



High-performance Regulator IC Series for PCs

Switching Regulators for DDR-SDRAM Cores





BD9533EKN

Description

BD9533EKN is a switching regulator controller with high output current which can achieve low output voltage (0.7V to 2.0V) from a wide input voltage range (4.5V to 28V). High efficiency for the switching regulator can be realized by utilizing an external N-MOSFET power transistor. A new technology called H³RegTM is a Rohm proprietary control method to realize ultra high transient response against load change. SLLM (Simple Light Load Mode) technology is also integrated to improve efficiency in light load mode, providing high efficiency over a wide load range. For protection and ease of use, the soft start function, variable frequency function, short circuit protection function with timer latch, over voltage protection function and REF synchronous function are all built in. This switching regulator is specially designed for the DDR-SDRAM core.

Features

- 1) H³RegTM Switching Regulator Controller
- 2) Light Load Mode and Continuous Mode Changeable
- 3) Thermal Shut Down (TSD), Under Voltage LockOut (UVLO), Over Current Protection (OCP), Over Voltage Protection (OVP), Short Circuit Protection (SCP) built-in
- 4) Soft start function to minimize rush current during startup
- 5) Switching Frequency Variable (f=200KHz~600KHz)
- 6) HQFN20V Package

Applications

Laptop PC, Desktop PC, LCD-TV, Digital Components

•Maximum Absolute Ratings (Ta=25°C)

| Parameter | Symbol | Limit | Unit |
|------------------------------|---------------------|-------------------|------|
| Input Voltage 1 | VCC | 7 ^1 | V |
| Input Voltage 2 | VDD | 7 ~1 | V |
| Input Voltage 3 | VIN | 30 ~1 | V |
| BOOT Voltage | BOOT | 35 ~ | V |
| BOOT-SW Voltage | BOOT-SW | 7 ^1 | V |
| HG-SW Voltage | HG-SW | 7 ~1 | V |
| LG Voltage | LG | VDD | V |
| REF Voltage | REF | VCC | V |
| Output Voltage | VOUT/Is+/Is- | VCC | V |
| ILIM/SCP/SS/FS/SLLM Voltage | ILIM/SCP/SS/FS/SLLM | VCC | V |
| VREG Voltage | VREG | VCC | V |
| EN Input Voltage | EN | 7 ^1 | V |
| Power Dissipation 1 | Pd1 | 0.5 *2 | W |
| Power Dissipation 2 | Pd2 | 0.75 -3 | W |
| Power Dissipation 3 | Pd3 | 1.75 ^4 | W |
| Power Dissipation 4 | Pd4 | 2.00 5 | W |
| Operating Temperature Range | Topr | -10~+100 | Ĵ |
| Storage Temperature Range | Tstg | -55 ~ +150 | °C |
| Maximum Junction Temperature | Tjmax | +150 | °C |

Operating Conditions (Ta=25°C)

| Parameter | Symbol | MIN | MAX | Unit |
|------------------------|---------|------|-----|------|
| Input Voltage 1 | VCC | 4.5 | 5.5 | V |
| Input Voltage 2 | VDD | 4.5 | 5.5 | V |
| Input Voltage 3 | VIN | 4.5 | 28 | V |
| BOOT Voltage | BOOT | 4.5 | 33 | V |
| SW Voltage | SW | -0.7 | 28 | V |
| BOOT-SW Voltage | BOOT-SW | 4.5 | 5.5 | V |
| SLLM Input Voltage | SLLM | 0 | 5.5 | V |
| EN Input Voltage | EN | 0 | 5.5 | V |
| Output setting voltage | REF | 0.7 | 2.0 | V |
| Is Input Voltage | ls+/ls- | 0.7 | 2.7 | V |
| MIN ON Time | Tonmin | - | 200 | nsec |

 $[\]bigstar$ This product should not be used in a radioactive environment.

^{*1} Not to exceed Pd.

*2 Reduced by 4mW for each increase in Ta of 1°C over 25°C (when don't mounted on a heat radiation board)

*3 Reduced by 6mW for increase in Ta of 1°C over 25°C. (when mounted on a board 70.0mm×70mm×1.6mm Glass-epoxy PCB.)

*4 Reduced by 14mW for increase in Ta of 1°C over 25°C. (when mounted on a board 70.0mm×70mm×1.6mm Glass-epoxy PCB.)

*5 Reduced by 16mW for increase in Ta of 1°C over 25°C. (when mounted on a board 70.0mm×70mm×1.6mm Glass-epoxy PCB.)

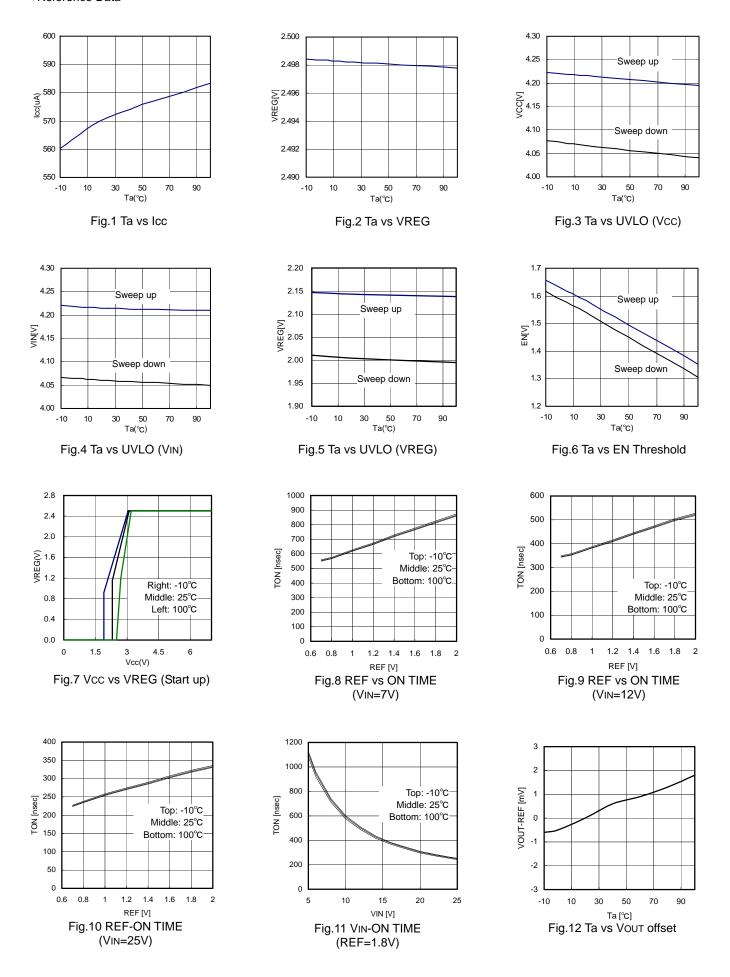
•ELECTRICAL CHARACTERISTICS

 $(unless\ otherwise\ noted,\ Ta=25^{\circ}C\ \ VCC=5V, VDD=5V, EN/SLLM=5V, VIN=12V, REF=1.8V, R_{FS}=68k\Omega)$

