

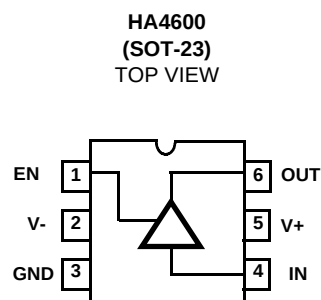
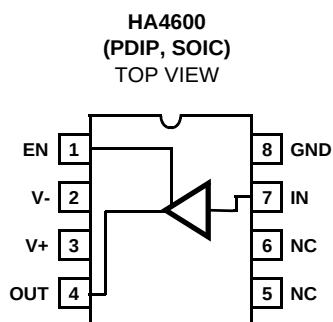
480MHz, SOT-23, Video Buffer with Output Disable

The HA4600 is a very wide bandwidth, unity gain buffer ideal for professional video switching, HDTV, computer monitor routing, and other high performance applications. The circuit features very low power dissipation (105mW Enabled, 1mW Disabled), excellent differential gain and phase, and very high off isolation. When disabled, the output is switched to a high impedance state, making the HA4600 ideal for routing matrix equipment and video multiplexers.

The HA4600 also features fast switching and symmetric slew rates. A typical application for the HA4600 is interfacing Intersil's wide range of video crosspoint switches.

For applications requiring a tally output (enable indicator), please refer to the HA4201 data sheet.

Pinouts



Truth Table

EN	OUT
0	High Z
1	Active

Features

- Micro Package Available SOT-23
- Low Power Dissipation 105mW
- Symmetrical Slew Rates 1700V/μs
- 0.1dB Gain Flatness 250MHz
- Off Isolation (100MHz) 85dB
- Differential Gain and Phase 0.01%/0.01 Degrees
- High ESD Rating >1800V
- TTL Compatible Enable Input
- Improved Replacement for GB4600
- Pb-Free Plus Anneal Available (RoHS Compliant)

Applications

- Professional Video Switching and Routing
- Video Multiplexers
- HDTV
- Computer Graphics
- RF Switching and Routing
- PCM Data Routing

Ordering Information

PART NUMBER	PART MARKING	TEMP. RANGE (°C)	PACKAGE	PKG. DWG. #
HA4600CP	HA4600CP	0 to 70	8 Ld PDIP	E8.3
HA4600CPZ (Note)	HA4600CPZ	0 to 70	8 Ld PDIP* (Pb-free)	E8.3
HA4600CB	4600CB	0 to 70	8 Ld SOIC	M8.15
HA4600CB96	4600CB	0 to 70	8 Ld SOIC Tape and Reel	M8.15
HA4600CBZ (Note)	4600CBZ	0 to 70	8 Ld SOIC (Pb-free)	M8.15
HA4600CBZ96 (Note)	4600CBZ	0 to 70	8 Ld SOIC Tape and Reel (Pb-free)	M8.15
HA4600CH96	46	0 to 70	6 Ld SOT-23 Tape and Reel	P6.064
HA4600CHZ96 (Note)	46Z	0 to 70	6 Ld SOT-23 Tape and Reel (Pb-free)	P6.064

*Pb-free PDIPs can be used for through hole wave solder processing only. They are not intended for use in Reflow solder processing applications.

NOTE: Intersil Pb-free plus anneal products employ special Pb-free material sets; molding compounds/die attach materials and 100% matte tin plate termination finish, which are RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

Absolute Maximum Ratings

Voltage Between V+ and V- 12V
 Input Voltage V_{SUPPLY}
 Digital Input Current (Note 2) $\pm 25\text{mA}$
 Output Current 20mA
 ESD Rating
 Human Body Model (Per MIL-STD-883 Method 3015.7) 1800V

Operating Conditions

Temperature Range 0°C to 70°C

Thermal Information

Thermal Resistance (Typical, Note 1) θ_{JA} (°C/W)
 PDIP Package* 130
 SOIC Package 170
 SOT-23 Package 210
 Maximum Junction Temperature (Die) 175°C
 Maximum Junction Temperature (Plastic Package) 150°C
 Maximum Storage Temperature Range -65°C to 150°C
 Maximum Lead Temperature (Soldering 10s) 300°C
 (SOIC and SOT-23 - Lead Tips Only)

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CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTES:

1. θ_{JA} is measured with the component mounted on an evaluation PC board in free air.
2. If an input signal is applied before the supplies are powered up, the input current must be limited to this maximum value.

