



Dual Low Power Operational Amplifier, Single or Dual Supply

OP-221

FEATURES

- Excellent TCV_{OS} Match $2\mu V/^{\circ}C$ Max
- Low Input Offset Voltage $150\mu V$ Max
- Low Supply Current $550\mu A$ Max
- Single Supply Operation $+5V$ to $+30V$
- Low Input Offset Voltage Drift $0.75\mu V/^{\circ}C$
- High Open-Loop Gain $1500V/mV$ Min
- High PSRR $3\mu V/V$
- Wide Common-Mode Voltage Range V^- to within $1.5V$ of V^+
- Pin Compatible with 1458, LM158, LM2904
- Available in Die Form

ORDERING INFORMATION[†]

$T_A = +25^{\circ}C$ V_{OS} MAX (μV)	PACKAGE			OPERATING TEMPERATURE RANGE
	TO-99	CERDIP 8-PIN	PLASTIC 8-PIN	
150	OP-221AJ/883	OP221AZ*	—	MIL
150	—	OP221EZ	—	IND
300	OP221BJ	—	—	MIL
500	OP221CJ	—	—	MIL
500	OP221GJ	OP221GZ	OP221GP	XIND
500	—	—	OP221GS	XIND

* For devices processed in total compliance to MIL-STD-883, add /883 after part number. Consult factory for 883 data sheet.

† Burn-in is available on commercial and industrial temperature range parts in CerDIP, plastic DIP, and TO-can packages.

GENERAL DESCRIPTION

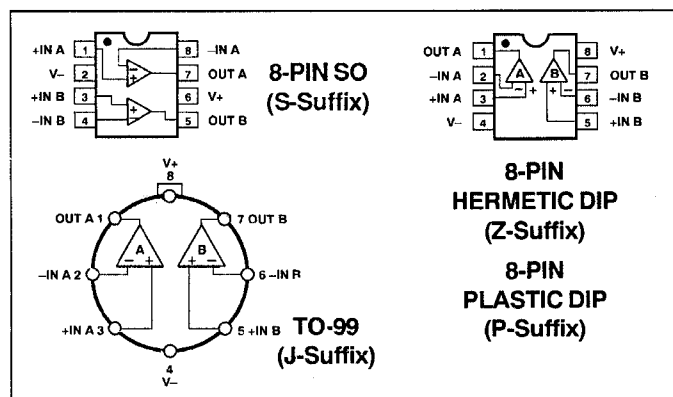
The OP-221 is a monolithic dual operational amplifier that can be used either in single or dual supply operation. The

wide supply voltage range, wide input voltage range, and low supply current drain of the OP-221 make it well-suited for operation from batteries or unregulated power supplies.

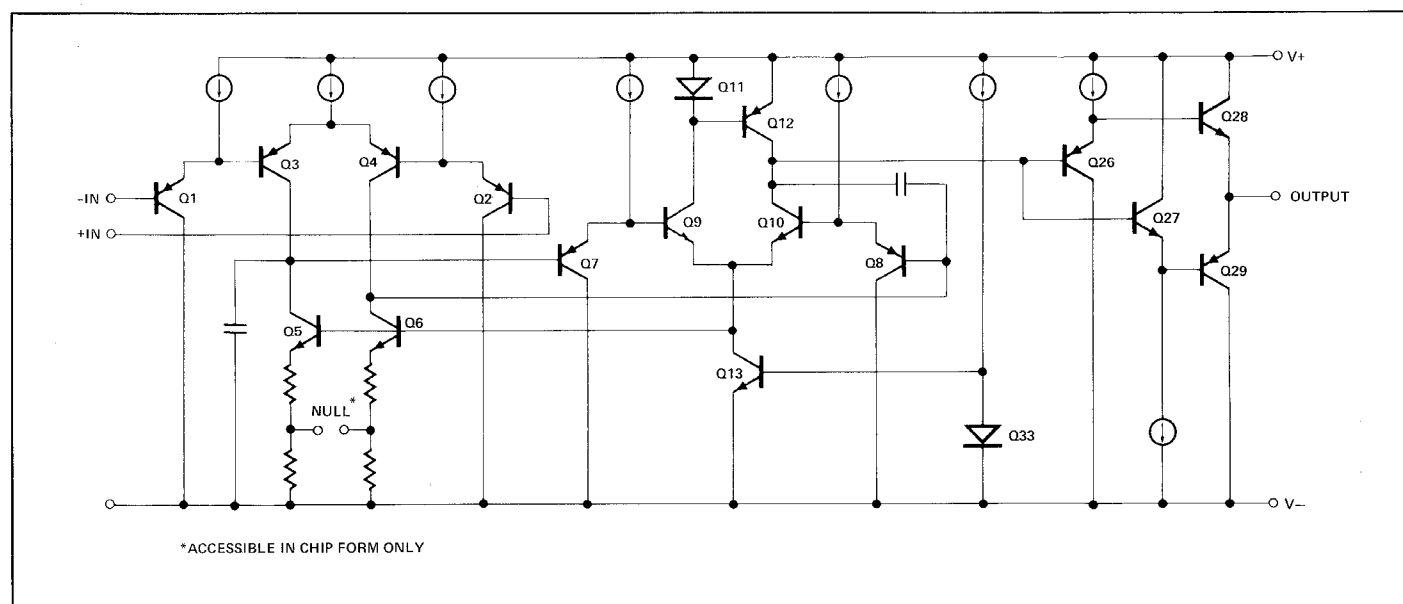
The excellent specifications of the individual amplifiers combined with the tight matching and temperature tracking between channels provide high performance in instrumentation amplifier designs. The individual amplifiers feature very low input offset voltage, low offset voltage drift, low noise voltage, and low bias current. They are fully compensated and protected.

