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**Quadruple Comparators** 



ADE-204-047 (Z) Rev. 0 Dec. 2000

#### **Description**

The HA17901 and HA17339 series products are comparators designed for use in power or control systems.

These IC operate from a single power-supply voltage over a wide range of voltages, and feature a reduced power-supply current since the power-supply voltage is determined independently.

These comparators have the unique characteristic of ground being included in the common-mode input voltage range, even when operating from a single-voltage power supply. These products have a wide range of applications, including limit comparators, simple A/D converters, pulse/square-wave/time delay generators, wide range VCO circuits, MOS clock timers, multivibrators, and high-voltage logic gates.

#### **Features**

Wide power-supply voltage range: 2 to 36V

• Extremely low current drain: 0.8mA

• Low input bias current: 25nA

• Low input offset current: 5nA

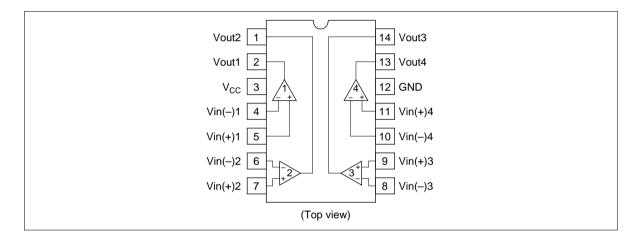
• Low input offset voltage: 2mV

- The common-mode input voltage range includes ground.
- Low output saturation voltage: 1mV (5µA), 70mV (1mA)
- Output voltages compatible with CMOS logic systems

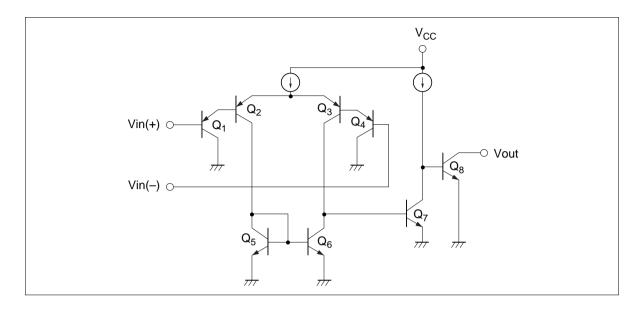
### **Ordering Information**

Type No.	Application	Package	
HA17901PJ	Car use	DP-14	
HA17901FPJ		FP-14DA	
HA17901FPK		FP-14DA	
HA17901P	Industrial use	DP-14	
HA17901FP		FP-14DA	
HA17339	Commercial use	DP-14	
HA17339F		FP-14DA	

### **Pin Arrangement**



### **Circuit Structure** (1/4)



#### **Absolute Maximum Ratings** ( $Ta = 25^{\circ}C$ )

Item	Symbol	17901 P	17901 PJ	17901 FP	17901 FPJ	17901 FPK	17339	17339 F	Unit
Power- supply voltage	V <sub>cc</sub>	36	36	36	36	36	36	36	V
Differential input voltage	Vin(diff)	±V <sub>CC</sub>	±V <sub>CC</sub>	±V <sub>cc</sub>	±V <sub>CC</sub>	±V <sub>CC</sub>	±V <sub>cc</sub>	±V <sub>CC</sub>	V
Input voltage	Vin	-0.3 to +V <sub>CC</sub>	-0.3 to +V <sub>cc</sub>	-0.3 to +V <sub>cc</sub>	-0.3 to +V <sub>cc</sub>	-0.3 to +V <sub>CC</sub>	-0.3 to	-0.3 to +V <sub>CC</sub>	V
Output current	lout*2	20	20	20	20	20	20	20	mA
Allowable power dissipation	P <sub>T</sub>	625*1	625*1	625* <sup>3</sup>	625*3	625* <sup>3</sup>	625*1	625* <sup>3</sup>	mW
Operating temperature	Topr	–20 to +75	–40 to +85	–20 to +75	–40 to +85	-40 to +125	–20 to +75	-20 to +75	°C
Storage temperature	Tstg	–55 to +125	–55 to +125	-55 to +125	–55 to +125	–55 to +150	–55 to +125	-55 to +125	°C
Output pin voltage	Vout	36	36	36	36	36	36	36	V

Notes: 1. These are the allowable values up to Ta = 50°C. Derate by 8.3mW/°C above that temperature.

