

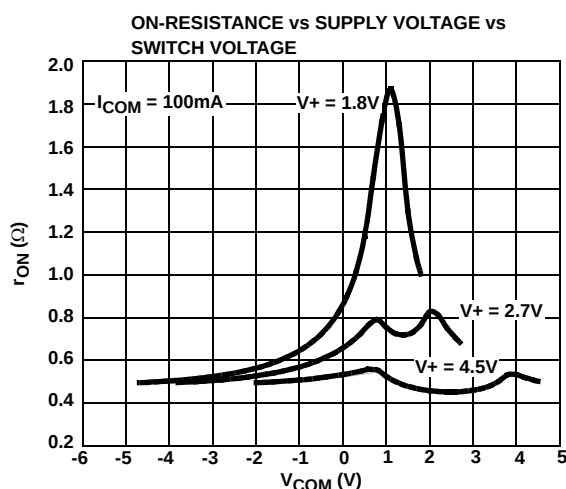
Negative Signal Swing, Sub-ohm, Dual SPDT with Click and Pop Elimination Single Supply Switch

The Intersil ISL54062 device is a low ON-resistance, low voltage, bidirectional, dual single-pole/double-throw (SPDT) analog switch. It is designed to operate from a single +1.8V to +6.5V supply and pass signals that swing down to 6.5V below the positive supply rail. Targeted applications include battery powered equipment that benefit from low r_{ON} (0.56Ω), low power consumption (20nA) and fast switching speeds ($t_{ON} = 55ns$, $t_{OFF} = 18ns$). The digital inputs are 1.8V logic-compatible up to a +3V supply. The ISL54062 also features integrated circuitry to eliminate click and pop noise to an audio speaker. The ISL54062 is offered in a small form factor package, alleviating board space limitations. It is available in a tiny 10 Ld 1.8x1.4mm μ TQFN or 10 Ld 3x3mm TDFN package.

The ISL54062 is a committed dual single-pole/double-throw (SPDT) that consist of two normally open (NO) and two normally closed (NC) switches with independent logic control. This configuration can be used as a dual 2-to-1 multiplexer.

TABLE 1. FEATURES AT A GLANCE

ISL54062	
Number of Switches	2
SW	SPDT or 2-to-1 MUX
4.3V r_{ON}	0.57Ω
4.3V t_{ON}/t_{OFF}	43ns/23ns
2.7V r_{ON}	0.82Ω
2.7V t_{ON}/t_{OFF}	55ns/18ns
1.8V r_{ON}	1.8Ω
1.8V t_{ON}/t_{OFF}	145ns/28ns
Packages	10 Ld μ TQFN, 10 Ld TDFN



Features

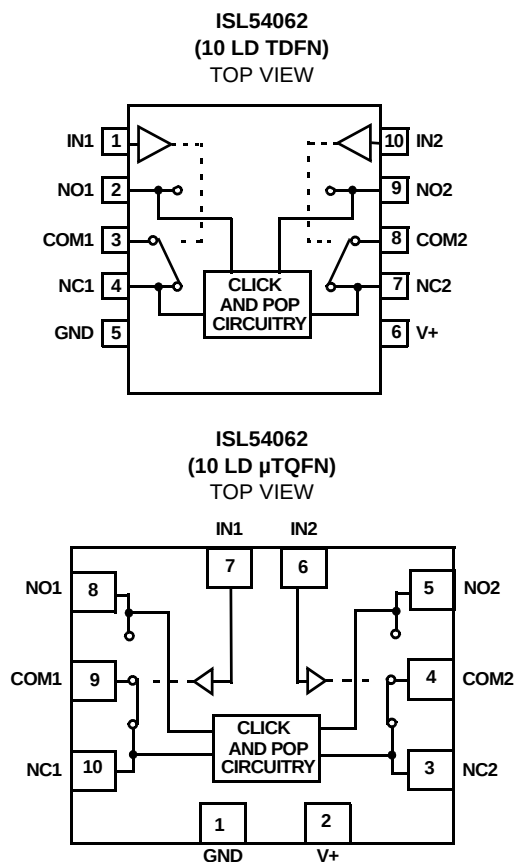
- Pb-free (RoHS Compliant)
- Negative Signal Swing (Max 6.5V Below V+)
- Audio Click and Pop Elimination Circuitry
- ON-Resistance (r_{ON})
 - $V+ = +4.5V$ 0.55Ω
 - $V+ = +4.3V$ 0.57Ω
 - $V+ = +2.7V$ 0.82Ω
 - $V+ = +1.8V$ 1.8Ω
- r_{ON} Matching Between Channels 10mΩ
- r_{ON} Flatness Across Signal Range 0.35Ω
- Low THD+N @ 32Ω Load 0.02%
- Single Supply Operation +1.8V to +6.5V
- Low Power Consumption @ 3V (P_D) 24nW
- Fast Switching Action ($V+ = +4.3V$)
 - t_{ON} 43ns
 - t_{OFF} 23ns
- ESD HBM Rating >6kV
- Guaranteed Break-before-Make
- 1.8V Logic Compatible (+3V supply)
- Low I+ Current when V_{INH} is not at the V+ Rail
- Available in 10 Ld μ TQFN 1.8x1.4mm and 10 Ld 3x3mm TDFN

Applications

- Audio and Video Switching
- Battery powered, Handheld, and Portable Equipment
 - MP3 and Multimedia Players
 - Cellular/mobile Phones
 - Pagers
 - Laptops, Notebooks, Palmtops
- Portable Test and Measurement
- Medical Equipment

Related Literature

- Technical Brief TB363 "Guidelines for Handling and Processing Moisture Sensitive Surface Mount Devices (SMDs)"
- Application Note AN557 "Recommended Test Procedures for Analog Switches"

Pinout (Note 1)

NOTE:

1. Switches Shown for INx = Logic "0".

Truth Table

IN1	IN2	NC1	NO1	NC2	NO2
0	0	ON	OFF	ON	OFF
0	1	ON	OFF	OFF	ON
1	0	OFF	ON	ON	OFF
1	1	OFF	ON	OFF	ON

NOTE: Logic "0" $\leq 0.5V$. Logic "1" $\geq 1.4V$ with a 3V supply.**Pin Descriptions**

PIN	FUNCTION
V+	IC Power Supply (+1.8V to +6.5V). Decouple V+ to ground by placing a 0.1 μ F capacitor at the V+ and GND supply lines as near as the IC as possible.
GND	Ground Connection
INx	Digital Control Input
COMx	Analog Switch Common Pin
NOx	Analog Switch Normally Open Pin
NCx	Analog Switch Normally Closed Pin

Ordering Information

PART NUMBER	PART MARKING	TEMP. RANGE (°C)	PACKAGE (Pb-Free)	PKG. DWG. #
ISL54062IRTZ (Note 3)	4062	-40 to +85	10 Ld 3x3 TDFN	L10.3x3A
ISL54062IRTZ-T (Notes 2, 3)	4062	-40 to +85	10 Ld 3x3 TDFN (Tape and Reel)	L10.3x3A
ISL54062IRUZ-T (Notes 2, 4)	8	-40 to +85	10 Ld Thin μ TQFN (Tape and Reel)	L10.1.8x1.4A

NOTES:

2. Please refer to TB347 for details on reel specifications.
3. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
4. These Intersil Pb-free plastic packaged products employ special Pb-free material sets; molding compounds/die attach materials and NiPdAu plate - e4 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

Absolute Maximum Ratings

V+ to GND	-0.5 to 7.0V
Input Voltages	
NOx, NCx (Note 5)	(V+ - 7V) to ((V+) + 0.5V)
INx (Note 5)	-0.5 to ((V+) + 0.5V)
Output Voltages	
COMx (Note 5)	(V+ - 7V) to ((V+) + 0.5V)
Continuous Current NOx, NCx, or COMx	±300mA
Peak Current NOx, NCx, or COMx (Pulsed 1ms, 10% Duty Cycle, Max)	±500mA
ESD Rating:	
Human Body Model	>6kV
Machine Model	>400V
Charged Device Model	>1.5kV

Thermal Information

Thermal Resistance (Typical)	θ_{JA} (°C/W)	θ_{JC} (°C/W)
10 Ld 3x3 TDFN Package (Notes 6, 8)	55	18
10 Ld μ TQFN Package (Note 7)	155	N/A
Maximum Junction Temperature (Plastic Package)	+150°C	
Maximum Storage Temperature Range	-65°C to +150°C	
Pb-Free Reflow Profile	see link below http://www.intersil.com/pbfree/Pb-FreeReflow.asp	

Operating Conditions

Temperature Range	-40°C to +85°C
Power Supply Range	+1.8V to +6.5V

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- Signals on NC, NO, IN, or COM exceeding V+ or GND by specified amount are clamped by internal diodes. Limit forward diode current to maximum current ratings.
- θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief TB379.
- θ_{JA} is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief TB379 for details.
- For θ_{JC} , the "case temp" location is the center of the exposed metal pad on the package underside.

