

DC-to-DC Converter Control Circuits

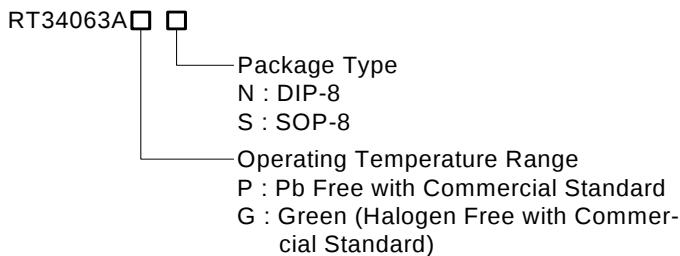
General Description

The RT34063A Series is a monolithic control circuit containing the primary functions required for DC-to-DC converters.

These devices consist of an internal temperature compensated reference, comparator, controlled duty cycle oscillator with an active current limit circuit, driver and high current output switch.

This series was specially designed to be incorporated in step-down and step-up and Voltage-inverting applications with a minimum number of external components.

Ordering Information



Note :

RichTek Pb-free and Green products are :

- ▶RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶Suitable for use in SnPb or Pb-free soldering processes.
- ▶100%matte tin (Sn) plating.

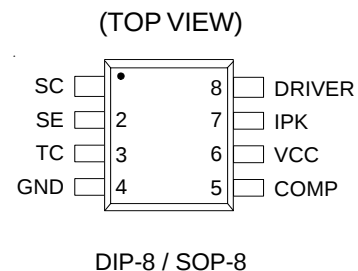
Features

- Operation from 3.0V to 30V Input
- Low Standby Current
- Current Limiting
- Internal Switch Current to 1.5A
- Output Voltage Adjustable
- Frequency Operation to 100kHz
- Precision 2% Reference
- RoHS Compliant and 100% Lead (Pb)-Free

Applications

- Saver for Cellular Phones
- DC-DC Converter Module

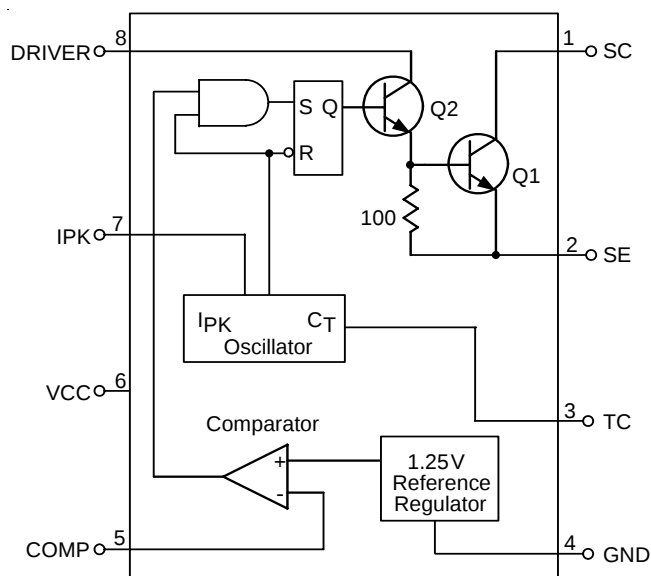
Pin Configurations



Functional Pin Description

Pin Name	Pin Function
SC	1.5A Switch Collector
SE	Darlington Switch Emitter
TC	Oscillator Timing Capacitor
GND	Power GND
COMP	Feedback Comparator Inverting Input
VCC	Power Supply Input
IPK	Highside Current Sense Input $V_{CC}-V_{IPK} = 330mV$
DRIVER	Driver Collector