<sup>2.</sup> These products can be destroyed if the output and  $V_{\text{cc}}$  are shorted together. The maximum output current is the allowable value for continuous operation.

<sup>3.</sup> See notes of SOP Package Usage in Reliability section.

### **Electrical Characteristics 1** ( $V_{CC} = 5V$ , Ta = 25°C)

Item	Symbol	Min	Тур	Max	Unit	Test Condition
Input offset voltage	V <sub>IO</sub>	_	2	7	mV	Output switching point: when $V_0 = 1.4V$ , $R_S = 0\Omega$
Input bias current	I <sub>IB</sub>	_	25	250	nA	I <sub>IN(+)</sub> or I <sub>IN(-)</sub>
Input offset current	I <sub>IO</sub>	_	5	50	nA	$I_{\mathrm{IN}(+)}-I_{\mathrm{IN}(-)}$
Common-mode input voltage*1	V <sub>CM</sub>	0	_	V <sub>cc</sub> – 1.5	V	
Supply current	I <sub>cc</sub>	_	8.0	2	mA	$R_L = \infty$
Voltage Gain	$A_{VD}$	_	200	_	V/mV	$R_L = 15k\Omega$
Response time*2	$t_{R}$	_	1.3	_	μs	$V_{RL} = 5V, R_L = 5.1k\Omega$
Output sink current	losink	6	16	_	mA	$V_{IN(-)} = 1V, \ V_{IN(+)} = 0, \ V_{O} \le 1.5V$
Output saturation voltage	V <sub>o</sub> sat	_	200	400	mV	$V_{IN(-)} = 1V, V_{IN(+)} = 0, Iosink = 3mA$
Output leakage current	I <sub>LO</sub>	_	0.1	_	nA	$V_{IN(+)} = 1V, \ V_{IN(-)} = 0, \ V_O = 5V$

Notes: 1. Voltages more negative than -0.3V are not allowed for the common-mode input voltage or for either one of the input signal voltages.

2. The stipulated response time is the value for a 100 mV input step voltage that has a 5mV overdrive.

### **Electrical Characteristics 2** ( $V_{CC} = 5V$ , Ta = -41 to +125°C)

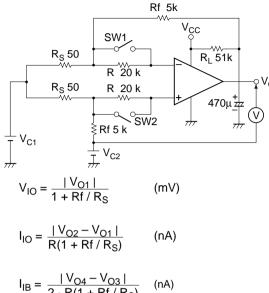
Item	Symbol	Min	Тур	Max	Unit	Test Condition
Input offset voltage	V <sub>IO</sub>	_	_	7	mV	Output switching point: when $V_0 = 1.4V$ , $R_s = 0\Omega$
Input offset current	I <sub>IO</sub>	_	_	200	nA	$I_{\text{IN}(\cdot)} - I_{\text{IN}(\cdot)}$
Input bias current	I <sub>IB</sub>	_	_	500	nA	
Common-mode input voltage*1	$V_{\text{CM}}$	0	_	$V_{CC} - 2.0$	V	
Output saturation voltage	$V_{\text{O sat}}$	_	_	440	mV	$V_{IN(-)} \ge 1V$ , $V_{IN(+)} = 0$ , $Iosink \le 4mA$
Output leakage current	I <sub>LO</sub>	_	1.0	_	μΑ	$V_{IN(-)} = 0V, \ V_{IN(+)} \ge 1V, \ V_O = 30V$
Supply current	I <sub>cc</sub>	_	_	4.0	mA	All comparators: $R_L = \infty$ , All channels ON

Note: 1. Voltages more negative than -0.3V are not allowed for the common-mode input voltage or for either one of the input signal voltages.



