

## LM5102 High Voltage Half-Bridge Gate Driver with Programmable Delay

Check for Samples: [LM5102](#)

### FEATURES

- Drives both a High Side and Low Side N-Channel MOSFET
- Independently Programmable High and Low Side Rising Edge Delay
- Bootstrap Supply Voltage Range up to 118V DC
- Fast Turn-Off Propagation Delay (25 ns Typical)
- Drives 1000 pF Loads with 15 ns Rise and Fall Times
- Supply Rail Under-Voltage Lockout
- Low Power Consumption
- Timer Can be Terminated Midway through Sequence

### TYPICAL APPLICATIONS

- Current Fed Push-Pull Power Converters
- Half and Full Bridge Power Converters
- Synchronous Buck Converters
- Two Switch Forward Power Converters
- Forward with Active Clamp Converters

### DESCRIPTION

The LM5102 High Voltage Gate Driver is designed to drive both the high side and the low side N-Channel MOSFETs in a synchronous buck or a half bridge configuration. The floating high-side driver is capable of working with supply voltages up to 100V. The outputs are independently controlled. The rising edge of each output can be independently delayed with a programming resistor. An integrated high voltage diode is provided to charge the high side gate drive bootstrap capacitor. A robust level shifter operates at high speed while consuming low power and providing clean level transitions from control logic to the high side gate driver. Under-voltage lockout is provided on both the low side and the high side power rails. This device is available in the standard VSSOP-10 pin and the WSON-10 pin packages.

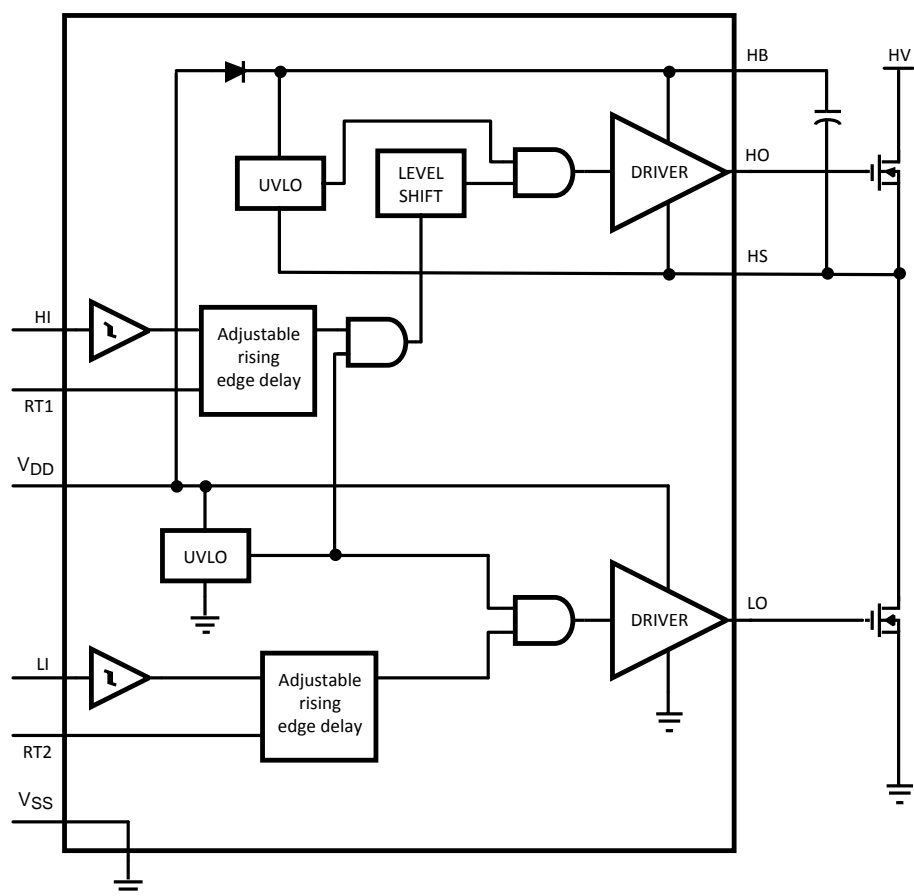
### Packages

- VSSOP-10
- WSON-10 (4 mm x 4 mm)

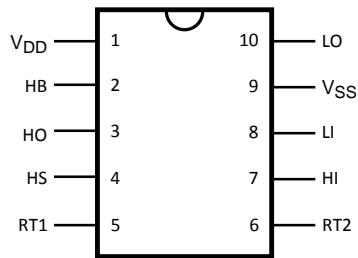


Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

## Simplified Block Diagram



## Connection Diagram



**Figure 1. 10-Lead VSSOP, WSON**

### PIN FUNCTIONS

PIN		NAME	DESCRIPTION	APPLICATION INFORMATION
VSSOP	WSON <sup>(1)</sup>			
1	1	V <sub>DD</sub>	Positive gate drive supply	Locally decouple to VSS using low ESR/ESL capacitor, located as close to IC as possible.
2	2	HB	High-side gate driver bootstrap rail	Connect the positive terminal of bootstrap capacitor to the HB pin and connect negative terminal of bootstrap capacitor to HS. The Bootstrap capacitor should be placed as close to IC as possible.
3	3	HO	High-side gate driver output	Connect to gate of high side MOSFET with short low inductance path.
4	4	HS	High-side MOSFET source connection	Connect bootstrap capacitor negative terminal and source of high side MOSFET.
5	5	RT1	High-side output edge delay programming	Resistor from RT1 to ground programs the leading edge delay of the high side gate driver. The resistor should be placed close to the IC to minimize noise coupling from adjacent traces.
6	6	RT2	Low-side output edge delay programming	Resistor from RT2 to ground programs the leading edge delay of the low side gate driver. The resistor should be placed close to the IC to minimize noise coupling from adjacent traces.
7	7	HI	High-side driver control input	TTL compatible thresholds. Unused inputs should be tied to ground and not left open
8	8	LI	Low-side driver control input	
9	9	V <sub>SS</sub>	Ground return	All signals are referenced to this ground.
10	10	LO	Low-side gate driver output	Connect to the gate of the low side MOSFET with a short low inductance path.

