- Low Supply-Voltage Range, 1.8 V to 3.6 V
- Ultralow-Power Consumption:
  - Active Mode: 400 μA at 1 MHz, 3.0 V
  - Standby Mode: 1.6 μA
  - Off Mode (RAM Retention): 0.1 μA
- Five Power-Saving Modes
- Wake-Up From Standby Mode in Less Than 6 μs
- Frequency-Locked Loop, FLL+
- 16-Bit RISC Architecture, 125-ns Instruction Cycle Time
- Three Independent 16-bit Sigma-Delta A/D Converters With Differential PGA Inputs
- 16-Bit Timer\_A With Three Capture/Compare Registers
- Integrated LCD Driver for 128 Segments
- Serial Communication Interface (USART), Asynchronous UART or Synchronous SPI Selectable by Software
- Brownout Detector

- Supply Voltage Supervisor/Monitor With Programmable Level Detection
- Serial Onboard Programming, No External Programming Voltage Needed Programmable Code Protection by Security Fuse
- Bootstrap Loader in Flash Devices
- Family Members Include:
  - MSP430F423:

8KB + 256B Flash Memory, 256B RAM

MSP430F425:

16KB + 256B Flash Memory, 512B RAM

- MSP430F427:

32KB + 256B Flash Memory, 1KB RAM

- Available in 64-Pin Quad Flat Pack (QFP)
- For Complete Module Descriptions, Refer to the MSP430x4xx Family User's Guide, Literature Number SLAU056

#### description

The Texas Instruments MSP430 family of ultralow power microcontrollers consist of several devices featuring different sets of peripherals targeted for various applications. The architecture, combined with five low power modes, is optimized to achieve extended battery life in portable measurement applications. The device features a powerful 16-bit RISC CPU, 16-bit registers, and constant generators that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows wake-up from low-power modes to active mode in less than 6 µs.

The MSP430F42x series are microcontroller configurations with three independent 16-bit sigma-delta A/D converters, each with an integrated differential programmable gain amplifier input stage. Also included is a built-in 16-bit timer, 128 LCD segment drive capability, hardware multiplier, and 14 I/O pins.

Typical applications include high resolution applications such as handheld metering equipment, weigh scales, and energy meters.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications. These devices have limited built-in ESD protection.



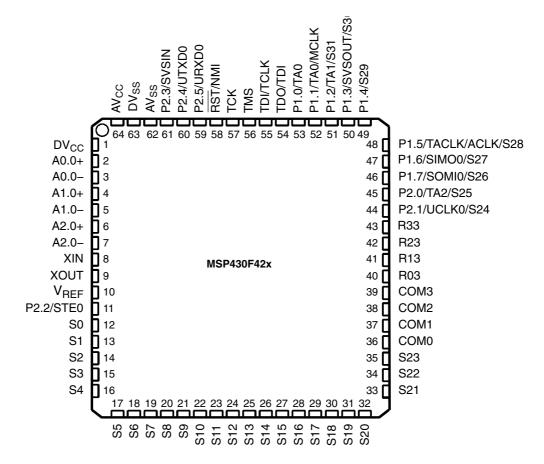
Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



#### **AVAILABLE OPTIONS**

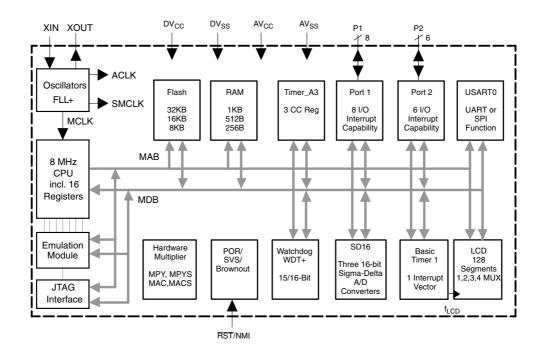
	PACKAGED DEVICES	
T <sub>A</sub>	PLASTIC 64-PIN QFP (PM)	
–40°C to 85°C	MSP430F423IPM MSP430F425IPM	
-40°C to 85°C	MSP430F425IPM MSP430F427IPM	

### pin designation†



<sup>&</sup>lt;sup>†</sup> Open connection recommended for all unused analog inputs.

### functional block diagram



### **MSP430F42x Terminal Functions**

TERMINAL	L		
PN		I/O	DESCRIPTION
NAME	NO.		
DV <sub>CC</sub>	1		Digital supply voltage, positive terminal.
A0.0+	2	I	Internal connection to SD16 Channel 0, input 0 +. (see Note 1)
A0.0-	3	ı	Internal connection to SD16 Channel 0, input 0 –. (see Note 1)
A1.0+	4	I	Internal connection to SD16 Channel 1, input 0 +. (see Note 1)
A1.0-	5	I	Internal connection to SD16 Channel 1, input 0 –. (see Note 1)
A2.0+	6	ı	Internal connection to SD16 Channel 2, input 0 +. (see Note 1)
A2.0-	7	I	Internal connection to SD16 Channel 2, input 0 –. (see Note 1)
XIN	8	ı	Input port for crystal oscillator XT1. Standard or watch crystals can be connected.
XOUT	9	0	Output terminal of crystal oscillator XT1
$V_{REF}$	10	I/O	Input for an external reference voltage / internal reference voltage output (can be used as mid-voltage)
P2.2/STE0	11	I/O	General-purpose digital I/O / slave transmit enable—USART0/SPI mode
S0	12	0	LCD segment output 0
S1	13	0	LCD segment output 1
S2	14	0	LCD segment output 2
S3	15	0	LCD segment output 3
S4	16	0	LCD segment output 4
S5	17	0	LCD segment output 5
S6	18	0	LCD segment output 6
S7	19	0	LCD segment output 7
S8	20	0	LCD segment output 8
S9	21	0	LCD segment output 9
S10	22	0	LCD segment output 10
S11	23	0	LCD segment output 11
S12	24	0	LCD segment output 12
S13	25	0	LCD segment output 13
S14	26	0	LCD segment output 14
S15	27	0	LCD segment output 15
S16	28	0	LCD segment output 16
S17	29	0	LCD segment output 17
S18	30	0	LCD segment output 18
S19	31	0	LCD segment output 19
S20	32	0	LCD segment output 20
S21	33	0	LCD segment output 21
S22	34	0	LCD segment output 22
S23	35	0	LCD segment output 23
COM0	36	0	Common output, COM0-3 are used for LCD backplanes.
COM1	37	0	Common output, COM0-3 are used for LCD backplanes.
COM2	38	0	Common output, COM0-3 are used for LCD backplanes.
COM3	39	0	Common output, COM0-3 are used for LCD backplanes.
R03	40	Ī	Input port of fourth positive (lowest) analog LCD level (V5)

NOTE 1: Open connection recommended for all unused analog inputs.



# **MSP430F42x Terminal Functions (Continued)**

TERMINAL				
PN		I/O	DESCRIPTION	
NAME	NO.			
R13	41	ı	Input port of third most positive analog LCD level (V4 or V3)	
R23	42	ı	Input port of second most positive analog LCD level (V2)	
R33	43	0	Output port of most positive analog LCD level (V1)	
P2.1/UCLK0/S24	44	I/O	General-purpose digital I/O / external clock input-USART0/UART or SPI mode, clock output—USART0/SPI mode / LCD segment output 24 (See Note 1)	
P2.0/TA2/S25	45	I/O	General-purpose digital I/O / Timer_A Capture: CCI2A input, Compare: Out2 output / LCD segment output 25 (See Note 1)	
P1.7/SOMI0/S26	46	I/O	General-purpose digital I/O / slave out/master in of USART0/SPI mode / LCD segment output 26 (See Note 1)	
P1.6/SIMO0/S27	47	I/O	General-purpose digital I/O / slave in/master out of USART0/SPI mode / LCD segment output 27 (See Note 1)	
P1.5/TACLK/ ACLK/S28	48	I/O	General-purpose digital I/O / Timer_A and SD16 clock signal TACLK input / ACLK output (divided by 1, 2, 4, or 8) / LCD segment output 28 (See Note 1)	
P1.4/S29	49	I/O	General-purpose digital I/O / LCD segment output 29 (See Note 1)	
P1.3/SVSOUT/ S30	50	I/O	General-purpose digital I/O / SVS: output of SVS comparator / LCD segment output 30 (See Note 1)	
P1.2/TA1/S31	51	I/O	General-purpose digital I/O / Timer_A, Capture: CCI1A, CCI1B input, Compare: Out1 output / LCD segmen output 31 (See Note 1)	
P1.1/TA0/MCLK	52	I/O	General-purpose digital I/O / Timer_A, Capture: CCI0B input / MCLK output.  Note: TA0 is only an input on this pin / BSL receive	
P1.0/TA0	53	I/O	General-purpose digital I/O / Timer_A, Capture: CCI0A input, Compare: Out0 output / BSL transmit	
TDO/TDI	54	I/O	Test data output port. TDO/TDI data output or programming data input terminal.	
TDI/TCLK	55	ı	Test data input or test clock input. The device protection fuse is connected to TDI.	
TMS	56	I	Test mode select. TMS is used as an input port for device programming and test.	
TCK	57	ı	Test clock. TCK is the clock input port for device programming and test.	
RST/NMI	58	-	Reset input or nonmaskable interrupt input port	
P2.5/URXD0	59	I/O	General-purpose digital I/O / receive data in—USART0/UART mode	
P2.4/UTXD0	60	I/O	General-purpose digital I/O / transmit data out—USART0/UART mode	
P2.3/SVSIN	61	I/O	General-purpose digital I/O / Analog input to brownout, supply voltage supervisor	
AV <sub>SS</sub>	62		Analog supply voltage, negative terminal. Supplies SD16, SVS, brownout, oscillator, and LCD resistive divider circuitry.	
DV <sub>SS</sub>	63		Digital supply voltage, negative terminal	
AV <sub>CC</sub>	64		Analog supply voltage, positive terminal. Supplies SD16, SVS, brownout, oscillator, and LCD resistive divider circuitry; must not power up prior to $DV_CC$ .	

NOTE 1: LCD function selected automatically when applicable LCD module control bits are set, not with PxSEL bits.



#### short-form description

#### **CPU**

The MSP430 CPU has a 16-bit RISC architecture that is highly transparent to the application. All operations, other than program-flow instructions, are performed as register operations in conjunction with seven addressing modes for source operand and four addressing modes for destination operand.

The CPU is integrated with 16 registers that provide reduced instruction execution time. The register-to-register operation execution time is one cycle of the CPU clock.

Four of the registers, R0 to R3, are dedicated as program counter, stack pointer, status register, and constant generator respectively. The remaining registers are general-purpose registers.

Peripherals are connected to the CPU using data, address, and control buses, and can be handled with all instructions.

#### instruction set

The instruction set consists of 51 instructions with three formats and seven address modes. Each instruction can operate on word and byte data. Table 1 shows examples of the three types of instruction formats; the address modes are listed in Table 2.