Electrical Specifications - 5V Supply

Test Conditions: V+ = +4.5V to +5.5V, GND = 0V, V_{INH} = 2.4V, V_{INL} = 0.8V (Note 9), Unless Otherwise Specified.

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
ANALOG SWITCH CHARACTERISTICS						
ON-Resistance, r_{ON}	$V+ = 4.5V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V+ - 6.5)$ to $V+$ (see Figure 5)	25	-	0.55	-	Ω
		Full	-	0.68	-	Ω
r_{ON} Matching Between Channels, Δr_{ON}	$V+ = 4.5V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} =$ Voltage at max r_{ON} , (Note 13)	25	-	15	-	$m\Omega$
		Full	-	30	-	$m\Omega$
r_{ON} Flatness, $R_{FLAT(ON)}$	$V+ = 4.5V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V+ - 6.5)$ to $V+$, (Note 12)	25	-	0.11	-	Ω
		Full	-	0.14	-	Ω
COM ON Leakage Current, $I_{COM(ON)}$	$V+ = 5V$, $V_{COM} = -1.5V, 5V$, V_{NO} or $V_{NC} =$ Float	25	-	49	-	nA
		Full	-	0.7	-	μA
DYNAMIC CHARACTERISTICS						
Turn-ON Time, t_{ON}	$V+ = 4.5V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	35	-	ns
		Full	-	50	-	ns
Turn-OFF Time, t_{OFF}	$V+ = 4.5V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	16	-	ns
		Full	-	22	-	ns
Break-Before-Make Time Delay, t_D	$V+ = 5.5V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 3)	Full	-	18	-	ns
Charge Injection, Q	$V_G = 0V$, $R_G = 0\Omega$, $C_L = 1.0nF$ (see Figure 2)	25	-	170	-	pC
OFF-Isolation	$R_L = 50\Omega$, $C_L = 5pF$, $f = 100kHz$, $V_{COM} = 1V_{RMS}$ (see Figure 4)	25	-	60	-	dB
Crosstalk (Channel-to-Channel)	$R_L = 50\Omega$, $C_L = 5pF$, $f = 1MHz$, $V_{COM} = 1V_{RMS}$ (see Figure 6)	25	-	-75	-	dB
Total Harmonic Distortion	$f = 20Hz$ to $20kHz$, $V_{COM} = 0.5V_{P-P}$, $R_L = 32\Omega$	25	-	0.02	-	%
-3dB Bandwidth	$V_{COM} = 1V_{RMS}$, $R_L = 50\Omega$, $C_L = 5pF$	25	-	60	-	MHz
NO x or NCx OFF Capacitance, C_{OFF}	$f = 1MHz$	25	-	36	-	pF

Electrical Specifications - 5V Supply Test Conditions: $V_+ = +4.5V$ to $+5.5V$, $GND = 0V$, $V_{INH} = 2.4V$, $V_{INL} = 0.8V$ (Note 9), Unless Otherwise Specified. **(Continued)**

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
COMx ON Capacitance, C _{COM(ON)}	f = 1MHz (see Figure 7)	25	-	88	-	pF
POWER SUPPLY CHARACTERISTICS						
Positive Supply Current, I+	V+ = 5.5V, V _{IN} = 0V or V+	25	-	0.02	0.1	μA
		Full	-	2.5	-	μA
DIGITAL INPUT CHARACTERISTICS						
Input Voltage Low, V _{INL}		Full	-	-	0.8	V
Input Voltage High, V _{INH}		Full	2.4	-	-	V
Input Current, I _{INH} , I _{INL}	V+ = 5.5V, V _{IN} = 0V or V+	25	-0.1	-	0.1	μA
		Full	-	0.89	-	μA

Electrical Specifications - 4.3V Supply Test Conditions: $V_+ = +3.9V$ to $+4.5V$, $GND = 0V$, $V_{INH} = 1.6V$, $V_{INL} = 0.5V$ (Note 9), Unless Otherwise Specified.

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
ANALOG SWITCH CHARACTERISTICS						
ON-Resistance, r_{ON}	$V+ = 4.3V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V+ - 6.5V)$ to $V+$, (See Figure 5)	25	-	0.57	-	Ω
		Full	-	0.68	-	Ω
r_{ON} Matching Between Channels, Δr_{ON}	$V+ = 4.3V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} =$ Voltage at max r_{ON} , (Note 12)	25	-	15	-	$m\Omega$
		Full	-	30	-	$m\Omega$
r_{ON} Flatness, $R_{FLAT(ON)}$	$V+ = 4.3V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V+ - 6.5V)$ to $V+$, (Note 13)	25	-	0.1	-	Ω
		Full	-	0.14	-	Ω
COM ON Leakage Current, $I_{COM(ON)}$	$V+ = 4.3V$, $V_{COM} = -1.2V, 4.3V$, V_{NO} or $V_{NC} =$ Float	25	-0.1	-	0.1	μA
		Full	-	1.1	-	μA
DYNAMIC CHARACTERISTICS						
Turn-ON Time, t_{ON}	$V+ = 3.9V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	43	-	ns
		Full	-	50	-	ns
Turn-OFF Time, t_{OFF}	$V+ = 3.9V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	23	-	ns
		Full	-	23	-	ns
Break-Before-Make Time Delay, t_D	$V+ = 4.5V$, V_{NO} or $V_{NC} = 3.0V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 3)	Full	-	22	-	ns
Charge Injection, Q	$C_L = 1.0nF$, $V_G = 0V$, $R_G = 0\Omega$ (see Figure 2)	25	-	200	-	pC
OFF-Isolation	$R_L = 50\Omega$, $C_L = 5pF$, $f = 100kHz$, $V_{COM} = 1V_{RMS}$ (see Figure 4)	25	-	60	-	dB
Crosstalk (Channel-to-Channel)	$R_L = 50\Omega$, $C_L = 5pF$, $f = 1MHz$, $V_{COM} = 1V_{RMS}$ (see Figure 6)	25	-	-75	-	dB
Total Harmonic Distortion	$f = 20Hz$ to $20kHz$, $V_{COM} = 2V_{P-P}$, $R_L = 32\Omega$	25	-	0.04	-	%
NOx or NCx OFF Capacitance, C_{OFF}	$f = 1MHz$	25	-	36	-	pF
COMx ON Capacitance, $C_{COM(ON)}$	$f = 1MHz$ (see Figure 7)	25	-	88	-	pF

Electrical Specifications - 4.3V Supply Test Conditions: $V_+ = +3.9V$ to $+4.5V$, $GND = 0V$, $V_{INH} = 1.6V$, $V_{INL} = 0.5V$ (Note 9), Unless Otherwise Specified. (Continued)

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
POWER SUPPLY CHARACTERISTICS						
Positive Supply Current, I_+	$V_+ = +4.5V$, $V_{IN} = 0V$ or V_+	25	-	0.003	0.1	μA
		Full	-	2.6	-	μA
Positive Supply Current, I_+	$V_+ = +4.2V$, $V_{IN} = 2.85V$	25	-	0.89	12	μA
DIGITAL INPUT CHARACTERISTICS						
Input Voltage Low, V_{INL}		Full	-	-	0.5	V
Input Voltage High, V_{INH}		Full	1.6	-	-	V
Input Current, I_{INH} , I_{INL}	$V_+ = 4.5V$, $V_{IN} = 0V$ or V_+	25	-0.5	-	0.5	μA
		Full	-	0.5	-	μA

Electrical Specifications - 3V Supply Test Conditions: $V_+ = +2.7V$ to $+3.3V$, $GND = 0V$, $V_{INH} = 1.4V$, $V_{INL} = 0.5V$ (Note 9), Unless Otherwise Specified.