Function Block Diagram



Typical Application Circuit

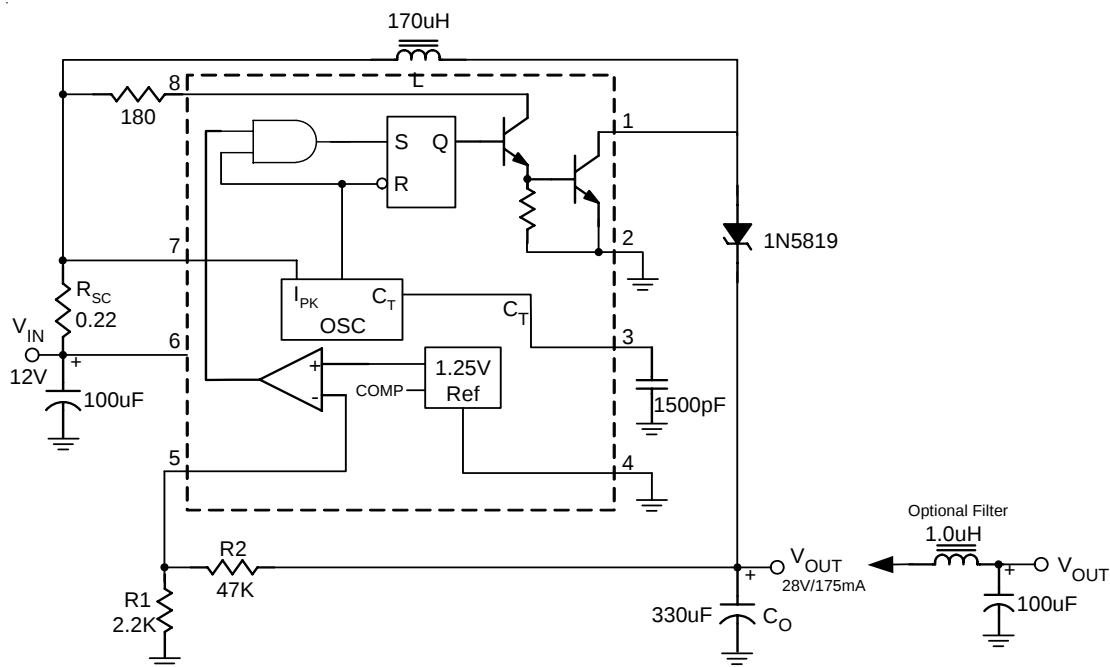
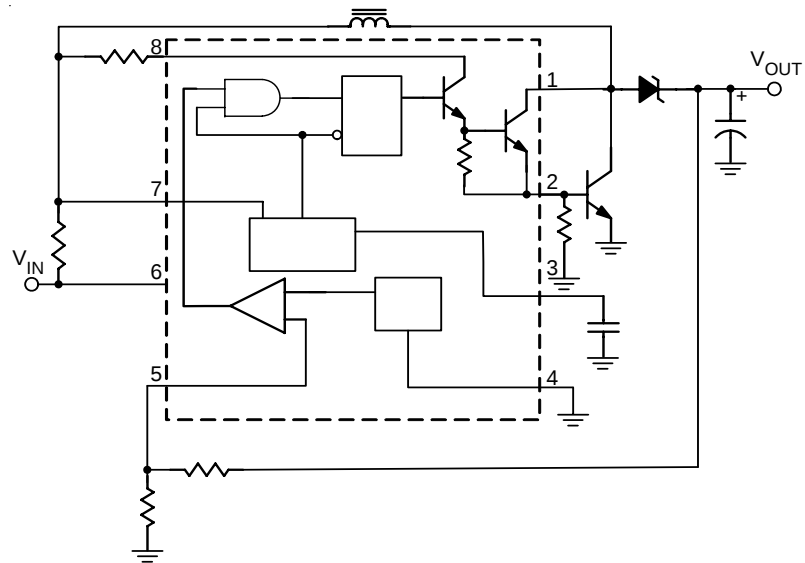
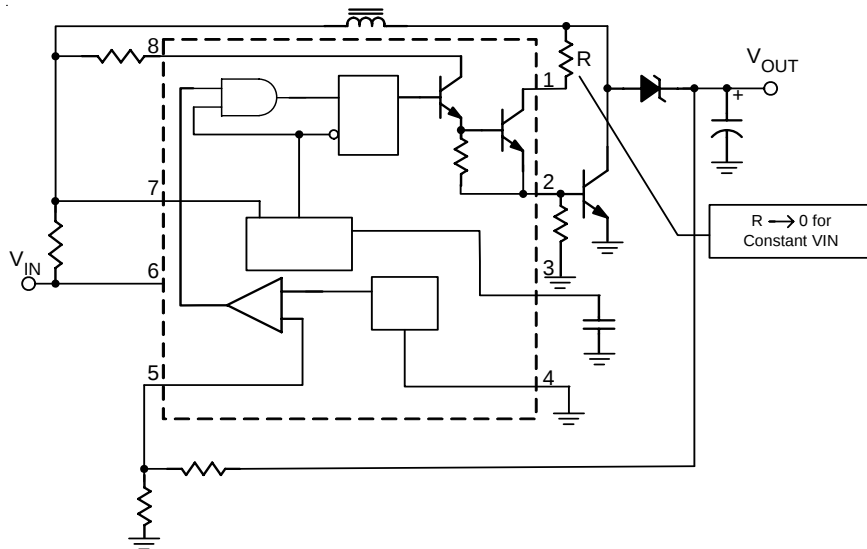


Figure 1. Step-up Converter



(a) External NPN Switch



(b) External NPN Saturated Switch (See Note)

Figure 2. External Current Boost Connections for I_C Peak Greater than 1.5A

Note: If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents ($I_C \approx 300\text{mA}$) and high driver currents ($I_B \approx 30\text{mA}$), it may take up to $2.0\mu\text{s}$ to come out of saturation. This condition will shorten the off time at frequencies $\geq 30\text{kHz}$, and is magnified at high temperature. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended.

