

### DESCRIPTION

The MP20042 is a dual-channel, micropower, low noise, low dropout and high PSRR linear regulator. The output voltage of MP20042 ranges from 1.2V to 5.0V and 1% accuracy by operating from a +2.5V to +6.0V input. The MP20042 can supply up to 200mA of load current at each channel.

The MP20042 uses an internal PMOS as the pass element, which consumes 114 $\mu$ A supply current (both LDOs on) at no load condition. The EN1 and EN2 pins control each output respectively. When both channels shutdown simultaneously, the chip will be turned off and consume nearly zero operation current which is suitable for battery-power devices. The MP20042 features current limiting and over temperature protection.

It is available in a 2mm x 2mm QFN8 package.

### FEATURES

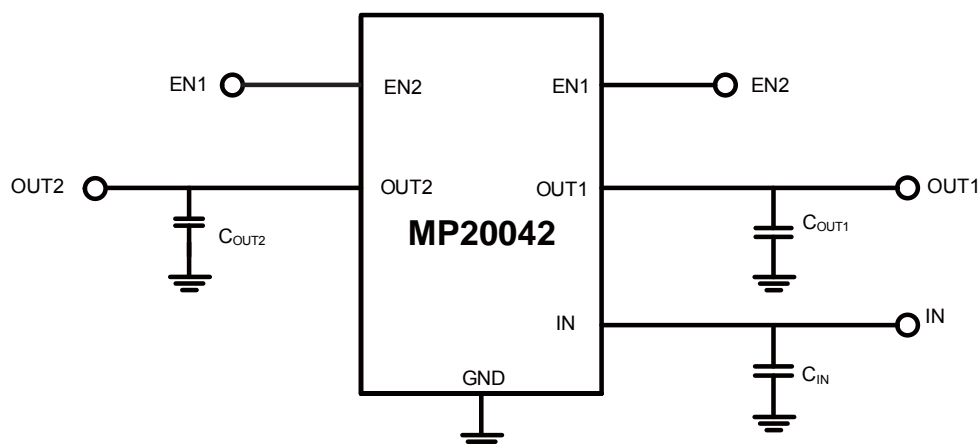
- Two LDOs in a 2mmx2mm QFN8 Package
- Up to 200mA Output Current (Per Channel)
- Dual Enable Pins Control Each Output
- 72dB PSRR at 1kHz
- 11 $\mu$ V<sub>RMS</sub> Low Noise Output
- 110mV Dropout at 100mA Load
- Very Fast Transient Responses with Small Output Capacitor
- Current Limiting and Thermal Protection

### APPLICATIONS

- Cellular Phones
- Battery-powered Equipment
- Laptop, Notebook, and Palmtop Computers
- Hand-held Equipment
- Wireless LAN

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### TYPICAL APPLICATION



## ORDERING INFORMATION\*

Part Number	V <sub>OUT1</sub>	V <sub>OUT2</sub>	Package	Temperature	Top Marking
MP20042DG-JG-LF-Z	2.5V	1.8V	QFN8 (2mmx2mm)	-40°C to +85°C	7E
MP20042DG-ZS-LF-Z	5.0V	3.3V			7L
MP20042DG-PN-LF-Z	3.0V	2.85V			8L
MP20042DG-DD-LF-Z	1.85V	1.85V			9L
MP20042DG-SJ-LF-Z	3.3V	2.5V			2M
MP20042DG-MG-LF-Z	2.8V	1.8V			3M

\* Other output voltage versions between 1.2V and 5.0V are available in 100mV increments. Contact factory for availability.

