

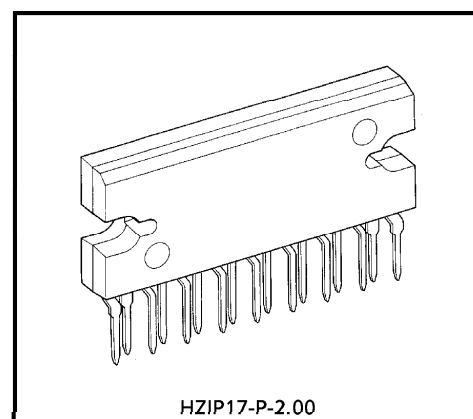
TA8233H

30W BTL×2CH AUDIO POWER AMPLIFIER

The thermal resistance θ_{j-T} of TA8233H package designed for low thermal resistance, has high efficiency of heat radiation.

The temperature rise of chip can be reduced, and the influence from the degradation of the features due to the temperature rise at the high output can also be reduced.

This stereo audio power IC, designed for car audio use, has two built-in channels to reduce the characteristic difference between L and R channels. It also contains output short detection circuit, output clip detection and various kind of protection.



HZIP17-P-2.00

Weight : 9.8g (Typ.)

FEATURES

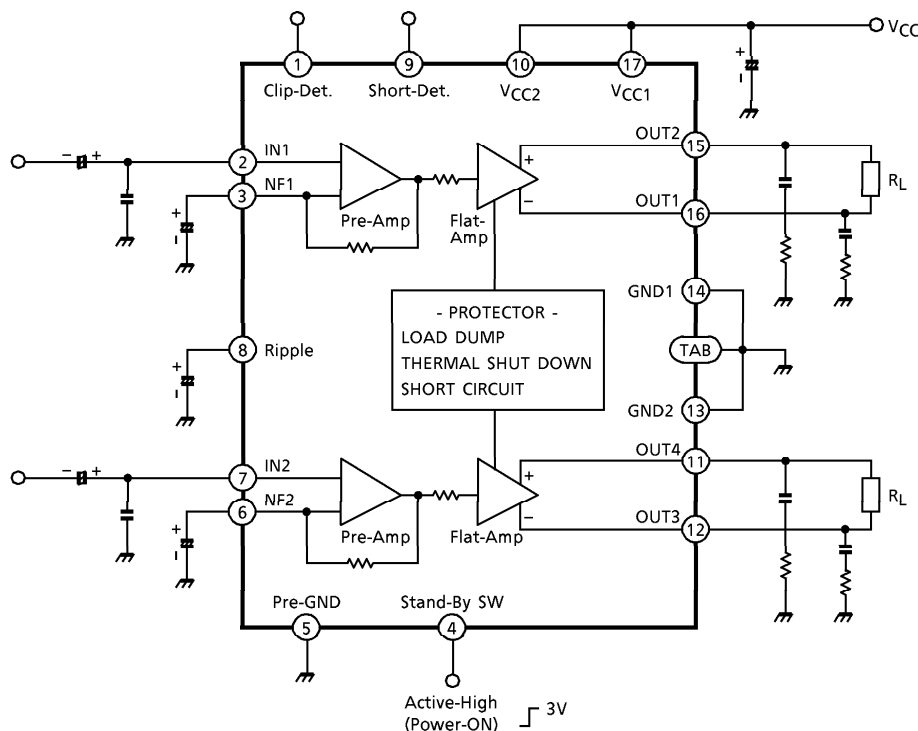
- High Power
 - : $P_{OUT}(1) = 30W$ (Typ.) / Channel
($V_{CC} = 14.4V$, $f = 1kHz$, $THD = 10\%$, $R_L = 2\Omega$)
 - $P_{OUT}(2) = 26W$ (Typ.) / Channel
($V_{CC} = 13.2V$, $f = 1kHz$, $THD = 10\%$, $R_L = 2\Omega$)
 - $P_{OUT}(3) = 19W$ (Typ.) / Channel
($V_{CC} = 13.2V$, $f = 1kHz$, $THD = 10\%$, $R_L = 4\Omega$)
- Low Thermal Resistance : $\theta_{j-T} = 1.5^\circ C / W$ (Infinite Heat Sink)
- Low Distortion Ratio : $THD = 0.04\%$ (Typ.)
($V_{CC} = 13.2V$, $f = 1kHz$, $P_{OUT} = 1W$, $R_L = 4\Omega$, $G_V = 50dB$)
- Low Noise : $V_{NO} = 0.30mV_{rms}$ (Typ.)
($V_{CC} = 13.2V$, $R_L = 4\Omega$, $G_V = 50dB$, $R_g = 0\Omega$, $BW = 20Hz \sim 20kHz$)
- Built-in Stand-by Function
 - : (With pin④ set at LOW, Power is turned OFF.) $I_{SB} = 1\mu A$ (Typ.)
- Built-in Output Clip Detection Circuit
 - : (pin① : Open Collector (Active Low))
- Built-in Output Short Detection Circuit
 - : (pin⑨ : Open Collector (Active Low))
- Built-in Various Protection Circuits
 - : Thermal Shut Down, Over Voltage, Out $\rightarrow V_{CC}$ Short, Out $\rightarrow GND$ Short and OUT-OUT Short.
- Operating Supply Voltage : $V_{CC(opr)} = 9 \sim 18V$

980508EBA2

● TOSHIBA is continually working to improve the quality and the reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to observe standards of safety, and to avoid situations in which a malfunction or failure of a TOSHIBA product could cause loss of human life, bodily injury or damage to property. In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent products specifications. Also, please keep in mind the precautions and conditions set forth in the TOSHIBA Semiconductor Reliability Handbook.

