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FCBS0550

Smart Power Module (SPM)

Features

- UL Certified No.E209204(SPM27-BA package)
- 500V-5A 3-phase MOSFET inverter bridge including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- Single-grounded power supply due to built-in HVIC
- Isolation rating of 2500Vrms/min.
- Very low leakage current due to using ceramic substrate

Applications

- AC 200V three-phase inverter drive for small power ac motor drives
- Home appliances applications like refrigerator.

General Description

It is an advanced smart power module (SPM) that Fairchild has newly developed and designed to provide very compact and high performance ac motor drives mainly targeting low-power inverter-driven application like refrigerator. It combines optimized circuit protection and drive matched to low-loss MOSFETs. System reliability is further enhanced by the integrated under-voltage lock-out and short-circuit protection. The high speed built-in HVIC provides opto-coupler-less single-supply MOSFET gate driving capability that further reduce the overall size of the inverter system design. Each phase current of inverter can be monitored separately due to the divided negative dc terminals.

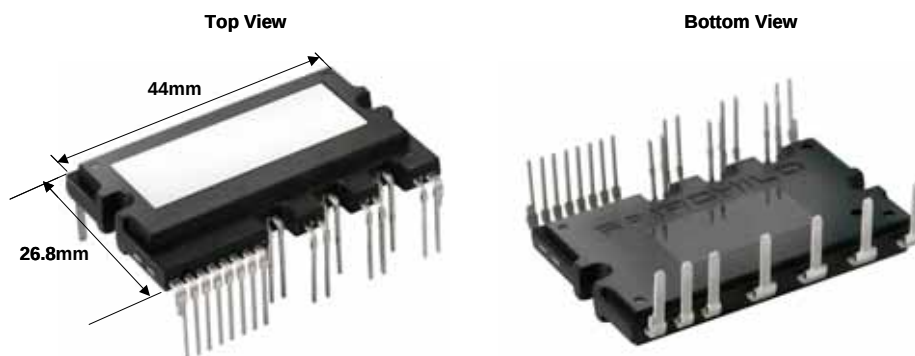


Figure 1.

Integrated Power Functions

- 500V-5A MOSFET inverter for three-phase DC/AC power conversion (Please refer to Fig. 3)

Integrated Drive, Protection and System Control Functions

- For inverter high-side MOSFETs: Gate drive circuit, High voltage isolated high-speed level shifting
Control circuit under-voltage (UV) protection
Note) Available bootstrap circuit example is given in Figs. 10 and 11.
- For inverter low-side MOSFETs: Gate drive circuit, Short circuit protection (SC)
Control supply circuit under-voltage (UV) protection
- Fault signaling: Corresponding to a UV fault (Low-side supply), SC fault
- Input interface: 3.3/5V CMOS/LSTTL compatible, Schmitt trigger input

Pin Configuration

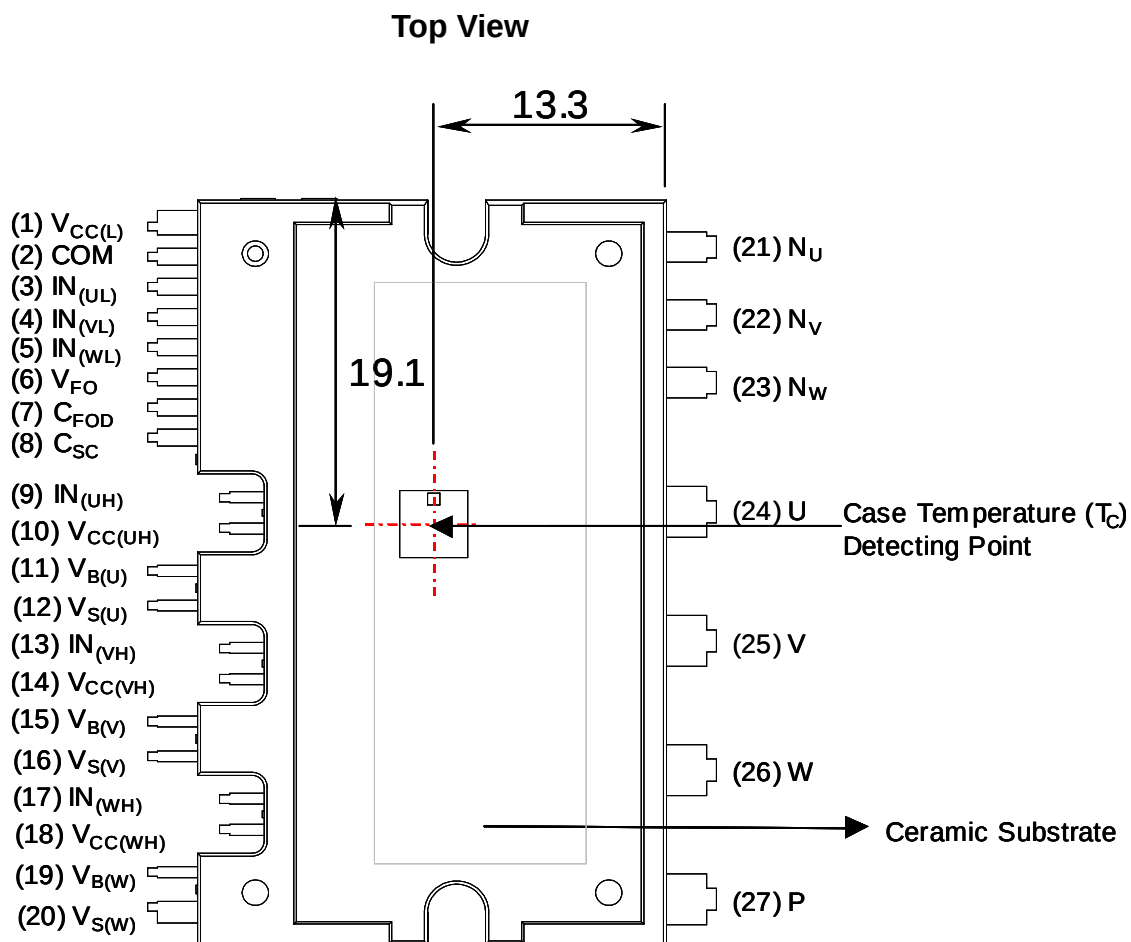
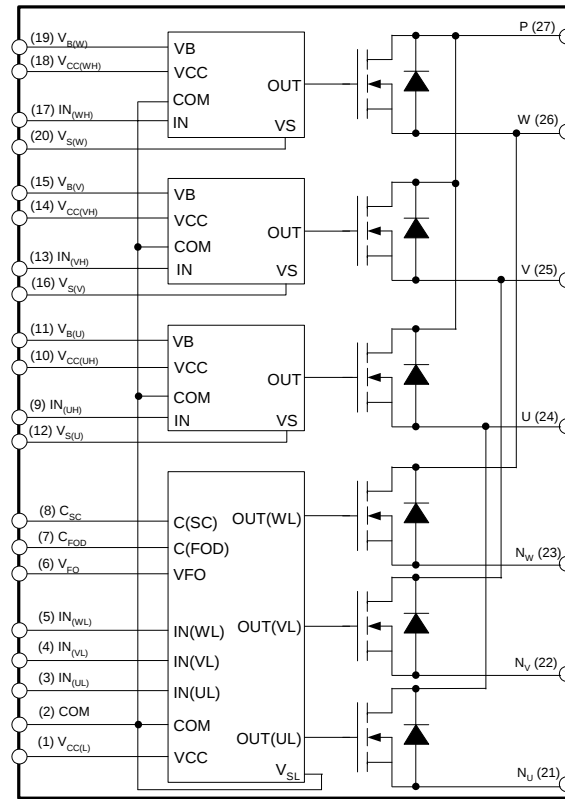