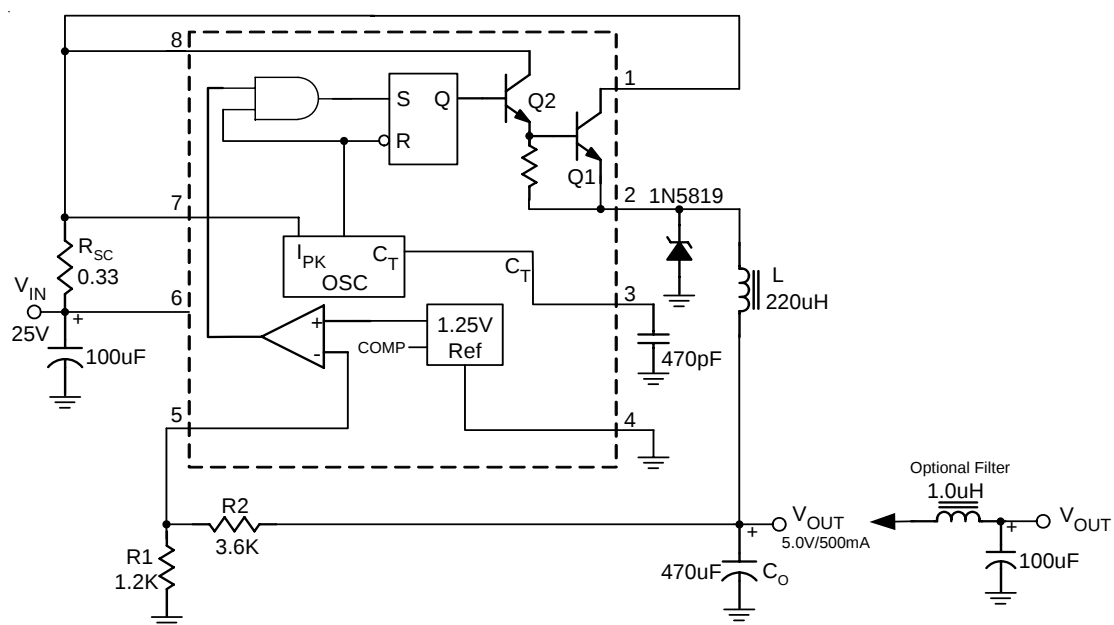
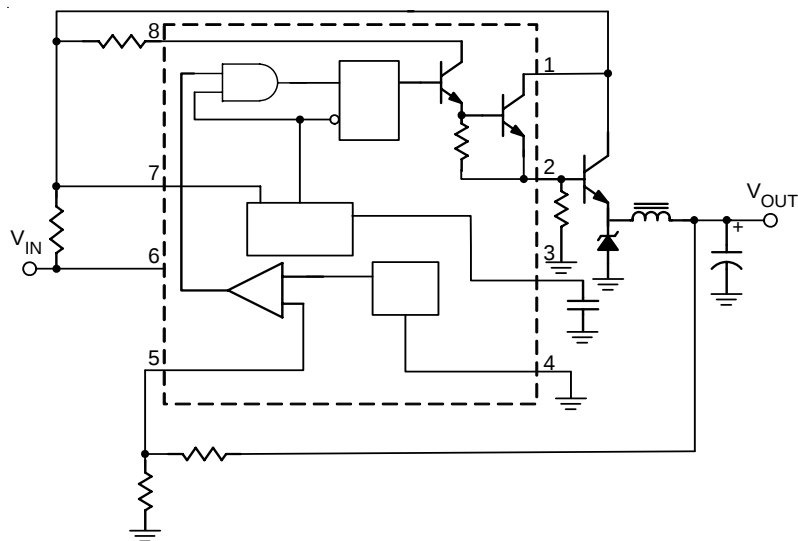
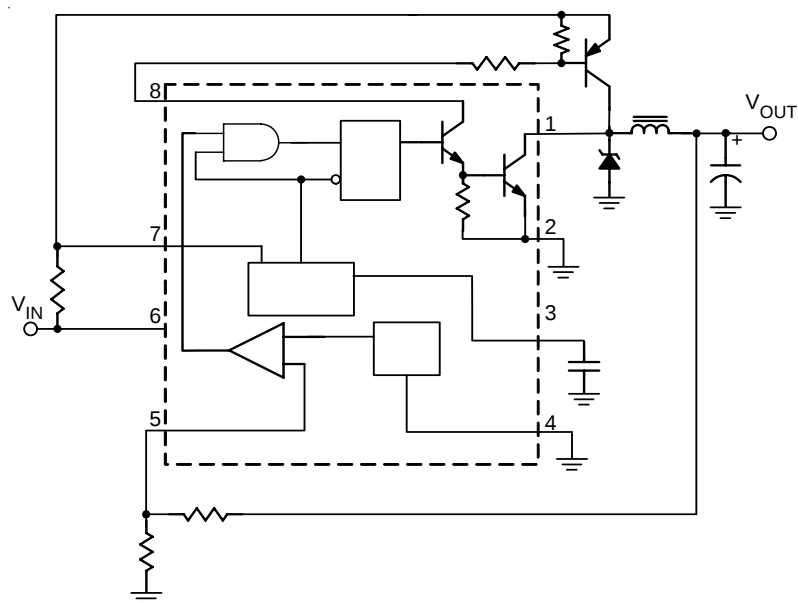


