

FEATURES

- Drives 6 LEDs from 2.6 V to 5.5 V (Li-Ion) input supply
- 1×/1.5×/2× fractional charge pump to maximize power efficiency
- 1% max LED current matching
- Up to 88% power efficiency over Li-Ion range
- Powers main and sub display LEDs with individual shutdown
- Package footprint only 9 mm² (3 mm × 3 mm)
- Package height only 0.9 mm
- Low power shutdown mode
- Shutdown function
- Soft-start limiting in-rush current

APPLICATIONS

- Cellular phones with main and sub displays
- White LED backlighting
- Camera flash/strobes and movie lights
- Micro TFT color displays
- DSC
- PDA's

GENERAL DESCRIPTION

The ADM8845 uses charge pump technology to provide the power required to drive up to six LEDs. The LEDs are used for backlighting a color LCD display, having regulated constant current for uniform brightness intensity. The main display can have up to four LEDs, and the sub display can have one or two LEDs. The digital CTRL1 and CTRL2 input control pins control the shutdown operation and the brightness of the main and sub displays.

To maximize power efficiency, the charge pump can operate in either a 1×, 1.5×, or 2× mode. The charge pump automatically switches between 1×/1.5×/2× modes based, on the input voltage, to maintain sufficient drive for the LED anodes at the highest power efficiency.

Improved brightness matching of the LEDs is achieved by a feedback pin to sense individual LED current with a maximum matching accuracy of 1%.

FUNCTIONAL BLOCK DIAGRAM

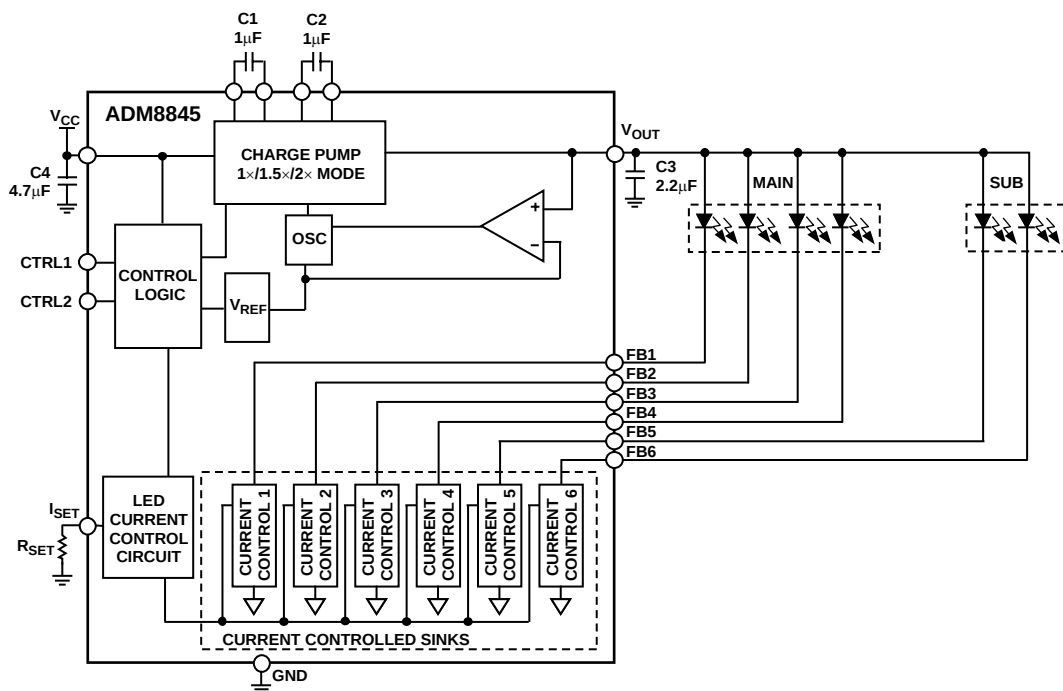


Figure 1.

Rev. 0

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REVISION HISTORY

10/04—Revision 0: Initial Version

SPECIFICATIONS

$V_{CC} = 2.6\text{ V to }5.5\text{ V}$; $T_A = -40^\circ\text{C to }+85^\circ\text{C}$, unless otherwise noted; $C1, C2 = 1.0\text{ }\mu\text{F}$; $C3 = 2.2\text{ }\mu\text{F}$; $C4 = 4.7\text{ }\mu\text{F}$

Table 1.

Parameter	Min	Typ	Max	Unit	Test Conditions
INPUT VOLTAGE, V_{CC}	2.6		5.5	V	
SUPPLY CURRENT, I_{CC}		2.6	5	mA	All six LEDs disabled, $V_{CC} = 3.3\text{ V}$, $R_{SET} = 7.08\text{ k}\Omega$ $CTRL1 = 1$, $CRTL2 = 1$
SHUTDOWN CURRENT			5	μA	
CHARGE PUMP FREQUENCY		1.5		MHz	
CHARGE PUMP MODE THRESHOLDS					
1.5× to 2× Accuracy		3.33	4	V %	
2× to 1.5× Accuracy		3.36	4	V %	
Hysteresis		40		mV	
1× to 1.5× Accuracy		4.77	4	V %	
1.5× to 1× Accuracy		4.81	4	V %	
Hysteresis		40		mV	
I_{SET} PIN					
LED : LED Matching	−1		+1	%	$I_{LED} = 20\text{ mA}$, $V_{FB} = 0.4\text{ V}$
LED : I_{SET} Accuracy	−1		+1	%	$I_{LED} = 20\text{ mA}$, $R_{SET} = 7.08\text{ k}\Omega$, $V_{FB} = 0.4\text{ V}$, $V_{CC} = 3.6\text{ V}$, $T_A = 25^\circ\text{C}$
I_{SET} Pin Voltage		1.18		V	
I_{LED} to I_{SET} Ratio		120			
MIN COMPLIANCE ON FBx PIN		0.2	0.3	V	$I_{SET} = 20\text{ mA}$
Charge Pump Output Resistance		1.2	1.8	Ω	1× mode
		3.5	5.1	Ω	1.5× mode
		8.0	14	Ω	2× mode
LED Current			30	mA	See Note 1 and Figure 21
PWM	0.1		200	kHz	
DIGITAL INPUTS					
Input High	0.5 V_{CC}			V	
Input Low			0.3 V_{CC}	V	
Input Leakage Current			1	μA	
CHARGE PUMP POWER EFFICIENCY		88		%	$CTRL1 = 1$, $CRTL2 = 1$, $V_{CC} = 3.4\text{ V}$, $V_{FB} = 0.2\text{ V}$, $I_{FB} = 20\text{ mA}$
V_{OUT} RIPPLE		30		mV	$V_{CC} = 3.6\text{ V}$, $I_{LED} = 20\text{ mA}$, all six LEDs enabled

¹ Guaranteed by design. Not 100% production tested.

ABSOLUTE MAXIMUM RATINGS

$T_A = 25^\circ\text{C}$, unless otherwise noted.

Table 2.

Parameter	Rating
Supply Voltage, V_{CC}	–0.3 V to +6.0 V
I_{SET}	–0.3 V to +2.0 V
CTRL1, CTRL2	–0.3 V to +6.0 V
V_{OUT} Shorted ¹	Indefinite
Feedback pins FB1 to FB6	–0.3 V to +6.0 V
Operating Temperature Range	
Six LEDs Enabled with 30 mA/LED ²	–40°C to +65°C
Six LEDs Enabled with 20 mA/LED ²	–40°C to +85°C
V_{OUT} ³	180 mA
Storage Temperature Range	–65°C to +125°C
Power Dissipation	2 mW
ESD Class	1

¹ Short through LED.

² LED current should be derated above $T_A > 65^\circ\text{C}$, refer to Figure 21.

