

## FEATURES

- Maximum Offset Voltage of  $3\mu\text{V}$
- Maximum Offset Voltage Drift of  $30\text{nV}/^\circ\text{C}$
- Small Footprint, Low Profile MS8/GN16 Packages
- Single Supply Operation:  $2.7\text{V}$  to  $\pm 5.5\text{V}$
- Noise:  $1.5\mu\text{V}_{\text{P-P}}$  (0.01Hz to 10Hz Typ)
- Voltage Gain: 140dB (Typ)
- PSRR: 130dB (Typ)
- CMRR: 130dB (Typ)
- Supply Current: 0.75mA (Typ) per Amplifier
- Extended Common Mode Input Range
- Output Swings Rail-to-Rail
- Operating Temperature Range  $-40^\circ\text{C}$  to  $125^\circ\text{C}$

## APPLICATIONS


- Thermocouple Amplifiers
- Electronic Scales
- Medical Instrumentation
- Strain Gauge Amplifiers
- High Resolution Data Acquisition
- DC Accurate RC Active Filters
- Low Side Current Sense

## DESCRIPTION

The LTC<sup>®</sup>2051/LTC2052 are dual/quad zero-drift operational amplifiers available in the MS8 and SO-8/GN16 and S14 packages. They operate from a single  $2.7\text{V}$  supply and support  $\pm 5\text{V}$  applications. The current consumption is  $750\mu\text{A}$  per op amp.

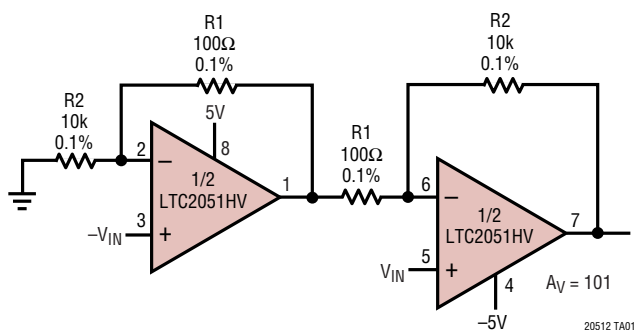
The LTC2051/LTC2052, despite their miniature size, feature uncompromising DC performance. The typical input offset voltage and offset drift are  $0.5\mu\text{V}$  and  $10\text{nV}/^\circ\text{C}$ . The almost zero DC offset and drift are supported with a power supply rejection ratio (PSRR) and common mode rejection ratio (CMRR) of more than 130dB.

The input common mode voltage ranges from the negative supply up to typically 1V from the positive supply. The LTC2051/LTC2052 also have an enhanced output stage capable of driving loads as low as  $2\text{k}\Omega$  to both supply rails. The open-loop gain is typically 140dB. The LTC2051/LTC2052 also feature a  $1.5\mu\text{V}_{\text{P-P}}$  DC to 10Hz noise and a 3MHz gain-bandwidth product.

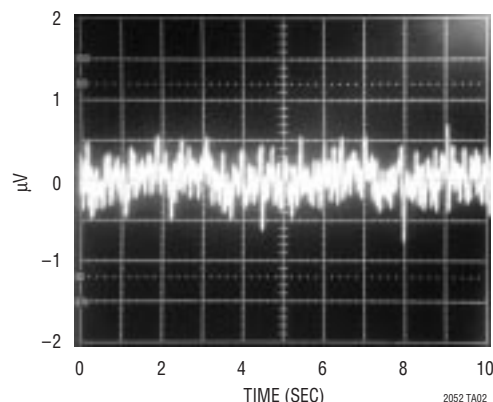
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## TYPICAL APPLICATION

High Performance Low Cost Instrumentation Amplifier



Input Referred Noise 0.1Hz to 10Hz



# LTC2051/LTC2052

## ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage ( $V^+$ to $V^-$ )	Operating Temperature Range .....	$-40^{\circ}\text{C}$ to $125^{\circ}\text{C}$
LTC2051/LTC2052 ..... 7V	Specified Temperature Range	
LTC2051HV/LTC2052HV ..... 12V	(Note 3) .....	$-40^{\circ}\text{C}$ to $125^{\circ}\text{C}$
Input Voltage (Note 5) ..... ( $V^+ + 0.3\text{V}$ ) to ( $V^- - 0.3\text{V}$ )	Storage Temperature Range .....	$-65^{\circ}\text{C}$ to $150^{\circ}\text{C}$
Output Short-Circuit Duration ..... Indefinite	Lead Temperature (Soldering, 10 sec) .....	$300^{\circ}\text{C}$

## PACKAGE/ORDER INFORMATION

<p>TOP VIEW</p> <p>MS8 PACKAGE 8-LEAD PLASTIC MSOP <math>T_{JMAX} = 125^{\circ}\text{C}</math>, <math>\theta_{JA} = 250^{\circ}\text{C/W}</math></p>		<p>TOP VIEW</p> <p>MS10 PACKAGE 10-LEAD PLASTIC MSOP <math>T_{JMAX} = 125^{\circ}\text{C}</math>, <math>\theta_{JA} = 250^{\circ}\text{C/W}</math></p>		<p>TOP VIEW</p> <p>S8 PACKAGE 8-LEAD PLASTIC SO <math>T_{JMAX} = 125^{\circ}\text{C}</math>, <math>\theta_{JA} = 190^{\circ}\text{C/W}</math></p>	
ORDER PART NUMBER	MS8 PART MARKING	ORDER PART NUMBER	MS10 PART MARKING	ORDER PART NUMBER	S8 PART MARKING
LTC2051CMS8	LTMN	LTC2051CMS10	LTMQ	LTC2051CS8	2051
LTC2051IMS8	LTMP	LTC2051IMS10	LTMR	LTC2051IS8	2051I
LTC2051HVCMS8	LTPJ	LTC2051HVCMS10	LTRB	LTC2051HVCS8	2051HV
LTC2051HVIMS8	LTPK	LTC2051HVIMS10	LTRC	LTC2051HVIS8	051HVI
LTC2051HMS8	LTVF			LTC2051HS8	2051H
LTC2051HVHMS8	LTVH			LTC2051HVHS8	051HVH
<p>TOP VIEW</p> <p>GN PACKAGE 16-LEAD PLASTIC SSOP <math>T_{JMAX} = 125^{\circ}\text{C}</math>, <math>\theta_{JA} = 110^{\circ}\text{C/W}</math></p>		ORDER PART NUMBER	<p>TOP VIEW</p> <p>S PACKAGE 14-LEAD PLASTIC SO <math>T_{JMAX} = 125^{\circ}\text{C}</math>, <math>\theta_{JA} = 110^{\circ}\text{C/W}</math></p>		ORDER PART NUMBER
		LTC2052CGN			LTC2052CS
		LTC2052IGN			LTC2052IS
		LTC2052HVCGN			LTC2052HVCS
		LTC2052HVGIGN			LTC2052HVIS
		LTC2052HGN			LTC2052HS
		LTC2052HVHGN			LTC2052HVHS
		GN PART MARKING			
		2052			
		2052I			
		2052HV			
		052HVI			
		2052H			
		052HVH			

