

FM24C64

64Kb FRAM® Serial Memory



Features

64K bit Ferroelectric Nonvolatile RAM

- Organized as 8,192 x 8 bits
- High endurance 10 Billion (10^{10}) read/writes
- 10 year data retention at 55°C
- No write delay
- Advanced high-reliability ferroelectric process

Fast Two-wire Serial Interface

- Up to 1 MHz maximum bus frequency
- Direct hardware replacement for EEPROM
- Supports legacy timing for 100 kHz & 400 kHz

Low Power Operation

- True 5V operation
- 150 µA Active current (100 kHz)
- 10 µA standby current

Industry Standard Configuration

- Industrial temperature -40°C to +85°C
- 8-pin SOP or DIP

Description

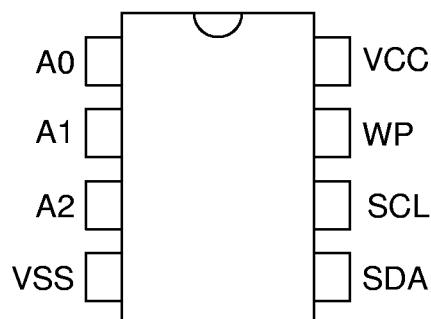
The FM24C64 is a 64-kilobit nonvolatile memory employing an advanced ferroelectric process. A ferroelectric random access memory or FRAM is nonvolatile but operates in other respects as a RAM. It provides reliable data retention for 10 years while eliminating the complexities, overhead, and system level reliability problems caused by EEPROM and other nonvolatile memories.

The FM24C64 performs write operations at bus speed. No write delays are incurred. Data is written to the memory array mere hundreds of nanoseconds after it has been successfully transferred to the device. The next bus cycle may commence immediately. In addition, the product offers substantial write endurance compared with other nonvolatile memories. The FM24C64 is capable of supporting up to 1E10-read/write cycles -- far more than most systems will require from a serial memory.

These capabilities make the FM24C64 ideal for nonvolatile memory applications requiring frequent or rapid writes. Examples range from data collection where the number of write cycles may be critical, to demanding industrial controls where the long write time of EEPROM can cause data loss. The combination of features allows more frequent data writing with less overhead for the system.

The FM24C64 provides substantial benefits to users of serial EEPROM, yet these benefits are available in a hardware drop-in replacement. The FM24C64 is provided in industry standard 8-pin packages using a familiar two-wire protocol. They are guaranteed over an industrial temperature range of -40°C to +85°C.

Pin Configuration



| Pin Names | Function |
|-----------|-----------------------|
| A0-A2 | Device Select Address |
| SDA | Serial Data/address |
| SCL | Serial Clock |
| WP | Write Protect |
| VSS | Ground |
| VCC | Supply Voltage 5V |

Ordering Information

| | |
|-----------|-------------------|
| FM24C64-P | 8-pin plastic DIP |
| FM24C64-S | 8-pin SOP |

This data sheet contains design specifications for product development. This product is still under development, these specifications may change in any manner without notice.