- (1) For the WSON package, it is recommended that the exposed pad on the bottom of the LM5100 / LM5101 be soldered to ground plane on the PC board, and the ground plane should extend out from beneath the IC to help dissipate the heat..



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**Absolute Maximum Ratings<sup>(1)(2)</sup>**

$V_{DD}$ to $V_{SS}$	–0.3V to +18V
$V_{HB}$ to $V_{HS}$	–0.3V to +18V
LI or HI Inputs to $V_{SS}$	–0.3V to $V_{DD} + 0.3V$
LO Output	–0.3V to $V_{DD} + 0.3V$
HO Output	$V_{HS} - 0.3V$ to $V_{HB} + 0.3V$
$V_{HS}$ to $V_{SS}$	–1V to +100V
$V_{HB}$ to $V_{SS}$	118V
RT1 & RT2 to $V_{SS}$	–0.3V to 5V
Junction Temperature	+150°C
Storage Temperature Range	–55°C to +150°C
ESD Rating HBM <sup>(3)</sup>	2 kV

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the component may occur. Operating Ratings are conditions under which operation of the device is specified. Operating Ratings do not imply performance limits. For performance limits and associated test conditions, see [Electrical Characteristics](#).
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) The human body model is a 100 pF capacitor discharged through a 1.5kΩ resistor into each pin. 2 kV for all pins except Pin 2, Pin 3 and Pin 4 which are rated at 500V.

**Recommended Operating Conditions**

$V_{DD}$	+9V to +14V
HS	–1V to 100V
HB	$V_{HS} + 8V$ to $V_{HS} + 14V$
HS Slew Rate	< 50V/ns
Junction Temperature	–40°C to +125°C

**Electrical Characteristics**

Specifications in standard typeface are for  $T_J = +25^\circ\text{C}$ , and those in **boldface type** apply over the full **operating junction temperature range**. Unless otherwise specified,  $V_{DD} = V_{HB} = 12V$ ,  $V_{SS} = V_{HS} = 0V$ ,  $RT1 = RT2 = 100k\Omega$ . No Load on LO or HO.

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Units
<b>SUPPLY CURRENTS</b>						
$I_{DD}$	$V_{DD}$ Quiescent Current	LI = HI = 0V		0.4	<b>0.6</b>	mA
$I_{DDO}$	$V_{DD}$ Operating Current	$f = 500\text{ kHz}$		1.5	<b>3</b>	mA
$I_{HB}$	Total HB Quiescent Current	LI = HI = 0V		0.06	<b>0.2</b>	mA
$I_{HBO}$	Total HB Operating Current	$f = 500\text{ kHz}$		1.3	<b>3</b>	mA
$I_{HBS}$	HB to $V_{SS}$ Current, Quiescent	$V_{HS} = V_{HB} = 100V$		0.05	<b>10</b>	μA
$I_{HBSO}$	HB to $V_{SS}$ Current, Operating	$f = 500\text{ kHz}$		0.08		mA
<b>INPUT PINS</b>						
$V_{IL}$	Low Level Input Voltage Threshold		<b>0.8</b>	1.8		V
$V_{IH}$	High Level Input Voltage Threshold			1.8	<b>2.2</b>	V
$R_I$	Input Pulldown Resistance		<b>100</b>	200	<b>500</b>	kΩ
<b>TIME DELAY CONTROLS</b>						
$V_{RT}$	Nominal Voltage at RT1, RT2		<b>2.7</b>	3	<b>3.3</b>	V
$I_{RT}$	RT Pin Current Limit	$RT1 = RT2 = 0V$	<b>0.75</b>	1.5	<b>2.25</b>	mA
$V_{th}$	Timer Termination Threshold			1.8		V
$T_{DL1}, T_{DH1}$	Rising edge turn-on delay, $RT = 10\text{ k}\Omega$		<b>75</b>	105	<b>150</b>	ns
$T_{DL2}, T_{DH2}$	Rising edge turn-on delay, $RT = 100\text{ k}\Omega$		<b>530</b>	630	<b>750</b>	ns

- (1) Min and Max limits are 100% production tested at 25°C. Limits over the operating temperature range are specified through correlation using Statistical Quality Control (SQC) methods. Limits are used to calculate TI's Average Outgoing Quality Level (AOQL).

## Electrical Characteristics (continued)

Specifications in standard typeface are for  $T_J = +25^\circ\text{C}$ , and those in **boldface type** apply over the full **operating junction temperature range**. Unless otherwise specified,  $V_{DD} = V_{HB} = 12\text{V}$ ,  $V_{SS} = V_{HS} = 0\text{V}$ ,  $RT1 = RT2 = 100\text{k}\Omega$ . No Load on LO or HO.

