

1MHz, PWM Controlled,  
Step-Up DC/DC Converter, Ceramic Capacitor Compatible

## ■ GENERAL DESCRIPTION

The XC9119D10A series is 1MHz, PWM controlled step-up DC/DC converter, designed to allow the use of ceramic capacitors. With a built-in  $2.0\Omega$  switching transistor, the XC9119D10A series can easily provide a step-up operation by using only a coil, a diode, a capacitor, and a resistor, connected externally.

Since output voltage up to 19.5V (Maximum Lx operating voltage: 20V) can be derived with reference voltage supply of 1.0V ( $\pm 2.0\%$ ) and external components, the series can easily supply high voltage for various general-purpose power supplies, LCD panels and organic EL displays.

With a high switching frequency of 1.0MHz, a low profile and small board area solution can be achieved using a chip coil and an ultra small ceramic output capacitor.

With the current limit function (400mA (TYP.):  $V_{DD}=3.6V$ ), a peak current, which flows through built-in driver transistors can be limited. Soft-start time can be adjusted by external resistors and capacitors. The stand-by function enables the output to be turned off (CE 'L'), that is, the supply current will be less than  $1.0\mu A$ .

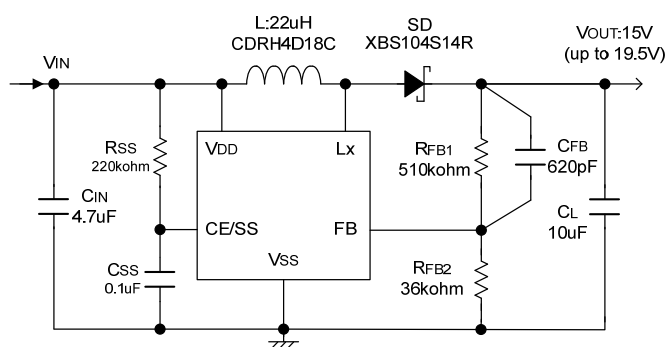
## ■ APPLICATIONS

- Organic electroluminescence display (OLED)
- Power supplies for LCDs
- Multi-function power supplies

## ■ FEATURES

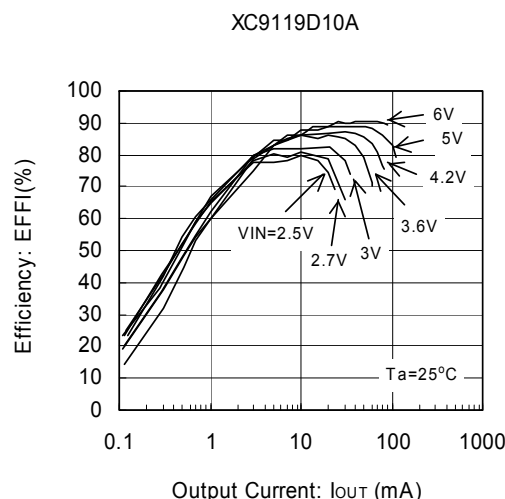
<b>Operating Voltage Range</b>	: 2.5V ~ 6.0V
<b>Output Voltage Range</b>	: Up to 19.5V externally set-up : Reference voltage $1.0V \pm 2.0\%$
<b>Oscillation Frequency</b>	: $1.0MHz \pm 20\%$
<b>ON Resistance</b>	: $2.0\Omega$ ( $V_{DD}=3.6V$ , $V_{DS}=0.4V$ )
<b>Efficiency</b>	: 86% ( $V_{OUT}=15V$ , $V_{DD}=3.6V$ , $I_{OUT}=10mA$ )
<b>Control</b>	: PWM control
<b>Stand-by function</b>	: $I_{STB}=1.0\mu A$ (MAX.)
<b>Load Capacitor</b>	: Low ESR ceramic capacitor
<b>Ultra Small Packages</b>	: SOT-25, USP-6C
<b>Lx Limit Current</b>	: 400mA ( $V_{DD}=3.6V$ )

## ■ TYPICAL APPLICATION CIRCUIT

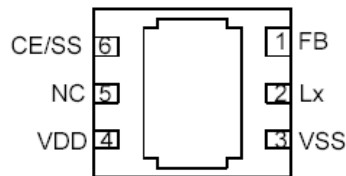
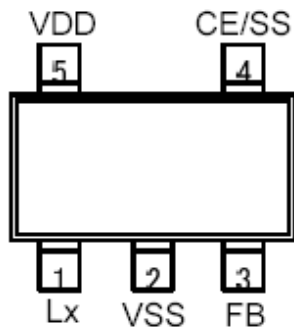


## ■ TYPICAL PERFORMANCE CHARACTERISTICS

OEfficiency vs. Output Current



## PIN CONFIGURATION



\*The dissipation pad for the USP-6C package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat resistance. If the pad needs to be connected to other pins, it should be connected to the VSS pin.

## PIN ASSIGNMENT

PIN NUMBER		PIN NAME	FUNCTION
SOT-25	USP-6C		
1	2	Lx	Switch
2	3	Vss	Ground
3	1	FB	Voltage Feedback
4	6	CE/SS	Chip Enable/ Soft Start
5	4	VDD	Power Input
-	5	NC	No Connection

## CE PIN FUNCTION

CE/SS PIN	OPERATIONAL STATE
H	Operation
L	Shut-down

## PRODUCT CLASSIFICATION

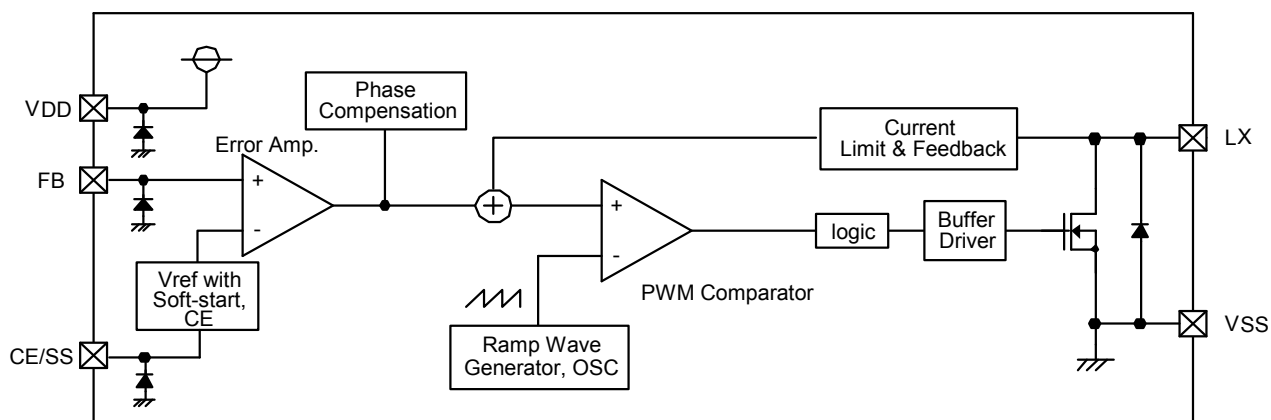
### Ordering Information

XC9119D①②③④⑤-⑥<sup>(\*)</sup>

DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
①②	Reference Voltage	10	FB voltage
③	Oscillation Frequency	A	1MHz
④⑤-⑥	Packages (Order Unit)	MR	SOT-25 (3,000/Reel)
		MR-G	SOT-25 (3,000/Reel)
		ER	USP-6C (3,000/Reel)
		ER-G	USP-6C (3,000/Reel)

<sup>(\*)</sup> The "-G" suffix denotes Halogen and Antimony free as well as being fully EU RoHS compliant.