Ramtron International Corporation

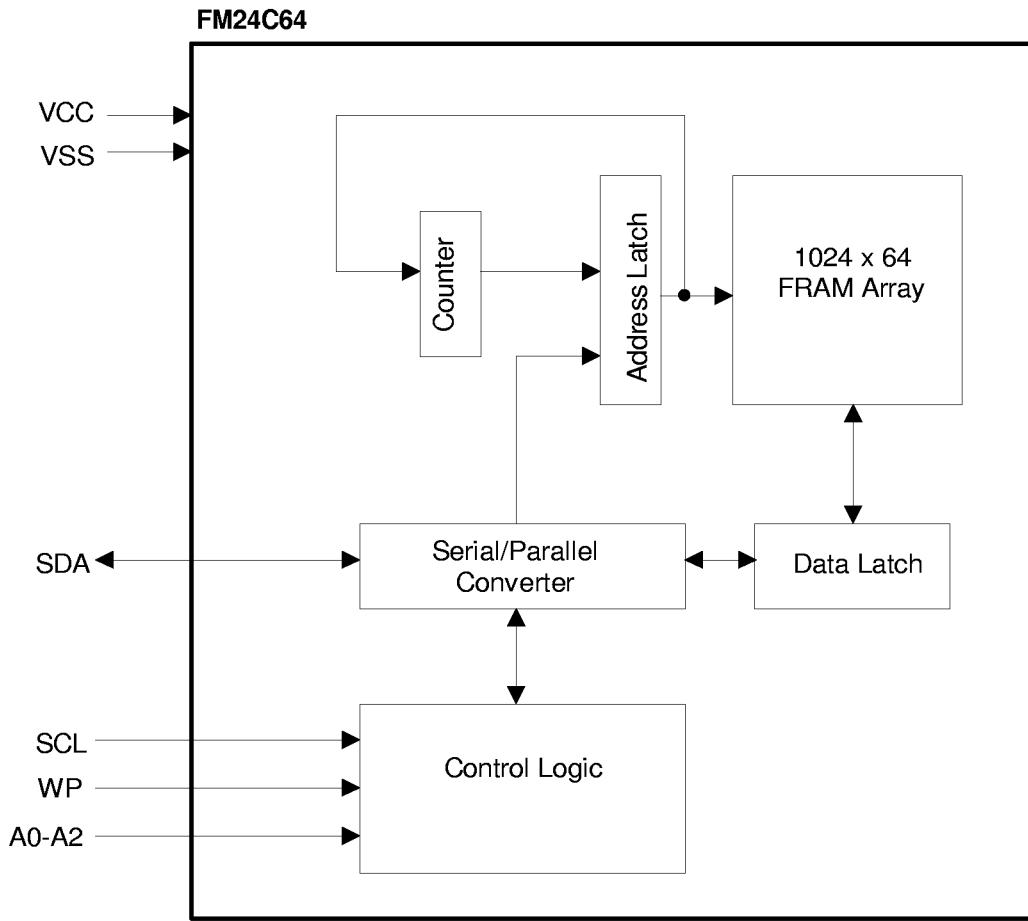
1850 Ramtron Drive

Colorado Springs, CO 80921

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www.ramtron.com

Figure 1. Block Diagram



Pin Description

| Pin Name | Pin Number | I/O | Pin Description |
|----------|------------|-----|--|
| A0-A2 | 1-3 | I | Address 0-2. These pins are used to select one of up to 8 devices of the same type on the same two-wire bus. To select the device, the address value on the three pins must match the corresponding bits contained in the device address. |
| VSS | 4 | I | Ground |
| SDA | 5 | I/O | Serial Data Address. This is a bi-directional line for the two-wire interface. It is open-drain and is intended to be wire-ORed with other devices on the two-wire bus. The input buffer incorporates a schmitt trigger for noise immunity and the output driver includes slope control for falling edges. A pull-up resistor is required. |
| SCL | 6 | I | Serial Clock. The serial clock line for the two-wire interface. Data is clocked out of the part on the falling edge, and in on the rising edge. The SCL input also incorporates a schmitt trigger input for noise immunity. |
| WP | 7 | I | Write Protect. When tied to VCC, addresses in the upper quadrant of the logical memory map will be write-protected. Write access to the lower three-quarters of the address area is permitted. When WP is connected to ground, all addresses may be written. This pin must not be left floating. |
| VCC | 8 | I | Supply Voltage. 5V |

Overview

The FM24C64 is a serial FRAM memory. The memory array is logically organized as a 8,192 x 8 bit memory array and is accessed using an industry standard two-wire interface. Functional operation of the FRAM is similar to serial EEPROMs. The major difference between the FM24C64 and a serial EEPROM with the same pin-out relates to its superior write performance.

Memory Architecture

When accessing the FM24C64, the user addresses 8,192 locations each with 8 data bits. These data bits are shifted serially. The 8,192 addresses are accessed using the two-wire protocol, which includes a slave address (to distinguish other non-memory devices), and an extended 16-bit address. Only the lower 13 bits are used by the decoder for accessing the memory. The upper three address bits should be set to 0 for compatibility with larger devices in the future.

The access time for memory operation is essentially zero beyond the time needed for the serial protocol. That is, the memory is read or written at the speed of the two-wire bus. Unlike an EEPROM, it is not necessary to poll the device for a ready condition since writes occur at bus speed. That is, by the time a new bus transaction can be shifted into the part, a write operation will be complete. This is explained in more detail in the interface section below.

Users expect several obvious system benefits from the FM24C64 due to its fast write cycle and high endurance as compared with EEPROM. However there are less obvious benefits as well. For example in a high noise environment, the fast-write operation is less susceptible to corruption than an EEPROM since it is completed quickly. By contrast, an EEPROM requiring milliseconds to write is vulnerable to noise during much of the cycle.

Note that the FM24C64 contains no power management circuits other than a simple internal power-on reset. It is the user's responsibility to ensure that VCC is within data sheet tolerances to prevent incorrect operation.

Two-wire Interface

The FM24C64 employs a bi-directional two-wire bus protocol using few pins or board space. Figure 2 illustrates a typical system configuration using the FM24C64 in a microcontroller-based system. The industry standard two-wire bus is familiar to many users but is described in this section.

By convention, any device that is sending data onto the bus is the transmitter while the target device for this data is the receiver. The device that is controlling the bus is the master. The master is responsible for generating the clock signal for all operations. Any device on the bus that is being controlled is a slave. The FM24C64 always is a slave device.

The bus protocol is controlled by transition states in the SDA and SCL signals. There are four conditions including start, stop, data bit, or acknowledge. Figure 3 illustrates the signal conditions that specify the four states. Detailed timing diagrams are in the electrical specifications.

