

## **LMC7660 Switched Capacitor Voltage Converter**

Check for Samples: LMC7660

### **FEATURES**

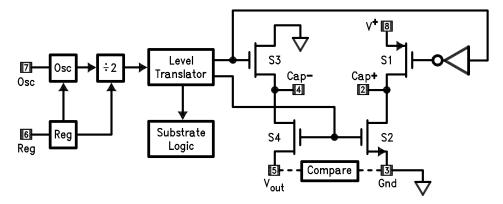
- **Operation Over Full Temperature and Voltage** Range without an External Diode
- Low Supply Current, 200 µA Max
- Pin-for-pin Replacement for the 7660
- Wide Operating Range 1.5V to 10V
- 97% Voltage Conversion Efficiency
- 95% Power Conversion Efficiency
- Easy to Use, Only 2 External Components
- **Extended Temperature Range**

## DESCRIPTION

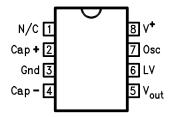
The LMC7660 is a CMOS voltage converter capable of converting a positive voltage in the range of +1.5V to +10V to the corresponding negative voltage of -1.5V to -10V. The LMC7660 is a pin-for-pin replacement for the industry-standard 7660. The converter features: operation over full temperature and voltage range without need for an external diode, low quiescent current, and high power efficiency.

The LMC7660 uses its built-in oscillator to switch 4 power MOS switches and charge two inexpensive electrolytic capacitors.

## **Block Diagram**



## **Pin Configuration**



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## ABSOLUTE MAXIMUM RATINGS(1)(2)

Supply Voltage	10.5V
	-0.3V to (V <sup>+</sup> + 0.3V)
lanut Valtaga an Bin C. 7(3)	for V <sup>+</sup> < 5.5V
Input Voltage on Pin 6, 7 <sup>(3)</sup>	$(V^+ - 5.5V)$ to $(V^+ + 0.3V)$
	for $V^+ > 5.5V$
Current into Pin 6 (3)	20 μΑ
Output Short Circuit Duration (V <sup>+</sup> ≤ 5.5V)	Continuous
Power Dissipation (4)	
PDIP Package	1.4W
SOIC Package	0.6W
T <sub>J</sub> Max <sup>(4)</sup>	150°C
$\theta_{\mathrm{JA}}^{(4)}$	
PDIP Package	90°C/W
SOIC Package	160°C/W
Storage Temp. Range	-65°C ≤ T ≤ 150°C
Lead Temperature	
(Soldering, 5 sec.)	260°C
ESD Tolerance (5)	± 2000V

- (1) Absolute Maximum ratings indicate limits beyond which damage to the device may occur. DC and AC electrical specifications do not apply when operating the device beyond its rated operating conditions. See Note (1) under Electrical Characteristics for conditions.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) Connecting any input terminal to voltages greater than V<sup>+</sup> or less than ground may cause destructive latchup. It is recommended that no inputs from sources operating from external supplies be applied prior to "power-up" of the LMC7660.
- (4) For operation at elevated temperature, these devices must be derated based on a thermal resistance of  $\theta_{ia}$  and  $T_i$  max,  $T_i = T_A + \theta_{ia} P_D$ .
- (5) The test circuit consists of the human body model of 100 pF in series with 1500Ω.

### ELECTRICAL CHARACTERISTICS(1)

Symbol	Parameter	Conditions	Тур	LMC7660IN/ LMC7660IM	Units
•				Limit (2)	Limits
Is	Supply Current	R <sub>L</sub> = ∞	120	200	μΑ
				400	max
V <sup>+</sup> H	Supply Voltage	R <sub>L</sub> = 10 kΩ, Pin 6 Open	3 to 10	3 to 10	V
	Range High (3)	Voltage Efficiency ≥ 90%		3 to 10	
V <sup>+</sup> L	Supply Voltage	$R_L$ = 10 kΩ, Pin 6 to Gnd.	1.5 to 3.5	1.5 to 3.5	V
	Range Low	Voltage Efficiency ≥ 90%		1.5 to 3.5	