Electrical Specifications $V_{SUPPLY} = \pm 5\text{V}$, $R_L = 10\text{k}\Omega$, $V_{EN} = 2.0\text{V}$, Unless Otherwise Specified

PARAMETER	TEST CONDITIONS	TEMP. (°C)	MIN	TYP	MAX	UNITS
DC SUPPLY CHARACTERISTICS						
Supply Voltage		Full	± 4.5	± 5.0	± 5.5	V
Supply Current ($V_{OUT} = 0\text{V}$)	$V_{EN} = 2\text{V}$	25, 70	-	10.5	13	mA
	$V_{EN} = 2\text{V}$	0	-	-	14.5	mA
	$V_{EN} = 0.8\text{V}$	25, 70	-	100	115	μA
	$V_{EN} = 0.8\text{V}$	0	-	100	125	μA
ANALOG DC CHARACTERISTICS						
Output Voltage Swing without Clipping	$V_{OUT} = V_{IN} \pm V_{IO} \pm 20\text{mV}$	25, 70	± 2.7	± 2.8	-	V
		0	± 2.4	± 2.5	-	V
Output Current		Full	15	20	-	mA
Input Bias Current		Full	-	30	50	μA
Output Offset Voltage		25	-10	-	10	mV
Output Offset Voltage Drift (Note 3)		Full	-	25	50	$\mu\text{V}/^\circ\text{C}$
SWITCHING CHARACTERISTICS						
Turn-On Time		25	-	160	-	ns
Turn-Off Time		25	-	320	-	ns
DIGITAL DC CHARACTERISTICS						
Input Logic High Voltage		Full	2	-	-	V
Input Logic Low Voltage		Full	-	-	0.8	V
EN Input Current	0V to 4V	Full	-2	-	2	μA
AC CHARACTERISTICS						
Insertion Loss	1V _{P-P}	Full	-	0.04	0.05	dB

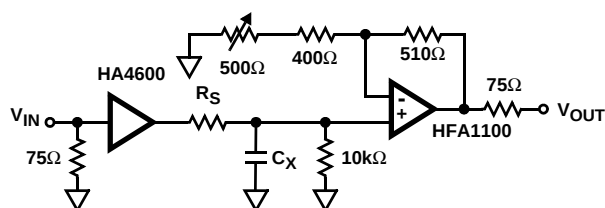
Electrical Specifications $V_{SUPPLY} = \pm 5V$, $R_L = 10k\Omega$, $V_{EN} = 2.0V$, Unless Otherwise Specified **(Continued)**

PARAMETER	TEST CONDITIONS	TEMP. (°C)	MIN	TYP	MAX	UNITS
-3dB Bandwidth	$R_S = 82\Omega$, $C_L = 10pF$	25	-	480	-	MHz
	$R_S = 43\Omega$, $C_L = 15pF$	25	-	380	-	MHz
	$R_S = 36\Omega$, $C_L = 21pF$	25	-	370	-	MHz
$\pm 0.1dB$ Flat Bandwidth	$R_S = 82\Omega$, $C_L = 10pF$	25	-	250	-	MHz
	$R_S = 43\Omega$, $C_L = 15pF$	25	-	175	-	MHz
	$R_S = 36\Omega$, $C_L = 21pF$	25	-	170	-	MHz
Input Resistance		Full	200	400	-	k Ω
Input Capacitance		Full	-	1.0	-	pF
Enabled Output Resistance		Full	-	15	-	Ω
Disabled Output Capacitance	$V_{EN} = 0.8V$	Full	-	2.0	-	pF
Differential Gain (Note 3)	4.43MHz	25	-	0.01	0.02	%
Differential Phase (Note 3)	4.43MHz	25	-	0.01	0.02	Degrees
Off Isolation	1V _{P-P} , 100MHz, $V_{EN} = 0.8V$, $R_L = 10\Omega$	Full	-	85	-	dB
Slew Rate (1.5V _{P-P} , +SR/-SR)	$R_S = 82\Omega$, $C_L = 10pF$	25	-	1750/1770	-	V/ μs
	$R_S = 43\Omega$, $C_L = 15pF$	25	-	1460/1360	-	V/ μs
	$R_S = 36\Omega$, $C_L = 21pF$	25	-	1410/1360	-	V/ μs
Total Harmonic Distortion (Note 3)		Full	-	0.01	0.1	%
Disabled Output Resistance		Full	-	12	-	M Ω

NOTE:

- This parameter is not tested. The limits are guaranteed based on lab characterization, and reflect lot-to-lot variation.

AC Test Circuit



NOTE: $C_L = C_X + \text{Test Fixture Capacitance}$.

