



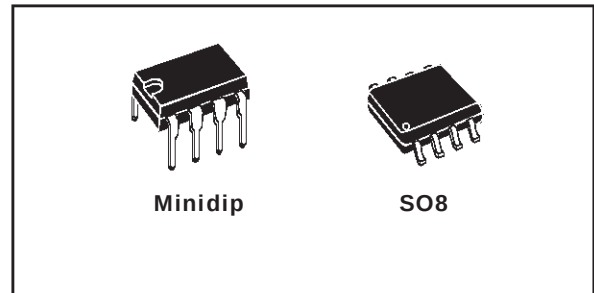
UC3842T UC3843T UC3844T UC3845T

HIGH PERFORMANCE CURRENT MODE PWM CONTROLLER

- TRIMMED OSCILLATOR FOR PRECISE FREQUENCY CONTROL
- OSCILLATOR FREQUENCY GUARANTEED AT 250kHz
- CURRENT MODE OPERATION TO 500kHz
- AUTOMATIC FEED FORWARD COMPENSATION
- LATCHING PWM FOR CYCLE-BY-CYCLE CURRENT LIMITING
- INTERNALLY TRIMMED REFERENCE WITH UNDERVOLTAGE LOCKOUT
- HIGH CURRENT TOTEM POLE OUTPUT
- UNDERVOLTAGE LOCKOUT WITH HYSTERESIS
- LOW START-UP AND OPERATING CURRENT

DESCRIPTION

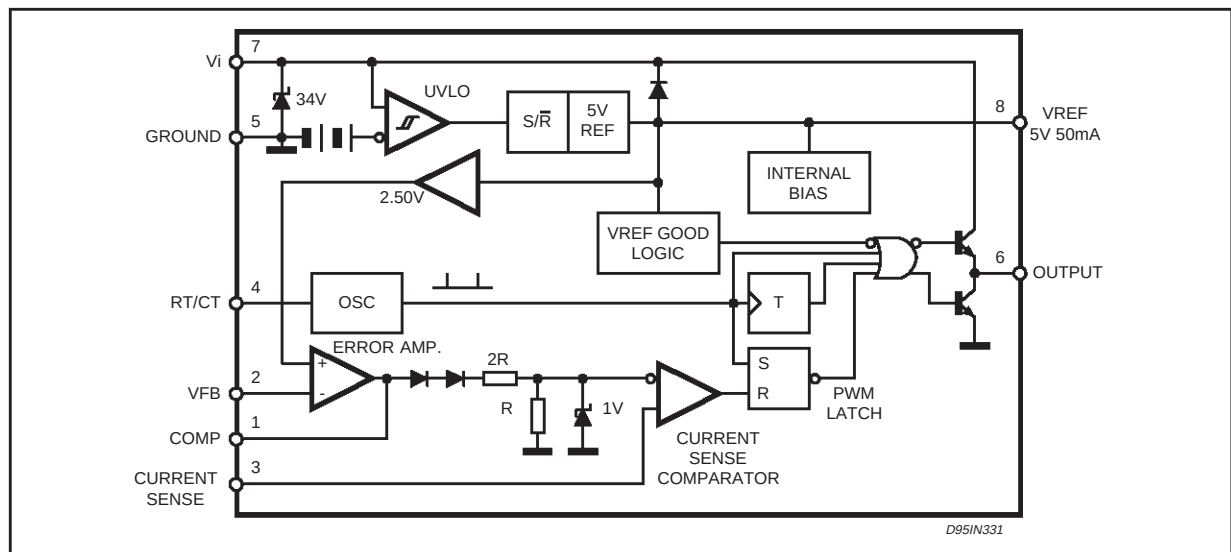
The UC384XT family of control ICs provides the necessary features to implement off-line or DC to DC fixed frequency current mode control schemes with a minimal external parts count. Internally implemented circuits include a trimmed oscillator for precise DUTY CYCLE CONTROL under voltage lockout featuring start-up current less than 0.5mA, a precision reference trimmed for accuracy at the error amp input, logic to insure latched operation, a PWM



comparator which also provides current limit control, and a totem pole output stage designed to source or sink high peak current. The output stage, suitable for driving N-Channel MOSFETs, is low in the off-state.

Differences between members of this family are the under-voltage lockout thresholds and maximum duty cycle ranges. The UC3842T and UC3844T have UVLO thresholds of 16V (on) and 10V (off), ideally suited to off-line applications. The corresponding thresholds for the UC3843T and UC3845T are 8.5 V and 7.9 V. The UC3842T and UC3843T can operate to duty cycles approaching 100%. A range of zero to < 50 % is obtained by the UC3844T and UC3845T by the addition of an internal toggle flip flop which blanks the output off every other clock cycle.

BLOCK DIAGRAM (toggle flip flop used only in UC3844T and UC3845T)



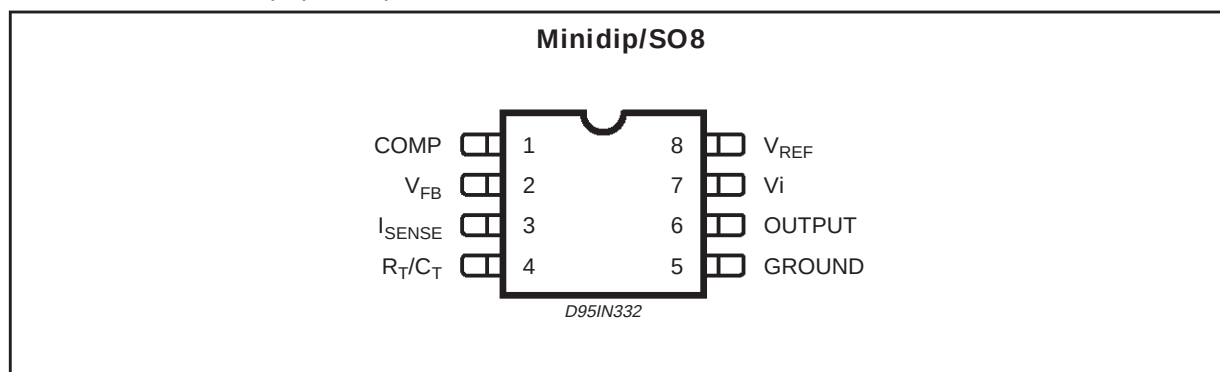
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ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_i	Supply Voltage (low impedance source)	30	V
V_i	Supply Voltage ($I_i < 30\text{mA}$)	Self Limiting	
I_o	Output Current	± 1	A
E_o	Output Energy (capacitive load)	5	μJ
	Analog Inputs (pins 2, 3)	-0.3 to 5.5	V
	Error Amplifier Output Sink Current	10	mA
P_{tot}	Power Dissipation at $T_{\text{amb}} \leq 25^\circ\text{C}$ (Minidip)	1.25	W
P_{tot}	Power Dissipation at $T_{\text{amb}} \leq 25^\circ\text{C}$ (SO8)	800	mW
T_{stg}	Storage Temperature Range	-65 to 150	$^\circ\text{C}$
T_L	Lead Temperature (soldering 10s)	300	$^\circ\text{C}$

* All voltages are with respect to pin 5, all currents are positive into the specified terminal.