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
ANALOG SWITCH CHARACTERISTICS						
ON-Resistance, r_{ON}	$V_+ = 2.7V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V_+ - 6.5V)$ to V_+ (see Figure 5)	25	-	0.82	-	Ω
		Full	-	0.94	-	Ω
r_{ON} Matching Between Channels, Δr_{ON}	$V_+ = 2.7V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} =$ Voltage at max r_{ON} (Note 13)	25	-	10	-	m Ω
		Full	-	30	-	m Ω
r_{ON} Flatness, $R_{FLAT(ON)}$	$V_+ = 2.7V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V_+ - 6.5V)$ to V_+ (Notes 12, 14)	25	-	0.35	0.5	Ω
		Full	-	0.4	0.55	Ω
DYNAMIC CHARACTERISTICS						
Turn-ON Time, t_{ON}	$V_+ = 2.7V$, V_{NO} or $V_{NC} = 1.5V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	55	-	ns
		Full	-	82	-	ns
Turn-OFF Time, t_{OFF}	$V_+ = 2.7V$, V_{NO} or $V_{NC} = 1.5V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	18	-	ns
		Full	-	24	-	ns
Break-Before-Make Time Delay, t_D	$V_+ = 3.3V$, V_{NO} or $V_{NC} = 1.5V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 3)	Full	-	30	-	ns
Charge Injection, Q	$C_L = 1.0nF$, $V_G = 0V$, $R_G = 0\Omega$ (see Figure 2)	25	-	150	-	pC
OFF-Isolation	$R_L = 50\Omega$, $C_L = 5pF$, $f = 100kHz$, $V_{COM} = 1V_{RMS}$ (see Figure 4)	25	-	60	-	dB
Crosstalk (Channel-to-Channel)	$R_L = 50\Omega$, $C_L = 5pF$, $f = 1MHz$, $V_{COM} = 1V_{RMS}$ (see Figure 6)	25	-	-75	-	dB
Total Harmonic Distortion	$f = 20Hz$ to $20kHz$, $V_{COM} = 0.5V_{P-P}$, $R_L = 32\Omega$	25	-	0.04	-	%
NOx or NCx OFF Capacitance, C_{OFF}	$f = 1MHz$	25	-	36	-	pF
COMx ON Capacitance, $C_{COM(ON)}$	$f = 1MHz$ (see Figure 7)	25	-	88	-	pF
DIGITAL INPUT CHARACTERISTICS						
Input Voltage Low, V_{INL}		25	-	-	0.5	V
Input Voltage High, V_{INH}		25	1.4	-	-	V
Input Current, I_{INH} , I_{INL}	$V_+ = 3.3V$, $V_{IN} = 0V$ or V_+	25	-0.5	-	0.5	μA
		Full	-	0.4	-	μA

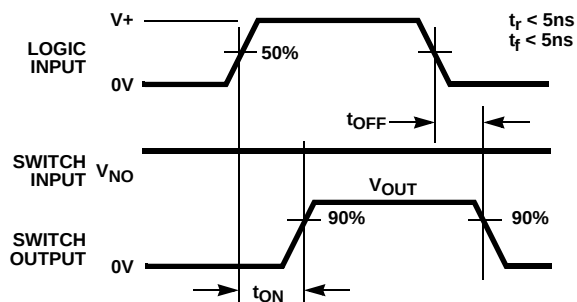
Electrical Specifications - 1.8V Supply Test Conditions: $V_+ = +1.8V$, $GND = 0V$, $V_{INH} = 1.0V$, $V_{INL} = 0.4V$ (Note 9), Unless Otherwise Specified.

PARAMETER	TEST CONDITIONS	TEMP (°C)	MIN (Notes 10, 11)	TYP	MAX (Notes 10, 11)	UNITS
ANALOG SWITCH CHARACTERISTICS						
ON-Resistance, r_{ON}	$V_+ = 1.8V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V_+ - 6.5V)$ to V_+ , (see Figure 5)	25	-	1.87	-	Ω
		Full	-	1.97	-	Ω
r_{ON} Matching Between Channels, Δr_{ON}	$V_+ = 1.8V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} =$ Voltage at max r_{ON} (Note 13)	25	-	16	-	$m\Omega$
		Full	-	30	-	$m\Omega$
r_{ON} Flatness, $R_{FLAT(ON)}$	$V_+ = 1.8V$, $I_{COM} = 100mA$, V_{NO} or $V_{NC} = (V_+ - 6.5V)$ to V_+ (Note 12)	25	-	1.34	-	Ω
		Full	-	1.43	-	Ω
DYNAMIC CHARACTERISTICS						
Turn-ON Time, t_{ON}	$V_+ = 1.8V$, V_{NO} or $V_{NC} = 1.8V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	145	-	ns
		Full	-	150	-	ns
Turn-OFF Time, t_{OFF}	$V_+ = 1.8V$, V_{NO} or $V_{NC} = 1.8V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 1)	25	-	20	-	ns
		Full	-	22	-	ns
Break-Before-Make Time Delay, t_D	$V_+ = 1.8V$, V_{NO} or $V_{NC} = 1.8V$, $R_L = 50\Omega$, $C_L = 35pF$ (see Figure 3)	Full	-	130	-	ns
Charge Injection, Q	$C_L = 1.0nF$, $V_G = 0V$, $R_G = 0\Omega$ (see Figure 2)	25	-	40	-	pC
-3dB Bandwidth	$V_{COM} = 1V_{RMS}$, $R_L = 50\Omega$, $C_L = 5pF$	25	-	60	-	MHz
NOx or NCx OFF Capacitance, C_{OFF}	$f = 1MHz$	25	-	36	-	pF
COMx ON Capacitance, $C_{COM(ON)}$	$f = 1MHz$ (see Figure 7)	25	-	88	-	pF
DIGITAL INPUT CHARACTERISTICS						
Input Voltage Low, V_{INL}		25	-	-	0.4	V
Input Voltage High, V_{INH}		25	1.0	-	-	V
Input Current, I_{INH} , I_{INL}	$V_+ = 2.0V$, $V_{IN} = 0V$ or V_+	25	-0.5	-	0.5	μA
Input Current, I_{INH} , I_{INL}	$V_+ = 2.0V$, $V_{IN} = 0V$ or V_+	Full	-	0.38	-	μA

NOTES:

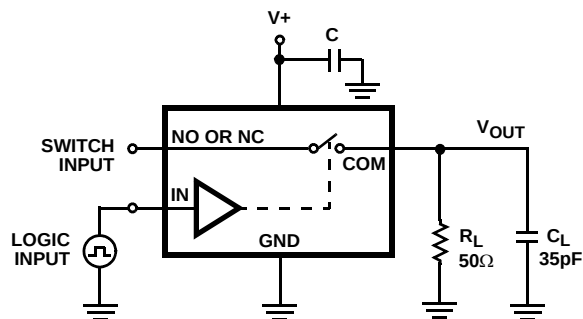
9. V_{IN} = input voltage to perform proper function.
10. The algebraic convention, whereby the most negative value is a minimum and the most positive a maximum, is used in this data sheet.
11. Parameters with MIN and/or MAX limits are 100% tested at +25°C, unless otherwise specified. Temperature limits established by characterization and are not production tested.
12. Flatness is defined as the difference between maximum and minimum value of on-resistance over the specified analog signal range.
13. r_{ON} matching between channels is calculated by subtracting the channel with the highest max r_{ON} value from the channel with lowest max r_{ON} value, between NC1 and NC2 or between NO1 and NO2.
14. Limits established by characterization and are not production tested.

Test Circuits and Waveforms



Logic input waveform is inverted for switches that have the opposite logic sense.

FIGURE 1A. MEASUREMENT POINTS



Repeat test for all switches. C_L includes fixture and stray capacitance.

$$V_{OUT} = V_{(NO \text{ or } NC)} \frac{R_L}{R_L + r_{ON}}$$

FIGURE 1B. TEST CIRCUIT

FIGURE 1. SWITCHING TIMES

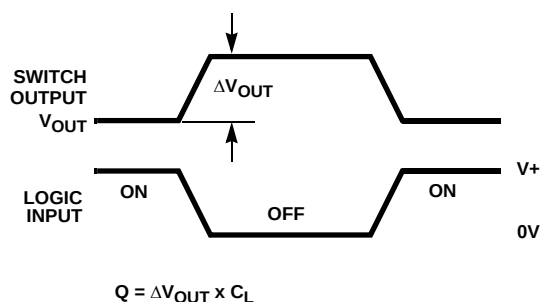
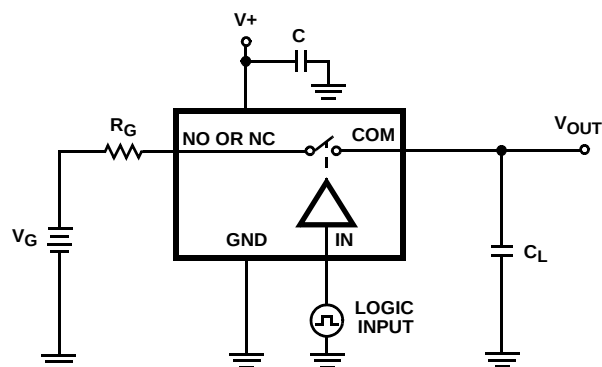


FIGURE 2A. MEASUREMENT POINTS



Repeat test for all switches.