- (1) Boldface numbers apply at temperature extremes. All other numbers apply at T<sub>A</sub> = 25°C, V<sup>+</sup> = 5V, C<sub>osc</sub> = 0, and apply for the LMC7660 unless otherwise specified. Test circuit is shown in Figure 1.
- (2) Limits at room temperature are specified and 100% production tested. Limits in **boldface** are specified over the operating temperature range (but not 100% tested), and are not used to calculate outgoing quality levels.
- (3) The LMC7660 can operate without an external diode over the full temperature and voltage range. The LMC7660 can also be used with the external diode Dx, when replacing previous 7660 designs.



## **ELECTRICAL CHARACTERISTICS**(1) (continued)

Symbol	Parameter	rameter Conditions		LMC7660IN/ LMC7660IM	Units	
-			Тур	Limit (2)	Limits	
R <sub>out</sub>	Output Source	I <sub>L</sub> = 20 mA	55	100	Ω	
	Resistance			120	max	
		$V = 2V$ , $I_L = 3 \text{ mA}$	110	200	Ω	
		Pin 6 Short to Gnd.		300	max	
F <sub>osc</sub>	Oscillator		10		kHz	
	Frequency					
P <sub>eff</sub>	Power Efficiency	$R_L = 5 \text{ k}\Omega$	97	95	%	
				90	min	
V <sub>o eff</sub>	Voltage Conversion	R <sub>L</sub> = ∞	99.9	97	%	
0 0	Efficiency			95	min	
I <sub>osc</sub>	Oscillator Sink or	Pin 7 = Gnd. or V <sup>+</sup>	3		μΑ	
	Source Current					

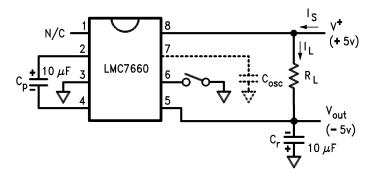
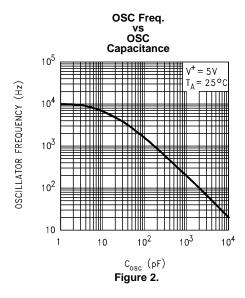
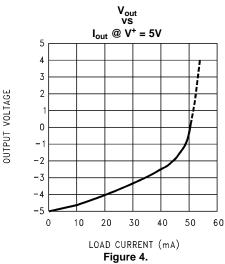