Figure 2. Typical System Configuration

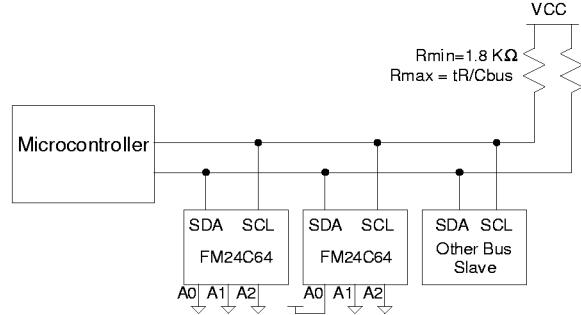
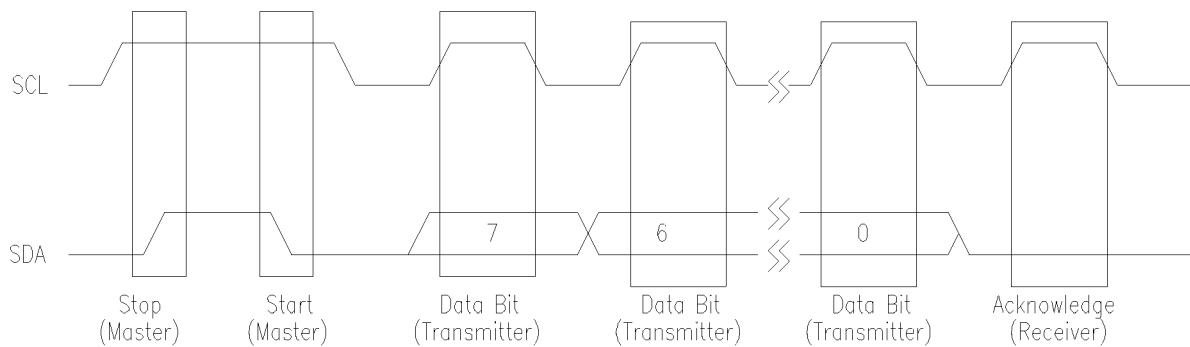


Figure 3. Data Transfer Protocol

Start Condition

A start condition is indicated when the bus master drives SDA from high to low while the SCL signal is high. All commands should be preceded by a start condition. An operation in progress can be aborted by asserting a start condition at any time. Aborting an operation using the start condition will ready the FM24C64 for a new operation.

If during operation the power supply drops below the specified VCC minimum, the system should issue a start condition prior to performing another operation.

Stop Condition

A stop condition is indicated when the bus master drives SDA from low to high while the SCL signal is high. All operations using the FM24C64 should end with a stop condition. If an operation is in progress when a stop is asserted, the operation will be aborted. The master must have control of SDA (not a memory read) in order to assert a stop condition.

Data/Address Transfer

All data transfers (including addresses) take place while the SCL signal is high. Except under the two conditions described above, the SDA signal should not change while SCL is high.

Acknowledge

The acknowledge takes place after the 8th data bit has been transferred in any transaction. During this state the transmitter should release the SDA bus to allow the receiver to drive it. The receiver drives the SDA signal low to acknowledge receipt of the byte. If the receiver does not drive SDA low, the condition is a no-acknowledge and the operation is aborted.

The receiver would fail to acknowledge for two distinct reasons. First is that a byte transfer fails. In this case, the no-acknowledge ceases the current

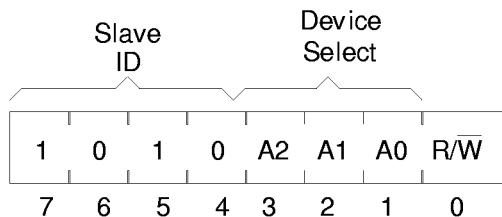
operation so that the part can be addressed again. This allows the last byte to be recovered in the event of a communication error.

Second and most common, the receiver does not acknowledge to deliberately end an operation. For example, during a read operation, the FM24C64 will continue to place data onto the bus as long as the receiver sends acknowledges (and clocks). When a read operation is complete and no more data is needed, the receiver must not acknowledge the last byte. If the receiver acknowledges the last byte, this will cause the FM24C64 to attempt to drive the bus on the next clock while the master is sending a new command such as stop.

Slave Address

The first byte that the FM24C64 expects after a start condition is the slave address. As shown in Figure 4, the slave address contains the device type, the device select address bits, and a bit that specifies if the transaction is a read or a write.

Bits 7-4 are the device type and should be set to 1010b for the FM24C64. These bits allow other types of function types to reside on the 2-wire bus within an identical address range. Bits 3-1 are the address select bits. They must match the corresponding value on the external address pins to select the device. Up to eight, FM24C64s can reside on the same two-wire bus by assigning a different address to each. Bit 0 is the read/write bit. A 0 indicates a write operation.

Figure 4. Slave Address

Addressing Overview

After the FM24C64 (as receiver) acknowledges the device address, the master can place the memory address on the bus for a write operation. The address requires two bytes. The first is the MSB. Since the device uses only 13 address bits, the value of the upper three bits are don't care. Following the MSB is the LSB with the remaining eight address bits. The address value is latched internally. Each access causes the latched address value to be incremented automatically. The current address is the value that is held in the latch -- either a newly written value or the address following the last access. The current address will be held for as long as power remains or until a new value is written. Reads always use the current address. A random read address can be loaded by beginning a write operation as explained below.

After transmission of each data byte, just prior to the acknowledge, the FM24C64 increments the internal address latch. This allows the next sequential byte to be accessed with no additional addressing. After the last address (1FFFh) is reached, the address latch will roll over to 0000h. There is no limit to the number of bytes that can be accessed with a single read or write operation.

Data Transfer

After the address information has been transmitted, data transfer between the bus master and the FM24C64 can begin. For a read operation the FM24C64 will place 8 data bits on the bus then wait for an acknowledge from the master. If the acknowledge occurs, the FM24C64 will transfer the next sequential byte. If the acknowledge is not sent, the FM24C64 will end the read operation. For a write operation, the FM24C64 will accept 8 data bits from the master then send an acknowledge. All data transfer occurs msb (most significant bit) first.