FIGURE 2B. TEST CIRCUIT

FIGURE 2. CHARGE INJECTION

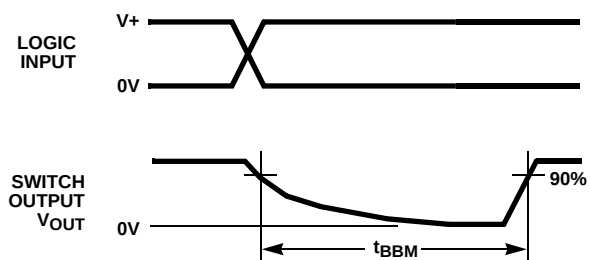
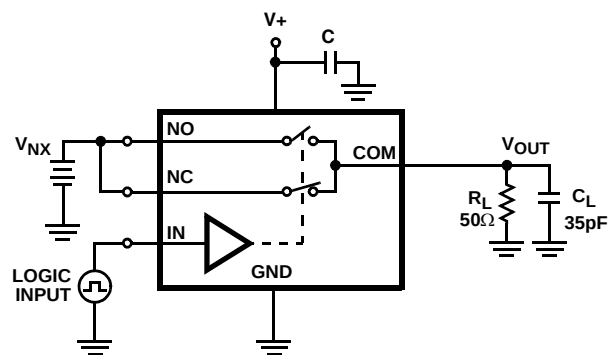


FIGURE 3A. MEASUREMENT POINTS

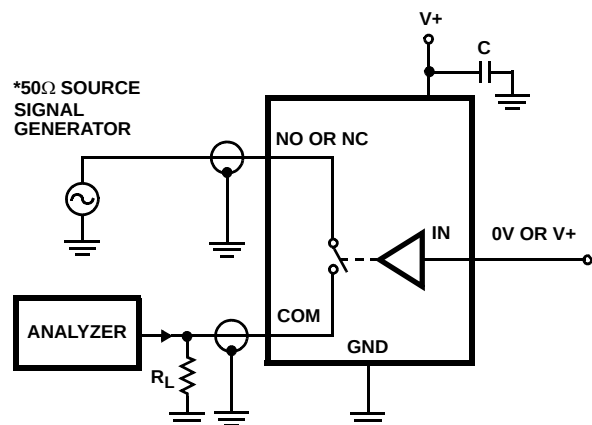


Repeat test for all switches. C_L includes fixture and stray capacitance.

FIGURE 3B. TEST CIRCUIT

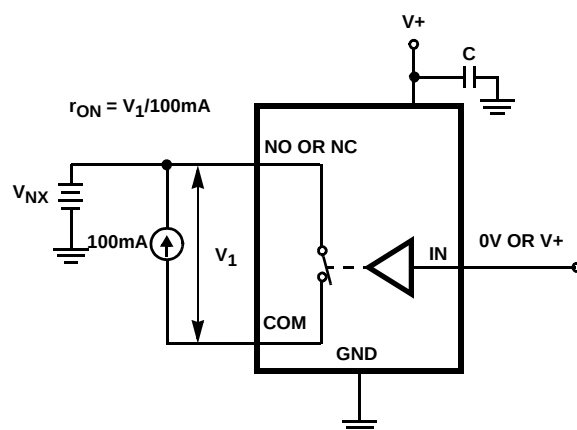
FIGURE 3. BREAK-BEFORE-MAKE TIME

Test Circuits and Waveforms (Continued)



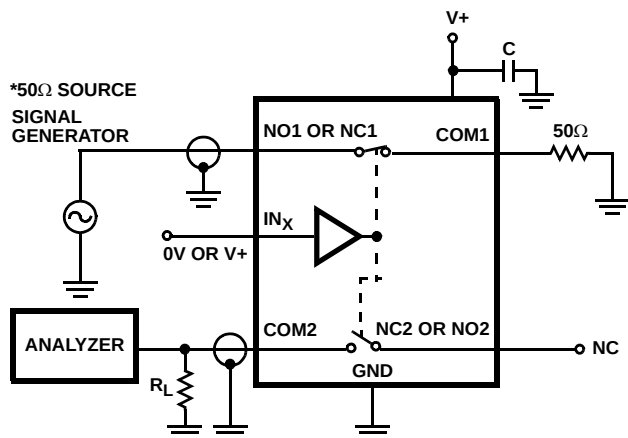
Signal direction through switch is reversed, worst case values are recorded. Repeat test for all switches.

FIGURE 4. OFF-ISOLATION TEST CIRCUIT



Repeat test for all switches.

FIGURE 5. r_{ON} TEST CIRCUIT



Signal direction through switch is reversed, worst case values are recorded. Repeat test for all switches.

FIGURE 6. CROSSTALK TEST CIRCUIT

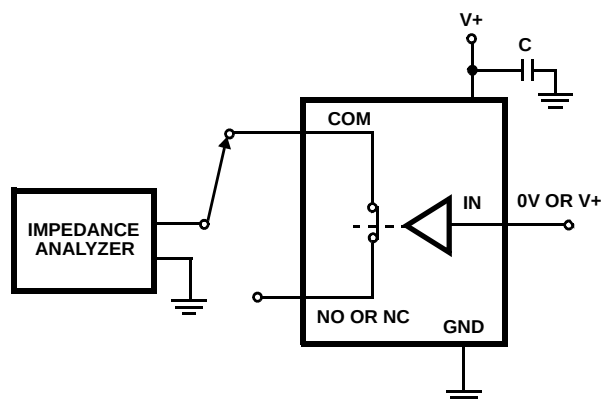
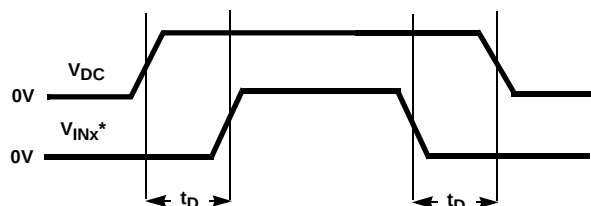


FIGURE 7. ON CAPACITANCE TEST CIRCUIT

Test Circuits and Waveforms (Continued)



* V_{INx} waveform for Click and Pop Elimination on NOx terminal.
For Click and Pop Elimination on NCx terminal invert V_{INx} .

$t_D = 200\text{ms}$ measured at 50% points.

FIGURE 8A. CLICK AND POP WAVEFORM

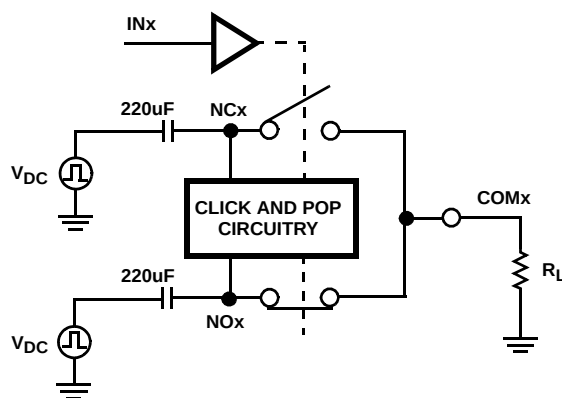


FIGURE 8B. CLICK AND POP TEST CIRCUIT

FIGURE 8. CLICK AND POP ELIMINATION

Detailed Description

The ISL54062 is a bi-directional, dual single pole-double throw (SPDT) analog switch that offers precise switching from a single 1.8V to 6.5V supply with low ON-resistance (0.83Ω), high speed operation ($t_{ON} = 55\text{ns}$, $t_{OFF} = 18\text{ns}$) and negative signal swing capability. The device is especially well suited for portable battery powered equipment due to its low operating supply voltage (1.8V), low power consumption (20nA @ 3V), and a tiny $1.8\text{mm} \times 1.4\text{mm}$ μTQFN package or a 3×3 TDFN package. The low ON-resistance and r_{ON} flatness provide very low insertion loss and signal distortion for applications that require signal switching with minimal interference by the switch.

Supply Sequencing and Overvoltage Protection

With any CMOS device, proper power supply sequencing is required to protect the device from excessive input currents which might permanently damage the IC. The ISL54062 contains ESD protection diodes on each pin of the IC (see Figure 9). These diodes connect to either a +Ring or -Ring for ESD protection. To prevent forward biasing the ESD diodes to the +Ring, V_+ must be applied before any input signals, and the input signal voltages must remain between recommended operating range.

If these conditions cannot be guaranteed, then precautions must be implemented to prohibit the current and voltage at the logic pin and signal pins from exceeding the maximum ratings of the switch. The following two methods can be used to provided additional protection to limit the current in the event that the voltage at a logic pin or switch terminal goes above the V_+ rail.

Logic inputs can be protected by adding a $1\text{k}\Omega$ resistor in series with the logic input (see Figure 9). The resistor limits

the input current below the threshold that produces permanent damage.