**Table 1. Instruction Word Formats** 

Dual operands, source-destination	e.g. ADD R4,R5	R4 + R5> R5
Single operands, destination only	e.g. CALL R8	PC>(TOS), R8> PC
Relative jump, un/conditional	e.g. JNE	Jump-on-equal bit = 0

**Table 2. Address Mode Descriptions** 

ADDRESS MODE	s	D	SYNTAX	EXAMPLE	OPERATION
Register	•	•	MOV Rs,Rd	MOV R10,R11	R10> R11
Indexed	•	•	MOV X(Rn),Y(Rm)	MOV 2(R5),6(R6)	M(2+R5)> M(6+R6)
Symbolic (PC relative)	•	•	MOV EDE,TONI		M(EDE)> M(TONI)
Absolute	•	•	MOV &MEM,&TCDAT		M(MEM)> M(TCDAT)
Indirect	•		MOV @Rn,Y(Rm)	MOV @R10,Tab(R6)	M(R10)> M(Tab+R6)
Indirect autoincrement	•		MOV @Rn+,Rm	MOV @R10+,R11	M(R10)> R11 R10 + 2> R10
Immediate	•		MOV #X,TONI	MOV #45,TONI	#45> M(TONI)

NOTE: S = source D = destination



#### operating modes

The MSP430 has one active mode and five software-selectable low-power modes of operation. An interrupt event can wake up the device from any of the five low-power modes, service the request, and restore back to the low-power mode on return from the interrupt program.

The following six operating modes can be configured by software:

- Active mode (AM)
  - All clocks are active
- Low-power mode 0 (LPM0)
  - CPU is disabled ACLK and SMCLK remain active, MCLK is available to modules FLL+ loop control remains active
- Low-power mode 1 (LPM1)
  - CPU is disabled
     ACLK and SMCLK remain active, MCLK is available to modules
     FLL+ loop control is disabled
- Low-power mode 2 (LPM2)
  - CPU is disabled MCLK, FLL+ loop control, and DCOCLK are disabled DCO's dc-generator remains enabled ACLK remains active
- Low-power mode 3 (LPM3)
  - CPU is disabled MCLK, FLL+ loop control, and DCOCLK are disabled DCO's dc-generator is disabled ACLK remains active
- Low-power mode 4 (LPM4)
  - CPU is disabled
     ACLK is disabled
     MCLK, FLL+ loop control, and DCOCLK are disabled
     DCO's dc-generator is disabled
     Crystal oscillator is stopped



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#### interrupt vector addresses

The interrupt vectors and the power-up starting address are located in the address range of 0FFFFh–0FFE0h. The vector contains the 16-bit address of the appropriate interrupt-handler instruction sequence.

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY
Power-up External Reset Watchdog Flash memory PC Out-of-Range (see Note 4)	WDTIFG KEYV (see Note 1)	Reset	0FFFEh	15, highest
NMI Oscillator Fault Flash memory access violation	NMIIFG (see Notes 1 and 3) OFIFG (see Notes 1 and 3) ACCVIFG (see Notes 1 and 3)	(Non)maskable (Non)maskable (Non)maskable	0FFFCh	14
			0FFFAh	13
SD16	SD16CCTLx SD16OVIFG, SD16CCTLx SD16IFG (see Notes 1 and 2)	Maskable	0FFF8h	12
			0FFF6h	11
Watchdog Timer	WDTIFG	Maskable	0FFF4h	10
USART0 Receive	URXIFG0	Maskable	0FFF2h	9
USART0 Transmit	UTXIFG0	Maskable	0FFF0h	8
			0FFEEh	7
Timer_A3	TACCR0 CCIFG (see Note 2)	Maskable	0FFECh	6
Timer_A3	TACCR1 and TACCR2 CCIFGs, and TACTL TAIFG (see Notes 1 and 2)	Maskable	0FFEAh	5
I/O port P1 (eight flags)	P1IFG.0 to P1IFG.7 (see Notes 1 and 2)	Maskable	0FFE8h	4
			0FFE6h	3
			0FFE4h	2
I/O port P2 (eight flags)	P2IFG.0 to P2IFG.7 (see Notes 1 and 2)	Maskable	0FFE2h	1
Basic Timer1	BTIFG	Maskable	0FFE0h	0, lowest

NOTES: 1. Multiple source flags

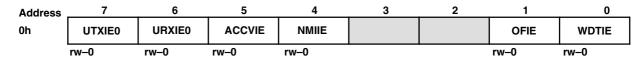
- 2. Interrupt flags are located in the module.
- 3. (Non)maskable: the individual interrupt-enable bit can disable an interrupt event, but the general interrupt-enable cannot.
- 4. A reset is generated if the CPU tries to fetch instructions from within the module register memory address range (0h–01FFh) or from within unused address ranges (from 0600h to 0BFFh).



#### special function registers

Most interrupt and module enable bits are collected into the lowest address space. Special function register bits that are not allocated to a functional purpose are not physically present in the device. Simple software access is provided with this arrangement.

#### interrupt enable 1 and 2



WDTIE: Watchdog-timer interrupt enable. Inactive if watchdog mode is selected. Active if watchdog

timer is configured in interval timer mode.

OFIE: Oscillator-fault-interrupt enable
NMIIE: Nonmaskable-interrupt enable

ACCVIE: Flash access violation interrupt enable

URXIE0: USART0: UART and SPI receive-interrupt enable UTXIE0: USART0: UART and SPI transmit-interrupt enable



BTIE: Basic Timer1 interrupt enable

#### interrupt flag register 1 and 2



WDTIFG: Set on watchdog timer overflow (in watchdog mode) or security key violation. Reset on V<sub>CC</sub>

power up or a reset condition at the RST/NMI pin in reset mode.

OFIFG: Flag set on oscillator fault NMIIFG: Set via RST/NMI pin

URXIFG0: USART0: UART and SPI receive flag
UTXIFG0: USART0: UART and SPI transmit flag

Address 7 6 5 4 3 2 1 0

BTIFG rw-0

BTIFG: Basic Timer1 interrupt flag

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### module enable registers 1 and 2

Address	7	6	5	4	3	2	1	0
04h	UTXE0	URXE0 USPIE0						
	rw-0	rw-0						

URXE0: USART0: UART mode receive enable UTXE0: USART0: UART mode transmit enable

USPIE0: USART0: SPI mode transmit and receive enable

Address 7 6 5 4 3 2 1 0 05h

Legend: rw-0,1: rw-(0,1): Bit Can Be Read and Written. It Is Reset or Set by PUC. Bit Can Be Read and Written. It Is Reset or Set by POR. SFR Bit Not Present in Device.

### memory organization

		MSP430F423	MSP430F425	MSP430F427
Memory Interrupt vector Code memory	Size Flash Flash	8KB 0FFFFh – 0FFE0h 0FFFFh – 0E000h	16KB 0FFFFh – 0FFE0h 0FFFFh – 0C000h	32KB 0FFFFh – 0FFE0h 0FFFFh – 08000h
Information memory	Size	256 Byte 010FFh – 01000h	256 Byte 010FFh – 01000h	256 Byte 010FFh – 01000h
Boot memory	Size	1kB 0FFFh – 0C00h	1kB 0FFFh – 0C00h	1kB 0FFFh – 0C00h
RAM	Size	256 Byte 02FFh – 0200h	512 Byte 03FFh – 0200h	1KB 05FFh – 0200h
Peripherals	16-bit 8-bit 8-bit SFR	01FFh – 0100h 0FFh – 010h 0Fh – 00h	01FFh – 0100h 0FFh – 010h 0Fh – 00h	01FFh – 0100h 0FFh – 010h 0Fh – 00h

#### bootstrap loader (BSL)

The MSP430 bootstrap loader (BSL) enables users to program the flash memory or RAM using a UART serial interface. Access to the MSP430 memory via the BSL is protected by user-defined password. For complete description of the features of the BSL and its implementation, see the Application report *Features of the MSP430 Bootstrap Loader*, Literature Number SLAA089.

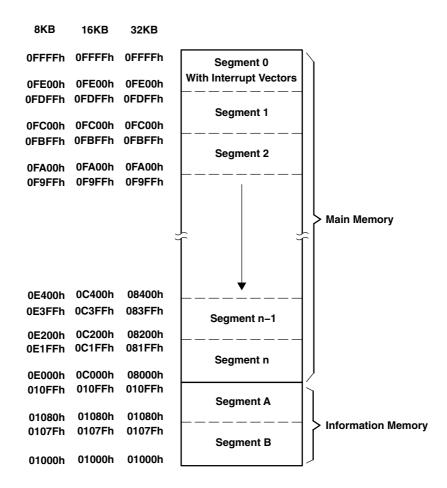
BSL Function	PM Package Pins
Data Transmit	53 - P1.0
Data Receive	52 - P1.1



#### flash memory

The flash memory can be programmed via the JTAG port, the bootstrap loader, or in-system by the CPU. The CPU can perform single-byte and single-word writes to the flash memory. Features of the flash memory include:

- Flash memory has n segments of main memory and two segments of information memory (A and B) of 128 bytes each. Each segment in main memory is 512 bytes in size.
- Segments 0 to n may be erased in one step, or each segment may be individually erased.
- Segments A and B can be erased individually, or as a group with segments 0-n.
   Segments A and B are also called information memory.
- New devices may have some bytes programmed in the information memory (needed for test during manufacturing). The user should perform an erase of the information memory prior to the first use.



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#### peripherals

Peripherals are connected to the CPU through data, address, and control busses and can be handled using all instructions. For complete module descriptions, see the *MSP430x4xx Family User's Guide*, TI literature number SLAU056.

#### oscillator and system clock

The clock system in the MSP430F42x family of devices is supported by the FLL+ module that includes support for a 32768-Hz watch crystal oscillator, an internal digitally-controlled oscillator (DCO), and a high frequency crystal oscillator. The FLL+ clock module is designed to meet the requirements of both low system cost and low power consumption. The FLL+ features digital frequency locked loop (FLL) hardware that, in conjunction with a digital modulator, stabilizes the DCO frequency to a programmable multiple of the watch crystal frequency. The internal DCO provides a fast turn-on clock source and stabilizes in less than 6 μs. The FLL+ module provides the following clock signals:

- Auxiliary clock (ACLK), sourced from a 32768-Hz watch crystal or a high frequency crystal.
- Main clock (MCLK), the system clock used by the CPU.
- Sub-Main clock (SMCLK), the sub-system clock used by the peripheral modules.
- ACLK/n, the buffered output of ACLK, ACLK/2, ACLK/4, or ACLK/8.

#### brownout, supply voltage supervisor

The brownout circuit is implemented to provide the proper internal reset signal to the device during power on and power off. The supply voltage supervisor (SVS) circuitry detects if the supply voltage drops below a user selectable level and supports both supply voltage supervision (the device is automatically reset) and supply voltage monitoring (SVM, the device is not automatically reset).

The CPU begins code execution after the brownout circuit releases the device reset. However,  $V_{CC}$  may not have ramped to  $V_{CC(min)}$  at that time. The user must insure the default FLL+ settings are not changed until  $V_{CC}$  reaches  $V_{CC(min)}$ . If desired, the SVS circuit can be used to determine when  $V_{CC}$  reaches  $V_{CC(min)}$ .

#### digital I/O

There are two 8-bit I/O ports implemented—ports P1 and P2 (only six P2 I/O signals are available on external pins):

- All individual I/O bits are independently programmable.
- Any combination of input, output, and interrupt conditions is possible.
- Edge-selectable interrupt input capability for all the eight bits of port P1 and six bits of P2.
- Read/write access to port-control registers is supported by all instructions.