## ORDERING GUIDE\*\*

MP20042DG-□□-LF-Z



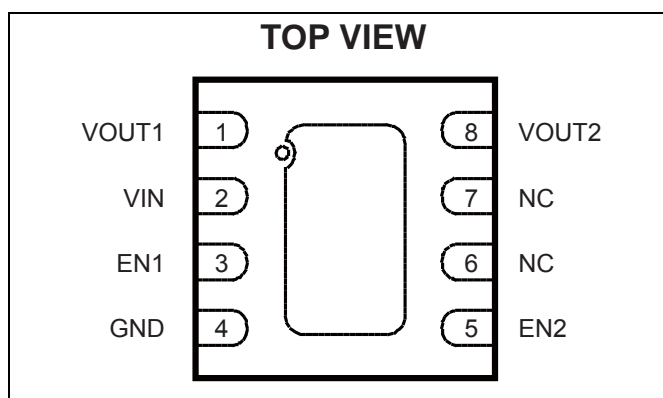
\*\* For RoHS Compliant Packaging, add suffix - LF (e.g. MP20042DG-□□-LF); For Tape and Reel, add suffix -Z (e.g. MP20042DG-□□-LF-Z)

## OUTPUT VOLTAGE SELECTOR GUIDE\*\*\*

Code	V <sub>OUT</sub>	Code	V <sub>OUT</sub>
C	1.2	T	2.65
B	1.3	L	2.7
F	1.5	<b>M</b>	<b>2.8</b>
W	1.6	<b>N</b>	<b>2.85</b>
<b>G</b>	<b>1.8</b>	V	2.9
<b>D</b>	<b>1.85</b>	<b>P</b>	<b>3.0</b>
Y	1.9	Q	3.1
H	2.0	X	3.15
E	2.1	R	3.2
<b>J</b>	<b>2.5</b>	<b>S</b>	<b>3.3</b>
K	2.6	<b>Z</b>	<b>5.0</b>

\*\*\* Code in **Bold** are standard versions. For other output voltages between 1.2V and 5.0V contact factory for availability. Minimum order quantity on non-standard versions is 25,000 units.

## PACKAGE REFERENCE



## ABSOLUTE MAXIMUM RATINGS <sup>(1)</sup>

Supply Input Voltage .....	6.5V
Continuous Power Dissipation (T <sub>A</sub> = +25°C) <sup>(2)</sup>	1.25W
Operation Temperature Range ...	-40°C to 85°C
Storage Temperature Range .....	-65°C to 150°C
Lead Temperature (Soldering, 10sec) .....	260°C

## Recommended Operating Conditions <sup>(3)</sup>

Supply Input Voltage.....	2.5V to 6.0V
Enable Input Voltage .....	0V to 6.0V
Junction Temperature Range .	-40°C to +125°C

Thermal Resistance <sup>(4)</sup>	$\theta_{JA}$	$\theta_{JC}$
2x2 QFN8 .....	80 .....	16 .. °C/W

### Notes:

- 1) Exceeding these ratings may cause permanent damage to the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub>(MAX), the junction-to-ambient thermal resistance  $\theta_{JA}$ , and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub>(MAX)=(T<sub>J</sub>(MAX)-T<sub>A</sub>)/  $\theta_{JA}$ . Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- 3) The device is not guaranteed to function outside its operating conditions.
- 4) Measured on JESD51-7 4-layer board.

## ELECTRICAL CHARACTERISTICS

$V_{IN}=3.6V$ ,  $V_{OUT1}=2.5V$ ,  $V_{OUT2}=1.8V$ ,  $C_{IN}=C_{OUT1}=C_{OUT2}=2.2\mu F$ ,  $EN1=EN2=V_{IN}$ , Typical Value at  $T_A=25^{\circ}C$  for each LDO unless otherwise noted.

