## **PRELIMINARY**

April 1994

## HPC16083/HPC26083/HPC36083/HPC46083/ HPC16003/HPC26003/HPC36003/HPC46003 High-Performance microControllers

## **General Description**

The HPC16083 and HPC16003 are members of the HPC™ family of High Performance microControllers. Each member of the family has the same core CPU with a unique memory and I/O configuration to suit specific applications. The HPC16083 has 8k bytes of on-chip ROM. The HPC16003 has no on-chip ROM and is intended for use with external direct memory. Each part is fabricated in National's advanced microCMOS technology. This process combined with an advanced architecture provides fast, flexible I/O control, efficient data manipulation, and high speed computation

The HPC devices are complete microcomputers on a single chip. All system timing, internal logic, ROM, RAM, and I/O are provided on the chip to produce a cost effective solution for high performance applications. On-chip functions such as UART, up to eight 16-bit timers with 4 input capture registers, vectored interrupts, WATCHDOG™ logic and MICRO-WIRE/PLUS™ provide a high level of system integration. The ability to address up to 64k bytes of external memory enables the HPC to be used in powerful applications typically performed by microprocessors and expensive peripheral chips. The term "HPC16083" is used throughout this datasheet to refer to the HPC16083 and HPC16003 devices unless otherwise specified.

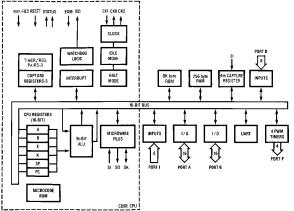
The microCMOS process results in very low current drain and enables the user to select the optimum speed/power product for his system. The IDLE and HALT modes provide further current savings. The HPC is available in 68-pin PLCC, LDCC, PGA and 80-Pin PQFP packages.

#### **Features**

- HPC family—core features:
  - 16-bit architecture, both byte and word
  - 16-bit data bus, ALU, and registers
  - 64k bytes of external direct memory addressing
  - FAST—200 ns for fastest instruction when using 20.0 MHz clock, 134 ns at 30 MHz
  - High code efficiency—most instructions are single byte
  - 16 x 16 multiply and 32 x 16 divide
  - Eight vectored interrupt sources
  - Four 16-bit timer/counters with 4 synchronous outputs and WATCHDOG logic
  - MICROWIRE/PLUS serial I/O interface
  - CMOS—very low power with two power save modes: IDLE and HALT
- UART—full duplex, programmable baud rate
- Four additional 16-bit timer/counters with pulse width modulated outputs
- Four input capture registers
- 52 general purpose I/O lines (memory mapped)
- 8k bytes of ROM, 256 bytes of RAM on chip
- ROMless version available (HPC16003)
- Commercial (0°C to +70°C), industrial (-40°C to +85°C), automotive (-40°C to +105°C) and military (-55°C to +125°C) temperature ranges

For applications requiring more RAM and ROM see HPC16064 data sheet.

Block Diagram (HPC16083 with 8k ROM shown)



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## **Absolute Maximum Ratings**

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Total Allowable Source or Sink Current
Storage Temperature Range
Lead Temperature (Soldering, 10 sec)

100 mA
-65°C to +150°C
300°C

 $\begin{array}{lll} \text{V}_{\text{CC}} \text{ with Respect to GND} & -0.5 \text{V to } 7.0 \text{V} \\ \text{All Other Pins} & (\text{V}_{\text{CC}} + 0.5) \text{V to } (\text{GND} - 0.5) \text{V} \\ \text{Note: } \textit{Absolute maximum ratings indicate limits beyond} \end{array}$ 

Note: Absolute maximum ratings indicate limits beyond which damage to the device may occur. DC and AC electrical specifications are not ensured when operating the device at absolute maximum ratings.

**DC Electrical Characteristics**  $V_{CC} = 5.0V \pm 10\%$  unless otherwise specified,  $T_A = 0^{\circ}C$  to  $+70^{\circ}C$  for HPC46083/HPC46003,  $-40^{\circ}C$  to  $+85^{\circ}C$  for HPC36083/HPC36003,  $-40^{\circ}C$  to  $+105^{\circ}C$  for HPC26083/HPC26003,  $-55^{\circ}C$  to  $+125^{\circ}C$  for HPC16083/HPC16003

Symbol	Parameter	Test Conditions	Min	Max	Units
I <sub>CC1</sub>	Supply Current	V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 30 MHz (Note 1)		65	mA
·		V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 20 MHz (Note 1)		47	mA
		V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 2.0 MHz (Note 1)		10	mA
I <sub>CC2</sub>	IDLE Mode Current	V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 30 MHz (Note 1)		5.0	mA
_		V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 20 MHz, (Note 1)		3.0	mA
		V <sub>CC</sub> = 5.5V, f <sub>in</sub> = 2.0 MHz, (Note 1)		1	mA
I <sub>CC3</sub>	HALT Mode Current	$V_{CC} = 5.5V$ , $f_{in} = 0$ kHz, (Note 1)		200	μΑ
		$V_{CC} = 2.5V$ , $f_{in} = 0$ kHz, (Note 1)		50	μΑ
INPUT VO	LTAGE LEVELS FOR SCHMITT TRIC	GGERED INPUTS $\overline{\text{RESET}}$ , NMI AND $\overline{\text{WO}}$ ;	AND ALSO CKI		
V <sub>IH1</sub>	Logic High		0.9 V <sub>CC</sub>		V
V <sub>IL1</sub>	Logic Low			0.1 V <sub>CC</sub>	V
INPUT VO	DLTAGE LEVELS FOR ALL OTHER IN	IPUTS			
V <sub>IH2</sub>	Logic High		0.7 V <sub>CC</sub>		V
$V_{IL_2}$	Logic Low			0.2 V <sub>CC</sub>	V
I <sub>LI1</sub>	Input Leakage Current	V <sub>IN</sub> = 0 and V <sub>IN</sub> = V <sub>CC</sub>		±2	μΑ
I <sub>LI2</sub>	Input Leakage Current RDY/HLD, EXUI	V <sub>IN</sub> = 0	-3	-50	μΑ
I <sub>LI3</sub>	Input Leakage Current B12	$\overline{\text{RESET}} = 0, V_{\text{IN}} = V_{\text{CC}}$	0.5	7	mA
C <sub>I</sub>	Input Capacitance	(Note 2)		10	pF
C <sub>IO</sub>	I/O Capacitance	(Note 2)		20	pF
OUTPUT	VOLTAGE LEVELS				
$V_{OH_1}$	Logic High (CMOS)	$I_{OH} = -10 \mu\text{A} \text{ (Note 2)}$	V <sub>CC</sub> - 0.1		V
$V_{OL_1}$	Logic Low (CMOS)	I <sub>OH</sub> = 10 μA (Note 2)		0.1	V
$V_{OH_2}$	Port A/B Drive, CK2	$I_{OH} = -7 \text{ mA}$	2.4		V
$V_{OL_2}$	(A <sub>0</sub> -A <sub>15</sub> , B <sub>10</sub> , B <sub>11</sub> , B <sub>12</sub> , B <sub>15</sub> )	$I_{OL} = 3 \text{ mA}$		0.4	V
$V_{OH_3}$	Other Port Pin Drive, WO (open	$I_{OH} = -1.6 \text{ mA (except } \overline{WO}\text{)}$	2.4		V
V <sub>OL3</sub>	drain), (B <sub>0</sub> -B <sub>9</sub> , B <sub>13</sub> , B <sub>14</sub> , P <sub>0</sub> -P <sub>3</sub> )	$I_{OL} = 0.5 \text{ mA}$		0.4	V
V <sub>OH4</sub>	ST1 and ST2 Drive	$I_{OH} = -6 \text{ mA}$	2.4		V
V <sub>OL4</sub>		I <sub>OL</sub> = 1.6 mA		0.4	V
V <sub>OH5</sub>	Port A/B Drive (A <sub>0</sub> -A <sub>15</sub> ,	$I_{OH} = -1 \text{ mA}$	2.4		V
V <sub>OL5</sub>	B <sub>10</sub> , B <sub>11</sub> , B <sub>12</sub> , B <sub>15</sub> ) when used as External Address/Data Bus	I <sub>OL</sub> = 3 mA		0.4	V
V <sub>RAM</sub>	RAM Keep-Alive Voltage	(Note 3)	2.5	V <sub>CC</sub>	V
loz	TRI-STATE® Leakage Current	V <sub>IN</sub> = 0 and V <sub>IN</sub> = V <sub>CC</sub>		±5	μΑ

Note 1:  $I_{CC_1}$ ,  $I_{CC_2}$ ,  $I_{CC_3}$  measured with no external drive ( $I_{OH}$  and  $I_{OL} = 0$ ,  $I_{IH}$  and  $I_{IL} = 0$ ).  $I_{CC_1}$  is measured with  $\overline{\text{RESET}} = V_{SS}$ ,  $I_{CC_3}$  is measured with  $\overline{\text{NMI}} = V_{CC}$ . CKI driven to  $V_{IH1}$  and  $V_{IL1}$ , with rise and fall times less than 10 ns.

Note 2: This is guaranteed by design and not tested.

Note 3: Test duration is 100 ms.

## **AC Electrical Characteristics**

(See Notes 1 and 4 and Figure 1 thru Figure 5)  $V_{CC}=5.0V\pm10\%$  unless otherwise specified,  $T_A=0^{\circ}C$  to  $+70^{\circ}C$  for HPC46083/HPC46003,  $-40^{\circ}C$  to  $+85^{\circ}C$  for HPC36083/HPC36003,  $-40^{\circ}C$  to  $+105^{\circ}C$  for HPC26083/HPC26003,  $-55^{\circ}C$  to  $+125^{\circ}C$  for HPC16083/HPC16003

	Symbol and Formula	Parameter	Min	Max	Units	Note
	$f_{\mathbb{C}}$	CKI Operating Frequency	2	20	MHz	
	$t_{C1} = 1/f_{C}$	CKI Clock Period	50	500	ns	
	t <sub>CKIH</sub>	CKI High Time	22.5		ns	
	tCKIL	CKI Low Time	22.5		ns	
	$t_{\rm C} = 2/f_{\rm C}$	CPU Timing Cycle	100		ns	
, s	$t_{WAIT} = t_{C}$	CPU Wait State Period	100		ns	
Clocks	<sup>t</sup> DC1C2R	Delay of CK2 Rising Edge after CKI Falling Edge	0	55	ns	(Note 2)
	t <sub>DC1C2F</sub>	Delay of CK2 Falling Edge after				
	DC102F	CKI Falling Edge	0	55	ns	(Note 2)
	$f_U = f_C/8$	External UART Clock Input Frequency		2.5**	MHz	
	f <sub>MW</sub>	External MICROWIRE/PLUS				
	-10100	Clock Input Frequency		1.25	MHz	
Timers	$f_{XIN} = f_C/22$	External Timer Input Frequency		0.91	MHz	
Ě	$t_{XIN} = t_{C}$	Pulse Width for Timer Inputs	100		ns	
		MODOWIDE OF THE MAN	100			
1	t <sub>UWS</sub>	MICROWIRE Setup Time—Master —Slave	100 20		ns	
SE						
	tuwh	MICROWIRE Hold Time—Master —Slave	20 50		ns	
MICROWIRE/ PLUS			50			
Σ	$t_{\sf UWV}$	MICROWIRE Output Valid Time—Master		50	ns	
		—Slave		150		
흥	$t_{SALE} = \frac{3}{4} t_{C} + 40$	HLD Falling Edge before ALE Rising Edge HLD Pulse Width	115		ns	
Ĕ	$t_{HWP} = t_{C} + 10$ $t_{HAE} = t_{C} + 100$	HLD Pulse Width HLDA Falling Edge after HLD Falling Edge	110	200	ns	(Note 3)
'n	$t_{HAD} = \frac{3}{4} t_{C} + 85$	HLDA Rising Edge after HLD Rising Edge		160	ns ns	(Note 3)
External Hold	$t_{BF} = \frac{1}{2} t_{C} + 66$	Bus Float after HLDA Falling Edge		116	ns	(Note 5)
Û	$t_{BE} = \frac{1}{2} t_{C} + 66$	Bus Enable after HLDA Rising Edge	116	110	ns	(Note 5)
	tuas	Address Setup Time to Falling Edge of URD	10		ns	
	t <sub>UAH</sub>	Address Hold Time from Rising Edge of URD	10		ns	
	t <sub>RPW</sub>	URD Pulse Width	100		ns	
6	t <sub>OE</sub>	URD Falling Edge to Output Data Valid	0	60	ns	
UPI Timing	t <sub>OD</sub>	Rising Edge of URD to Output Data Invalid	5	35	ns	(Note 6)
I	t <sub>DRDY</sub>	RDRDY Delay from Rising Edge of URD		70	ns	
د	t <sub>WDW</sub>	UWR Pulse Width	40		ns	
	t <sub>UDS</sub>	Input Data Valid before Rising Edge of UWR	10		ns	
	t <sub>UDH</sub>	Input Data Hold after Rising Edge of UWR	20		ns	
	t <sub>A</sub>	WRRDY Delay from Rising Edge of UWR		70	ns	

<sup>\*\*</sup>This maximum frequency is attainable provided that this external baud clock has a duty cycle such that the high period includes two (2) falling edges of the CK2 clock.