#### NOTE

Six bits of port P2 (P2.0 to P2.5) are available on external pins, but all control and data bits for port P2 are implemented.

#### **Basic Timer1**

The Basic Timer1 has two independent 8-bit timers which can be cascaded to form a 16-bit timer/counter. Both timers can be read and written by software. The Basic Timer1 can be used to generate periodic interrupts and clock for the LCD module.

#### **LCD** drive

The LCD driver generates the segment and common signals required to drive an LCD display. The LCD controller has dedicated data memory to hold segment drive information. Common and segment signals are generated as defined by the mode. Static, 2-MUX, 3-MUX, and 4-MUX LCDs are supported by this peripheral.



#### WDT+ watchdog timer

The primary function of the watchdog timer (WDT+) module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be configured as an interval timer and can generate interrupts at selected time intervals.

#### timer A3

Timer\_A3 is a 16-bit timer/counter with three capture/compare registers. Timer\_A3 can support multiple capture/compares, PWM outputs, and interval timing. Timer\_A3 also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

	Timer_A3 Signal Connections					
Input Pin Number	Device Input Signal	Module Input Name	Module Block	Module Output Signal	Output Pin Number	
48 - P1.5	TACLK	TACLK				
	ACLK	ACLK	-	NA		
	SMCLK	SMCLK	Timer			
48 - P1.5	TACLK	INCLK				
53 - P1.0	TA0	CCI0A			53 - P1.0	
52 - P1.1	TA0	CCI0B	0000	T4.0		
	DV <sub>SS</sub>	GND	CCR0	TA0		
	DV <sub>CC</sub>	V <sub>CC</sub>				
51 - P1.2	TA1	CCI1A			51 - P1.2	
51 - P1.2	TA1	CCI1B	0004	<b>T</b>		
	DV <sub>SS</sub>	GND	CCR1	TA1		
	DV <sub>CC</sub>	V <sub>CC</sub>				
45 - P2.0	TA2	CCI2A			45 - P2.0	
	ACLK (internal)	CCI2B	0000	TA2		
	DV <sub>SS</sub>	GND	CCR2			
	DV <sub>CC</sub>	V <sub>CC</sub>				

#### **USARTO**

The MSP430F42x devices have one hardware universal synchronous/asynchronous receive transmit (USART0) peripheral module that is used for serial data communication. The USART supports synchronous SPI (3 or 4 pin) and asynchronous UART communication protocols, using double-buffered transmit and receive channels.

#### hardware multiplier

The multiplication operation is supported by a dedicated peripheral module. The module performs  $16 \times 16$ ,  $16 \times 8$ ,  $8 \times 16$ , and  $8 \times 8$  bit operations. The module is capable of supporting signed and unsigned multiplication as well as signed and unsigned multiply and accumulate operations. The result of an operation can be accessed immediately after the operands have been loaded into the peripheral registers. No additional clock cycles are required.

#### **SD16**

The SD16 module integrates three independent 16-bit sigma-delta A/D converters, internal temperature sensor and built-in voltage reference. Each channel is designed with a fully differential analog input pair and programmable gain amplifier input stage.

## peripheral file map

	PERIPHERALS WITH WORD ACCESS		
Watchdog	Watchdog Timer control	WDTCTL	0120h
Timer_A3	Timer_A interrupt vector	TAIV	012Eh
	Timer_A control	TACTL	0160h
	Capture/compare control 0	TACCTL0	0162h
	Capture/compare control 1	TACCTL1	0164h
	Capture/compare control 2	TACCTL2	0166h
	Timer_A register	TAR	0170h
	Capture/compare register 0	TACCR0	0172h
	Capture/compare register 1	TACCR1	0174h
	Capture/compare register 2	TACCR2	0176h
Hardware Multiplier	Sum extend	SUMEXT	013Eh
	Result high word	RESHI	013Ch
	Result low word	RESLO	013Ah
	Second operand	OP2	0138h
	Multiply signed + accumulate/operand1	MACS	0136h
	Multiply + accumulate/operand1	MAC	0134h
	Multiply signed/operand1	MPYS	0132h
	Multiply unsigned/operand1	MPY	0130h
Flash	Flash control 3	FCTL3	012Ch
	Flash control 2	FCTL2	012Ah
	Flash control 1	FCTL1	0128h
SD16	General Control	SD16CTL	0100h
(see also: Peripherals with Byte Access)	Channel 0 Control	SD16CCTL0	0102h
Willi Byte Access)	Channel 1 Control	SD16CCTL1	0104h
	Channel 2 Control	SD16CCTL2	0106h
	Reserved		0108h
	Reserved		010Ah
	Reserved		010Ch
	Reserved		010Eh
	Interrupt vector word register	SD16IV	0110h
	Channel 0 conversion memory	SD16MEM0	0112h
	Channel 1 conversion memory	SD16MEM1	0114h
	Channel 2 conversion memory	SD16MEM2	0116h
	Reserved		0118h
	Reserved		011Ah
	Reserved		011Ch
	Reserved		011Eh



# peripheral file map (continued)

	PERIPHERALS WITH BYTE ACCESS		
SD16	Channel 0 Input Control	SD16INCTL0	0B0h
(see also: Peripherals	Channel 1 Input Control	SD16INCTL1	0B1h
with Word Access)	Channel 2 Input Control	SD16INCTL2	0B2h
	Reserved		0B3h
	Reserved		0B4h
	Reserved		0B5h
	Reserved		0B6h
	Reserved		0B7h
	Channel 0 preload	SD16PRE0	0B8h
	Channel 1 preload	SD16PRE1	0B9h
	Channel 2 preload	SD16PRE2	0BAh
	Reserved		0BBh
	Reserved		0BCh
	Reserved		0BDh
	Reserved		0BEh
	Reserved		0BFh
LCD	LCD memory 20	LCDM20	0A4h
	:	:	:
	LCD memory 16	LCDM16	0A0h
	LCD memory 15	LCDM15	09Fh
	:	:	:
	LCD memory 1	LCDM1	091h
	LCD control and mode	LCDCTL	090h
USART0	Transmit buffer	U0TXBUF	077h
	Receive buffer	U0RXBUF	076h
	Baud rate	U0BR1	075h
	Baud rate	U0BR0	074h
	Modulation control	U0MCTL U0RCTL	073h
	Receive control Transmit control	UOTCTL	072h 071h
	USART control	U0CTL	07111 070h
Brownout, SVS	SVS control register	SVSCTL	056h
FLL+ Clock	FLL+ Control1	FLL_CTL1	054h
	FLL+ Control0	FLL_CTL0	053h
	System clock frequency control	SCFQCTL	052h
	System clock frequency integrator	SCFI1	051h
	System clock frequency integrator	SCFI0	050h
Basic Timer1	BT counter2	BTCNT2	047h
	BT counter1	BTCNT1	046h
	BT control	BTCTL	040h



## peripheral file map (continued)

	PERIPHERALS WITH BYTE ACCESS (C	CONTINUED)	
Port P2	Port P2 selection	P2SEL	02Eh
	Port P2 interrupt enable	P2IE	02Dh
	Port P2 interrupt-edge select	P2IES	02Ch
	Port P2 interrupt flag	P2IFG	02Bh
	Port P2 direction	P2DIR	02Ah
	Port P2 output	P2OUT	029h
	Port P2 input	P2IN	028h
Port P1	Port P1 selection	P1SEL	026h
	Port P1 interrupt enable	P1IE	025h
	Port P1 interrupt-edge select	P1IES	024h
	Port P1 interrupt flag	P1IFG	023h
	Port P1 direction	P1DIR	022h
	Port P1 output	P1OUT	021h
	Port P1 input	P1IN	020h
Special Functions	SFR module enable 2	ME2	005h
	SFR module enable 1	ME1	004h
	SFR interrupt flag 2	IFG2	003h
	SFR interrupt flag 1	IFG1	002h
	SFR interrupt enable2	IE2	001h
	SFR interrupt enable1	IE1	000h

#### absolute maximum ratings†

Voltage applied at V <sub>CC</sub> to V <sub>SS</sub>	–0.3 V to + 4.1 V
Voltage applied to any pin (see Note 1)	0.3 V to V <sub>CC</sub> + 0.3 V
Diode current at any device terminal	±2 mA
Storage temperature (unprogrammed device)	–55°C to 150°C
Storage temperature (programmed device)	–40°C to 85°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

#### recommended operating conditions

PARA	AMETER		MIN	NOM	MAX	UNITS
Supply voltage during program execution; SD16 $V_{CC}$ (AV $_{CC}$ = DV $_{CC}$ = V $_{CC}$ ) (see Note 1)	MSP430F42x	1.8		3.6	٧	
Supply voltage during program execution; SD16 PORON = 1. $V_{CC}$ (AV $_{CC}$ = DV $_{CC}$ = $V_{CC}$ ) (see N	MSP430F42x	2.0		3.6	V	
Supply voltage during program execution; SD16 during programming of flash memory. $V_{\rm CC}$ (AV $_{\rm C}$	MSP430F42x	2.7		3.6	V	
Supply voltage, $V_{SS}$ (AV <sub>SS</sub> = DV <sub>SS</sub> = V <sub>SS</sub> )		0		0	V	
Operating free-air temperature range, TA		MSP430F42x	-40		85	°C
	LF selected, XTS_FLL=0	Watch crystal		32768		Hz
LFXT1 crystal frequency, f <sub>(LFXT1)</sub> (see Note 3)	XT1 selected, XTS_FLL=1	Ceramic resonator	450		8000	kHz
,	XT1 selected, XTS_FLL=1	Crystal	1000		8000	kHz
	V <sub>CC</sub> = 1.8 V	DC		4.15		
Processor frequency (signal MCLK), f <sub>(System)</sub>	V <sub>CC</sub> = 3.6 V	DC		8	MHz	

- NOTES: 1. It is recommended to power AV<sub>CC</sub> and DV<sub>CC</sub> from the same source. A maximum difference of 0.3 V between AV<sub>CC</sub> and DV<sub>CC</sub> can be tolerated during power up and operation.
  - 2. The minimum operating supply voltage is defined according to the trip point where POR is going active by decreasing supply voltage. POR is going inactive when the supply voltage is raised above minimum supply voltage plus the hysteresis of the SVS circuitry.
  - 3. The LFXT1 oscillator in LF-mode requires a watch crystal.

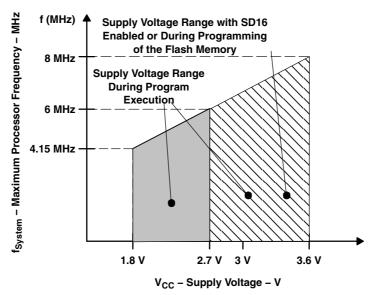


Figure 1. Frequency vs Supply Voltage



NOTE 1: All voltages referenced to V<sub>SS</sub>. The JTAG fuse-blow voltage, V<sub>FB</sub>, is allowed to exceed the absolute maximum rating. The voltage is applied to the TDI/TCLK pin when blowing the JTAG fuse.

