

Dual and Quad, 1.8V, 13 μ A Precision Rail-to-Rail Op Amps

FEATURES

- **Ideal for Battery-Powered Applications**
 - **Low Voltage: 1.8V Operation**
 - **Low Current: 16 μ A/Amplifier Max**
 - **Small Packages: DFN, MSOP, SSOP**
 - **Shutdown to 1.5 μ A Max (LT6001DD)**
- **Low Offset Voltage: 500 μ V Max**
- **Rail-to-Rail Input and Output**
- **Fully Specified on 1.8V and 5V Supplies**
- **Operating Temperature Range: –40°C to 85°C**
- **Available in 10-Lead and 16-Lead DFN, 8-Lead MSOP and 16-Lead SSOP Packages**

APPLICATIONS

- Gas Sensing
- Portable Instrumentation
- Battery- or Solar-Powered Systems
- Low Voltage Signal Processing
- Micropower Active Filters

DESCRIPTION

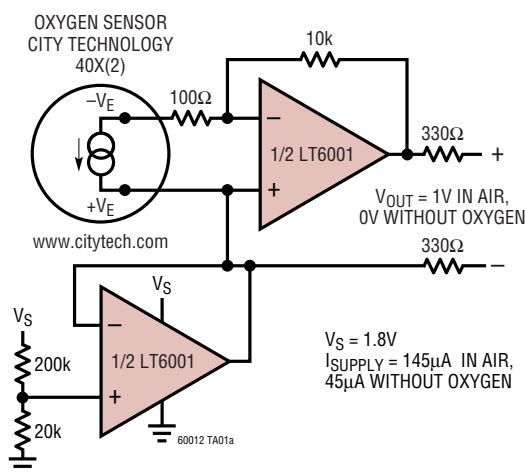
The LT[®]6001/LT6002 are dual and quad precision rail-to-rail input and output operational amplifiers. Designed to maximize battery life in always-on applications, the devices will operate on supplies down to 1.8V while drawing only 13 μ A quiescent current. The low supply current and low voltage operation is combined with precision specifications; input offset is guaranteed less than 500 μ V. The performance on 1.8V supplies is fully specified and guaranteed over temperature. A shutdown feature in the 10-lead dual version can be used to extend battery life by allowing the amplifiers to be switched off during periods of inactivity.

The LT6001 is available in the 8-Pin MSOP package; a 10-lead version with the shutdown feature is available in a tiny, dual fine pitch leadless package (DFN). The quad LT6002 is available in the 16-pin SSOP package and the 16-pin DFN package. These devices are specified over the commercial and industrial temperature range.

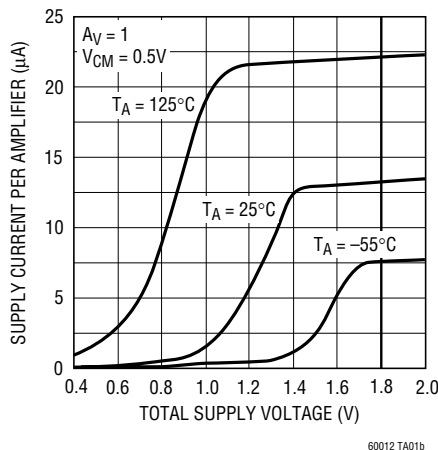
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TYPICAL APPLICATION

Micropower Oxygen Sensor



Start-Up Characteristics
Supply Current vs Supply Voltage



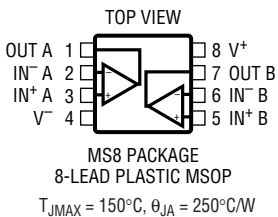
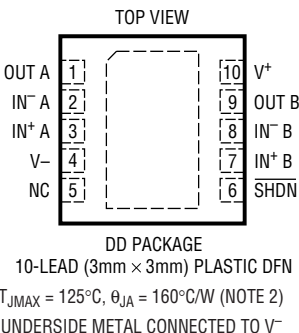
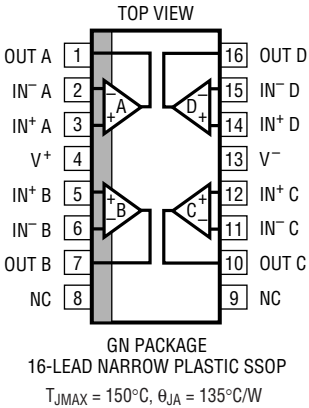
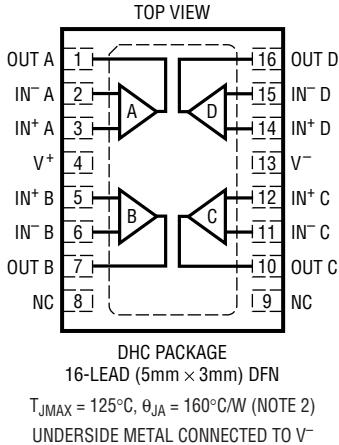
LT6001/LT6002

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-) 18V
 Input Current $\pm 10\text{mA}$
 SHDN Pin Voltage (Note 7) V^- to V^+
 Output Short Current Duration (Note 2) Indefinite
 Operating Temperature Range (Note 3) ... -40°C to 85°C
 Specified Temperature Range (Note 4) -40°C to 85°C
 Junction Temperature 150°C

Junction Temperature (DFN Packages) 125°C
 Storage Temperature Range -65°C to 150°C
 Storage Temperature Range
 DFN Packages -65°C to 125°C
 Lead Temperature (Soldering, 10 sec)
 MSOP, SSOP Packages 300°C

PACKAGE/ORDER INFORMATION

 MS8 PACKAGE 8-LEAD PLASTIC MSOP $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 250^\circ\text{C/W}$		 DD PACKAGE 10-LEAD (3mm x 3mm) PLASTIC DFN $T_{JMAX} = 125^\circ\text{C}$, $\theta_{JA} = 160^\circ\text{C/W}$ (NOTE 2) UNDERSIDE METAL CONNECTED TO V^-	
ORDER PART NUMBER	MS8 PART MARKING*	ORDER PART NUMBER	DD PART MARKING*
LT6001CMS8 LT6001IMS8	LTBVD LTBVD	LT6001CDD LT6001IDD	LBVH LBVH
 GN PACKAGE 16-LEAD NARROW PLASTIC SSOP $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 135^\circ\text{C/W}$		 DHC PACKAGE 16-LEAD (5mm x 3mm) DFN $T_{JMAX} = 125^\circ\text{C}$, $\theta_{JA} = 160^\circ\text{C/W}$ (NOTE 2) UNDERSIDE METAL CONNECTED TO V^-	
ORDER PART NUMBER	GN PART MARKING	ORDER PART NUMBER	DHC PART MARKING*
LT6002CGN LT6002IGN	6002 6002I	LT6002CDHC LT6002IDHC	6002 6002
Order Options Tape and Reel: Add #TR Lead Free: Add #PBF Lead Free Tape and Reel: Add #TRPBF Lead Free Part Marking: http://www.linear.com/leadfree/			