PIN CONNECTION (top view)



PIN FUNCTIONS

No	Function	Description
1	COMP	This pin is the Error Amplifier output and is made available for loop compensation.
2	V_{FB}	This is the inverting input of the Error Amplifier. It is normally connected to the switching power supply output through a resistor divider.
3	I_{SENSE}	A voltage proportional to inductor current is connected to this input. The PWM uses this information to terminate the output switch conduction.
4	R_T/C_T	The oscillator frequency and maximum Output duty cycle are programmed by connecting resistor R_T to V_{ref} and capacitor C_T to ground. Operation to 500kHz is possible.
5	GROUND	This pin is the combined control circuitry and power ground.
6	OUTPUT	This output directly drives the gate of a power MOSFET. Peak currents up to 1A are sourced and sunk by this pin.
7	V_{CC}	This pin is the positive supply of the control IC.
8	V_{ref}	This is the reference output. It provides charging current for capacitor C_T through resistor R_T .

ORDERING NUMBERS

SO8	Minidip
UC3842TD	UC3842TN
UC3843TD	UC3843TN
UC3844TD	UC3844TN
UC3845TD	UC3845TN

THERMAL DATA

Symbol	Description	Minidip	SO8	Unit
$R_{th\ j-amb}$	Thermal Resistance Junction-ambient. max.	100	150	°C/W

ELECTRICAL CHARACTERISTICS ([note 1] Unless otherwise stated, these specifications apply for $0 \leq T_{amb} \leq 105^{\circ}\text{C}$; $V_i = 15\text{V}$ (note 5); $R_T = 10\text{K}$; $C_T = 3.3\text{nF}$)

Symbol	Parameter	Test Conditions	Value			Unit
			Min.	Typ.	Max.	
REFERENCE SECTION						
ΔV_{REF}	Line Regulation	$12V \leq V_i \leq 25V$		2	20	mV
ΔV_{REF}	Load Regulation	$1 \leq I_o \leq 20mA$		3	25	mV
$\Delta V_{REF}/\Delta T$	Temperature Stability	(Note 2)		0.2		mV/°C
	Total Output Variation	Line, Load, Temperature	4.85		5.15	V
e_N	Output Noise Voltage	$10Hz \leq f \leq 10KHz$ $T_j = 25^{\circ}C$ (note 2)		50		μV
	Long Term Stability	$T_{amb} = 125^{\circ}C$, 1000Hrs (note 2)		5	25	mV
I_{SC}	Output Short Circuit		-30	-100	-180	mA
OSCILLATOR SECTION						
f_{osc}	Frequency	$T_j = 25^{\circ}C$ $T_A = T_{low}$ to T_{high} $T_j = 25^{\circ}C$ ($R_T = 6.2k$, $C_T = 1nF$)	49 48 225	52 – 250	55 56 275	KHz KHz KHz
$\Delta f_{osc}/\Delta V$	Frequency Change with Volt.	$V_{CC} = 12V$ to $25V$	–	0.2	1	%
$\Delta f_{osc}/\Delta T$	Frequency Change with Temp.	$T_A = T_{low}$ to T_{high}	–	1	–	%
V_{OSC}	Oscillator Voltage Swing	(peak to peak)	–	1.6	–	V
I_{dischg}	Discharge Current ($V_{OSC} = 2V$)	$T_A = T_{low}$ to T_{high}	7.3	–	8.8	mA
ERROR AMP SECTION						
V_2	Input Voltage	$V_{PIN1} = 2.5V$	2.42	2.50	2.58	V
I_b	Input Bias Current	$V_{FB} = 5V$		-0.1	-2	μA
	A_{VOL}	$2V \leq V_o \leq 4V$	65	90		dB
BW	Unity Gain Bandwidth	$T_j = 25^{\circ}C$	0.7	1		MHz
PSRR	Power Supply Rejec. Ratio	$12V \leq V_i \leq 25V$	60	70		dB
I_o	Output Sink Current	$V_{PIN2} = 2.7V$ $V_{PIN1} = 1.1V$	2	12		mA
I_o	Output Source Current	$V_{PIN2} = 2.3V$ $V_{PIN1} = 5V$	-0.5	-1		mA
	V_{OUT} High	$V_{PIN2} = 2.3V$; $R_L = 15K\Omega$ to Ground	5	6.2		V
	V_{OUT} Low	$V_{PIN2} = 2.7V$; $R_L = 15K\Omega$ to Pin 8		0.8	1.1	V
CURRENT SENSE SECTION						
G_V	Gain	(note 3 & 4)	2.85	3	3.15	V/V
V_3	Maximum Input Signal	$V_{PIN1} = 5V$ (note 3)	0.9	1	1.1	V
SVR	Supply Voltage Rejection	$12 \leq V_i \leq 25V$ (note 3)		70		dB
I_b	Input Bias Current			-2	-10	μA
	Delay to Output			100	300	ns

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ELECTRICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test Conditions	Value			Unit
			Min.	Typ.	Max.	
OUTPUT SECTION						
V _{OL}	Output Low Level	I _{SINK} = 20mA		0.1	0.4	V
		I _{SINK} = 200mA		1.6	2.2	V
V _{OH}	Output High Level	I _{SOURCE} = 20mA	13	13.5		V
		I _{SOURCE} = 200mA	12	13.5		V
V _{OLS}	UVLO Saturation	V _{CC} = 6V; I _{SINK} = 1mA		0.1	1.1	V
t _r	Rise Time	T _J = 25°C C _L = 1nF (2)		50	150	ns
t _f	Fall Time	T _J = 25°C C _L = 1nF (2)		50	150	ns
UNDER-VOLTAGE LOCKOUT SECTION						
	Start Threshold	UC3842T/4T	15	16	17	V
		UC3843T/5T	7.8	8.4	9.0	V
	Min Operating Voltage After Turn-on	UC3842T/4T	9	10	11	V
		UC3843T/5T	7.0	7.6	8.2	V
PWM SECTION						
	Maximum Duty Cycle	UC3842T/3T	94	96	100	%
		UC3844T/5T	47	48	50	%
	Minimum Duty Cycle				0	%
TOTAL STANDBY CURRENT						
I _{st}	Start-up Current	V _i = 6.5V for UC3843T/45T		0.3	0.5	mA
		V _i = 14V for UC3842T/44T		0.3	0.5	mA
I _i	Operating Supply Current	V _{PIN2} = V _{PIN3} = 0V		12	17	mA
V _{iz}	Zener Voltage	I _i = 25mA	30	36		V

- Notes :**
1. Max package power dissipation limits must be respected; low duty cycle pulse techniques are used during test maintain T_j as close to T_{amb} as possible.
 2. These parameters, although guaranteed, are not 100% tested in production.
 3. Parameter measured at trip point of latch with V_{PIN2} = 0.
 4. Gain defined as :

$$A = \frac{\Delta V_{PIN1}}{\Delta V_{PIN3}} ; 0 \leq V_{PIN3} \leq 0.8V$$
 5. Adjust V_i above the start threshold before setting at 15 V.

Figure 4: Oscillator Discharge Current vs. Temperature.

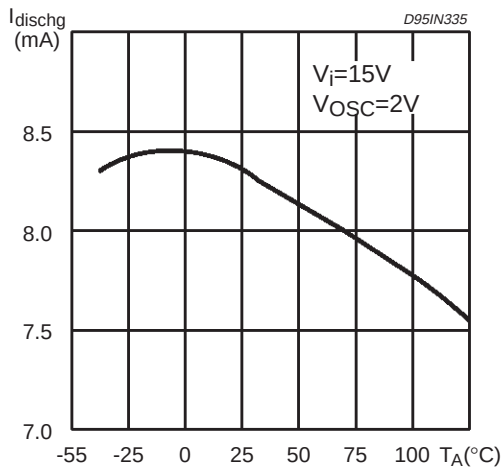


Figure 5: Maximum Output Duty Cycle vs. Timing Resistor.