# electrical characteristics over recommended operating free-air temperature (unless otherwise noted)

### supply current into AV<sub>CC</sub> + DV<sub>CC</sub> excluding external current (see Note 1)

	PARAMETER	TEST CON	DITIONS	MIN	NOM	MAX	UNIT
I <sub>(AM)</sub>	Active mode, $f_{(MCLK)} = f_{(SMCLK)} = f_{(DCO)} = 1 \text{ MHz},$ $f_{(ACLK)} = 32,768 \text{ Hz}, \text{ XTS\_FLL} = 0$ (program executes in flash)	$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	V <sub>CC</sub> = 3 V		400	500	μΑ
I <sub>(LPM0)</sub>	Low-power mode, (LPM0/LPM1) $f_{(MCLK)} = f_{(SMCLK)} = f_{(DCO)} = 1 \text{ MHz},$ $f_{(ACLK)} = 32,768 \text{ Hz}, XTS\_FLL = 0$ $FN_8 = FN_4 = FN_3 = FN_2 = 0 \text{ (see Note 2)}$	$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}$	V <sub>CC</sub> = 3 V		130	150	μΑ
I <sub>(LPM2)</sub>	Low-power mode, (LPM2) (see Note 2)	$T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$	V <sub>CC</sub> = 3 V		10	22	μΑ
		T <sub>A</sub> = -40°C			1.5	2.0	
١.	Low request mode (LDMO) (one Note O)	T <sub>A</sub> = 25°C	., .,		1.6	2.1	
I <sub>(LPM3)</sub>	Low-power mode, (LPM3) (see Note 2)	T <sub>A</sub> = 60°C	V <sub>CC</sub> = 3 V		1.7	2.2	μ <b>A</b>
		T <sub>A</sub> = 85°C			2.0	2.0 2.6	
		$T_A = -40^{\circ}C$		·	0.1	0.5	
I <sub>(LPM4)</sub>	Low-power mode, (LPM4) (see Note 2)	T <sub>A</sub> = 25°C	V <sub>CC</sub> = 3 V		0.1	0.5	μΑ
` ′		T <sub>A</sub> = 85°C			0.8	2.5	

NOTES: 1. All inputs are tied to 0 V or V<sub>CC</sub>. Outputs do not source or sink any current.

The current consumption in LPM2, LPM3, and LPM4 are measured with active Basic Timer1 and LCD (ACLK selected).

The current consumption of the SD16 and the SVS module are specified in their respective sections.

LPMx currents measured with WDT disabled.

The currents are characterized with a KDS Daishinku DT-38 (6 pF) crystal.

2. Current for brownout included.

### current consumption of active mode versus system frequency

$$I_{(AM)} = I_{(AM)} [1 \text{ MHz}] \times f_{(System)} [MHz]$$

#### current consumption of active mode versus supply voltage

$$I_{(AM)} = I_{(AM)[3\ V]} + 170\ \mu A/V \times (V_{CC} - 3\ V)$$



#### Schmitt-trigger inputs – Ports P1 and P2; RST/NMI; JTAG: TCK, TMS, TDI/TCLK, TDO/TDI

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
$V_{IT+}$	Positive-going input threshold voltage	V <sub>CC</sub> = 3 V	1.5	1.98	V
$V_{IT-}$	Negative-going input threshold voltage	V <sub>CC</sub> = 3 V	0.9	1.3	V
$V_{hys}$	Input voltage hysteresis (V <sub>IT+</sub> – V <sub>IT-</sub> )	$V_{CC} = 3 V$	0.45	1	V

#### inputs Px.x, TAx

PARAMETER		TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
t <sub>(int)</sub> External interrupt timing		Port P1, P2: P1.x to P2.x, External trigger signal	3 V	1.5			cycle
		for the interrupt flag, (see Note 1)	3 V	50			ns
t <sub>(cap)</sub>	Timer_A, capture timing	TAx	3 V	50			ns
f <sub>(TAext)</sub>	Timer_A clock frequency externally applied to pin	TACLK, INCLK $t_{(H)} = t_{(L)}$	3 V			10	MHz
f <sub>(TAint)</sub>	Timer_A clock frequency	SMCLK or ACLK signal selected	3 V			10	MHz

NOTES: 1. The external signal sets the interrupt flag every time the minimum  $t_{(int)}$  cycle and time parameters are met. It may be set even with trigger signals shorter than  $t_{(int)}$ . Both the cycle and timing specifications must be met to ensure the flag is set.  $t_{(int)}$  is measured in MCLK cycles.

#### leakage current (see Note 1)

	PARAMETER		TEST CONDITIONS			NOM	MAX	UNIT
I <sub>lkg(P1.x)</sub>	Leakage	Port P1	Port 1: V <sub>(P1.x)</sub> (see Note 2)	V - 2 V			±50	nA
I <sub>lkg(P2.x)</sub>	current	Port P2	Port 2: V <sub>(P2.x)</sub> (see Note 2)	$V_{CC} = 3 V$			±50	IIA

NOTES: 1. The leakage current is measured with VSS or VCC applied to the corresponding pin(s), unless otherwise noted.

2. The port pin must be selected as an input.

#### outputs - Ports P1 and P2

	PARAMETER	TEST	TEST CONDITIONS			TYP MAX	UNIT
	High lavel autout valtage	$I_{OH(max)} = -1.5 \text{ mA},$	$V_{CC} = 3 V$ ,	See Note 1	V <sub>CC</sub> -0.25	$V_{CC}$	V
VOH	V <sub>OH</sub> High-level output voltage	$I_{OH(max)} = -6 \text{ mA},$	$V_{CC} = 3 V$ ,	See Note 2	V <sub>CC</sub> -0.6	$V_{CC}$	V
V	Low-level output voltage	$I_{OL(max)} = 1.5 \text{ mA},$	$V_{CC} = 3 V$ ,	See Note 1	$V_{SS}$	V <sub>SS</sub> +0.25	V
$V_{OL}$	Low-level output voltage	$I_{OL(max)} = 6 \text{ mA},$	$V_{CC} = 3 V$ ,	See Note 2	$V_{SS}$	V <sub>SS</sub> +0.6	V

- NOTES: 1. The maximum total current, I<sub>OH(max)</sub> and I<sub>OL(max)</sub>, for all outputs combined, should not exceed ±12 mA to satisfy the maximum specified voltage drop.
  - 2. The maximum total current, I<sub>OH(max)</sub> and I<sub>OL(max)</sub>, for all outputs combined, should not exceed ±48 mA to satisfy the maximum specified voltage drop.

#### output frequency

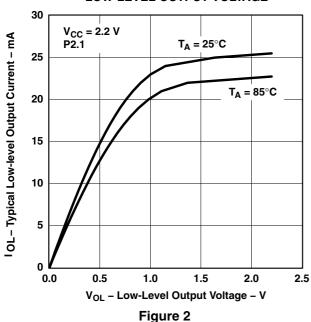
	PARAMETER	TEST C	TEST CONDITIONS		TYP	MAX	UNIT
f <sub>Px.y</sub>	$(1 \le x \le 2, \ 0 \le y \le 7)$	$C_L = 20 \text{ pF},$ $I_L = \pm 1.5 \text{mA}$	V <sub>CC</sub> = 3 V	DC		12	MHz
f <sub>ACLK</sub> , f <sub>MCLK</sub> , f <sub>SMCLK</sub>	P1.1/TA0/MCLK P1.5/TACLK/ACLK/S28	C <sub>L</sub> = 20 pF	V <sub>CC</sub> = 3 V			12	MHz
		P1.5/TACLK/ACLK/ S28, C <sub>L</sub> = 20 pF	$f_{ACLK} = f_{LFXT1} = f_{XT1}$	40%		60%	
			$f_{ACLK} = f_{LFXT1} = f_{LF}$	30%		70%	
t <sub>Xdc</sub>	Duty cycle of output frequency	V <sub>CC</sub> = 3 V	f <sub>ACLK</sub> = f <sub>LFXT1</sub>		50%		
7,00	Daily Gyolo of Galpat Hoquology	$\begin{aligned} &\text{P1.1/TA0/MCLK,}\\ &\text{C}_{\text{L}} = 20 \text{ pF,}\\ &\text{V}_{\text{CC}} = 3 \text{ V} \end{aligned}$	f <sub>MCLK</sub> = f <sub>DCOCLK</sub>	50%– 15 ns	50%	50%+ 15 ns	



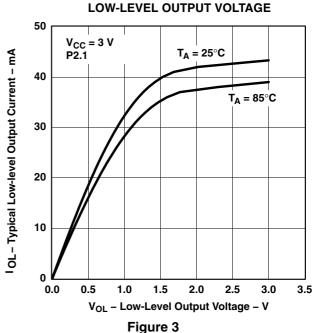
electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

outputs - Ports P1 and P2 (continued)

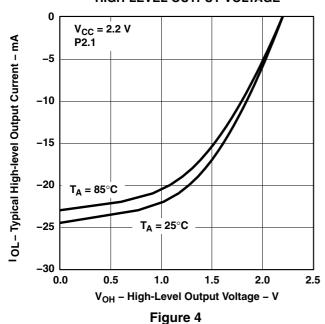
# TYPICAL LOW-LEVEL OUTPUT CURRENT vs LOW-LEVEL OUTPUT VOLTAGE



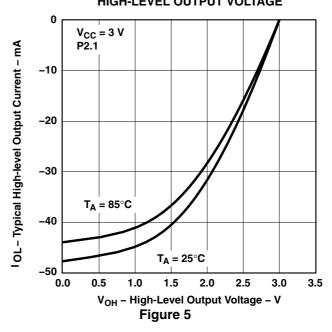
# TYPICAL LOW-LEVEL OUTPUT CURRENT vs



# TYPICAL HIGH-LEVEL OUTPUT CURRENT vs HIGH-LEVEL OUTPUT VOLTAGE



# TYPICAL HIGH-LEVEL OUTPUT CURRENT vs HIGH-LEVEL OUTPUT VOLTAGE



NOTE: One output loaded at a time

# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

#### wake-up LPM3

PARAMETER		TEST (	CONDITIONS	MIN	TYP	MAX	UNIT
		f = 1 MHz				6	
t <sub>d(LPM3)</sub> Delay time	Delay time	f = 2 MHz	V <sub>CC</sub> = 3 V			6	μs
` ′		f = 3 MHz				6	

#### RAM (see Note 1)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VRAMh	CPU halted (see Note 1)	1.6			V

NOTE 1: This parameter defines the minimum supply voltage when the data in the program memory RAM remain unchanged. No program execution should take place during this supply voltage condition.

