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MAXIM

Low-Noise, Low-Dropout, 200mA Linear Regulator in UCSP

MAX8532

General Description

The MAX8532 offers the benefits of low-dropout voltage and ultra-low power regulation in a subminiaturized UCSP, making it ideal for space-restricted portable equipment. The device operates from a 2.5V to 6.5V input and delivers up to 200mA, with low dropout of 100mV (typ) at 100mA. Designed with an internal P-channel MOSFET pass transistor, the supply current is kept at a low 80μA, independent of the load current and dropout voltage. Other features include short-circuit protection and thermal-shutdown protection.

The MAX8532 includes a reference bypass pin for low output noise (40μV_{RMS}) and a logic-controlled shutdown input. The device is available in a tiny 6-pin UCSP.

Applications

Cellular and Cordless Phones
PDAs and Palmtop Computers
Notebook Computers
Digital Cameras
PCMCIA Cards
Wireless LAN Cards
Hand-Held Instruments

Features

- ◆ Guaranteed 200mA Output Current
- ◆ Low 100mV (typ) Dropout at 100mA
- ◆ Low 40μV_{RMS} Output Noise
- ◆ Low 80μA Operating Supply Current
- ◆ 62dB PSRR
- ◆ <1μA Shutdown Current
- ◆ Thermal-Overload and Short-Circuit Protection
- ◆ Output Current Limit
- ◆ Tiny 1.16mm x 1.57mm x 0.66mm UCSP (3 x 2 Grid)

Ordering Information

PART	TEMP RANGE	OUT VOLTAGE	PIN-PACKAGE
MAX8532EBT_*	-40°C to +85°C	1.5V to 3.3V	6 UCSP

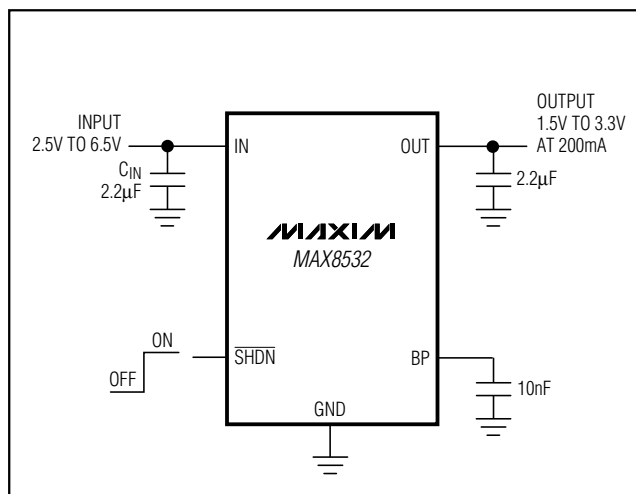
*"_" = Output voltage code (see the Output Voltage Selector Guide).

Output Voltage Selector Guide

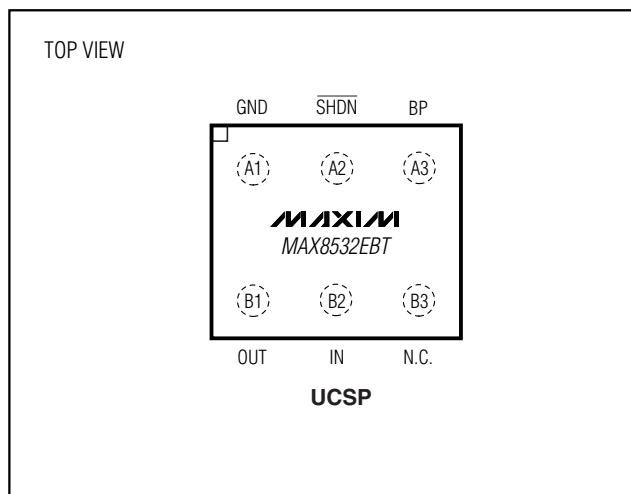
PART	V _{OUT} (V)	TOP MARK
MAX8532EBTJ	2.85	ACP
MAX8532EBTG	3	ACU

Note: Contact the factory for other output voltages between 1.5V and 3.3V. The minimum order quantity is 25,000 units.