Memory Operation

The FM24C64 is designed to operate in a manner very similar to other 2-wire interface memory products. The major differences result from the higher performance write capability of FRAM technology. These improvements result in some differences between the FM24C64 and a similar configuration EEPROM during writes. The complete operation for both writes and reads is explained below.

Write Operation

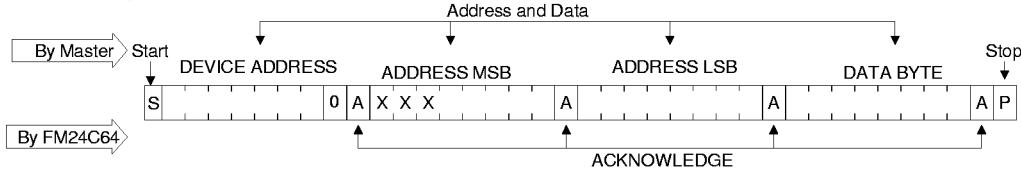
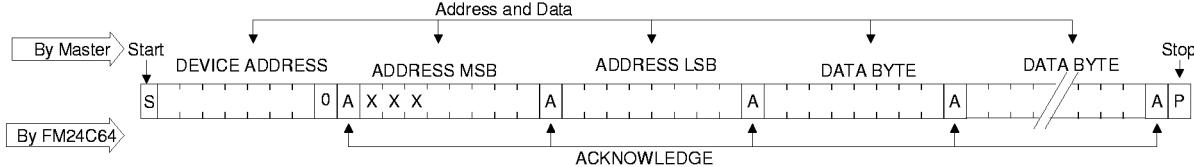
All writes begin with a device address, then a memory address. The bus master indicates a write operation by setting the lsb of the device address to a 0. After addressing, the bus master sends each byte of data to the memory and the memory generates an acknowledge condition. Any number of sequential bytes may be written. If the end of the address range is reached internally, the address counter will wrap from 1FFFh to 0000h.

Unlike other nonvolatile memory technologies, there is no effective write delay with FRAM. Since the read and write access times of the underlying memory are the same, the user experiences no delay through the bus. The entire memory cycle occurs in less time than a single bus clock. Therefore, any operation including read or write can occur immediately following a write. Acknowledge polling, a technique used with EEPROMs to determine if a write is complete is unnecessary and will always return a ready condition.

Internally, an actual memory write occurs after the 8th data bit is transferred. It will be complete before the acknowledge is sent. Therefore if the user desires to abort a write without altering the memory contents, this should be done using start or stop condition prior to the 8th data bit. The FM24C64 uses no page buffering.

Portions of the memory array can be write protected using the WP pin. Setting the WP pin to a high condition (VCC) will write-protect addresses in the upper quadrant from 1800h to 1FFFh. The FM24C64 will not acknowledge data bytes that are written to protected addresses. In addition, the address counter will not increment if writes are attempted to these addresses. Setting WP to a low state (VSS) will deactivate this feature. WP should not be left floating.

Figure 5 below illustrates both a single-byte and multiple-write.

Figure 5 Byte Write**Figure 6 Multiple Byte Write**

Read Operation

There are two basic types of read operations. They are current address read and selective address read. In a current address read, the FM24C64 uses the internal address latch to supply the address. In a selective read, the user performs a procedure to set the address to a specific value.

Current Address & Sequential Read

As mentioned above the FM24C64 uses an internal latch to supply the address for a read operation. A current address read uses the existing value in the address latch as a starting place for the read operation. The system reads from the address immediately following that of the last operation.

To perform a current address read, the bus master supplies a device address with the lsb set to 1. This indicates that a read operation is requested. After receiving the complete device address, the FM24C64 will begin shifting out data from the current address on the next clock. The current address is the value held in the internal address latch.

Beginning with the current address, the bus master can read any number of bytes. Thus, a sequential read is simply a current address read with multiple byte transfers. After each byte the internal address counter will be incremented.

Each time the bus master acknowledges a byte, this indicates that the FM24C64 should read out the next sequential byte.

There are four ways to properly terminate a read operation. Failing to properly terminate the read will

most likely create a bus contention as the FM24C64 attempts to read out additional data onto the bus. The four valid methods are as follows.