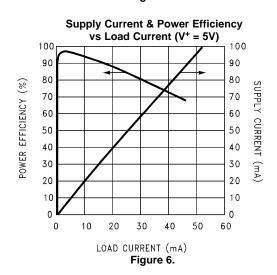
Figure 1. LMC7660 Test Circuit

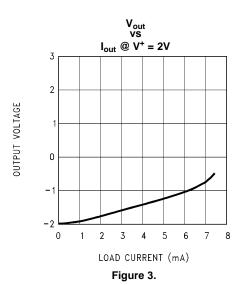


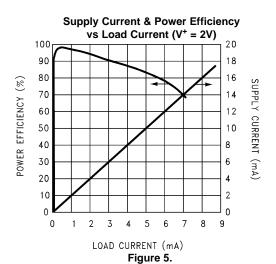
### TYPICAL PERFORMANCE CHARACTERISTICS

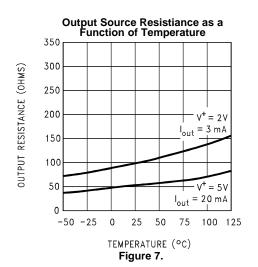








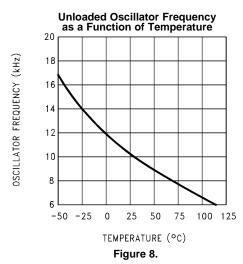


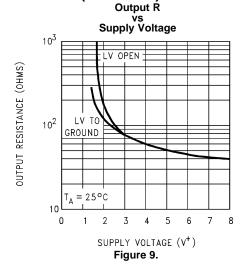


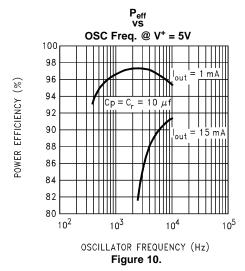
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## **TYPICAL PERFORMANCE CHARACTERISTICS (continued)**









#### **APPLICATION INFORMATION**

#### **CIRCUIT DESCRIPTION**

The LMC7660 contains four large CMOS switches which are switched in a sequence to provide supply inversion  $V_{out} = -V_{in}$ . Energy transfer and storage are provided by two inexpensive electrolytic capacitors. Figure 11 shows how the LMC7660 can be used to generate  $-V^+$  from  $V^+$ . When switches S1 and S3 are closed,  $C_p$  charges to the supply voltage  $V^+$ . During this time interval, switches S2 and S4 are open. After  $C_p$  charges to  $V^+$ , S1 and S3 are opened, S2 and S4 are then closed. By connecting S2 to ground,  $C_p$  develops a voltage  $-V^+/2$  on  $C_r$ . After a number of cycles  $C_r$  will be pumped to exactly  $-V^+$ . This transfer will be exact assuming no load on  $C_r$ , and no loss in the switches.

In the circuit of Figure 11, S1 is a P-channel device and S2, S3, and S4 are N-channel devices. Because the output is biased below ground, it is important that the  $p^-$  wells of S3 and S4 never become forward biased with respect to either their sources or drains. A substrate logic circuit specifies that these  $p^-$  wells are always held at the proper voltage. Under all conditions S4  $p^-$  well must be at the lowest potential in the circuit. To switch off S4, a level translator generates  $V_{GS4} = 0V$ , and this is accomplished by biasing the level translator from the S4  $p^-$  well.

An internal RC oscillator and  $\div$  2 circuit provide timing signals to the level translator. The built-in regulator biases the oscillator and divider to reduce power dissipation on high supply voltage. The regulator becomes active at about V<sup>+</sup> = 6.5V. Low voltage operation can be improved if the LV pin is shorted to ground for V<sup>+</sup>  $\le$  3.5V. For V<sup>+</sup>  $\ge$  3.5V, the LV pin must be left open to prevent damage to the part.

### POWER EFFICIENCY AND RIPPLE

It is theoretically possible to approach 100% efficiency if the following conditions are met:

- 1. The drive circuitry consumes little power.
- 2. The power switches are matched and have low R<sub>on</sub>.
- 3. The impedance of the reservoir and pump capacitors are negligibly small at the pumping frequency.

The LMC7660 closely approaches 1 and 2 above. By using a large pump capacitor  $C_p$ , the charge removed while supplying the reservoir capacitor is small compared to  $C_p$ 's total charge. Small removed charge means small changes in the pump capacitor voltage, and thus small energy loss and high efficiency. The energy loss by  $C_p$  is:

$$E = \frac{1}{2}C_{p}(V1^{2} - V2^{2})$$
 (1)

By using a large reservoir capacitor, the output ripple can be reduced to an acceptable level. For example, if the load current is 5 mA and the accepted ripple is 200 mV, then the reservoir capacitor can omit approximately be calculated from:

$$Is = C_r \frac{dv}{dt}$$

$$\sim C_r \times \frac{V_{ripple p-p}}{4/F_{osc}} \qquad C_r = \frac{0.5 \text{ mA}}{0.5 \text{V/ms}} = 10 \ \mu\text{F}$$
(2)

### **PRECAUTIONS**

- 1. Do not exceed the maximum supply voltage or junction temperature.
- 2. Do not short pin 6 (LV terminal) to ground for supply voltages greater than 3.5V.
- 3. Do not short circuit the output to V<sup>+</sup>.
- 4. External electrolytic capacitors C<sub>r</sub> and C<sub>p</sub> should have their polarities connected as shown in Figure 1.

