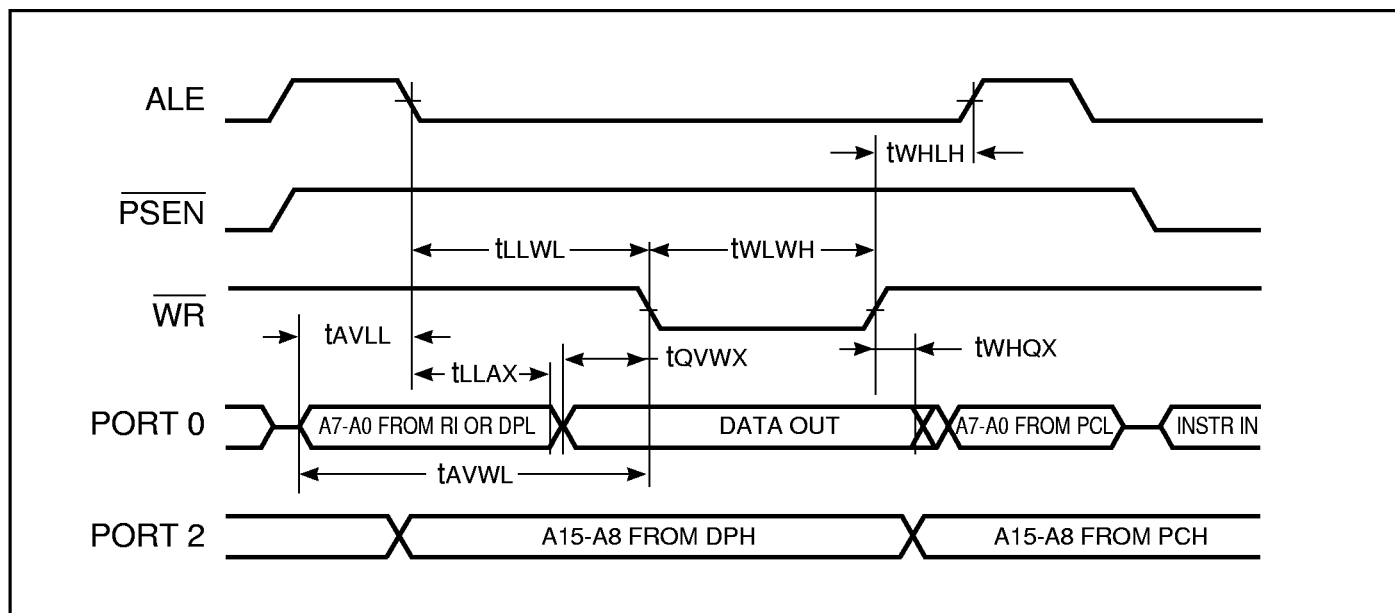


Figure 33. External Data Memory Write Cycle

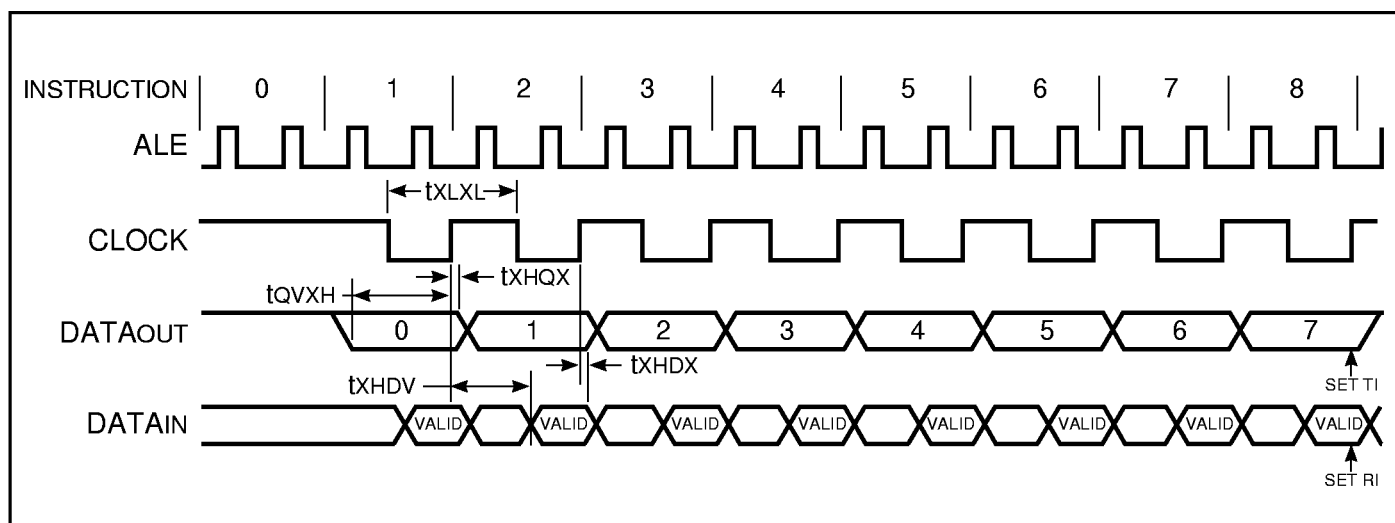


Figure 34. Shift Register Mode Timing Waveform

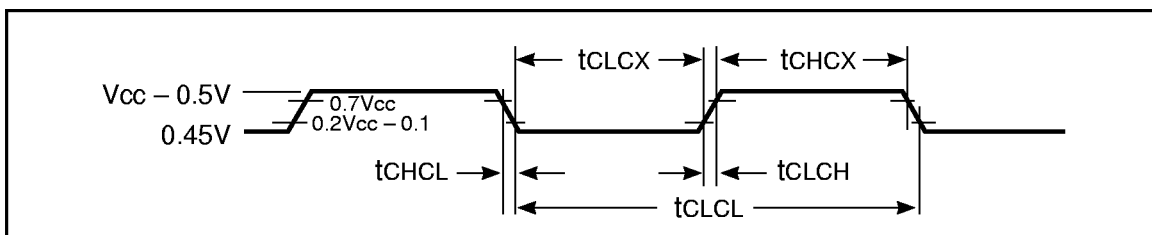


Figure 35. External Clock Drive Waveform

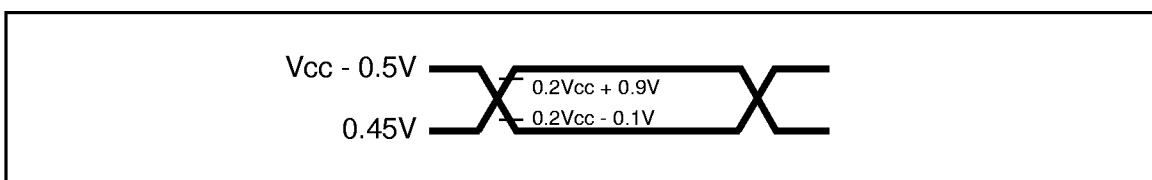


Figure 36. AC Test Point

Note:

1. AC inputs during testing are driven at $V_{CC} - 0.5V$ for logic "1" and 0.45V for logic "0". Timing measurements are made at V_{IH} min for logic "1" and max for logic "0".

ORDERING INFORMATION

Speed	Order Part Number	Package