This method is not acceptable for the signal path inputs. Adding a series resistor to the switch input defeats the purpose of using a low r_{ON} switch. Alternatively, connecting external Schottky diodes from the V_+ rail to the signal pins will shunt the fault current through the Schottky diode instead of through the internal ESD diodes, thereby protecting the switch. These Schottky diodes must be sized to handle the expected fault current.

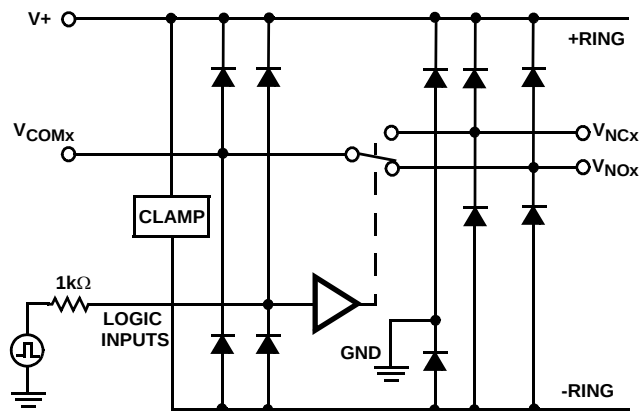


FIGURE 9. OVERVOLTAGE PROTECTION

Power-Supply Considerations

The ISL54062 construction is typical of most single supply CMOS analog switches which have two supply pins: V_+ and GND. V_+ and GND provide the CMOS switch bias and sets their analog voltage limits. Unlike switches with a 5.5V maximum supply voltage, the ISL54062's 6.5V maximum supply voltage provides plenty of head room for the 10% tolerance of 5V supplies due to overshoot and noise spikes.

The minimum recommended supply voltage is 1.8V. It is important to note that the input signal range, switching times, and ON-resistance degrade at lower supply voltages. Refer to the “Electrical Specifications” tables, beginning on page 3, and “Typical Performance Curves”, beginning on page 11, for details.

V+ and GND also power the internal logic and level shifters. The level shifters convert the input logic levels to V+ and GND signals levels to drive the analog switch gate terminals. A high frequency decoupling capacitor placed as close to the V+ and GND pin as possible is recommended for proper operation of the switch. A value of 0.1μF is highly recommended.

Negative Signal Swing Capability

The ISL54062 contains circuitry that allows the analog switch signal to swing below ground. The device has an analog signal range of 6.5V below V₊ up to the V₊ rail (see Figure 16) while maintaining low r_{ON} performance. For example, if V₊ = 5V, then the analog input signal range is from -1.5V to +5V. If V₊ = 2.7V then the range is from -3.8V to +2.7V.

Click and Pop Operation

The ISL54062 contains circuitry that prevents audible click and pop noises that may occur when audio sources are powered on or off. Single supply audio sources are biased at a DC offset that can generate transients during power on/off. A DC blocking capacitor is needed to remove the DC bias at the speaker load. For 32Ω headphones, a $220\mu\text{F}$ capacitor is typically used to preserve the audio bandwidth. The power on/off transients are AC coupled by the $220\mu\text{F}$ capacitor to the speaker load causing a click and pop noise.

The ISL54062 has shunt switches on the NO and NC pins to eliminate click and pop transients (see Figure 10). These switches are driven complimentary to the main switch. When NC is connected to COM, the shunt switch is active on the NO pin (and vice versa). The shunt switches connect an impedance (140Ω typical, see Figure 24) from the NO/NC pin to ground to discharge any transients that may appear on the NO or NC pins.

When a DC bias becomes active at the source, the NO and NC terminals will also have a DC offset due to capacitor dv/dt principle. The DC offset will be discharged through the shunt impedance on the NO and NC terminals instead of the speaker, eliminating click and pop noise.

*Under high impedance loads such as the input impedance of pre-amplifiers ($20k\Omega$), the COM terminal voltage may rise due to small leakage currents charging the COM capacitance. This is not seen when low impedance loads such as headphones (32Ω) are used because the small leakage currents does not result in significant potential drop across the load. If the user desires to reduce the voltage build up on the COM pin, a $1k\Omega$ resistor to ground may be

placed on the COM pin. This impedance is small enough to reduce the voltage build up significantly while not increasing the power dissipation dramatically. Current consumption considerations will need to be taken for driving a smaller load impedance under this scenario.

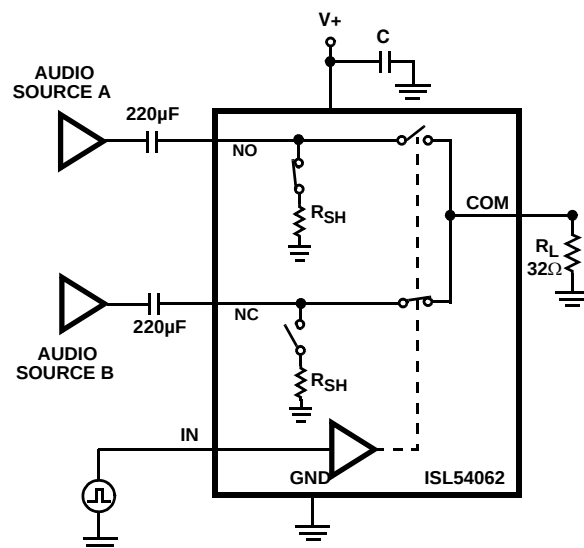


FIGURE 10. CLICK AND POP OPERATION

Click and Pop Elimination with INx Pin

Audio click and pop elimination can be driven with the Input Select (INx) pin. When $INx = 0$, the NOx terminals are connected to the shunt impedance. When $INx = 1$, the NCx terminals are connected to the shunt impedance. In this situation, only one of the source transient voltages will be shunted to ground, depending on the Input Select state. The Input Select pin should be driven 200ms after any source transients occurs to prevent audible transients at the speaker load.

Logic-Level Thresholds

This switch family is 1.8V CMOS compatible ($0.45\text{V } V_{OLMAX}$ and $1.35\text{V } V_{OHMIN}$) over a supply range of 1.8V to 3.3V (see Figure 18). At 3.3V the V_{IL} level is 0.5V maximum. This is still below the 1.8V CMOS guaranteed low output maximum level of 0.45V, but noise margin is reduced to approximately 50mV. At 3.3V the V_{IH} level is 1.4V minimum. While this is above the 1.8V CMOS guaranteed high output minimum of 1.35V, under most operating conditions the switch will recognize this as a valid logic high.

The digital input stages draw supply current whenever the digital input voltage is not at one of the supply rails. Driving the digital input signals from GND to V+ with a fast transition time minimizes power dissipation. The ISL54062 has been designed to minimize the supply current whenever the digital input voltage is not driven to the supply rails (0V to V+). For example, driving the device with 2.85V logic high while operating with a 4.2V supply the device draws only 1μA of current.

High-Frequency Performance

In 50Ω systems, the ISL54062 has an ON switch -3dB bandwidth of 60MHz (see Figure 21). The frequency response is very consistent over a wide V+ range, and for varying analog signal levels.

An OFF switch acts like a capacitor across the open terminals and AC couples higher frequencies, resulting in signal feed-through from a switch's input to its output. Off-Isolation is the resistance to this feed-through. Crosstalk indicates the amount of feed-through from one switch channel to another switch channel. Figure 22 details the high Off-Isolation and Crosstalk rejection provided by this part. At 100kHz, Off-Isolation is about 60dB in 50Ω systems, decreasing approximately 20dB per decade as frequency increases. At 1MHz, Crosstalk is about -75dB in 50Ω systems, decreasing approximately 20dB per decade as frequency increases.

