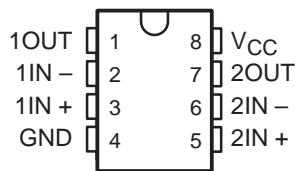


TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

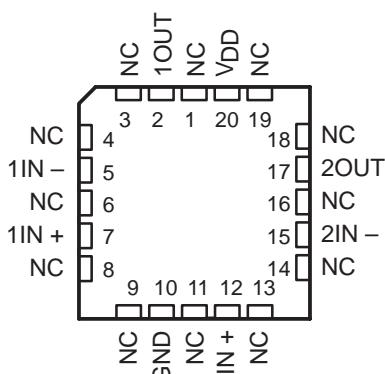
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- Trimmed Offset Voltage: TLC27M7 . . . 500 μ V Max at 25°C, $V_{DD} = 5$ V
- Input Offset Voltage Drift . . . Typically 0.1 μ V/Month, Including the First 30 Days
- Wide Range of Supply Voltages Over Specified Temperature Ranges:
0°C to 70°C . . . 3 V to 16 V
-40°C to 85°C . . . 4 V to 16 V
-55°C to 125°C . . . 4 V to 16 V
- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)
- Low Noise . . . Typically 32 nV/ $\sqrt{\text{Hz}}$ at $f = 1$ kHz
- Low Power . . . Typically 2.1 mW at 25°C, $V_{DD} = 5$ V
- Output Voltage Range Includes Negative Rail
- High Input impedance . . . 10^{12} Ω Typ
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latch-Up Immunity

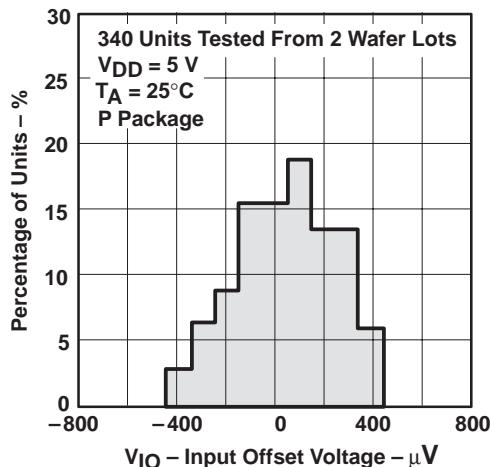
**D, JG, P OR PW PACKAGE
(TOP VIEW)**



**FK PACKAGE
(TOP VIEW)**



**DISTRIBUTION OF TLC27M7
INPUT OFFSET VOLTAGE**



AVAILABLE OPTIONS

T_A	V_{IO}^{max} AT 25°C	PACKAGE				
		SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	TSSOP (PW)
0°C to 70°C	500 μ V	TLC27M7CD	—	—	TLC27M7CP	—
	2 mV	TLC27M2BCD	—	—	TLC27M2BCP	—
	5 mV	TLC27M2ACD	—	—	TLC27M2ACP	—
	10 mV	TLC27M2CD	—	—	TLC27M2CP	TLC27M2CPW
-40°C to 85°C	500 μ V	TLC27M7ID	—	—	TLC27M7IP	—
	2 mV	TLC27M2BID	—	—	TLC27M2BIP	—
	5 mV	TLC27M2AID	—	—	TLC27M2AIP	—
	10 mV	TLC27M2ID	—	—	TLC27M2IP	TLC27M2IPW
-55°C to 125°C	500 μ V	TLC27M7MD	TLC27M7MFK	TLC27M7MJG	TLC27M7MP	—
	10 mV	TLC27M2MD	TLC27M2MFK	TLC27M2MJG	TLC27M2MP	—

The D and PW package is available taped and reeled. Add R suffix to the device type (e.g., TLC27M7CDR).

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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



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TLC27M2, TLC27M2A, TLC27M2B, TLC27M7

LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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description

The TLC27M2 and TLC27M7 dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these cost-effective devices ideal for applications which have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M2 (10 mV) to the high-precision TLC27M7 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M2 and TLC27M7. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up.

The TLC27M2 and TLC27M7 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C.

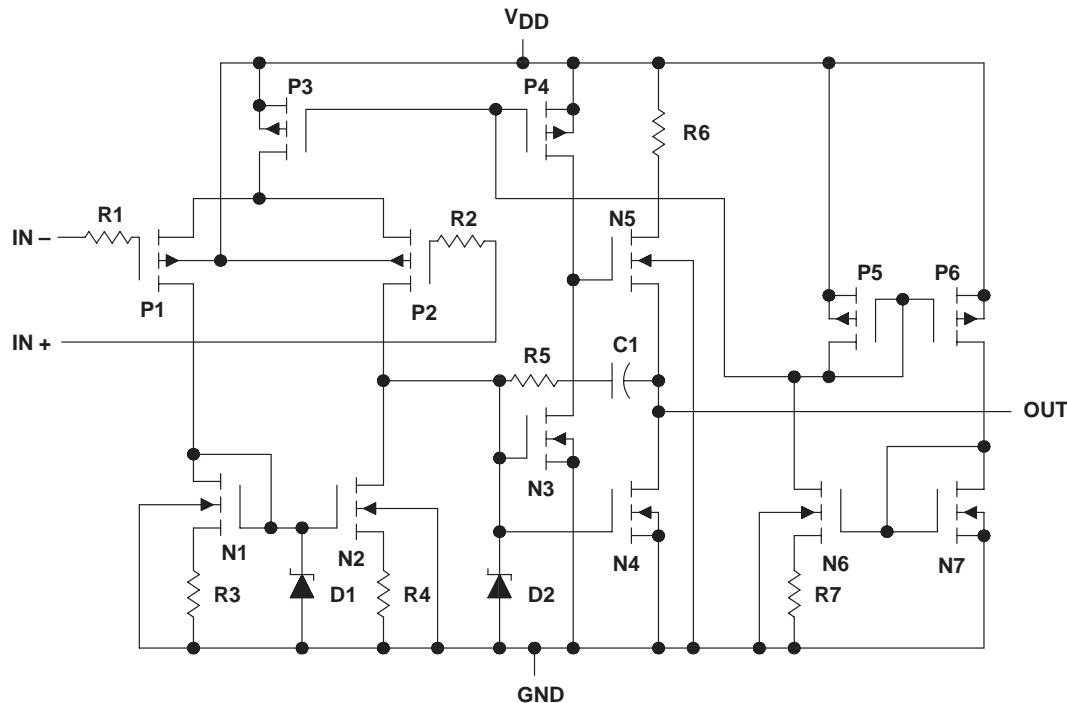


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TLV27M2, TLV27M2A, TLV27M2B, TLV27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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equivalent schematic (each amplifier)



TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage, V_{ID} (see Note 2)	$\pm V_{DD}$
Input voltage range, V_I (any input)	-0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O (each output)	± 30 mA
Total current into V_{DD}	45 mA
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : C suffix	0°C to 70°C
I suffix	-40°C to 85°C
M suffix	-55°C to 125°C
Storage temperature range	-65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.

1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at IN+ with respect to IN-.
3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	

recommended operating conditions

	C SUFFIX		I SUFFIX		M SUFFIX		UNIT	
	MIN	MAX	MIN	MAX	MIN	MAX		
Supply voltage, V_{DD}	3	16	4	16	4	16	V	
Common-mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$		-0.2	3.5	-0.2	3.5	0	3.5
	$V_{DD} = 10\text{ V}$		-0.2	8.5	-0.2	8.5	0	8.5
Operating free-air temperature, T_A	0	70	-40	85	-55	125	°C	

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
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electrical characteristics at specified free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A †	TLC27M2C			UNIT
				MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M2C	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$,	25°C	1.1	10
			$R_L = 100 \text{ k}\Omega$		Full range		12
		TLC27M2AC	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$,	25°C	0.9	5
			$R_L = 100 \text{ k}\Omega$		Full range		6.5
	TLC27M2BC	TLC27M2BC	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$,	25°C	220	2000
			$R_L = 100 \text{ k}\Omega$		Full range		3000
		TLC27M7C	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$,	25°C	185	500
			$R_L = 100 \text{ k}\Omega$		Full range		1500
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		1.7	$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5$ V,	$V_{IC} = 2.5$ V	25°C		0.1	pA
				70°C		7 300	
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5$ V,	$V_{IC} = 2.5$ V	25°C		0.6	pA
				70°C		40 600	
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2 to 4	-0.3 to 4.2	V
				Full range	-0.2 to 3.5		
V_{OH}	High-level output voltage	$V_{ID} = 100 \text{ mV}$,	$R_L = 100 \text{ k}\Omega$	25°C	3.2	3.9	V
				0°C	3	3.9	
				70°C	3	4	
V_{OL}	Low-level output voltage	$V_{ID} = -100 \text{ mV}$,	$I_{OL} = 0$	25°C	0	50	mV
				0°C	0	50	
				70°C	0	50	
AVD	Large-signal differential voltage amplification	$V_O = 0.25 \text{ V to } 2 \text{ V}$,	$R_L = 100 \text{ k}\Omega$	25°C	25	170	V/mV
				0°C	15	200	
				70°C	15	140	
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICR\text{min}}$		25°C	65	91	dB
				0°C	60	91	
				70°C	60	92	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5 \text{ V to } 10 \text{ V}$,	$V_O = 1.4 \text{ V}$	25°C	70	93	dB
				0°C	60	92	
				70°C	60	94	
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5 \text{ V}$,	$V_{IC} = 2.5 \text{ V}$, No load	25°C	210	560	μA
				0°C	250	640	
				70°C	170	440	

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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electrical characteristics at specified free-air temperature, $V_{DD} = 10$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS		T_A^\dagger	TLC27M2C			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M2C	$V_O = 1.4$ V, $R_S = 50$ Ω ,	25°C	1.1	10		mV
			$V_{IC} = 0$, $R_L = 100$ k Ω	Full range		12		
		TLC27M2AC	$V_O = 1.4$ V, $R_S = 50$ Ω ,	25°C	0.9	5		μ V
			$R_L = 100$ k Ω	Full range		6.5		
	TLC27M2BC	TLC27M2BC	$V_O = 1.4$ V, $R_S = 50$ Ω ,	25°C	224	2000		μ V
			$R_L = 100$ k Ω	Full range		3000		
		TLC27M7C	$V_O = 1.4$ V, $R_S = 50$ Ω ,	25°C	190	800		
			$R_L = 100$ k Ω	Full range		1900		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C		2.1		μ V/°C
I_{IO}	Input offset current (see Note 4)	$V_O = 5$ V,	$V_{IC} = 5$ V	25°C	0.1			pA
				70°C	7	300		
I_{IB}	Input bias current (see Note 4)	$V_O = 5$ V,	$V_{IC} = 5$ V	25°C	0.7			pA
				70°C	50	600		
V_{ICR}	Common-mode input voltage range (see Note 5)			25°C	-0.2	-0.3		V
					to 9	to 9.2		
V_{OH}	High-level output voltage	$V_{ID} = 100$ mV,	$R_L = 100$ k Ω	25°C	8	8.7		V
				0°C	7.8	8.7		
				70°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100$ mV,	$I_{OL} = 0$	25°C	0	50		mV
				0°C	0	50		
				70°C	0	50		
AVD	Large-signal differential voltage amplification	$V_O = 1$ V to 6 V,	$R_L = 100$ k Ω	25°C	25	275		V/mV
				0°C	15	320		
				70°C	15	230		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICR\min}$		25°C	65	94		dB
				0°C	60	94		
				70°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5$ V to 10 V,	$V_O = 1.4$ V	25°C	70	93		dB
				0°C	60	92		
				70°C	60	94		
I_{DD}	Supply current (two amplifiers)	$V_O = 5$ V, No load	$V_{IC} = 5$ V,	25°C	285	600		μ A
				0°C	345	800		
				70°C	220	560		

† Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
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electrical characteristics at specified free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A †	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT	
				MIN	TYP	MAX		
V_{IO}	Input offset voltage	TLC27M2I	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$, $R_L = 100$ k Ω	25°C	1.1	10	mV
					Full range		13	
		TLC27M2AI	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$, $R_L = 100$ k Ω	25°C	0.9	5	
					Full range		7	
		TLC27M2BI	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$, $R_L = 100$ k Ω	25°C	220	2000	μ V
					Full range		3500	
		TLC27M7I	$V_O = 1.4$ V, $R_S = 50 \Omega$,	$V_{IC} = 0$, $R_L = 100$ k Ω	25°C	185	500	
					Full range		2000	
α_{VIO}	Average temperature coefficient of input offset voltage				25°C to 85°C		1.7	μ V/°C
I_{IO}	Input offset current (see Note 4)		$V_O = 2.5$ V,	$V_{IC} = 2.5$ V	25°C	0.1		pA
					85°C	24	1000	
I_{IB}	Input bias current (see Note 4)		$V_O = 2.5$ V,	$V_{IC} = 2.5$ V	25°C	0.6		pA
					85°C	200	2000	
V_{ICR}	Common-mode input voltage range (see Note 5)				25°C	-0.2 to 4	-0.3 to 4.2	V
					Full range	-0.2 to 3.5		
V_{OH}	High-level output voltage		$V_{ID} = 100$ mV,	$R_L = 100$ k Ω	25°C	3.2	3.9	V
					-40°C	3	3.9	
					85°C	3	4	
V_{OL}	Low-level output voltage		$V_{ID} = -100$ mV,	$I_{OL} = 0$	25°C	0	50	mV
					-40°C	0	50	
					85°C	0	50	
AVD	Large-signal differential voltage amplification		$V_O = 0.25$ V to 2 V,	$R_L = 100$ k Ω	25°C	25	170	V/mV
					-40°C	15	270	
					85°C	15	130	
$CMRR$	Common-mode rejection ratio		$V_{IC} = V_{ICR\min}$		25°C	65	91	dB
					-40°C	60	90	
					85°C	60	90	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)		$V_{DD} = 5$ V to 10 V,	$V_O = 1.4$ V	25°C	70	93	dB
					-40°C	60	91	
					85°C	60	94	
I_{DD}	Supply current (two amplifiers)		$V_O = 2.5$ V,	$V_{IC} = 2.5$ V, No load	25°C	210	560	μ A
					-40°C	315	800	
					85°C	160	400	

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7 LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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electrical characteristics at specified free-air temperature, $V_{DD} = 10$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	TLC27M2I $V_O = 1.4$ V, $R_S = 50$ Ω ,	$25^\circ C$	1.1	10		mV
		Full range		13		
	TLC27M2AI $V_O = 1.4$ V, $R_S = 50$ Ω ,	$25^\circ C$	0.9	5		μ V
		Full range		7		
	TLC27M2BI $V_O = 1.4$ V, $R_S = 50$ Ω ,	$25^\circ C$	224	2000		
		Full range		3500		
	TLC27M7I $V_O = 1.4$ V, $R_S = 50$ Ω ,	$25^\circ C$	190	800		
		Full range		2900		
α_{VIO}	Average temperature coefficient of input offset voltage	25°C to 85°C		2.1		μ V/°C
I_{IO}	Input offset current (see Note 4)	25°C	0.1			pA
		85°C	26	1000		
I_{IB}	Input bias current (see Note 4)	25°C	0.7			pA
		85°C	220	200	0	
V_{ICR}	Common-mode input voltage range (see Note 5)	25°C	-0.2	-0.3		V
		to 9	to 9.2			
		Full range	-0.2			V
			to 8.5			
V_{OH}	High-level output voltage	25°C	8	8.7		V
		-40°C	7.8	8.7		
		85°C	7.8	8.7		
V_{OL}	Low-level output voltage	25°C	0	50		mV
		-40°C	0	50		
		85°C	0	50		
A_{VD}	Large-signal differential voltage amplification	25°C	25	275		V/mV
		-40°C	15	390		
		85°C	15	220		
$CMRR$	Common-mode rejection ratio	25°C	65	94		dB
		-40°C	60	93		
		85°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	25°C	70	93		dB
		-40°C	60	91		
		85°C	60	94		
I_{DD}	Supply current	25°C	285	600		μ A
		-40°C	450	900		
		85°C	205	520		