Parameter	Symbol	Condition	Min	Typ	Max	Units
Output Voltage Accuracy (Load regulation) <sup>(5)</sup>	$\Delta V_{OUT}$	$I_{LOAD} = 1mA$ to $200mA$	-1		+1	%
Maximum Output Current	$I_{MAX}$	Continuous	200			mA
Current Limit	$I_{LIM}$	$R_{Load}=1\Omega$		450		mA
Quiescent Current	$I_G$	No Load		114		$\mu A$
Dropout Voltage <sup>(6)</sup>	$V_{DROP}$	$I_{OUT} = 100mA$		110		mV
		$I_{OUT} = 200mA$		250		mV
Line Regulation <sup>(7)</sup>	$\Delta V_{LINE}$	$V_{IN}=(V_{OUT}+0.4V$ or $2.5V)$ to $6V$ , $I_{OUT}=1mA$	-0.05		+0.05	%/V
EN Input High Threshold	$V_{IH}$	$V_{IN} = 2.5V$ to $6.0V$	1.6			V
EN Input Low Threshold	$V_{IL}$	$V_{IN} = 2.5V$ to $6.0V$			0.45	V
EN Input Bias Current	$I_{SD}$	$EN = V_{IN}=6.5V$			300	nA
Shutdown Supply Current	$I_{GSD}$	$EN1 = EN2 = GND$		0.03	1	$\mu A$
Thermal Shutdown Temperature	$T_{SD}$			140		$^{\circ}C$
Thermal Shutdown Hysteresis	$\Delta T_{SD}$			10		$^{\circ}C$
Output Voltage Noise		10Hz to 100kHz, $C_{OUT}=4.7\mu F$ , $I_{LOAD}=1mA$		11		$\mu V_{RMS}$
Output Voltage AC PSRR		100Hz, $C_{OUT} = 2.2\mu F$ , $I_{LOAD} = 100mA$		72		dB
		1kHz, $C_{OUT} = 2.2\mu F$ , $I_{LOAD} = 100mA$		72		dB
		100kHz, $C_{OUT} = 2.2\mu F$ , $I_{LOAD} = 100mA$		47		dB

### Notes:

5) Load Regulation = 
$$\frac{V_{OUT[I_{OUT(MAX)}]} - V_{OUT[I_{OUT(MIN)}]}}{V_{OUT(NOM)}} \times (\%)$$

6) Dropout Voltage is defined as the input to output differential when the output voltage drops 100mV below its nominal value.

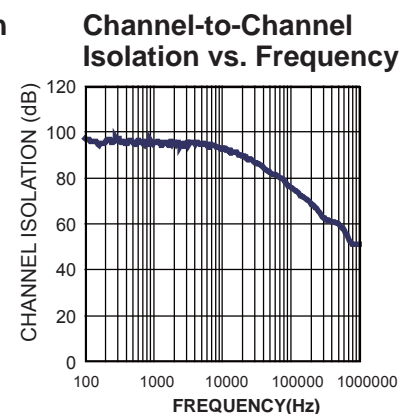
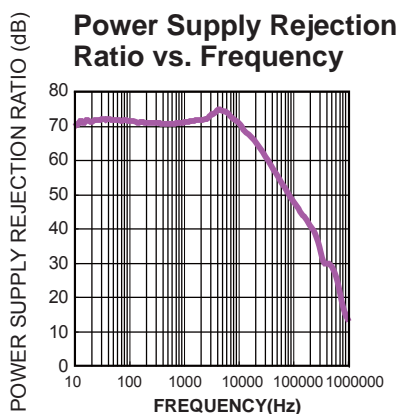
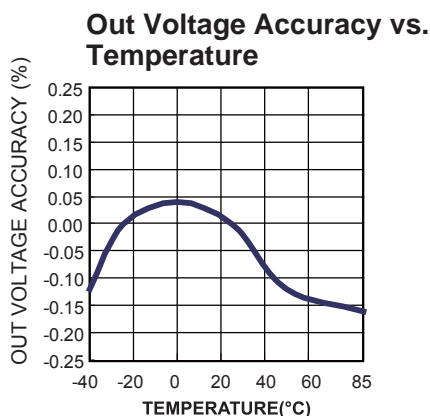
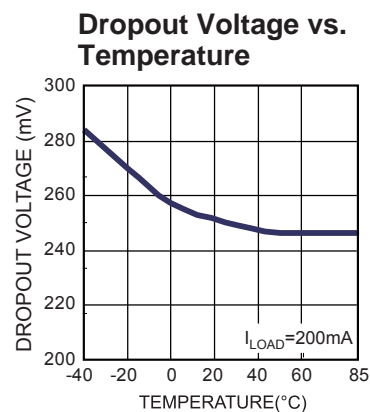
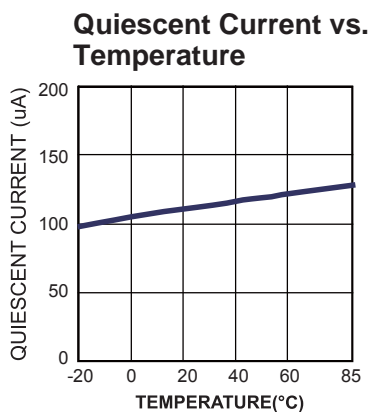
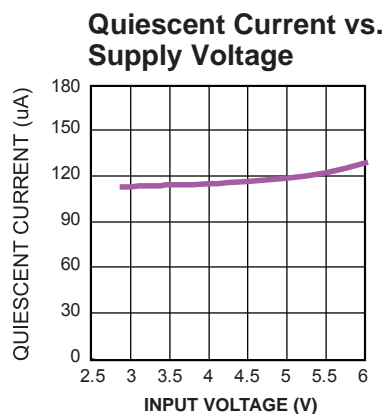
7) Line Regulation = 
$$\frac{V_{OUT[V_{IN(MAX)}]} - V_{OUT[V_{IN(MIN)}]}}{[V_{IN(MAX)} - V_{IN(MIN)}] \times V_{OUT(NOM)}} \times (\% / V)$$