Matching between channels is provided on all critical parameters including input offset voltage, tracking of offset voltage vs. temperature, non-inverting bias currents, and common-mode rejection.

PIN CONNECTIONS



SIMPLIFIED SCHEMATIC (Each Amplifier)



*ACCESSIBLE IN CHIP FORM ONLY

OP-221

ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±18V
Differential Input Voltage	30V or Supply Voltage
Input Voltage	Supply Voltage
Output Short-Circuit Duration	Indefinite
Storage Temperature Range	–65°C to +150°C
Operating Temperature Range	
OP-221A, B, C	–55°C to +125°C
OP-221E	–25°C to +85°C
OP-221G	–40°C to +85°C
Lead Temperature (Soldering, 60 sec)	+300°C
Junction Temperature (T_J)	–65°C to +150°C

PACKAGE TYPE	θ_{JA} (Note 2)	θ_{JC}	UNITS
TO-99 (J)	150	18	°C/W
8-Pin Hermetic DIP (Z)	148	16	°C/W
8-Pin Plastic DIP (P)	103	43	°C/W
8-Pin SO (S)	158	43	°C/W

NOTES:

1. Absolute maximum ratings apply to both DICE and packaged parts, unless otherwise noted.
2. θ_{JA} is specified for worst case mounting conditions, i.e., θ_{JA} is specified for device in socket for TO, CerDIP and P-DIP packages; θ_{JA} is specified for device soldered to printed circuit board for SO package.

ELECTRICAL CHARACTERISTICS at $V_S = \pm 2.5V$ to $\pm 15V$, $T_A = 25^\circ C$, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-221A/E			OP-221B			OP-221C/G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	V_{OS}		—	75	150	—	150	300	—	250	500	μV
Input Offset Current	I_{OS}	$V_{CM} = 0$	—	0.5	3	—	1	5	—	1.5	7	nA
Input Bias Current	I_B	$V_{CM} = 0$	—	50	80	—	60	100	—	70	120	nA
Input Voltage Range	IVR	$V_+ = 5V, V_- = 0V$ $V_S = \pm 15V$ (Note 2)	0/3.5 –15/13.5	—	—	0/3.5 –15/13.5	—	—	0/3.5 –15/13.5	—	—	V
Common-Mode Rejection Ratio	CMRR	$V_+ = 5V, V_- = 0V$ $0V \leq V_{CM} \leq 3.5V$ $V_S = \pm 15V$	90	100	—	85	90	—	75	85	—	dB
		$-15V \leq V_{CM} \leq 13.5V$	95	100	—	90	95	—	80	90	—	
Power Supply Rejection Ratio	PSRR	$V_S = \pm 2.5V$ to $\pm 15V$	—	3	10	—	10	32	—	32	100	$\mu V/V$
		$V_- = 0V, V_+ = 5V$ to 30V	—	6	18	—	18	57	—	57	180	
Large-Signal Voltage Gain	A_{VO}	$V_S = \pm 15V, R_L = 10k\Omega$ $V_O = \pm 10V$	1500	—	—	1000	—	—	800	—	—	V/mV
Output Voltage Swing	V_O	$V_+ = 5V, V_- = 0V,$ $R_L = 10k\Omega$	0.7/4.1	—	—	0.7/4.1	—	—	0.8/4	—	—	V
		$V_S = \pm 15V, R_L = 10k\Omega$	± 13.8	—	—	± 13.8	—	—	± 13.5	—	—	
Slew Rate	SR	$R_L = 10k\Omega$, (Note 1)	0.2	0.3	—	0.2	0.3	—	0.2	0.3	—	V/ μs
Bandwidth	BW		—	600	—	—	600	—	—	600	—	kHz
Supply Current (Both Amplifiers)	I_{SV}	$V_S = \pm 2.5V$, No Load	—	450	550	—	500	600	—	550	650	μA
		$V_S = \pm 15V$, No Load	—	600	800	—	800	850	—	850	900	