³ Based on long-term current density limitations.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL CHARACTERISTICS

16-Lead LFCSP Package:

$\theta_{JA} = 50^\circ\text{C/W}$

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



PIN CONFIGURATION AND FUNCTION DESCRIPTION

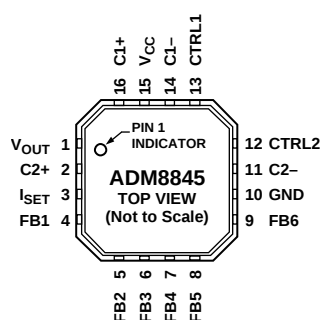


Figure 2. Pin Configuration

Table 3. Pin Function Descriptions

Pin No.	Mnemonic	Function
1	V _{OUT}	Charge Pump Output. A 2.2 μ F capacitor to ground is required on this pin. Connect V _{OUT} to the anodes of all the LEDs.
2	C2+	Flying Capacitor 2 Positive Connection.
3	I _{SET}	Bias Current Set Input. The current flowing through R _{SET} , I _{SET} , is gained up by 120 to give the I _{LED} current. Connect R _{SET} to GND to set the bias current as V _{SET} /R _{SET} . Note that V _{SET} = 1.18 V.
4–9	FB1–FB6	LED1–LED6 Cathode Connection and Charge Pump Feedback. The current flowing in these LEDs is 120 times the current flowing through R _{SET} , I _{SET} . When using fewer than six LEDs, this pin can be left unconnected or connected to GND.
10	GND	Device Ground Pin.
11	C2–	Flying Capacitor 2 Negative Connection.
12	CTRL2	Digital Input. 3 V CMOS Logic. Used with CTRL1 to control the shutdown operation of the main and sub LEDs.
13	CTRL1	Digital Input. 3 V CMOS Logic. Used with CTRL2 to control the shutdown operation of the main and sub LEDs.
14	C1–	Flying Capacitor 1 Negative Connection.
15	V _{CC}	Positive Supply Voltage Input. Connect this pin to a 2.6 V to 5.5 V supply with a 4.7 μ F decoupling capacitor.
16	C1+	Flying Capacitor 1 Positive Connection.
-	EP	Expose Paddle. Connect the exposed paddle to GND.

TYPICAL PERFORMANCE CHARACTERISTICS

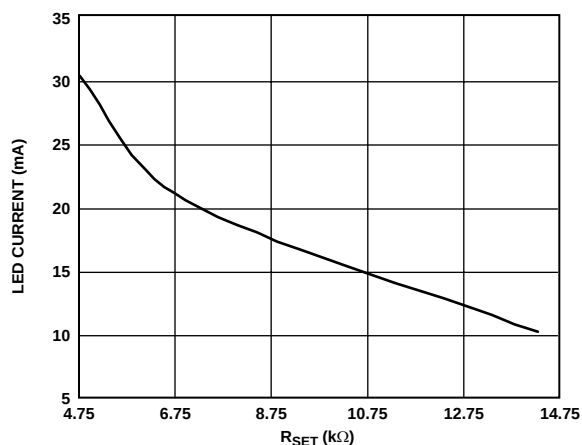


Figure 3. I_{LED} (mA) Current vs. R_{SET}

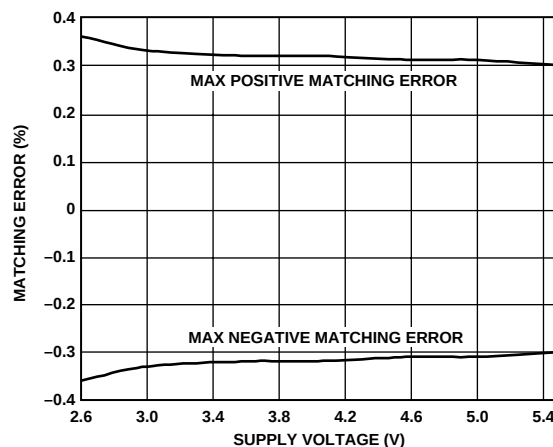


Figure 6. I_{LED} (mA) Matching Error (%) vs. Supply Voltage (V), $T_A = 25^\circ\text{C}$ and $I_{LED} = 20\text{ mA}$

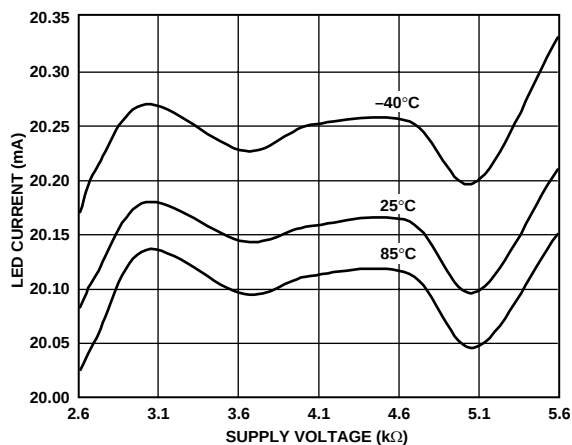


Figure 4. I_{LED} (mA) vs. Temperature ($^\circ\text{C}$), Six LEDs Enabled

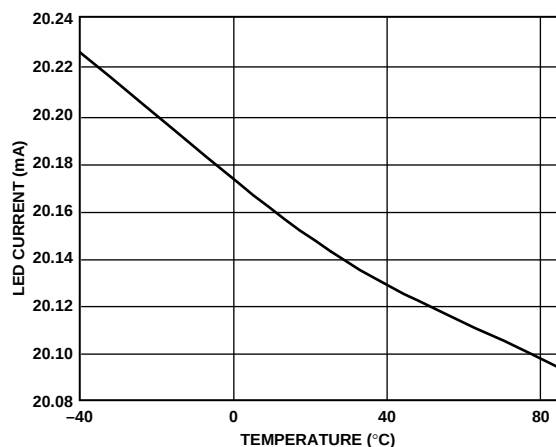


Figure 7. I_{LED} (mA) Variation over Temperature ($^\circ\text{C}$), $V_{CC} = 3.6\text{ V}$

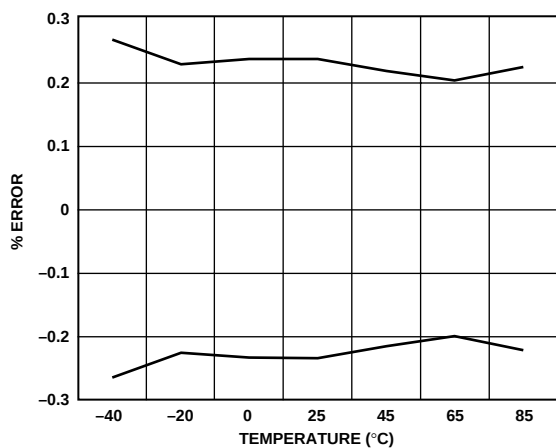


Figure 5. I_{LED} Matching (%) over Temperature ($^\circ\text{C}$), $V_{CC} = 3.6\text{ V}$, $I_{LED} = 20\text{ mA}$, Six LEDs Enabled

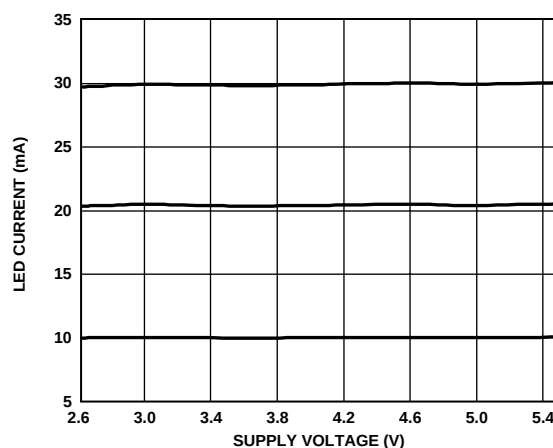


Figure 8. I_{LED} (mA) vs. Supply Voltage (V)

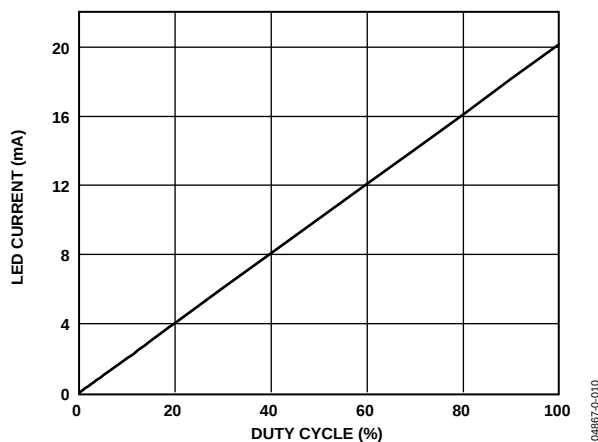


Figure 9. I_{LED} (mA) vs. PWM Dimming (Varying Duty Cycle), Six LEDs Enabled, Frequency = 1 kHz