BLOCK DIAGRAM

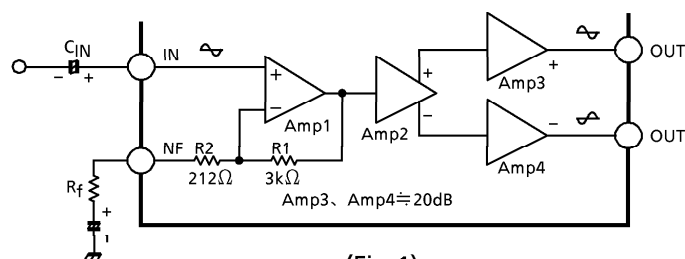
TA8233H ($G_V = 50\text{dB}$)



CAUTION AND APPLICATION METHOD

(Description is made only on the single channel.)

1. Voltage Gain Adjustment



(Fig.1)

Amp1 : Pre-Amp
Amp2 : Phase Amp
Amp3 : POWER Amp (Flat-Amp)
Amp4 : POWER Amp (Flat-Amp)

This IC has the amplifier constructions as shown in Fig. 1. The Pre-Amp (Amp 1) is provided to the primary stage, and the input voltage is amplified by the Flat Amps, Amp 3 and Amp 4 of each channel through the phase Amp (Amp 2).

Since the input offset is prevented by Pre-Amp when V_{CC} is set to ON, this circuit can remarkably reduce the pop noise.

980508EBA2'

- The products described in this document are subject to foreign exchange and foreign trade laws.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA CORPORATION for any infringements of intellectual property or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any intellectual property or other rights of TOSHIBA CORPORATION or others.
- The information contained herein is subject to change without notice.

The total closed loop gain G_V of this IC can be obtained by expression below when the closed loop voltage gain of Amp 1 is G_{V1} .

$$G_{V1} = 20 \log \frac{R1 + (R_f + R2)}{R_f + R2} \text{ (dB)} \dots\dots\dots (1)$$

The closed loop voltage gain of POWER Amp, Amp 3 and Amp 4 is fixed at $G_{V3} \doteq G_{V4} = 20\text{dB}$.

Therefore, the total closed circuit voltage gain G_V is obtained through BTL connection by the expression as below.

$$G_V = G_{V1} + G_{V3} + 6 \text{ (dB)} \dots\dots\dots (2)$$

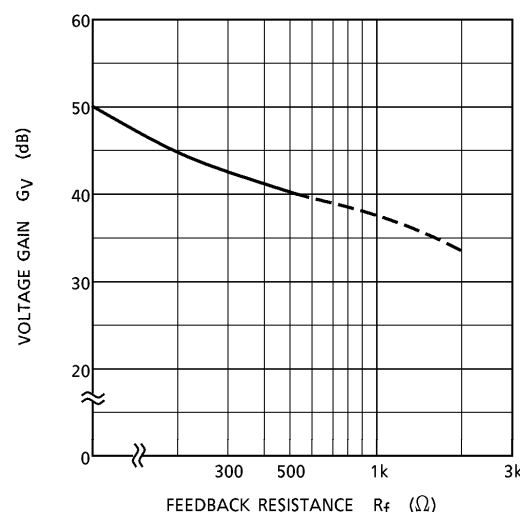
For example, when $R_f = 0\Omega$, G_V is obtained by the expressions (1) and (2) as below.

$$G_V \doteq 24 + 20 + 6 = 50\text{dB}$$

The voltage gain is reduced when R_f is increased.

(Fig.2)

With the voltage gain reduced, since (1) the oscillation stability is reduced, and (2) the pop noise changes when V_{CC} is set to ON, refer to the items 3 and 4.



(Fig.2)

2. Stand-by SW Function

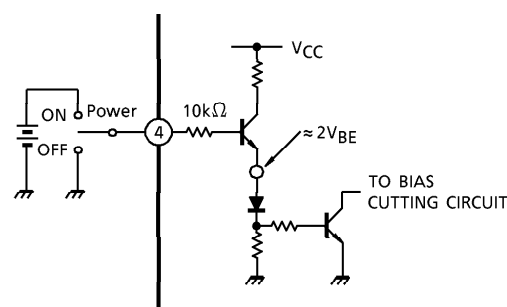
By means of controlling pin④ (Stand-by terminal) to High and Low, the power supply can be set to ON and OFF. The threshold voltage of pin④ is set at 2.1V ($3V_{BE}$), and the Power Supply current is about $1\mu\text{A}$ (Typ.) at the stand-by state.

Control Voltage of pin④ : $V_{(SB)}$

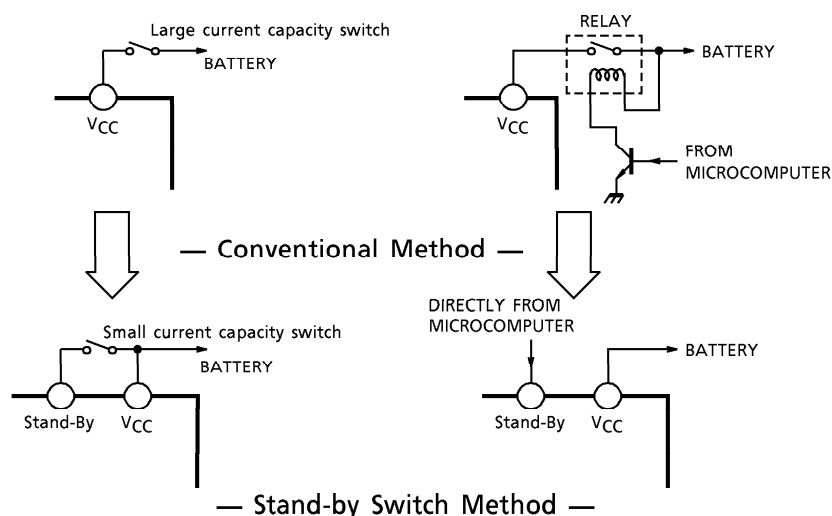
Stand-By	Power	$V_{(SB)}$ (V)
ON	OFF	0~2
OFF	ON	3~ V_{CC}