Figure 2.

Pin Descriptions

Pin Number	Pin Name	Pin Description
1	$V_{CC(L)}$	Low-side Common Bias Voltage for IC and MOSFETs Driving
2	COM	Common Supply Ground
3	$IN_{(UL)}$	Signal Input for Low-side U Phase
4	$IN_{(VL)}$	Signal Input for Low-side V Phase
5	$IN_{(WL)}$	Signal Input for Low-side W Phase
6	V_{FO}	Fault Output
7	C_{FOD}	Capacitor for Fault Output Duration Time Selection
8	C_{SC}	Capacitor (Low-pass Filter) for Short-Current Detection Input
9	$IN_{(UH)}$	Signal Input for High-side U Phase
10	$V_{CC(UH)}$	High-side Bias Voltage for U Phase IC
11	$V_{B(U)}$	High-side Bias Voltage for U Phase MOSFET Driving
12	$V_{S(U)}$	High-side Bias Voltage Ground for U Phase MOSFET Driving
13	$IN_{(VH)}$	Signal Input for High-side V Phase
14	$V_{CC(VH)}$	High-side Bias Voltage for V Phase IC
15	$V_{B(V)}$	High-side Bias Voltage for V Phase MOSFET Driving
16	$V_{S(V)}$	High-side Bias Voltage Ground for V Phase MOSFET Driving
17	$IN_{(WH)}$	Signal Input for High-side W Phase
18	$V_{CC(WH)}$	High-side Bias Voltage for W Phase IC
19	$V_{B(W)}$	High-side Bias Voltage for W Phase MOSFET Driving
20	$V_{S(W)}$	High-side Bias Voltage Ground for W Phase MOSFET Driving
21	N_U	Negative DC–Link Input for U Phase
22	N_V	Negative DC–Link Input for V Phase
23	N_W	Negative DC–Link Input for W Phase
24	U	Output for U Phase
25	V	Output for V Phase
26	W	Output for W Phase
27	P	Positive DC–Link Input

Internal Equivalent Circuit and Input/Output Pins



Note:

1. Inverter low-side is composed of three MOSFETs, and one control IC. It has gate driving and protection functions.
2. Inverter power side is composed of four inverter dc-link input terminals and three inverter output terminals.
3. Inverter high-side is composed of three MOSFETs and three drive ICs for each MOSFET.

Figure 3.

Absolute Maximum Ratings ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)**Inverter Part**

Symbol	Parameter	Conditions	Rating	Units
V_{PN}	Supply Voltage	Applied between P- N_U , N_V , N_W	400	V
$V_{PN(\text{Surge})}$	Supply Voltage (Surge)	Applied between P- N_U , N_V , N_W	450	V
V_{DSS}	Drain-Source Voltage		500	V
$\pm I_D$	Each MOSFET Drain Current	$T_C = 25^\circ\text{C}$, Peak Sinusoidal Current	5	A
$\pm I_{DP}$	Each MOSFET Drain Current (Peak)	$T_C = 25^\circ\text{C}$, Under 1ms Pulse Width	7	A
P_C	Collector Dissipation	$T_C = 25^\circ\text{C}$ per One Chip	25	W
T_J	Operating Junction Temperature	(Note 1)	-20 ~ 125	$^\circ\text{C}$