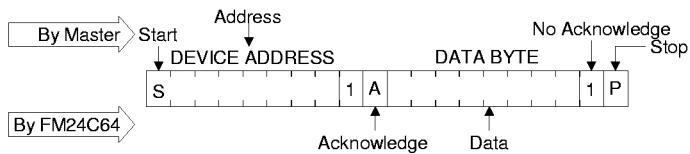
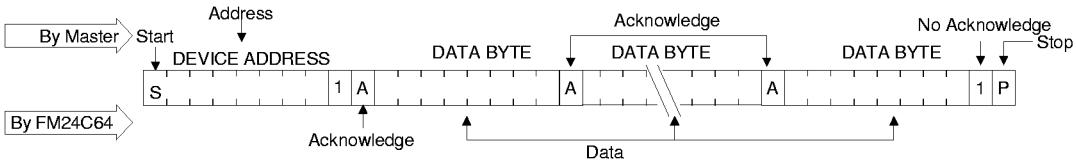
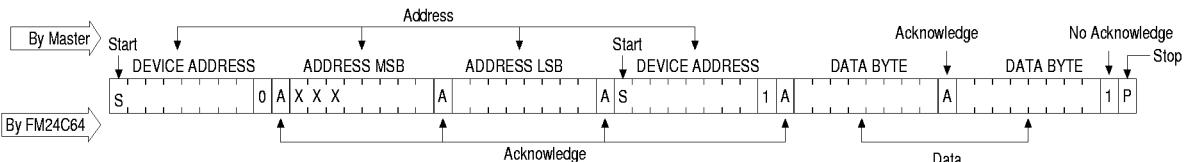
1. The bus master issues a no-acknowledge in the 9th clock cycle and a stop in the 10th clock cycle. This is illustrated in the diagrams below. This is preferred.
2. The bus master issues a no-acknowledge in the 9th clock cycle and a start in the 10th.
3. The bus master issues a stop in the 9th clock cycle.
4. The bus master issues a start in the 9th clock cycle.

If the internal address reaches 1FFFh, it will wrap around to 0000h on the next read cycle. Figures 7 and 8 below show the proper operation for current address reads.

Selective (Random) Read

There is a simple technique that allows a user to select a random address location as the starting point for a read operation. This involves using the first three bytes of a write operation to set the internal address followed by subsequent read operations.

To perform a selective read, the bus master sends out the device address with the lsb set to 0. This specifies a write operation. According to the write protocol, the bus master then sends the address bytes that are loaded into the internal address latch. After the FM24C64 acknowledges the address, the bus master issues a start condition. This simultaneously aborts the write operation and allows the read command to be issued with the device address lsb set to a 1. The operation is now a current address read.

Figure 7 Current Address Read**Figure 8 Sequential Read****Figure 9 Selective (Random) Read**

Data Retention and Endurance

Data retention is specified in the electrical specifications below. For purposes of clarity, this section contrasts the retention and endurance of FRAM with EEPROM.

Like other nonvolatile memories such as EEPROM, FRAM data retention actually is a function of temperature. Much like other memories, this data sheet provides retention information at one commonly used temperature. A more complete characteristic is provided in a separate application note. The retention performance of FRAM is very comparable to EEPROM in its characteristics. However the effect of endurance cycles on retention is different.

A typical EEPROM has a write endurance specification that is fixed. Surpassing the specified level of cycles on an EEPROM usually leads to a hard memory failure. By contrast, the effect of increasing cycles on FRAM produces an increase in the soft error rate. That is, there is a higher likelihood

of data loss but the memory continues to function properly.

While enough cycles would cause an apparent hard error, this is simply a very high soft error rate. This characteristic makes it problematic to assign a fixed endurance specification. Users should review the characteristic curve for retention after endurance. Individual system design tradeoffs should dictate the target endurance level.

Endurance is a soft specification. Therefore, the user may operate the device with different levels of endurance cycling for different portions of the memory. For example, critical data needing the highest reliability level could be stored in memory locations that receive comparatively few cycles. Data with shorter-term use could be located in an area receiving many more cycles. A scratchpad area, needing little if any retention can be cycled until there is virtually no retention capability remaining.

Internally, a FRAM operates with a read and restore mechanism similar to a DRAM. Therefore, endurance cycles are applied for each access; read or

write. The FRAM architecture is based on an array of rows and columns. Each access causes a cycle for an entire row. Therefore, data locations targeted for substantially differing numbers of cycles should not be located within the same row. In the FM24C64, a row is 64 bits wide. Each 8 bytes in the address marks the beginning of a new row.

Applications

The versatility of FRAM technology fits into many diverse applications. Clearly the strength of higher write endurance and faster writes make FRAM superior to EEPROM in all but one-time programmable applications. The advantage is most obvious in data collection environments where writes are frequent and data must be nonvolatile.

The attributes of fast writes and high write endurance combine in many innovative ways. A short list of ideas is provided here.

1. Data collection. In applications where data is collected and saved, FRAM provides a superior alternative to other solutions. It is more cost effective than battery backup for SRAM and provides better write attributes than EEPROM.
2. Configuration. Any nonvolatile memory can retain a configuration. However, if the configuration changes and power failure is a possibility, the higher write endurance of FRAM allows changes to be recorded without restriction. Any time the system state is altered, the change can be written. This avoids writing to memory on power down when the available time is short and power scarce.
3. High noise environments. Writing to EEPROM in a noisy environment can be challenging. When severe noise or power fluctuations are present, the long write time of EEPROM creates a window of

vulnerability during which the write can be corrupted. The fast write of FRAM is complete within a microsecond. This time is typically too short for noise or power fluctuation to disturb it.

4. Time to market. In a complex system, multiple software routines may need to access the nonvolatile memory. In this environment the time delay associated with programming EEPROM adds undue complexity to the software development. Each software routine must wait for complete programming before allowing access to the next routine. When time to market is critical, FRAM can eliminate this simple obstacle. As soon as a write is issued to the FM24C64, it is effectively done -- no waiting.