#### LCD

PARA	AMETER	TEST CON	DITIONS	MIN	TYP	MAX	UNIT
V <sub>(33)</sub>		Voltage at R33		2.5		V <sub>CC</sub> +0.2	
V <sub>(23)</sub>	7,	Voltage at R23	ן,, ביי [	('	$V_{33} - V_{03}) \times 2/3 +$	V <sub>03</sub>	١.,
V <sub>(13)</sub>	Analog voltage	Voltage at R13	$V_{CC} = 3 V$	(V <sub>(</sub>	<sub>33)</sub> -V <sub>(03)</sub> ) × 1/3 +	· V <sub>(03)</sub>	٧
V <sub>(33)</sub> – V <sub>(03)</sub>	1	Voltage at R33/R03		2.5		V <sub>CC</sub> +0.2	
I <sub>(R03)</sub>		R03 = V <sub>SS</sub>	No load at all			±20	
I <sub>(R13)</sub>	Input leakage	R13 = V <sub>CC</sub> /3	segment and common lines,			±20	nA
I <sub>(R23)</sub>	1	$R23 = 2 \times V_{CC}/3$	$V_{CC} = 3 \text{ V}$			±20	
V <sub>(Sxx0)</sub>				V <sub>(03)</sub>		V <sub>(03)</sub> – 0.1	
V <sub>(Sxx1)</sub>	Segment line		], , , [	V <sub>(13)</sub>		V <sub>(13)</sub> – 0.1	١,,
V <sub>(Sxx2)</sub>	voltage	$I_{(Sxx)} = -3 \mu A,$	$V_{CC} = 3 V$	V( <sub>23)</sub>		$V_{(23)} - 0.1$	V
V <sub>(Sxx3)</sub>				V( <sub>33)</sub>		$V_{(33)} + 0.1$	

#### **USART0** (see Note 1)

	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
$t_{(\tau)}$	USART0: deglitch time	V <sub>CC</sub> = 3 V, SYNC = 0, UART mode	150	280	500	ns

NOTE 1: The signal applied to the USART0 receive signal/terminal (URXD0) should meet the timing requirements of  $t_{(\tau)}$  to ensure that the URXS flip-flop is set. The URXS flip-flop is set with negative pulses meeting the minimum-timing condition of  $t_{(\tau)}$ . The operating conditions to set the flag must be met independently from this timing constraint. The deglitch circuitry is active only on negative transitions on the URXD0 line.

#### POR brownout, reset (see Notes 1 and 2)

PARAI	METER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>d(BOR)</sub>					2000	μs
V <sub>CC(start)</sub>		dV <sub>CC</sub> /dt ≤ 3 V/s (see Figure 6)		$0.7 \times V_{(B\_IT-)}$		V
V <sub>(B_IT-)</sub>	Brownout	dV <sub>CC</sub> /dt ≤ 3 V/s (see Figure 6, Figure 7, and Figure 8)			1.71	V
V <sub>hys(B_IT-)</sub>	Biownout	dV <sub>CC</sub> /dt ≤ 3 V/s (see Figure 6)	70	130	180	mV
t <sub>(reset)</sub>		Pulse length needed at $\overline{\text{RST/NMI}}$ pin to accepted reset internally, $V_{CC} = 3 \text{ V}$	2			μs

NOTES: 1. The current consumption of the brownout module is already included in the  $I_{CC}$  current consumption data. The voltage level  $V_{(B_-|T_-)} + V_{hys(B_-|T_-)}$  is  $\leq 1.8$  V.

During power up, the CPU begins code execution following a period of t<sub>d(BOR)</sub> after V<sub>CC</sub> = V<sub>(B\_IT-)</sub> + V<sub>hys(B\_IT-)</sub>.
 The default FLL+ settings must not be changed until V<sub>CC</sub> ≥ V<sub>CC(min)</sub>, where V<sub>CC(min)</sub> is the minimum supply voltage for the desired operating frequency. See the MSP430x4xx Family User's Guide (SLAU056) for more information on the brownout/SVS circuit.



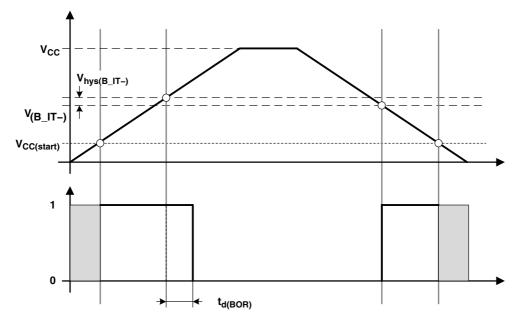


Figure 6. POR/Brownout Reset (BOR) vs Supply Voltage

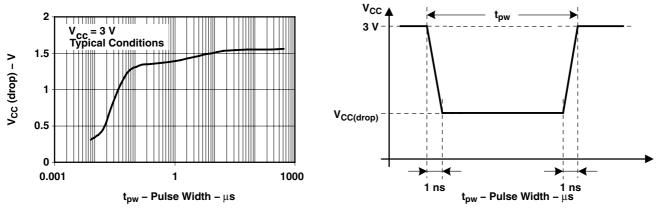


Figure 7. V<sub>CC(drop)</sub> Level With a Square Voltage Drop to Generate a POR/Brownout Signal

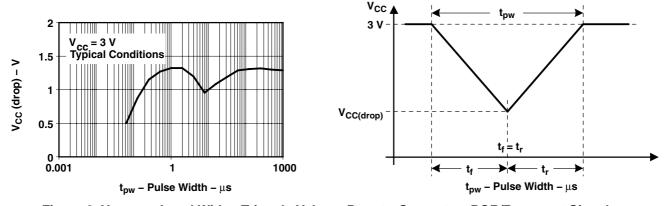


Figure 8.  $V_{CC(drop)}$  Level With a Triangle Voltage Drop to Generate a POR/Brownout Signal



# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

SVS (supply voltage supervisor/monitor) (see Note 1)

PARAMETER	TEST CONDITIONS		MIN	NOM	MAX	UNIT
	dV <sub>CC</sub> /dt > 30 V/ms (see Figure 9)		5		150	
<sup>T</sup> (SVSR)4	dV <sub>CC</sub> /dt ≤ 30 V/ms				2000	μs
t <sub>d(SVSon)</sub>	SVSon, switch from VLD=0 to VLD ≠ 0, V <sub>CC</sub> = 3 V		20		150	μs
t <sub>settle</sub>	VLD ≠ 0 <sup>‡</sup>				12	μs
	VLD ≠ 0, V <sub>CC</sub> /dt ≤ 3 V/s (see Figure 9)			1.55	1.7	V
		VLD = 1	70	120	155	mV
V <sub>hvs(SVS IT-)</sub>	V <sub>CC</sub> /dt ≤ 3 V/s (see Figure 9)	VLD = 2 14	V <sub>(SVS_IT-)</sub> x 0.004		V <sub>(SVS_IT-)</sub> x 0.008	
d(SVSR)4  d(SVSon)  settle  V(SVSstart)  Vhys(SVS_IT-)  CC(SVS)	$V_{CC}/dt \le 3$ V/s (see Figure 9), external voltage applied on P2.3	4.4		10.4	mV	
		VLD = 1	1.8	1.9	2.05	
		VLD = 2	1.94	2.1	2.25	1
		VLD = 3	2.05	2.2	2.37	1
		VLD = 4	2.14	2.3	2.48	1
		VLD = 5	2.24	2.4	2.6	1
		VLD = 6	2.33	2.5	2.71	1
tsettle V(SVSstart)  Vhys(SVS_IT-)	VLD =		2.46	2.65	2.86	
V	VCC/dt ≥ 3 V/s (see Figure 9)	VLD = 8	2.58	2.8	3	] <sub>v</sub>
V(SVS_II-)		VLD = 9	2.69	2.9	3.13	]
	VSR)4	2.83	3.05	3.29		
		VLD = 11	2.94	3.2	3.42	
		VLD = 12	3.11	3.35	3.61 <sup>†</sup>	
		VLD = 13	3.24	3.5	3.76 <sup>†</sup>	
		VLD = 14	3.43	3.7†	3.99†	
		VLD = 15	1.1	1.2	1.3	
I <sub>CC(SVS)</sub> (see Note 1)	VLD ≠ 0, V <sub>CC</sub> = 2.2 V/3 V		_	10	15	μА

<sup>&</sup>lt;sup>†</sup> The recommended operating voltage range is limited to 3.6 V.

NOTE 1: The current consumption of the SVS module is not included in the  $I_{CC}$  current consumption data.

<sup>&</sup>lt;sup>‡</sup> t<sub>settle</sub> is the settling time that the comparator o/p needs to have a stable level after VLD is switched VLD ≠ 0 to a different VLD value somewhere between 2 and 15. The overdrive is assumed to be > 50 mV.

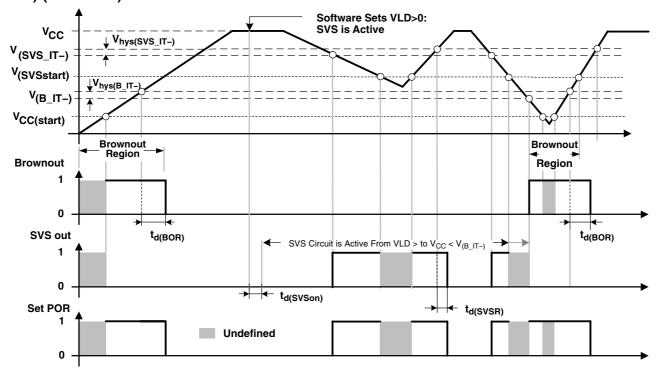


Figure 9. SVS Reset (SVSR) vs Supply Voltage

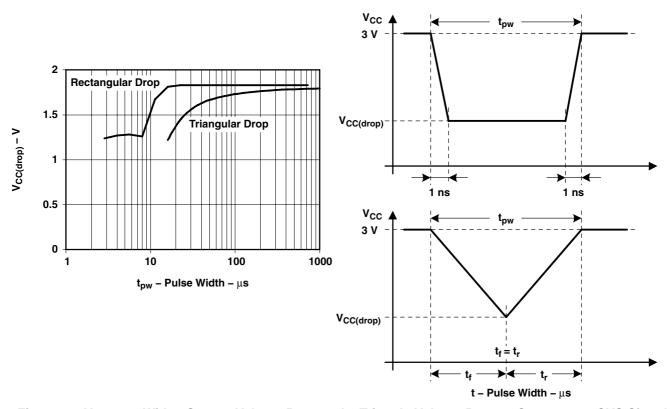


Figure 10. V<sub>CC(drop)</sub> With a Square Voltage Drop and a Triangle Voltage Drop to Generate an SVS Signal



#### DCO

PARAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>(DCOCLK)</sub>	$N_{(DCO)}$ =01Eh, FN_8=FN_4=FN_3=FN_2=0, D = 2; DCOPLUS= 0, $f_{Crystal}$ = 32.768 kHz	3 V		1		MHz
f <sub>(DCO=2)</sub>	FN_8=FN_4=FN_3=FN_2=0 ; DCOPLUS = 1	3 V	0.3	0.7	1.3	MHz
f <sub>(DCO=27)</sub>	FN_8=FN_4=FN_3=FN_2=0; DCOPLUS = 1	3 V	2.7	6.1	11.3	MHz
f <sub>(DCO=2)</sub>	FN_8=FN_4=FN_3=0, FN_2=1; DCOPLUS = 1	3 V	0.8	1.5	2.5	MHz
f <sub>(DCO=27)</sub>	FN_8=FN_4=FN_3=0, FN_2=1; DCOPLUS = 1	3 V	6.5	12.1	20	MHz
f <sub>(DCO=2)</sub>	FN_8=FN_4=0, FN_3= 1, FN_2=x; DCOPLUS = 1	3 V	1.3	2.2	3.5	MHz
f <sub>(DCO=27)</sub>	FN_8=FN_4=0, FN_3= 1, FN_2=x; DCOPLUS = 1	3 V	10.3	17.9	28.5	MHz
f <sub>(DCO=2)</sub>	FN_8=0, FN_4= 1, FN_3= FN_2=x; DCOPLUS = 1	3 V	2.1	3.4	5.2	MHz
f <sub>(DCO=27)</sub>	FN_8=0, FN_4=1, FN_3= FN_2=x; DCOPLUS = 1	3 V	16	26.6	41	MHz
f <sub>(DCO=2)</sub>	FN_8=1, FN_4=FN_3=FN_2=x; DCOPLUS = 1	3 V	4.2	6.3	9.2	MHz
f <sub>(DCO=27)</sub>	FN_8=1,FN_4=FN_3=FN_2=x; DCOPLUS = 1	3 V	30	46	70	MHz
0	Step size between adjacent DCO taps:	1 < TAP ≤ 20	1.06		1.11	
S <sub>n</sub>	$S_n = f_{DCO(Tap n+1)} / f_{DCO(Tap n)}$ , (see Figure 12 for taps 21 to 27)	TAP = 27	1.07		1.17	
D <sub>t</sub>	Temperature drift, N <sub>(DCO)</sub> = 01Eh, FN_8=FN_4=FN_3=FN_2=0 D = 2; DCOPLUS = 0	3 V	-0.2	-0.3	-0.4	%/°C
D <sub>V</sub>	Drift with $V_{CC}$ variation, $N_{(DCO)} = 01Eh$ , $FN_8 = FN_4 = FN_3 = FN_2 = 0$ D = 2; DCOPLUS = 0		0	5	15	%/V