Note:

1. The maximum junction temperature rating of the power chips integrated within the SPM is 150°C ($@T_C \leq 100^\circ\text{C}$). However, to insure safe operation of the SPM, the average junction temperature should be limited to $T_{J(\text{ave})} \leq 125^\circ\text{C}$ ($@T_C \leq 100^\circ\text{C}$)

Control Part

Symbol	Parameter	Conditions	Rating	Units
V_{CC}	Control Supply Voltage	Applied between $V_{CC(UH)}$, $V_{CC(VH)}$, $V_{CC(WH)}$, $V_{CC(L)}$ - COM	20	V
V_{BS}	High-side Control Bias Voltage	Applied between $V_{B(U)}$ - $V_{S(U)}$, $V_{B(V)}$ - $V_{S(V)}$, $V_{B(W)}$ - $V_{S(W)}$	20	V
V_{IN}	Input Signal Voltage	Applied between $IN_{(UH)}$, $IN_{(VH)}$, $IN_{(WH)}$, $IN_{(UL)}$, $IN_{(VL)}$, $IN_{(WL)}$ - COM	-0.3~17	V
V_{FO}	Fault Output Supply Voltage	Applied between V_{FO} - COM	-0.3~ $V_{CC}+0.3$	V
I_{FO}	Fault Output Current	Sink Current at V_{FO} Pin	5	mA
V_{SC}	Current Sensing Input Voltage	Applied between C_{SC} - COM	-0.3~ $V_{CC}+0.3$	V

Total System

Symbol	Parameter	Conditions	Rating	Units
T_{SC}	Short Circuit Withstanding Time	$V_{CC} = V_{BS} = 13.5 \sim 16.5\text{V}$, $T_J = 125^\circ\text{C}$, Non-repetitive, $V_{PN}=400\text{V}$, $R_{Shunt}=0\text{m}\Omega$	10	μs
T_C	Module Case Operation Temperature	$-20^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$, See Figure 2	-20 ~ 100	$^\circ\text{C}$
T_{STG}	Storage Temperature		-40 ~ 125	$^\circ\text{C}$
V_{ISO}	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to ceramic substrate	2500	V_{rms}

Thermal Resistance

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$R_{th(j-c)}$	Junction to Case Thermal Resistance	Inverter MOSFET part (per 1/6 module)	-	-	4	$^\circ\text{C/W}$

Note:

2. For the measurement point of case temperature(T_C), please refer to Figure 2.

Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FCBS0550	FCBS0550	SPM27BA	-	-	10

Electrical Characteristics ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)

Inverter Part

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$R_{DS(ON)}$	Static Drain-Source On Resistance	$V_{CC} = V_{BS} = 15\text{V}$ $V_{IN} = 5\text{V}$ $I_D = 2.5\text{A}$, $T_J = 25^\circ\text{C}$	-	1.35	1.75	Ω
V_{SD}	Drain-Source Diode Forward Voltage	$V_{CC} = V_{BS} = 15\text{V}$ $V_{IN} = 0\text{V}$ $I_D = 2.5\text{A}$, $T_J = 25^\circ\text{C}$	-	-	1.20	V
HS	t_{ON}	$V_{PN} = 300\text{V}$, $V_{CC} = V_{BS} = 15\text{V}$ $I_D = 2.5\text{A}$ $V_{IN} = 0\text{V} \leftrightarrow 5\text{V}$, Inductive Load (Note 3)	-	0.51	-	μs
	$t_{C(ON)}$		-	0.16	-	μs
	t_{OFF}		-	0.72	-	μs
	$t_{C(OFF)}$		-	0.10	-	μs
	t_{rr}		-	0.16	-	μs
LS	t_{ON}	$V_{PN} = 300\text{V}$, $V_{CC} = V_{BS} = 15\text{V}$ $I_D = 2.5\text{A}$ $V_{IN} = 0\text{V} \leftrightarrow 5\text{V}$, Inductive Load (Note 3)	-	0.52	-	μs
	$t_{C(ON)}$		-	0.18	-	μs
	t_{OFF}		-	0.74	-	μs
	$t_{C(OFF)}$		-	0.10	-	μs
	t_{rr}		-	0.16	-	μs
I_{DSS}	Drain - Source Leakage Current	$V_{DS} = V_{DSS}$	-	-	250	μA

Note:

- t_{ON} and t_{OFF} include the propagation delay time of the internal drive IC. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.