Typical Operating Circuit



Pin Configuration



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For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

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ABSOLUTE MAXIMUM RATINGS

IN, $\overline{\text{SHDN}}$, BP to GND -0.3V to +7V
 OUT to GND -0.3V to ($V_{\text{IN}} + 0.3\text{V}$)
 Output Short-Circuit Duration Indefinite
 Continuous Power Dissipation ($T_A = +70^\circ\text{C}$)
 6-Pin UCSP (derate 3.9mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$) 308mW

Operating Temperature Range -40°C to $+85^\circ\text{C}$
 Junction Temperature $+150^\circ\text{C}$
 Storage Temperature Range -65°C to $+150^\circ\text{C}$
 6-Pin UCSP Solder Profile (Note 1)

Note 1: For UCSP solder profile information, visit www.maxim-ic.com/1st_pages/UCSP.html.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

($V_{\text{IN}} = 3.8\text{V}$, $\overline{\text{SHDN}} = \text{IN}$, $C_{\text{BP}} = 10\text{nF}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$, unless otherwise noted. Typical values are at $T_A = +25^\circ\text{C}$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Voltage	V _{IN}			2.5		6.5	V
Undervoltage Lockout Threshold	V _{UVLO}	IN rising, hysteresis is 40mV (typ)		2.15	2.25	2.42	V
Output Voltage Accuracy		T _A = +25°C, I _{OUT} = 1mA		-1		+1	%
		T _A = -40°C to +85°C, I _{OUT} = 1mA		-2		+2	
		T _A = -40°C to +85°C, I _{OUT} = 0.1mA to 200mA		-3		+3	
Maximum Output Current	I _{OUT}			200			mA
Current Limit	I _{LIM}			210	330	550	mA
Ground Current	I _Q	No load			80	150	μA
		I _{OUT} = 100mA			100		
Dropout Voltage	V _{OUT} - V _{IN}	I _{OUT} = 100mA (Note 3)			100	200	mV
Line Regulation	ΔV _{LNR}	IN = (OUT + 0.1V) to 3.8V		-0.2		+0.2	%/V
Output Voltage Noise		10Hz to 100kHz, C _{OUT} = 10μF, I _{OUT} = 10mA			40		μV _{RMS}
Ripple Rejection	PSRR	100Hz, I _{OUT} = 30mA			62		dB
SHUTDOWN							
$\overline{\text{SHDN}}$ Supply Current	I _{OFF}	$\overline{\text{SHDN}}$ = 0, T _A = +25°C			0.01	1	μA
		$\overline{\text{SHDN}}$ = 0, T _A = +85°C			0.1		
$\overline{\text{SHDN}}$ Input Threshold	V _{IH}	Input high voltage		1.6			V
	V _{IL}	Input low voltage				0.4	
$\overline{\text{SHDN}}$ Input Bias Current	I _{SHDN}	$\overline{\text{SHDN}}$ = IN or GND	T _A = +25°C		0.7	100	nA
			T _A = +85°C		0.8		
THERMAL PROTECTION							
Thermal-Shutdown Temperature	T _{SHDN}	T _J rising			160		°C
Thermal-Shutdown Hysteresis	ΔT _{SHDN}				10		°C

Note 2: All units are 100% production tested at $T_A = +25^\circ\text{C}$. Limits over the operating temperature range are guaranteed by design.

Note 3: The dropout voltage is defined as $V_{\text{IN}} - V_{\text{OUT}}$, when V_{OUT} is 100mV below the value of V_{OUT} for $V_{\text{IN}} = V_{\text{OUT}} + 0.5\text{V}$.

Specification applies only when $V_{\text{OUT}} > 2.5\text{V}$.

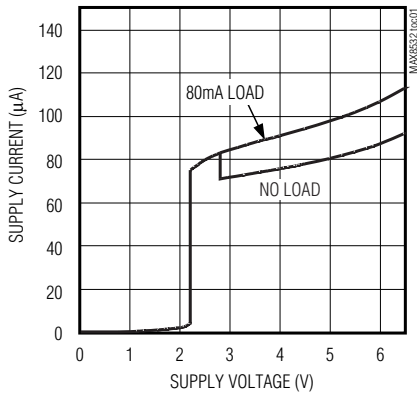
Low-Noise, Low-Dropout, 200mA Linear Regulator in UCSP