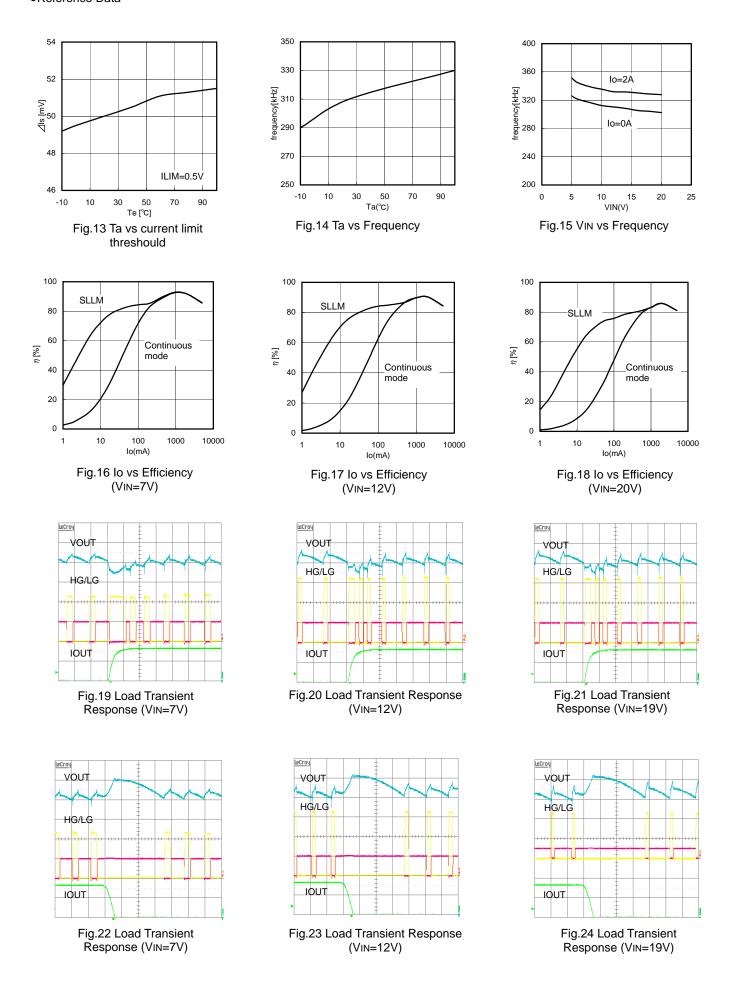
| Whole Device | less otherwise noted, Ta=25°C VC0 | | | andard Valu | | | T ['] | |
|---|-----------------------------------|------------|----------|-------------|----------|------|--------------------|--|
| VCC Bias Current Ioc 700 900 μA WIN Bias Current Iin - 100 200 μA NCC Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 200 μA EN=0V VIN Standby Current Iinst - 100 10 | Parameter | Symbol | | | | Unit | Condition | |
| VIN Bias Current | [Whole Device] | | | | | | | |
| Vic Standby Current Instb - 0 10 IA EN=OV ViN Standby Current Instb - 100 200 IA EN=OV EN Low Voltage Enlow GND - 0.8 V EN HighVoltage Enlow Enlow 2.3 - 5.5 V EN HighVoltage Enlow Enlow 2.3 - 5.5 V EN High Voltage Enlow Enlow 2.475 2.500 2.525 V Ireg=100µA Ta=-10°C to 100°C* T | | Icc | - | 700 | 900 | μΑ | | |
| VIN Standby Current Inistb - 100 200 μA EN=0V EN Low Voltage Enhigh S.3 - 5.5 V EN High Voltage Enhigh 2.3 - 5.5 V EN High Coltage Vreg 2.475 2.500 2.525 V Freg-10°C to 100°C- | | | - | | 200 | | | |
| Enlow Voltage | | | - | | | | | |
| EN HighVoltage Enhigh E.3 - 5.5 V Ireg=100μA VREG Voltage Vreg 2.475 2.500 2.525 V Ireg=100μA | | | - | 100 | 200 | μΑ | EN=0V | |
| RN Bias Current Ien | | | | - | | | | |
| VREG Voltage | EN HighVoltage | Enhigh | 2.3 | | | V | | |
| Vicage | EN Bias Current | len | - | 7 | 10 | μΑ | | |
| Linder Voltage Locked Out Voc. UVLO | V/PEC Voltage | Vrog | 2 475 | 2 500 | 2 525 | \/ | | |
| VCC threshold voltage Vcc, UVLO 4.1 4.3 4.5 V VCC:Sweep up VCC hysteresis voltage dVcc, UVLO 100 160 220 mV VCC:Sweep down VIN threshold voltage Vin, UVLO 4.1 4.3 4.5 V VIN:Sweep down VIN thysteresis voltage dVin, UVLO 100 160 220 mV VIN:Sweep down VREG threshold voltage Vreg UVLO 2.0 2.2 2.4 V VREG:Sweep up VREG thysteroid voltage dVreg, UVLO 100 160 220 mV VREG:Sweep up VREG thysteroid voltage dVreg, UVLO 100 160 220 mV VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep down VREG:Sweep up VREG:Sweep down VREG:Sweep down VREG:Sweep up VREG:Sweep down VREG:Sweep down VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep up VREG:Sweep down VREG:Sweep down VREG:Sweep down VREG:Sweep down VREG:Sweep down VREG:Sweep down VREG:Sweep up VREG:Sweep up VREG:Sweep down VREG:Sweep up VREG:Swee | | vieg | 2.475 | 2.500 | 2.525 | V | Ta=-10°C to 100°C* | |
| VCC bysteresis voltage Vin UVLO 100 160 220 mV VCC:Sweep down VIN threshold voltage Vin UVLO 4.1 4.3 4.5 V VIN:Sweep up VIN hysteresis voltage dVin UVLO 100 160 220 mV VIN:Sweep down VREG hysteresis voltage Vreg UVLO 2.0 2.2 2.4 V VREG:Sweep up VREG hysteresis voltage dVreg UVLO 100 160 220 mV VREG:Sweep up VREG hysteresis voltage dVreg UVLO 100 160 220 mV VREG:Sweep down H ² Reg M VR | | | | | | | 1 | |
| VIN hysteresis voltage | | | | | | | | |
| VREG threshold voltage Vreg_UVLO 100 160 220 mV VIN:Sweep down VREG threshold voltage Vreg_UVLO 2.0 2.2 2.4 V VREG:Sweep up VREG threshold voltage Vreg_UVLO 100 160 220 mV VREG:Sweep up VREG:Sweep down VREG:Son | | | | | | | | |
| VREG threshold voltage Vreg. UVLO 2.0 2.2 2.4 V VREG:Sweep up VREG hysteresis voltage dVreg. UVLO 100 160 220 mV VREG:Sweep down H*Reg!** | | | | | | - | VIN:Sweep up | |
| VREG hysteresis voltage dVreg UVLO 100 160 220 mV VREG:Sweep down H²Reg M² Fequency Fosc - 300 - kHz | | | | | | | VIN:Sweep down | |
| H*Reg M* | | | | | | | | |
| Frequency | VREG hysteresis voltage | dVreg_UVLO | 100 | 160 | 220 | mV | VREG:Sweep down | |
| ON Time | | _ | | | T | | 1 | |
| MAX ON Time | | | - | | - | | | |
| MIN OFF Time Toffmin - | | | 400 | | 600 | | | |
| FET Driver HG Higher side ON resistor | | | - | | - | µsec | | |
| HG Higher side ON resistor | | Toffmin | - | 450 | 550 | nsec | | |
| HG Lower side ON resistor | | | | | | | | |
| LG Higher side ON resistor LGhon - 2.0 4.0 Ω | | | - | | | | | |
| LG Lower side ON resistor LG Ion - 0.5 1.0 Ω | | | - | | | | | |
| Dead Time | | | - | | | | | |
| HG rising | | LGIon | - | 0.5 | 1.0 | Ω | | |
| LG rising LG dead - 50 - nsec | | | | | | | | |
| SCP | | | - | | - | nsec | | |
| SCP Detect Voltage Vscp | | LGdead | - | 50 | - | nsec | | |
| SCP threshold voltage | | | , | | | | | |
| Charge current | | | | | REF×0.75 | | | |
| Standby voltage Vscp_stb - - 50 mV | | | | 1.25 | 1.3 | V | | |
| OVP OVP Detect Voltage VovP REFx1.16 REFx1.2 REFx1.24 V | | | 1.5 | 2 | | | | |
| OVP Detect Voltage VOVP REFx1.16 REFx1.2 REFx1.24 V [Soft start] Charge current Iss 1.5 2 2.5 μA Standby voltage Vss_stb - - 50 mV [Over Current Protection Block] Current limit threshold1 Ilim1 43 50 57 mV ILIM=0.5V Ta=-10°C to 100°C* Current limit threshold2 Ilim2 160 200 240 mV ILIM=2.0V Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=0.5V Ta=-10°C to 100°C* Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V | | Vscp_stb | - | - | 50 | mV | | |
| Soft start Charge current | [OVP] | | , | | | | | |
| Charge current Iss 1.5 2 2.5 μA Standby voltage Vss_stb - - 50 mV [Over Current Protection Block] Current limit threshold1 Ilim1 43 50 57 mV ILIM=0.5V Ta=-10°C to 100°C* Current limit threshold2 Ilim2 160 200 240 mV ILIM=2.0V Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V [VOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA Is+ Input current I | OVP Detect Voltage | Vovp | REF×1.16 | REF×1.2 | REF×1.24 | V | | |
| Standby voltage Vss_stb - - 50 mV | | | | | | | | |
| Current Protection Block Current limit threshold1 | | | 1.5 | 2 | | | | |
| Current limit threshold1 Ilim1 43 50 57 mV ILIM=0.5V Ta=-10°C to 100°C* Current limit threshold2 Ilim2 160 200 240 mV ILIM=2.0V Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=2.0V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V [VOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V Continuous mode threshold Vthcon VCC-0.5 - VCC V | | Vss_stb | - | - | 50 | mV | | |
| Current limit threshold1 Ilim1 43 50 57 mV Ta=-10°C to 100°C* Current limit threshold2 Ilim2 160 200 240 mV ILIM=2.0V Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=0.5V IVOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is-=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V | [Over Current Protection Block] | - | 1 | | 1 | | | |
| Current limit threshold2 Ilim2 160 200 240 mV ILIM=2.0V Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V [VOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | Current limit threshold1 | llim1 | 43 | 50 | 57 | m\/ | | |
| Reverse current limit threshold1 Rellim1 - -50 - mV ILIM=0.5V Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V [VOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | | | | | | | | |
| Reverse current limit threshold2 Rellim2 - -200 - mV ILIM=2.0V [VOUT setting] VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | | | 160 | | 240 | | | |
| VOUT setting VOUT offset voltage1 | | | - | | - | | | |
| VOUT offset voltage1 Voutoff1 REF-7m REF REF+7m V Ta=-10°C to 100°C* VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | | Rellim2 | - | -200 | - | mV | ILIM=2.0V | |
| VOUT bias current Ivout -100 0 100 nA REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | | | | | T | | | |
| REF bias current Iref -100 0 100 nA Is+ Input current IIs+ -1 0 1 μA Is+=1.8V Is- Input current IIs- -1 0 1 μA Is-=1.8V [SLLM] Continuous mode threshold Vthcon VCC-0.5 - VCC V | | Voutoff1 | REF-7m | REF | REF+7m | V | Ta=-10°C to 100°C* | |
| $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$ | | | | 0 | | nΑ | | |
| $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$ | REF bias current | Iref | -100 | 0 | 100 | nA | | |
| Is- Input current Is- -1 0 1 μA Is-=1.8V | Is+ Input current | lls+ | | 0 | 1 | μΑ | Is+=1.8V | |
| Continuous mode threshold Vthcon VCC-0.5 - VCC V | Is- Input current | IIs- | -1 | 0 | 1 | | Is-=1.8V | |
| | | | | | | | | |
| | Continuous mode threshold | | VCC-0.5 | - | VCC | | | |
| | | VthSLLM | | - | 0.5 | V | | |