## **AC Electrical Characteristics**

(See Notes 1 and 4 and Figure 1 thru Figure 5)  $V_{CC}=5.0V\pm10\%$  unless otherwise specified,  $T_A=0^{\circ}C$  to  $+70^{\circ}C$  for HPC46083/HPC46003,  $-40^{\circ}C$  to  $+85^{\circ}C$  for HPC36083/HPC36003,  $-40^{\circ}C$  to  $+105^{\circ}C$  for HPC16083/HPC16003 (Continued)

	Symbol and Formula	Parameter	Min	Max	Units	Note
	<sup>t</sup> DC1ALER	Delay from CKI Rising Edge to ALE Rising Edge	0	35	ns	(Notes 1, 2)
	<sup>t</sup> DC1ALEF	Delay from CKI Rising Edge to ALE Falling Edge	0	35	ns	(Notes 1, 2)
Cycles	$t_{DC2ALER} = \frac{1}{4}t_{C} + 20$	Delay from CK2 Rising Edge to ALE Rising Edge		45	ns	(Note 2)
Address Cycles	$t_{DC2ALEF} = \frac{1}{4}t_{C} + 20$	Delay from CK2 Rising Edge to ALE Rising Edge		45	ns	(Note 2)
Ą	$t_{LL} = \frac{1}{2}t_C - 9$	ALE Pulse Width	41		ns	
	$t_{ST} = \frac{1}{4} t_C - 7$	Setup of Address Valid before ALE Falling Edge	18		ns	
	$t_{VP} = \frac{1}{4} t_C - 5$	Hold of Address Valid after ALE Falling Edge	20		ns	
	$t_{ARR} = \frac{1}{4}t_{C} - 5$	ALE Falling Edge to RD Falling Edge	20		ns	
s	$t_{ACC} = t_C + WS - 55$	Data Input Valid after Address Output Valid		145	ns	(Note 6)
Read Cycles	$t_{RD} = \frac{1}{2} t_{C} + WS - 65$	Data Input Valid after RD Falling Edge		95	ns	
3eac	$t_{RW} = \frac{1}{2} t_{C} + WS - 10$	RD Pulse Width	140		ns	
_	$t_{DR} = \frac{3}{4} t_{C} - 15$	Hold of Data Input Valid after RD Rising Edge	0	60	ns	
	$t_{RDA} = t_{C} - 15$	Bus Enable after RD Rising Edge	85		ns	
s	$t_{ARW} = \frac{1}{2} t_C - 5$	ALE Falling Edge to WR Falling Edge	45		ns	
)Cle	$t_{WW} = \frac{3}{4} t_{C} + WS - 15$	WR Pulse Width	160		ns	
Write Cycles	$t_V = \frac{1}{2} t_C + WS - 5$	Data Output Valid before WR Rising Edge	145		ns	
>	$t_{HW} = \frac{1}{4} t_{C} - 5$	Hold of Data Valid after WR Rising Edge	20		ns	
Ready Input	$t_{DAR} = \frac{1}{4} t_{C} + WS - 50$	Falling Edge of ALE to Falling Edge of RDY		75	ns	
œ=	$t_{RWP} = t_{C}$	RDY Pulse Width	100		ns	

Note: C<sub>L</sub> = 40 pF.

Note 1: These AC characteristics are guaranteed with external clock drive on CKI having 50% duty cycle and with less than 15 pF load on CKO with rise and fall times (t<sub>CKIR</sub> and T<sub>CKIL</sub>) on CKI input less than 2.5 ns.

Note 2: Do not design with these parameters unless CKI is driven with an active signal. When using a passive crystal circuit, its stability is not guaranteed if either CKI or CKO is connected to any external logic other than the passive components of the crystal circuit.

Note 3: t<sub>HAE</sub> is spec'd for case with HED falling edge occurring at the latest time it can be accepted during the present CPU cycle being executed. If HED falling edge occurs later, t<sub>HAE</sub> as long as (3t<sub>C</sub> + 4WS + 72 t<sub>C</sub> + 100) may occur depending on the following CPU instruction cycles, its wait state and ready input.

Note 4: WS (t<sub>WAIT</sub>) x (number of preprogrammed wait states). Minimum and maximum values are calculated at maximum operating frequency, t<sub>C</sub> = 20 MHz, with one wait programmed.

Note 5: Due to emulation restrictions—actual limits will be better.

Note 6: This is guaranteed by design and not tested.

AC Electrical Characteristics (Continued) (See Notes 1 and 4 and Figure 1 thru Figure 5)  $V_{CC}=5.0V\pm10\%$  unless otherwise specified,  $T_A=0^{\circ}C$  to  $+70^{\circ}C$  for HPC46083/HPC46003,  $-40^{\circ}C$  to  $+85^{\circ}C$  for HPC36083/HPC36003,  $-40^{\circ}C$  to  $+105^{\circ}C$  for HPC16083/HPC16003

	Symbol and Formula	Parameter	Min	Max	Units	Note
	$f_{C}$ $t_{C1} = 1/f_{C}$ $t_{CKIH}$ $t_{CKIL}$	CKI Operating Frequency CKI Clock Period CKI High Time CKI Low Time	2 33 15 16.6	30 500	MHz ns ns	
Clocks	$t_C = 2/f_C$ $t_{WAIT} = t_C$ $t_{DC1C2R}$	CPU Timing Cycle CPU Wait Sate Period Delay of CK2 Rising Edge after	66 66		ns ns	(NI=+= 0)
ฮั	t <sub>DC1C2F</sub>	CKI Falling Edge Delay of CK2 Falling Edge after CKI Falling Edge	0	55 55	ns ns	(Note 2)
	$f_U = f_C/8$ $f_{MW}$	External UART Clock Input Frequency External MICROWIRE/PLUS Clock Input Frequency		3.75** 1.875	MHz MHz	
Timers	$f_{XIN} = f_C/22$ $t_{XIN} = t_C$	External Timer Input Frequency Pulse Width for Timer Inputs	66	1.364	MHz ns	
3E/	t <sub>UWS</sub>	MICROWIRE Setup Time—Master —Slave	100 20		ns	
MICROWIRE/ PLUS	t <sub>UWH</sub>	MICROWIRE Hold Time—Master —Slave	20 50		ns	
MIC	$t_{\sf UWV}$	MICROWIRE Output Valid Time—Master —Slave		50 150	ns	
External Hold	$t_{SALE} = \frac{3}{4}t_{C} + 40$ $t_{HWP} = t_{C} + 10$ $t_{HAE} = t_{C} + 85$ $t_{HAD} = \frac{3}{4}t_{C} + 85$ $t_{BF} = \frac{1}{2}t_{C} + 66$ $t_{BE} = \frac{1}{2}t_{C} + 66$	HLD Falling Edge before ALE Rising Edge HLD Pulse Width HLDA Falling Edge after HLD Falling Edge HLDA Rising Edge after HLD Rising Edge Bus Float after HLDA Falling Edge Bus Enable after HLDA Rising Edge	90 76 99	151 135 99	ns ns ns ns ns	(Note 3) (Note 5) (Note 5)
	tuas	Address Setup Time to Falling Edge of URD	10		ns	, ,
Ī	t <sub>UAH</sub>	Address Hold Time from Rising Edge of URD	10		ns	
	t <sub>RPW</sub>	URD Pulse Width	100		ns	
	t <sub>OE</sub>	URD Falling Edge to Output Data Valid	0	60	ns	
JPI Timing	t <sub>OD</sub>	Rising Edge of URD to Output Data Invalid	5	35	ns	(Note 6)
l l	t <sub>DRDY</sub>	RDRDY Delay from Rising Edge of URD		70	ns	
	$t_{WDW}$	UWR Pulse Width	40		ns	
	t <sub>UDS</sub>	Input Data Valid before Rising Edge of UWR	10		ns	
	t <sub>UDH</sub>	Input Data Hold after Rising Edge of UWR	15		ns	
	t <sub>A</sub>	WRRDY Delay from Rising Edge of UWR		70	ns	

<sup>\*\*</sup>This maximum frequency is attainable provided that this external baud clock has a duty cycle such that the high period includes two (2) falling edges of the CK2 clock.

## **AC Electrical Characteristics**

(See Notes 1 and 4 and Figure 1 thru Figure 5)  $V_{CC} = 5.0V \pm 10\%$  unless otherwise specified,  $T_A = 0^{\circ}C$  to  $+70^{\circ}C$  for HPC46083/HPC46003,  $-40^{\circ}C$  to  $+85^{\circ}C$  for HPC36083/HPC36003,  $-40^{\circ}C$  to  $+105^{\circ}C$  for HPC26083/HPC26003,  $-55^{\circ}C$  to  $+125^{\circ}C$  for HPC16083/HPC16003 (Continued)

	Symbol and Formula	Parameter	Min	Max	Units	Notes
	<sup>t</sup> DC1ALER	Delay from CKI Rising Edge to ALE Rising Edge	0	35	ns	(Notes 1, 2)
es	<sup>t</sup> DC1ALEF	Delay from CKI Rising Edge to ALE Falling Edge	0	35	ns	(Notes 1, 2)
Cycles	$t_{DC2ALER} = \frac{1}{4}t_{C} + 20$	Delay from CK2 Rising Edge to ALE Rising Edge		37	ns	(Note 2)
	$t_{DC2ALEF} = \frac{1}{4} t_{C} + 20$	Delay from CK2 Falling Edge to ALE Falling Edge		37	ns	(Note 2)
Address	$t_{LL} = \frac{1}{2} t_{C} - 9$	ALE Pulse Width	24		ns	
¥	$t_{ST} = \frac{1}{4} t_{C} - 7$	Setup of Address Valid before ALE Falling Edge	9		ns	
	$t_{VP} = \frac{1}{4} t_{C} - 5$	Hold of Address Valid after ALE Falling Edge	11		ns	
	$t_{ARR} = \frac{1}{4}t_{C} - 5$	ALE Falling Edge to RD Falling Edge	12		ns	
es	$t_{ACC} = t_C + WS - 32$	Data Input Valid after Address Output Valid		100	ns	(Note 6)
Cycles	$t_{RD} = \frac{1}{2} t_{C} + WS - 39$	Data Input Valid after RD Falling Edge		60	ns	
Read	$t_{RW} = \frac{1}{2} t_{C} + WS - 14$	RD Pulse Width	85		ns	
8	$t_{DR} = \frac{3}{4} t_{C} - 15$	Hold of Data Input Valid after RD Rising Edge	0	35	ns	
	$t_{RDA} = t_{C} - 15$	Bus Enable after RD Rising Edge	51		ns	
	$t_{ARW} = \frac{1}{2} t_{C} - 5$	ALE Falling Edge to WR Falling Edge	28		ns	
Write Cycles	$t_{WW} = \frac{3}{4} t_{C} + WS - 15$	WR Pulse Width	101		ns	
خَ≷	$t_V = \frac{1}{2} t_C + WS - 5$	Data Output Valid before WR Rising Edge	94		ns	
	$t_{HW} = \frac{1}{4} t_C - 10$	Hold of Data Valid after WR Rising Edge	7		ns	
Ready Input	$t_{DAR} = \frac{1}{4} t_{C} + WS - 50$	Falling Edge of ALE to Falling Edge of RDY		33	ns	
P. Re	$t_{RWP} = t_{C}$	RDY Pulse Width	66		ns	

Note:  $C_L = 40 \text{ pF}.$ 

Note 1: These AC characteristics are guaranteed with external clock drive on CKI having 50% duty cycle and with less than 15 pF load on CKO wih rise and fall times (t<sub>CKIR</sub> and t<sub>CKIL</sub>) on CKI input less than 2.5 ns.

Note 2: Do not design with these parameters unless CKI is driven with an active signal. When using a passive crystal circuit, its stability is not guaranteed if either CKI or CKO is connected to any external logic other than the passive components of the crystal circuit.

Note 3: t<sub>HAE</sub> is spec'd for case with  $\overline{HLD}$  falling edge occurring at the latest time it can be accepted during the present CPU cycle being executed. If  $\overline{HLD}$  falling edge occurs later, t<sub>HAE</sub> as long as (3t<sub>C</sub> + 4WS + 72 t<sub>C</sub> + 100) may occur depending on the following CPU instruction cycles, its wait states and ready input.

Note 4: WS  $t_{WAIT} \times$  (number of pre-programmed wait states). Minimum and maximum values are calculated from maximum operating frequency,  $t_C = 30$  MHz, with one wait state programmed.

Note 5: Due to emulation restrictions—actual limits will be better.

Note 6: This is guaranteed by design and not tested.

## **CKI Input Signal Characteristics**

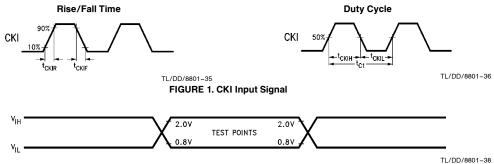
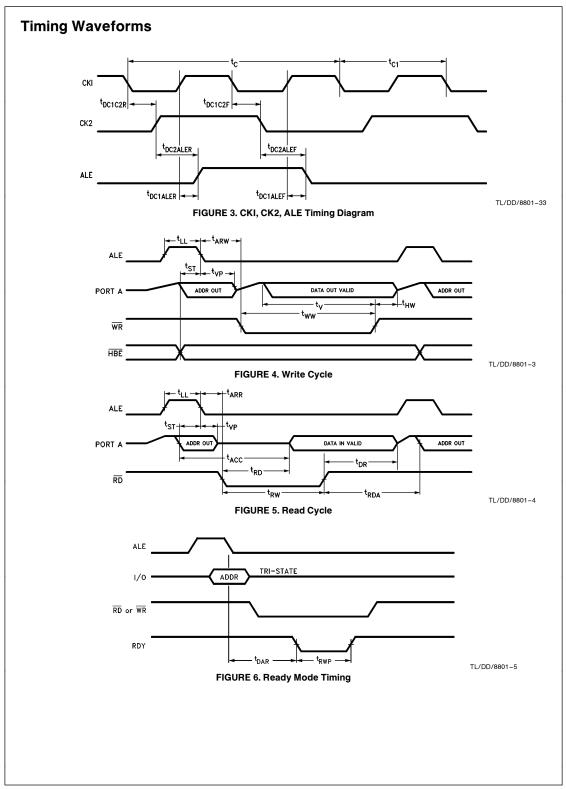
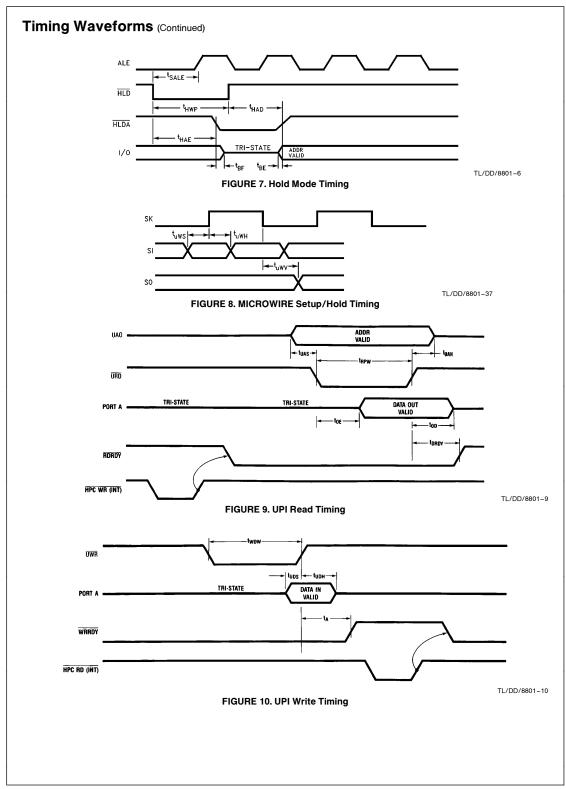


FIGURE 2. Input and Output for AC Tests

Note: AC testing inputs are driven at V<sub>IH</sub> for a logic "1" and V<sub>IL</sub> for a logic "0". Output timing measurements are made at 2.0V for a logic "1" and 0.8V for a logic "0".