Figure 3. Step-down Converter



(a) External NPN Switch



(b) External PNP Saturated Switch

Figure 4. External Current Boost Connections for I_C Peak Greater than 1.5A

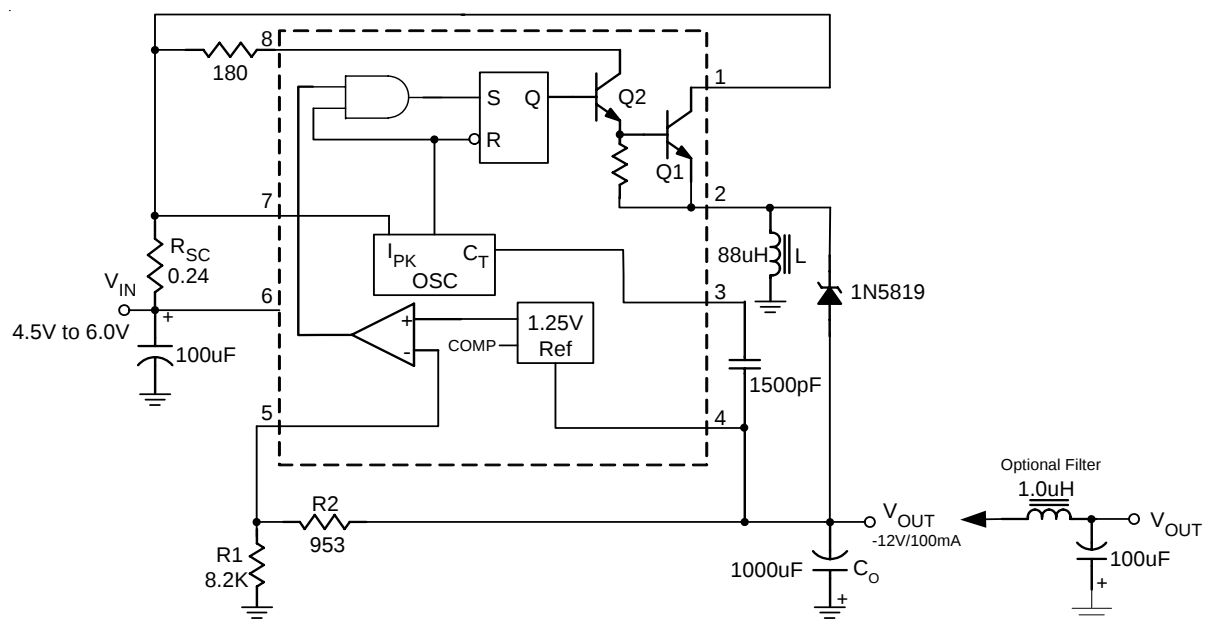
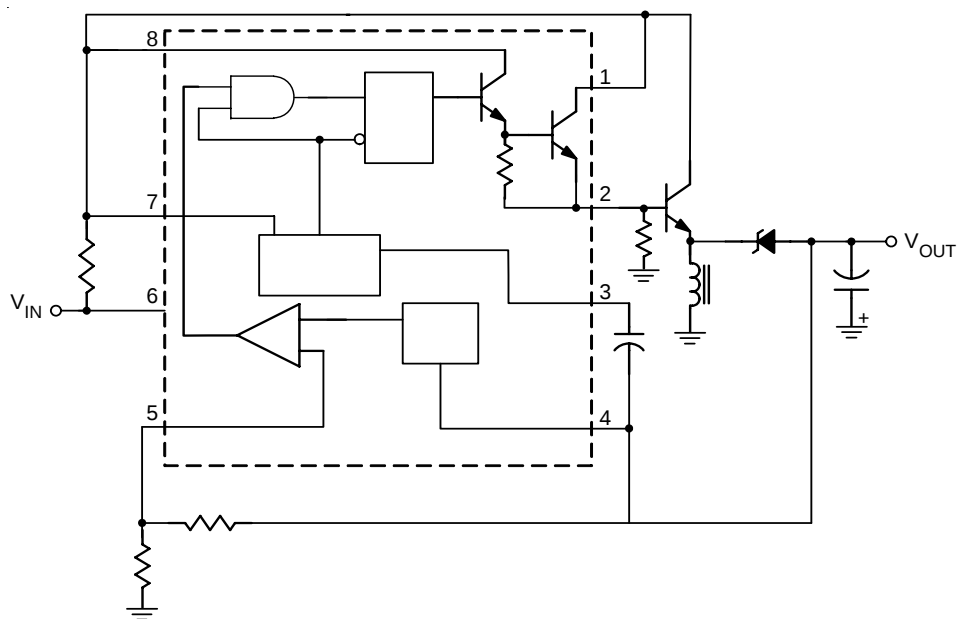
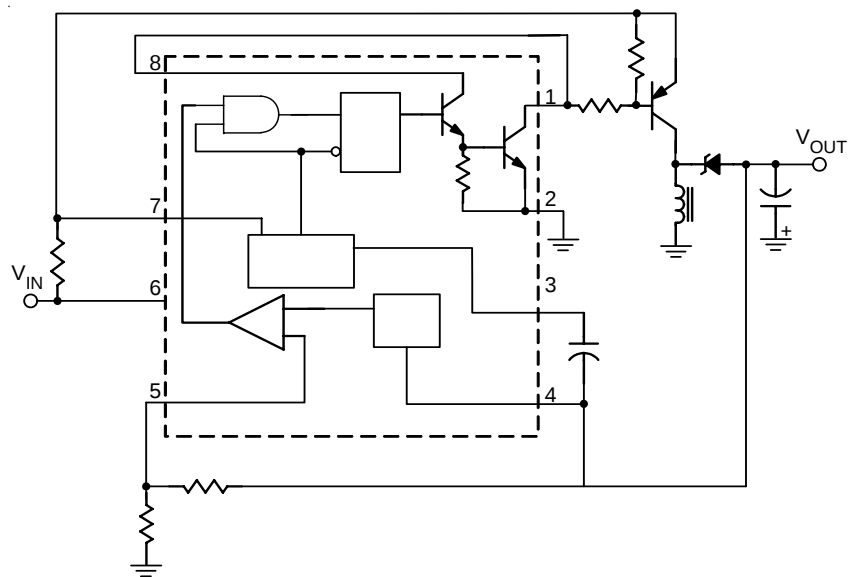


