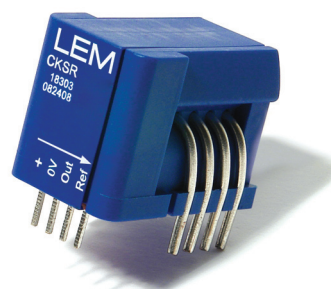


## Current Transducer CKSR series

$I_{PN} = 6, 15, 25, 50 \text{ A}$

Ref: CKSR 6-NP, CKSR 15-NP, CKSR 25-NP, CKSR 50-NP

For the electronic measurement of current: DC, AC, pulsed..., with galvanic isolation between the primary and the secondary circuit.



### Features

- Closed loop (compensated) multi-range current transducer
- Voltage output
- Single supply
- Isolated plastic case material recognized according to UL 94-V0
- Compact design for PCB mounting.

### Advantages

- Very low temperature coefficient of offset
- Very good dv/dt immunity
- Higher creepage / clearance distances
- Reduced height
- Reference pin with two modes: Ref IN and Ref OUT
- Extended measuring range for unipolar measurement.

### Applications

- AC variable speed and servo motor drives
- Static converters for DC motor drives
- Battery supplied applications
- Uninterruptible Power Supplies (UPS)
- Switched Mode Power Supplies (SMPS)
- Power supplies for welding applications
- Solar inverters.

### Standards

- EN 50178: 1997
- IEC 60950-1: 2006
- IEC 61010-1: 2010
- IEC 61326-1: 2012
- UL 508: 2004.

### Application Domain

- Industrial.

**Absolute maximum ratings**

Parameter	Symbol	Unit	Value
Supply voltage	$V_C$	V	7
Primary conductor temperature		°C	110
Maximum primary current	$I_{P\ max}$	A	$20 \times I_{PN}$
ESD rating, Human Body Model (HBM)		kV	4

Stresses above these ratings may cause permanent damage. Exposure to absolute maximum ratings for extended periods may degrade reliability.

**Isolation characteristics**

Parameter	Symbol	Unit	Value	Comment
RMS voltage for AC isolation test 50/60Hz/1 min	$V_d$	kV	4.3	
Impulse withstand voltage 1.2/50 $\mu$ s	$\hat{V}_w$	kV	8	
Partial discharge extinction voltage @ 10 pC (rms)	$V_e$	V	1000	
Clearance distance (pri. - sec.)	<b>dCl</b>	mm	8.2	Shortest distance through air
Creepage distance (pri. - sec.)	<b>dCp</b>	mm	8.2	Shortest internal path along device body
Case material	-	-	V0 according to UL 94	
Comparative tracking index	<b>CTI</b>	V	600	
Application example	-	-	300 V CAT III PD2	Reinforced isolation, non uniform field according to EN 61010
Application example	-	-	600 V CAT III PD2	Reinforced isolation, non uniform field according to EN 50178
Application example	-	-	1000 V CAT III PD2	Simple isolation, non uniform field according to EN 50178
According to UL 508: primary potential involved in V rms AC or DC	-	V	600	For use in a pollution degree 2 environment

**Environmental and mechanical characteristics**

Parameter	Symbol	Unit	Min	Typ	Max	Comment
Ambient operating temperature	$T_A$	°C	-40		105	
Ambient storage temperature	$T_S$	°C	-55		105	
Mass	<b>m</b>	g		9		

**Electrical data CKSR 6-NP**

At  $T_A = 25^\circ\text{C}$ ,  $V_C = +5\text{ V}$ ,  $N_P = 1$  turn,  $R_L = 10\text{ k}\Omega$ , internal reference, unless otherwise noted. (see Min., Max., Typ. definition paragraph in page 12).

Parameter	Symbol	Unit	Min	Typ	Max	Comment
Primary nominal current rms	$I_{PN}$	A		6		Apply derating according to fig. 25
Primary current, measuring range	$I_{PM}$	A	-20		20	
Number of primary turns	$N_P$	-		1,2,3,4		
Supply voltage	$V_C$	V	4.75	5	5.25	
Current consumption	$I_C$	mA		$15 + \frac{I_P (\text{mA})}{N_S}$	$20 + \frac{I_P (\text{mA})}{N_S}$	$N_S = 1731$ turns
Reference voltage @ $I_P = 0\text{ A}$	$V_{REF}$	V	2.495	2.5	2.505	Internal reference
External reference voltage	$V_{REF}$	V	0		4	
Output voltage	$V_{OUT}$	V	0.375		4.625	
Output voltage @ $I_P = 0\text{ A}$	$V_{OUT}$	V		$V_{REF}$		
Electrical offset voltage	$V_{OE}$	mV	-5.3		5.3	100% tested $V_{OUT} - V_{REF}$
Electrical offset current referred to primary	$I_{OE}$	mA	-51		51	100% tested
Temperature coefficient of $V_{REF}$	$TCV_{REF}$	ppm/K		$\pm 5$	$\pm 50$	Internal reference
Temperature coefficient of $V_{OUT}$ @ $I_P = 0\text{ A}$	$TCV_{OUT}$	ppm/K		$\pm 6$	$\pm 14$	ppm/K of 2.5 V - $40^\circ\text{C} \dots 105^\circ\text{C}$ (at $\pm 6$ Sigma)
Theoretical sensitivity	$G_{th}$	mV/A		104.2		$625\text{ mV} / I_{PN}$
Sensitivity error	$\varepsilon_G$	%	-0.7		0.7	100% tested
Temperature coefficient of $G$	$TCG$	ppm/K			$\pm 40$	- $40^\circ\text{C} \dots 105^\circ\text{C}$
Linearity error	$\varepsilon_L$	% of $I_{PN}$	-0.1		0.1	
Magnetic offset current ( $10 \times I_{PN}$ ) referred to primary	$I_{OM}$	A	-0.1		0.1	
Output current noise (spectral density) rms 100 Hz .. 100 kHz referred to primary	$i_{no}$	$\mu\text{A}/\text{Hz}^{1/2}$		20		$R_L = 1\text{ k}\Omega$
Peak-peak output ripple at oscillator frequency $f = 450\text{ kHz}$ (typ.)	-	mV		40	160	$R_L = 1\text{ k}\Omega$
Reaction time @ 10 % of $I_{PN}$	$t_{ra}$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 18\text{ A}/\mu\text{s}$
Response time @ 90 % of $I_{PN}$	$t_r$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 18\text{ A}/\mu\text{s}$
Frequency bandwidth ( $\pm 1\text{ dB}$ )	<b>BW</b>	kHz	200			$R_L = 1\text{ k}\Omega$
Frequency bandwidth ( $\pm 3\text{ dB}$ )	<b>BW</b>	kHz	300			$R_L = 1\text{ k}\Omega$
Overall accuracy	$X_G$	% of $I_{PN}$			1.7	
Overall accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	$X_G$	% of $I_{PN}$			2.2 (2.4)	
Accuracy	<b>X</b>	% of $I_{PN}$			0.8	
Accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	<b>X</b>	% of $I_{PN}$			1.4 (1.6)	

**Electrical data CKSR 15-NP**

At  $T_A = 25^\circ\text{C}$ ,  $V_C = +5\text{ V}$ ,  $N_P = 1$  turn,  $R_L = 10\text{ k}\Omega$ , internal reference, unless otherwise noted. (see Min., Max., Typ. definition paragraph in page 12).