The following is the Military 883 Electrical Specification for HPC16083 and HPC16003. For latest information on RETS 16083X contact NSC local sales office.

## **DC Electrical Specifications** Test Conditions $V_{CC} = 5V \pm 10\%$ (Unless Otherwise Specified) (Note 1)

Symbol	Parameter	neter Conditions		RP 1 5°C	SBGRP 2 + 125°C			RP 3 5°C	Units	Notes
•			Min	Max	Min	Max	Min	Max		
V <sub>IH1</sub>	Logical "1" Input	RESET, NMI, CKI and WO	0.9		0.9		0.9		V	
$V_{\text{IH2}}$	Voltage	B <sub>10</sub> -B <sub>13</sub> , B <sub>15</sub> All Inputs except Port A	(V <sub>CC</sub> ) 0.7 (V <sub>CC</sub> )		(V <sub>CC</sub> ) 0.7 (V <sub>CC</sub> )		(V <sub>CC</sub> ) 0.7 (V <sub>CC</sub> )		v	
$V_{IH3}$		Port A, V <sub>CC</sub> = 5.5V	4.65		4.65		4.65		V	(Note 2)
		Port A, V <sub>CC</sub> = 4.5V	3.95		3.95		3.95		V	(Note 2)
V <sub>IL1</sub>	Logical "0" Input Voltage	RESET, NMI, CKI and WO		0.1 (V <sub>CC</sub> )		0.1 (V <sub>CC</sub> )		0.1 (V <sub>CC</sub> )	٧	
$V_{IL2}$		All Inputs except Port A		0.2 (V <sub>CC</sub> )		0.2 (V <sub>CC</sub> )		0.2 (V <sub>CC</sub> )	V	
$V_{\text{IL3}}$		Port A, V <sub>CC</sub> = 5.5V Port A, V <sub>CC</sub> = 4.5V		0.7		0.7		0.7	V V	(Note 3) (Note 3)
V <sub>OH2</sub>	Logical "1" Output Voltage	$I_{OH} = -7 \text{ mA } (A_0 - A_{15}, B_{10} - B_{12}, B_{15}, CK2)$	2.4		2.4		2.4		٧	
$V_{OH3}$		$I_{OH3} = -1.6 \text{ mA } (B_0 - B_9, B_{13} - B_{14}, P_0 - P_3), \overline{WO}$ (Open Drain)	2.4		2.4		2.4		V	
V <sub>OH4</sub>		$I_{OH} = -6 \text{ mA (ST1, ST2)}$	2.4		2.4		2.4		V	
V <sub>OH5</sub>		$I_{OH} = -1$ mA ( $A_0$ - $A_{15}$ , $B_{10}$ - $B_{12}$ , $B_{15}$ ) When Used as an External Address/Data Bus	2.4		2.4		2.4		٧	
V <sub>OL2</sub>	Logical "0" Output	$I_{OL} = 3 \text{ mA (CK2, } A_0 - A_{15}, B_{10} - B_{12}, B_{15})$		0.4		0.4		0.4	٧	
$V_{OL3}$	Voltage	$  I_{OL} = 0.5 \text{ mA } (B_0 - B_9, B_{13} - B_{14}, P_0 - P_3)  $   WO (Open Drain)		0.4		0.4		0.4	V	
$V_{OL4}$ $V_{OL5}$		I <sub>OL</sub> = 1.6 mA (ST1, ST2) I <sub>OL</sub> = 3 mA (A <sub>0</sub> -A <sub>15</sub> , B <sub>10</sub> -B <sub>12</sub> , B <sub>15</sub> )		0.4		0.4		0.4	٧	
		When Used as an External Address/Data Bus		0.4		0.4		0.4	V	
loz	TRI-STATE Leakage	$V_{SS} \le V_{IN} \le V_{CC}$ ( $\overline{WO}$ , Port A, Port B), $V_{CC} = 5.5V$		±5		±5		±5	μΑ	
I <sub>LI1</sub>	Input Leakage Current	$\begin{aligned} &V_{SS} \leq V_{IN} \leq V_{CC}, V_{CC} = 5.5V \\ &(I_1 - I_6, D_0 - D_7, CKI, \\ &RESET, EXM, EI) \end{aligned}$		±2		±2		±2	μΑ	(Note 7)
I <sub>LI2</sub>	Input Pullup Current	$V_{IN} = 0$ (I <sub>0</sub> , I <sub>7</sub> , RDY/HLD, EXUI), $V_{CC} = 5.5V$	-50	-3	-50	-3	-50	-3	μΑ	(Note 7)
I <sub>LI3</sub>	Port B <sub>12</sub> Pulldown during Reset	$V_{IN} = V_{CC}$ , Port B <sub>12</sub> , $V_{CC} = 5.5V$	1	7	1	7	1	7	mA	
VRAM	RAM Keep Alive Voltage	Test Duration is 10 ms	2.5		2.5		2.5		٧	
I <sub>CC1</sub>	Supply Current Dynamic	$F_{IN} = 20 \text{ MHz}, \overline{RESET} = V_{SS},$ $I_{OH} = 0 \text{ mA}, I_{OL} = 0 \text{ mA}, V_{CC} = 5.5V$		55		55		55	mA	
I <sub>CC2</sub>	Idle Mode Current	F <sub>IN</sub> = 20 MHz, External Clock		3.5		3.5		3.5	mA	
Icc	Halt Mode Current	NMI = V <sub>CC</sub>		2		2		2	mA	
CI/O	Input/Output Capacitance	f <sub>test</sub> = 1.0 MHz, I/O Pin to Ground		20					pF	(Note 4)
			SBG	RP4						
CI	Input Capacitance	$f_{test} = 1.0 \text{ MHz},$ Input Pin to Ground		10					pF	(Note 4)

Note 1: Electrical end point testing (when required) for Groups C & D shall consist only of subgroups 1, 2, 9 and 10.

 $<sup>\</sup>textbf{Note 2:} \ \mathsf{Port} \ \mathsf{A} \ \mathsf{V}_{IH} \ \mathsf{test} \ \mathsf{limit} \ \mathsf{includes} \ \mathsf{700} \ \mathsf{mV} \ \mathsf{offset} \ \mathsf{caused} \ \mathsf{by} \ \mathsf{output} \ \mathsf{loads} \ \mathsf{being} \ \mathsf{on} \ \mathsf{during} \ \mathsf{Data} \ \mathsf{Drive} \ \mathsf{Time}.$ 

 $<sup>\</sup>textbf{Note 3:} \ \text{Port A V}_{\text{IL}} \ \text{test limit includes 400 mV offset caused by output loads being on during Data Drive Time.}$ 

Note 4: Verified at initial qual only.

Note 7: Future revisions of this device will not have pullups on pins  $I_0$ ,  $I_7$  which will be tested to  $I_{L11}$  conditions.

## $\textbf{AC Electrical Specifications} \ \ \text{Test Conditions V}_{CC} = 4.5 \text{V and } 5.5 \text{V (Unless Otherwise Specified) (Note 1)}$

Symbol	Parameter	Conditions	l .	RP 9 5°C		RP 10 125°C			Units	Notes
•			Min Max I		Min	Max	Min	Max		
f <sub>C</sub> = CKI Freq.	Operating Frequency		2	20	2	20	2	20	MHz	(Note 5)
$t_{CI} = 1/FC$	Clock Period		50		50		50		ns	(Note 5)
$t_{C} = 2/FC$	Timing Cycle		100		100		100		ns	(Note 5)
$t_{LL} = \frac{1}{2} t_{C} - 9$	ALE Pulse Width		41		41		41		ns	(Note 6)
$t_{ST} = \frac{1}{4}t_C - 7$	Address Valid to ALE Falling Edge		18		18		18		ns	(Note 6)
$t_{WAIT} = t_C = WS$	Wait State Period		100		100		100		ns	(Note 5)
$FMW = 0.0625 f_{C}$	External MICROWIRE/PLUS CLK Input Frequency			1.25		1.25		1.25	MHz	(Note 6)
$f_{U} = 0.125 f_{C}$	External UART Clock Input Frequency			2.5		2.5		2.5	MHz	(Note 5)
t <sub>DCIC2</sub>	CK2 Delay From CK1			55		55		55	ns	(Note 6)
$t_{ARR} = \frac{1}{4}t_{C} - 5$	ALE Falling Edge to RD Falling Edge		20		20		20		ns	(Note 6)
$t_{RW} = \frac{1}{2}$ $t_{C} + WS - 10$	RD Pulse Width		140		140		140		ns	(Note 6)
$t_{DR} = 3.4 t_{C} - 15$	Data Hold after Rising Edge of RD		0	60	0	60	0	60	ns	(Note 6)
$t_{RD} = \frac{1}{2}$ $t_{C} + WS - 65$	RD Falling Edge to Data in Valid			85		85		85	ns	(Note 6)
$t_{RDA} = t_C - 15$	RD Rising Edge to Address Valid		85		85		85		ns	(Note 6)
$t_{VP} = \frac{1}{4} t_{C} - 5$	Address Hold from ALE Falling Edge		20		20		20		ns	(Note 6)
$t_{ARW} = \frac{1}{2}t_{C} - 5$	ALE Trailing Edge to WR Falling Edge		45		45		45		ns	(Note 6)
$t_{WW} = \frac{3}{4} t_{C} + WS - 15$	WR Pulse Width		160		160		160		ns	(Note 6)
$t_{HW} = \frac{1}{4} t_{C} - 5$	Data Hold after Trailing Edge of WR		20		20		20		ns	(Note 6)
$t_V = \frac{1}{2}t_C + WS - 5$	Data Valid before Rising Edge of WR		145		145		145		ns	(Note 6)
$t_{DAR} = \frac{1}{4}t_{C} + WS - 50$	Falling Edge of ALE to Falling Edge of RDY			75		75		75	ns	(Note 6)

# AC Electrical Specifications Test Conditions V<sub>CC</sub> = 4.5V and 5.5V (Unless Otherwise Specified) (Note 1) (Continued)

Symbol	Parameter Cor	Conditions	SBGRP 9 + 25°C		SBGRP 10 + 125°C		SBGRP 11 -55°C		Units	Notes
,			Min	Max	Min	Max	Min	Max	1	
$t_{RWP} = t_{C}$	RDY Pulse Width		100		100		100		ns	(Note 6)
$t_{SALE} = \frac{3}{4}t_{C} + 40$	Falling Edge of HLD to to Rising Edge of ALE		115		115		115		ns	(Note 6)
$t_{HWP} = t_C + 10$	HLD Pulse Width		110		110		110		ns	(Note 6)
$t_{HAD} = \frac{3}{4} t_{C} + 85$	Rising Edge on HLD to Rising Edge on HLDA			160		160		160	ns	(Note 6)
$t_{HAE} = t_C + 100$	Falling Edge on HLD to Falling Edge on HLDA			200		200		200	ns	(Note 6)
$t_{BF} = \frac{1}{2} t_{C} + 66$	BUS Float before Falling Edge on HLDA			116		116		116	ns	(Note 6)
$t_{BE} = \frac{1}{2} t_{C} + 66$	BUS Enable from Rising Edge of HLDA		116		116		116		ns	(Note 6)
t <sub>UAS</sub>	Address Setup Time to Falling Edge of URD		10		10		10		ns	(Note 6)
t <sub>UAH</sub>	Address Hold Time from Rising Edge of URD		10		10		10		ns	(Note 6)
t <sub>RPW</sub>	URD Pulse Width		100		100		100		ns	(Note 6)
t <sub>OE</sub>	URD Falling Edge to Data Out Valid			60		60		60	ns	(Note 6)
t <sub>RDRDY</sub>	RDY Delay from Rising Edge of URD			70		70		70	ns	(Note 6)
t <sub>WDW</sub>	UWR Pulse Width		40		40		40		ns	(Note 6)
t <sub>UDS</sub>	Data Invalid before Trailing Edge of UWR		10		10		10		ns	(Note 6)
t <sub>UDH</sub>	Data In Hold after Rising Edge of UWR		15		15		15		ns	(Note 6)
t <sub>A</sub>	WRRDY Delay from Rising Edge of UWR			70		70		70	ns	(Note 6)

Note 1: Electrical end point testing (when required) for groups C & D shall consist only of subgroups 1, 2, 9 and 10.