Adjustage of Stand-by SW

- (1) Since V_{CC} can directly be controlled to ON or OFF by the microcomputer, the switching relay can be omitted.
- (2) Since the control current is microscopic, the switching relay of small current capacity is satisfactory for switching



(Fig.3) With pin④ set to High, Power is turned ON



3. Preventive Measure Against Oscillation

For preventing the oscillation, it is advisable to use C₄, the condenser of polyester film having small characteristic fluctuation of the temperature and the frequency.

The condenser (C₆) between input and GND is effective for preventing oscillation which is generated with a feedback signal from a output stage.

The resistance R to be series applied to C₄ is effective for phase correction of high frequency, and improves the oscillation allowance.

- (1) Voltage gain to be used (G_V Setting)
- (2) Capacity value of condenser
- (3) Kind of condenser
- (4) Layout of printed board

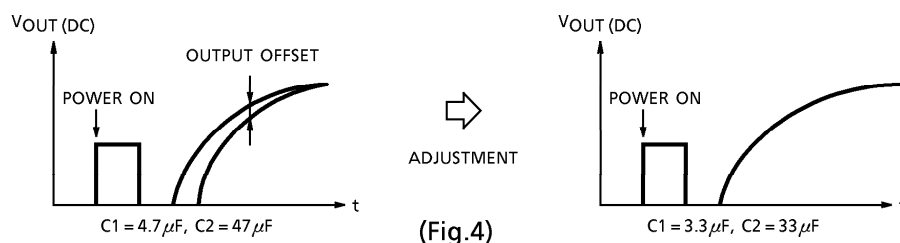
In case of its use with the voltage gain G_V reduced or with the feedback amount increased, care must be taken because the phase-inversion is caused by the high frequency resulting in making the oscillation liable generated.

4. Adjustment of Output Offset (When the power supply turn on)

As this IC is constructed with DC circuit on the Pre-Amp stage, it is necessary to lower a input offset or output offset by agreement with the each leading edge time constant of the input voltage in the Pre-Amp stage and NF terminal voltage.

Concretely, monitor the output DC voltage and vary the capacity value in input condenser and NF condenser (See Fig.4)

(Reference) In case of setting the condition ($G_V = 40\text{dB}$) with $R_f = 470\Omega$



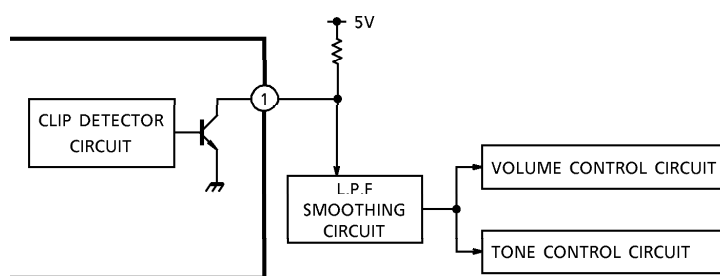
(Fig.4)

5. Output Clip Detection Function (pin①)

The output clip detection terminal of pin① has the open collector output structure on chip as shown in Fig.5. In case that the output waveform is clipping, the clip detection circuit is operated and NPN Tr. is turned on.

It is possible to improve the tone quality with the current of flowing into pin① and with controlling the volume, tone control circuit through L.P.F smoothing circuit as shown in Fig.5. In case of being unused this function, use this IC as open connection on pin①.

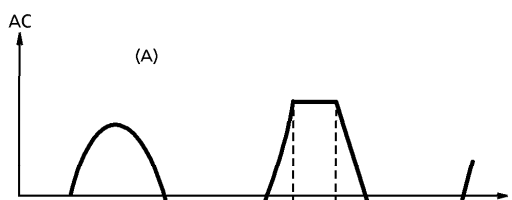
(Application)



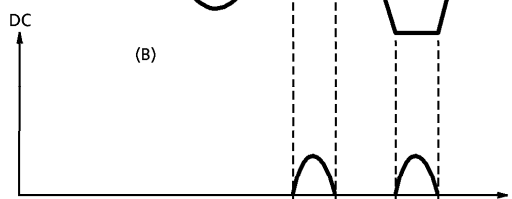
pin① : Open Collector Output (Active Low)

(Fig.5)

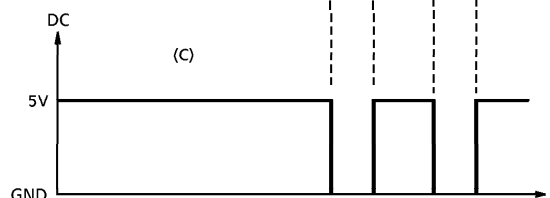
(A) OUTPUT (AC WAVE FORM)



(B) CLIP DETECTOR CIRCUIT (INTERNAL)



(C) CLIP DET. TERMINAL (pin①)

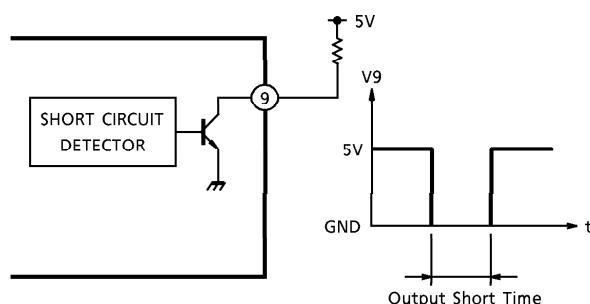


6. Output to V_{CC} , Output to GND Short Detection Function (pin⑨)

The Output Short Detection Terminal of pin⑨ has open collector output structure on chip as shown in Fig.6.