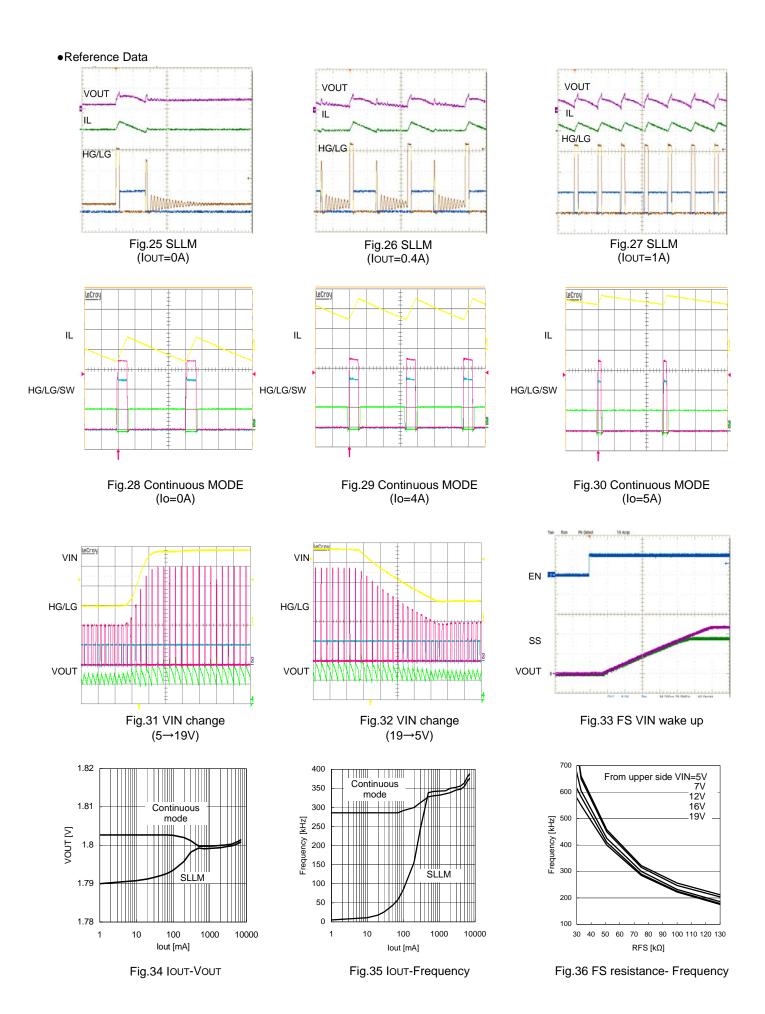
^{*} Design Guarantee

•Reference Data

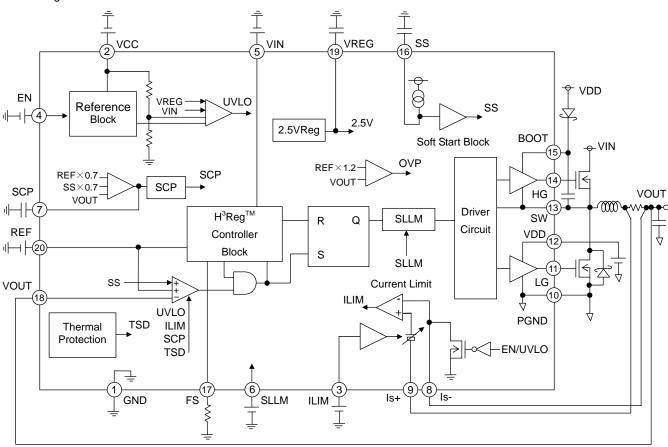


•Reference Data

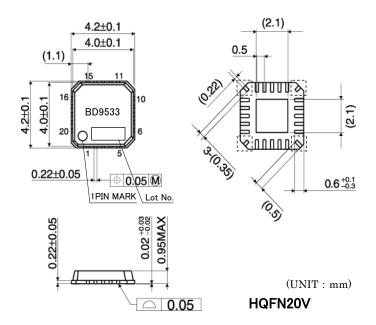




•Block Diagram



•PHYSICAL DIMENSIONS



※ Mounting is not recommended to the dotted line part.

