



LA4815M

Monolithic Linear IC Monaural Power Amplifier

Overview

The LA4815M incorporates a 1-channel power amplifier with a wide operating supply voltage range built into a surface-mounted package. This IC also has a mute function and requires only a few external components, making it suitable for low-cost set design. There is also a surface-mounted package type with heat sink (LA4815VH).

Applications

Intercoms, door phones, transceivers, radios, toys, home appliances with voice guidance, etc.

Features

- Built-in 1-channel power amplifier
 - Output power 1 = 370mW typ. (V_{CC} = 6V, R_L = 8Ω, THD = 10%)
 - Output power 2 = 620mW typ. (V_{CC} = 6V, R_L = 4Ω, THD = 10%)
 - Output power 3 = 230mW typ. (V_{CC} = 5V, R_L = 8Ω, THD = 10%)
 - Output power 4 = 1,000mW typ. (V_{CC} = 12V, R_L = 16Ω, THD = 10%)
- Mute function
- Selectable voltage gain : 2 types
 - 26dB/40dB
 - * Gain values between 26 and 40dB can also be set by adding external components (two resistors).
- Only a few external components
 - 4 components/total
- Wide supply voltage range
 - 4 to 13V (When using 9V or more, another package product, LA4815VH, is recommended.)

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Specifications**Maximum Ratings** at $T_a = 25^\circ\text{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
|------------------------------|--------------|------------------------|-------------|------------------|
| Maximum power supply voltage | V_{CC} max | | 15 | V |
| Allowable power dissipation | P_d max | * Mounted on the board | 0.9 | W |
| Maximum junction temperature | T_j max | | 150 | $^\circ\text{C}$ |
| Operating temperature | T_{opr} | | -30 to +75 | $^\circ\text{C}$ |
| Storage temperature | T_{stg} | | -40 to +150 | $^\circ\text{C}$ |

* Mounted on SANYO evaluation board : Double-sided board with dimensions of 50mm \times 50mm \times 1.6mm (glass epoxy)

Operating Conditions at $T_a = 25^\circ\text{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
|--|-------------|------------|---------|----------|
| Recommended power supply voltage | V_{CC} | | 6 | V |
| Recommended load resistance | R_L | | 4 to 32 | Ω |
| Allowable operating supply voltage range | V_{CC} op | | 4 to 13 | V |

* The supply voltage level to be used must be determined with due consideration given to the allowable power dissipation of the IC.

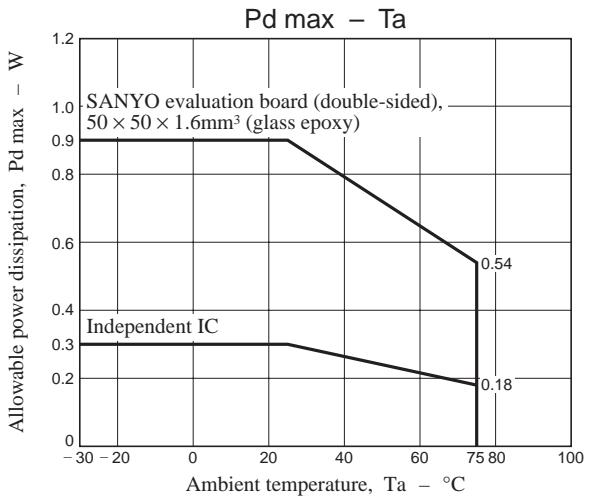
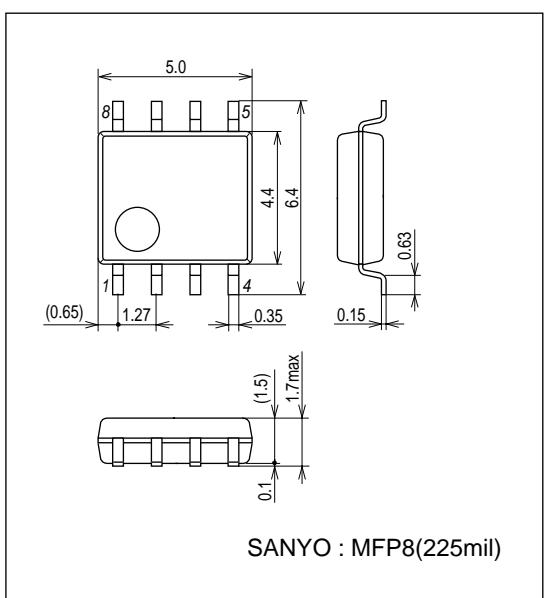
Electrical Characteristics at $T_a = 25^\circ\text{C}$, $V_{CC} = 6\text{V}$, $R_L = 8\Omega$, $f_{in} = 1\text{kHz}$

| Parameter | Symbol | Conditions | Ratings | | | Unit |
|----------------------------|--------------|--|---------|------|------|------------------|
| | | | min | typ | max | |
| Quiescent current drain-1 | I_{CCOP1} | No signal | | 4.6 | 8.3 | mA |
| Quiescent current drain-2 | I_{CCOP2} | No signal, pin 7 = LOW | | 1.6 | | mA |
| Maximum output power-1 | P_{OMAX1} | THD = 10% | 250 | 370 | | mW |
| Maximum output power-2 | P_{OMAX2} | THD = 10%, $R_L = 4\Omega$ | | 620 | | mW |
| Voltage gain-1 | $VG1$ | $V_{IN} = -30\text{dBV}$ | 23.9 | 25.9 | 27.9 | dB |
| Voltage gain-2 | $VG2$ | $V_{IN} = -40\text{dBV}$, pin 1/pin8 = GND | 37 | 39.5 | 42 | dB |
| Total harmonic distortion | THD | $V_{IN} = -30\text{dBV}$ | | 0.11 | 0.7 | % |
| Mute attenuation | MT | $V_{IN} = -10\text{dBV}$, pin 7 = LOW | -90 | -115 | | dBV |
| Output noise voltage | V_{NOUT} | $R_g = 620\Omega$, 20 to 20kHz | | 40 | 100 | μVrms |
| Ripple rejection ratio | SVRR | $R_g = 620\Omega$, $f_r = 100\text{Hz}$, $V_r = -20\text{dBV}$ | | 44 | | dB |
| Mute control voltage-LOW | V_{7cntL} | Mute mode | | | 0.3 | V |
| Mute control voltage-HIGH1 | V_{7cntH1} | Mute released, $V_{CC} = 6.5\text{V}$ or lower | 1.8 | | | V |
| Mute control voltage-HIGH2 | V_{7cntH2} | Mute released, $V_{CC} = 6.5\text{V}$ or higher | 2.4 | | | V |
| Input resistance | R_i | | | 100 | | $\text{k}\Omega$ |

Package Dimensions

unit : mm (typ)

