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National Semiconductor

November 1994

LMC6081

Precision CMOS Single Operational Amplifier

General Description

The LMC6081 is a precision low offset voltage operational amplifier, capable of single supply operation. Performance characteristics include ultra low input bias current, high voltage gain, rail-to-rail output swing, and an input common mode voltage range that includes ground. These features, plus its low offset voltage, make the LMC6081 ideally suited for precision circuit applications.

Other applications using the LMC6081 include precision full-wave rectifiers, integrators, references, and sample-and-hold circuits.

This device is built with National's advanced Double-Poly Silicon-Gate CMOS process.

For designs with more critical power demands, see the LMC6061 precision micropower operational amplifier.

For a dual or quad operational amplifier with similar features, see the LMC6082 or LMC6084 respectively.

PATENT PENDING

Features

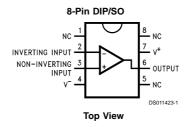
(Typical unless otherwise stated)

- Low offset voltage: 150 µV
- Operates from 4.5V to 15V single supply
- Ultra low input bias current: 10 fA
- Output swing to within 20 mV of supply rail, 100k load
- Input common-mode range includes V-
- High voltage gain: 130 dB
- Improved latchup immunity

Applications

- Instrumentation amplifier
- Photodiode and infrared detector preamplifier
- Transducer amplifiers
- Medical instrumentation
- D/A converter
- Charge amplifier for piezoelectric transducers

Connection Diagram



Ordering Information

Package	Temperatu	ıre Range	NSC	Transport
	Military	ary Industrial		Media
	-55°C to +125°C	-40°C to +85°C		
8-Pin	LMC6081AMN	LMC6081AIN	N08E	Rail
Molded DIP		LMC6081IN		
8-Pin		LMC6081AIM	M08A	Rail
Small Outline		LMC6081IM		Tape and Reel

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DS011423

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Differential Input Voltage ±Supply Voltage Voltage at Input/Output Pin $(V^{+}) +0.3V,$ $(V^{-}) - 0.3V$

Supply Voltage (V⁺ – V⁻) 16V Output Short Circuit to V+ (Note 10) Output Short Circuit to V-(Note 2)

Lead Temperature

(Soldering, 10 Sec.) 260°C Storage Temp. Range –65°C to +150°C

Junction Temperature 150°C ESD Tolerance (Note 4) 2 kV

Current at Input Pin ±10 mA Current at Output Pin ±30 mA Current at Power Supply Pin 40 mA Power Dissipation (Note 3)

Operating Ratings (Note 1)

Temperature Range

LMC6081AM $-55^{\circ}\text{C} \le \text{T}_{\text{J}} \le +125^{\circ}\text{C}$ LMC6081AI, LMC6081I $-40^{\circ}C \leq T_{J} \leq +85^{\circ}C$ Supply Voltage $4.5V \le V^{+} \le 15.5V$

Thermal Resistance (θ_{JA}), (Note 11)

115°C/W N Package, 8-Pin Molded DIP M Package, 8-Pin Surface Mount 193°C/W

Power Dissipation (Note 9)

DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for T_J = 25°C. **Boldface** limits apply at the temperature extremes. V^+ = 5V, V^- = 0V, V_{CM} = 1.5V, V_O = 2.5V and R_L > 1M unless otherwise specified.