Typical Operating Characteristics

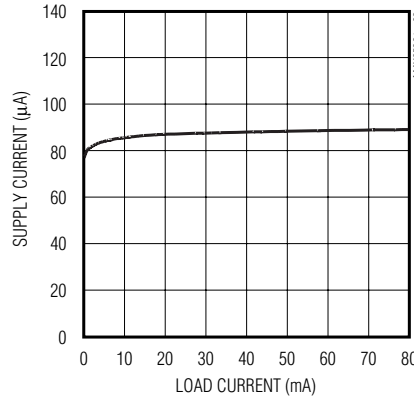
($V_{OUT} = 2.85V$, load = 80mA, $V_{IN} = 3.8V$, $C_{OUT} = 2.2\mu F$, $C_{BP} = 0.01\mu F$, and $C_{IN} = 2.2\mu F$. $T_A = +25^\circ C$, unless otherwise noted.)

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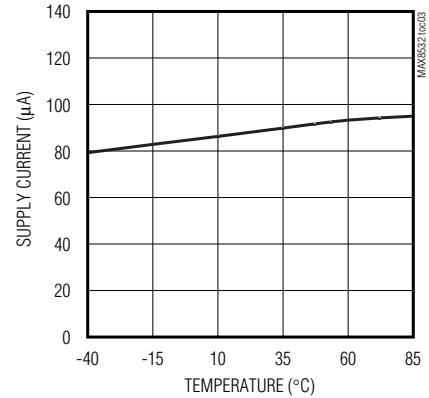
**SUPPLY CURRENT
vs. SUPPLY VOLTAGE**



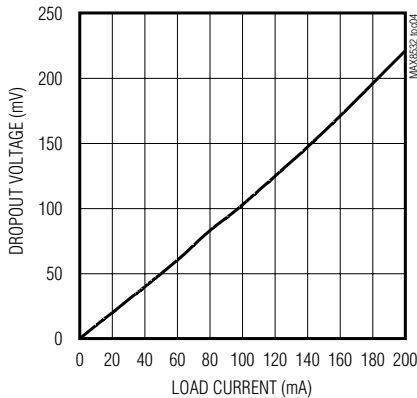
**SUPPLY CURRENT
vs. LOAD CURRENT**



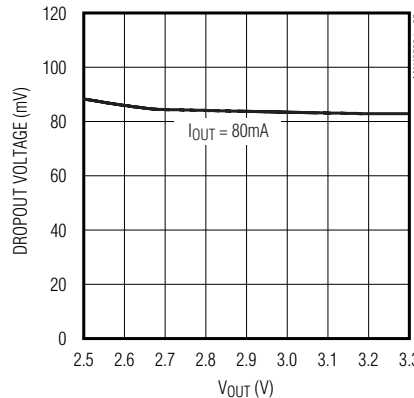
**SUPPLY CURRENT
vs. TEMPERATURE**



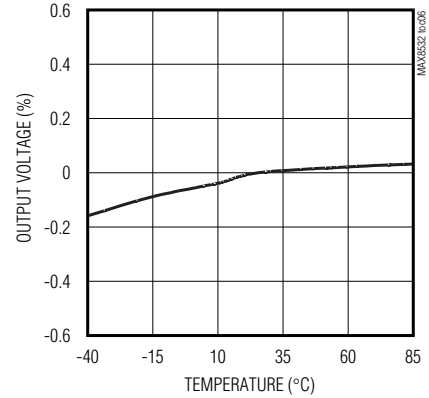
**DROPOUT VOLTAGE
vs. LOAD CURRENT**



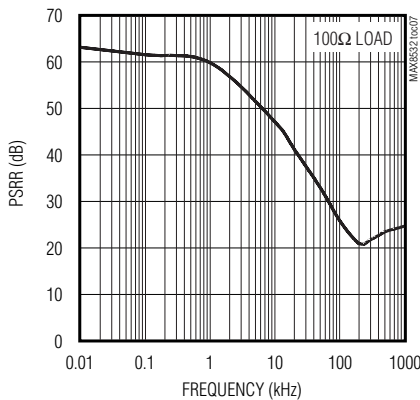
**LDO DROPOUT VOLTAGE
vs. V_{OUT}**



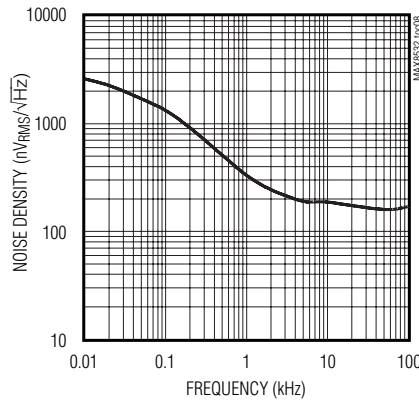
**OUTPUT VOLTAGE ACCURACY
vs. TEMPERATURE**