† Full range is -40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7
LinCMOS™ PRECISION DUAL OPERATIONAL AMPLIFIERS

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electrical characteristics at specified free-air temperature, $V_{DD} = 5$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A †	TLC27M2M TLC27M7M			UNIT
				MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M2M $V_O = 1.4$ V, $R_S = 50$ Ω , $V_{IC} = 0$, $R_L = 100$ k Ω	25°C	1.1	10	12	mV
			Full range				
	TLC27M7M $V_O = 1.4$ V, $R_S = 50$ Ω , $V_{IC} = 0$, $R_L = 100$ k Ω	25°C	185	500	3750		
		Full range					
αV_{IO}	Average temperature coefficient of input offset voltage		25°C to 125°C	1.7			$\mu\text{V}/^\circ\text{C}$
I_{IO}	Input offset current (see Note 4)	$V_O = 2.5$ V, $V_{IC} = 2.5$ V	25°C	0.1			pA
			125°C	1.4	15		nA
I_{IB}	Input bias current (see Note 4)	$V_O = 2.5$ V, $V_{IC} = 2.5$ V	25°C	0.6			pA
			125°C	9	35		nA
V_{ICR}	Common-mode input voltage range (see Note 5)		25°C	0	−0.3		V
			to	to			
			4	4.2			
			Full range	0			V
V_{OH}	High-level output voltage	$V_{ID} = 100$ mV, $R_L = 100$ k Ω	25°C	3.2	3.9		V
			−55°C	3	3.9		
			125°C	3	4		
V_{OL}	Low-level output voltage	$V_{ID} = −100$ mV, $I_{OL} = 0$	25°C	0	50		mV
			−55°C	0	50		
			125°C	0	50		
AVD	Large-signal differential voltage amplification	$V_O = 0.25$ V to 2 V, $R_L = 100$ k Ω	25°C	25	170		V/mV
			−55°C	15	290		
			125°C	15	120		
$CMRR$	Common-mode rejection ratio	$V_{IC} = V_{ICR\min}$	25°C	65	91		dB
			−55°C	60	89		
			125°C	60	91		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5$ V to 10 V, $V_O = 1.4$ V	25°C	70	93		dB
			−55°C	60	91		
			125°C	60	94		
I_{DD}	Supply current (two amplifiers)	$V_O = 2.5$ V, $V_{IC} = 2.5$ V, No load	25°C	210	560		μA
			−55°C	340	880		
			125°C	140	360		

† Full range is −55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

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electrical characteristics at specified free-air temperature, $V_{DD} = 10$ V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A^\dagger	TLC27M2M TLC27M7M			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	TLC27M2M $V_O = 1.4$ V, $R_S = 50$ Ω , $R_L = 100$ k Ω	25°C	1.1	10	12	mV
		Full range				
	TLC27M7M $V_O = 1.4$ V, $R_S = 50$ Ω , $R_L = 100$ k Ω	25°C	190	800	4300	
		Full range				
α_{VIO} Average temperature coefficient of input offset voltage		25°C to 125°C		2.1		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current (see Note 4)	$V_O = 5$ V, $V_{IC} = 5$ V	25°C	0.1			pA
		125°C	1.8	15		
	$V_O = 5$ V, $V_{IC} = 5$ V	25°C	0.7			
		125°C	10	35		
V_{ICR} Common-mode input voltage range (see Note 5)		25°C	0	-0.3		V
		to 9	to 9.2			
		Full range	0			
			to 8.5			
V_{OH} High-level output voltage	$V_{ID} = 100$ mV, $R_L = 100$ k Ω	25°C	8	8.7		V
		-55°C	7.8	8.6		
		125°C	7.8	8.8		
		25°C	0	50		
V_{OL} Low-level output voltage	$V_{ID} = -100$ mV, $I_{OL} = 0$	-55°C	0	50		mV
		125°C	0	50		
		25°C	25	275		
AVD Large-signal differential voltage amplification	$V_O = 1$ V to 6 V, $R_L = 100$ k Ω	-55°C	15	420		V/mV
		125°C	15	190		
		25°C	65	94		
$CMRR$ Common-mode rejection ratio	$V_{IC} = V_{ICR\min}$	-55°C	60	93		dB
		125°C	60	93		
		25°C	70	93		
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5$ V to 10 V, $V_O = 1.4$ V	-55°C	60	91		dB
		125°C	60	94		
		25°C	285	600		
I_{DD} Supply current (two amplifiers)	$V_O = 5$ V, $V_{IC} = 5$ V, No load	-55°C	490	1000		μA
		125°C	180	480		

† Full range is -55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.
5. This range also applies to each input individually.

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operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ μs
			0°C	0.46		
		$V_{I(PP)} = 2.5\text{ V}$	70°C	0.36		
			25°C	0.40		
		$R_S = 20\text{ }\Omega$, See Figure 2	0°C	0.43		
			70°C	0.34		
V_n Equivalent input noise voltage		25°C	32			nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth		25°C	55			kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	0°C	60		
			70°C	50		
			25°C	525		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$	0°C	600		kHz
			70°C	400		
			25°C	40°		
			0°C	41°		
			70°C	39°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2C TLC27M2AC TLC27M2BC TLC27M7C			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ μs
			0°C	0.67		
		$V_{I(PP)} = 5.5\text{ V}$	70°C	0.51		
			25°C	0.56		
		$R_S = 20\text{ }\Omega$, See Figure 2	0°C	0.61		
			70°C	0.46		
V_n Equivalent input noise voltage		25°C	32			nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth		25°C	35			kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	0°C	40		
			70°C	30		
			25°C	635		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$	0°C	710		kHz
			70°C	510		
			25°C	43°		
			0°C	44°		
			70°C	42°		

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operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.43		V/ μs
			-40°C	0.51		
		$V_I(\text{PP}) = 2.5\text{ V}$	85°C	0.35		
			25°C	0.40		
			-40°C	0.48		
			85°C	0.32		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\text{ }\Omega$,	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	55		kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	-40°C	75		
			85°C	45		
			25°C	525		MHz
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$, See Figure 3	-40°C	770		
			85°C	370		
			25°C	40°		
			-40°C	43°		
			85°C	38°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2I TLC27M2AI TLC27M2BI TLC27M7I			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.62		V/ μs
			-40°C	0.77		
		$V_I(\text{PP}) = 5.5\text{ V}$	85°C	0.47		
			25°C	0.56		
			-40°C	0.70		
			85°C	0.44		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\text{ }\Omega$,	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	35		kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$,	-40°C	45		
			85°C	25		
			25°C	635		MHz
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$, See Figure 3	-40°C	880		
			85°C	480		
			25°C	43°		
			-40°C	46°		
			85°C	41°		

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operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2M TLC27M7M			UNIT
			MIN	TYP	MAX	
			25°C	0.43	0.54	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.43	0.54	V/ μs
			-55°C	0.54	0.65	
			125°C	0.29	0.35	
		$V_I(\text{PP}) = 2.5\text{ V}$	25°C	0.40	0.45	
			-55°C	0.49	0.55	
			125°C	0.28	0.32	
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20\text{ }\Omega$	25°C	32	35	nV/ $\sqrt{\text{Hz}}$
			-55°C	30	33	
			125°C	28	31	
B _{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$	25°C	55	60	kHz
			-55°C	80	85	
			125°C	40	45	
		$C_L = 20\text{ pF}$	25°C	525	550	
			-55°C	850	880	
			125°C	330	350	
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$	25°C	40°	42°	°
			-55°C	44°	46°	
			125°C	36°	38°	

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	TA	TLC27M2M TLC27M7M			UNIT
			MIN	TYP	MAX	
			25°C	0.62	0.75	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_I(\text{PP}) = 1\text{ V}$	25°C	0.62	0.75	V/ μs
			-55°C	0.81	0.95	
			125°C	0.38	0.45	
		$V_I(\text{PP}) = 5.5\text{ V}$	25°C	0.56	0.65	
			-55°C	0.73	0.85	
			125°C	0.35	0.42	
V _n Equivalent input noise voltage	f = 1 kHz, See Figure 2	$R_S = 20\text{ }\Omega$	25°C	32	35	nV/ $\sqrt{\text{Hz}}$
			-55°C	30	33	
			125°C	28	31	
B _{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 1	$C_L = 20\text{ pF}$	25°C	35	40	kHz
			-55°C	50	55	
			125°C	20	25	
		$C_L = 20\text{ pF}$	25°C	635	680	
			-55°C	960	1000	
			125°C	440	480	
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 3	$f = B_1$	25°C	43°	45°	°
			-55°C	47°	49°	
			125°C	39°	41°	

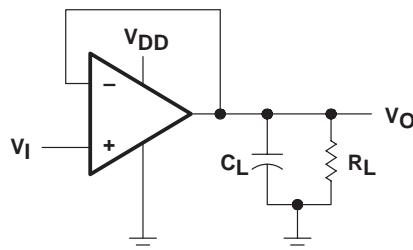
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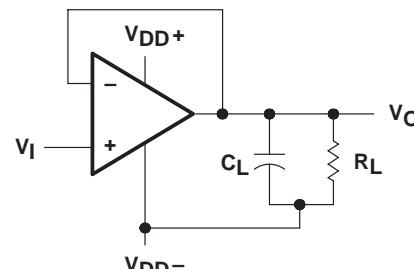
PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC27M2 and TLC27M7 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

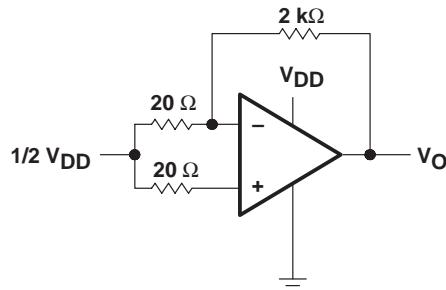


(a) SINGLE SUPPLY

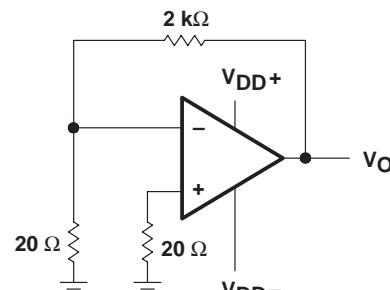


(b) SPLIT SUPPLY

Figure 1. Unity-Gain Amplifier

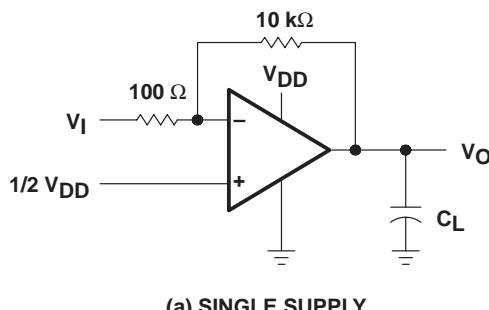


(a) SINGLE SUPPLY

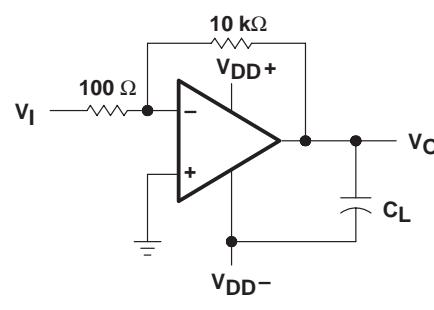


(b) SPLIT SUPPLY

Figure 2. Noise-Test Circuit



(a) SINGLE SUPPLY



(b) SPLIT SUPPLY

Figure 3. Gain-of-100 Inverting Amplifier

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TL27M2 and TL27M7 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution—many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

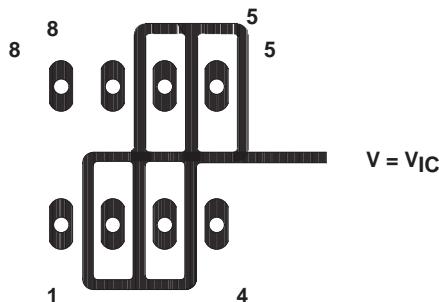


Figure 4. Isolation Metal Around Device Inputs
(JG and P packages)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

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PARAMETER MEASUREMENT INFORMATION

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage, since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

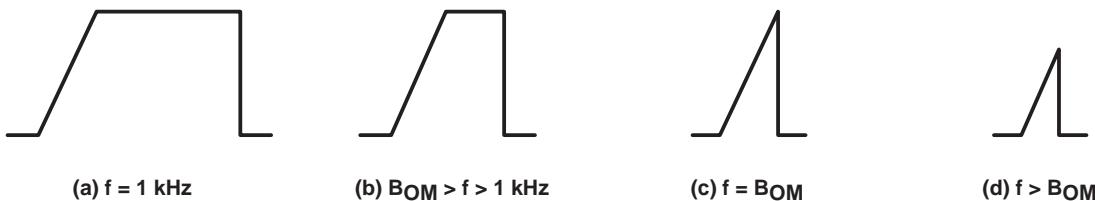


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

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TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
V_{IO}	Input offset voltage	Distribution 6, 7
α_{VIO}	Temperature coefficient	Distribution 8, 9
V_{OH}	High-level output voltage	vs High-level output current 10, 11 vs Supply voltage 12 vs Free-air temperature 13
V_{OL}	Low-level output voltage	vs Common-mode input voltage 14, 15 vs Differential input voltage 16 vs Free-air temperature 17 vs Low-level output current 18, 19
A_{VD}	Differential voltage amplification	vs Supply voltage 20 vs Free-air temperature 21 vs Frequency 32, 33
I_{IB}/I_{IO}	Input bias and input offset current	vs Free-air temperature 22
V_{IC}	Common-mode input voltage	vs Supply voltage 23
I_{DD}	Supply current	vs Supply voltage 24 vs Free-air temperature 25
SR	Slew rate	vs Supply voltage 26 vs Free-air temperature 27
	Normalized slew rate	vs Free-air temperature 28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency 29
B_1	Unity-gain bandwidth	vs Free-air temperature 30 vs Supply voltage 31
ϕ_m	Phase margin	vs Supply voltage 34 vs Free-air temperature 35 vs Capacitive loads 36
V_n	Equivalent input noise voltage	vs Frequency 37
ϕ	Phase shift	vs Frequency 32, 33

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TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC27M2 INPUT OFFSET VOLTAGE