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Units
<b>UNDER VOLTAGE PROTECTION</b>						
$V_{DDR}$	$V_{DD}$ Rising Threshold		<b>6.0</b>	6.9	<b>7.4</b>	V
$V_{DDH}$	$V_{DD}$ Threshold Hysteresis			0.5		V
$V_{HBR}$	HB Rising Threshold		<b>5.7</b>	6.6	<b>7.1</b>	V
$V_{HBH}$	HB Threshold Hysteresis			0.4		V
<b>BOOT STRAP DIODE</b>						
$V_{DL}$	Low-Current Forward Voltage	$I_{VDD-HB} = 100\text{ }\mu\text{A}$		0.60	<b>0.9</b>	V
$V_{DH}$	High-Current Forward Voltage	$I_{VDD-HB} = 100\text{ mA}$		0.85	<b>1.1</b>	V
$R_D$	Dynamic Resistance	$I_{VDD-HB} = 100\text{ mA}$		0.8	<b>1.5</b>	$\Omega$
<b>LO GATE DRIVER</b>						
$V_{OLL}$	Low-Level Output Voltage	$I_{LO} = 100\text{ mA}$		0.25	<b>0.4</b>	V
$V_{OHL}$	High-Level Output Voltage	$I_{LO} = -100\text{ mA}$ , $V_{OHL} = V_{DD} - V_{LO}$		0.35	<b>0.55</b>	V
$I_{OHL}$	Peak Pullup Current	$V_{LO} = 0\text{V}$		1.6		A
$I_{OLL}$	Peak Pulldown Current	$V_{LO} = 12\text{V}$		1.8		A
<b>HO GATE DRIVER</b>						
$V_{OLH}$	Low-Level Output Voltage	$I_{HO} = 100\text{ mA}$		0.25	<b>0.4</b>	V
$V_{OHH}$	High-Level Output Voltage	$I_{HO} = -100\text{ mA}$ , $V_{OHH} = V_{HB} - V_{HO}$		0.35	<b>0.55</b>	V
$I_{OHH}$	Peak Pullup Current	$V_{HO} = 0\text{V}$		1.6		A
$I_{OLH}$	Peak Pulldown Current	$V_{HO} = 12\text{V}$		1.8		A
<b>THERMAL RESISTANCE</b>						
$\theta_{JA}$ <sup>(2)</sup>	Junction to Ambient	VSSOP		200		$^\circ\text{C/W}$
		WSN-10 <sup>(3)</sup>		40		

(2) The  $\theta_{JA}$  is not a given constant for the package and depends on the printed circuit board design and the operating environment.

(3) 4 layer board with Cu finished thickness 1.5/1/1/1.5 oz. Maximum die size used. 5x body length of Cu trace on PCB top. 50 x 50mm ground and power planes embedded in PCB. See Application Note AN-1187.

## Switching Characteristics

Specifications in standard typeface are for  $T_J = +25^\circ\text{C}$ , and those in **boldface type** apply over the full **operating junction temperature range**. Unless otherwise specified,  $V_{DD} = V_{HB} = 12\text{V}$ ,  $V_{SS} = V_{HS} = 0\text{V}$ , No Load on LO or HO.

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Units
$t_{LPHL}$	Lower Turn-Off Propagation Delay LM5102 (LI Falling to LO Falling)			27	<b>56</b>	ns
$t_{HPLH}$	Upper Turn-Off Propagation Delay LM5102 (HI Falling to HO Falling)			27	<b>56</b>	ns
$t_{RC}, t_{FC}$	Either Output Rise/Fall Time	$C_L = 1000\text{ pF}$		15		ns
$t_R, t_F$	Either Output Rise/Fall Time (3V to 9V)	$C_L = 0.1\text{ }\mu\text{F}$		0.6		$\mu\text{s}$
$t_{BS}$	Bootstrap Diode Turn-Off Time	$I_F = 20\text{ mA}$ , $I_R = 200\text{ mA}$		50		ns

(1) Min and Max limits are 100% production tested at  $25^\circ\text{C}$ . Limits over the operating temperature range are specified through correlation using Statistical Quality Control (SQC) methods. Limits are used to calculate TI's Average Outgoing Quality Level (AOQL).

## Typical Performance Characteristics

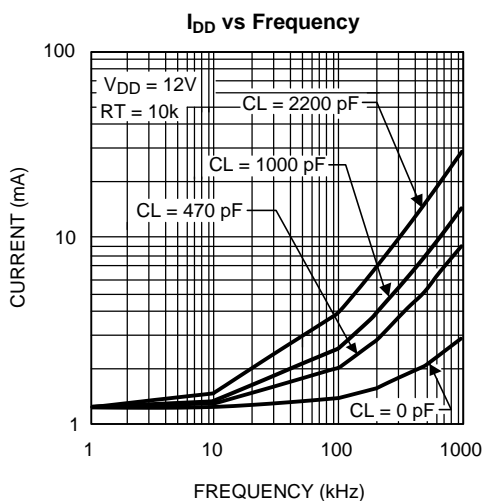


Figure 2.

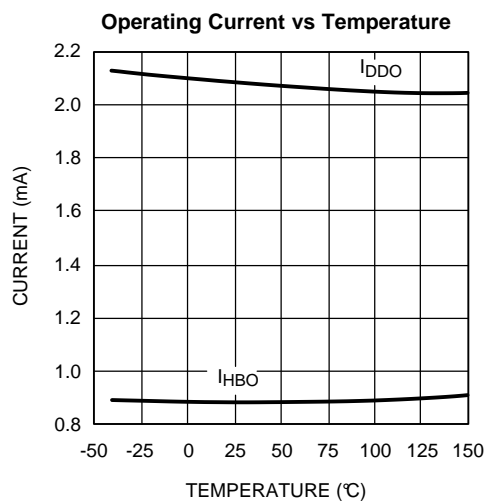


Figure 3.