In unusual case that output terminal of Power Amp. is condition of output to V_{CC} or output to GND short, it is possible to protect all the system of apparatus as well as power IC protection.

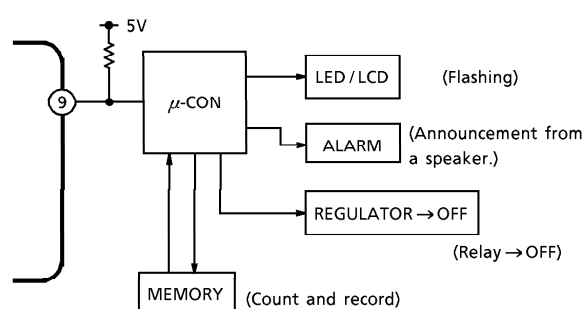
In case of being unused this function, use this IC as open-connection on pin⑨.



pin⑨ : Open Collector Output (Active Low)

(Fig.6)

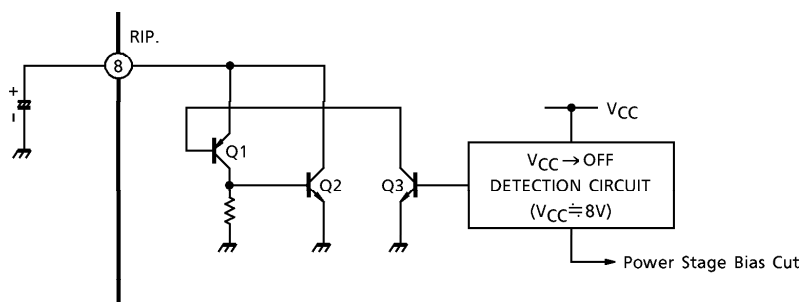
(Application)



(Fig.7)

7. Rapid Ripple Discharging Circuit at the Time of V_{CC} OFF

This circuit is effective in such a mode where the V_{CC} and the stand-by terminals become high or Low simultaneously ; for instance, for a pop noise produced when the power is turned ON or OFF repeatedly by operating the ignition key.



(Fig.8)

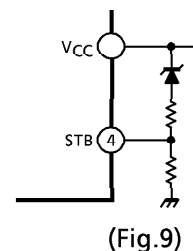
When V_{CC} is OFF, this detection circuit detect $V_{CC} \doteq 8V$ internally on IC.

(1) The power stage bias circuit is cut.

(2) pin⑧ : Ripple Capacitor is rapidly discharged by turning Q3 ON and then Q1 and Q2 ON.

(Precaution 1)

When the stand-by terminal was put to the low level after the ripple rapid discharging circuit was operated ($V_{CC} \approx 8V$) at the time when V_{CC} was turned OFF, a pop noise may be generated. Therefore, V_{CC} which makes the Stand-by terminal low shall be set at 8V or above so that (1) the Stand-by terminal is put at the low level and (2) the ripple rapid discharging circuit is turned ON when V_{CC} is turned OFF (in order of (1) and (2)). An example of application is shown in Fig.9.



(Precaution 2)

If the falling time constant of the V_{CC} line is large (the fall is gentle), the pop noise may become worse.

In this case, it is possible to prevent the pop noise from becoming worse by reducing the ability of "Ripple Rapid Discharging Circuit at the time of V_{CC} OFF" according to the increase of the capacity of ripple capacitor of pin⑧.

However, it shall be kept in mind that the time for turning the power ON becomes longer as the result of this method.

8. External Part List and Description

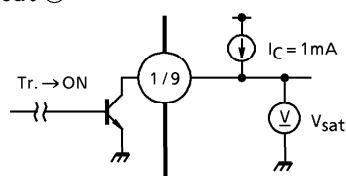
SYM-BOL	RECOMMENDED VALUE	FEATURE	INFLUENCE		REMARKS
			SMALLER THAN RECOMMENDED VALUE	LARGER THAN RECOMMENDED VALUE	
C1	4.7 μ F	DC blocking	Related to pop noise at $V_{CC} \rightarrow$ ON.		Related to gain.
C2	47 μ F	Feedback condenser	Related to pop noise at $V_{CC} \rightarrow$ ON.		Refer to item 4.
			Determination of low cut-off frequency $C2 = \frac{1}{2\pi \cdot f_L \cdot R_f}$		
C3	220 μ F	Ripple reduction	Time constant is small at $V_{CC} \rightarrow$ ON or OFF.	Time constant is Large at $V_{CC} \rightarrow$ ON or OFF.	
C4	0.12 μ F	Oscillation prevention	Made liable to oscillate.	Oscillation allowance	Refer to item 3.
C5	1000 μ F	Ripple filter	For filtering power supply hum and ripple. Large at using AC rectified power supply. Small at using DC power supply.		
C6	1000pF	Oscillation prevention	Oscillation allowance improved. Noise Reduction		Refer to item 3.