Note 5: Tested in functional patterns. Not directly measured.

Note 6:  $C_L = 70$  pF. Input and output levels are per DC characteristics.

#### **Pin Descriptions**

The HPC16083 is available in 68-pin PLCC, LDCC, PGA, and 80-pin PQFP packages.

## I/O PORTS

Port A is a 16-bit bidirectional I/O port with a data direction register to enable each separate pin to be individually defined as an input or output. When accessing external memory, port A is used as the multiplexed address/data bus.

Port B is a 16-bit port with 12 bits of bidirectional I/O similar in structure to Port A. Pins B10, B11, B12 and B15 are general purpose outputs only in this mode. Port B may also be configured via a 16-bit function register BFUN to individually allow each pin to have an alternate function.

B0:	TDX	UART Data Output
B1:		
B2:	CKX	UART Clock (Input or Output)
B3:	T2IO	Timer2 I/O Pin
B4:	T3IO	Timer3 I/O Pin
B5:	SO	MICROWIRE/PLUS Output
B6:	SK	MICROWIRE/PLUS Clock (Input or Output)
B7:	HLDA	Hold Acknowledge Output
B8:	TS0	Timer Synchronous Output
B9:	TS1	Timer Synchronous Output
B10:	UA0	Address 0 Input for UPI Mode
B11:	WRRDY	Write Ready Output for UPI Mode
B12:		
B13:	TS2	Timer Synchronous Output

## Pin Descriptions (Continued)

B14: TS3 Timer Synchronous Output
B15: RDRDY Read Ready Output for UPI Mode
When accessing external memory, four bits of port B

are used as follows:

B10: ALE Address Latch Enable Output
B11: WR Write Output

B12: High Byte Enable Output/Input

(sampled at reset)

B15: RD Read Output

Port I is an 8-bit input port that can be read as general purpose inputs and is also used for the following functions:

10:

I1: NMI Nonmaskable Interrupt Input

I2: INT2 Maskable Interrupt/Input Capture/URD
 I3: INT3 Maskable Interrupt/Input Capture/UWR
 INT4 Maskable Interrupt/Input Capture
 SI MICROWIRE/PLUS Data Input

I6: RDX UART Data Input

17:

Port D is an 8-bit input port that can be used as general purpose digital inputs.

Port P is a 4-bit output port that can be used as general purpose data, or selected to be controlled by timers 4 through 7 in order to generate frequency, duty cycle and pulse width modulated outputs.

#### **POWER SUPPLY PINS**

V<sub>CC1</sub> and

V<sub>CC2</sub> Positive Power Supply GND Ground for On-Chip Logic DGND Ground for Output Buffers

Note: There are two electrically connected V<sub>CC</sub> pins on the chip, GND and DGND are electrically isolated. Both V<sub>CC</sub> pins and both ground pins must be used.

#### **CLOCK PINS**

CKI The Chip System Clock Input

CKO The Chip System Clock Output (inversion of CKI)

Pins CKI and CKO are usually connected across an external crystal.

CK2 Clock Output (CKI divided by 2)

#### OTHER PINS

WO This is an active low open drain output that signals an illegal situation has been detected by the

Watch Dog logic.

ST1 Bus Cycle Status Output: indicates first opcode

fetch.

ST2 Bus Cycle Status Output: indicates machine

states (skip, interrupt and first instruction cycle).

RESET is an active low input that forces the chip to restart and sets the ports in a TRI-STATE mode.

RDY/HLD has two uses, selected by a software bit. It's either a READY input to extend the bus cycle for slower memories, or a HOLD request input to put the bus in a high impedance state for DMA purposes.

NC (no connection) do not connect anything to this

pin.

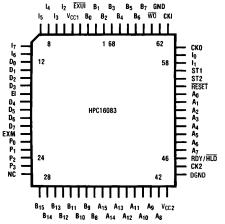
EXM External memory enable (active high) disables internal ROM and maps it to external memory.

El External interrupt with vector address FFF1:FFF0. (Rising/falling edge or high/low level sensitive). Alternately can be configured as 4th input capture.

EXUI External interrupt which is internally OR'ed with the UART interrupt with vector address FFF3:FFF2 (Active Low).

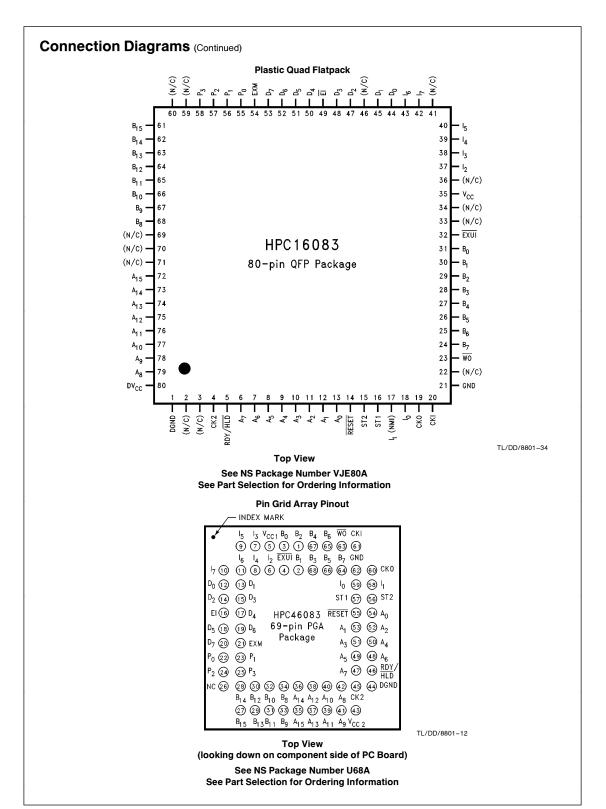
## **Connection Diagrams**

#### **Plastic and Ceramic Leaded Chip Carriers**



TL/DD/8801-11 **Top View** 

See NS Package Number EL68A or V68A See Part Selection for Ordering Information



## Ports A & B

The highly flexible A and B ports are similarly structured. The Port A (see *Figure 11*), consists of a data register and a direction register. Port B (see *Figures 12, 13, 14*) has an alternate function register in addition to the data and direction registers. All the control registers are read/write registers.

The associated direction registers allow the port pins to be individually programmed as inputs or outputs. Port pins selected as inputs, are placed in a TRI-STATE mode by resetting corresponding bits in the direction register.

A write operation to a port pin configured as an input causes the value to be written into the data register, a read operation returns the value of the pin. Writing to port pins configured as outputs causes the pins to have the same value, reading the pins returns the value of the data register.

Primary and secondary functions are multiplexed onto Port B through the alternate function register (BFUN). The secondary functions are enabled by setting the corresponding bits in the BFUN register.

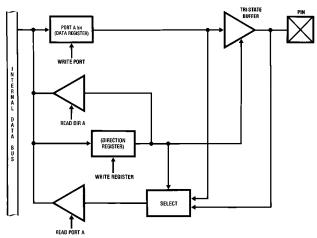


FIGURE 11. Port A: I/O Structure

TL/DD/8801-13

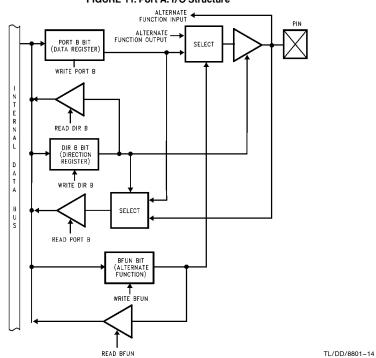
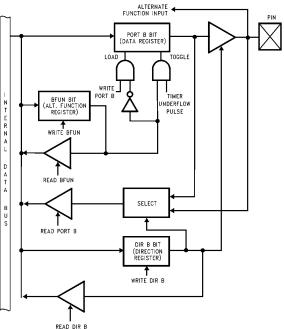
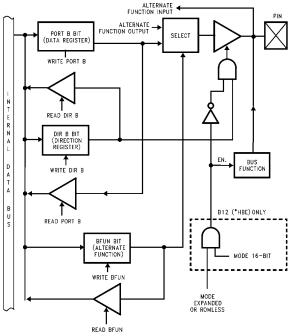


FIGURE 12. Structure of Port B Pins B0, B1, B2, B5, B6 and B7 (Typical Pins)

## Ports A & B (Continued)



READ DIR B TL/DD/8801-15
FIGURE 13. Structure of Port B Pins B3, B4, B8, B9, B13 and B14 (Timer Synchronous Pins)



READ BFUN TL/DD/8801
FIGURE 14. Structure of Port B Pins B10, B11, B12 and B15 (Pins with Bus Control Roles)

TL/DD/8801-16

## **Operating Modes**

To offer the user a variety of I/O and expanded memory options, the HPC16083 has four operating modes. The ROMless HPC16003 has one mode of operation. The various modes of operation are determined by the state of both the EXM pin and the EA bit in the PSW register. The state of the EXM pin determines whether on-chip ROM will be accessed or external memory will be accessed within the address range of the on-chip ROM. The on-chip ROM range of the HPC16083 is E000 to FFFF (8k bytes). The HPC16003 has no on-chip ROM and is intended for use with external memory for program storage. A logic "0" state on the EXM pin will cause the HPC device to address on-chip ROM when the Program Counter (PC) contains addresses within the on-chip ROM address range. A logic "1" state on the EXM pin will cause the HPC device to address memory that is external to the HPC when the PC contains on-chip ROM addresses. The EXM pin should always be pulled high (logic "1") on the HPC16003 because no on-chip ROM is available. The function of the EA bit is to determine the legal addressing range of the HPC device. A logic "0" state in the EA bit of the PSW register does two things-addresses are limited to the on-chip ROM range and on-chip RAM and Register range, and the "illegal address detection" feature of the WATCHDOG logic is engaged. A logic "1" in the EA bit enables accesses to be made anywhere within the 64k byte address range and the "illegal address detection" feature of the WATCHDOG logic is disabled. The EA bit should be set to "1" by software when using the HPC16003 to disable the "illegal address detection" feature of WATCH-DOG

All HPC devices can be used with external memory. External memory may be any combination of RAM and ROM. Both 8-bit and 16-bit external data bus modes are available. Upon entering an operating mode in which external memory is used, port A becomes the Address/Data bus. Four pins of port B become the control lines ALE,  $\overline{\text{RD}}$ ,  $\overline{\text{WR}}$  and  $\overline{\text{HBE}}$ . The High Byte Enable pin ( $\overline{\text{HBE}}$ ) is used in 16-bit mode to select high order memory bytes. The  $\overline{\text{RD}}$  and  $\overline{\text{WR}}$  signals are only generated if the selected address is off-chip. The 8-bit mode is selected by pulling  $\overline{\text{HBE}}$  high at reset. If  $\overline{\text{HBE}}$  is left floating or connected to a memory device chip select at reset, the 16-bit mode is entered. The following sections describe the operating modes of the HPC16003 and HPC16003.

Note: The HPC devices use 16-bit words for stack memory. Therefore, when using the 8-bit mode, User's Stack must be in internal RAM.

### **HPC16083 Operating Modes**

#### SINGLE CHIP NORMAL MODE

In this mode, the HPC16083 functions as a self-contained microcomputer (see Figure 15) with all memory (RAM and

ROM) on-chip. It can address internal memory only, consisting of 8k bytes of ROM (E000 to FFFF) and 256 bytes of onchip RAM and registers (0000 to 01FF). The "illegal address detection" feature of the WATCHDOG is enabled in the Single-Chip Normal mode and a WATCHDOG Output (WO) will occur if an attempt is made to access addresses that are outside of the on-chip ROM and RAM range of the device. Ports A and B are used for I/O functions and not for addressing external memory. The EXM pin and the EA bit of the PSW register must both be logic "0" to enter the Single-Chip Normal mode.

#### **EXPANDED NORMAL MODE**

The Expanded Normal mode of operation enables the HPC16083 to address external memory in addition to the on-chip ROM and RAM (see Table I). WATCHDOG illegal address detection is disabled and memory accesses may be made anywhere in the 64k byte address range without triggering an illegal address condition. The Expanded Normal mode is entered with the EXM pin pulled low (logic "0") and setting the EA bit in the PSW register to "1".

#### SINGLE-CHIP ROMLESS MODE

In this mode, the on-chip mask programmed ROM of the HPC16083 is not used. The address space corresponding to the on-chip ROM is mapped into external memory so 8k bytes of external memory may be used with the HPC16083 (see Table I). The WATCHDOG circuitry detects illegal addresses (addresses not within the on-chip ROM and RAM range). The Single-Chip ROMless mode is entered when the EXM pin is pulled high (logic "1") and the EA bit is logic "0".

#### **EXPANDED ROMLESS MODE**

This mode of operation is similar to Single-Chip ROMless mode in that no on-chip ROM is used, however, a full 64k bytes of external memory may be used. The "illegal address detection" feature of WATCHDOG is disabled. The EXM pin must be pulled high (logic "1") and the EA bit in the PSW register set to "1" to enter this mode.

TABLE I. HPC16083 Operating Modes

Operating Mode	EXM Pin	EA Bit	Memory Configuration
Single-Chip Normal	0	0	E000:FFFF on-chip
Expanded Normal	0	1	E000:FFFF on-chip 0200:DFFF off-chip
Single-Chip ROMless	1	0	E000:FFFF off-chip
Expanded ROMless	1	1	0200:FFFF off-chip

**Note:** In all operating modes, the on-chip RAM and Registers (0000:01FF) may be accessed.

## **HPC16003 Operating Modes**

## **EXPANDED ROMLESS MODE (HPC16003)**

Because the HPC16003 has no on-chip ROM, it has only one mode of operation, the Expanded ROMless Mode. The EXM pin must be pulled high (logic "1") on power up, the EA bit in the PSW register should be set to a "1". The HPC16003 is a ROMless device and is intended for use with external memory. The external memory may be any combination of ROM and RAM. Up to 64k bytes of external memory may be accessed. It is necessary to vector on reset to an address between F000 and FFFF, therefore the user should have external memory at these addresses. The EA bit in the PSW register must immediately be set to "1" at the beginning of the user's program to disable illegal address detection in the WATCHDOG logic.