Consult LTC Marketing for parts specified with wider operating temperature ranges.

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## AVAILABLE OPTIONS

PART NUMBER	AMPS/PACKAGE	SPECIFIED TEMP RANGE	SPECIFIED VOLTAGE	PACKAGE
LTC2051CS8	2	0°C to 70°C	3V, 5V	SO-8
LT2051CMS8	2	0°C to 70°C	3V, 5V	8-Lead MSOP
LT2051CMS10	2	0°C to 70°C	3V, 5V	10-Lead MSOP
LT2051HVCS8	2	0°C to 70°C	3V, 5V, ±5V	SO-8
LTC2051HVCMS8	2	0°C to 70°C	3V, 5V, ±5V	8-Lead MSOP
LTC2051HVCMS10	2	0°C to 70°C	3V, 5V, ±5V	10-Lead MSOP
LTC2051IS8	2	–40°C to 85°C	3V, 5V	SO-8
LT2051IMS8	2	–40°C to 85°C	3V, 5V	8-Lead MSOP
LT2051IMS10	2	–40°C to 85°C	3V, 5V	10-Lead MSOP
LT2051HVIS8	2	–40°C to 85°C	3V, 5V, ±5V	SO-8
LTC2051HVIMS8	2	–40°C to 85°C	3V, 5V, ±5V	8-Lead MSOP
LTC2051HVIMS10	2	–40°C to 85°C	3V, 5V, ±5V	10-Lead MSOP
LTC2051HS8	2	–40°C to 125°C	3V, 5V	SO-8
LT2051HMS8	2	–40°C to 125°C	3V, 5V	8-Lead MSOP
LT2051HVHS8	2	–40°C to 125°C	3V, 5V, ±5V	SO-8
LT2051HVHMS8	2	–40°C to 125°C	3V, 5V, ±5V	8-Lead MSOP
LTC2052CS	4	0°C to 70°C	3V, 5V	14-Lead SO
LTC2052CGN	4	0°C to 70°C	3V, 5V	16-Lead SSOP
LTC2052HVCS	4	0°C to 70°C	3V, 5V, ±5V	14-Lead SO
LTC2052HVCGN	4	0°C to 70°C	3V, 5V, ±5V	16-Lead SSOP
LTC2052IS	4	–40°C to 85°C	3V, 5V	14-Lead SO
LTC2052IGN	4	–40°C to 85°C	3V, 5V	16-Lead SSOP
LTC2052HVIS	4	–40°C to 85°C	3V, 5V, ±5V	14-Lead SO
LTC2052HVIGN	4	–40°C to 85°C	3V, 5V, ±5V	16-Lead SSOP
LTC2052HS	4	–40°C to 125°C	3V, 5V	14-Lead SO
LTC2052HGN	4	–40°C to 125°C	3V, 5V	16-Lead SSOP
LTC2052HVHS	4	–40°C to 125°C	3V, 5V, ±5V	14-Lead SO
LTC2052HVHGN	4	–40°C to 125°C	3V, 5V, ±5V	16-Lead SSOP

# LTC2051/LTC2052

**ELECTRICAL CHARACTERISTICS** (LTC2051/LTC2052, LTC2051HV/LTC2052HV) The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ .  $V_S = 3\text{V}$ ,  $5\text{V}$  unless otherwise noted. (Note 3)