#### **Test Circuits**

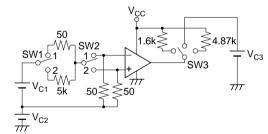
1. Input offset voltage  $(V_{IO})$ , input offset current  $(I_{IO})$ , and Input bias current  $(I_{IB})$  test circuit



SW1	SW2	Vout	
On	On	V <sub>O1</sub>	$V_{C1} = \frac{1}{2} V_{CC}$
Off	Off	V <sub>O2</sub>	$v_{C1} - \frac{1}{2}v_{CC}$
On	Off	V <sub>O3</sub>	$V_{C2} = 1.4V$
Off	On	$V_{O4}$	

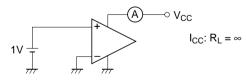
$$I_{IB} = \frac{\mid V_{O4} - V_{O3} \mid}{2 \cdot R(1 + Rf / R_S)} \quad (nA)$$

2. Output saturation voltage ( $V_0$  sat) output sink current (Iosink), and common-mode input voltage ( $V_{CM}$ ) test circuit

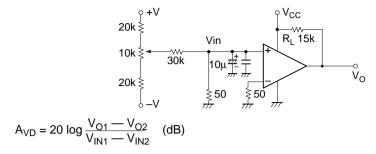


Item	$V_{C1}$	$V_{C2}$	$V_{C3}$	SW1	SW2	SW3	Unit
V <sub>O</sub> sat	2V	0V	_	1	1	1 at V <sub>CC</sub> = 5V 3 at	V ,
losink	2\/	OV	1.5V	1	1	V <sub>CC</sub> = 15\	mA
		UV	1.5 v	'	'	_	
V <sub>CM</sub>	2V	–1 to V <sub>CC</sub>	_	2	Switched between 1 and 2	3	V

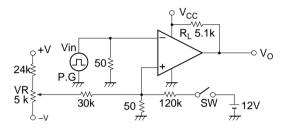
3. Supply current  $(I_{CC})$  test circuit



4. Voltage gain  $(A_{VD})$  test circuit  $(R_L = 15k\Omega)$ 

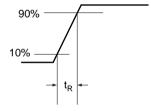


5. Response time  $(t_R)$  test circuit

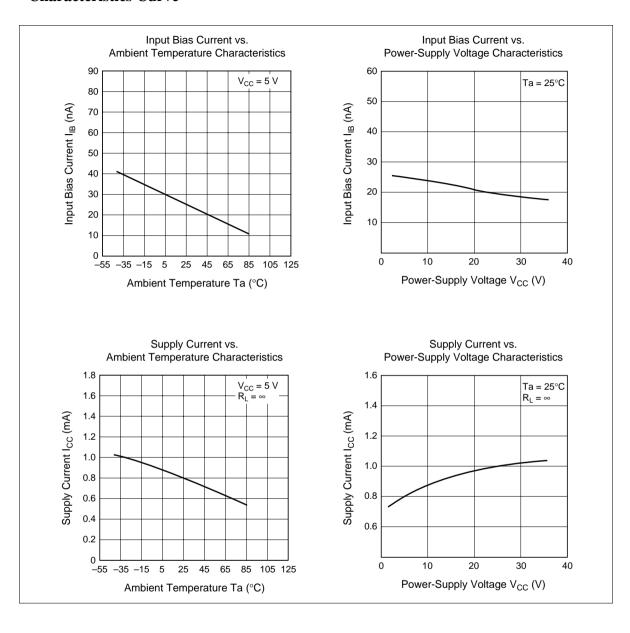


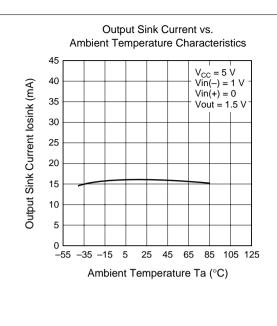
 $t_R$ :  $R_L = 5.1 k\Omega$ , a 100mV input step voltage that has a 5mV overdrive

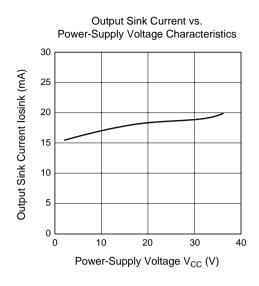
- With  $V_{IN}$  not applied, set the switch SW to the off position and adjust  $V_R$  so that  $V_O$  is in the vicinity of 1.4V.
- Apply  $V_{IN}$  and turn the switch SW on.