MAXIMUM RATINGS (Ta = 25°C)

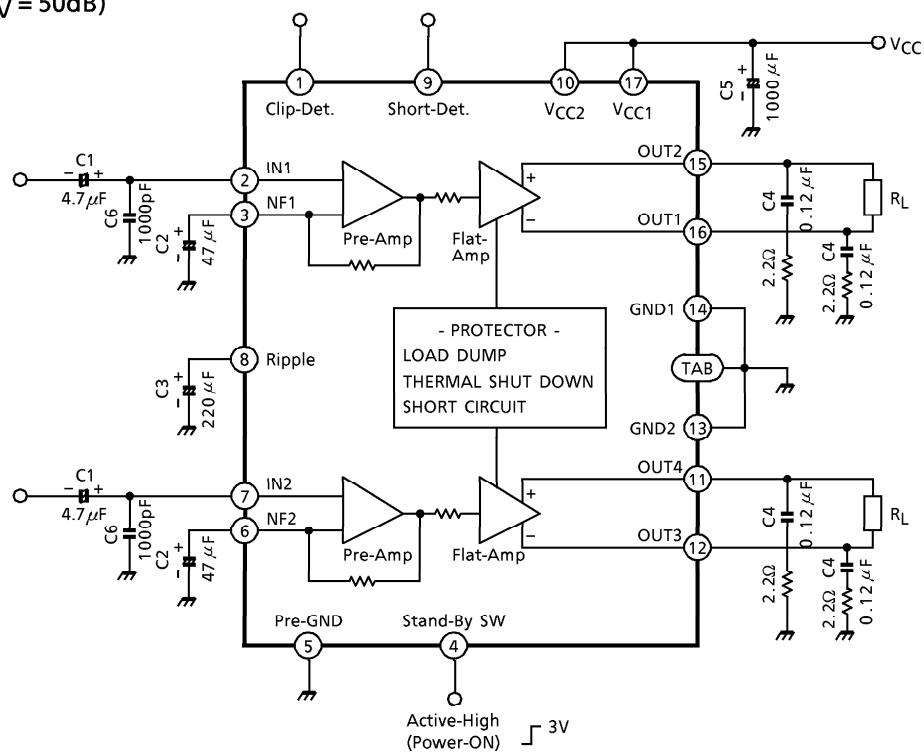
CHARACTERISTIC	SYMBOL	RATING	UNIT
Peak Supply Voltage (0.2s)	V_{CC} (surge)	50	V
DC Supply Voltage	V_{CC} (DC)	25	V
Operating Supply Voltage	V_{CC} (opr)	18	V
Output Current (Peak)	I_O (peak)	9	A
Power Dissipation	P_D	50	W
Operating Temperature	T_{opr}	-30~85	°C
Storage Temperature	T_{stg}	-55~150	°C

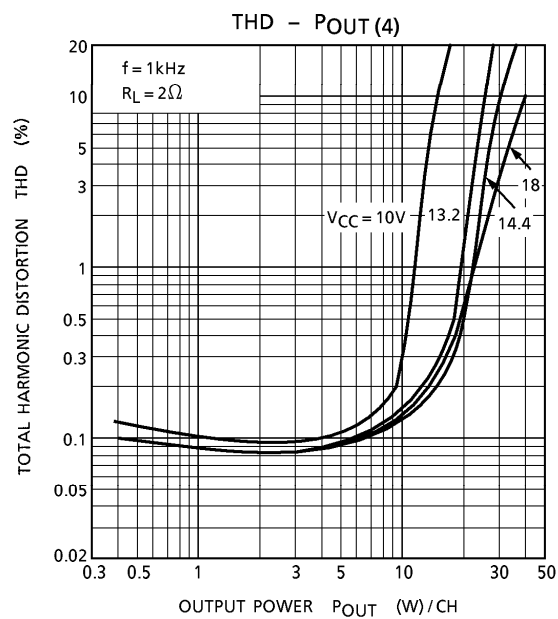
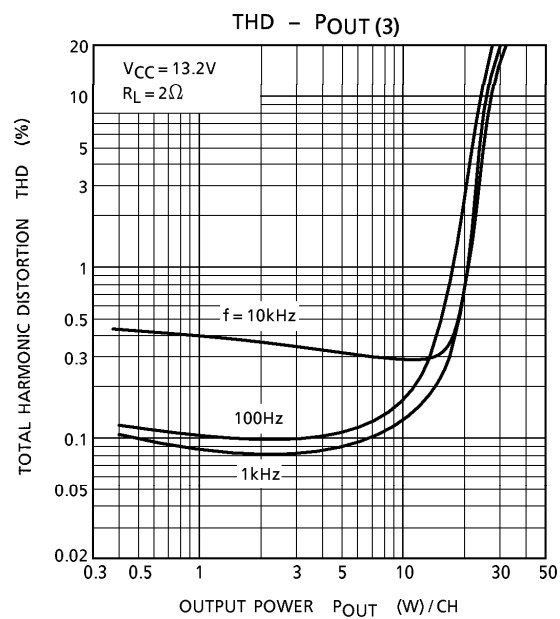
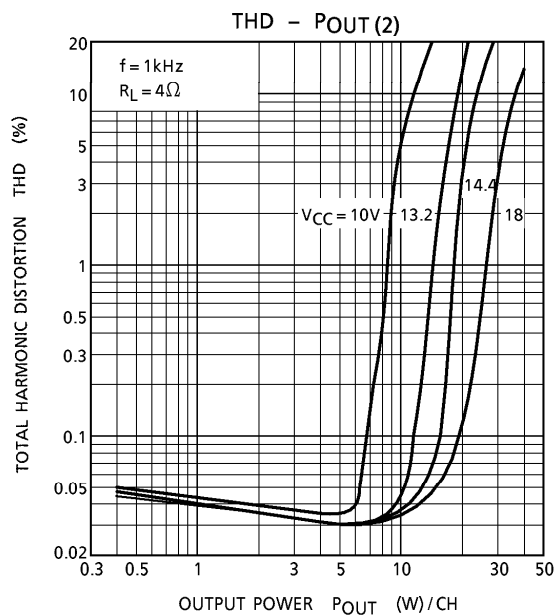
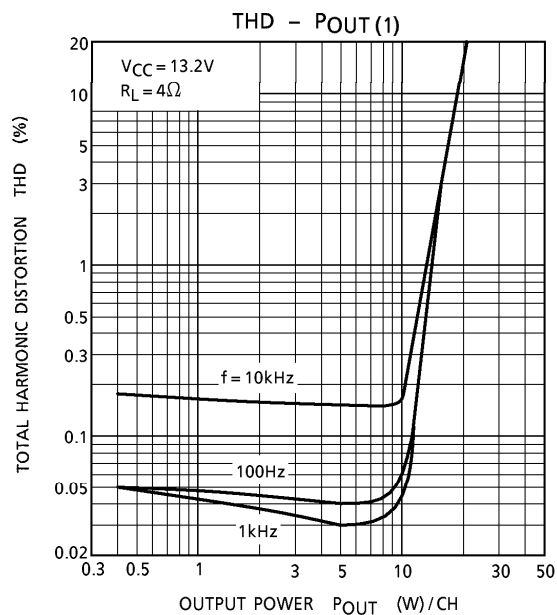
ELECTRICAL CHARACTERISTICS (Unless otherwise specified, $V_{CC} = 13.2V$, $R_L = 4\Omega$, $f = 1kHz$, $T_a = 25^\circ C$)