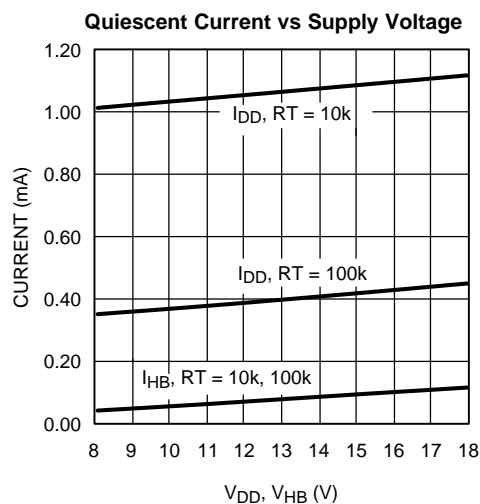


Figure 4.

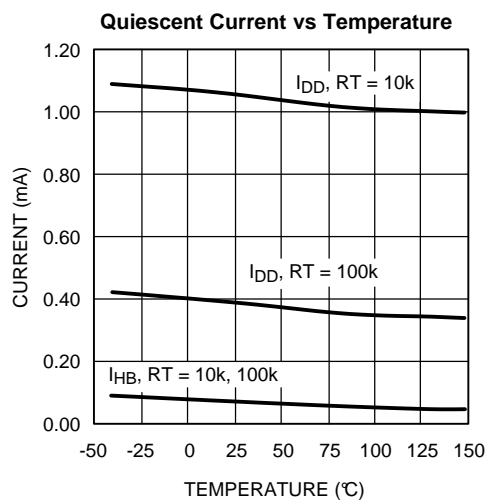


Figure 5.

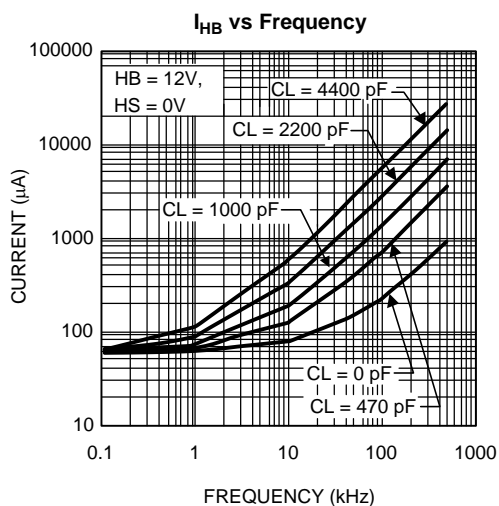


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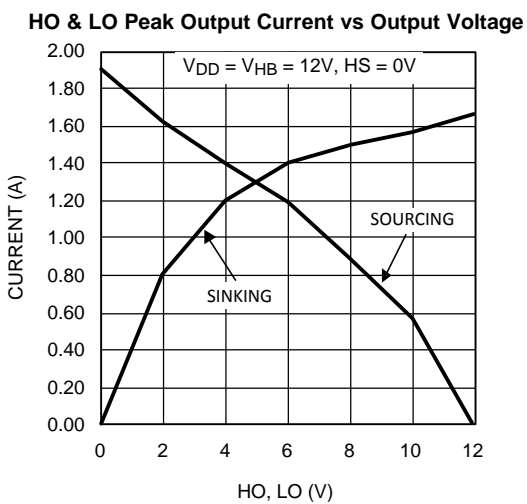


Figure 7.

## Typical Performance Characteristics (continued)

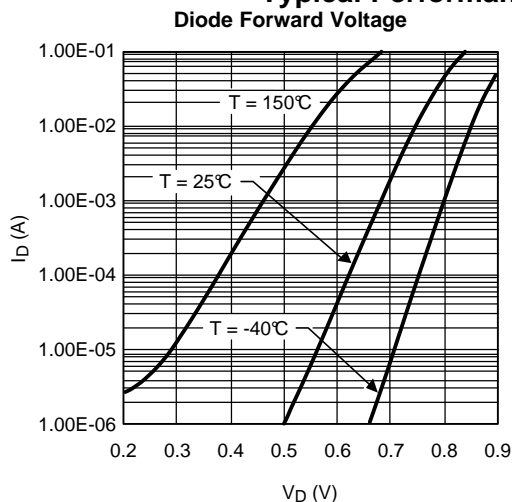


Figure 8.

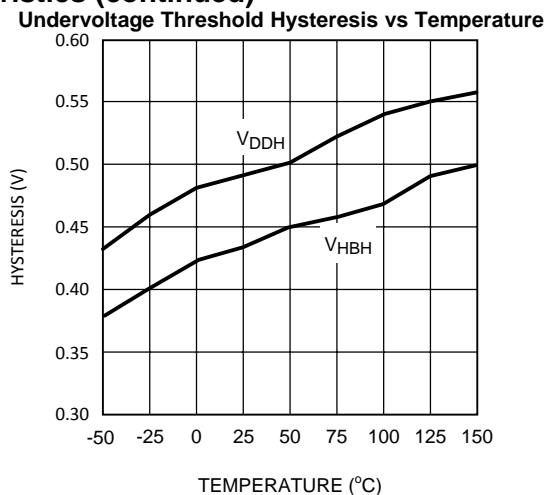


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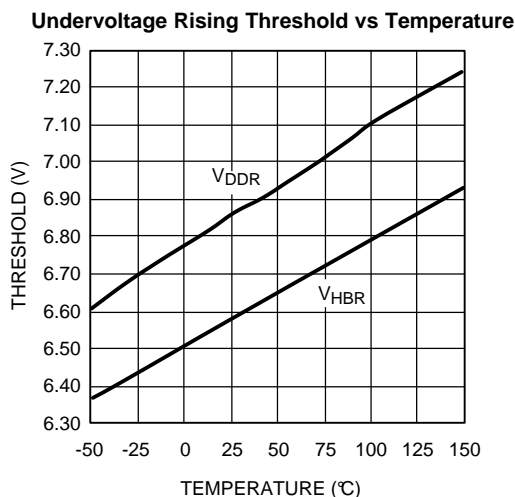


Figure 10.