5. RF/ID. In the area of contactless memory, FRAM provides an ideal solution. Since RF/ID memory is powered by an RF field, the long programming time and high current consumption needed to write EEPROM is unattractive. FRAM provides a superior solution. The FM24C64 is suitable for multi-chip RF/ID products.

6. Maintenance tracking. In sophisticated systems, the operating history and system state during a failure is important knowledge. Maintenance can be expedited when this information has been recorded. Due to the high write endurance, FRAM makes an ideal system log. In addition, the convenient 2-wire interface of the FM24C64 allows memory to be distributed throughout the system using minimal additional resources.

Electrical Specifications

Absolute Maximum Ratings

| Description | Ratings |
|---|-----------------|
| Ambient storage or operating temperature | -40°C to + 85°C |
| Voltage on any pin with respect to ground | -1.0V to +7.0V |
| D.C. output current on any pin | 5 mA |
| Lead temperature (Soldering, 10 seconds) | 300° C |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only, and the functional operation of the device at these or any other conditions above those listed in the operational section of this specification is not implied. Exposure to absolute maximum ratings conditions for extended periods may affect device reliability

DC Operating Conditions TA = -40° C to + 85° C, VCC = 4.5V to 5.5V unless otherwise specified

| Symbol | Parameter | Min | Typ | Max | Units | Notes |
|------------------|---|-----------|-----------------|-------------------|----------------|-------|
| V _{CC} | Main Power Supply | 4.5 | 5.0 | 5.5 | V | 1 |
| I _{CC} | VCC Supply Current @ SCL = 100 kHz @ SCL = 400 kHz @ SCL = 1 MHz | | 115 400 1 | 150 500 1.2 | µA µA mA | 2 |
| I _{SB} | Standby Current | | 1 | 10 | uA | 3 |
| I _{LI} | Input Leakage Current | | | 10 | uA | 4 |
| I _{LO} | Output Leakage Current | | | 10 | uA | 4 |
| V _{IL} | Input Low Voltage | -0.3 | | VCC x 0.3 | V | 1,5 |
| V _{IH} | Input High Voltage | VCC x 0.7 | | VCC + 0.5 | V | 1,5 |
| V _{OL} | Output Low Voltage @ IOL = 3 mA | | | 0.4 | V | 1 |
| V _{HYS} | Input Hysteresis | VCC x .05 | | | V | 1,5 |

Notes

1. Referenced to VSS.
2. SCL toggling between VCC-0.3V and VSS, other inputs VSS or VCC-0.3V
3. SCL = SDA = VCC. All inputs VSS or VCC. Stop command issued.
4. VIN or VOUT = VSS to VCC
5. This parameter is characterized but not tested.

AC Parameters TA = -40° C to +85° C, VCC = 4.5V to 5.5V, CL = 100 pF unless otherwise specified

| Symbol | Parameter | Min | Max | Min | Max | Min | Max | Units | Notes |
|--------------|---|-----|------|-----|-----|------|------|-------|-------|
| t_{SCL} | SCL Clock Frequency | 0 | 100 | 0 | 400 | 0 | 1000 | kHz | |
| t_{LOW} | Clock Low Period | 4.7 | | 1.3 | | 0.6 | | μs | |
| t_{HIGH} | Clock High Period | 4.0 | | 0.6 | | 0.4 | | μs | |
| t_{AA} | SCL Low to SDA Data Out Valid | | 3 | | 0.9 | | 0.55 | μs | |
| t_{BUF} | Bus Free Before New Transmission | 4.7 | | 1.3 | | 0.5 | | μs | |
| $t_{HD:STA}$ | Start Condition Hold Time | 4.0 | | 0.6 | | 0.25 | | μs | |
| $t_{SU:STA}$ | Start Condition Setup for Repeated Start | 4.7 | | 0.6 | | 0.25 | | μs | |
| $t_{HD:DAT}$ | Data In Hold | 0 | | 0 | | 0 | | nS | |
| $t_{SU:DAT}$ | Data In Setup | 250 | | 100 | | 100 | | nS | |
| t_R | Input Rise Time | | 1000 | | 300 | | 300 | nS | 1 |
| t_F | Input Fall Time | | 300 | | 300 | | 100 | nS | 1 |
| $t_{SU:STO}$ | Stop Condition Setup | 4.0 | | 0.6 | | 0.25 | | μs | |
| t_{DH} | Data Output Hold (from SCL @ VIL) | 0 | | 0 | | 0 | | nS | |
| t_{SP} | Noise Suppression Time Constant on SCL, SDA | | 50 | | 50 | | 50 | nS | |

Notes : All SCL specifications as well as start and stop conditions apply to both read and write operations.