Parameter	Symbol	Unit	Min	Typ	Max	Comment
Primary nominal current rms	$I_{PN}$	A		15		Apply derating according to fig. 26
Primary current, measuring range	$I_{PM}$	A	-51		51	
Number of primary turns	$N_P$	-		1,2,3,4		
Supply voltage	$V_C$	V	4.75	5	5.25	
Current consumption	$I_C$	mA		$15 + \frac{I_P (\text{mA})}{N_S}$	$20 + \frac{I_P (\text{mA})}{N_S}$	$N_S = 1731$ turns
Reference voltage @ $I_P = 0\text{ A}$	$V_{REF}$	V	2.495	2.5	2.505	Internal reference
External reference voltage	$V_{REF}$	V	0		4	
Output voltage	$V_{OUT}$	V	0.375		4.625	
Output voltage @ $I_P = 0\text{ A}$	$V_{OUT}$	V		$V_{REF}$		
Electrical offset voltage	$V_{OE}$	mV	-2.21		2.21	100% tested $V_{OUT} - V_{REF}$
Electrical offset current referred to primary	$I_{OE}$	mA	-53		53	100% tested
Temperature coefficient of $V_{REF}$	$TCV_{REF}$	ppm/K		$\pm 5$	$\pm 50$	Internal reference
Temperature coefficient of $V_{OUT}$ @ $I_P = 0\text{ A}$	$TCV_{OUT}$	ppm/K		$\pm 2.3$	$\pm 6$	ppm/K of 2.5 V - $40^\circ\text{C} \dots 105^\circ\text{C}$ (at $\pm 6$ Sigma)
Theoretical sensitivity	$G_{th}$	mV/A		41.67		$625\text{ mV} / I_{PN}$
Sensitivity error	$\varepsilon_G$	%	-0.7		0.7	100% tested
Temperature coefficient of $G$	$TCG$	ppm/K			$\pm 40$	- $40^\circ\text{C} \dots 105^\circ\text{C}$
Linearity error	$\varepsilon_L$	% of $I_{PN}$	-0.1		0.1	
Magnetic offset current ( $10 \times I_{PN}$ ) referred to primary	$I_{OM}$	A	-0.1		0.1	
Output current noise (spectral density) rms 100 Hz .. 100 kHz referred to primary	$i_{no}$	$\mu\text{A}/\text{Hz}^{1/2}$		20		$R_L = 1\text{ k}\Omega$
Peak-peak output ripple at oscillator frequency $f = 450\text{ kHz}$ (typ.)	-	mV		15	60	$R_L = 1\text{ k}\Omega$
Reaction time @ 10 % of $I_{PN}$	$t_{ra}$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 44\text{ A}/\mu\text{s}$
Response time @ 90 % of $I_{PN}$	$t_r$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 44\text{ A}/\mu\text{s}$
Frequency bandwidth ( $\pm 1\text{ dB}$ )	<b>BW</b>	kHz	200			$R_L = 1\text{ k}\Omega$
Frequency bandwidth ( $\pm 3\text{ dB}$ )	<b>BW</b>	kHz	300			$R_L = 1\text{ k}\Omega$
Overall accuracy	$X_G$	% of $I_{PN}$			1.2	
Overall accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	$X_G$	% of $I_{PN}$			1.5 (1.7)	
Accuracy	<b>X</b>	% of $I_{PN}$			0.8	
Accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	<b>X</b>	% of $I_{PN}$			1.2 (1.3)	

**Electrical data CKSR 25-NP**

At  $T_A = 25^\circ\text{C}$ ,  $V_C = +5\text{ V}$ ,  $N_P = 1$  turn,  $R_L = 10\text{ k}\Omega$ , internal reference, unless otherwise noted. (see Min., Max., Typ. definition paragraph in page 12).

Parameter	Symbol	Unit	Min	Typ	Max	Comment
Primary nominal current rms	$I_{PN}$	A		25		Apply derating according to fig. 27
Primary current, measuring range	$I_{PM}$	A	-85		85	
Number of primary turns	$N_P$	-		1,2,3,4		
Supply voltage	$V_C$	V	4.75	5	5.25	
Current consumption	$I_C$	mA		$15 + \frac{I_P (\text{mA})}{N_S}$	$20 + \frac{I_P (\text{mA})}{N_S}$	$N_S = 1731$ turns
Reference voltage @ $I_P = 0\text{ A}$	$V_{REF}$	V	2.495	2.5	2.505	Internal reference
External reference voltage	$V_{REF}$	V	0		4	
Output voltage	$V_{OUT}$	V	0.375		4.625	
Output voltage @ $I_P = 0\text{ A}$	$V_{OUT}$	V		$V_{REF}$		
Electrical offset voltage	$V_{OE}$	mV	-1.35		1.35	100% tested $V_{OUT} - V_{REF}$
Electrical offset current referred to primary	$I_{OE}$	mA	-54		54	100% tested
Temperature coefficient of $V_{REF}$	$TCV_{REF}$	ppm/K		$\pm 5$	$\pm 50$	Internal reference
Temperature coefficient of $V_{OUT}$ @ $I_P = 0\text{ A}$	$TCV_{OUT}$	ppm/K		$\pm 1.4$	$\pm 4$	ppm/K of 2.5 V - $40^\circ\text{C} \dots 105^\circ\text{C}$ (at $\pm 6$ Sigma)
Theoretical sensitivity	<b>Gth</b>	mV/A		25		625 mV/ $I_{PN}$
Sensitivity error	$\varepsilon_G$	%	-0.7		0.7	100% tested
Temperature coefficient of <b>G</b>	<b>TCG</b>	ppm/K			$\pm 40$	- $40^\circ\text{C} \dots 105^\circ\text{C}$
Linearity error	$\varepsilon_L$	% of $I_{PN}$	-0.1		0.1	
Magnetic offset current ( $10 \times I_{PN}$ ) referred to primary	$I_{OM}$	A	-0.1		0.1	
Output current noise (spectral density) rms 100 Hz .. 100 kHz referred to primary	$i_{no}$	$\mu\text{A}/\text{Hz}^{1/2}$		20		$R_L = 1\text{ k}\Omega$
Peak-peak output ripple at oscillator frequency $f = 450\text{ kHz}$ (typ.)	-	mV		10	40	$R_L = 1\text{ k}\Omega$
Reaction time @ 10 % of $I_{PN}$	$t_{ra}$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 68\text{ A}/\mu\text{s}$
Response time @ 90 % of $I_{PN}$	$t_r$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 68\text{ A}/\mu\text{s}$
Frequency bandwidth ( $\pm 1\text{ dB}$ )	<b>BW</b>	kHz	200			$R_L = 1\text{ k}\Omega$
Frequency bandwidth ( $\pm 3\text{ dB}$ )	<b>BW</b>	kHz	300			$R_L = 1\text{ k}\Omega$
Overall accuracy	$X_G$	% of $I_{PN}$			1	
Overall accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	$X_G$	% of $I_{PN}$			1.35 (1.45)	
Accuracy	<b>X</b>	% of $I_{PN}$			0.8	
Accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	<b>X</b>	% of $I_{PN}$			1.15 (1.25)	

**Electrical data CKSR 50-NP**

At  $T_A = 25^\circ\text{C}$ ,  $V_C = +5\text{ V}$ ,  $N_P = 1$  turn,  $R_L = 10\text{ k}\Omega$ , internal reference, unless otherwise noted. (see Min., Max., Typ. definition paragraph in page 12).