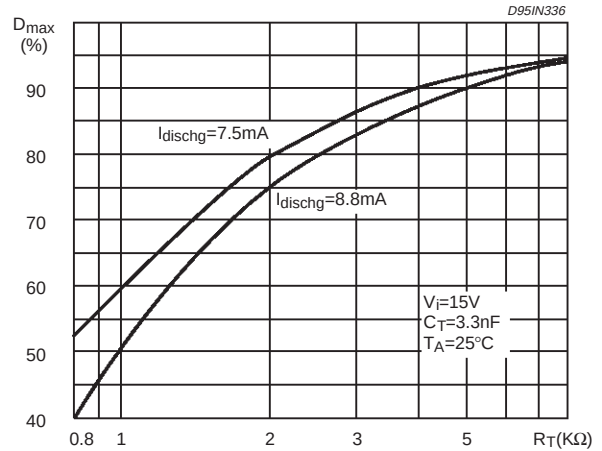


Figure 6: Error Amp Open-Loop Gain and Phase vs. Frequency.

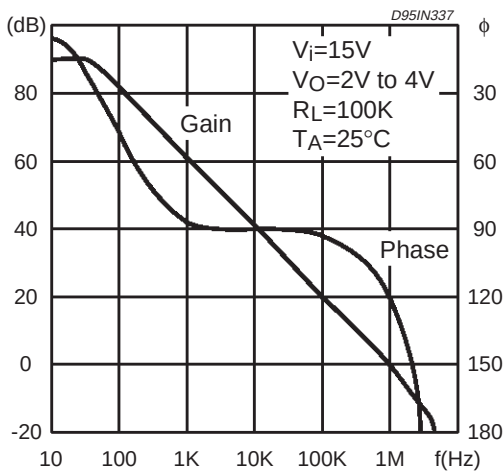


Figure 7: Current Sense Input Threshold vs. Error Amp Output Voltage.

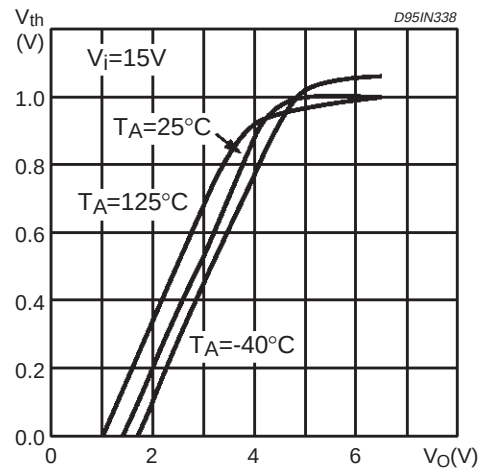


Figure 8: Reference Voltage Change vs. Source Current.

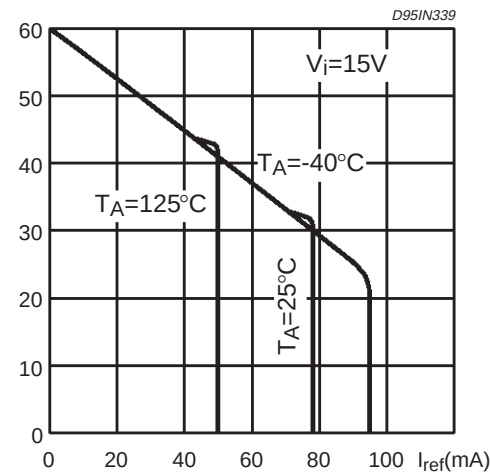


Figure 9: Reference Short Circuit Current vs. Temperature.

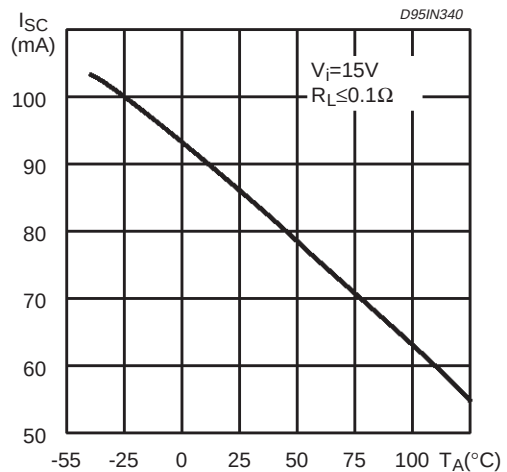


Figure 10: Output Saturation Voltage vs. Load Current.

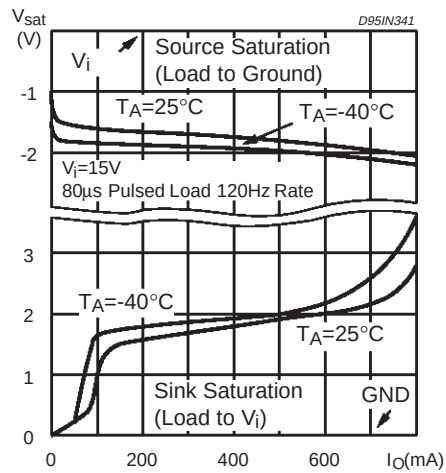


Figure 12: Output Waveform.

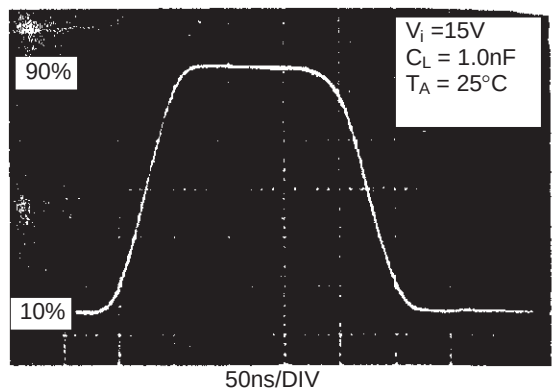


Figure 14: Oscillator and Output Waveforms.

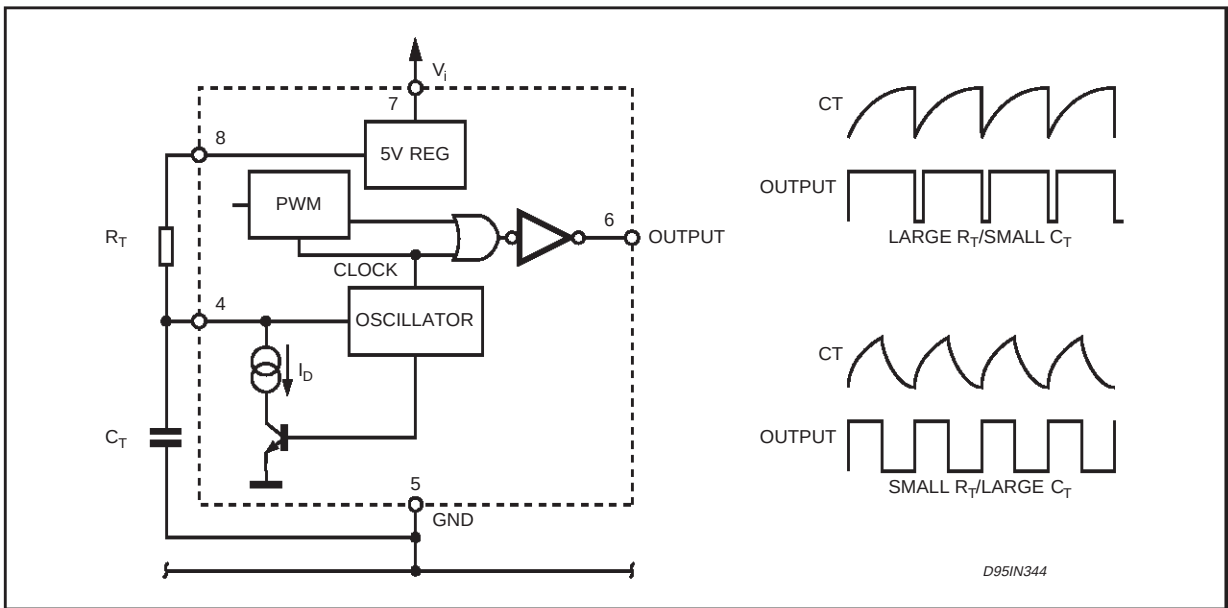


Figure 15 : Error Amp Configuration.