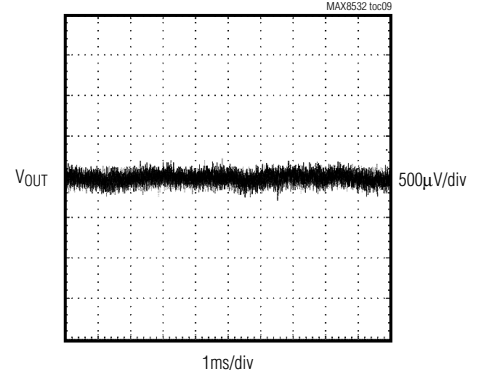
PSRR vs. FREQUENCY



**OUTPUT NOISE SPECTRAL DENSITY
vs. FREQUENCY**



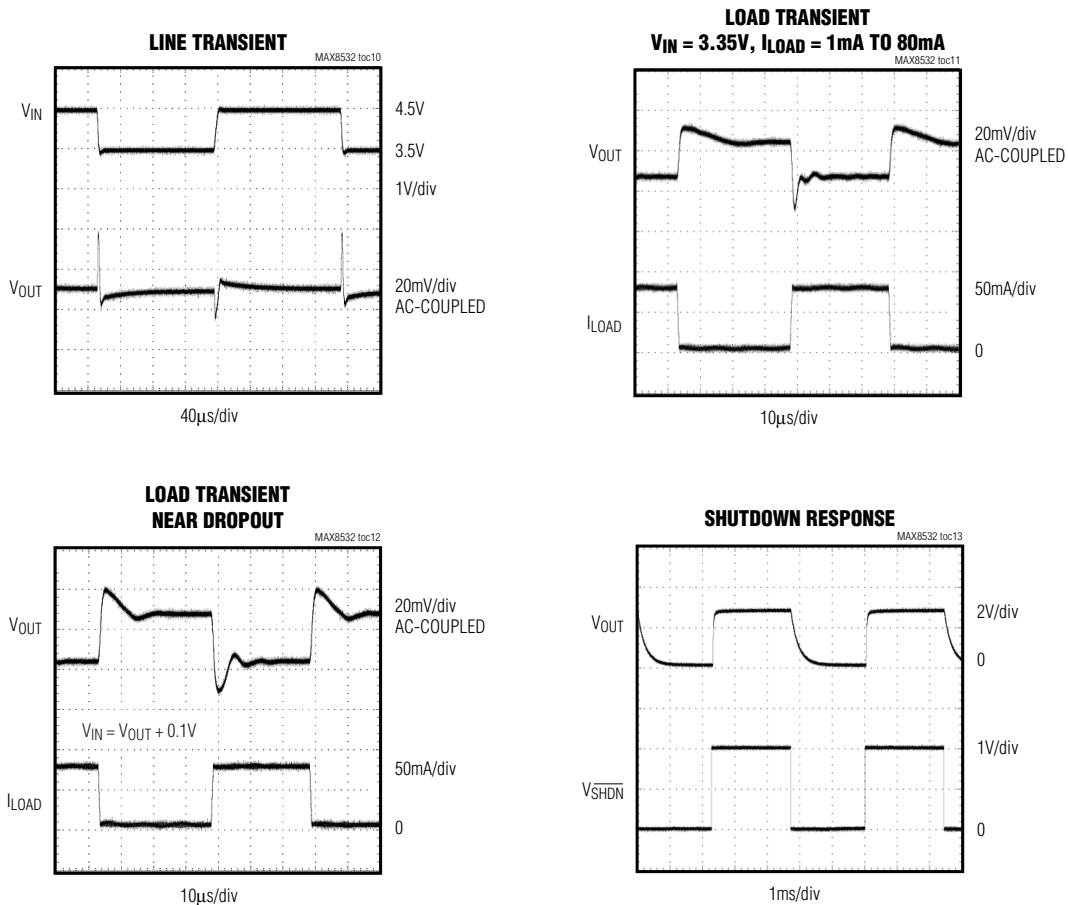
OUTPUT NOISE (10Hz TO 100kHz)