3032D

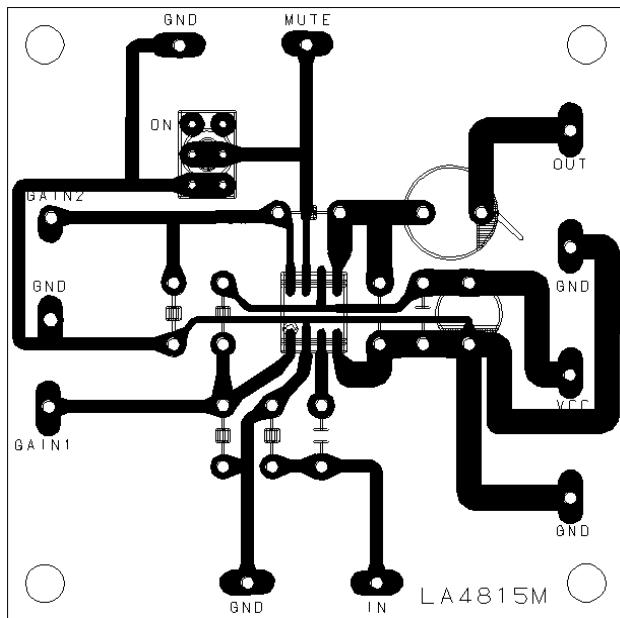


Evaluation board

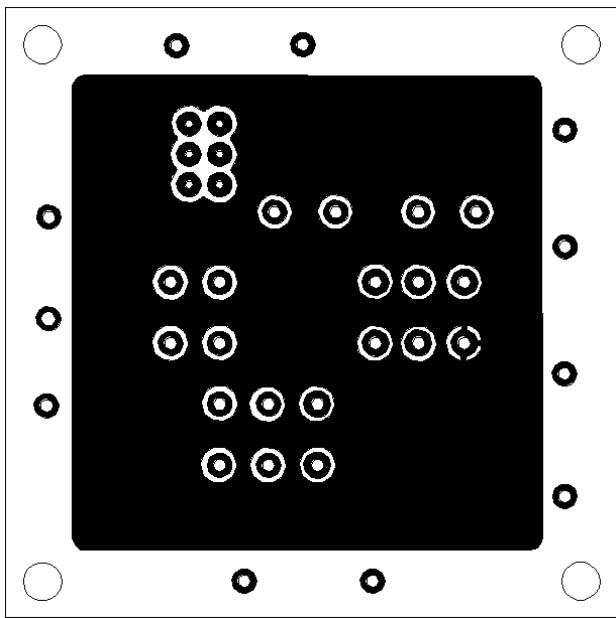
1. Double-sided circuit board

Dimensions : 50mm × 50mm × 1.6mm

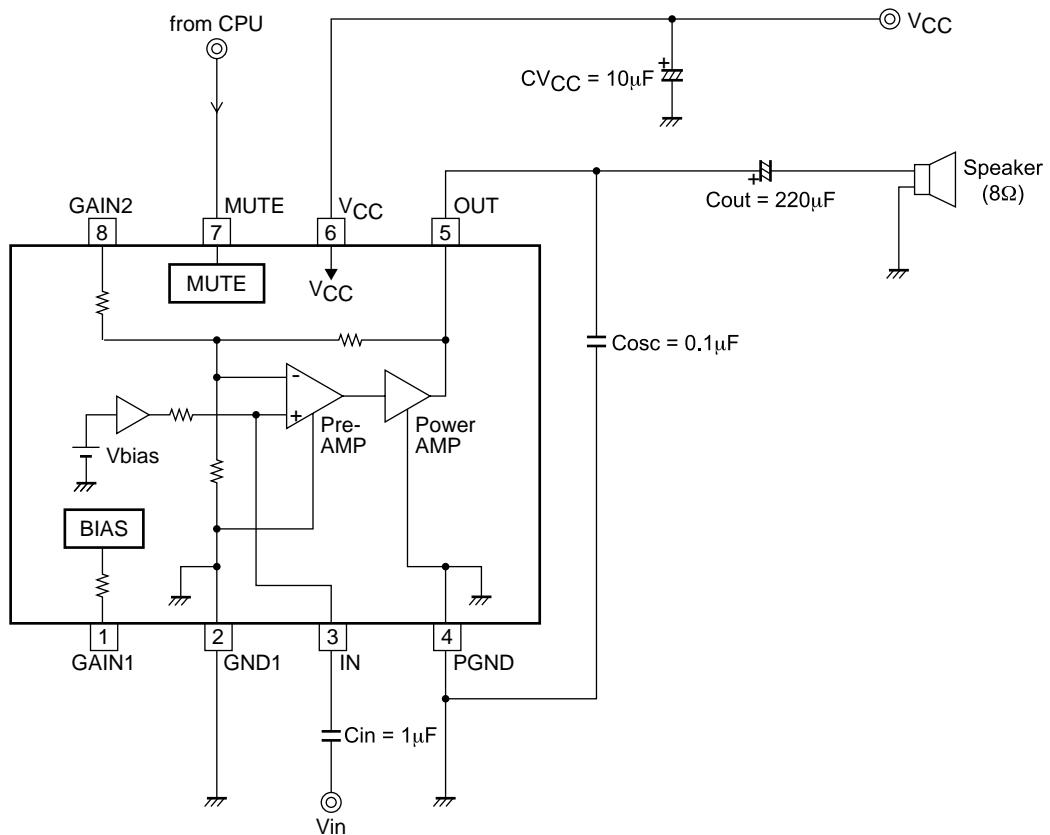
Top Layer(Top view)



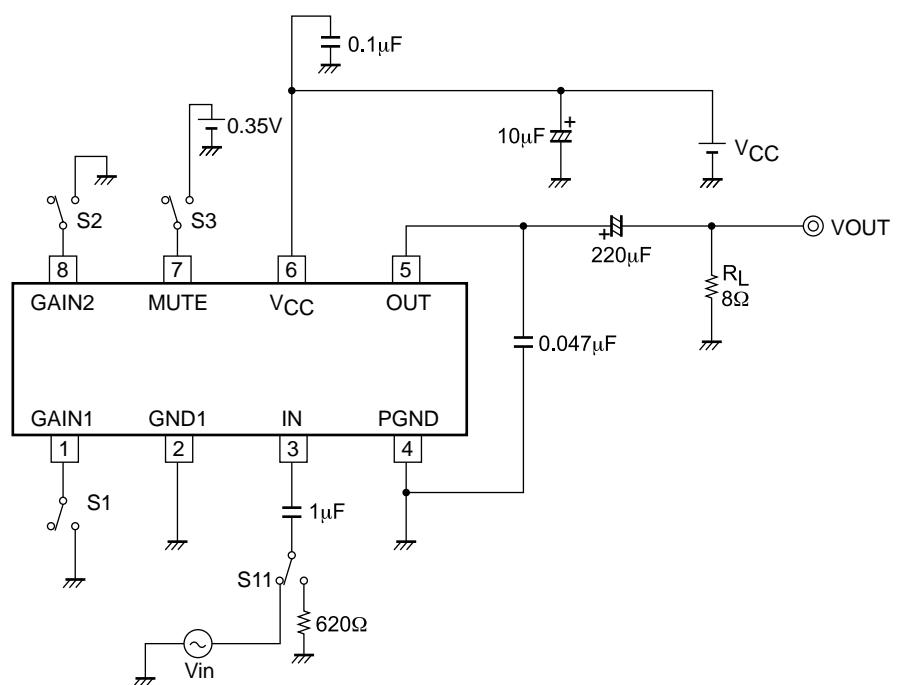
Bottom Layer(Top view)