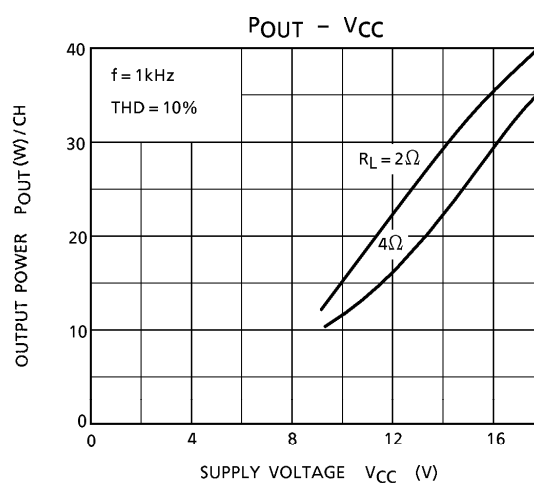
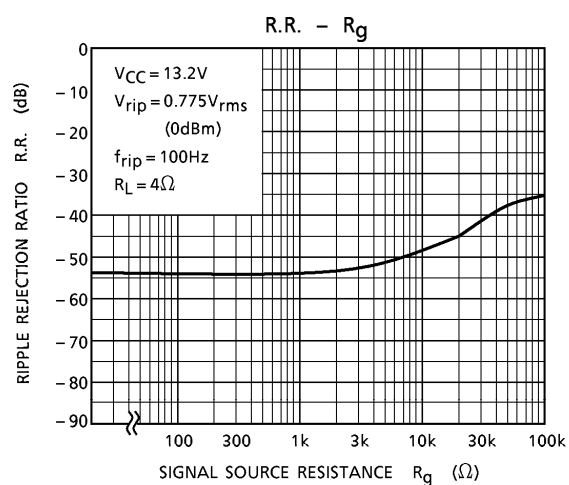
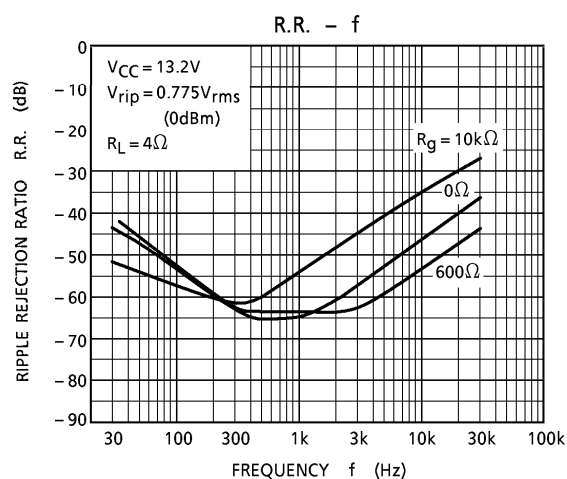
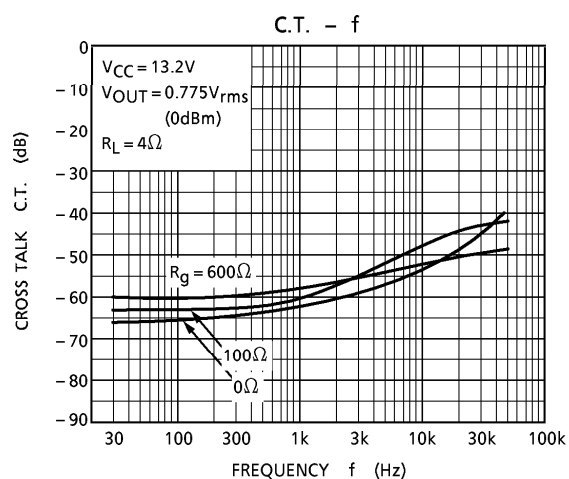
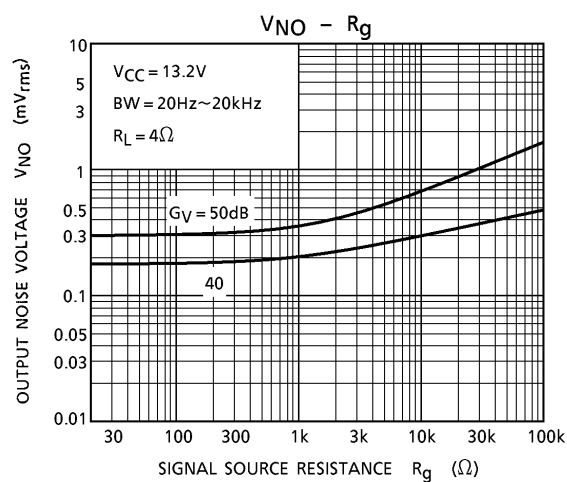
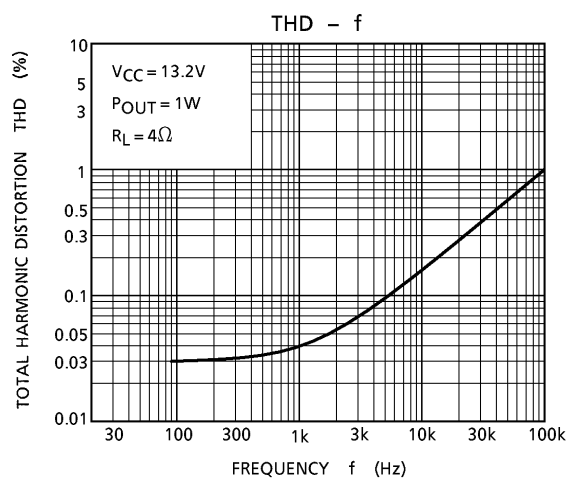
CHARACTERISTIC	SYMBOL	TEST CIR-CUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Quiescent Supply Current	I_{CCO}	—	$V_{IN} = 0$	—	120	250	mA
Output Power	P_{OUT} (1)	—	$V_{CC} = 14.4V$, $R_L = 2\Omega$ THD = 10%	—	30	—	W
	P_{OUT} (2)	—	$R_L = 2\Omega$, THD = 10%	17	26	—	
	P_{OUT} (3)	—	THD = 10%	16	19	—	
Total Harmonic Distortion Ratio	THD	—	$P_{OUT} = 1W$	—	0.04	0.4	%
Voltage Gain	G_V	—		48	50	52	dB
Voltage Gain Ratio	ΔG_V	—		-1.0	0	1.0	dB
Output Noise Voltage	V_{NO}	—	$R_g = 0\Omega$, BW = 20Hz~20kHz	—	0.3	0.7	mV _{rms}
Ripple Rejection Ratio	R.R.	—	$f_{ripple} = 100Hz$, $R_g = 600\Omega$	40	54	—	dB
Input Resistance	R_{IN}	—		—	30	—	k Ω
Output Offset Voltage	V_{offset}	—	$V_{IN} = 0$	-100	0	100	mV
Current at Stand-by State	I_{SB}	—		—	1	10	μA
Cross Talk	C.T.	—	$R_g = 600\Omega$ $V_{OUT} = 0.775V_{rms}$ (0dBm)	—	60	—	dB
Pin④ Control Voltage	V_{SB}	—	Stand-By → OFF (Power → ON)	2.5	—	V_{CC}	V
Pin① (CLIP DET) Saturation Voltage	V_{sat} ①	—	$I_C = 1mA$	—	100	—	mV
Pin⑨ (SHORT DET) Saturation Voltage	V_{sat} ⑨	—	$I_C = 1mA$	—	100	—	mV