Parameter	Symbol	Unit	Min	Typ	Max	Comment
Primary nominal current rms	$I_{PN}$	A		50		Apply derating according to fig. 28
Primary current, measuring range	$I_{PM}$	A	-150		150	
Number of primary turns	$N_P$	-		1,2,3,4		
Supply voltage	$V_C$	V	4.75	5	5.25	
Current consumption	$I_C$	mA		$15 + \frac{I_P (\text{mA})}{N_S}$	$20 + \frac{I_P (\text{mA})}{N_S}$	$N_S = 966$ turns
Reference voltage @ $I_P = 0\text{ A}$	$V_{REF}$	V	2.495	2.5	2.505	Internal reference
External reference voltage	$V_{REF}$	V	0		4	
Output voltage	$V_{OUT}$	V	0.375		4.625	
Output voltage @ $I_P = 0\text{ A}$	$V_{OUT}$	V		$V_{REF}$		
Electrical offset voltage	$V_{OE}$	mV	-0.725		0.725	100% tested $V_{OUT} - V_{REF}$
Electrical offset current referred to primary	$I_{OE}$	mA	-58		58	100% tested
Temperature coefficient of $V_{REF}$	$TCV_{REF}$	ppm/K		$\pm 5$	$\pm 50$	Internal reference
Temperature coefficient of $V_{OUT}$ @ $I_P = 0\text{ A}$	$TCV_{OUT}$	ppm/K		$\pm 0.7$	$\pm 3$	ppm/K of 2.5 V - $40^\circ\text{C} \dots 105^\circ\text{C}$ (at $\pm 6$ Sigma)
Theoretical sensitivity	$G_{th}$	mV/A		12.5		$625\text{ mV} / I_{PN}$
Sensitivity error	$\varepsilon_G$	%	-0.7		0.7	100% tested
Temperature coefficient of $G$	$TCG$	ppm/K			$\pm 40$	- $40^\circ\text{C} \dots 105^\circ\text{C}$
Linearity error	$\varepsilon_L$	% of $I_{PN}$	-0.1		0.1	
Magnetic offset current ( $10 \times I_{PN}$ ) referred to primary	$I_{OM}$	A	-0.1		0.1	
Output current noise (spectral density) rms 100 Hz .. 100 kHz referred to primary	$i_{no}$	$\mu\text{A}/\text{Hz}^{1/2}$		20		$R_L = 1\text{ k}\Omega$
Peak-peak output ripple at oscillator frequency $f = 450\text{ kHz}$ (typ.)	-	mV		5	20	$R_L = 1\text{ k}\Omega$
Reaction time @ 10 % of $I_{PN}$	$t_{ra}$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 100\text{ A}/\mu\text{s}$
Response time @ 90 % of $I_{PN}$	$t_r$	$\mu\text{s}$			0.3	$R_L = 1\text{ k}\Omega$ , $di/dt = 100\text{ A}/\mu\text{s}$
Frequency bandwidth ( $\pm 1\text{ dB}$ )	<b>BW</b>	kHz	200			$R_L = 1\text{ k}\Omega$
Frequency bandwidth ( $\pm 3\text{ dB}$ )	<b>BW</b>	kHz	300			$R_L = 1\text{ k}\Omega$
Overall accuracy	$X_G$	% of $I_{PN}$			0.9	
Overall accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	$X_G$	% of $I_{PN}$			1.2 (1.3)	
Accuracy	<b>X</b>	% of $I_{PN}$			0.8	
Accuracy @ $T_A = 85^\circ\text{C}$ ( $105^\circ\text{C}$ )	<b>X</b>	% of $I_{PN}$			1.1 (1.3)	