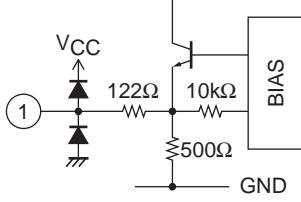
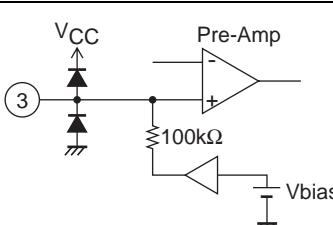
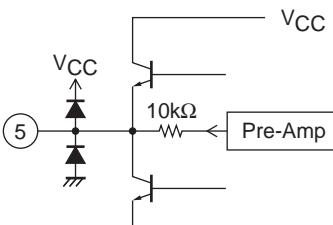
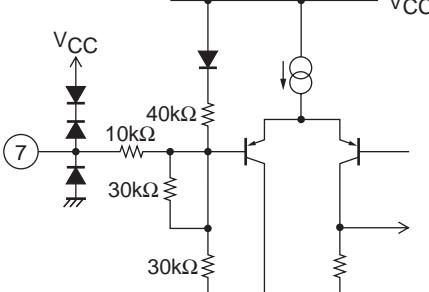
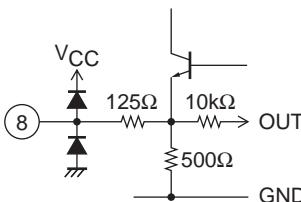
Block Diagram and Sample Application Circuit



Test Circuit



Pin Functions

| Pin No. | Pin Name | Pin Voltage $V_{CC} = 6V$ | Description | Equivalent Circuit |
|---------|----------|------------------------------|---|---|
| 1 | GAIN1 | 0.2 | Gain switching pin. • 26dB mode when left open. • 40dB mode when connected to ground. (Both pins 1 and 8 must be reconfigured at the same time.) |  |
| 2 | GND1 | 0 | Preamplifier system ground pin. | |
| 3 | IN | 1.57 | Input pin. |  |
| 4 | PGND | 0 | Power amplifier ground pin. | |
| 5 | OUT | 2.94 | Power amplifier output pin. |  |
| 6 | VCC | 6 | Power supply pin. | |
| 7 | MUTE | 2.3 | Mute control pin. • Mute ON \Rightarrow Low • Mute OFF \Rightarrow High |  |
| 8 | GAIN2 | 0.2 | Gain switching pin. • 26dB mode when left open. • 40dB mode when connected to ground. (Both pins 1 and 8 must be reconfigured at the same time.) |  |

Notes on Using the IC

1. Voltage gain settings (Pins 1 and 8)

The voltage gain of the power amplifier is fixed by the internal resistors.

- Pins 1 and 8 be left open : Approximately 26dB
- Pins 1 and 8 connected to GND : Approximately 39.5dB

Note that the voltage gain can be changed using two resistors. (See Fig. 1)

- Voltage gain setting : According to the resistor connected between Pin 8 and Pin 2 (GND1)

$$\text{* Voltage gain} = 20\log(20 \times (625 + R_{vg1})/(125 + R_{vg1}))$$

- Output DC voltage setting : According to the resistor connected between Pin 1 and Pin 2 (GND1)

$$\text{* } R_{vg1} = R_{vg2} \text{ must be satisfied.}$$

In addition, the voltage gain can also be lowered to approximately 20dB (when using 5V or 6V power supply) by an application such as shown in Fig. 2 below.

- Voltage gain setting : According to the resistor connected between Pin 8 and Pin 5 (OUT)

$$\text{* Voltage gain} = 20\log(20 \times (125 + R_{vg3})/(10,125 + R_{vg3}))$$

- Output DC voltage setting : According to the resistor connected between Pin 1 and Pin 6 (V_{CC})

$$\text{* Set the resistor values so that the Pin 5 (OUT) DC voltage is approximately half the supply voltage.}$$

Example : When $R_{vg3} = 10k\Omega$, $R_{vg4} = 22k\Omega$ (when $V_{CC} = 6V$)

However, note that using this method to greatly lower the voltage gain deteriorates the characteristics, so the voltage gain should be lowered only to approximately 20dB. In addition, when using a high supply voltage (7V or more), the clipped waveform may invert, so this voltage gain reduction method must not be used in these cases.

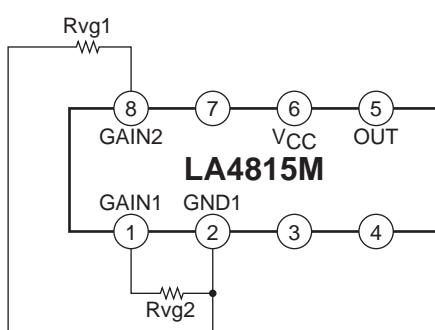


Figure 1

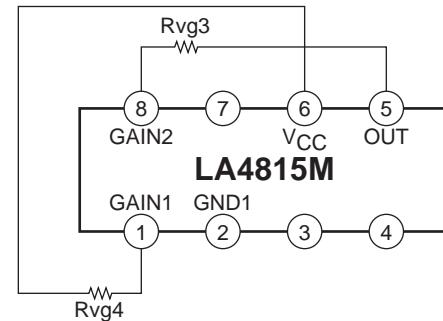


Figure 2

2. Signal source impedance : r_g

The signal source impedance value r_g affects the ripple rejection ratio together with input coupling capacitor C_{in} , so r_g should be as small as possible. Therefore, when attenuating the signal at the C_{in} front end as shown in Fig. 4, the constants should be set in consideration of these characteristics. Using the smallest resistor R_{g1} value possible is recommended.

In addition, when setting the signal level, the voltage gain should be set on the LA4815M side and the input front-end should be configured using only the input coupling capacitor, C_{in} , as shown in Fig. 5 in order to maximize the ripple rejection ratio.

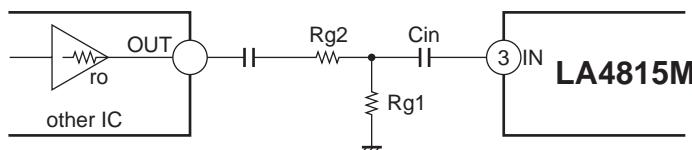


Figure 4

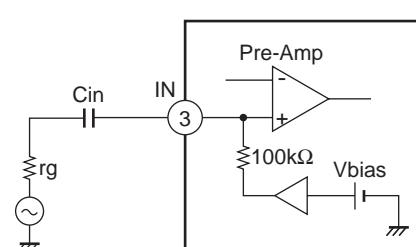


Figure 3

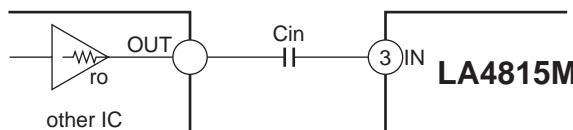


Figure 5

3. Mute control pin (Pin 7)

The internal power amplifier circuit can be disabled and audio mute is turned on by controlling the voltage applied to Pin 7. Control can be performed directly using the CPU output port, but digital noise from the CPU may worsen the LA4815M noise floor. Therefore, inserting a series resistor, Rm1 (1 to 2.2k Ω) as shown in Fig. 6, is recommended.

- Mute ON : Low
- Mute OFF : High or open

In addition, the Pin 7 DC voltage is dependent on the supply voltage, so a reverse current flows to the CPU power supply line when the Pin 7 voltage is higher than the CPU supply voltage. In these cases, connect a resistor, Rm2 (see Fig. 7) between Pin 7 and GND to lower the Pin 7 DC voltage as shown in Fig. 6.