## PIN FUNCTIONS

Pin #	Name	Description
1	VOUT1	Channel 1 Output Voltage
2	VIN	Supply Input Pin
3	EN1	Channel 1 Enable (Active High). Do Not Float This Pin.
4	GND	Common Ground
5	EN2	Channel 2 Enable (Active High). Do Not Float This Pin.
6, 7	NC	
8	VOUT2	Channel 2 Output Voltage

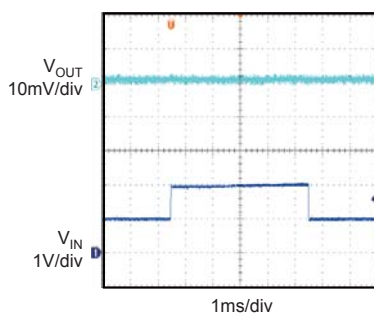
## TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN}=3.6V$ ,  $V_{OUT1}=2.5V$ ,  $V_{OUT2}=1.8V$ ,  $C_{IN}=C_{OUT1}=C_{OUT2}=2.2\mu F$ ,  $EN1=EN2=V_{IN}$ , Typical Value at  $T_A = 25^\circ C$  for Both Channel Enabled.



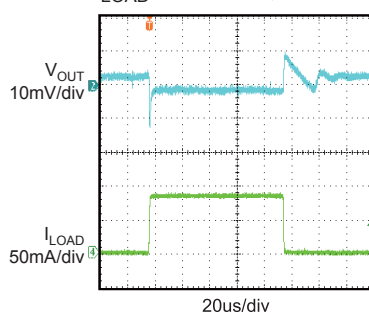
### Line Transient Response

$V_{OUT}=3.5V$  to  $4.5V$ ,  $C_{IN}=0\mu F$ ,  $C_{OUT}=2.2\mu F$   
No Load



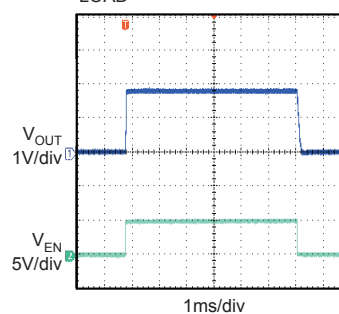
### Load Transient Response

$V_{IN}=3.6V$ ,  $V_{OUT}=1.8V$   
 $I_{LOAD}=0$  to  $80mA$ , with Resistor Load