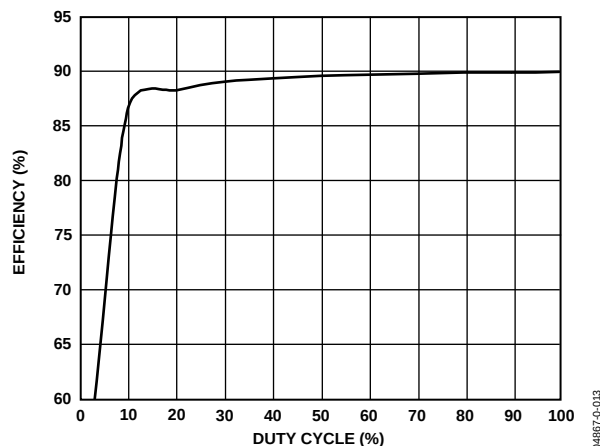


Figure 12. LED Efficiency vs. Varying Duty Cycle of 1 kHz PWM Signal, Six LEDs Enabled, 20 mA/LED

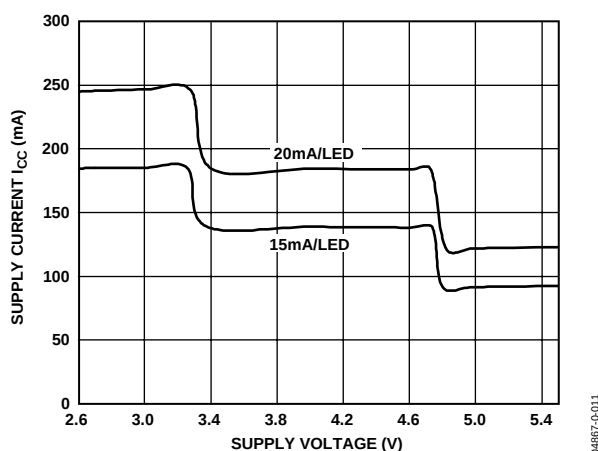


Figure 10. Input Current vs. Supply Voltage, Six LEDs Enabled

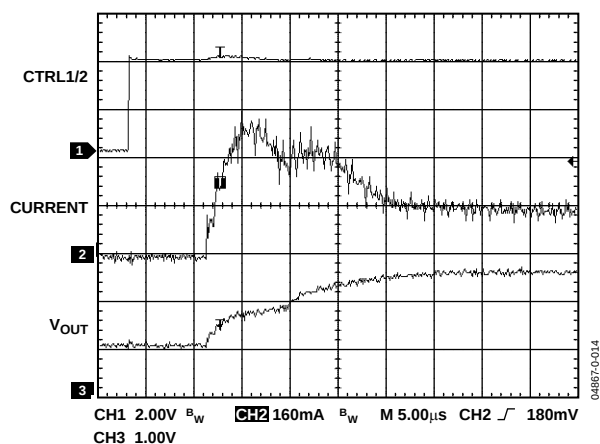


Figure 13. Soft Start Showing the Initial In-Rush Current and V_{OUT} Variation, Six LEDs @ 20 mA/LED, $V_{CC} = 3.6$ V

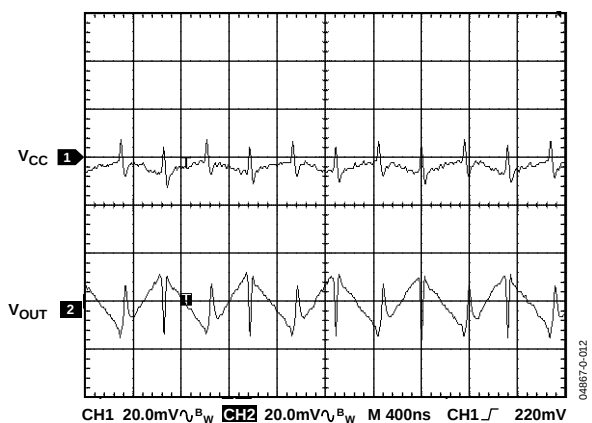


Figure 11. 1.5x Mode Operating Waveforms

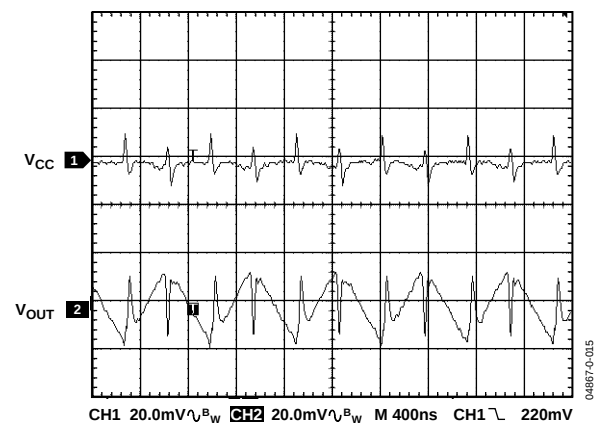


Figure 14. .2x Mode Operating Waveform

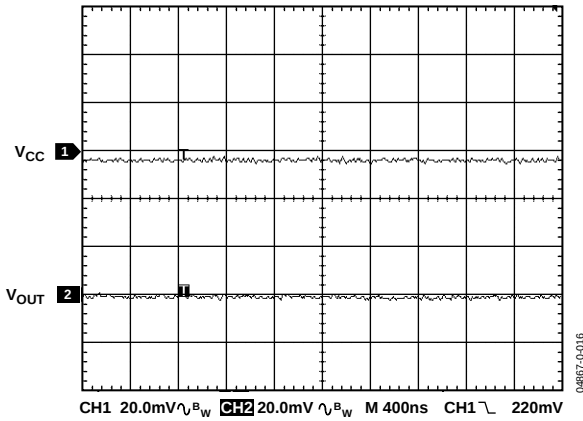


Figure 15. 1x Mode Operating Waveforms

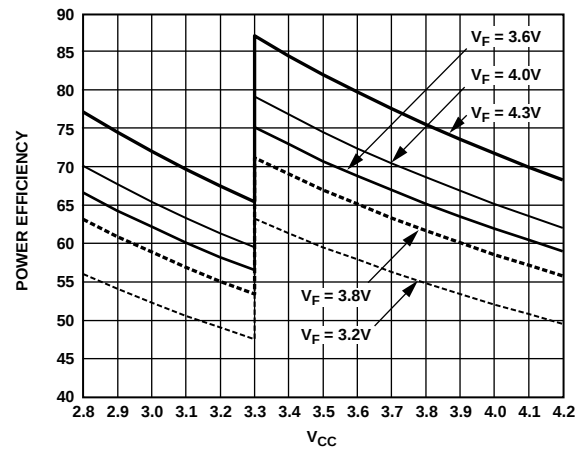


Figure 18. Power Efficiency vs. Supply Voltage over Li-Ion Range, Six LEDs @ 20 mA/LED

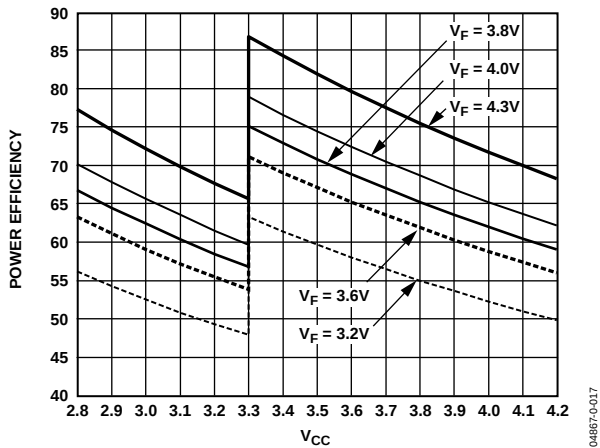


Figure 16. Power Efficiency vs. Supply Voltage over Li-Ion Range, Six LEDs @ 15 mA/LED