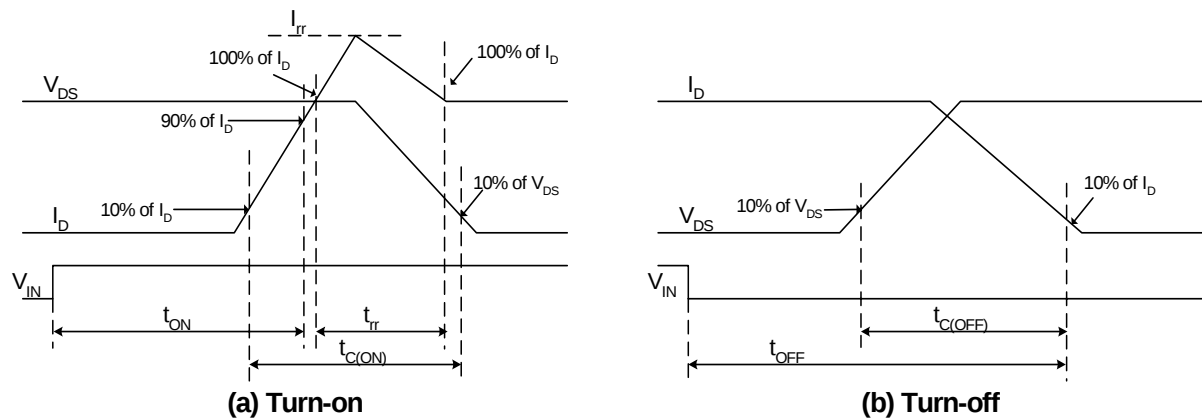


Figure 4. Switching Time Definition

Electrical Characteristics ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)**Control Part**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
I_{QCCL}	Quiescent V_{CC} Supply Current	$V_{CC} = 15\text{V}$ $I_{N(U), V_L, W_L} = 0\text{V}$	-	-	23	mA
I_{QCCH}		$V_{CC} = 15\text{V}$ $I_{N(UH, VH, WH)} = 0\text{V}$	-	-	100	μA
I_{QBS}	Quiescent V_{BS} Supply Current	$V_{BS} = 15\text{V}$ $I_{N(UH, VH, WH)} = 0\text{V}$	-	-	500	μA
V_{FOH}	Fault Output Voltage	$V_{SC} = 0\text{V}$, V_{FO} Circuit: 4.7k Ω to 5V Pull-up	4.5	-	-	V
V_{FOL}		$V_{SC} = 1\text{V}$, V_{FO} Circuit: 4.7k Ω to 5V Pull-up	-	-	0.8	V
$V_{SC(ref)}$	Short Circuit Trip Level	$V_{CC} = 15\text{V}$ (Note 4)	0.45	0.5	0.55	V
UV_{CCD}	Supply Circuit Under-Voltage Protection	Detection Level	10.7	11.9	13.0	V
UV_{CCR}		Reset Level	11.2	12.4	13.2	V
UV_{BSD}		Detection Level	10.1	11.3	12.5	V
UV_{BSR}		Reset Level	10.5	11.7	12.9	V
t_{FOD}	Fault-out Pulse Width	$C_{FOD} = 33\text{nF}$ (Note 5)	1.0	1.8	-	ms
$V_{IN(ON)}$	ON Threshold Voltage	Applied between $I_{N(UH)}$, $I_{N(VH)}$, $I_{N(WH)}$, $I_{N(UL)}$, $I_{N(VL)}$, $I_{N(WL)}$ - COM	2.9	-	-	V
$V_{IN(OFF)}$	OFF Threshold Voltage		-	-	0.8	V

Note:

4. Short-circuit current protection is functioning only at the low-sides.

5. The fault-out pulse width t_{FOD} depends on the capacitance value of C_{FOD} according to the following approximate equation : $C_{FOD} = 18.3 \times 10^{-6} \times t_{FOD}[\text{F}]$ **Recommended Operating Conditions**

Symbol	Parameter	Conditions	Value			Units
			Min.	Typ.	Max.	
V_{PN}	Supply Voltage	Applied between P - N_U , N_V , N_W	-	300	400	V
V_{CC}	Control Supply Voltage	Applied between $V_{CC(UH)}$, $V_{CC(VH)}$, $V_{CC(WH)}$, $V_{CC(L)}$ - COM	13.5	15	16.5	V
V_{BS}	High-side Bias Voltage	Applied between $V_{B(U)}$ - $V_{S(U)}$, $V_{B(V)}$ - $V_{S(V)}$, $V_{B(W)}$ - $V_{S(W)}$	13.0	15	18.5	V
dV_{CC}/dt , dV_{BS}/dt	Control supply variation		-1	-	1	V/ μs
t_{dead}	Blanking Time for Preventing Arm-short	For Each Input Signal	2	-	-	μs
f_{PWM}	PWM Input Signal	$-20^\circ\text{C} \leq T_C \leq 100^\circ\text{C}$, $-20^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$	-	-	20	kHz
V_{SEN}	Voltage for Current Sensing	Applied between N_U , N_V , N_W - COM (Including surge voltage)	-4		4	V

Mechanical Characteristics and Ratings

Parameter	Conditions		Limits			Units
			Min.	Typ.	Max.	
Mounting Torque	Mounting Screw: - M3	Recommended 0.62N•m	0.51	0.62	0.72	N•m
Device Flatness		Note Fig. 5	0	-	+120	μm
Weight			-	15.4	-	g

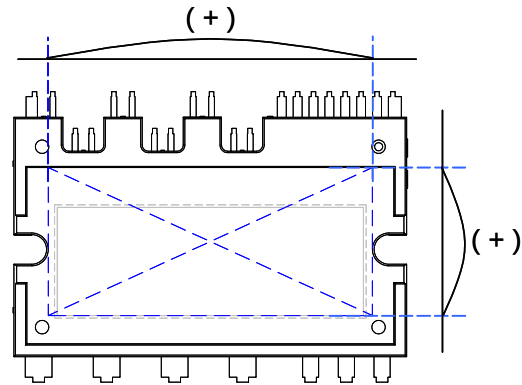
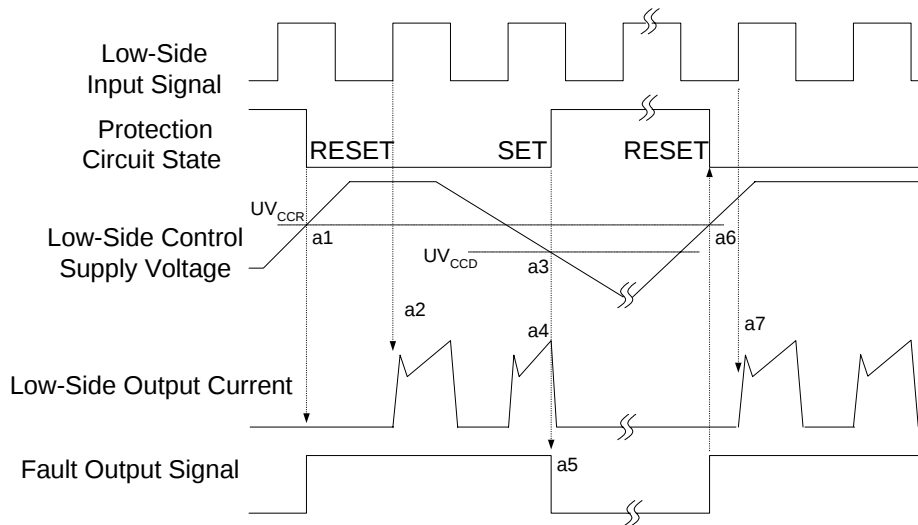