*Temperature grades are identified on the shipping container. Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full specified temperature range, otherwise specifications are $T_A = 25^\circ\text{C}$. $V_S = 1.8\text{V}$, 0V , $V_{CM} = V_{OUT} = 0.5\text{V}$, for the LT6001DD, $V_{SHDN} = V^+$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6001MS8, LT6002GN $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	200	500	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		700	μV
					850	μV
		LT6001DD, LT6002DHC $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	250	650	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		850	μV
		$V_{CM} = V^-$ to V^+ LT6001MS8, LT6002GN	●	400	1000	μV
					1300	μV
		$V_{CM} = V^-$ to V^+ LT6001DD, LT6002DHC	●	500	1200	μV
					1500	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 5)	$V_{CM} = 0.5\text{V}$	●	2	5	$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current	$V_{CM} = 0.5\text{V}$	●	-5	-2	nA
		$V_{CM} = V^-$	●	-5	-2	nA
		$V_{CM} = V^+$	●		4	10
I_{OS}	Input Offset Current	$V_{CM} = 0.5\text{V}$	●	0.2	1	nA
		$V_{CM} = V^-$	●	0.2	1	nA
		$V_{CM} = V^+$	●	0.4	2	nA
	Input Noise Voltage	0.1Hz to 10Hz		1.2		μV_{P-P}
e_n	Input Voltage Noise Density	$f = 1\text{kHz}$		75		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Current Noise Density	$f = 1\text{kHz}$		25		$\text{fA}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Common Mode ($V_{CM} = 0\text{V}$ to 0.6V)		3.5		$\text{G}\Omega$
		Differential		10	25	$\text{M}\Omega$
C_{IN}	Input Capacitance			5		pF
CMRR	Common Mode Rejection Ratio	$V_{CM} = 0\text{V}$ to 0.6V , $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	82	96	dB
		$V_{CM} = 0.1\text{V}$ to 0.6V , $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●	82	96	dB
		$V_{CM} = 0\text{V}$ to 1.8V	●	62	78	dB
	Input Voltage Range		●	0	1.8	V
PSRR	Power Supply Rejection Ratio	$V_S = 1.8\text{V}$ to 6V $V_{CM} = V_O = 0.5\text{V}$	●	80	90	dB
	Minimum Supply	$V_{CM} = V_O = 0.5\text{V}$	●	1.8		V

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full specified temperature range, otherwise specifications are $T_A = 25^\circ\text{C}$. $V_S = 1.8\text{V}$, 0V , $V_{CM} = V_{OUT} = 0.5\text{V}$, for the LT6001DD, $V_{SHDN} = V^+$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
A_{VOL}	Large-Signal Gain	$V_O = 0.25\text{V to } 1.25\text{V}$				
		$R_L = 100\text{k to GND}$	25	65		V/mV
		$R_L = 100\text{k to GND}$	20			V/mV
		$R_L = 10\text{k to GND}$	40	125		V/mV
V_{OL}	Output Swing Low (Note 6)	$R_L = 10\text{k to GND}$	25			V/mV
		Input Overdrive = 30mV				
		No Load		30	60	mV
		$I_{SINK} = 100\mu\text{A}$		120	200	mV
V_{OH}	Output Swing High (Note 6)	Input Overdrive = 30mV				
		No Load		30	60	mV
		$I_{SOURCE} = 100\mu\text{A}$		140	225	mV
		$R_L = 10\text{k to GND}$		160	250	mV
I_{SC}	Short-Circuit Current	Short to GND	2	4		mA
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	1			mA
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	0.4			mA
		Short to V^+	0.7	2		mA
I_S	Supply Current per Amplifier	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$				mA
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$				mA
				13	16	μA
					22	μA
I_{SHDN}	Total Supply Current in Shutdown (Note 7)	$V_{SHDN} = 0.3\text{V}$				μA
				0.8	1.5	μA
		$V_{SHDN} = 1.8\text{V}$		0	20	nA
		$V_{SHDN} = 0\text{V}$	-300	-200		nA
V_L	Shutdown Output Leakage Current (Note 7)	$V_{SHDN} = 0.3\text{V}$ ($V^- \leq V_{OUT} \leq V^+$)		20		nA
V_L	SHDN Pin Input Low Voltage (Note 7)				0.3	V
V_H	SHDN Pin Input High Voltage (Note 7)		1.5V			V
t_{ON}	Turn On Time (Note 7)	$V_{SHDN} = 0\text{V to } 1.8\text{V}$, $R_L = 10\text{k}$		400		μs
t_{OFF}	Turn Off Time (Note 7)	$V_{SHDN} = 1.8\text{V to } 0\text{V}$, $R_L = 10\text{k}$		100		μs
GBW	Gain Bandwidth Product (Note 8)	Freq = 1kHz	32	50		kHz
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	28			kHz
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	24			kHz
SR	Slew Rate	$A_V = -1$, $V_{OUT} = 0.25\text{V to } 1.5\text{V}$	9	15		V/ms
		Measure 0.5V to 1.25V, $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	7			V/ms
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	5			V/ms
FPBW	Full Power Bandwidth (Note 9)	$V_{OUT} = 1.25V_{P-P}$	2.3	3.8		kHz

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full specified temperature range, otherwise specifications are $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V , $V_{CM} = V_{OUT} = 1/2$ Supply, for the LT6001DD, $V_{SHDN} = V^+$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6001MS8, LT6002GN $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	200	500	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		700	μV
					850	μV
		LT6001DD, LT6002DHC $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	250	650	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		850	μV
					1000	μV
		$V_{CM} = V^-$ to V^+ LT6001MS8, LT6002GN	●	400	1000	μV
					1300	μV
		$V_{CM} = V^-$ to V^+ LT6001DD, LT6002DHC	●	500	1200	μV
					1500	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 5)	$V_{CM} = V_S/2$	●	2	5	$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current	$V_{CM} = V_S/2$	●	-6	-2	nA
		$V_{CM} = V^-$	●	-6	-2	nA
		$V_{CM} = V^+$	●		4	12
I_{OS}	Input Offset Current	$V_{CM} = V_S/2$	●	0.2	1.2	nA
		$V_{CM} = V^-$	●	0.2	1.2	nA
		$V_{CM} = V^+$	●	0.4	2.4	nA
	Input Noise Voltage	0.1Hz to 10Hz		1.2		μV_{P-P}
e_n	Input Voltage Noise Density	$f = 1\text{kHz}$		75		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Current Noise Density	$f = 1\text{kHz}$		25		$\text{fA}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Common Mode ($V_{CM} = 0\text{V}$ to 3.8V)		3.5		$\text{G}\Omega$
		Differential	●	8.5	25	$\text{M}\Omega$
C_{IN}	Input Capacitance			5		pF
CMRR	Common Mode Rejection Ratio	$V_{CM} = 0\text{V}$ to 3.8V , $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	90	105	dB
		$V_{CM} = 0.1\text{V}$ to 3.8V , $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●	90	105	dB
		$V_{CM} = 0\text{V}$ to 5V	●	70	86	dB
	Input Voltage Range		●	0	5	V
PSRR	Power Supply Rejection Ratio	$V_S = 1.8\text{V}$ to 6V $V_{CM} = V_O = 0.5\text{V}$	●	80	90	dB
	Minimum Supply		●	1.8		V
A_{VOL}	Large-Signal Gain	$V_O = 0.5\text{V}$ to 4.5V				
		$R_L = 100\text{k}$ to $V_S/2$	●	30	60	V/mV
		$R_L = 100\text{k}$ to $V_S/2$	●	25		V/mV
		$R_L = 10\text{k}$ to $V_S/2$	●	16	25	V/mV
		$R_L = 10\text{k}$ to $V_S/2$	●	10		V/mV
		$R_L = 10\text{k}$ to GND	●	160	1000	V/mV
		$R_L = 10\text{k}$ to GND	●	80		V/mV

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full specified temperature range, otherwise specifications are $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V , $V_{CM} = V_{OUT} = 1/2$ Supply, for the LT6001DD, $V_{SHDN} = V^+$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OL}	Output Swing Low (Note 6)	Input Overdrive = 30mV				
		No Load		30	60	mV
		$I_{SINK} = 100\mu\text{A}$		120	200	mV
		$I_{SINK} = 500\mu\text{A}$		180	300	mV
V_{OH}	Output Swing High (Note 6)	Input Overdrive = 30mV				
		No Load		30	60	mV
		$I_{SOURCE} = 100\mu\text{A}$		140	225	mV
		$R_L = 10\text{k}$ to GND		160	400	mV
I_{SC}	Short-Circuit Current	Short to GND				
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	5	10		mA
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	4			mA
			3			mA
		Short to V^+				
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	3.5	7.5		mA
I_S	Supply Current per Amplifier	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$				
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		15	18	μA
					24	μA
					27	μA
		Total Supply Current in Shutdown (Note 7)		3	5	μA
I_{SHDN}	SHDN Pin Current (Note 7)	$V_{SHDN} = 5\text{V}$		0	20	nA
		$V_{SHDN} = 0\text{V}$	-1000	-650		nA
	Shutdown Output Leakage Current (Note 7)	$V_{SHDN} = 0.3\text{V}$ ($V^- \leq V_{OUT} \leq V^+$)		20		nA
V_L	SHDN Pin Input Low Voltage (Note 7)				0.3	V
V_H	SHDN Pin Input High Voltage (Note 7)		4.7			V
t_{ON}	Turn On Time (Note 7)	$V_{SHDN} = 0\text{V}$ to 5V , $R_L = 10\text{k}$		400		μs
t_{OFF}	Turn Off Time (Note 7)	$V_{SHDN} = 5\text{V}$ to 0V , $R_L = 10\text{k}$		100		μs
GBW	Gain Bandwidth Product	Freq = 1kHz		40	60	kHz
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		35		kHz
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		30		kHz
SR	Slew Rate	$A_V = -1$, $V_{OUT} = 0.5\text{V}$ to 4.5V		11	18	V/ms
		Measure 1V to 4V, $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		8		V/ms
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		6		V/ms
FPBW	Full Power Bandwidth (Note 9)	$V_{OUT} = 4V_{P-P}$	0.87	1.4		kHz

Note 1: Absolute maximum ratings are those values beyond which the life of the device may be impaired.