1 This parameter is periodically sampled and not 100% tested.

Capacitance TA = 25° C, f=1.0 MHz, VCC = 5V

| Symbol | Parameter | Max | Units | Notes |
|----------|--------------------------------|-----|-------|-------|
| C_{IO} | Input/output capacitance (SDA) | 8 | pF | 1 |
| C_{IN} | Input capacitance | 6 | pF | 1 |

Notes

1 This parameter is periodically sampled and not 100% tested.

AC Test Conditions

Input Pulse Levels VCC * 0.1 to VCC * 0.9
 Input rise and fall times 10 nS
 Input and output timing levels VCC*0.5

Equivalent AC Load Circuit

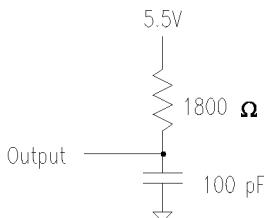
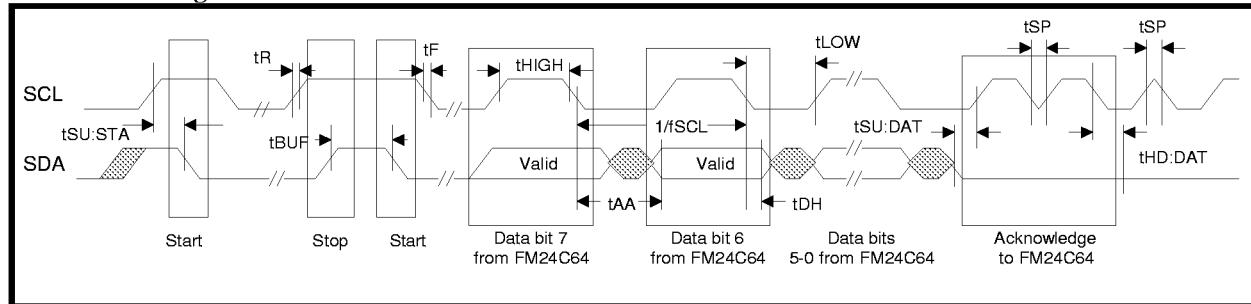
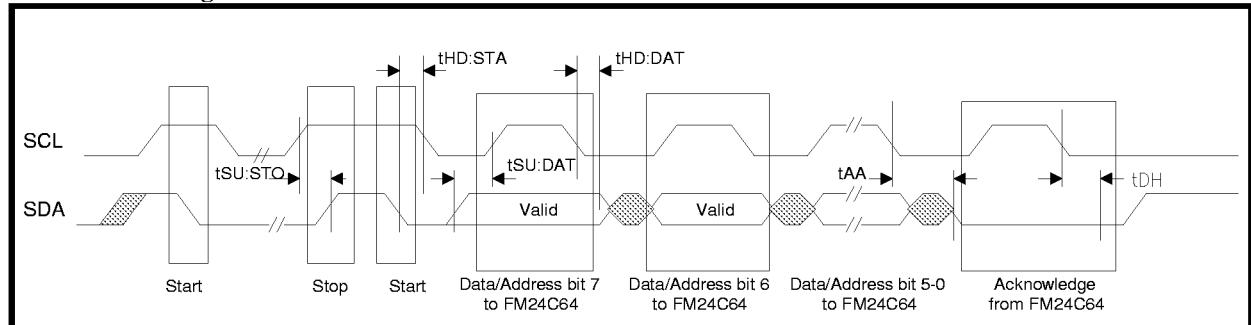


Diagram Notes

All start and stop timing parameters apply to both read and write cycles. Clock specifications are identical for read and write cycles. Write timing parameters apply to slave address, word address, and write data bits. Functional relationships are illustrated in the relevant data sheet sections. These diagrams illustrate the timing parameters only.

Read Bus Timing**Write Bus Timing**

Data Retention TA = -40° C to +85° C, VCC = 4.5V to 5.5V unless otherwise specified

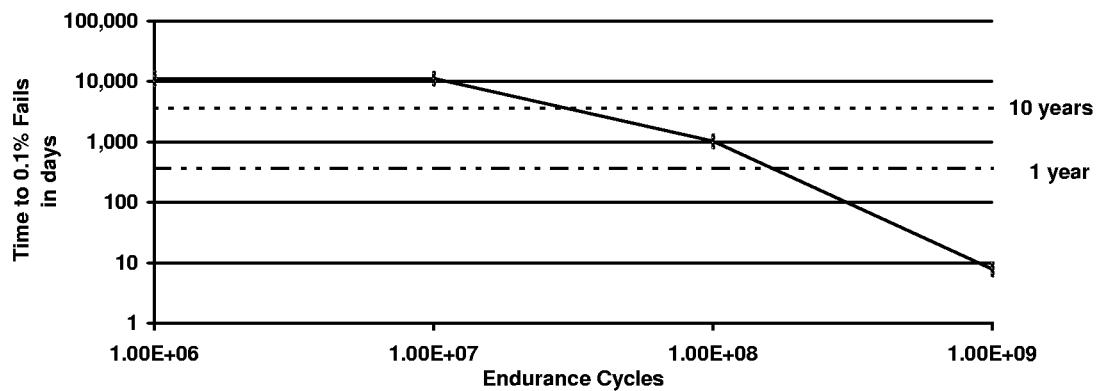
| Parameter | Min | Units | Notes |
|----------------|-----|-------|-------|
| Data Retention | 10 | years | 1.2 |