Figure 5. Voltage Inverting Converter



(a) External NPN Switch



(b) External PNP Saturated Switch

Figure 6. External Current Boost Connections for Peak Greater than 1.5A

Design Formula Table

Calculation	Step-up	Step-down	Voltage-Inverting
t_{on}/t_{off}	$\frac{V_{OUT} + V_F - V_{IN(MIN)}}{V_{IN(MIN)} - V_{SAT}}$	$\frac{V_{OUT} + V_F}{V_{IN(MIN)} - V_{SAT} - V_{OUT}}$	$\frac{ V_{OUT} + V_F}{V_{IN} - V_{SAT}}$
$(t_{on} + t_{off})$	$\frac{1}{f}$	$\frac{1}{f}$	$\frac{1}{f}$
t_{off}	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$
t_{on}	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$
C_T	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$
$I_{pk(SWITCH)}$	$2 I_{OUT(MAX)} \left(\frac{t_{on}}{t_{off}} + 1 \right)$	$2 I_{OUT(MAX)}$	$2 I_{OUT(MAX)} \left(\frac{t_{on}}{t_{off}} + 1 \right)$
R_{SC}	$0.3/I_{pk(SWITCH)}$	$0.3/I_{pk(SWITCH)}$	$0.3/I_{pk(SWITCH)}$
$L(\min)$	$\left(\frac{(V_{IN(MIN)} - V_{SAT})}{I_{pk(SWITCH)}} \right) t_{on(MAX)}$	$\left(\frac{(V_{IN(MIN)} - V_{SAT})}{I_{pk(SWITCH)}} \right) t_{on(MAX)}$	$\left(\frac{(V_{IN(MIN)} - V_{SAT})}{I_{pk(SWITCH)}} \right) t_{on(MAX)}$
C_O	$9 \frac{I_{OUT} t_{on}}{V_{ripple(pp)}}$	$\frac{I_{pk(SWITCH)}(t_{on} + t_{off})}{8V_{ripple(pp)}}$	$9 \frac{I_{OUT} t_{on}}{V_{ripple(pp)}}$

V_{SAT} : Saturation voltage of the output switch.

V_F : Forward voltage drop of the output rectifier.

The following power supply characteristics must be chosen:

V_{IN} : Nominal input voltage.

V_{OUT} : Desired output voltage $|V_{OUT}| = 1.25(1 + \frac{R_2}{R_1})$

I_{OUT} : Desired output current.

f : Minimum desired output switching frequency at the selected values of V_{in} and I_O .

$V_{ripple(pp)}$: Desired peak-to-peak output ripple voltage. In practice, the calculated capacitor value needs to be increased due to its equivalent series resistance and board layout. The ripple voltage should be kept to a low value since it directly affects the line and load regulation.

Absolute Maximum Ratings

- Power Supply Voltage ----- 30V
- Feedback Input Voltage Range ----- -0.3 to +30V
- Switch Collector Voltage ----- 30V
- Switch Emitter Voltage ----- 30V
- Switch Collector to Emitter Voltage ----- 30V
- Driver Collector Voltage ----- 30V
- Driver Collector Current (see Note) ----- 100mA
- Switch Current ----- 1.5A
- Power Dissipation, P_D @ $T_A = 25^\circ\text{C}$
 - DIP-8 ----- 1.25W
 - SOP-8 ----- 0.625W
- Package Thermal Resistance
 - DIP-8, θ_{JA} ----- 100°C/W
 - SOP-8, θ_{JA} ----- 160°C/W
- Operating Junction Temperature ----- $+125^\circ\text{C}$
- Storage Temperature Range ----- 65 to $+150^\circ\text{C}$

Note: Maximum package power dissipation limits must be observed.

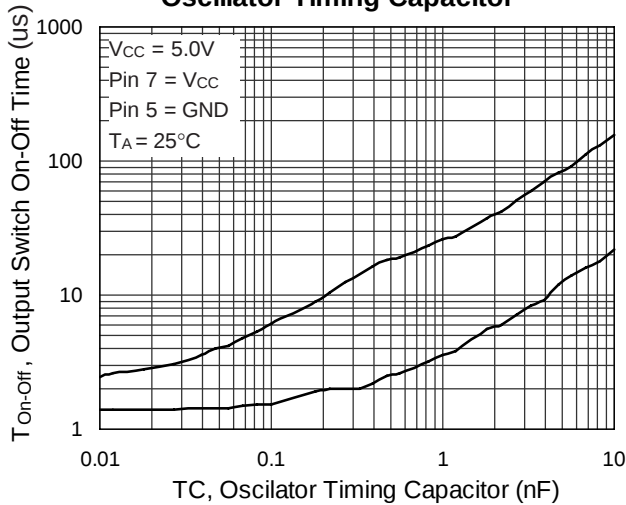
Electrical Characteristics

($V_{CC} = 5V$, $T_A = 25^\circ\text{C}$, unless otherwise specified)