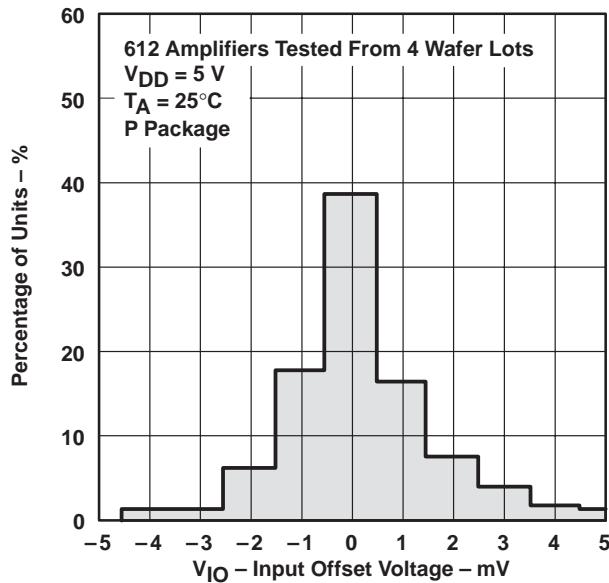


Figure 6

DISTRIBUTION OF TLC27M2 INPUT OFFSET VOLTAGE

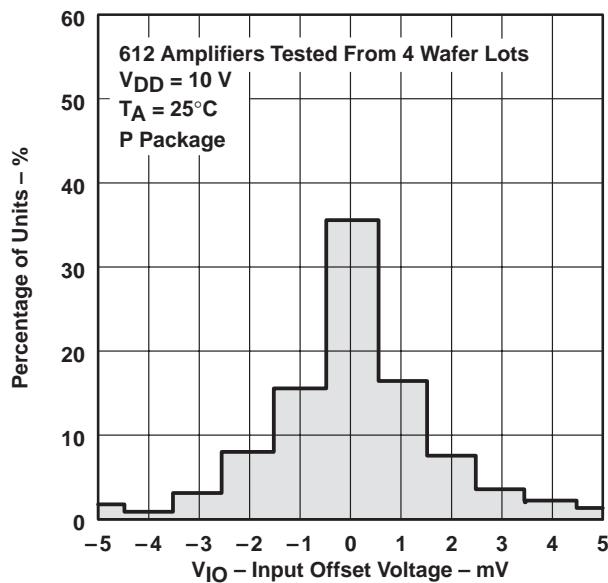


Figure 7

DISTRIBUTION OF TLC27M2 AND TLC27M7 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

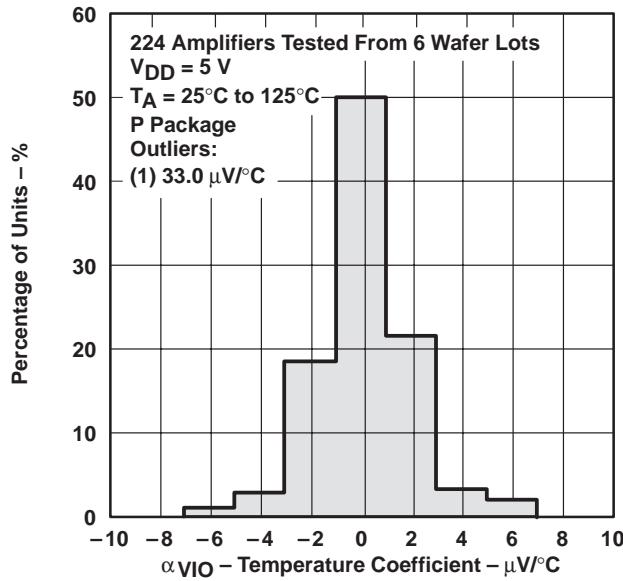


Figure 8

DISTRIBUTION OF TLC27M2 AND TLC27M7 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