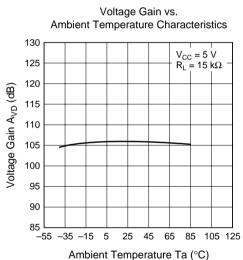


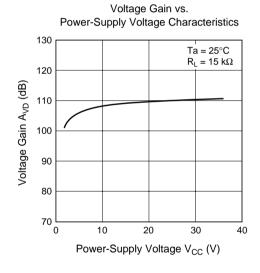
#### **Characteristics Curve**











#### **HA17901 Application Examples**

The HA17901 houses four independent comparators in a single package, and operates over a wide voltage range at low power from a single-voltage power supply. Since the common-mode input voltage range starts at the ground potential, the HA17901 is particularly suited for single-voltage power supply applications. This section presents several sample HA17901 applications.

#### **HA17901 Application Notes**

#### 1. Square-Wave Oscillator

The circuit shown in figure one has the same structure as a single-voltage power supply astable multivibrator. Figure 2 shows the waveforms generated by this circuit.

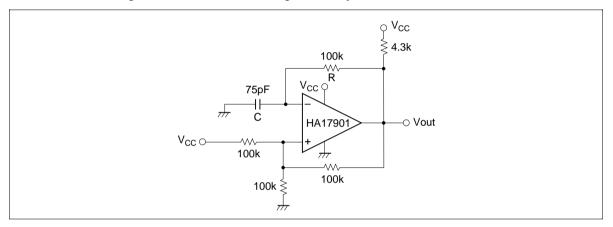


Figure 1 Square-Wave Oscillator

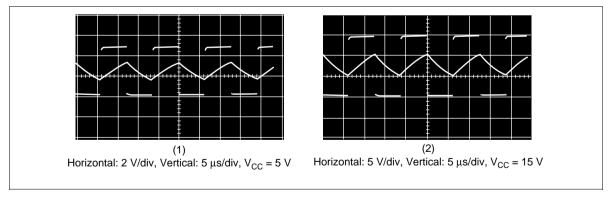


Figure 2 Operating Waveforms

#### 2. Pulse Generator

The charge and discharge circuits in the circuit from figure 1 are separated by diodes in this circuit. (See figure 3.) This allows the pulse width and the duty cycle to be set independently. Figure 4 shows the waveforms generated by this circuit.