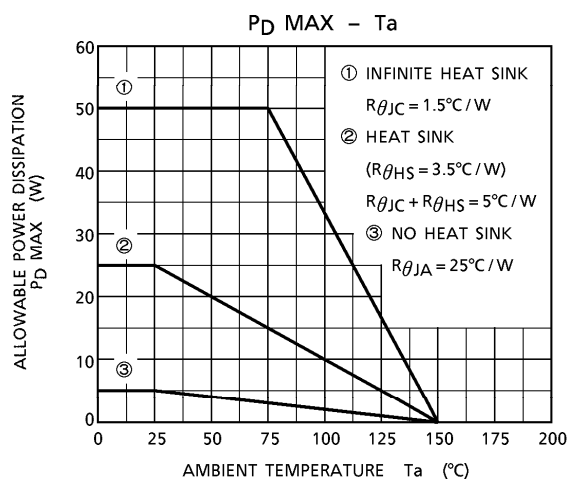
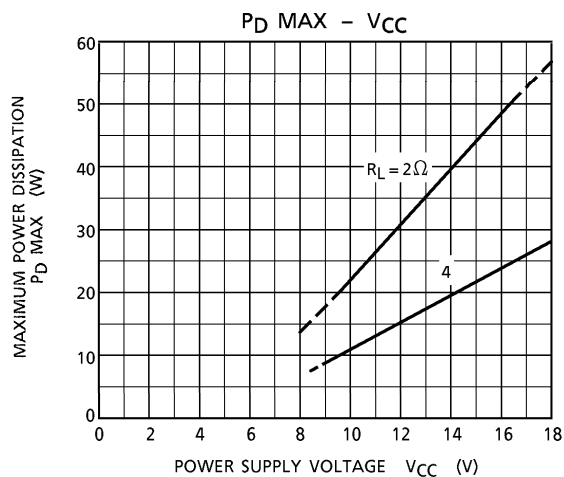
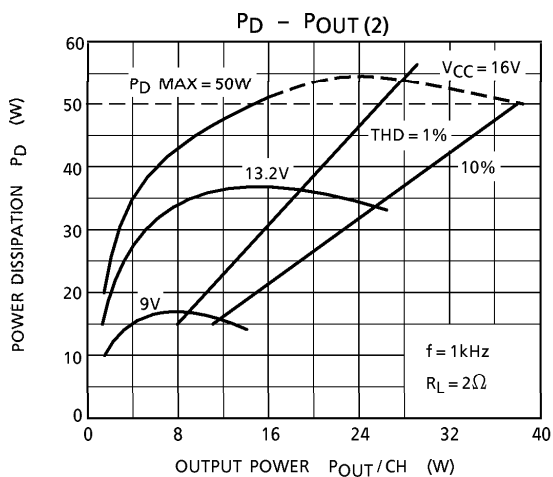
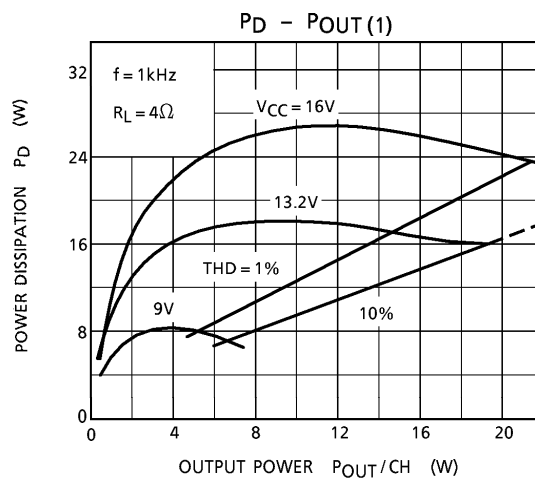
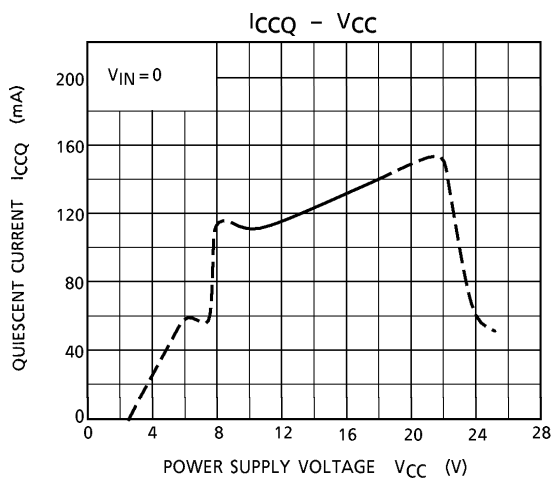
 V_{sat} ①, V_{sat} ⑨ Test Circuit