PARAMETER	CONDITIONS		LTC2051C/LTC2052C LTC2051I/LTC2052I			LTC2051H/LTC2052H			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	(Note 2)			$\pm 0.5$	$\pm 3$		$\pm 0.5$	$\pm 3$	$\mu\text{V}$
Average Input Offset Drift	(Note 2)	●		0.01	$\pm 0.03$		0.01	$\pm 0.05$	$\mu\text{V}/^\circ\text{C}$
Long-Term Offset Drift				50			50		$\text{nV}/\sqrt{\text{mo}}$
Input Bias Current (Note 4)	$V_S = 3\text{V}$			$\pm 8$	$\pm 50$		$\pm 8$	$\pm 50$	$\text{pA}$
	$V_S = 3\text{V}$	●			$\pm 100$			$\pm 3000$	$\text{pA}$
	$V_S = 5\text{V}$			$\pm 25$	$\pm 75$		$\pm 25$	$\pm 75$	$\text{pA}$
	$V_S = 5\text{V}$	●			$\pm 150$			$\pm 3000$	$\text{pA}$
Input Offset Current (Note 4)	$V_S = 3\text{V}$				$\pm 100$			$\pm 100$	$\text{pA}$
	$V_S = 3\text{V}$	●			$\pm 150$			$\pm 700$	$\text{pA}$
	$V_S = 5\text{V}$				$\pm 150$			$\pm 150$	$\text{pA}$
	$V_S = 5\text{V}$	●			$\pm 200$			$\pm 700$	$\text{pA}$
Input Noise Voltage	$R_S = 100\Omega$ , DC to 10Hz			1.5			1.5		$\mu\text{V}_{\text{P-P}}$
Common Mode Rejection Ratio	$V_{\text{CM}} = \text{GND to } V^+ - 1.3$ , $V_S = 3\text{V}$	●	115	130		115	130		$\text{dB}$
			110	130		110	130		$\text{dB}$
	$V_{\text{CM}} = \text{GND to } V^+ - 1.3$ , $V_S = 5\text{V}$	●	120	130		120	130		$\text{dB}$
			115	130		115	130		$\text{dB}$
Power Supply Rejection Ratio		●	120	130		120	130		$\text{dB}$
			115	130		115	130		$\text{dB}$
Large-Signal Voltage Gain	$R_L = 10\text{k}$ , $V_S = 3\text{V}$		120	140		120	140		$\text{dB}$
		●	115	140		115	140		$\text{dB}$
	$R_L = 10\text{k}$ , $V_S = 5\text{V}$		125	140		125	140		$\text{dB}$
		●	120	140		120	140		$\text{dB}$
Output Voltage Swing High	$R_L = 2\text{k to GND}$	●	$V^+ - 0.15$	$V^+ - 0.06$		$V^+ - 0.15$	$V^+ - 0.06$		$\text{V}$
	$R_L = 10\text{k to GND}$	●	$V^+ - 0.05$	$V^+ - 0.02$		$V^+ - 0.05$	$V^+ - 0.02$		$\text{V}$
Output Voltage Swing Low	$R_L = 2\text{k to GND}$	●		2	15		2	15	$\text{mV}$
	$R_L = 10\text{k to GND}$	●		2	15		2	15	$\text{mV}$
Slew Rate				2			2		$\text{V}/\mu\text{s}$
Gain Bandwidth Product				3			3		$\text{MHz}$
Supply Current (Per Amplifier)	No Load, $V_S = 3\text{V}$ , $V_{\text{SHDN}} = V_{\text{IH}}$	●		0.75	1.0		0.75	1.1	$\text{mA}$
	No Load, $V_S = 5\text{V}$ , $V_{\text{SHDN}} = V_{\text{IH}}$	●		0.85	1.2		0.85	1.3	$\text{mA}$
Supply Current, Shutdown	$V_{\text{SHDN}} = V_{\text{IL}}$ , $V_S = 3\text{V}$	●		2	5		2	5	$\mu\text{A}$
	$V_{\text{SHDN}} = V_{\text{IL}}$ , $V_S = 5\text{V}$	●		4	10		4	10	$\mu\text{A}$
Shutdown Pin Input Low Voltage ( $V_{\text{IL}}$ )		●			$V^- + 0.5$			$V^- + 0.5$	$\text{V}$
Shutdown Pin Input High Voltage ( $V_{\text{IH}}$ )		●	$V^+ - 0.5$			$V^+ - 0.5$			$\text{V}$
Shutdown Pin Input Current	$V_{\text{SHDN}} = V_{\text{IL}}$ , $V_S = 3\text{V}$	●		-1	-3		-1	-3	$\mu\text{A}$
	$V_{\text{SHDN}} = V_{\text{IL}}$ , $V_S = 5\text{V}$	●		-2	-5		-2	-5	$\mu\text{A}$
Internal Sampling Frequency				7.5			7.5		$\text{kHz}$

# ELECTRICAL CHARACTERISTICS (LTC2051HV/LTC2052HV) The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$ . $V_S = \pm 5\text{V}$ unless otherwise noted. (Note 3)

PARAMETER	CONDITIONS		LTC2051C/LTC2052C LTC2051I/LTC2052I			LTC2051H/LTC2052H			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	(Note 2)			$\pm 1$	$\pm 3$		$\pm 1$	$\pm 3$	$\mu\text{V}$
Average Input Offset Drift	(Note 2)	●		0.01	$\pm 0.03$		0.01	$\pm 0.05$	$\mu\text{V}/^\circ\text{C}$
Long-Term Offset Drift				50			50		$\text{nV}/\sqrt{\text{mo}}$
Input Bias Current (Note 4)		●		$\pm 90$	$\pm 150$ $\pm 300$		$\pm 90$	$\pm 150$ $\pm 3000$	$\text{pA}$ $\text{pA}$
Input Offset Current (Note 4)		●			$\pm 300$ $\pm 500$			$\pm 300$ $\pm 700$	$\text{pA}$ $\text{pA}$
Input Noise Voltage	$R_S = 100\Omega$ , DC to 10Hz			1.5			1.5		$\mu\text{V}_{\text{P-P}}$
Common Mode Rejection Ratio	$V_{\text{CM}} = V^-$ to $V^+ - 1.3$	●	125 120	130 130		125 120	130 130		$\text{dB}$ $\text{dB}$
Power Supply Rejection Ratio		●	120 115	130 130		120 115	130 130		$\text{dB}$ $\text{dB}$
Large-Signal Voltage Gain	$R_L = 10\text{k}$	●	125 120	140 140		125 120	140 140		$\text{dB}$ $\text{dB}$
Maximum Output Voltage Swing	$R_L = 2\text{k}$ to GND $R_L = 10\text{k}$ to GND	● ●	$\pm 4.75$ $\pm 4.90$	$\pm 4.92$ $\pm 4.98$		$\pm 4.50$ $\pm 4.85$	$\pm 4.92$ $\pm 4.98$		$\text{V}$ $\text{V}$
Slew Rate				2			2		$\text{V}/\mu\text{s}$
Gain Bandwidth Product				3			3		$\text{MHz}$
Supply Current (Per Amplifier)	No Load, $V_{\text{SHDN}} = V_{\text{IH}}$	●		1	1.5		1	1.5	$\text{mA}$
Supply Current, Shutdown	$V_{\text{SHDN}} = V_{\text{IL}}$	●		15	30		15	30	$\mu\text{A}$
Shutdown Pin Input Low Voltage ( $V_{\text{IL}}$ )		●			$V^- + 0.5$			$V^- + 0.5$	$\text{V}$
Shutdown Pin Input High Voltage ( $V_{\text{IH}}$ )		●	$V^+ - 0.5$			$V^+ - 0.5$			$\text{V}$
Shutdown Pin Input Current	$V_{\text{SHDN}} = V_{\text{IL}}$	●		-7	-15		-7	-15	$\mu\text{A}$
Internal Sampling Frequency				7.5			7.5		$\text{kHz}$

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

**Note 2:** These parameters are guaranteed by design. Thermocouple effects preclude measurements of these voltage levels during automated testing.

