

Cost-Effective, 2A Peak Sink/Source Bus Termination Regulator

General Description

The RT9199 is a simple, cost-effective and high-speed linear regulator designed to generate termination voltage in Double Data Rate (DDR) memory system to comply with the devices requirements. The regulator is capable of actively sinking or sourcing up to peak 2A for DDRII or 1.5A for DDRIII while regulating an output voltage to within 20mV. The output termination voltage can be tightly regulated to track $1/2V_{DDQ}$ by two external voltage divider resistors or the desired output voltage can be pro-grammed by externally forcing the REFEN pin voltage.

The RT9199 also incorporates a high-speed differential amplifier to provide ultra-fast response in line/load transient. Other features include extremely low initial offset voltage, excellent load regulation, current limiting in bi-directions and on-chip thermal shut-down protection.

The RT9199 are available in both SOP-8 and SOP-8 (Exposed Pad) surface mount packages.

Ordering Information

Package Type
S: SOP-8
SP: SOP-8 (Exposed Pad-Option 2)

Lead Plating System
P: Pb Free
G: Green (Halogen Free and Pb Free)
Z: ECO (Ecological Element with

Note:

Richtek products are:

RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.

Halogen Free and Pb free)

▶ Suitable for use in SnPb or Pb-free soldering processes.

Features

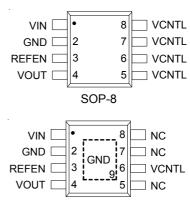
- Ideal for DDRII / DDRIII V_{TT} Applications
- Sink and Source Current:
 - ▶ Peak 2A for DDRII
 - ▶ Peak 1.5A for DDRIII
- Integrated Power MOSFETs
- Generate Termination Voltage for DDR Memory Interfaces
- High Accuracy Output Voltage at Full-Load
- Output Adjustment by Two External Resistors
- Low External Component Count
- Shutdown for Suspend to RAM (STR) Functionality with High-Impedance Output
- Current Limiting Protection
- On-Chip Thermal Protection
- RoHS Compliant and Halogen Free

Applications

- Desktop PCs, Notebooks, and Workstations
- Graphics Card Memory Termination
- Set Top Boxes, Digital TVs, Printers
- Embedded Systems
- Active Termination Buses
- DDR/II Memory Systems

Pin Configurations

(TOP VIEW)



SOP-8 (Exposed Pad)

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Marking Information

RT9199xS

RT9199 xSYMDNN RT9199xS: Product Number

x:PorGorZ

YMDNN: Date Code

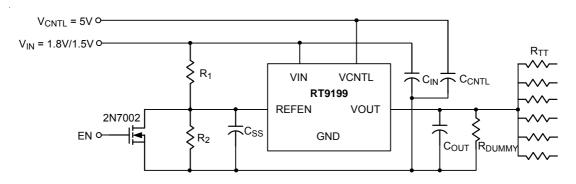
RT9199xSP

RT9199 xSPYMDNN RT9199xSP: Product Number

x:PorGorZ

YMDNN: Date Code

Typical Application Circuit



 $R_1 = R_2 = 100k\Omega$, $R_{TT} = 50\Omega / 33\Omega / 25\Omega$

 $C_{\text{OUT(MIN)}}$ = $10 \mu F$ (Ceramic) + $1000 \mu F$ under the worst case testing condition

 R_{DUMMY} = 1k Ω as for V_{OUT} discharge when V_{IN} is not presented but V_{CNTL} is presented