Note that when not using the mute function, Pin 7 must be left open.

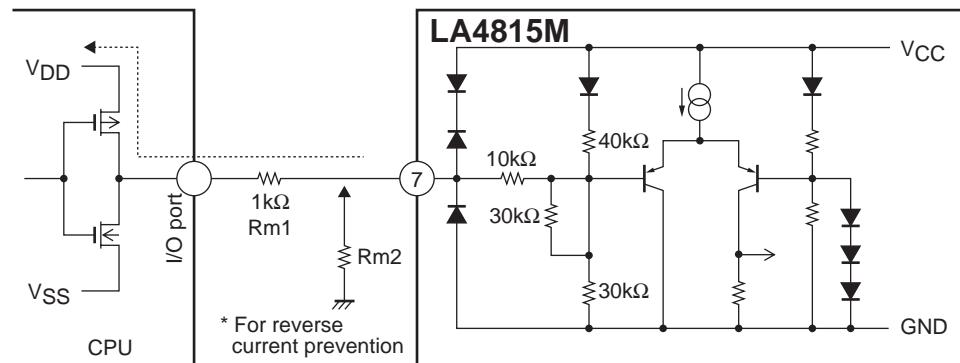


Figure 6

Reverse current prevention resistor value : Rm2 (reference value) ← When V7 is set to approximately 2.5V

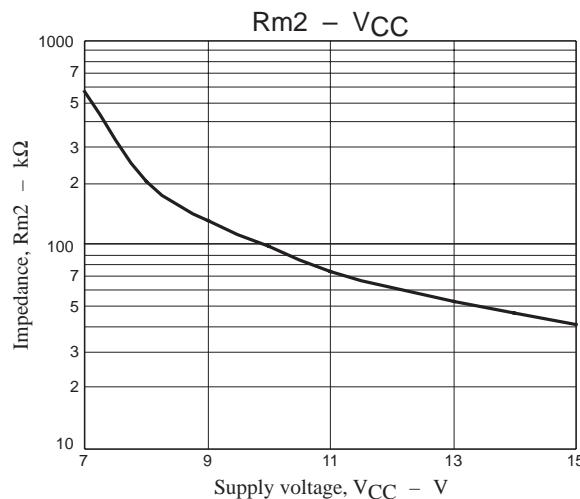


Figure 7

4. Mute control timing

When performing mute control, exercise control at the timing shown in Fig. 8.

During power-on : Twu = 0 to 50ms

* Pins 6 and 7 can also rise simultaneously.

During power-off : Twd = 100 to 200ms

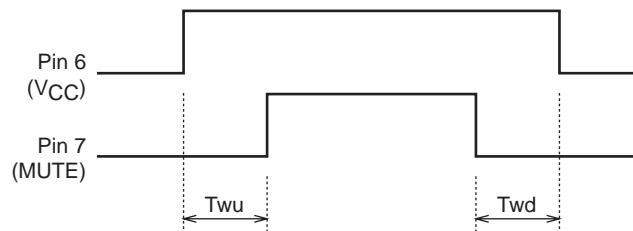


Figure 8

5. Popping noise reduction during power-off

The power supply line can be directly controlled ON and OFF without using the mute function. However, when using a high supply voltage, the shock noise and aftersound during power-off tends to worsen. One method of coping with this is to connect a capacitor between Pin 6 (VCC) and Pin 7 (MUTE) as shown in Fig. 9 so that the auto mute function operates during power-off.

Recommended value = $1\mu\text{F}$

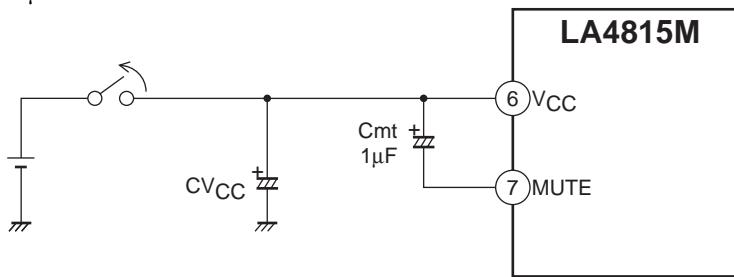


Figure 9

6. Input coupling capacitor (Cin)

C_{in} is an input coupling capacitor, and is used for DC cutting. However, this capacitor is also used to improve the ripple rejection ratio, which changes according to the capacitance value (recommended value = $1\mu\text{F}$). In addition, this capacitor also affects the transient response characteristics during power-on and when mute is canceled, so the constant should be set in consideration of these characteristics.

Design reference value = approximately 0.33 to $3.3\mu\text{F}$

- Ripple rejection ratio : Increasing the capacitance value increases the rate, and reducing the value reduces the rate.
- Rise response speed : Increasing the capacitance value reduces the speed, and reducing the value increases the speed.
- Popping noise : Increasing the capacitance value reduces the noise, and reducing the value increases the noise.

7. Output coupling capacitor (Cout)

C_{out} is an output coupling capacitor used for DC cutting. However, this capacitor, C_{out} , in combination with load impedance R_L , forms a high-pass filter and attenuates the low frequency signal. Take into account the cutoff frequency when determining the capacitance value. In addition, normally a chemical capacitor is used for this capacitor, but the capacitance value of chemical capacitors decreases at low temperatures, so the value should be set in accordance with this characteristic.

The cutoff frequency is expressed by the following formula.

$$f_c = 1/(2\pi \times R_L \times C_{out})$$

8. Output phase compensation capacitor (Cosc)

The C_{osc} capacitor is used to prevent output oscillation. Use a ceramic capacitor (recommended value = $0.1\mu\text{F}$) with good high frequency characteristics, and locate this capacitor as close to the IC as possible.

9. Power supply capacitor (CVCC)

The C_{VCC} capacitor is used to suppress the ripple component of the power supply line. Normally a chemical capacitor (recommended value = $10\mu\text{F}$) is used for this capacitor. However, chemical capacitors have poor high frequency characteristics, so when using a CPU, DSP or other IC that generates digital noise in the set, it is recommended that a power supply bypass capacitor (ceramic capacitor, recommended value = approximately $0.1\mu\text{F}$) be added to reject high-frequency components. Locate this bypass capacitor as close to the IC as possible.

10. Signal mixing methods

The following methods can be used to mix a beep, key tone or other signal into the audio signal. Note that when input to Pin 8 is selected, amplification of signals input from Pin 3 changes according to impedance Z_8 connected to Pin 8.

