

**HYBRID VOLTAGE REGULATORS** CJCA001 CJCA002 CJCA007 CJCA008

# VARIABLE OUTPUT HYBRID VOLTAGE REGULATORS

± 8V TO ± 56V

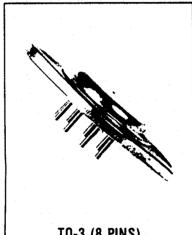
## **5 AMPERES**

#### **FFATURES**

POSITIVE. NEGATIVE SUPPLY OPERATION **5A CURRENT RATING 60V LINE VOLTAGE CAPABILITY** LINE AND LOAD REGULATION OUTPUT VOLTAGE ADJUSTED FROM 8 TO 56 VOLTS BY A RESISTOR DIVIDED NETWORK. ALL DEVICES ARE PLANAR OXIDE PASSIVATED. BERILLIA CERAMIC ISOLATION PADS.

#### **APPLICATIONS**

- INDUSTRIAL CONTROLS
- INSTRUMENTATION
- COMPUTERS, DATA TERMINALS
- MILITARY EQUIPMENT, SPACE AND TELECOMMUNICATIONS
- DISTRIBUTED POWER SYSTEMS



TO-3 (8 PINS)

#### **MAXIMUM RATINGS**

MEDICAL ELECTRONICS

DC MOTOR SUPPLIES

•		CJCA001 CJCA002 CJCA007 CJCA008
± V <sub>in</sub>	INPUT VOLTAGE (FIGS. 2, 3 PINS 1-7)	60 V
lopk	PEAK LOAD CURRENT (FIGS. 2, 3 PIN 2)	5 A
TA	OPERATING TEMPERATURE	55°C to +150°C
T <sub>stg</sub>	STORAGE TEMPERATURE	55°C to +150°C
R <sub>Ø</sub> JC	THERMAL RESISTANCE, JUNCTION TO CASE	3°C/W
PD	POWER DISSIPATION (100°C)	50 W

REGULATORS	CJCA	001	002	007	008	UNITS
Output Voltage Range		+8 to +56	-8 to -56	-	-	V <sub>dc</sub>
with FET Internal Current Source			-	+8 to +56	-8 to -56	V <sub>dc</sub>

CJCA 007 and 008 incorporate a FET constant current source, which provides current mode regulation automatically. A minimum inputoutput voltage differential of 4 volts is recommended to bias the FET into its constant current region. At lower voltages the FET becomes resistive, and regulation reverts back to basic mode.



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**ELECTRICAL CHARACTERISTICS**  $(T_C = 25^{\circ}C. |\pm V_{IM}| = 28V. |\pm V_{Q}| = 20V. |\pm I_{Q}| = 14$  Unless otherwise moted)

CHARACTERISTICS			SYMBOL	MIN.	MAX.	UNITS
INPUT VOLTAGE			±V <sub>in</sub>			
	CJCA001	CJCA002		10	60	v
	CJCA007	CJCA008		12	60	v
OUTPUT VOLTAGE RANGE			± V <sub>0</sub>	8	56	v
INPUT-OUTPUT DIFFERENTIAL			A V			
	CJCA001	CJCA002		2	50	v
	CJCA007	CJCA008		4	50	٧
LOAD REGULATION			Δ V <sub>0</sub>			
$(I_0 = 0A-5.0A)$			V <sub>o</sub>		0.5	%
LINE REGULATION			Δ V <sub>o</sub>			
$( V_{i0}  \pm 20\%, I_0 = 1.0A)$	CJCA001	CJCA002	V <sub>o</sub>		2.0	%
	CJCA007	CJCA008			0.5	%
RIPPLE ATTENUATION						
(fig. 11, fig. 12)	CJCA001	CJCA002		32		db
	CJCA007	CJCA008		62		db
TEMPERATURE COEFFICIENT			ΔV <sub>o</sub>			
$(-55^{\circ}C \le T_{A} \le + 125^{\circ}C)$			V <sub>o Δ</sub> T		± 0.02	%/°C

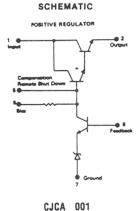


FIGURE 1

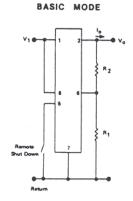


FIGURE 2

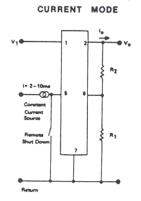
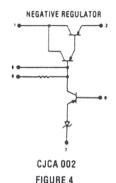


FIGURE 3

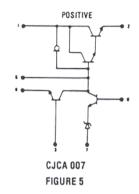


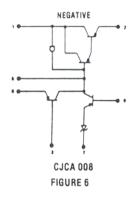
# HYBRID VOLTAGE REGULATORS CJCA001 CJCA002 CJCA007 CJCA008

#### REGULATORS WITH FET INTERNAL CURRENT SOURCE AND LIMIT



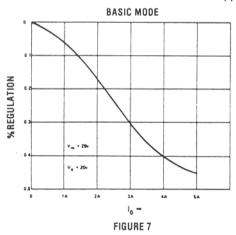
Hook up per Fig. 2 and 3, but  $V_1$  and  $V_0$  are negative voltage