Note 2: A heat sink may be required to keep the junction temperature below the absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted. The θ_{JA} specified for the DD and DHC packages is with minimal PCB heat spreading metal. Using expanded metal area on all layers of a board reduces this value.

Note 3: The LT6001C/LT6001I and LT6002C/LT6002I are guaranteed functional over the temperature range of -40°C to 85°C .

Note 4: The LT6001C/LT6002C is guaranteed to meet specified performance from 0°C to 70°C . The LT6001I/LT6002I are designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures. The LT6001I/LT6002I is guaranteed to meet specified performance from -40°C to 85°C .

Note 5: This parameter is not 100% tested.

Note 6: Output voltage swings are measured between the output and power supply rails.

Note 7: Specifications apply to the LT6001DD with shutdown.

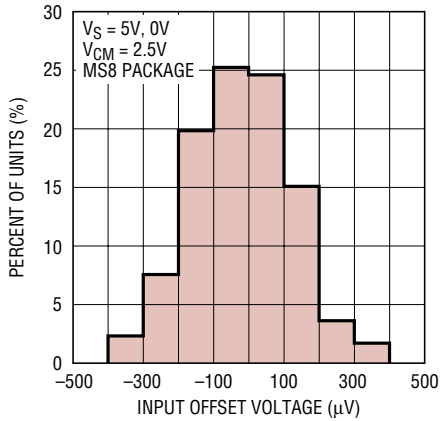
Note 8: Guaranteed by correlation to slew rate at $V_S = 1.8\text{V}$ and GBW at $V_S = 5\text{V}$.

Note 9: Full-power bandwidth is calculated from the slew rate:

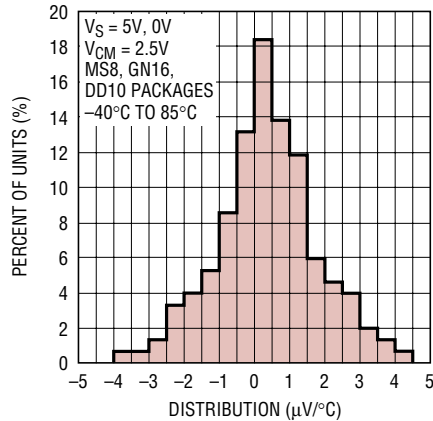
$$\text{FPBW} = \text{SR} / (2\pi V_P)$$

TYPICAL PERFORMANCE CHARACTERISTICS

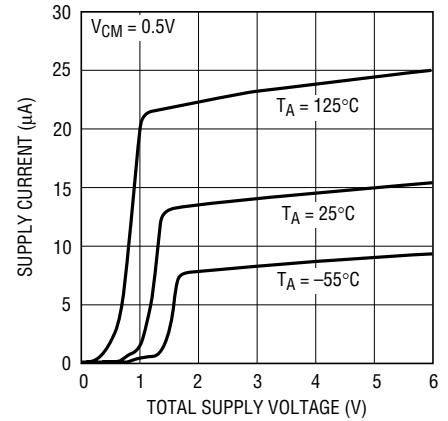
V_{OS} Distribution



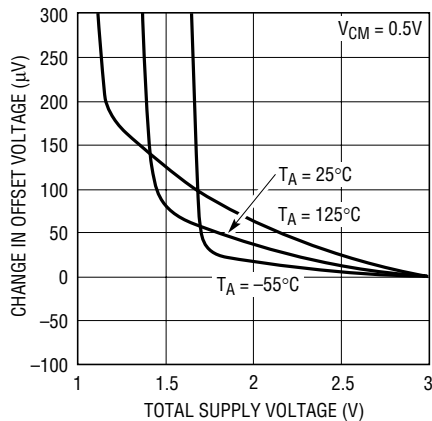
TC V_{OS} Distribution



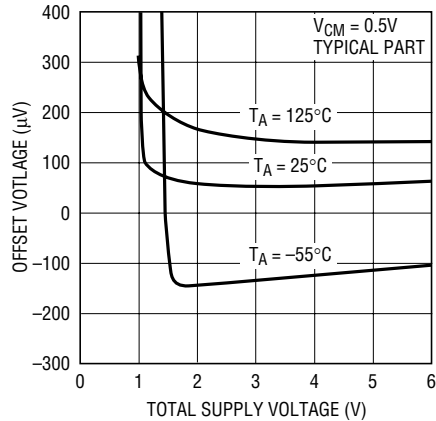
Supply Current vs Supply Voltage



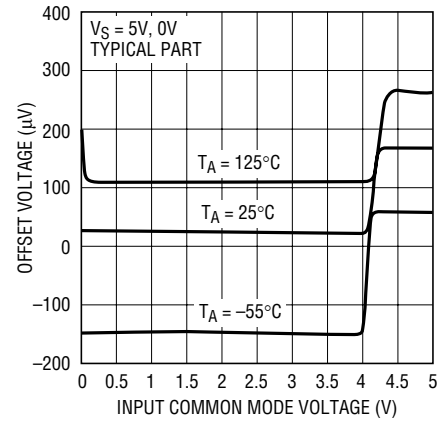
Change in Input Offset Voltage vs Total Supply Voltage



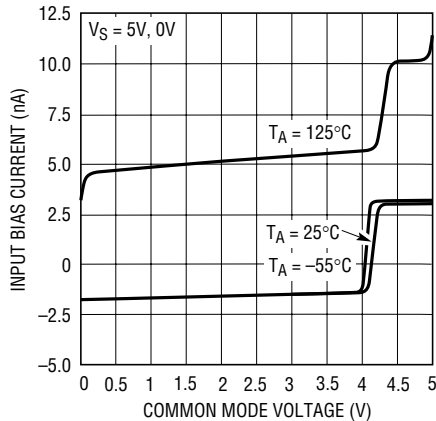
Input Offset Voltage vs Total Supply Voltage



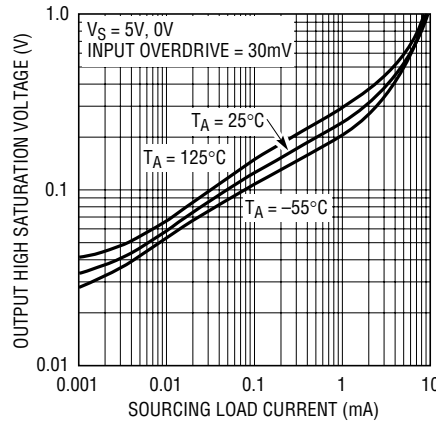
Input Offset Voltage vs Input Common Mode Voltage



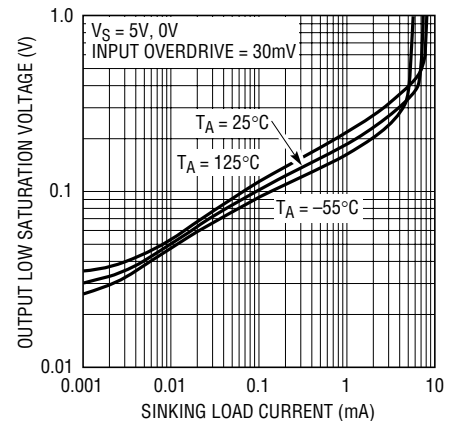
Input Bias Current vs Common Mode Voltage