### **REPLACING PREVIOUS 7660 DESIGNS**

To prevent destructive latchup, previous 7660 designs require a diode in series with the output when operated at elevated temperature or supply voltage. Although this prevented the latchup problem of these designs, it lowered the available output voltage and increased the output series resistance.

The TI LMC7660 has been designed to solve the inherent latch problem. The LCM7660 can operate over the entire supply voltage and temperature range without the need for an output diode. When replacing existing designs, the LMC7660 can be operated with diode Dx.



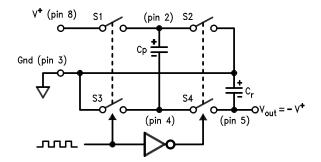


Figure 11. Idealized Voltage Converter

#### **TYPICAL APPLICATIONS**

#### CHANGING OSCILLATOR FREQUENCY

It is possible to dramatically reduce the quiescent operating current of the LMC7660 by lowering the oscillator frequency. The oscillator frequency can be lowered from a nominal 10 kHz to several hundred hertz, by adding a slow-down capacitor  $C_{osc}$  (Figure 12). As shown in the Typical Performance Curves the supply current can be lowered to the 10  $\mu$ A range. This low current drain can be extremely useful when used in  $\mu$ Power and battery back-up equipment. It must be understood that the lower operating frequency and supply current cause an increased impedance of  $C_r$  and  $C_p$ . The increased impedance, due to a lower switching rate, can be offset by raising  $C_r$  and  $C_p$  until ripple and load current requirements are met.

#### SYNCHRONIZING TO AN EXTERNAL CLOCK

Figure 13 shows an LMC7660 synchronized to an external clock. The CMOS gate overrides the internal oscillator when it is necessary to switch faster or reduce power supply interference. The external clock still passes through the ÷2 circuit in the 7660, so the pumping frequency will be ½ the external clock frequency.

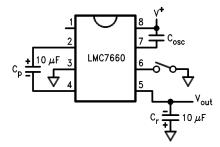


Figure 12. Reduce Supply Current by Lowering Oscillator Frequency

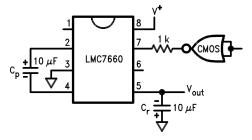


Figure 13. Synchronizing to an External Clock



#### LOWERING OUTPUT IMPEDANCE

Paralleling two or more LMC7660's lowers output impedance. Each device must have it's own pumping capacitor  $C_p$ , but the reservoir capacitor  $C_r$  is shared as depicted in Figure 14. The composite output resistance is:

$$R_{out} = \frac{R_{out} \text{ of one LMC7660}}{\text{Number of devices}}$$
(3)

#### **INCREASING OUTPUT VOLTAGE**

Stacking the LMC7660s is an easy way to produce a greater negative voltage. It should be noted that the input current required for each stage is twice the load current on that stage as shown in Figure 15. The effective output resistance is approximately the sum of the individual R<sub>out</sub> values, and so only a few levels of multiplication can be used.

It is possible to generate -15V from +5V by connecting the second 7660's pin 8 to +5V instead of ground as shown in Figure 16. Note that the second 7660 sees a full 20V and the input supply should not be increased beyond +5V.