NOTES:

1. Sample tested.
2. Guaranteed by CMRR test limits.

ELECTRICAL CHARACTERISTICS at $V_S = \pm 2.5V$ to $\pm 15V$, $-55^\circ C \leq T_A \leq +125^\circ C$ for OP-221A, B, and C, $-25^\circ C \leq T_A \leq +85^\circ C$ for OP-221E, $-40^\circ C \leq T_A \leq +85^\circ C$ for OP-221G, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-221A/E			OP-221B			OP-221C/G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Average Input Offset Voltage Drift (Note 1)	TCV_{OS}		—	0.75	1.5	—	1.2	2	—	2	3	$\mu V/^\circ C$
Input Offset Voltage	V_{OS}		—	150	300	—	250	450	—	400	700	μV
Input Offset Current	I_{OS}	$V_{CM} = 0$	—	1	5	—	1.5	7	—	2	10	nA
Input Bias Current	I_B	$V_{CM} = 0$	—	55	100	—	65	120	—	80	140	nA
Input Voltage Range	IVR	$V_+ = 5V, V_- = 0V$ $V_S = \pm 15V$ (Note 2)	0/3.2 –15/13.2	—	—	0/3.2 –15/13.2	—	—	0/3.2 –15/13.2	—	—	V
Common-Mode Rejection Ratio	CMRR	$V_+ = 5V, V_- = 0V$ $0V \leq V_{CM} \leq 3.2V$ $V_S = \pm 15V$	85	90	—	80	85	—	70	80	—	dB
		$-15V \leq V_{CM} \leq 13.2V$	90	95	—	85	90	—	75	85	—	

ELECTRICAL CHARACTERISTICS at $V_S = \pm 2.5V$ to $\pm 15V$, $-55^\circ C \leq T_A \leq +125^\circ C$ for OP-221A, B, and C, $-25^\circ C \leq T_A \leq +85^\circ C$ for OP-221E, $-40^\circ C \leq T_A \leq +85^\circ C$ for OP-221G, unless otherwise noted. (Continued)

PARAMETER	SYMBOL	CONDITIONS	OP-221A/E			OP-221B			OP-221C/G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Power Supply Rejection Ratio	PSRR	$V_S = \pm 2.5V$ to $\pm 15V$	—	6	18	—	18	57	—	57	180	$\mu V/V$
		$V_- = 0V$, $V_+ = 5V$ to $30V$	—	10	32	—	32	100	—	100	320	
Large-Signal Voltage Gain	A_{VO}	$V_S = \pm 15V$, $R_L = 10k\Omega$ $V_O = \pm 10V$	1000	—	—	800	—	—	600	—	—	V/mV
Output Voltage Swing	V_O	$V_+ = 5V$, $V_- = 0V$, $R_L = 10k\Omega$	0.8/3.8	—	—	0.8/3.8	—	—	0.9/3.7	—	—	V
		$V_S = \pm 15V$, $R_L = 10k\Omega$	± 13.5	± 14	—	± 13.5	± 14	—	± 13.2	—	—	
Supply Current (Both Amplifiers)	I_{SY}	$V_S = \pm 2.5V$, No Load	—	500	650	—	550	700	—	600	750	μA
		$V_S = \pm 15V$, No Load	—	700	900	—	900	950	—	950	1000	

NOTES:

1. Sample tested.
2. Guaranteed by CMRR test limits.

MATCHING CHARACTERISTICS at $V_S = \pm 15V$, $T_A = 25^\circ C$, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-221A/E			OP-221B			OP-221C/G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage Match	ΔV_{OS}		—	50	200	—	150	400	—	250	600	μV
Average Noninverting Bias Current	I_{B^+}		—	—	80	—	—	100	—	—	120	nA
Noninverting Input Offset Current	I_{OS^+}		—	2	5	—	2	5	—	4	10	nA
Common-Mode Rejection Ratio Match (Note 1)	$\Delta CMRR$	$V_{CM} = -15V$ to $+13.5V$	92	—	—	87	—	—	72	—	—	dB
Power Supply Rejection Ratio Match (Note 2)	$\Delta PSRR$	$V_S = \pm 2.5V$ to $\pm 15V$	—	—	14	—	—	44	—	—	140	$\mu V/V$

MATCHING CHARACTERISTICS at $V_S = \pm 15V$, $-55^\circ C \leq T_A \leq +125^\circ C$ for OP-221A, B, and C, $-25^\circ C \leq T_A \leq +85^\circ C$ for OP-221E, $-40^\circ C \leq T_A \leq +85^\circ C$ for OP-221G, unless otherwise noted. Grades E and G are sample tested.