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Typical Operating Characteristics (continued)

($V_{OUT} = 2.85V$, load = 80mA, $V_{IN} = 3.8V$, $C_{OUT} = 2.2\mu F$, $C_{BP} = 0.01\mu F$, and $C_{IN} = 2.2\mu F$. $T_A = +25^\circ C$, unless otherwise noted.)



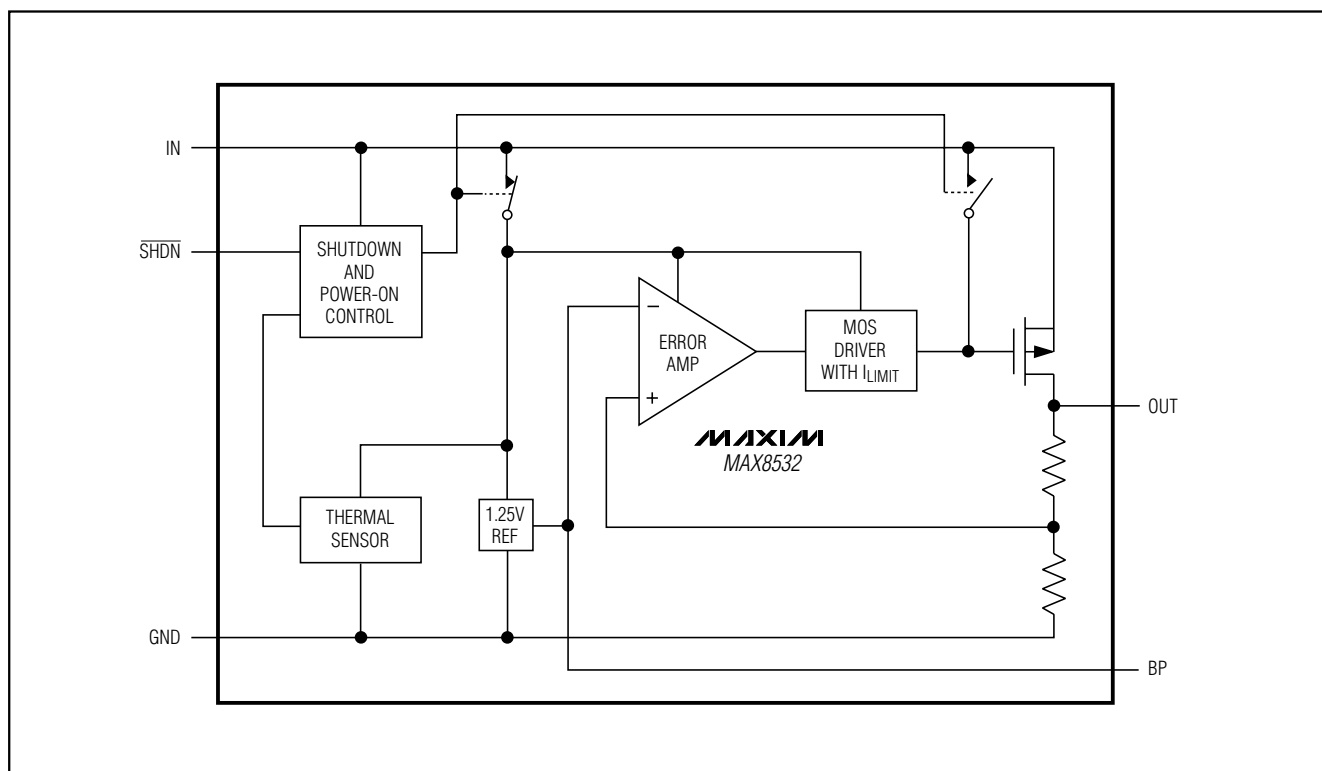
Pin Description

PIN	NAME	FUNCTION
B3	N.C.	Not Connected
B2	IN	Regulator Input
B1	OUT	Regulator Output. Guaranteed 200mA output current.
A1	GND	Ground
A2	$\overline{SHDN}$	Shutdown Input. A logic low shuts down the regulator. Connect to IN for normal operation.
A3	BP	Reference Noise Bypass. Bypass with a 0.01μF ceramic capacitor for reduced noise.

