

HIGH VOLTAGE FASTSWITCHING NPN POWER TRANSISTOR

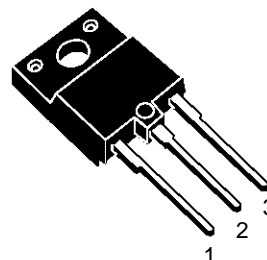
- SGS-THOMSON PREFERRED SALESTYPE
- HIGH VOLTAGE CAPABILITY
- U.L. RECOGNISED ISOWATT218 PACKAGE (U.L. FILE # E81734 (N))

APPLICATIONS:

- HORIZONTAL DEFLECTION FOR MONITORS
- SWITCH MODE POWER SUPPLIES

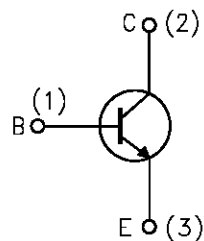
DESCRIPTION

The BUH715 is manufactured using Multiepitaxial Mesa technology for cost-effective high performance and uses a Hollow Emitter structure to enhance switching speeds. .



ISOWATT218

INTERNAL SCHEMATIC DIAGRAM



SC06960

ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Value | Unit |
|-----------|---|------------|------------------|
| V_{CBO} | Collector-Base Voltage ($I_E = 0$) | 1500 | V |
| V_{CEO} | Collector-Emitter Voltage ($I_B = 0$) | 700 | V |
| V_{EBO} | Emitter-Base Voltage ($I_C = 0$) | 10 | V |
| I_C | Collector Current | 10 | A |
| I_{CM} | Collector Peak Current ($t_p < 5$ ms) | 20 | A |
| I_B | Base Current | 5 | A |
| I_{BM} | Base Peak Current ($t_p < 5$ ms) | 10 | A |
| P_{tot} | Total Dissipation at $T_C = 25^\circ\text{C}$ | 57 | W |
| T_{stg} | Storage Temperature | -65 to 150 | $^\circ\text{C}$ |
| T_j | Max. Operating Junction Temperature | 150 | $^\circ\text{C}$ |

THERMAL DATA

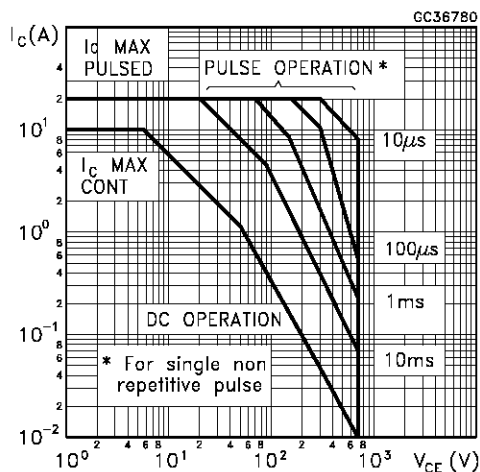
| | | | | |
|-----------------------|----------------------------------|-----|-----|------|
| R _{thj-case} | Thermal Resistance Junction-case | Max | 2.2 | °C/W |
|-----------------------|----------------------------------|-----|-----|------|

ELECTRICAL CHARACTERISTICS (T_{case} = 25 °C unless otherwise specified)