 C_{SS} = 1 μ F, C_{IN} = 470 μ F (Low ESR), C_{CNTL} = 47 μ F

Test Circuit

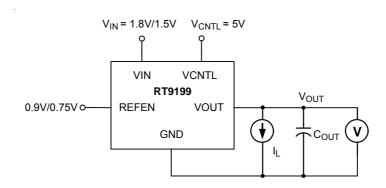


Figure 1. Output Voltage Tolerance, ΔV_{LOAD}

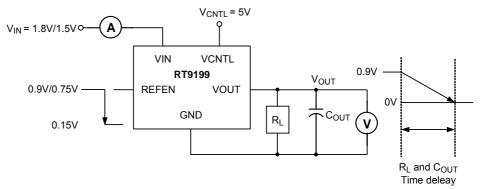


Figure 2. Current in Shutdown Mode, ISTBY

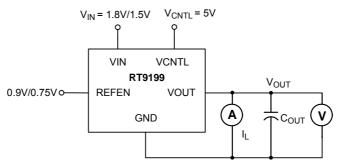


Figure 3. Current Limit for High Side, ILIM

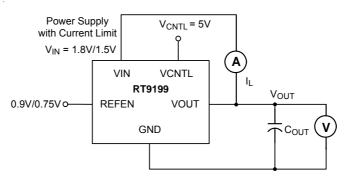


Figure 4. Current Limit for Low Side, ILIM

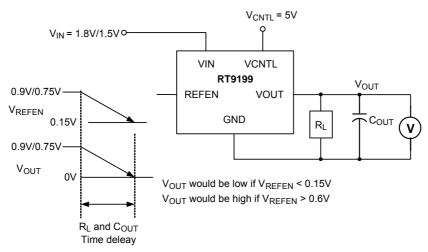


Figure 5. REFEN Pin Shutdown Threshold, VIH & VIL

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Functional Pin Description

VIN

Input voltage which supplies current to the output pin. Connect this pin to a well-decoupled supply voltage. To prevent the input rail from dropping during large load transient, a large, low ESR capacitor is recommended to use. The capacitor should be placed as close as possible to the VIN pin.

GND (Exposed Pad)

Common Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.

VCNTL

VCNTL supplies the internal control circuitry and provides the drive voltage. The driving capability of output current is proportioned to the VCNTL. Connect this pin to 5V bias supply to handle large output current with at least $1\mu F$ capacitor from this pin to GND. An important note is that VIN should be kept lower or equal to VCNTL.

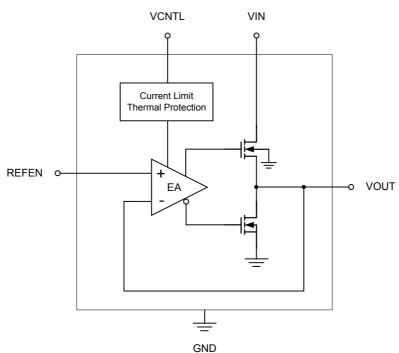
REFEN

Reference voltage input and active low shutdown control pin. Two resistors dividing down the VIN voltage on the pin to create the regulated output voltage. Pulling the pin to ground turns off the device by an open-drain, such as 2N7002, signal N-MOSFET.

VOUT

Regulator output. VOUT is regulated to REFEN voltage that is used to terminate the bus resistors. It is capable of sinking and sourcing current while regulating the output rail. To maintain adequate large signal transient response, typical value of $1000\mu F$ Al electrolytic capacitor with $10\mu F$ ceramic capacitors are recommended to reduce the effects of current transients on VOUT.

Function Block Diagram





Absolute Maximum Ratings (Note 1)

• Input Voltage, V _{IN}	6V
Control Voltage, V _{CNTL}	6V
 Power Dissipation, P_D @ T_A = 25°C 	
SOP-8	0.909W
SOP-8 (Exposed Pad)	2.33W
Package Thermal Resistance (Note 2)	
SOP-8, θ_{JA}	110°C/W
SOP-8, θ _{JC}	60°C/W
SOP-8 (Exposed Pad), θ_{JA}	42.9°C/W
SOP-8 (Exposed Pad), θ_{JC}	12.6°C/W
Junction Temperature	125°C
• Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range	–65°C to 150°C
ESD Susceptibility (Note 3)	
HBM (Human Body Model)	2kV
MM (Machine Model)	200V

Recommended Operating Conditions (Note 4)