Typical Performance Curves $T_A = +25^\circ\text{C}$, Unless Otherwise Specified

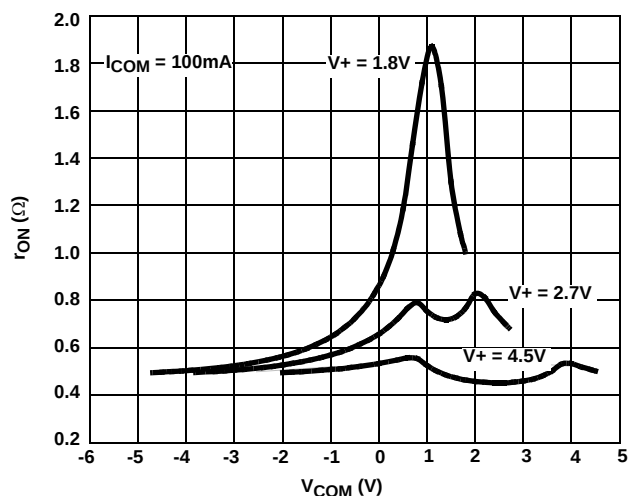


FIGURE 11. ON-RESISTANCE vs SUPPLY VOLTAGE vs SWITCH VOLTAGE

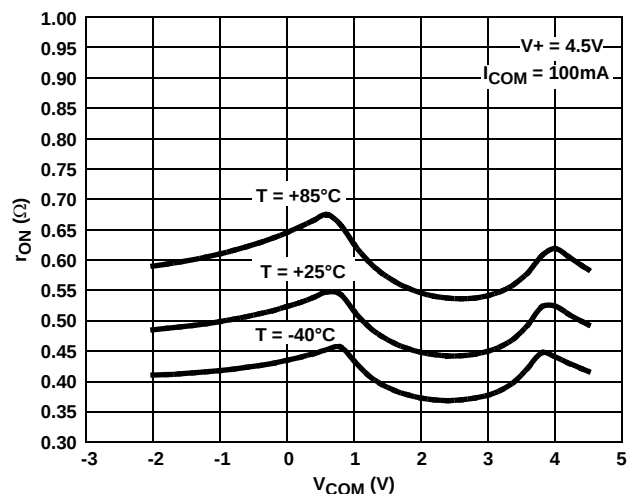


FIGURE 12. ON-RESISTANCE vs SWITCH VOLTAGE

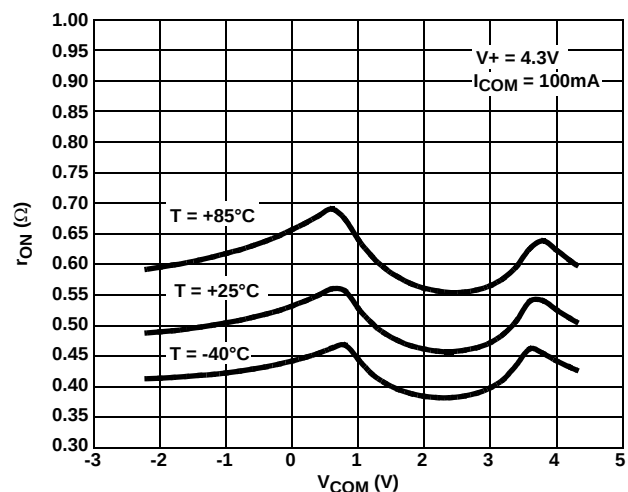


FIGURE 13. ON-RESISTANCE vs SWITCH VOLTAGE

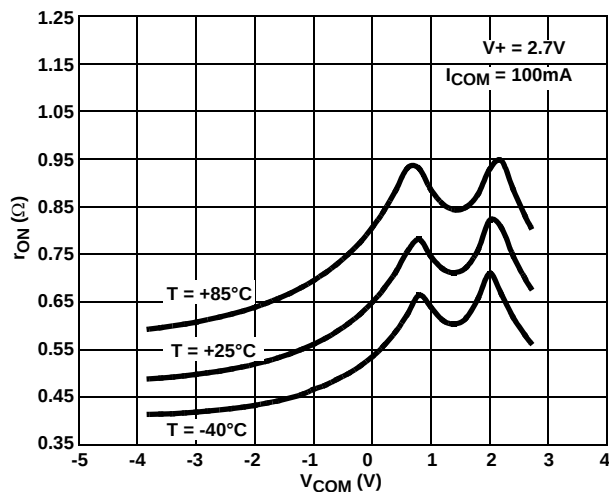


FIGURE 14. ON-RESISTANCE vs SWITCH VOLTAGE

Leakage Considerations

Reverse ESD protection diodes are internally connected between each analog-signal pin, V+ and GND. One of these diodes conducts if any analog signal exceeds the recommended analog signal range.

Virtually all the analog switch leakage current comes from the ESD diodes and reversed biased junctions in the switch cell. Although the ESD diodes on a given signal pin are identical and therefore fairly well balanced, they are reverse biased differently. Each is biased to either the +Ring or -Ring and the analog input signal. This means their leakages will vary as the signal varies. The difference in the two diode leakages to the +Ring or -Ring and the reverse biased junctions at the internal switch cell constitutes the analog-signal-path leakage current.

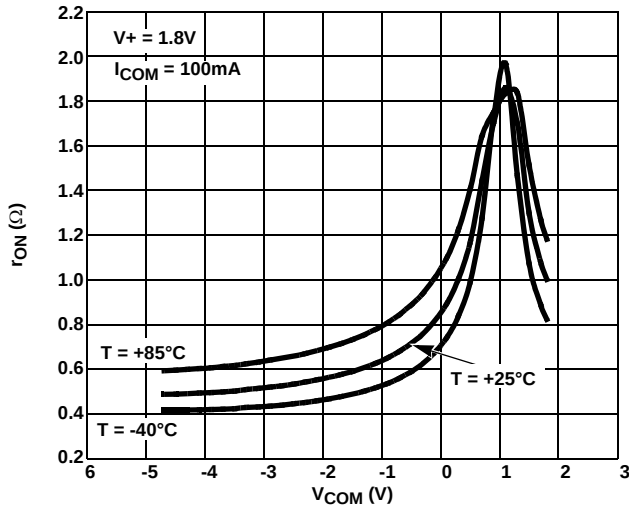
Typical Performance Curves $T_A = +25^\circ\text{C}$, Unless Otherwise Specified (Continued)