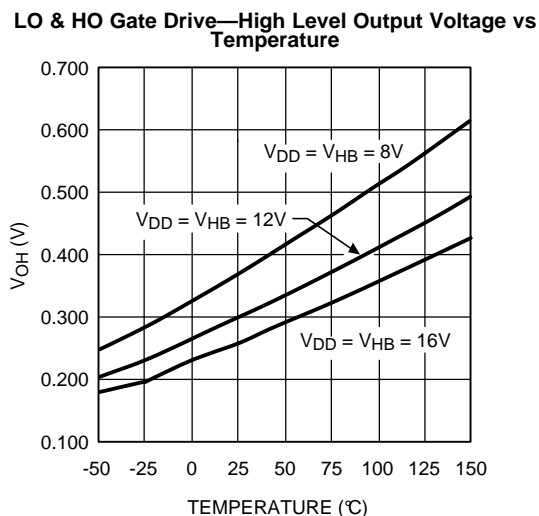


Figure 11.

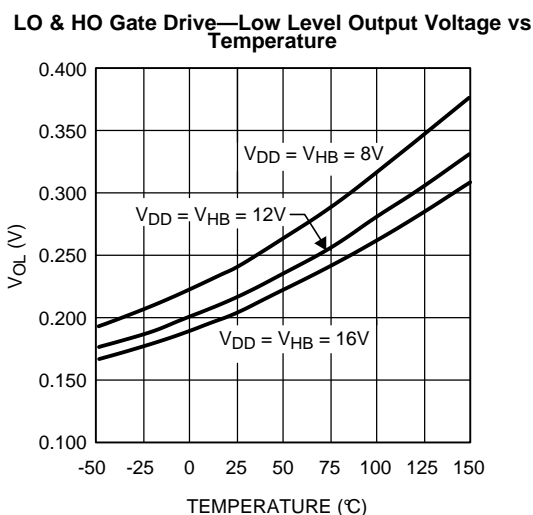


Figure 12.

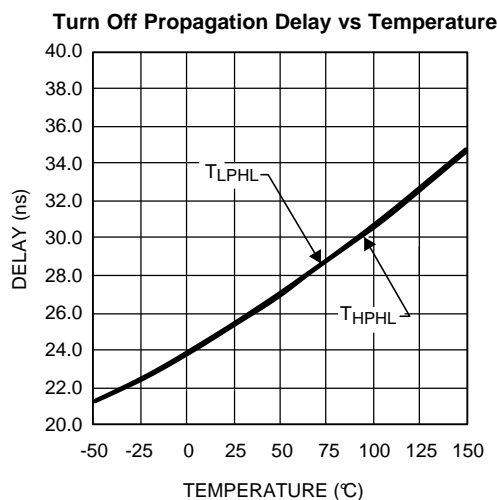


Figure 13.

## Typical Performance Characteristics (continued)

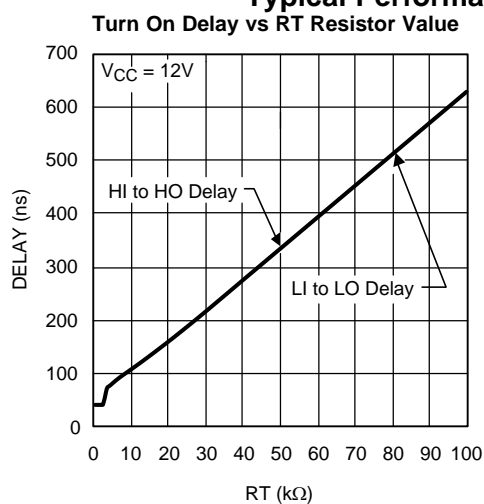


Figure 14.

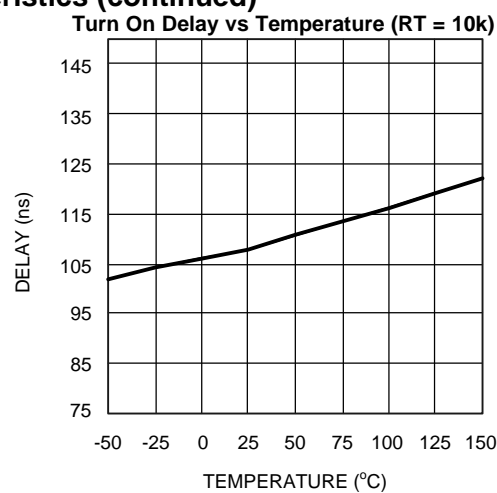


Figure 15.