## ■ BLOCK DIAGRAM



## ■ ABSOLUTE MAXIMUM RATINGS

Ta = 25°C

PARAMETER		SYMBOL	RATINGS	UNITS
VDD Pin Voltage		VDD	VSS – 0.3 ~ 7.0	V
Lx Pin Voltage		VLx	VSS – 0.3 ~ 22.0	V
FB Pin Voltage		VFB	VSS – 0.3 ~ 7.0	V
CE Pin Voltage		VCE	VSS – 0.3 ~ 7.0	V
Lx Pin Current		ILx	1000	mA
Power Dissipation	SOT-25	Pd	250	mW
	USP-6C		120	
Operating Ambient Temperature		Topr	- 40 ~ + 85	°C
Storage Temperature		Tstg	- 55 ~ +125	°C

## ELECTRICAL CHARACTERISTICS

XC9119D10AMR

Ta = 25 °C

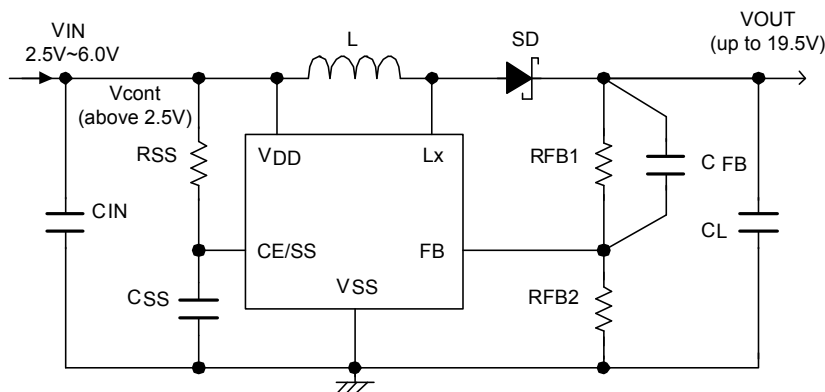
PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
FB Voltage	V <sub>FB</sub>	-	0.980	1.000	1.020	V	①
Line Regulation	$\frac{\Delta V_{FB}}{\Delta V_{IN} \cdot V_{FB}}$	2.5 ≤ V <sub>DD</sub> ≤ 6.0V	-	0.05	0.20	%/V	①
Supply Voltage	V <sub>DD</sub>	-	2.5	-	6.0	V	①
Operation Start-up Voltage	V <sub>ST1</sub>	I <sub>OUT</sub> =0mA	-	-	2.5	V	②
Supply Current 1	I <sub>DD1</sub>	V <sub>IN</sub> =V <sub>CE</sub> =3.0V, V <sub>FB</sub> =0V, V <sub>pull</sub> =5.0V R <sub>pull</sub> =100Ω	-	450	700	μA	②
Supply Current 2	I <sub>DD2</sub>	V <sub>FB</sub> =2.0V	-	55	110	μA	②
Stand-by Current	I <sub>STB</sub>	V <sub>CE</sub> =0V	-	-	1.0	μA	③
Oscillation Frequency	f <sub>OSC</sub>	Same as I <sub>DD1</sub>	0.8	1.0	1.2	MHz	②
Maximum Duty Ratio	MAXDTY	Same as I <sub>DD1</sub>	86	92	98	%	②
Efficiency (*1)	EFFI	V <sub>IN</sub> =V <sub>DD</sub> =3.6V, V <sub>OUT</sub> =15V, I <sub>OUT</sub> =10mA	-	86	-	%	①
Current Limit	I <sub>LIM</sub>	V <sub>DD</sub> =3.6V	310	400	750	mA	④
Lx Operating Voltage Range	V <sub>Lx</sub>	V <sub>OUT</sub> =18V	-	-	20.0	V	①
Lx Switch On Resistance	R <sub>SWON</sub>	V <sub>DD</sub> =3.6V, V <sub>Lx</sub> =0.4V, R <sub>pull</sub> =10Ω	-	2.0	4.0	Ω	②
Lx Leak Current	I <sub>LxL</sub>	Same as I <sub>STB</sub>	-	-	1	μA	③
CE "High" Voltage	V <sub>CEH</sub>	Applied voltage to CE when Lx pin voltage holding "H" level	0.65	-	-	V	②
CE "Low" Voltage	V <sub>CEL</sub>	Applied voltage to CE when Lx pin voltage holding "H" level	-	-	0.20	V	②
Soft-Start Threshold Voltage	V <sub>SST</sub>	V <sub>FB</sub> =0.95V, Applied voltage to CE when Lx voltage holding "H" level	1.3	1.6	1.9	V	②
CE "High" Current	I <sub>CEH</sub>	Same as I <sub>DD2</sub>	-0.1	-	0.1	μA	③
CE "Low" Current	I <sub>CEL</sub>	Same as I <sub>STB</sub>	-0.1	-	0.1	μA	③
FB "High" Current	I <sub>FBH</sub>	Same as I <sub>DD2</sub>	-0.1	-	0.1	μA	③
FB "Low" Current	I <sub>FBL</sub>	Same as I <sub>STB</sub>	-0.1	-	0.1	μA	③

Test Condition: Unless otherwise stated, V<sub>IN</sub>=3.0V, V<sub>CE</sub>=3.0V, V<sub>FB</sub>=0V, V<sub>pull</sub>=5.0V, R<sub>pull</sub>=100Ω.

NOTE:

\*1: EFFI={(output voltage x output current) / (input voltage) x (input current)} x 100

## TYPICAL APPLICATION CIRCUIT



## ■ OPERATIONAL EXPLANATION

The XC9119D10A series consists of a reference voltage source, ramp wave circuit, error amplifier, PWM comparator, phase compensation circuit, driver transistor, current limiter circuit and others. The series ICs compare, using the error amplifier, the voltage of the internal reference voltage source with the feedback voltage from the FB pin. Phase compensation is performed on the resulting error amplifier output, to input a signal to the PWM comparator to determine the turn-on time during switching. The PWM comparator compares, in terms of voltage level, the signal from the error amplifier with the ramp wave from the ramp wave circuit, and delivers the resulting output to the buffer drive circuit to cause the Lx pin to output a switching duty cycle. This process is continuously performed to ensure stable output voltage. The current feedback circuit detects the N-channel MOS driver transistor's current for each switching operation, and modulates the error amplifier output signal to provide multiple feedback signals. This enables a stable feedback loop even when a low ESR capacitor, such as a ceramic capacitor, is used, ensuring stable output voltage.

### <Reference Voltage Source>

The reference voltage source provides the reference voltage to ensure stable output voltage of the IC.

### <Ramp Wave Circuit>

The ramp wave circuit determines switching frequency. The 1MHz (TYP.) of frequency is fixed internally. Clock pulses generated in this circuit are used to produce ramp waveforms needed for PWM operation.

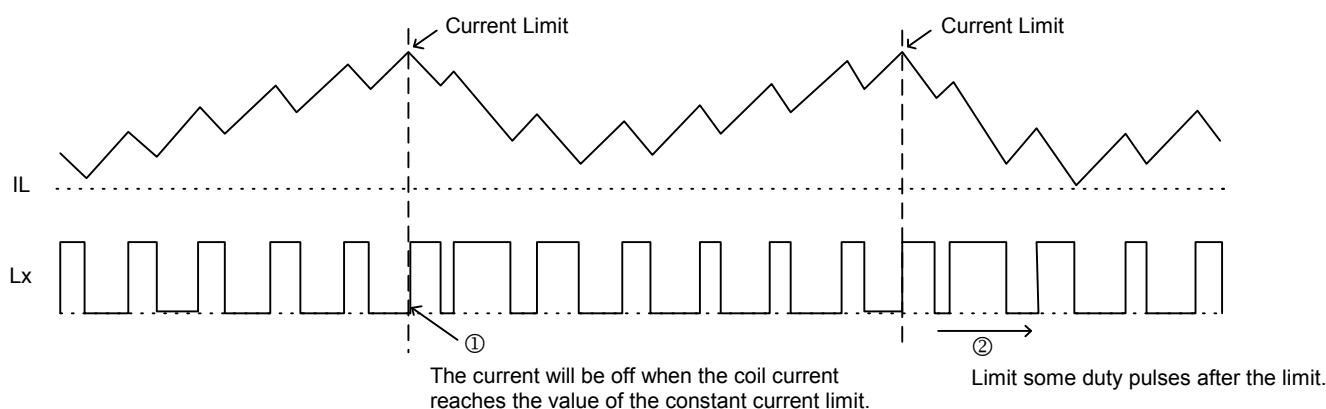
### <Error Amplifier>

The error amplifier is designed to monitor output voltage. The amplifier compares the reference voltage with the FB pin voltage. When a voltage lower than the reference voltage is fed back, the output voltage of the error amplifier increases. Gain and frequency characteristics of the error amplifier output are fixed internally as an optimize signal.