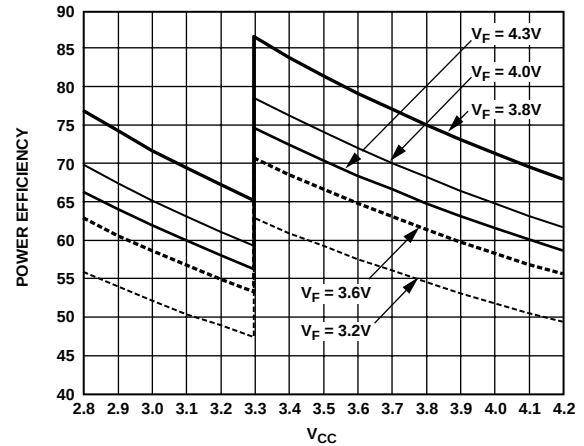


Figure 19. Power Efficiency vs. Supply Voltage over Li-Ion Range, Four LEDs @ 20 mA/LED

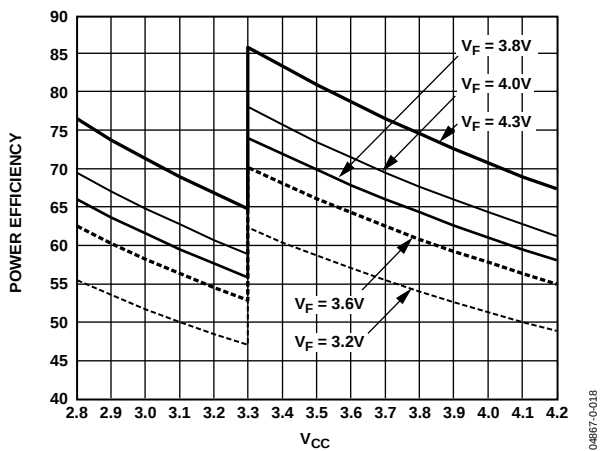


Figure 17. Power Efficiency vs. Supply Voltage over Li-Ion Range, Four LEDs @ 15 mA/LED

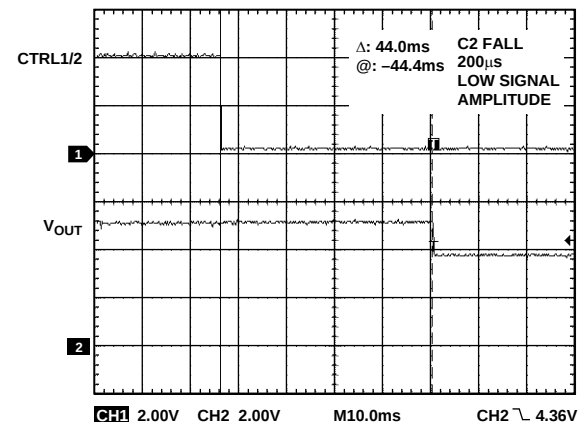


Figure 20. TPC Delay

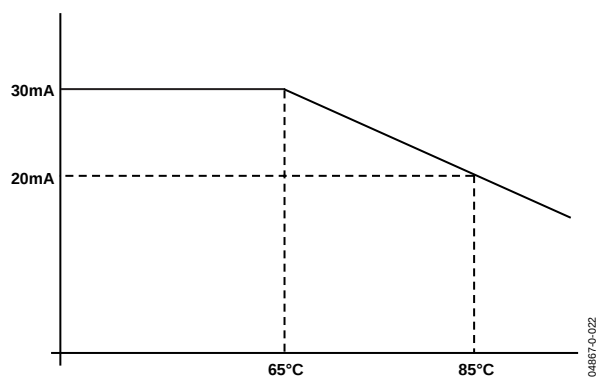


Figure 21. Maximum I_{LED} (mA) vs. Ambient Temperature,
Six LEDs Connected

THEORY OF OPERATION

The ADM8845 charge pump driver for LCD white LED backlights implements a multiple gain charge pump (1×, 1.5×, 2×) to maintain the correct voltage on the anodes of the LEDs over a 2.6 V to 5.5 V (Li-Ion) input supply voltage. The charge pump automatically switches between 1×/1.5×/2× modes based, on the input voltage, to maintain sufficient drive for the LED anodes, with V_{CC} input voltages as low as 2.6 V. It also includes regulation of the charge pump output voltage for supply voltages up to 5.5 V. The ADM8845's six LEDs are arranged into two groups, main and sub. The main display can have up to four LEDs (FB1 to FB4), and the sub display can have one or two LEDs (FB5 and FB6) (see Figure 23). Two digital input control pins, CTRL1 and CTRL2, control the shutdown operation and the brightness of the main and sub displays (see Table 4).

Table 4. Shutdown Truth Table

CTRL1	CTRL2	LED Shutdown Operation
0	0	Sub Display Off / Main Display Off
0	1	Sub Display Off / Main Display On
1	0	Sub Display On / Main Display Off
1	1	Sub Display On / Main Display On

An external resistor, R_{SET}, is connected between the I_{SET} pin and GND. This resistor sets up a reference current, I_{SET}, which is internally gained up by 120 within the ADM8845 to produce the I_{LED} currents of up to 30 mA/LED (I_{LED} = I_{SET} × 120 and I_{SET} = 1.18 V/R_{SET}). The ADM8845 uses six individual current sinks to individually sense each LED current with a maximum matching performance of 1%. This current matching performance ensures uniform brightness across a color display.

The ADM8845 lets the user control the brightness of the white LEDs with a digital PWM signal applied to CTRL1 and/or CTRL2. The duty cycle of the applied PWM signal determines the brightness of the main and/or sub display backlight white LEDs. The ADM8845 also allows the brightness of the white LEDs to be controlled using a dc voltage (see Figure 22). Soft-start circuitry limits the in-rush current flow at power-up. The ADM8845 is fabricated using CMOS technology for minimal power consumption and is packaged in a 16-lead lead frame chip scale package.

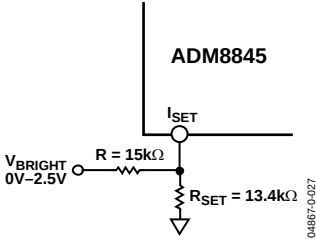


Figure 22. PWM Brightness Control Using a DC Voltage Applied to V_{BRIGHT}

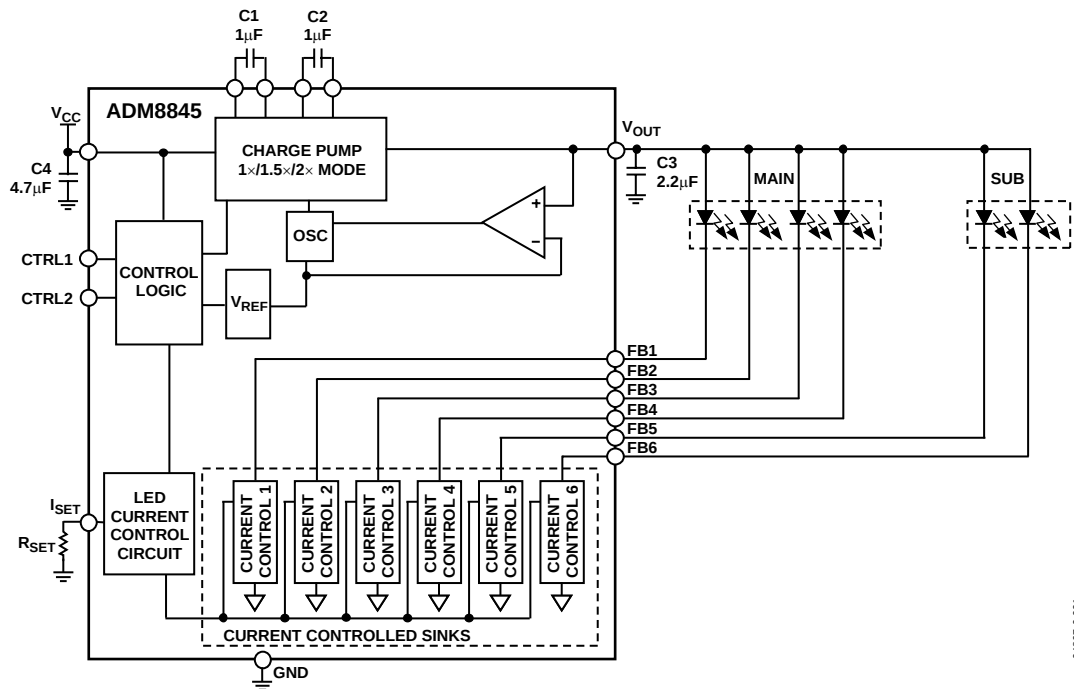


Figure 23. Functional Block Diagram

OUTPUT CURRENT CAPABILITY

The ADM8845 can drive up to 30 mA of current to each of the six LEDs given an input voltage of 2.6 V to 5.5 V. The LED currents have a maximum current matching of 1% between any two LED currents. An external resistor, R_{SET} , sets the output current, approximated by the following equation:

$$R_{SET} = 120 \times (1.18 \text{ V}/I_{LED})$$

To regulate the LED currents properly, sufficient headroom voltage (compliance) must be present. The compliance refers to the minimum amount of voltage that must be present across the internal current sinks to ensure that the desired current and matching performance is realizable. To ensure that the desired current is obtained, use the following equation to find the minimum input voltage required:

$$V_{OUT} - V_F \geq \text{Compliance}$$

where V_F is the LED forward voltage. For 20 mA/LED, the compliance is 0.20 V typ and 0.30 V max (see Table 5).

Table 5. I_{LED} , R_{SET} , and Compliance Table

I_{LED}	R_{SET}	Typ. Compliance
15 mA	9.44 k Ω	0.17 V
20 mA	7.08 k Ω	0.20 V
30 mA	4.72 k Ω	0.34 V

When the ADM8845 charge pump is loaded with 180 mA (six LEDs at 30 mA/LED), the ambient operating temperature is reduced (see Figure 21).

AUTOMATIC GAIN CONTROL

The automatic gain control block controls the operation of the charge pump by selecting the appropriate gain for the charge pump. This maintains sufficient drive for the LED anodes at the highest power efficiency over a 2.6 V to 5.5 V input supply range. The charge pump switching thresholds are described in Table 6.

Table 6. Charge Pump Switching Thresholds

Gain	Threshold
1.5 $\times$ to 2 $\times$	3.33 V
2 $\times$ to 1.5 $\times$	3.36 V
1 $\times$ to 1.5 $\times$	4.77 V
1.5 $\times$ to 1 $\times$	4.81 V

CURRENT MATCHING

The 1% maximum current matching performance is defined by the following equations:

$$I_{AVG} = (I_{MAX} + I_{MIN})/2$$

$$\text{Max Matching Error} = [(I_{MAX} - I_{AVG})/I_{AVG}] \times 100$$

or

$$\text{Min Matching Error} = [(I_{MIN} - I_{AVG})/I_{AVG}] \times 100$$

where I_{MAX} is the largest I_{LED} current, and I_{MIN} is the smallest I_{LED} current.

BRIGHTNESS CONTROL WITH A DIGITAL PWM SIGNAL

PWM brightness control provides the widest brightness control method by pulsing the white LEDs on and off using the digital input control pins, CTRL1 and/or CTRL2. PWM brightness control also removes any chromaticity shifts associated with changing the white LED current, because the LEDs operate either at zero current or full current (set by R_{SET}).

The digital PWM signal applied with a frequency of 100 Hz to 200 kHz turns the current control sinks on and off using CTRL1 and/or CTRL2. The average current through the LEDs changes with the PWM signal duty cycle. If the PWM frequency is much less than 100 Hz, flicker could be seen in the LEDs. For the ADM8845, zero duty cycle turns off the LEDs, and a 50% duty cycle results in an average LED current I_{LED} being half the programmed LED current. For example, if R_{SET} is set to program 20 mA/LED, a 50% duty cycle results in an average I_{LED} of 10 mA/LED, I_{LED} being half the programmed LED current.

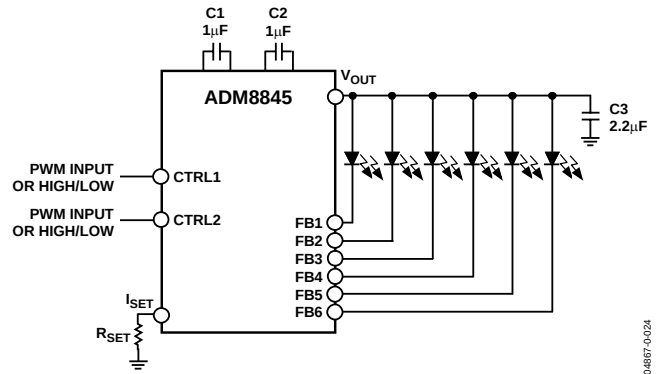


Figure 24. Digital PWM Brightness Control Application Diagram

By applying a digital PWM signal to the digital input control pins, CTRL1 and/or CTRL2 can adjust the brightness of the sub and/or main displays. The ADM8845's six white LEDs are organized into two groups, main display (FB1 to FB4) and sub display (FB4 to FB6); refer to the Theory of Operation section.

ADM8845

The ADM8845's main and sub display brightness can be controlled together or separately by applying a digital PWM signal to both CTRL1 and CTRL2 pins. The duty cycle of the applied digital PWM signal determines the brightness of the main and sub displays together. Varying the duty cycle of the applied PWM signal also varies the brightness of the main and sub displays from 0% to 100%.

By holding CTRL1 low and applying a digital PWM signal to CTRL2, the sub display is turned off and the main display is turned on. Then the brightness of the main display is determined by the duty cycle of the applied digital PWM signal.

By applying a digital PWM signal to CTRL1 and holding CTRL2 low, the sub display is turned on and the main display is turned off. Then the brightness of the sub display is determined by the duty cycle of the applied digital PWM signal.

By applying a digital PWM signal to CTRL1 and holding CTRL2 high, the sub display is turned on and the main display is turned on. Then the brightness of the sub display is determined by the duty cycle of the applied digital PWM signal. The brightness of the main display is set to the maximum (maximum is set by R_{SET}).

By holding CTRL1 high and applying a digital PWM signal to CTRL2, the sub display is turned on and the main display is turned on. Then the brightness of the main display is determined by the duty cycle of the applied digital PWM signal. The brightness of the sub display is set to the maximum (maximum is set by R_{SET}).

When CTRL1 and CTRL2 go low, the LED current control sinks shutdown. Shutdown of the charge pump is delayed by 15 ms. This timeout period (t_{CP}) allows the ADM8845 to determine if a digital PWM signal is present on CTRL1 and CTRL2 or if the user has selected a full chip shutdown (see Figure 25).

If digital PWM brightness control of the LEDs is not required, a constant Logic Level 1 (V_{CC}) or 0 (GND) must be applied.

The six white LED in the ADM8845 are arranged in two groups, sub and main. It is possible to configure the six LEDs as in Table 7; refer also to Figure 25.

Table 7. Digital Inputs Truth Table

CTRL1	CTRL2	LED Operation
0	0	Sub Display Off / Main Display Off (Full Shutdown) ^{1,2}
0	1	Sub Display Off / Main Display On ^{1,3}
1	0	Sub Display On / Main Display Off ^{1,2}
1	1	Sub Display On / Main Display On (Full On) ^{1,3}
0	PWM	Sub Display Off/ Digital PWM Brightness Control on Main Display ^{4,5}
PWM	0	Digital PWM Brightness Control on Sub Display / Main Display Off ^{2,4}
1	PWM	Sub Display On/ Digital PWM Brightness Control on Main Display ^{1,5}
PWM	1	Digital PWM Brightness Control on Sub Display / Main Display On ⁵
PWM	PWM	Digital PWM Brightness Control on Sub and Main Display

¹ Sub Display On means the display is on with the maximum brightness set by the R_{SET} resistor. CTRL1 = 1 means a constant logic level (V_{CC}) is applied to CTRL1.

² Main Display Off means the main display only is off. CTRL2 = 0 means a constant logic level (GND) is applied to CTRL2.