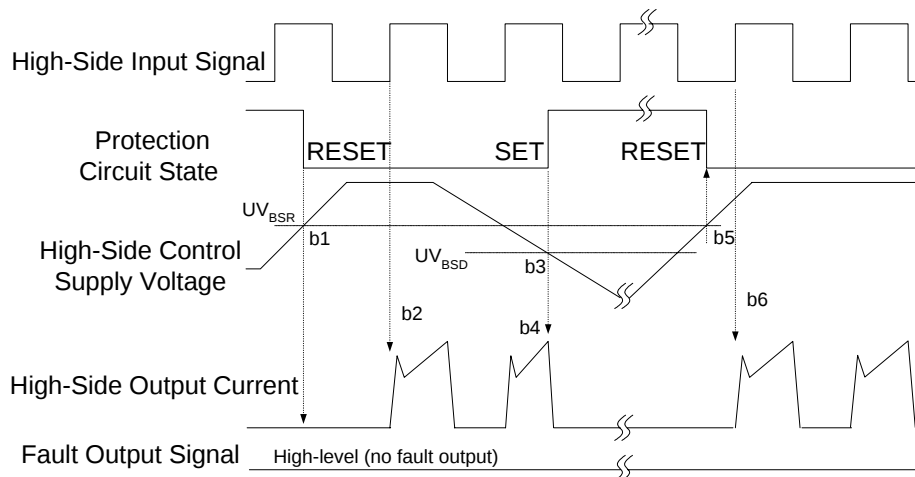
Figure 5. Flatness Measurement Position

Time Charts of SPMs Protective Function



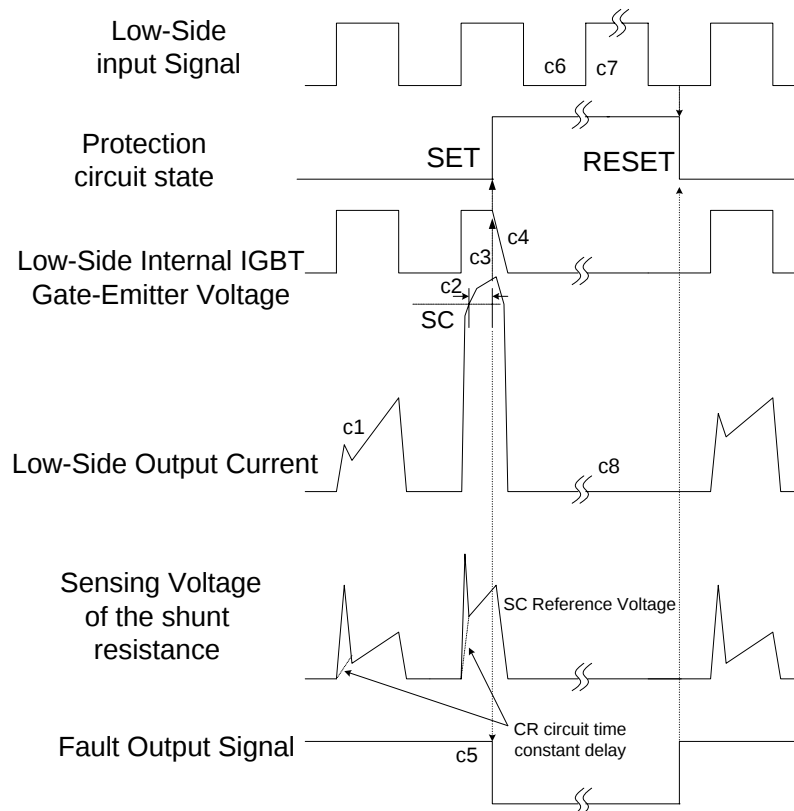
- a1 : Control supply voltage rises: After the voltage rises UV_{CCR} , the circuits start to operate when next input is applied.
- a2 : Normal operation: MOSFET ON and carrying current.
- a3 : Under voltage detection (UV_{CCD}).
- a4 : MOSFET OFF in spite of control input condition.
- a5 : Fault output operation starts.
- a6 : Under voltage reset (UV_{CCR}).
- a7 : Normal operation: MOSFET ON and carrying current.