### <Current Limit >

The current limit circuit of the XC9119D10A series monitors the current flowing through the N-channel MOS driver transistor connected to the Lx pin, and features a combination of the constant-current type current limit mode and the duty cycle limit of the next pulse.

- ①When the driver current is greater than a specific level, the constant-current type current limit function operates to turn off the pulses from the Lx pin at any given timing.
- ②The IC controls the next pulse to be smaller than the first pulse.



### <CE Pin Function>

The operation of the XC9119D10A series will enter into the shut down mode when a low level signal is input to the CE pin. During the shut down mode, the supply current is 0μA (TYP.), with high impedance at the Lx pin. The IC starts its operation with a high level signal to the CE pin. The input to the CE pin is a CMOS input and the sink current is 0μA (TYP.). The hysteresis between the chip enable and the chip disable is 50mV (TYP.).

### <Soft-Start Time>

Soft-start function operates when capacitors and resistors are connected to the CE/SS pin. With the Vref voltage limited by the CE/SS pin start-up voltage and applying the input to the error amps, the operation maintains a balance between the two inputs of the error amps. and controls the Lx pin's ON time so that it doesn't increase more than is necessary. Depending of current limit function, load current, step-up ratio, and external components, the IC takes about 500μs to 5ms to attain the setting voltage after applying the CE 'H' voltage even though the RSS is 0Ω and a soft start capacitor C<sub>SS</sub> is not connected. (For a numerical constant, please refer to Note on Use.) For longer soft-start time, please connect RSS and C<sub>SS</sub>. Soft-start function operates while the CE pin voltage is between 0V to around 1.9V. Please be noted that if the CE/SS pin voltage does not start from 0V but is in intermediate potential when the power is turned on etc., soft start function may lose an effect and that will cause a high inrush current and ripple voltage.

## OPERATIONAL EXPLANATION (Continued)

### <CE/SS (Pin No. 4): Chip Enable / Soft-Start Pin>

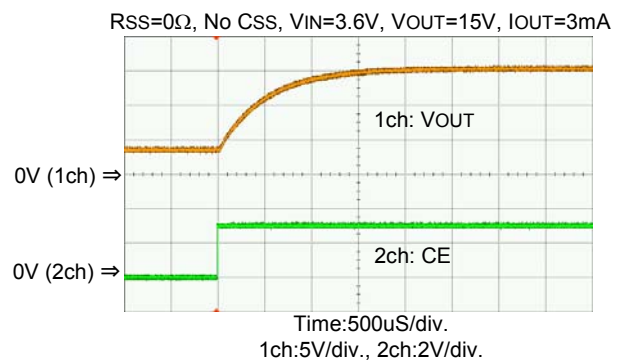
Pin No. 4 can be used as in either chip enable (CE) pin or soft-start (SS) pin. The IC takes about 5ms at most to attain the setting voltage after starting operation (CE 'H') even though the  $R_{ss}$  is  $0\Omega$  and the  $C_{ss}$  is not connected.

Soft-start function is good for setting a longer time than the start-up time when the  $R_{ss}$  is  $0\Omega$  and the  $C_{ss}$  is not connected. Soft-start operates while the CE pin voltage increases from 0V to around 1.9V. The following equation is used with the values of  $V_{cont}$  voltage, the  $R_{ss}$  and the  $C_{ss}$ .

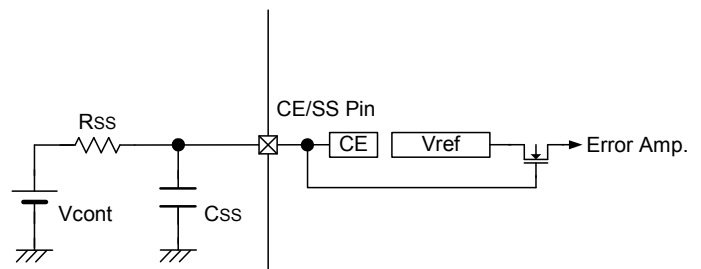
$$T = -C_{ss} \times R_{ss} \times \ln \left\{ (V_{cont} - 1.6) / V_{cont} \right\}$$

#### ● Start-up waveform

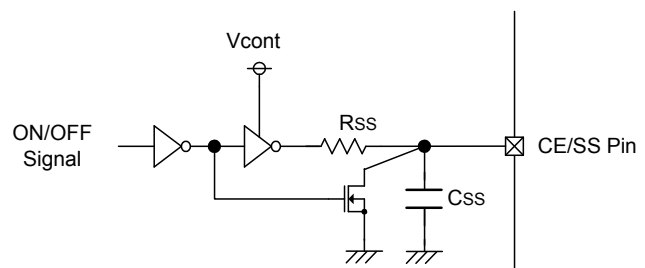
when the  $R_{ss}$  is  $0\Omega$  and the  $C_{ss}$  is not connected



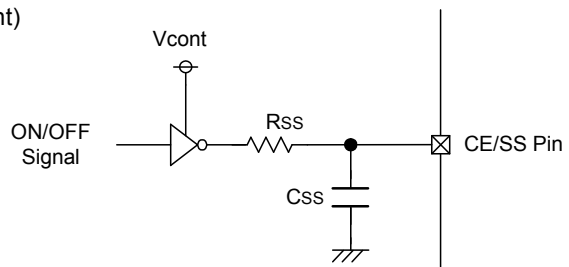
Ex.) When  $C_{ss}=0.1\mu F$ ,  $R_{ss}=220k\Omega$ ,  $V_{cont}=5V$ ,  
 $T = -0.1 \times 10^{-6} \times 220 \times 10^3 \times \ln((5-1.6)/5) = 8.48ms$



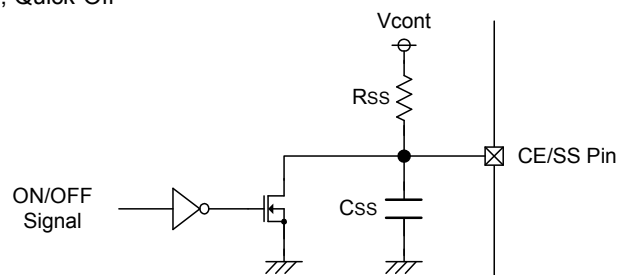
Ex.) Reference Circuit 1: N-ch Open Drain



Ex.) Reference Circuit 2: CMOS Logic (Low Supply Current)



Ex.) Reference Circuit 3: CMOS Logic (Low Supply Current), Quick-Off



## ■ OPERATIONAL EXPLANATION (Continued)

<Lx (Pin No. 1): Switch Pin>

Please connect the anode of a Schottky barrier diode and inductor to the Lx pin.

<FB (Pin No. 3): Voltage Feedback Pin>

The reference voltage is 1.0V (TYP.). Output voltage is approximated by the following equation according to the value for two resistors (RFB1 and RFB2). The sum of the two resistors should be 1MΩ or less.

$$V_{OUT} = R_{FB1} / R_{FB2} + 1$$

Output voltage should be set as to fill  $V_{OUT} < (\text{Maximum value of } V_{Lx}) - (V_F \text{ of Schottky diode})$ .

Please adjust the CFB value of the speed-up capacitor for phase compensation so that  $f_{zfb} = 1 / (2 \pi \times C_{FB} \times R_{FB1})$  will be about 500Hz. According to the usage, adjusting the inductance value, the load capacity value, and so on to the most suitable operation.