◆Pin Number • Pin Name

| Pin No. | Pin Name | | |
|---|--------------------|--|--|
| 1 | GND | | |
| 2 | VCC | | |
| | VCC | | |
| 3 | ILIM | | |
| 4 | EN | | |
| 5 | VIN | | |
| 6 | SLLM | | |
| 7 | SCP | | |
| 1 2 3 4 5 6 7 8 9 | IS- IS+ PGND | | |
| 9 | IS+ | | |
| 10 | PGND | | |
| 11 | IG | | |
| 12 | VDD | | |
| 13 | SW | | |
| 14 | HG | | |
| 15 | BOOT | | |
| 15 16 | SS | | |
| 17 | FS | | |
| 18 | VOUT | | |
| 19 | VREG | | |
| 20 | REF | | |
| - | FIN | | |
| _ | 1 1111 | | |

Please short FIN to the 1PIN

Pin Descriptions

GND(1pin)

This is the ground pin for IC internal circuits. It is equivalent to FIN voltage.

VCC(2pin)

This is the power supply pin for IC internal circuits, except the FET driver. The maximum circuit current is 900uA. The input supply voltage range is 4.5V to 5.5V. It is recommended that a 0.1uF bypass capacitor be put in this pin.

ILIM(3pin)

BD9533EKN detects the voltage between Is+ pin and Is- pin and limits the output current (OCP). Voltage equivalent to 1/10 of the ILIM voltage is the voltage drop of external current sense resistor. A very low current sense resistor or inductor DCR can also be used for this platform.

EN(4pin)

When EN pin voltage is at least 2.3V, the status of this switching regulator becomes active. Conversely, the status switches off when EN pin voltage goes lower than 0.8V and circuit current becomes 10uA.

VIN(5pin)

The duty cycle is determined by input voltage and output voltage. In other words, the output voltage is affected by input voltage. Therefore, when VIN voltage fluctuates, the output voltage becomes also unstable. Since the VIN line is also the input voltage of the switching regulator, stability depends on the impedance of the voltage supply. It is recommended to establish a bypass capacitor or CR filter suitable for the actual application.

· SLLM(6pin)

This is the switch shift pin for Simple Light Load Mode. The efficiency in SLLM is improved when SLLM pin voltage goes lower than 0.5V.

· SCP(7pin)

This is the pin to adjust the timer latch time for short circuit protection. The timer circuit is active when the pin voltage becomes 70% of REF, and the output switches OFF and latched after the specified time. When the UVLO circuit is active or EN is low, this latch function is cancelled.

• Is-(8pin) ,Is+(9pin)

These pins are connected to both sides of the current sense resistor to detect output current. The voltage drop between ls+ and ls- is compared with the voltage equivalent to 1/10 of ILIM voltage. When this voltage drop reaches the specified voltage level, the output voltage goes OFF.

- PGND(10pin)

This is the power ground pin connected to the source of the low side FET.

LG(11pin)

This is the voltage supply to drive the Gate of the low side FET. This voltage swings between VDD and PGND. High-speed Gate driving for the low side FET is achieved due to the low on-resistance (2 ohm when LG is high, 0.5 ohm when LG is low) of the driver.

VDD(12pin)

This is the power supply pin to drive the LOW side FET. It is recommended that a 1uF bypass capacitor be established to compensate for rush current during the FET ON/OFF transition.

SW(13pin)

This is the source pin for the high side FET. The maximum absolute ratings are 30V (from GND). SW voltage swings between VIN and GND.

HG(14pin)

This is the voltage supply to drive the Gate of the high side FET. This voltage swings between BOOT and SW. High-speed Gate driving for the high side FET is achieved due to the low on-resistance (3 ohm when HG is high, 2 ohm when HG is low) of the driver.

• BOOT(15pin)

This is the voltage supply to drive the high side FET. The maximum absolute ratings are 35V (from GND) and 7V (from SW). BOOT voltage swings between VIN+Vcc and Vcc during active operation.

SS(16pin)

This is the adjustment pin to set the soft start time. SS voltage is low during standby status. When EN is ON, the soft start time can be determined by the SS charge current and capacitor between SS-GND. Until SS reaches REF voltage, the output voltage is equivalent to SS voltage.

FS(17pin)

This is the pin to adjust the switching frequency based on the resistor value. The frequency range is f=200KHz - 600KHz.

VOUT(18pin)

This is the output voltage sense pin. It is also possible to adjust the output voltage using external resistor divider based on the equation, REF≒VOUT.

· VREG(19pin)

This is the reference voltage output pin. The voltage is 2.5V, with 100uA current ability. It is recommended that a 1uF capacitor (X5R or X7R) be established between VREG and GND. When REF is not adjusted from the external voltage supply, the REF voltage can be adjusted using the external resistor divider of VREF.

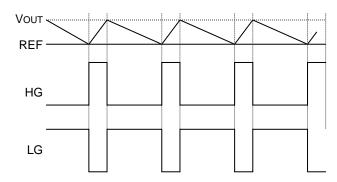
REF(20pin)

This is the output voltage adjustment pin. It is very convenient for synchronizing external voltage supply. The IC controls the output voltage (REF=VOUT).

Explanation of Operation

The BD9533EKN is a synchronous buck regulator controller incorporating ROHM's proprietary H³RegTM CONTROLLA control system. When Vout drops due to a rapid load change, the system quickly restores Vout by extending the Ton time interval. Thus, it serves to improve the regulator's transient response. Activating the Light Load Mode will also exercise Simple Light Load Mode (SLLM) control when the load is light, to further increase efficiency.

H³Reg[™] control (Normal operation)

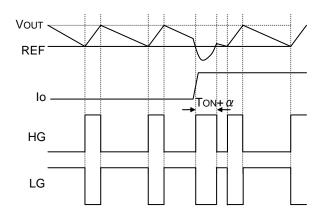


When Vout falls to a threshold voltage (REF), the drop is detected, activating the H³RegTM CONTROLLA system.

Ton=
$$\frac{\text{REF}}{\text{Vin}} \times \frac{1}{\text{f}} [\text{sec}] \cdot \cdot \cdot (1)$$

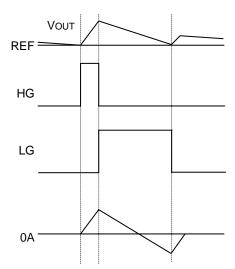
HG output is determined with the formula above. LG outputs until the status of VouT is lower than REF after the status of HG is off.

(VOUT drops due to a rapid load change)



When Vout drops due to a rapid load change, and the voltage remains below REF after the programmed Ton time interval has elapsed, the system quickly restores Vout by extending the Ton time, improving the transient response.