# Typical performance characteristics CKSR 6-NP

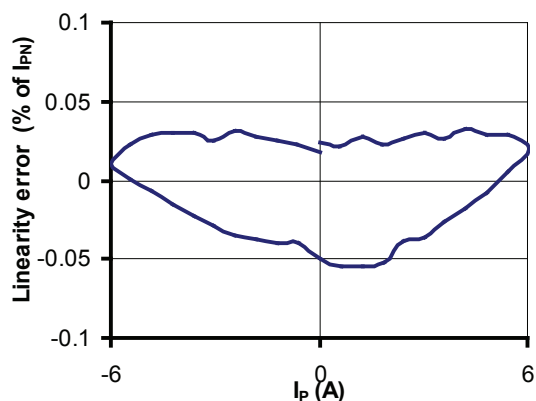


Figure 1: Linearity error

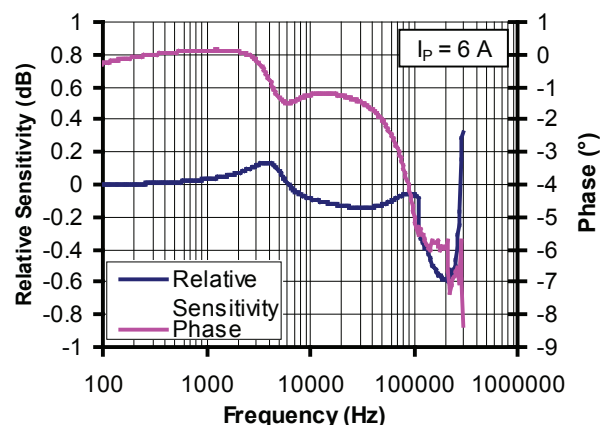


Figure 2: Frequency response

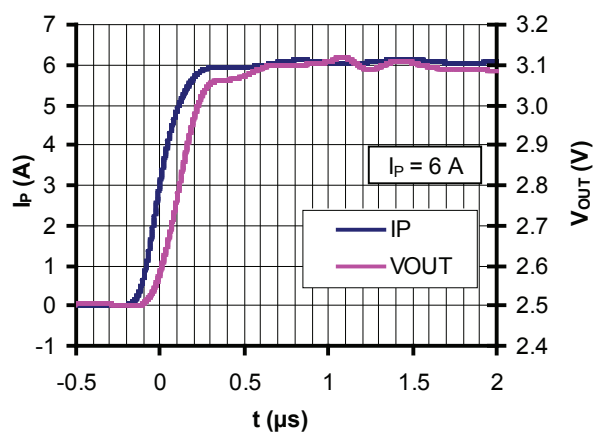


Figure 3: Step response

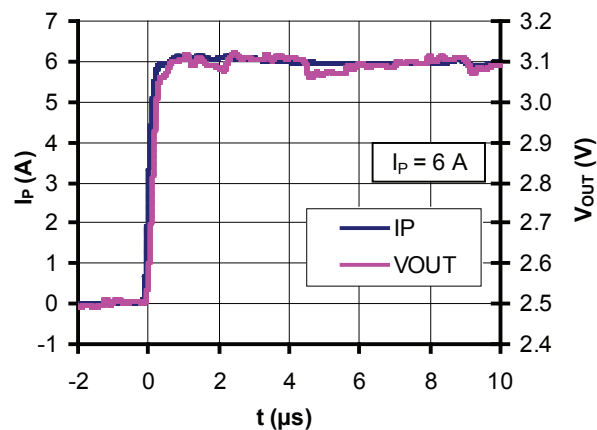


Figure 4: Step response

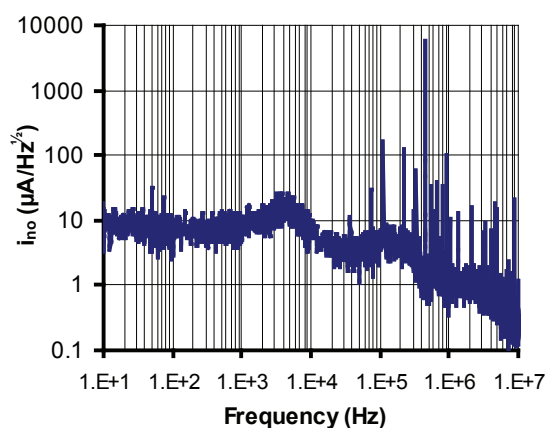


Figure 5: Input referred noise

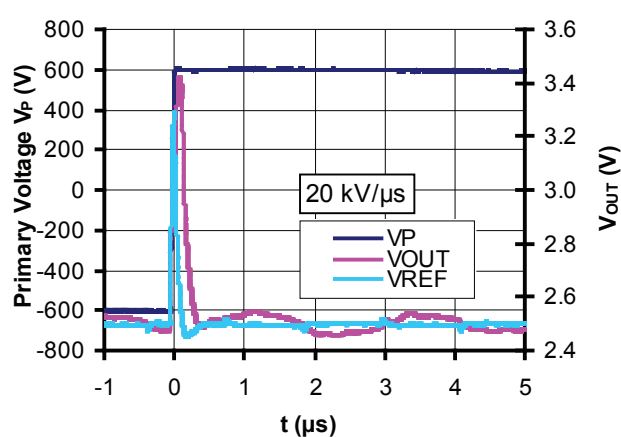


Figure 6: dv/dt

### Typical performance characteristics CKSR 15-NP

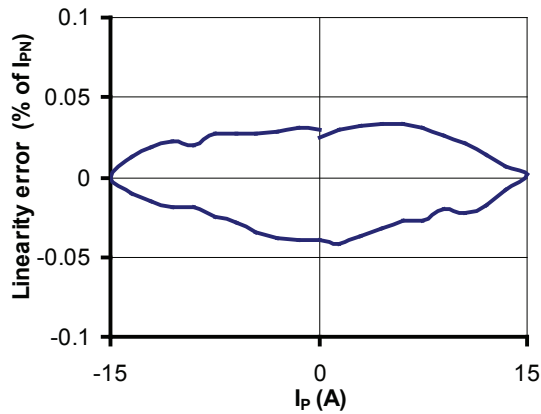


Figure 7: Linearity error

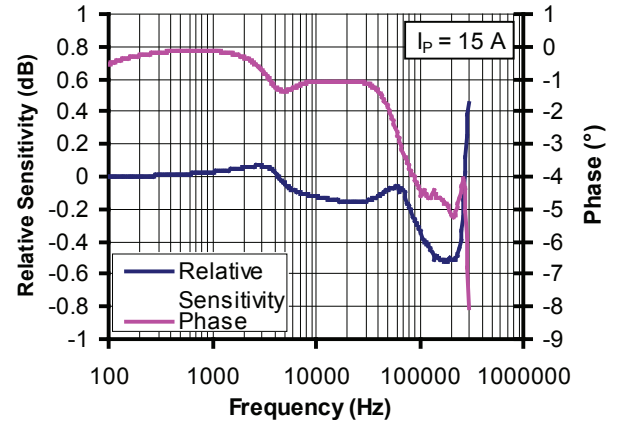


Figure 8: Frequency response

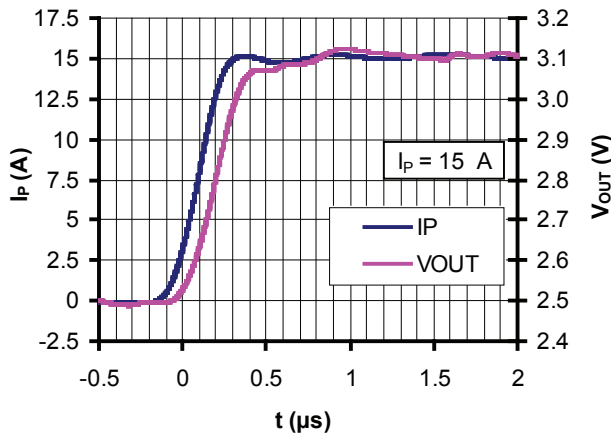


Figure 9: Step response

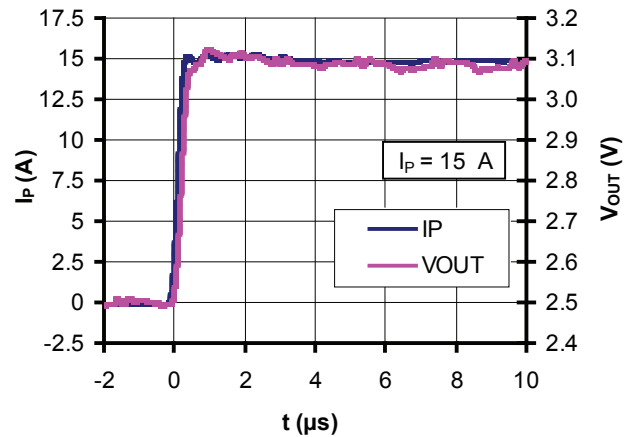


Figure 10: Step response

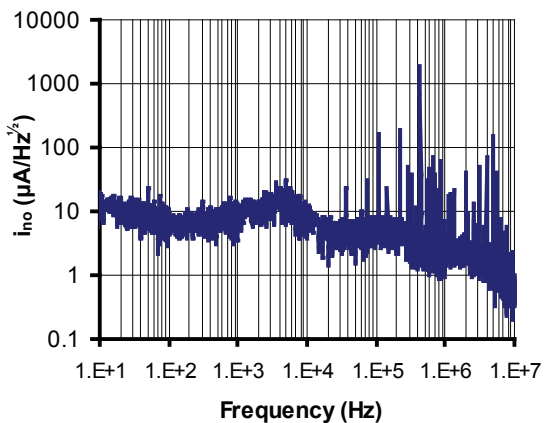


Figure 11: Input referred noise

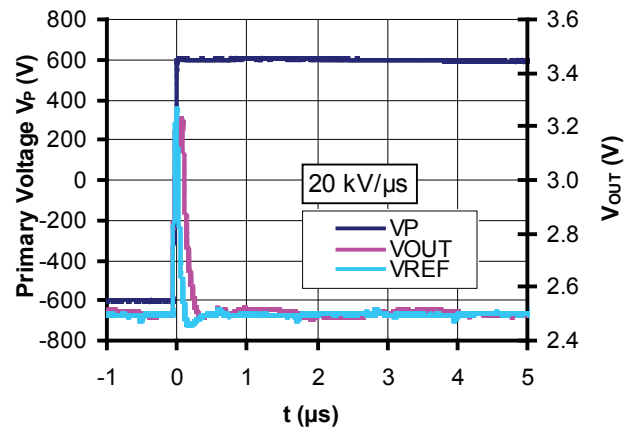


Figure 12: dv/dt



### Typical performance characteristics CKSR 25-NP

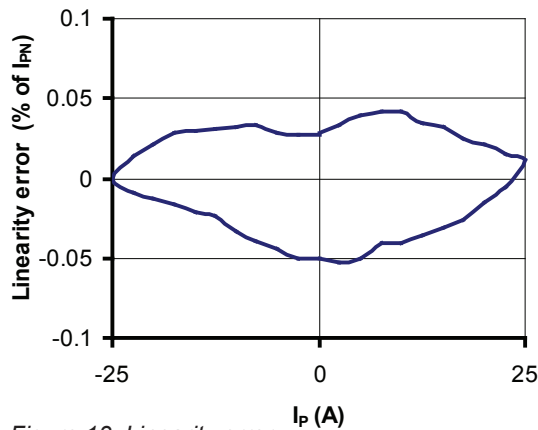


Figure 13: Linearity error

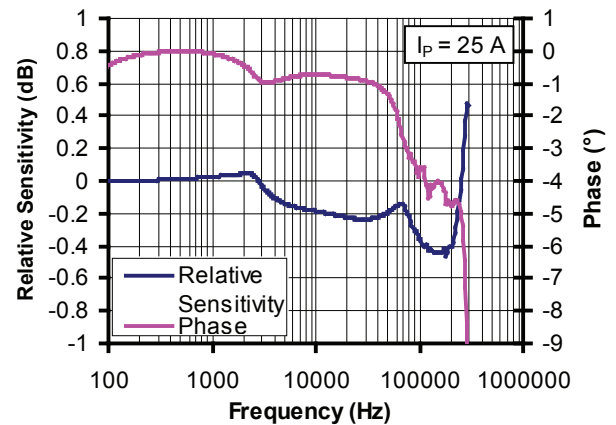


Figure 14: Frequency response

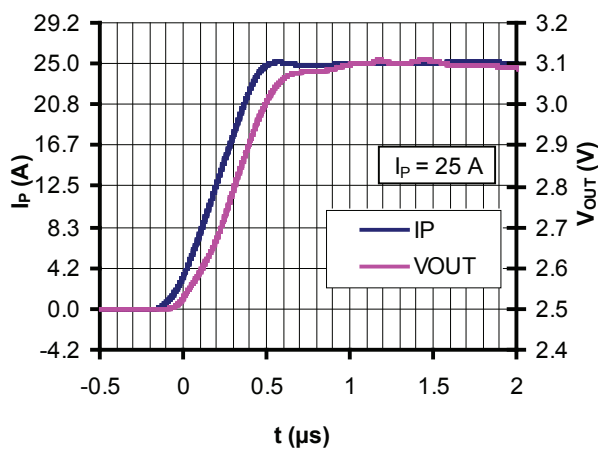


Figure 15: Step response

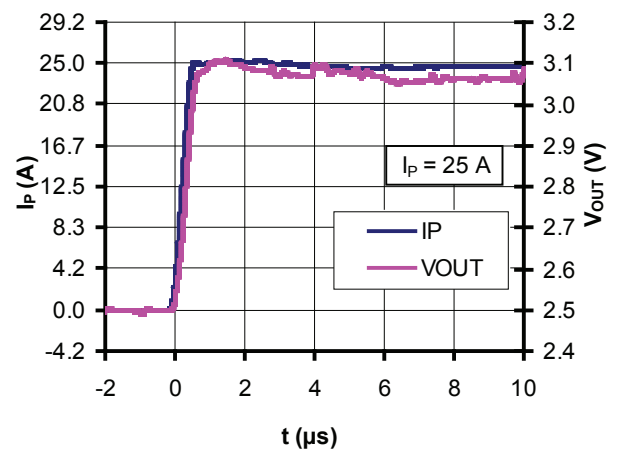


Figure 16: Step response

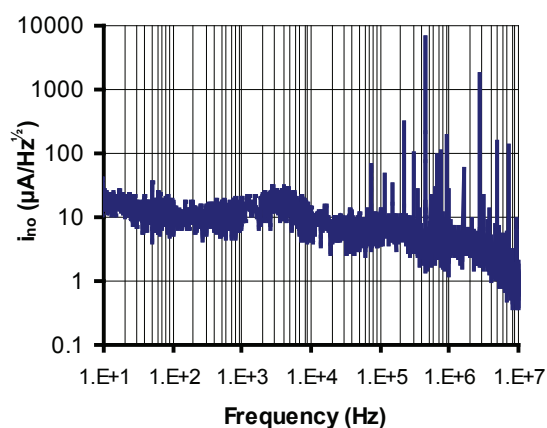


Figure 17: Input referred noise

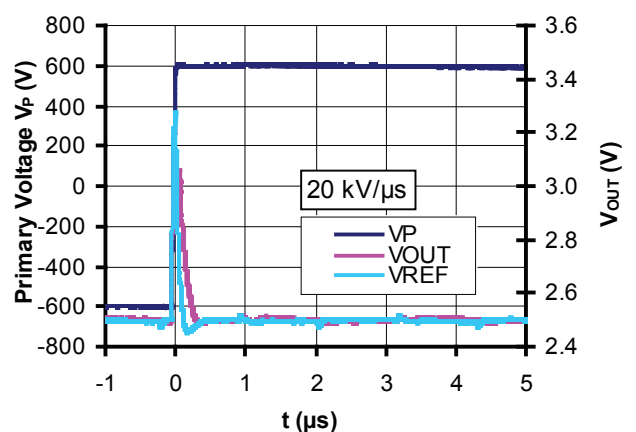


Figure 18: dv/dt

### Typical performance characteristics CKSR 50-NP

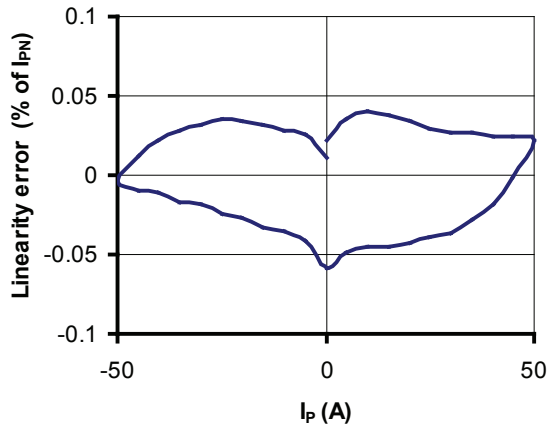


Figure 19: Linearity error

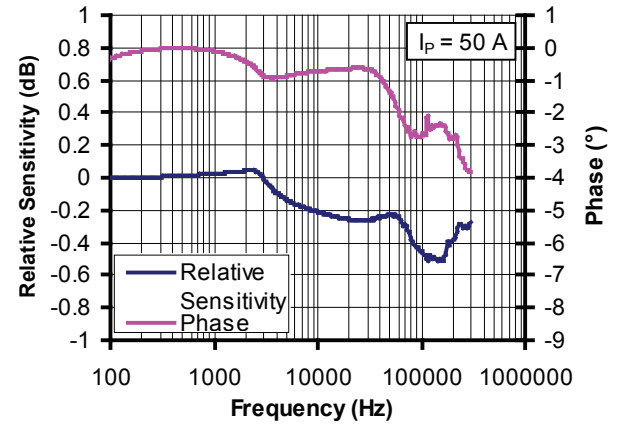


Figure 20: Frequency response

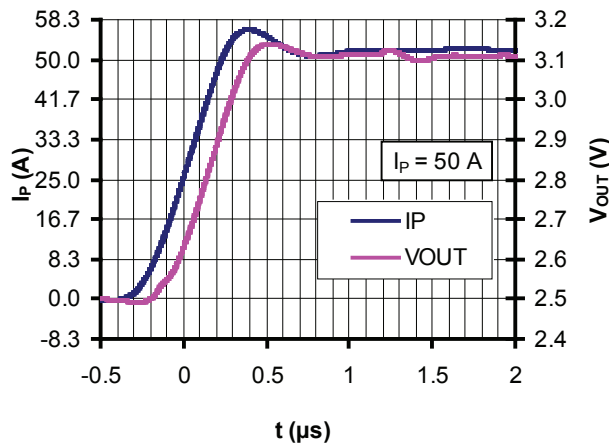


Figure 21: Step response

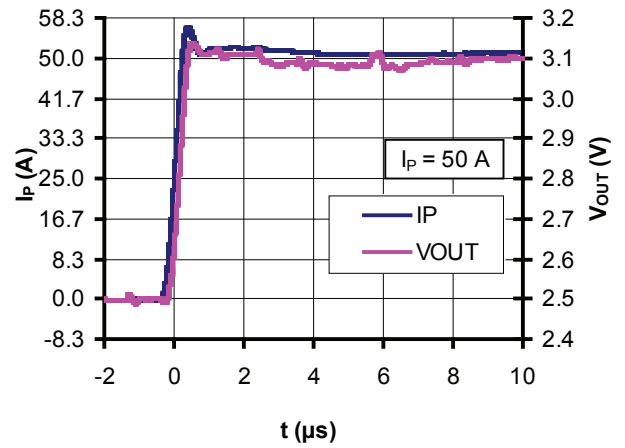


Figure 22: Step response

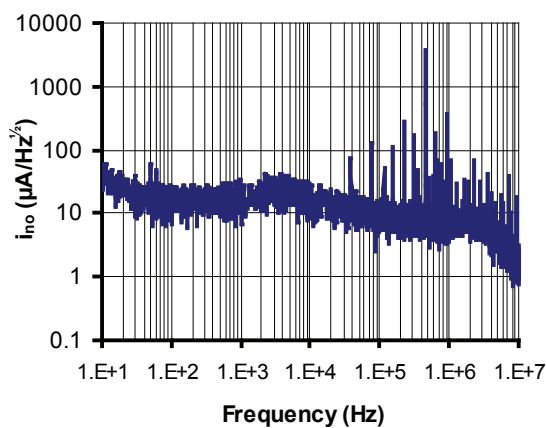


Figure 23: Input referred noise

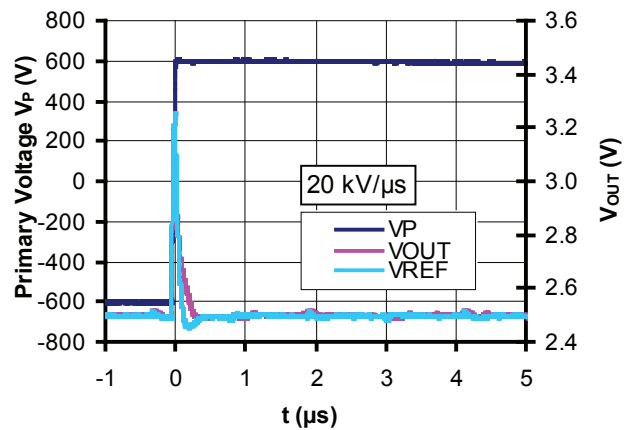


Figure 24: dv/dt

## Maximum continuous DC primary current

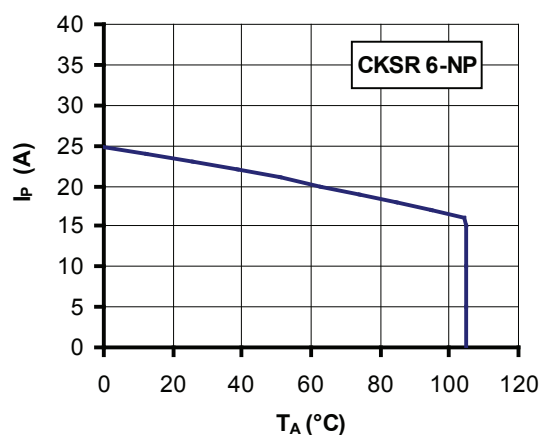


Figure 25:  $I_P$  vs  $T_A$  for CKSR 6-NP

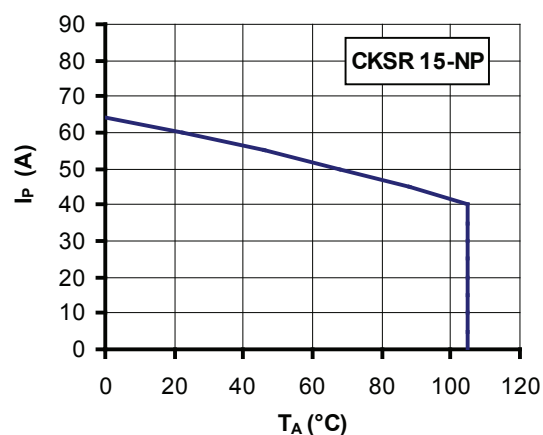


Figure 26:  $I_P$  vs  $T_A$  for CKSR 15-NP