10-1. Mixing method using resistors in the Pin 3 input front end

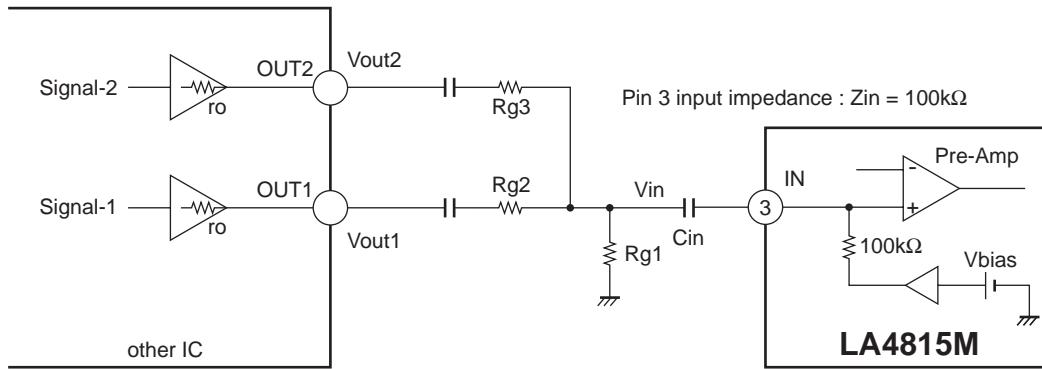


Figure 10

10-2. Method using input to Pin 8

- First signal system (Signal-1) voltage gain : $Vg1$
 $Vg1 = 20\log(Vout/Vin1) = 20\log(4 \times (125 + Z8) (500 + (125 \times Z8/(125 + Z8)))/(25 \times Z8))$
* $Z8 = R1 + ro$
- Second signal system (Signal-2) voltage gain : $Vg2$
 $Vg2 = 20\log(Vout/Vin2) = 20\log(10000/(125 + R1))$
* $fc2 = 1/(2\pi \times Cin2 \times (R1 + 125))$

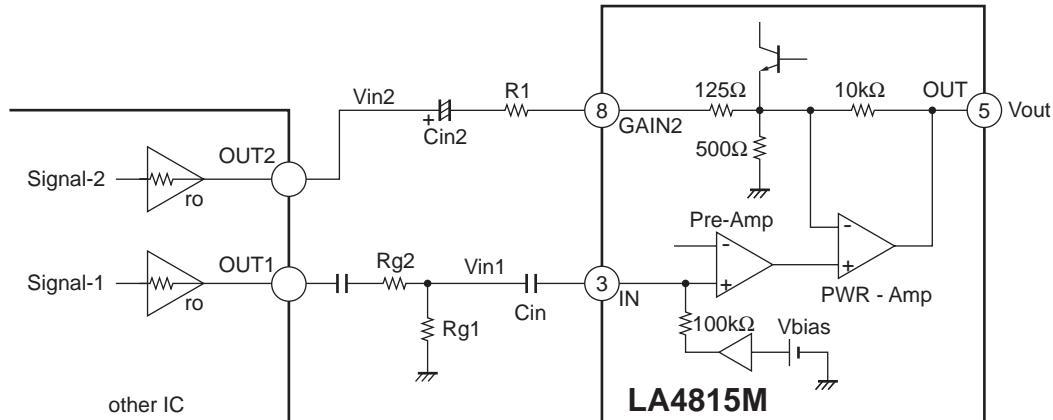


Figure 11

11. Short-circuit between pins

Turning on the power supply with some pins short-circuited may cause deterioration or breakdown. Therefore, when mounting the IC on a board, check to make sure that no short-circuit is formed between pins by solder or other foreign substances before turning on the power supply.

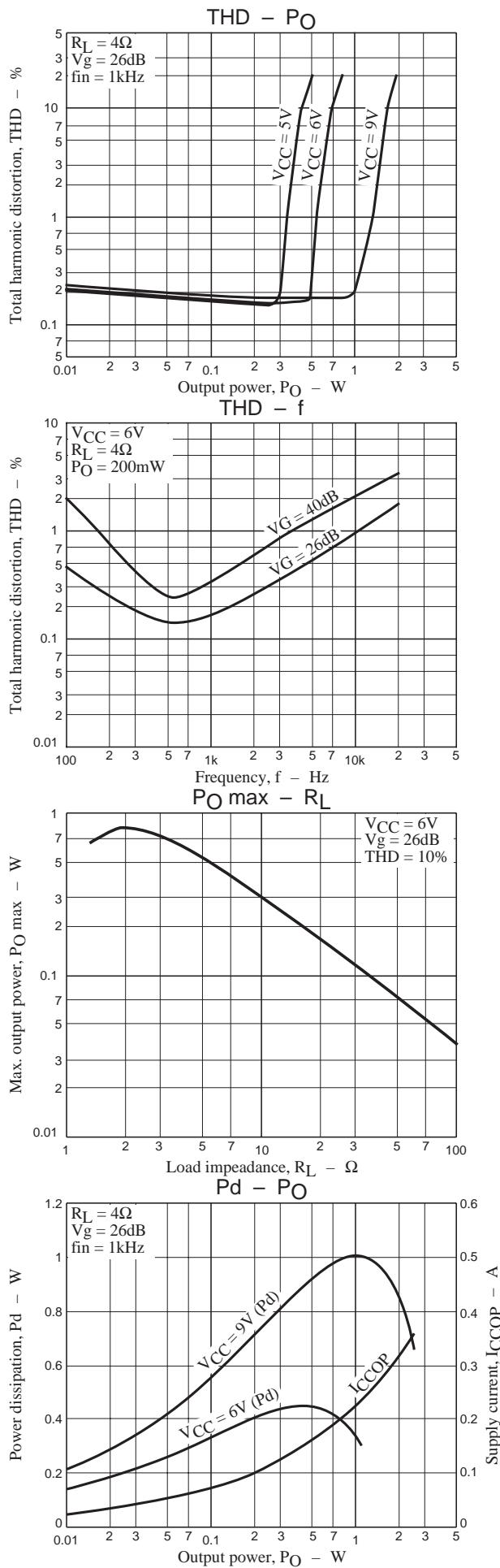
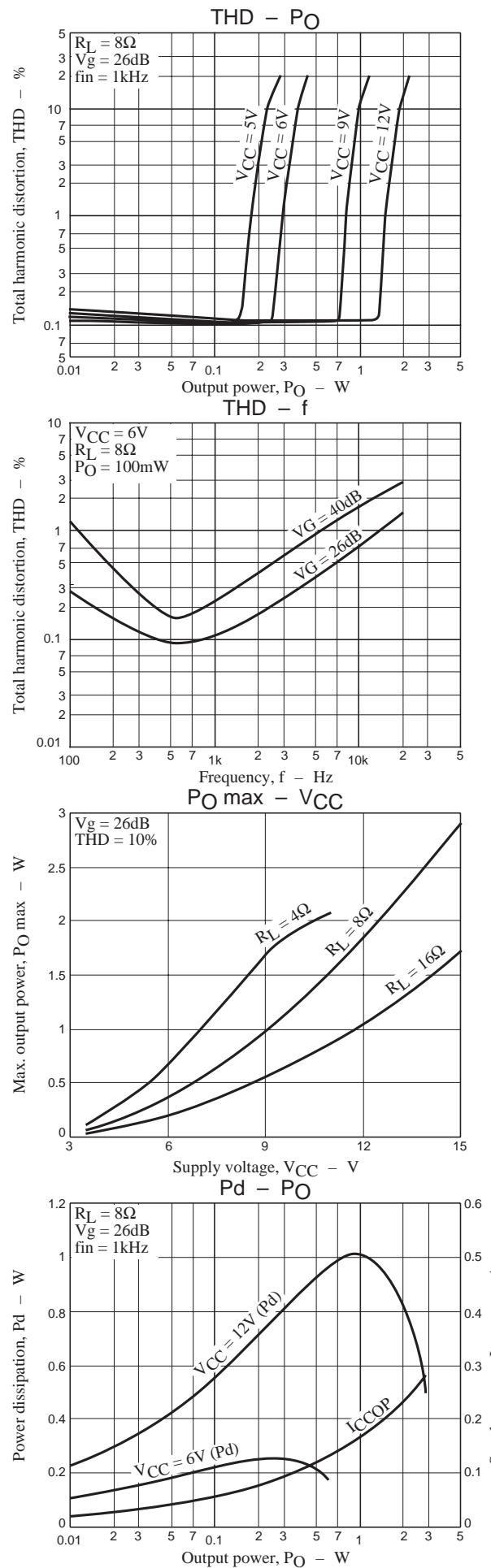
12. Load short circuit

Leaving the IC for a long time in the condition with a load short circuit may cause deterioration or breakdown. Therefore, never short-circuit the load.