FIGURE 15. ON-RESISTANCE vs SWITCH VOLTAGE

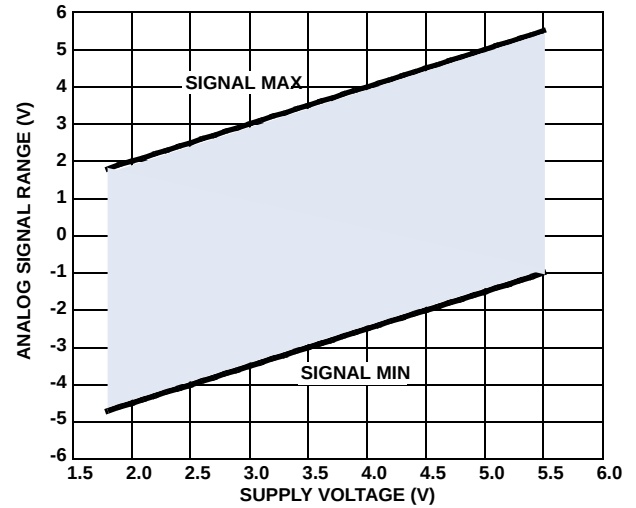


FIGURE 16. ANALOG SIGNAL RANGE vs SUPPLY VOLTAGE

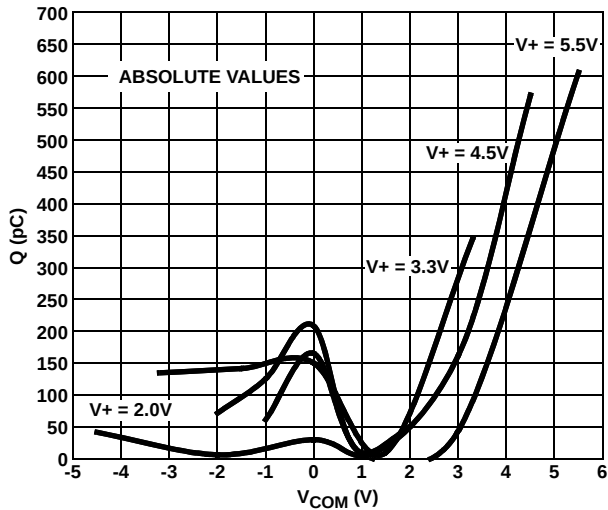


FIGURE 17. CHARGE INJECTION vs SWITCH VOLTAGE

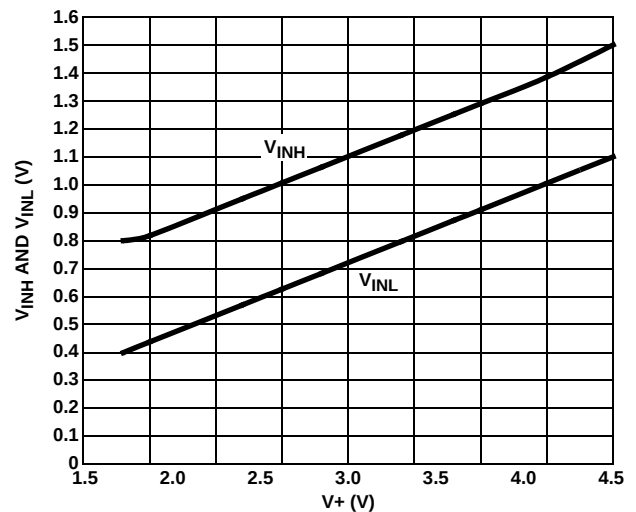


FIGURE 18. DIGITAL SWITCHING POINT vs SUPPLY VOLTAGE

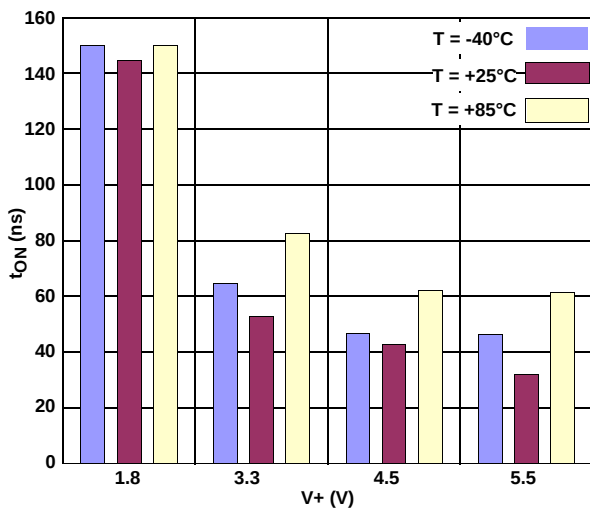


FIGURE 19. TURN-ON TIME vs SUPPLY VOLTAGE

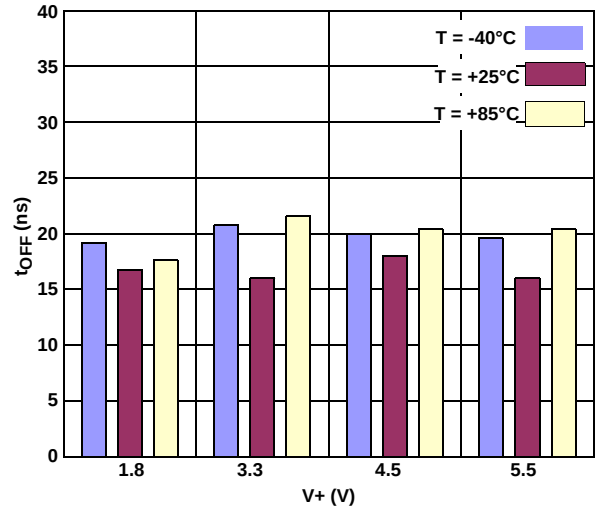


FIGURE 20. TURN-OFF TIME vs SUPPLY VOLTAGE

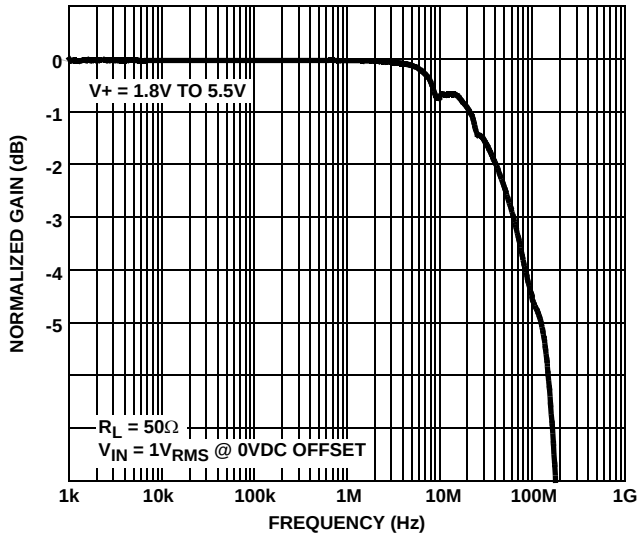
Typical Performance Curves $T_A = +25^\circ\text{C}$, Unless Otherwise Specified (Continued)


FIGURE 21. FREQUENCY RESPONSE

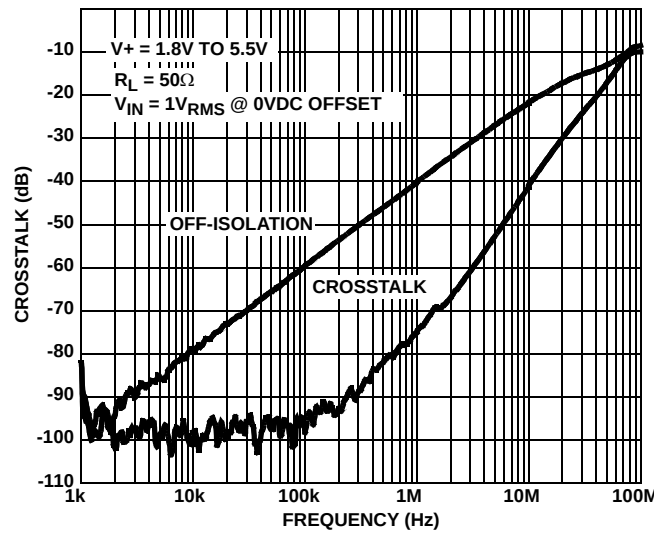


FIGURE 22. CROSSTALK AND OFF-ISOLATION

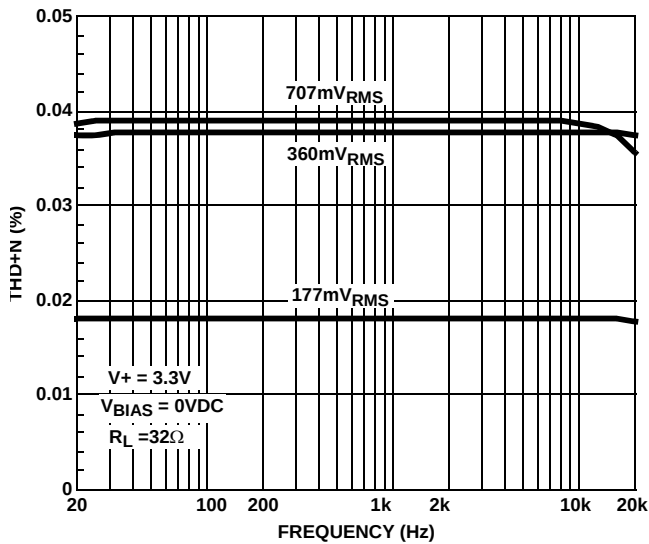


FIGURE 23. TOTAL HARMONIC DISTORTION vs FREQUENCY

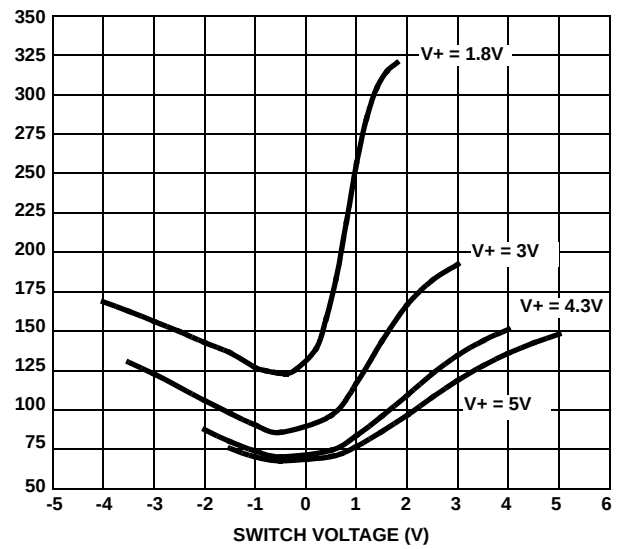


FIGURE 24. SHUNT RESISTANCE vs SWITCH VOLTAGE

Typical Performance Curves $T_A = +25^\circ\text{C}$, Unless Otherwise Specified (Continued)

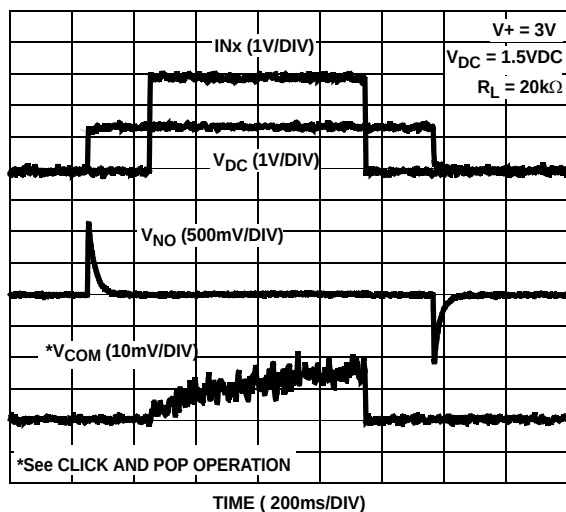


FIGURE 25. CLICK AND POP ELIMINATION 20k Ω LOAD
200ms DELAY

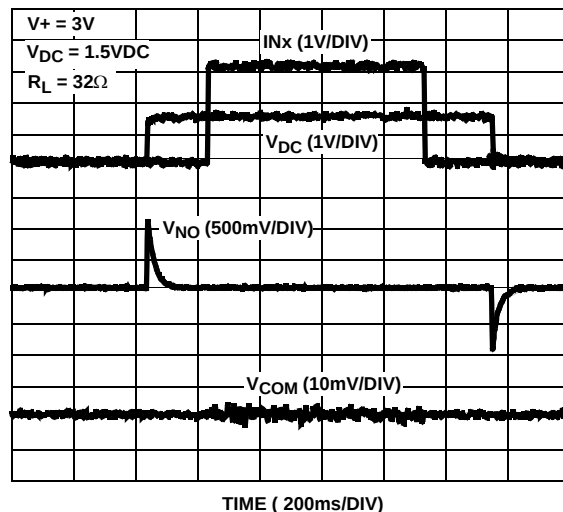


FIGURE 26. CLICK AND POP ELIMINATION 32 Ω LOAD 200ms
DELAY

Die Characteristics

SUBSTRATE POTENTIAL (POWERED UP):

