



LB1939T

2 Channel H Bridge Constant Voltage/ Constant Current Driver IC

Overview

The LB1939T is a two-phase excitation bipolar stepping motor driver that features low voltage operation, a low saturation voltage, and low power consumption. It supports constant voltage and constant current drive, can control two iris motors, and is optimal for shutter, iris, and AF drive in 3 V battery operated still digital cameras and other battery operated equipment.

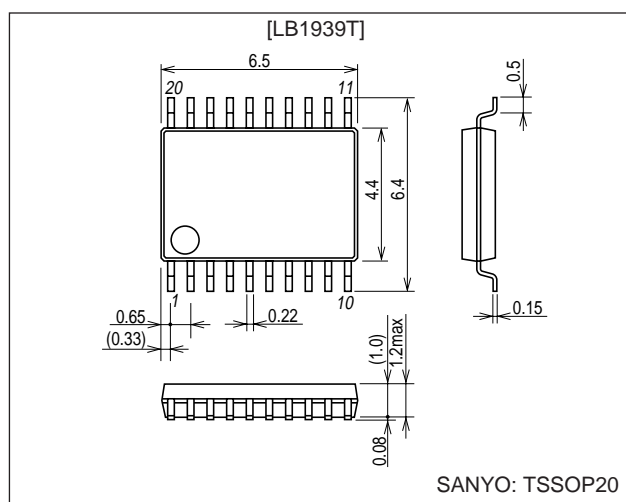
Features

- Low-voltage drive
 - Dual power supply operation: $V_S = 1.6$ to 7.5 V, $V_{DD} = 1.9$ to 6.5 V
 - Single power supply operation: $V_S = V_{DD} = 1.9$ to 7.5 V
- Low saturation voltage output: $V_{sat} = 0.3$ V at $I_O = 200$ mA
- Supports constant voltage and constant current drive
- Built-in reference voltage circuit ($V_{ref} = 0.9$ V)
- Miniature, thin form package (Thickness $t = 1.1$ mm)

Package Dimensions

unit: mm

3246-TSSOP20



Specifications

Absolute Maximum Ratings at $T_a = 25^\circ\text{C}$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V_{Bmax}	$VS1, VS2, V_{DD}$	-0.3 to $+10.5$	V
Applied output voltage	V_{OUT}	OUT1, 2, 3, 4	-0.3 to $+10.5$	V
Maximum output current: OUT1, 2, 3, and 4	I_{Omax}	$t \leq 10$ ms	400	mA
Applied input voltage	V_{IN}	ENA, IN, VC	10.5	V
Allowable power dissipation	P_{dmax}	When mounted on a printed circuit board*	0.8	W
Operating temperature	T_{opr}		-20 to $+85$	$^\circ\text{C}$
Storage temperature	T_{stg}		-55 to $+150$	$^\circ\text{C}$

Note: Circuit board: $114.3 \times 76.1 \times 1.6$ mm³ glass epoxy board

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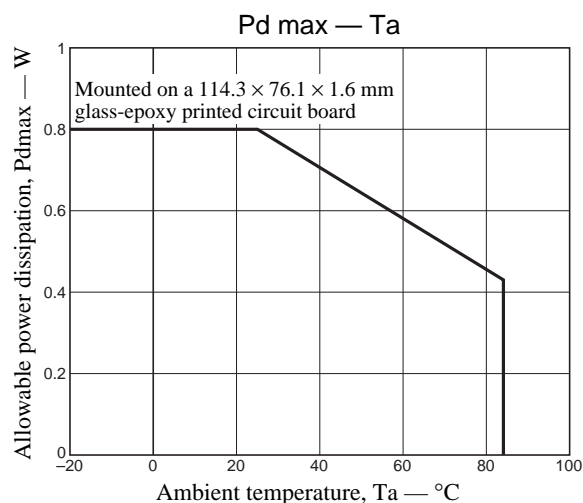
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Allowable Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Operation guaranteed voltage range 1	V _{OPR1}	V _{DD} system, V _S = 2.0 V	1.9		6.5	V
Operation guaranteed voltage range 2	V _{OPR2}	V _S system, V _{DD} = 5.0 V	1.6		7.5	V
Input low-level threshold voltage	V _{IL}	ENA1, ENA2, IN1, IN2	-0.3		+1.0	V
Input high-level threshold voltage	V _{IH}	ENA1, ENA2, IN1, IN2	2.0		6.0	V