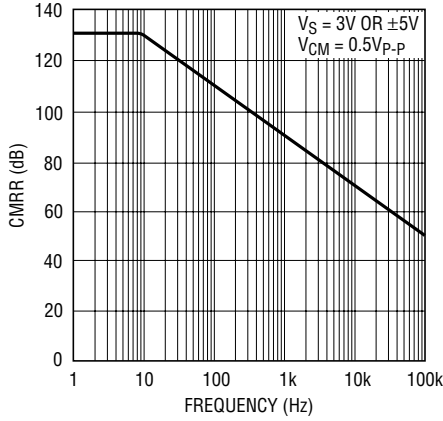
**Note 3:** All versions of the LTC2051/LTC2052 are designed, characterized and expected to meet the extended temperature limits of  $-40^\circ\text{C}$  and  $125^\circ\text{C}$ . The LTC2051C/LTC2052C/LTC2051HVC/LTC2052HVC are guaranteed to meet the temperature limits of  $0^\circ\text{C}$  and  $70^\circ\text{C}$ . The LTC2051I/LTC2052I/LTC2051HVI/LTC2052HVI are guaranteed to meet temperature limits of  $-40^\circ\text{C}$  and  $85^\circ\text{C}$ . The LTC2051H/LTC2051HVH and LTC2052H/LTC2052HVH are guaranteed to meet the temperature limits of  $-40^\circ\text{C}$  and  $125^\circ\text{C}$ .

**Note 4:** The bias current measurement accuracy depends on the proximity of the negative supply bypass capacitors to the device under test. Because of this, only the bias current of channel B (LTC2051) and channels A and B (LTC2052) are 100% tested to the data sheet specifications. The bias currents of the remaining channels are 100% tested to relaxed limits, however, their values are guaranteed by design to meet the data sheet limits.

**Note 5:** This parameter is guaranteed to meet specified performance through design and characterization. It has not been tested.