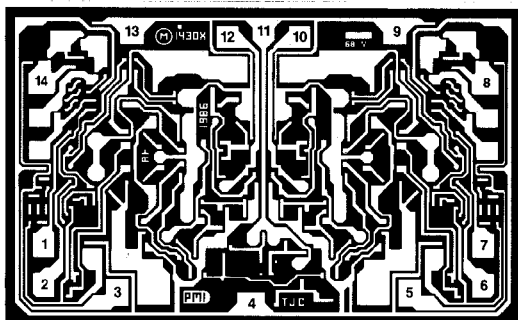
PARAMETER	SYMBOL	CONDITIONS	OP-221A/E			OP-221B			OP-221C/G			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage Match	ΔV_{OS}		—	100	400	—	250	600	—	400	800	μV
Average Noninverting Bias Current	I_{B^+}	$V_{CM} = 0$	—	—	100	—	—	120	—	—	140	nA
Input Offset Voltage Tracking	$TC\Delta V_{OS}$		—	1	2	—	1	3	—	3	5	$\mu V/^\circ C$
Noninverting Input Offset Current	I_{OS^+}	$V_{CM} = 0$	—	3	7	—	3	7	—	6	12	nA
Common-Mode Rejection Ratio Match (Note 1)	$\Delta CMRR$	$V_{CM} = -15V$ to $+13.2V$	87	90	—	82	85	—	72	80	—	dB
Power Supply Rejection Ratio Match (Note 2)	$\Delta PSRR$		—	—	26	—	—	78	—	—	250	$\mu V/V$

NOTES:

1. $\Delta CMRR$ is $20 \log_{10} V_{CM}/\Delta CME$, where V_{CM} is the voltage applied to both noninverting inputs and ΔCME is the difference in common-mode input-referred error.
2. $\Delta PSRR$ is: Input-Referred Differential Error
 ΔV_S

OP-221

DICE CHARACTERISTICS



DIE SIZE 0.097 × 0.063 inch, 6111 sq. mils
(2.464 × 1.600 mm, 3.94 sq. mm)

NOTE: All V+ PADS ARE INTERNALLY CONNECTED.

1. INVERTING INPUT (A)
2. NONINVERTING INPUT (A)
3. BALANCE (A)
4. V⁻
5. BALANCE (B)
6. NONINVERTING INPUT (B)
7. INVERTING INPUT (B)
8. BALANCE (B)
9. V⁺
10. OUT (B)
11. V⁺
12. OUT (A)
13. V⁺
14. BALANCE (A)

WAFER TEST LIMITS at $V_S = \pm 2.5V$ to $\pm 15V$, $T_A = 25^\circ C$, unless otherwise noted.

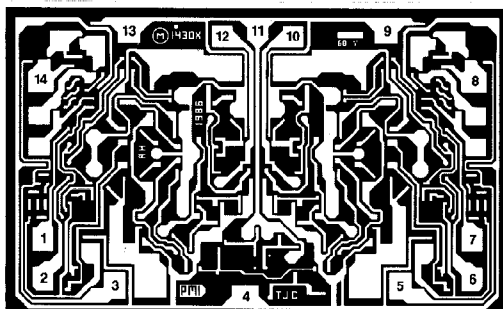
PARAMETER	SYMBOL	CONDITIONS	OP-221N LIMIT	OP-221G LIMIT	OP-221GR LIMIT	UNITS
Input Offset Voltage	V_{OS}		200	350	500	μV MAX
Input Offset Current	I_{OS}	$V_{CM} = 0$	3.5	5.5	7	nA MAX
Input Bias Current	I_B	$V_{CM} = 0$	85	105	120	nA MAX
Input Voltage Range	IVR	$V_+ = 5V, V_- = 0V$ $V_S = \pm 15V$	0/3.5 -15/13.5	0/3.5 -15/13.5	0/3.5 -15/13.5	V MIN/MAX V MIN
Common-Mode Rejection Ratio	CMRR	$V_- = 0V, V_+ = 5V, 0V \leq V_{CM} \leq 3.5V$ $V_S = \pm 15V, -15V \leq V_{CM} \leq 13.5V$	88 93	83 88	75 80	dB MIN
Power Supply Rejection Ratio	PSRR	$V_S = \pm 2.5V$ to $\pm 15V$ $V_- = 0V, V_+ = 5V$ to $30V$	12.5 22.5	40 70	100 180	$\mu V/V$ MAX
Large-Signal Voltage Gain	A_{VO}	$V_S = \pm 15V$ $R_L = 10k\Omega$	1500	1000	800	V/mV MIN
Output Voltage Swing	V_O	$V_+ = 5V, V_- = 0V, R_L = 10k\Omega$ $V_S = \pm 15V, R_L = 10k\Omega$	0.7/4.1 ± 13.8	0.7/4.1 ± 13.8	0.8/4 ± 13.5	V MIN/MAX V MIN
Supply Current (Both Amplifiers)	I_{SY}	$V_S = \pm 2.5V$, No Load $V_S = \pm 15V$, No Load	560 810	610 860	650 900	μA MAX

NOTE:

Electrical tests are performed at wafer probe to the limits shown. Due to variations in assembly methods and normal yield loss, yield after packaging is not guaranteed for standard product dice. Consult factory to negotiate specifications based on dice lot qualification through sample lot assembly and testing.