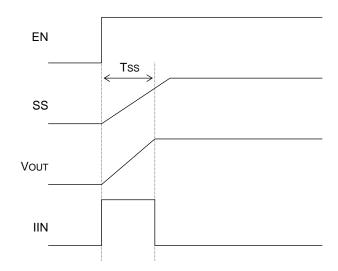
(SLLM)



In SLLM (SLLM=0V), SLLM function is operated when LG pin is OFF and the coil current is lower than 0A (the current goes from VouT to SW). And it stops to output next HG.

When Vout goes lower than REF voltage again, the status of HG is ON.

- Timing Chart
- Soft Start Function



Soft start is exercised with the EN pin set high. Current control takes effect at startup, enabling a moderate output voltage "ramping start." Soft start timing and incoming current are calculated with formulas (2) and (3) below.

Soft start time

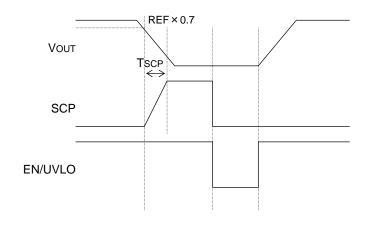
Tss=
$$\frac{\text{REF} \times \text{Css}}{2 \,\mu \, \text{A(typ)}}$$
 [sec] · · · (2)

Incoming current

IIN=
$$\frac{\text{Co} \times \text{Vout}}{\text{Tss}}$$
 [A] · · · (3)

(Css: Soft start capacitor; Co: Output capacitor)

• Timer Latch Type Short Circuit Protection

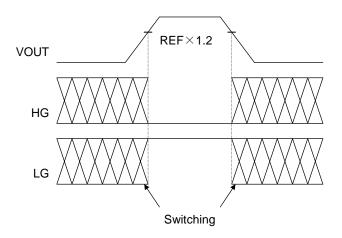


Short protection kicks in when output falls to or below REF X 0.7. When the programmed time period elapses, output is latched OFF to prevent destruction of the IC. Output voltage can be restored either by reconnecting the EN pin or disabling UVLO. Short circuit protection time is programmed using formula (4) below.

Short protection time setting

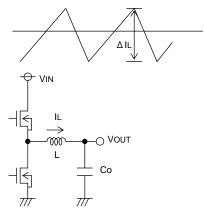
Tscp=
$$\frac{1.25(V) \times CscP}{2 \mu A(typ)}$$
 [sec] · · · (4)

· Over Voltage Protection



External Component Selection

1. Inductor (L) selection



Output ripple current

The inductor value is a major influence on the output ripple current. As formula (5) below indicates, the greater the inductor or the switching frequency, the lower the ripple current.

$$\Delta I = \frac{(V \text{In-Vout}) \times V \text{out}}{\Delta I \times V \text{In} \times f} [A] \cdot \cdot \cdot (5)$$

The proper output ripple current setting is about 30% of maximum output current.

$$\Delta \text{ IL=0.3 \times IouTmax. [A]} \cdot \cdot \cdot (6)$$

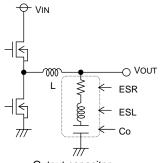
$$L = \frac{(\text{Vin-Vout}) \times \text{Vout}}{\Delta \text{ IL} \times \text{Vin} \times \text{f}} \quad [H] \cdot \cdot \cdot (7)$$

(Δ IL: output ripple current; f: switch frequency)

*Passing a current larger than the inductor's rated current will cause magnetic saturation in the inductor and decrease system efficiency. In selecting the inductor, be sure to allow enough margin to assure that peak current does not exceed the inductor rated current value.

*To minimize possible inductor damage and maximize efficiency, choose a inductor with a low (DCR, ACR) resistance.

2. Output Capacitor (Co) Selection



Output capacitor

When determining the proper output capacitor, be sure to factor in the equivalent series resistance required to smooth out ripple volume and maintain a stable output voltage range.

Output ripple voltage is determined as in formula (8) below.

$$\Delta VOUT = \Delta IL \times ESR + ESL \times \Delta IL/TON [V] \cdot \cdot \cdot (8)$$

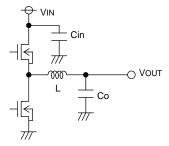
(Δ IL: Output ripple current; ESR: Co equivalent series resistance, ESR:equivalent series inductance)

Please give due consideration to the conditions in formula (9) below for output capacity, bearing in mind that output rise time must be established within the soft start time frame.

$$Co \le \frac{Tss \times (Limit-IOUT)}{VOLIT} \cdot \cdot \cdot (9)$$
 Tss: Soft start time Limit: Over current detection

Note: Improper capacitor may cause startup malfunctions.

3. Input Capacitor (Cin) Selection



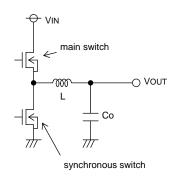
Input Capacitor

The input capacitor selected must have low enough ESR resistance to fully support large ripple output, in order to prevent extreme over current. The formula for ripple current IRMS is given in (10) below.

IRMS=IOUT ×
$$\frac{\sqrt{\text{VOUT}(\text{VIN-VOUT})}}{\text{VIN}}$$
 [A] · · · (10)
Where Vcc=2 × Vout, IRMS= $\frac{\text{IOUT}}{2}$

A low ESR capacitor is recommended to reduce ESR loss and maximize efficiency.

4. MOSFET Selection



Loss on the main MOSFET

Pmain=Pron+Pgate+Ptran

$$= \frac{\text{Vout}}{\text{Vin}} \times \text{Ron} \times \text{Iout}^2 + \text{Ciss} \times \text{f} \times \text{Vdd} + \frac{\text{Vin}^2 \times \text{Crss} \times \text{Iout} \times \text{f}}{\text{Idrive}} \quad \cdot \quad \cdot \quad (11)$$

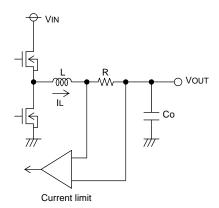
(Ron: On-resistance of FET; Ciss: FET gate capacity; f: Switching frequency Crss: FET inverse transfer function; I_{DRIVE} : Gate peak current)

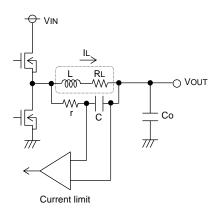
Loss on the synchronous MOSFET

Psyn=Pron+Pgate

$$= \frac{\text{Vin-Vout}}{\text{Vin}} \times \text{Ron} \times \text{Iout}^2 + \text{Ciss} \times \text{f} \times \text{Vdd} \cdot \cdot \cdot (12)$$

5. Setting Detection Resistance





The over current protection function detects the output ripple current peak value. This parameter (setting value) is determined as in formula (13) below.