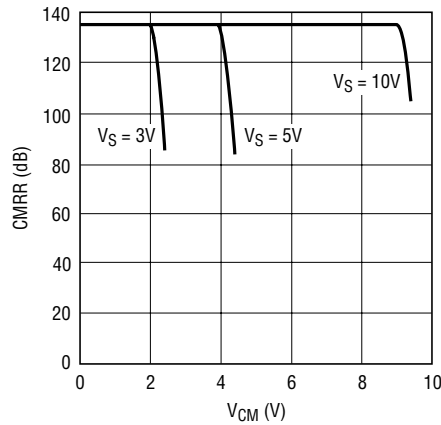
## TYPICAL PERFORMANCE CHARACTERISTICS

**Common Mode Rejection Ratio vs Frequency**



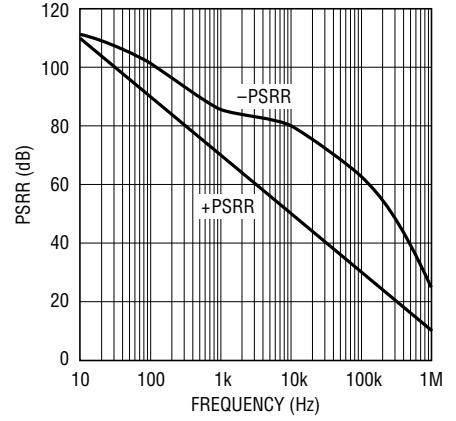
20512 G01

**DC CMRR vs Common Mode Input Range**



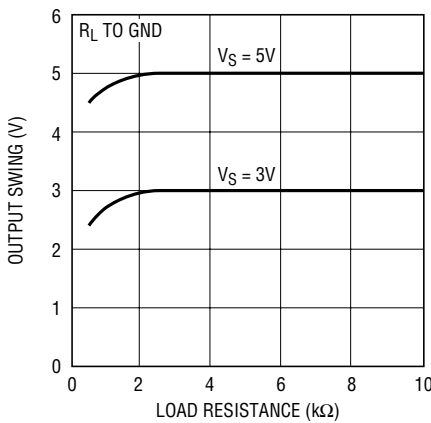
20512 G02

**PSRR vs Frequency**



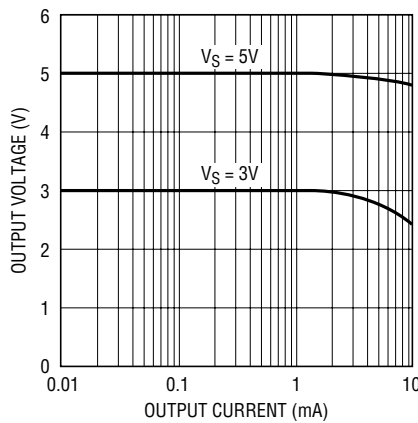
20512 G03

**Output Voltage Swing vs Load Resistance**



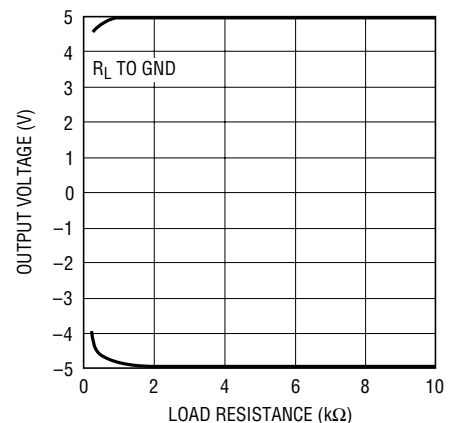
20512 G04

**Output Swing vs Output Current**



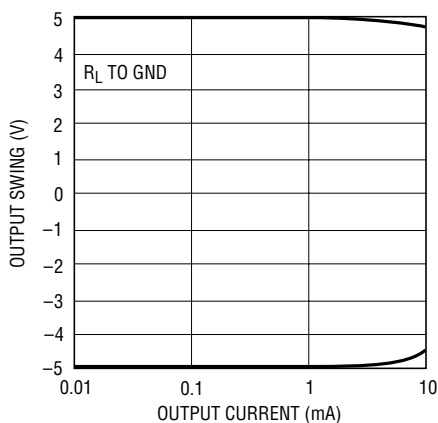
20512 G05

**Output Swing vs Load Resistance  $\pm 5V$**



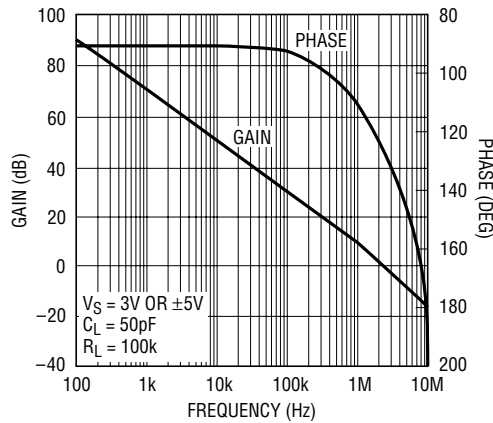
20512 G06

**Output Swing vs Output Current,  $\pm 5V$  Supply**



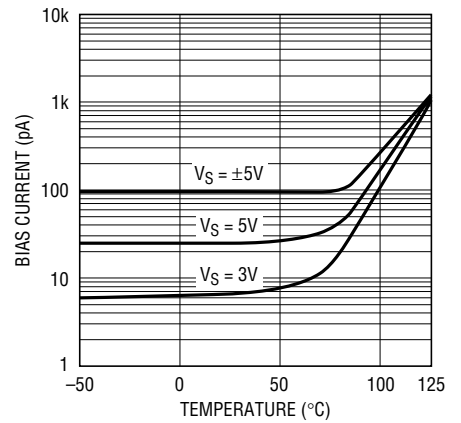
20512 G07

**Gain/Phase vs Frequency**



20512 G08

**Bias Current vs Temperature**

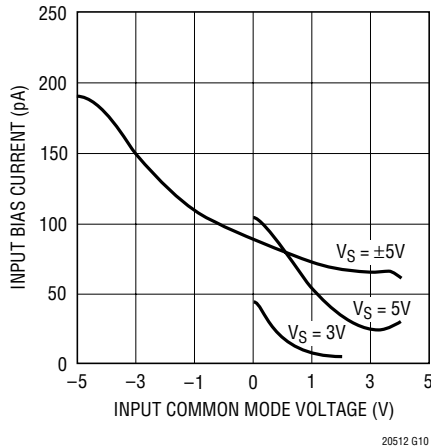


20512 G09

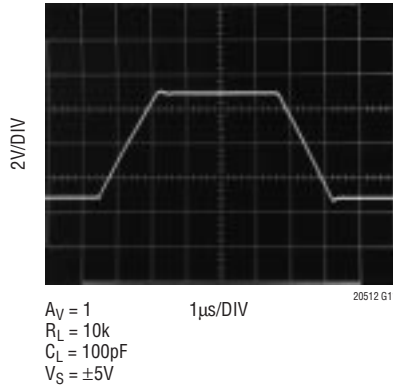
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# TYPICAL PERFORMANCE CHARACTERISTICS

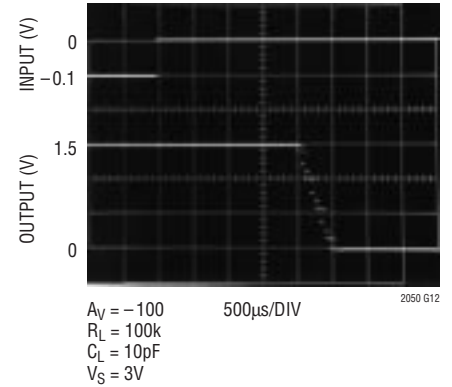
**Input Bias Current  
vs Input Common Mode Voltage**