OP-221

DICE CHARACTERISTICS



DIE SIZE 0.097 × 0.063 inch, 6111 sq. mils
(2.464 × 1.600 mm, 3.94 sq. mm)

NOTE: All V+ PADS ARE INTERNALLY CONNECTED.

1. INVERTING INPUT (A)
2. NONINVERTING INPUT (A)
3. BALANCE (A)
4. V-
5. BALANCE (B)
6. NONINVERTING INPUT (B)
7. INVERTING INPUT (B)
8. BALANCE (B)
9. V+
10. OUT (B)
11. V+
12. OUT (A)
13. V+
14. BALANCE (A)

WAFER TEST LIMITS at $V_S = \pm 2.5V$ to $\pm 15V$, $T_A = 25^\circ C$, unless otherwise noted.

PARAMETER	SYMBOL	CONDITIONS	OP-221N LIMIT	OP-221G LIMIT	OP-221GR LIMIT	UNITS
Input Offset Voltage	V_{OS}		200	350	500	μV MAX
Input Offset Current	I_{OS}	$V_{CM} = 0$	3.5	5.5	7	nA MAX
Input Bias Current	I_B	$V_{CM} = 0$	85	105	120	nA MAX
Input Voltage Range	IVR	$V+ = 5V, V- = 0V$ $V_S = \pm 15V$	0/3.5 -15/13.5	0/3.5 -15/13.5	0/3.5 -15/13.5	V MIN/MAX V MIN
Common-Mode Rejection Ratio	CMRR	$V- = 0V, V+ = 5V, 0V \leq V_{CM} \leq 3.5V$ $V_S = \pm 15V, -15V \leq V_{CM} \leq 13.5V$	88 93	83 88	75 80	dB MIN
Power Supply Rejection Ratio	PSRR	$V_S = \pm 2.5V$ to $\pm 15V$ $V- = 0V, V+ = 5V$ to $30V$	12.5 22.5	40 70	100 180	$\mu V/V$ MAX
Large-Signal Voltage Gain	A_{VO}	$V_S = \pm 15V$ $R_L = 10k\Omega$	1500	1000	800	V/mV MIN
Output Voltage Swing	V_O	$V+ = 5V, V- = 0V, R_L = 10k\Omega$ $V_S = \pm 15V, R_L = 10k\Omega$	0.7/4.1 ± 13.8	0.7/4.1 ± 13.8	0.8/4 ± 13.5	V MIN/MAX V MIN
Supply Current (Both Amplifiers)	I_{SY}	$V_S = \pm 2.5V$, No Load $V_S = \pm 15V$, No Load	560 810	610 860	650 900	μA MAX

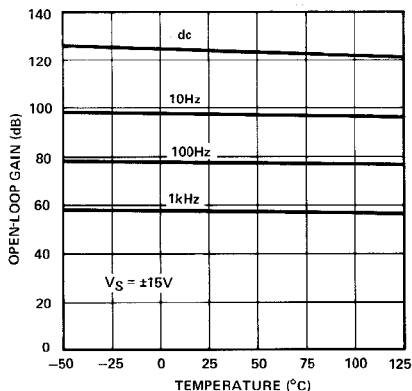
NOTE:

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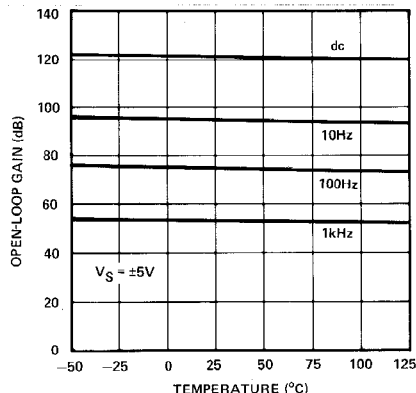
OP-221

TYPICAL PERFORMANCE CHARACTERISTICS

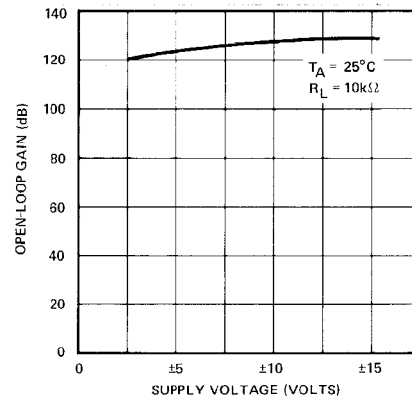
**OPEN-LOOP GAIN AT $\pm 15V$
vs TEMPERATURE**