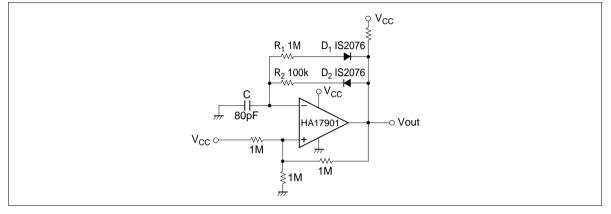


Figure 3 Pulse Generator

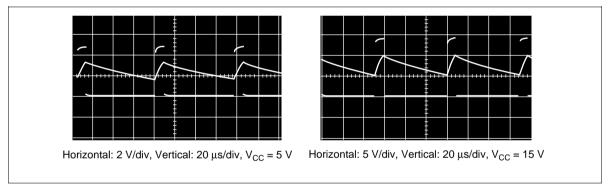


Figure 4 Operating Waveforms

#### 3. Voltage Controlled Oscillator

In the circuit in figure 5, comparator  $A_1$  operates as an integrator,  $A_2$  operates as a comparator with hysteresis, and  $A_3$  operates as the switch that controls the oscillator frequency. If the output Vout1 is at the low level, the  $A_3$  output will go to the low level and the A1 inverting input will become a lower level than the A1 noninverting input. The A1 output will integrate this state and its output will increase towards the high level. When the output of the integrator  $A_1$  exceeds the level on the comparator  $A_2$  inverting input,  $A_2$  inverts to the high level and both the output Vout1 and the  $A_3$  output go to the high level. This causes the integrator to integrate a negative state, resulting in its output decreasing towards the low level. Then, when the  $A_1$  output level becomes lower than the level on the  $A_2$  noninverting input, the output Vout1 is once again inverted to the low level. This operation generates a square wave on Vout1 and a triangular wave on Vout2.

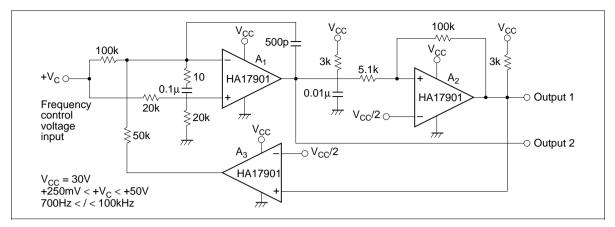


Figure 5 Voltage Controlled Oscillator

#### 4. Basic Comparator

The circuit shown in figure 6 is a basic comparator. When the input voltage  $V_{IN}$  exceeds the reference voltage  $V_{REF}$ , the output goes to the high level.

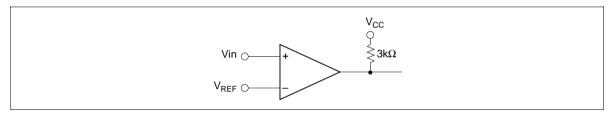


Figure 6 Basic Comparator

#### 5. Noninverting Comparator (with Hysteresis)

Assuming  $+V_{IN}$  is 0V, when  $V_{REF}$  is applied to the inverting input, the output will go to the low level (approximately 0V). If the voltage applied to  $+V_{IN}$  is gradually increased, the output will go high when the value of the noninverting input,  $+V_{IN} \times R_2/(R_1+R_2)$ , exceeds  $+V_{REF}$ . Next, if  $+V_{IN}$  is gradually lowered, Vout will be inverted to the low level once again when the value of the noninverting input,  $(Vout-V_{IN})\times R_1/(R_1+R_2)$ , becomes lower than  $V_{REF}$ . With the circuit constants shown in figure 7, assuming  $V_{CC}=15V$  and  $+V_{REF}=6V$ , the following formula can be derived, i.e.  $+V_{IN}\times 10M/(5.1M+10M)>6V$ , and Vout will invert from low to high when  $+V_{IN}$  is >9.06V.

$$(Vout - V_{IN}) \times \frac{R_1}{R_1 + R_2} + V_{IN} < 6V$$
(Assuming Vout = 15V)

When  $+V_{IN}$  is lowered, the output will invert from high to low when  $+V_{IN} < 1.41$ V. Therefore this circuit has a hysteresis of 7.65V. Figure 8 shows the input characteristics.

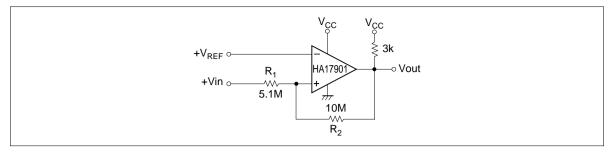


Figure 7 Noninverting Comparator

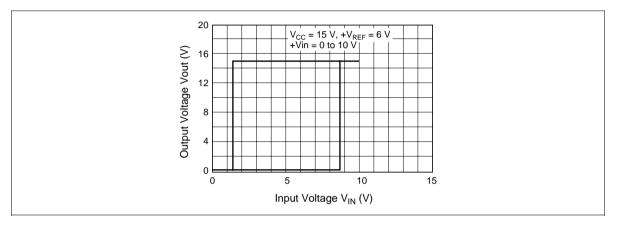


Figure 8 Noninverting Comparator I/O Transfer Characteristics

#### 6. Inverting Comparator (with Hysteresis)

In this circuit, the output Vout inverts from high to low when  $+V_{IN} > (V_{CC} + Vout)/3$ . Similarly, the output Vout inverts from low to high when  $+V_{IN} < V_{CC}/3$ . With the circuit constants shown in figure 9, assuming  $V_{CC} = 15V$  and  $V_{CC} = 15V$ , this circuit will have a 5V hysteresis. Figure 10 shows the I/O characteristics for the circuit in figure 9.