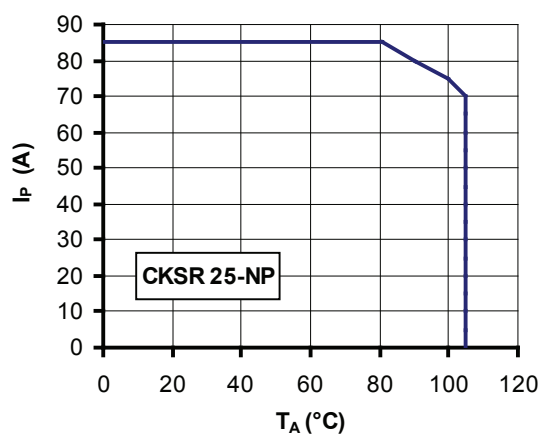


Figure 27:  $I_P$  vs  $T_A$  for CKSR 25-NP

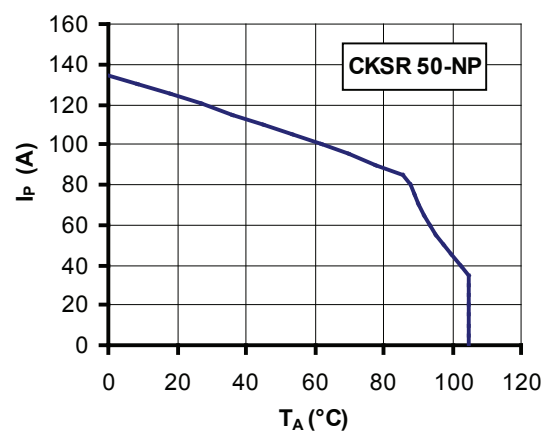


Figure 28:  $I_P$  vs  $T_A$  for CKSR 50-NP

The maximum continuous DC primary current plot shows the boundary of the area for which all the following conditions are true:

- $I_P < I_{PM}$
- Junction temperature  $T_J < 125\text{ °C}$
- Primary conductor temperature  $< 110\text{ °C}$
- Resistor power dissipation  $< 0.5 \times \text{rated power}$

## Frequency derating

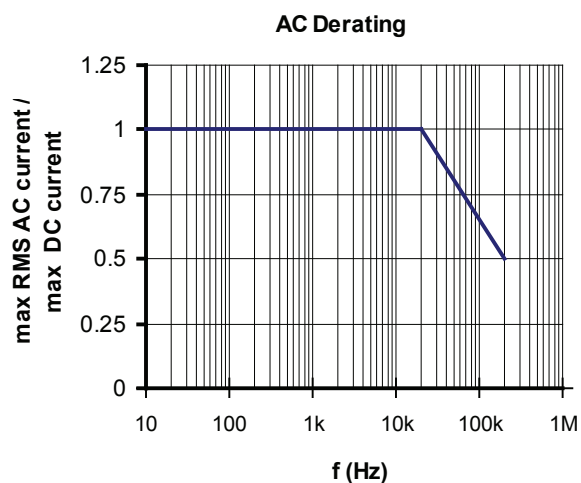


Figure 29: Maximum RMS AC primary current / maximum DC primary current vs frequency

## Performance parameters definition

### Ampere-turns and amperes

The transducer is sensitive to the primary current linkage  $\Theta_p$  (also called ampere-turns).

$$\Theta_p = N_p I_p (\text{At})$$

Where  $N_p$  is the number of primary turn (1, 2, 3 or 4 depending on the connection of the primary jumpers)

Caution: As most applications will use the transducer with only one single primary turn ( $N_p = 1$ ), much of this datasheet is written in terms of primary current instead of current linkages. However, the ampere-turns (A-t) unit is used to emphasize that current linkages are intended and applicable.

### Transducer simplified model

The static model of the transducer at temperature  $T_A$  is:

$$V_{OUT} = G \Theta_p + \text{error}$$

In which error =

$$V_{OE} + V_{OT}(T_A) + \varepsilon_G \cdot \Theta_p \cdot G + \varepsilon_L(\Theta_{Pmax}) \cdot \Theta_{Pmax} \cdot G + TCG \cdot (T_A - 25) \cdot \Theta_p \cdot G$$

With:

- $\Theta_p = N_p I_p$  :the input ampere-turns (At)  
Please read above warning.
- $\Theta_{Pmax}$  :the maxi input ampere-turns that have been applied to the transducer (At)
- $V_{OUT}$  :the secondary voltage (V)
- $T_A$  :the ambient temperature ( $^{\circ}\text{C}$ )
- $V_{OE}$  :the electrical offset voltage (V)
- $V_{OT}(T_A)$  :the temperature variation of  $V_O$  at temperature  $T_A$  (V)
- $G$  :the sensitivity of the transducer (V/At)
- $\varepsilon_G$  :the sensitivity error