### EN Pin Shut Down Response

$V_{IN}=3.6V$ ,  $V_{OUT}=1.8V$ ,  $V_{EN}=0$  to  $5V$   
 $I_{LOAD}=50mA$ , with Resistor Load



The schematic diagram illustrates a two-channel precision current source. It features two input terminals, VIN and EN1, and two output terminals, VOUT1 and EN2. The circuit is divided into two symmetrical channels. Each channel includes a current limit block, a thermal protection block, and a current limit block. The output of each channel is connected to a load resistor and a diode. The output voltage is VOUT1 and VOUT2. The input voltage is VIN. The output current is EN1 and EN2. The circuit also includes a reference voltage source VREF and a bias source. The output current is limited by the current limit blocks and the thermal protection block. The output voltage is determined by the load resistor and the diode. The input voltage is VIN. The output current is EN1 and EN2.

### Figure1—Functional Block Diagram

The MP20042 integrates two low noise, low dropout, low quiescent current and high PSRR linear regulators. It is intended for use in devices that require very low voltage, low quiescent current power such as wireless LAN, battery-powered equipment and hand-held equipment. The MP20042 uses internal PMOSs as the pass elements and features internal thermal shutdown and an internal current limit circuit.

Dropout voltage is the minimum input to output differential voltage required for the regulator to maintain an output voltage within 100mV of its nominal value. Because the PMOS pass element behaves as a low-value resistor, the dropout voltage of MP20042 is very low.

The MP20042 can be switched ON or OFF by a logic input at the EN pin. A high voltage at this pin will turn the device on. When the EN pin is low, the regulator output is off. The EN pin should be tied to VIN to keep the regulator output always on if the application does not require the shutdown feature. Do not float the EN pin.

The MP20042 includes two independent current limit structures which monitor and control each PMOS's gate voltage limiting the guaranteed maximum output current to 200mA.

Thermal protection turns off the PMOS when the junction temperature exceeds +140°C, allowing the IC to cool. When the IC's junction temperature drops by 10°C, the PMOS will be turned on again. Thermal protection limits total power dissipation in the MP20042. For reliable operation, junction temperature should be limited to 125 °C maximum.

The output response of load-transient consists of a DC shift and transient response. Because of the excellent load regulation of MP20042, the DC shift is very small. The output voltage transient depends on the output capacitor's value and the ESR. Increasing the capacitance and decreasing the ESR will improve the transient response. Typical output voltage transient spike of MP20042 for a step change in the load current from 0mA to 80mA is tens mV.

## APPLICATION INFORMATION

### Power Dissipation

The power dissipation for any package depends on the thermal resistance of the case and circuit board, the temperature difference between the junction and ambient air, and the rate of airflow. The power dissipation across the device can be represented by the equation:

$$P = (V_{IN} - V_{OUT}) \times I_{OUT}$$

The allowable power dissipation can be calculated using the following equation:

$$P_{(MAX)} = (T_{Junction} - T_{Ambient}) / \theta_{JA}$$

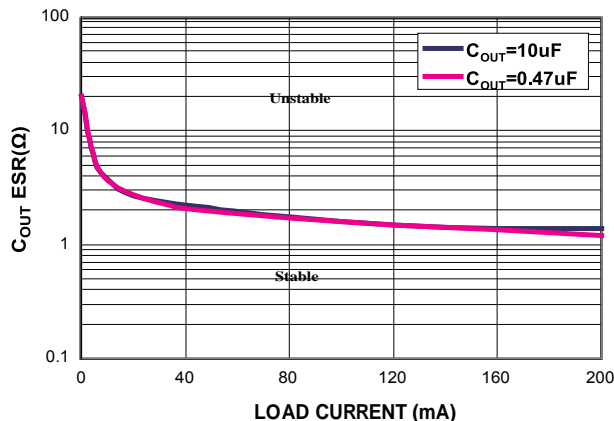
Where  $(T_{Junction} - T_{Ambient})$  is the temperature difference between the junction and the surrounding environment,  $\theta_{JA}$  is the thermal resistance from the junction to the ambient environment. Connect the GND pin of MP20042 to ground using a large pad or ground plane helps to channel heat away.