12 MHz	IS89C51-12PL IS89C51-12W	PLCC – Plastic Leaded Chip Carrier 600-mil Plastic DIP
20 MHz	IS89C51-20PL IS89C51-20W	PLCC – Plastic Leaded Chip Carrier 600-mil Plastic DIP
24 MHz	IS89C51-24PL IS89C51-24W	PLCC – Plastic Leaded Chip Carrier 600-mil Plastic DIP
40 MHz	IS89C51-40PL IS89C51-40W	PLCC – Plastic Leaded Chip Carrier 600-mil Plastic DIP

ISSI®

Integrated Silicon Solution, Inc.

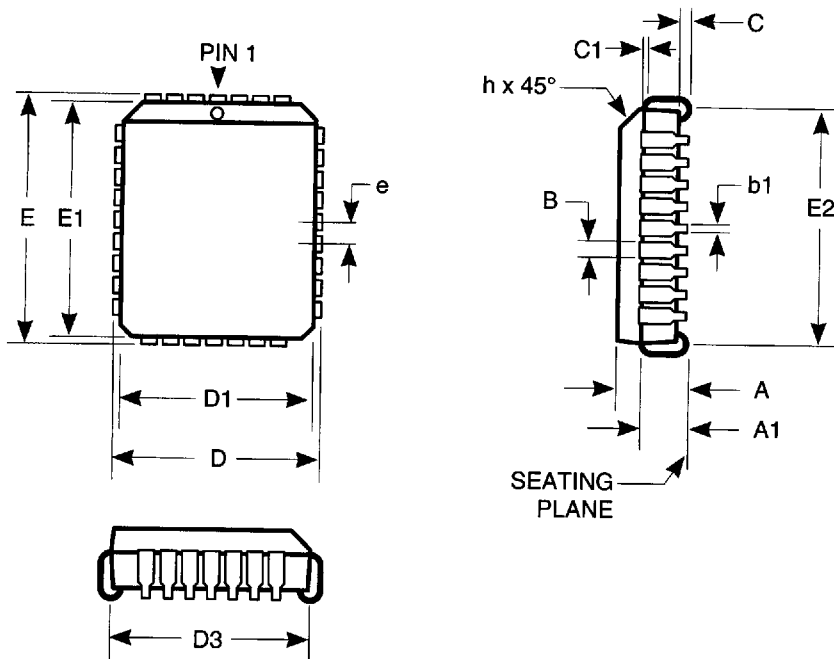
2231 Lawson Lane
 Santa Clara, CA 95054
 Fax: (408) 588-0806
 Toll Free: 1-800-379-4774
<http://www.issi.com>