• Input Voltage, V_{IN} ------ 1.4V to 5.5V

ullet Control Voltage, V_{CNTL} ------ 5V \pm 5%

Electrical Characteristics

 $(V_{IN}$ = 1.8V/1.5V, V_{CNTL} = 5V, V_{REFEN} = 0.9V/0.75V, C_{OUT} = 10 μ F (Ceramic), T_A = 25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions		Тур	Max	Unit	
Input							
VCNTL Operation Current	I _{CNTL}	I _{OUT} = 0A		1	2.5	mA	
Standby Current (Note 5)	I _{STBY}	$V_{REFEN} < 0.2V$ (Shutdown), $R_{LOAD} = 180\Omega$		2	90	μΑ	
Output (DDR II)							
Output Offset Voltage (Note 6)	Vos	I _{OUT} = 0A	-10		10	mV	
	A \ /	V_{IN} = 1.8V, V_{REFIN} = 0.9V, I_{OUT} = ±1.8A	-20		20	⊢ mV l	
Load Regulation (Note 7)	ΔV_{LOAD}	$V_{IN} = 1.5V$, $V_{REFIN} = 0.75V$, $I_{OUT} = \pm 1.4A$	-10		10		
Protection							
Current limit	1	V _{IN} = 1.8V, V _{REFIN} = 0.9V	2 3.5		3.5		
Current limit	ILIMIT	V _{IN} = 1.5V, V _{REFIN} = 0.75V	1.5		3.5	Α	
Thermal Shutdown Temperature	T _{SD}	V _{CNTL} = 5V		170		°C	
Thermal Shutdown Hysteresis	ΔT_{SD}	V _{CNTL} = 5V		35		°C	
REFEN Shutdown							
Chutdour Throchold	VIH	Enable				V	
Shutdown Threshold	VIL	Shutdown			0.15		

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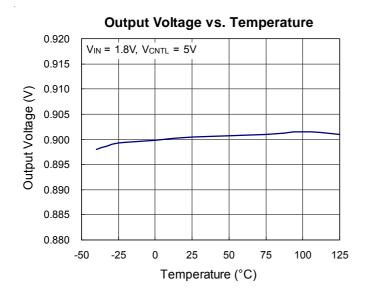
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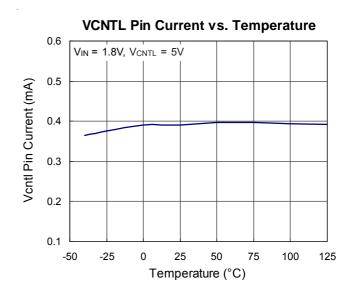


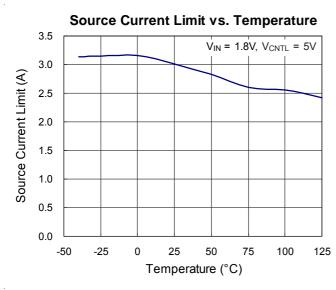
- Note 1. Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2. θ_{JA} is measured in the natural convection at T_A = 25°C on a high effective thermal conductivity test board (4 Layers, 2S2P) of JEDEC 51-7 thermal measurement standard. The case point of θ_{JC} is on the exposed pad for package.
- Note 3. Devices are ESD sensitive. Handling precaution recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.
- Note 5. Standby current is the input current drawn by a regulator when the output voltage is disabled by a shutdown signal on REFEN pin ($V_{IL} < 0.15V$). It is measured with $V_{IN} = V_{CNTL} = 5V$.
- Note 6. Vos offset is the voltage measurement defined as VouT subtracted from VREFEN.
- Note 7. Regulation is measured at constant junction temperature by using a 5ms current pulse. Devices are tested for load regulation in the load range from 0A to 2A peak.