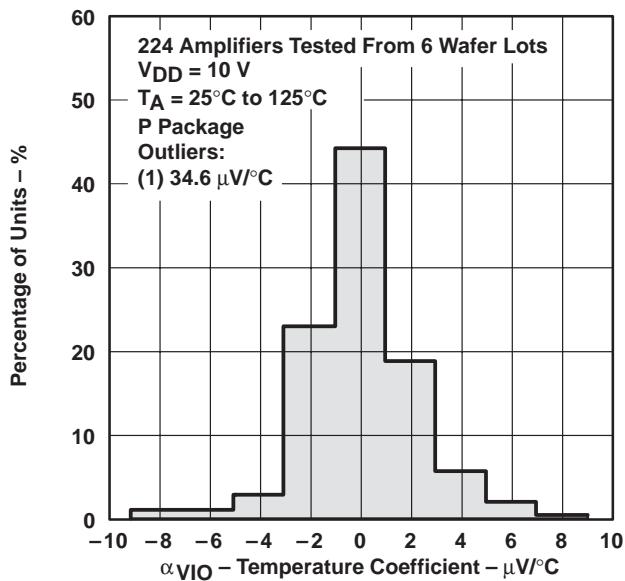


Figure 9

TYPICAL CHARACTERISTICS†

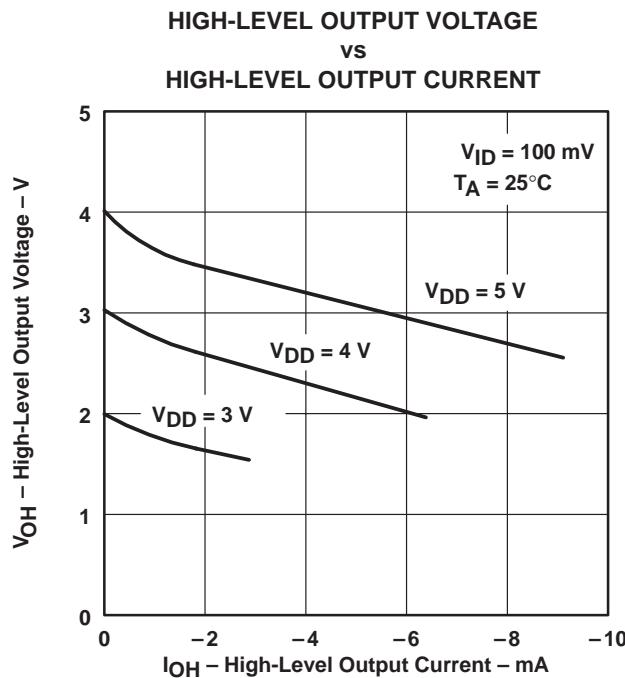


Figure 10

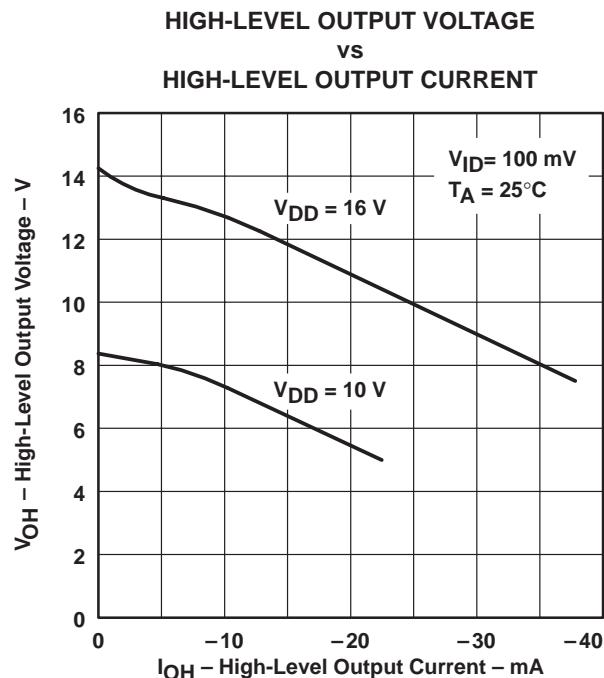


Figure 11

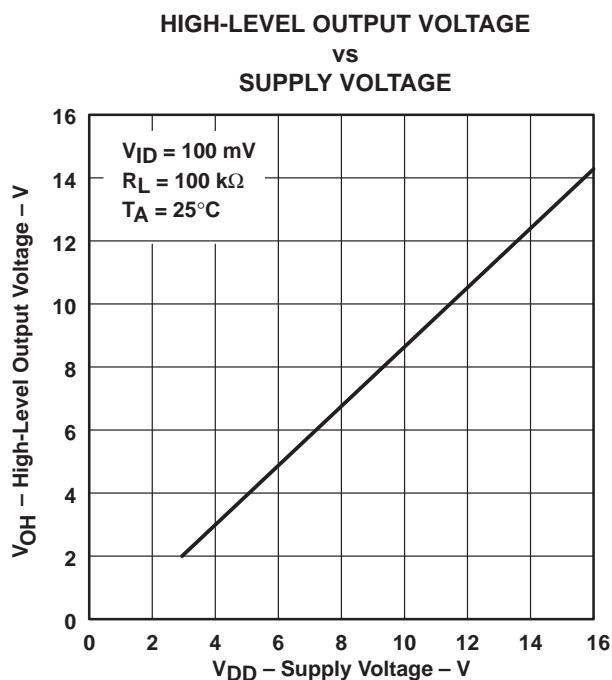


Figure 12

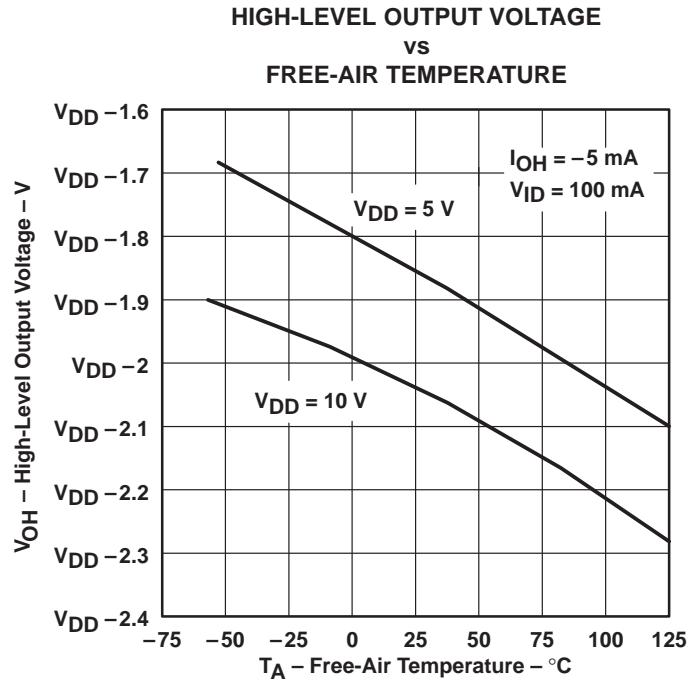


Figure 13

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS†

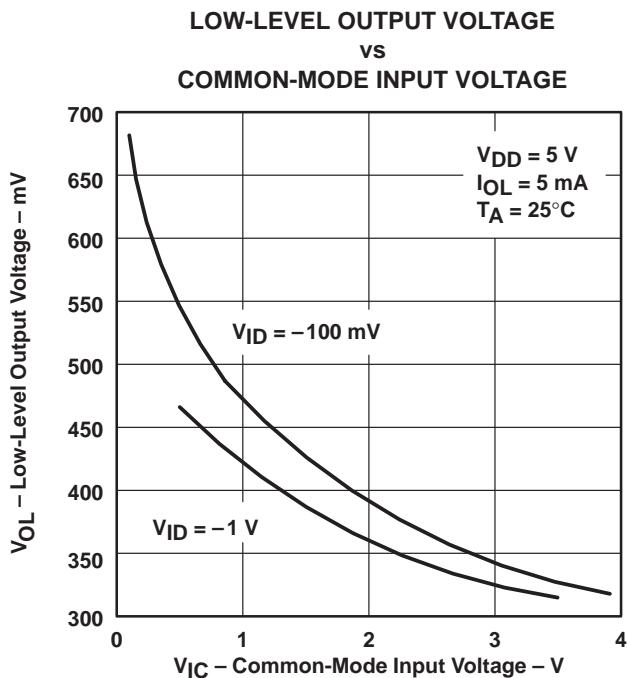


Figure 14

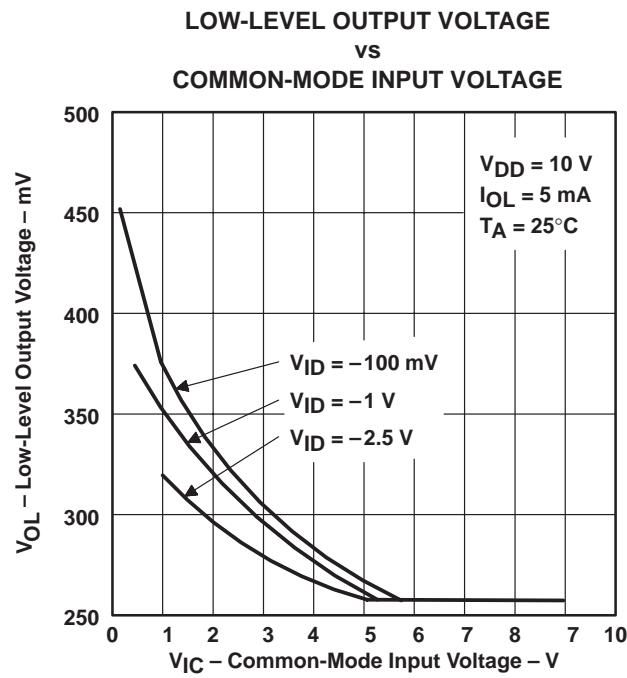


Figure 15

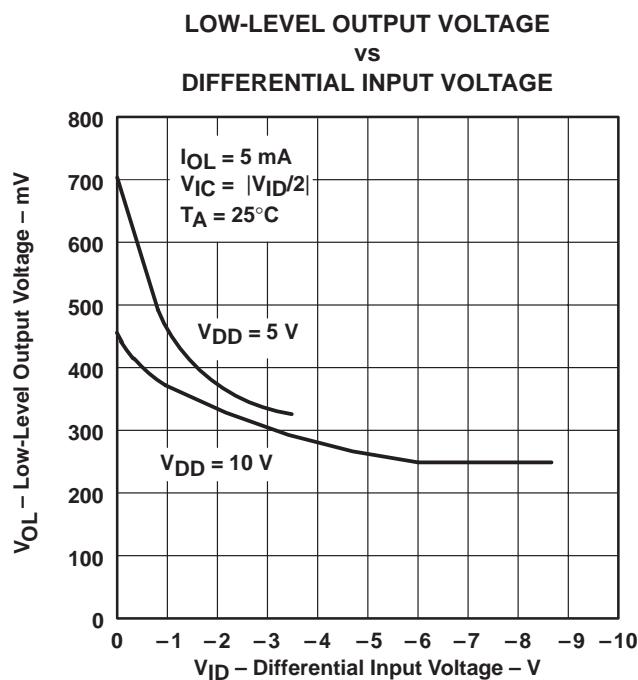


Figure 16

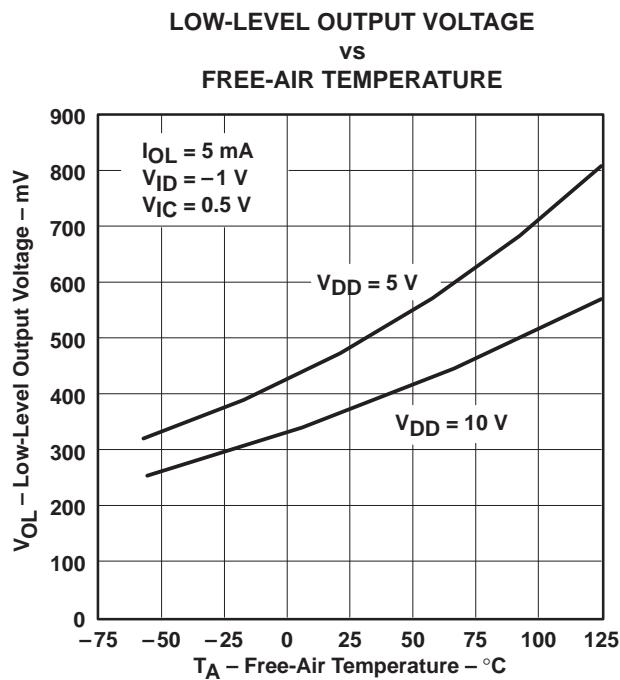


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

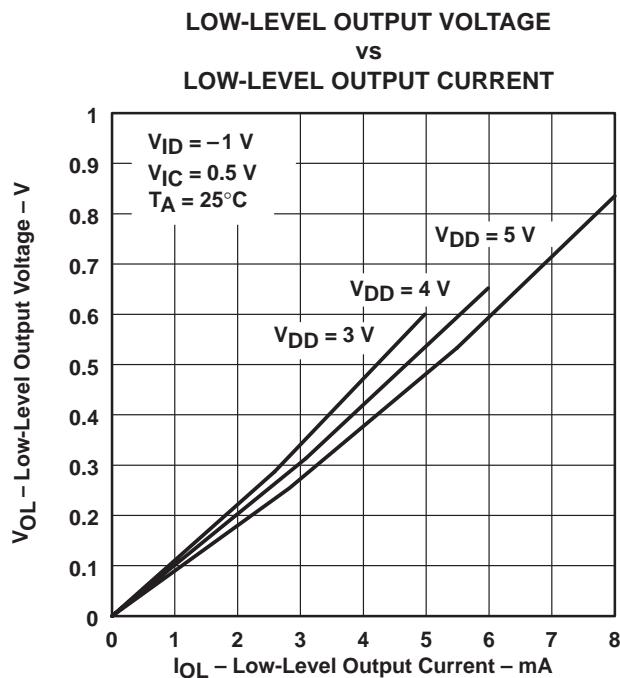


Figure 18

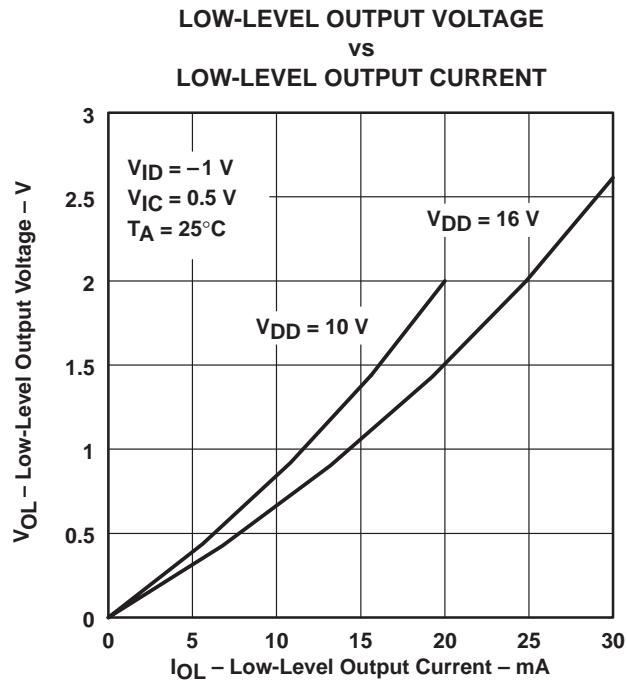


Figure 19

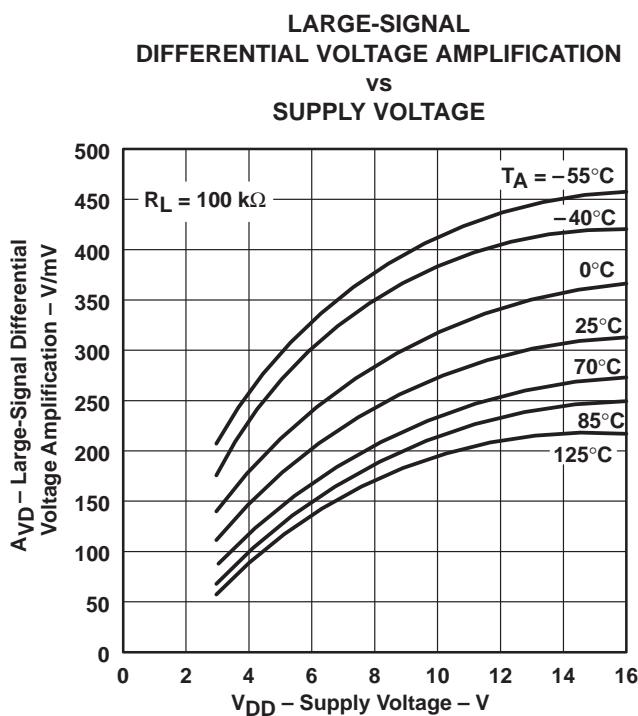


Figure 20

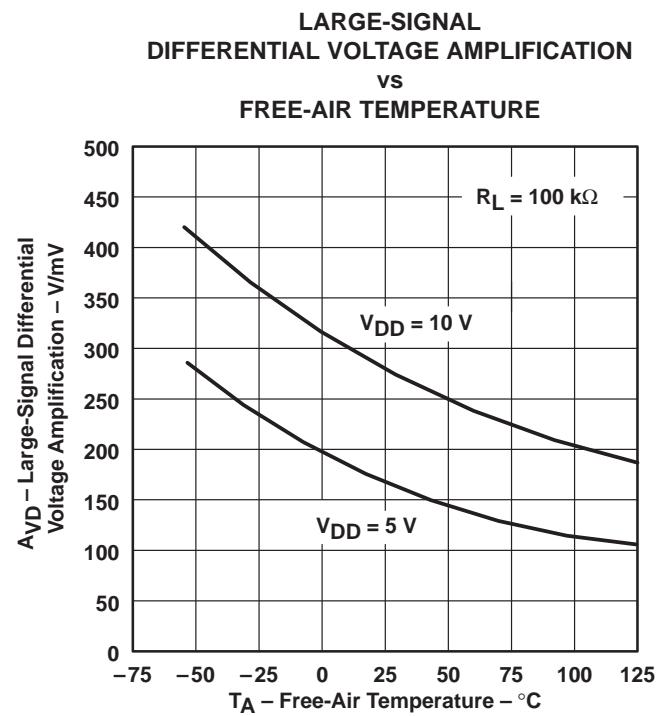


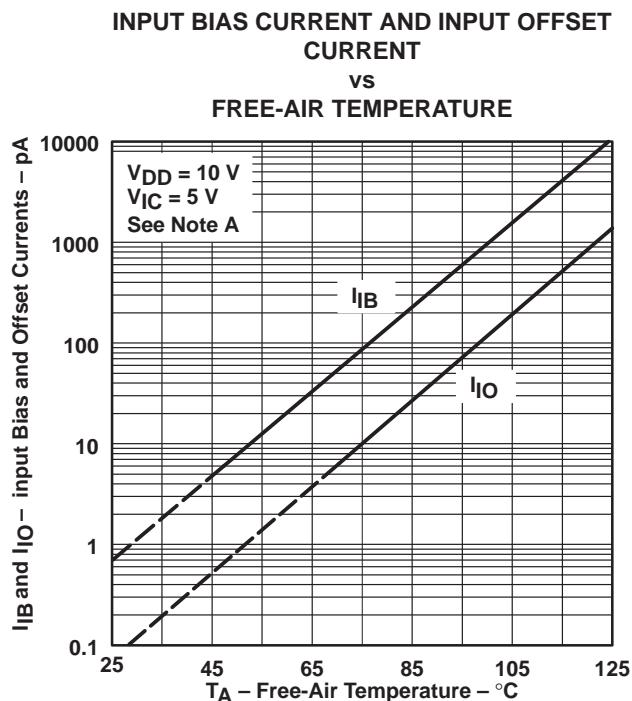
Figure 21

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS†



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22

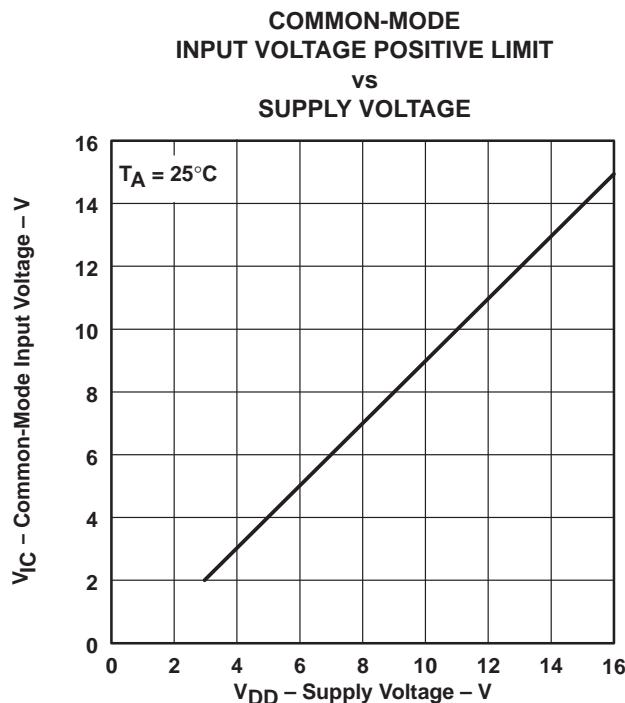


Figure 23

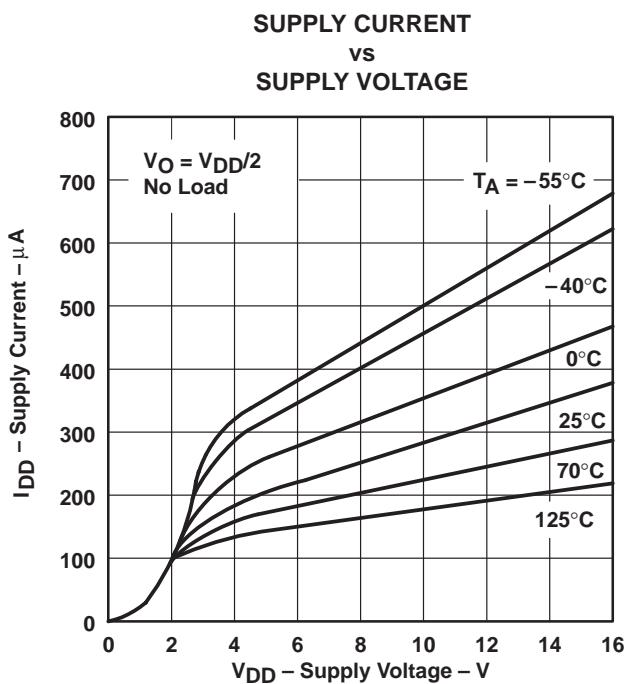


Figure 24

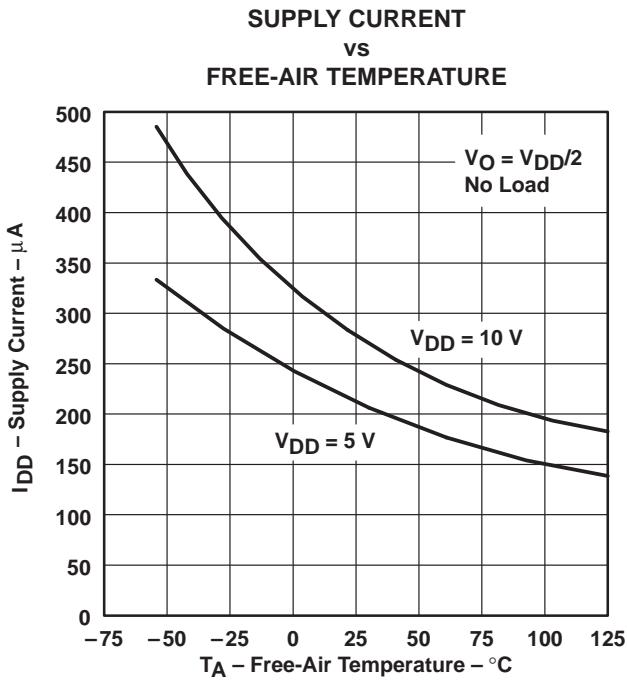


Figure 25

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