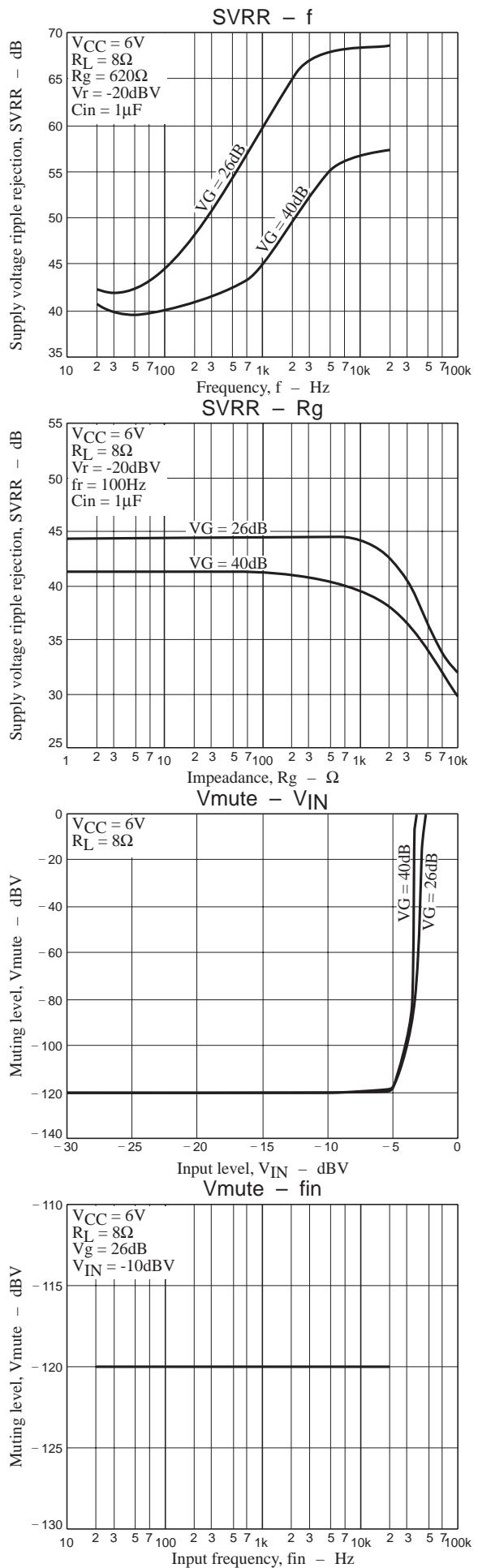
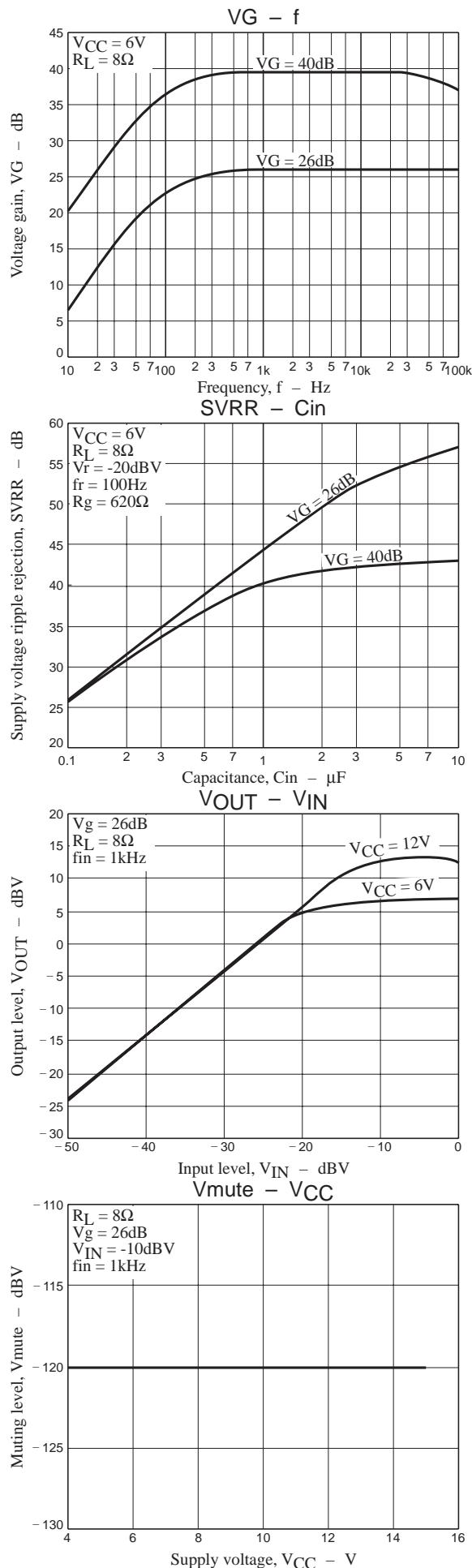
13. Maximum ratings

When used under conditions near the maximum ratings, even a slight fluctuation in the conditions may cause the maximum ratings to be exceeded, possibly resulting in a breakdown or other accidents. Therefore, always provide enough margin for fluctuations in the supply voltage and other conditions, and use within a range not exceeding the maximum ratings.