³ Main Display On means the display is on with the maximum brightness set by the R_{SET} resistor. CTRL2 = 1 means a constant logic level (V_{CC}) is applied to CTRL2.

⁴ Sub Display Off means the sub display LEDs only is off. CTRL1 = 0 means a constant logic level (GND) is applied to CTRL1.

⁵ PWM means a digital PWM signal is applied to the CTRL1 and/or the CTRL2 pin with a frequency from 100 Hz to 200 kHz.

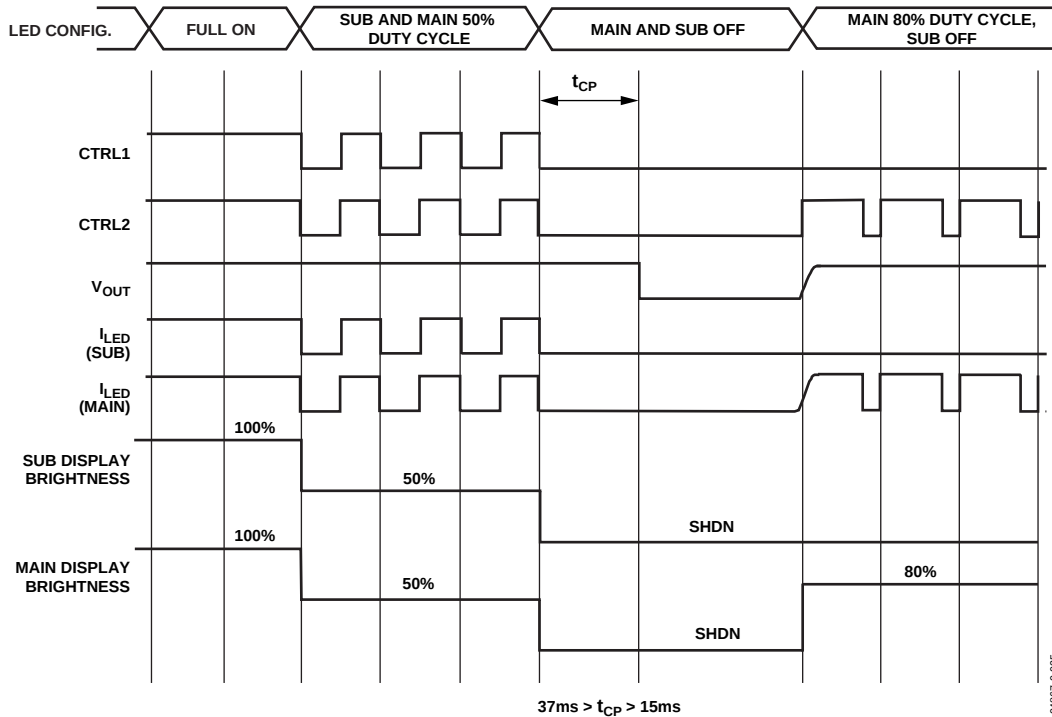


Figure 25. Application Timing

LED BRIGHTNESS CONTROL USING A PWM SIGNAL APPLIED TO V_{PWM}

Adding two external resistors and a capacitor, as shown on Figure 26, also can be used to control PWM brightness. This PWM brightness control method can be used instead of CTRL1 and/or CTRL2 digital PWM brightness control. With this configuration, The CTRL1 and CTRL2 digital logic pins can be used to control shutdown of the white LEDs, while V_{PWM} can be used to control the brightness of all the white LEDs by applying a high frequency PWM signal (amplitude 0 V to 2.5 V) to drive an R-C-R filter on the I_{SET} pin of the ADM8845. A 0% PWM duty cycle corresponds to 20 mA/LED, while a 100% PWM duty cycle corresponds to a 0 mA/LED. At PWM frequencies above 5 kHz, C5 may be reduced (see Figure 26). To have 20 mA flowing in each LED, the amplitude of the PWM signal must be 0 V and 2.5 V only.

$$I_{LED} = \frac{I_{SET} - \text{Voltage}}{R_{SET} \times 2R} \times 120 \times \frac{(1 - \text{Duty Cycle})}{100}$$

$$R_{SET} + 2R$$

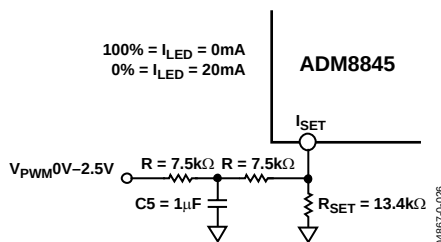


Figure 26. PWM Brightness Control Using Filtered-PWM Signal

LED BRIGHTNESS CONTROL USING A DC VOLTAGE APPLIED TO V_{BRIGHT}

Adding one resistor, as shown in Figure 22, this configuration can also be used to control brightness the white LEDs by using a dc voltage applied to the V_{BRIGHT} node. Figure 27 shows an application example of LED brightness control using a dc voltage with a amplitude of 0 V to 2.5 V, applied to V_{BRIGHT}.

The equation for I_{LED} is

$$I_{SET} = [(1/R_{SET} + 1/R)(V_{SET})] - [(1/R)(V_{BRIGHT})]$$

$$I_{LED} = 120 \times I_{SET}$$

where R = 15 kΩ and V_{SET} = voltage at I_{SET} pin (1.18 V).

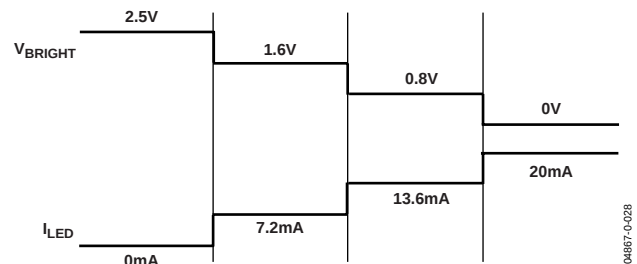


Figure 27. PWM Brightness Control Application Diagram Using a DC Voltage Applied to V_{BRIGHT}

APPLICATIONS

LAYOUT CONSIDERATIONS AND NOISE

Because of the ADM8845's switching behavior, PCB trace layout is an important consideration. To ensure optimum performance, a ground plane should be used, and all capacitors (C1, C2, C3, C4) must be located with minimal track lengths to the pins of the ADM8845.

WHITE LED SHORTING

If an LED is shorted, the ADM8845 continues to drive the remaining LEDs with I_{LED} per LED ($I_{LED} = I_{SET} \times 120 \text{ mA}$). This is because the ADM8845 uses six internal current sinks to produce the LED current. If an LED is shorted, the ADM8845 continues to sink ($I_{SET} \times 120 \text{ mA}$) as programmed by R_{SET} through the shorted LED.

DRIVING FEWER THAN SIX LEDs

The ADM8845 can be operated with fewer than six LEDs in parallel by simply leaving the unused FBx pins floating or connected to GND. For example, Figure 28 shows five LEDs being powered by the ADM8845, and Figure 29 shows three main LEDs and one sub LED.

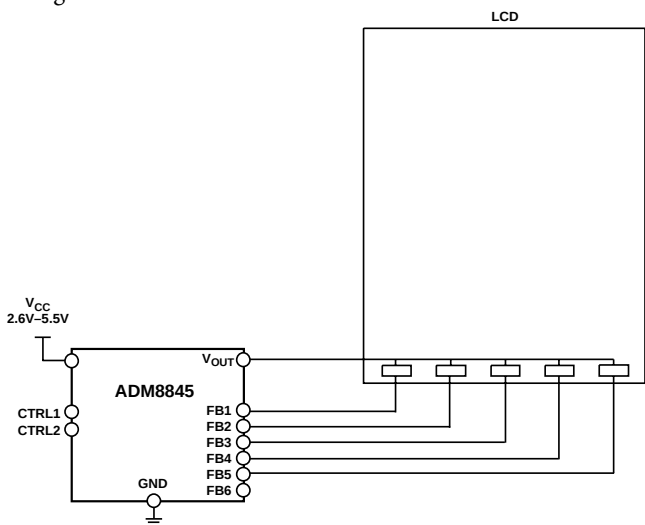


Figure 28. Driving Five White LEDs

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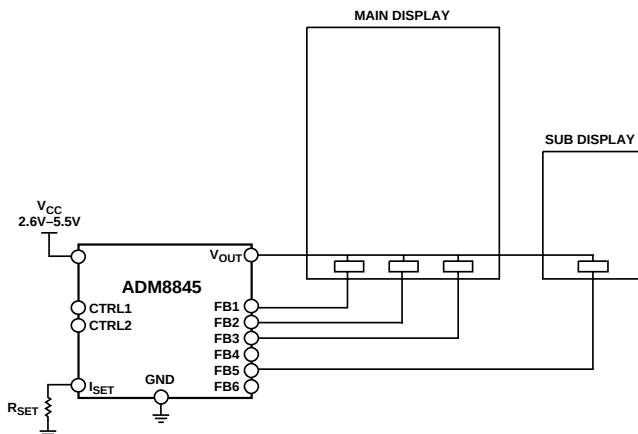