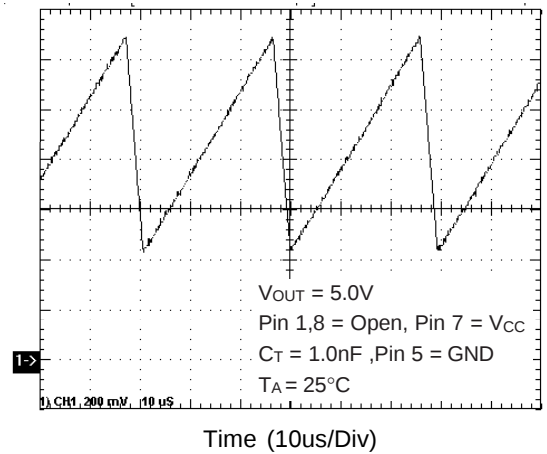
Parameter	Symbol	Test Conditions	Min	Typ	Max	Units
Oscillator						
Frequency		$V_{PIN5} = 0V$, $C_T = 1.0nF$	26	38	48	kHz
Charge Current	I_{CHG}	$5.0V \leq V_{CC} \leq 30V$	25	36	43	μA
Discharge Current	I_{DISCHG}	$5.0V \leq V_{CC} \leq 30V$	160	250	290	μA
Discharge to Charge Current Ratio		Pin 7 to VCC	5.5	6.9	7.9	--
Current Limit Sense Voltage	V_{LIMIT}	$I_{CHG} = I_{DISCHG}$	280	330	380	mV
Output Switch						
Saturation Voltage, Darlington Connection		$I_{SW} = 1.0A$, Pins1, 8 connected	--	1.0	1.3	V
Saturation Voltage, Darlington Connection		$I_{SW} = 1.0A$, $R_{PIN8} = 82\Omega$ to VCC, Forced $\beta \approx 20$	--	0.45	0.7	V
DC Current Gain		$I_{SW} = 1.0A$, $V_{CE} = 5.0V$	50	75	--	--
Collector Off-state Current		$V_{CE} = 30V$	--	0.01	100	μA
Comparator						
Threshold Voltage			1.225	1.25	1.275	V
Threshold Voltage Line Regulation		$3.0V \leq V_{CC} \leq 30V$	--	1.4	5.0	mV
Input Bias Current	I_{BIAS}	$V_{IN} = 0V$	--	-20	-400	nA
Total Device						
Supply Current	I_{CC}	$V_{CC} = 5.0V$ to $30V$, $C_T = 1.0nF$, Pin7 = VCC, $V_{PIN5} > V_{TH}$, Pin2 = GND, Remaining pins open	--	3.0	4.5	mA

Typical Operating Characteristics

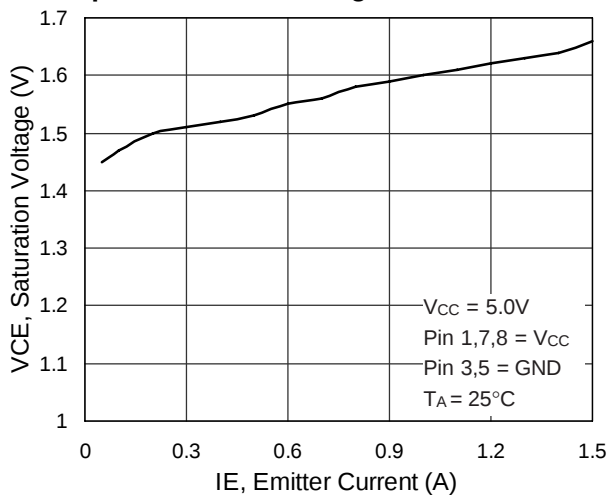
Output Switch On-Off Time vs. Oscillator Timing Capacitor