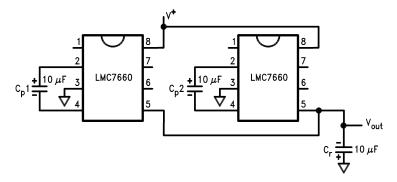


Figure 14. Lowering Output Resistance by Paralleling Devices

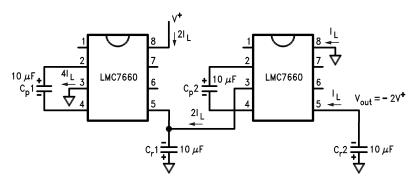


Figure 15. Higher Voltage by Cascade

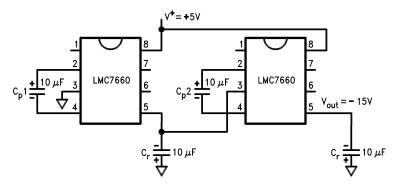


Figure 16. Getting -15V from +5V

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#### **SPLIT V<sup>+</sup> IN HALF**

Figure 17 is one of the more interesting applications for the LMC7660. The circuit can be used as a precision voltage divider (for very light loads), alternately it is used to generate a  $\frac{1}{2}$  supply point in battery applications. In the  $\frac{1}{2}$  cycle when S1 and S3 are closed, the supply voltage divides across the capacitors in a conventional way proportional to their value. In the  $\frac{1}{2}$  cycle when S2 and S4 are closed, the capacitors switch from a series connection to a parallel connection. This forces the capacitors to have the same voltage; the charge redistributes to maintain precisely V<sup>+</sup>/2, across C<sub>p</sub> and C<sub>r</sub>. In this application all devices are only V<sup>+</sup>/2, and the supply voltage can be raised to 20V giving exactly 10V at V<sub>out</sub>.

#### **GETTING UP ... AND DOWN**

The LMC7660 can also be used as a positive voltage multiplier. This application, shown in Figure 18, requires 2 additional diodes. During the first ½ cycle S2 charges  $C_p1$  through D1; D2 is reverse biased. In the next ½ cycle S2 is open and S1 is closed. Since  $C_p1$  is charged to  $V^+ - V_{D1}$  and is referenced to  $V^+$  through S1, the junction of D1 and D2 is at  $V^+ + (V^+ - V_{D1})$ . D1 is reverse biased in this interval. This application uses only two of the four switches in the 7660. The other two switches can be put to use in performing a negative conversion at the same time as shown in Figure 19. In the ½ cycle that D1 is charging  $C_p1$ ,  $C_p2$  is connected from ground to  $-V_{out}$  via S2 and S4, and  $C_r2$  is storing  $C_p2$ 's charge. In the interval that S1 and S3 are closed,  $C_p1$  pumps the junction of D1 and D2 above  $V^+$ , while  $C_p2$  is refreshed from  $V^+$ .

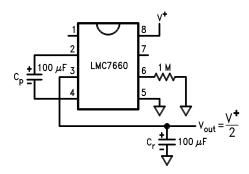


Figure 17. Split V+ in Half

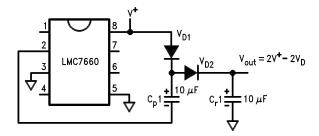


Figure 18. Positive Voltage Multiplier

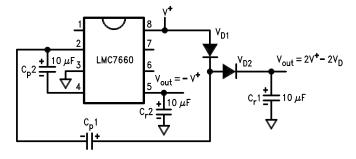


Figure 19. Combined Negative Converter and Positive Multiplier



#### THERMOMETER SPANS 180°C

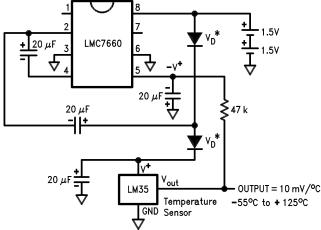
Using the combined negative and positive multiplier of Figure 20 with an LM35 it is possible to make a  $\mu$ Power thermometer that spans a 180°C temperature range. The LM35 temperature sensor has an output sensitivity of 10 mV/°C, while drawing only 50  $\mu$ A of quiescent current. In order for the LM35 to measure negative temperatures, a pull down to a negative voltage is required. Figure 20 shows a thermometer circuit for measuring temperatures from -55°C to +125°C and requiring only two 1.5V cells. End of battery life can be extended by replacing the up converter diodes with Schottky's.