GND (DFN Paddle Connection: Tie to GND or Float)

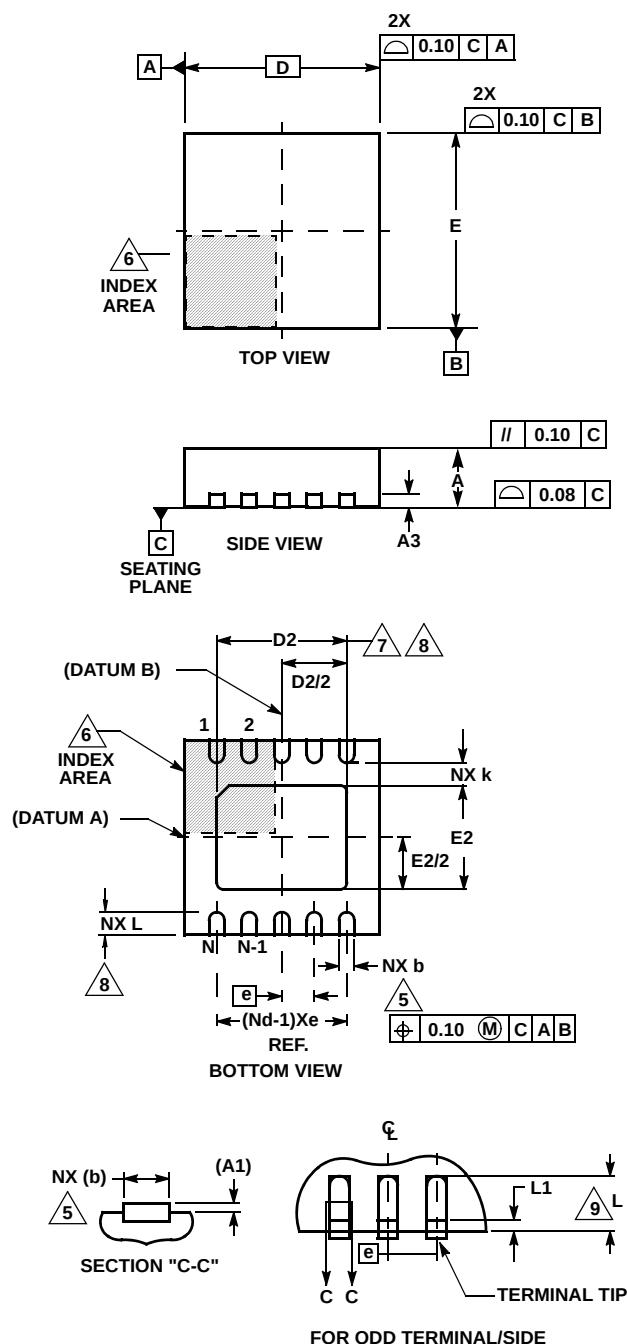
TRANSISTOR COUNT:

432

PROCESS:

Submicron CMOS

Thin Dual Flat No-Lead Plastic Package (TDFN)



L10.3x3A

10 LEAD THIN DUAL FLAT NO-LEAD PLASTIC PACKAGE

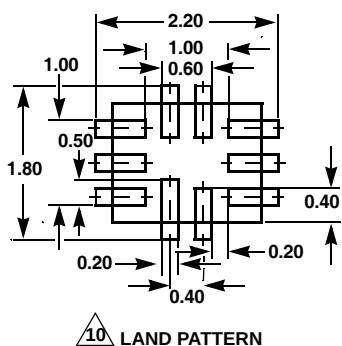
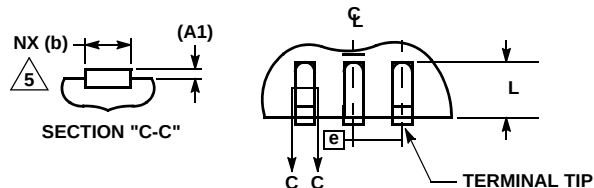
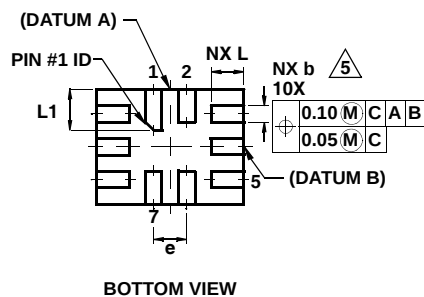
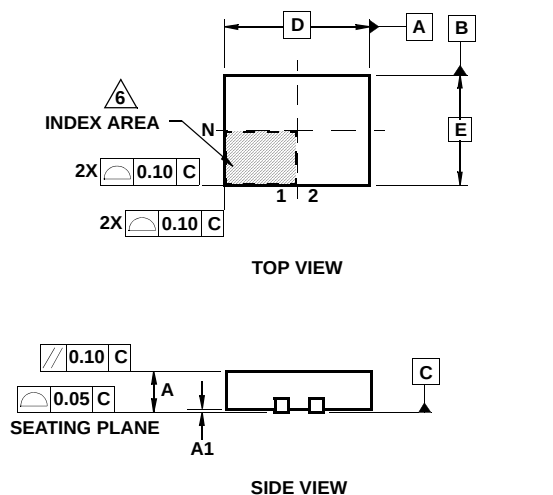
SYMBOL	MILLIMETERS			NOTES
	MIN	NOMINAL	MAX	
A	0.70	0.75	0.80	-
A1	-	-	0.05	-
A3	0.20 REF			-
b	0.20	0.25	0.30	5, 8
D	2.95	3.0	3.05	-
D2	2.25	2.30	2.35	7, 8
E	2.95	3.0	3.05	-
E2	1.45	1.50	1.55	7, 8
e	0.50 BSC			-
k	0.25	-	-	-
L	0.25	0.30	0.35	8
N	10			2
Nd	5			3

Rev. 3 3/06

NOTES:

1. Dimensioning and tolerancing conform to ASME Y14.5-1994.
2. N is the number of terminals.
3. Nd refers to the number of terminals on D.
4. All dimensions are in millimeters. Angles are in degrees.
5. Dimension b applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.
7. Dimensions D2 and E2 are for the exposed pads which provide improved electrical and thermal performance.
8. Nominal dimensions are provided to assist with PCB Land Pattern Design efforts, see Intersil Technical Brief TB389.
9. Compliant to JEDEC MO-229-WEED-3 except for D2 dimensions.

Ultra Thin Quad Flat No-Lead Plastic Package (UTQFN)



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Intersil Corporation's quality certifications can be viewed at www.intersil.com/design/quality

L10.1.8x1.4A**10 LEAD ULTRA THIN QUAD FLAT NO-LEAD PLASTIC PACKAGE**

SYMBOL	MILLIMETERS			NOTES
	MIN	NOMINAL	MAX	
A	0.45	0.50	0.55	-
A1	-	-	0.05	-
A3	0.127 REF			-
b	0.15	0.20	0.25	5
D	1.75	1.80	1.85	-
E	1.35	1.40	1.45	-
e	0.40 BSC			-
L	0.35	0.40	0.45	-
L1	0.45	0.50	0.55	-
N	10			2
Nd	2			3
Ne	3			3
θ	0	-	12	4

Rev. 3 6/06

NOTES:

1. Dimensioning and tolerancing conform to ASME Y14.5-1994.
2. N is the number of terminals.
3. Nd and Ne refer to the number of terminals on D and E side, respectively.
4. All dimensions are in millimeters. Angles are in degrees.
5. Dimension b applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.
7. Maximum package warpage is 0.05mm.
8. Maximum allowable burrs is 0.076mm in all directions.
9. JEDEC Reference MO-255.
10. For additional information, to assist with the PCB Land Pattern Design effort, see Intersil Technical Brief TB389.

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