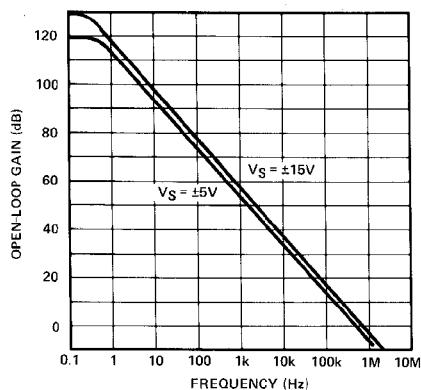
**OPEN-LOOP GAIN AT $\pm 5V$
vs TEMPERATURE**



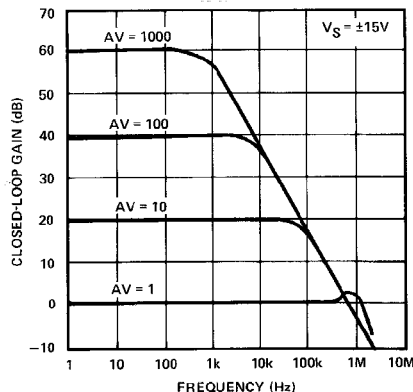
**OPEN-LOOP GAIN
vs SUPPLY VOLTAGE**



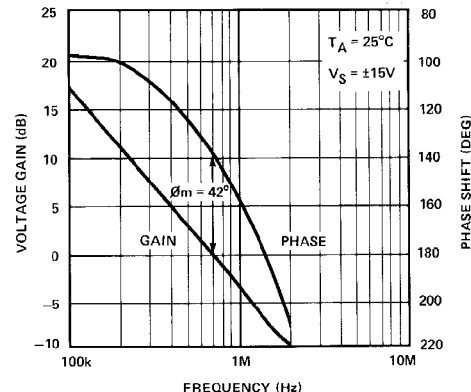
**OPEN-LOOP GAIN
vs FREQUENCY**



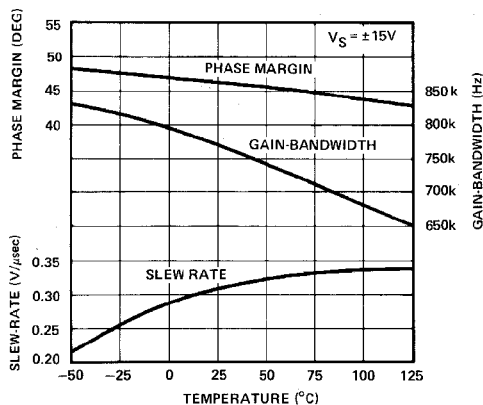
**CLOSED-LOOP GAIN
vs FREQUENCY**



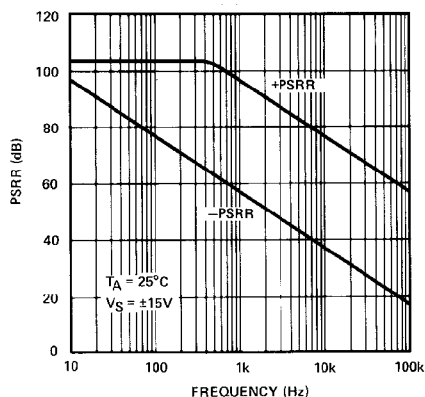
**GAIN AND PHASE SHIFT
vs FREQUENCY**



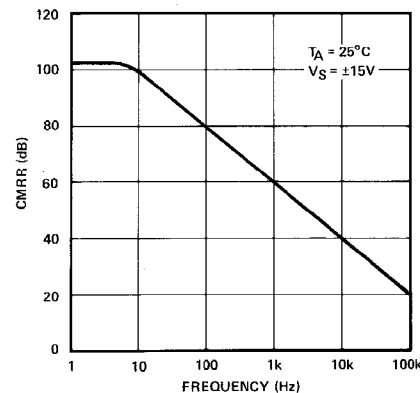
PHASE MARGIN, GAIN-BANDWIDTH, AND SLEW RATE vs TEMPERATURE