Typical example:

V <sub>OUT</sub> (V)	R <sub>FB1</sub> (kΩ)	R <sub>FB2</sub> (kΩ)	C <sub>FB</sub> (pF)
3.3	300	130	1000
5.0	300	75	1000
7.0	180	30	1800
10.0	270	30	1200
15.0	510	36	510
18.0	510	30	510

<VDD (Pin No. 5): Power Supply Pin>

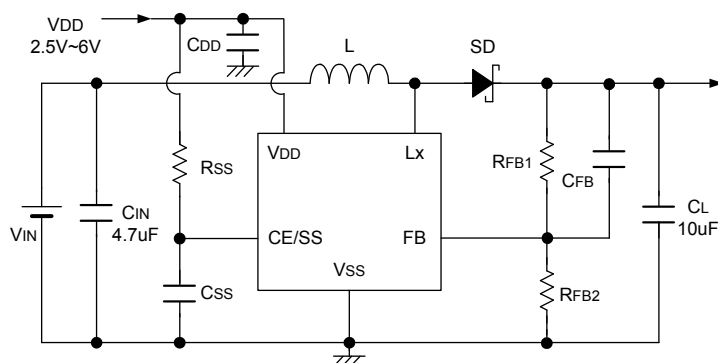
Please connect an input by-pass capacitor (C<sub>IN</sub>).

## ● Application Information

<Obtaining VDD from other source than V<sub>IN</sub>>

In case that the input voltage V<sub>IN</sub> and power source VDD in the step-up circuit are isolated, the circuit starts step-up operations with the input voltage less than 2.5V when voltage from 2.5V to 6.0V is applied to the power source. Please connect more than 1μF of CDD between the VDD pin and the VSS pin as close as possible.

Ex.) When VDD=3.6V, V<sub>IN</sub>=1.8V, V<sub>OUT</sub>=5.0V (R<sub>FB1</sub>=300kΩ, R<sub>FB2</sub>=75kΩ, C<sub>FB</sub>=1000pF, C<sub>L</sub>=10μF), the IC can operate up to I<sub>OUT</sub>=40mA.

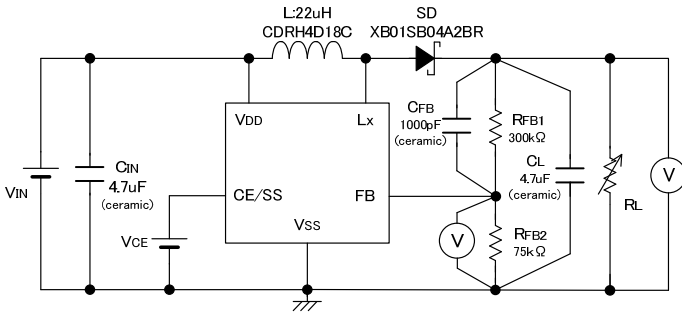


## ■ NOTES ON USE

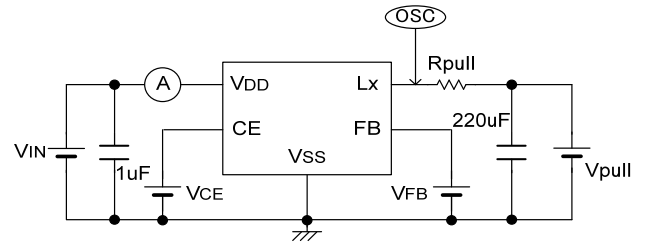
1. For temporary, transitional voltage drop or voltage rising phenomenon, the IC is liable to malfunction should the ratings be exceeded.
2. Please do not exceed the value of stated absolute maximum ratings.
3. The DC/DC converter performance is greatly influenced by not only the ICs' characteristics, but also by those of the external components. Care must be taken when selecting the external components.
4. Make sure that the PCB GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
5. Please mount each external component as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
6. Please set up the output voltage value so that the Lx pin voltage does not exceed 20V.
7. Torex places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using Torex products in their systems.

## TEST CIRCUITS

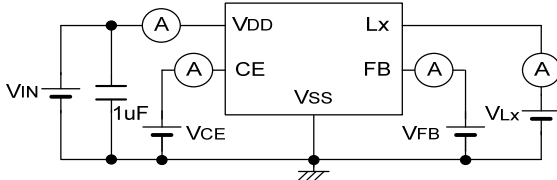
Circuit ①



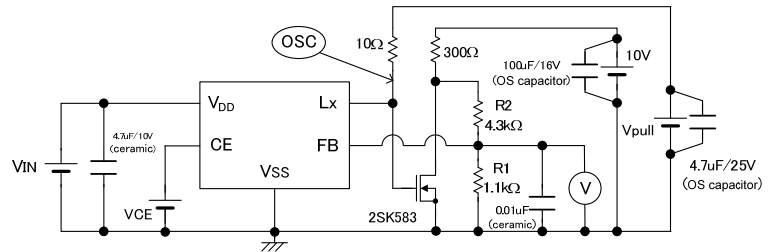
Circuit ②



Circuit ③



Circuit ④



### 1. The measurement method of Lx On resistance RSWON

Using the circuit ②, Lx On resistance can be measured by adjusting Vpull voltage to set Lx voltage VLx x 0.4V when the driver transistor is ON. The oscilloscope is used for measuring the Lx voltage when the driver transistor is ON.

$$R_{SWON} = 0.4 / \{(V_{pull} - 0.4) / 10\}$$

### 2. The measurement method of current limit ILIM

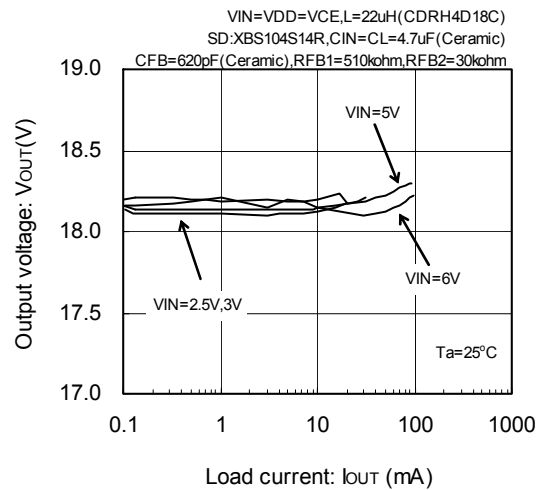
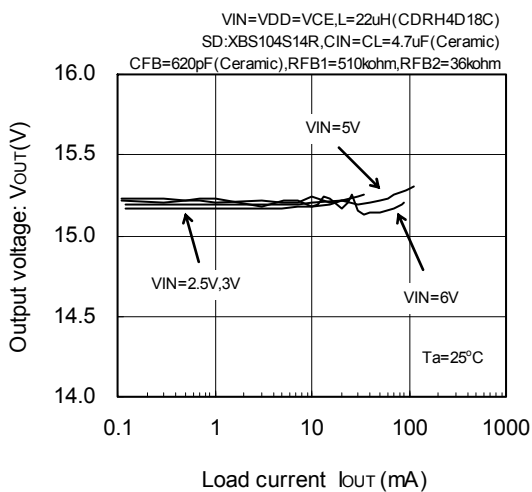
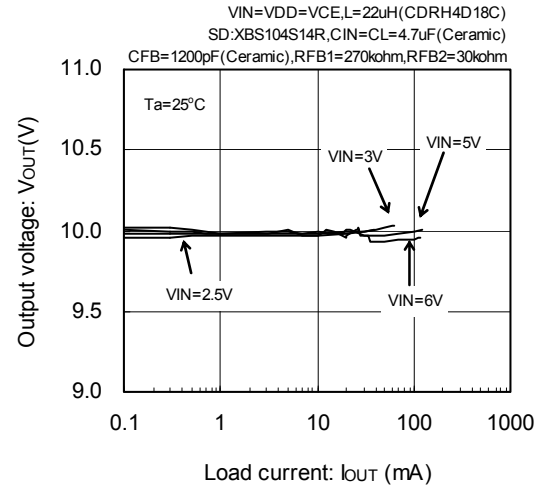
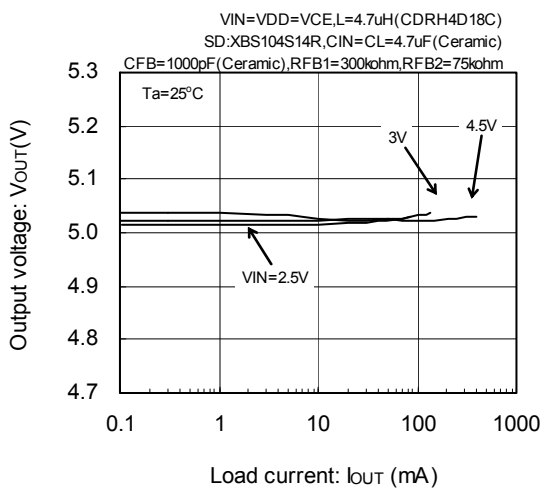
Using the circuit ④, current limit ILIM can be calculate by the equation including Vpull voltage when FB voltage is decreased while Vpull voltage is adjusted and Lx voltage VLx when the driver transistor is ON. The oscilloscope is used for measuring the Lx voltage when the driver transistor is ON.