				Тур	LMC6081AM	LMC6081AI	LMC6081I	
Symbol	Parameter	Condi	tions	(Note 5)	Limit	Limit	Limit	Units
					(Note 6)	(Note 6)	(Note 6)	
Vos	Input Offset Voltage			150	350	350	800	μV
					1000	800	1300	Max
TCVos	Input Offset Voltage			1.0				μV/°C
	Average Drift							
I _B	Input Bias Current			0.010				pА
					100	4	4	Max
I _{os}	Input Offset Current			0.005				pА
					100	2	2	Max
R _{IN}	Input Resistance			>10				Tera Ω
CMRR	Common Mode	$0V \le V_{CM} \le 12.0V$		85	75	75	66	dB
	Rejection Ratio	V+ = 15V			72	72	63	Min
+PSRR	Positive Power Supply	5V ≤ V ⁺ ≤ 15V		85	75	75	66	dB
	Rejection Ratio	V _O = 2.5V			72	72	63	Min
-PSRR	Negative Power Supply	0V ≤ V ⁻ ≤ −10V		94	84	84	74	dB
	Rejection Ratio				81	81	71	Min
V _{CM}	Input Common-Mode	V+ = 5V and 15V		-0.4	-0.1	-0.1	-0.1	V
	Voltage Range	for CMRR ≥ 60 dB			0	0	0	Max
				V+ - 1.9	V+ - 2.3	V+ - 2.3	V+ - 2.3	V
					V+ - 2.6	V+ - 2.5	V+ - 2.5	Min
A _V	Large Signal	$R_L = 2 k\Omega$	Sourcing	1400	400	400	300	V/mV
	Voltage Gain	(Note 7)			300	300	200	Min
			Sinking	350	180	180	90	V/mV
					70	100	60	Min
		$R_L = 600\Omega$	Sourcing	1200	400	400	200	V/mV
		(Note 7)			150	150	80	Min
			Sinking	150	100	100	70	V/mV
					35	50	35	Min

DC Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for T_J = 25°C. **Boldface** limits apply at the temperature extremes. V⁺ = 5V, V⁻ = 0V, V_{CM} = 1.5V, V_O = 2.5V and R_L > 1M unless otherwise specified.

			Тур	LMC6081AM	LMC6081AI	LMC6081I	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Limit	Units
				(Note 6)	(Note 6)	(Note 6)	1
Vo	Output Swing	V ⁺ = 5V	4.87	4.80	4.80	4.75	V
		$R_L = 2 k\Omega$ to 2.5V		4.70	4.73	4.67	Min
			0.10	0.13	0.13	0.20	٧
				0.19	0.17	0.24	Max
		V ⁺ = 5V	4.61	4.50	4.50	4.40	٧
		$R_L = 600\Omega$ to 2.5V		4.24	4.31	4.21	Min
			0.30	0.40	0.40	0.50	٧
				0.63	0.50	0.63	Max
		V+ = 15V	14.63	14.50	14.50	14.37	٧
		$R_L = 2 k\Omega$ to 7.5V		14.30	14.34	14.25	Min
			0.26	0.35	0.35	0.44	V
				0.48	0.45	0.56	Max
		V+ = 15V	13.90	13.35	13.35	12.92	٧
		$R_L = 600\Omega$ to 7.5V		12.80	12.86	12.44	Min
			0.79	1.16	1.16	1.33	٧
				1.42	1.32	1.58	Max
lo	Output Current	Sourcing, V _O = 0V	22	16	16	13	mA
	V+ = 5V			8	10	8	Min
		Sinking, V _O = 5V	21	16	16	13	mA
				11	13	10	Min
I _O	Output Current	Sourcing, V _O = 0V	30	28	28	23	mA
	V+ = 15V			18	22	18	Min
		Sinking, V _O = 13V	34	28	28	23	mA
		(Note 10)		19	22	18	Min
Is	Supply Current	$V^{+} = +5V, V_{O} = 1.5V$	450	750	750	750	μA
				900	900	900	Max
		$V^+ = +15V, V_O = 7.5V$	550	850	850	850	μA
				950	950	950	Max

AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for T_J = 25°C, **Boldface** limits apply at the temperature extremes. V⁺ = 5V, V⁻ = 0V, V_{CM} = 1.5V, V_O = 2.5V and R_L > 1M unless otherwise specified.