PSRR vs FREQUENCY

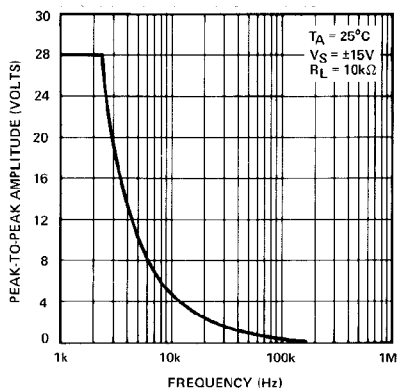


CMRR vs FREQUENCY

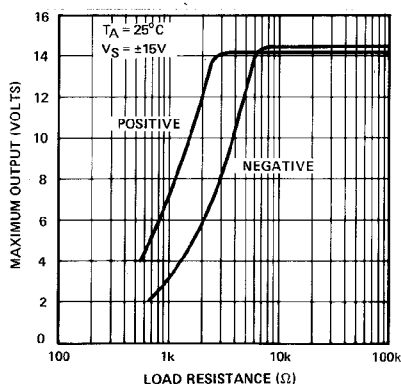


TYPICAL PERFORMANCE CHARACTERISTICS

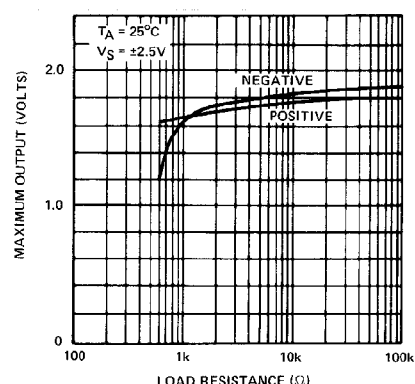
**MAXIMUM OUTPUT SWING
vs FREQUENCY**



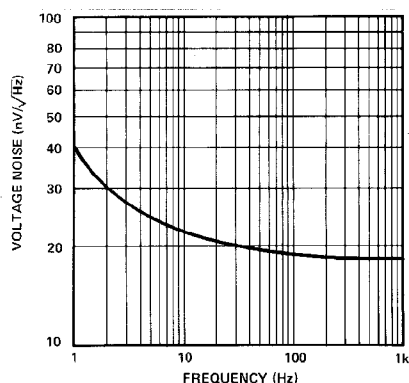
**MAXIMUM OUTPUT VOLTAGE
vs LOAD RESISTANCE**



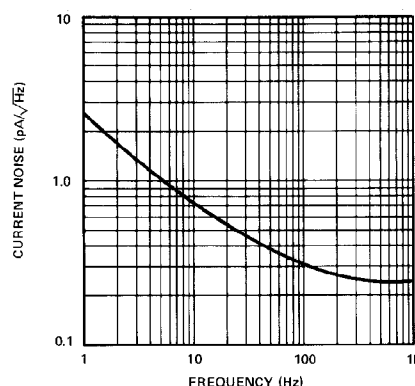
**MAXIMUM OUTPUT VOLTAGE
vs LOAD RESISTANCE**



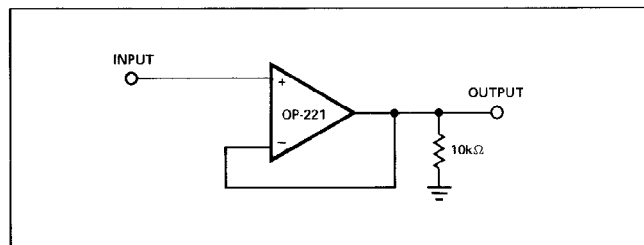
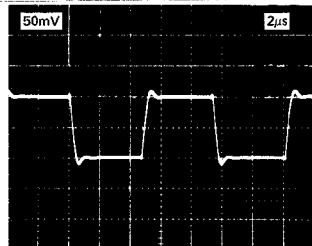
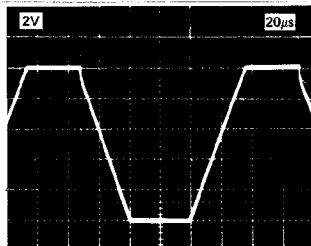
**VOLTAGE NOISE DENSITY (e_n)
vs FREQUENCY**



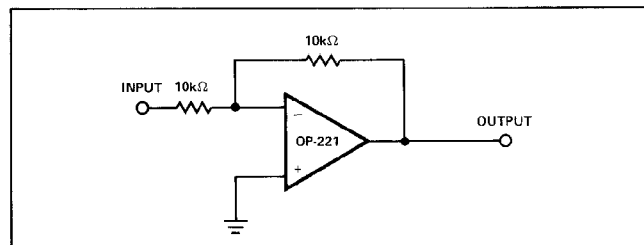
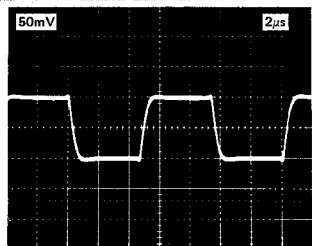
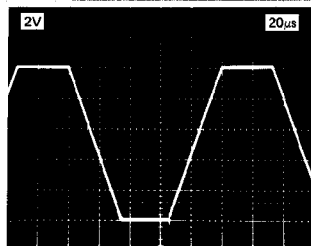
**CURRENT NOISE DENSITY (i_n)
vs FREQUENCY**