Figure 29. Driving Three Main LEDs and One Sub LED

041877-0-030

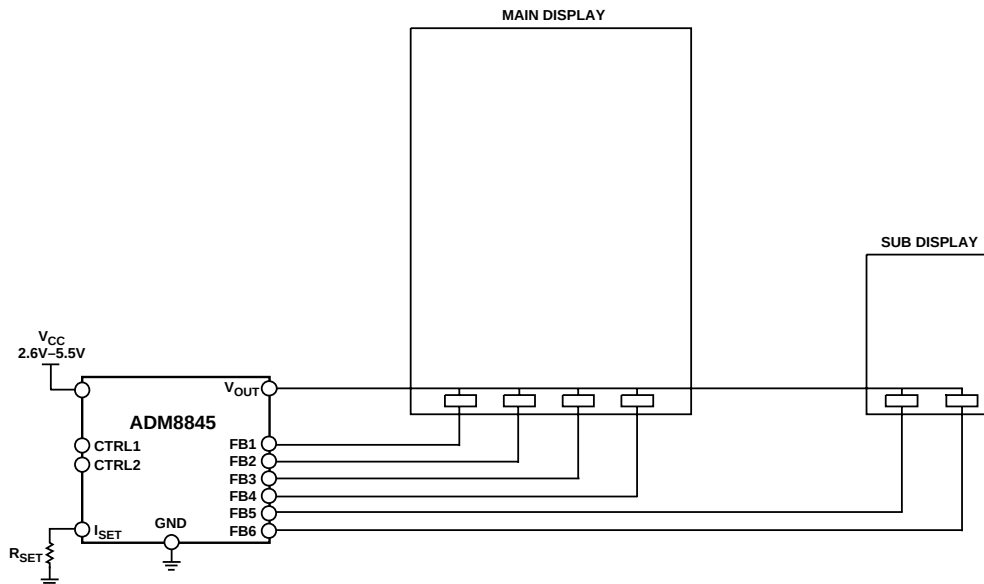


Figure 30. Typical Application Diagram

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DRIVING FLASH LEDS

The ADM8845 can be operated with any two FBx pins operated in parallel to double the combined LED current supplied by the ADM8845. For example, if three flash LEDs need to be driven with 60 mA/LED, the ADM8845 can be configured as in Figure 31 (see also Figure 21).

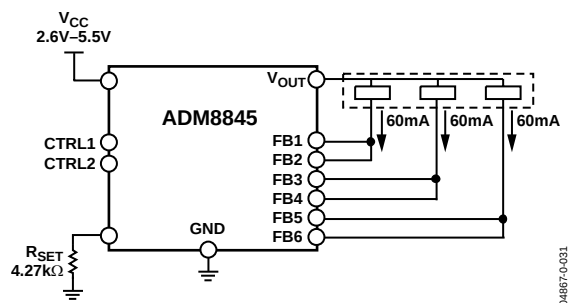


Figure 31. Driving Three Flash LEDs

DRIVING CAMERA LIGHT, MAIN, AND SUB LEDS

The ADM8845 can also be configured to power a camera light that is composed of four white LEDs in parallel, packaged into one package. FB1 to FB4 now power the camera light, and FB5 and FB6 power the main display. The sub display LED is powered from the ADM8845 by using an external current mirror to control the current flowing through the sub white LED (see Figure 32). All white LEDs have 15 mA/LED, therefore total load on the ADM8845 charge pump is 105 mA, and the maximum load on the ADM8845 charge pump is 180 mA (see Figure 21).

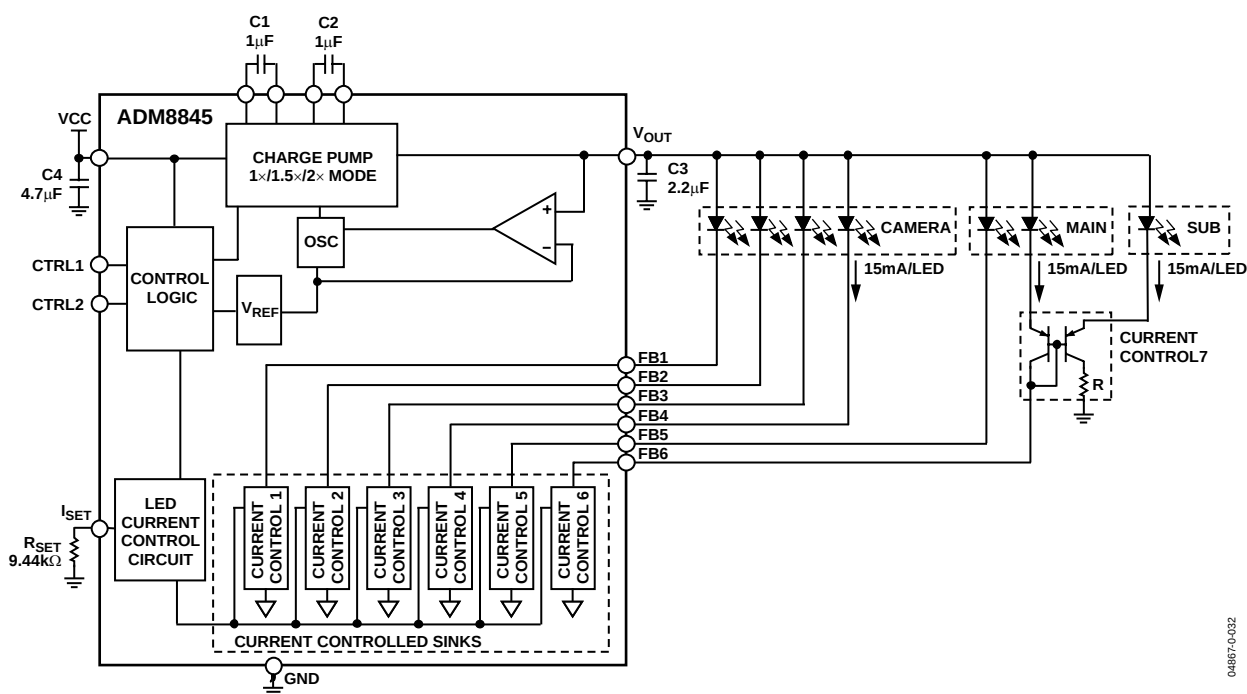


Figure 32. Driving Camera Light, Two Main LEDs, and One Sub LED

DRIVING FOUR BACKLIGHT WHITE LEDs AND FLASH LEDs

The ADM8845 also can be configured to power four backlight white LEDs and a camera flash, packaged into one package. FB1 to FB4 power the backlight light, FB5 and FB6 powers the two of the flash LEDs, and the third is powered an external current mirror to control the current flowing through the third flash LED (see Figure 33). All the backlight white LEDs have 15 mA/LED, and the flash current is 20 mA/LED. The total load on the ADM8845 charge pump is 120 mA; the maximum load on the ADM8845 charge pump is 180 mA (see Figure 21).

CTRL1 controls the flash on/off, and CTRL2 controls the backlight on/off and brightness control. Because the R_{SET} resistor sets the current that each of the six current control blocks can sink, a PWM signal is used to change the current in the backlight from 20 mA to 5 mA/LED. The CTRL2 duty cycle is 15/20 to give 15 mA/backlight LED.