TEST CIRCUIT

TA8233H ($G_V = 50\text{dB}$)

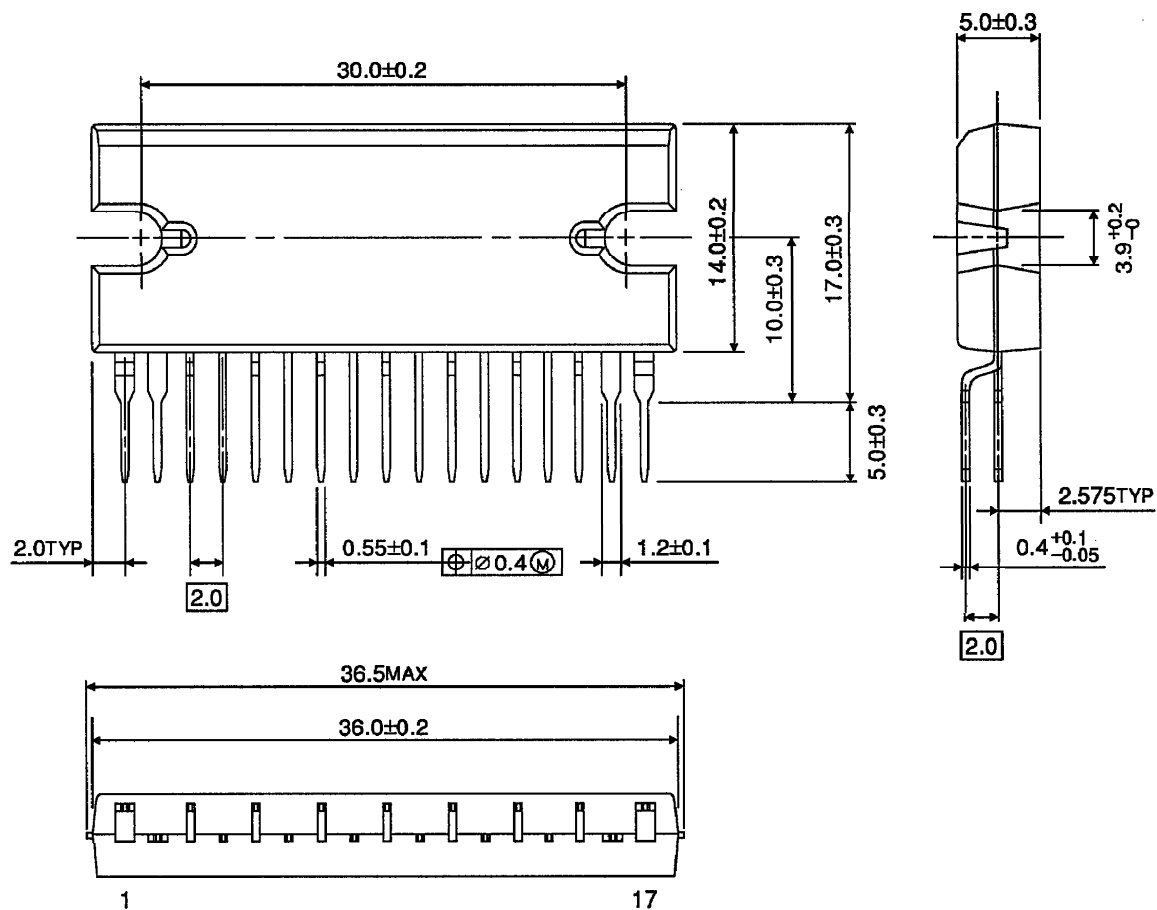






OUTLINE DRAWING
HZIP17-P-2.00

Unit : mm



Weight : 9.8g (Typ.)