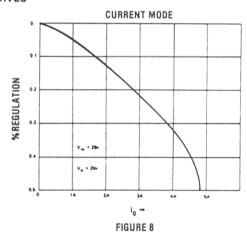




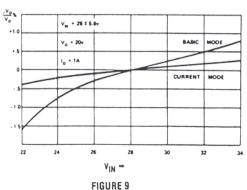
## LOAD REGULATION

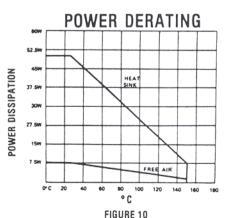
#### TYPICAL CURVES





## LINE REGULATION







#### **HYBRID VOLTAGE REGULATORS** CJCA002 CJCA007 CJCA008 CJCA001

## RIPPLE ATTENUATION

BASIC MODE

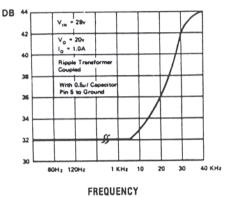
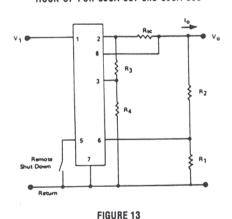


FIGURE 11

HOOK UP FOR CJCA 007 and CJCA 008



See note 2 for above application

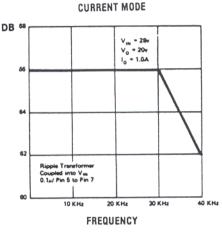


FIGURE 12

# D.C. SAFE OPERATING AREA FOR PASS TRANSISTORS

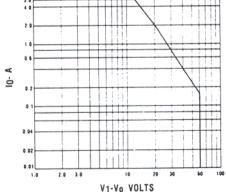


FIGURE 14

#### NOTES:

1. Reference diode is 6.2  $\pm$  .3 volts, and the VBE of the feedback transistor is 0.6  $\pm$  .1 volt, giving  $V_{6-7}=\,6.4$  to 7.2 volts, temperature compensated. External divider  $R_1$  and  $R_2$  in Figs. 2, 3 and 13 determines the output voltage Vo.

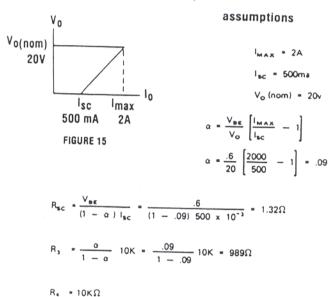
$$V_0 = V_{6-7} \left( \frac{R_1 + R_2}{R_1} \right) \approx 7v \left( \frac{R_1 + R_2}{R_1} \right)$$



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- 1. (cont'd.) Recommended maximum value for R<sub>1</sub> is  $7K\Omega(I_{min.}=1ma)$ , typical usage being  $1K\Omega$  to  $5K\Omega$ . R<sub>2</sub> must be determined experimentally for each regulator for a particular value of V<sub>0</sub>, due to the variations in V<sub>6-7</sub>.
- 2. Current limiting capability is provided on the CJCA 007 and CJCA 008. Rsc may be selected to limit  $I_0$  and  $I_{SC}$  as shown in Fig. 13. The user must be sure that the dissipation rating is not exceeded in worst-case operation (see Fig. 10). Also, for  $V_1 V_0$  (=  $V_1$  for a shorted load) greater than 20 volts, a further limitaton is second breakdown of the pass transistor (see Fig. 14).

#### Example:



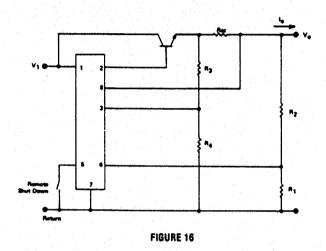
Resistors  ${\rm R}_1$  and  ${\rm R}_2$  are found from note 1.

- Current limiting for CJCA 001 and CJCA 002 can be provided by adding a NPN(PNP) transistor and resistors hooked-up as shown in Figs. 5, 6 and 13. Resistor values are found in notes 1 and 2.
- 4. Oscillaton suppression may be best obtained if needed by connecting a capacitor from pin 5 to pin 7; 0.01uf is usually sufficent. A capacitor at the input (pin 1 to 7) or output (pin 2 to 7) may also be helpful. When using a capacitor on the ouput, and not current limiting, care must be taken that the capacitance is not so large that an excessive turn-on surge current develops, damaging the pass transistor.
- Inductive surges on the output may be suppressed by adding a reverse-biased diode on the output returned to ground.

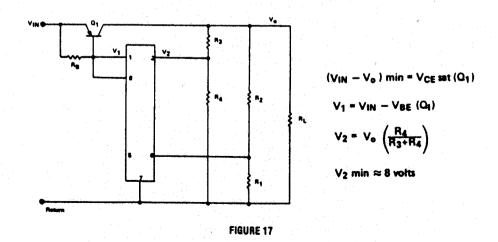


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6. Output current and power capability may be increased by driving an external power pass transistor. Figure 16 below illustrates for the regulator CJCA 007. For the negative regulators, reverse the polarities and use external pnp pass transistor. Be sure to maintain safe operating area conditions for both the regulator and the external transistors. All resistor values are found in notes 1 and 2 and Figs. 5, 6 and 13.



 Output current and power may be increased, and input-output voltage differential reduced, by the use of one or more external pnp pass transistors for the positive regulators. Figure 17 below illustrates for CJCA 001.



Thus, Vo is not restricted by the input-output differential of the regulator, which is  $V_1 - V_2$ .



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#### 8. Regulator Failures

- a. Regulator failures are caused by overdissipating the series pass transitor. Excessive heating in the pass transistor causes it to short out. A good heat sink must be used. Highest power dissipation occurs when the regulator output is shorted. Foldback current limiting should be used giving less power dissipation than at full load.
- b. Use conservatively voltage rated capacitors on the input to the regulator in order to minimize ripple. If the input capacitor is operated with excessive ripple and near the maximim dc voltage rating, the capacitor will sputter (short momentarily) causing the output capacitor of the regulator to discharge back through the reverse-biased pass transistor or the control circuitry with destructive effects.
- c. Avoid severe voltage transients on the unregulated input. Subsequent transients can destroy the regulator because of load failures due to previous transients.