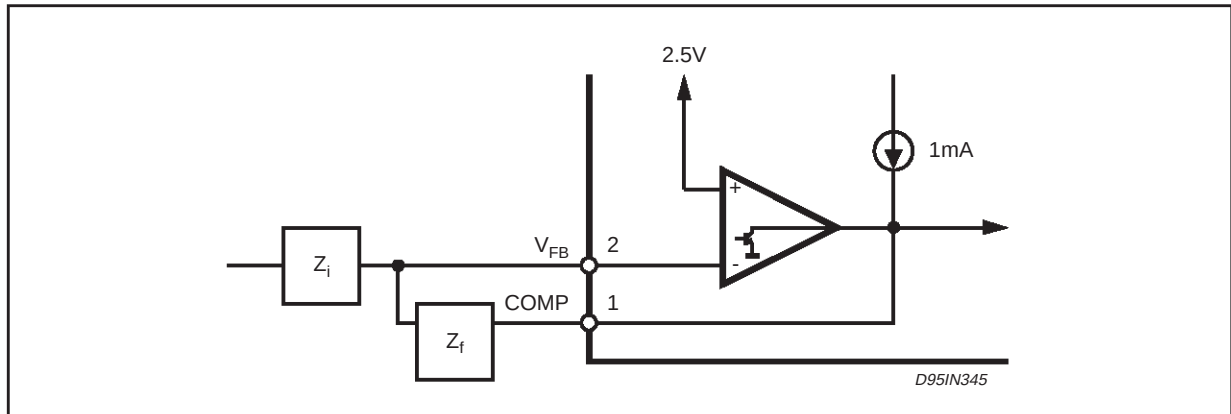


Figure 16 : Under Voltage Lockout.

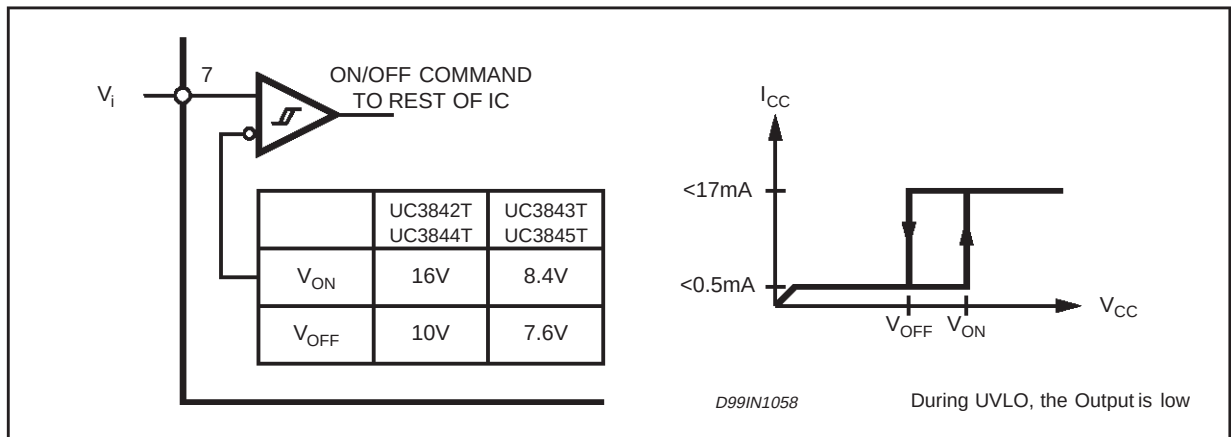
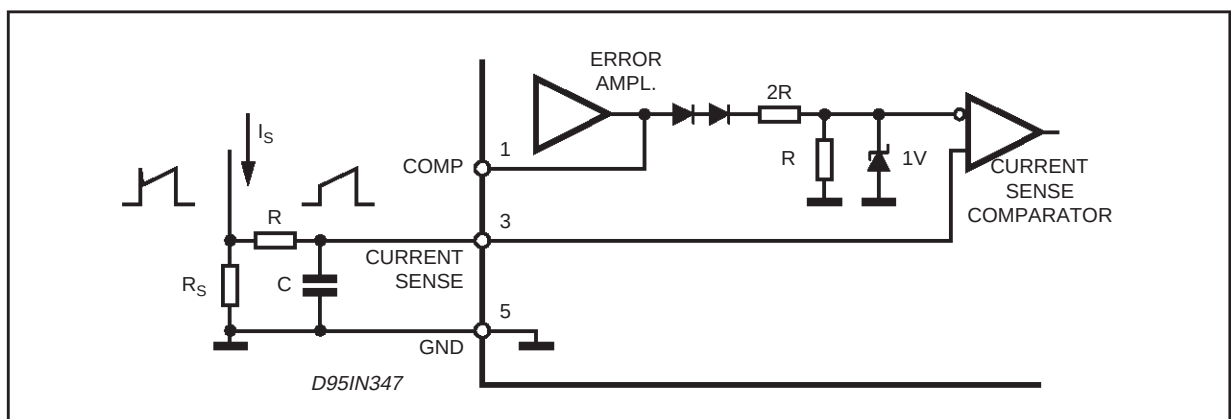


Figure 17 : Current Sense Circuit .



Peak current (i_s) is determined by the formula

$$I_{S \text{ max}} \approx \frac{1.0 \text{ V}}{R_s}$$

A small RC filter may be required to suppress switch transients.

Figure 18 : Slope Compensation Techniques.