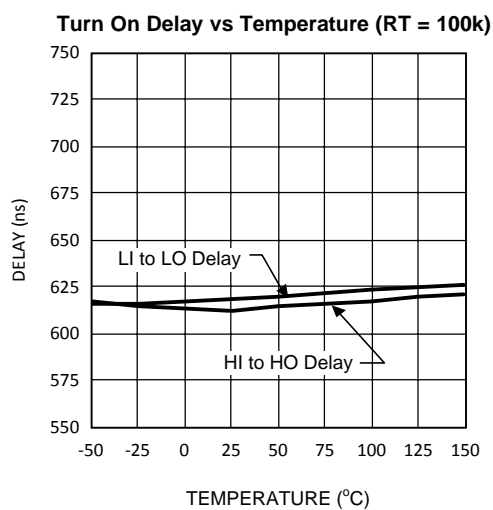


Figure 16.



## LM5102 Waveforms

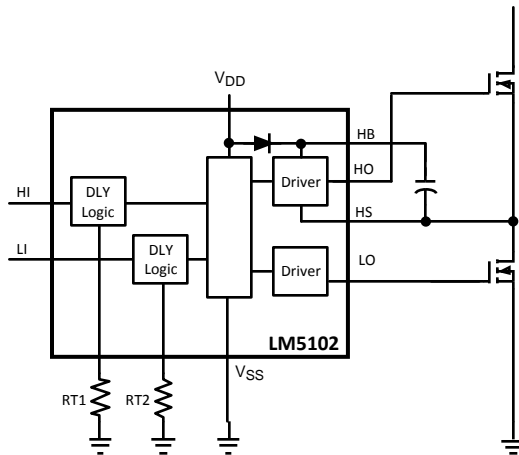


Figure 17.

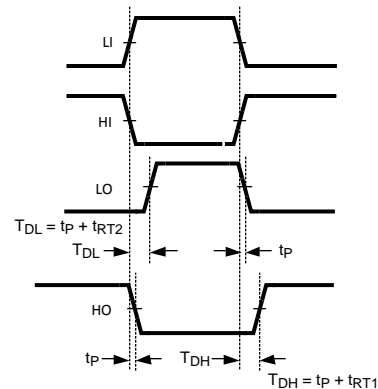


Figure 18. Application Timing Waveforms

## Operational Notes

The LM5102 offers a unique flexibility with independently programmable delay of the rising edge for both high and low side driver outputs independently. The delays are set with resistors at the RT1 and RT2 pins, and can be adjusted from 100 ns to 600 ns. This feature reduces component count, board space and cost compared to discrete solutions for adjusting driver dead time. The wide delay programming range provides the flexibility to optimize drive signal timing for a wide range of MOSFETs and applications.

The RT pins are biased at 3V and current limited to 1 mA maximum programming current. The time delay generator will accommodate resistor values from 5k to 100k with turn-on delay times that are proportional to the RT resistance. In addition, each RT pin is monitored by a comparator that will bypass the turn-on delay if the RT pin is pulled below the timer elimination threshold (1.8V typical). Grounding the RT pins programs the LM5102 to drive both outputs with minimum turn-on delay.

## STARTUP AND UVLO

Both top and bottom drivers include under-voltage lockout (UVLO) protection circuitry which monitors the supply voltage ( $V_{DD}$ ) and bootstrap capacitor voltage ( $V_{HB} - V_{HS}$ ) independently. The UVLO circuit inhibits each driver until sufficient supply voltage is available to turn-on the external MOSFETs, and the built-in hysteresis prevents chattering during supply voltage transitions. When the supply voltage is applied to  $V_{DD}$  pin of LM5102, the top and bottom gates are held low until  $V_{DD}$  exceeds UVLO threshold, typically about 6.9V. Any UVLO condition on the bootstrap capacitor will disable only the high side output (HO).

## LAYOUT CONSIDERATIONS

The optimum performance of high and low side gate drivers cannot be achieved without taking due considerations during circuit board layout. Following points are emphasized.

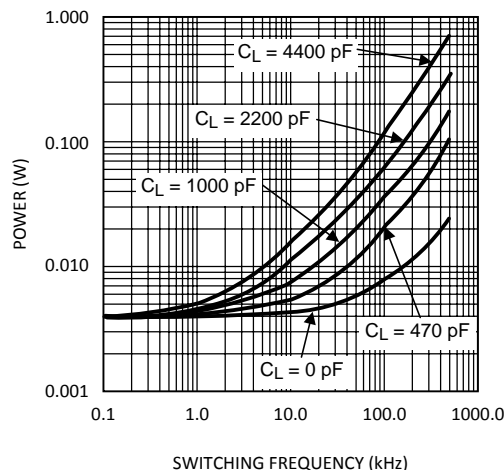
1. A low ESR/ESL capacitor must be connected close to the IC, and between  $V_{DD}$  and  $V_{SS}$  pins and between HB and HS pins to support high peak currents being drawn from  $V_{DD}$  during turn-on of the external MOSFET.
2. To prevent large voltage transients at the drain of the top MOSFET, a low ESR electrolytic capacitor must be connected between MOSFET drain and ground ( $V_{SS}$ ).
3. In order to avoid large negative transients on the switch node (HS) pin, the parasitic inductances in the source of top MOSFET and in the drain of the bottom MOSFET (synchronous rectifier) must be minimized.
4. Grounding considerations:
  - The first priority in designing grounding connections is to confine the high peak currents from charging and discharging the MOSFET gate in a minimal physical area. This will decrease the loop inductance and minimize noise issues on the gate terminal of the MOSFET. The MOSFETs should be placed as close as possible to the gate driver.
  - The second high current path includes the bootstrap capacitor, the bootstrap diode, the local ground referenced bypass capacitor and low side MOSFET body diode. The bootstrap capacitor is recharged on the cycle-by-cycle basis through the bootstrap diode from the ground referenced  $V_{DD}$  bypass capacitor. The recharging occurs in a short time interval and involves high peak current. Minimizing this loop length and area on the circuit board is important to ensure reliable operation.
5. The resistors on the RT1 and RT2 timer pins must be placed very close to the IC and separated from high current paths to avoid noise coupling to the time delay generator which could disrupt timer operation.