Output Saturation Voltage vs Load Current (Output High)

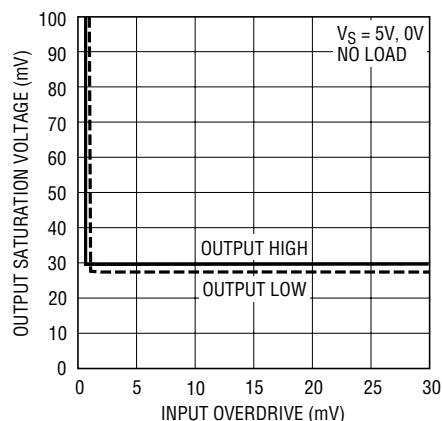


Output Saturation Voltage vs Load Current (Output Low)



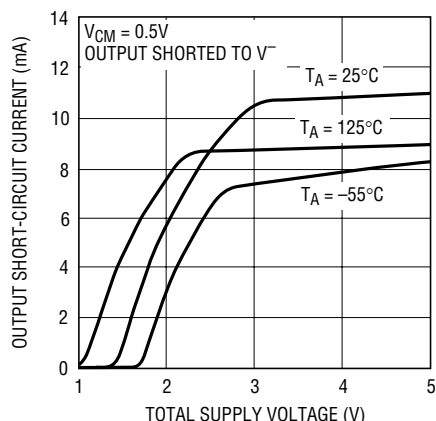
TYPICAL PERFORMANCE CHARACTERISTICS

Output Saturation Voltage vs Input Overdrive



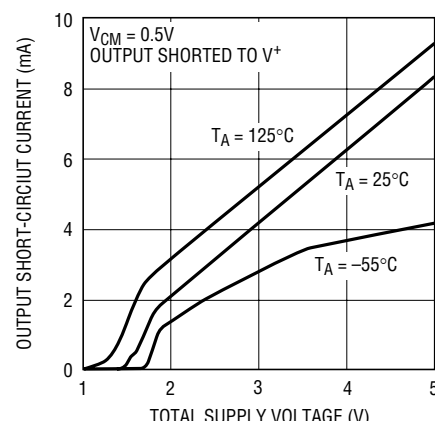
60012 G10

Output Short-Circuit Current vs Total Supply Voltage (Sourcing)



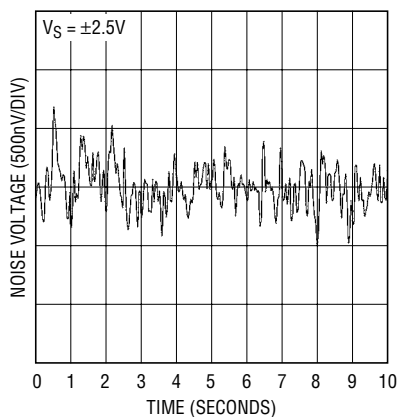
60012 G11

Output Short-Circuit Current vs Total Supply Voltage (Sinking)



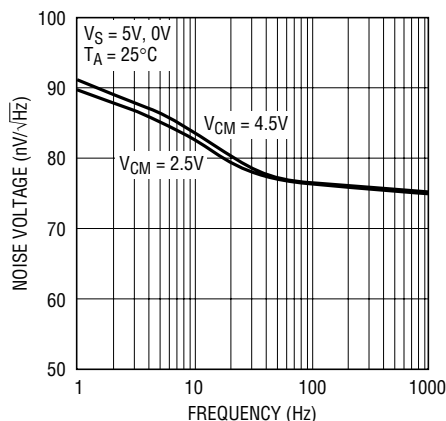
60012 G12

0.1Hz to 10Hz Output Voltage Noise



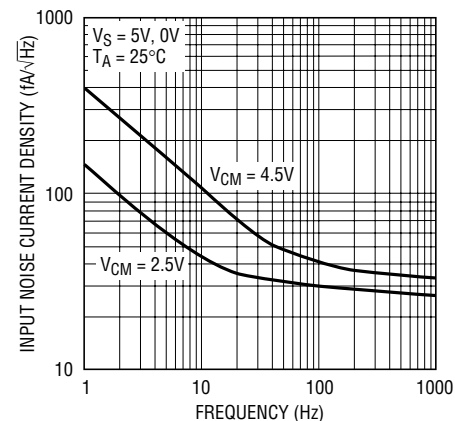
60012 G13

Noise Voltage Density vs Frequency



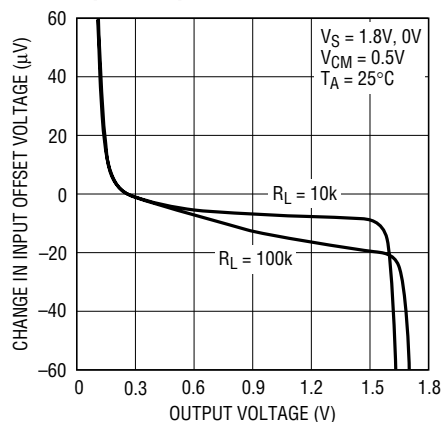
60012 G14

Input Noise Current vs Frequency



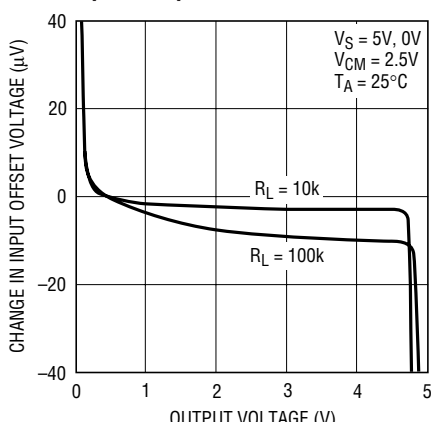
60012 G15

Open-Loop Gain



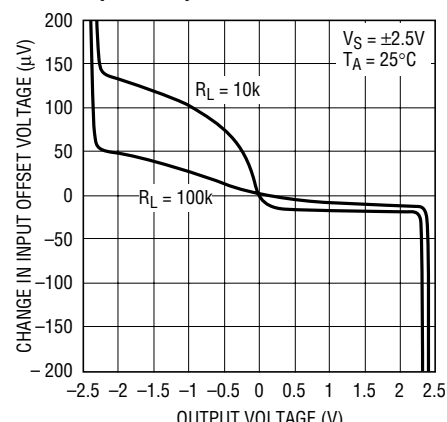
60012 G16

Open-Loop Gain



60012 G17

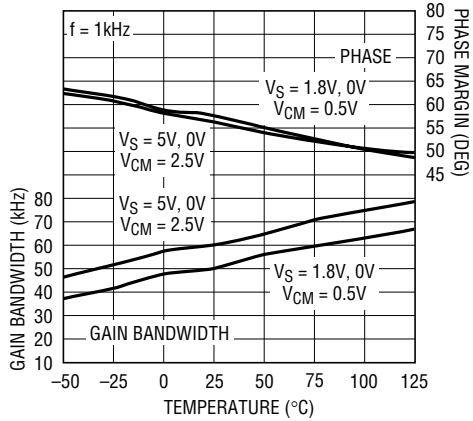
Open-Loop Gain



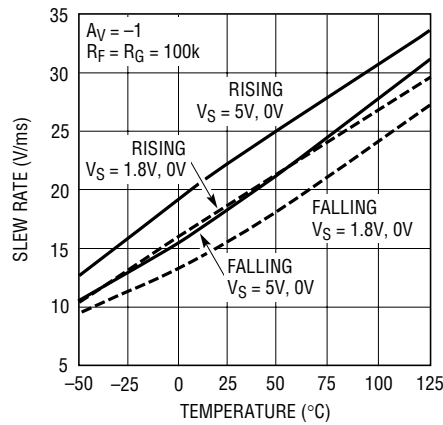
20012 G18

TYPICAL PERFORMANCE CHARACTERISTICS

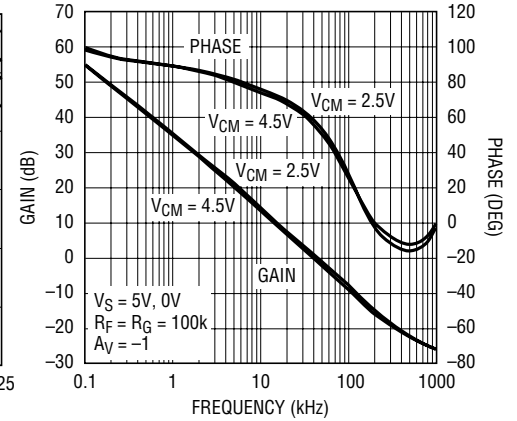
Gain Bandwidth and Phase Margin vs Temperature