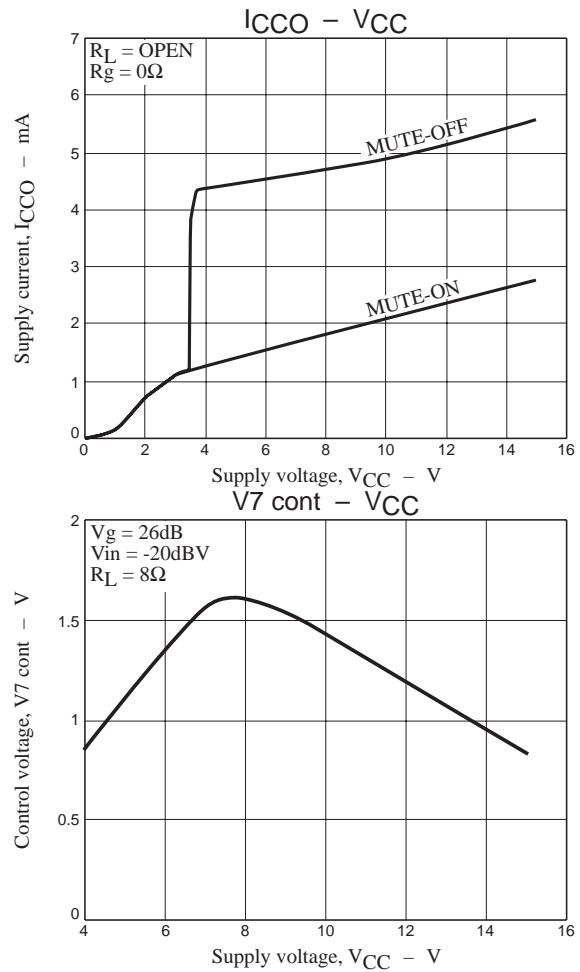
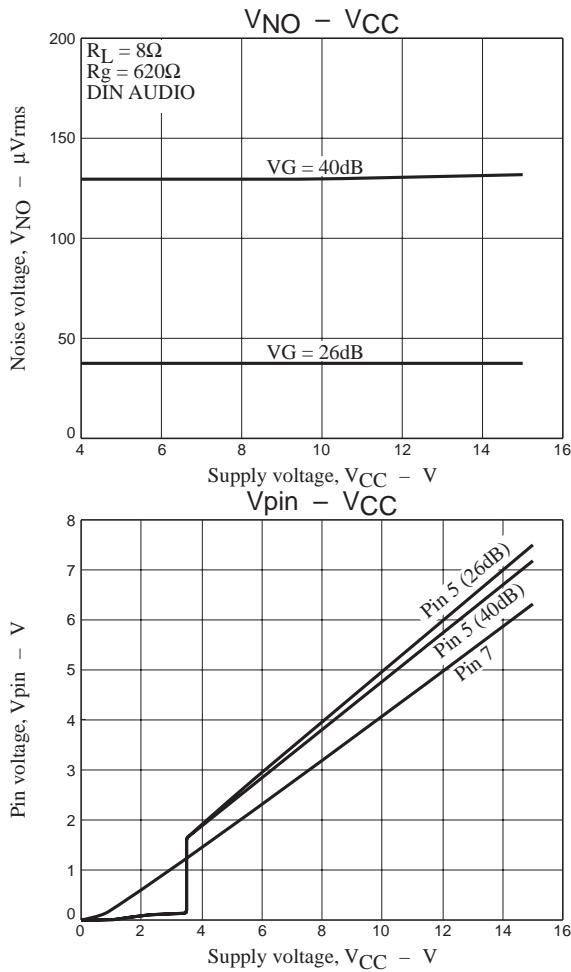
General characteristics (1)



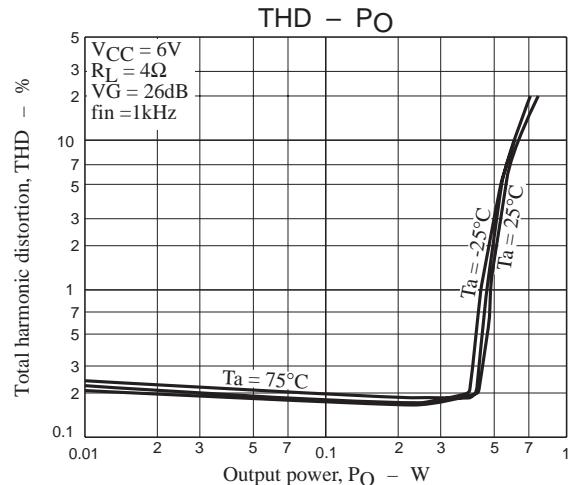
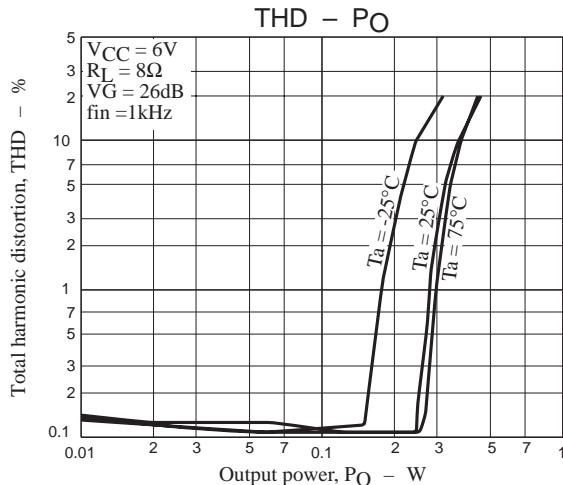
General characteristics (2)