Notes

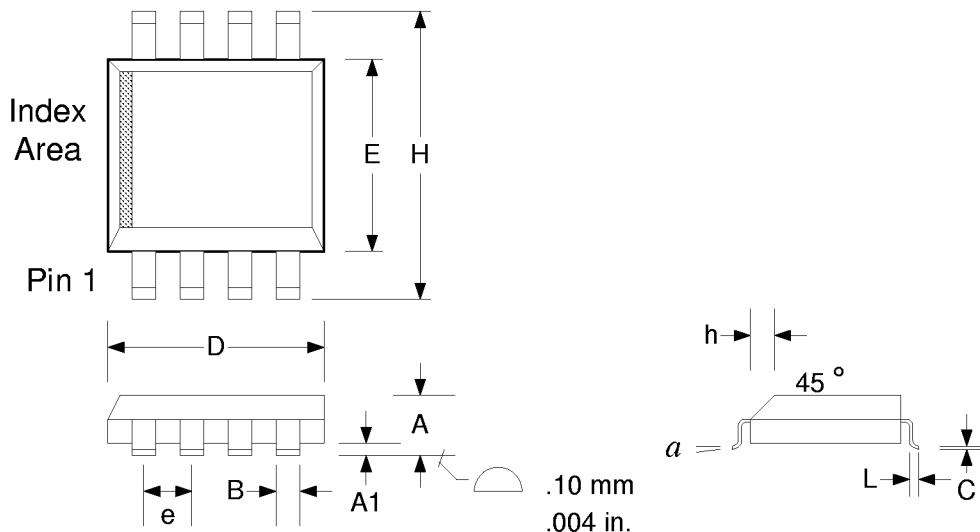
1. Data retention is specified at 55° C. The relationship between retention, temperature, and the associated reliability level is characterized in a separate reliability report.
2. Data retention reliability may be a function of cumulative read/write cycles as illustrated in the curve below.

Endurance

The impact of endurance cycles is to increase the soft error rate in data retention. The chart below illustrates the relationship. It shows the time to 0.1% parts with a data retention error. Also note that each read/write cycle affects an entire memory row for purposes of endurance.

Error Rate Versus Read/Write Cycles

8-pin SOP JEDEC MS-012



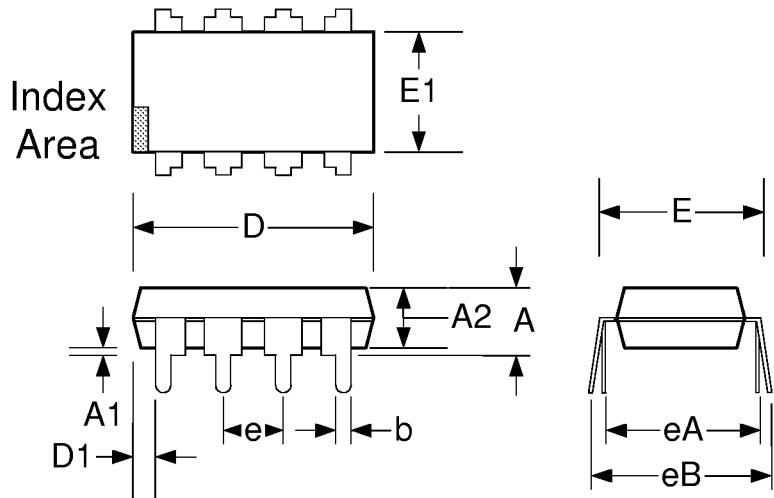
Selected Dimensions

Refer to JEDEC MS-012 for complete dimensions and notes.

Controlling dimensions is in millimeters. Conversions to inches are not exact.

| Symbol | Dim | Min | Nom. | Max |
|----------|-----------|--------------|----------------------|--------------|
| A | mm in. | 1.35 .053 | | 1.75 .069 |
| A1 | mm in. | .10 .004 | | .25 .010 |
| B | mm in. | .33 .013 | | .51 .020 |
| C | mm in. | .19 .007 | | .25 .010 |
| D | mm in. | 4.80 .189 | | 5.00 .197 |
| E | mm in. | 3.80 .150 | | 4.00 .157 |
| e | mm in. | | 1.27 BSC .050 BSC | |
| H | mm in. | 5.80 .228 | | 6.20 .244 |
| h | mm in. | .25 .010 | | .50 .197 |
| L | mm in. | .40 .016 | | 1.27 .050 |
| α | | 0° | | 8° |

8-pin DIP JEDEC MS-001



Selected Dimensions

Refer to JEDEC MS-001 for complete dimensions and notes.

Controlling dimensions is in inches. Conversions to millimeters are not exact.

| Symbol | Dim | Min | Nom. | Max |
|--------|-----------|---------------|----------------------|----------------|
| A | in. mm | | | .210 5.33 |
| A1 | in. mm | 0.015 .381 | | |
| A2 | in. mm | 0.115 2.92 | 0.130 3.30 | 0.195 4.95 |
| b | in. mm | 0.014 .356 | 0.018 .457 | 0.022 .508 |
| D | in. mm | 0.355 9.02 | 0.365 9.27 | 0.400 10.2 |
| D1 | in. mm | 0.005 .127 | | |
| E | in. mm | 0.300 7.62 | 0.310 7.87 | 0.325 8.26 |
| E1 | in. mm | 0.240 6.10 | 0.250 6.35 | 0.280 7.11 |
| e | in. mm | | .100 BSC 2.54 BSC | |
| eA | in. mm | | .300 BSC 7.62 BSC | |
| eB | in. mm | | | 0.430 10.92 |
| L | in. mm | 0.115 2.92 | 0.130 3.30 | 0.150 3.81 |