ILMIT=
$$\frac{\text{VILIM} \times 0.1}{\text{R}}$$
 [A] · · · (13)

(VILIM: ILIM voltage; R: Detection resistance)

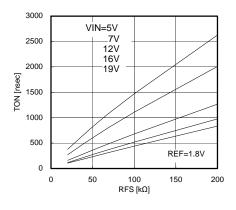
When the over current protection is detected by DCR of coil L, this parameter (setting value) is determined as in formula (14) below. (Application circuit:P20)

ILMIT=VILIM × 0.1 ×
$$\frac{r \times C}{L}$$
 [A] · · · (14)

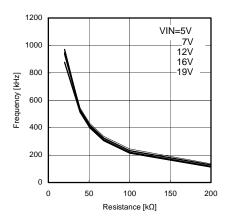
$$(RL = \frac{L}{r \times C})$$

(VILIM: ILIM voltage RL: the DCR value of coil)

6. Setting frequency



Calculated frequency using TON above



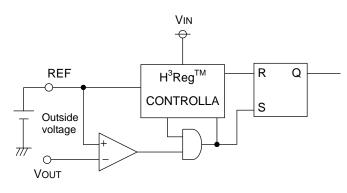
On Time (TON) at steady state is determined by resistance value connected to FS pin as shown in the figure. The frequency can be calculated with TON, input voltage VIN, the reference voltage REF as follows.

$$F = \frac{REF}{Vin \times Ton} \cdot \cdot \cdot (15)$$

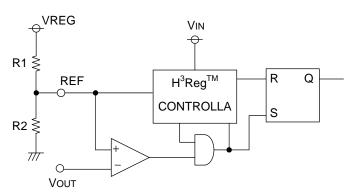
However, the actual frequency becomes lower than the formula due to expanded TON time caused by the rise/fall time of the MOSFET (This rise/fall time is affected by the switching speed). TON is also influenced by Dead Time under light load in continuous mode. In this case the frequency becomes lower than setting frequency.

The setting frequency noted above is under heavy load (when the inductor current does not go back to VIN through the body diodes of the external MOSFET's).

7. Setting standard voltage (REF)



It is available to synchronize setting the reference voltage (REF) with outside supply voltage [V] by using outside power supply voltage.



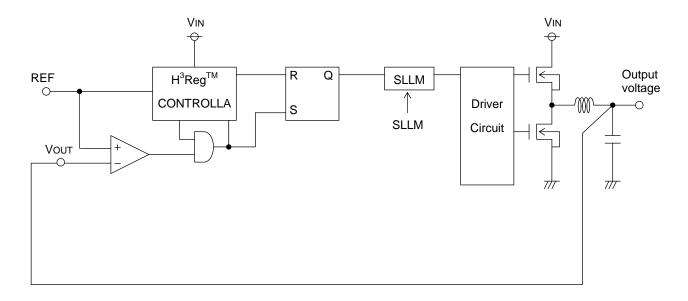
It is available to set the reference voltage (REF) by the resistance division value from VREG in case it is not set REF from an external power supply.

$$REF = \frac{R2}{R1+R2} \times VREG [V] \cdot \cdot \cdot (16)$$

8. Setting output voltage

This IC is operated that output voltage is REF≒VOUT.

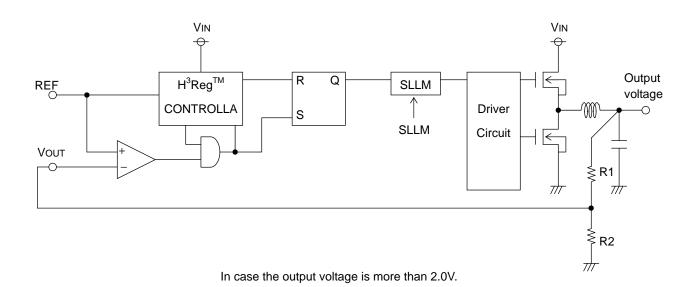
And it is operated that output voltage is feed back to FB pin in case the output voltage is 0.7V to 2.0V.



In case the output voltage range is 0.7V to 2.0V.

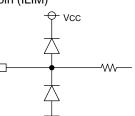
It is operated that the resistance division value of the output voltage is feed back to VouT pin in case the output voltage is more than 2.0V.

output voltage
$$=\frac{R1+R2}{R2}$$
 ×REF [V] · · · (17)

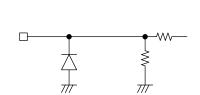


●I/O Equivalent Circuit

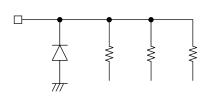




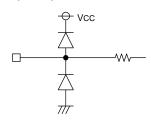




5pin (VIN)

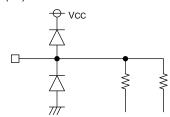


6pin (SLLM)

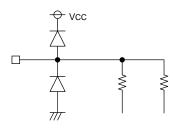


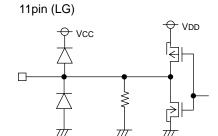
7pin (SCP)

8pin (Is-)



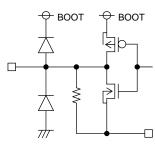
9pin (Is+)



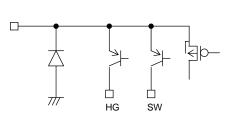


13pin (SW) ⋺ воот 🜡

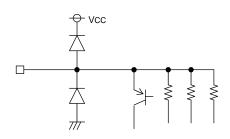
14pin(HG)



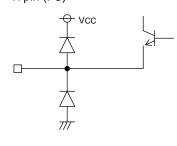
15pin (BOOT)



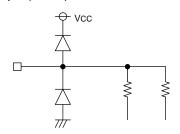
16pin (SS)



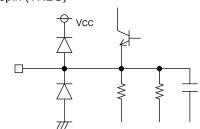
17pin (FS)



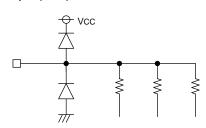
18pin (VOUT)

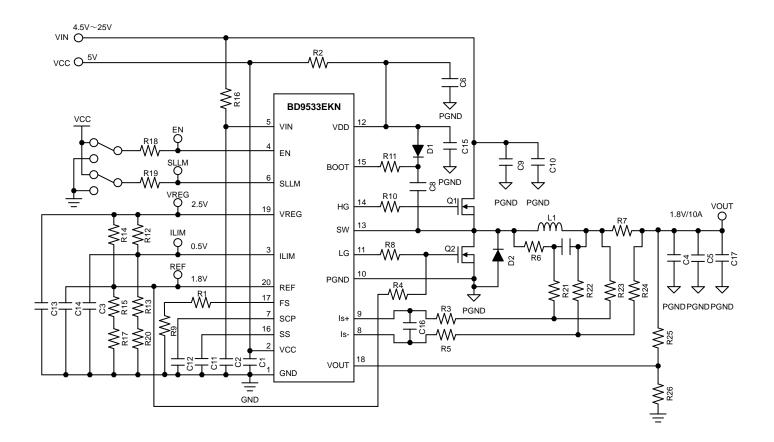


19pin (VREG)



20pin (REF)