$$I_{LIM} = (V_{pull} - V_{Lx}) / R_{pull}$$

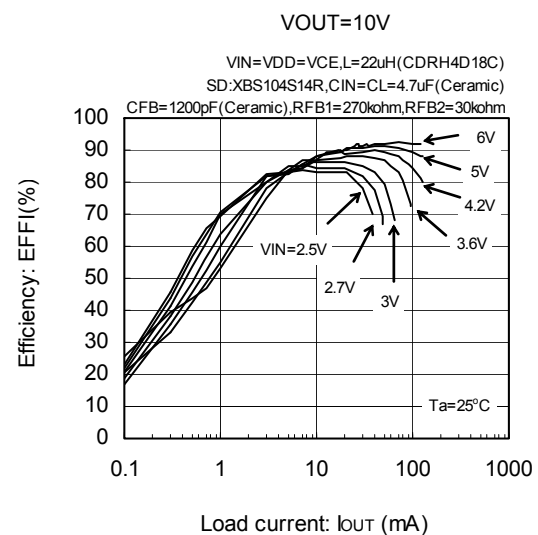
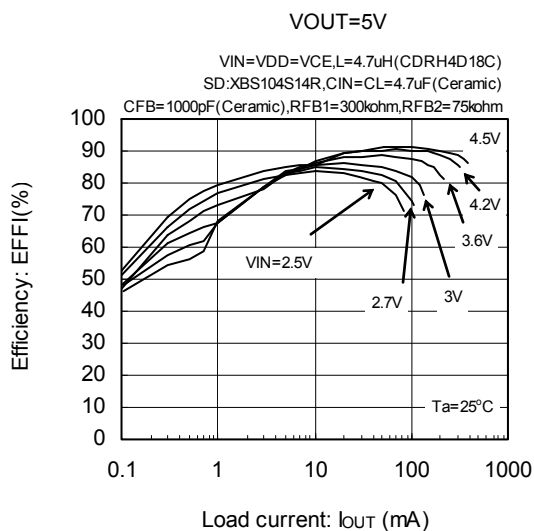


## ■ TYPICAL PERFORMANCE CHARACTERISTICS

(1) Output Voltage vs. Output Current

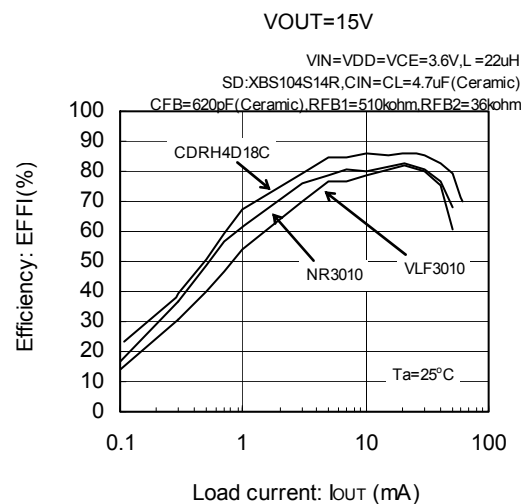
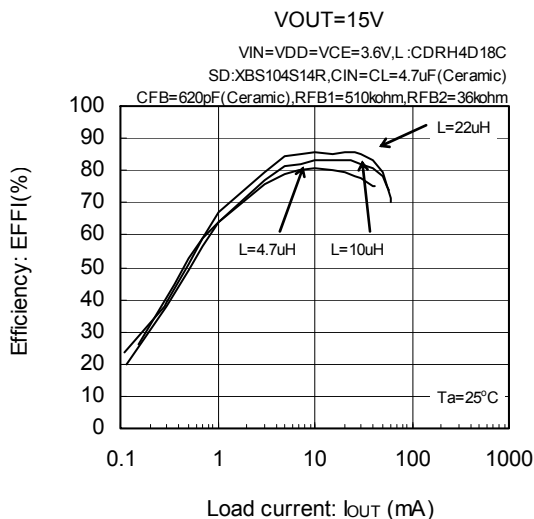
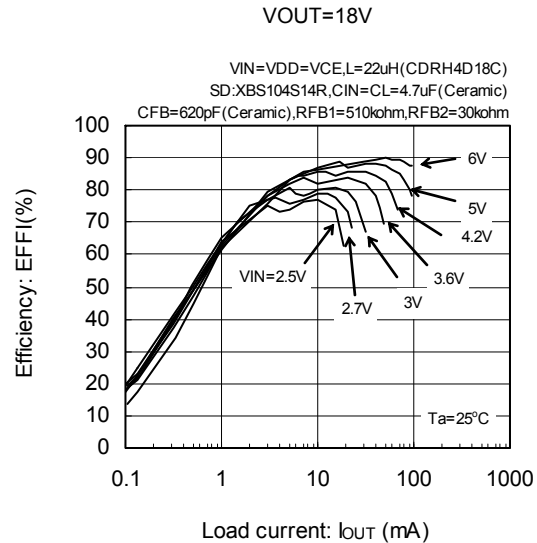
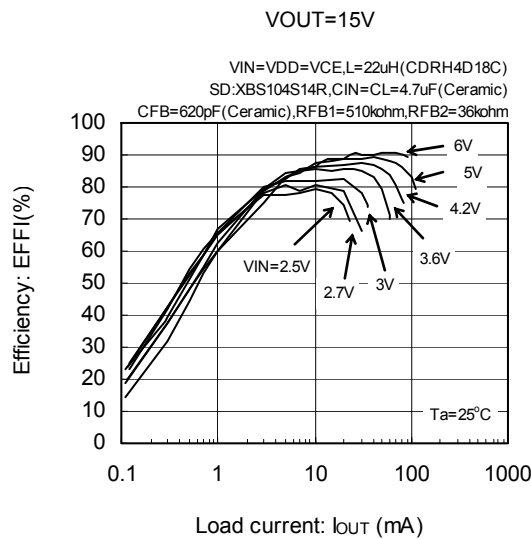


(2) Efficiency vs. Output Current

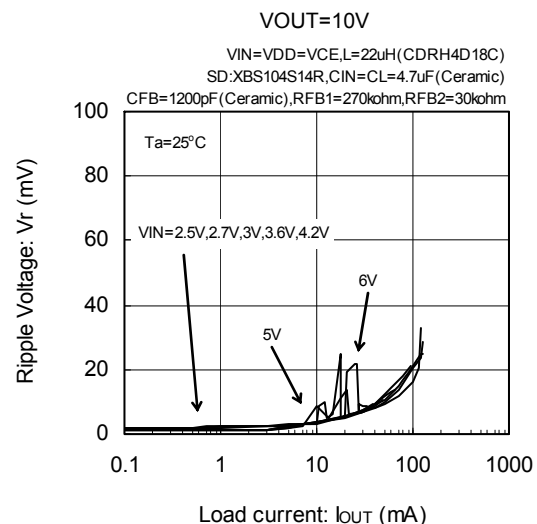
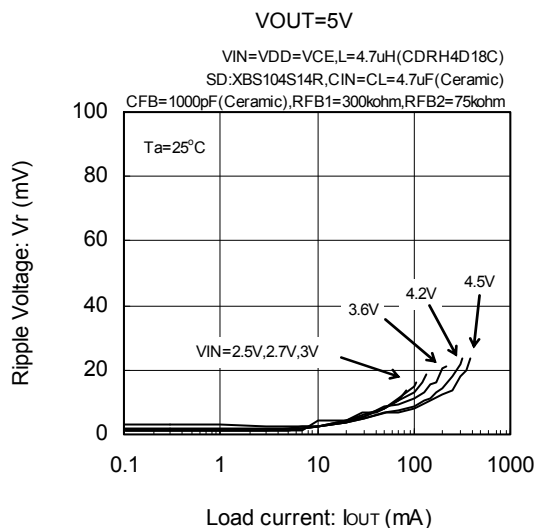


## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

### (2) Efficiency vs. Output Current (Continued)

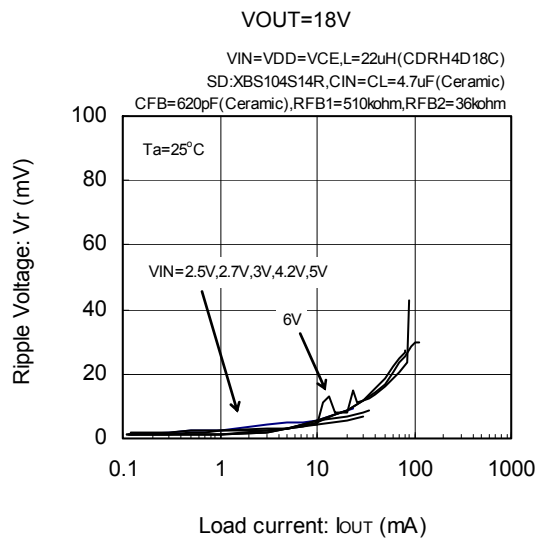
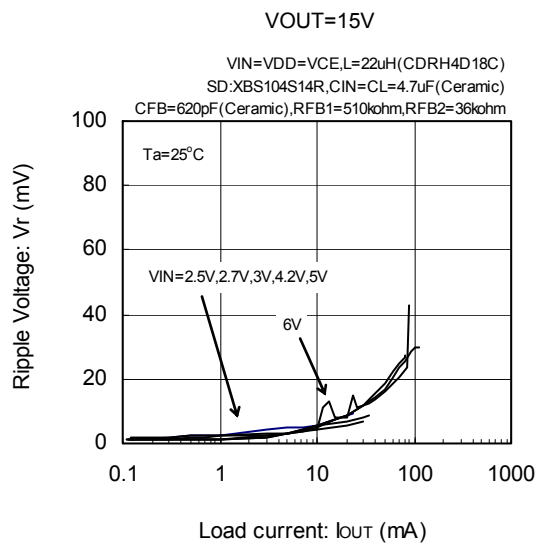


### (3) Ripple Voltage vs. Output Current

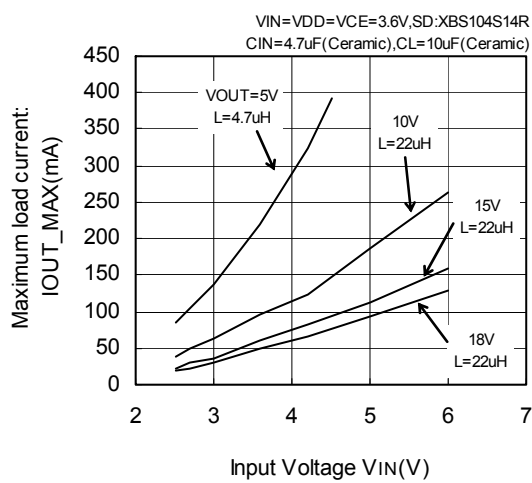


## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

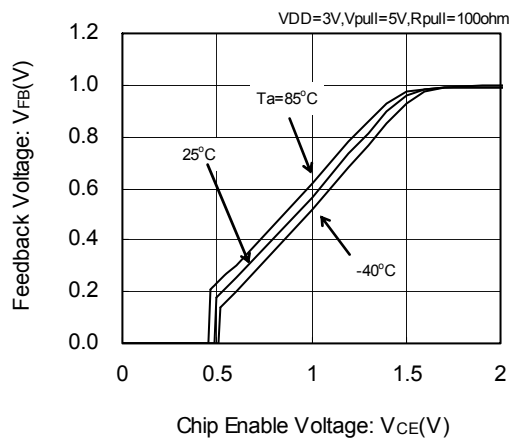
(3) Ripple Voltage vs. Output Current (Continued)



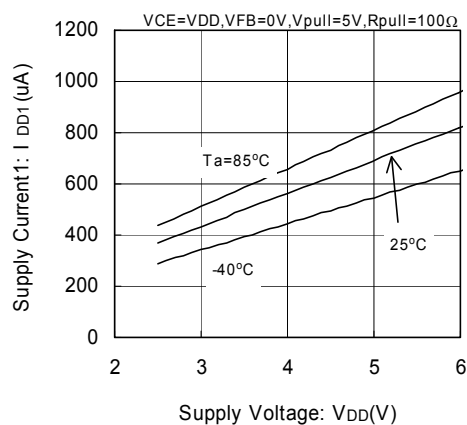
(4) Maximum Output Current vs. Input Voltage



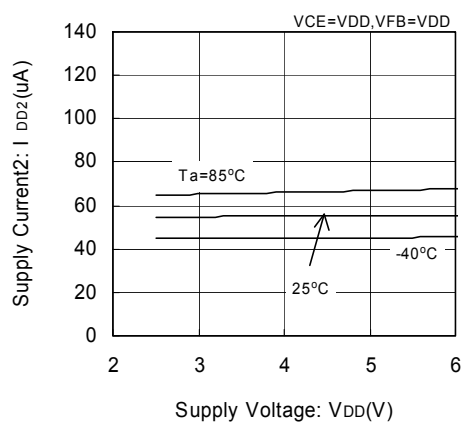
(5) Feedback Voltage vs. Chip Enable Voltage



(6) Supply Current 1 vs. Supply Voltage

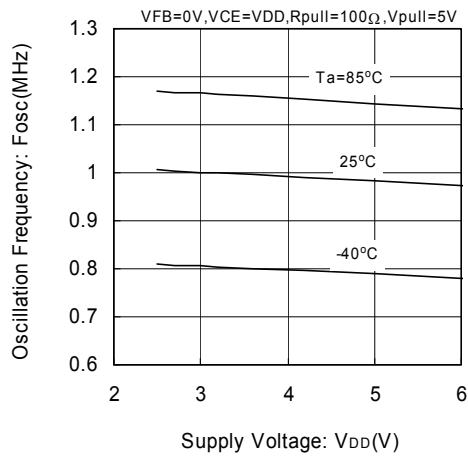


(7) Supply Current 2 vs. Supply Voltage

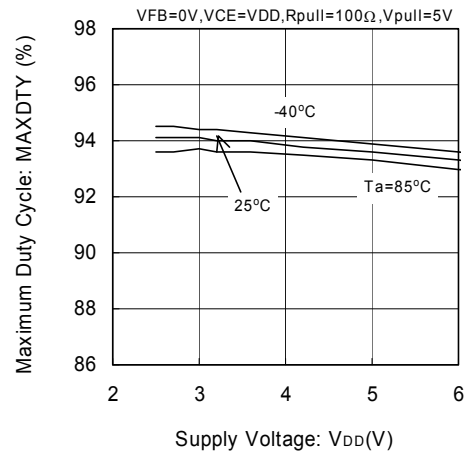


## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

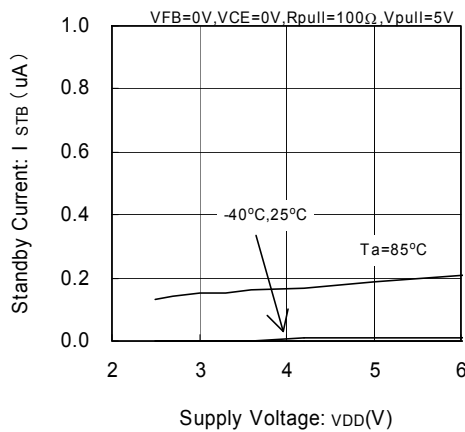
(8) Oscillation Frequency vs. Supply Voltage