**TABLE II. HPC16003 Operating Modes** 

Operating	EXM	EA	Memory
Mode	Pin	Bit	Configuration
Expanded ROMless	1	1	0200:FFFF off-chip

Note: The on-chip RAM and Registers (0000:01FF) of the HPC16003 may be accessed at all times.

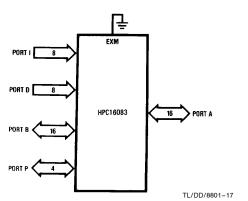
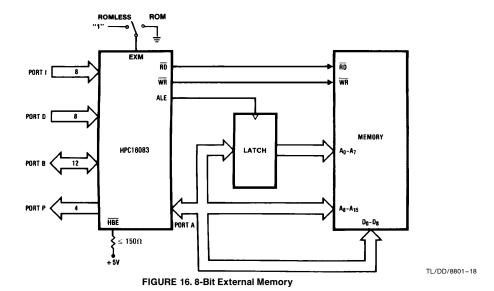


FIGURE 15. Single-Chip Mode



17

## **HPC16003 Operating Modes** (Continued)

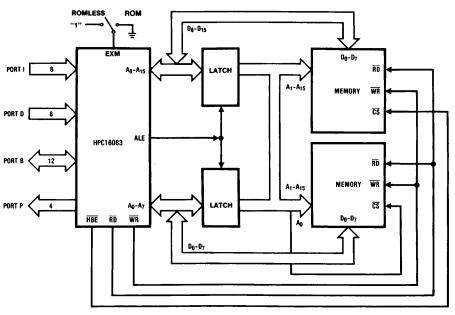


FIGURE 17. 16-Bit External Memory

TL/DD/8801-19

#### **Wait States**

The internal ROM can be accessed at the maximum operating frequency with one wait state. With 0 wait states, internal ROM accesses are limited to  $^2$ 3 f<sub>C</sub> max.

The HPC16083 provides four software selectable Wait States that allow access to slower memories. The Wait States are selected by the state of two bits in the PSW register. Additionally, the RDY input may be used to extend the instruction cycle, allowing the user to interface with slow memories and peripherals.

## **Power Save Modes**

Two power saving modes are available on the HPC16083: HALT and IDLE. In the HALT mode, all processor activities are stopped. In the IDLE mode, the on-board oscillator and timer T0 are active but all other processor activities are stopped. In either mode, all on-board RAM, registers and I/O are unaffected.

#### **HALT MODE**

The HPC16083 is placed in the HALT mode under software control by setting bits in the PSW. All processor activities, including the clock and timers, are stopped. In the HALT mode, power requirements for the HPC16083 are minimal and the applied voltage (V<sub>CC</sub>) may be decreased without altering the state of the machine. There are two ways of exiting the HALT mode: via the  $\overline{\text{RESET}}$  or the NMI. The  $\overline{\text{RESET}}$  input reinitializes the processor. Use of the NMI input will generate a vectored interrupt and resume operation from that point with no initialization. The HALT mode can be enabled or disabled by means of a control register HALT enable. To prevent accidental use of the HALT mode the HALT enable register can be modified only once.

#### **IDLE MODE**

The HPC16083 is placed in the IDLE mode through the PSW. In this mode, all processor activity, except the onboard oscillator and Timer T0, is stopped. As with the HALT mode, the processor is returned to full operation by the RESET or NMI inputs, but without waiting for oscillator stabilization. A timer T0 overflow will also cause the HPC16083 to resume normal operation.

### **HPC16083 Interrupts**

Complex interrupt handling is easily accomplished by the HPC16083's vectored interrupt scheme. There are eight possible interrupt sources as shown in Table III.

TABLE III. Interrupts

Vector Address	Interrupt Source	Arbitration Ranking							
FFFF:FFFE	RESET	0							
FFFD:FFFC	FFD:FFFC Nonmaskable external on rising edge of I1 pin								
FFFB:FFFA	External interrupt on I2 pin	2							
FFF9:FFF8	External interrupt on I3 pin	3							
FFF7:FFF6	External interrupt on I4 pin	4							
FFF5:FFF4	Overflow on internal timers	5							
FFF3:FFF2	Internal on the UART transmit/receive complete or external on EXUI	6							
FFF1:FFF0	External interrupt on El pin	7							

### Interrupt Arbitration

The HPC16083 contains arbitration logic to determine which interrupt will be serviced first if two or more interrupts occur simultaneously. The arbitration ranking is given in Table III. The interrupt on RESET has the highest rank and is serviced first

## **Interrupt Processing**

Interrupts are serviced after the current instruction is completed except for the RESET, which is serviced immediately. RESET and EXUI are level-LOW-sensitive interrupts and EI is programmable for edge-(RISING or FALLING) or level-(HIGH or LOW) sensitivity. All other interrupts are edge-sensitive. NMI is positive-edge sensitive. The external interrupts on I2, I3 and I4 can be software selected to be rising or falling edge. External interrupt (EXUI) is shared with the UART interrupt. This interrupt is level-low sensitive. To select this interrupt disable the ERI and ETI UART interrupt bits in the ENUI register. To select the UART interrupt leave this pin floating or tie it high.

### **Interrupt Control Registers**

The HPC16083 allows the various interrupt sources and conditions to be programmed. This is done through the various control registers. A brief description of the different control registers is given below.

#### **INTERRUPT ENABLE REGISTER (ENIR)**

RESET and the External Interrupt on I1 are non-maskable interrupts. The other interrupts can be individually enabled or disabled. Additionally, a Global Interrupt Enable Bit in the ENIR Register allows the Maskable interrupts to be collectively enabled or disabled. Thus, in order for a particular interrupt to request service both the individual enable bit and the Global Interrupt bit (GIE) have to be set.

#### **INTERRUPT PENDING REGISTER (IRPD)**

The IRPD register contains a bit allocated for each interrupt vector. The occurrence of specified interrupt trigger conditions causes the appropriate bit to be set. There is no indication of the order in which the interrupts have been received. The bits are set independently of the fact that the

interrupts may be disabled. IRPD is a Read/Write register. The bits corresponding to the maskable, external interrupts are normally cleared by the HPC16083 after servicing the interrupts.

For the interrupts from the on-board peripherals, the user has the responsibility of resetting the interrupt pending flags through software.

The NMI bit is read only and I2, I3, and I4 are designed as to only allow a zero to be written to the pending bit (writing a one has no affect). A LOAD IMMEDIATE instruction is to be the only instruction used to clear a bit or bits in the IRPD register. This allows a mask to be used, thus ensuring that the other pending bits are not affected.

#### INTERRUPT CONDITION REGISTER (IRCD)

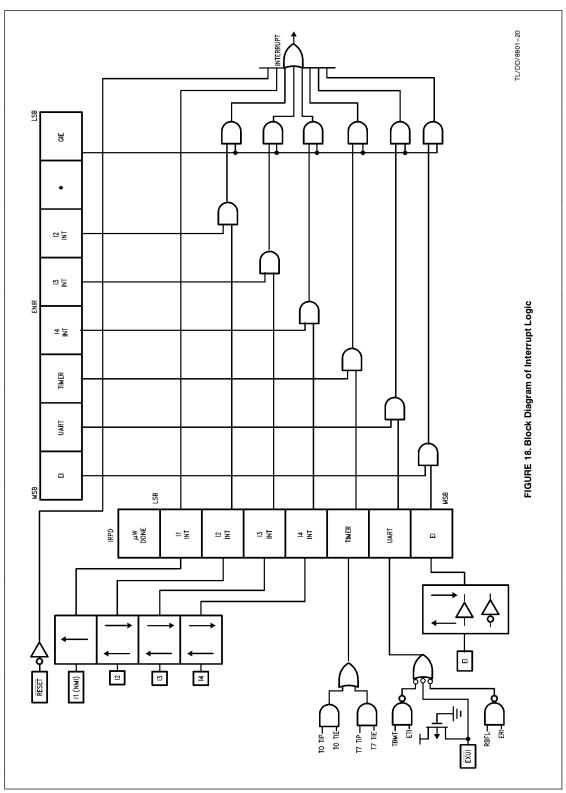
Three bits of the register select the input polarity of the external interrupt on I2, I3, and I4.

## **Servicing the Interrupts**

The Interrupt, once acknowledged, pushes the program counter (PC) onto the stack thus incrementing the stack pointer (SP) twice. The Global Interrupt Enable bit (GIE) is copied into the CGIE bit of the PSW register; it is then reset, thus disabling further interrupts. The program counter is loaded with the contents of the memory at the vector address and the processor resumes operation at this point. At the end of the interrupt service routine, the user does a RETI instruction to pop the stack and re-enable interrupts if the CGIE bit is set, or RET to just pop the stack if the CGIE bit is clear, and then returns to the main program. The GIE bit can be set in the interrupt service routine to nest interrupts if desired. Figure 18 shows the Interrupt Enable Logic.

## RESET

The RESET input initializes the processor and sets ports A and B in the TRI-STATE condition and port P in the LOW state. RESET is an active-low Schmitt trigger input. The processor vectors to FFFF:FFFE and resumes operation at the address contained at that memory location (which must correspond to an on board location). The Reset vector address must be between E000 and FFFF when using the HPC16003.



#### **Timer Overview**

The HPC16083 contains a powerful set of flexible timers enabling the HPC16083 to perform extensive timer functions; not usually associated with microcontrollers.

The HPC16083 contains nine 16-bit timers. Timer T0 is a free-running timer, counting up at a fixed CKI/16 (Clock Input/16) rate. It is used for WATCHDOG logic, high speed event capture, and to exit from the IDLE mode. Consequently, it cannot be stopped or written to under software control. Timer T0 permits precise measurements by means of the capture registers I2CR, I3CR, and I4CR. A control bit in the register TMMODE configures timer T1 and its associated register R1 as capture registers I3CR and I2CR. The capture registers I2CR, I3CR, and I4CR respectively, record the value of timer T0 when specific events occur on the interrupt pins I2, I3, and I4. The control register IRCD programs the capture registers to trigger on either a rising edge or a falling edge of its respective input. The specified edge can also be programmed to generate an interrupt (see Figure 19)

The HPC16083 provides an additional 16-bit free running timer, T8, with associated input capture register EICR (External Interrupt Capture Register) and Configuration Register, EICON. EICON is used to select the mode and edge of the EI pin. EICR is a 16-bit capture register which records the value of T8 (which is identical to T0) when a specific event occurs on the EI pin.

The timers T2 and T3 have selectable clock rates. The clock input to these two timers may be selected from the following two sources: an external pin, or derived internally by dividing the clock input. Timer T2 has additional capability of being clocked by the timer T3 underflow. This allows the user to cascade timers T3 and T2 into a 32-bit timer/counter. The control register DIVBY programs the clock input to timers T2 and T3 (see Figure 20).

The timers T1 through T7 in conjunction with their registers form Timer-Register pairs. The registers hold the pulse duration values. All the Timer-Register pairs can be read from or written to. Each timer can be started or stopped under

software control. Once enabled, the timers count down, and upon underflow, the contents of its associated register are automatically loaded into the timer.

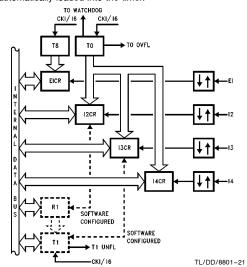


FIGURE 19. Timers T0, T1 and T8 with Four Input Capture Registers

#### SYNCHRONOUS OUTPUTS

The flexible timer structure of the HPC16083 simplifies pulse generation and measurement. There are four synchronous timer outputs (TS0 through TS3) that work in conjunction with the timer T2. The synchronous timer outputs can be used either as regular outputs or individually programmed to toggle on timer T2 underflows (see *Figure 20*). Timer/register pairs 4–7 form four identical units which can generate synchronous outputs on port P (see *Figure 21*).

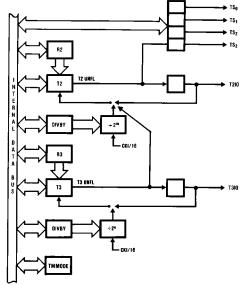
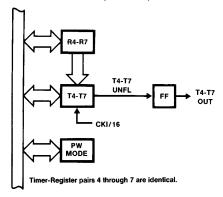


FIGURE 20. Timers T2-T3 Block

### Timer Overview (Continued)



TL/DD/8801-23 FIGURE 21. Timers T4-T7 Block

Maximum output frequency for any timer output can be obtained by setting timer/register pair to zero. This then will produce an output frequency equal to  $\frac{1}{2}$  the frequency of the source used for clocking the timer.

## **Timer Registers**

There are four control registers that program the timers. The divide by (DIVBY) register programs the clock input to timers T2 and T3. The timer mode register (TMMODE) contains control bits to start and stop timers T1 through T3. It also contains bits to latch, acknowledge and enable interrupts from timers T0 through T3. The control register PWMODE similarly programs the pulse width timers T4 through T7 by allowing them to be started, stopped, and to latch and enable interrupts on underflows. The PORTP register contains bits to preset the outputs and enable the synchronous timer output functions.

#### **Timer Applications**

The use of Pulse Width Timers for the generation of various waveforms is easily accomplished by the HPC16083.

Frequencies can be generated by using the timer/register pairs. A square wave is generated when the register value is a constant. The duty cycle can be controlled simply by changing the register value.



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## FIGURE 22. Square Wave Frequency Generation

Synchronous outputs based on Timer T2 can be generated on the 4 outputs TS0-TS3. Each output can be individually programmed to toggle on T2 underflow. Register R2 contains the time delay between events. *Figure 23* is an example of synchronous pulse train generation.