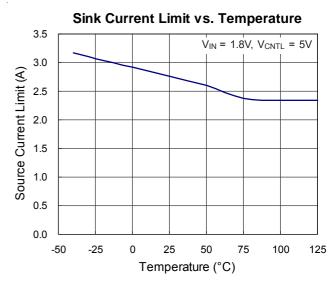


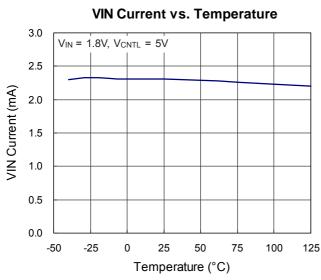
Typical Operating Characteristics

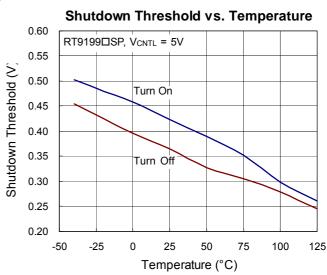








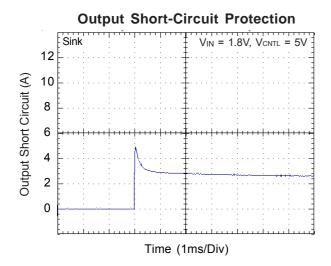


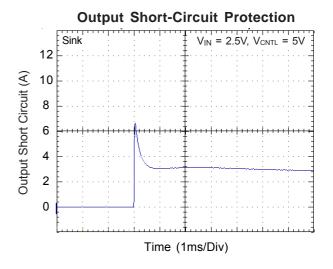


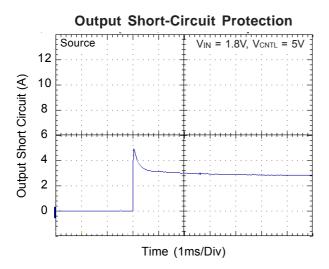
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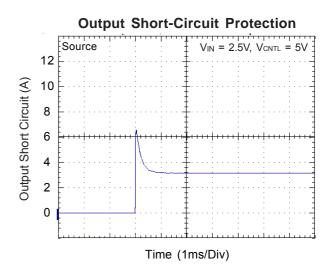
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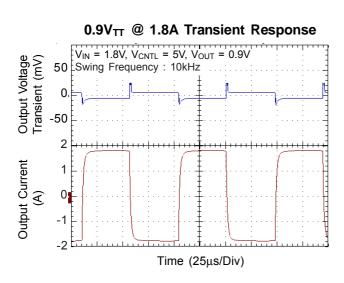


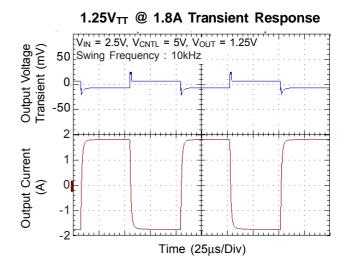












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Application Information

Consideration while designing the resistance of voltage divider

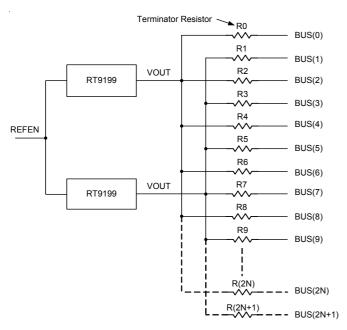
Refer to the "Typical Application Circuit". Make sure the current sinking capability of pull-down NMOS is enough for the chosen voltage divider to pull-down the voltage at REFEN pin below 0.15V to shutdown the device.

In addition, the capacitor C_{SS} and voltage divider form the low-pass filter. There are two reasons doing this design; one is for output voltage soft-start while another is for noise immunity.

How to reduce power dissipation on Notebook PC or the dual channel DDR SDRAM application?

In notebook application, using RichTek's Patent "Distributed Bus Terminator Topology" with choosing RichTek's product is encouraged.

Distributed Bus Terminating Topology



General Regulator

The RT9199 could also serves as a general linear regulator. The RT9199 accepts an external reference voltage at REFEN pin and provides output voltage regulated to this reference voltage as shown in Figure 6, where

 $V_{OUT} = V_{REFEN} \times R2/(R1+R2)$

As other linear regulator, dropout voltage and thermal issue should be specially considered. Figure 7 shows the $R_{DS(ON)}$ over temperature of RT9199. The minimum dropout voltage could be obtained by the product of $R_{DS(ON)}$ and output current. For thermal consideration, please refer to the relative sections.

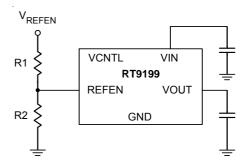
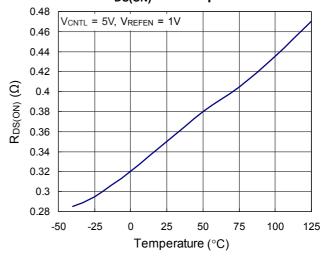


Figure 6

R_{DS(ON)} vs. Temperature



Input Capacitor and Layout Consideration

Place the input bypass capacitor as close as possible to the RT9199. A low ESR capacitor larger than 470uF is recommended for the input capacitor. Use short and wide traces to minimize parasitic resistance and inductance. Inappropriate layout may result in large parasitic inductance and cause undesired oscillation between RT9199 and the preceding power converter.