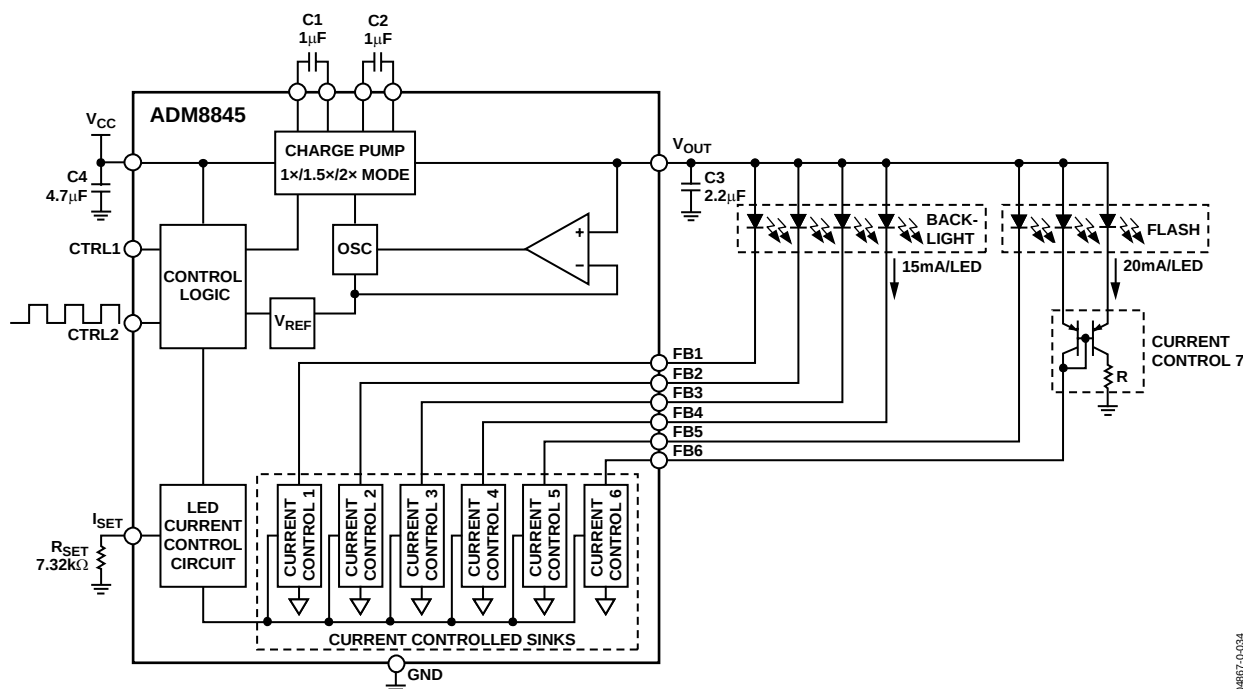


Figure 33. Driving Four Backlight LEDs and Flash LED

04867-0-034

POWER EFFICIENCY

The ADM8845 power efficiency (η) equations are

$$\eta = P_{OUT}/P_{IN}$$

$$P_{IN} = ((V_{CC} \times I_{LOAD} \times Gain) + (I_Q \times V_{CC}))$$

$$P_{OUT} = 6 \times (V_F \times I_{LED})$$

where:

I_Q is the quiescent current of the ADM8845, 2.6 mA.

V_F is the LED forward voltage.

$Gain$ is equal to charge pump mode (1 $\times$, 1.5 $\times$, 2 $\times$).

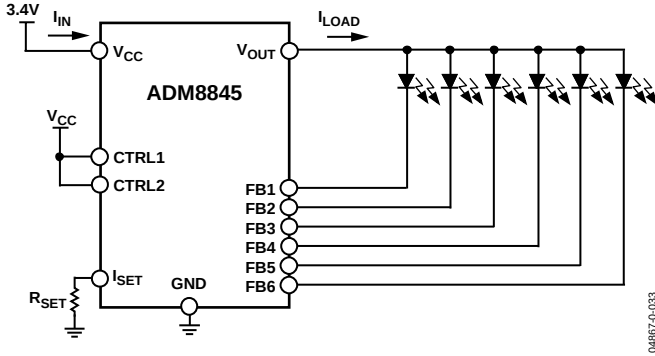


Figure 34. Charge Pump Power Efficiency Diagram

Examples 1 and 2 show calculations of the ADM8845 power efficiency; also see Figure 34.

Example 1

The ADM8845 driving six white LED with 20 mA/LED at $V_{CC} = 3.4$ V (1.5 $\times$ mode), LED $V_F = 4.5$ V.

$$P_{IN} = ((V_{CC} \times I_{LOAD} \times Gain) + (V_{CC} \times I_Q))$$

$$P_{IN} = ((3.4 \times 120 \text{ mA} \times 1.5) + (3.4 \times 2.6 \text{ mA}))$$

$$P_{IN} = ((0.612) + (0.00884))$$

$$P_{IN} = 0.62084$$

$$P_{OUT} = 6(V_F \times I_{LED})$$

$$P_{OUT} = 6(4.5 \text{ V} \times 20 \text{ mA})$$

$$P_{OUT} = 0.54$$

$$\eta = P_{OUT}/P_{IN}$$

$$\eta = 0.54/0.62084$$

$$\eta = 87 \%$$

Example 2

The ADM8845 driving six white LED with 20 mA/LED at $V_{CC} = 3.4$ (1.5 $\times$ mode), LED $V_F = 3.6$ V.

$$P_{IN} = ((V_{CC} \times I_{LOAD} \times Gain) + (V_{CC} \times I_Q))$$

$$P_{IN} = ((3.4 \times 120 \text{ mA} \times 1.5) + (3.4 \times 2.6 \text{ mA}))$$

$$P_{IN} = ((0.612) + (0.00884))$$

$$P_{IN} = 0.62084$$

$$P_{OUT} = 6(V_F \times I_{LED})$$

$$P_{OUT} = 6(3.6 \text{ V} \times 20 \text{ mA})$$

$$P_{OUT} = 0.432$$

$$\eta = P_{OUT}/P_{IN}$$

$$\eta = 0.432/0.62084$$

$$\eta = 70 \%$$

OUTLINE DIMENSIONS

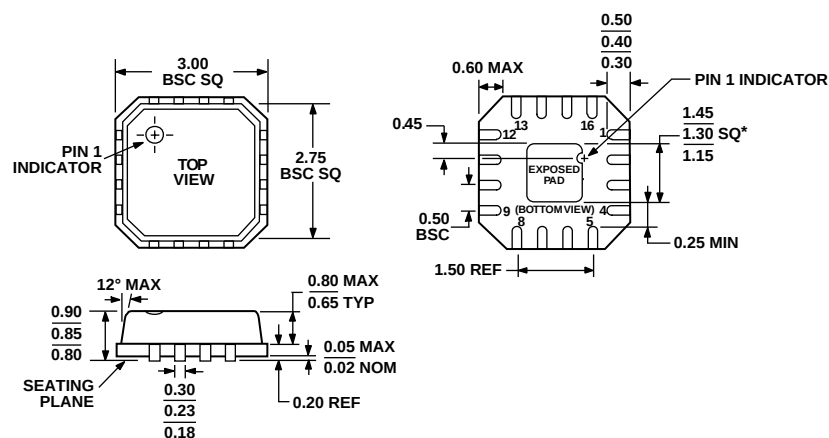


Figure 35. 16-Lead Lead Frame Chip Scale Package [LFCSP]
(CP-16)

Dimensions shown in millimeters

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option	Branding
ADM8845ACP-REEL	-40°C to + 85°C	16-Lead LFCSP	CP-16	MOP
ADM8845ACP-REEL7	-40°C to + 85°C	16-Lead LFCSP	CP-16	MOP
ADM8845ACPZ-REEL ¹	-40°C to + 85°C	16-Lead LFCSP	CP-16	MOP
ADM8845ACPZ-REEL7 ¹	-40°C to + 85°C	16-Lead LFCSP	CP-16	MOP
ADM8845ACPZ-WP ^{1,2}	-40°C to + 85°C	16-Lead LFCSP	CP-16	MOP
EVAL-ADM8845EB		Evaluation Board		

¹ Z = Pb-free part.

² WP = waffle pack.

NOTES

ADM8845

NOTES