### **REGULATING -VOLT**

It is possible to regulate the output of the LMC7660 and still maintain  $\mu$ Power performance. This is done by enclosing the LMC7660 in a loop with a LP2951. The circuit of Figure 21 will regulate  $V_{out}$  to -5V for  $I_L = 10$  mA, and  $V_{in} = 6V$ . For  $V_{in} > 7V$ , the output stays in regulation up to  $I_L = 25$  mA. The error flag on pin 5 of the LP2951 sets low when the regulated output at pin 4 drops by about 5%. The LP2951 can be shutdown by taking pin 3 high; the LMC7660 can be shutdown by shorting pin 7 and pin 8.

The LP2951 can be reconfigured to an adjustable type regulator, which means the LMC7660 can give a regulated output from -2.0V to -10V dependent on the resistor ratios R1 and R2, as shown in Figure 22,  $V_{ref} = 1.235V$ :

$$V_{\text{out}} = V_{\text{ref}} \left( 1 + \frac{R1}{R2} \right) \tag{4}$$



\*For lower voltage operation, use Schottky rectifiers

Figure 20. μPower Thermometer Spans 180°C, and Pulls Only 150 μA

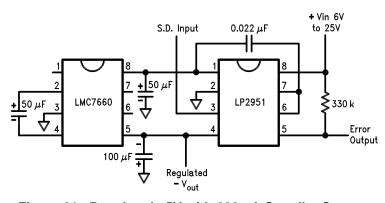
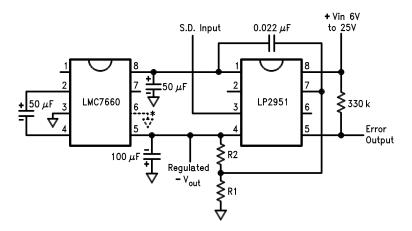


Figure 21. Regulated -5V with 200 µA Standby Current





$$V_{out} = V_{ref} \left( 1 + \frac{R1}{R2} \right)$$

Figure 22. LMC7660 and LP2951 Make a Negative Adjustable Regulator

V<sub>ref</sub> = 1.235V \*Low voltage operation

## SNOSBZ9C - APRIL 1997 - REVISED APRIL 2013



## **REVISION HISTORY**

	Pag	ge	
•	Changed layout of National Data Sheet to TI format		11

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25-Feb-2014

#### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LMC7660IM	NRND	SOIC	D	8	95	TBD	Call TI	Call TI		LMC76 60IM	
LMC7660IM/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	SN   CU SN	Level-1-260C-UNLIM		LMC76 60IM	Samples
LMC7660IMX	NRND	SOIC	D	8	2500	TBD	Call TI	Call TI		LMC76 60IM	
LMC7660IMX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	SN   CU SN	Level-1-260C-UNLIM		LMC76 60IM	Samples
LMC7660IN	LIFEBUY	PDIP	Р	8	40	TBD	Call TI	Call TI		LMC 7660IN	
LMC7660IN/NOPB	ACTIVE	PDIP	Р	8	40	Green (RoHS & no Sb/Br)	SN   CU SN	Level-1-NA-UNLIM		LMC 7660IN	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



## **PACKAGE OPTION ADDENDUM**

25-Feb-2014

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

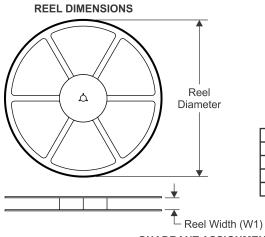
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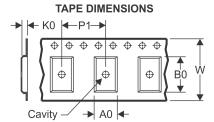
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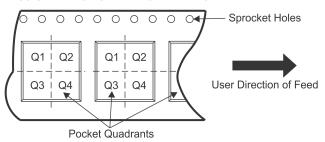
## TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