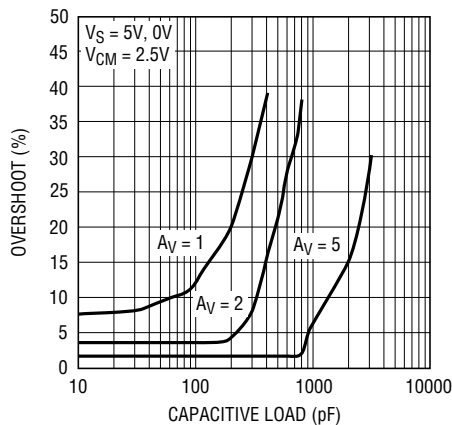
Slew Rate vs Temperature



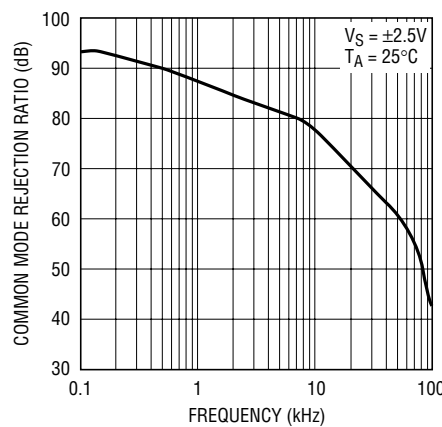
Gain and Phase vs Frequency



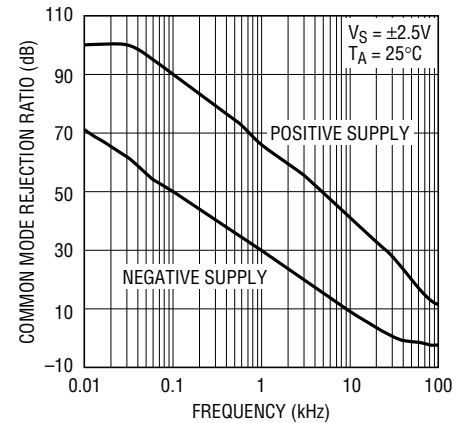
**Capacitive Load Handling
Overshoot vs Capacitive Load**



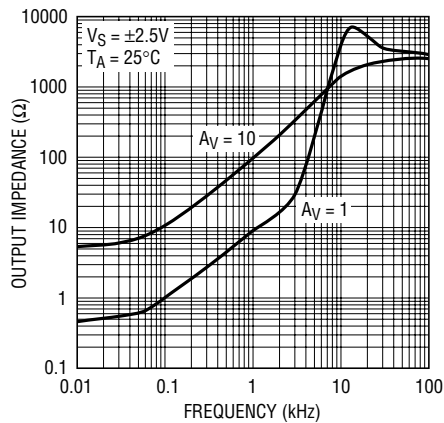
**Common Mode Rejection Ratio
vs Frequency**



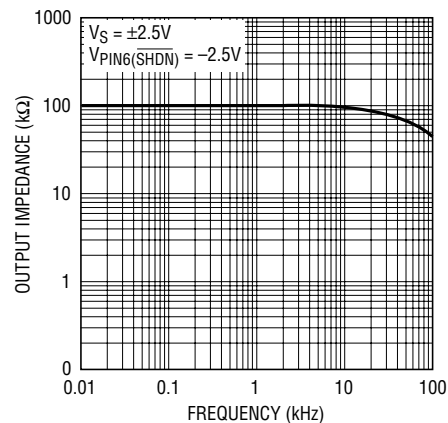
**Power Supply Rejection Ratio
vs Frequency**



Output Impedance vs Frequency

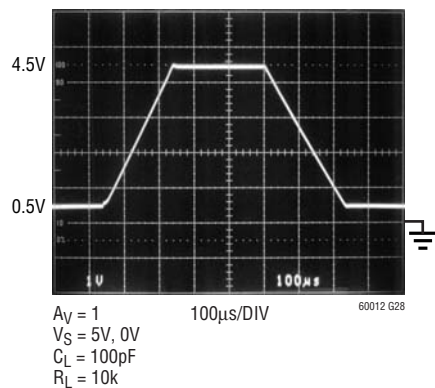


**Disabled Output Impedance
vs Frequency (LT6001DD)**

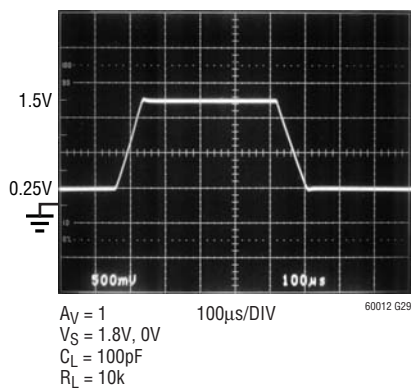


TYPICAL PERFORMANCE CHARACTERISTICS

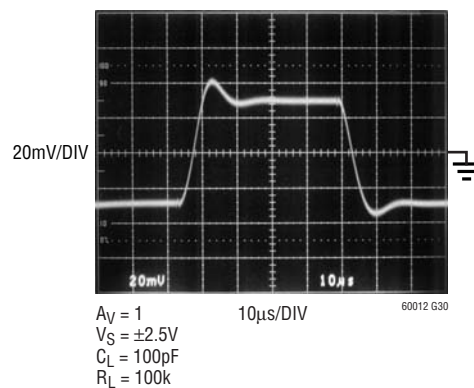
Large-Signal Response



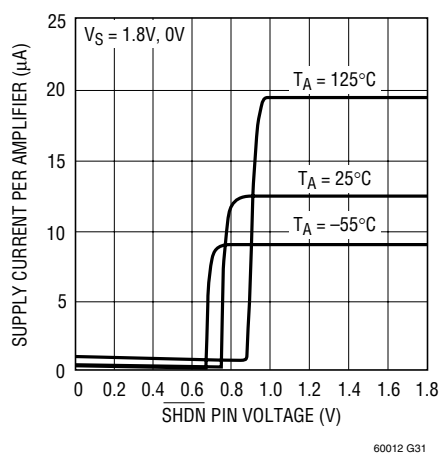
Large-Signal Response



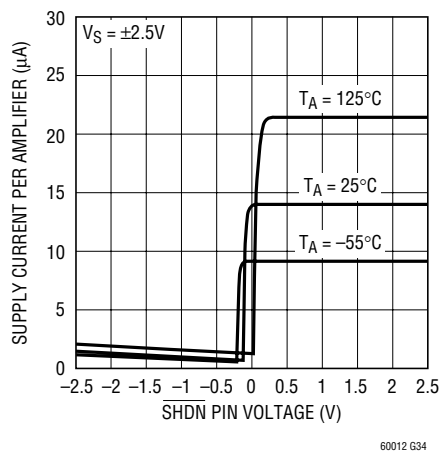
Small-Signal Response



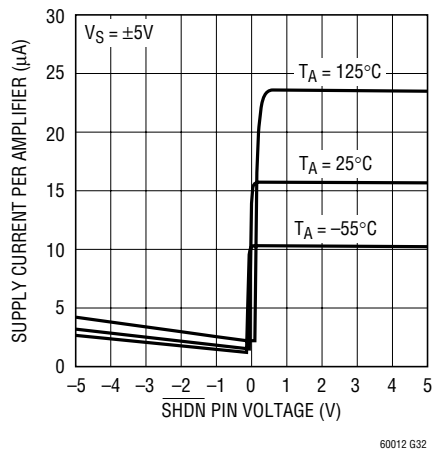
Supply Current vs SHDN Pin Voltage (LT6001DD)