			Тур	LMC6081AM	LMC6081AI	LMC6081	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Limit	Units
				(Note 6)	(Note 6)	(Note 6)	
SR	Slew Rate	(Note 8)	1.5	0.8	0.8	0.8	V/µs
				0.5	0.6	0.6	Min
GBW	Gain-Bandwidth Product		1.3				MHz
φ _m	Phase Margin		50				Deg
e _n	Input-Referred	F = 1 kHz	22				nV/v∕ Hz
	Voltage Noise						
i _n	Input-Referred	F = 1 kHz	0.0002				pA/√ Hz
	Current Noise						'
T.H.D.	Total Harmonic Distortion	$F = 10 \text{ kHz}, A_V = -10$					
		$R_L = 2 k\Omega, V_O = 8 V_{PP}$	0.01				%
		±5V Supply					

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed.

Note 2: Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Output currents in excess of ±30 mA over long term may adversely affect reliability.

Note 3: The maximum power dissipation is a function of $T_{J(Max)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(Max)} - T_A) / \theta_{JA}$.

Note 4: Human body model, 1.5 $k\Omega$ in series with 100 pF.

Note 5: Typical values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

Note 7: V^+ = 15V, V_{CM} = 7.5V and R_L connected to 7.5V. For Sourcing tests, 7.5V \leq $V_O \leq$ 11.5V. For Sinking tests, 2.5V \leq $V_O \leq$ 7.5V.

Note 8: V+ = 15V. Connected as Voltage Follower with 10V step input. Number specified is the slower of the positive and negative slew rates.

Note 9: For operating at elevated temperatures the device must be derated based on the thermal resistance θ_{JA} with P_D = $(T_J - T_A)/\theta_{JA}$.

Note 10: Do not connect output to V⁺, when V⁺ is greater than 13V or reliability will be adversely affected.

Note 11: All numbers apply for packages soldered directly into a PC board.

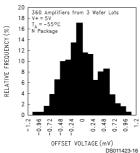
Typical Performance Characteristics $V_S = \pm 7.5V$, $T_A = 25^{\circ}C$, Unless otherwise specified

Distribution of LMC6081 Input Offset Voltage (T_A = +25°C)

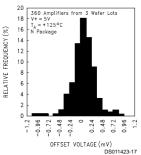
20

> -0.96 -0.72 -0.48

Distribution of LMC6081 Input Offset Voltage (T_A = -55°C)



Distribution of LMC6081 Input Offset Voltage $(T_A = +125^{\circ}C)$



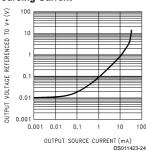
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0.24 0.48 0.72 0.96

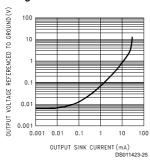
OFFSET VOLTAGE (mV)
DS011423-15

Typical Performance Characteristics $V_S = \pm 7.5V$, $T_A = 25^{\circ}C$, Unless otherwise specified (Continued) Input Bias Current **Supply Current** Input Voltage vs Output Voltage vs Temperature vs Supply Voltage 800 40 SUPPLY CURRENT (µA) 20 600 INPUT VOLTAGE (µV) = 25°C 10 100 fA 400 10 fA 200 -30 10 12 -10 -8 -6 -4 -2 0 2 4 TOTAL SUPPLY VOLTAGE (V) DS011423-19 TEMPERATURE (°C) DS011423-18 Power Supply Rejection Ratio vs Frequency Input Voltage Noise vs Frequency **Common Mode** Rejection Ratio vs Frequency 120 120 100 100 100 VOLTAGE NOISE (nV//Hz) 80 60 o ∟ 10 10k 100k 100 1k 10 100 1k 10k FREQUENCY (Hz) FREQUENCY (Hz) DS011423-21 DS011423-22

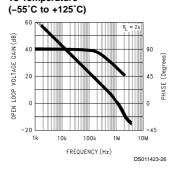
Output Characteristics Sourcing Current



Output Characteristics Sinking Current



Gain and Phase Response vs Temperature



FREQUENCY (Hz)

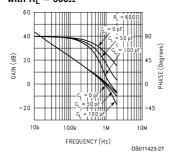
DS011423-23

OUTPUT VOLTAGE (V)
DS011423-20

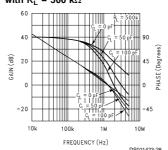
Typical Performance Characteristics $V_S = \pm 7.5V$, $T_A = 25^{\circ}C$, Unless otherwise

specified (Continued)