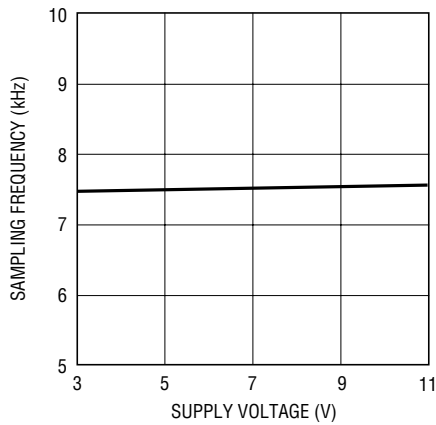
**Transient Response**



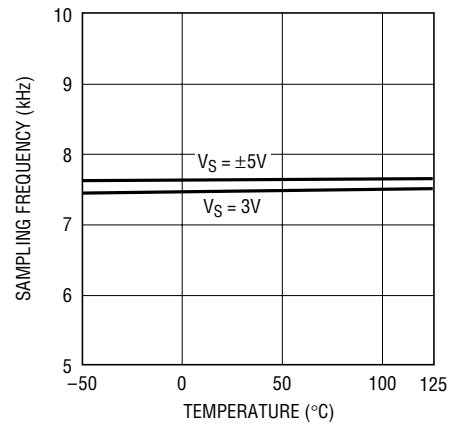
**Input Overload Recovery**



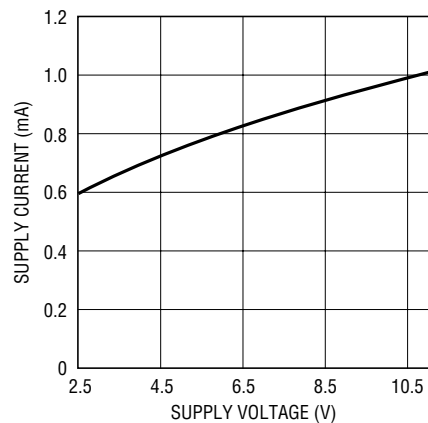
**Sampling Frequency  
vs Supply Voltage**



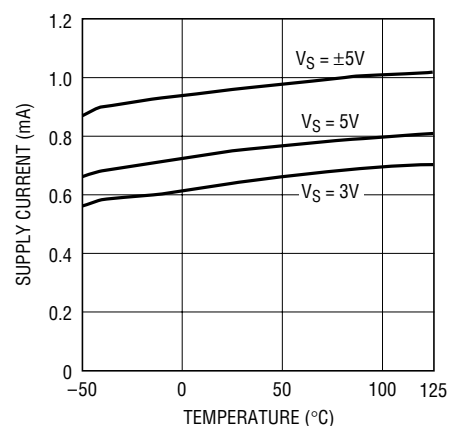
**Sampling Frequency  
vs Temperature**



**Supply Current (Per Amplifier)  
vs Supply Voltage**



**Supply Current (Per Amplifier)  
vs Temperature**



## APPLICATIONS INFORMATION

### Shutdown

The LTC2051 includes a shutdown pin in the 10-lead MSOP. When this active low pin is high or allowed to float, the device operates normally. When the shutdown pin is pulled low, the device enters shutdown mode; supply current drops to  $3\mu\text{A}$ , all clocking stops and the output assumes a high impedance state.

### Clock Feedthrough, Input Bias Current

The LTC2051/LTC2052 use autozeroing circuitry to achieve an almost zero DC offset over temperature, common mode voltage and power supply voltage. The frequency of the clock used for autozeroing is typically 7.5kHz. The term clock feedthrough is broadly used to indicate visibility of this clock frequency in the op amp output spectrum. There are typically two types of clock feedthrough in autozeroed op amps like the LTC2051/LTC2052.

The first form of clock feedthrough is caused by the settling of the internal sampling capacitor and is input referred; that is, it is multiplied by the closed-loop gain of the op amp. This form of clock feedthrough is independent of the magnitude of the input source resistance or the magnitude of the gain setting resistors. The LTC2051/LTC2052 have a residue clock feedthrough of less than  $1\mu\text{V}_{\text{RMS}}$  input referred at 7.5kHz.

The second form of clock feedthrough is caused by the small amount of charge injection occurring during the sampling and holding of the op amps input offset voltage. The current spikes are multiplied by the impedance seen at the input terminals of the op amp, appearing at the output multiplied by the closed-loop gain of the op amp.

To reduce this form of clock feedthrough, use smaller valued gain setting resistors and minimize the source resistance at the input. If the resistance seen at the inputs is less than 10k, this form of clock feedthrough is less than  $1\mu\text{V}_{\text{RMS}}$  input referred at 7.5kHz, or less than the amount of residue clock feedthrough from the first form previously described.

Placing a capacitor across the feedback resistor reduces either form of clock feedthrough by limiting the bandwidth of the closed-loop gain.

Input bias current is defined as the DC current into the input pins of the op amp. The same current spikes that cause the second form of clock feedthrough previously described, when averaged, dominate the DC input bias current of the op amp below 70°C.

At temperatures above 70°C, the leakage of the ESD protection diodes on the inputs increase the input bias currents of both inputs in the positive direction, while the current caused by the charge injection stays relatively constant. At elevated temperatures (above 85°C) the leakage current begins to dominate and both the negative and positive pin's input bias currents are in the positive direction (into the pins).

### Input Pins, ESD Sensitivity

ESD voltages above 700V on the input pins of the op amp will cause the input bias currents to increase (more DC current into the pins). At these voltages, it is possible to damage the device to a point where the input bias current exceeds the maximums specified in this data sheet.



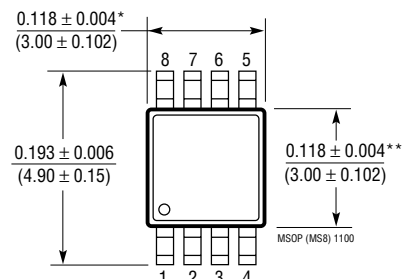
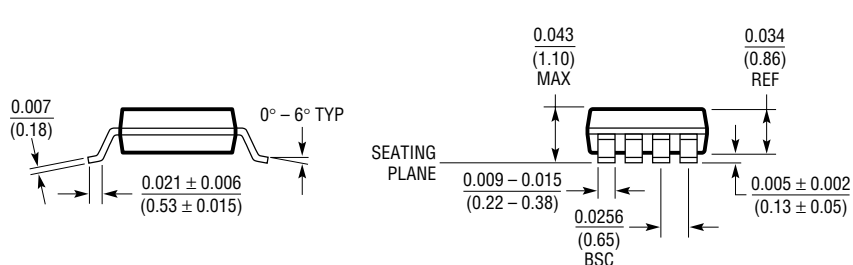
needed and the input source resistance is low. (For instance a 350Ω strain gauge bridge.) The LT1012 or equivalent should be used when low bias current (100pA) is also required in conjunction with DC to 10Hz low noise, low  $V_{OS}$  and  $V_{OS}$  drift. The measured typical input offset voltages are less than 1μV.