- $\varepsilon_L(\Theta_{Pmax})$  :the linearity error for  $\Theta_{Pmax}$

This model is valid for primary ampere-turns  $\Theta_p$  between  $-\Theta_{Pmax}$  and  $+\Theta_{Pmax}$  only.

### Min., Max., Typ. definition

Some parameters have a statistically normal distribution. The typical value published in the LEM datasheet is the mean or average value of the distribution. The typical value listed is the 1 sigma value. This means that in 68 % of the devices tested, the parameters is found to be  $\pm$  the typical value or better. LEM currently uses  $\pm 3$  sigma (99.73 %) to define minimum and maximum values. Usually, typical values are set when the part is characterized and never changes.

### Sensitivity and linearity

To measure sensitivity and linearity, the primary current (DC) is cycled from 0 to  $I_p$ , then to  $-I_p$  and back to 0 (equally spaced  $I_p/10$  steps).

The sensitivity  $G$  is defined as the slope of the linear regression line for a cycle between  $\pm I_{PN}$ .

The linearity error  $\varepsilon_L$  is the maximum positive or negative difference between the measured points and the linear regression line, expressed in % of  $I_{PN}$ .

### Magnetic offset

The magnetic offset current  $I_{OM}$  is the consequence of a current on the primary side ("memory effect" of the transducer's ferro-magnetic parts). It is included in the linearity figure but can be measured individually.

It is measured using the following primary current cycle.  $I_{OM}$  depends on the current value  $I_{P1}$ .

$$I_{OM} = \frac{V_{OUT}(t_1) - V_{OUT}(t_2)}{2} \cdot \frac{1}{Gth}$$

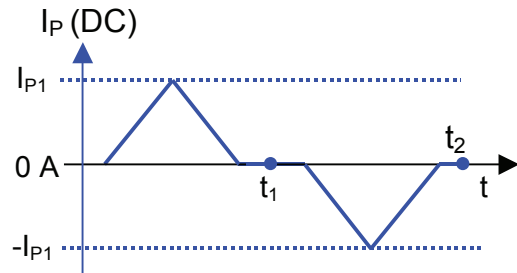


Figure 30: Current cycle used to measure magnetic and electrical offset (transducer supplied)

## Performance parameters definition (continued)

### Electrical offset

The electrical offset voltage  $V_{OE}$  can either be measured when the ferro-magnetic parts of the transducer are:

- completely demagnetized, which is difficult to realize,
- or in a known magnetization state, like in the current cycle shown in figure 30.