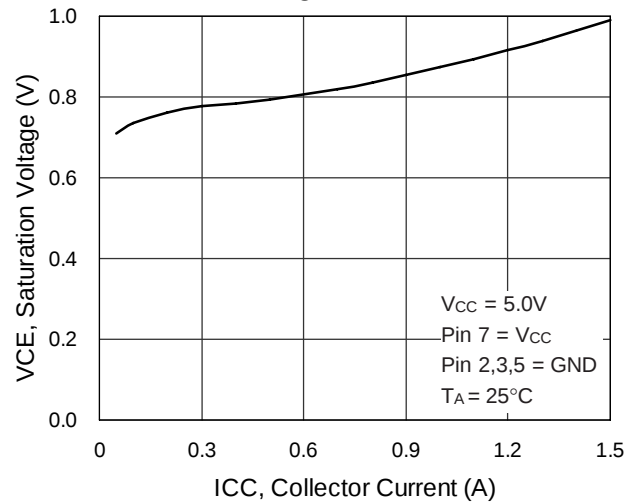
Timing Capacitor Waveform



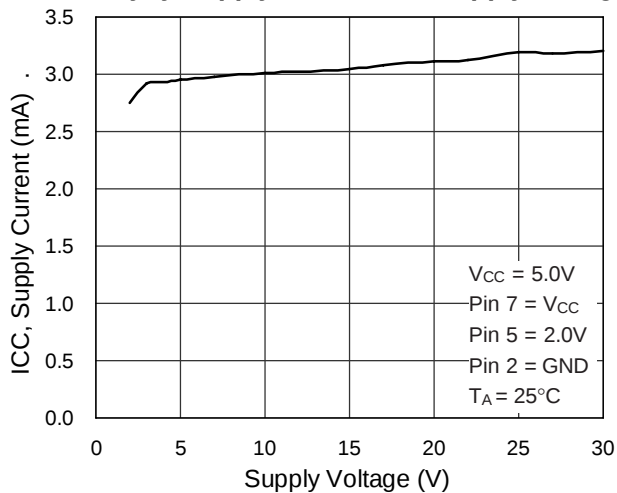
Emitter Follower Configuration Output Saturation Voltage vs. Emter Current



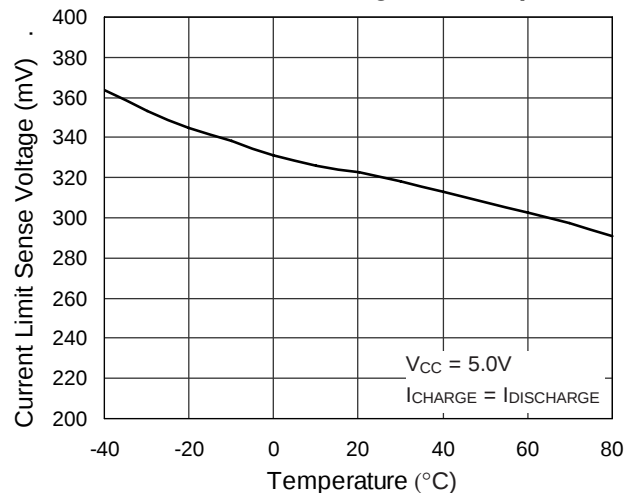
Common Emitter Configuration Output Switch Saturation Voltage vs. Collector Current



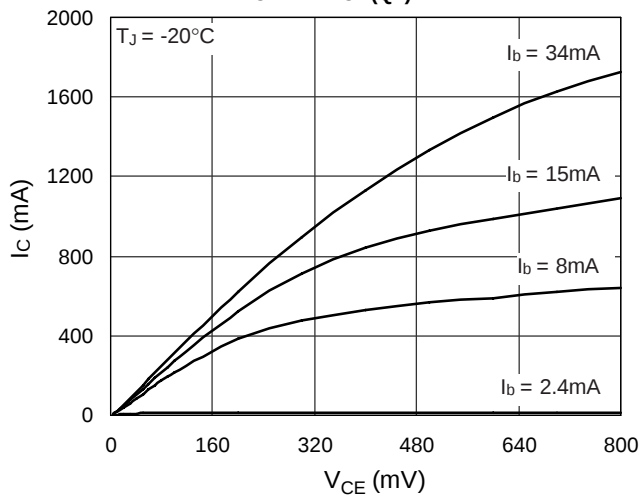
Standby Supply Current vs. Supply Voltage



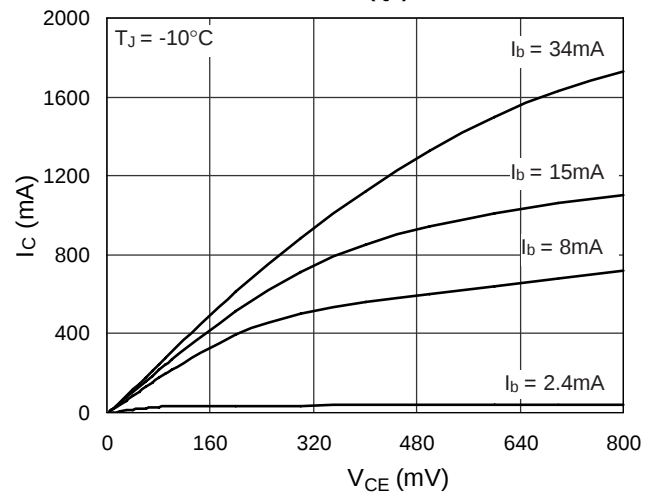
Current Limit Sense Voltage vs. Temperature