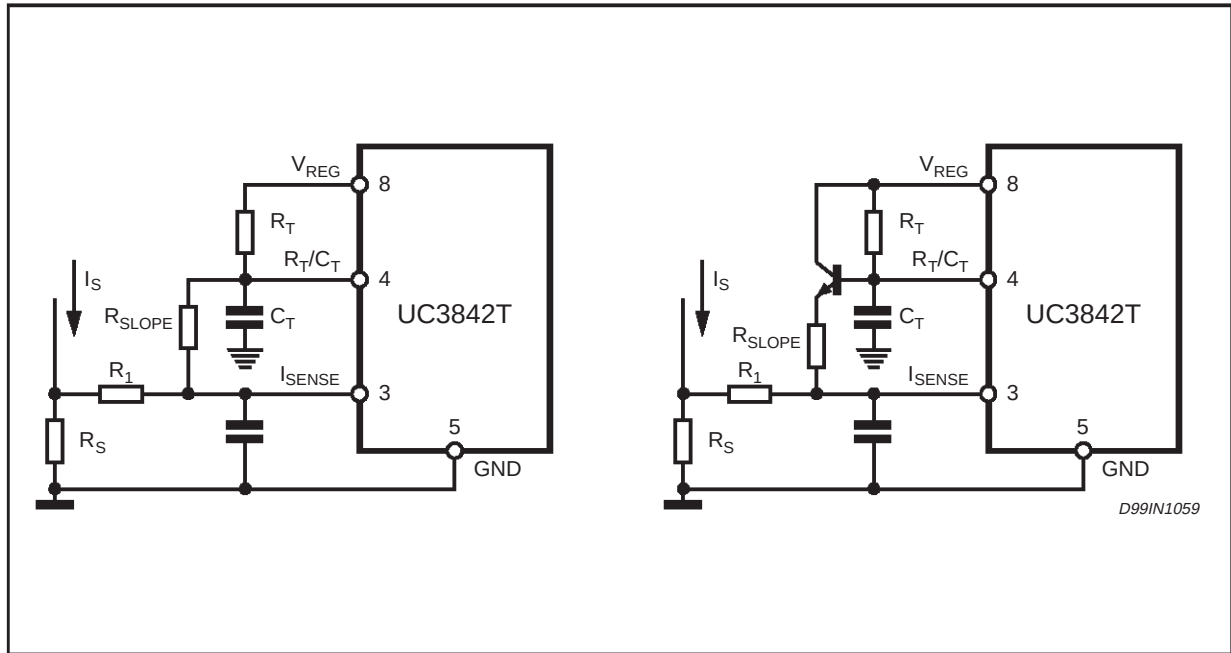


Figure 19 : Isolated MOSFET Drive and Current Transformer Sensing.

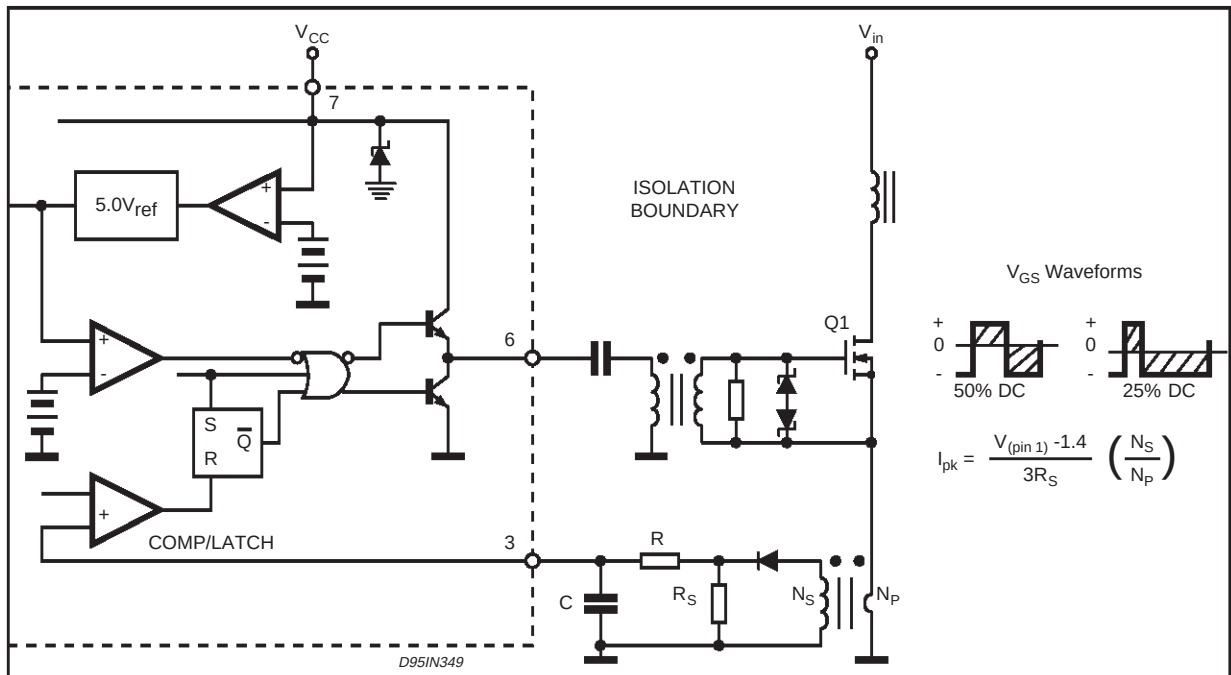


Figure 20 : Latched Shutdown.