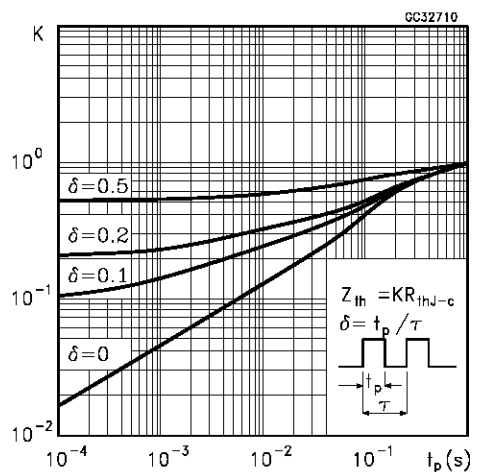
| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
|----------------------------------|---|---|--------|------------|------------|----------|
| I _{CES} | Collector Cut-off Current (V _{BE} = 0) | V _{CE} = 1500 V V _{CE} = 1500 V T _j = 125 °C | | | 1 2 | mA mA |
| I _{EBO} | Emitter Cut-off Current (I _C = 0) | V _{EB} = 5 V | | | 100 | μA |
| V _{CEO(sus)} | Collector-Emitter Sustaining Voltage | I _C = 100 mA | 700 | | | V |
| V _{EBO} | Emitter-Base Voltage (I _C = 0) | I _E = 10 mA | 10 | | | V |
| V _{CE(sat)*} | Collector-Emitter Saturation Voltage | I _C = 7 A I _B = 1.5 A | | | 1.5 | V |
| V _{BE(sat)*} | Base-Emitter Saturation Voltage | I _C = 7 A I _B = 1.5 A | | | 1.3 | V |
| h _{FE*} | DC Current Gain | I _C = 7 A V _{CE} = 5 V I _C = 7 A V _{CE} = 5 V T _j = 100 °C | 8 5 | | 16 | |
| t _s t _f | RESISTIVE LOAD Storage Time Fall Time | V _{CC} = 400 V I _C = 7 A I _{B1} = 1.5 A I _{B2} = 3.5 A | | 2.1 140 | 3.1 210 | μs ns |
| t _s t _f | INDUCTIVE LOAD Storage Time Fall Time | I _C = 7 A f = 15625 Hz I _{B1} = 1.5 A I _{B2} = -3.5 A V _{ceflyback} = 1050 sin(π/10 10 ⁶) t V | | 3.5 350 | | μs ns |
| t _s t _f | INDUCTIVE LOAD Storage Time Fall Time | I _C = 7 A f = 31250 Hz I _{B1} = 1.5 A I _{B2} = -3.5 A V _{ceflyback} = 1200 sin(π/5 10 ⁶) t V | | 3.5 320 | | μs ns |

* Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %

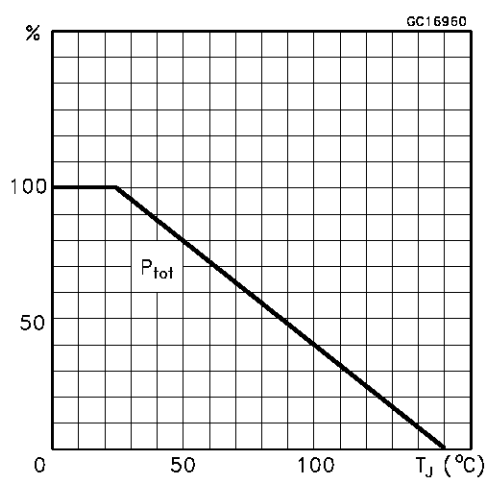
Safe Operating Area



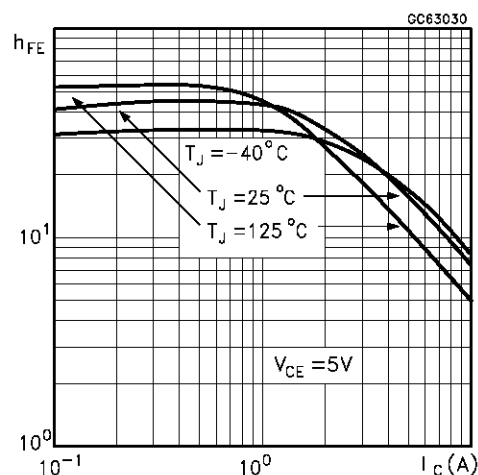
Thermal Impedance



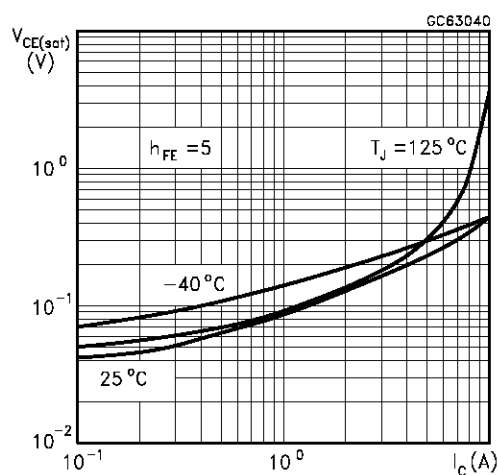
Derating Curve



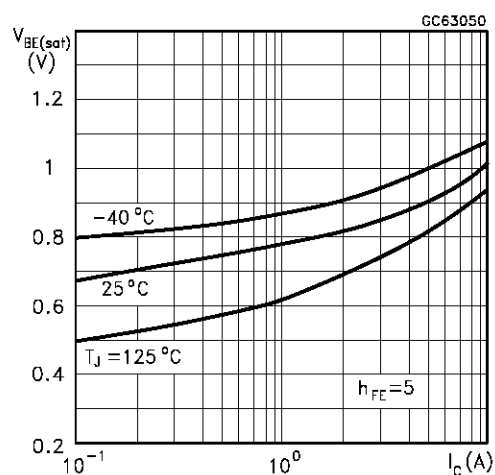
DC Current Gain



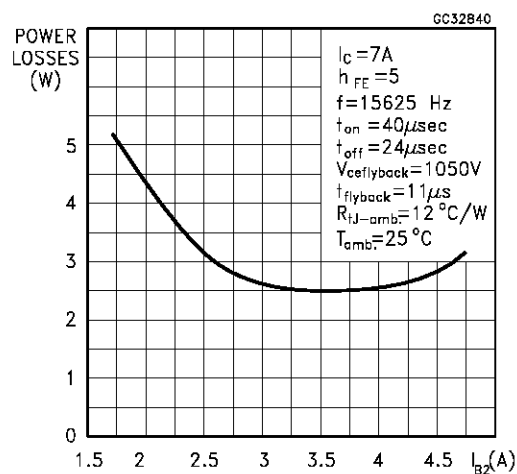
Collector Emitter Saturation Voltage



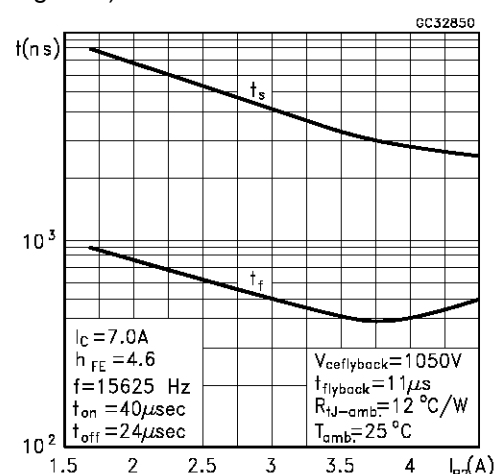
Base Emitter Saturation Voltage



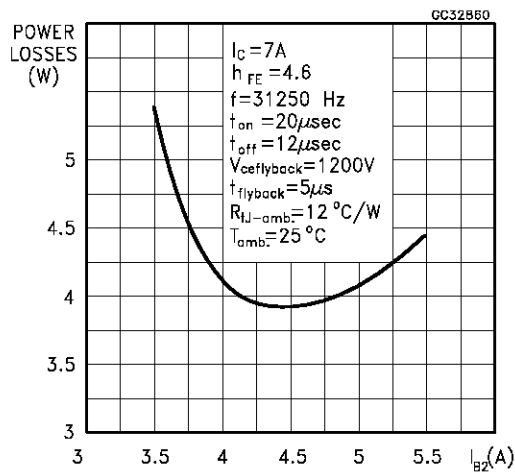
Power Losses at 16 KHz



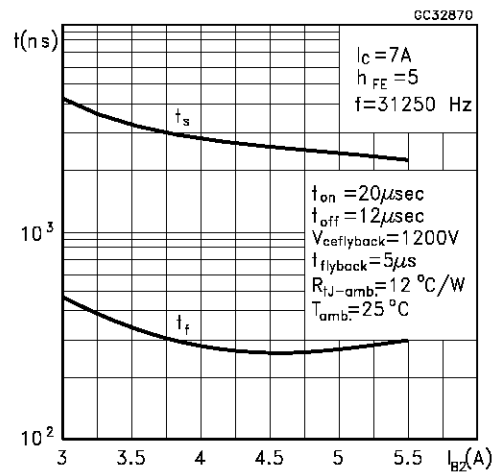
Switching Time Inductive Load at 16KHz
(see figure 2)



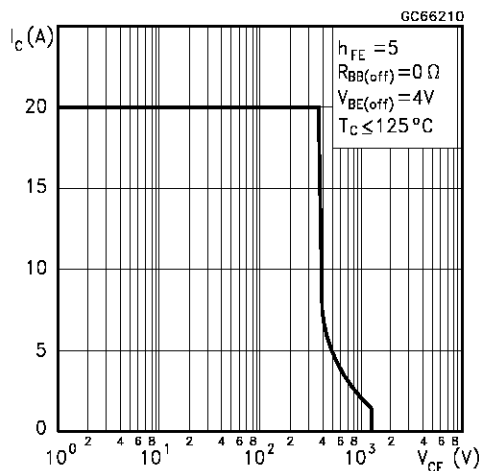
Power Losses at 32 KHz



Switching Time Inductive Load at 32 KHz (see figure 2)



Reverse Biased SOA



BASE DRIVE INFORMATION

In order to saturate the power switch and reduce conduction losses, adequate direct base current I_{B1} has to be provided for the lowest gain h_{FE} at 100 °C (line scan phase). On the other hand, negative base current I_{B2} must be provided to turn off the power transistor (retrace phase).