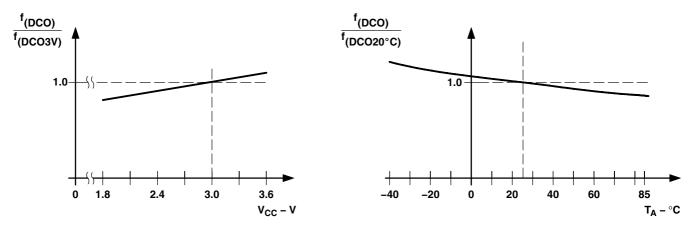


Figure 11. DCO Frequency vs Supply Voltage  $V_{\text{CC}}$  and vs Ambient Temperature

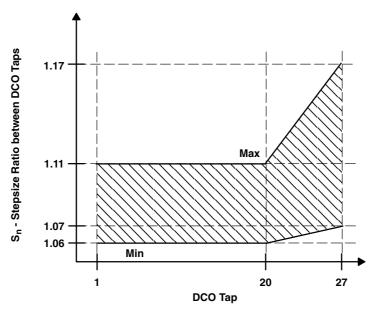


Figure 12. DCO Tap Step Size

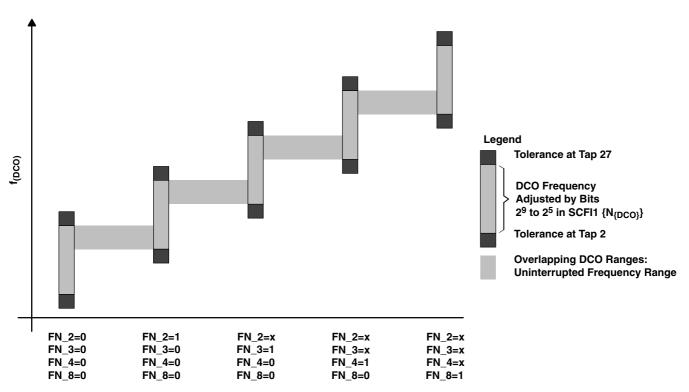


Figure 13. Five Overlapping DCO Ranges Controlled by FN\_x Bits



# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

#### crystal oscillator, LFXT1 oscillator (see Notes 1 and 2)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
		OSCCAPx = 0h	3 V		0		
	Integrated input capacitance	OSCCAPx = 1h	3 V		10		_
C <sub>XIN</sub>	(see Note 4)	OSCCAPx = 2h	3 V		14		pF
		OSCCAPx = 3h	3 V		18		
		OSCCAPx = 0h	3 V		0		
	Integrated output capacitance	OSCCAPx = 1h	3 V		10		_
C <sub>XOUT</sub>	(see Note 4)	OSCCAPx = 2h	3 V		14		pF
		OSCCAPx = 3h	3 V		18		
V <sub>IL</sub>	Installation of MINI	and Nation	0.1/	V <sub>SS</sub>		0.2×V <sub>CC</sub>	.,
V <sub>IH</sub>	Input levels at XIN	see Note 3	3 V	0.8×V <sub>CC</sub>		V <sub>CC</sub>	V

- NOTES: 1. The parasitic capacitance from the package and board may be estimated to be 2pF. The effective load capacitor for the crystal is (C<sub>XIN</sub> x C<sub>XOUT</sub>) / (C<sub>XIN</sub> + C<sub>XOUT</sub>). It is independent of XTS\_FLL.
  - 2. To improve EMI on the low-power LFXT1 oscillator, particularly in the LF mode (32 kHz), the following guidelines must be observed:
    - Keep as short a trace as possible between the 'F42x and the crystal.
    - Design a good ground plane around oscillator pins.
    - Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
    - Avoid running PCB traces underneath or adjacent to XIN an XOUT pins.
    - Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.
    - If conformal coating is used, ensure that it does not induce capacitive/resistive leakage between the oscillator pins.
    - Do not route the XOUT line to the JTAG header to support the serial programming adapter as shown in other documentation. This signal is no longer required for the serial programming adapter.
  - 3. Applies only when using an external logic-level clock source. XTS\_FLL must be set. Not applicable when using a crystal or resonator.
  - 4. External capacitance is recommended for precision real-time clock applications; OSCCAPx = 0h.



# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

### SD16, power supply and recommended operating conditions

	PARAMETER		TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
AV <sub>CC</sub>	Analog supply voltage	$AV_{CC} = DV_{CC}$ $AV_{SS} = DV_{SS} = 0V$			2.7		3.6	٧
		SD16LP = 0,	GAIN: 1, 2	3 V		650	950	
	Analog supply	$f_{SD16} = 1 \text{ MHz},$	GAIN: 4, 8, 16	3 V		730	1100	
I <sub>SD16</sub>	current: 1 active SD16 channel	SD16OSR = 256	GAIN: 32	3 V		1050	1550	μА
13016	including internal	SD16LP = 1,	GAIN: 1	3 V		620	930	μι
	reference	f <sub>SD16</sub> = 0.5 MHz, SD16OSR = 256	GAIN: 32	3 V		700	1060	
	Analog front-end	SD16LP = 0 (Low p	ower mode disabled)	3 V		1		N41.1-
f <sub>SD16</sub>	input clock frequency	SD16LP = 1 (Low p	ower mode enabled)	3 V		0.5		MHz

#### SD16, analog input range (see Note 1)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
		SD16GAINx = 1, SD16REFON = 1			±500		
	Differential input	SD16GAINx = 2, SD16REFON = 1			±250		
\ ,	voltage range for	SD16GAINx = 4, SD16REFON = 1			±125		\
$V_{ID}$	specified performance	SD16GAINx = 8, SD16REFON = 1			±62		mV
	(see Note 2)	SD16GAINx = 16, SD16REFON = 1			±31		
		SD16GAINx = 32, SD16REFON = 1			±15		
_	Input impedance	f <sub>SD16</sub> = 1MHz, SD16GAINx = 1	3 V		200		1.0
Z <sub>I</sub>	(one input pin to AV <sub>SS</sub> )	f <sub>SD16</sub> = 1MHz, SD16GAINx = 32	3 V		75		kΩ
_	Differential	f <sub>SD16</sub> = 1MHz, SD16GAINx = 1	3 V	300	400		1.0
Z <sub>ID</sub>	input impedance (IN+ to IN-)	f <sub>SD16</sub> = 1MHz, SD16GAINx = 32	3 V	100	150		kΩ
VI	Absolute input voltage range			AV <sub>SS</sub> - 1.0V		AV <sub>CC</sub>	V
V <sub>IC</sub>	Common-mode input voltage range			AV <sub>SS</sub> -		AV <sub>CC</sub>	V

NOTES: 1. All parameters pertain to each SD16 channel.



The analog input range depends on the reference voltage applied to V<sub>REF</sub>. If V<sub>REF</sub> is sourced externally, the full-scale range is defined by V<sub>FSR+</sub> = +(V<sub>REF</sub>/2)/GAIN and V<sub>FSR-</sub> = -(V<sub>REF</sub>/2)/GAIN. The analog input range should not exceed 80% of V<sub>FSR+</sub> or V<sub>FSR+</sub>.

#### electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

# SD16, analog performance ( $f_{SD16} = 1MHz$ , SD16OSRx = 256, SD16REFON = 1)

P.	ARAMETER	TEST CONDITIONS		V <sub>CC</sub>	MIN	TYP	MAX	UNIT
		SD16GAINx = 1,Signal Amplitude = 500mV		3 V	83.5	85		
		SD16GAINx = 2,Signal Amplitude = 250mV		3 V	81.5	84		
SINAD	Signal-to-noise +	SD16GAINx = 4,Signal Amplitude = 125mV	f <sub>IN</sub> = 50Hz,	3 V	76	79.5		dB
SINAD	distortion ratio	SD16GAINx = 8,Signal Amplitude = 62mV	100Hz	3 V	73	76.5		ав
		SD16GAINx = 16,Signal Amplitude = 31mV		3 V	69	73		
		SD16GAINx = 32,Signal Amplitude = 15mV		3 V	62	69		
		SD16GAINx = 1		3 V	0.97	1.00	1.02	
		SD16GAINx = 2		3 V	1.90	1.96	2.02	
•		SD16GAINx = 4		3 V	3.76	3.86	3.96	
G	Nominal gain	SD16GAINx = 8		3 V	7.36	7.62	7.84	
		SD16GAINx = 16		3 V	14.56	15.04	15.52	
		SD16GAINx = 32		3 V	27.20	28.35	29.76	
_	Offset error	SD16GAINx = 1		3 V			±0.2	%FSR
E <sub>OS</sub>	Oliset error	SD16GAINx = 32		3 V			±1.5	%F3N
ط⊏ /ط⊤	Offset error temperature	SD16GAINx = 1		3 V		±4	±20	ppm
dE <sub>OS</sub> /dT	coefficient	SD16GAINx = 32		3 V		±20	±100	FSR/°C
OMBB	Common-mode	SD16GAINx = 1, Common-mode input signal V <sub>ID</sub> = 500 mV, f <sub>IN</sub> = 50 Hz, 100 Hz	l:	3 V		>90		į
CMRR	rejection ratio	SD16GAINx = 32, Common-mode input signa $V_{ID}$ = 16 mV, $f_{IN}$ = 50 Hz, 100 Hz	al:	3 V		>75		dB
AC PSRR	AC power supply rejection ratio	SD16GAINx = 1, $V_{CC}$ = 3 V ± 100 mV, $f_{VCC}$ =	= 50 Hz	3 V		>80		dB
X <sub>T</sub>	Crosstalk			3 V		<-100		dB

### SD16, built-in temperature sensor

PA	RAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	TYP	MAX	UNIT
TC <sub>Sensor</sub>	Sensor temperature coefficient			1.18	1.32	1.46	mV/K
V <sub>Offset,sensor</sub>	Sensor offset voltage			-100		100	mV
	_	Temperature sensor voltage at T <sub>A</sub> = 85°C	3 V	435	475	515	
V <sub>Sensor</sub>	Sensor output voltage (see Note 2)	Temperature sensor voltage at T <sub>A</sub> = 25°C	3 V	355	395	435	mV
	Voltago (000 14010 2)	Temperature sensor voltage at T <sub>A</sub> = 0°C	3 V	320	360	400	