Keep input and output traces as short as possible, because trace inductance and capacitance can easily become the performance limiting items.

PC Board Layout

The frequency response of this circuit depends greatly on the care taken in designing the PC board. **The use of low inductance components such as chip resistors and chip capacitors is strongly recommended, while a solid ground plane is a must!**

Attention should be given to decoupling the power supplies. A large value (10 μF) tantalum in parallel with a small value (0.1 μF) chip capacitor works well in most cases.

Application Information

General

The HA4600 is a unity gain buffer that is optimized for high performance video applications. The output disable function makes it ideal for the matrix element in small, high input-to-output isolation switchers and routers. This buffer contains no feedback or gain setting resistors, so the output is a true high impedance load when the IC is disabled ($EN = 0$). The HA4600 also excels as an input buffer for routers with a large number of outputs (i.e. each input must connect to a large number of outputs) and delivers performance superior to most video amplifiers at a fraction of the cost. As an input buffer, the HA4600's low input capacitance and high input resistance provide excellent video terminations when used with an external 75Ω resistor.

Frequency Response

Most applications utilizing the HA4600 require a series output resistor, R_S , to tune the response for the specific load capacitance, C_L , driven. Bandwidth and slew rate degrade as C_L increases (as shown in the Electrical Specification table), so give careful consideration to component placement to minimize trace length. As an example, -3dB bandwidth decreases to 160MHz for $C_L = 100\text{pF}$, $R_S = 0\Omega$. In big matrix configurations where C_L is large, better frequency response is obtained by cascading two levels of crosspoints in the case of multiplexed outputs (see Figure 2), or distributing the load between two drivers if C_L is due to bussing and subsequent stage input capacitance.

Control Signals

EN - The ENABLE input is a TTL/CMOS compatible, active high input. When driven low this input forces the output to a true high impedance state and reduces the power dissipation by two orders of magnitude. The EN input has no on-chip pull-up resistor, so it must be connected to a logic high (recommend V+) if the enable function isn't utilized.

Switcher/Router Applications

Figure 1 illustrates one possible implementation of a wideband, low power, 4×4 switcher/router. A 4×4 switcher/router allows any of the four outputs to be driven by any one of the four inputs (e.g. each of the four inputs may connect to a different output, or an input may connect to multiple outputs). This application utilizes the HA4600 for the input buffer, the HA4404 (4×1 crosspoint switch) as the switch matrix, and the HFA1112 (programmable gain buffer) as the gain of two output driver. Figure 2 details a 16×1 switcher (basically a $16:1$ MUX) which uses the HA4600 in a cascaded stage configuration to minimize capacitive loading at each output node, thus increasing system bandwidth.

Power Up Considerations

No signals should be applied to the analog or digital inputs before the power supplies are activated. Latch-up may occur if the inputs are driven at the time of power up. To prevent

latch-up, the input currents during power up must not exceed the values listed in the Absolute Maximum Ratings.

Intersil's Crosspoint Family

Intersil offers a variety of 1×1 and 4×1 crosspoint switches. In addition to the HA4600, the 1×1 family includes the HA4201 which is an essentially similar device that includes a Tally output (enable indicator). The 4×1 family is comprised of the HA4314, HA4404, and HA4344. The HA4314 is a 14 lead basic 4×1 crosspoint. The HA4404 is a 16 lead device with Tally outputs to indicate the selected channel. The HA4344 is a 16 lead crosspoint with synchronized control lines (A0, A1, CS). With synchronization, the control information for the next channel switch can be loaded into the crosspoint without affecting the current state. On a subsequent clock edge the stored control state effects the desired channel switch.