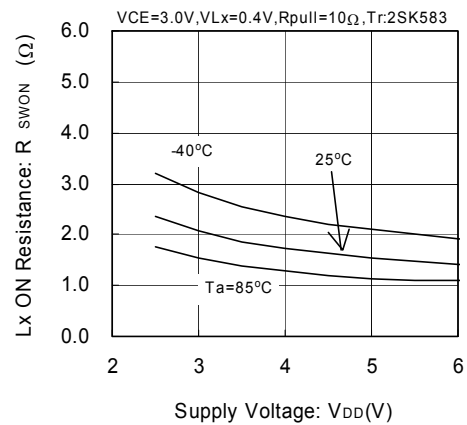
(9) Maximum Duty Cycle vs. Supply Voltage



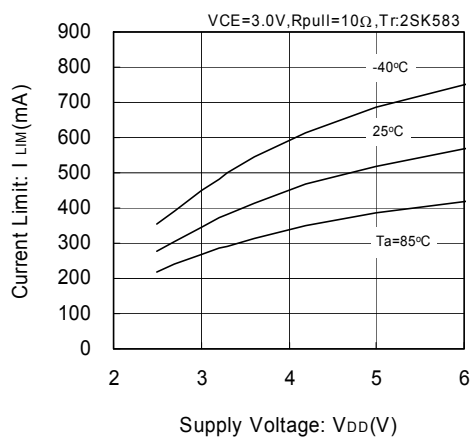
(10) Stand-by Current vs. Supply Voltage



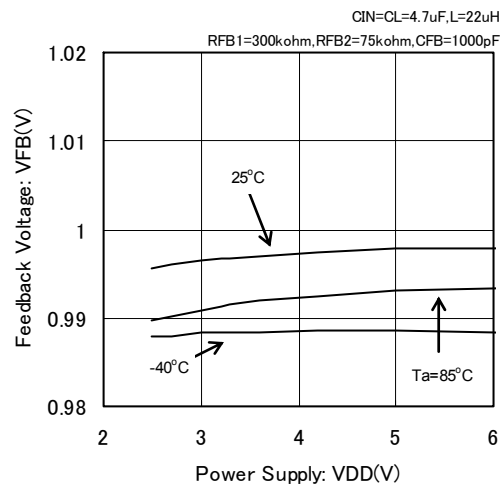
(11) Lx ON Resistance vs. Supply Voltage



(12) Current Limit vs. Supply Voltage

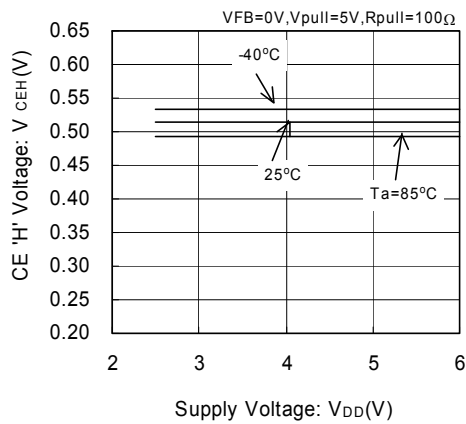


(13) Feedback Voltage vs. Supply Voltage

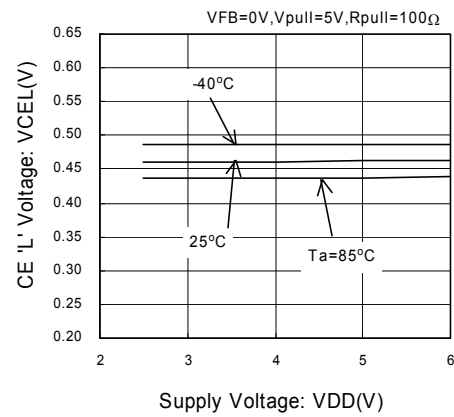


## ■ TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

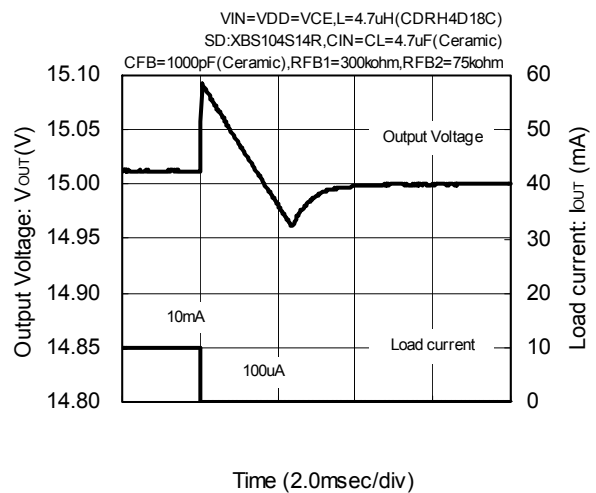
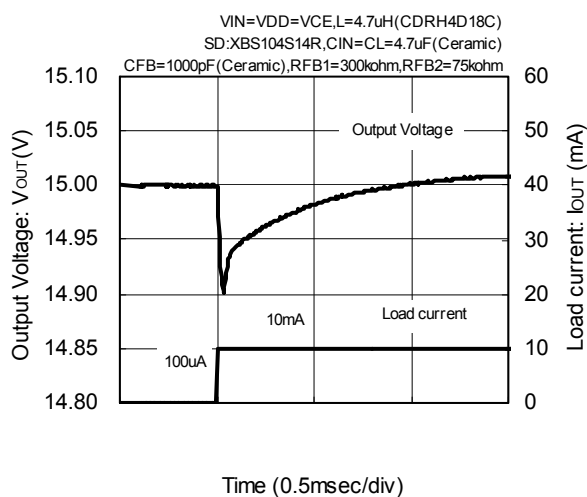
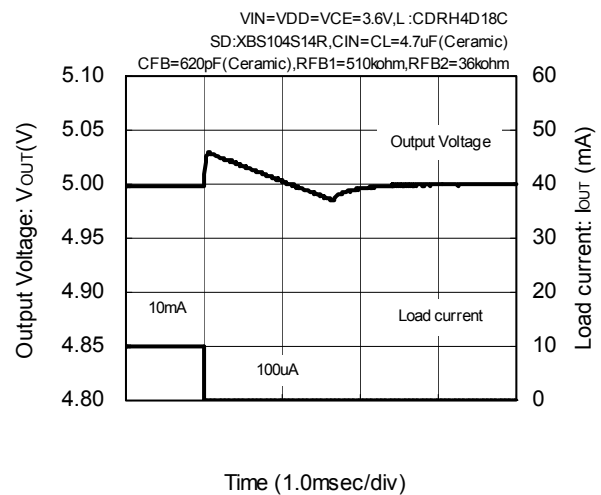
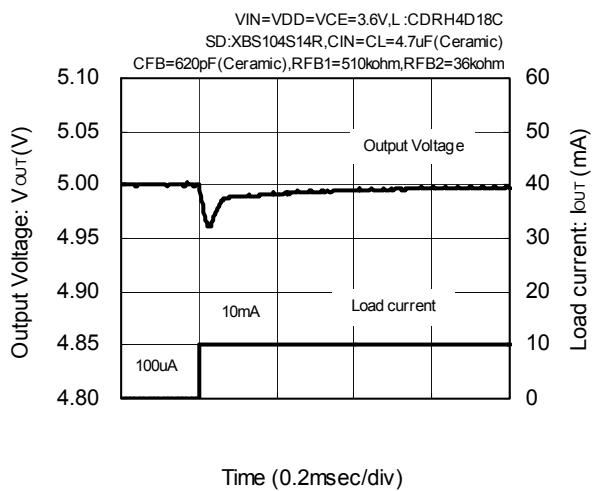
(14) CE 'H' Voltage vs. Supply Voltage