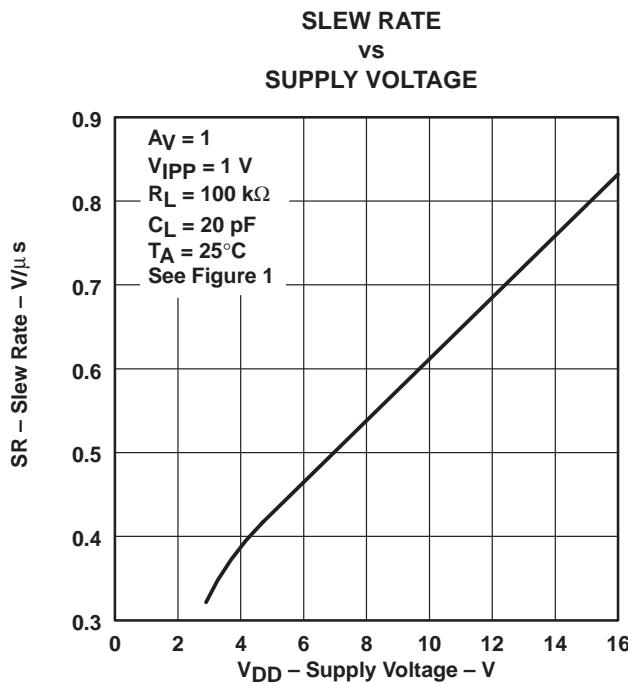


Figure 26

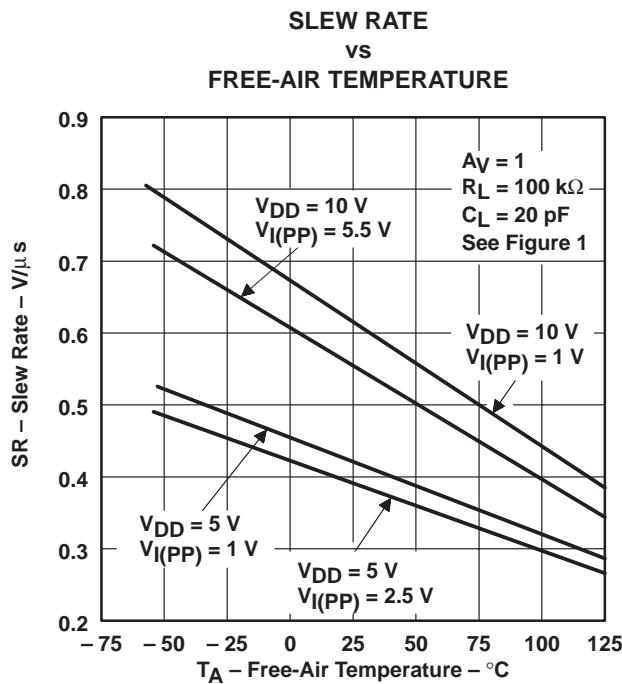


Figure 27

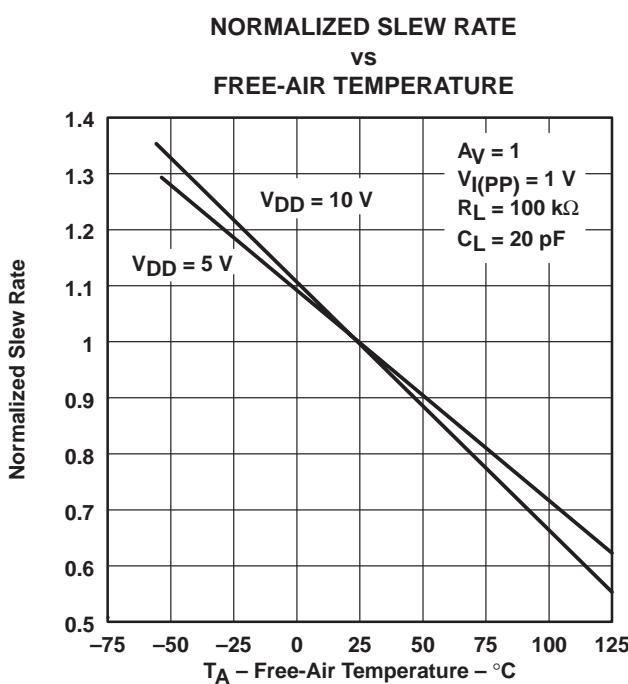


Figure 28

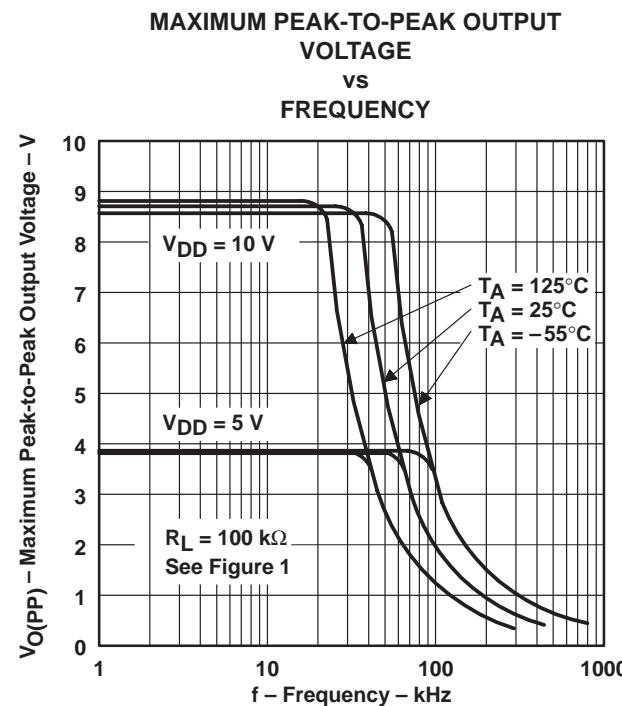


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS†

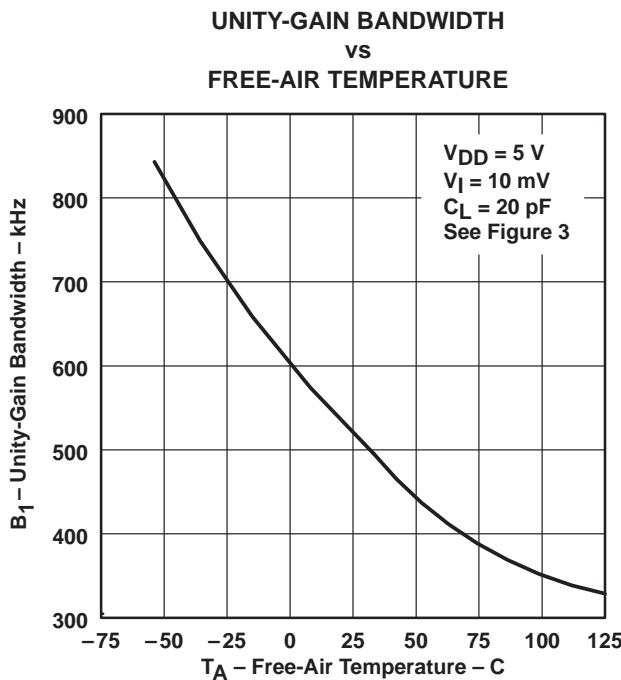


Figure 30

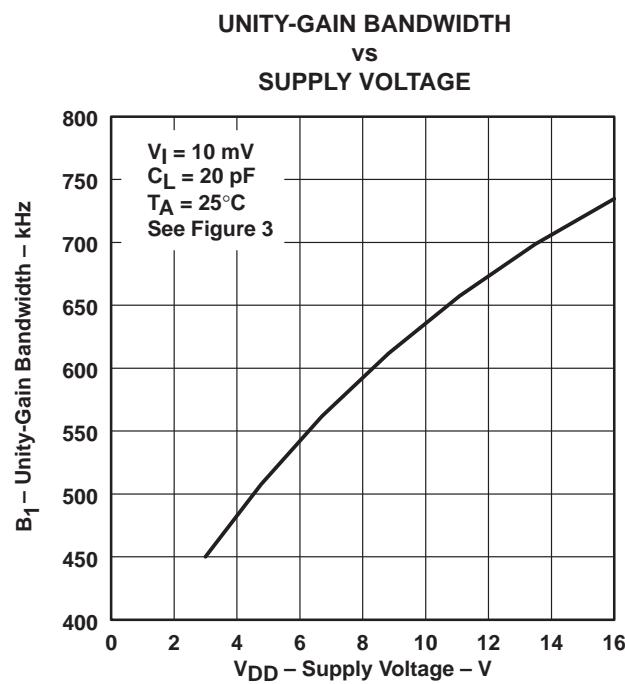


Figure 31

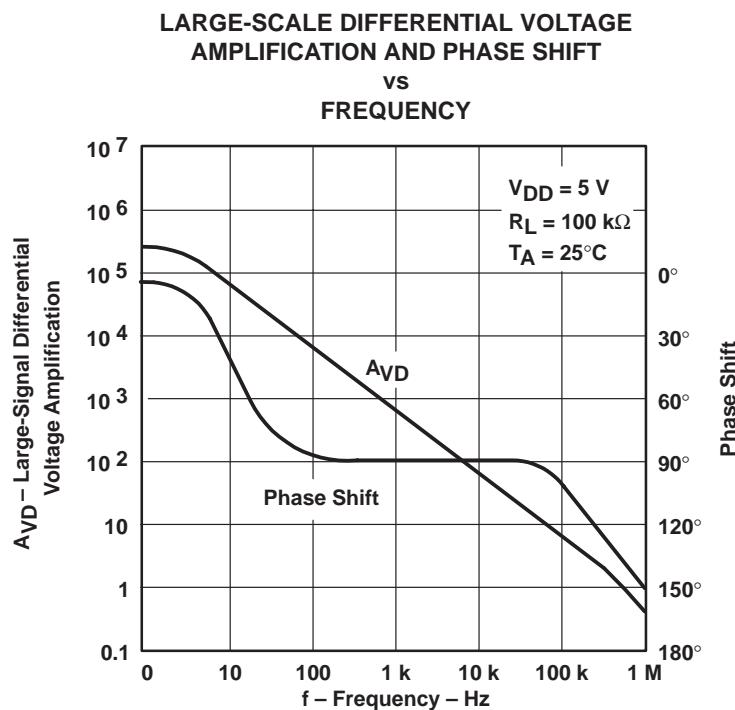


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†

**LARGE-SCALE DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY**

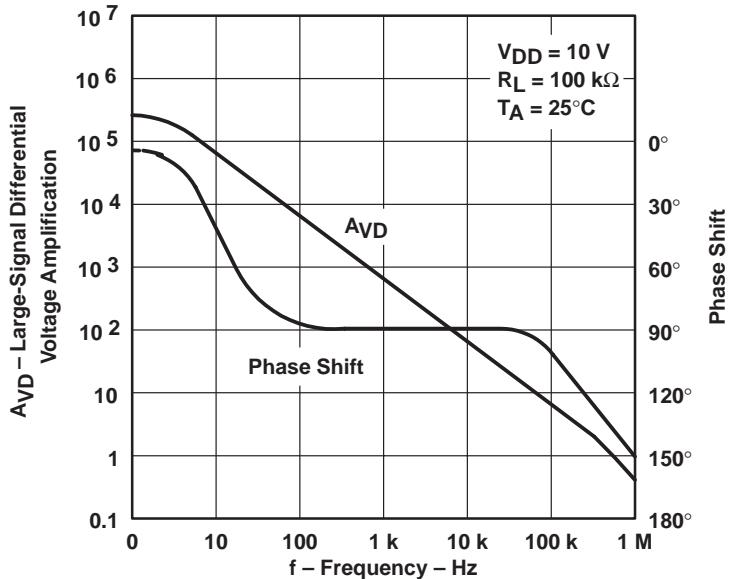


Figure 33

**PHASE MARGIN
VS
SUPPLY VOLTAGE**

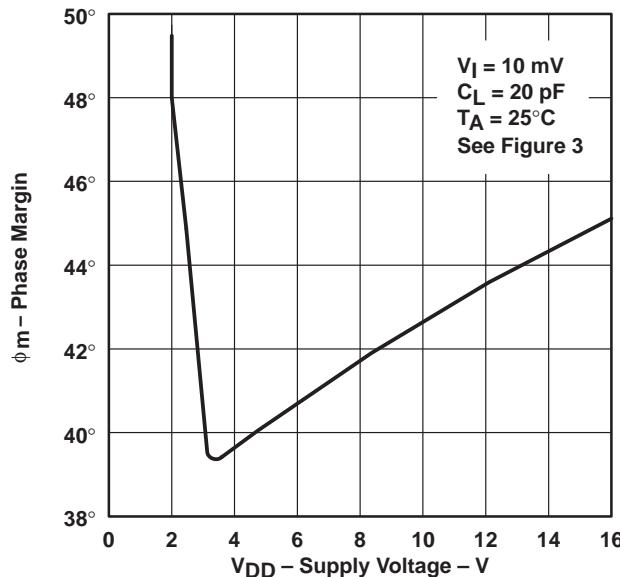


Figure 34

**PHASE MARGIN
VS
FREE-AIR TEMPERATURE**

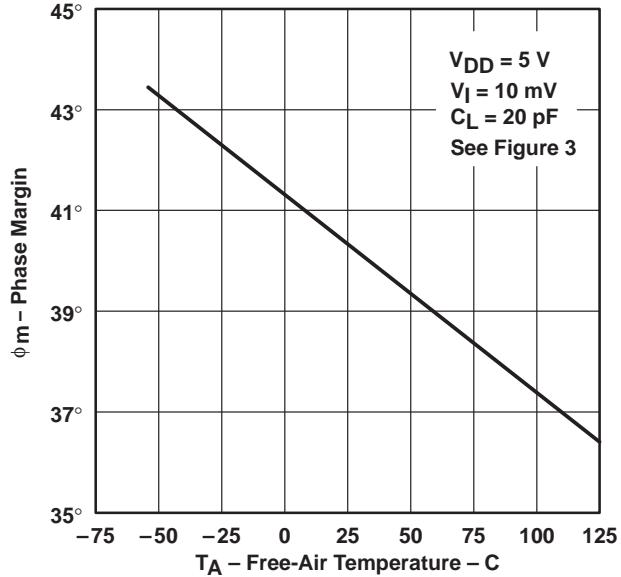


Figure 35

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

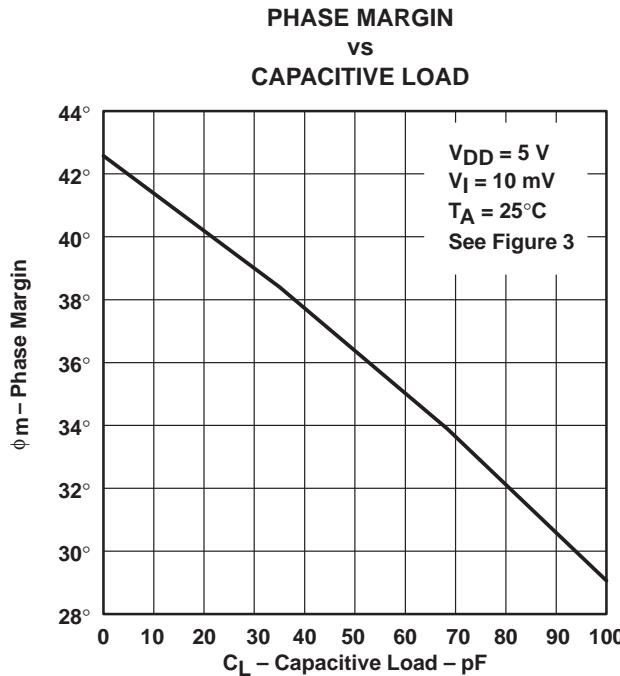


Figure 36

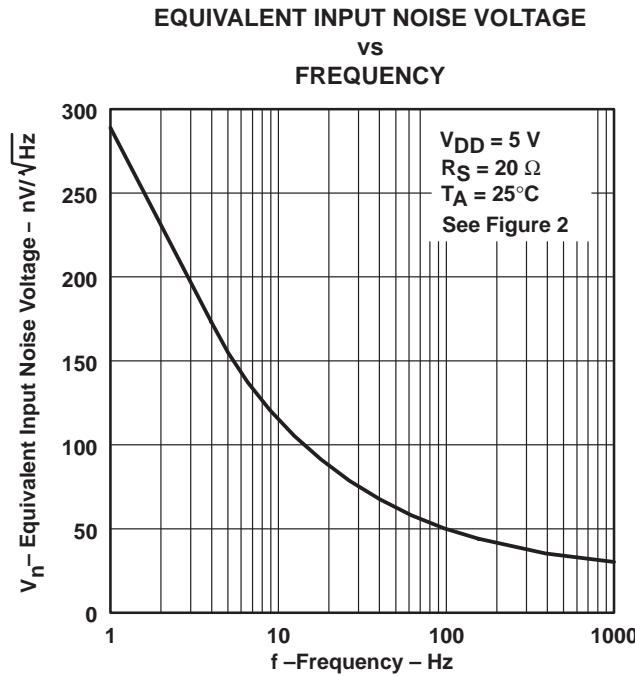


Figure 37

APPLICATION INFORMATION

single-supply operation

While the TLC27M2 and TLC27M7 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M2 and TLC27M7 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M2 and TLC27M7 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