Electrical Characteristics at Ta = 25°C, V_S = 3.0 V, V_{DD} = 5.0 V

Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Standby mode current drain	I _{STB}	V _S = V _{DD} = 6.5 V		0.1	1.0	μA
[Regulator Output Circuit]						
V _{REF} output voltage	V _{REF}	I _{OL} = 0 to 1 mA	0.85	0.9	0.95	V
SV _{DD} output voltage	V _{SVDD}	I _{OL} = 10 mA	4.7	4.8		V
[H Bridge Output Circuit]						
OUT pin output saturation voltage 1 (Saturation control mode)	V _{osat1}	V _{DD} = 5.0 V, V _C = V _{SVDD} , V _S = 2.0 V I _O = 200 mA (PNP transistor side)		0.20	0.30	V
OUT pin output saturation voltage 2 (Saturation control mode)	V _{osat2}	V _{DD} = 5.0 V, V _C = V _{SVDD} , V _S = 2.0 V I _O = 200 mA (NPN transistor side)		0.10	0.15	V
OUT pin output voltage 1 (Constant voltage control mode)	V _{OUT1}	V _{DD} = 6.0 V, V _C = 1.5 V, V _S = 3.5 V I _O = 200 mA (PNP transistor side)	2.8	2.9	3.0	V
OUT pin output voltage 2 (Constant voltage control mode)	V _{OUT2}	V _{DD} = 6.0 V, V _C = V _{REF} , V _S = 2.0 V I _O = 200 mA (PNP transistor side)	1.65	1.75	1.85	V
OUT pin output current 1 (Constant current control mode)	I _{OUT1}	V _{DD} = 6.0 V, V _C = 0.9 V, V _S = 3.5 V R _L = 5 Ω (between OUT and OUT), R _{FB} = 1 Ω	197	210	223	mA
OUT pin output current 2 (Constant current control mode)	I _{OUT2}	V _{DD} = 6.0 V, V _C = V _{REF} , V _S = 2.0 V R _L = 5 Ω (between OUT and OUT), R _{FB} = 1 Ω	189	210	231	mA
V _S system operating current drain 1	I _{S1}	V _C = V _{SVDD}		4	7	mA
V _S system operating current drain 2	I _{S2}	V _C = V _{REF}		1.5	3	mA
V _{DD} system operating current drain 1	I _{DD1}	V _C = V _{SVDD} ENA1 = 2 V		4	7	mA
V _{DD} system operating current drain 2	I _{DD2}	V _C = V _{REF} ENA1 = 2 V		4	7	mA
V _C input voltage range	V _C		0.1		7	V
V _C input current	I _{VC}	V _{DD} = 6.0 V, V _S = 2.0 V, V _C = 5.0 V	0	50	100	μA
[Control Input Circuit]						
Control pin maximum input current	I _{IH}	V _{IH} = 5.5 V		70	100	μA
	I _{IL}	V _{IL} = GND	-1		0	μA