(15) CE 'L' Voltage vs. Supply Voltage

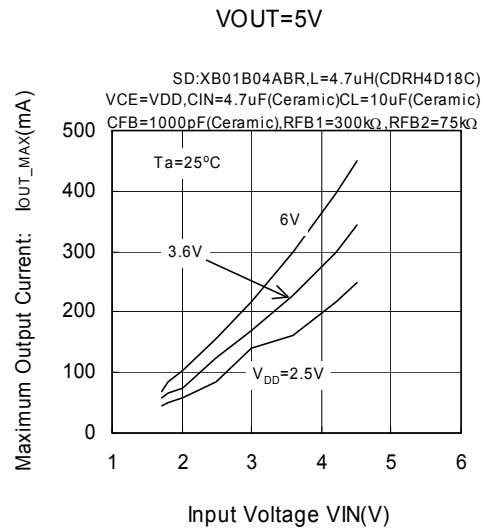
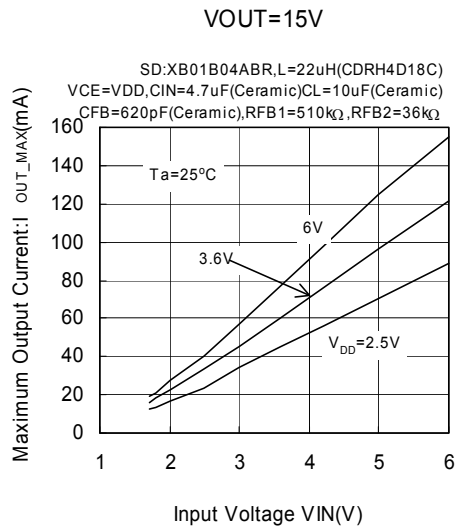


(16) Load Transient Response



## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

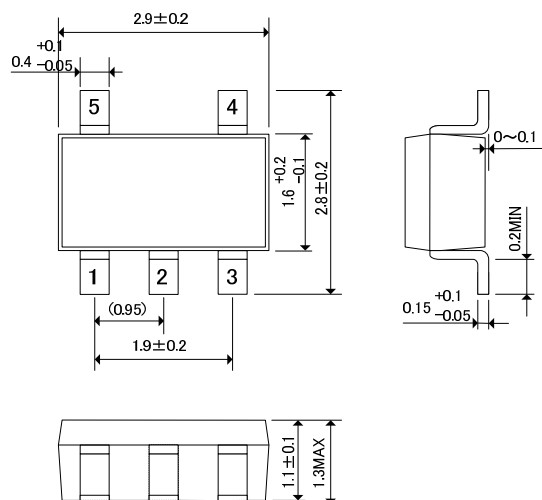
(17) Maximum Output Current vs. Input Voltage



## ■ PACKAGING INFORMATION

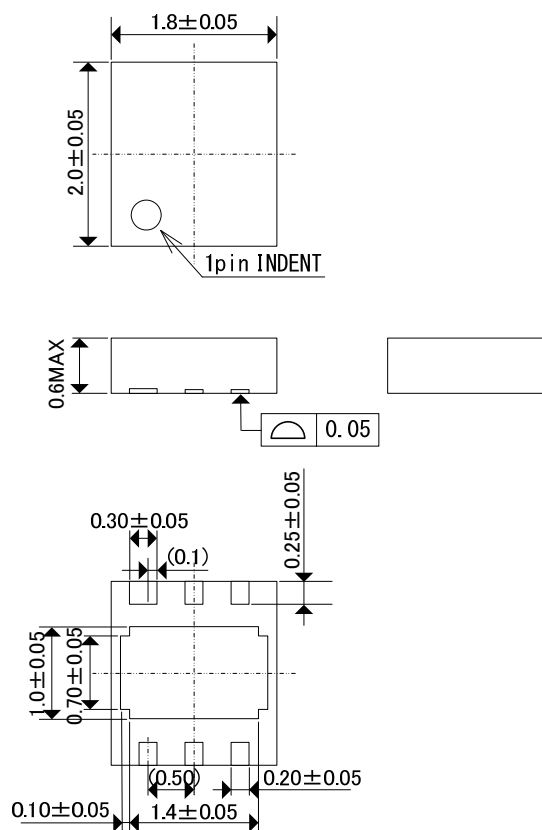
### ● SOT-25

(unit : mm)



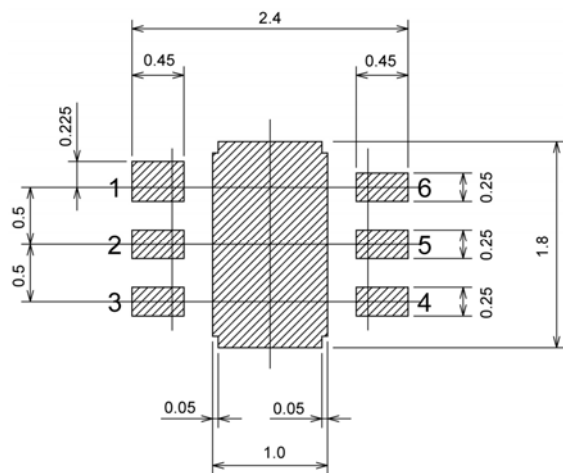
### ● USP-6C

(unit : mm)

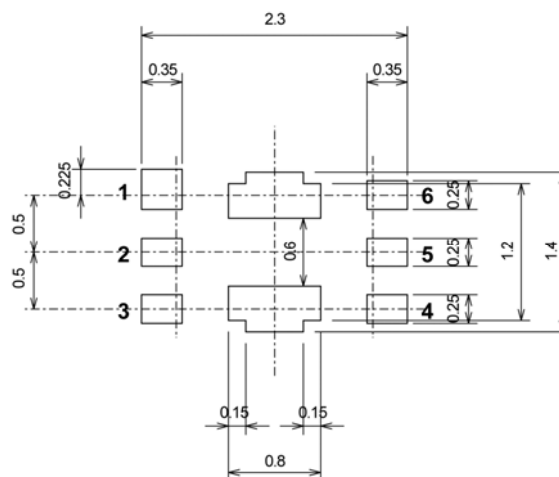


## ■ PACKAGING INFORMATION (Continued)

● USP-6C Reference Pattern Layout



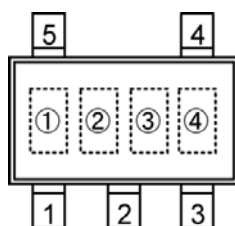
● USP-6C Reference Metal Mask Design





## ■ MARKING RULE

### ● SOT-25



SOT-25  
(TOP VIEW)

① represents product series

MARK	PRODUCT SERIES
<b>L</b>	XC9119xxxxMx

② represents Lx overvoltage limit

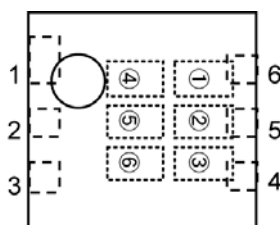
MARK	Lx OVERVOLTAGE LIMIT	PRODUCT SERIES
D	Not Available	XC9119DxxxMx

③ represents oscillation frequency

MARK	OSCILLATION FREQUENCY	PRODUCT SERIES
A	1MHz	XC9119xxxAMx

④ represents production lot number  
0 to 9 and A to Z, or inverted characters 0 to 9 and A to Z repeated.  
(G, I, J, O, Q, W excepted)

### ● USP-6C



USP-6C  
(TOP VIEW)

① represents product series

MARK	PRODUCT SERIES
V	XC9119xxxxDx

② represents Lx overvoltage limit

MARK	Lx OVERVOLTAGE LIMIT	PRODUCT SERIES
D	Not Available	XC9119DxxxDx

③④ represents FB voltage

MARK		FB VOLTAGE (V)	PRODUCT SERIES
③	④		
1	0	1.0	XC9119x10xDx

⑤ represents oscillation frequency

MARK	OSCILLATION FREQUENCY	PRODUCT SERIES
A	1MHz	XC9119xxxADx

⑥ represents production lot number  
0 to 9 and A to Z repeated (G, I, J, O, Q, W excepted)  
\* No character inversion used.

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