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Functional Diagram

MAX8532



Detailed Description

The MAX8532 is a low-power, low-dropout, low-quiescent current linear regulator designed primarily for battery-powered applications. For preset output voltages, see the *Output Voltage Selector Guide*. The device supplies up to 200mA for OUT. The MAX8532 consists of a 1.25V reference, error amplifier, P-channel pass transistor, reference bypass block, and internal feedback voltage divider.

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled lower, allowing more current to pass to the output and increasing the output voltage. If the feedback voltage is high, the pass-transistor gate is pulled up, allowing less current to pass to the output. The output voltage is fed back through an internal resistor voltage-divider connected to the OUT pin.

Shutdown

The MAX8532 has a single shutdown control input (SHDN). Drive SHDN low to shut down the output, reducing supply current to 10nA. Connect SHDN to a logic-high, or IN, for normal operation.

Internal P-Channel Pass Transistor

The MAX8532 features a 1 Ω P-channel MOSFET pass transistor. A P-channel MOSFET provides several advantages over similar designs using PNP pass transistors, including longer battery life. It requires no base drive, reducing quiescent current. PNP-based regulators waste considerable current in dropout when the pass transistor saturates and also use high base-drive currents under heavy loads. The MAX8532 does not suffer these problems and consumes only 90 μ A quiescent current whether in dropout, light-load, or heavy-load applications (see the *Typical Operating Characteristics*). Whereas a PNP-based regulator has dropout voltage independent of the load, a P-channel MOSFET's dropout voltage is proportional to load current, providing for low dropout voltage at heavy loads and extremely low dropout voltage at lighter loads.

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Current Limit

The MAX8532 contains an independent current limiter, which monitors and controls the pass transistor's gate voltage, limiting the output current to 210mA (min). The output can be shorted to ground indefinitely without damaging the part.

Thermal-Overload Protection

Thermal-overload protection limits total power dissipation in the MAX8532. When the junction temperature exceeds $T_J = +160^{\circ}\text{C}$, the thermal sensor signals the shutdown logic, turning off the pass transistor and allowing the IC to cool. The thermal sensor turns the pass transistor on again after the IC's junction temperature cools by 10°C , resulting in a pulsed output during continuous thermal-overload conditions.

Thermal-overload protection is designed to protect the MAX8532 in the event of fault conditions. For continual operation, do not exceed the absolute maximum junction temperature rating of $T_J = +150^{\circ}\text{C}$.

Operating Region and Power Dissipation

The MAX8532's maximum power dissipation depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the airflow rate. The power dissipation across the device is $P = I_{OUT} (V_{IN} - V_{OUT})$. Maximum power dissipation:

$$P_{MAX} = (T_J - T_A) / (\theta_{JB} + \theta_{BA})$$

where $T_J - T_A$ is the temperature difference between the MAX8532 die junction and the surrounding air, θ_{JB} (or θ_{JC}) is the thermal resistance of the package, and θ_{BA} is the thermal resistance through the printed circuit board, copper traces, and other materials, to the surrounding air.

Low-Noise Operation

An external $0.01\mu\text{F}$ bypass capacitor at BP, in conjunction with an internal resistor, creates a lowpass filter. The MAX8532 exhibits $40\mu\text{V}_{\text{RMS}}$ output voltage noise with $C_{BP} = 0.01\mu\text{F}$ and $C_{OUT} = 2.2\mu\text{F}$ (see the Output Noise Spectral Density vs. Frequency graph in the *Typical Operating Characteristics*).

Applications Information

Capacitor Selection and Regulator Stability

Use a $2.2\mu\text{F}$ capacitor on the MAX8532's input. Larger input capacitor values with lower ESR provide better supply-noise rejection and line-transient response. To reduce noise and improve load transients, use large output capacitors up to $10\mu\text{F}$. For stable operation over the full temperature range and with rated maximum load currents, use a minimum of $2.2\mu\text{F}$ (or $1\mu\text{F}$ for $<150\text{mA}$ loading) for OUT.

Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, use $4.7\mu\text{F}$ or more to ensure stability at temperatures below -10°C . With X7R or X5R dielectrics, $2.2\mu\text{F}$ is sufficient at all operating temperatures. These regulators are optimized for ceramic capacitors. Tantalum capacitors are not recommended.

PSRR and Operation from Sources Other than Batteries

The MAX8532 is designed to deliver low dropout voltages and low quiescent currents in battery-powered systems. Power-supply rejection is 62dB at low frequencies (see the Power-Supply Rejection Ratio vs. Frequency graph in the *Typical Operating Characteristics*).

When operating from sources other than batteries, improve supply-noise rejection and transient response by increasing the values of the input and output bypass capacitors and through passive filtering techniques.

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Load-Transient Considerations

The MAX8532 load-transient response graphs (see the *Typical Operating Characteristics*) show two components of the output response: a DC shift in the output voltage due to the different load currents, and the transient response. Increase the output capacitor's value and decrease its ESR to attenuate transient spikes.

Input/Output (Dropout Voltage)

A regulator's minimum input/output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX8532 uses a P-channel MOSFET pass transistor, its dropout voltage is a function of drain-to-source on-resistance ($R_{DS(ON)}$) multiplied by the load current (see the *Typical Operating Characteristics*).

Calculating the Maximum Output Power in UCSP

The maximum output power of the MAX8532 can be limited by the maximum power dissipation of the package.

Ascertain the maximum power dissipation by calculating the power dissipation of the package as a function of the input voltage, output voltage, and output current. The maximum power dissipation should not exceed the package's maximum power rating:

$$P = (V_{IN(MAX)} - V_{OUT}) \times I_{OUT}$$

where:

$V_{IN(MAX)}$ = Maximum input voltage

P_{MAX} = Maximum power dissipation of the package (308mW for UCSP)

V_{OUT} = Output voltage of OUT

I_{OUT} = Maximum output current of OUT

P should be less than P_{MAX} .

Chip Information

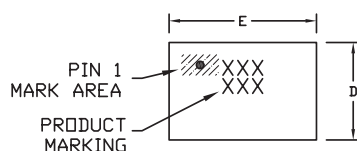
TRANSISTOR COUNT: 1320

PROCESS: BiCMOS

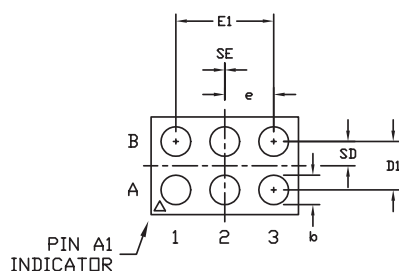
Low-Noise, Low-Dropout, 200mA Linear Regulator in UCSP

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)



TOP VIEW



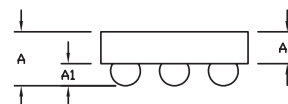
BOTTOM VIEW

COMMON DIMENSIONS	
A	0.60±0.05
A1	0.27±0.03
A2	0.33 REF.
b	Ø0.37 BASIC
D1	0.50 BASIC
E1	1.00 BASIC
e	0.50 BASIC
SD	0.25 BASIC
SE	0.00 BASIC

PKG. CODE	VARIABLE DIMENSIONS		DEPOPULATED SOLDER BALLS
	D	E	
B6-1	1.00±0.05	1.52±0.05	NONE
B6-2	1.00±0.05	1.52±0.05	B2
B6-3	1.05±0.05	1.57±0.05	NONE
B6-4	1.05±0.05	1.57±0.05	B2
B6-5	0.97±0.05	1.46±0.05	NONE
B6-6	1.16±0.05	1.57±0.05	NONE

NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. PRODUCT MARKING: NUMBER OF CHARACTERS AND LINES VARY PER PRODUCT.



SIDE VIEW

MAXIM			
PROPRIETARY INFORMATION			
TITLE: PACKAGE OUTLINE, 3x2 UCSP			
APPROVAL	DOCUMENT CONTROL NO. 21-0097	REV. F	1/1

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