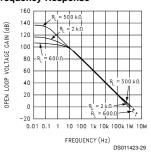
Gain and Phase Response vs Capacitive Load with R $_{L}$ = 600 $\!\Omega$



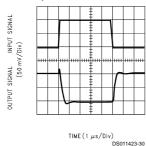
Gain and Phase Response vs Capacitive Load with R $_{L}$ = 500 $k\Omega$



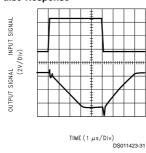
Open Loop Frequency Response



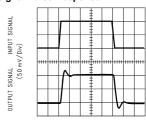
Inverting Small Signal Pulse Response



Inverting Large Signal Pulse Response

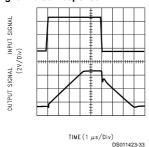


Non-Inverting Small Signal Pulse Response

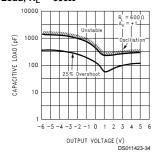


TIME (1 µs/Div) DS011423-32

Non-Inverting Large Signal Pulse Response

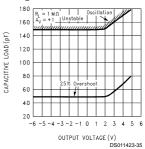


Stability vs Capacitive Load, $R_L = 600\Omega$



6

Stability vs Capacitive Load R_L = 1 $M\Omega$



Applications Hints

AMPLIFIER TOPOLOGY

The LMC6081 incorporates a novel op-amp design topology that enables it to maintain rail-to-rail output swing even when driving a large load. Instead of relying on a push-pull unity gain output buffer stage, the output stage is taken directly from the internal integrator, which provides both low output impedance and large gain. Special feed-forward compensation design techniques are incorporated to maintain stability over a wider range of operating conditions than traditional

micropower op-amps. These features make the LMC6081 both easier to design with, and provide higher speed than products typically found in this ultra-low power class.

COMPENSATING FOR INPUT CAPACITANCE

It is quite common to use large values of feedback resistance for amplifiers with ultra-low input current, like the LMC6081.

Applications Hints (Continued)

Although the LMC6081 is highly stable over a wide range of operating conditions, certain precautions must be met to achieve the desired pulse response when a large feedback resistor is used. Large feedback resistors and even small values of input capacitance, due to transducers, photodiodes, and circuit board parasitics, reduce phase margins. When high input impedances are demanded, guarding of the LMC6081 is suggested. Guarding input lines will not only reduce leakage, but lowers stray input capacitance as well. (See *Printed-Circuit-Board Layout for High Impedance Work*).

The effect of input capacitance can be compensated for by adding a capacitor, C_f , around the feedback resistors (as in *Figure 1*) such that:

$$\frac{1}{2\pi R_1 C_{IN}} \geq \frac{1}{2\pi R_2 C_f}$$

or

$$R_1 \ C_{IN} \leq R_2 \ C_f$$

Since it is often difficult to know the exact value of C_{IN} , C_{f} can be experimentally adjusted so that the desired pulse response is achieved. Refer to the LMC660 and LMC662 for a more detailed discussion on compensating for input capacitance.

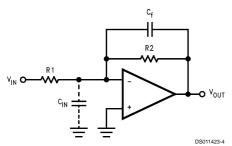


FIGURE 1. Cancelling the Effect of Input Capacitance

CAPACITIVE LOAD TOLERANCE

All rail-to-rail output swing operational amplifiers have voltage gain in the output stage. A compensation capacitor is normally included in this integrator stage. The frequency location of the dominant pole is affected by the resistive load on the amplifier. Capacitive load driving capability can be optimized by using an appropriate resistive load in parallel with the capacitive load (see typical curves).

Direct capacitive loading will reduce the phase margin of many op-amps. A pole in the feedback loop is created by the combination of the op-amp's output impedance and the capacitive load. This pole induces phase lag at the unity-gain crossover frequency of the amplifier resulting in either an oscillatory or underdamped pulse response. With a few external components, op amps can easily indirectly drive capacitive loads, as shown in *Figure 2*.