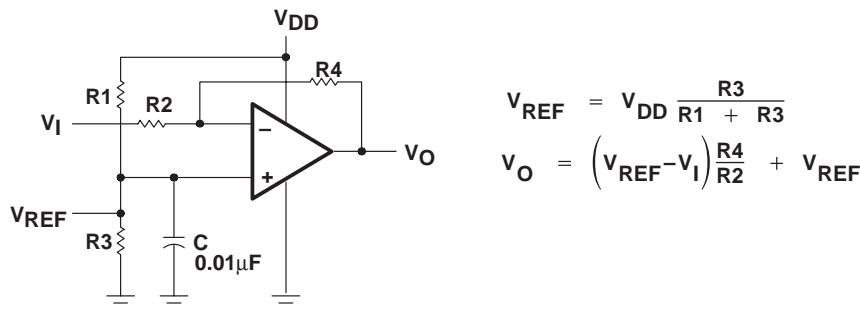


Figure 38. Inverting Amplifier With Voltage Reference

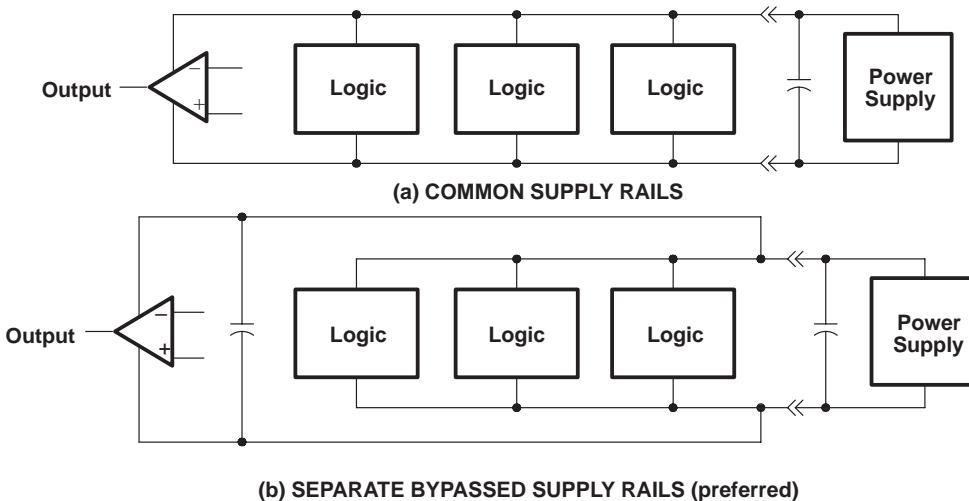


Figure 39. Common Versus Separate Supply Rails

TLC27M2, TLC27M2A, TLC27M2B, TLC27M7

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APPLICATION INFORMATION

input characteristics

The TLC27M2 and TLC27M7 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1$ V at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5$ V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M2 and TLC27M7 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1\mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M2 and TLC27M7 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M2 and TLC27M7 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50\text{ k}\Omega$, since bipolar devices exhibit greater noise currents.

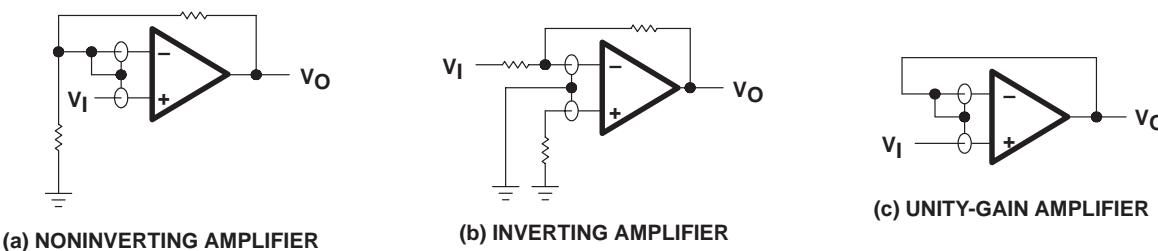


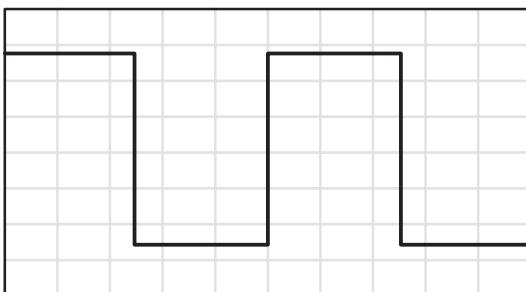
Figure 40. Guard-Ring Schemes

output characteristics

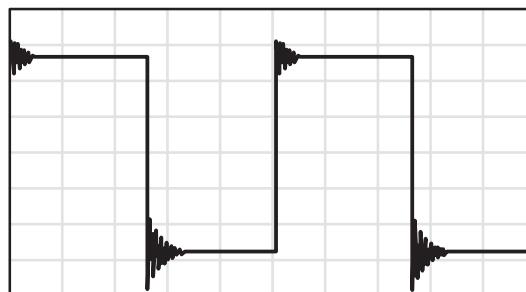
The output stage of the TLC27M2 and TLC27M7 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M2 and TLC27M7 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

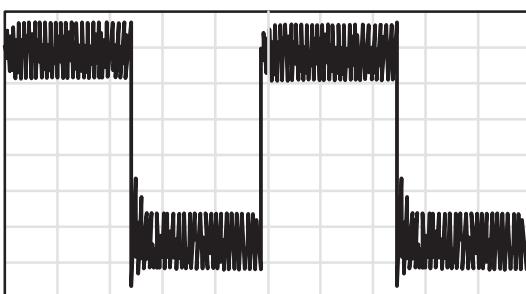
APPLICATION INFORMATION



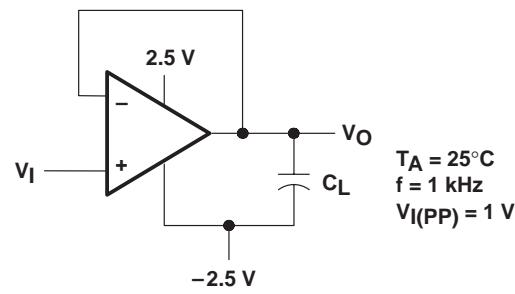
(a) $C_L = 20 \text{ pF}$, $R_L = \text{NO LOAD}$



(b) $C_L = 170 \text{ pF}$, $R_L = \text{NO LOAD}$



(c) $C_L = 190 \text{ pF}$, $R_L = \text{NO LOAD}$



(d) TEST CIRCUIT

Figure 41. Effect of Capacitive Loads and Test Circuit

output characteristics (continued)

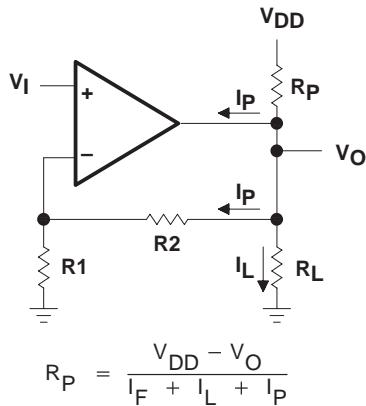
Although the TLC27M2 and TLC27M7 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_P) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60Ω and 180Ω , depending on how hard the op amp input is driven. With very low values of R_P , a voltage offset from 0 V at the output occurs. Second, pullup resistor R_P acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

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APPLICATION INFORMATION

output characteristics (continued)



I_P = Pullup current required by
the operational amplifier
(typically 500 μ A)

Figure 42. Resistive Pullup to Increase V_{OH}

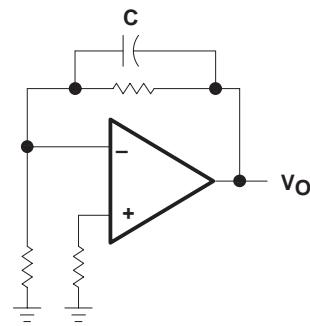


Figure 43. Compensation for Input Capacitance

feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic-discharge protection

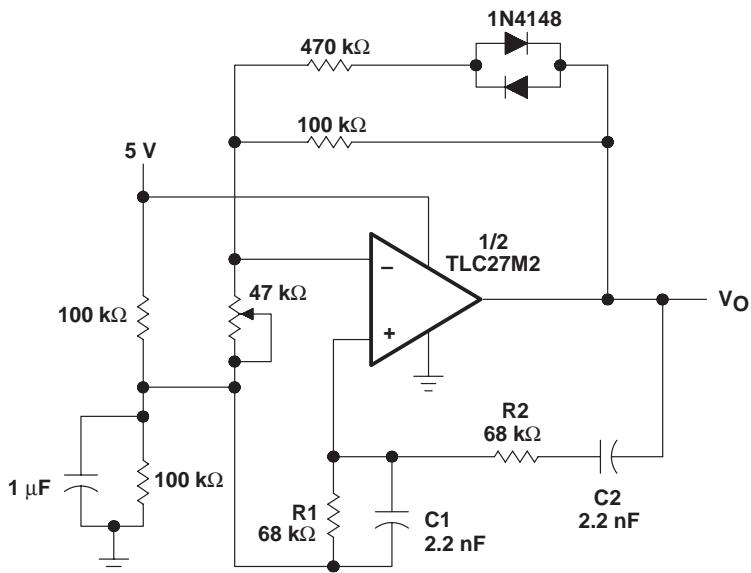
The TLC27M2 and TLC27M7 incorporate an internal electrostatic-discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27M2 and TLC27M7 inputs and outputs were designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

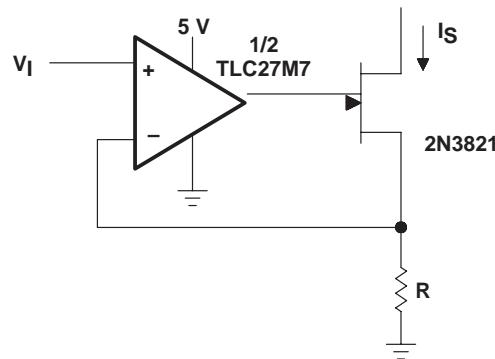
APPLICATION INFORMATION



NOTES: $V_O(PP) \approx 2$ V

$$f_O = \frac{1}{2\pi\sqrt{R1R2C1C2}}$$

Figure 44. Wien Oscillator



NOTES: $V_I = 0$ V to 3 V

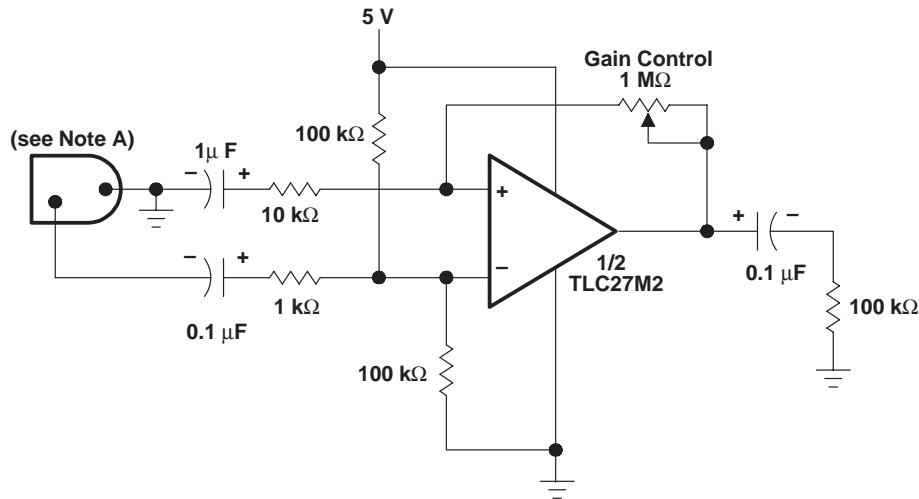
$$I_S = \frac{V_I}{R}$$

Figure 45. Precision Low-Current Sink

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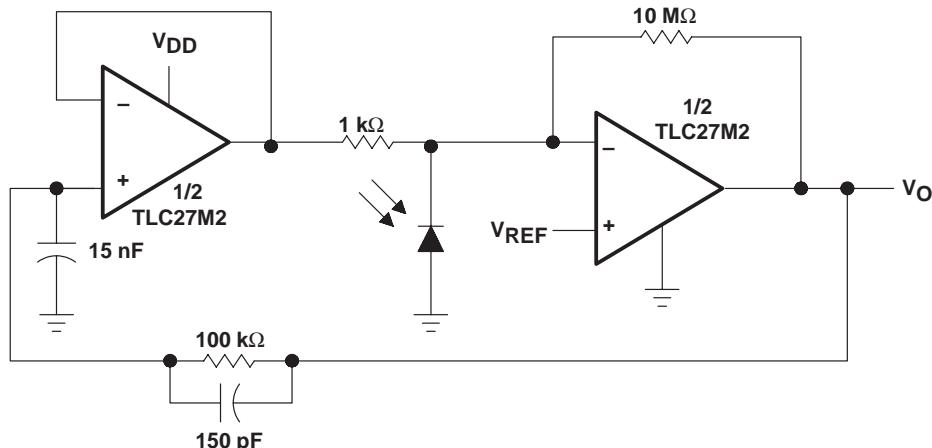
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APPLICATION INFORMATION



NOTE A: Low to medium impedance dynamic mike

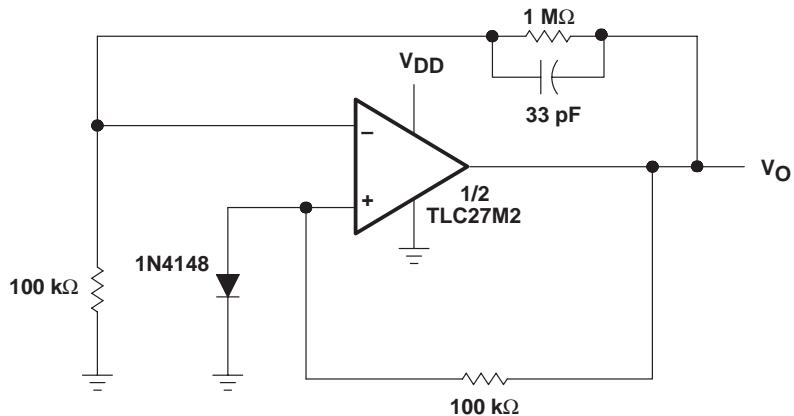
Figure 46. Microphone Preamplifier



NOTES: $V_{DD} = 4 \text{ V to } 15 \text{ V}$
 $V_{ref} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

Figure 47. Photo-Diode Amplifier With Ambient Light Rejection

APPLICATION INFORMATION



NOTES: $V_{DD} = 8 \text{ V to } 16 \text{ V}$
 $V_O = 5 \text{ V, } 10 \text{ mA}$