PACKAGING INFORMATION

ISSI

PLCC (Plastic Leaded Chip Carrier)

Package Code: PL



Plastic Leaded Chip Carrier (PL)

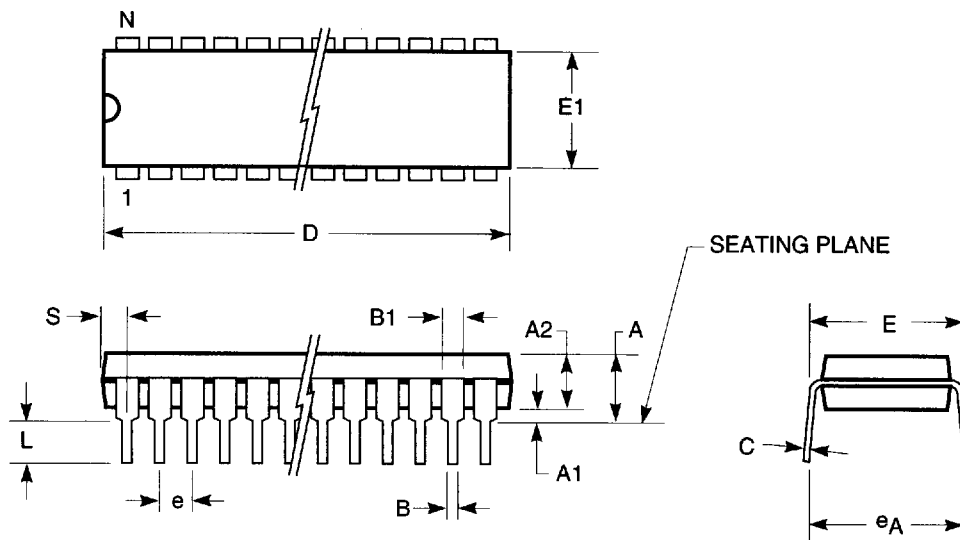
Inches				
Symbol	Min	Max	Min	Max
Ref. Std.				
No. Leads	32		44	
A	0.131	0.140	0.165	0.180
A1	0.075	0.095	0.090	0.120
B	0.026	0.032	0.026	0.032
b1	0.013	0.021	0.013	0.021
C	0.020	0.040	0.020	0.040
C1	0.008	0.014	0.008	0.014
D	0.485	0.495	0.685	0.695
D1	0.447	0.453	0.650	0.656
D2	0.390	0.430	—	—
E	0.585	0.595	0.685	0.695
E1	0.547	0.553	0.650	0.656
E2	0.490	0.530	0.659	0.680
e	0.050 BSC		0.050 BSC	
ND/NE	7/9		11/11	

Notes:

1. Controlling dimension: inches, unless otherwise specified.
2. BSC = Basic lead spacing between centers.
3. Dimensions D and E do not include mold flash protrusions.
4. Formed leads shall be planar with respect to one another within 0.004 inches at the seating plane.
5. ND and NE represent the number of leads in D and E directions, respectively.
6. D1 and E1 should be measured from the bottom of the package.

600-mil Plastic DIP

Package Code: W



600-mil Plastic DIP (W)

Inches

Symbol	Min	Max	Min	Max	Min	Max
Ref. Std.						
N	28		32		40	
A	0.160	0.185	0.165	0.180	0.165	0.200
A1	0.015	0.035	0.010	—	0.020	0.045
B	0.015	0.020	0.018		0.015	0.022
B1	0.050	0.065	0.050		0.045	0.067
C	0.008	0.012	0.010		0.008	0.015
D	1.420	1.460	1.645	1.655	1.415	1.460
E	0.600	0.620	0.590	0.610	0.600	0.620
E1	0.530	0.550	0.540	0.555	0.530	0.560
eA	0.610	0.670	0.620	0.680	0.600	0.680
e	0.100	BSC	0.100	BSC	0.100	BSC
L	0.120	0.150	0.120	0.140	0.120	0.138
S	0.055	0.080	0.065	0.085	0.055	0.085
α	0°	15°	2°	8°	—	—

Notes:

1. Controlling dimension: inches, unless otherwise specified.
2. BSC = Basic lead spacing between centers.
3. Dimensions D and E1 do not include mold flash protrusions and should be measured from the bottom of the package.
4. Formed leads shall be planar with respect to one another within 0.004 inches at the seating plane.