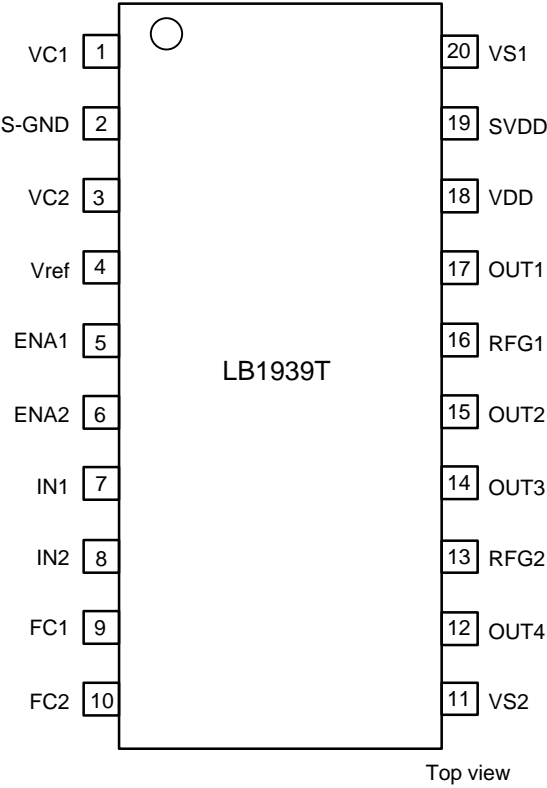


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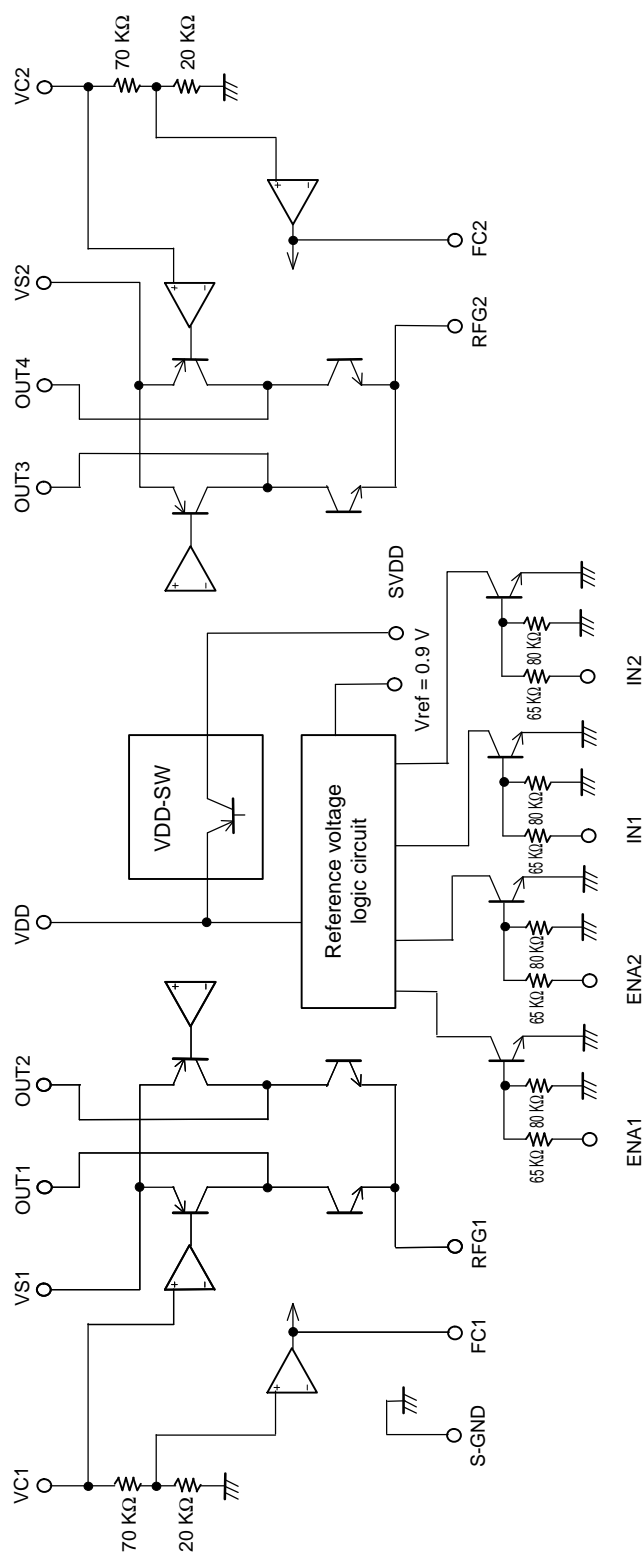
Truth Table

Input				Output					Mode
ENA		IN		OUT				SVDD	
1	2	1	2	1	2	3	4		
L	L								Standby mode (zero current drain)
H		H		L	H			on	Channel 1: reverse
		L		H	L			on	Channel 1: forward
	H		H			L	H	on	Channel 2: reverse
			L			H	L	on	Channel 2: forward
Blank entries indicate “don’t care” states.				Blank entries indicate off states.					