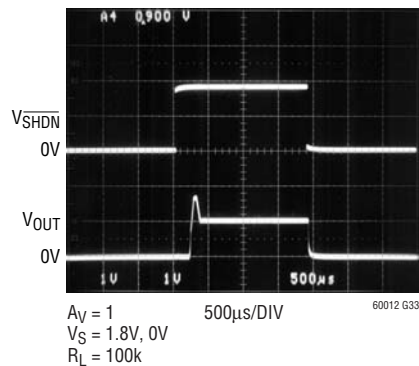
Supply Current vs SHDN Pin Voltage (LT6001DD)



Supply Current vs SHDN Pin Voltage (LT6001DD)



Shutdown Response (LT6001DD)



SIMPLIFIED SCHEMATIC

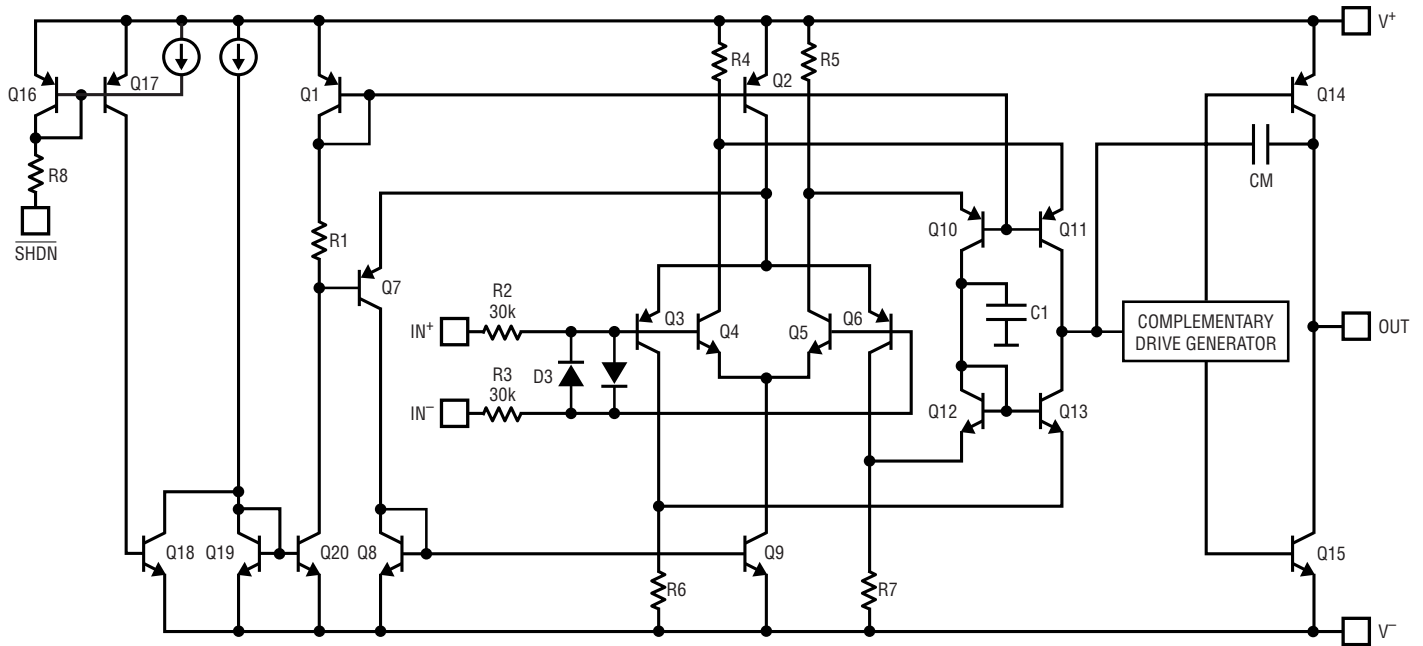


Figure 1

APPLICATIONS INFORMATION

Supply Voltage

The positive supply of the LT6001/LT6002 should be bypassed with a small capacitor (about 0.01 μ F) within an inch of the pin. When driving heavy loads, an additional 4.7 μ F electrolytic capacitor should be used. When using split supplies, the same is true for the negative supply pin.

Rail-to-Rail Characteristics

The LT6001/LT6002 are fully functional for an input signal range from the negative supply to the positive supply. Figure 1 shows a simplified schematic of the amplifier. The input stage consists of two differential amplifiers, a PNP stage Q3/Q6 and an NPN stage Q4/Q5 that are active over different ranges of the input common mode voltage. The PNP stage is active for common mode voltages, V_{CM} , between the negative supply to approximately 1V below the positive supply. As V_{CM} moves closer towards the positive supply, the transistor Q7 will steer Q2's tail current to the current mirror Q8/Q9, activating the NPN differential pair. The PNP pair becomes inactive for

the rest of the input common mode range up to the positive supply.

The second stage is a folded cascode and current mirror that converts the input stage differential signals into a single ended output. Capacitor C1 reduces the unity cross frequency and improves the frequency stability without degrading the gain bandwidth of the amplifier. The complementary drive generator supplies current to the output transistors that swing from rail to rail.

Input

The input bias current depends on which stage is active. The input bias current polarity depends on the input common mode voltage. When the PNP stage is active, the input bias currents flow out of the input pins. They flow in the opposite direction when the NPN stage is active. The offset error due to the input bias currents can be minimized by equalizing the noninverting and inverting source impedance.

APPLICATIONS INFORMATION

The input offset voltage changes depending on which input stage is active; input offset voltage is trimmed on both input stages, and is guaranteed to be 500 μ V max in the PNP stage. By trimming the input offset voltage of both input stages, the input offset voltage over the entire common mode range (CMRR) is typically 400 μ V, maintaining the precision characteristics of the amplifier.

The input stage of the LT6001/LT6002 incorporates phase reversal protection to prevent wrong polarity outputs from occurring when the inputs are driven up to 2V beyond the rails. 30k protective resistors are included in the input leads so that current does not become excessive when the inputs are forced beyond the supplies or when a large differential signal is applied.

Output

The output of the LT6001/LT6002 can swing to within 30mV of the positive rail with no load and within 30mV of the negative rail with no load. When monitoring input voltages within 30mV of the positive rail or within 30mV of the negative rail, gain should be taken to keep the output from clipping. The LT6001/LT6002 can typically source 10mA on a single 5V supply, sourcing current is reduced to 4mA on a single 1.8V supply as noted in the electrical characteristics.

The normally reverse-biased substrate diode from the output to V^- will cause unlimited currents to flow when the output is forced below V^- . If the current is transient and limited to 100mA, no damage will occur.

The LT6001/LT6002 are optimized for low voltage operation but will remain functional up to 18V total supply voltage. On a total supply voltage greater than 6V, the output may exhibit small amplitude, high frequency oscillations when sourcing more than 500 μ A into a load

impedance greater than 20k Ω . Adding a 470pF capacitor in series with a 150 Ω resistor between the output and ground will stabilize the output.

Start-Up and Output Saturation Characteristics

Micropower op amps are often not micropower during start-up characteristics or during output saturation. This can wreak havoc on limited current supplies, in the worst case there may not be enough supply current available to take the system up to nominal voltages. Also, when the output saturates, the part may draw excessive current and pull down the supplies, compromising rail-to-rail performance. Figure 1 shows the start-up characteristics of the LT6001 for three limiting cases. The circuits are shown in Figure 2. One circuit creates a positive offset forcing the output to come up saturated high. Another circuit creates a negative offset forcing the output to come up saturated low, while the last circuit brings the output up at 1/2 supply. In all cases, the supply current is well controlled and is not excessive when the output is on either rail.