Most of the dissipation, in the deflection application, occurs at switch-off. Therefore it is essential to determine the value of I_{B2} which minimizes power losses, fall time t_f and, consequently, T_j . A new set of curves have been defined to give total power losses, t_s and t_f as a function of I_{B2} at both 16 KHz, 32 KHz and 64KHz scanning frequencies for choosing the optimum negative drive. The test circuit is

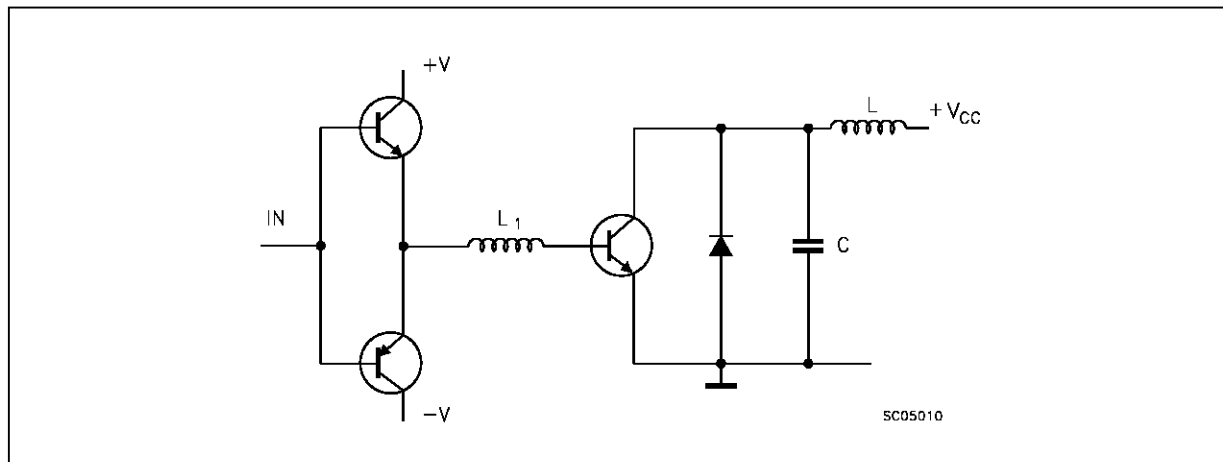
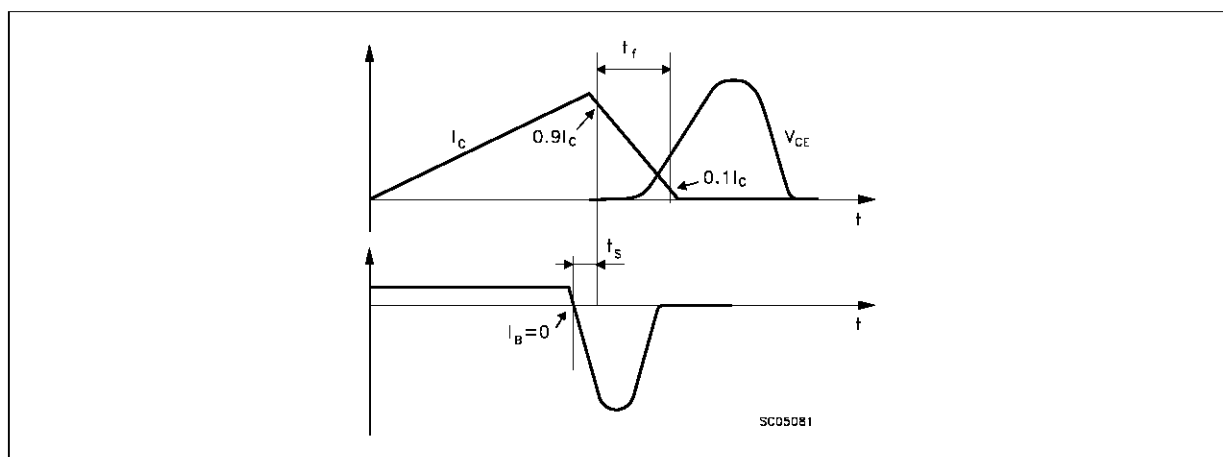
illustrated in figure 1.