Using the current cycle shown in figure 30, the electrical offset is:

$$V_{OE} = \frac{V_{OUT}(t_1) + V_{OUT}(t_2)}{2}$$

The temperature variation  $V_{OT}$  of the electrical offset voltage  $V_{OE}$  is the variation of the electrical offset from 25°C to the considered temperature:

$$V_{OT}(T) = V_{OE}(T) - V_{OE}(25^\circ\text{C})$$

Note: the transducer has to be demagnetized prior to the application of the current cycle (for example with a demagnetization tunnel).

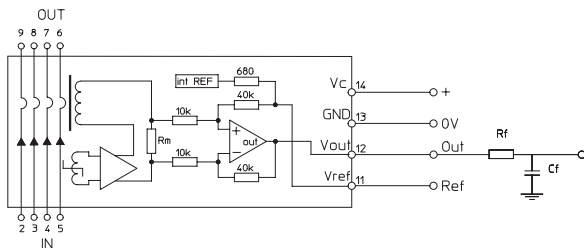


Figure 31: Test connection

### Overall accuracy

The overall accuracy at 25°C  $X_G$  is the error in the  $-I_{PN} \dots +I_{PN}$  range, relative to the rated value  $I_{PN}$ .

It includes:

- the electrical offset  $V_{OE}$
- the sensitivity error  $\varepsilon_G$
- the linearity error  $\varepsilon_L$  (to  $I_{PN}$ )

The magnetic offset is part of the overall accuracy. It is taken into account in the linearity error figure provided the transducer has not been magnetized by a current higher than  $I_{PN}$ .

### Response and reaction times

The response time  $t_r$  and the reaction time  $t_{ra}$  are shown in figure 32.

Both depend on the primary current  $di/dt$ . They are measured at nominal ampere-turns.

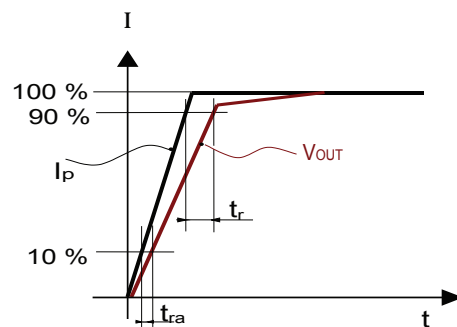


Figure 32: response time  $t_r$  and reaction time  $t_{ra}$

## Application information

## Filtering and decoupling

### Supply voltage $V_C$

The fluxgate oscillator draws current pulses of up to 30 mA at a rate of ca. 900 kHz. Significant 900 kHz voltage ripple on  $V_C$  can indicate a power supply with high impedance. At these frequencies the power supply rejection ratio is low, and the ripple may appear on the transducer output  $V_{OUT}$  and reference  $V_{REF}$ . The transducer has internal decoupling capacitors, but in the case of a power supply with high impedance, it is advised to provide local decoupling (100 nF or more, located close to the transducer)



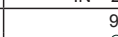
## Output $V_{OUT}$

The output  $\mathbf{V}_{OUT}$  has a very low output impedance of typically 2 Ohms; it can drive 100 pF directly. Adding series  $R_f = 100$  Ohms allows much larger capacitive loads. Empirical evaluation may be necessary to obtain optimum results. The minimum load resistance on  $\mathbf{V}_{OUT}$  is 1 kOhm.

## Total Primary Resistance

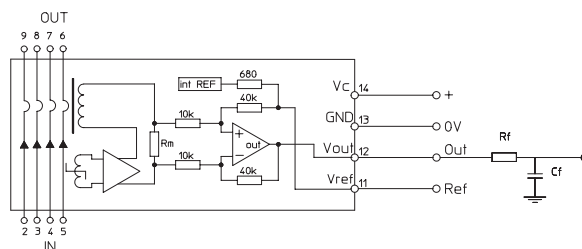
The primary resistance is  $0.72 \text{ m}\Omega$  per conductor

In the following table, examples of primary resistance according to the number of primary turns

Number of primary turns	Primary resistance $R_p$ [mΩ]	Recommended connections
1	0.18	
2	0.72	
4	2.88	

## Reference $V_{BEE}$

Ripple present on the reference output can be filtered with a low value of capacitance because of the internal 680 Ohm series resistance. The maximum filter capacitance value is 1  $\mu\text{F}$



## Application information (continued)

### External reference voltage

If the Ref pin of the transducer is not used it could be either left unconnected or filtered according to the previous paragraph "Reference  $V_{REF}$ ".

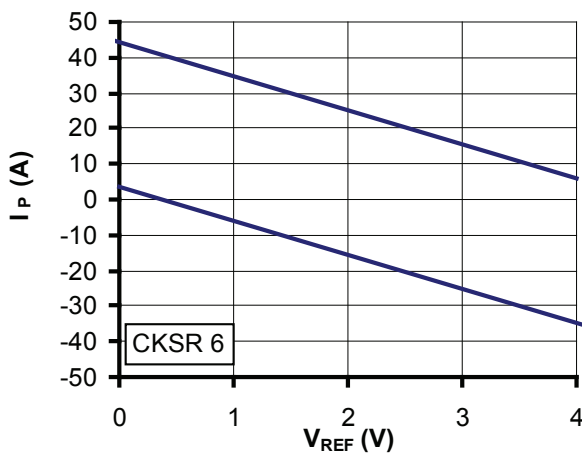
The Ref pin has two modes Ref IN and Ref OUT:

- In the Ref OUT mode the 2.5 V internal precision reference is used by the transducer as the reference point for bipolar measurements; this internal reference is connected to the Ref pin of the transducer through a 680 Ohms resistor. it tolerates sink or source currents up to  $\pm 5$  mA, but the 680 Ohms resistor prevents this current to exceed these limits.
- In the Ref IN mode, an external reference voltage is connected to the Ref pin; this voltage is specified in the range 0 to 4 V and is directly used by the transducer as the reference point for measurements.  
The external reference voltage  $V_{REF}$  must be able:

- either to source a typical current of  $\frac{V_{ref} - 2.5}{680}$ , the maximum value will be 2.2 mA typ. when  $V_{REF} = 4$  V.

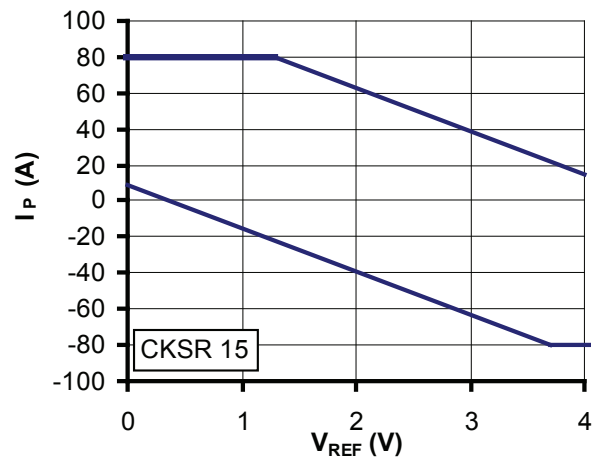
- or to sink a typical current of  $\frac{2.5 - V_{ref}}{680}$ , the maximum value will be 3.68 mA typ. when  $V_{REF} = 0$  V.

The following graphs show how the measuring range of each transducer version depends on the external reference voltage value  $V_{REF}$ .