### Input Capacitor Selection

Using a capacitor whose value is  $>0.47\mu F$  on the MP20042 input and the amount of capacitance can be increased without limit. Larger values will help improve line transient response with the drawback of increased size. Ceramic capacitors are preferred, but tantalum capacitors may also suffice.

### Output Capacitor Selection

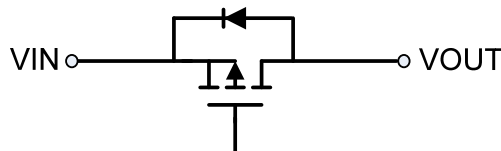
The MP20042 is designed specifically to work with very low ESR ceramic output capacitor in space-saving and performance consideration. A ceramic capacitor in the range of  $0.47\mu F$  and  $10\mu F$ , and with ESR lower than  $1.2\Omega$  is suitable for the MP20042 application circuit. Output capacitor of larger values will help to improve load transient response and reduce noise with the drawback of increased size.



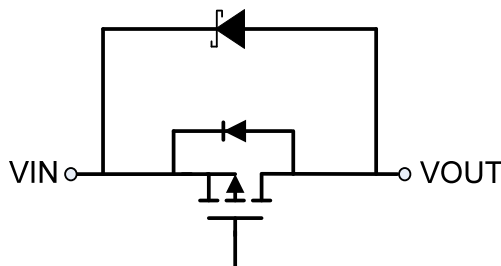
**Figure 2—Relationship between ESR and LDO Stability**

### Reverse Current Path

The PMOS used in the MP20042 has an inherent diode connected between input and output (see Figure 3). If  $V_{OUT} - V_{IN}$  is more than a diode-drop, this diode gets forward biased and starts to conduct. To avoid misoperation, an external Schottky connected in parallel with the internal parasitic diode prevents it from being turned on by limiting the voltage drop across it to about 0.3V (see Figure 4).