## POWER DISSIPATION CONSIDERATIONS

The total IC power dissipation is the sum of the gate driver losses and the bootstrap diode losses. The gate driver losses are related to the switching frequency ( $f$ ), output load capacitance on LO and HO ( $C_L$ ), and supply voltage ( $V_{DD}$ ) and can be roughly calculated as:

$$P_{DGATES} = 2 \cdot f \cdot C_L \cdot V_{DD}^2 \quad (1)$$

There are some additional losses in the gate drivers due to the internal CMOS stages used to buffer the LO and HO outputs. The following plot shows the measured gate driver power dissipation versus frequency and load capacitance. At higher frequencies and load capacitance values, the power dissipation is dominated by the power losses driving the output loads and agrees well with the above equation. This plot can be used to approximate the power losses due to the gate drivers.



**Figure 19. Gate Driver Power Dissipation (LO + HO)**  
 $V_{CC} = 12V$ , Neglecting Diode Losses

The bootstrap diode power loss is the sum of the forward bias power loss that occurs while charging the bootstrap capacitor and the reverse bias power loss that occurs during reverse recovery. Since each of these events happens once per cycle, the diode power loss is proportional to frequency. Larger capacitive loads require more current to recharge the bootstrap capacitor resulting in more losses. Higher input voltages ( $V_{IN}$ ) to the half bridge result in higher reverse recovery losses. The following plot was generated based on calculations and lab measurements of the diode recovery time and current under several operating conditions. This can be useful for approximating the diode power dissipation.

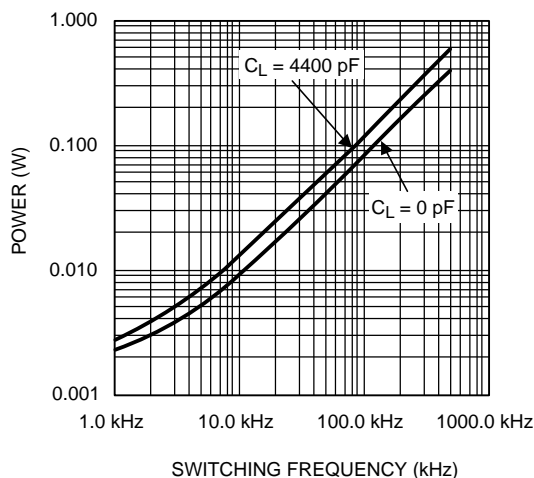


Figure 20. Diode Power Dissipation  $V_{IN} = 80V$

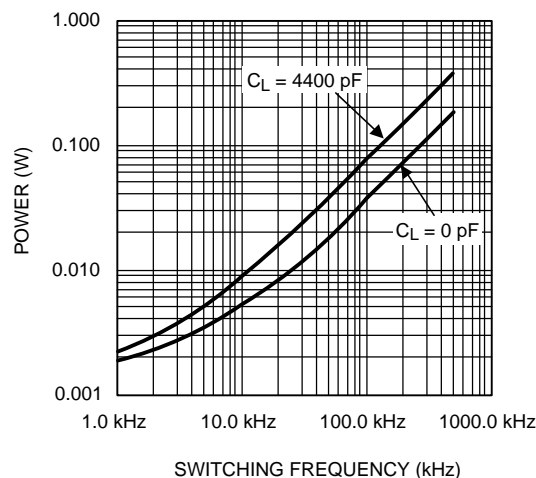


Figure 21. Diode Power Dissipation  $V_{IN} = 40V$

The total IC power dissipation can be estimated from the above plots by summing the gate drive losses with the bootstrap diode losses for the intended application. Because the diode losses can be significant, an external diode placed in parallel with the internal bootstrap diode (refer to Figure 22) and can be helpful in removing power from the IC. For this to be effective, the external diode must be placed close to the IC to minimize series inductance and have a significantly lower forward voltage drop than the internal diode.

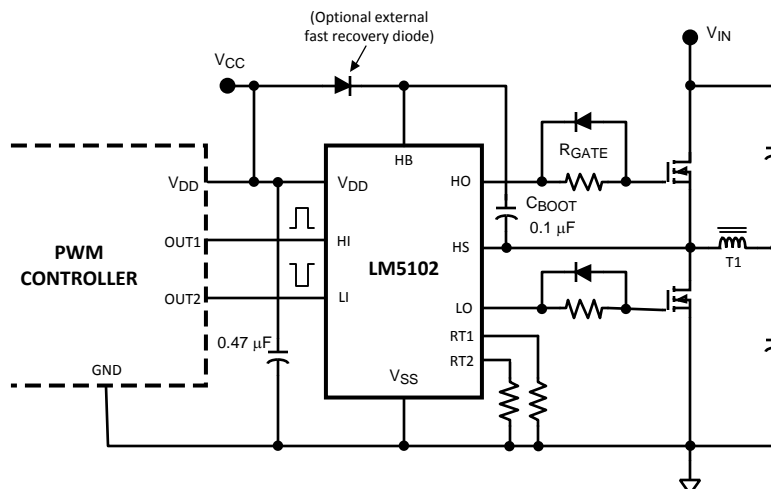


Figure 22. LM5102 Driving MOSFETs Connected in Half-Bridge Configuration

## REVISION HISTORY

Changes from Original (March 2013) to Revision A	Page
• Changed layout of National Data Sheet to TI format .....	<a href="#">11</a>

## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM5102MM	ACTIVE	VSSOP	DGS	10		TBD	Call TI	Call TI	-40 to 125	5102	<a href="#">Samples</a>
LM5102MM/NOPB	ACTIVE	VSSOP	DGS	10	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	5102	<a href="#">Samples</a>
LM5102MMX	ACTIVE	VSSOP	DGS	10		TBD	Call TI	Call TI	-40 to 125	5102	<a href="#">Samples</a>
LM5102MMX/NOPB	ACTIVE	VSSOP	DGS	10	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	5102	<a href="#">Samples</a>
LM5102SD	ACTIVE	WSO	DPR	10	1000	TBD	Call TI	Call TI	-40 to 125	5102SD	<a href="#">Samples</a>
LM5102SD/NOPB	ACTIVE	WSO	DPR	10	1000	Green (RoHS & no Sb/Br)	SN	Level-1-260C-UNLIM	-40 to 125	5102SD	<a href="#">Samples</a>
LM5102SDX	ACTIVE	WSO	DPR	10		TBD	Call TI	Call TI	-40 to 125	5102SD	<a href="#">Samples</a>
LM5102SDX/NOPB	ACTIVE	WSO	DPR	10	4500	Green (RoHS & no Sb/Br)	SN	Level-1-260C-UNLIM	-40 to 125	5102SD	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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**TAPE AND REEL INFORMATION**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM5102MM/NOPB	VSSOP	DGS	10	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LM5102MMX/NOPB	VSSOP	DGS	10	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LM5102SD	WSOP	DPR	10	1000	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1
LM5102SD/NOPB	WSOP	DPR	10	1000	178.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1
LM5102SDX/NOPB	WSOP	DPR	10	4500	330.0	12.4	4.3	4.3	1.3	8.0	12.0	Q1

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM5102MM/NOPB	VSSOP	DGS	10	1000	210.0	185.0	35.0
LM5102MMX/NOPB	VSSOP	DGS	10	3500	367.0	367.0	35.0
LM5102SD	WSO	DPR	10	1000	210.0	185.0	35.0
LM5102SD/NOPB	WSO	DPR	10	1000	210.0	185.0	35.0
LM5102SDX/NOPB	WSO	DPR	10	4500	367.0	367.0	35.0

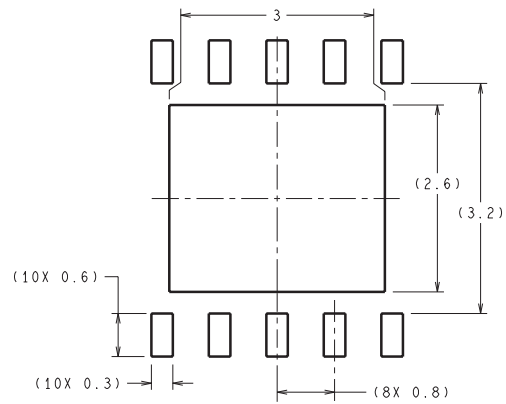


## DGS (S-PDSO-G10)

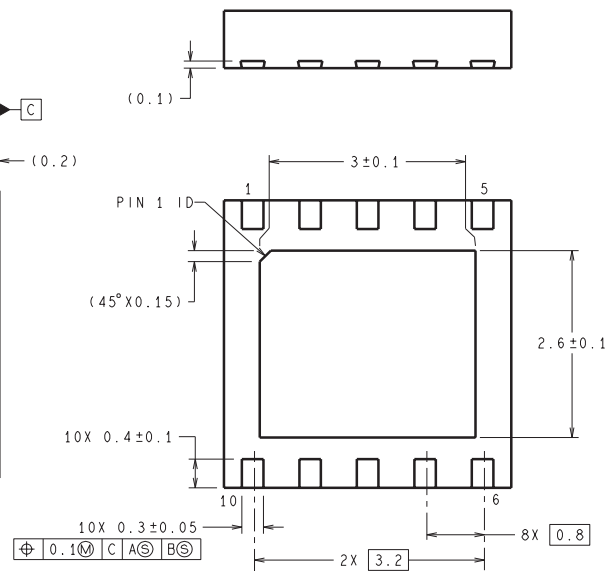
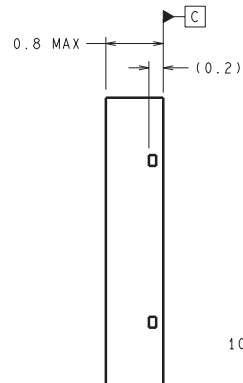
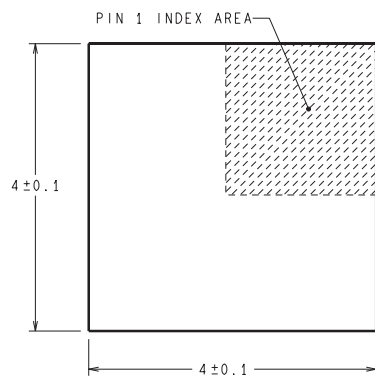
## PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Body dimensions do not include mold flash or protrusion.
  - Falls within JEDEC MO-187 variation BA.



### RECOMMENDED LAND PATTERN



**DIMENSIONS ARE IN MILLIMETERS**

SDC10A (Rev A)

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