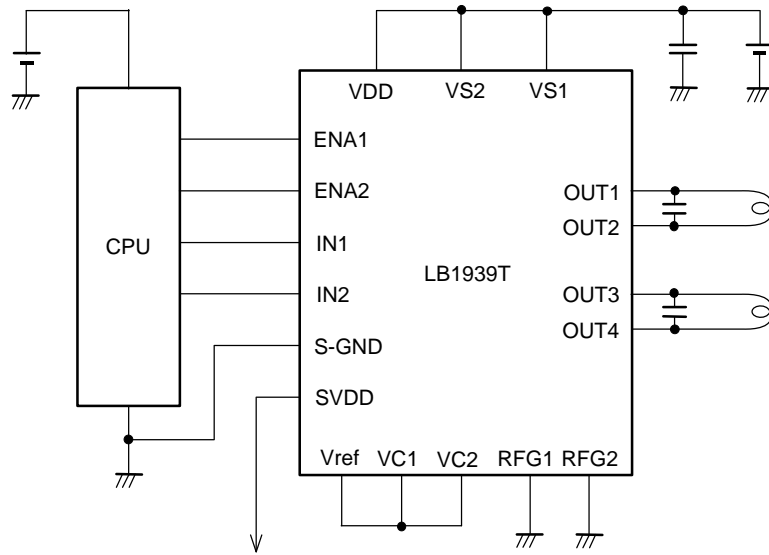
Pin Assignment



Block Diagram

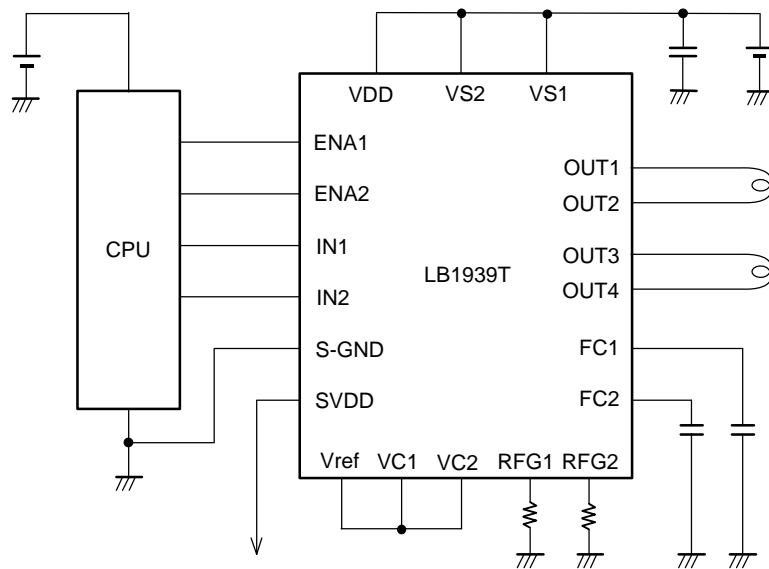


Application Circuit Example 1



Constant voltage control mode: OUT outputs a 1.75 V, which is $V_{ref} (0.9 \text{ V}) \times 1.95$.
 *: FC1 and FC2 are left open.

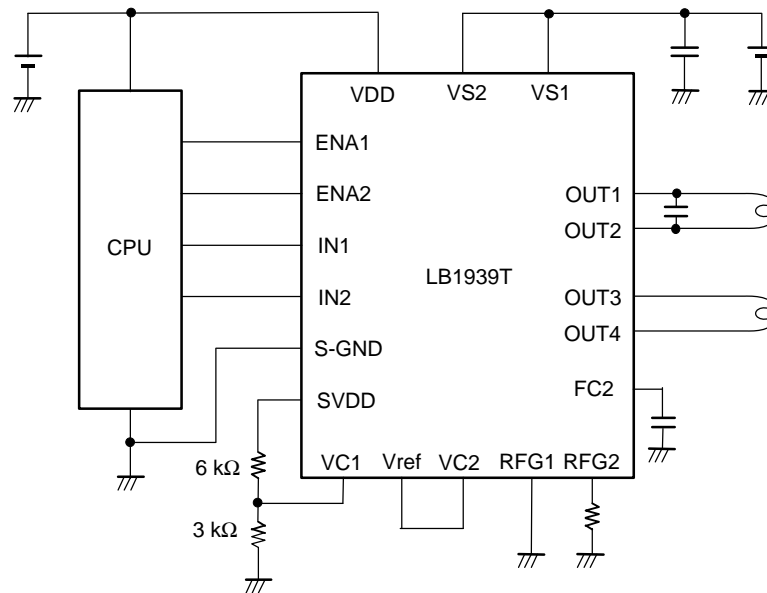
Application Circuit Example 2



Constant current control mode: The RFG voltage is controlled so that $V_{ref}/4.5 = 0.2 \text{ V}$.
 Therefore, when R_{fB} is 1Ω , the circuit operates in constant current drive with $I_{coil} = 0.2 \text{ V}/1 \Omega = 200 \text{ mA}$.

*: There are no magnitude constraints on the inputs (ENA, IN) and the supply voltages (V_{DD} , V_S).
 For example, the IC can be operated at $V_{IN} = 5 \text{ V}$, $V_{DD} = 3 \text{ V}$, and $V_S = 2 \text{ V}$.