The circuit diagram shows a precision rectifier. It consists of two LTC2051 comparators and an operational amplifier A1. The first comparator (left) has its non-inverting input (+) connected to the input signal and its inverting input (-) connected to the output of the second comparator. Its output (pin 1) is connected to the non-inverting input of A1. The second comparator (right) has its non-inverting input (+) connected to the output of the first comparator and its inverting input (-) connected to the output of A1. Its output (pin 7) is connected to the output of the first comparator. The output of A1 (pin 6) is labeled 'OUT'. The circuit is powered by a 5V supply and includes resistors R1, R2, R3, R4, R5 and capacitors C1, C2. The input signal is connected to the non-inverting input of the first comparator and the inverting input of A1. The output of the circuit is labeled 'OUT'.

A1	R1	R2	R3	R4	R5	C1	C2	$\bar{e}_{IN}$ (DC – 1Hz)	$\bar{e}_{IN}$ (DC – 10Hz)
LT1677	2.49k	3.01k	340k	10k	100k	0.01 $\mu$ F	0.001 $\mu$ F	0.15 $\mu$ V <sub>P-P</sub>	0.2 $\mu$ V <sub>P-P</sub>
LT1012	750 $\Omega$	57 $\Omega$	250k	10k	100k	0.01 $\mu$ F	0.001 $\mu$ F	0.3 $\mu$ V <sub>P-P</sub>	0.4 $\mu$ V <sub>P-P</sub>

## PACKAGE DESCRIPTION

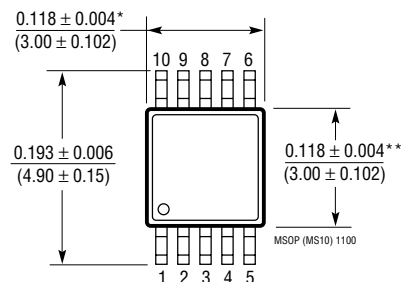
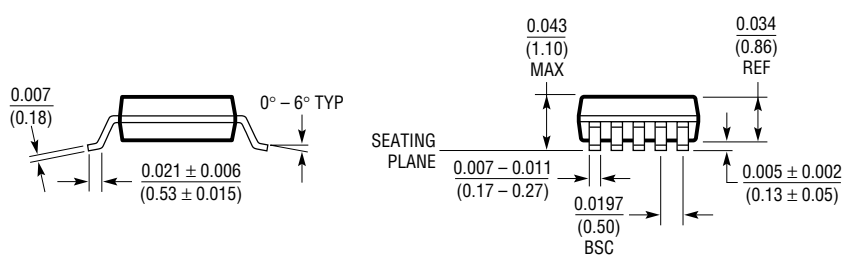
### MS8 Package 8-Lead Plastic MSOP (Reference LTC DWG # 05-08-1660)



\* DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED  $0.006^*$  (0.152mm) PER SIDE

\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS. INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED  $0.006^*$  (0.152mm) PER SIDE

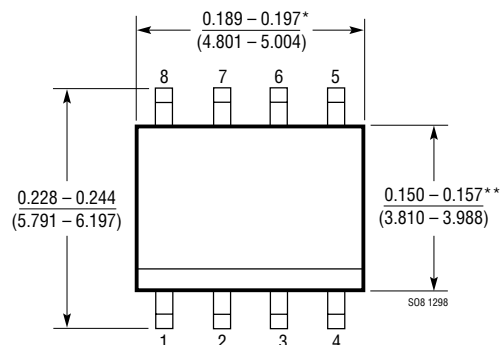
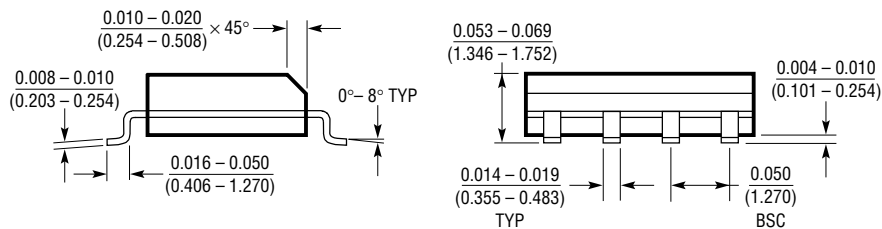
### MS10 Package 10-Lead Plastic MSOP (Reference LTC DWG # 05-08-1661)



\* DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED  $0.006^*$  (0.152mm) PER SIDE

\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS. INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED  $0.006^*$  (0.152mm) PER SIDE

### S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)

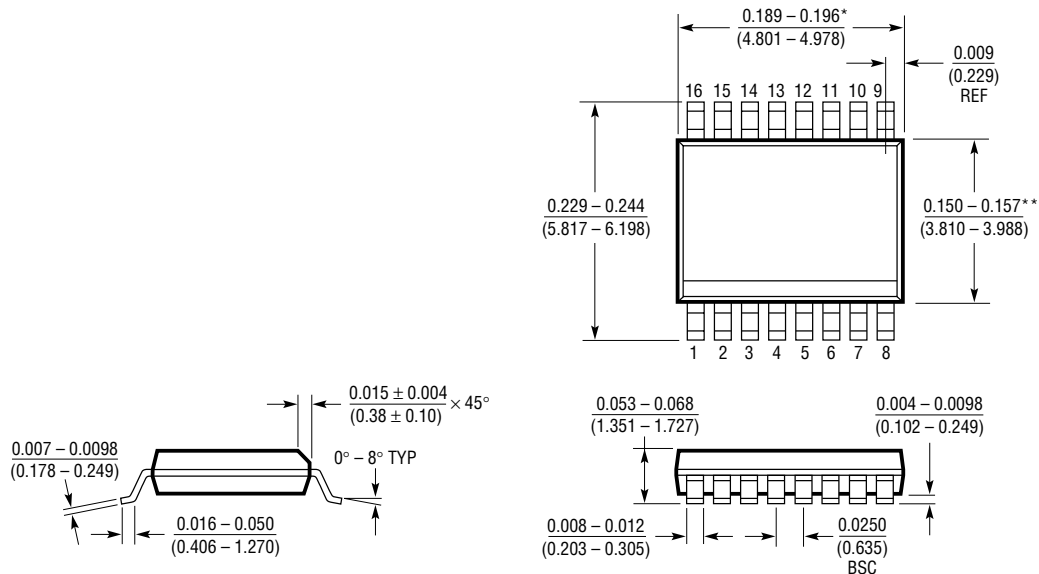


\* DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED  $0.006^*$  (0.152mm) PER SIDE