Figure 48. 5-V Low-Power Voltage Regulator

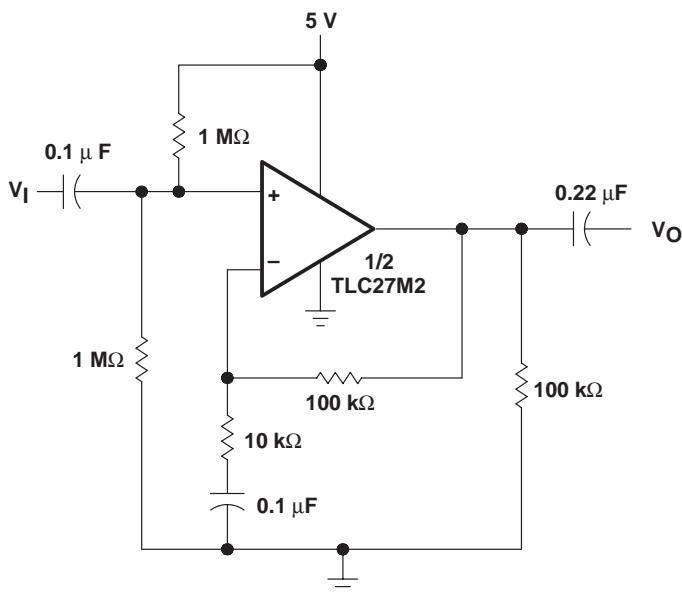


Figure 49. Single-Rail AC Amplifiers

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TLC27M2ACD	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2ACDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2ACDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2ACP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2ACPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2AID	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2AIDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2AIDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2AIDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2AIP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2AIPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2BCD	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BCDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BCDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BCDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BCP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2BCPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2BID	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BIDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BIDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BIDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2BIP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2BIP4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2CD	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TLC27M2CDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2CPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2CPSLE	OBsolete	SO	PS	8		TBD	Call TI	Call TI
TLC27M2CPSR	ACTIVE	SO	PS	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CPSRG4	ACTIVE	SO	PS	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CPW	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CPWLE	OBsolete	TSSOP	PW	8		TBD	Call TI	Call TI
TLC27M2CPWR	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2CPWRG4	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2ID	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2IPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M2IPW	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IPWR	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2IPWRG4	ACTIVE	TSSOP	PW	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M2MD	ACTIVE	SOIC	D	8		TBD	CU NIPDAU	Level-1-220C-UNLIM
TLC27M2MDR	ACTIVE	SOIC	D	8		TBD	Call TI	Call TI
TLC27M2MFKB	OBsolete	LCCC	FK	20		TBD	Call TI	Call TI
TLC27M2MJG	OBsolete	CDIP	JG	8		TBD	Call TI	Call TI
TLC27M2MJGB	OBsolete	CDIP	JG	8		TBD	Call TI	Call TI
TLC27M7CD	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7CDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7CDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7CDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
no Sb/Br)								
TLC27M7CP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M7CPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M7CPSR	ACTIVE	SO	PS	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7CPSRG4	ACTIVE	SO	PS	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7ID	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7IDG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7IDR	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7IDRG4	ACTIVE	SOIC	D	8		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLC27M7IP	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M7IPE4	ACTIVE	PDIP	P	8		Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TLC27M7MFKB	OBsolete	LCCC	FK	20		TBD	Call TI	Call TI
TLC27M7MJG	OBsolete	CDIP	JG	8		TBD	Call TI	Call TI
TLC27M7MJGB	OBsolete	CDIP	JG	8		TBD	Call TI	Call TI
TLC27M7MUB	OBsolete	CFP	U	10		TBD	Call TI	Call TI

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBsolete: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

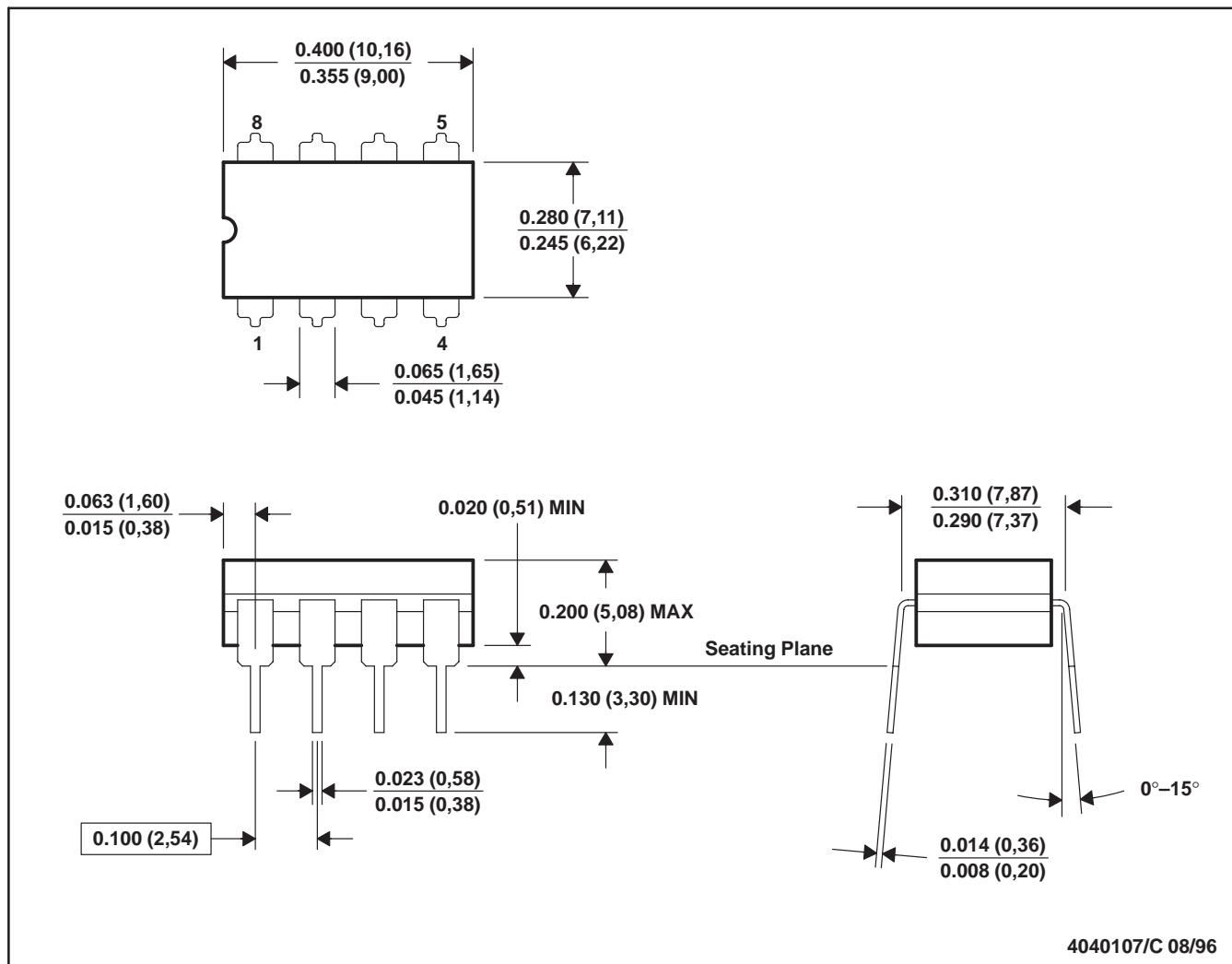
⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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JG (R-GDIP-T8)

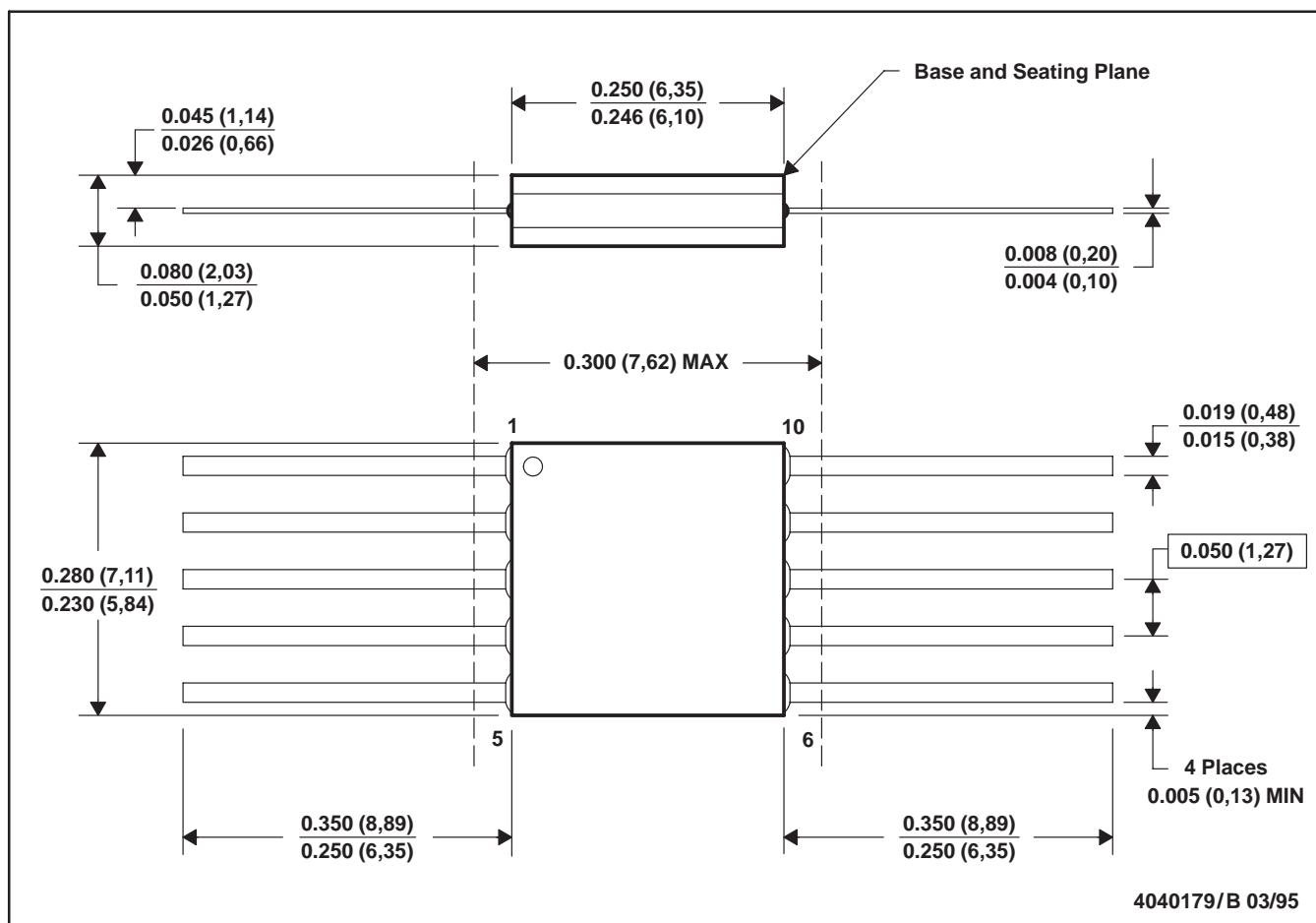
CERAMIC DUAL-IN-LINE



NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a ceramic lid using glass frit.
 D. Index point is provided on cap for terminal identification.
 E. Falls within MIL STD 1835 GDIP1-T8

U (S-GDFP-F10)

CERAMIC DUAL FLATPACK

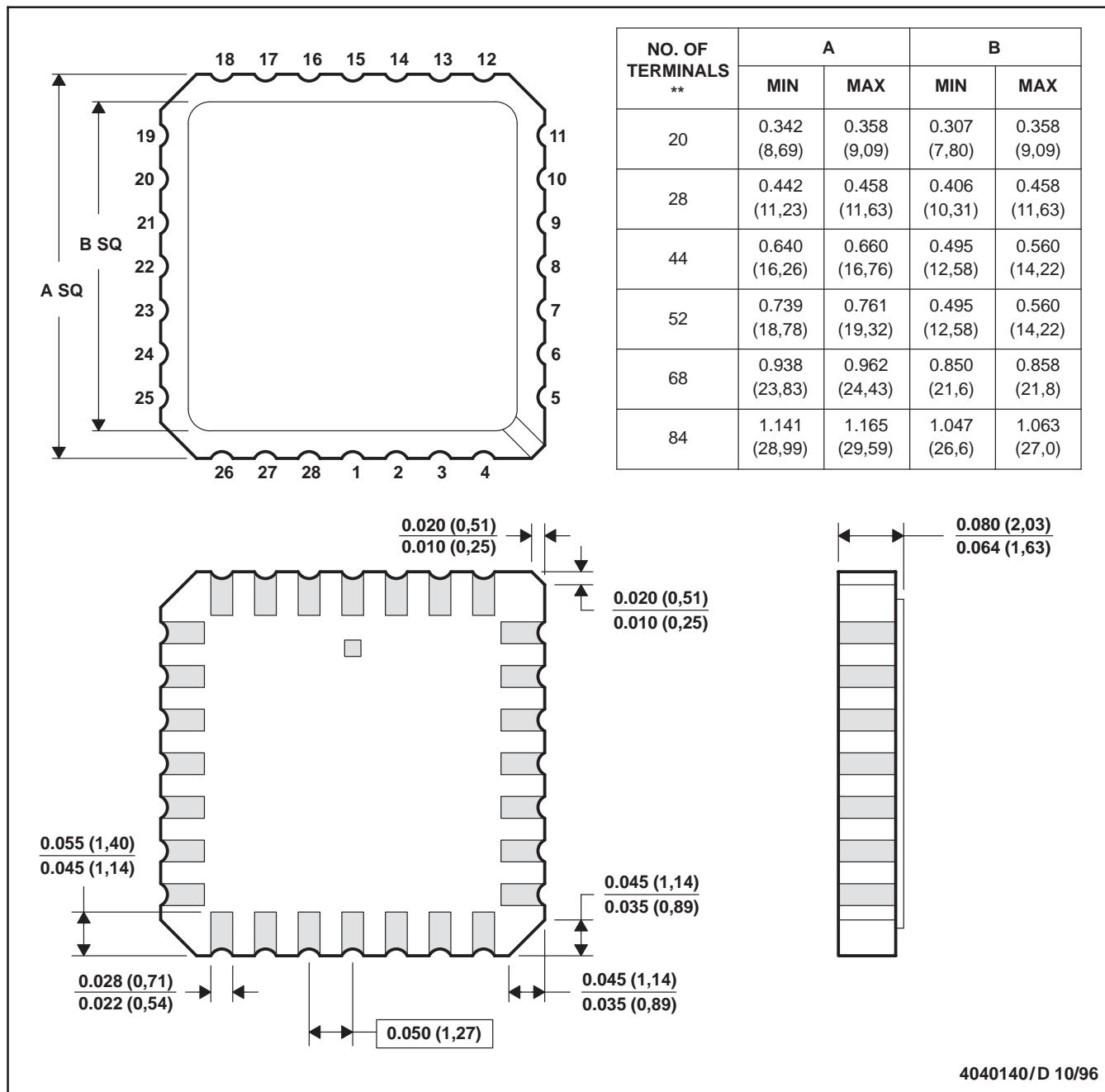


NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a ceramic lid using glass frit.
 D. Index point is provided on cap for terminal identification only.
 E. Falls within MIL STD 1835 GDFP1-F10 and JEDEC MO-092AA

FK (S-CQCC-N**)