Upper limit:  $I_p = -9.6 * V_{REF} + 44.4$  ( $V_{REF} = 0 \dots 4$  V)

Lower limit:  $I_p = -9.6 * V_{REF} + 3.6$  ( $V_{REF} = 0 \dots 4$  V)

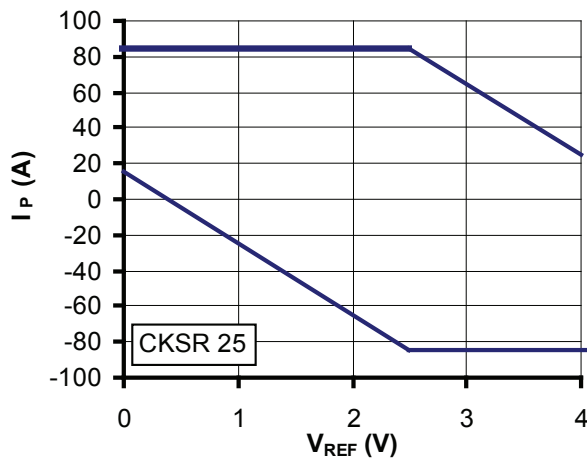


Upper limit:  $I_p = -24 * V_{REF} + 111$  ( $V_{REF} = 1.29 \dots 4$  V)

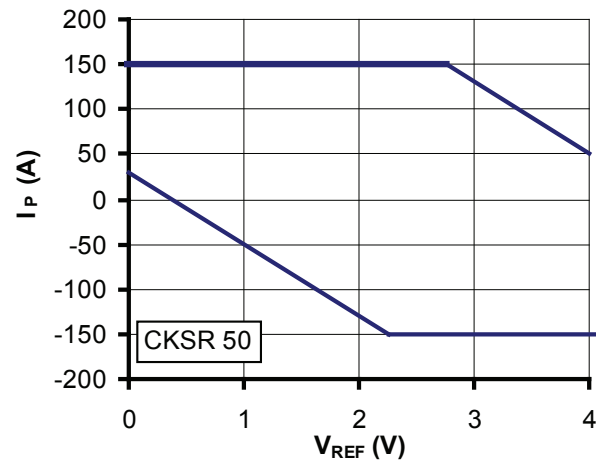
Upper limit:  $I_p = 80$  ( $V_{REF} = 0 \dots 1.29$  V)

Lower limit:  $I_p = -24 * V_{REF} + 9$  ( $V_{REF} = 0 \dots 3.7$  V)

Lower limit:  $I_p = -80$  ( $V_{REF} = 3.7 \dots 4$  V)

**External reference voltage (continued)**


Upper limit:  $I_P = -40 \cdot V_{REF} + 185$  ( $V_{REF} = 2.5 \dots 4$  V)  
 Upper limit:  $I_P = 85$  ( $V_{REF} = 0 \dots 2.5$  V)  
 Lower limit:  $I_P = -40 \cdot V_{REF} + 15$  ( $V_{REF} = 0 \dots 2.5$  V)  
 Lower limit:  $I_P = -85$  ( $V_{REF} = 2.5 \dots 4$  V)



Upper limit:  $I_P = -80 \cdot V_{REF} + 370$  ( $V_{REF} = 2.75 \dots 4$  V)  
 Upper limit:  $I_P = 150$  ( $V_{REF} = 0 \dots 2.75$  V)  
 Lower limit:  $I_P = -80 \cdot V_{REF} + 30$  ( $V_{REF} = 0 \dots 2.25$  V)  
 Lower limit:  $I_P = -150$  ( $V_{REF} = 2.25 \dots 4$  V)

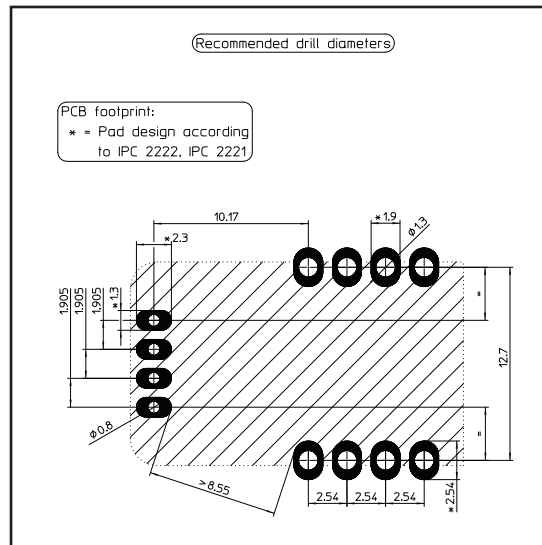
*Example with  $V_{REF} = 1.65$  V:*

- The 6 A version has a measuring range from - 12.24 A to + 28.5 A
- The 15 A version has a measuring range from - 30.6 A to + 71.4 A
- The 25 A version has a measuring range from - 51 A to + 85 A
- The 50 A version has a measuring range from - 102 A to + 150 A

*Example with  $V_{REF} = 0$  V:*

- The 6 A version has a measuring range from + 3.6 A to + 44.4 A
- The 15 A version has a measuring range from + 9 A to + 80 A
- The 25 A version has a measuring range from + 15 A to + 85 A
- The 50 A version has a measuring range from + 30 A to + 150 A



**CKSR Series, PCB footprint**

**Assembly on PCB**

- Recommended PCB hole diameter 1.3 mm for primary pin  
0.8 mm for secondary pin
- Maximum PCB thickness 2.4 mm
- Wave soldering profile maximum 260°C for 10 s  
No clean process only.

**Safety**

This transducer must be used in limited-energy secondary circuits according to IEC 61010-1.



This transducer must be used in electric/electronic equipment with respect to applicable standards and safety requirements in accordance with the manufacturer's operating instructions.



Caution, risk of electrical shock

When operating the transducer, certain parts of the module can carry hazardous voltage (eg. primary busbar, power supply).

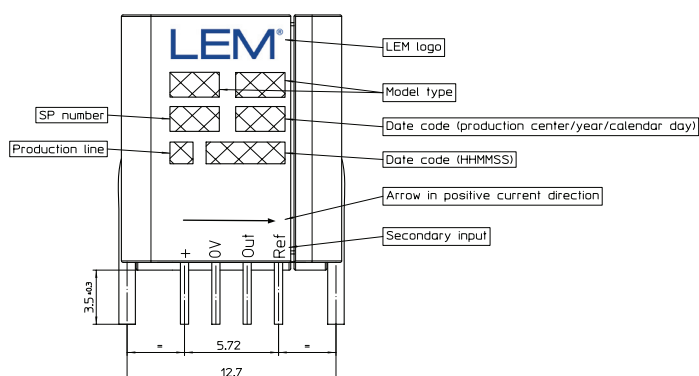
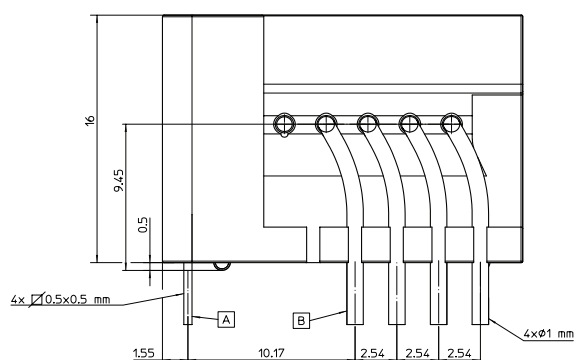
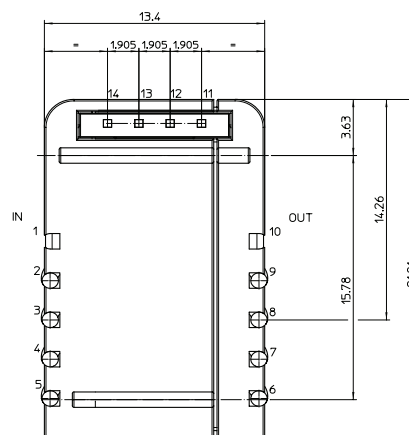
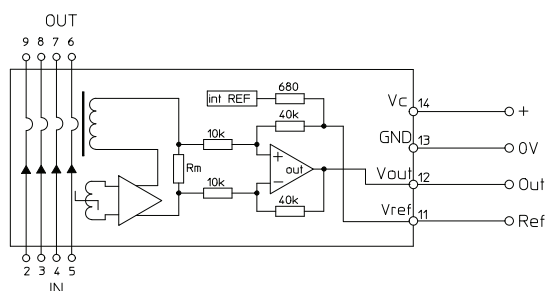
Ignoring this warning can lead to injury and/or cause serious damage.

This transducer is a build-in device, whose conducting parts must be inaccessible after installation.

A protective housing or additional shield could be used.

Main supply must be able to be disconnected.

## Connection



	Clearance	Creepage
A-B	8.2 mm*	8.2 mm

\* Clearance between pads on the pcb: 8.55 mm

Additional marking on the TOP of the transducer,  
model type, current range and SP number