NOTES: 1. The following formula can be used to calculate the temperature sensor output voltage:  $V_{Sensor,typ} = TC_{Sensor} (273 + T [^{\circ}C]) + V_{Offset,sensor} [mV] \\ 2. Results based on characterization and/or production test, not TC_{Sensor} or V_{Offset,sensor}.$ 



# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

#### SD16, built-in voltage reference

PA	RAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>REF</sub>	Internal reference voltage	SD16REFON = 1, SD16VMIDON = 0	3 V	1.14	1.20	1.26	V
I <sub>REF</sub>	Reference supply current	SD16REFON = 1, SD16VMIDON = 0	3 V		175	260	μΑ
тс	Temperature coefficient	SD16REFON = 1, SD16VMIDON = 0	3 V		20	50	ppm/K
C <sub>REF</sub>	V <sub>REF</sub> load capacitance	SD16REFON = 1, SD16VMIDON = 0 (see Note 1)			100		nF
I <sub>LOAD</sub>	V <sub>REF</sub> maximum load current	SD16REFON = 0	3 V			±200	nA
t <sub>ON</sub>	Turn-on time	SD16REFON = $0 \rightarrow 1$ , SD16VMIDON = $0$ , $C_{REF} = 100 \text{ nF}$	3 V		5		ms
DC PSR	DC power supply rejection, ΔV <sub>REF</sub> /ΔV <sub>CC</sub>	SD16REFON = 1, SD16VMIDON = 0, V <sub>CC</sub> = 2.5 V to 3.6 V			200		μV/V

NOTES: 1. There is no capacitance required on V<sub>REF</sub>. However, a capacitance of at least 100nF is recommended to reduce any reference voltage noise.

#### SD16, built-in reference output buffer

P/	ARAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>REF,BUF</sub>	Reference buffer output voltage	SD16REFON = 1, SD16VMIDON = 1	3 V		1.2		V
I <sub>REF,BUF</sub>	Reference Supply + Reference output buffer quiescent current	SD16REFON = 1, SD16VMIDON = 1	3 V		385	600	μΑ
C <sub>REF(O)</sub>	Required load capacitance on V <sub>REF</sub>	SD16REFON = 1, SD16VMIDON = 1		470			nF
I <sub>LOAD,Max</sub>	Maximum load current on V <sub>REF</sub>	SD16REFON = 1, SD16VMIDON = 1	3 V			±1	mA
	Maximum voltage variation vs. load current	I <sub>LOAD</sub>   = 0 to 1mA	3 V	-15		+15	mV
t <sub>ON</sub>	Turn-on time	SD16REFON = $0 \rightarrow 1$ , SD16VMIDON = 1, $C_{REF} = 470 \text{ nF}$	3 V		100		μs

#### SD16, external reference input

PA	RAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
$V_{REF(I)}$	Input voltage range	SD16REFON = 0	3 V	1.0	1.25	1.5	V
I <sub>REF(I)</sub>	Input current	SD16REFON = 0	3 V			50	nA



# electrical characteristics over recommended operating free-air temperature (unless otherwise noted) (continued)

#### **Flash Memory**

	PARAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	NOM	МАХ	UNIT
V <sub>CC(PGM/</sub> ERASE)	Program and Erase supply voltage			2.7		3.6	٧
f <sub>FTG</sub>	Flash Timing Generator frequency			257		476	kHz
I <sub>PGM</sub>	Supply current from DV <sub>CC</sub> during program		2.7 V/ 3.6 V		3	5	mA
I <sub>ERASE</sub>	Supply current from DV <sub>CC</sub> during erase		2.7 V/ 3.6 V		3	7	mA
t <sub>CPT</sub>	Cumulative program time	see Note 1	2.7 V/ 3.6 V			10	ms
t <sub>CMErase</sub>	Cumulative mass erase time	see Note 2	2.7 V/ 3.6 V	200			ms
	Program/Erase endurance			10 <sup>4</sup>	10 <sup>5</sup>		cycles
t <sub>Retention</sub>	Data retention duration	$T_J = 25^{\circ}C$		100			years
t <sub>Word</sub>	Word or byte program time				35		
t <sub>Block, 0</sub>	Block program time for 1st byte or word	1			30		
t <sub>Block, 1-63</sub>	Block program time for each additional byte or word	] ,, ,			21		
t <sub>Block, End</sub>	Block program end-sequence wait time	see Note 3			6		t <sub>FTG</sub>
t <sub>Mass Erase</sub> Mass erase time					5297		
t <sub>Seg Erase</sub>	Segment erase time				4819	·	

- NOTES: 1. The cumulative programming time must not be exceeded when writing to a 64-byte flash block. This parameter applies to all programming methods: individual word/byte write and block write modes.
  - The mass erase duration generated by the flash timing generator is at least 11.1ms (= 5297x1/f<sub>FTG</sub>,max = 5297x1/476kHz). To achieve the required cumulative mass erase time the Flash Controller's mass erase operation can be repeated until this time is met. (A worst case minimum of 19 cycles are required).
  - 3. These values are hardwired into the Flash Controller's state machine ( $t_{FTG} = 1/f_{FTG}$ ).

#### **JTAG Interface**

	PARAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	NOM	MAX	UNIT
,		N	2.2 V	0		5	MHz
TCK	TCK input frequency	see Note 1	3 V	0		10	MHz
R <sub>Internal</sub>	Internal pull-up resistance on TMS, TCK, TDI/TCLK	see Note 2	2.2 V/ 3 V	25	60	90	kΩ

NOTES: 1. f<sub>TCK</sub> may be restricted to meet the timing requirements of the module selected.

2. TMS, TDI/TCLK, and TCK pull-up resistors are implemented in all versions.

#### JTAG Fuse (see Note 1)

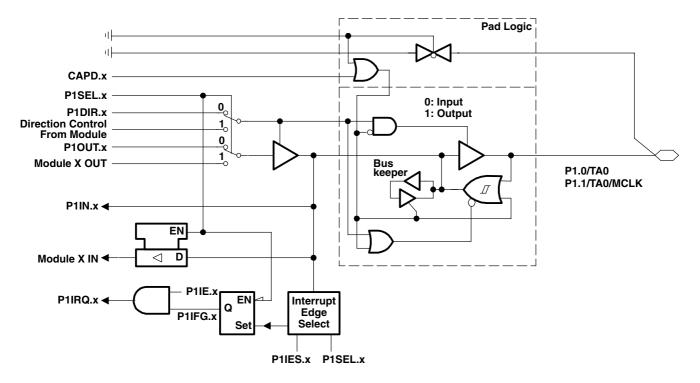
	PARAMETER	TEST CONDITIONS	v <sub>cc</sub>	MIN	NOM	MAX	UNIT
V <sub>CC(FB)</sub>	Supply voltage during fuse-blow condition	T <sub>A</sub> = 25°C		2.5			V
$V_{FB}$	Voltage level on TDI/TCLK for fuse-blow			6		7	V
I <sub>FB</sub>	Supply current into TDI/TCLK during fuse-blow					100	mA
t <sub>FB</sub>	Time to blow fuse					1	ms

NOTES: 1. Once the fuse is blown, no further access to the MSP430 JTAG/Test and emulation features is possible. The JTAG block is switched to bypass mode.



#### input/output schematic

#### Port P1, P1.0 to P1.1, input/output with Schmitt-trigger



NOTE:  $0 \le x \le 1$ .

Port Function is Active if CAPD.x = 0

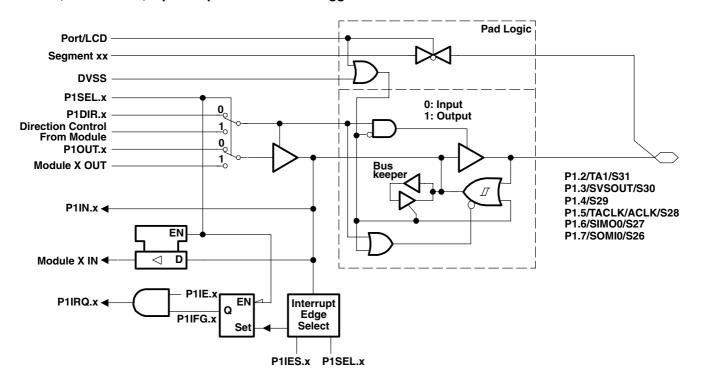
PnSEL.x	PnDIR.x	Direction Control From Module	PnOUT.x	Module X OUT	PnlN.x	Module X IN	PnIE.x	PnIFG.x	PnlES.x	CAPD.x
P1SEL.0	P1DIR.0	P1DIR.0	P1OUT.0	Out0 Sig.†	P1IN.0	CCI0A <sup>†</sup>	P1IE.0	P1IFG.0	P1IES.0	DVSS
P1SEL.1	P1DIR.1	P1DIR.1	P1OUT.1	MCLK	P1IN.1	CCI0B <sup>†</sup>	P1IE.1	P1IFG.1	P1IES.1	DVSS

<sup>†</sup> Timer\_A3



#### input/output schematic (continued)

#### Port P1, P1.2 to P1.7, input/output with Schmitt-trigger



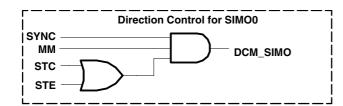
NOTE:  $2 \le x \le 7$ .

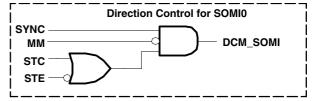
Port Function is Active if Port/LCD = 0

PnSEL.x	PnDIR.x	Direction Control From Module	PnOUT.x	Module X OUT	PnIN.x	Module X IN	PnIE.x	PnIFG.x	PnIES.x	Port/LCD	Segment
P1SEL.2	P1DIR.2	P1DIR.2	P1OUT.2	Out1 Sig.†	P1IN.2	CCI1A <sup>†</sup>	P1IE.2	P1IFG.2	P1IES.2		S31
P1SEL.3	P1DIR.3	P1DIR.3	P1OUT.3	SVSOUT	P1IN.3	unused	P1IE.3	P1IFG.3	P1IES.3	0: LCDM < 0E0h	S30
P1SEL.4	P1DIR.4	P1DIR.4	P1OUT.4	DVSS	P1IN.4	unused	P1IE.4	P1IFG.4	P1IES.4	1: LCDM ≥ 0E0h	S29
P1SEL.5	P1DIR.5	P1DIR.5	P1OUT.5	ACLK	P1IN.5	TACLK†	P1IE.5	P1IFG.5	P1IES.5		S28
P1SEL.6	P1DIR.6	DCM_SIMO	P1OUT.6	SIMO0(o)‡	P1IN.6	SIMO0(i)‡	P1IE.6	P1IFG.6	P1IES.6	0: LCDM < 0C0h	S27
P1SEL.7	P1DIR.7	DCM_SOMI	P1OUT.7	SOMI0(o) <sup>‡</sup>	P1IN.7	SOMI0(i)‡	P1IE.7	P1IFG.7	P1IES.7	1: LCDM ≥ 0C0h	S26

<sup>†</sup> Timer\_A3

<sup>‡</sup> USART0

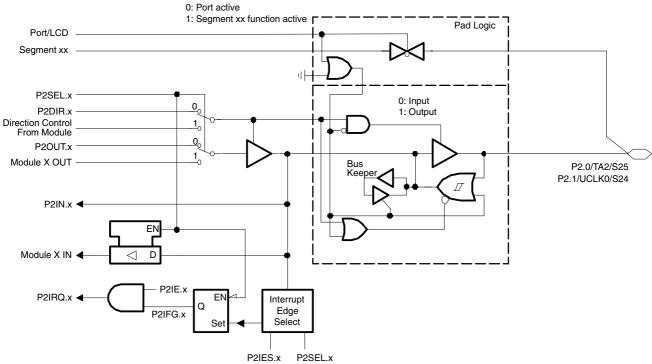






#### input/output schematic (continued)

### port P2, P2.0 to P2.1, input/output with Schmitt-trigger

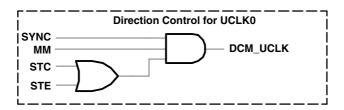


NOTE:  $0 \le x \le 1$ .