Figure 6. Under-Voltage Protection (Low-side)



- b1 : Control supply voltage rises: After the voltage reaches UV_{BSR} , the circuits start to operate when next input is applied.
- b2 : Normal operation: MOSFET ON and carrying current.
- b3 : Under voltage detection (UV_{BSD}).
- b4 : MOSFET OFF in spite of control input condition, but there is no fault output signal.
- b5 : Under voltage reset (UV_{BSR})
- b6 : Normal operation: MOSFET ON and carrying current

Figure 7. Under-Voltage Protection (High-side)



(with the external shunt resistance and CR connection)

c1 : Normal operation: MOSFET ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard MOSFET gate interrupt.

c4 : MOSFET turns OFF.

c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor C_{FO} .

c6 : Input "L" : MOSFET OFF state.

c7 : Input "H": MOSFET ON state, but during the active period of fault output the MOSFET doesn't turn ON.

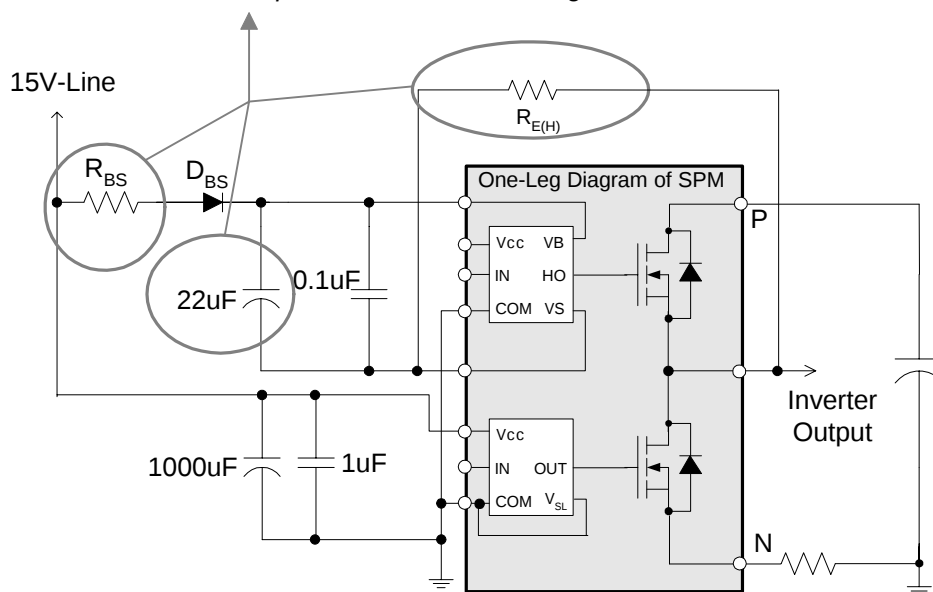
c8 : MOSFET OFF state

Figure 8. Short-Circuit Current Protection (Low-side Operation only)

1. RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The SPM input signal section integrates 3.3k Ω (typ.) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.
2. The logic input is compatible with standard CMOS or LSTTL outputs.

Figure 9. Recommended CPU I/O Interface Circuit

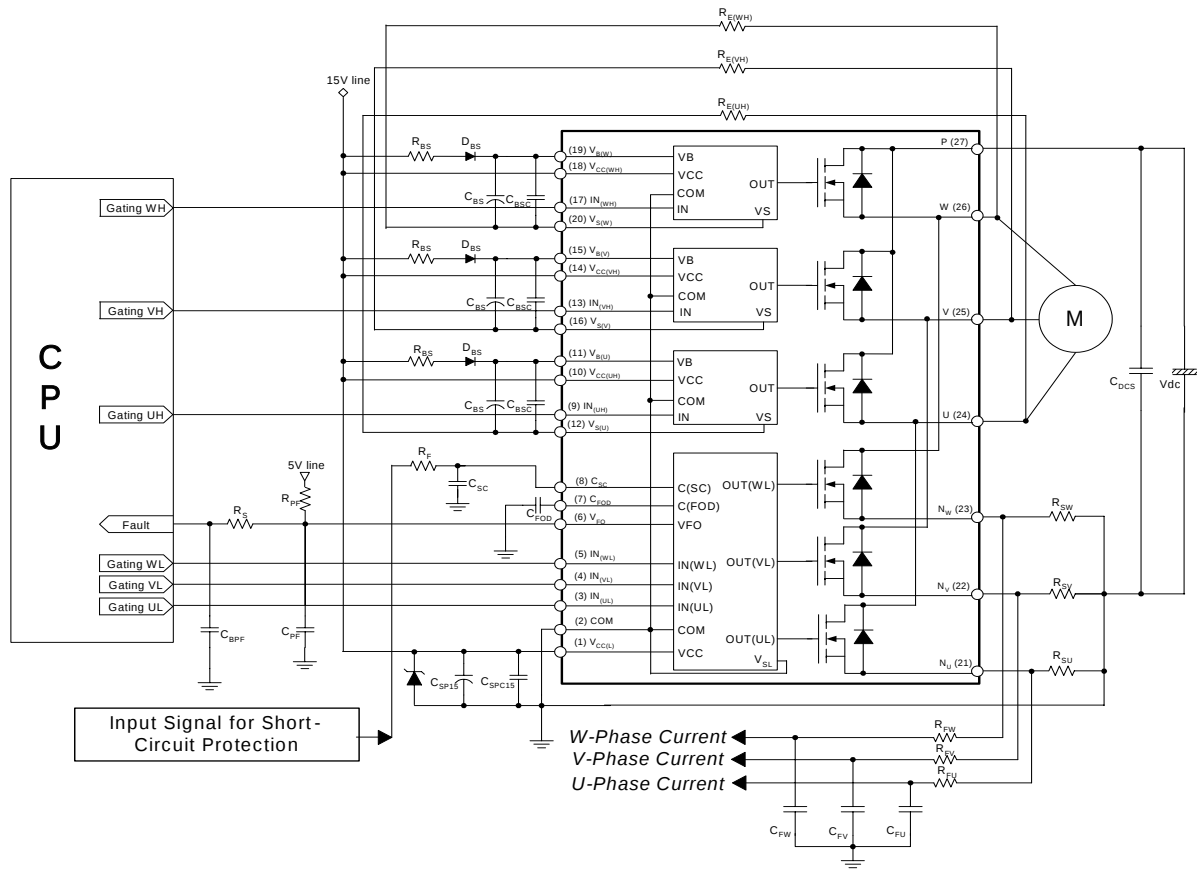
These Values depend on PWM Control Algorithm