LEADLESS CERAMIC CHIP CARRIER

28 TERMINAL SHOWN

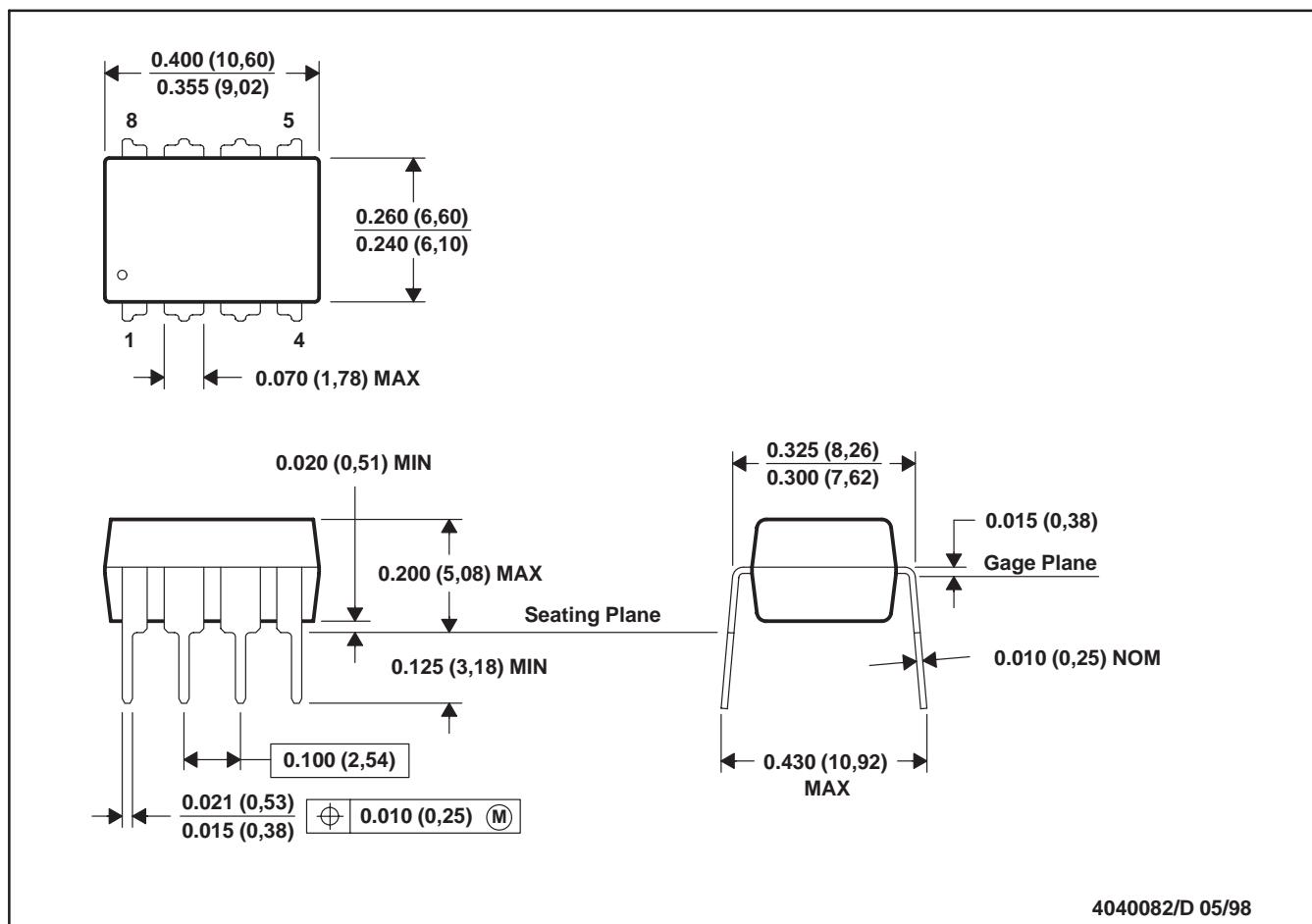


NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a metal lid.
 D. The terminals are gold plated.
 E. Falls within JEDEC MS-004

4040140/D 10/96

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE

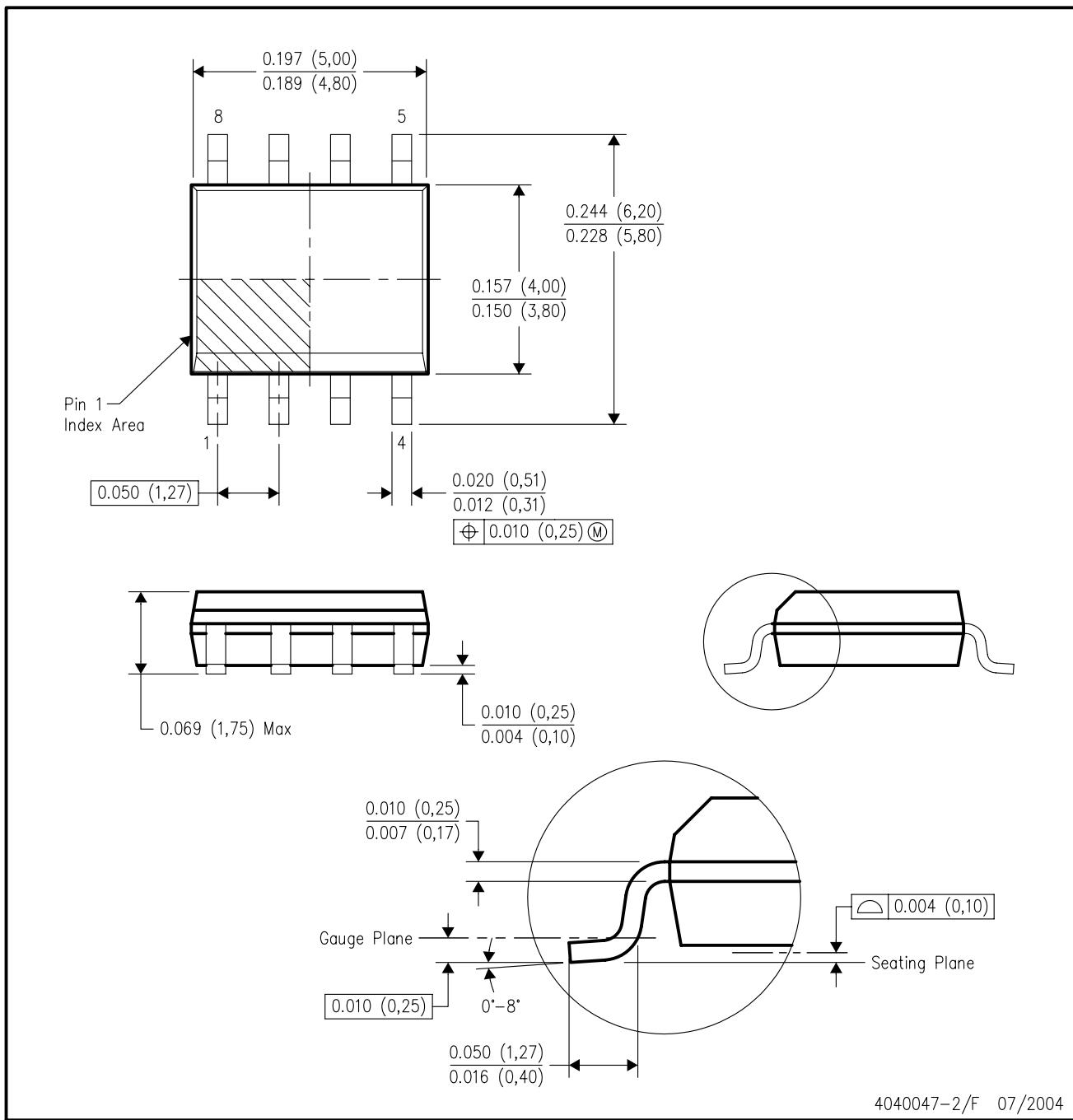


NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. Falls within JEDEC MS-001

For the latest package information, go to http://www.ti.com/sc/docs/package/pkg_info.htm

D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



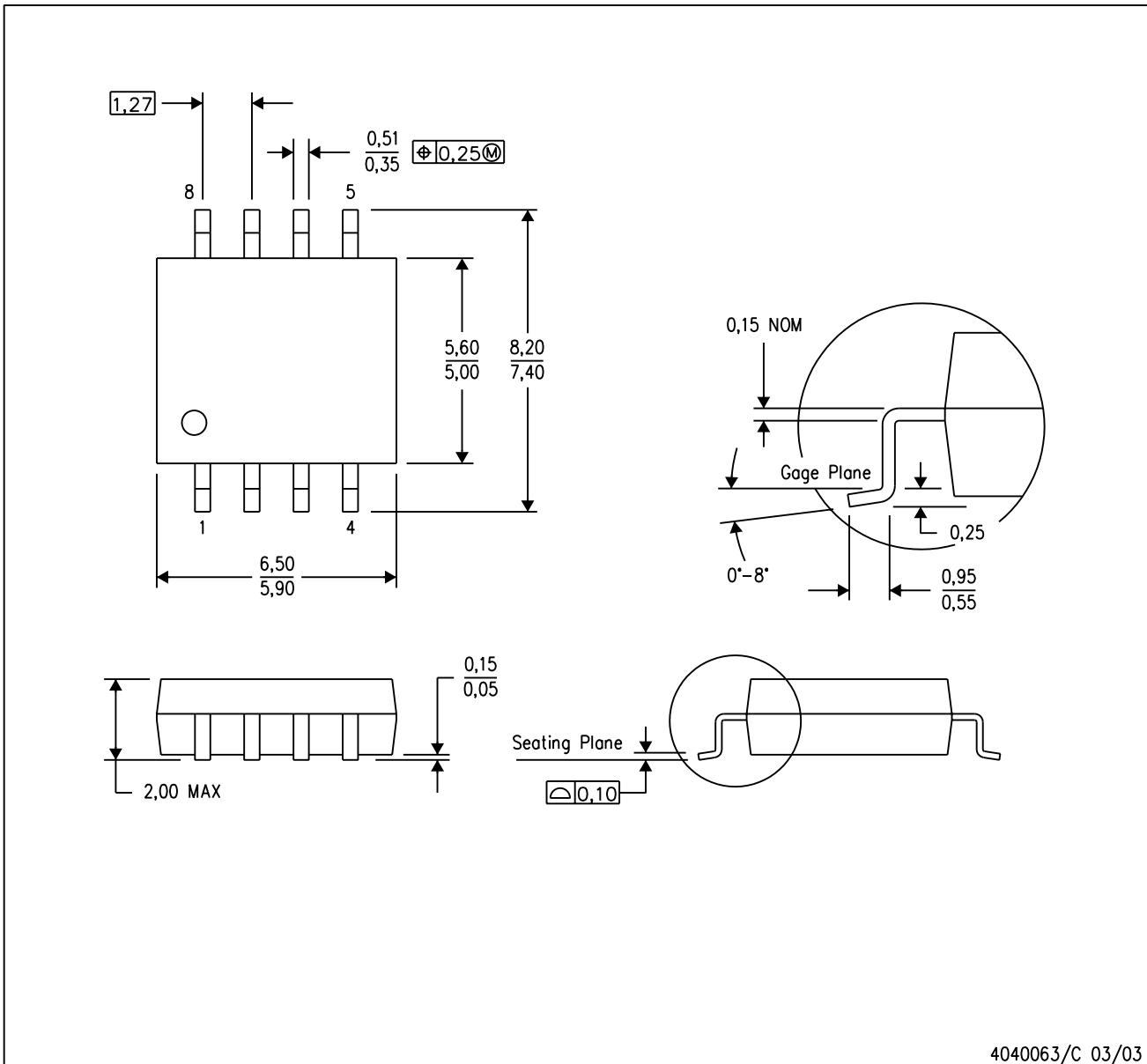
NOTES:

- All linear dimensions are in inches (millimeters).
- This drawing is subject to change without notice.
- Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0.15).
- Falls within JEDEC MS-012 variation AA.

MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



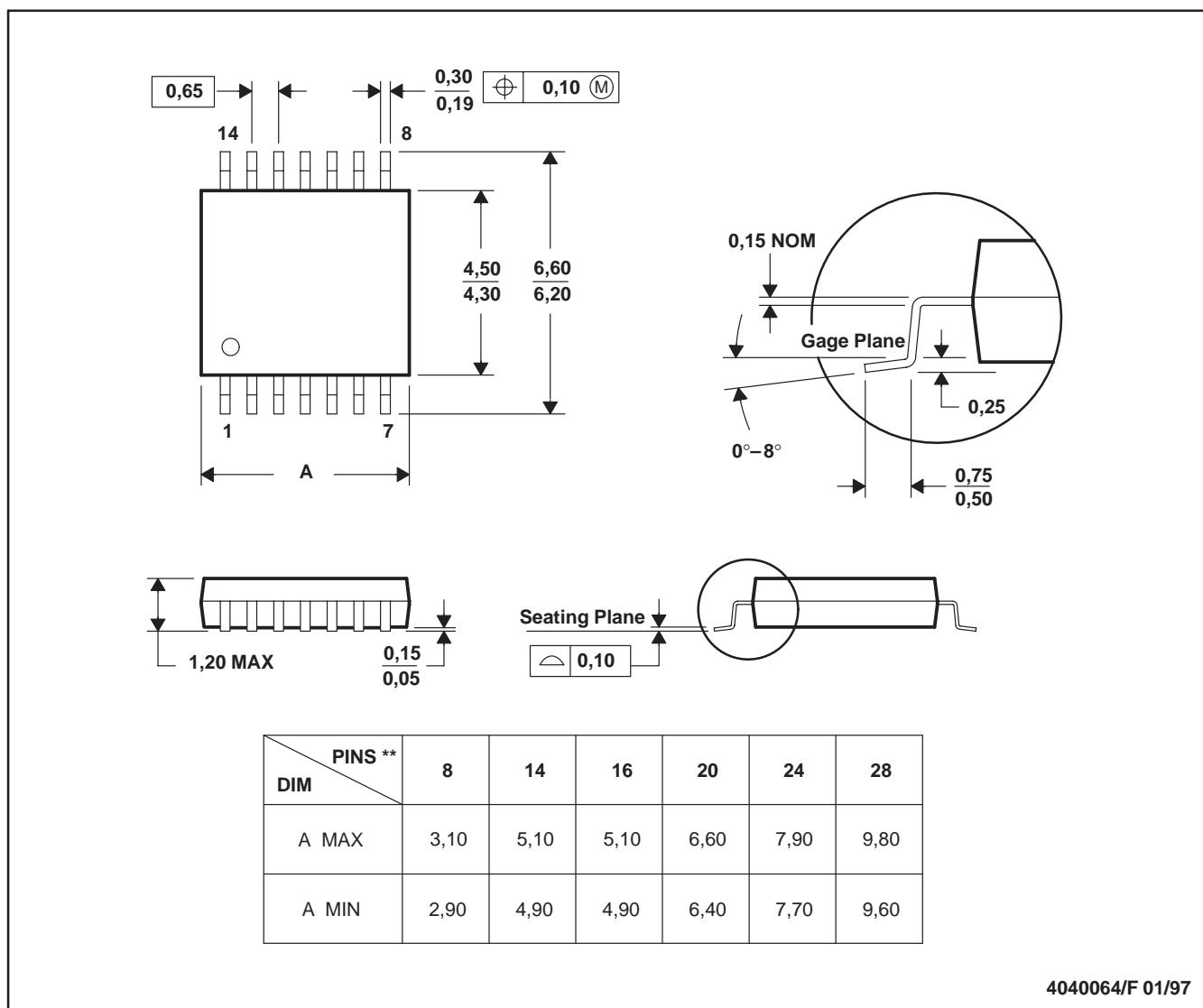
NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

PW (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14 PINS SHOWN



NOTES:

- All linear dimensions are in millimeters.
- This drawing is subject to change without notice.
- Body dimensions do not include mold flash or protrusion not to exceed 0,15.
- Falls within JEDEC MO-153

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Mailing Address: Texas Instruments
Post Office Box 655303 Dallas, Texas 75265

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