NONINVERTING STEP RESPONSE



INVERTING STEP RESPONSE



OP-221

SPECIAL NOTES ON THE APPLICATION OF DUAL MATCHED OPERATIONAL AMPLIFIERS

ADVANTAGES OF DUAL MONOLITHIC OPERATIONAL AMPLIFIERS

Dual matched operational amplifiers provide the engineer with a powerful tool for designing instrumentation amplifiers and many other differential-input circuits. These designs are based on the principle that careful matching between two operational amplifiers can minimize the effect of DC errors in the individual amplifiers.

Reference to the circuit shown in Figure 1, a differential-in, differential-out amplifier, shows how the reductions in error can be accomplished. Assuming the resistors used are ideally matched, the gain of each side will be identical. If the offset voltages of each amplifier are perfectly matched, then the net differential voltage at the amplifier's output will be zero. Note that the output offset error of this amplifier is not a function of the offset voltage of the individual amplifiers, but only a function of the *difference* (degree of matching) between the amplifiers' offset voltages. This error-cancellation principle holds for a considerable number of input referred error parameters — offset voltage, offset voltage drift, inverting and noninverting bias currents, common-mode and power supply rejection ratios. Note also that the impedances of each input, both common-mode and differential-mode, are high and tightly matched, an important feature not practical with single operational amplifier circuits.

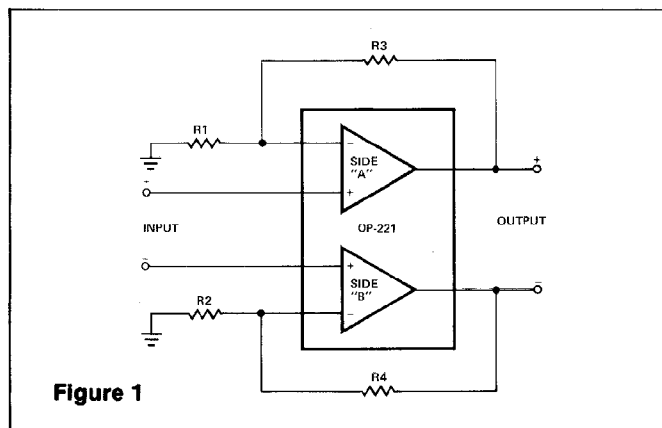


Figure 1

INSTRUMENTATION AMPLIFIER APPLICATIONS

Two-Op-Amp Configuration

The two-op-amp circuit (Figure 2), is recommended where the common-mode input voltage range is relatively limited; the common-mode and differential voltage both appear at V1.

The high open-loop gain of the OP-221 is very important in achieving good CMRR in this configuration. Finite open-loop gain of A1 (A_{01}) causes undesired feedthrough of the common-mode input. For $A_d/A_{01} \ll 1$, the common-mode error (CME) at the output due to this effect is approximately $(2 A_d/A_{01}) \times V_{CM}$. This circuit features independent adjustment of CMRR and differential gain.

Three-Op-Amp Configuration

The three-op-amp circuit (Figure 3), has increased common-mode voltage range because the common-mode voltage is not amplified as it is in Figure 2. The CMR of this amplifier is directly proportional to the match of the CMR of the input op amps. CMRR can be raised even further by trimming the output stage resistors.

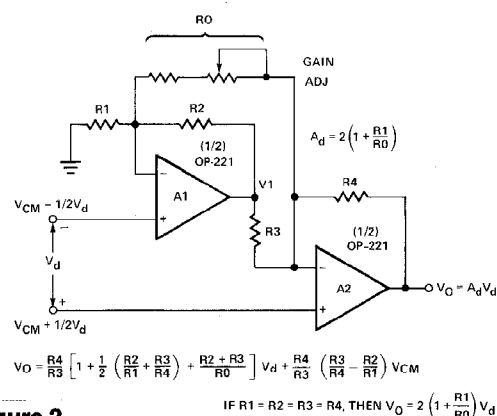


Figure 2

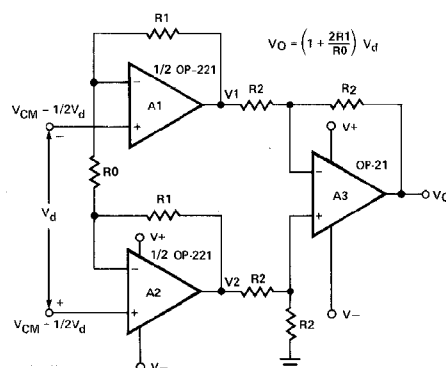


Figure 3