Port Function is Active if Port/LCD = 0

PnSel.x	PnDIR.x	Dir. Control from module	PnOUT.x	Module X OUT	PnIN.x	Module X IN	PnIE.x	PnIFG.x	PnIES.x	Port/LCD	Segment
P2Sel.0	P2DIR.0	P2DIR.0	P2OUT.0	Out2sig.†	P2IN.0	CCI2A †	P2IE.0	P2IFG.0	P2IES.0	0: LCDM < 0E0h	S25
P2Sel.1	P2DIR.1	DCM_UCLK	P2OUT.1	UCLK0(o) <sup>‡</sup>	P2IN.1	UCLK0(i) <sup>‡</sup>	P2IE.1	P2IFG.1	P2IES.1	1: LCDM ≥ 0E0h	S24

<sup>†</sup> Timer\_A3

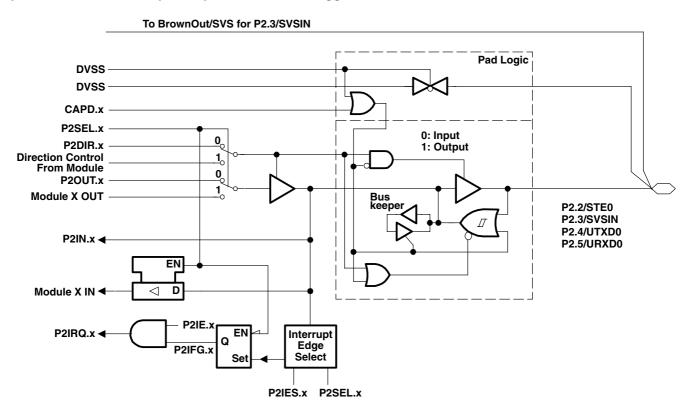




<sup>‡</sup> USART0

#### input/output schematic (continued)

## port P2, P2.2 to P2.5, input/output with Schmitt-trigger



NOTE:  $2 \le x \le 5$ 

Port function is active if CAPD.x = 0

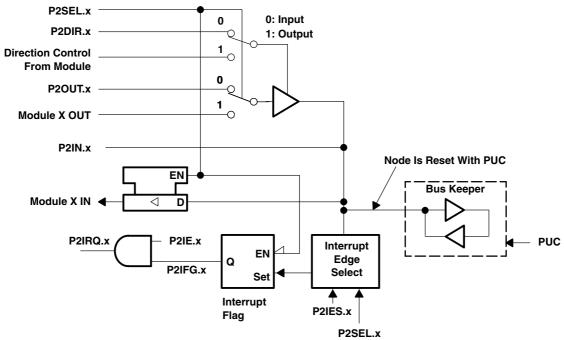
PnSEL.x	PnDIR.x	Direction Control From Module	PnOUT.x	Module X OUT	PnIN.x	Module X IN	PnIE.x	PnIFG.x	PnIES.x	CAPD.x
P2SEL.2	P2DIR.2	DVSS	P2OUT.2	DVSS	P2IN.2	STE0†	P2IE.2	P2IFG.2	P2IES.2	DVSS
P2SEL.3	P2DIR.3	P2DIR.3	P2OUT.3	DVSS	P2IN.3	unused	P2IE.3	P2IFG.3	P2IES.3	SVSCTL VLD = 1111b
P2SEL.4	P2DIR.4	DVCC	P2OUT.4	UTXD0 <sup>†</sup>	P2IN.4	unused	P2IE.4	P2IFG.4	P2IES.4	DVSS
P2SEL.5	P2DIR.5	DVSS	P2OUT.5	DVSS	P2IN.5	URXD0†	P2IE.5	P2IFG.5	P2IES.5	DVSS

<sup>†</sup> USART0



#### input/output schematic (continued)

#### Port P2, unbonded GPIOs P2.6 and P2.7



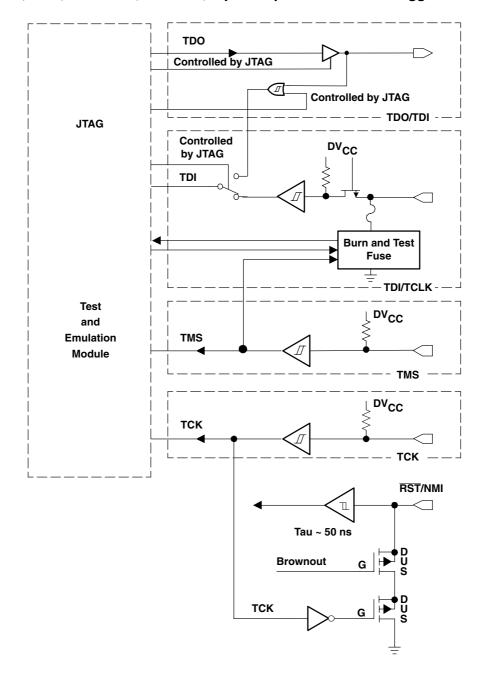
NOTE: x = Bit/identifier, 6 to 7 for port P2 without external pins

P2Sel.x	P2DIR.x	DIRECTION CONTROL FROM MODULE	P2OUT.x	MODULE X OUT	P2IN.x	MODULE X IN	P2IE.x	P2IFG.x	P2IES.x
P2Sel.6	P2DIR.6	P2DIR.6	P2OUT.6	$DV_SS$	P2IN.6	unused	P2IE.6	P2IFG.6	P2IES.6
P2Sel.7	P2DIR.7	P2DIR.7	P2OUT.7	DV <sub>SS</sub>	P2IN.7	unused	P2IE.7	P2IFG.7	P2IES.7

NOTE: Unbonded GPIOs 6 and 7 of port P2 can be used as interrupt flags. Only software can affect the interrupt flags. They work as software interrupts.



## JTAG pins TMS, TCK, TDI/TCLK, TDO/TDI, input/output with Schmitt-trigger or output



#### JTAG fuse check mode

MSP430 devices that have the fuse on the TDI/TCLK terminal have a fuse check mode that tests the continuity of the fuse the first time the JTAG port is accessed after a power-on reset (POR). When activated, a fuse check current, I<sub>TF</sub>, of 1.8 mA at 3 V can flow from the TDI/TCLK pin to ground if the fuse is not burned. Care must be taken to avoid accidentally activating the fuse check mode and increasing overall system power consumption.

Activation of the fuse check mode occurs with the first negative edge on the TMS pin after power up or if the TMS is being held low during power up. The second positive edge on the TMS pin deactivates the fuse check mode. After deactivation, the fuse check mode remains inactive until another POR occurs. After each POR the fuse check mode has the potential to be activated.

The fuse check current only flows when the fuse check mode is active and the TMS pin is in a low state (see Figure 14). Therefore, the additional current flow can be prevented by holding the TMS pin high (default condition).

The JTAG pins are terminated internally, and therefore do not require external termination.

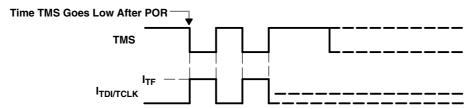


Figure 14. Fuse Check Mode Current, MSP430F42x



# MSP430F42x MIXED SIGNAL MICROCONTROLLER

SLAS421A - APRIL 2004 - REVISED JUNE 2007

# **Data Sheet Revision History**

Literature Number	Summary
SLAS421	Production datasheet release
SLAS421A	Updated functional block diagram (page 3)  Clarified test conditions in recommended operating conditions table (page 17)  Changed "Supply voltage during program execution; SD16 disabled, SVS enabled, and PORON = 1" MIN value from 2.2 V to 2.0 V (page 17)  Clarified test conditions for I <sub>(LPM0)</sub> in supply current into AV <sub>CC</sub> + DV <sub>CC</sub> table (page 18)  Clarified test conditions in USART0 table (page 21)  Changed PSRR to AC PSRR in SD16 analog performance table (page 29)  Added DC PSR in SD16, built-in voltage reference table (page 30)  Added t <sub>ON</sub> parameter to SD16, built-in voltage reference and SD16, built-in reference output buffer tables (page 30)  Changed t <sub>CPT</sub> maximum value from 4 ms to 10 ms in Flash memory table (page 31)

NOTE: Page and figure numbers refer to the respective document revision.

#### PACKAGE OPTION ADDENDUM

www.ti.com 30-Apr-2009

#### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
MSP430A090IPMR	ACTIVE	LQFP	PM	64	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430A092IPMR	ACTIVE	LQFP	PM	64	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F423IPM	ACTIVE	LQFP	PM	64	160	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F423IPMR	ACTIVE	LQFP	PM	64	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F425IPM	ACTIVE	LQFP	PM	64	160	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F425IPMR	ACTIVE	LQFP	PM	64	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F427IPM	ACTIVE	LQFP	PM	64	160	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
MSP430F427IPMR	ACTIVE	LQFP	PM	64	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR

<sup>&</sup>lt;sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

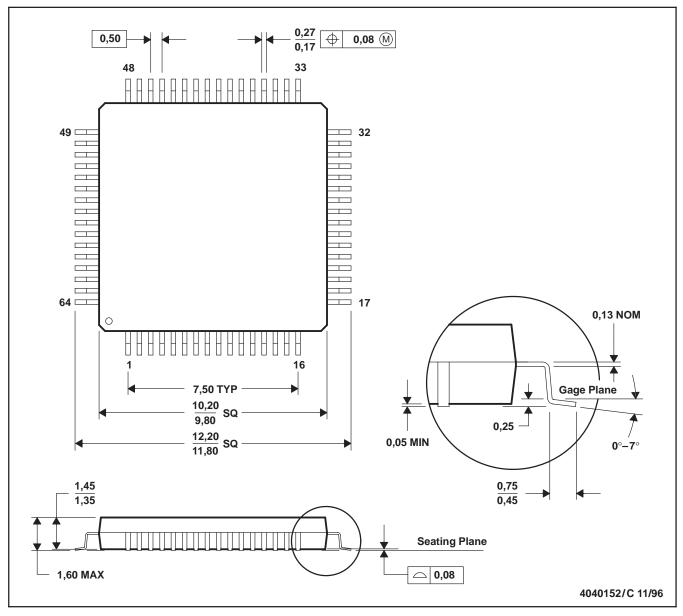
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### PM (S-PQFP-G64)

#### PLASTIC QUAD FLATPACK

1



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-026
- D. May also be thermally enhanced plastic with leads connected to the die pads.

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