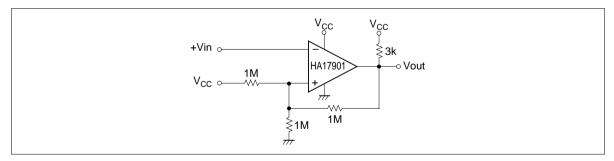


Figure 9 Inverting Comparator

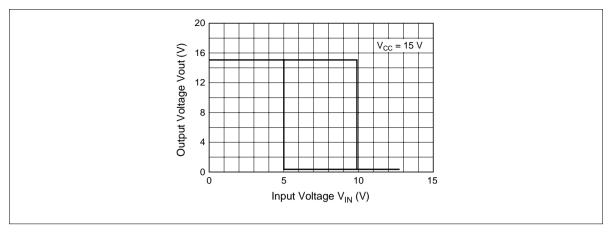


Figure 10 Inverting Comparator I/O Transfer Characteristics

7. Zero-Cross Detector (Single-Voltage Power Supply)

In this circuit, the noninverting input will essentially beheld at the potential determined by dividing  $V_{CC}$  with  $100k\Omega$  and  $10k\Omega$  resistors. When  $V_{IN}$  is 0V or higher, the output will be low, and when  $V_{IN}$  is negative, Vout will invert to the high level. (See figure 11.)

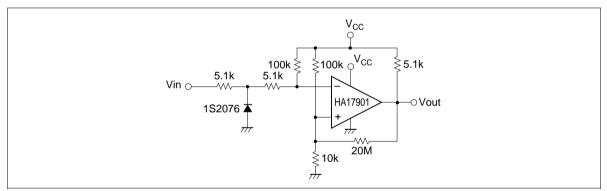
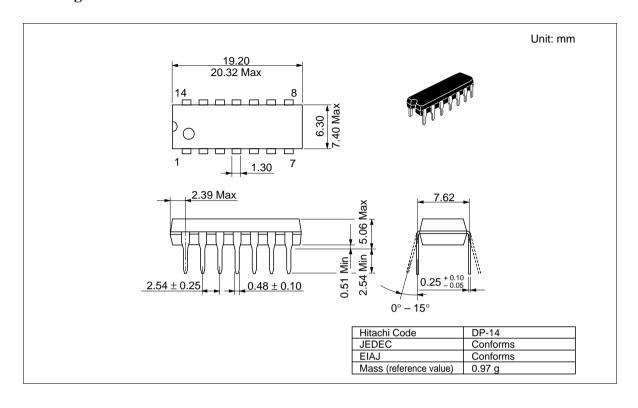
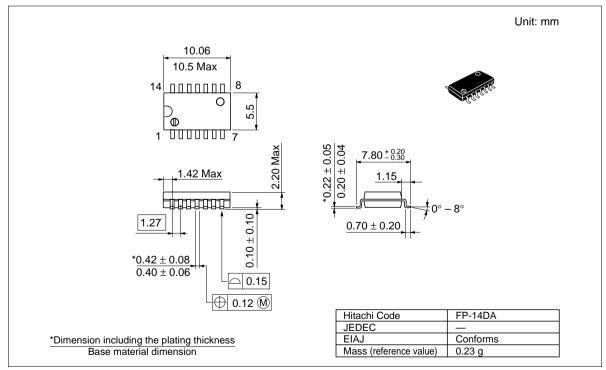


Figure 11 Zero-Cross Detector

#### **Package Dimensions**





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