Note:

1. It would be recommended that the bootstrap diode, D_{BS} , has soft and fast recovery characteristics.
2. The bootstrap resistor (R_{BS}) should be 3 times greater than $R_{E(H)}$. The recommended value of $R_{E(H)}$ is 5.6Ω , but it can be increased up to 20Ω (maximum) for a slower dv/dt of high-side.
3. The ceramic capacitor placed between V_{CC} -COM should be over $1\mu F$ and mounted as close to the pins of the SPM as possible.

Fig. 10. Recommended Bootstrap Operation Circuit and Parameters

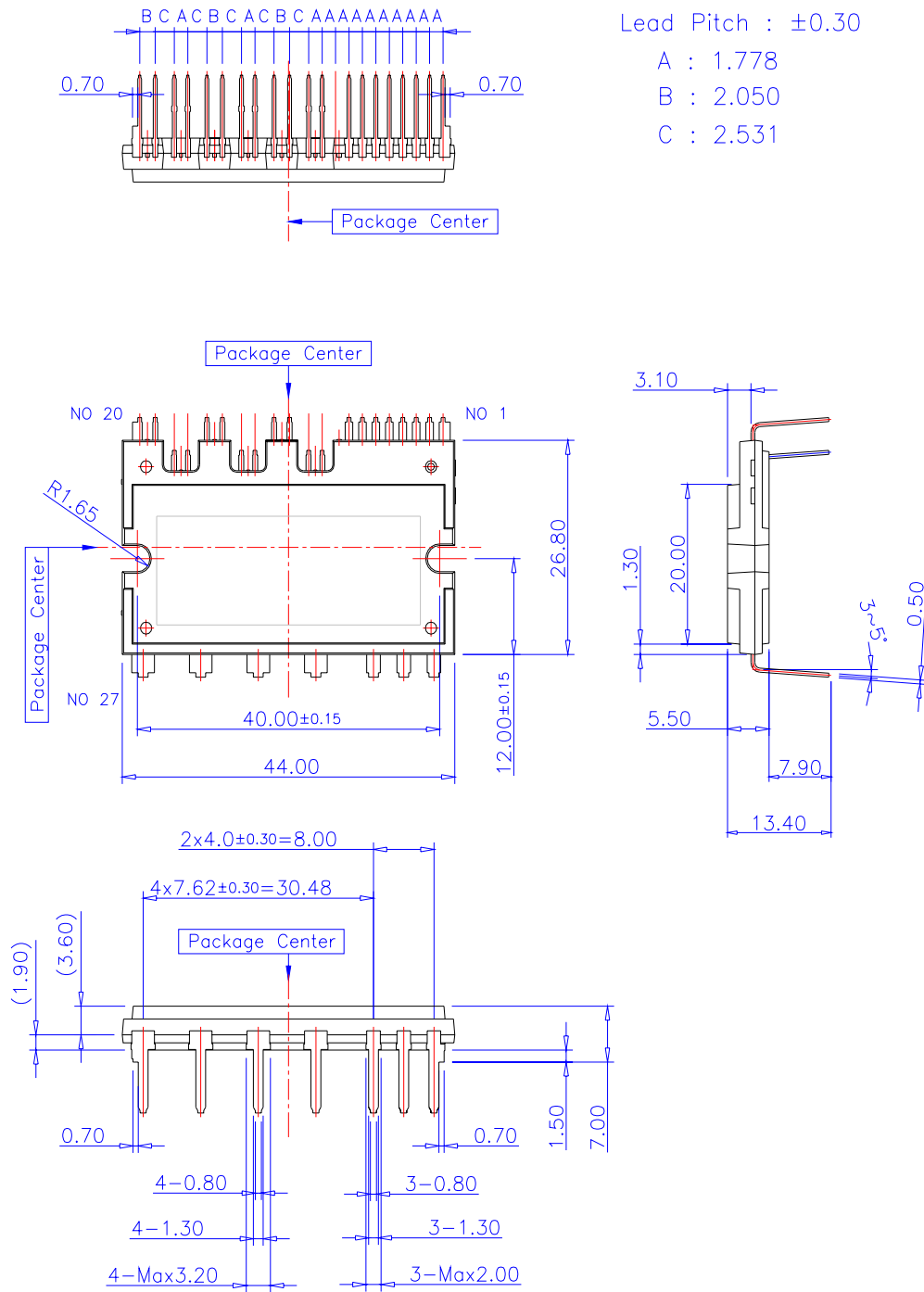


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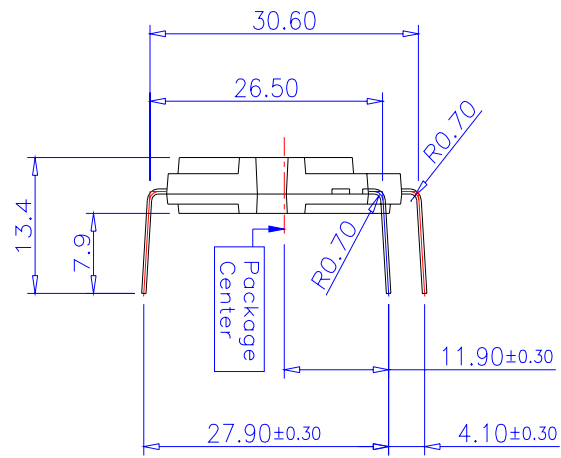
1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)
2. By virtue of integrating an application specific type HVIC inside the SPM, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
3. V_{FO} output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7k Ω resistance. Please refer to Figure 9.
4. C_{SP15} of around 7 times larger than bootstrap capacitor C_{BS} is recommended.
5. V_{FO} output pulse width should be determined by connecting an external capacitor (C_{FOD}) between C_{FOD} (pin7) and COM(pin2). (Example : if $C_{FOD} = 33$ nF, then $t_{FO} = 1.8$ ms (typ.)) Please refer to the note 5 for calculation method.