Figure 7

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Thermal Consideration

RT9199 regulators have internal thermal limiting circuitry designed to protect the device during overload conditions. For continued operation, do not exceed absolute maximum operation junction temperature 125°C. The power dissipation definition in device is:

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{O}$$

The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surroundings airflow and temperature difference between junction to ambient. The maximum power dissipation can be calculated by following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

Where $T_{J(MAX)}$ is the maximum operation junction temperature 125°C, TA is the ambient temperature and the θ_{JA} is the junction to ambient thermal resistance. The junction to ambient thermal resistance for SOP-8 package (Exposed Pad) is 42.9°C/W, on standard JEDEC 51-7 (4 layers, 2S2P) thermal test board. The maximum power dissipation at T_A = 25°C can be calculated by following formula:

$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / 42.9^{\circ}C/W = 2.33W$$

Figure 8 shows the package sectional drawing of SOP-8 (Exposed Pad). Every package has several thermal dissipation paths. As show in Figure 9, the thermal resistance equivalent circuit of SOP-8 (Exposed Pad). The path 2 is the main path due to these materials thermal conductivity. We define the exposed pad is the case point of the path 2.

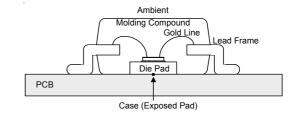


Figure 8. SOP-8 (Exposed Pad) Package Sectional Drawing

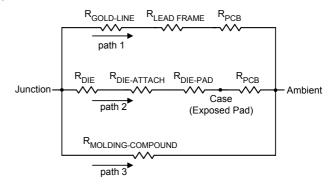
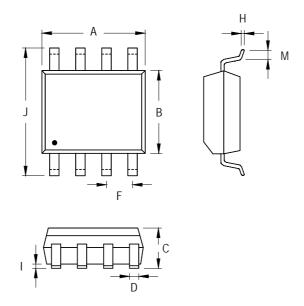


Figure 9. Thermal Resistance Equivalent Circuit

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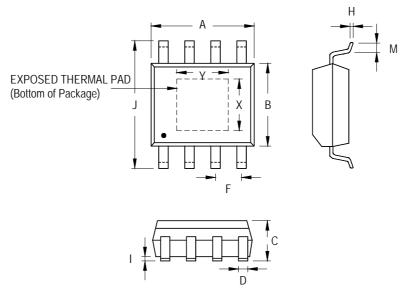
Outline Information



O le a l	Dimensions In Millimeters		Dimensions In Inches		
Symbol	Min	Max	Min	Max	
Α	4.801	5.004	0.189	0.197	
В	3.810	3.988	0.150	0.157	
С	1.346	1.753	0.053	0.069	
D	0.330	0.508	0.013	0.020	
F	1.194	1.346	0.047	0.053	
Н	0.170	0.254	0.007	0.010	
I	0.050	0.254	0.002	0.010	
J	5.791	6.200	0.228	0.244	
М	0.400	1.270	0.016	0.050	

8-Lead SOP Plastic Package





Symbol		Dimensions In Millimeters		Dimensions In Inches		
		Min	Max	Min	Max	
А		4.801	5.004	0.189	0.197	
В		3.810	4.000	0.150	0.157	
С		1.346	1.753	0.053	0.069	
D		0.330	0.510	0.013	0.020	
F		1.194	1.346	0.047	0.053	
Н		0.170	0.254	0.007	0.010	
I		0.000	0.152	0.000	0.006	
J		5.791	6.200	0.228	0.244	
М		0.406	1.270	0.016	0.050	
Option 1	Х	2.000	2.300	0.079	0.091	
	Υ	2.000	2.300	0.079	0.091	
Option 2	Х	2.100	2.500	0.083	0.098	
	Υ	3.000	3.500	0.118	0.138	

8-Lead SOP (Exposed Pad) Plastic Package

Richtek Technology Corporation

14F, No. 8, Tai Yuen 1st Street, Chupei City Hsinchu, Taiwan, R.O.C.

Tel: (8863)5526789

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