**Figure 3—Inherent Diode Connected between Each Regulator Input and Output**

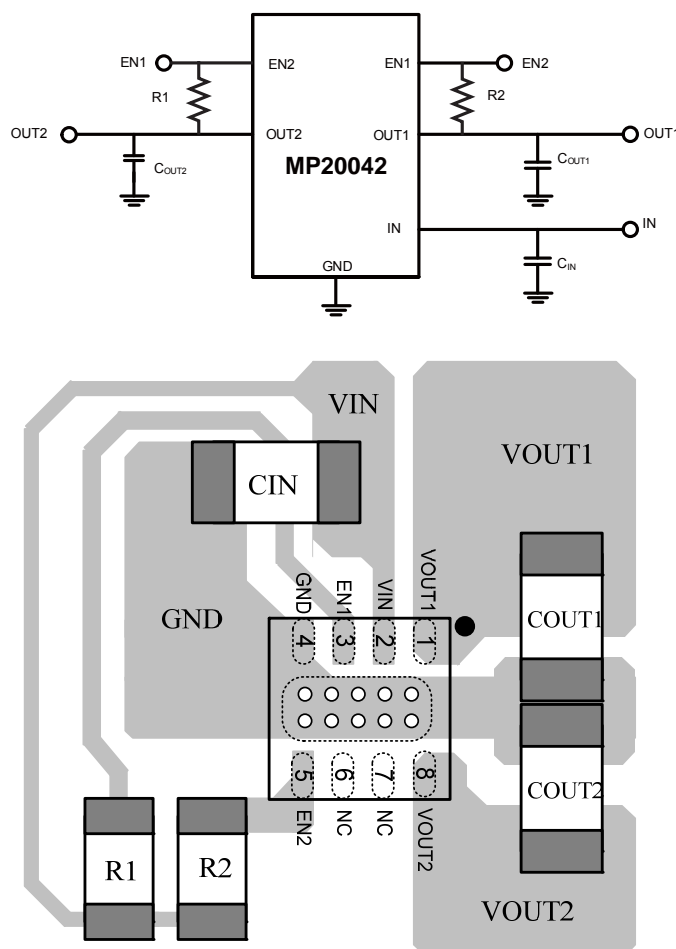


**Figure 4—External Schottky Diode Connected in Parallel with the Internal Parasitic Diode**

### PCB layout guide

PCB layout is very important to achieve good regulation, ripple rejection, transient response and thermal performance. It is highly recommended to duplicate EVB layout for optimum performance. If change is necessary, please follow these guidelines and take figure ? for reference.

- 1) Input and output bypass ceramic capacitors are suggested to be put close to the IN Pin and OUT Pin respectively.
- 2) Ensure all feedback connections are short and direct. Place the feedback resistors and compensation components as close to the chip as possible.
- 3) Connect IN, OUT and especially GND respectively to a large copper area to cool the chip to improve thermal performance and long-term reliability.

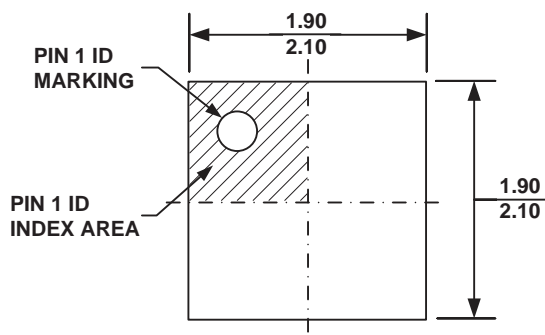


Top Layer  
Figure 5—PCB Layout

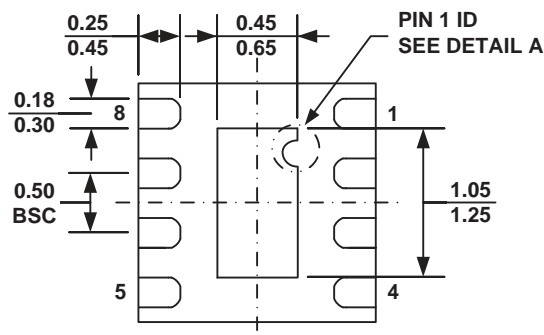


# PACKAGE INFORMATION

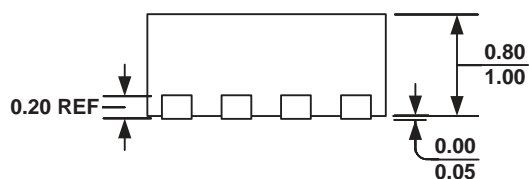
## 2mm x 2mm QFN8



**TOP VIEW**



**BOTTOM VIEW**

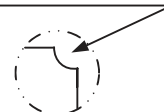


**SIDE VIEW**

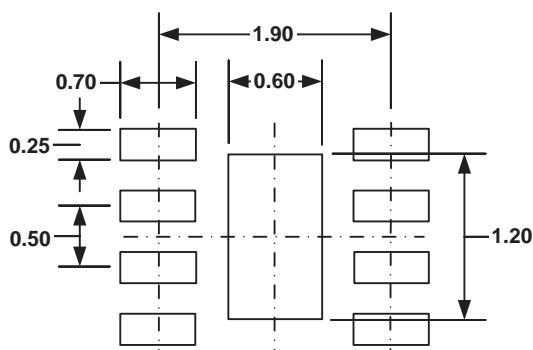
**PIN 1 ID OPTION A**  
R0.20 TYP.



**PIN 1 ID OPTION B**  
R0.20 TYP.



**DETAIL A**



**RECOMMENDED LAND PATTERN**

### NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETER MAX.
- 4) DRAWING CONFORMS TO JEDEC MO-229, VARIATION VCCD-3.
- 5) DRAWING IS NOT TO SCALE.

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