## WATCHDOG Logic

The WATCHDOG Logic monitors the operations taking place and signals upon the occurrence of any illegal activity. The illegal conditions that trigger the WATCHDOG logic are potentially infinite loops and illegal addresses. Should the

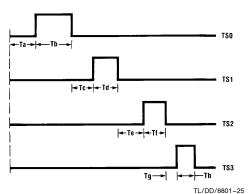


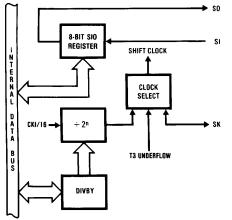
FIGURE 23. Synchronous Pulse Generation

WATCHDOG register not be written to before Timer T0 overflows twice, or more often than once every 4096 counts, an infinite loop condition is assumed to have occurred. An illegal condition also occurs when the processor generates an illegal address when in the Single-Chip modes.\* Any illegal condition forces the WATCHDOG Output ( $\overline{WO}$ ) pin low. The  $\overline{WO}$  pin is an open drain output and can be connected to the  $\overline{RESET}$  or NMI inputs or to the users external logic.

\*Note: See Operating Modes for details.

#### MICROWIRE/PLUS

MICROWIRE/PLUS is used for synchronous serial data communications (see *Figure 24*). MICROWIRE/PLUS has an 8-bit parallel-loaded, serial shift register using SI as the input and SO as the output. SK is the clock for the serial shift register (SIO). The SK clock signal can be provided by an internal or external source. The internal clock rate is programmable by the DIVBY register. A DONE flag indicates when the data shift is completed.



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#### FIGURE 24. MICROWIRE/PLUS

The MICROWIRE/PLUS capability enables it to interface with any of National Semiconductor's MICROWIRE peripherals (i.e., A/D converters, display drivers, EEPROMs).

## **MICROWIRE/PLUS Operation**

The HPC16083 can enter the MICROWIRE/PLUS mode as the master or a slave. A control bit in the IRCD register determines whether the HPC16083 is the master or slave. The shift clock is generated when the HPC16083 is configured as a master. An externally generated shift clock on the SK pin is used when the HPC16083 is configured as a slave. When the HPC16083 is a master, the DIVBY register programs the frequency of the SK clock. The DIVBY register allows the SK clock frequency to be programmed in 15 selectable steps from 64 Hz to 1 MHz with CKI at 16.0 MHz.

The contents of the SIO register may be accessed through any of the memory access instructions. Data waiting to be transmitted in the SIO register is clocked out on the falling edge of the SK clock. Serial data on the SI pin is clocked in on the rising edge of the SK clock.

## **MICROWIRE/PLUS Application**

Figure 25 illustrates a MICROWIRE/PLUS arrangement for an automotive application. The microcontroller-based sys-

tem could be used to interface to an instrument cluster and various parts of the automobile. The diagram shows two HPC16083 microcontrollers interconnected to other MICROWIRE peripherals. HPC16083 #1 is set up as the master and initiates all data transfers. HPC16083 #2 is set up as a slave answering to the master.

The master microcontroller interfaces the operator with the system and could also manage the instrument cluster in an automotive application. Information is visually presented to the operator by means of a LCD display controlled by the COP472 display driver. The data to be displayed is sent serially to the COP472 over the MICROWIRE/PLUS link. Data such as accumulated mileage could be stored and retrieved from the EEPROM COP494. The slave HPC16083 could be used as a fuel injection processor and generate timing signals required to operate the fuel valves. The master processor could be used to periodically send updated values to the slave via the MICROWIRE/PLUS link. To speed up the response, chip select logic is implemented by connecting an output from the master to the external interrupt input on the slave.

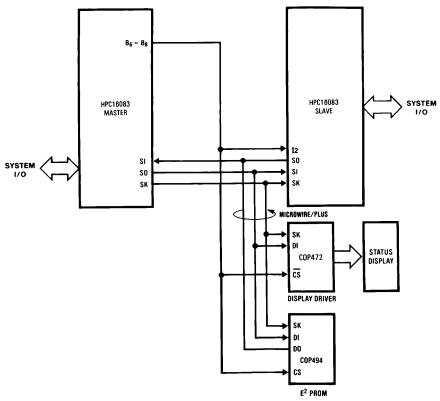


FIGURE 25. MICROWIRE/PLUS Application

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### **HPC16083 UART**

The HPC16083 contains a software programmable UART. The UART (see Figure 26) consists of a transmit shift register, a receiver shift register and five addressable registers, as follows: a transmit buffer register (TBUF), a receiver buffer register (RBUF), a UART control and status register (ENU), a UART receive control and status register (ENUR) and a UART interrupt and clock source register (ENUI). The ENU register contains flags for transmit and receive functions; this register also determines the length of the data frame (8 or 9 bits) and the value of the ninth bit in transmission. The ENUR register flags framing and data overrun errors while the UART is receiving. Other functions of the ENUR register include saving the ninth bit received in the data frame and enabling or disabling the UART's Wake-up Mode of operation. The determination of an internal or external clock source is done by the ENUI register, as well as selecting the number of stop bits and enabling or disabling transmit and receive interrupts.

The baud rate clock for the Receiver and Transmitter can be selected for either an internal or external source using two bits in the ENUI register. The internal baud rate is programmed by the DIVBY register. The baud rate may be selected from a range of 8 Hz to 128 kHz in binary steps or T3 underflow. By selecting a 9.83 MHz crystal, all standard baud rates from 75 baud to 38.4 kBaud can be generated. The external baud clock source comes from the CKX pin. The Transmitter and Receiver can be run at different rates by selecting one to operate from the internal clock and the other from an external source.

The HPC16083 UART supports two data formats. The first format for data transmission consists of one start bit, eight data bits and one or two stop bits. The second data format for transmission consists of one start bit, nine data bits, and one or two stop bits. Receiving formats differ from transmission only in that the Receiver always requires only one stop bit in a data frame.

## **UART Wake-up Mode**

The HPC16083 UART features a Wake-up Mode of operation. This mode of operation enables the HPC16083 to be networked with other processors. Typically in such environments, the messages consist of addresses and actual data. Addresses are specified by having the ninth bit in the data frame set to 1. Data in the message is specified by having the ninth bit in the data frame reset to 0.

The UART monitors the communication stream looking for addresses. When the data word with the ninth bit set is received, the UART signals the HPC16083 with an interrupt. The processor then examines the content of the receiver buffer to decide whether it has been addressed and whether to accept subsequent data.

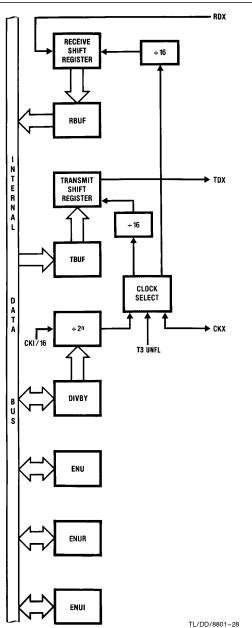


FIGURE 26. UART Block Diagram

## **Universal Peripheral Interface**

The Universal Peripheral Interface (UPI) allows the HPC16083 to be used as an intelligent peripheral to another processor. The UPI could thus be used to tightly link two HPC16083's and set up systems with very high data exchange rates. Another area of application could be where a HPC16083 is programmed as an intelligent peripheral to a host system such as the Series 32000® microprocessor. Figure 27 illustrates how a HPC16083 could be used an an intelligent peripherial for a Series 32000-based application.

The interface consists of a Data Bus (port A), a Read Strobe  $(\overline{URD})$ , a Write Strobe  $(\overline{UWR})$ , a Read Ready Line  $(\overline{RDRDY})$ , a Write Ready Line  $(\overline{WRRDY})$  and one Address Input (UA0). The data bus can be either eight or sixteen bits wide.

The  $\overline{URD}$  and  $\overline{UWR}$  inputs may be used to interrupt the HPC16083. The  $\overline{RDRDY}$  and  $\overline{WRRDY}$  outputs may be used to interrupt the host processor.

The UPI contains an Input Buffer (IBUF), an Output Buffer (OBUF) and a Control Register (UPIC). In the UPI mode, port A on the HPC16083 is the data bus. UPI can only be used if the HPC16083 is in the Single-Chip mode.

### **Shared Memory Support**

Shared memory access provides a rapid technique to exchange data. It is effective when data is moved from a peripheral to memory or when data is moved between blocks of memory. A related area where shared memory access proves effective is in multiprocessing applications where two CPUs share a common memory block. The HPC16083 supports shared memory access with two pins. The pins are the RDY/HLD input pin and the HLDA output pin. The user can software select either the Hold or Ready function by the state of a control bit. The HLDA output is multiplexed onto port B.

The host uses DMA to interface with the HPC16083. The host initiates a data transfer by activating the  $\overline{\text{HLD}}$  input of the HPC16083. In response, the HPC16083 places its system bus in a TRI-STATE Mode, freeing it for use by the host. The host waits for the acknowledge signal ( $\overline{\text{HLDA}}$ ) from the HPC16083 indicating that the sytem bus is free. On receiving the acknowledge, the host can rapidly transfer data into, or out of, the shared memory by using a conventional DMA controller. Upon completion of the message transfer, the host removes the HOLD request and the HPC16083 resumes normal operations.

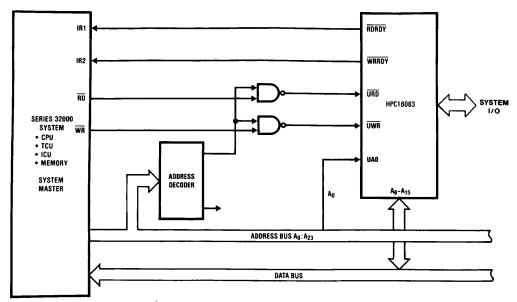
Figure 28 illustrates an application of the shared memory interface between the HPC16083 and a Series 32000 system. To insure proper operation, the interface logic shown is recommended as the means for enabling and disabling the user's bus.

### Memory

The HPC16083 has been designed to offer flexibility in memory usage. A total address space of 64 kbytes can be addressed with 8 kbytes of ROM and 256 bytes of RAM available on the chip itself. The ROM may contain program instructions, constants or data. The ROM and RAM share the same address space allowing instructions to be executed out of RAM.

Program memory addressing is accomplished by the 16-bit program counter on a byte basis. Memory can be addressed directly by instructions or indirectly through the B, X and SP registers. Memory can be addressed as words or bytes. Words are always addressed on even-byte boundaries. The HPC16083 uses memory-mapped organization to support registers, I/O and on-chip peripheral functions.

The HPC16083 memory address space extends to 64 kbytes and registers and I/O are mapped as shown in Table



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FIGURE 27. HPC16083 as a Peripheral: (UPI Interface to Series 32000 Application)

## **Shared Memory Support** (Continued)

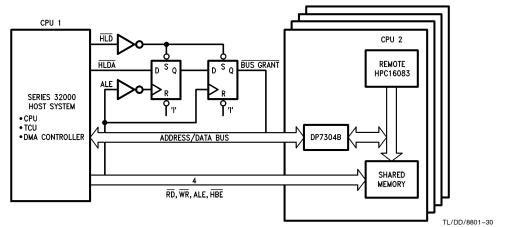


FIGURE 28. Shared Memory Application: HPC16083 Interface to Series 32000 System

#### TABLE IV. HPC16083 Memory Map

FFFF:FFF0 FFEF:FFD0 FFCF:FFCE : : E001:E000	Interrupt Vectors JSRP Vectors On-Chip ROM	USER MEMORY
DFFF:DFFE : : 0201:0200	External Expansion Memory	
01FF:01FE : : 01C1:01C0	On-Chip RAM	USER RAM
0195:0194	WATCHDOG Address	WATCHDOG Logic
0192 0191:0190 018F:018E 018D:018C 018B:018A 0189:0188 0187:0186 0185:0184 0183:0182 0181:0180	TOCON Register TMMODE Register DIVBY Register T3 Timer R3 Register T2 Timer R2 Register I2CR Register/ R1 I3CR Register/ T1 I4CR Register	Timer Block T0:T3
015E:015F 015C 0153:0152 0151:0150 014F:014E 014D:014C 014B:014A 0149:0148 0147:0146 0145:0144 0143:0142 0141:0140	EICR EICON Port P Register PWMODE Register R7 Register T7 Timer R6 Register T6 Timer R5 Register T5 Timer R4 Register T4 Timer	Timer Block T4:T7

Memory Map		
0128 0126 0124 0122 0120	ENUR Register TBUF Register RBUF Register ENUI Register ENU Register	UART
0104	Port D Input Register	
00F5:00F4 00F3:00F2 00F1:00F0	BFUN Register DIR B Register DIR A Register / IBUF	PORTS A & B CONTROL
00E6	UPIC Register	UPI CONTROL
00E3:00E2 00E1:00E0	Port B Port A / OBUF	PORTS A & B
00DE:00DF 00DD:00DC 00D8 00D6 00D4 00D2 00D0	(reserved) HALT Enable Register Port I Input Register SIO Register IRCD Register IRPD Register ENIR Register	PORT CONTROL & INTERRUPT CONTROL REGISTERS
00CF:00CE 00CD:00CC 00CB:00CA 00C9:00C8 00C7:00C6 00C5:00C4 00C3:00C2 00C0	X Register B Register K Register A Register PC Register SP Register (reserved) PSW Register	HPC CORE REGISTERS
00BF:00BE : : 0001:0000	On-Chip RAM	USER RAM

## **Design Considerations**

Designs using the HPC family of 16-bit high speed CMOS microcontrollers need to follow some general guidelines on usage and board layout.

Floating inputs are a frequently overlooked problem. CMOS inputs have extremely high impedance and, if left open, can float to any voltage. You should thus tie unused inputs to  $V_{CC}$  or ground, either through a resistor or directly. Unlike the inputs, unused outputs should be left floating to allow the output to switch without drawing any DC current.