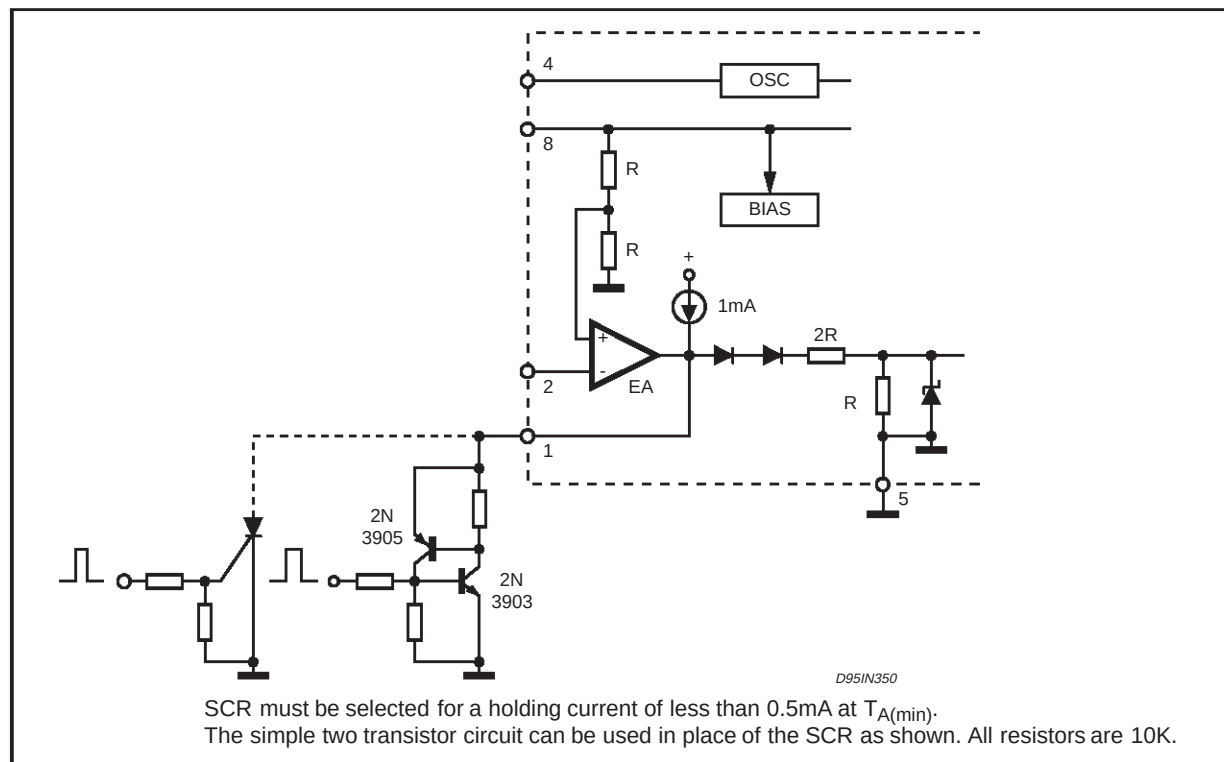


Figure 21: Error Amplifier Compensation

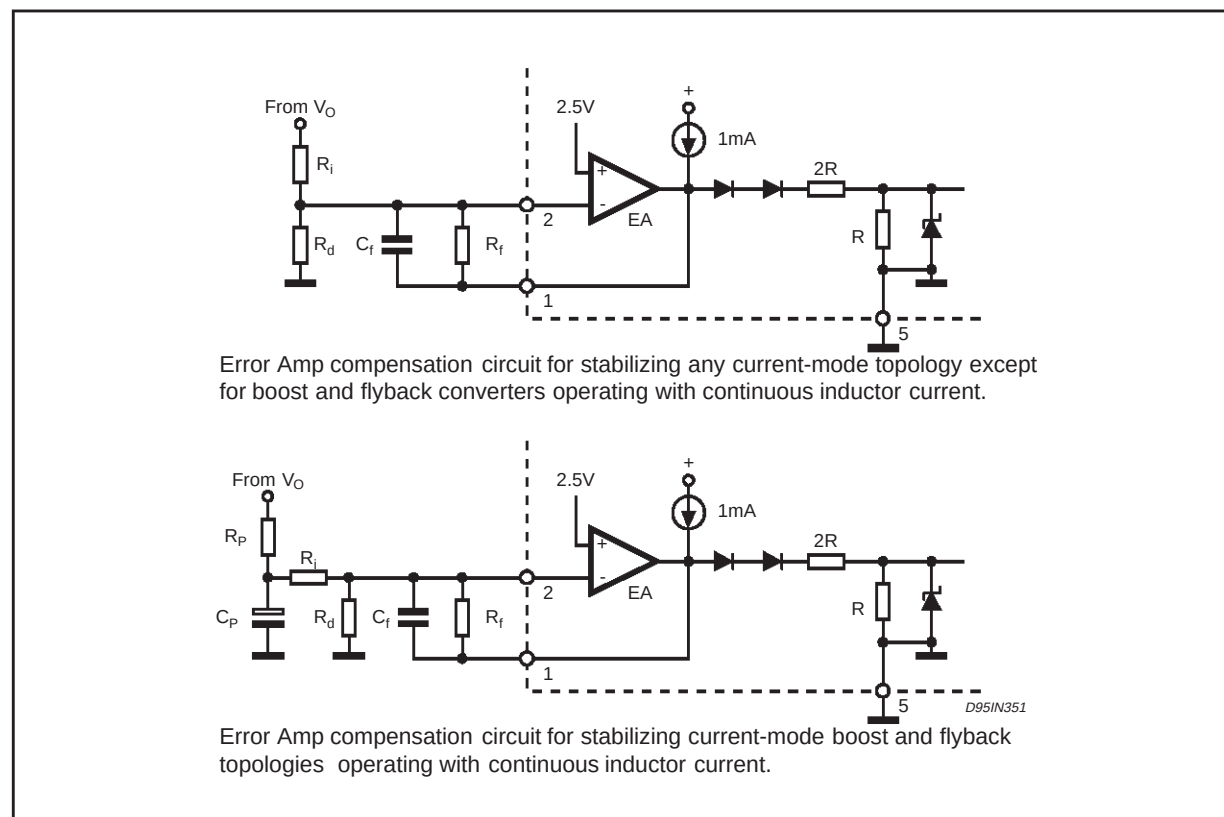


Figure 22: External Clock Synchronization.

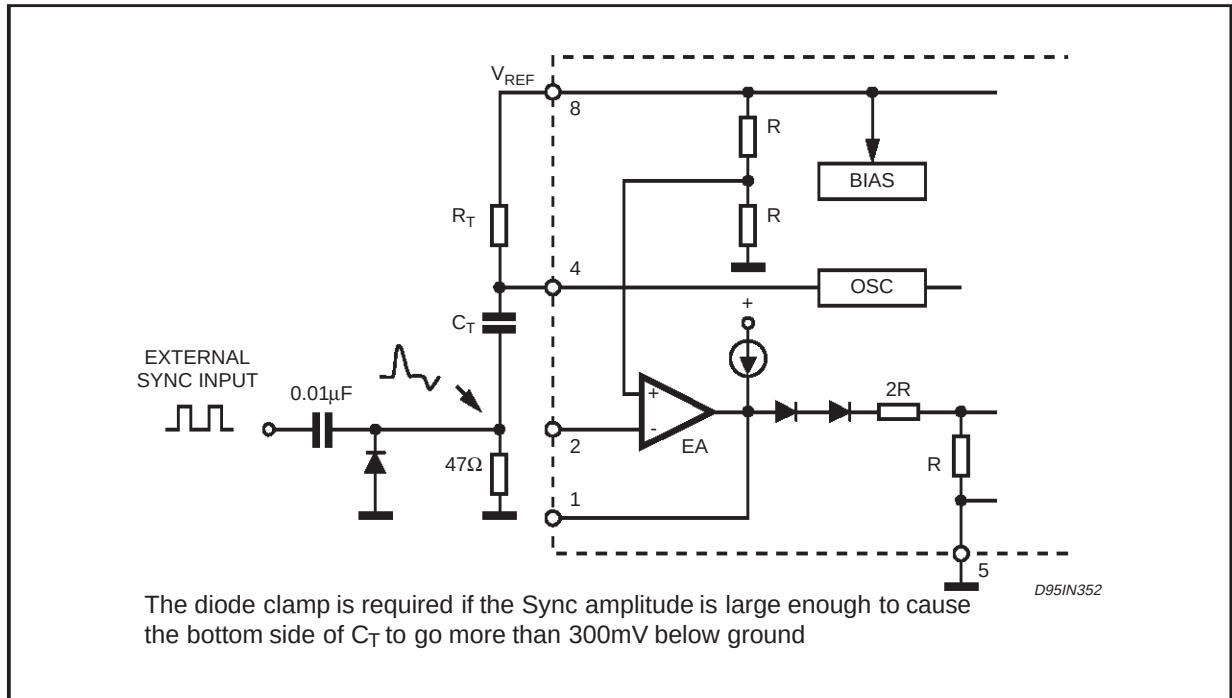


Figure 23: External Duty Cycle Clamp and Multi Unit Synchronization.