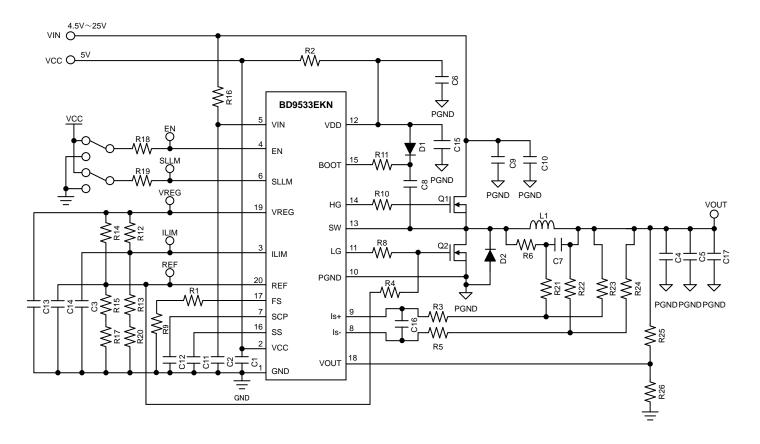




| Part No | Value | Company | Part name |
|------------|--------|---------|----------------|
| U1 | | ROHM | BD9533EKN |
| Q1 | | ROHM | RSS100N03 |
| Q2 | | ROHM | RSS100N03 |
| D1 | | ROHM | RB521S-30 |
| D2 | | ROHM | RB051L-40 |
| C1 | 1uF | KYOCERA | CM105B105K06A |
| C2 | 10nF | MURATA | GRM39X7R103K50 |
| C3 | 100pF | MURATA | GRM39C0G101J50 |
| C4 | | | |
| C5 | | | |
| C6 | 10uF | KYOCERA | CM21B106K06 |
| C7 | | | |
| C8 | 0.1uF | KYOCERA | CM05B104K25A |
| C9 | 10uF | KYOCERA | CM316B106M16A |
| C10 | 10uF | KYOCERA | CM316B106M16A |
| C11 | 1000pF | MURATA | GRM39X7R102K50 |
| C12 | 1500pF | MURATA | GRM39X7R152K50 |
| C13 | 1uF | KYOCERA | CM105B105K06A |
| C14 | | | |
| C15 | 10uF | KYOCERA | CM21B106M06A |
| C16 | | | |
| C17 | 470uF | SANYO | 2R5TPE470ML |
| R1 | 68ΚΩ | ROHM | MCR03 |
| R2 | 0Ω | ROHM | MCR03 |
| R3 | 0Ω | ROHM | MCR03 |

| Part No | Value | Company | Part name |
|------------|-------|---------|---------------|
| R4 | | | |
| R5 | 0Ω | ROHM | MCR03 |
| R6 | | | |
| R7 | 5mΩ | ROHM | PMR100 |
| R8 | 0Ω | ROHM | MCR03 |
| R9 | 0Ω | ROHM | MCR03 |
| R10 | 0Ω | ROHM | MCR03 |
| R11 | 0Ω | ROHM | MCR03 |
| R12 | 200kΩ | ROHM | MCR03 |
| R13 | 51kΩ | ROHM | MCR03 |
| R14 | 68kΩ | ROHM | MCR03 |
| R15 | 180kΩ | ROHM | MCR03 |
| R16 | 1kΩ | ROHM | MCR03 |
| R17 | 0Ω | ROHM | MCR03 |
| R18 | 10kΩ | ROHM | MCR03 |
| R19 | 10kΩ | ROHM | MCR03 |
| R20 | 0Ω | ROHM | MCR03 |
| R21 | | | |
| R22 | | | |
| R23 | 0Ω | ROHM | MCR03 |
| R24 | 0Ω | ROHM | MCR03 |
| R25 | 0Ω | ROHM | MCR03 |
| R26 | | | |
| L1 | 1.8uH | SUMIDA | CDEP104-1R8ML |
| | | | |

•Evaluation Board Circuit (Frequency=300kHz application circuit for detecting DCR current in Continuous mode/SLLM)



•Evaluation Board Parts List

| Part No | Value | Company | Part name |
|------------|--------|---------|----------------|
| U1 | | ROHM | BD9533EKN |
| Q1 | | ROHM | RSS100N03 |
| Q2 | | ROHM | RSS100N03 |
| D1 | | ROHM | RB521S-30 |
| D2 | | ROHM | RB051L-40 |
| C1 | 1uF | KYOCERA | CM105B105K06A |
| C2 | 10nF | MURATA | GRM39X7R103K50 |
| C3 | 100pF | MURATA | GRM39C0G101J50 |
| C4 | | | |
| C5 | | | |
| C6 | 10uF | KYOCERA | CM21B106K06 |
| C7 | 0.1uF | KYOCERA | CM05B104K25A |
| C8 | 0.1uF | KYOCERA | CM05B104K25A |
| C9 | 10uF | KYOCERA | CM316B106M16A |
| C10 | 10uF | KYOCERA | CM316B106M16A |
| C11 | 1000pF | MURATA | GRM39X7R102K50 |
| C12 | 1500pF | MURATA | GRM39X7R152K50 |
| C13 | 1uF | KYOCERA | CM105B105K06A |
| C14 | | | |
| C15 | 10uF | KYOCERA | CM21B106M06A |
| C16 | | | |
| C17 | 330uF | SANYO | 6TPB330M |
| R1 | 68ΚΩ | ROHM | MCR03 |
| R2 | 0Ω | ROHM | MCR03 |
| R3 | 0Ω | ROHM | MCR03 |

| Part No | Value | Company | Part name |
|------------|-------|------------------------|-----------|
| R4 | | | |
| R5 | 0Ω | ROHM | MCR03 |
| R6 | 1kΩ | ROHM | MCR03 |
| R7 | | | |
| R8 | 0Ω | ROHM | MCR03 |
| R9 | 0Ω | ROHM | MCR03 |
| R10 | Ω0 | ROHM | MCR03 |
| R11 | 0Ω | ROHM | MCR03 |
| R12 | 51kΩ | ROHM | MCR03 |
| R13 | 200kΩ | ROHM | MCR03 |
| R14 | 68kΩ | ROHM | MCR03 |
| R15 | 180kΩ | ROHM | MCR03 |
| R16 | 1kΩ | ROHM | MCR03 |
| R17 | Ω0 | ROHM | MCR03 |
| R18 | 10kΩ | ROHM | MCR03 |
| R19 | 10kΩ | ROHM | MCR03 |
| R20 | 0Ω | ROHM | MCR03 |
| R21 | 0Ω | ROHM | MCR03 |
| R22 | 0Ω | ROHM | MCR03 |
| R23 | | | |
| R24 | | | |
| R25 | 0Ω | ROHM | MCR03 |
| R26 | | | |
| L1 | 3.3uH | NEC/TOKIN MPLC0730L3R3 | |
| | | | |

Operation Notes

1. Absolute maximum ratings

An excess in the absolute maximum ratings, such as supply voltage, temperature range of operating conditions, etc., can break down the devices, thus making impossible to identify breaking mode, such as a short circuit or an open circuit. If any over rated values will expect to exceed the absolute maximum ratings, consider adding circuit protection devices, such as fuses.

2. Connecting the power supply connector backward

Connecting of the power supply in reverse polarity can damage IC. Take precautions when connecting the power supply lines. An external direction diode can be added.

3. Power supply lines

Design PCB layout pattern to provide low impedance GND and supply lines. To obtain a low noise ground and supply line, separate the ground section and supply lines of the digital and analog blocks. Furthermore, for all power supply terminals to ICs, connect a capacitor between the power supply and the GND terminal. When applying electrolytic capacitors in the circuit, not that capacitance characteristic values are reduced at low temperatures.

4. GND voltage

The potential of GND pin must be minimum potential in all operating conditions.

5. Thermal design

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) in actual operating conditions.

6. Inter-pin shorts and mounting errors

Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error or if pins are shorted together.

7. Actions in strong electromagnetic field

Use caution when using the IC in the presence of a strong electromagnetic field as doing so may cause the IC to malfunction.

8. ASO

When using the IC, set the output transistor so that it does not exceed absolute maximum ratings or ASO.

9. Thermal shutdown circuit

The IC incorporates a built-in thermal shutdown circuit (TSD circuit). The thermal shutdown circuit (TSD circuit) is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation. Do not continue to use the IC after operating this circuit or use the IC in an environment where the operation of this circuit is assumed.

| | TSD on temperature [°C] (typ.) | Hysteresis temperature [°C] (typ.) |
|-----------|--------------------------------|------------------------------------|
| BD9533EKN | 175 | 15 |

10. Testing on application boards

When testing the IC on an application board, connecting a capacitor to a pin with low impedance subjects the IC to stress. Always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

11. Regarding input pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes can occur inevitable in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Accordingly, methods by which parasitic diodes operate, such as applying a voltage that is lower than the GND (P substrate) voltage to an input pin, should not be used.