To reduce voltage transients, keep the supply line's parasitic inductances as low as possible by reducing trace lengths, using wide traces, ground planes, and by decoupling the supply with bypass capacitors. In order to prevent additional voltage spiking, this local bypass capacitor must exhibit low inductive reactance. You should therefore use high frequency ceramic capacitors and place them very near the IC to minimize wiring inductance.

- Keep V<sub>CC</sub> bus routing short. When using double sided or multilayer circuit boards, use ground plane techniques.
- Keep ground lines short, and on PC boards make them as wide as possible, even if trace width varies. Use separate ground traces to supply high current devices such as relay and transmission line drivers.
- In systems mixing linear and logic functions and where supply noise is critical to the analog components' performance, provide separate supply buses or even separate supplies.
- If you use local regulators, bypass their inputs with a tantalum capacitor of at least 1 μF and bypass their outputs with a 10 μF to 50 μF tantalum or aluminum electrolytic capacitor.
- If the system uses a centralized regulated power supply, use a 10 μF to 20 μF tantalum electrolytic capacitor or a 50 μF to 100 μF aluminum electrolytic capacitor to decouple the V<sub>CC</sub> bus connected to the circuit board.
- Provide localized decoupling. For random logic, a rule of thumb dictates approximately 10 nF (spaced within 12 cm) per every two to five packages, and 100 nF for every 10 packages. You can group these capacitances, but it's more effective to distribute them among the ICs. If the design has a fair amount of synchronous logic with outputs that tend to switch simultaneously, additional decoupling might be advisable. Octal flip flop and buffers in bus-oriented circuits might also require more decoupling. Note that wire-wrapped circuits can require more decoupling than ground plane or multilayer PC boards.

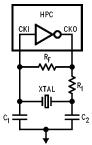


FIGURE 29. Recommended Crystal Circuit

A recommended crystal oscillator circuit to be used with the HPC is shown below. See table for recommended component values. The recommended values given in the table below have yielded consistent results and are made to match a crystal with a 18 pF load capacitance, with some small allowance for layout capacitance.

A recommended layout for the oscillator network should be as close to the processor as physically possible, entirely within 1" distance. This is to reduce lead inductance from long PC traces, as well as interference from other components, and reduce trace capacitance. The layout contains a large ground plane either on the top or bottom surface of the board to provide signal shielding, and a convenient location to ground both the HPC, and the case of the crystal.

It is very critical to have an extremely clean power supply for the HPC crystal oscillator. Ideally one would like a  $V_{CC}$  and ground plane that provide low inductance power lines to the chip. The power planes in the PC board should be decoupled with three decoupling capacitors as close to the chip as possible. A 1.0  $\mu F$ , a 0.1  $\mu F$ , and a 0.001  $\mu F$  dipped mica or ceramic cap mounted as close to the HPC as is physically possible on the board, using the shortest leads, or surface mount components. This should provide a stable power supply, and noiseless ground plane which will vastly improve the performance of the crystal oscillator network.

**HPC Oscillator Table** 

XTAL Frequency (MHz)	R <sub>1</sub> (Ω)
≤2	1500
4	1200
6	910
8	750
10	600
12	470
14	390
16	300
18	220
20	180
22	150
24	120
26	100
28	75
30	62

 $R_{\text{F}}=$  3.3  $M\Omega$ 

 $C_1 = 27 pF$ 

 $C_2 = 33 \text{ pF}$ 

XTAL Specifications: The crystal used was an M-TRON Industries MP-1 Series XTAL. "AT" cut parallel resonant

 $C_L = 18 pF$ 

Series Resistance is:

25Ω @ 25 MHz 40Ω @ 10 MHz

40Ω @ 2 MHz

#### **HPC16083 CPU**

The HPC16083 CPU has a 16-bit ALU and six 16-bit registers

#### Arithmetic Logic Unit (ALU)

The ALU is 16 bits wide and can do 16-bit add, subtract and shift or logic AND, OR and exclusive OR in one timing cycle. The ALU can also output the carry bit to a 1-bit C register.

#### Accumulator (A) Register

The 16-bit A register is the source and destination register for most I/O, arithmetic, logic and data memory access operations.

#### Address (B and X) Registers

The 16-bit B and X registers can be used for indirect addressing. They can automatically count up or down to sequence through data memory.

#### Boundary (K) Register

The 16-bit K register is used to set limits in repetitive loops of code as register B sequences through data memory.

#### Stack Pointer (SP) Register

The 16-bit SP register is the pointer that addresses the stack. The SP register is incremented by two for each push or call and decremented by two for each pop or return. The stack can be placed anywhere in user memory and be as deep as the available memory permits.

#### Program (PC) Register

The 16-bit PC register addresses program memory.

## **Addressing Modes**

## ADDRESSING MODES—ACCUMULATOR AS DESTINATION

#### Register Indirec

This is the "normal" mode of addressing for the HPC16083 (instructions are single-byte). The operand is the memory addressed by the B register (or X register for some instructions).

#### Direct

The instruction contains an 8-bit or 16-bit address field that directly points to the memory for the operand.

#### Indirect

The instruction contains an 8-bit address field. The contents of the WORD addressed points to the memory for the operand.

#### Indexed

The instruction contains an 8-bit address field and an 8- or 16-bit displacement field. The contents of the WORD addressed is added to the displacement to get the address of the operand.

#### **Immediate**

The instruction contains an 8-bit or 16-bit immediate field that is used as the operand.

#### Register Indirect (Auto Increment and Decrement)

The operand is the memory addressed by the X register. This mode automatically increments or decrements the X register (by 1 for bytes and by 2 for words).

## Register Indirect (Auto Increment and Decrement) with Conditional Skip

The operand is the memory addressed by the B register. This mode automatically increments or decrements the B register (by 1 for bytes and by 2 for words). The B register is then compared with the K register. A skip condition is generated if B goes past K.

## ADDRESSING MODES—DIRECT MEMORY AS DESTINATION

#### **Direct Memory to Direct Memory**

The instruction contains two 8- or 16-bit address fields. One field directly points to the source operand and the other field directly points to the destination operand.

#### **Immediate to Direct Memory**

The instruction contains an 8- or 16-bit address field and an 8- or 16-bit immediate field. The immediate field is the operand and the direct field is the destination.

#### Double Register Indirect Using the B and X Registers

Used only with Reset, Set and IF bit instructions; a specific bit within the 64 kbyte address range is addressed using the B and X registers. The address of a byte of memory is formed by adding the contents of the B register to the most significant 13 bits of the X register. The specific bit to be modified or tested within the byte of memory is selected using the least significant 3 bits of register X.

#### **HPC Instruction Set Description**

Mnemonic	Description	Action	
ARITHMETIC INSTRU	CTIONS		
ADD	Add	MA+MemI → MA	carry → C
ADC	Add with carry	$MA + MemI + C \rightarrow MA$	carry → C
ADDS	Add short imm8	$MA + imm8 \rightarrow MA$	carry $\rightarrow$ C
DADC	Decimal add with carry	$MA + Meml + C \rightarrow MA$ (Decimal)	carry → C
SUBC	Subtract with carry	$MA-MemI+C \rightarrow MA$	carry $\rightarrow$ C
DSUBC	Decimal subtract w/carry	$MA - Meml + C \rightarrow MA$ (Decimal)	carry → C
MULT	Multiply (unsigned)	$MA*MemI \rightarrow MA \& X, 0 \rightarrow K, 0 \rightarrow C$	
DIV	Divide (unsigned)	$MA/Meml \rightarrow MA, rem. \rightarrow X, 0 \rightarrow K,$	$0 \rightarrow C$
DIVD	Divide Double Word (unsigned)	$(X \& MA)/Meml \rightarrow MA, rem \rightarrow X, 0 -$	$\rightarrow$ K, carry $\rightarrow$ C
IFEQ	If equal	Compare MA & Meml, Do next if equal	
IFGT	If greater than	Compare MA & Meml, Do next if MA >	Meml
AND	Logical and	MA and Meml → MA	
OR	Logical or	MA or Meml → MA	
XOR	Logical exclusive-or	MA xor Meml → MA	
MEMORY MODIFY INS	STRUCTIONS		
INC	Increment	Mem + 1 → Mem	
DECSZ	Decrement, skip if 0	Mem −1 → Mem, Skip next if Mem =	0

Mnemonic	Description	Action
INSTRUCTIONS		
SBIT	Set bit	1 → Mem.bit
RBIT	Reset bit	0 → Mem.bit
IFBIT	If bit	If Mem.bit is true, do next instr.
MORY TRANSFER INST		ii Moiii.bit is true, do next iiisti.
LD	Load	Meml → MA
	Load, incr/decr X	$Mem(X) \longrightarrow A, X \pm 1 \text{ (or 2)} \longrightarrow X$
ST	Store to Memory	$A \longrightarrow Mem$
X	Exchange	A ←→ Mem
	Exchange, incr/decr X	$A \longleftrightarrow Mem(X), X \pm 1 \text{ (or 2)} \longrightarrow X$
PUSH	Push Memory to Stack	$W \rightarrow W(SP), SP + 2 \rightarrow SP$
POP	Pop Stack to Memory	$SP-2 \rightarrow SP, W(SP) \rightarrow W$
LDS	Load A, incr/decr B,	$Mem(B) \rightarrow A, B \pm 1 \text{ (or 2)} \rightarrow B,$
LDS		Skip next if B greater/less than K
VC.	Skip on condition	
XS	Exchange, incr/decr B,	$Mem(B) \longleftrightarrow A,B \pm 1 \text{ (or 2)} \to B,$
	Skip on condition	Skip next if B greater/less than K
SISTER LOAD IMMEDIA	ATE INSTRUCTIONS	
LD B	Load B immediate	$imm \rightarrow B$
LD K	Load K immediate	$imm \rightarrow K$
LDX	Load X immediate	$imm \rightarrow X$
LD BK	Load B and K immediate	$imm \rightarrow B, imm \rightarrow K$
CUMULATOR AND C IN	STRUCTIONS	
CLR A	Clear A	$0 \rightarrow A$
INC A	Increment A	$A + 1 \rightarrow A$
DEC A	Decrement A	$A-1 \rightarrow A$
COMP A	Complement A	1's complement of A → A
	•	
SWAP A	Swap nibbles of A	$A15:12 \leftarrow A11:8 \leftarrow A7:4 \longleftrightarrow A3:0$
RRC A	Rotate A right thru C	$C \to A15 \to \dots \to A0 \to C$
RLC A	Rotate A left thru C	$C \leftarrow A15 \leftarrow \dots \leftarrow A0 \leftarrow C$
SHR A	Shift A right	$0 \to A15 \to \dots \to A0 \to C$
SHL A	Shift A left	$C \leftarrow A15 \leftarrow \ldots \leftarrow A0 \leftarrow 0$
SC	Set C	1 → C
RC	Reset C	$0 \rightarrow C$
IFC	IF C	Do next if C = 1
IFNC	IF not C	Do next if $C = 0$
ANSFER OF CONTROL	INSTRUCTIONS	
JSRP	Jump subroutine from table	$PC \rightarrow [SP], SP + 2 \rightarrow SP$
00111	damp dabi damid nom table	$W(table #) \rightarrow PC$
JSR	Jump subroutine relative	$PC \rightarrow [SP],SP+2 \rightarrow SP,PC+\# \rightarrow PC$
JOH	Jump subroduite relative	/ /
ICDI	leann a dean de la com	(#is +1025 to -1023)
JSRL	Jump subroutine long	$PC \rightarrow [SP], SP + 2 \rightarrow SP, PC + \# \rightarrow PC$
JP	Jump relative short	$PC + \# \rightarrow PC(\# \text{ is } +32 \text{ to } -31)$
JMP	Jump relative	$PC + \# \rightarrow PC(\#is + 257 \text{ to } -255)$
JMPL	Jump relative long	$PC + \# \longrightarrow PC$
JID	Jump indirect at PC + A	$PC+A+1 \rightarrow PC$
JIDW		then Mem(PC) + PC $\rightarrow$ PC
NOP	No Operation	$PC + 1 \rightarrow PC$
RET	Return	$SP-2 \rightarrow SP,[SP] \rightarrow PC$
RETSK	Return then skip next	$SP-2 \rightarrow SP,[SP] \rightarrow PC, \& skip$
RETI	Return from interrupt	$SP-2 \rightarrow SP,[SP] \rightarrow PC$ , interrupt re-enabled
Note: W is 16-bit word of memo	•	· · ·
	lirect memory (8 or 16-bit)	
Mem is 8-bit byte or 16-b		
	ory or 8 or 16-bit immediate data	
imm is 8-bit or 16-bit imm	•	
imm8 is 8-bit immediate of	lata only	

## **Memory Usage**

Number Of Bytes For Each Instruction (number in parenthesis is 16-Bit field)

Using Accumulator A						To Direc	t Memory			
	Reg I (B)	Indir. (X)	Direct	Indir.	Index	Immed.	Dir *	ect **	Imn *	ned. **
LD	1	1	2(4)	3	4(5)	2(3)	3(5)	5(6)	3(4)	5(6)
X	1	1	2(4)	3	4(5)	_	_	_	_	_
ST	1	1	2(4)	3	4(5)	_	_	_	_	
ADC	1	2	3(4)	3	4(5)	4(5)	4(5)	5(6)	4(5)	5(6)
ADDS	_	_	l –	_	_	2	_	_	_	_
SBC	1	2	3(4)	3	4(5)	4(5)	4(5)	5(6)	4(5)	5(6)
DADC	1	2	3(4)	3	4(5)	4(5)	4(5)	5(6)	4(5)	5(6)
DSBC	1	2	3(4)	3	4(5)	4(5)	4(5)	5(6)	4(5)	5(6)
ADD	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
MULT	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
DIV	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
DIVD	1	2	3(4)	3	4(5)	_	4(5)	5(6)	4(5)	5(6)
IFEQ	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
IFGT	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
AND	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
OR	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)
XOR	1	2	3(4)	3	4(5)	2(3)	4(5)	5(6)	4(5)	5(6)

#### Instructions that modify memory directly

	(B)	(X)	Direct	Indir	Index	B&X
SBIT	1	2	3(4)	3	4(5)	1
RBIT	1	2	3(4)	3	4(5)	1
IFBIT	1	2	3(4)	3	4(5)	1
DECSZ	3	2	2(4)	3	4(5)	
INC	3	2	2(4)	3	4(5)	

	Immed.
LD B,*	2(3)
LD X,*	2(3)
LD K,*	2(3)
LD BK,*,*	3(5)

**Immediate Load Instructions** 

# Register Indirect Instructions with Auto Increment and Decrement

Register B With Skip			
	(B+)	(B-)	
LDS A,*	1	1	
XS A,*	1	1	

Register X			
	( <b>X</b> + )	(X -)	
LD A,*	1	1	
X A *	1	1	

#### Instructions Using A and C

	_	
CLR	Α	1
INC	Α	1
DEC	Α	1
COMP	Α	1
SWAP	Α	1
RRC	Α	1
RLC	Α	1
SHR	Α	1
SHL	Α	1
SC		1
RC		1
IFC		1
IFNC		1

## **Transfer of Control Instructions**

JSRP	1
JSR	2
JSRL	3
JP	1
JMP	2
JMPL	3
JID	1
JIDW	1
NOP	1
RET	1
RETSK	1
RETI	1

#### **Stack Reference Instructions**

	Direct
PUSH	2
POP	2

<sup>\*8-</sup>bit direct address
\*\*16-bit direct address

### **Code Efficiency**

One of the most important criteria of a single chip microcontroller is code efficiency. The more efficient the code, the more features that can be put on a chip. The memory size on a chip is fixed so if code is not efficient, features may have to be sacrificed or the programmer may have to buy a larger, more expensive version of the chip.