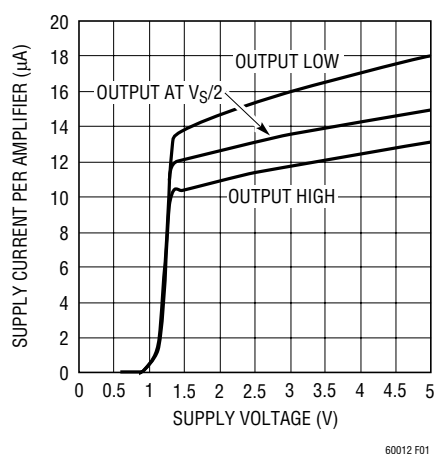


Figure 1. Start-Up Characteristics

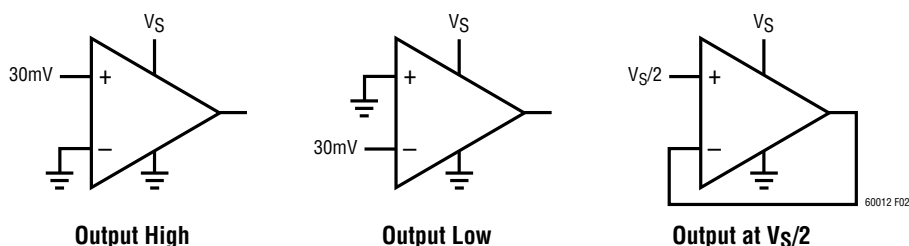


Figure 2. Circuits for Start-Up Characteristics

APPLICATIONS INFORMATION

The LT6001 can swing to a respectable 30mV within each rail and draw virtually no excessive supply current. Figure 3 compares the LT6001 to a competitive part. Both op amps are in unity gain and their outputs are driven into each rail. The supply current is shown when the op amps are in linear operation and when they are driven into each rail. As can be seen from Figure 3, the supply current of the competitive part increases 3-fold or 5-fold depending on which rail the output goes to whereas the LT6001 draws virtually no excessive current.

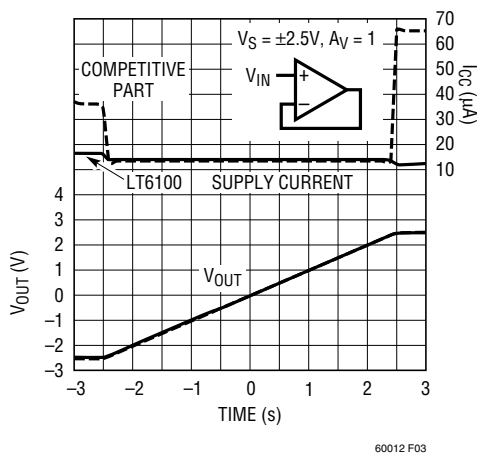


Figure 3. V_{CC} and I_{CC} vs Input Current

Gain

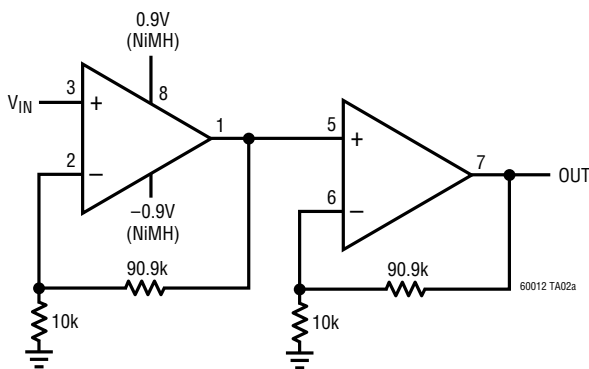
The open-loop gain is almost independent of load when the output is sourcing current. This optimizes performance in single supply applications where the load is returned to ground. The typical performance curve of Open-Loop Gain for various loads shows the details.

Shutdown

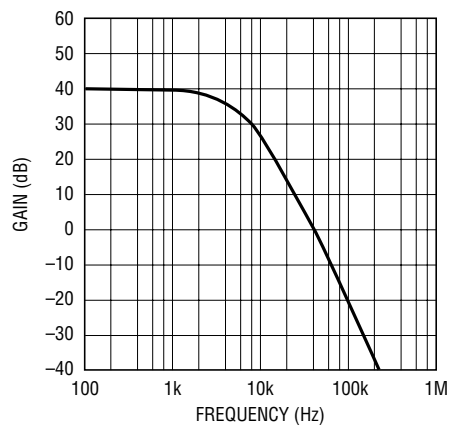
The 10-lead LT6001 includes a shutdown feature that disables the part reducing quiescent current and makes the output high impedance. The LT6001 can be shut down by bringing the $\overline{\text{SHDN}}$ pin within 0.3V of V^- , disabling both op-amps. The LT6001 is guaranteed to shut down if the $\overline{\text{SHDN}}$ pin is brought within 0.3V of V^- . The exact switchover point will be a function of the supply voltage. See the Typical Performance Characteristics curves Supply Current vs Shutdown Pin Voltage. When shut down the total supply current is about $0.8\mu\text{A}$ (both amplifiers) and the output leakage current is 20nA ($V^- \leq V_{OUT} \leq V^+$). For normal operation the $\overline{\text{SHDN}}$ pin should be tied to V^+ . It can be left floating, however, parasitic leakage currents over $1\mu\text{A}$ at the $\overline{\text{SHDN}}$ pin may inadvertently place the part into shutdown.

TYPICAL APPLICATION

Gain of 100 Amplifier
(400kHz GBW on $30\mu\text{A}$ Supply)

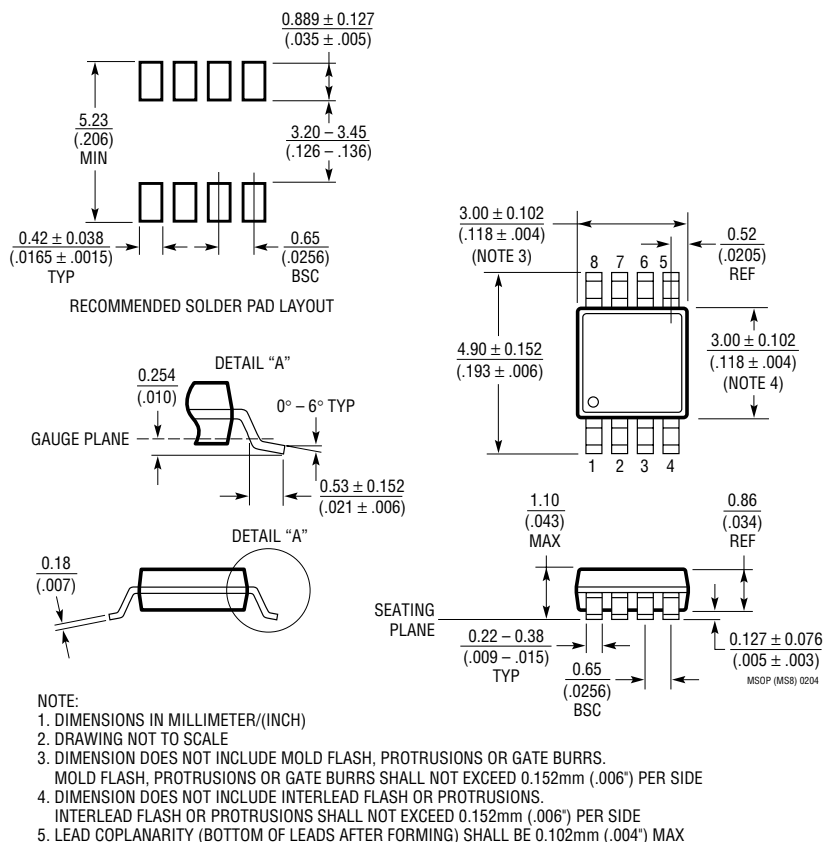


Gain vs Frequency

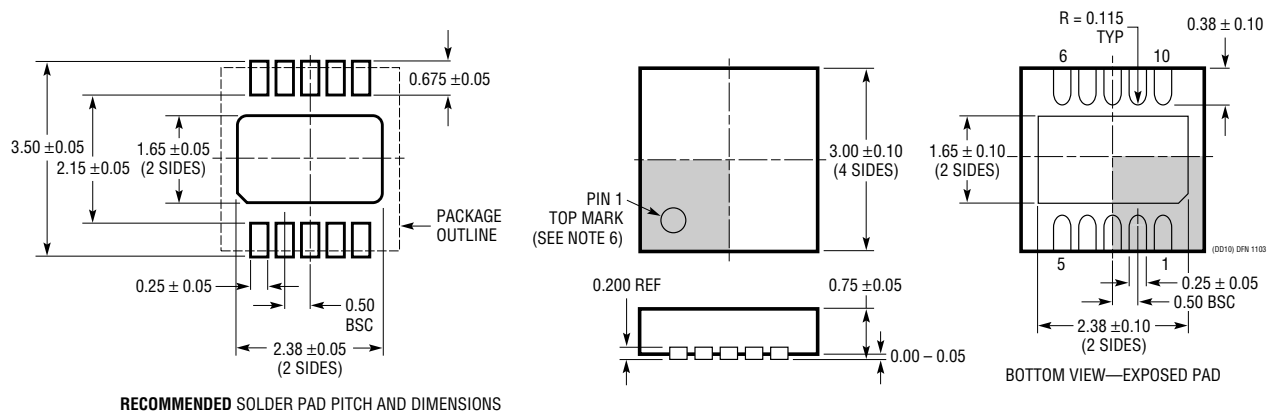


PACKAGE DESCRIPTION

MS8 Package 8-Lead Plastic MSOP (Reference LTC DWG # 05-08-1660)