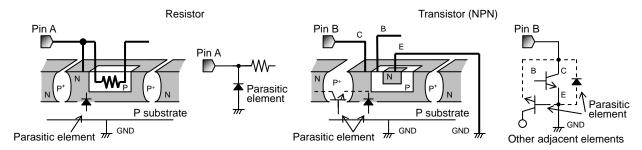
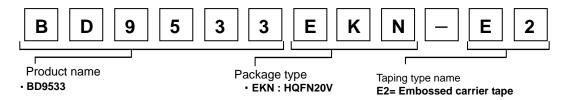


Fig. 37 Example of IC structure

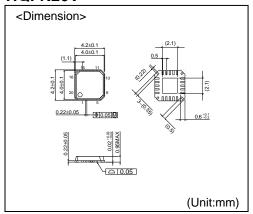
12. Ground Wiring Pattern

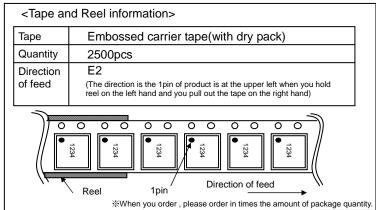
When using both small signal and large current GND patterns, it is recommended to isolate the two ground patterns, placing a single ground point at the ground potential of application so that the pattern wiring resistance and voltage variations caused by large currents do not cause variations in the small signal ground voltage. Be careful not to change the GND wiring pattern of any external components, either.

•Type Designations (Selections) for Ordering



HQFN20V





- The contents described herein are correct as of September, 2008
- The contents described herein are subject to change without notice. For updates of the latest information, please contact and confirm with ROHM CO. LTD
- Any part of this application note must not be duplicated or copied without our permission
- Application circuit diagrams and circuit constants contained herein are shown as examples of standard use and operation. Please pay careful attention to the peripheral conditions when designing circuits and deciding upon circuit constants in the set.
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- The products described herein utilize silicon as the main material.
 The products described herein are not designed to be X ray proof.

The products listed in this catalog are designed to be used with ordinary electronic equipment or devices (such as audio visual equipment, office-automation equipment, communications devices, electrical appliances and electronic toys).

Should you intend to use these products with equipment or devices which require an extremely high level of reliability and the malfunction of which would directly endanger human life (such as medical instruments, transportation equipment, aerospace machinery, nuclear-reactor controllers, fuel controllers and other safety devices), please be sure to consult with our sales representative in advance.

Excellence in Electronics



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21 Saiin Mizosaki-cho, Ukyo-ku, Kyoto 615-8585, Japan TEL: +81-75-311-2121 FAX: +81-75-315-0172 URL http://www.rohm.com

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TEL: +1-858-625-3630 FAX: +1-858-625-3670 TEL: +1-770-754-5972 FAX: +1-770-754-0691 FAX: +1-7/0-754-0691 FAX: +1-928-438-7164 FAX: +1-847-368-1008 FAX: +1-469-362-7973 FAX: +1-303-708-0858 FAX: +1-248-348-9942 FAX: +1-615-620-6702 TEL: +1-978-371-0382 TEL: +1-847-368-1006 TEL: +1-469-287-5366 TEL: +1-303-708-0908 TEL: +1-248-348-9920 TEL: +1-615-620-6700 TEL: +1-615-620-6700
TEL: +52-33-3123-2001
TEL: +49-2154-9210
TEL: +49-8999-216168
TEL: +49-711-7272-370
TEL: +33-1-5697-3060 FAX: +1-615-620-6702 FAX: +52-33-3123-2002 FAX: +49-2154-921400 FAX: +49-8999-216176 FAX: +49-711-7272-3720 FAX: +33-1-5697-3080 TEL: +44-1-908-306700 FAX: +44-1-908-235788 TFI: +45-3694-4739 FAX: +45-3694-4789 TEL: +45-3694-4739
TEL: +358-9725-54491
TEL: +358-2-7332234
TEL: +358-8-5372930
TEL: +34-9375-24320
TEL: +36-1-4719338
TEL: +48-22-5757213 FAX: +45-3694-4/89 FAX: +358-9-7255-4499 FAX: +358-2-7332237 FAX: +358-8-5372931 FAX: +34-9375-24410 FAX: +36-1-4719339 FAX: +48-22-5757001

Hangzhou Nanjing Ningbo Qingdao Suzhou Wuxi Shenzhen Dongguan Fuzhou Guangzhou Huizhou Xiamen Zhuhai Hong Kong Taipei Kaohsiung Singapore Philippines Thailand Triana... Kuala Lumpur Penang Kyoto Yokohama

Tianiin

TEL: +86-22-23029181 FAX: +86-22-23029183 TEL: +86-22-29029181 FAX: +86-22-29029183 FEL: +86-51-8679-2727 FAX: +86-51-6247-2066 TEL: +86-57-87658072 FAX: +86-57-87658071 TEL: +86-57-887658071 FAX: +86-57-87658071 TEL: +86-532-5779-312 FAX: +86-532-5779-653 TEL: +86-512-8807-1300 FAX: +86-512-8807-2300 FAX: +86-512-8 TEL: +86-510-82702693 FAX: +86-510-82702992 FEL: +86-55-8307-3003 FAX: +86-55-8307-3003 TEL: +86-769-8393-3320 FAX: +86-769-8398-4140 TEL: +86-91-8801-8698 FAX: +86-51-8801-8690 FAX: +86-23-825-5965 TEL: +86-752-205-1054 FAX: +86-752-205-1059 TEL: +86-592-238-5705 FAX: +86-592-239-8380 TEL: +86-756-3232-480 FAX: +86-756-3232-460 TEL: +86-73-3232-460
TEL: +852-2-740-6262
TEL: +886-2-2500-6956
TEL: +886-7-237-0881
TEL: +65-6332-2322 FAX: +86-7-238-7332 FAX: +852-2-375-8971 FAX: +886-2-2503-2869 FAX: +886-7-238-7332 FAX: +65-6332-5662 TEL: +63-2-807-6872 FAX: +63-2-809-1422 TEL: +66-2-254-4890 FAX: +66-2-256-6334 TEL: +60-3-7958-8355 TEL: +60-4-2286453 TEL: +81-75-365-1218 TEL: +81-45-476-2290 FAX: +60-3-7958-8377 FAX: +60-4-2286452 FAX: +81-75-365-1228 FAX: +81-45-476-2295

Notice

Precaution on using ROHM Products

Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

(Note1) Medical Equipment Classification of the Specific Applications

| JÁPAN | USA | EU | CHINA |
|---------|---------|------------|----------|
| CLASSI | СГУССШ | CLASS II b | СГУССШ |
| CLASSIV | CLASSII | CLASSIII | — CLASSⅢ |

- 2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:
 - [a] Installation of protection circuits or other protective devices to improve system safety
 - [b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure
- 3. Our Products are designed and manufactured for use under standard conditions and not under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
 - [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
 - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
 - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

Precautions Regarding Application Examples and External Circuits

- If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
- You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
 - [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
 - [b] the temperature or humidity exceeds those recommended by ROHM
 - the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label

QR code printed on ROHM Products label is for ROHM's internal use only.

Precaution for Disposition

When disposing Products please dispose them properly using an authorized industry waste company.

Precaution for Foreign Exchange and Foreign Trade act

Since our Products might fall under controlled goods prescribed by the applicable foreign exchange and foreign trade act, please consult with ROHM representative in case of export.

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