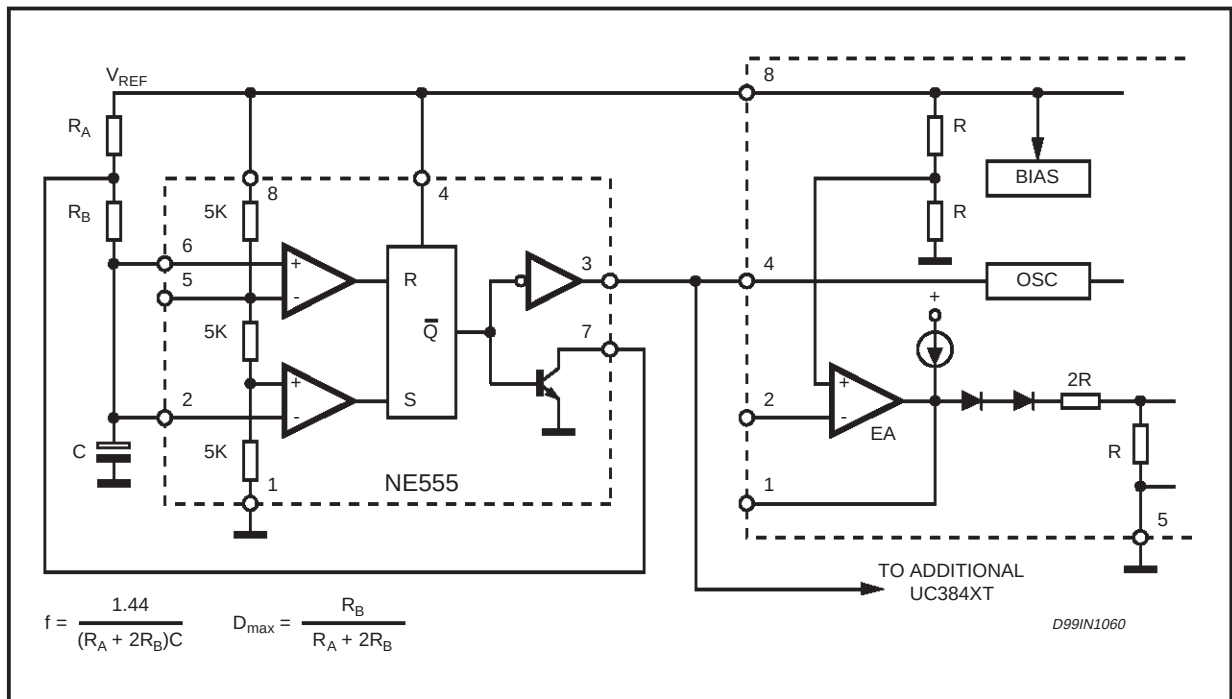


Figure 24: Soft-Start Circuit

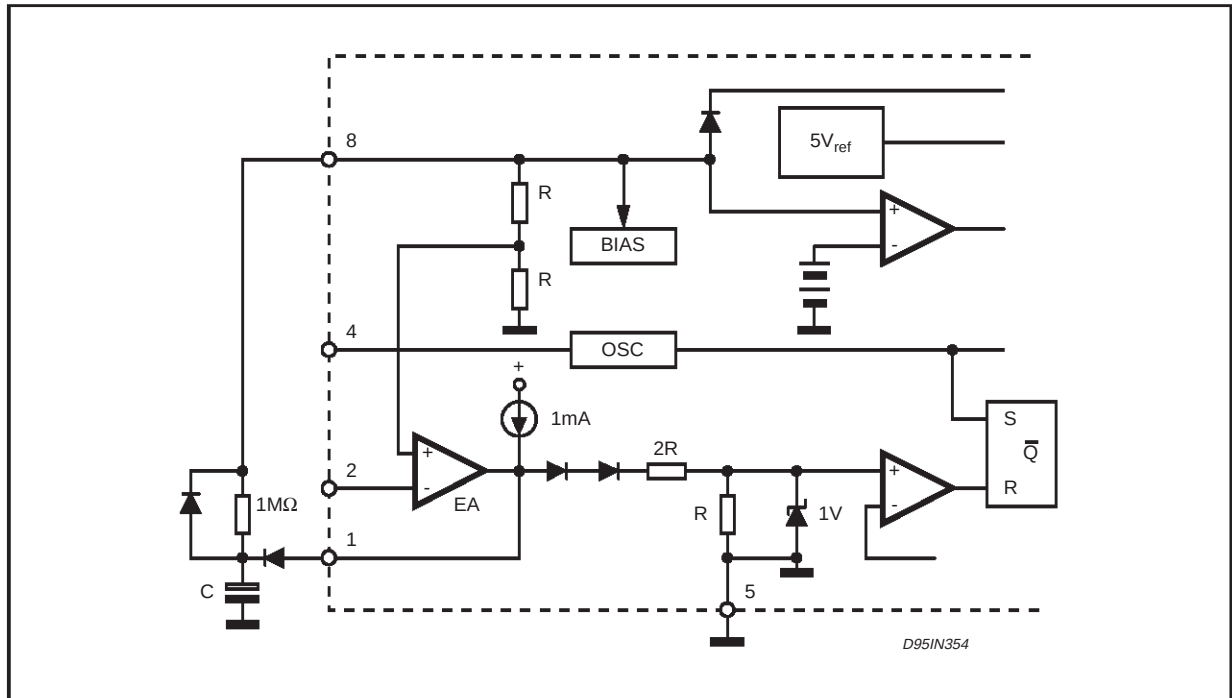
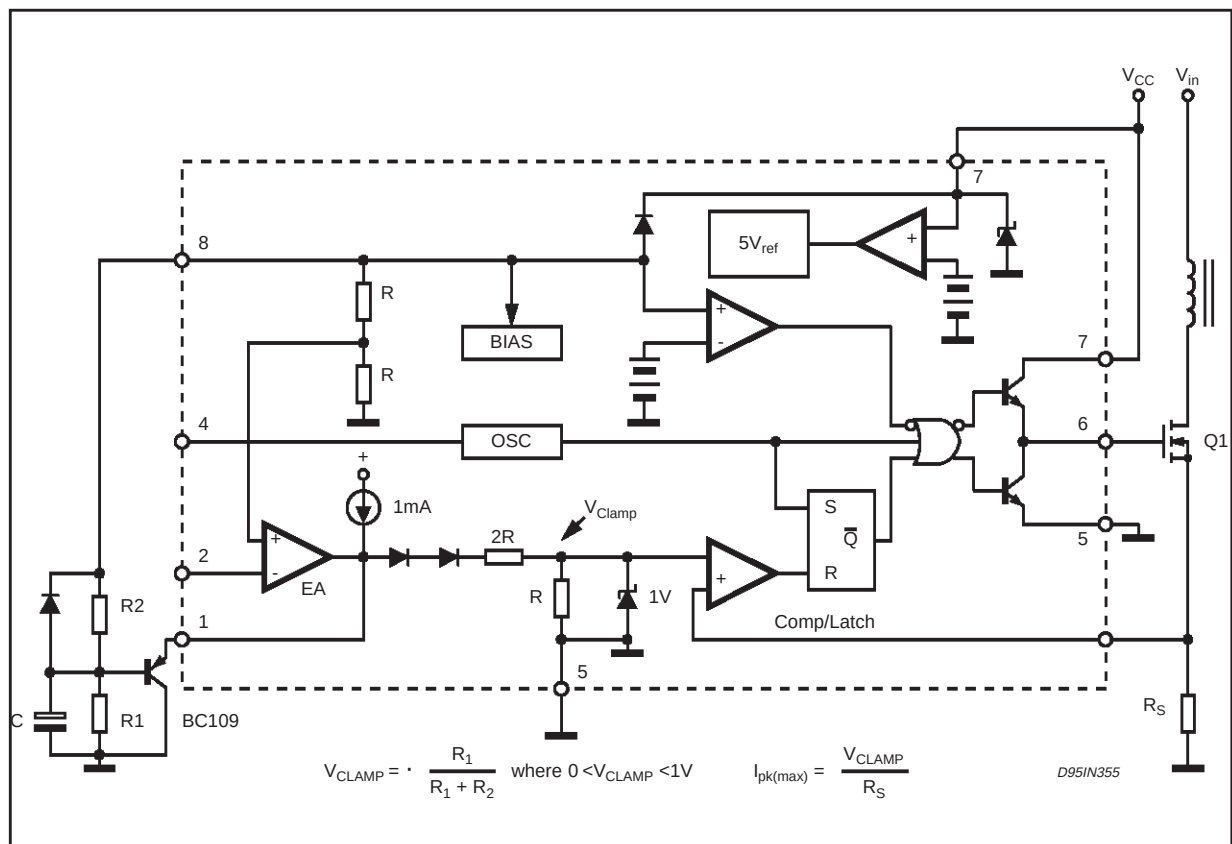


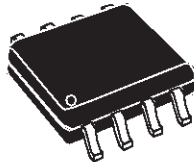
Figure 25: Soft-Start and Error Amplifier Output Duty Cycle Clamp.