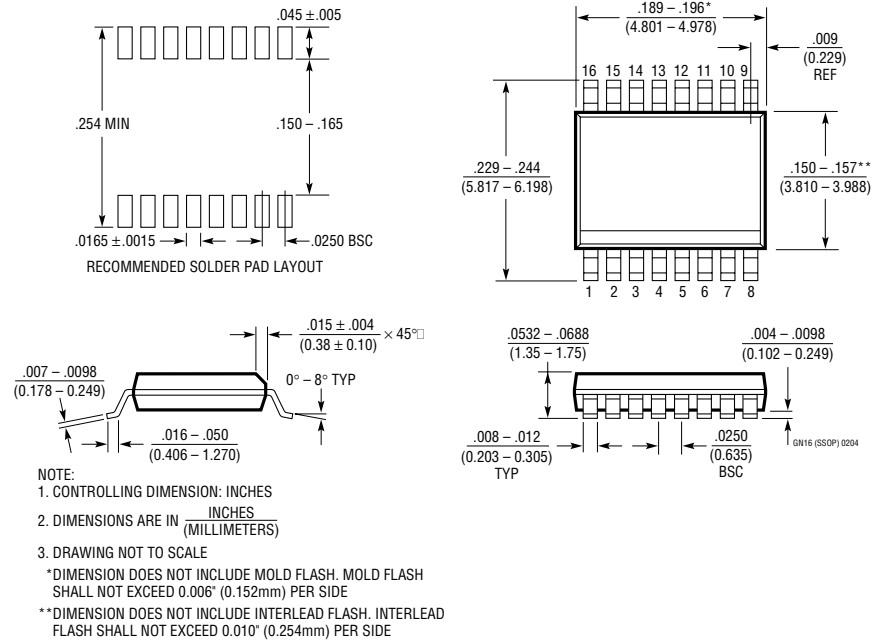


DD Package 10-Lead (3mm × 3mm) Plastic DFN (Reference LTC DWG # 05-08-1699)

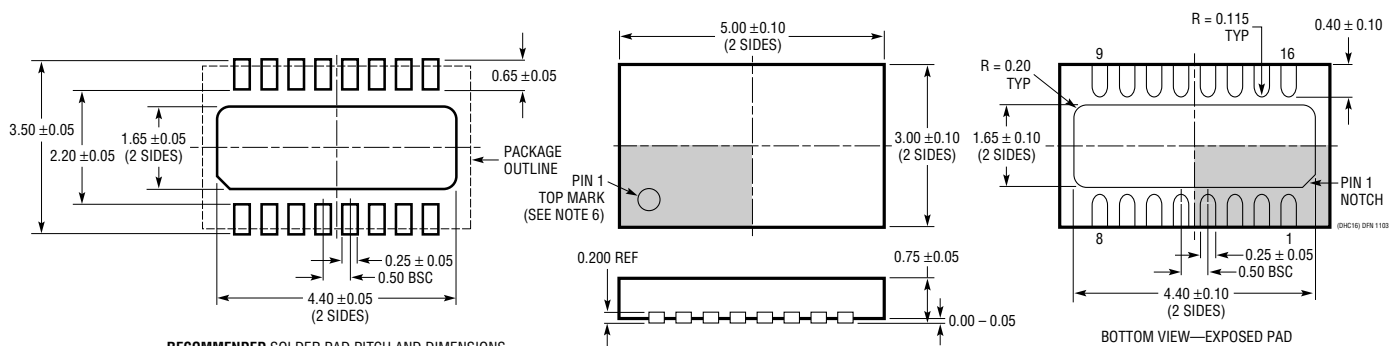


PACKAGE DESCRIPTION

GN Package 16-Lead Narrow Plastic SSOP (Reference LTC DWG # 05-08-1641)

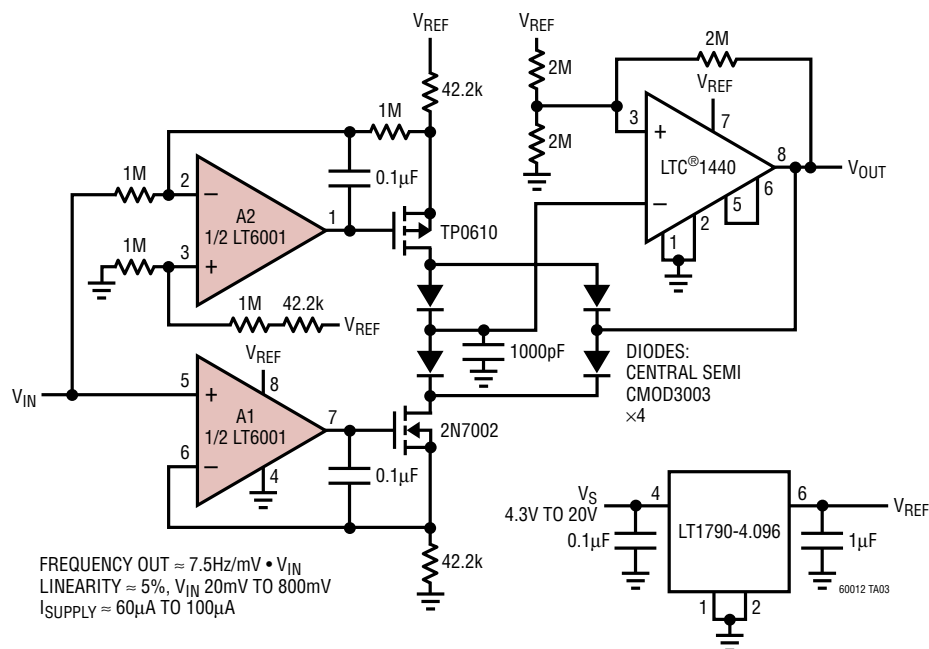


DHC Package 16-Lead (5mm × 5mm) Plastic DFN (Reference LTC DWG # 05-08-1706)



TYPICAL APPLICATION

Low Power V-to-F Converter



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT2178/LT2179	17μA Dual/Quad Single Supply Op Amps	120μV $V_{OS(MAX)}$, Gain Bandwidth = 60kHz
LT1490A/LT1491A	50μA Dual/Quad Over-The-Top® Rail-to-Rail Input and Output Op Amps	950μV $V_{OS(MAX)}$, Gain Bandwidth = 200kHz
LT1494/LT1495/LT1496	1.5μA Max Single/Dual/Quad Over-The-Top Precision Rail-to-Rail Input and Output Op Amps	375μV $V_{OS(MAX)}$, Gain Bandwidth = 2.7kHz
LT1672/LT1673/LT1674	2μA Max, $A_V \geq 5$, Single/Dual/Quad Over-The-Top Precision Rail-to-Rail Input and Output Op Amps	Gain of 5 Stable, Gain Bandwidth = 12kHz
LT1782	Micropower, Over-The-Top, SOT-23, Rail-to-Rail Input and Output Op Amps	SOT-23, 800μV $V_{OS(MAX)}$, $I_S = 55\mu A_{(MAX)}$, Gain Bandwidth = 200kHz, Shutdown Pin

Over-The-Top is a registered trademark of Linear Technology Corporation.