Application Circuit Example 3



Channel 1 operates in constant voltage control mode: OUT outputs $V_{DD} \times 3K/(3K + 6K) \times 1.95$
 Channel 2 operates in constant current control mode: The RFG voltage is controlled so that $V_{ref}/4.5 = 0.2$ V.
 * : FC1 is left open.

Notes on Constant Current Control Settings

The LB1939T constant current control circuit has the structure shown in the figure at the right. The voltage input to the VC pin is resistor divided internally (by 70 kΩ and 20 kΩ resistors) to 1/4.5 and input to the plus (+) input of the constant current control amplifier as reference.

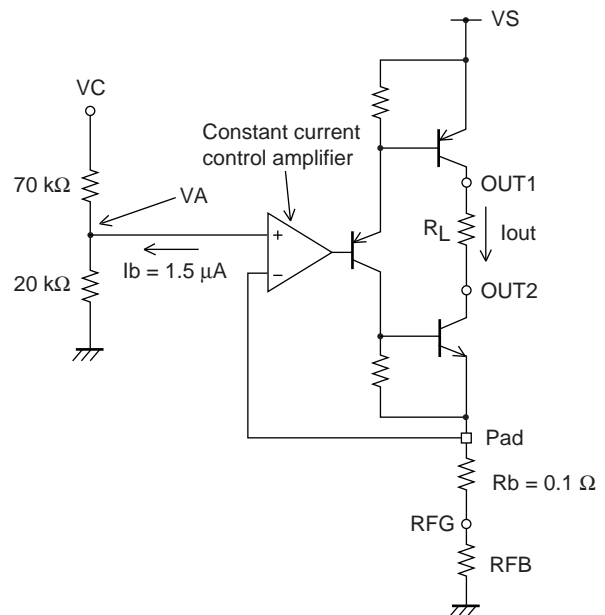
The minus (-) input of this constant current control amplifier is connected, through the wire bond resistor Rb (= 0.1 Ω), to the RFG pin. The constant current control circuit operates by comparing the voltage generated by the external current detection resistor connected to the RFG pin and the reference voltage mentioned above.

Note that the voltage at VA will be that given by the following formula since the bias current Ib (= 1.5 μA) flows from the constant current control amplifier plus (+) input during constant current control operation.

$$\begin{aligned} VA &= VC/4.5 + Ib \times 20 \text{ k}\Omega \\ &= VC/4.5 + 0.03 \end{aligned}$$

Therefore, the logical expression for setting the constant current Iout is as follows.

$$\begin{aligned} I_{out} &= VA/(RFB + Rb) \\ &= (VC/4.5 + 0.03) / (RFB + Rb) \dots\dots(1) \end{aligned}$$



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Constant Current Control Usage Notes

This IC supports both constant current control and constant voltage control modes. However, since both of these control circuits operate at all times, certain of the limitations imposed by the constant voltage control circuit apply may when using constant current control.

For example, if constant current control is used with the application circuit shown in figure 2, if $V_C = 0.9 \text{ V}$ ($= V_{\text{ref}}$) and $R_{\text{FB}} = 1 \text{ } \Omega$, then the output current can be calculated as follows from (1) on the previous page.

$$\begin{aligned} I_{\text{out}} &= (0.9/4.5 + 0.03) / (1 + 0.1) \\ &= 0.23/1.1 \\ &\approx 0.209\text{A} \end{aligned}$$

Here, if the value driven load resistance R_L is r , since the RFG pin voltage is 0.23 V and the npn transistor output saturation voltage is 0.1 V (typical), the pnp transistor output pin voltage can be calculated as follows.

$$\begin{aligned} V_{\text{out}} &= (\text{RFG pin voltage}) + (\text{npn transistor output saturation voltage}) + (\text{voltage across the load terminals}) \\ &= 0.23 + 0.1 + 0.209 \times r \\ &= 0.3 + 0.209r \end{aligned}$$

At the same time, however, this IC's internal constant voltage control circuit controls the output voltage as follows.

$$V_{\text{out}}' = V_C \times 1.95 \approx 1.75 \text{ V}$$

Therefore, it will not be possible to use the constant current control mode if the value of r is set so that V_{out} is greater than V_{out}' . That is, the condition

$$0.33 + 0.209r > 1.75$$

implies that

$$r > 6.79$$

This means that constant current control can be used when the value of the load resistance used is strictly less than $6.79 \text{ } \Omega$.

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