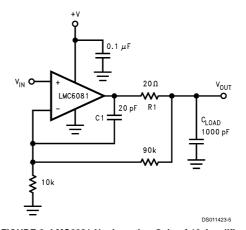


FIGURE 2. LMC6081 Noninverting Gain of 10 Amplifier, Compensated to Handle Capacitive Loads

In the circuit of *Figure 2*, R1 and C1 serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

Capacitive load driving capability is enhanced by using a pull up resistor to V⁺ (*Figure 3*). Typically a pull up resistor conducting 500 μA or more will significantly improve capacitive load responses. The value of the pull up resistor must be determined based on the current sinking capability of the amplifier with respect to the desired output swing. Open loop gain of the amplifier can also be affected by the pull up resistor (see electrical characteristics).

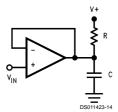


FIGURE 3. Compensating for Large Capacitive Loads with a Pull Up Resistor

PRINTED-CIRCUIT-BOARD LAYOUT FOR HIGH-IMPEDANCE WORK

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It is generally recognized that any circuit which must operate with less than 1000 pA of leakage current requires special layout of the PC board. When one wishes to take advantage of the ultra-low bias current of the LMC6081, typically less than 10 fA, it is essential to have an excellent layout. Fortunately, the techniques of obtaining low leakages are quite simple. First, the user must not ignore the surface leakage of the PC board, even though it may sometimes appear acceptably low, because under conditions of high humidity or dust or contamination, the surface leakage will be appreciable.

To minimize the effect of any surface leakage, lay out a ring of foil completely surrounding the LMC6081's inputs and the terminals of capacitors, diodes, conductors, resistors, relay terminals, etc. connected to the op-amp's inputs, as in Fig-

Applications Hints (Continued)

ure~4. To have a significant effect, guard rings should be placed on both the top and bottom of the PC board. This PC foil must then be connected to a voltage which is at the same voltage as the amplifier inputs, since no leakage current can flow between two points at the same potential. For example, a PC board trace-to-pad resistance of $10^{12}\Omega_{\rm c}$ which is normally considered a very large resistance, could leak 5 pA if the trace were a 5V bus adjacent to the pad of the input. This would cause a 100 times degradation from the LMC6081's actual performance. However, if a guard ring is held within 5 mV of the inputs, then even a resistance of $10^{11}\Omega$ would cause only 0.05 pA of leakage current. See Figure~5 for typical connections of guard rings for standard op-amp configurations.

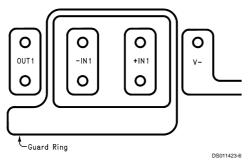
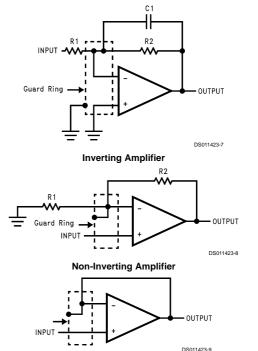
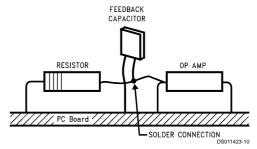


FIGURE 4. Example of Guard Ring in P.C. Board Layout



Follower
FIGURE 5. Typical Connections of Guard Rings

The designer should be aware that when it is inappropriate to lay out a PC board for the sake of just a few circuits, there is another technique which is even better than a guard ring on a PC board: Don't insert the amplifier's input pin into the board at all, but bend it up in the air and use only air as an insulator. Air is an excellent insulator. In this case you may have to forego some of the advantages of PC board construction, but the advantages are sometimes well worth the effort of using point-to-point up-in-the-air wiring. See Figure 6.



(Input pins are lifted out of PC board and soldered directly to components. All other pins connected to PC board).