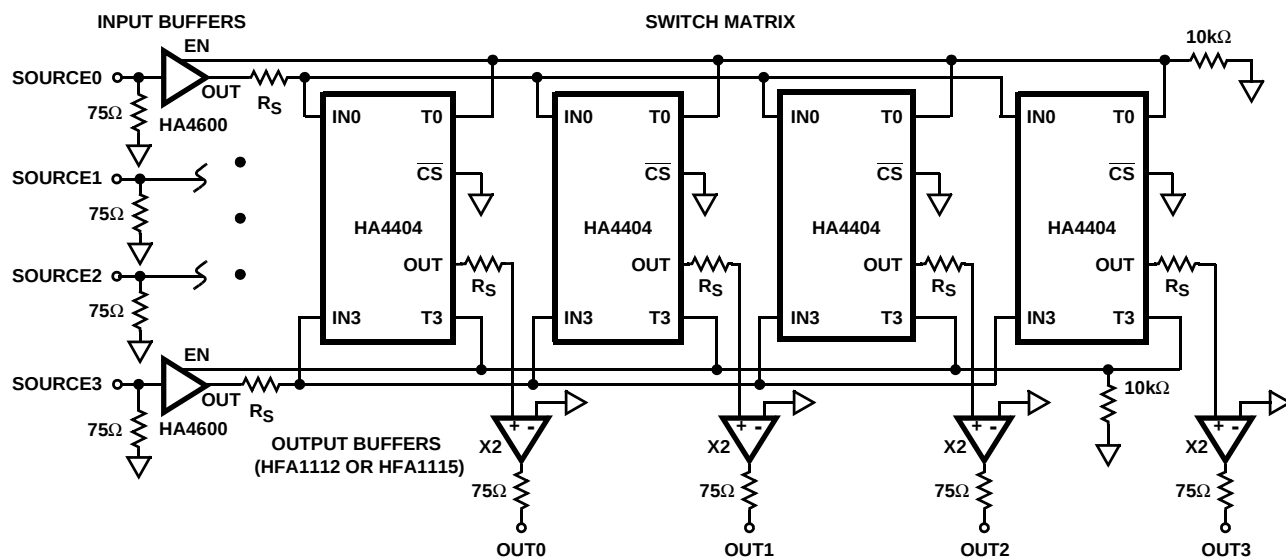
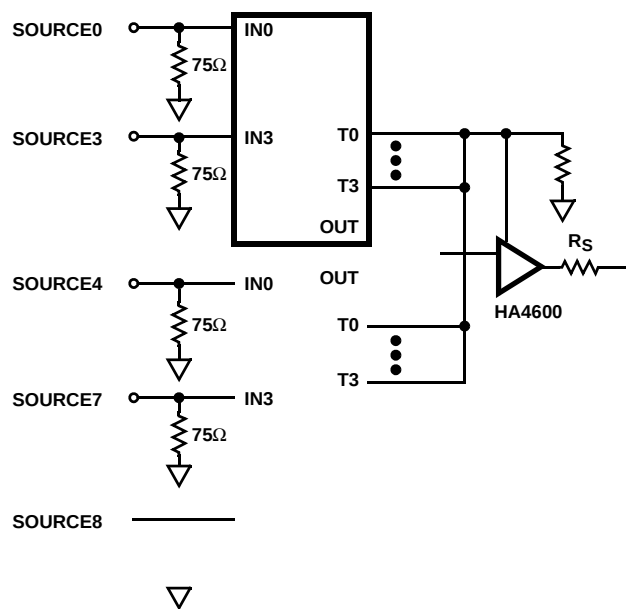


FIGURE 1. 4 x 4 SWITCHER/ROUTER APPLICATION



Typical Performance Curves $V_{\text{SUPPLY}} = \pm 5\text{V}$, $T_A = 25^\circ\text{C}$, $R_L = 10\text{k}\Omega$, Unless Otherwise Specified