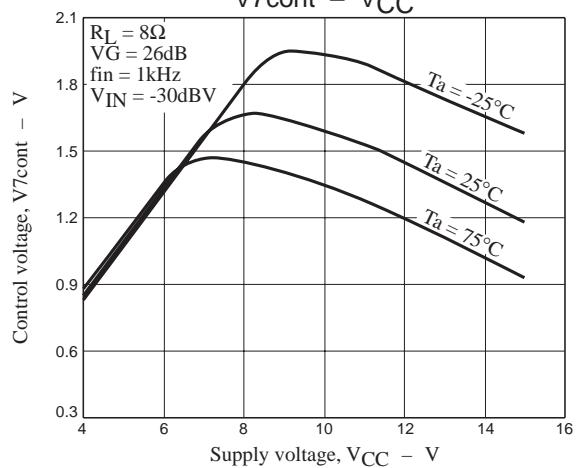
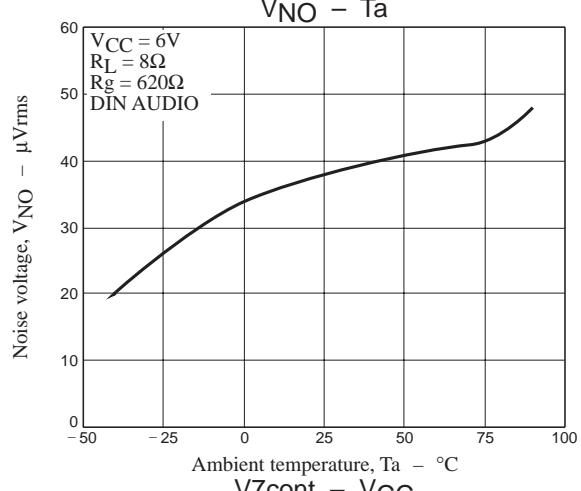
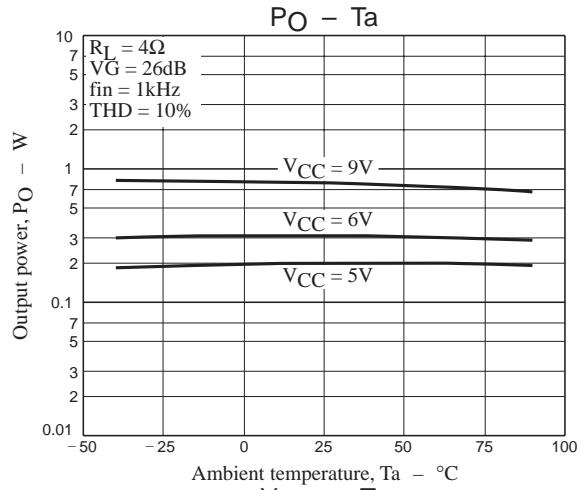
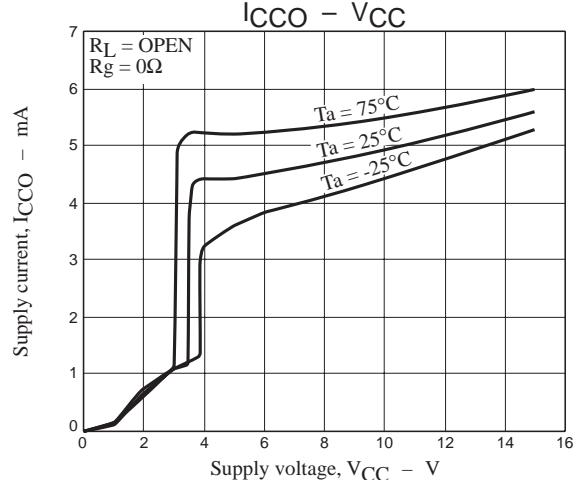
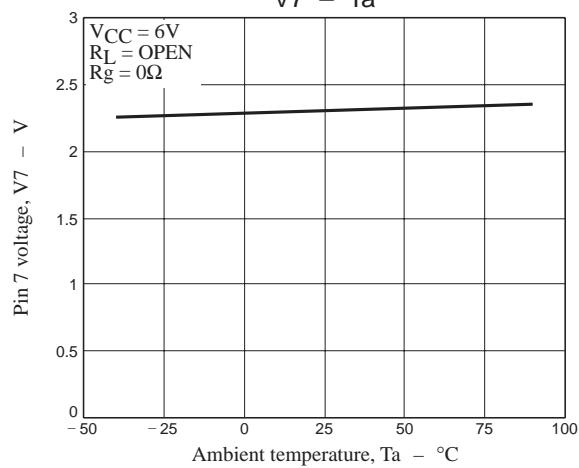
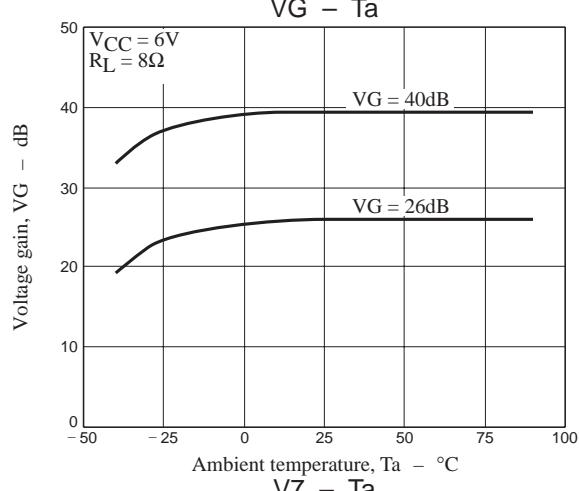
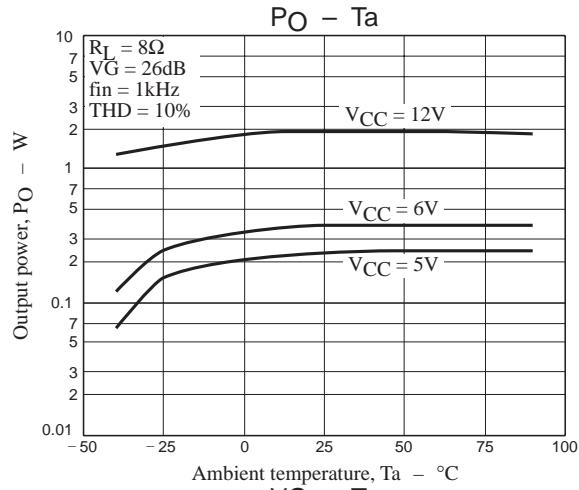
General characteristics (3)



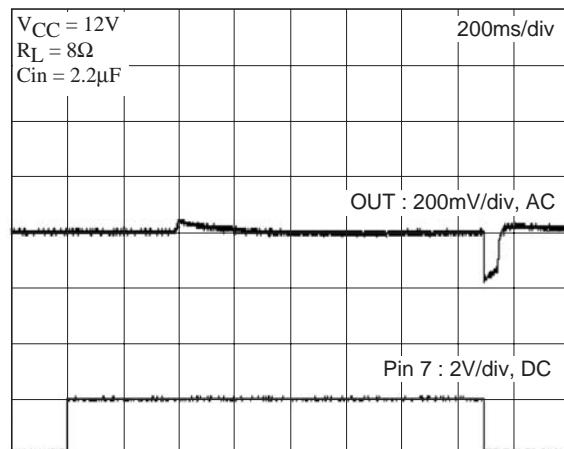
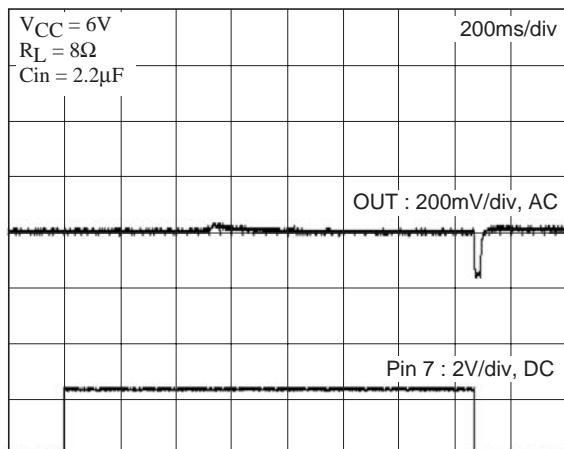
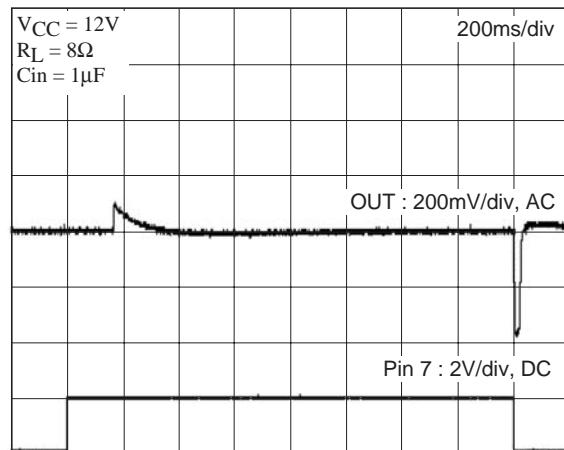
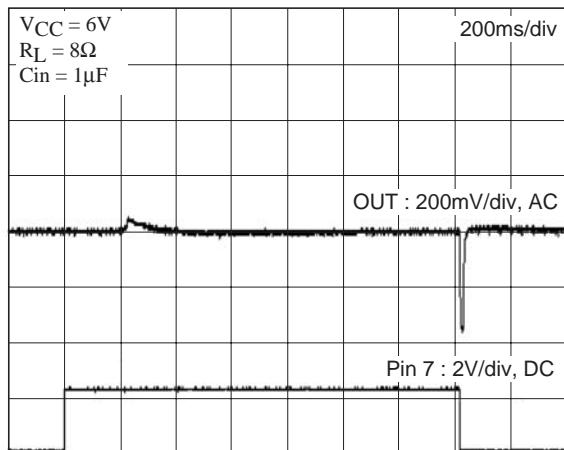
Temperature characteristics (1)



Temperature characteristics (2)



Muting on and off transient characteristics



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