6. Input signal is High-Active type. There is a 3.3k Ω resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turn-off/turn-on threshold voltage.
7. To prevent errors of the protection function, the wiring around R_F and C_{SC} should be as short as possible.
8. In the short-circuit protection circuit, please select the $R_F C_{SC}$ time constant in the range 1.5~2 μ s.
9. Each capacitor should be mounted as close to the pins of the SPM as possible.
10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&COM pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 μ F between the P&COM pins is recommended.
11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays.
12. C_{SP15} should be over 1 μ F and mounted as close to the pins of the SPM as possible.

Fig. 11. Typical Application Circuit

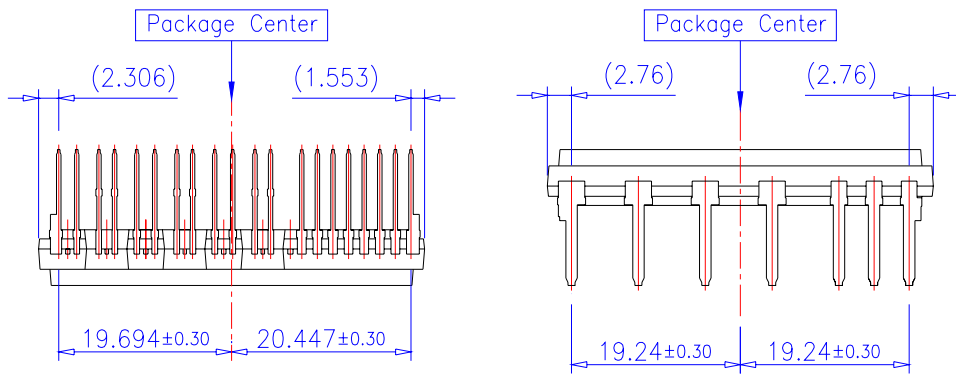
Detailed Package Outline Drawings



Detailed Package Outline Drawings (Continued)

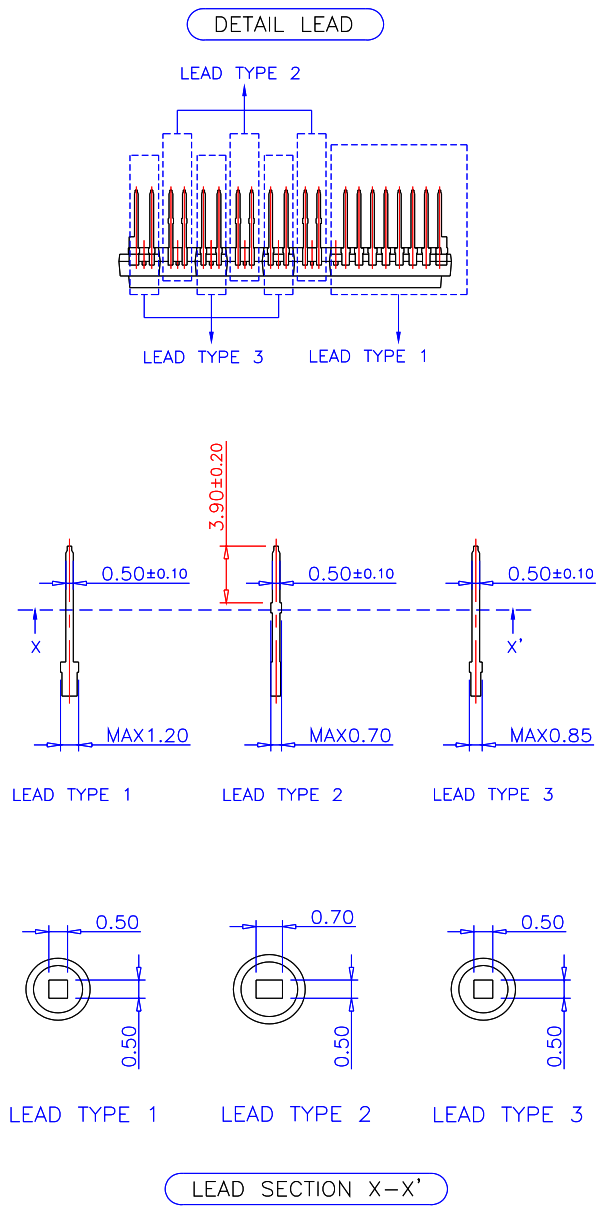


Lead Forming Dimension



PKG Center to Lead Distance

Detailed Package Outline Drawings (Continued)



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