FIGURE 6. Air Wiring

Latchup

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CMOS devices tend to be susceptible to latchup due to their internal parasitic SCR effects. The (I/O) input and output pins look similar to the gate of the SCR. There is a minimum cur-

Latchup (Continued)

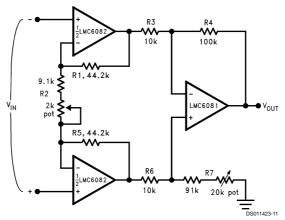
rent required to trigger the SCR gate lead. The LMC6061 and LMC6081 are designed to withstand 100 mA surge current on the I/O pins. Some resistive method should be used to isolate any capacitance from supplying excess current to the I/O pins. In addition, like an SCR, there is a minimum holding current for any latchup mode. Limiting current to the supply pins will also inhibit latchup susceptibility.

Typical Single-Supply Applications

 $(V^{+} = 5.0 V_{DC})$

The extremely high input impedance, and low power consumption, of the LMC6081 make it ideal for applications that

require battery-powered instrumentation amplifiers. Examples of these types of applications are hand-held pH probes, analytic medical instruments, magnetic field detectors, gas detectors, and silicon based pressure transducers. Figure 7 shows an instrumentation amplifier that features high differential and common mode input resistance (>10 $^{14}\Omega$), 0.01% gain accuracy at $A_{\rm V}=1000$, excellent CMRR with 1 k Ω imbalance in bridge source resistance. Input current is less than 100 fA and offset drift is less than 2.5 μ V/ $^{\prime}$ C. R_2 provides a simple means of adjusting gain over a wide range without degrading CMRR. R_7 is an initial trim used to maximize CMRR without using super precision matched resistors. For good CMRR over temperature, low drift resistors should be used.



If $R_1 = R_5$, $R_3 = R_6$, and $R_4 = R_7$; then

$$\frac{V_{OUT}}{V_{IN}} = \frac{R_2 + 2R_1}{R_2} \times \frac{R_4}{R_3}$$

 $A_V \approx 100$ for circuit shown (R₂ = 9.822k).

FIGURE 7. Instrumentation Amplifier

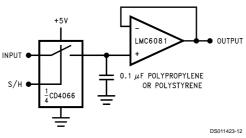


FIGURE 8. Low-Leakage Sample and Hold

Typical Single-Supply Applications (Continued)

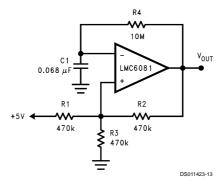
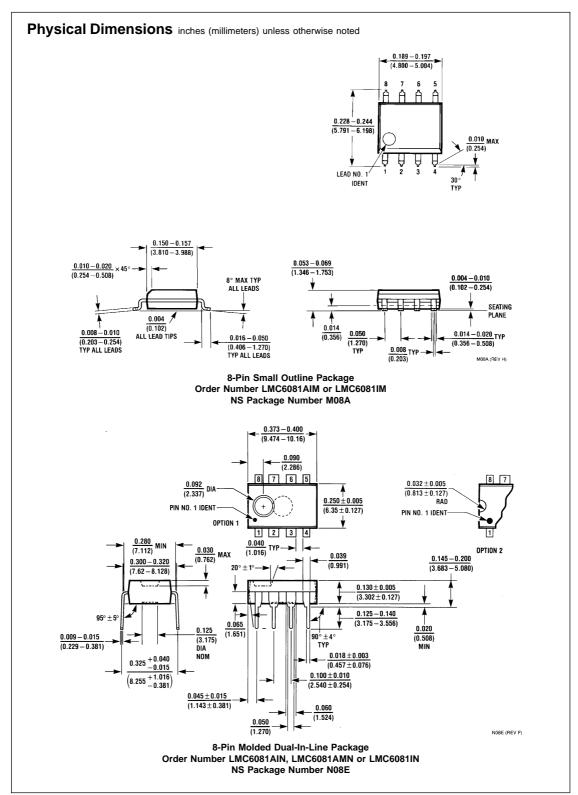


FIGURE 9. 1 Hz Square Wave Oscillator



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- 2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



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