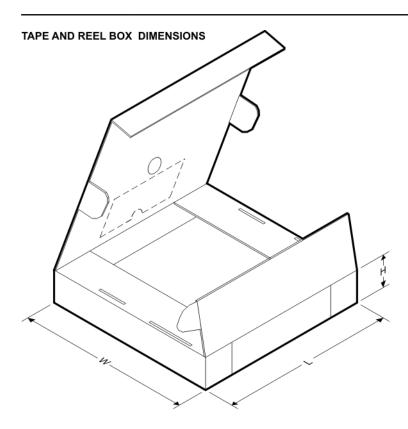
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMC7660IMX	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LMC7660IMX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1

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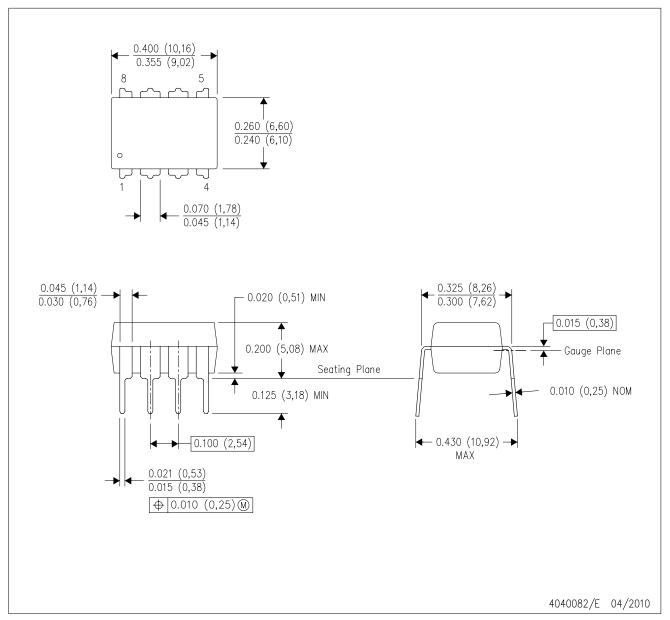


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMC7660IMX	SOIC	D	8	2500	367.0	367.0	35.0
LMC7660IMX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0

# P (R-PDIP-T8)

## PLASTIC DUAL-IN-LINE PACKAGE



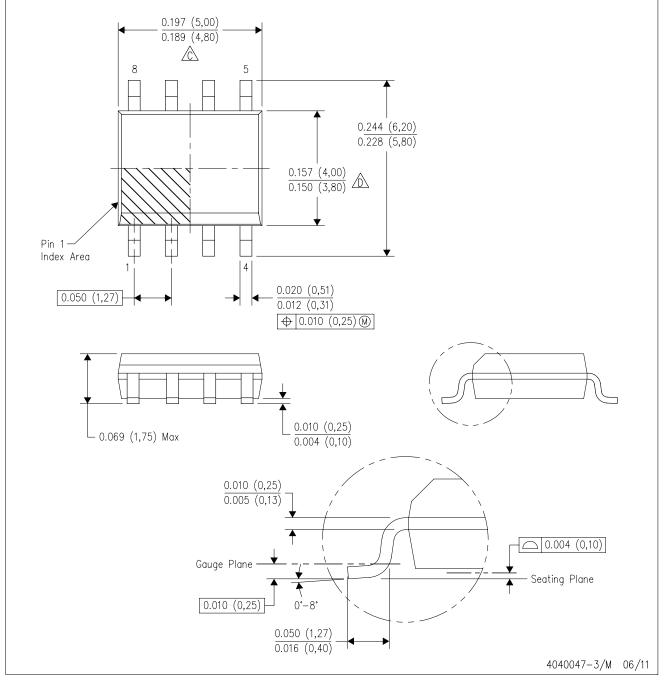
NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



## D (R-PDSO-G8)

## PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



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