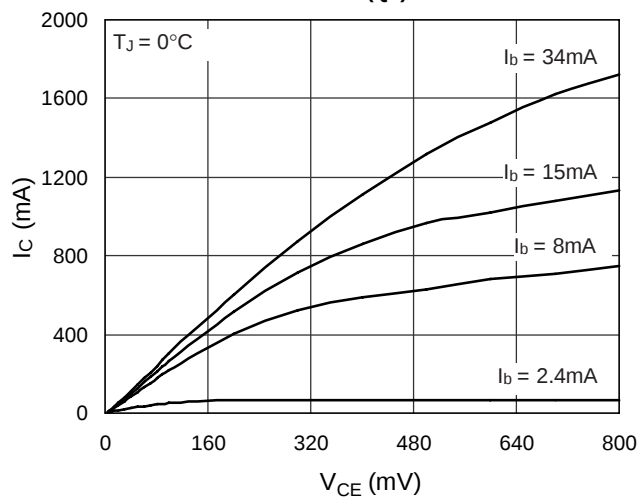
I_C vs. $V_{CE(Q1)}$ Curve



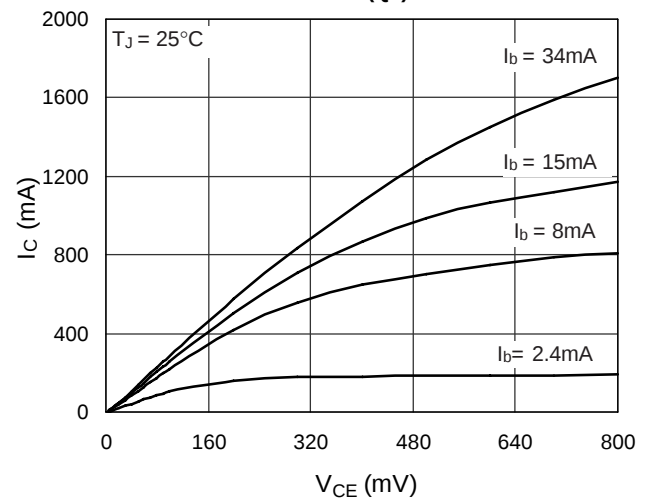
I_C vs. $V_{CE(Q1)}$ Curve



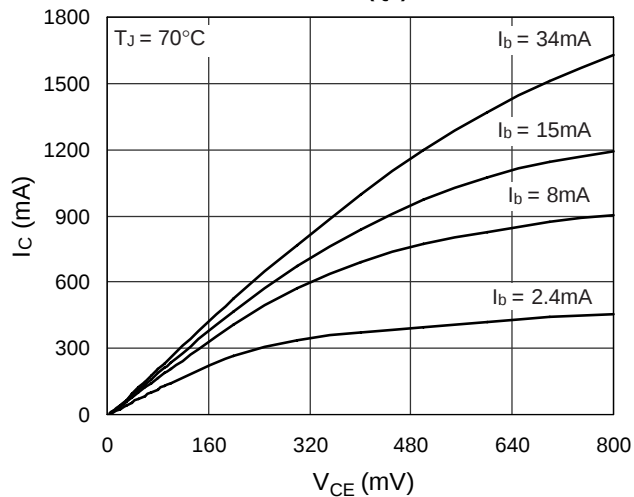
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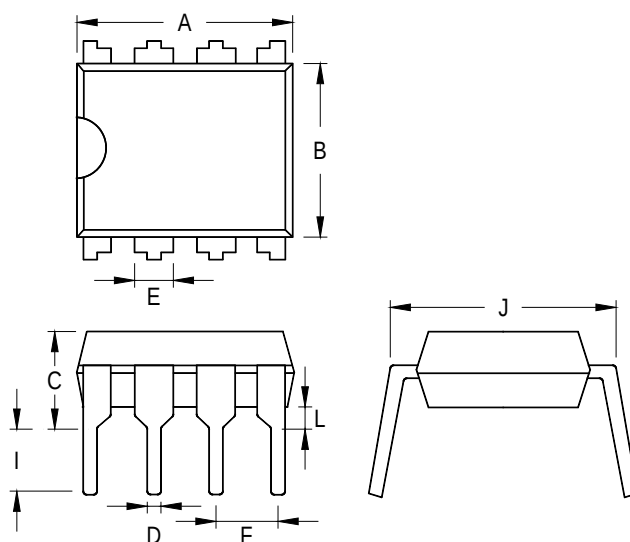
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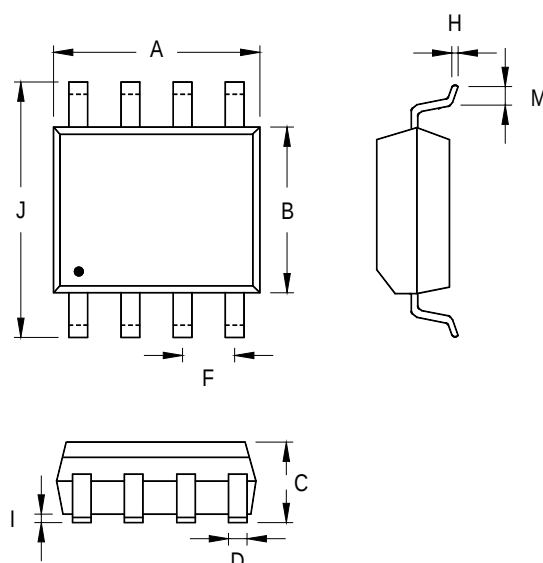


Outline Dimension



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	9.068	9.627	0.357	0.379
B	6.198	6.604	0.244	0.260
C	3.556	4.318	0.140	0.170
D	0.356	0.559	0.014	0.022
E	1.397	1.651	0.055	0.065
F	2.337	2.743	0.092	0.108
I	3.048	3.556	0.120	0.140
J	7.366	8.255	0.290	0.325
L	0.381		0.015	

8-Lead DIP Plastic Package



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	4.801	5.004	0.189	0.197
B	3.810	3.988	0.150	0.157
C	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
H	0.170	0.254	0.007	0.010
I	0.050	0.254	0.002	0.010
J	5.791	6.200	0.228	0.244
M	0.400	1.270	0.016	0.050

8-Lead SOP Plastic Package

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