\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED  $0.010^*$  (0.254mm) PER SIDE

## PACKAGE DESCRIPTION

### GN Package 16-Lead Plastic SSOP (Narrow .150 Inch) (Reference LTC DWG # 05-08-1641)

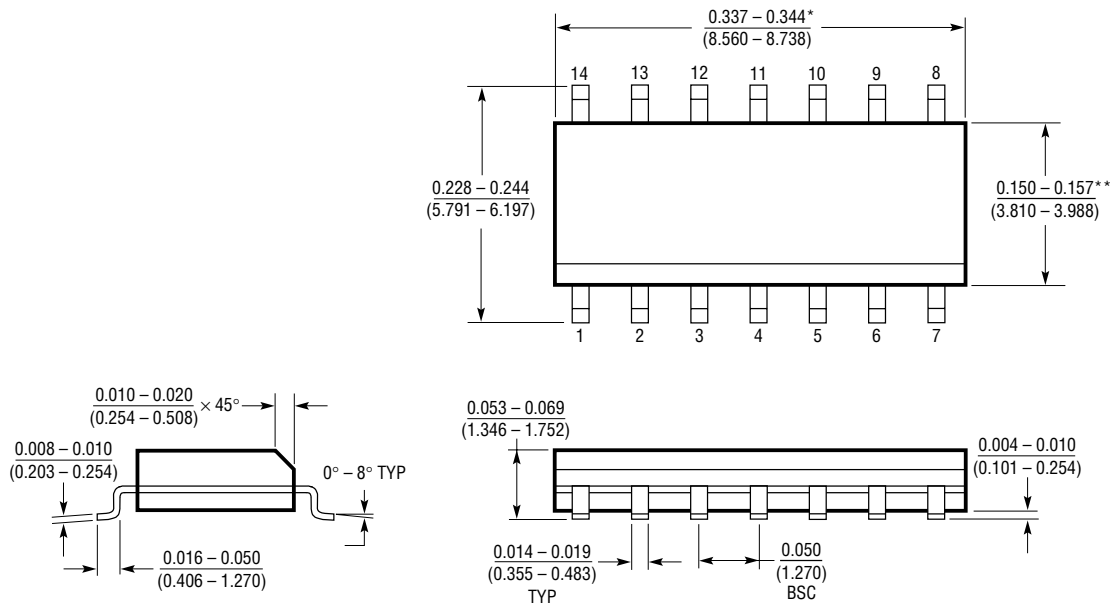


\* DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

GN16 (SSOP) 1098

### S Package 14-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



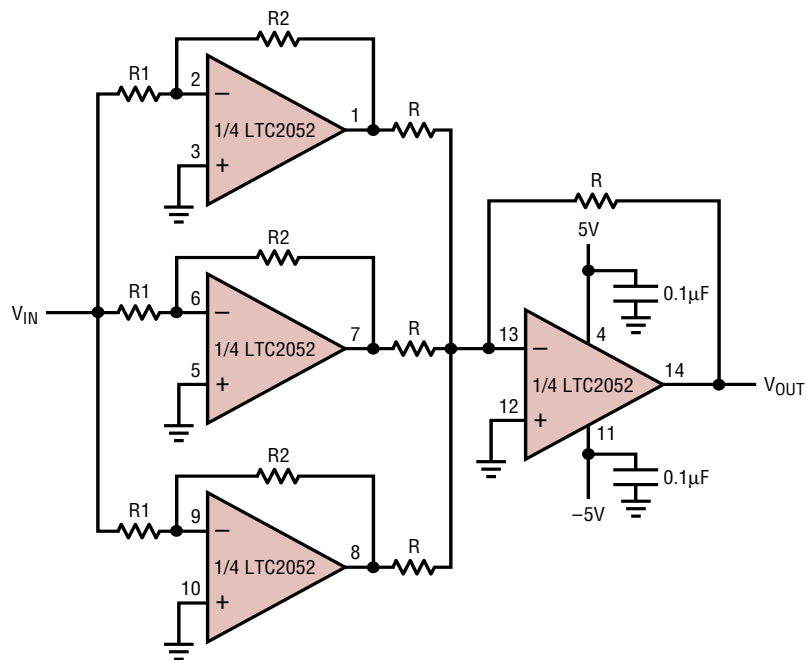
\* DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

S14 1298

TYPICAL APPLICATION

Paralleling Amplifiers to Improve Noise



$$\frac{V_{OUT}}{V_{IN}} = 3 \frac{R_2}{R_1} ; \text{ INPUT DC - 10Hz NOISE } \cong 0.8\mu\text{V}_{P-P} = \frac{\text{NOISE OF EACH PARALLEL OP AMP}}{\sqrt{3}}$$

20512 F02

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC1051/LTC1053	Precision Zero-Drift Op Amp	Dual/Quad
LTC1151	±15V Zero-Drift Op Amp	Dual High Voltage Operation ±18V
LTC1152	Rail-to-Rail Input and Output Zero-Drift Op Amp	Single Zero-Drift Op Amp with Rail-to-Rail Input and Output and Shutdown
LTC2050	Zero-Drift Op Amp in SOT-23	Single Supply Operation 2.7V to ±5V, Shutdown