The HPC16083 has been designed to be extremely codeefficient. The HPC16083 looks very good in all the standard coding benchmarks; however, it is not realistic to rely only on benchmarks. Many large jobs have been programmed onto the HPC16083, and the code savings over other popular microcontrollers has been considerable.

Reasons for this saving of code include the following:

#### SINGLE BYTE INSTRUCTIONS

The majority of instructions on the HPC16083 are singlebyte. There are two especially code-saving instructions:

JP is a 1-byte jump. True, it can only jump within a range of plus or minus 32, but many loops and decisions are often within a small range of program memory. Most other micros need 2-byte instructions for any short jumps.

JSRP is a 1-byte call subroutine. The user makes a table of the 16 most frequently called subroutines and these calls will only take one byte. Most other micros require two and even three bytes to call a subroutine. The user does not have to decide which subroutine addresses to put into the table; the assembler can give this information.

#### **EFFICIENT SUBROUTINE CALLS**

The 2-byte JSR instructions can call any subroutine within plus or minus 1k of program memory.

#### MULTIFUNCTION INSTRUCTIONS FOR DATA MOVE-MENT AND PROGRAM LOOPING

The HPC16083 has single-byte instructions that perform multiple tasks. For example, the XS instruction will do the following:

- 1. Exchange A and memory pointed to by the B register
- 2. Increment or decrement the B register
- 3. Compare the B register to the K register
- 4. Generate a conditional skip if B has passed K

The value of this multipurpose instruction becomes evident when looping through sequential areas of memory and exiting when the loop is finished.

#### BIT MANIPULATION INSTRUCTIONS

Any bit of memory, I/O or registers can be set, reset or tested by the single byte bit instructions. The bits can be addressed directly or indirectly. Since all registers and I/O are mapped into the memory, it is very easy to manipulate specific bits to do efficient control.

#### **DECIMAL ADD AND SUBTRACT**

This instruction is needed to interface with the decimal user world.

It can handle both 16-bit words and 8-bit bytes.

The 16-bit capability saves code since many variables can be stored as one piece of data and the programmer does not have to break his data into two bytes. Many applications store most data in 4-digit variables. The HPC16083 supplies 8-bit byte capability for 2-digit variables and literal variables.

#### **MULTIPLY AND DIVIDE INSTRUCTIONS**

The HPC16083 has 16-bit multiply, 16-bit by 16-bit divide, and 32-bit by 16-bit divide instructions. This saves both code and time. Multiply and divide can use immediate data or data from memory. The ability to multiply and divide by immediate data saves code since this function is often needed for scaling, base conversion, computing indexes of arrays, etc.

## **Development Support**

#### HPC MICROCONTROLLER DEVELOPMENT SYSTEM

National Semiconductor's HPC microcontroller development is supported through a combination of third party hardware and software, coupled with NSC in-house developed software consisting of compilers, assemblers, linkers, cross converters and debuggers. The code modules can then be transferred to many EPROM programming systems.

#### **CUSTOMER SUPPORT**

National Semiconductor's Customer Response Center (CRC) provides samples, literature, prices, product information. The CRC's engineering staff is prepared to answer questions regarding specific design and application questions regarding specific design and application questions. Call any weekday 7:00 AM to 7:00 PM central time (US) to 1-800-272-9959 or contact your regional business center.

## **Development Support (Continued)**

#### DIAL-A-HELPER

Dial-A-Helper is a service provided by the Microcontroller Applications group. Dial-A-Helper is an Electronic Bulletin Board Information system and additionally, provides the capability of remotely accessing the development system at a customer site.

#### INFORMATION SYSTEM

The Dial-A-Helper system provides access to an automated information storage and retrieval system that may be accessed over standard dial-up telephone lines 24 hours a day. The system capabilities include a MESSAGE SECTION (electronic mail) for communications to and from the Microcontroller Applications Group and a FILE SECTION which consists of several file areas where valuable application software and utilities can be found. The minimum require-

ment for accessing Dial-A-Helper is a Hayes compatible modem.

If the user has a PC with a communications package then files from the FILE SECTION can be down loaded to disk for later use.

#### Order P/N: MDS-DIAL-A-HLP

Information system package contains: DIAL-A-HELPER Users Manual Public Domain Communications Software

#### **FACTORY APPLICATIONS SUPPORT**

Dial-A-Helper also provides immediate factory applications support.

#### **Development Tools Selection Table**

Development Tools Selection Table		
Order Number	Description	Manual Number
NSC		
HPC-DEV-IBMA	User's manuals and disks for Assembler/Linker/Librarian package for the IBM PC	424410836-001
HPC-DEV-IBMC	User's manuals and disks for C Compiler and Assembler/Linker/Librarian package for the	424410883-001
	IBM PC	424410836-001
HPC-DEV-HDB	User's manuals and disks for Source Symbolic Debugger, C Compiler and Assembler/Linker/	424421640-001
	Librarian Package for the IBM PC	424410883-001
	For use with the HP system only	424410836-001
Signum		
USP-HPC	Base Unit—User's manual and screen debugger	
POD-HPC164	30 MHz POD and interface board for HPC46164	
POD-HPC064	30 MHz POD and interface board for HPC46064	
POD-HPC083	30 MHz POD and interface board for HPC46083	
POD-HPC100	40 MHz POD and interface board for HPC46100	
POD-HPC164-3	20 MHz 3.3V POD and interface board for HPC43164	
POD-HPC064-3	20 MHz 3.3V POD and interface board for HPC43064	
POD-HPC100-3	30 MHz 3.3V POD and interface board for HPC43100	
Hewlett Packard		
64700A	Card cage	
64706A	48 Channel Analyzer	
64775S	Software interface	
OPT006	Software interface to IBM PC	
64775G	HPC16083 Emulator with 128K RAM	
64775H	HPC16064 Emulator with 128K RAM	
64775J	HPC16400E Emulator with 128K RAM	
64701A	LAN Interface (Optional)	

#### Contact your local NSC sales office for ordering information

The Signum system comes with power supply, base unit software, RS232 link to host and emulator pod for the HPC Family member ordered. It also includes an interface connector that fits between the POD and the Target board. This system does not support deelopment of HPC46400E based systems. Source symbolic debug capability for both assembly and C language is included in the screen debugger.

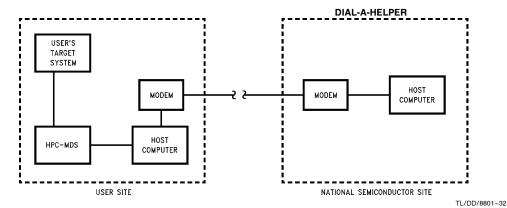
The HP model 64775 emulator/analyzer provides in system emulation up to 20 MHz, 0 wait state memory, and 30 MHz, 1 wait state memory for all devices except the HPC46400E, which is 20 MHz, 1 wait state. A reverse assembler is also available.

The recommended configuration for the IBM PC compatible host is a 386 or higher running DOS 3.0 or higher with 4 MB of extended memory. An RS232 serial port capable of running at 19.2K baud and a three button mouse is recommended for the Signum System interface.

## **Development Support (Continued)**

Voice: (408) 721-5582 Modem: (408) 739-1162

Baud: 300 or 1200 Baud
Set-Up: Length: 8-bit
Parity: None
Stop Bit: 1
Operation: 24 hrs, 7 days



## **Part Selection**

The HPC family includes devices with many different options and configurations to meet various application needs. The number HPC16083 has been generically used throughout this datasheet to represent the whole family of parts. The following chart explains how to order various options available when ordering HPC family members.

Note: All options may not currently be available.

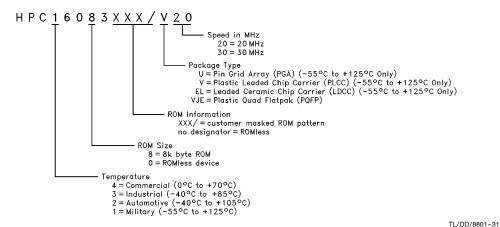
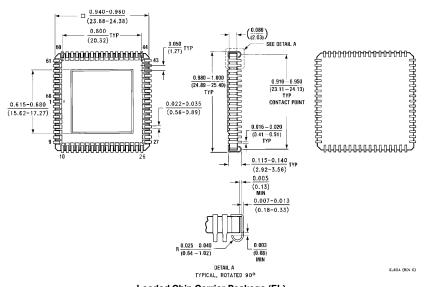


FIGURE 30. HPC Family Part Numbering Scheme

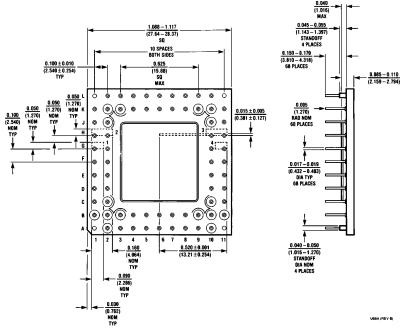
### **Examples**

HPC46003V20 — ROMless, Commercial temp. (0°C to 70°C), PLCC HPC16083XXX/U20 — 8k masked ROM, Military temp. ( $-55^{\circ}$ C to  $+125^{\circ}$ C), PGA HPC26083XXX/V20 — 8k masked ROM, Automotive temp. ( $-40^{\circ}$ C to  $+105^{\circ}$ C), PLCC

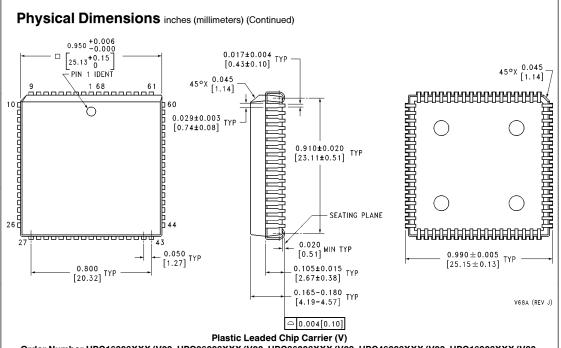
## Physical Dimensions inches (millimeters)



Leaded Chip Carrier Package (EL)
Order Number HPC16083XXX/L20, HPC16083XXX/L30, HPC16003EL20, HPC26003EL20, HPC36003EL20, HPC46003EL20, HPC16003EL30, HPC26003EL30, HPC36003EL30 or HPC46003EL30
NS Package Number EL68A



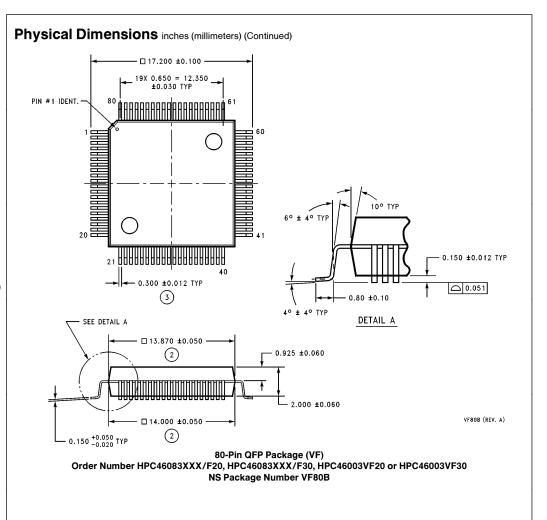
Pin Grid Array Pinout (U)
Order Number HPC16083XXX/U20, HPC16083XXX/U30, HPC16003U20 or HPC16003U30
NS Package Number U68A



Plastic Leaded Chip Carrier (V)

Order Number HPC16083XXX/V20, HPC26083XXX/V20, HPC36083XXX/V20, HPC46083XXX/V20, HPC16083XXX/V30, HPC26083XXX/V30, HPC36083XXX/V30, HPC36003V20, HPC36003V20, HPC36003V20, HPC46003V20, HPC46003V20, HPC46003V30, HPC36003V30, HPC36003V30 or HPC46003V30

NS Package Number V68A



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- 2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



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