Inductance L_1 serves to control the slope of the negative base current I_{B2} to recombine the excess carrier in the collector when base current is still present, this would avoid any tailing phenomenon in the collector current.

The values of L and C are calculated from the following equations:

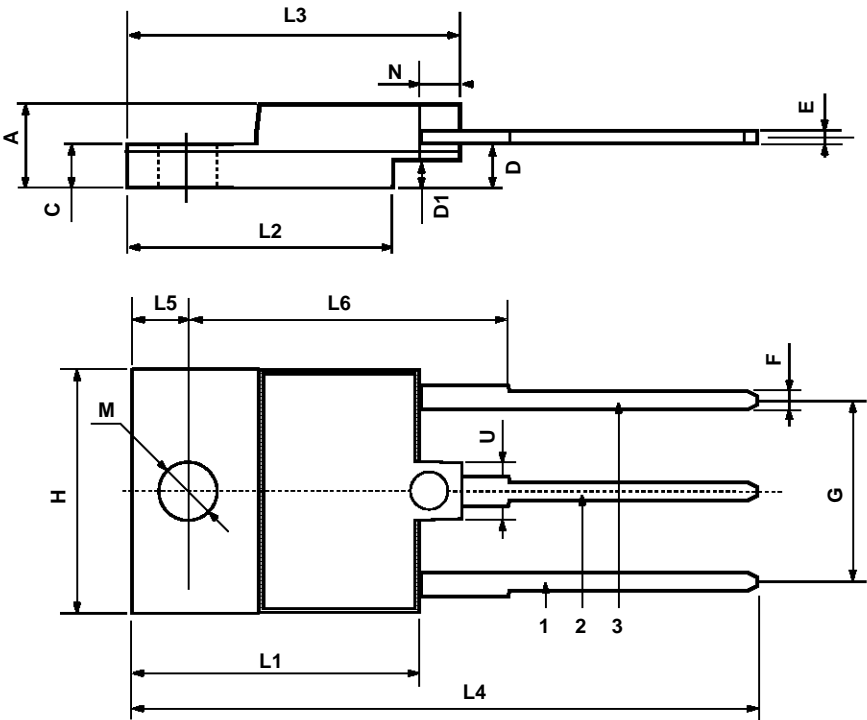
$$\frac{1}{2} L (I_C)^2 = \frac{1}{2} C (V_{CEfly})^2 \quad \omega = 2 \pi f = \frac{1}{\sqrt{LC}}$$

Where I_C = operating collector current, V_{CEfly} = flyback voltage, f = frequency of oscillation during retrace.

Figure 1: Inductive Load Switching Test Circuits.**Figure 2:** Switching Waveforms in a Deflection Circuit

ISOWATT218 MECHANICAL DATA

| DIM. | mm | | | inch | | |
|------|-------|------|-------|-------|-------|-------|
| | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | 5.35 | | 5.65 | 0.210 | | 0.222 |
| C | 3.3 | | 3.8 | 0.130 | | 0.149 |
| D | 2.9 | | 3.1 | 0.114 | | 0.122 |
| D1 | 1.88 | | 2.08 | 0.074 | | 0.081 |
| E | 0.45 | | 1 | 0.017 | | 0.039 |
| F | 1.05 | | 1.25 | 0.041 | | 0.049 |
| G | 10.8 | | 11.2 | 0.425 | | 0.441 |
| H | 15.8 | | 16.2 | 0.622 | | 0.637 |
| L1 | 20.8 | | 21.2 | 0.818 | | 0.834 |
| L2 | 19.1 | | 19.9 | 0.752 | | 0.783 |
| L3 | 22.8 | | 23.6 | 0.897 | | 0.929 |
| L4 | 40.5 | | 42.5 | 1.594 | | 1.673 |
| L5 | 4.85 | | 5.25 | 0.190 | | 0.206 |
| L6 | 20.25 | | 20.75 | 0.797 | | 0.817 |
| M | 3.5 | | 3.7 | 0.137 | | 0.145 |
| N | 2.1 | | 2.3 | 0.082 | | 0.090 |
| U | | 4.6 | | | 0.181 | |



P025C

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