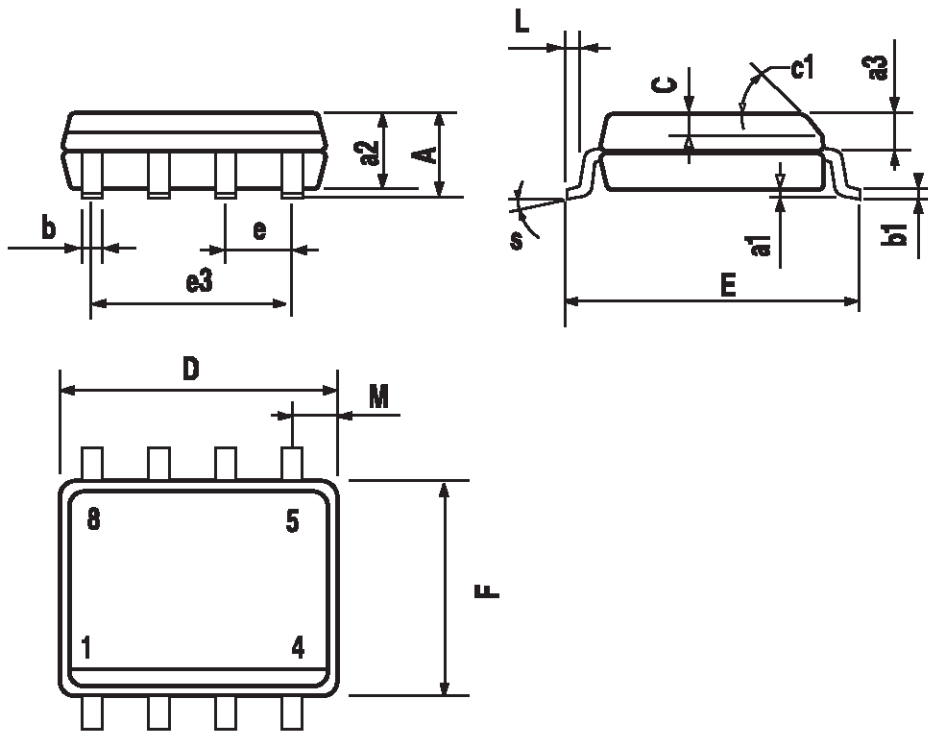
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			1.75			0.069
a1	0.1		0.25	0.004		0.010
a2			1.65			0.065
a3	0.65		0.85	0.026		0.033
b	0.35		0.48	0.014		0.019
b1	0.19		0.25	0.007		0.010
C	0.25		0.5	0.010		0.020
c1	45° (typ.)					
D (1)	4.8		5.0	0.189		0.197
E	5.8		6.2	0.228		0.244
e		1.27			0.050	
e3		3.81			0.150	
F (1)	3.8		4.0	0.15		0.157
L	0.4		1.27	0.016		0.050
M			0.6			0.024
S	8° (max.)					

(1) D and F do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm (.006inch).

OUTLINE AND MECHANICAL DATA



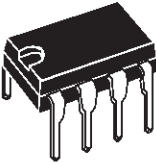
SO8



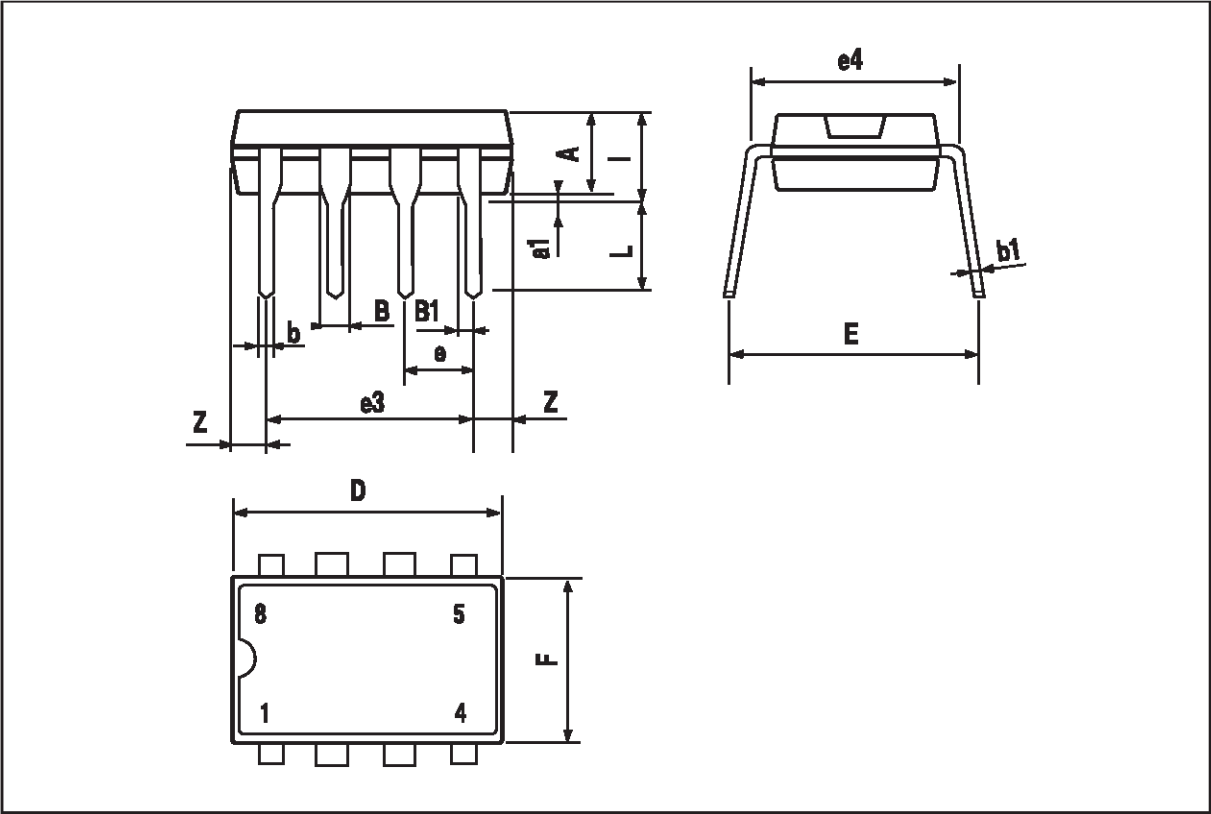
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DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A		3.32			0.131	
a1	0.51			0.020		
B	1.15		1.65	0.045		0.065
b	0.356		0.55	0.014		0.022
b1	0.204		0.304	0.008		0.012
D			10.92			0.430
E	7.95		9.75	0.313		0.384
e		2.54			0.100	
e3		7.62			0.300	
e4		7.62			0.300	
F			6.6			0.260
I			5.08			0.200
L	3.18		3.81	0.125		0.150
Z			1.52			0.060

OUTLINE AND MECHANICAL DATA



Minidip



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