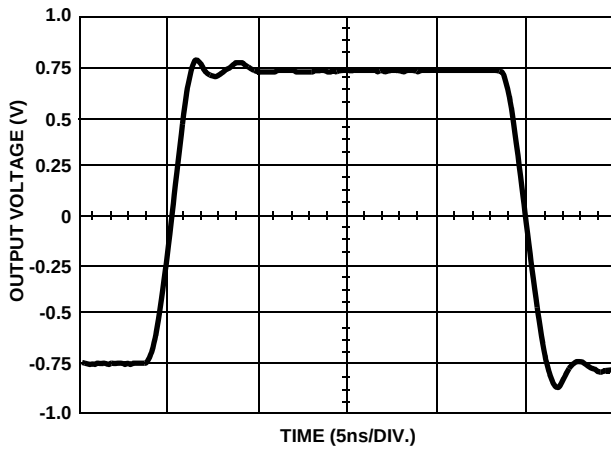


FIGURE 3. LARGE SIGNAL PULSE RESPONSE

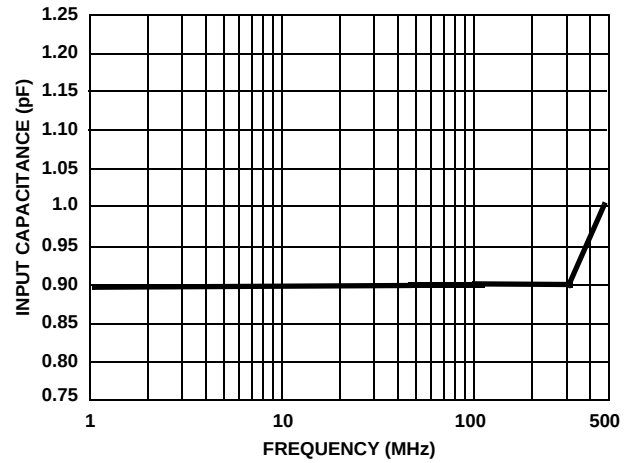


FIGURE 4. INPUT CAPACITANCE vs FREQUENCY

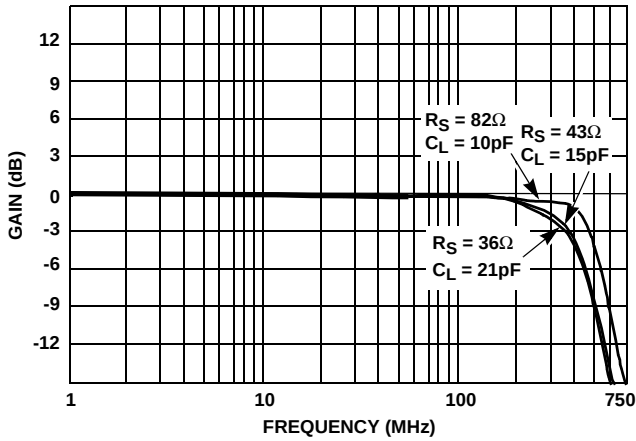


FIGURE 5. FREQUENCY RESPONSE

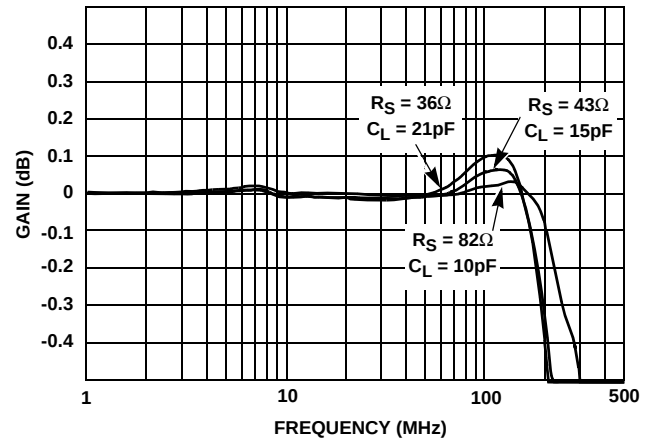


FIGURE 6. GAIN FLATNESS

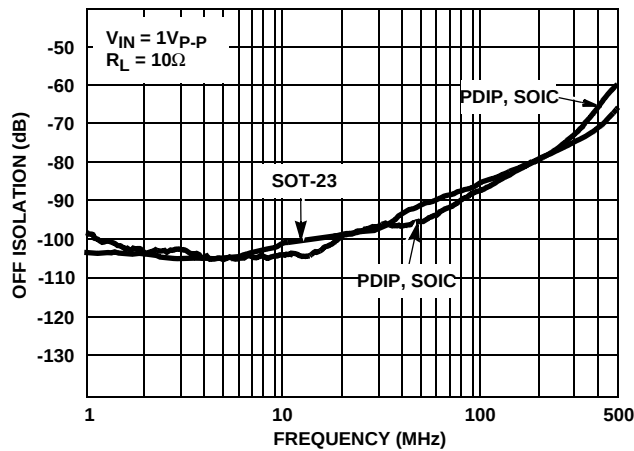
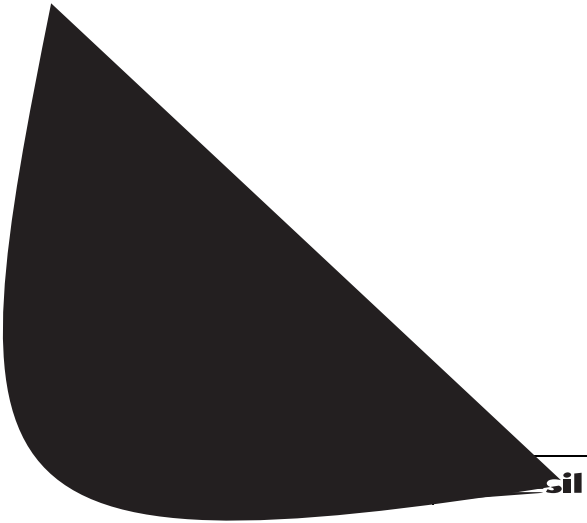


FIGURE 7. OFF ISOLATION



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